

Principles and Experiments of Communications

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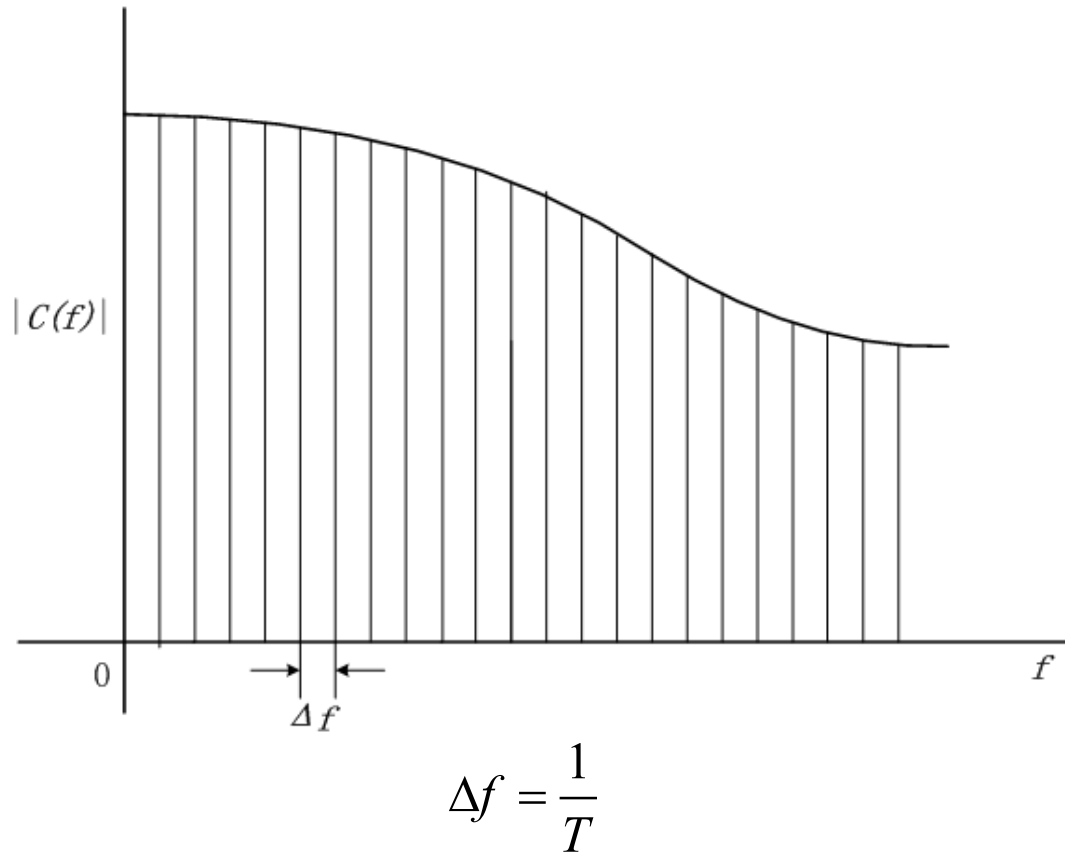
Dept. of Electronic Engineering
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Textbook: Chapter 11

Lecture 06: Multicarrier modulation and OFDM

Multicarrier Modulation

- Multipath channels cause intersymbol interference (ISI) when the reciprocal of the symbol rate is smaller than the dispersion of the nonideal channel.
- A solution:
 - Subdivide the available channel bandwidth into a number of equal-bandwidth subchannels;
 - The frequency response characteristics of each subchannel are nearly ideal.

Subdivision of Channel Bandwidth



where T is the symbol interval

Orthogonal Frequency Division Multiplexing (1/2)

- Each subchannel associate a carrier

$$x_k(t) = \cos 2\pi f_k t, \quad k = 0, 1, \dots, N-1$$

- Orthogonal Frequency Division Multiplexing (OFDM)
 - The subcarriers are **orthogonal** over the symbol interval;
 - Independently of the relative phase relationship between subcarriers.

$$\int_0^T \cos(2\pi f_i t + \phi_i) \cos(2\pi f_j t + \phi_j) dt = 0$$

Orthogonal Frequency Division Multiplexing (2/2)

- ISI can be made arbitrarily small through the selection of N :
 - Select sufficiently large N
 - Since N is large sufficiently, the symbol interval can be larger than the time duration of the channel-time dispersion

Modulation

- Each OFDM symbol is the sum of signals modulated by subcarriers.
- OFDM symbol can express as

$$s(t) = \text{Re} \left\{ \sum_{i=-N/2}^{N/2-1} d_{i+N/2} \exp \left[j2\pi \left(f_c - \frac{i+0.5}{2} \right) (t - t_s) \right] \right\} \quad t_s \leq t \leq t_s + T$$

equivalent baseband signal

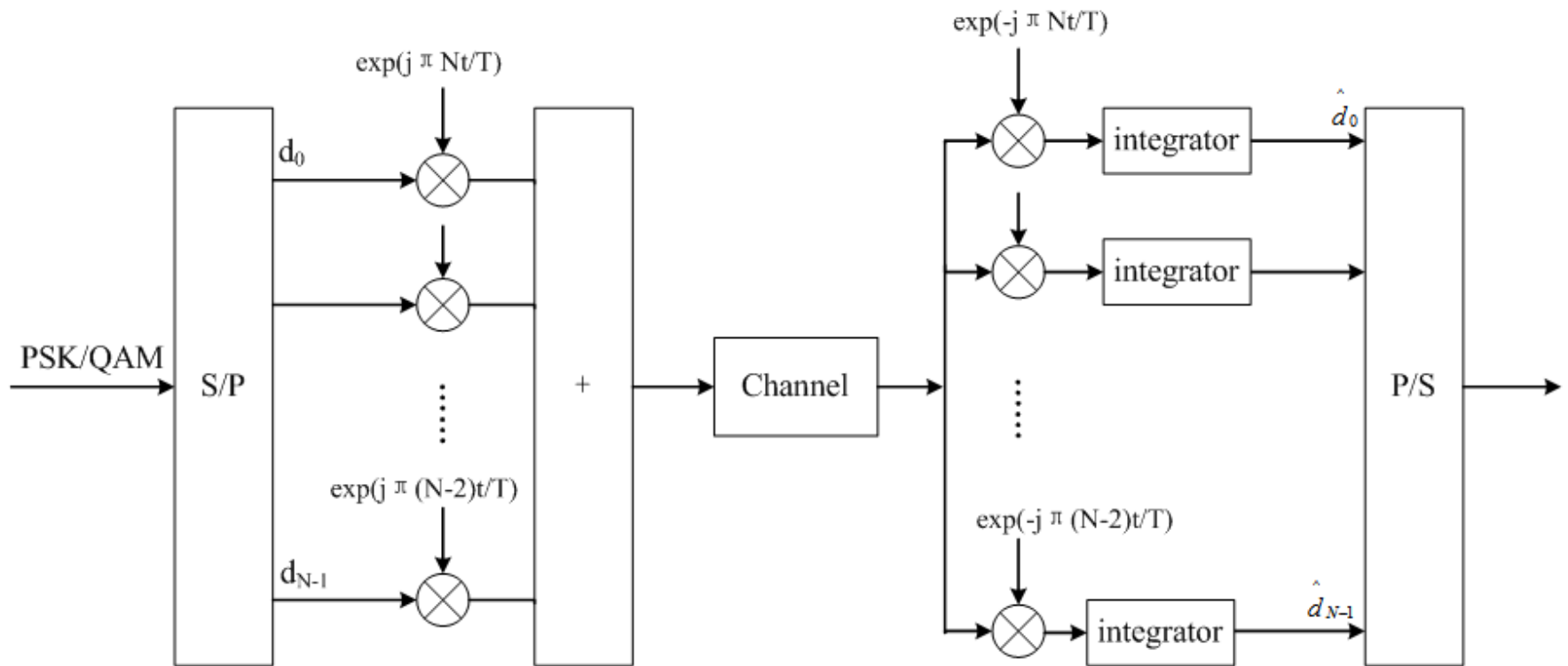
$$s(t) = \sum_{i=-N/2}^{N/2-1} d_{i+N/2} \exp \left[j2\pi \frac{i}{T} (t - t_s) \right] \quad t_s \leq t \leq t_s + T$$

Demodulation

- For the k th signal

$$\begin{aligned}\hat{d}_k &= \frac{1}{T} \int_{t_s}^{t_s+T} \exp\left[j2\pi \frac{k-N/2}{T}(t-t_s)\right] \sum_{i=-N/2}^{N/2-1} d_{i+N/2} \exp\left[-j2\pi \frac{i}{T}(t-t_s)\right] dt \\ &= \frac{1}{T} \sum_{i=-N/2}^{N/2-1} d_{i+N/2} \int_{t_s}^{t_s+T} \exp\left[j2\pi \frac{i-k+N/2}{T}(t-t_s)\right] dt \\ &= d_k\end{aligned}$$

Modulation and Demodulation of OFDM System



FFT and IFFT

- Inverse DFT and IDFT are critical in the implementation of an OFDM system

$$\text{IDFT } x[n] = \frac{1}{N} \sum_{k=0}^{N-1} x[k] e^{-j \frac{2\pi}{N} kn}$$

$$\text{DFT } x[k] = \sum_{n=0}^{N-1} x[n] e^{j \frac{2\pi}{N} kn}$$

- IFFT and FFT algorithms are the fast implementation for the IDFT and DFT

Modulation Using IDFT

- OFDM equivalent baseband signal can implement with IDFT.

$$s(t) = \sum_{i=-N/2}^{N/2-1} d_{i+N/2} \exp\left[j2\pi \frac{i}{T}(t-t_s)\right] \quad t_s \leq t \leq t_s + T$$

let $t_s = 0$, $t = kT/N, k = 0, 1, \dots, N-1$

$$s(k) = s\left(\frac{kT}{N}\right) = \sum_{i=0}^{N-1} d_i \exp\left(j \frac{2\pi ki}{N}\right) \quad 0 \leq k \leq N-1$$

$s(k)$ is IDFT of d_i

Demodulation Using DFT

- At the receiver

$$d_i = \sum_{k=0}^{N-1} s(k) \exp\left(-j \frac{2\pi ki}{N}\right) \quad 0 \leq i \leq N-1$$

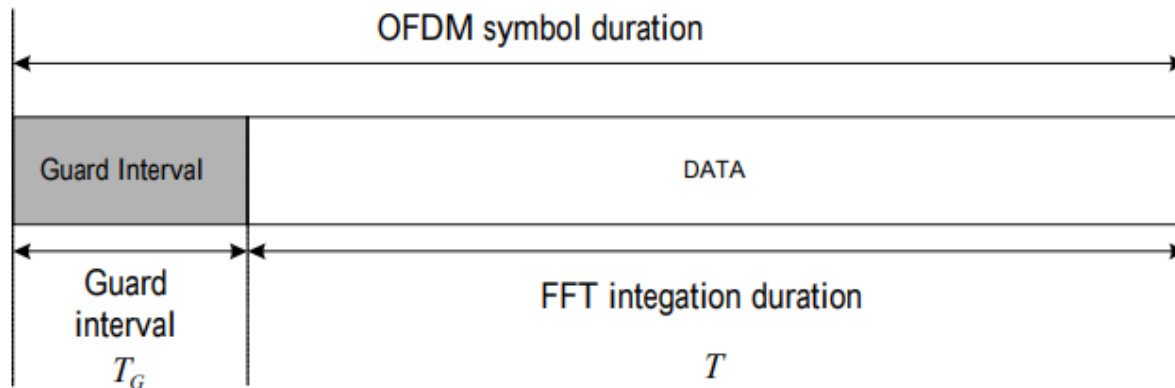
- Modulating the transmitted signal and demodulating the received signal is equivalent to the computation of the IDFT and its inverse.

OFDM via the FFT Algorithm

- Algorithm complexity
 - DFT/IDFT: $O(N^2)$
 - FFT/IFFT: $O\left(\frac{N}{2}\log_2(N)\right)$
- Implement the OFDM system via the FFT algorithm instead of DFT algorithm can be achieved more easily.

Eliminate ISI (1/2)

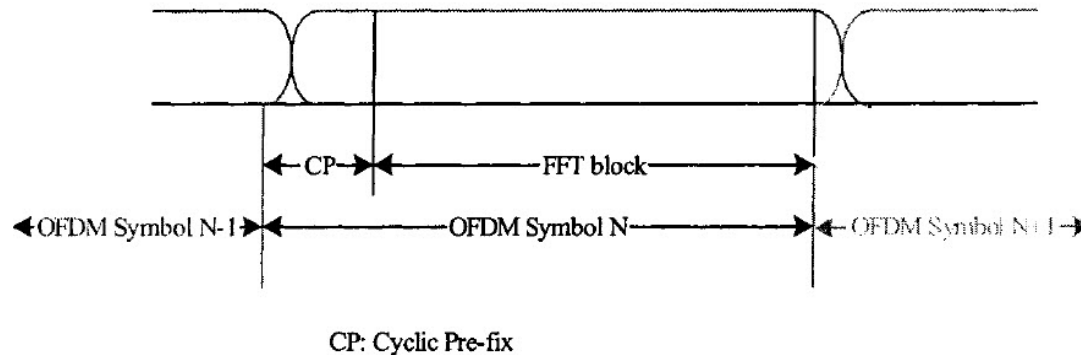
- To eliminate ISI, we need to insert **guard interval** between OFDM symbols.



- If the guard interval is larger than the delay spread, the multipath component of a symbol does not interfere with adjacent symbols.

Eliminate ISI (2/2)

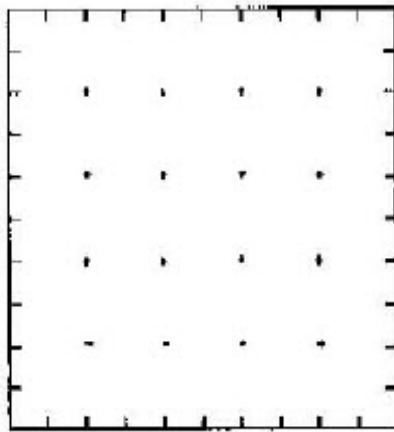
- An alternative approach is to append **cyclic prefix** to each OFDM symbol



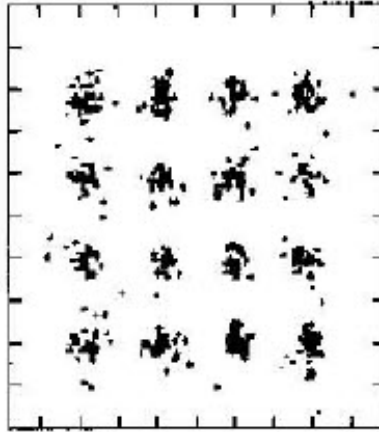
- Since the ISI in any pair of successive signal transmission block **only affects the cyclic prefix**, we discard the affected samples of received signal to eliminate ISI

Cyclic Prefix and Delay Spread

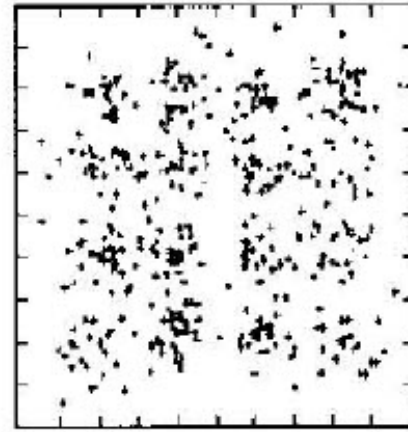
- 3 cases:
 - (a) Delay Spread < Cyclic Prefix
 - (b) Delay Spread is 3% over Cyclic Prefix
 - (c) Delay Spread is 10% over Cyclic Prefix



(a)

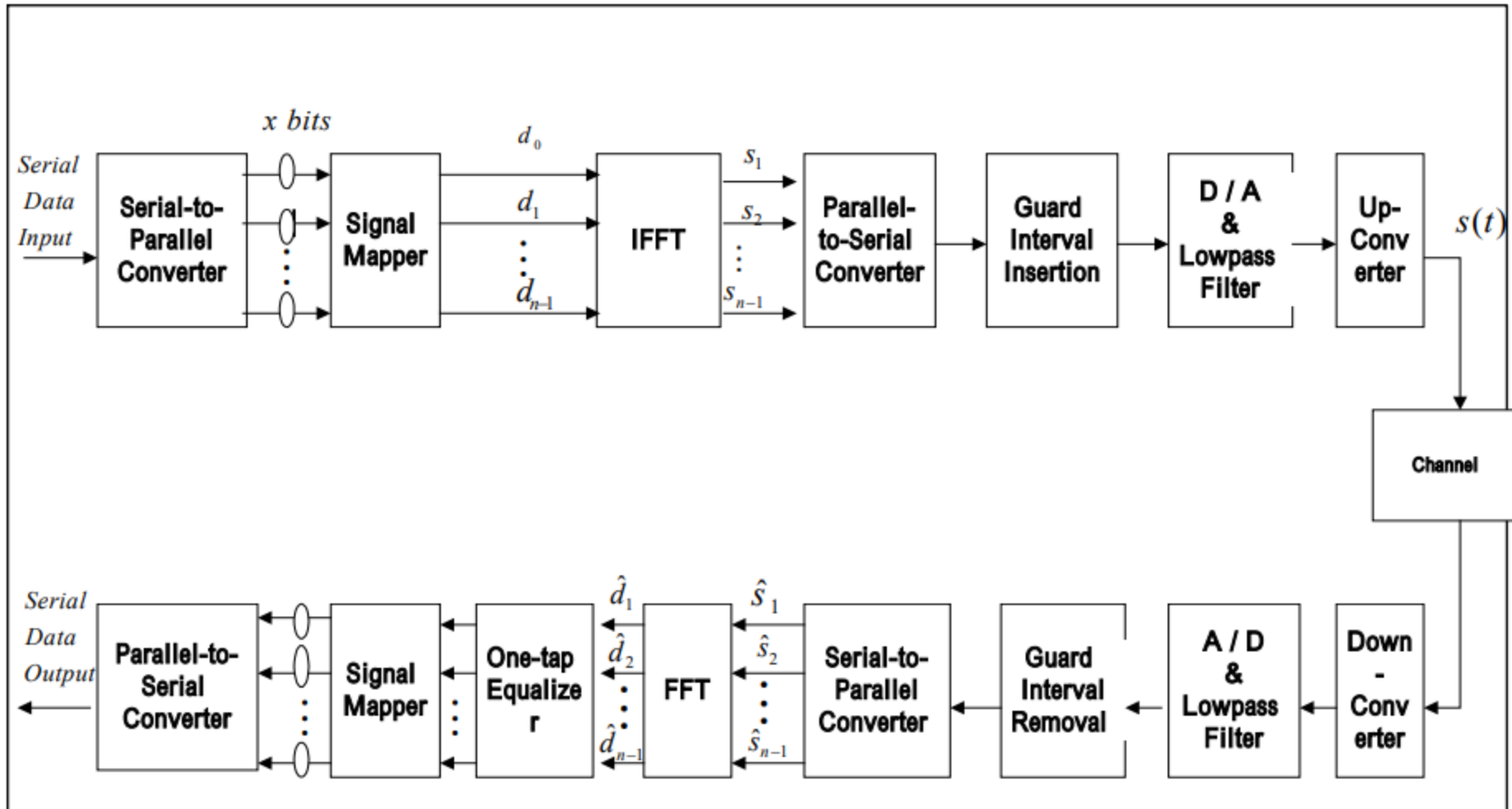


(b)



