

## Lecture 24, Tuesday, March 1, 2022

- Frequency response of LTI systems:  $H(\omega)$

- In general

$$f(t) = \operatorname{Re} \{ F e^{j\omega t} \} \rightarrow \boxed{\text{LTI}} \rightarrow y(t) = \operatorname{Re} \{ Y e^{j\omega t} \},$$

then the phasors follow

$$F \rightarrow \boxed{H(\omega)} \rightarrow Y = FH(\omega).$$

- \* The output phasor is simply the product of the input phasor,  $F$ , and the frequency response of the LTI system,  $H(\omega)$ , hence

$$H(\omega) = \frac{Y}{F} = \frac{\text{output phasor}}{\text{input phasor}}$$

- General first and second order filters

- Lowpass

$$|H(\omega)| \approx \frac{1}{\sqrt{1 + \omega^2}} \quad |H(0)| \approx 1 \quad |H(\infty)| \approx 0 \quad |H(\omega)| \searrow \text{ as } \omega \rightarrow \infty$$

- Highpass

$$|H(\omega)| \approx \frac{\omega}{\sqrt{1 + \omega^2}} \quad |H(0)| \approx 0 \quad |H(\infty)| \approx 1 \quad |H(\omega)| \nearrow \text{ as } \omega \rightarrow \infty$$

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- Bandpass

$$|H(\omega)| \approx \frac{\omega}{\sqrt{\omega^2 + (1 - \omega^2)^2}} \quad |H(0)| \approx 0 \quad |H(\infty)| \approx 0 \quad |H(\omega)| \nearrow \text{and then } \searrow \text{ as } \omega \rightarrow \infty$$

- $H(\omega)$  is only meaningful for dissipative LTI systems
- Properties of  $H(\omega)$  for real-valued systems:

- Conjugate symmetry

$$H(-\omega) = H^*(\omega)$$

- Even amplitude response

$$|H(-\omega)| = |H(\omega)|$$

- Odd phase response

$$\angle H(-\omega) = -\angle H(\omega)$$

- Real-valued D.C. response

$$H(0) \in \mathbb{R}$$

- Steady-state response to complex exponential

$$f(t) = Ae^{j(\omega t + \theta)} \longrightarrow \boxed{H(\omega)} \longrightarrow y(t) = H(\omega)Ae^{j(\omega t + \theta)}$$

- NOTE: correction in solutions to phase response in example # 3.