



Plug N Drive™ Integrated Power Module for Appliance Motor Drive

Description

International Rectifier's IRAMS06UP60A is an Integrated Power Module developed and optimized for electronic motor control in appliance applications specifically for VF compressor drives for refrigerators and freezer or in heating and ventilation as electronic fan controls. The IRAMS06UP60A offers an extremely compact, high performance AC motor-driver in a single isolated package for a very simple design. A built-in temperature monitor and over-current protection, along with the short-circuit rated IGBTs and integrated under-voltage lockout function, deliver high level of protection and fail-safe operation. The integration of the bootstrap diodes for the high-side driver section, and the single polarity power supply required to drive the internal circuitry, simplify the utilization of the module and deliver further cost reduction advantages.

Features

- Integrated Gate Drivers and Bootstrap Diodes
- Temperature Monitor
- · Overcurrent shutdown
- · Fully Isolated Package.
- Low V_{CE(on)} Non Punch Through IGBT Technology
- · Undervoltage lockout for all channels
- Matched propagation delay for all channels
- Schmitt-triggered input logic
- Cross-conduction prevention logic
- · Lower di/dt gate driver for better noise immunity
- Motor Power range 0.1~0.5kW / 85~253 Vac
- Isolation 2000V_{RMS}/1min

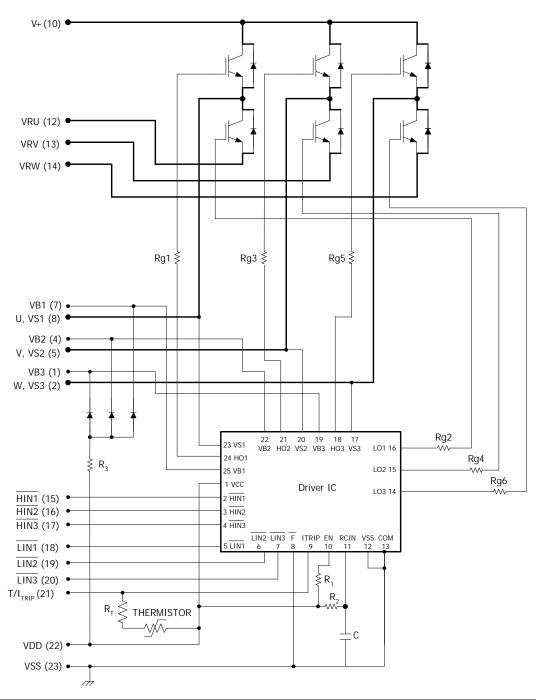


Absolute Maximum Ratings

| Parameter | Description | Max. Value | Units |
|--|---------------------------------------|-------------|------------------|
| V _{CES} | Maximum IGBT Blocking Voltage | 600 | V |
| V ⁺ | Positive Bus Input Voltage | 450 | V |
| I _o @ T _C =25°C | RMS Phase Current | 6 | |
| I _o @ T _C =100°C | RMS Phase Current | 3 | Α |
| I_{pk} | Maximum Peak Phase Current (tp<100ms) | 10 | |
| Fp | Maximum PWM Carrier Frequency | 20 | kHz |
| P_d | Maximum Power dissipation per Phase | 7.5 | W |
| V _{iso} | Isolation Voltage (1min) | 2000 | V _{RMS} |
| T _J (IGBT & Diodes) | Operating Junction temperature Range | -40 to +150 | °C |
| T _{J (Driver IC)} | Operating Junction temperature Range | -40 to +150 | C |
| Т | Mounting torque Range (M3 screw) | 0.8 to 1.0 | Nm |



