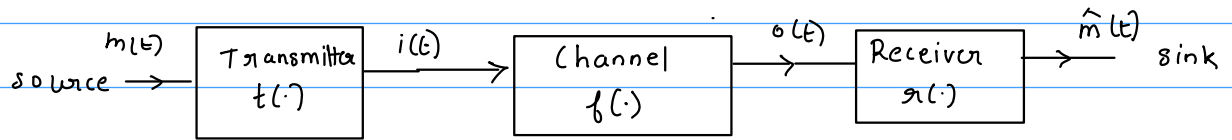


## Review



Design  $t(\cdot)$  and  $r(\cdot)$

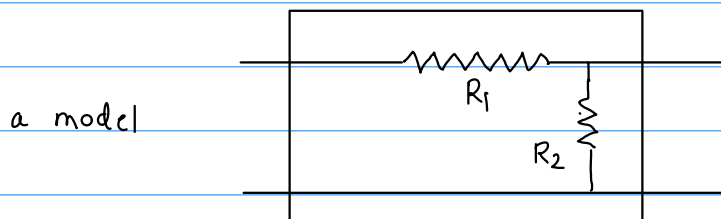
such that  $e(m(t), \hat{m}(t)) \leq \epsilon$ .

given  $f(\cdot)$

*real-world*

\* Channels are well modelled as LTI systems or LTI filters.

e.g. can you model a wire as a LTI filter?

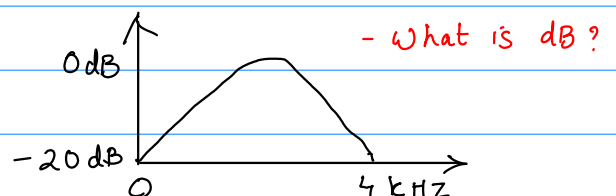


an important idea here is that  $t(\cdot)$  and  $r(\cdot)$  are used to nullify the effect of the channel (at least in our current understanding).

Important points to take away from this class :

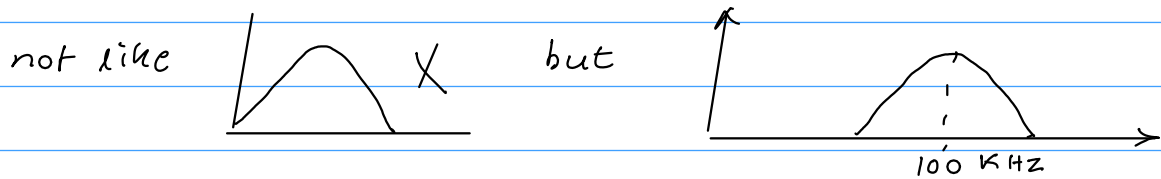
- Channels can be modelled as LTI filters.
- There are model fitting procedures to find out LTI filter models ( $h(t)$  or  $H(f)$ ) for a channel.
- The  $t(\cdot)$  and  $r(\cdot)$  functions are used to nullify the effect of the channel so that  $e(m(t), \hat{m}(t)) \leq \epsilon$ .
- Wired channels are modelled as low pass filters, wireless channels as band pass filters.

e.g. Telephone channel  $|H(f)|$ ?

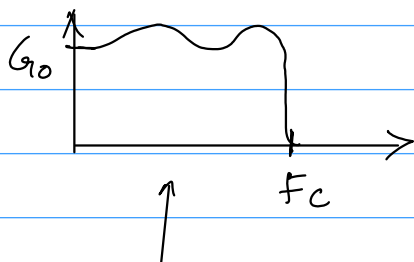


another example DS2 line model.

there are bandpass wired channels too.



Usually wired channels are modelled as low pass filters

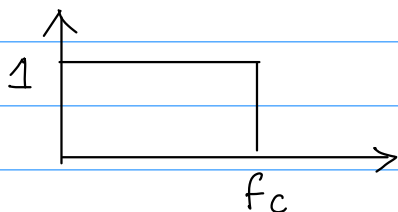


- other ways in which a LPF is characterized.

