

A software package for exact linear system identification

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System identification: $w_d \mapsto \hat{\mathcal{B}} \in \mathcal{M}$

Notation

- $w_d = (u_d, y_d)$ — given data, in this talk a vector time series
- $\hat{\mathcal{B}}$ — to be found model for w_d , in this talk an LTI system
- \mathcal{M} — model class, in this talk the set of LTI systems \mathcal{L}

System identification

- defines a mapping $w_d \mapsto \mathcal{B}$
- derives effective algorithms that realize the mapping, and
- develops efficient software that implements the algorithms

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Outline

Introduction: exact and approximate identification

Algorithms for exact identification

Software package

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Exact identification: two points of view

Find the true data generating system

- assume that $w_d \in \tilde{\mathcal{B}} \in \mathcal{L}$
- find back $\tilde{\mathcal{B}}$ from w_d (and an upper bound of the order)
- this is possible provided $\tilde{\mathcal{B}}$ is controllable and an input component of w_d is persistently exciting

Find the least complex LTI system that fits w_d

- no assumption about w_d
- find $\hat{\mathcal{B}} \in \mathcal{L}$ with minimal # of inputs and order, s.t. $w_d \in \hat{\mathcal{B}}$
- $\hat{\mathcal{B}}$ — **most powerful unfalsified model (MPUM)** for w_d in \mathcal{L}

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Exact identification: not a practical SYSID problem

w_d can always be fitted exactly

- take all variables as inputs
- for finite $w_d \in (\mathbb{R}^w)^T$, take the order sufficiently large

\leadsto trivial solution

Of interest are nontrivial solutions, i.e., we want

\mathcal{M} to be a set of **bounded complexity LTI systems** $\mathcal{L}_{m,n_{\max}}$,
of inputs $\leq m$ and order $\leq n_{\max}$.

However,

- $w_d \in \bar{\mathcal{B}} \in \mathcal{L}_{m,n_{\max}}$ is a too restrictive assumption

alternatively

- the MPUM generically does not exist in $\mathcal{L}_{m,n_{\max}}$

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Approximate identification: suboptimal methods

exact identification is more than an academic problem

it leads to suboptimal approximate identification methods

an exact ID method can be used for approximate SYSID by

- rank computation \leftrightarrow **numerical rank computation**
- solution of equations \leftrightarrow **least squares**

MATLAB does this substitution automatically where necessary

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Approximate identification: optimization point of view

- the model need not fit the data exactly
- choose a **distance measure** $M(w_d, \mathcal{B})$ between w_d and \mathcal{B}
- minimize $M(w_d, \mathcal{B})$ over all models in \mathcal{M}

Computing $M(w_d, \mathcal{B})$ is equivalent to

- finding the “best” approximation of w_d in \mathcal{B} ,
- smoothing or filtering (if causality is imposed) w_d by \mathcal{B} ,
- projecting w_d on \mathcal{B} .

$M(w_d, \mathcal{B})$ can be computed in various ways:

smoothing, spectral factorization, Cholesky factorization, ...

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LTI model representations

- **Kernel representation** (parameter $R(z) := \sum_{i=0}^1 R_i z^i$)

$$R_0 w(t) + R_1 w(t+1) + \dots + R_{\perp} w(t+1) = 0$$

- **Impulse response represent** (parameter $h: \mathbb{Z} \rightarrow \mathbb{R}^{p \times m}$)

$$w = \text{col}(u, y), \quad y(t) = \sum_{\tau=-\infty}^t h(\tau) u(t-\tau)$$

- **Input/state/output representation** (parameter (A, B, C, D))

$$w = \text{col}(u, y), \quad \begin{aligned} x(t+1) &= Ax(t) + Bu(t) \\ y(t) &= Cx(t) + Du(t) \end{aligned}$$

$p := \dim(y) = \text{rowdim}(R)$ is the # of outputs

$m := \dim(u)$ is the # of inputs, $\perp := \text{degree}(R)$ is the lag

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Algorithms for exact identification

1. $w_d \mapsto R(z)$
2. $w_d \mapsto$ impulse response H
3. $w_d \mapsto (A, B, C, D)$ (possibly balanced)
 - 3.1 $w_d \mapsto R(\xi) \mapsto (A, B, C, D)$ or $w_d \mapsto H \mapsto (A, B, C, D)$
 - 3.2 $w_d \mapsto \mathcal{O}_{1_{\max}+1}(A, C) \mapsto (A, B, C, D)$
 - 3.3 $w_d \mapsto (x_d(1), \dots, x_d(n_{\max} + m + 1)) \mapsto (A, B, C, D)$

Various ways to implement the mapping $w_d \mapsto (A, B, C, D)$.

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$$w_d \mapsto H$$

Assuming $(u_d, y_d) \in \mathcal{B}$, \mathcal{B} controllable, and u_d persist. exciting, there is G , such that $H = \mathcal{H}_t(y_d)G$.

$w_d \mapsto H$ reduces to the problem of finding a particular G .

$$\begin{bmatrix} \mathcal{H}_{l+t}(u_d) \\ \mathcal{H}_{l+t}(y_d) \end{bmatrix} G = \begin{bmatrix} 0 \\ [I] \\ 0 \\ H \end{bmatrix} \quad \begin{array}{l} \leftarrow l \text{ zero samples} \\ \leftarrow t \text{ samples long impulse} \\ \leftarrow l \text{ zero samples} \\ \leftarrow t \text{ samples impulse response} \end{array}$$

Block algorithm $w_d \mapsto H$

1. Solve the system of equations in blue for G .
2. Substitute G in the equations in red $\rightsquigarrow H$.

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$$w_d \mapsto R(z)$$

The difference equation representation

$$R_0 w_d(t) + R_1 w_d(t+1) + \dots + R_l w_d(t+l) = 0, \quad \text{for } t = 1, \dots, T-l$$

is equivalent to the linear system of equations

$$\underbrace{\begin{bmatrix} R_0 & R_1 & \dots & R_l \end{bmatrix}}_R \underbrace{\begin{bmatrix} w_d(1) & w_d(2) & \dots & w_d(T-l) \\ w_d(2) & w_d(3) & \dots & w_d(T-l+1) \\ \vdots & \vdots & & \vdots \\ w_d(l+1) & w_d(l+2) & \dots & w_d(T) \end{bmatrix}}_{\mathcal{H}_{l+1}(w_d)} = 0.$$

Finding R , requires to compute the left kernel of $\mathcal{H}_{l+1}(w_d)$.

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Simulation example $w_d \mapsto H$

Simulation setup

- \mathcal{B} is of order $n = 4$, lag $l = 2$, with $m = 2$ inputs, and $p = 2$ outputs
- w_d is a trajectory of \mathcal{B} with length $T = 500$

Compared algorithms

- the block algorithm
- an iterative refinement of the block algorithm
- the function `impulse` from the Identification Toolbox

Approximation error $e = \|H - \hat{H}\|_F$ and execution time

method	error, e	time, sec.
block algorithm	10^{-14}	0.293

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$$w_d \mapsto (A, B, C, D)$$

Possible paths to go

- $w_d \mapsto H(0 : 2l_{\max})$ or $R(z) \xrightarrow{\text{realization}} (A, B, C, D)$
- $w_d \mapsto \mathcal{O}_{l_{\max}+1}(A, C) \xrightarrow{(1)} (A, B, C, D)$
- $w_d \mapsto (x_d(1), \dots, x_d(n_{\max} + m + 1)) \xrightarrow{(2)} (A, B, C, D)$

(1) and (2) are easy

$$\mathcal{O}_{l_{\max}+1}(A, C) \mapsto (A, C) \quad \text{and} \quad (u_d, y_d, A, C) \mapsto (B, C, x_{\text{ini}}) \quad (1)$$

$$\begin{bmatrix} x_d(2) & \cdots & x_d(n_{\max} + m + 1) \\ y_d(1) & \cdots & y_d(n_{\max} + m) \end{bmatrix} = \begin{bmatrix} A & B \\ C & D \end{bmatrix} \begin{bmatrix} x_d(1) & \cdots & x_d(n_{\max} + m) \\ u_d(1) & \cdots & u_d(n_{\max} + m) \end{bmatrix} \quad (2)$$

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$$w_d \mapsto \mathcal{O}_{l_{\max}+1}(A, C)$$

- the columns of $\mathcal{O}_{l_{\max}+1}(A, C)$ are lin. indep. free resp. of \mathcal{B}
- under the conditions of FL, such resp. can be computed

$$\begin{bmatrix} \mathcal{H}_t(u_d) \\ \mathcal{H}_t(y_d) \end{bmatrix} G = \begin{bmatrix} 0 \\ Y_0 \end{bmatrix} \quad \begin{array}{l} \leftarrow \text{zero inputs} \\ \leftarrow \text{free responses} \end{array}$$

(G should be maximal rank)

- once we have a maximal rank matrix of free responses Y_0

$$Y_0 = \mathcal{O}_{l_{\max}+1}(A, C) \underbrace{\begin{bmatrix} x_{\text{ini},1} & \cdots & x_{\text{ini},j} \end{bmatrix}}_{X_{\text{ini}}} \quad \text{rank revealing factorization}$$

- the factorization fixes the basis for $\mathcal{O}_{l_{\max}+1}(A, C)$ and X_{ini}

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$$\mathcal{O}_{l_{\max}+1}(A, C) \mapsto (A, B, C, D)$$

First C and A

C is the first block entry of $\mathcal{O}_{l_{\max}+1}(A, C)$ and A is given by

$$(\sigma^* \mathcal{O}_{l_{\max}+1}(A, C))A = (\sigma \mathcal{O}_{l_{\max}+1}(A, C)) \quad \text{shift equation}$$

(σ^* removes the last block entry and σ removes the first block entry)

Then D, B, and $x_d(1)$

Once C and A are known, the system of equations

$$y_d(t) = CA^t x_d(1) + \sum_{\tau=1}^{t-1} CA^{t-1-\tau} B u_d(\tau) + D \delta(t+1),$$

is linear in D, B, $x_d(1)$ and can be solved explicitly.

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$$w_d \mapsto (x_d(1), \dots, x_d(n_{\max} + m + 1))$$

Main idea

If the free resp. are sequential, i.e., if Y_0 is block-Hankel, then X_{ini} is a state sequence of \mathcal{B} .

Computation of sequential free responses

$$\begin{bmatrix} U_p \\ Y_p \\ U_f \end{bmatrix} G = \begin{bmatrix} U_p \\ Y_p \\ 0 \end{bmatrix} \quad \left. \begin{array}{l} \text{sequential ini. conditions} \\ \leftarrow \text{zero inputs} \end{array} \right\} \quad Y_f \quad G = Y_0$$

$$Y_0 = \mathcal{O}_{l_{\max}+1}(A, C) \begin{bmatrix} x_d(1) & \cdots & x_d(n_{\max} + m + 1) \end{bmatrix} \quad \text{rank revealing factorization}$$

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Building blocks for the algorithms

Function	Description
w2r	from data (time series w) to a kernel repr.
r2pq	from a kernel repr. to an LMF representation
pq2ss	from an LMF repr. to an I/S/O representation
uy2h	from data to the impulse response
h2ss	from the impulse resp. to an I/S/O repr.
uy2y0	from data to sequential free responses
y02ox	from free responses to an observability matrix and a state sequence
h2ox	from the impulse response to an observability matrix and a state sequence
uy02ss	from data and an observability matrix to an I/S/O representation
uyx2ss	from data and a state seq. to an I/S/O repr.
hy02xbal	from the impulse response and sequential free responses to a balanced state sequence

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Conclusions

- choice of representation
- decomposition of the identification problem into **standard easy to solve subproblems**
- various ways to achieve the mapping $w_d \mapsto \mathcal{B}$
- can be used as suboptimal approximate ID methods
- open question:

when w_d is not exact, which choice of representation and computational algorithm gives best approximate system?

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Thank you

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