Internal Electrical Schematic - IRAMS06UP60A





IRAMS06UP60A

Inverter Section Electrical Characteristics @ $T_J = 25$ °C

| Symbol | Parameter | Min | Тур | Max | Units | Conditions |
|-----------------------------------|--|-----|------|------|-------|---|
| V _{(BR)CES} | Collector-to-Emitter Breakdown Voltage | 600 | | | V | $V_{IN} = 5V$, $I_C = 250 \mu A$ |
| $\Delta V_{(BR)CES}$ / ΔT | Temperature Coeff. Of Breakdown Voltage | | 0.3 | | V/°C | V _{IN} =5V, I _C =1.0mA (25°C - 150°C) |
| V | Collector-to-Emitter Saturation | | 1.9 | 2.4 | V | $I_C=3A$, $V_{DD}=15V$ |
| V _{CE(ON)} | Voltage | | 2.2 | 2.6 | V | I _C =3A, V _{DD} =15V, T _J =150°C |
| 1 | Zero Gate Voltage Collector-to- Emitter Current | | 15 | 45 | | V _{IN} =5V, V ⁺ =600V |
| I _{CES} | | | 60 | 170 | μΑ | $V_{IN}=5V$, $V^{+}=600V$, $T_{J}=150$ °C |
| I _{Ik_module} | Zero Gate Voltage Phase-to- phase Current | | | 50 | μА | V _{IN} =5V, V ⁺ =600V |
| V | Diode Forward Voltage Drop | | 1.45 | 1.85 | V | I _C =3A |
| V _{FM} | | | 1.25 | 1.65 | V | $I_{c}=3A$ $I_{c}=3A, T_{J}=150^{\circ}C$ |

Inverter Section Switching Characteristics @ T_J = 25°C

| Symbol | Parameter | Min | Тур | Max | Units | Conditions |
|------------------|--------------------------------------|-------------|-----|-----|---|---|
| E _{on} | Turn-On Switching Loss | | 130 | 235 | | $I_C=3A, V^+=400V$ |
| E _{off} | Turn-Off Switching Loss | | 65 | 120 | μJ | V_{DD} =15V, L=1mH |
| E _{tot} | Total Switching Loss | | 195 | 355 | | See CT1 T _J =25°C |
| E _{on} | Turn-on Swtiching Loss | | 200 | 345 | | T _J =150°C |
| E _{off} | Turn-off Switching Loss | | 90 | 150 | μJ | Energy losses include "tail" and |
| E _{tot} | Total Switching Loss | | 290 | 495 | | diode reverse recovery |
| Erec | Diode Reverse Recovery energy | | 50 | 110 | μJ | $T_J = 150 ^{\circ}\text{C}, \ V^+ = 400 ^{\circ}\text{V}_{DD} = 15 ^{\circ}\text{V},$ |
| t _{rr} | Diode Reverse Recovery time | | 150 | 200 | ns | I _F =3A, L=1mH |
| RBSOA | Reverse Bias Safe Operating Area | FULL SQUARE | | | T_J =150°C, I_C =3A, V_P =600V V ⁺ =480V, V_{DD} =+15V to 0V See CT3 | |
| SCSOA | Short Circuit Safe Operating Area | 10 | | | μs | $\begin{split} &T_J {=}150^{\circ}\text{C}, \ V_P {=}600\text{V}, \\ &V^{+} {=}360\text{V}, \\ &V_{DD} {=}{+}15\text{V} \ to \ 0\text{V} \end{split} \qquad \text{See CT2}$ |

Thermal Resistance

| Symbol | Parameter | Min | Тур | Max | Units | Conditions |
|----------------------|---|-----|-----|-----|-------|---|
| $R_{th(J-C)}$ | Junction to case thermal resistance, each IGBT under inverter operation. | | | 6.5 | °C/W | |
| R _{th(J-C)} | Junction to case thermal resistance, each Diode under inverter operation. | | | 9 | °C/W | Flat, greased surface. Heatsink compound thermal conductivity - 1W/mK |
| R _{th(C-S)} | Thermal Resistance case to sink | | 0.1 | | °C/W | |

IRAMS06UP60A



Absolute Maximum Ratings Driver Function

Absolute Maximum Ratings indicate substaines limits beyond which damage to the device may occur. All voltage parameters are absolute voltages referenced to $V_{\rm SS}$. (Note 1)

| Symbol | Definition | Min | Max | Units |
|---------------------|---|------|-----|-------|
| V _{S1,2,3} | High Side offset voltage | -0.3 | 600 | V |
| V _{B1,2,3} | High Side floating supply voltage | -0.3 | 20 | V |
| V _{DD} | Low Side and logic fixed supply voltage | -0.3 | 20 | V |
| V _{IN} | Input voltage LIN, HIN, T/I _{TRIP} | -0.3 | 7 | V |
| T _J | Juction Temperature | -40 | 150 | °C |

Recommended Operating Conditions Driver Function

The Input/Output logic timing diagram is shown in Figure 1. For proper operation the device should be used within the recommended conditions. All voltages are absolute referenced to V_{SS} . The V_S offset is tested with all supplies biased at 15V differential (Note 1). All input pin (V_{IN}) and I_{TRIP} are clamped with a 5.2V zener diode and pull-up resistor to V_{DD}

| Symbol | Definition | Min | Max | Units | |
|---------------------|--|--------------|--------------------|-------|--|
| V _{B1,2,3} | High side floating supply voltage | $V_{S} + 12$ | V _S +20 | V | |
| V _{S1,2,3} | High side floating supply offset voltage | Note 2 | 450 | V | |
| V_{DD} | Low side and logic fixed supply voltage | 12 | 20 | V | |
| V _{ITRIP} | T/I _{TRIP} input voltage | V_{SS} | V _{SS} +5 | V | |
| V _{IN} | Logic input voltage LIN, HIN | V_{SS} | V _{SS} +5 | V | |

Static Electrical Characteristics Driver Function

 V_{BIAS} (V_{CC} , $V_{BS1,2,3}$)=15V, unless otherwise specified. The V_{IN} and I_{IN} parameters are referenced to V_{SS} and are applicable to all six channels. (Note 1)

| Symbol | Definition | Min | Тур | Max | Units |
|--|---|------|------|------|-------|
| $V_{IN,th+}$ | Positive going input threshold | 3.0 | | | V |
| $V_{IN,th}$ | Negative going input threshold | | | 0.8 | V |
| V _{CCUV+} V _{BSUV+} | V_{CC} and V_{BS} supply undervoltage Positive going threshold | 10.6 | 11.1 | 11.6 | V |
| V _{CCUV-} V _{BSUV-} | V_{CC} and V_{BS} supply undervoltage Negative going threshold | 10.4 | 10.9 | 11.4 | V |
| V _{CCUVH} V _{BSUVH} | V_{CC} and V_{BS} supply undervoltage I_{lockout} hysteresis | | 0.2 | | V |
| I _{QBS} | Quiescent V _{BS} supply current | | 70 | 120 | μΑ |
| I _{QCC} | Quiscent V _{CC} supply current | | 1.6 | 2.3 | mA |
| I _{LK} | Offset Supply Leakage Current | | | 50 | μΑ |
| I _{IN+} | Input bias current (OUT=LO) | | 100 | 220 | μΑ |
| I _{IN+} | Input bias current (OUT=HI) | | 200 | 300 | μΑ |
| V(I _{TRIP}) | I _{TRIP} threshold Voltage (OUT=HI or OUT=LO) | 3.85 | 4.3 | 4.75 | V |



Dynamic Electrical Characteristics

 $V_{DD} \! = \! V_{BS} \! = \! V_{BIAS} \! = \! 15V, \; I_0 \! = \! 1A, \; V_D \! = \! 9V, \; PWM_{IN} \! = \! 2kHz, \; V_{IN_ON} \! = \! V_{IN_th+}, \; V_{IN_OFF} \! = \! V_{IN_th-}, \; V_{IN_th-$

T_A=25°C, unless otherwise specified

| Symbol | Definition | Min | Тур | Max | Units |
|---------------------|--|-----|-----|-----|-------|
| T _{ON} | Input to output propagation turn-on delay time (see fig.11) | - | 470 | - | ns |
| T _{OFF} | Input to output propagation turn-off delay time (see fig. 11) | - | 615 | - | ns |
| D _T | Dead Time | - | 290 | - | ns |
| I/T _{Trip} | T/I _{Trip} to six switch to turn-off propagation delay (see fig. 2) | - | 750 | - | ns |
| T _{FCLTRL} | Post I _{Trip} to six switch to turn-off clear time (see fig. 2) | - | 9 | - | ms |