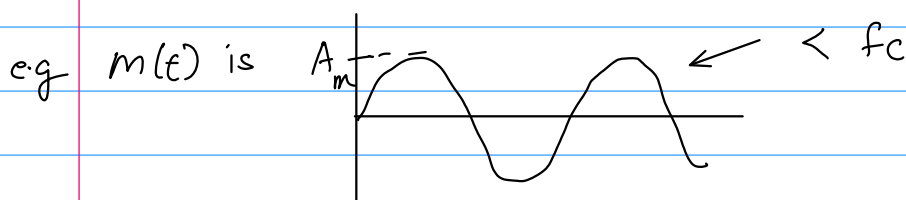
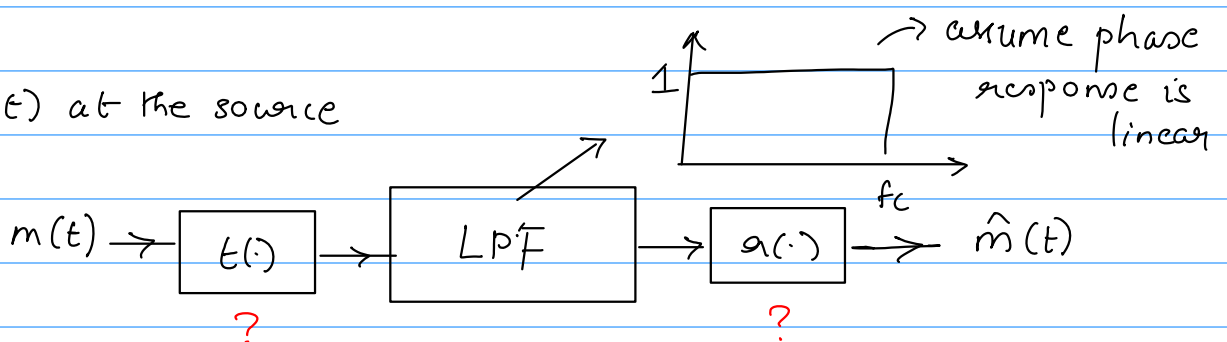
-  $H(f)$  of the LPF

actual low pass filters

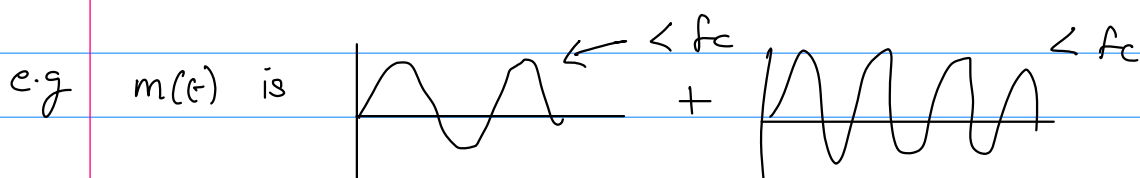
We replace this by an ideal filter to study how to design  $t(\cdot)$  and  $\pi(\cdot)$



$m(t)$  at the source

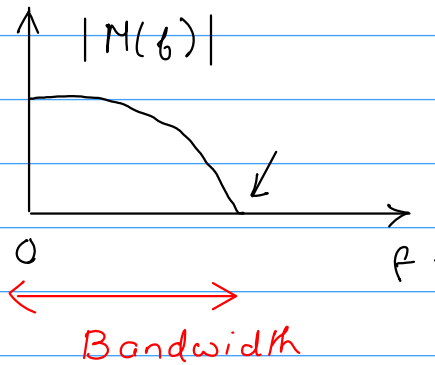


so  $t(\cdot)$  and  $\pi(\cdot)$  are identity functions



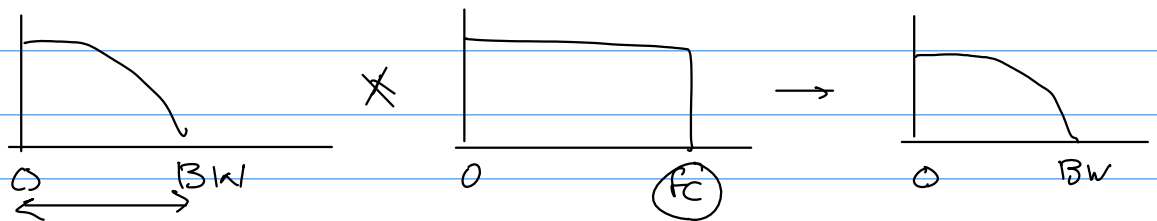
$m(t)$  will have some characteristics which are important for designing  $t(\cdot)$  and  $x(\cdot)$

$$m(t), E_m = \int_{-\infty}^{\infty} |m(t)|^2 dt < \infty, m(t) \xrightarrow{F} \underline{M(f)} \checkmark$$

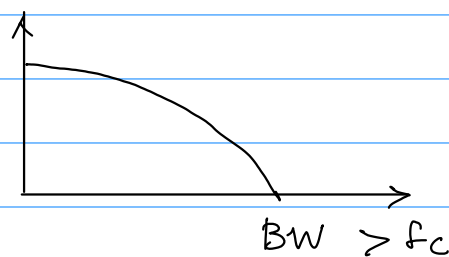


$m(t)$  will a low pass signal  
baseband

What is  $t(\cdot)$  and  $x(\cdot)$  if  $f_c > \text{Bandwidth}$ ?



e.g suppose  $m(t)$   
baseband?



- try and get another channel
- how to reduce BW?

$$\begin{aligned} x(t) &\xrightarrow{F} X(f) \\ x(at) &\xrightarrow{F} X(f/a) \end{aligned}$$

exercise: