

Lecture

Adaptive Filters



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Lecture 3: Linear Prediction



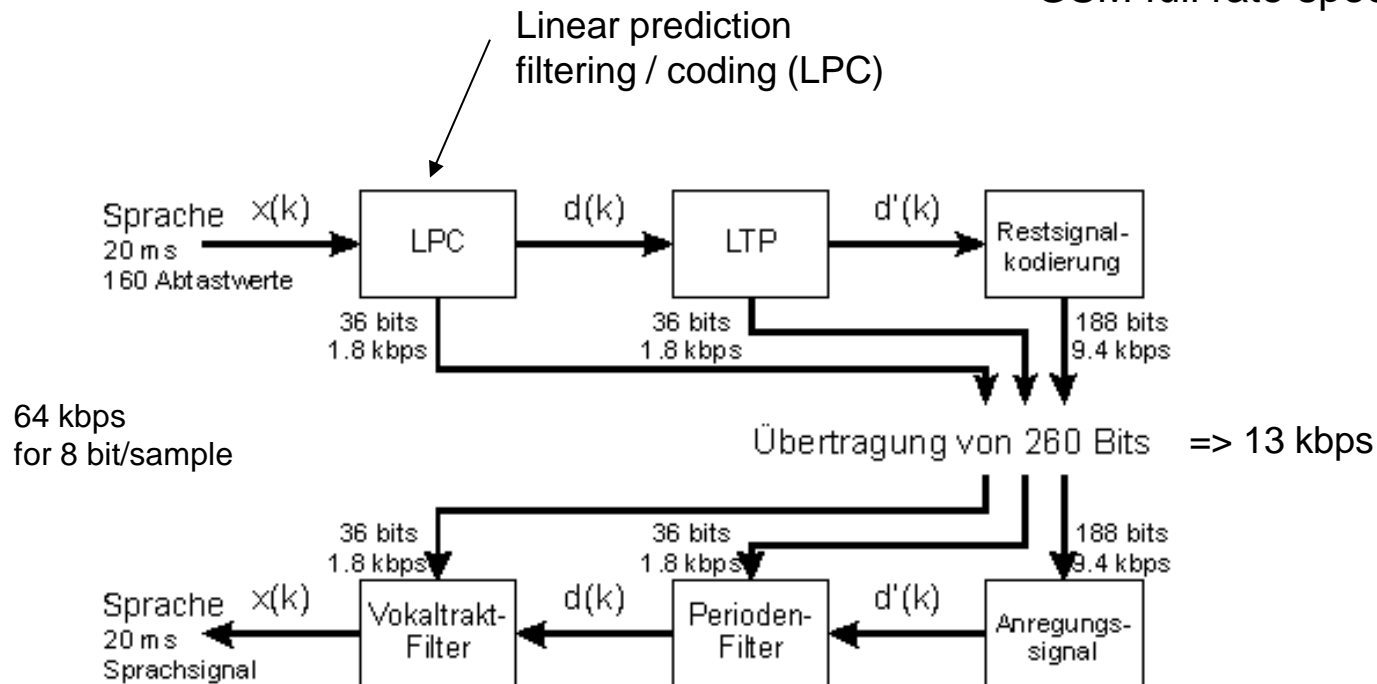
- ❑ Linear prediction is one very strong instrument of adaptive signal processing
- ❑ Examples:
 - ❑ Source coding: eliminate redundant information
 - ❑ Speech signal processing: determine vocal tract filter
- ❑ Derivation of the optimum prediction filter coefficients based on known Wiener filter
- ❑ Prediction error analysis: power and correlation
- ❑ Examples of reduced amplitudes => source coding
- ❑ Examples of spectral envelope estimation
- ❑ Levinson-Durbin recursion
- ❑ Lattice filter structures

Linear prediction: Application example (I): Source coding

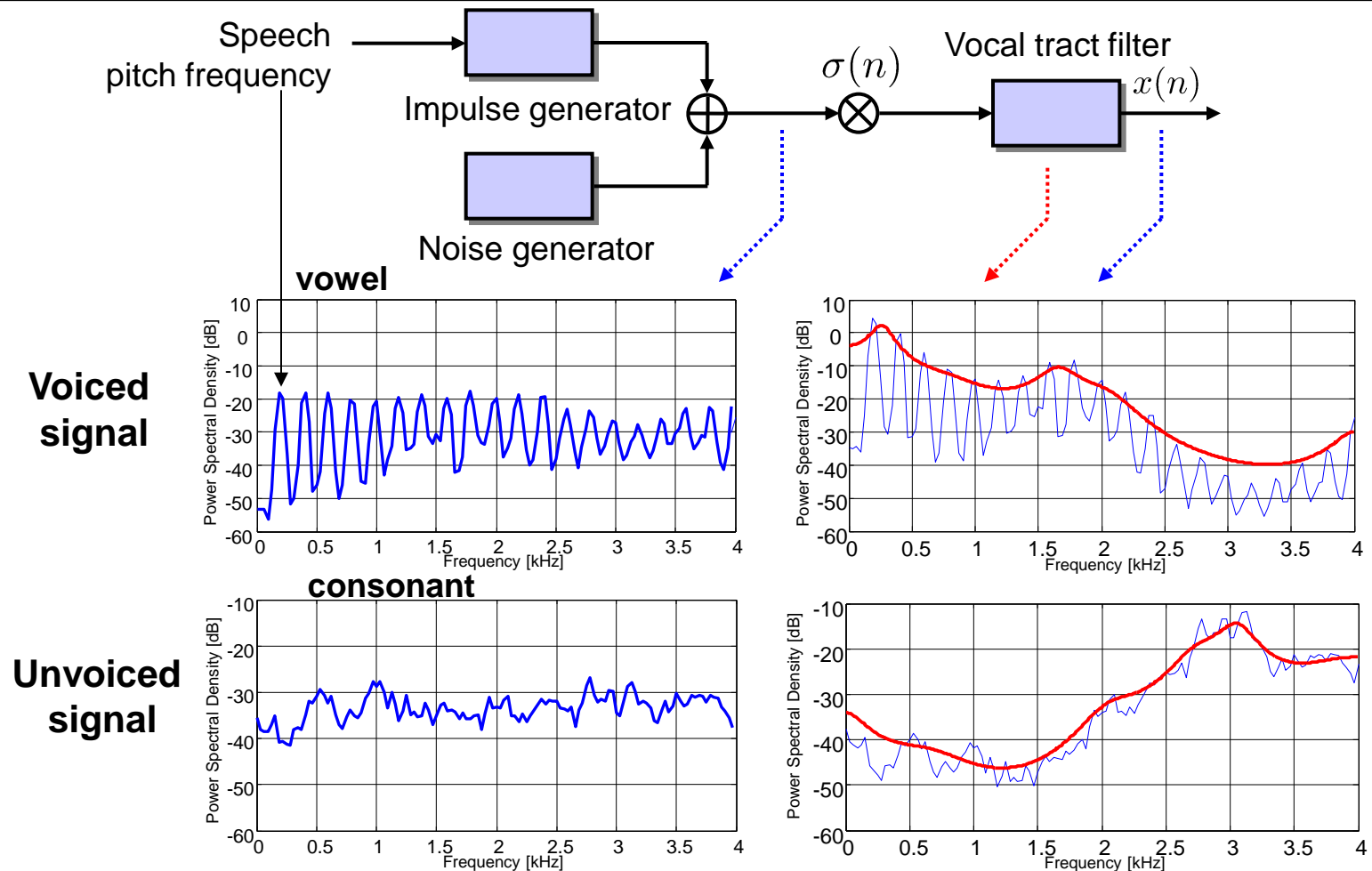
Target:

Reduce the amount of data to be transmitted by removing redundant information

GSM full rate speech coder



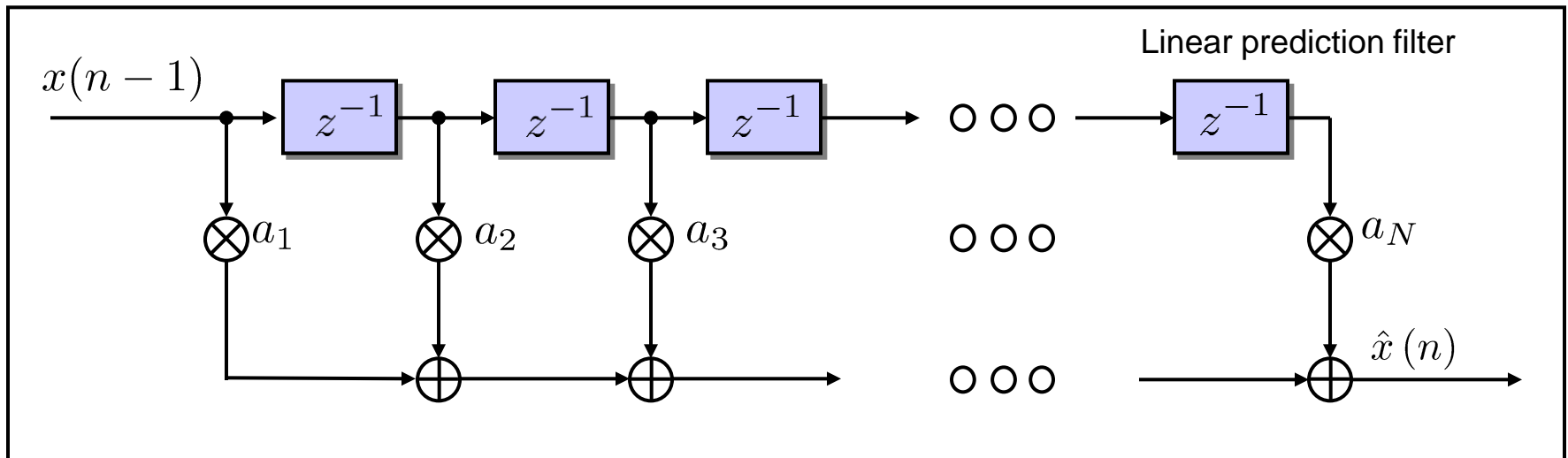
Linear prediction: Application example (II): Estimation of vocal tract filters



Approach

Prediction of the current signal sample based on the last N signal samples:

$$\hat{x}(n) = \sum_{i=1}^N a_i x(n-i)$$



With:

- $\hat{x}(n)$: Estimation for $x(n)$
- a_i : prediction coefficients

- N : Length / order of the prediction filter

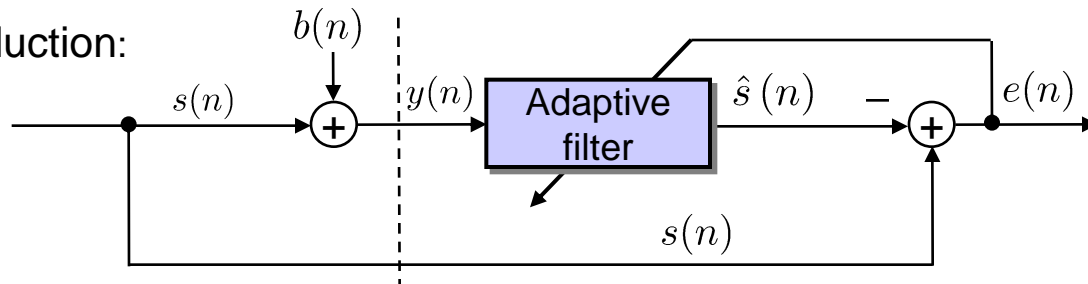
Wiener Filter

pergunta? pq colocar



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□ Noise reduction:



$$E\{e^2(n)\} = r_{ss}(0) - 2\hat{\mathbf{h}}^T \mathbf{r}_{ys}(0) + \hat{\mathbf{h}}^T \mathbf{R}_{yy} \hat{\mathbf{h}}$$

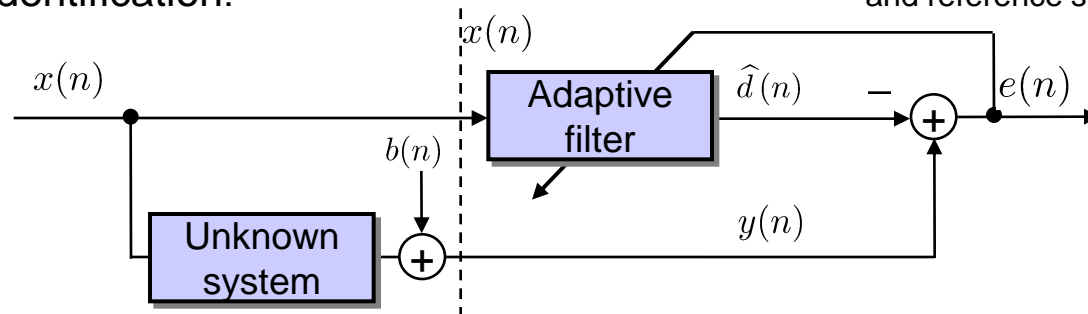
$$E_{\min} = r_{ss}(0) - \mathbf{r}_{ys}^T(0) \mathbf{R}_{yy}^{-1} \mathbf{r}_{ys}(0)$$

autocorrelation of
adaptive filter input

cross-correlation of
adaptive filter input
and reference signal

$$\hat{\mathbf{h}}_{\text{opt}} = \mathbf{R}_{yy}^{-1} \mathbf{r}_{ys}(0)$$

□ System identification:



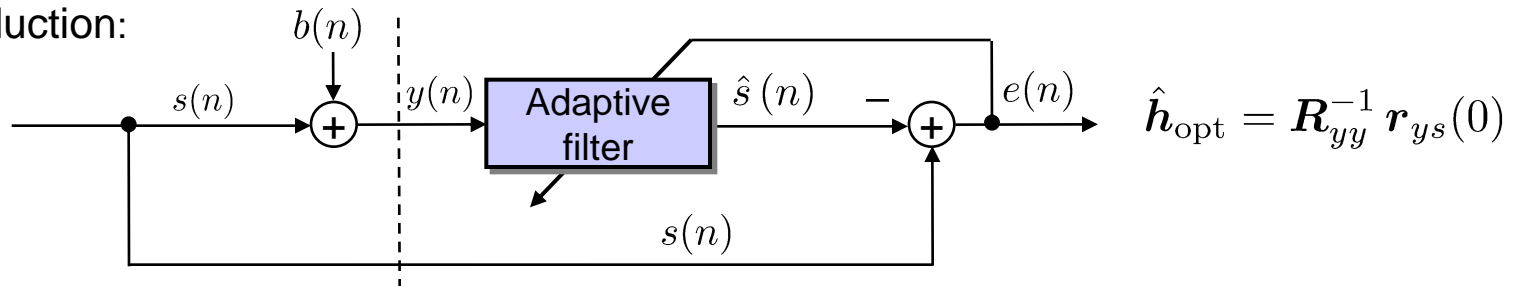
$$E\{e^2(n)\} = r_{yy}(0) - 2\hat{\mathbf{h}}^T \mathbf{r}_{xy}(0) + \hat{\mathbf{h}}^T \mathbf{R}_{xx} \hat{\mathbf{h}}$$

$$E_{\min} = r_{yy}(0) - \mathbf{r}_{xy}^T(0) \mathbf{R}_{xx}^{-1} \mathbf{r}_{xy}(0)$$

$$\hat{\mathbf{h}}_{\text{opt}} = \mathbf{R}_{xx}^{-1} \mathbf{r}_{xy}(0)$$

Wiener Filter

□ Noise reduction:



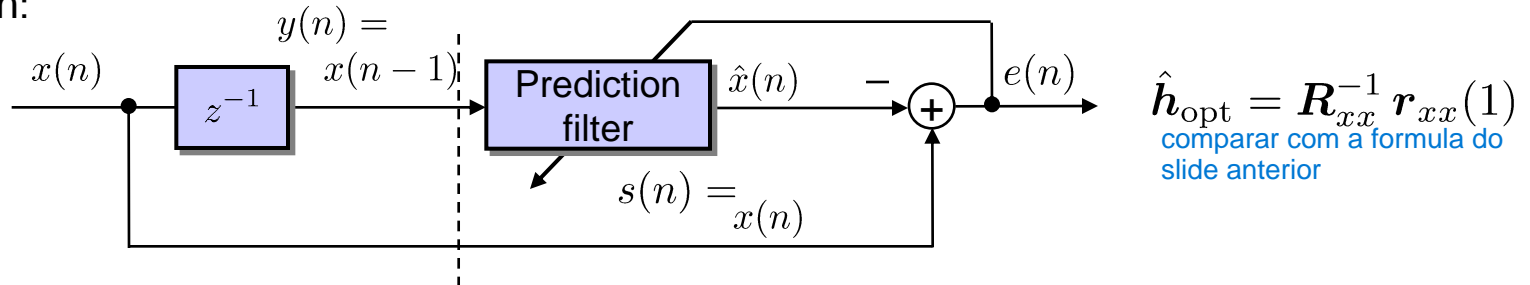
$$\mathbf{r}_{ys}(0) = [r_{ys}(0), r_{ys}(1), \dots, r_{ys}(N-1)]^T$$

$$\mathbf{r}_{ys}(0) = [\mathbb{E}\{y(n)s(n)\}, \mathbb{E}\{y(n)s(n+1)\}, \dots, \mathbb{E}\{y(n)s(n+N-1)\}]^T$$

pergunta: pq atrasar em 1 amostra só pra poder calcular ela mesma?

a potência do erro é menor e também se distribui melhor na frequência - outras análises interessantes - quantização do erro é menor

□ Prediction:

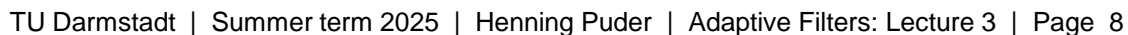


$$\mathbf{r}_{ys}(0) = [\mathbb{E}\{x(n-1)x(n)\}, \mathbb{E}\{x(n-1)x(n+1)\}, \dots, \mathbb{E}\{x(n-1)x(n+N-1)\}]^T$$

$$= [r_{xx}(1), r_{xx}(2), \dots, r_{xx}(N)]^T = \mathbf{r}_{xx}(1)$$

pergunta: pq pensar o vetor de autocorrelação é usado começando em 1 e indo até N?

$$\begin{bmatrix} a_1 \\ a_2 \\ \vdots \\ a_N \end{bmatrix} = \begin{bmatrix} r_{xx}(0) & r_{xx}(1) & \cdots & r_{xx}(N-1) \\ r_{xx}(1) & r_{xx}(0) & \cdots & r_{xx}(N-2) \\ \vdots & \vdots & \ddots & \vdots \\ r_{xx}(N-1) & r_{xx}(N-2) & \cdots & r_{xx}(0) \end{bmatrix}^{-1} \begin{bmatrix} r_{xx}(1) \\ r_{xx}(2) \\ \vdots \\ r_{xx}(N) \end{bmatrix}$$



Example: White input signal

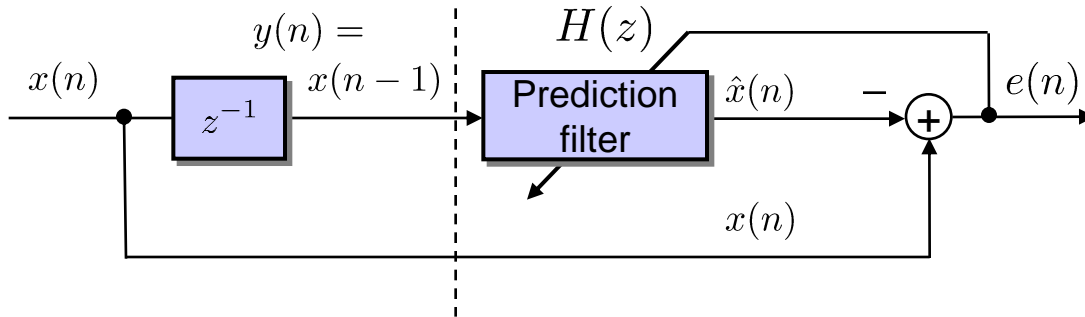
- Input signal $x(n)$: white noise with the power σ_0^2 (mean value = 0)
- Prediction order: $N = 3$
- Prediction by one sample: $L = 1$

Resulting in:

$$\mathbf{R}_{xx} = \begin{bmatrix} \sigma_0^2 & 0 & 0 \\ 0 & \sigma_0^2 & 0 \\ 0 & 0 & \sigma_0^2 \end{bmatrix} \quad \mathbf{R}_{xx}^{-1} = \frac{1}{\sigma_0^2} \begin{bmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 1 \end{bmatrix}$$
$$\mathbf{r}_{xx}(1) = [0, 0, 0]^T$$
$$\mathbf{a} = \mathbf{R}_{xx}^{-1} \mathbf{r}_{xx}(1) = [0, 0, 0]^T \quad \text{i.e., no prediction is possible, signal does not contain redundant information.}$$

toda a informação de $x(n)$ NÃO pode ser prevista a partir de amostras anteriores -> $x(N)$ carrega nenhuma informação
É melhor não fazer nada quando o sinal é branco, pq nada pode ser previsto de $x(n)$ a partir das amostras anteriores

Prediction: Error signal power



$$\hat{\mathbf{h}}_{\text{opt}} = \mathbf{R}_{xx}^{-1} \mathbf{r}_{xx}(1)$$

$$H(z) = \sum_{i=0}^{N-1} \hat{h}_{\text{opt},i} z^{-i} = \sum_{i=0}^{N-1} a_{i+1} z^{-i}$$

$$e(n) = x(n) - \sum_{i=0}^{N-1} \hat{h}_{\text{opt},i} x(n-i-1) = x(n) - \sum_{i=1}^N a_i x(n-i)$$

Wiener filter (known from last lecture):

$$\mathbb{E} \{e^2(n)\} = r_{yy}(0) - 2 \hat{\mathbf{h}}^T \mathbf{r}_{xy}(0) + \hat{\mathbf{h}}^T \mathbf{R}_{xx} \hat{\mathbf{h}}$$

$$\mathbb{E}_{\min} = r_{yy}(0) - \mathbf{r}_{xy}^T(0) \mathbf{R}_{xx}^{-1} \mathbf{r}_{xy}(0)$$

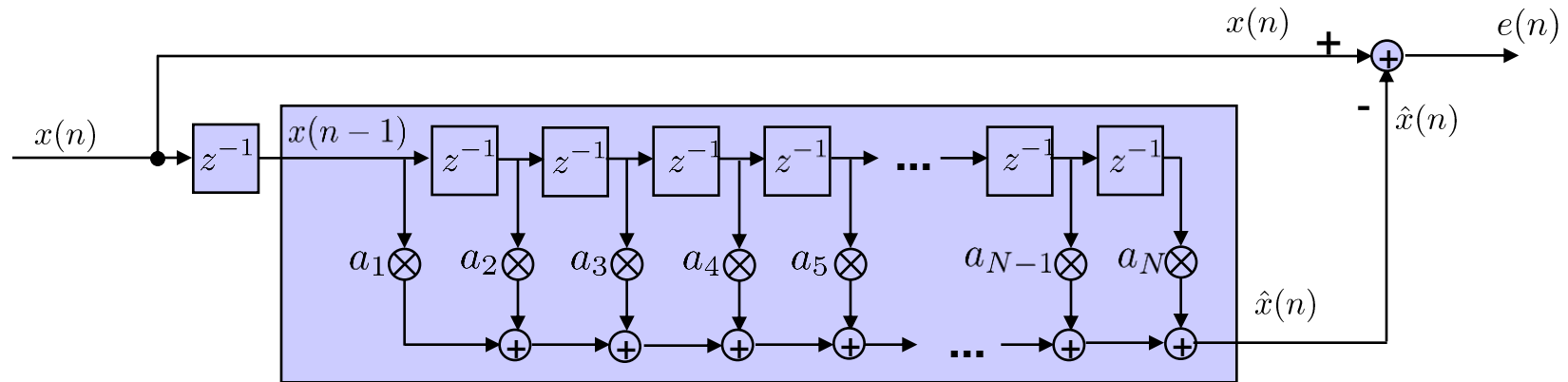
Predictor:

$$\mathbb{E} \{e^2(n)\} = r_{xx}(0) - 2 \mathbf{a}^T \mathbf{r}_{xx}(1) + \mathbf{a}^T \mathbf{R}_{xx} \mathbf{a}$$

$$\mathbb{E}_{\min} = r_{xx}(0) - \mathbf{r}_{xx}^T(1) \mathbf{R}_{xx}^{-1} \mathbf{r}_{xx}(1)$$

$$\mathbb{E}_{\min} = r_{xx}(0) - \mathbf{r}_{xx}^T(1) \mathbf{a}$$

Prediction error gain



$$E_{\min} = r_{xx}(0) - \mathbf{r}_{xx}^T(1) \mathbf{R}_{xx}^{-1} \mathbf{r}_{xx}(1)$$

=> error signal power equal or reduced compared to input signal power.

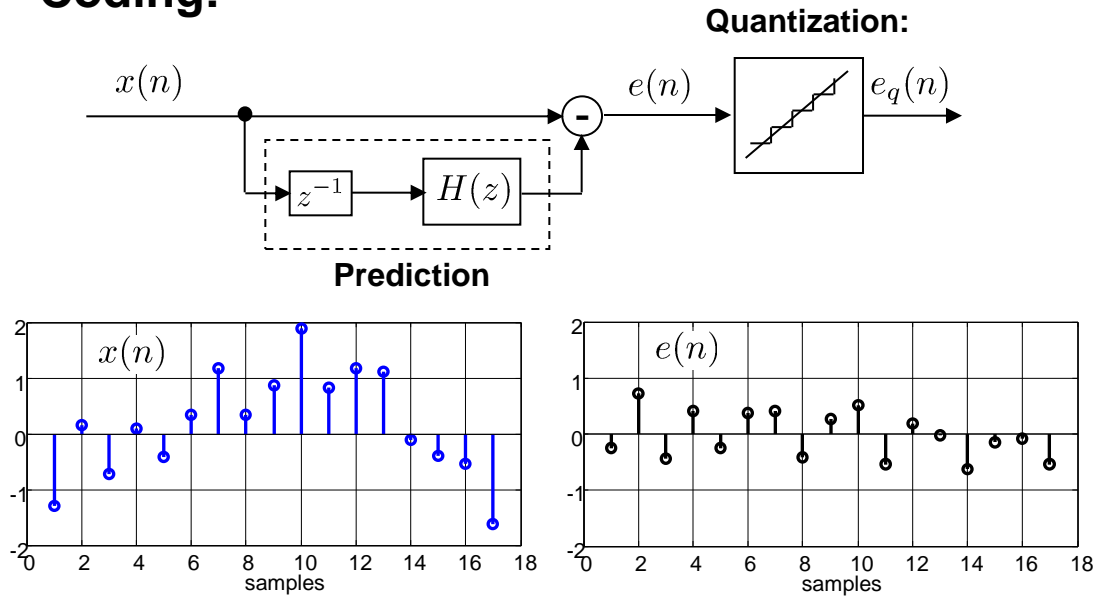
Prediction error gain:

$$\frac{E\{x^2(n)\}}{E_{\min}} = \frac{r_{xx}(0)}{r_{xx}(0) - \mathbf{r}_{xx}^T(1) \mathbf{R}_{xx}^{-1} \mathbf{r}_{xx}(1)}$$

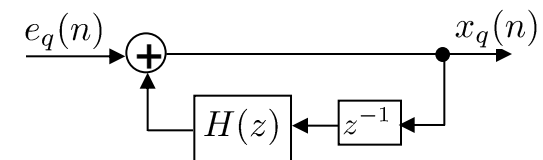
RANGE: [1, +inf [

Linear prediction application example: source coding

Coding:



Decoding:

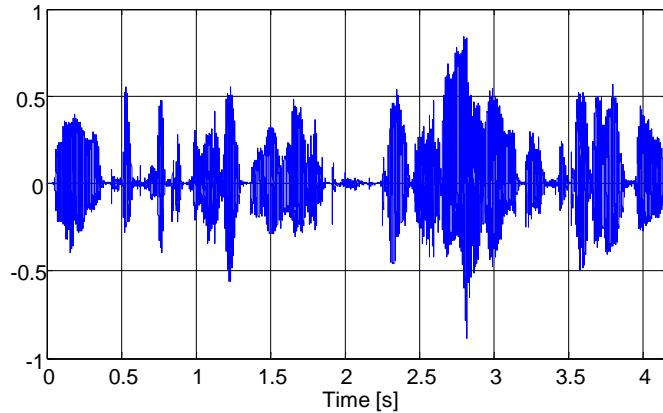


Reduced amplitude of signal to code
 \Rightarrow reduced number of coding bit required

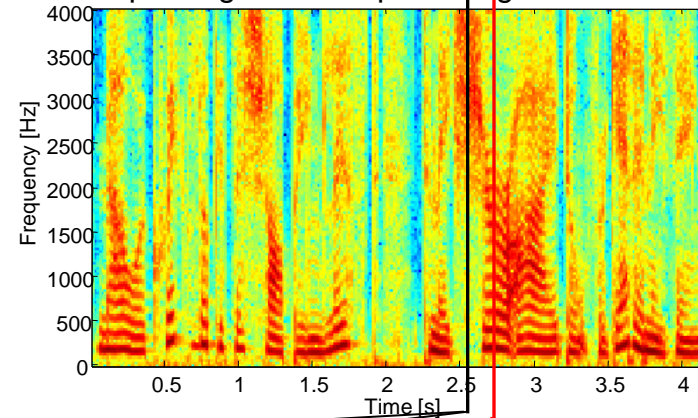
Example: Power reduced by 6 dB \Rightarrow 1 Bit less per signal sample required at quantization