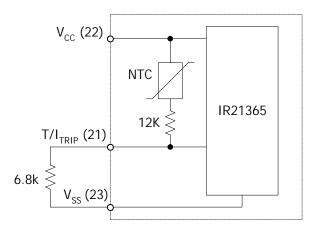
Internal NTC - Thermistor Characteristics

| Parameter | | Тур | Units | Conditions |
|---------------------------|------------|-------------------------|-------|----------------------------------|
| R ₂₅ | Resistance | 100 +/- 3% | kΩ | $T_C = 25^{\circ}C$ |
| R ₁₂₅ | Resistance | 2.522 + 17.3 % /- 14.9% | kΩ | $T_C = 125^{\circ}C$ |
| B B-constant (25-50°C) | | 4250 +/- 2% | k | $R_2 = R_1 e^{[B(1/T2 - 1/T1)]}$ |
| Temperature Range | | -40 / 125 | °C | |
| Typ. Dissipation constant | | 1 | mW/°C | $T_C = 25^{\circ}C$ |

Note 1: For more details, see IR21365 data sheet

Note 2: Logic operational for V_s from V^- -5V to V^- +600V. Logic state held for V_s from V^- -5V to V^- -V_{BS}. (please refer to DT97-3 for more details)

Thermistor Built-in IRAMS06UP60A



Note 3: The Maximum recommended sense voltage at the T/I_{TRIP} terminal under normal operating conditions is 3.3V.

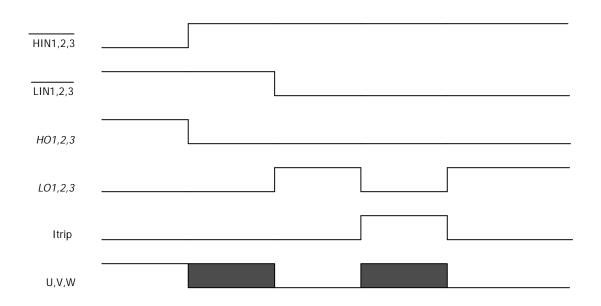
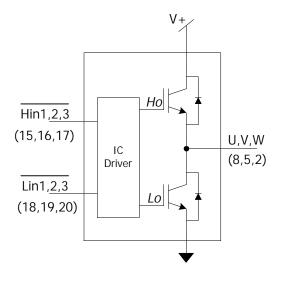


Figure 1. Input/Output Timing Diagram

Note 5: The shaded area indicates that both high-side and low-side switches are off and therefore the half-bridge output voltage would be determined by the direction of current flow in the load.



| Itrip | HIN1,2,3 | LIN1,2,3 | U,V,W |
|-------|----------|----------|-------|
| 0 | 0 | 1 | V+ |
| 0 | 1 | 0 | 0 |
| 0 | 1 | 1 | Χ |
| 1 | Х | Χ | Х |

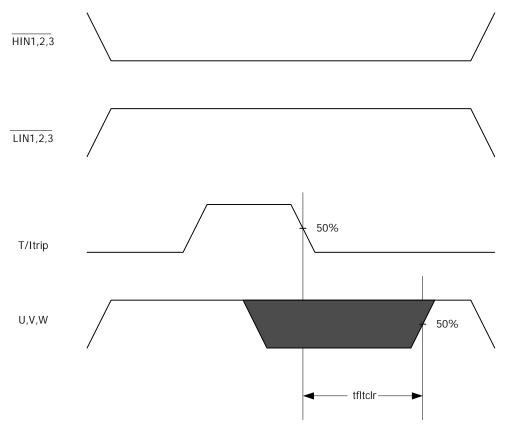


Figure 2. I_{Trip} Timing Waveform

Note 6: The shaded area indicates that both high-side and low-side switches are off and therefore the half-bridge output voltage would be determined by the direction of current flow in the load.

