

**Indian Institute of Space Science and Technology**  
**AV336 - Digital Signal Processing Lab**  
**Department of Avionics**

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**Labsheet 9**

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1. Visualization of the bilinear transform: The bilinear transform is a map between the  $s$ -plane and the  $z$ -plane and is defined as  $z = \frac{1+sT/2}{1-sT/2}$ . The inverse map is defined as  $s = \frac{2}{T} \frac{1-z^{-1}}{1+z^{-1}}$ . In this task you will visualize how an area in the  $s$  plane is mapped to  $z$ -plane and vice-versa.
  - (a) Select  $N \times M$  points  $s_{i,j}$  uniformly in any rectangular region that you wish in the  $s$ -plane. (Hint: you can use meshgrid for this purpose). On the  $z$ -plane indicate where these points are mapped to under the bilinear transform.
  - (b) Similarly select  $N \times M$  points  $z_{i,j}$  uniformly in any rectangular region that you wish in the  $z$ -plane. (Hint: you can use meshgrid for this purpose). On the  $s$ -plane indicate where these points are mapped to under the bilinear transform.
  - (c) Verify (using a sufficiently dense set of points) whether the  $j\Omega$  axis is mapped to the unit circle under the bilinear transform.
2. Filter design using least squares inverse design: Suppose the desired frequency response  $H_d(e^{j\omega})$  can be realized by the causal system with the following  $z$ -transform,

$$H_d(z) = \frac{1 + z^{-1}}{1 - 0.5z^{-1}}.$$

Suppose we approximate  $H_d(z)$  using a  $H(z)$  of the form

$$\frac{b_0}{1 - a_1z^{-1} - a_2z^{-2}}.$$

Using least squares inverse design, obtain the values of  $b_0$ ,  $a_1$  and  $a_2$ . We note that in least squares inverse design, a set of linear equations have to be written down which constrain the values of  $b_0$ ,  $a_1$  and  $a_2$ . Explore the effect of the number of linear equations that you have on your answer.

3. Filter design using Butterworth analog filter design and impulse invariance: Suppose we have the following desired requirements  $H_d(e^{j\omega})$  on the magnitude of digital filter:
  - (a) Passband edge  $= 0.2\pi$
  - (b) Stopband edge (starting freq)  $= 0.4\pi$
  - (c) Magnitude gain in passband to be  $\in [1, 1 - \delta_p]$ , where  $\delta_p = 0.05$
  - (d) Magnitude gain in stopband to be  $\in [0, \delta_s]$ , where  $\delta_s = 0.001$

Note that no constraints are being put on the phase response of the filter here. In the following, for the impulse invariance based design assume that  $T = 1$ .

- (a) Assuming that there is no aliasing and that  $H_d(e^{j\omega})$  has been obtained from sampling of an analog signal  $h_a(t)$  uniformly at rate  $\frac{1}{T}$ , what is  $H_a(j\Omega)$  (the CTFT of  $h_a(t)$ )?
- (b) We note that  $H_a(j\Omega)$  can be interpreted as the specification for the design of an analog filter. Obtain a Butterworth filter that is a good approximation to  $H_a(j\Omega)$ .
- (c) Write down the location of the poles of the analog Butterworth filter  $H_a(s)$ ?
- (d) Under the impulse invariance condition, where are these poles mapped to in the  $z$ -plane. Write down the locations of the poles.

Exploration:

- (a) Repeat the design process for  $T = \frac{1}{4}$ . Is there any change in the frequency response of the realized digital filter.
- (b) Repeat the design process but by not compensating for aliasing in the stopband attenuation. How much is stopband attenuation in the final design? Does it meet the given requirements on  $H_d(e^{j\omega})$ ?
- (c) Using internet resources or Matlab help, find out what the inbuilt function “butter” does. How will you use “butter” for the design problem above?
- (d) Using internet resources or Matlab help, find out what the inbuilt function “filter” does. Suppose  $x[n] = 2\cos(0.1\pi n) + 5\cos(0.6\pi n)$  for  $n \in \{0, \dots, 499\}$ . Simulate what happens when the filter that you have designed above is used to filter  $x[n]$  in order to obtain  $y[n]$ . Plot  $y[n]$  as well as its DTFT.