Linear prediction: source coding

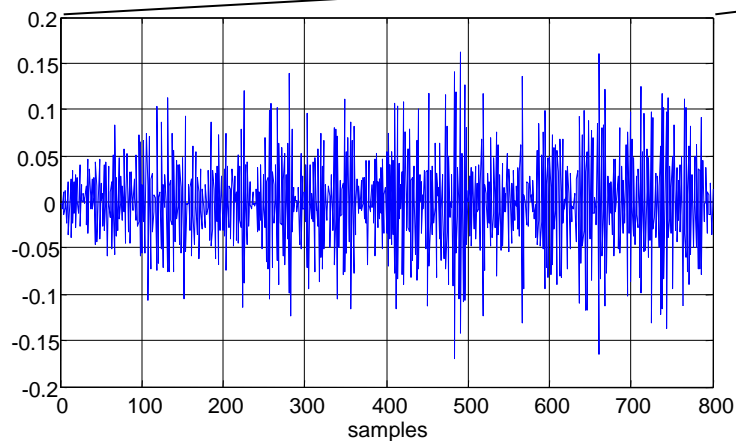
Speech signal:



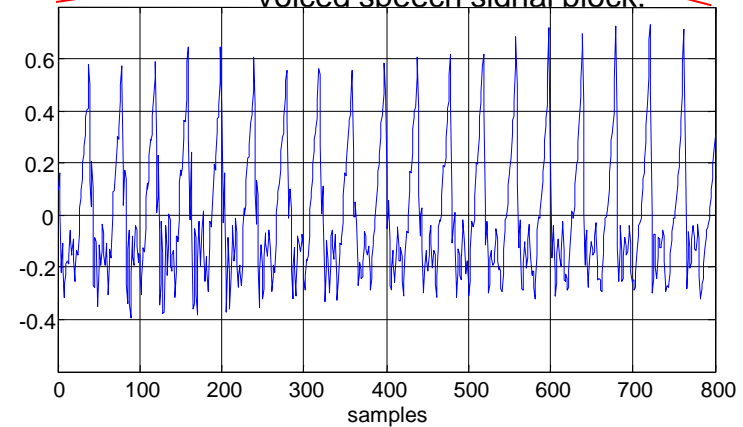
Spectrogram of a speech signal:



Unvoiced speech signal block:

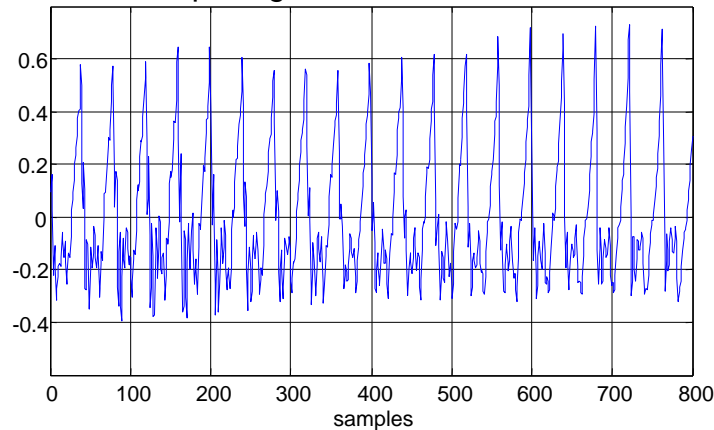


Voiced speech signal block:

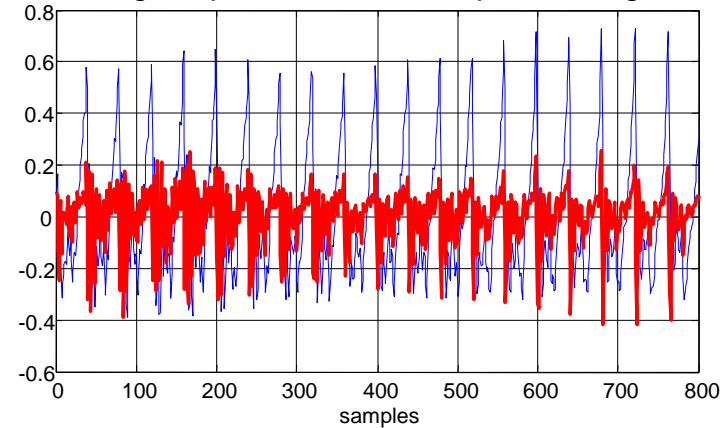


Linear prediction: source coding

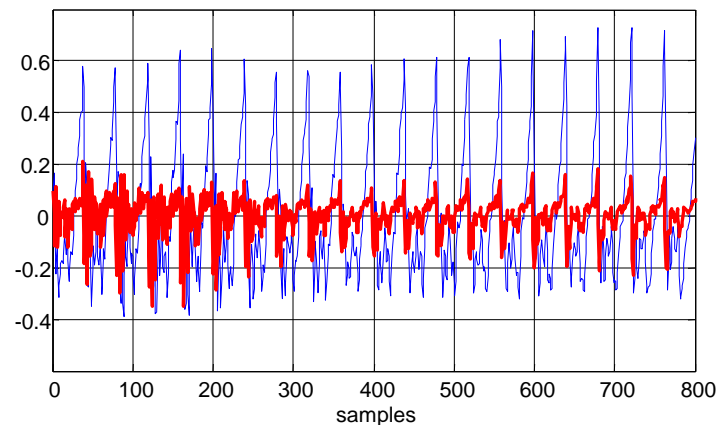
Voiced input signal block:



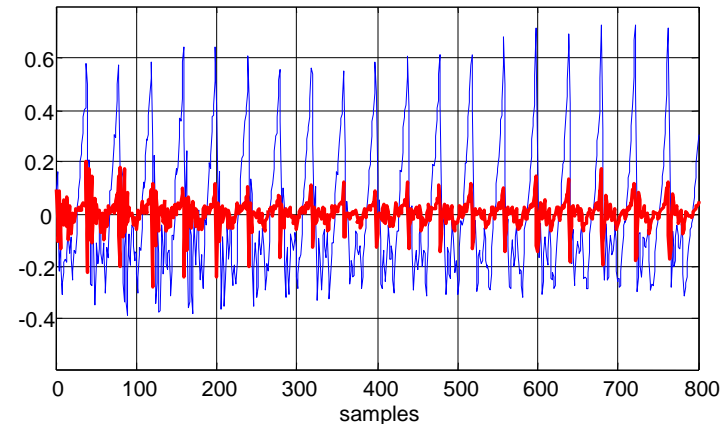
Error signal: prediction order 1; prediction gain: 7.2 dB



Error signal: prediction order 2; prediction gain: 10.4 dB

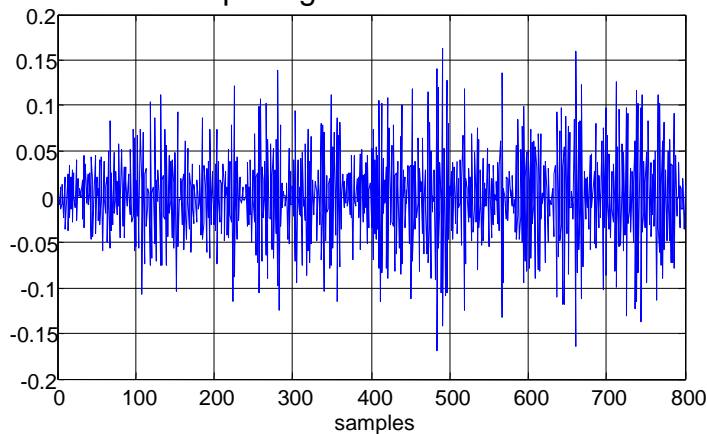


Error signal: prediction order 10; prediction gain: 13.6 dB

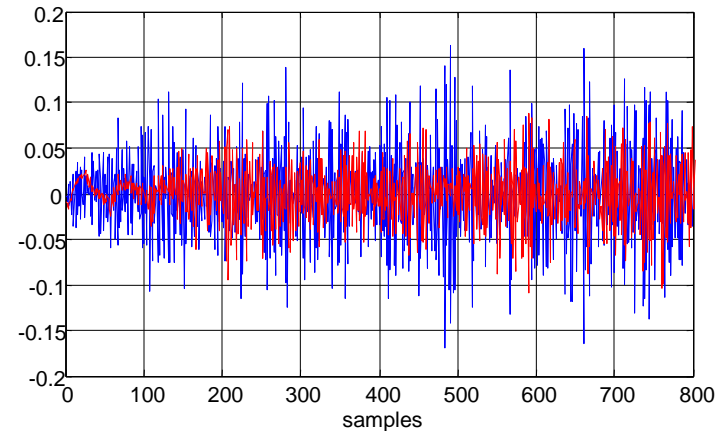


Linear prediction: source coding

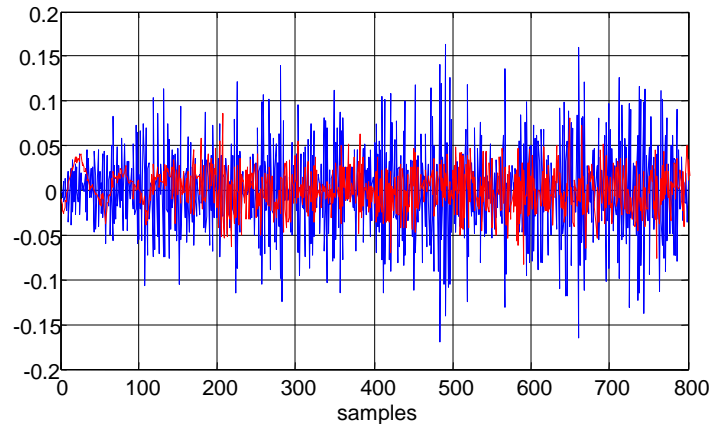
Unvoiced input signal block:



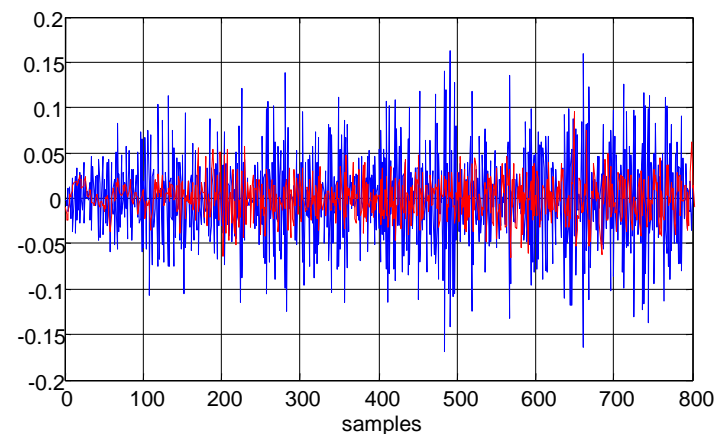
Error signal: prediction order 1; prediction gain: 3.0 dB



Error signal: prediction order 2; prediction gain: 7.5 dB



Error signal: prediction order 6; prediction gain: 8.9 dB





Correlation analysis of the prediction error signal

quando não há correlação/quando não há redundância -> sinal branco

se vc consegue remover toda a redundância então não haveria correlação entre as amostras de erro -> erro seria branco

$$\begin{aligned} r_{ee}(l) &= E\left\{\left(x(n) - \sum_{i=1}^N a_i x(n-i)\right) \left(x(n+l) - \sum_{j=1}^N a_j x(n-j+l)\right)\right\} \\ &= r_{xx}(l) - \sum_{i=1}^N a_i r_{xx}(l-i) - \sum_{i=1}^N a_i r_{xx}(l+i) + \sum_{i=1}^N \sum_{j=1}^N a_i a_j r_{xx}(l+i-j) \\ &= r_{xx}(l) - \underbrace{\sum_{i=1}^N a_i r_{xx}(l-i)}_{=0 \text{ for } l=1, \dots, N} - \sum_{i=1}^N a_i \left[\underbrace{r_{xx}(l+i) - \sum_{j=1}^N a_j r_{xx}(l+i-j)}_{=0 \text{ for } l+i=1, \dots, N} \right] \end{aligned}$$

with the relation for optimum prediction coefficients:

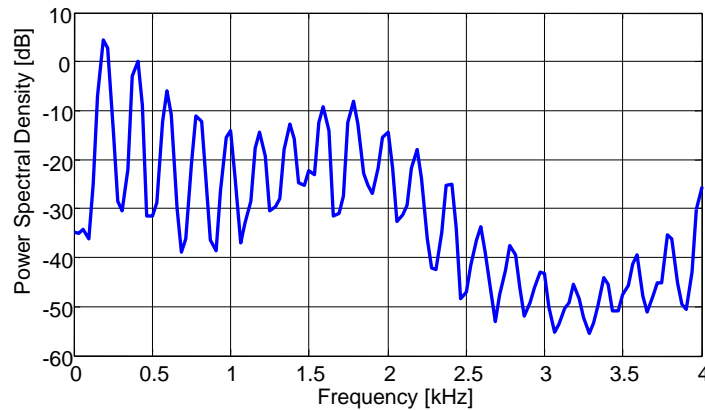
$$r_{xx}(l) \stackrel{!}{=} \sum_{i=1}^N a_i r_{xx}(l-i) \quad \text{for } l=1, \dots, N$$

$$r_{ee}(l) = 0 \quad \text{for } |l| > 0 \text{ and } N \rightarrow \infty$$

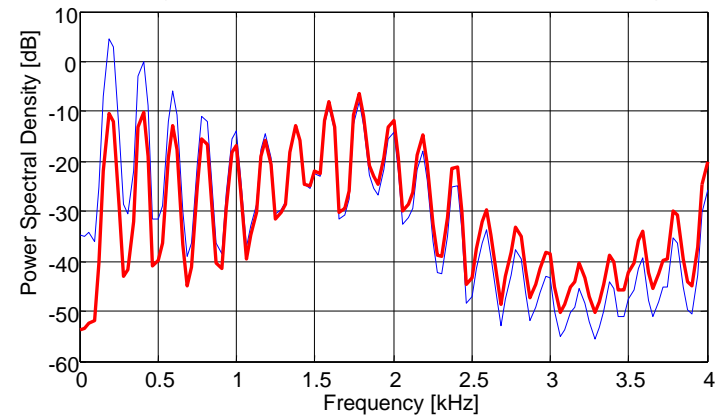
=> whitening of output signal spectrum

Linear prediction: spectral envelope calculation

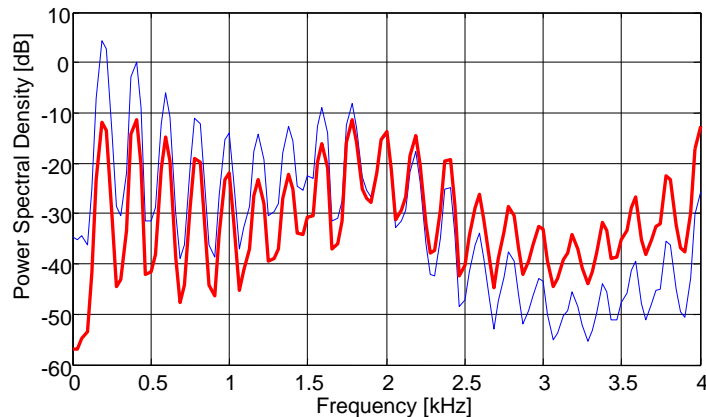
Voiced input signal block:



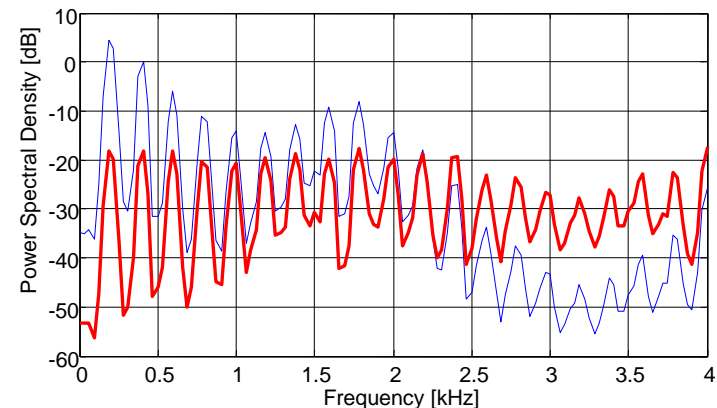
Error signal: prediction order 1; prediction gain: 7.2 dB



Error signal: prediction order 2; prediction gain: 10.4 dB

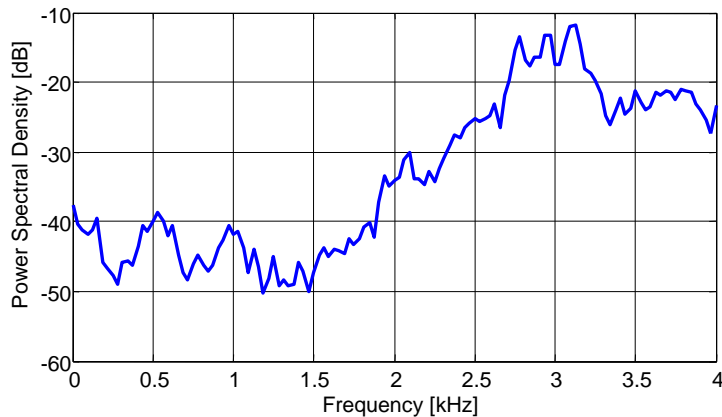


Error signal: prediction order 10; prediction gain: 13.6 dB

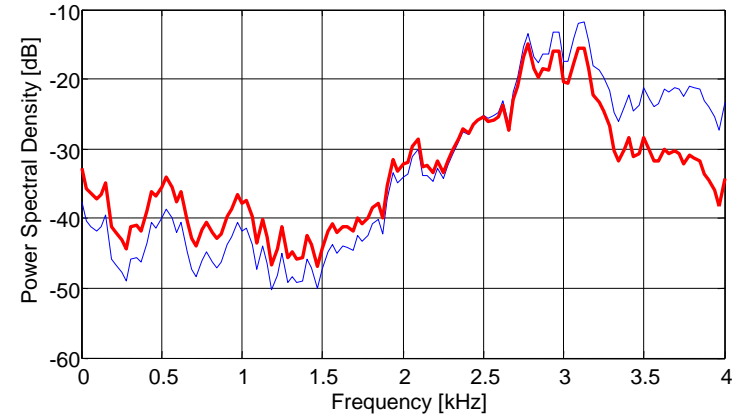


Linear prediction: spectral envelope calculation

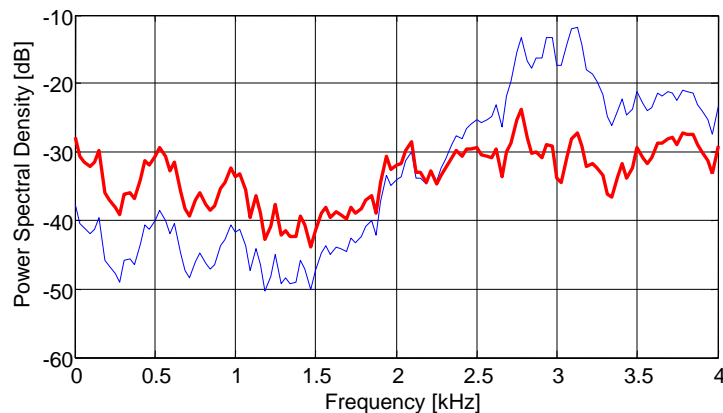
Unvoiced input signal block:



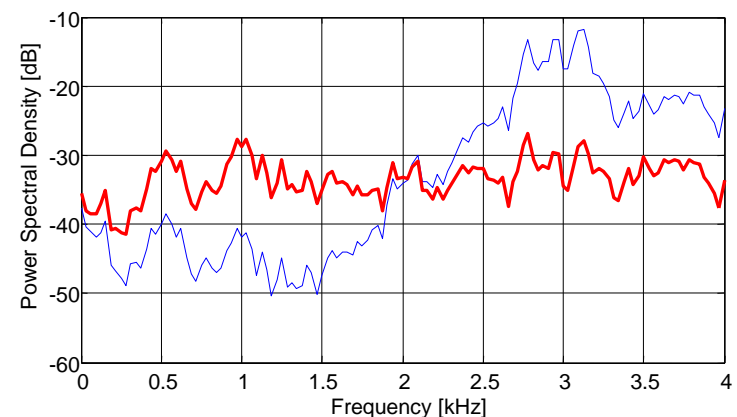
Error signal: prediction order 1; prediction gain: 3.0 dB



Error signal: prediction order 2; prediction gain: 7.5 dB



Error signal: prediction order 6; prediction gain: 8.9 dB

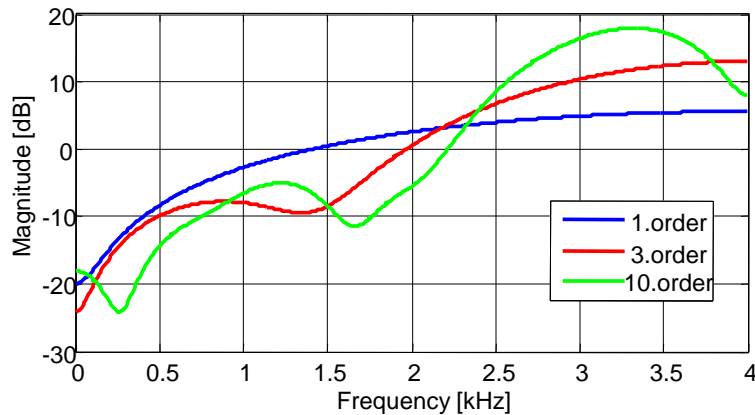




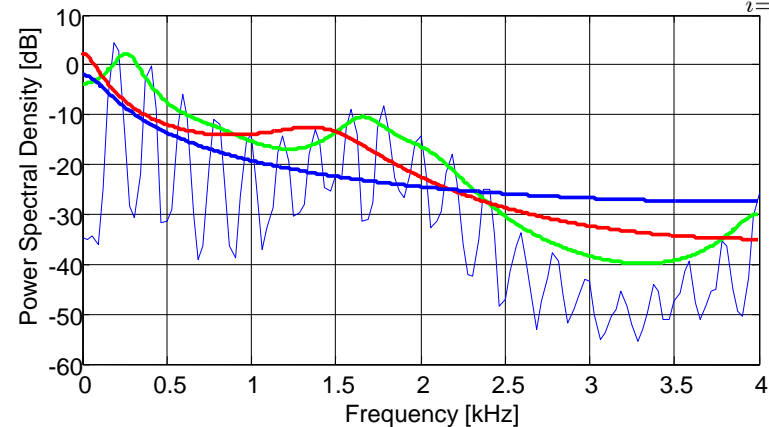
Linear prediction: spectral envelope calculation

PEF modela o INVERSO do envelope ----- O inverso do PEF modela o envelope ->
Explicação: PEF faz com que o sinal*PEF = sinal branco = constante

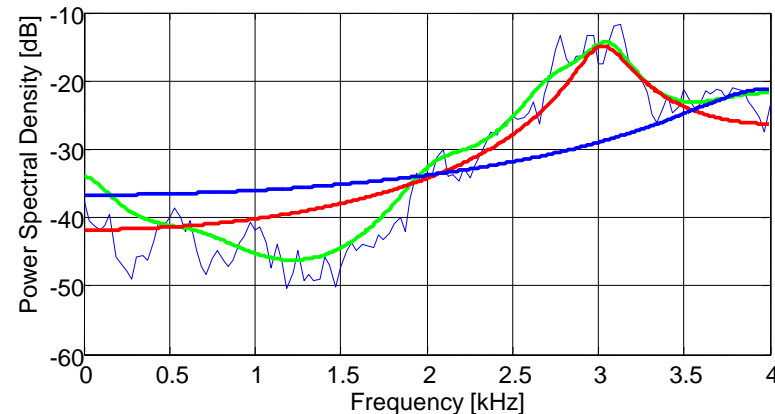
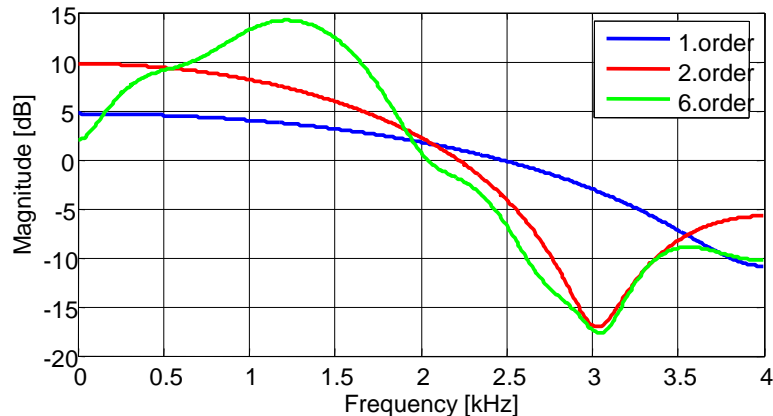
Voiced input signal block: $H_{\text{PEF}}(e^{j\Omega}) = 1 - \sum_{i=1}^N a_i e^{-j\Omega i}$
Prediction error filter:
moving average



Inverse prediction error filter / approx. the PSD: $H_{\text{inv. PEF}}(e^{j\Omega}) = \frac{1}{1 - \sum_{i=1}^N a_i e^{-j\Omega i}}$



Unvoiced input signal block:
Prediction error filter:



Estimation of the autocorrelation function

Problem:

Ensemble averages are unknown in most applications.

Remedy:

Assume ergodic processes: replace ensemble averages by time averages:

$$E\{x(n) x(n + l)\} \quad \sum_n x(n) x(n + l)$$

geralmente não é o caso, mas...

?da pra assumir localmente?

geralmente não, não tente fazer isso kkkk use outros processos de estimation

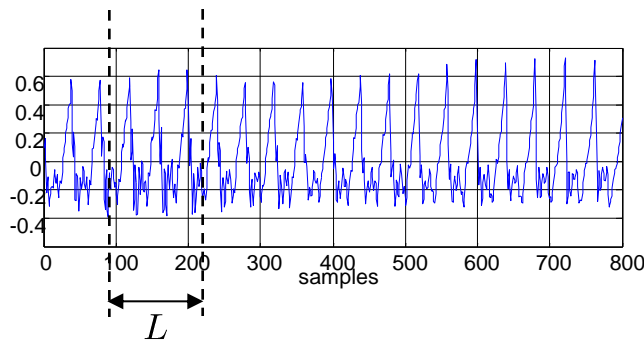
Estimation procedures:

There are some estimation procedures, differing by the properties of the estimated autocorrelation function (biased / unbiased; ACF matrix which is positive-definite)

Estimation of the autocorrelation function

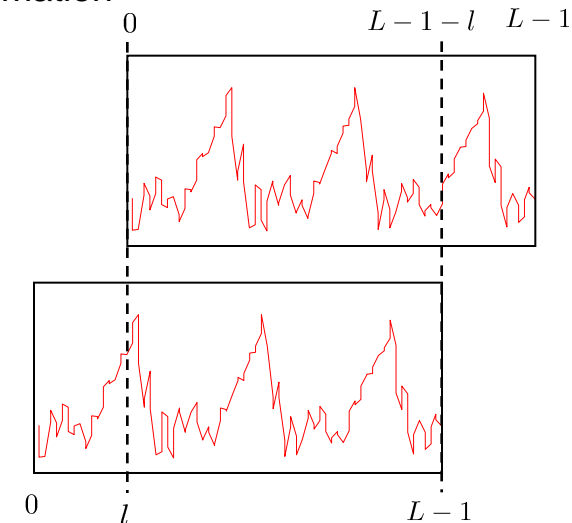
Example – „autocorrelation method“:

Windowing:



Correlation:

=> Time shifted blocks; element-wise multiplication; summation



Calculation:

$$\hat{r}_{xx}(l) = \begin{cases} \frac{1}{L} \sum_{n=0}^{L-1-l} x(n) x(n+l), & \text{for } l \geq 0, \\ \frac{1}{L} \sum_{n=-l}^{L-1} x(n) x(n+l), & \text{for } l < 0 \end{cases}$$

Note: L should be chosen larger than the prediction order N!

Estimation of the autocorrelation function

Properties of the „autocorrelation method“:

- ❑ Biased estimation => Bias calculation:

$$E\{\hat{r}_{xx}(l)\} = \begin{cases} \frac{1}{L} \sum_{n=0}^{L-1-l} E\{x(n)x(n+l)\}, & \text{for } l \geq 0, \\ \frac{1}{L} \sum_{n=-l}^{L-1} E\{x(n)x(n+l)\}, & \text{for } l < 0 \end{cases} = \frac{L-|l|}{L} r_{xx}(l) \leq r_{xx}(l)$$

embora tenha esse problema, esse bias até que ajuda
(acho que no sentido de que a autocorrelação para l maior geralmente é menor mesmo)

- ❑ Properties: $\hat{r}_{xx}(l) = 0$, for $|l| \geq L$
 $\hat{r}_{xx}(l) = \hat{r}_{xx}(-l)$
 $\hat{r}_{xx}(l) \leq \hat{r}_{xx}(0)$

para um prediction order de 10, seria interessante uma janela de mais ou menos 100 ou mais

- ❑ The ACF matrix estimated with the autocorrelation method is positive-definite.
- ❑ The ACF matrix estimated with the autocorrelation method has Toeplitz structure.

pergunta: seria diferente se nós na verdade usássemos o fator correto de normalização?? é exatamente o que ta sendo dito abaixo kkkk

Other methods:

- ❑ Covariance method
- ❑ Modified covariance method

=> Unbiased estimates;
But: ACF matrices have no Toeplitz structure
and are not positive-definite

Levinson-Durbin recursion - Motivation

Problem:

Solving the matrix equation

$$\mathbf{a} = \mathbf{R}_{xx}^{-1} \mathbf{r}_{xx}(1)$$

requires a large computational effort proportional to N^2 in case of a Toeplitz structure and proportional to N^3 otherwise. Additional problems may occur during matrix inversion in case of a bad conditioning of the ACF matrix.

Target:

Robust method which solves the above equation without calculating a matrix inversion of: \mathbf{R}_{xx}

Solution:

Taking advantage of the Toeplitz structure of the matrix \mathbf{R}_{xx} .

- ☐ Recursion with the prediction order
- ☐ Combining forward and backward prediction

Literature (original):

- ☐ J. Durbin: *The Fitting of Time Series Models*, Rev. Int. Stat. Inst., No. 28, pages 233 - 244, 1960
- ☐ N. Levinson: *The Wiener RMS Error Criterion in Filter Design and Prediction*, J. Math. Phys., Nr. 25, Pages 261 - 268, 1947

Levinson-Durbin recursion – General Overview

□ Recursion: Three steps: Initialization, Iteration, Stop

□ **Initialization:** Predictor of order 1:

$$\mathbf{a} = \mathbf{R}_{xx}^{-1} \mathbf{r}_{xx}(1) \quad \Rightarrow \text{Order 1:} \quad a_1 = \frac{r_{xx}(1)}{r_{xx}(0)}$$

□ **Iteration:** Increasing the order:

Order 1: $\mathbf{a}^{(1)} = \overset{\uparrow(1)}{a_1^{(1)}} = \frac{r_{xx}(1)}{r_{xx}(0)}$

Not identical!

Differentiation by upscript (1), (2), ... (N)

Order 2: $\mathbf{a}^{(2)} = \begin{bmatrix} \overset{\uparrow(2)}{a_1^{(2)}} \\ a_2^{(2)} \end{bmatrix} = \begin{bmatrix} r_{xx}(0) & r_{xx}(1) \\ r_{xx}(1) & r_{xx}(0) \end{bmatrix}^{-1} \begin{bmatrix} r_{xx}(1) \\ r_{xx}(2) \end{bmatrix}$

However: Recursion does not solve the Yule-Walker equation (matrix inversion)

but uses $\mathbf{a}^{(N)} = [a_1^{(N)}, a_2^{(N)}, \dots, a_N^{(N)}]^T$ and $r_{xx}(l)$ to generate

$$\mathbf{a}^{(N+1)} = [a_1^{(N+1)}, a_2^{(N+1)}, \dots, a_N^{(N+1)}, a_{N+1}^{(N+1)}]^T$$

□ **Stop:** If iteration target is reached

Levinson-Durbin recursion – Extension of the matrix

(according to Monson H. Hayes: “Statistical Digital Signal Processing and Modeling”)

ele não quer que explique as deduções na prova - talvez não precise decorar com detalhe



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□ Prediction equation:

$$\begin{bmatrix} r_{xx}(0) & r_{xx}(1) & \cdots & r_{xx}(N-1) \\ r_{xx}(1) & r_{xx}(0) & \cdots & r_{xx}(N-2) \\ \vdots & \vdots & \ddots & \vdots \\ r_{xx}(N-1) & r_{xx}(N-2) & \cdots & r_{xx}(0) \end{bmatrix} \begin{bmatrix} a_1 \\ a_2 \\ \vdots \\ a_N \end{bmatrix} = \begin{bmatrix} r_{xx}(1) \\ r_{xx}(2) \\ \vdots \\ r_{xx}(N) \end{bmatrix}$$

$$\begin{bmatrix} r_{xx}(1) & r_{xx}(0) & r_{xx}(1) & \cdots & r_{xx}(N-1) \\ r_{xx}(2) & r_{xx}(1) & r_{xx}(0) & \cdots & r_{xx}(N-2) \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ r_{xx}(N) & r_{xx}(N-1) & r_{xx}(N-2) & \cdots & r_{xx}(0) \end{bmatrix} \begin{bmatrix} -1 \\ a_1 \\ a_2 \\ \vdots \\ a_N \end{bmatrix} = \begin{bmatrix} 0 \\ 0 \\ \vdots \\ 0 \end{bmatrix}$$

with: $E_{\min} = r_{xx}(0) - \mathbf{r}_{xx}^T(1) \mathbf{a}$ a power value

$$\Rightarrow \begin{bmatrix} r_{xx}(0) & r_{xx}(1) & r_{xx}(2) & \cdots & r_{xx}(N) \\ r_{xx}(1) & r_{xx}(0) & r_{xx}(1) & \cdots & r_{xx}(N-1) \\ r_{xx}(2) & r_{xx}(1) & r_{xx}(0) & \cdots & r_{xx}(N-2) \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ r_{xx}(N) & r_{xx}(N-1) & r_{xx}(N-2) & \cdots & r_{xx}(0) \end{bmatrix} \begin{bmatrix} -1 \\ a_1 \\ a_2 \\ \vdots \\ a_N \end{bmatrix} = \begin{bmatrix} -E_{\min} \\ 0 \\ 0 \\ \vdots \\ 0 \end{bmatrix}$$

Levinson-Durbin recursion - Extension of the matrix



$$\begin{bmatrix} r_{xx}(0) & r_{xx}(1) & r_{xx}(2) & \cdots & r_{xx}(N) \\ r_{xx}(1) & r_{xx}(0) & r_{xx}(1) & \cdots & r_{xx}(N-1) \\ r_{xx}(2) & r_{xx}(1) & r_{xx}(0) & \cdots & r_{xx}(N-2) \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ r_{xx}(N) & r_{xx}(N-1) & r_{xx}(N-2) & \cdots & r_{xx}(0) \end{bmatrix} \begin{bmatrix} -1 \\ a_1 \\ a_2 \\ \vdots \\ a_N \end{bmatrix} = \begin{bmatrix} -E_{\min} \\ 0 \\ 0 \\ \vdots \\ 0 \end{bmatrix}$$

$$E_{\min} = r_{xx}(0) - \mathbf{r}_{xx}^T(1) \mathbf{a} \quad \mathbf{a} = [a_1 \ a_2 \ \dots \ a_N]^T$$

□ Extension of the order:

$$\begin{bmatrix} r_{xx}(0) & r_{xx}(1) & r_{xx}(2) & \cdots & r_{xx}(N) & r_{xx}(N+1) \\ r_{xx}(1) & r_{xx}(0) & r_{xx}(1) & \cdots & r_{xx}(N-1) & r_{xx}(N) \\ r_{xx}(2) & r_{xx}(1) & r_{xx}(0) & \cdots & r_{xx}(N-2) & r_{xx}(N-1) \\ \vdots & \vdots & \vdots & \ddots & \vdots & \vdots \\ r_{xx}(N) & r_{xx}(N-1) & r_{xx}(N-2) & \cdots & r_{xx}(0) & r_{xx}(1) \\ r_{xx}(N+1) & r_{xx}(N) & r_{xx}(N-1) & \cdots & r_{xx}(1) & r_{xx}(0) \end{bmatrix} \begin{bmatrix} -1 \\ a_1 \\ a_2 \\ \vdots \\ a_N \\ 0 \end{bmatrix} = \begin{bmatrix} -E_{\min} \\ 0 \\ 0 \\ \vdots \\ 0 \\ -V \end{bmatrix}$$

Last row of the matrix:

$$V = r_{xx}(N+1) - \mathbf{r}_{xx}^T(1) \tilde{\mathbf{a}} \quad \tilde{\mathbf{a}} = [a_N \ a_{N-1} \ \dots \ a_1]^T$$

Levinson-Durbin recursion

$$\begin{bmatrix} r_{xx}(0) & r_{xx}(1) & r_{xx}(2) & \cdots & r_{xx}(N) & r_{xx}(N+1) \\ r_{xx}(1) & r_{xx}(0) & r_{xx}(1) & \cdots & r_{xx}(N-1) & r_{xx}(N) \\ r_{xx}(2) & r_{xx}(1) & r_{xx}(0) & \cdots & r_{xx}(N-2) & r_{xx}(N-1) \\ \vdots & \vdots & \vdots & \ddots & \vdots & \vdots \\ r_{xx}(N) & r_{xx}(N-1) & r_{xx}(N-2) & \cdots & r_{xx}(0) & r_{xx}(1) \\ r_{xx}(N+1) & r_{xx}(N) & r_{xx}(N-1) & \cdots & r_{xx}(1) & r_{xx}(0) \end{bmatrix} \begin{bmatrix} -1 \\ a_1 \\ a_2 \\ \vdots \\ a_N \\ 0 \end{bmatrix} = \begin{bmatrix} -E_{\min} \\ 0 \\ 0 \\ \vdots \\ 0 \\ -V \end{bmatrix}$$

□ Reorder the equations (flip all rows and columns), profit from Toeplitz structure:

$$\begin{bmatrix} r_{xx}(0) & r_{xx}(1) & r_{xx}(2) & \cdots & r_{xx}(N) & r_{xx}(N+1) \\ r_{xx}(1) & r_{xx}(0) & r_{xx}(1) & \cdots & r_{xx}(N-1) & r_{xx}(N) \\ r_{xx}(2) & r_{xx}(1) & r_{xx}(0) & \cdots & r_{xx}(N-2) & r_{xx}(N-1) \\ \vdots & \vdots & \vdots & \ddots & \vdots & \vdots \\ r_{xx}(N) & r_{xx}(N-1) & r_{xx}(N-2) & \cdots & r_{xx}(0) & r_{xx}(1) \\ r_{xx}(N+1) & r_{xx}(N) & r_{xx}(N-1) & \cdots & r_{xx}(1) & r_{xx}(0) \end{bmatrix} \begin{bmatrix} 0 \\ a_N \\ a_{N-1} \\ \vdots \\ a_1 \\ -1 \end{bmatrix} = \begin{bmatrix} -V \\ 0 \\ 0 \\ \vdots \\ 0 \\ -E_{\min} \end{bmatrix}$$

Levinson-Durbin recursion

Note with (N), the N-th order of prediction:

$$\mathbf{a}^{(N)} = [a_1^{(N)} \ a_2^{(N)} \ \dots \ a_N^{(N)}]^T \quad \mathbf{a}^{(N+1)} = [a_1^{(N+1)} \ a_2^{(N+1)} \ \dots \ a_{N+1}^{(N+1)}]^T$$

Linear combination of previous matrix equations:

$$\begin{bmatrix} r_{xx}(0) & r_{xx}(1) & r_{xx}(2) & \dots & r_{xx}(N) & r_{xx}(N+1) \\ r_{xx}(1) & r_{xx}(0) & r_{xx}(1) & \dots & r_{xx}(N-1) & r_{xx}(N) \\ r_{xx}(2) & r_{xx}(1) & r_{xx}(0) & \dots & r_{xx}(N-2) & r_{xx}(N-1) \\ \vdots & \vdots & \vdots & \ddots & \vdots & \vdots \\ r_{xx}(N) & r_{xx}(N-1) & r_{xx}(N-2) & \dots & r_{xx}(0) & r_{xx}(1) \\ r_{xx}(N+1) & r_{xx}(N) & r_{xx}(N-1) & \dots & r_{xx}(1) & r_{xx}(0) \end{bmatrix} \left\{ \begin{bmatrix} -1 \\ a_1^{(N)} \\ a_2^{(N)} \\ \vdots \\ a_N^{(N)} \\ 0 \end{bmatrix} - \Gamma^{(N+1)} \begin{bmatrix} 0 \\ a_N^{(N)} \\ a_{N-1}^{(N)} \\ \vdots \\ a_1^{(N)} \\ -1 \end{bmatrix} \right\} = \begin{bmatrix} -E_{\min}^{(N)} \\ 0 \\ 0 \\ \vdots \\ 0 \\ -V^{(N)} \end{bmatrix} - \Gamma^{(N+1)} \begin{bmatrix} -V^{(N)} \\ 0 \\ 0 \\ \vdots \\ 0 \\ -E_{\min}^{(N)} \end{bmatrix}$$

Matrix equation for predictor of order (N+1):

agora ele só iguala

$$\begin{bmatrix} r_{xx}(0) & r_{xx}(1) & r_{xx}(2) & \dots & r_{xx}(N) & r_{xx}(N+1) \\ r_{xx}(1) & r_{xx}(0) & r_{xx}(1) & \dots & r_{xx}(N-1) & r_{xx}(N) \\ r_{xx}(2) & r_{xx}(1) & r_{xx}(0) & \dots & r_{xx}(N-2) & r_{xx}(N-1) \\ \vdots & \vdots & \vdots & \ddots & \vdots & \vdots \\ r_{xx}(N) & r_{xx}(N-1) & r_{xx}(N-2) & \dots & r_{xx}(0) & r_{xx}(1) \\ r_{xx}(N+1) & r_{xx}(N) & r_{xx}(N-1) & \dots & r_{xx}(1) & r_{xx}(0) \end{bmatrix} \begin{bmatrix} -1 \\ a_1^{(N+1)} \\ a_2^{(N+1)} \\ \vdots \\ a_N^{(N+1)} \\ a_{N+1}^{(N+1)} \end{bmatrix} = \begin{bmatrix} -E_{\min}^{(N+1)} \\ 0 \\ 0 \\ \vdots \\ 0 \\ 0 \end{bmatrix}$$

$$\Gamma^{(N+1)} = \frac{V^{(N)}}{E_{\min}^{(N)}} = \frac{r_{xx}(N+1) - \mathbf{r}_{xx}^T(1) \tilde{\mathbf{a}}^{(N)}}{r_{xx}(0) - \mathbf{r}_{xx}^T(1) \mathbf{a}^{(N)}}$$

$$a_{N+1}^{(N+1)} = \Gamma^{(N+1)}$$

$$\begin{bmatrix} a_1^{(N+1)} \\ a_2^{(N+1)} \\ \vdots \\ a_N^{(N+1)} \end{bmatrix} = \begin{bmatrix} a_1^{(N)} \\ a_2^{(N)} \\ \vdots \\ a_N^{(N)} \end{bmatrix} - \Gamma^{(N+1)} \begin{bmatrix} a_N^{(N)} \\ a_{N-1}^{(N)} \\ \vdots \\ a_1^{(N)} \end{bmatrix}$$