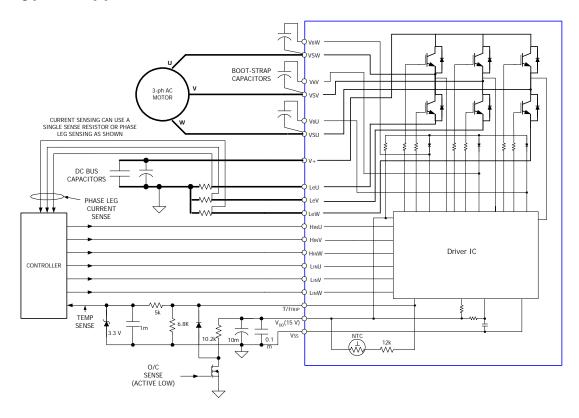


Module Pin-Out Description

| Pin | Name | Description |
|-----|---------|---|
| 1 | VB3 | High Side Floating Supply Voltage 3 |
| 2 | W,VS3 | Output 3 - High Side Floating Supply Offset Voltage |
| 3 | na | none |
| 4 | VB2 | High Side Floating Supply voltage 2 |
| 5 | V,VS2 | Output 2 - High Side Floating Supply Offset Voltage |
| 6 | na | none |
| 7 | VB1 | High Side Floating Supply voltage 1 |
| 8 | U,VS1 | Output 1 - High Side Floating Supply Offset Voltage |
| 9 | na | none |
| 10 | V+ | Positive Bus Input Voltage |
| 11 | na | none |
| 12 | LE1 | Low Side Emitter Connection - Phase 1 |
| 13 | LE2 | Low Side Emitter Connection - Phase 2 |
| 14 | LE3 | Low Side Emitter Connection - Phase 3 |
| 15 | HIN1 | Logic Input High Side Gate Driver - Phase 1 |
| 16 | HIN2 | Logic Input High Side Gate Driver - Phase 2 |
| 17 | HIN3 | Logic Input High Side Gate Driver - Phase 3 |
| 18 | LIN1 | Logic Input Low Side Gate Driver - Phase 1 |
| 19 | LIN2 | Logic Input Low Side Gate Driver - Phase 2 |
| 20 | LIN3 | Logic Input Low Side Gate Driver - Phase 3 |
| 21 | T/Itrip | Temperature Monitor and Shut-down Pin |
| 22 | VCC | +15V Main Supply |
| 23 | VSS | Negative Main Supply |



Typical Application Connection IRAMS06UP60A



- 1. Electrolytic bus capacitors should be mounted as close to the module bus terminals as possible to reduce ringing and EMI problems. Additional high frequency ceramic capacitor mounted close to the module pins will further improve performance.
- 2. In order to provide good decoupling between V_{CC} -Gnd and V_{B} - V_{SS} terminals, the capacitors shown connected between these terminals should be located very close to the module pins. Additional high frequency capacitors, typically 0.1mF, are strongly recommended.
- 3. Value of the boot-strap capacitors depends upon the switching frequency. Their selection should be made based on IR design tip DN 98-2a, application note AN-1044 or Figure 9.
- 4. Low inductance shunt resistors shuld be used for phase leg current sensing. Similarly, the length of the traces between pins 12, 13 and 14 to the corresponding shunt resistors should be kept as small as possible.
- 5. Over-current sense signal can be obtained from external hardware detecting excessive instantaneous current in inverter.

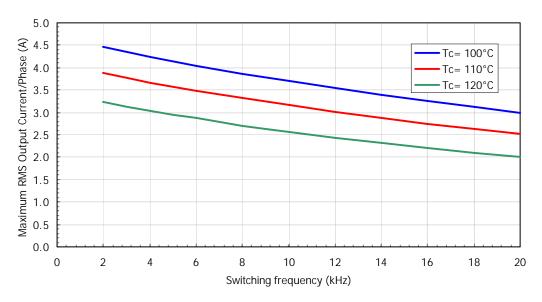


Figure 3. Maximum sinusoidal phase current as function of switching frequency V+=400V, T_j =150°C, Modulation Depth=0.8, PF=0.6

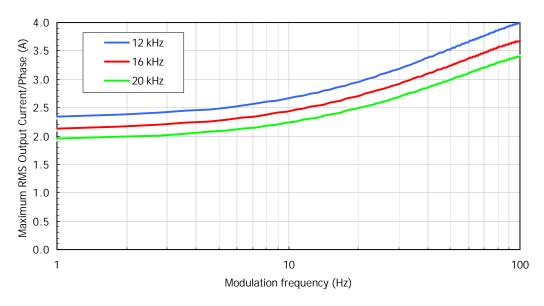


Figure 4. Maximum sinusoidal phase current as function of modulation frequency V+=400V, T_j =150°C, T_c =100°C, Modulation Depth=0.8, PF=0.6

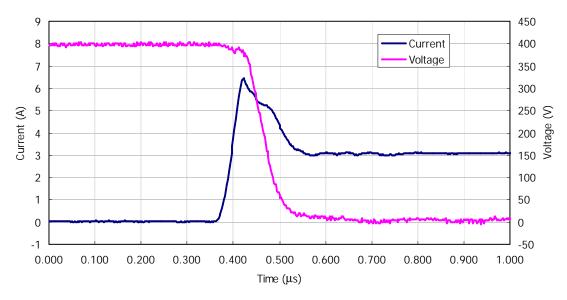


Figure 5. IGBT Turn-on. Typical turn-on waveform $@T_j=125^{\circ}C$, V+=400V

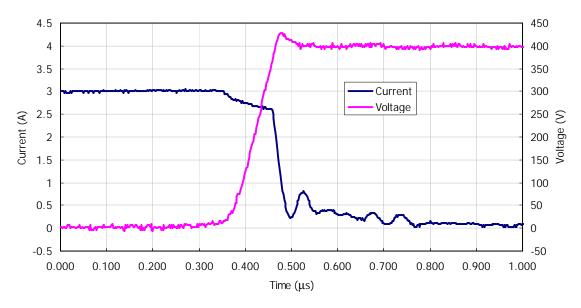


Figure 6. IGBT Turn-off. Typical turn-off waveform $@T_j = 125$ °C, V + = 400V

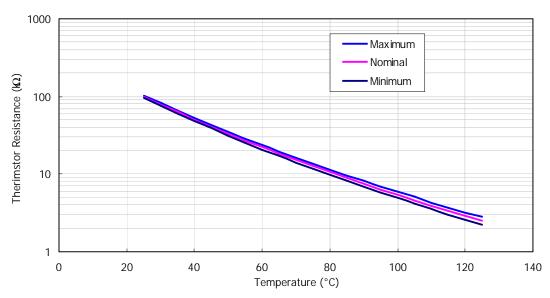


Figure 7. Variation of thermistor resistance with temperature



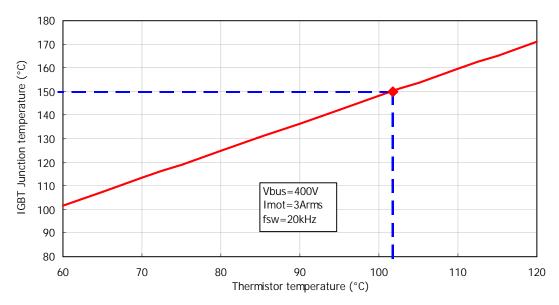


Figure 8. Estimated maximum IGBT junction temperature with thermistor temperature

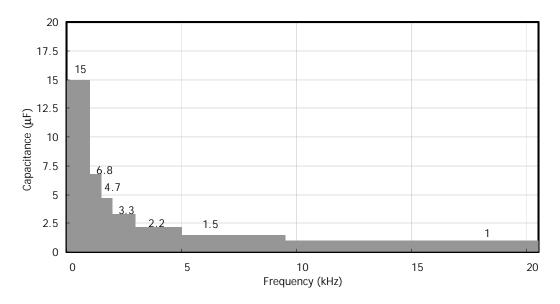


Figure 9. Recommended minimum Bootstrap Capacitor value Vs Switching Frequency

Figure 11. Switching Parameter Definitions

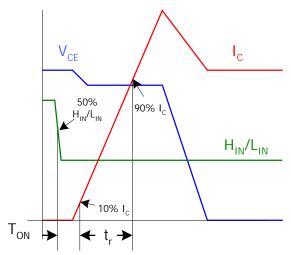


Figure 11a. Input to Output propagation turn-on delay time

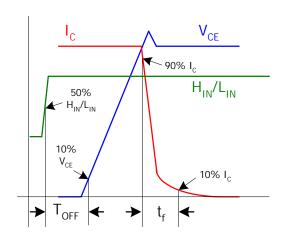


Figure 11b. Input to Output propagation turn-off delay time

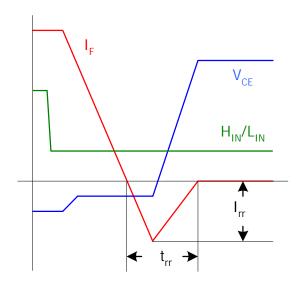


Figure 11c. Diode Reverse Recovery

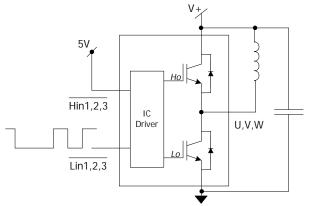


Figure CT1. Switching Loss Circuit

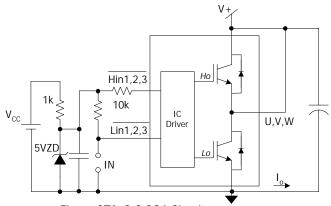
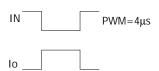
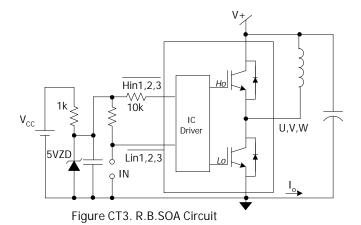


Figure CT2. S.C.SOA Circuit



V_P=Peak Voltage on the IGBT die

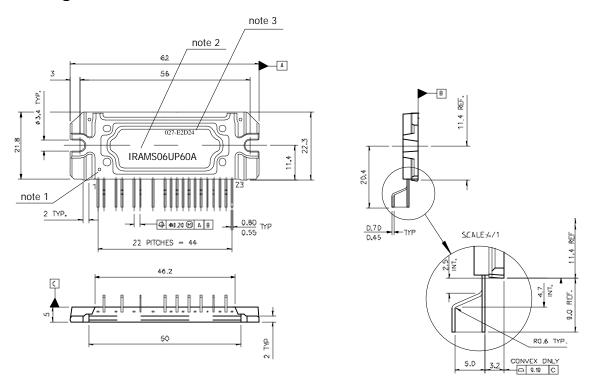


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 $V_P = Peak\ Voltage\ on\ the\ IGBT\ die$



Package Outline



Standard pin leadforming option

Notes:

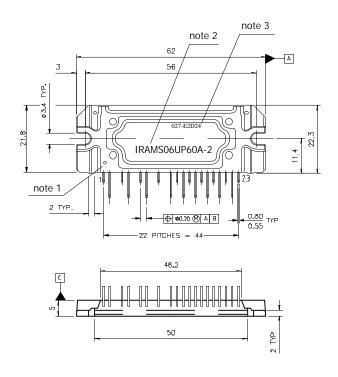
Dimensions in mm

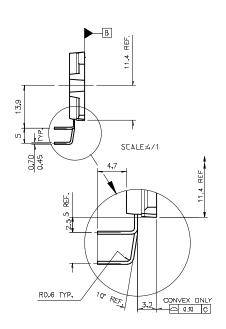
- 1- Marking for pin 1 identification
- 2- Product Part Number
- 3- Lot and Date code marking

For mounting instruction, see AN1049



Package Outline





Pin leadforming option -2

Notes:

Dimensions in mm

- 1- Marking for pin 1 identification
- 2- Product Part Number
- 3- Lot and Date code marking

Data and Specifications are subject to change without notice



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