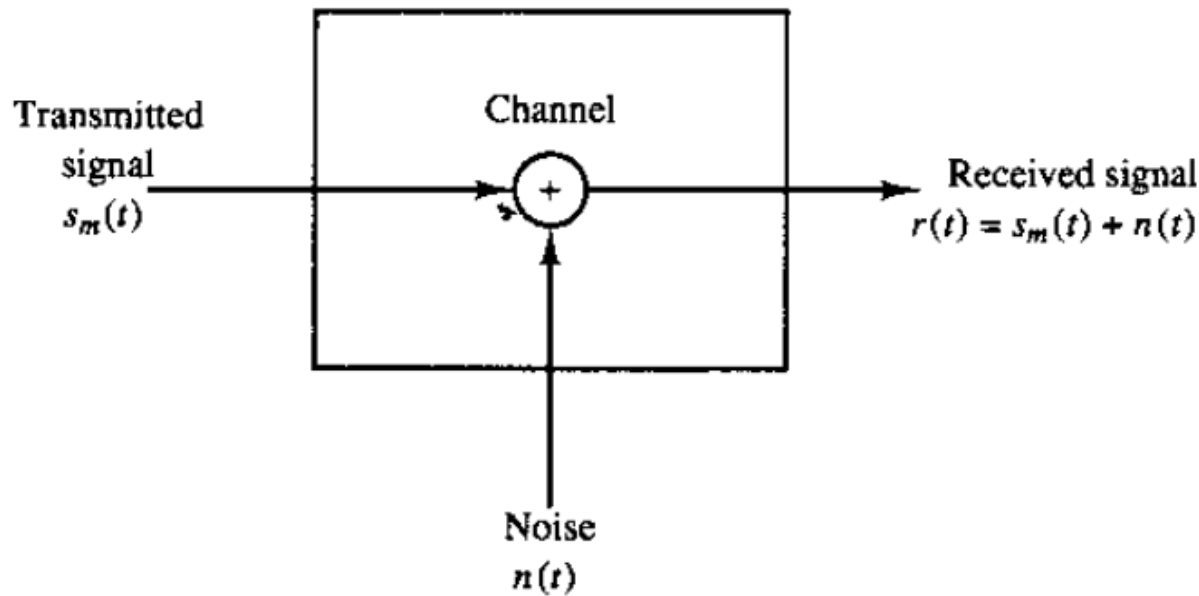
(c)

Multicarrier OFDM Communication System

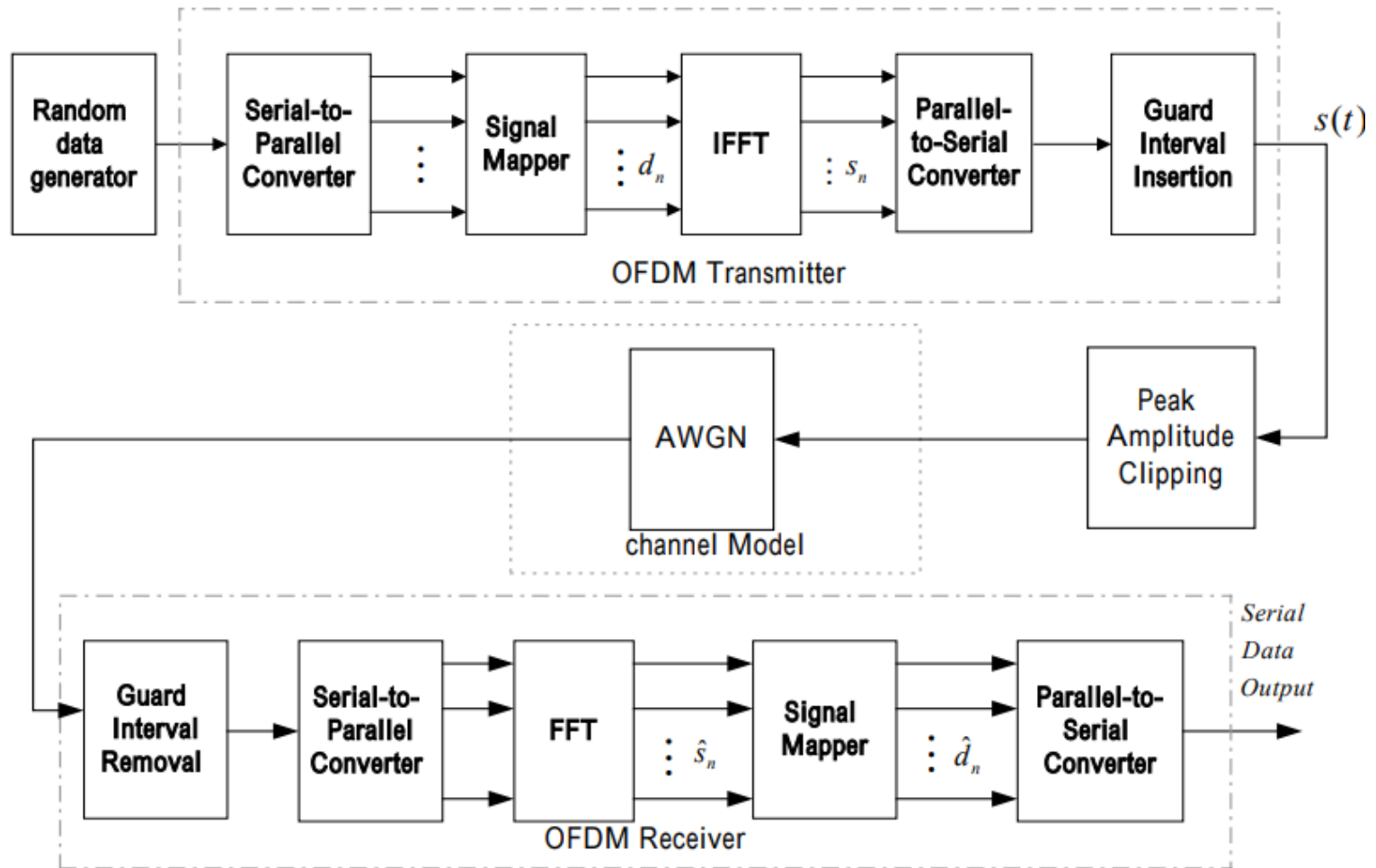


The AWGN Channel Model

- The channel is assumed to corrupt the signal by the addition of white Gaussian noise as shown in the following figure.



OFDM System Performance over AWGN Channel (1/3)



OFDM System Performance over AWGN Channel (2/3)

- QAM

$$p_{e,M-QAM} = \frac{1}{\log_2 M} \left\{ 1 - \frac{1}{M} \left[\left(\sqrt{M} - 2 \right)^2 \cdot p(c|I) + 4 \left(\sqrt{M} - 2 \right) p(c|II) + 4 p(c|III) \right] \right\}$$

$$a = \sqrt{\frac{3E_s}{2(M-1)}}$$

$$p(c|I) = \left[1 - 2Q \left(\sqrt{\frac{2a^2}{N_0}} \right) \right]^2$$

$$p(c|II) = \left[1 - 2Q \left(\sqrt{\frac{2a^2}{N_0}} \right) \right] \left[1 - Q \left(\sqrt{\frac{2a^2}{N_0}} \right) \right]$$

$$p(c|III) = \left[1 - Q \left(\sqrt{\frac{2a^2}{N_0}} \right) \right]^2$$

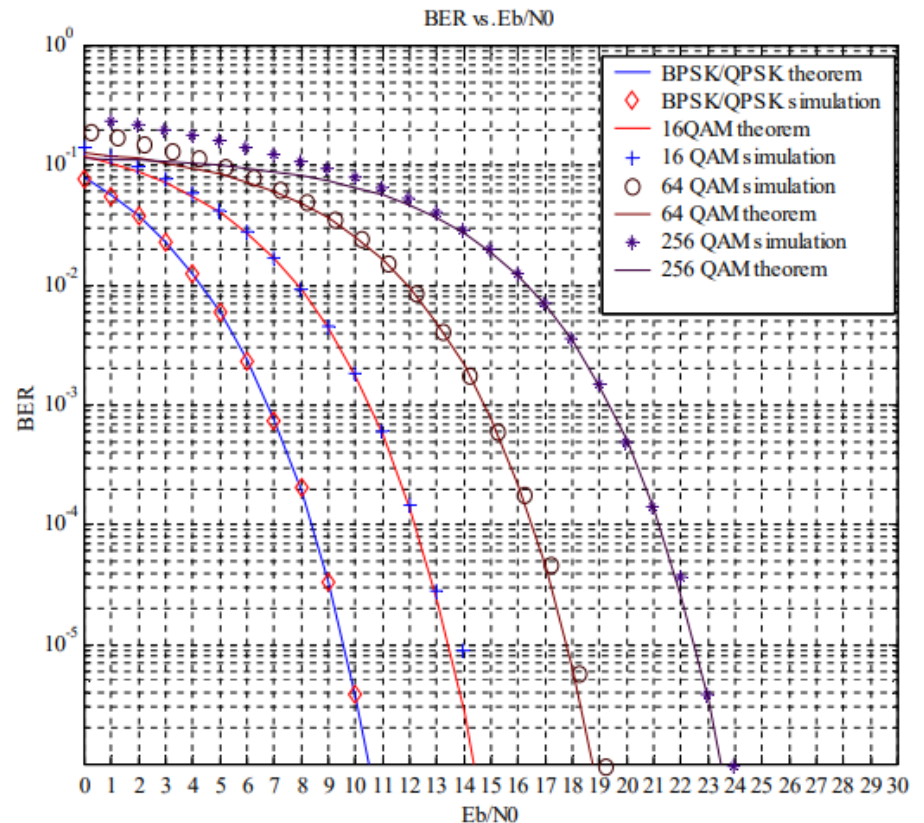
- M-ary PSK

$$p_e \approx \text{erfc} \left(\sqrt{\frac{E_s}{N_0}} \sin \left(\frac{\pi}{M} \right) \right)$$