Levinson-Durbin recursion - Results

▣ **Parcor coefficient, (N+1)th coefficient of the predictor filter of order (N+1):**

$$\Gamma^{(N+1)} = \frac{V^{(N)}}{E_{\min}^{(N)}} = \frac{r_{xx}(N+1) - \mathbf{r}_{xx}^T(1) \tilde{\mathbf{a}}^{(N)}}{r_{xx}(0) - \mathbf{r}_{xx}^T(1) \mathbf{a}^{(N)}}$$

$$a_{N+1}^{(N+1)} = \Gamma^{(N+1)}$$

▣ **All other coefficients can be updated based on the Parcor coefficient and the previous prediction coefficients:**

$$\begin{bmatrix} a_1^{(N+1)} \\ a_2^{(N+1)} \\ \vdots \\ a_N^{(N+1)} \end{bmatrix} = \begin{bmatrix} a_1^{(N)} \\ a_2^{(N)} \\ \vdots \\ a_N^{(N)} \end{bmatrix} - \Gamma^{(N+1)} \begin{bmatrix} a_N^{(N)} \\ a_{N-1}^{(N)} \\ \vdots \\ a_1^{(N)} \end{bmatrix} = \mathbf{a}^{(N)} - a_{N+1}^{(N+1)} \tilde{\mathbf{a}}^{(N)}$$

with:

$$\mathbf{a}^{(N)} = [a_1 \ a_2 \ \dots \ a_N]^T$$

$$\tilde{\mathbf{a}}^{(N)} = [a_N \ a_{N-1} \ \dots \ a_1]^T$$

Levinson-Durbin recursion - Results

Update of the prediction error power:

$$E_{\min}^{(N+1)} = E_{\min}^{(N)} - \Gamma^{(N+1)} V^{(N)} \quad \text{with: } V^{(N)} = \Gamma^{(N+1)} E_{\min}^{(N)}$$

$$E_{\min}^{(N+1)} = E_{\min}^{(N)} (1 - |\Gamma^{(N+1)}|^2)$$

The Levinson-Durbin recursion guarantees:

$$|\Gamma^{(N+1)}| \leq 1$$

Start of the recursion:

$$\Gamma^{(1)} = \frac{V^{(0)}}{E_{\min}^{(0)}} = \frac{r_{xx}(1)}{r_{xx}(0)}$$

$$a_1^{(1)} = \Gamma^{(1)} = \frac{r_{xx}(1)}{r_{xx}(0)}$$

with: $E_{\min} = r_{xx}(0) - \mathbf{r}_{xx}^T(1) \mathbf{a}$

$$E_{\min}^{(1)} = r_{xx}(0) - r_{xx}(1) a_1^{(1)}$$

□ The property $|\Gamma^{(N+1)}| \leq 1$ guarantees phase minimum prediction error filters (i.e., stable inverse Prediction error filters)

=> Will be shown with lattice structure realization

Levinson-Durbin recursion - Summary

Initialization:

❑ Predictor:

$$a_1^{(1)} = \tilde{a}_1^{(1)} = r_{xx}(1)/r_{xx}(0)$$

❑ Error signal power (minimum):

$$E_{\min}^{(0)} = r_{xx}(0)$$

Recursion:

pergunta de prova: Para uma construção de lattice structure, só precisa do reflection coefficient da pra economizar computação no cálculo

❑ Reflection coefficient:

$$a_{N+1}^{(N+1)} = \frac{r_{xx}(N+1) - \mathbf{r}_{xx}^T(1) \tilde{\mathbf{a}}^{(N)}}{r_{xx}(0) - \mathbf{r}_{xx}^T(1) \mathbf{a}^{(N)}}$$

❑ Forward prediction:

$$[a_1^{(N+1)} a_2^{(N+1)} \dots a_N^{(N+1)}]^T = \mathbf{a}^{(N)} - a_{N+1}^{(N+1)} \tilde{\mathbf{a}}^{(N)}$$

❑ Backward prediction:

$$\tilde{a}_i^{(N)} = a_{N-i}^{(N)}$$

❑ Error power (minimum):

$$E_{\min}^{(N+1)} = E_{\min}^{(N)} (1 - |a_{N+1}^{(N+1)}|^2)$$

Stop criteria:

❑ Numeric criterion:

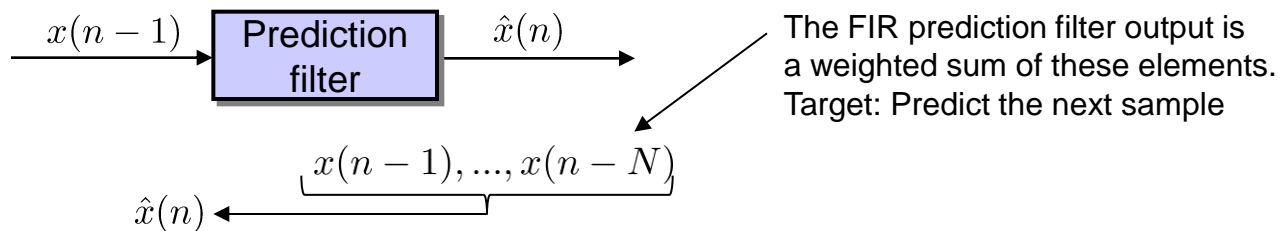
When $|a_{N+1}^{(N+1)}|^2 < \epsilon$, use the previous recursion step and stop the recursion.

❑ Order criterion:

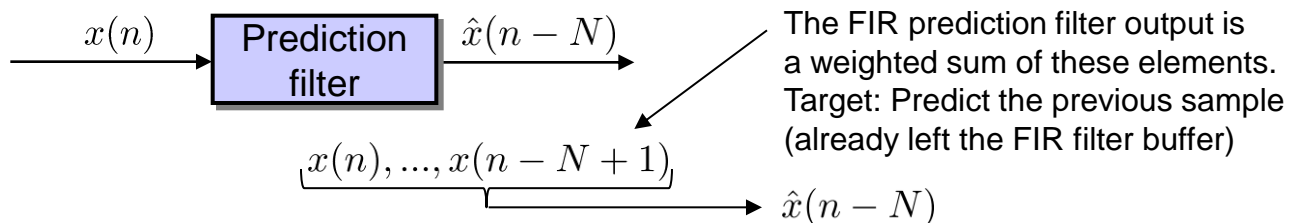
In case N has reached the desired order \Rightarrow recursion stop.

Backward Prediction: Comparison to forward prediction

□ Forward Prediction:

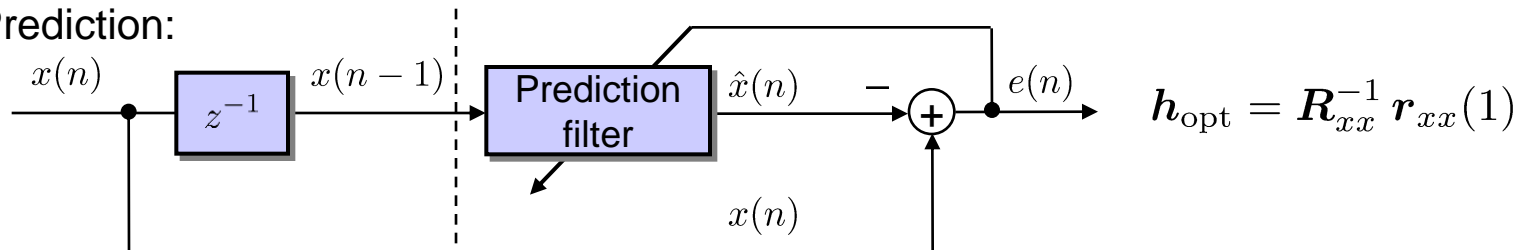


□ Backward Prediction:

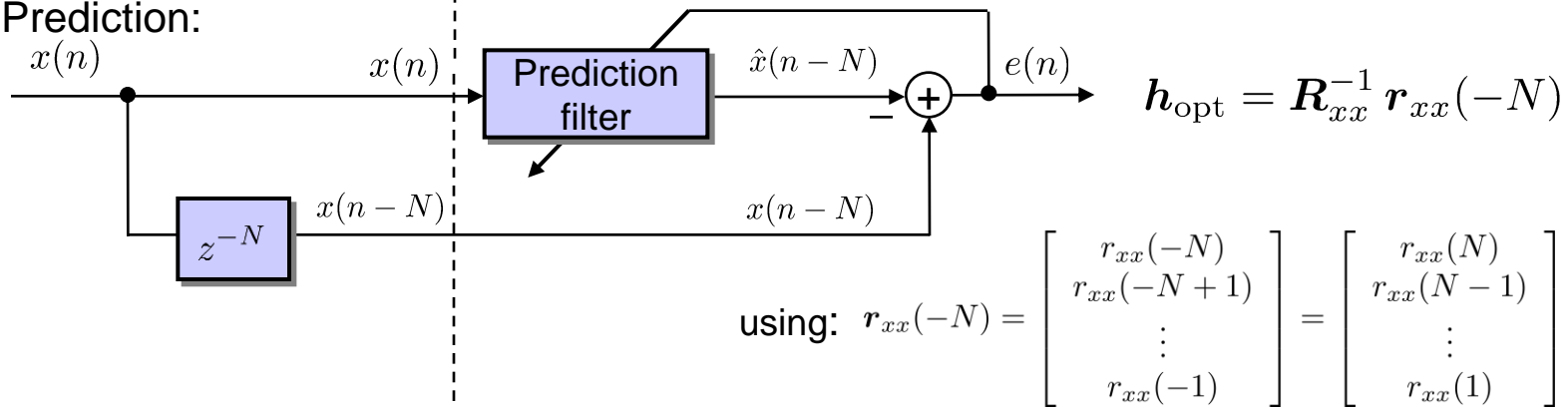


Backward Prediction

Forward Prediction:



Backward Prediction:



results in:

$$\begin{bmatrix} r_{xx}(0) & r_{xx}(1) & \cdots & r_{xx}(N-1) \\ r_{xx}(1) & r_{xx}(0) & \cdots & r_{xx}(N-2) \\ \vdots & \vdots & \ddots & \vdots \\ r_{xx}(N-1) & r_{xx}(N-2) & \cdots & r_{xx}(0) \end{bmatrix} \begin{bmatrix} a_{r,1} \\ a_{r,2} \\ \vdots \\ a_{r,N} \end{bmatrix} = \begin{bmatrix} r_{xx}(N) \\ r_{xx}(N-1) \\ \vdots \\ r_{xx}(1) \end{bmatrix}$$

Backward Prediction

$$\begin{bmatrix} r_{xx}(0) & r_{xx}(1) & \cdots & r_{xx}(N-1) \\ r_{xx}(1) & r_{xx}(0) & \cdots & r_{xx}(N-2) \\ \vdots & \vdots & \ddots & \vdots \\ r_{xx}(N-1) & r_{xx}(N-2) & \cdots & r_{xx}(0) \end{bmatrix} \begin{bmatrix} a_{r,1} \\ a_{r,2} \\ \vdots \\ a_{r,N} \end{bmatrix} = \begin{bmatrix} r_{xx}(N) \\ r_{xx}(N-1) \\ \vdots \\ r_{xx}(1) \end{bmatrix}$$

Flip the rows and columns:

$$\begin{bmatrix} r_{xx}(0) & r_{xx}(1) & \cdots & r_{xx}(N-1) \\ r_{xx}(1) & r_{xx}(0) & \cdots & r_{xx}(N-2) \\ \vdots & \vdots & \ddots & \vdots \\ r_{xx}(N-1) & r_{xx}(N-2) & \cdots & r_{xx}(0) \end{bmatrix} \begin{bmatrix} a_{r,N} \\ a_{r,N-1} \\ \vdots \\ a_{r,1} \end{bmatrix} = \begin{bmatrix} r_{xx}(1) \\ r_{xx}(2) \\ \vdots \\ r_{xx}(N) \end{bmatrix}$$

□ Forward Prediction:

$$\mathbf{a} = [a_1 \ a_2 \ \dots \ a_N]^T$$

$$e^{(N)}(n) = x(n) - \sum_{i=1}^N a_i^{(N)} x(n-i)$$

□ Backward Prediction:

$$\mathbf{a}_r = \tilde{\mathbf{a}} = [a_N \ a_{N-1} \ \dots \ a_1]^T$$

$$\begin{aligned} e_r^{(N)}(n-N) &= x(n-N) - \sum_{i=1}^N a_{r,i}^{(N)} x(n-i+1) \\ &= x(n-N) - \sum_{i=1}^N a_{N+1-i}^{(N)} x(n-i+1) \end{aligned}$$

Backward Prediction: Recursive error calculation

$$\begin{aligned}
 \hat{x}^{(N)}(n) &= \sum_{i=1}^N a_i^{(N)} x(n-i) \\
 &= \sum_{i=1}^{N-1} (a_i^{(N-1)} - a_N^{(N)} a_{N-i}^{(N-1)}) x(n-i) + a_N^{(N)} x(n-N) \\
 &= \underbrace{\sum_{i=1}^{N-1} a_i^{(N-1)} x(n-i)}_{\substack{\text{Forward prediction} \\ \text{of order } N-1 \\ = \hat{x}^{(N-1)}(n)}} + a_N^{(N)} \underbrace{\left(x(n-N) - \sum_{i=1}^{N-1} a_{N-i}^{(N-1)} x(n-i) \right)}_{\substack{\text{additional} \\ \text{signal value} \\ \text{Backward prediction} \\ \text{of order } N-1 \\ = \hat{x}_r^{(N-1)}(n-N)}} \\
 &\quad \underbrace{\hspace{10em}}_{\text{Innovation}} \\
 &= \hat{x}^{(N-1)}(n) + a_N^{(N)} \left(x(n-N) - \hat{x}_r^{(N-1)}(n-N) \right)
 \end{aligned}$$

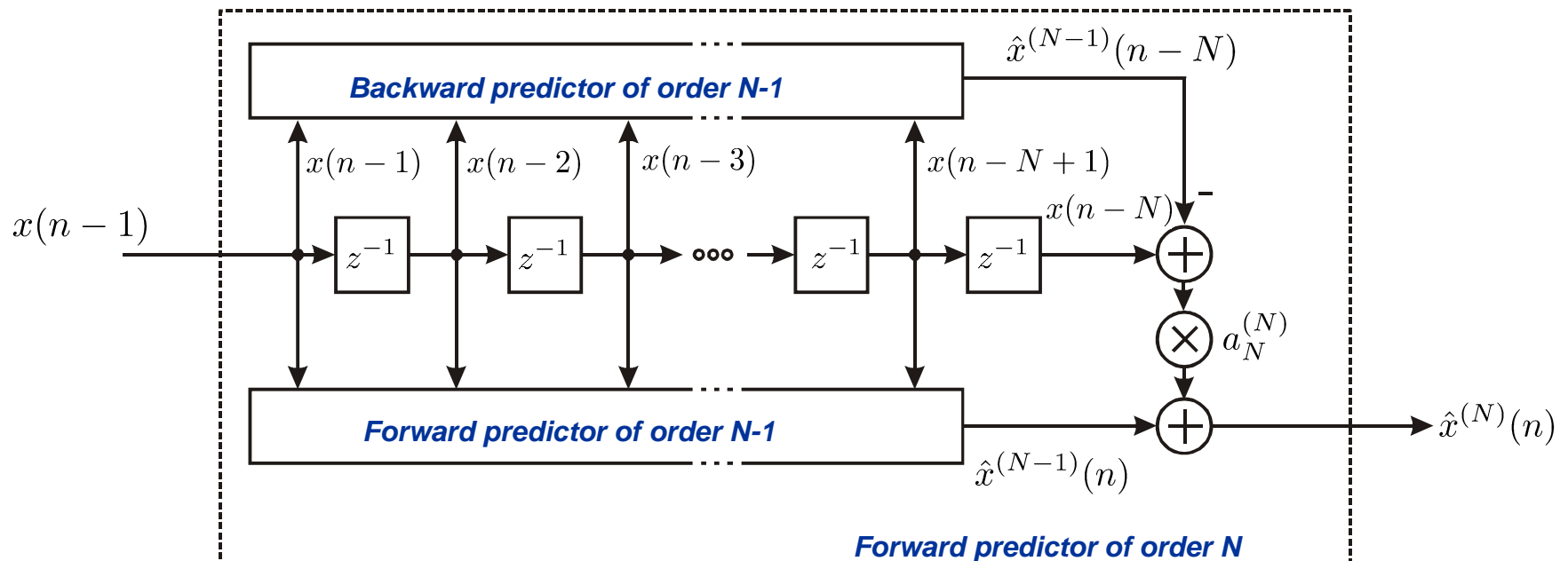
Backward Prediction: Recursive error calculation

Short Form:

$$\hat{x}^{(N)}(n) = \hat{x}^{(N-1)}(n) + a_N^{(N)} \left(x(n - N) - \hat{x}_r^{(N-1)}(n - N) \right)$$

*New estimate = Old estimate + weighting * (new – prediction of new)*

Structure of the order recursion:



Backward Prediction: Recursive error calculation



$$\begin{aligned}
 e^{(N)}(n) &= x(n) - \hat{x}^{(N)}(n) \\
 &= x(n) - \hat{x}^{(N-1)}(n) - a_N^{(N)} \left(x(n-N) - \hat{x}_r^{(N-1)}(n-N) \right) \\
 &= e^{(N-1)}(n) - a_N^{(N)} e_r^{(N-1)}(n-N)
 \end{aligned}$$

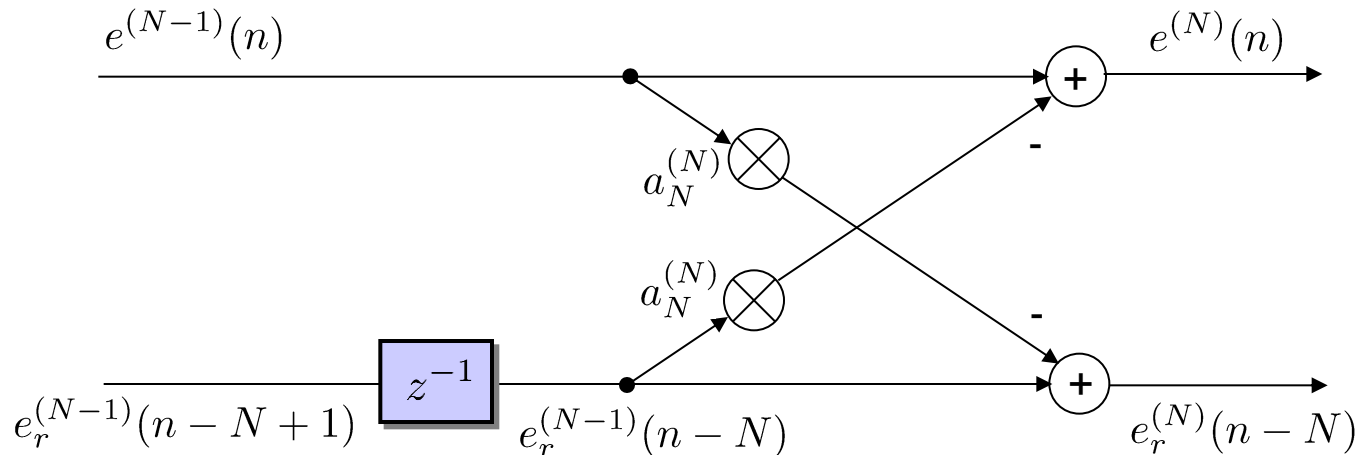
$$\begin{aligned}
 e_r^{(N)}(n-N) &= x(n-N) - \hat{x}^{(N)}(n-N) \\
 &= x(n-N) - \sum_{i=1}^N a_{N+1-i}^{(N)} x(n-i+1) \\
 &= x(n-N) - \sum_{i=1}^N a_i^{(N)} x(n-N+i) \quad \downarrow \text{Flip summation index} \\
 &= x(n-N) - \sum_{i=1}^{N-1} [a_i^{(N-1)} - a_N^{(N)} a_{N-i}^{(N-1)}] x(n-N+i) - a_N^{(N)} x(n) \\
 &= x(n-N) - \sum_{i=1}^{N-1} a_i^{(N-1)} x(n-N+i) - a_N^{(N)} \left[x(n) - \sum_{i=1}^{N-1} a_{N-i}^{(N-1)} x(n-N+i) \right] \\
 &= e_r^{(N-1)}(n-N) - a_N^{(N)} e^{(N-1)}(n)
 \end{aligned}$$

Backward Prediction: Recursive error calculation

□ Previous equations resulting in the following recursion:

$$e^{(N)}(n) = e^{(N-1)}(n) - a_N^{(N)} e_r^{(N-1)}(n - N)$$

$$e_r^{(N)}(n - N) = e_r^{(N-1)}(n - N) - a_N^{(N)} e^{(N-1)}(n)$$



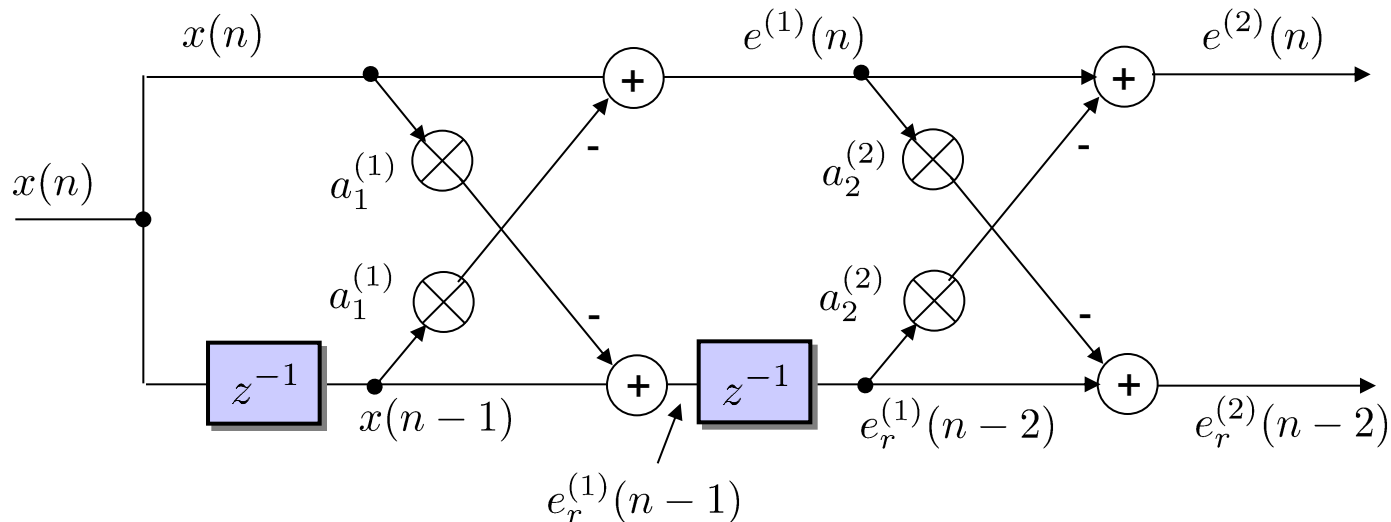
Lattice filter structure of the prediction error filter

$N = 1$:

$$e^{(1)}(n) = e^{(0)}(n) - a_1^{(1)} e_r^{(0)}(n-1) \longrightarrow e^{(1)}(n) = x(n) - a_1^{(1)} x(n-1)$$

$$e_r^{(1)}(n-1) = e_r^{(0)}(n-1) - a_1^{(1)} e^{(0)}(n) \longrightarrow e_r^{(1)}(n-1) = x(n-1) - a_1^{(1)} x(n)$$

□ Lattice structure of the prediction error filter:



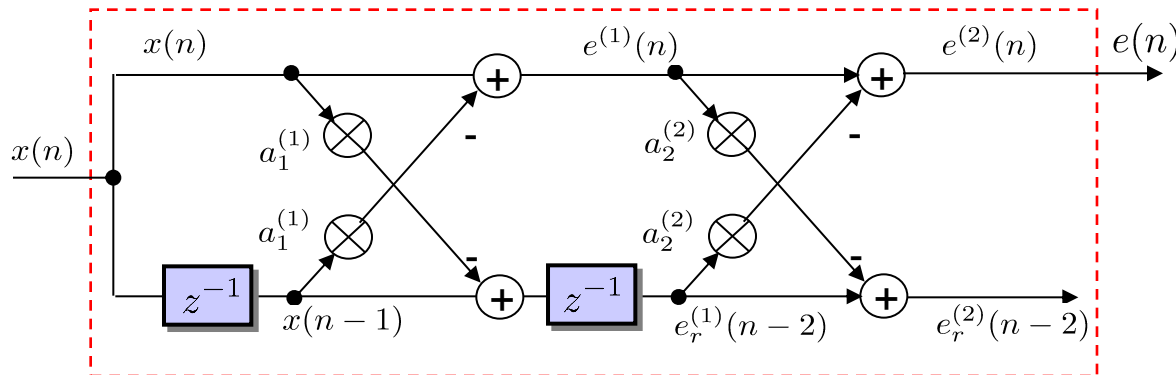
Different realizations of the prediction error filter

ele parou aqui

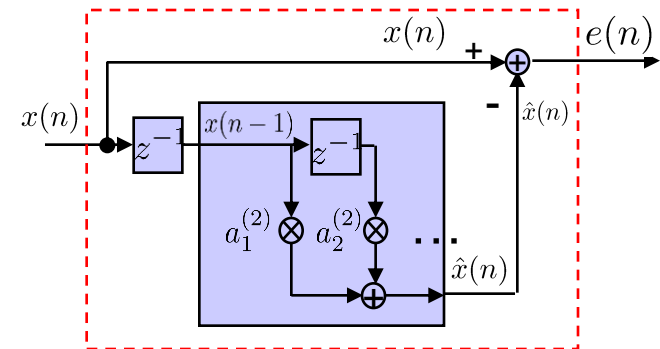
~~PERGUNTA: esse modelo parece pior em termos de hardware pq vc precisa de 2x mais multiplicadores~~ - RESPONDIDO

r - vai depender da sua aplicação, mas essencialmente as multiplicações são feitas com SHIFT-AND-ADD

❑ Lattice structure:



❑ Normal FIR filter:



não dá pra checar se é minimum phase
não dá pra checar se o inverso é estável

❑ Advantage of the lattice structure:

Since the Parcor coefficients are of magnitude < 1 , the stability of the inverse prediction error filter (IIR filter !) is guaranteed

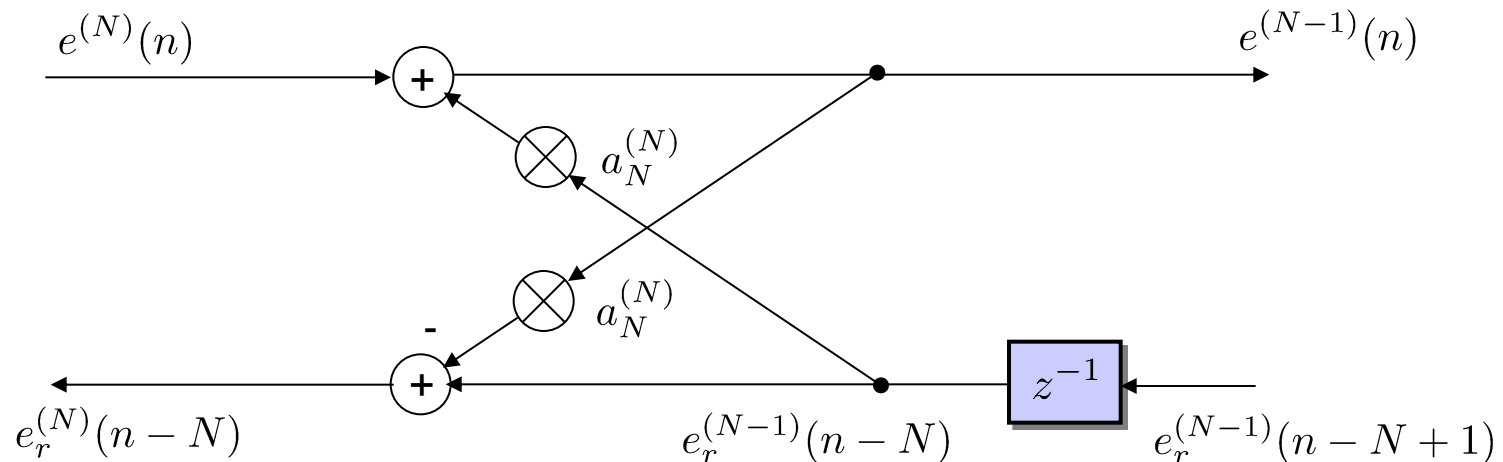
IIR inverse prediction lattice filter structure:

□ **FIR equations:** $e^{(N)}(n) = e^{(N-1)}(n) - a_N^{(N)} e_r^{(N-1)}(n - N)$

$$e_r^{(N)}(n - N) = e_r^{(N-1)}(n - N) - a_N^{(N)} e^{(N-1)}(n)$$

□ **IIR equations:** $e^{(N-1)}(n) = e^{(N)}(n) + a_N^{(N)} e_r^{(N-1)}(n - N)$

$$e_r^{(N)}(n - N) = e_r^{(N-1)}(n - N) - a_N^{(N)} e^{(N-1)}(n)$$



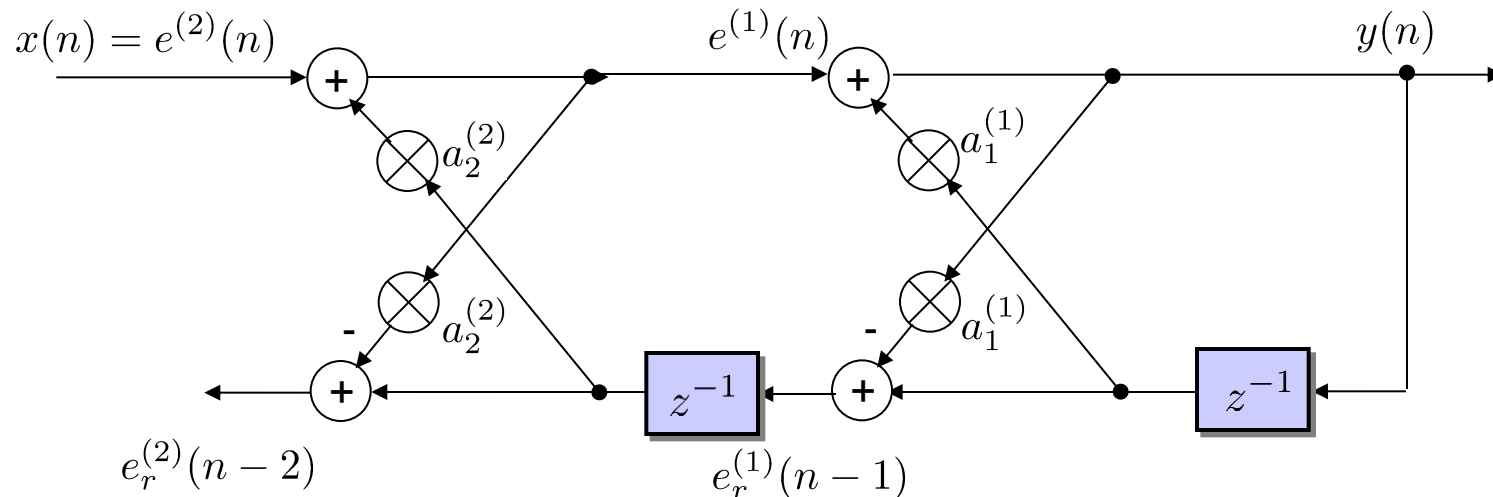
IIR inverse prediction lattice filter structure

$N = 1$:

$$e^{(0)}(n) = e^{(1)}(n) + a_1^{(1)} e_r^{(0)}(n-1) \longrightarrow y(n) = e^{(1)}(n) + a_1^{(1)} y(n-1)$$

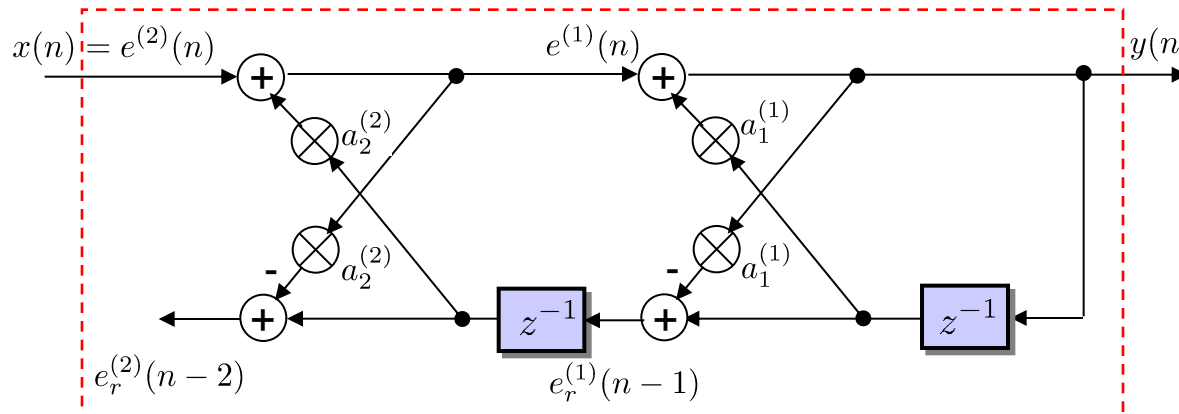
$$e_r^{(1)}(n-1) = e_r^{(0)}(n-1) - a_1^{(1)} e^{(0)}(n) \longrightarrow e_r^{(1)}(n-1) = y(n-1) - a_1^{(1)} y(n)$$

□ IIR filter of second order in lattice structure:

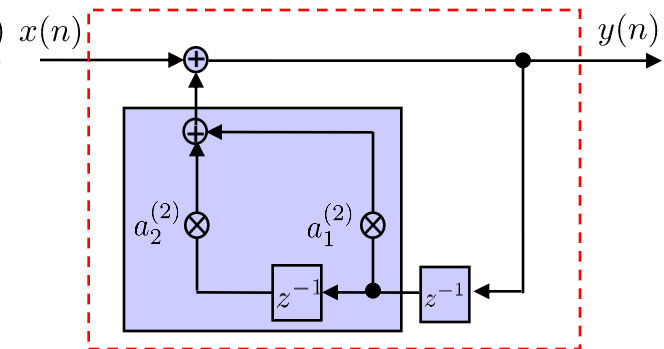


Different realizations of IIR inverse prediction filters

□ Lattice structure:



□ Normal IIR filter:



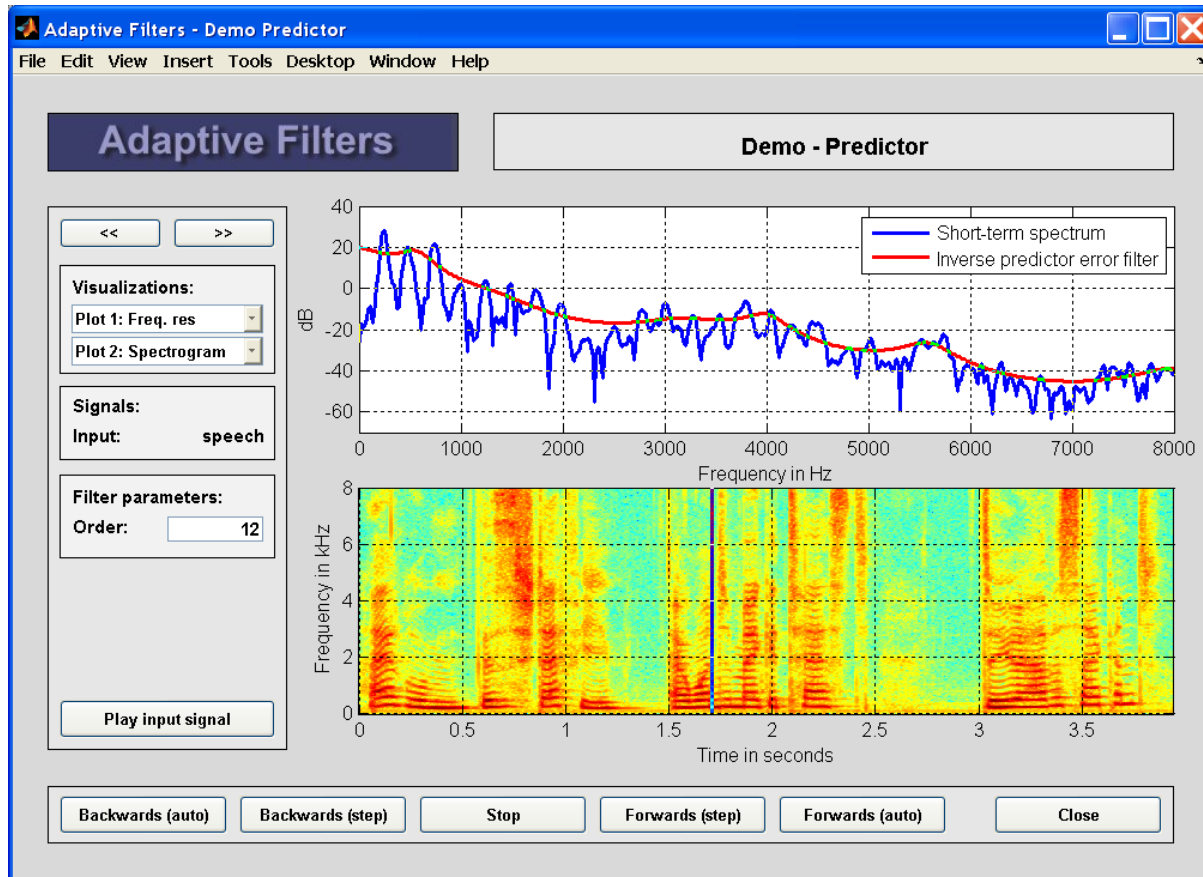
□ Advantage of the lattice structure:

Since the Parcor coefficients are of magnitude < 1 , the stability of the IIR filter is guaranteed.

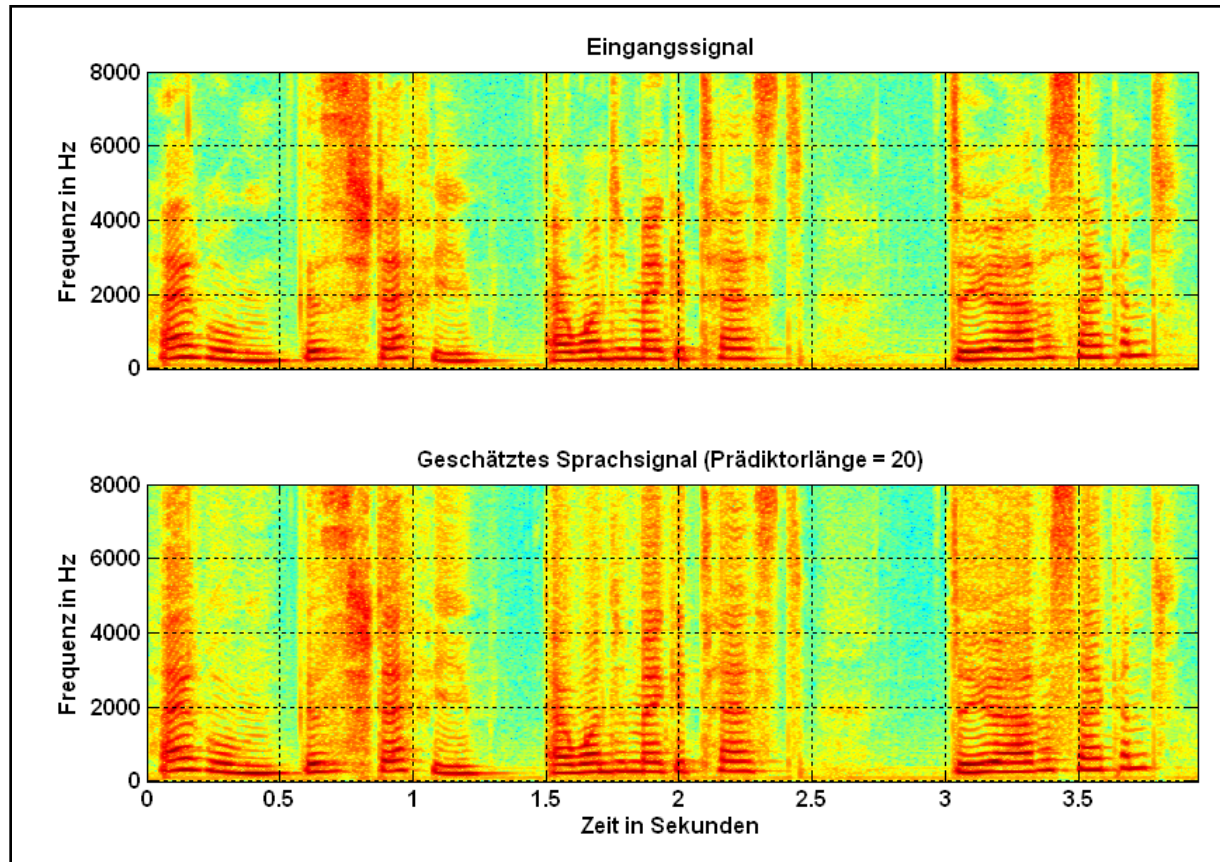
os coefs $a(j)_i$ podem ser maiores que 1, só $a(i)_i$ tem que ter magnitude menor que 1

Matlab example

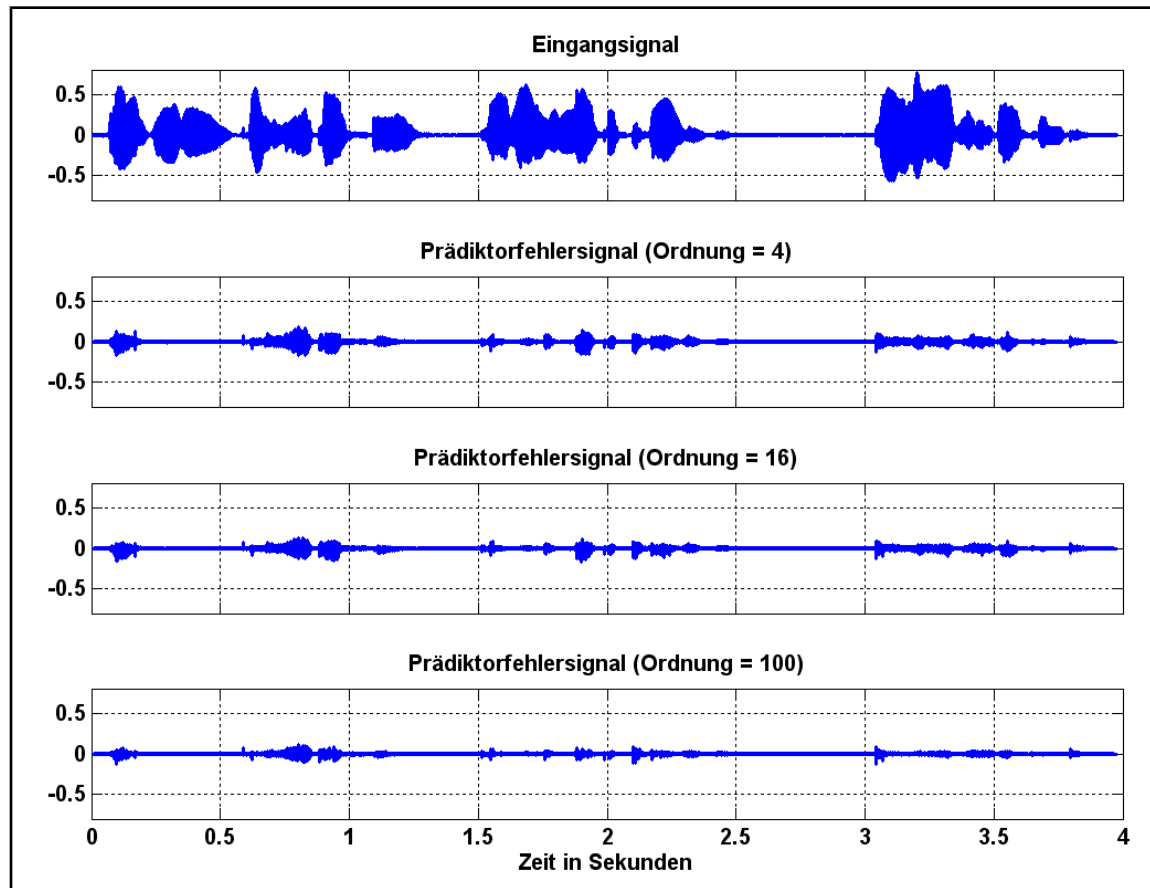
PEF FILTER: [1 -a1 -a2 ...



Matlab example – estimated speech signal



Matlab example – prediction error signal



This week:

- ❑ Linear prediction as tool to predict next signal sample bases on previous samples => utilizing redundancy
- ❑ Application are removal of redundancy for efficient source signal coding and spectral envelope estimation
- ❑ The Levinson-Durbin recursion (order recursion!) allows an efficient calculation of the prediction coefficients.

Next week:

- ❑ Applications of linear prediction