OFDM System Performance over AWGN Channel (3/3)

- BER versus SNR curves for the OFDM system in AWGN channel using BPSK/QPSK, 16QAM, 64QAM, 256QAM

Simulation parameter	Value
Channel	AWGN
FFT size	1024
Subcarrier #	1024
Modulation	BPSK, QPSK, 16QAM, 64QAM, 256QAM
Guard Type	Cyclic Prefix
SNR	0 - 30 dB
Delay spread	No
Doppler frequency	No
Reflected power	No
Power Clipped	No



SNR per subchannel

- The SNR per subchannel can be defined as:

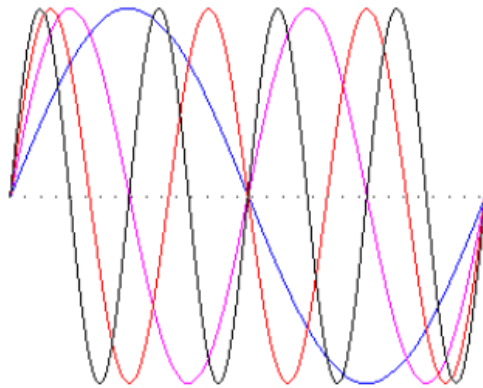
$$\text{SNR}_k = \frac{TP_k |C_k|^2}{\sigma_{nk}^2}$$

- In subchannels with high SNR, we transmit more bits/symbol. Thus, the bit rate on each subchannel can be optimized

Spectra of OFDM Signal (1/2)

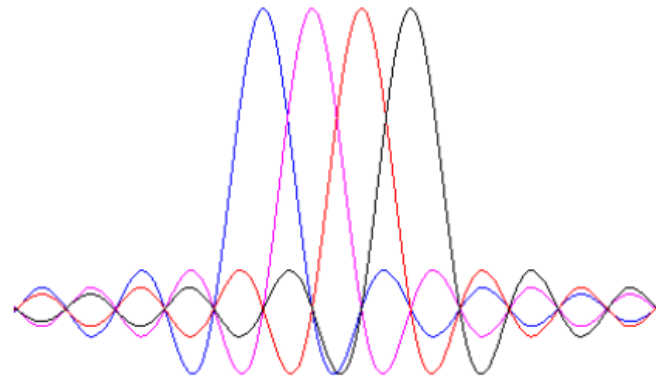
- OFDM signals are **orthogonal** in the time domain but have significant **overlap** in frequency domain.

Time domain



Example of four subcarriers within one OFDM symbol

Frequency domain



Spectra of individual subcarriers

Spectra of OFDM Signal (2/2)

- The large spectral overlap of the OFDM signals causes intersubchannel interference (**ICI**) when the communication channel is a fading channel
- OFDM may not be as robust as a single carrier system in radio communications where the receiving terminal is moving at high speed

Peak-to-Average Power Ratio

- Major problem: **high PAPR**
 - Large peaks occurs when the signals in the K subchannels add constructively in phase.
 - Such large signal peaks may saturate the power amplifier at the transmission, thus, causing intermodulation distortion in the transmitted signal.
- Solution:
 - Insert different phase shifts in each of the subcarriers;
 - Modulate a small subset of the subcarriers with dummy symbol.

Applications of OFDM

- Digital audio broadcasting (DAB)
- Digital video broadcasting (DVB)
- Digital subscriber lines (DSL)
- Wireless local area networks (LANs)

Advantages and disadvantages of OFDM

- Advantages:
 - High data rate
 - Immunity to delay spread
 - Efficient bandwidth
- Disadvantages:
 - The problem of Synchronization
 - Need FFT units at transmitter, receiver
 - The problem of high peak to average power ratio (PAPR)