

# The most powerful unfalsified model

Ivan Markovsky

University of Southampton

# Outline

- Exact identification problems
- Identifiability conditions
- Algorithms
  - from data to kernel representation
  - impulse response identification
  - N4SID-type algorithms
  - MOESP-type algorithms

# Exact identification problems

$(w_d \mapsto \mathcal{B} \text{ such that } w_d \in \mathcal{B})$

# An exact identification problem

## Problem P1 (Exact identification)

Given two vector time series

$$u_d = (u_d(1), \dots, u_d(T)) \in (\mathbb{R}^m)^T \quad \text{“inputs”}$$

$$y_d = (y_d(1), \dots, y_d(T)) \in (\mathbb{R}^p)^T \quad \text{“outputs”}$$

find  $n \in \mathbb{N}$  and LTI system  $\mathcal{B}$  of order  $n$ , with  $m$  inputs and  $p$  outputs, s.t.

$$w_d := (u_d, y_d) \in \mathcal{B},$$

i.e.,  $w_d$  is a trajectory of  $\mathcal{B}$ .

How can we check that “ $w_d \in \mathcal{B}$ ”?

## Checking that $w_d \in \mathcal{B} = \ker(R(\sigma))$

$$w_d \in \mathcal{B} \iff R(\sigma)w_d = 0$$

$$\iff R_0 w_d(t) + R_1 w_d(t+1) + \dots + R_\ell w_d(t+\ell) = 0$$

for  $t = 1, \dots, T - \ell$

$$\iff \begin{bmatrix} R_0 & R_1 & \dots & R_\ell & & & \\ & R_0 & R_1 & \dots & R_\ell & & \\ & & \ddots & \ddots & & \ddots & \\ & & & R_0 & R_1 & \dots & R_\ell \end{bmatrix} \begin{bmatrix} w_d(1) \\ w_d(2) \\ \vdots \\ w_d(T) \end{bmatrix} = 0$$

$$\iff \begin{bmatrix} R_0 & R_1 & \dots & R_\ell \end{bmatrix} \begin{bmatrix} w_d(1) & w_d(2) & \dots & w_d(T-\ell) \\ w_d(2) & w_d(3) & \dots & \\ \vdots & \vdots & & \vdots \\ w_d(\ell+1) & w_d(\ell+2) & \dots & w_d(T) \end{bmatrix} = 0$$

## Checking that $w_d \in \mathcal{B} = \text{image}(M(\sigma))$

$w_d \in \mathcal{B} \iff$  there is  $v$ , such that  $w_d = M(\sigma)v$

$\iff$  there is  $v$ , such that for  $t = 1, \dots, T$

$$w_d(t) = M_0 v(t) + M_1 v(t+1) + \dots + M_\ell v(t+\ell)$$

$\iff$  there is solution  $v$  of the system

$$\begin{bmatrix} w_d(1) \\ w_d(2) \\ \vdots \\ w_d(T) \end{bmatrix} = \begin{bmatrix} M_0 & M_1 & \cdots & M_\ell & & & \\ & M_0 & M_1 & \cdots & M_\ell & & \\ & & \ddots & \ddots & & \ddots & \\ & & & M_0 & M_1 & \cdots & M_\ell \end{bmatrix} \begin{bmatrix} v(1) \\ v(2) \\ \vdots \\ v(T+\ell) \end{bmatrix}$$

## Checking that $w_d \in \mathcal{B} = \mathcal{B}_{i/s/o}(A, B, C, D)$

Let  $\mathcal{B}$  be defined by a minimal input/state/output representation

$$\mathcal{B} := \mathcal{B}_{i/s/o}(A, B, C, D) = \{ (u, y) \mid \sigma x = Ax + Bu, y = Cx + Du \}$$

$$(u_d, y_d) \in \mathcal{B}_{i/s/o}(A, B, C, D) \iff \text{there exists } x_{ini} \in \mathbb{R}^n, \text{ such that}$$

$$y_d = \underbrace{\begin{bmatrix} C \\ CA \\ CA^2 \\ \vdots \\ CA^{T-1} \end{bmatrix}}_{\mathcal{O}_T(A, C)} x_{ini} + \begin{bmatrix} D & & & \\ CB & D & & \\ CAB & CB & D & \\ \vdots & \ddots & \ddots & \ddots \\ CA^{T-1}B & \dots & CAB & CB & D \end{bmatrix} u_d$$

( $y_d$  is the response of  $\mathcal{B}$  under input  $u_d$  and initial condition  $x_{ini}$ )

## Comments

- P1 is an **exact fitting problem**, a most basic SYSID problem
- easily generalizable to a **set of  $N$  time series**  
 $u_{d,1}, \dots, u_{d,N} \in (\mathbb{R}^m)^T$  and  $y_{d,1}, \dots, y_{d,N} \in (\mathbb{R}^p)^T$
- the **realization problem**

impulse response  $\mapsto (A, B, C, D)$

is a special case of P1 for a set of  $m$  time series

- while  $m$  is given, **finding  $n$  is part of the problem**  
any observable system of order  $n \geq p$  is a (trivial) solution
- we are actually interested in a **solution of a minimal order**



# Revised exact identification problem

## Problem P1' (Exact identification)

Given two vector time series

$$u_d = (u_d(1), \dots, u_d(T)) \in (\mathbb{R}^m)^T \quad \text{"inputs"}$$

$$y_d = (y_d(1), \dots, y_d(T)) \in (\mathbb{R}^p)^T \quad \text{"outputs"}$$

find the smallest  $n \in \mathbb{N}$  and LTI system  $\mathcal{B}$  of order  $n$ , with  $m$  inputs and  $p$  outputs, such that

$$w_d = (u_d, y_d) \in \mathcal{B}.$$

# Set of LTI systems with a bounded complexity

**Notation:**  $\mathcal{L}_{m,\ell}^{w,n}$  is the set of all LTI systems with

- $w$  (external) variables
- **at most**  $m$  inputs
- minimal state dimension **at most**  $n$  and
- lag (= observability index) **at most**  $\ell$

For  $t \geq n$ , **the set  $\mathcal{B}|_t$  of all  $t$  samples long traj. of  $\mathcal{B}$**  has dimension

$$\dim(\mathcal{B}|_t) \leq tm + n \leq tm + p\ell$$

(where  $p(\ell - 1) \leq n \leq p\ell$ )

$\implies (m, n)$  and  $(m, \ell)$  specify the **complexity** of the model class  $\mathcal{L}_{m,\ell}^{w,n}$

# Another exact identification problem

## Problem P2 (Exact identification)

Given a vector time series

$$w_d = (w_d(1), \dots, w_d(T)) \in (\mathbb{R}^w)^T$$

find the smallest  $m \in \mathbb{N}$  and  $\ell \in \mathbb{N}$  and LTI system  $\mathcal{B} \in \mathcal{L}_{m,\ell}^w$ , s.t.  $w_d \in \mathcal{B}$ .

## Comments:

- no separation between inputs and outputs
- the complexity is defined by  $(m, \ell)$

# Most powerful unfalsified model

The most powerful unfalsified model in the model class  $\mathcal{L}_{m,\ell}^w$  of a time series  $w_d \in (\mathbb{R}^w)^T$  is the system  $\mathcal{B}_{\text{mpum}}$  that is

1. **in the model class**, *i.e.*,  $\mathcal{B}_{\text{mpum}} \in \mathcal{L}_{m,\ell}^w$ ,
2. **unfalsified**, *i.e.*,  $w_d \in \mathcal{B}_{\text{mpum}}|_T$ , and
3. **most powerful** among all LTI unfalsified systems, *i.e.*,

$$\mathcal{B}' \in \mathcal{L}_{m,\ell}^w \text{ and } w_d \in \mathcal{B}'|_T \implies \mathcal{B}_{\text{mpum}}|_T \subseteq \mathcal{B}'|_T.$$

MPUM **may not exist**, but if it does, then it is **unique**

# Identifiability

# Identifiability question

P2 is the problem of computing the MPUM of  $w_d$  in  $\mathcal{L}^w$

The following related question is of interest:

Suppose that

$$w_d \in \overline{\mathcal{B}} \in \mathcal{L}^w$$

and **upper bounds**  $n_{\max}$ ,  $\ell_{\max}$  of the order  $n$  and lag  $\ell$  of  $\overline{\mathcal{B}}$  are given.

Under what conditions  $\mathcal{B}_{\text{mpum}}(w_d)$  is equal to the system  $\mathcal{B}$ ?

the answer is given by the following lemma

# Fundamental Lemma

Let  $\overline{\mathcal{B}} \in \mathcal{L}_m^{w,n}$  be controllable and let  $w_d := (u_d, y_d) \in \overline{\mathcal{B}}|_T$ .

Then, if  $u_d$  is persistently exciting of order  $L + n$ ,

$$\text{image} \left( \begin{bmatrix} w_d(1) & w_d(2) & w_d(3) & \cdots & w_d(T-L+1) \\ w_d(2) & w_d(3) & w_d(4) & \cdots & w_d(T-L+2) \\ w_d(3) & w_d(4) & w_d(5) & \cdots & w_d(T-L+3) \\ \vdots & \vdots & \vdots & & \vdots \\ w_d(L) & w_d(L+1) & w_d(L+2) & \cdots & w_d(T) \end{bmatrix} \right) = \overline{\mathcal{B}}|_L$$

$\implies$  under the conditions of the FL, any  $L$  samples long response  $y$  of  $\mathcal{B}$  can be obtained as  $y = \mathcal{H}_L(y_d)g$ , for certain  $g \rightsquigarrow$  algorithms

$\implies$  with  $L = \ell_{\max} + 1$ , the FL gives conditions for identifiability

## Persistence of excitation

$u_d = (u_d(1), \dots, u_d(T))$  is **persistently exciting of order  $L$**  if

$$\mathcal{H}_L(u_d) := \begin{bmatrix} u_d(1) & u_d(2) & u_d(3) & \cdots & u_d(T-L+1) \\ u_d(2) & u_d(3) & u_d(4) & \cdots & u_d(T-L+2) \\ u_d(3) & u_d(4) & u_d(5) & \cdots & u_d(T-L+3) \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ u_d(L) & u_d(L+1) & u_d(L+2) & \cdots & u_d(T) \end{bmatrix}$$

is full  
row rank

System theoretic interpretation:

$u_d$  is persistently  
exciting of order  $L$



there is no LTI system with  
# of inputs  $< m$  and lag  $< L$   
for which  $u_d$  is a trajectory



# Algorithms for exact identification ( $w_d \mapsto$ representation of the MPUM)

# Overview of algorithms

1.  $w_d \mapsto R(\xi)$
  2.  $w_d \mapsto \text{impulse response } H$
  3.  $w_d \mapsto (A, B, C, D)$  (possibly balanced)
- 
- 3.1  $w_d \mapsto R(\xi) \mapsto (A, B, C, D)$  or  $w_d \mapsto H \mapsto (A, B, C, D)$
  - 3.2  $w_d \mapsto \mathcal{O}_{\ell_{\max}+1}(A, C) \mapsto (A, B, C, D)$
  - 3.3  $w_d \mapsto (x_d(1), \dots, x_d(n_{\max} + m + 1)) \mapsto (A, B, C, D)$

$$w_d \mapsto R(\xi)$$

under the assumptions of the FL, image  $(\mathcal{H}_{\ell_{\max}+1}(w_d)) = \mathcal{B}|_{\ell_{\max}+1}$   
 $\implies$  **a basis for  $\ker(\mathcal{H}_{\ell_{\max}+1}(w_d))$  defines a kernel repr. of  $\mathcal{B}$**

let

$$\begin{bmatrix} \tilde{R}_0 & \tilde{R}_1 & \cdots & \tilde{R}_{\ell_{\max}} \end{bmatrix} \mathcal{H}_{\ell_{\max}+1}(w_d) = 0, \quad \text{where } \tilde{R}_i \in \mathbb{R}^{g \times w}$$

and define  $\tilde{R}(\xi) = \sum_{i=0}^{\ell_{\max}} \xi^i \tilde{R}_i$

then  $\mathcal{B} = \ker(\tilde{R}(\sigma))$  is, in general, a **nonminimal kernel representation**

$$w_d \mapsto R(\xi)$$

$\tilde{R}$  can be made minimal by standard polynomial linear algebra alg.  
find a unimodular matrix  $U$ , such that

$$U\tilde{R} = \begin{bmatrix} R \\ 0 \end{bmatrix} \quad \text{and } R \text{ is full row rank}$$

then  $\ker(R(\sigma)) = 0$  is minimal

### Refinements:

- efficient **recursive computation** (exploiting the Hankel structure)
- as a byproduct **find an input/output partition** of the variables
- find a **shortest lag** kernel representation (*i.e.*,  $R$  row proper)

$$w_d \mapsto H$$

Under the conditions of FL, there is  $G$ , such that  $H = \mathcal{H}_t(y_d)G$   
the problem reduces to the one of finding a particular  $G$ . Define

$$\begin{bmatrix} \mathcal{H}_{\ell_{\max}+t}(u_d) \\ \mathcal{H}_{\ell_{\max}+t}(y_d) \end{bmatrix} =: \begin{bmatrix} U_p \\ U_f \\ Y_p \\ Y_f \end{bmatrix} \quad \begin{array}{lll} \text{row dim}(U_p) & = & \text{row dim}(Y_p) = \ell_{\max} \\ \text{row dim}(U_f) & = & \text{row dim}(Y_f) = t \end{array}$$

Let  $u_d$  be p.e. of order  $t + \ell_{\max} + n_{\max}$ . Then there is  $G$ , such that

$$\begin{bmatrix} U_p \\ Y_p \\ U_f \end{bmatrix} G = \begin{bmatrix} 0 \\ 0 \\ \begin{bmatrix} I_m \\ 0 \end{bmatrix} \end{bmatrix} \left. \begin{array}{l} \} \text{zero ini. conditions} \\ \leftarrow \text{impulse input} \end{array} \right\} \quad (1)$$

$$Y_f G = H$$

$$w_d \mapsto H$$

**Block algorithm** for computation of  $(H(0), \dots, H(t-1))$ :

1. Input:  $u_d$ ,  $y_d$ ,  $\ell_{\max}$ , and  $t$ .
2. Solve the system of eqs (1). Let  $\bar{G}$  be the computed solution.
3. Compute  $H = Y_f \bar{G}$ .
4. Output: the first  $t$  samples of the impulse response  $H$ .

### Refinements:

- solve (1) efficiently by **exploiting the Hankel structure**
- do the computations iteratively for pieces of  $H \rightsquigarrow$  **iterative alg.**
- **automatically choose  $t$** , for a sufficient decay of  $H$

$$w_d \mapsto (A, B, C, D)$$

- $w_d \mapsto H(0 : 2\ell_{\max})$  or  $R(\xi) \xrightarrow{\text{realization}} (A, B, C, D)$
- $w_d \mapsto \mathcal{O}_{\ell_{\max}+1}(A, C) \xrightarrow{(2)} (A, B, C, D)$
- $w_d \mapsto (x_d(1), \dots, x_d(n_{\max} + m + 1)) \xrightarrow{(3)} (A, B, C, D)$

(2) and (3) are easy:

$$\mathcal{O}_{\ell_{\max}+1}(A, C) \mapsto (A, C) \quad \text{and} \quad (u_d, y_d, A, C) \mapsto (B, C, x_{\text{ini}}) \quad (2)$$

$$\begin{bmatrix} x_d(2) & \cdots & x_d(n_{\max} + m + 1) \\ y_d(1) & \cdots & y_d(n_{\max} + m) \end{bmatrix} = \begin{bmatrix} A & B \\ C & D \end{bmatrix} \begin{bmatrix} x_d(1) & \cdots & x_d(n_{\max} + m) \\ u_d(1) & \cdots & u_d(n_{\max} + m) \end{bmatrix} \quad (3)$$

$$\mathcal{O}_{\ell_{\max}+1}(A, C) \mapsto (A, B, C, D)$$

$C$  is the **first block entry** of  $\mathcal{O}_{\ell_{\max}+1}(A, C)$  and  $A$  is given by

$$(\sigma^* \mathcal{O}_{\ell_{\max}+1}(A, C))A = (\sigma \mathcal{O}_{\ell_{\max}+1}(A, C)) \quad \text{shift equation}$$

( $\sigma^*$  removes the last block entry and  $\sigma$  removes the first block entry)

Once  $C$  and  $A$  are known, the system of equations

$$y_d(t) = CA^t x_d(1) + \sum_{\tau=1}^{t-1} CA^{t-1-\tau} B u_d(\tau) + D \delta(t+1), \text{ for } t = 1, \dots, \ell_{\max} + 1$$

is **linear in  $D, B, x_d(1)$**  (can be solved using Kronecker products)



$$w_d \mapsto \mathcal{O}_{\ell_{\max}+1}(A, C)$$

The columns of  $\mathcal{O}_{\ell_{\max}+1}(A, C)$  are  $n$  linearly indep. free responses of  $\mathcal{B}$

Under the conditions of FL, **such resp. can be computed from data**

$$\begin{bmatrix} \mathcal{H}_t(u_d) \\ \mathcal{H}_t(y_d) \end{bmatrix} G = \begin{bmatrix} 0 \\ Y_0 \end{bmatrix} \quad \begin{array}{l} \leftarrow \text{zero inputs} \\ \leftarrow \text{free responses} \end{array}$$

in order to obtain lin. indep. free responses,  $G$  should be maximal rank

Once we have a maximal rank matrix of free responses  $Y_0$

$$Y_0 = \mathcal{O}_{\ell_{\max}+1}(A, C) \underbrace{\begin{bmatrix} x_{\text{ini},1} & \cdots & x_{\text{ini},j} \end{bmatrix}}_{X_{\text{ini}}} \quad \begin{array}{l} \text{rank revealing} \\ \text{factorization} \end{array}$$

$\rightsquigarrow \mathcal{O}_{\ell_{\max}+1}(A, C)$  and  $X_{\text{ini}}$ , the factorization fixes the state space basis

$$w_d \mapsto (x_d(1), \dots, x_d(n_{\max} + m + 1))$$

If the free responses are **sequential**, i.e., if  $Y_0$  is block-Hankel, then  $X_{\text{ini}}$  is a state sequence of  $\mathcal{B}$

Computation of **sequential free responses** is achieved as follows

$$\left[ \begin{array}{c} U_p \\ Y_p \\ U_f \end{array} \right] G = \left[ \begin{array}{c} U_p \\ Y_p \\ 0 \end{array} \right] \left. \begin{array}{l} \} \text{ sequential ini. conditions} \\ \leftarrow \text{ zero inputs} \end{array} \right\} \quad (4)$$

$$Y_f \quad G = Y_0$$

**Note:** now we use the splitting of the data into “past” and “future”

$$Y_0 = \mathcal{O}_{\ell_{\max}+1}(A, C) \begin{bmatrix} x_d(1) & \cdots & x_d(n_{\max} + m + 1) \end{bmatrix} \quad \begin{array}{l} \text{rank revealing} \\ \text{factorization} \end{array}$$

# Refinements

- Solve (4) efficiently **exploiting the Hankel structure**
- Iteratively compute pieces of  $Y_0$ 
  - $\rightsquigarrow$  **iterative algorithm**
    - requires smaller persistency of excitation of  $u_d$
    - could be more efficient

(Solve a few smaller systems of eqns instead of a single bigger one)

## MOESP type algorithms

**Orth. projection** of the rows of  $\mathcal{H}_{n_{\max}}(y_d)$  on  $\left(\text{rowspan}(\mathcal{H}_{n_{\max}}(u_d))\right)^\perp$

$$Y_0 := \mathcal{H}_{n_{\max}}(y_d) \Pi_{u_d}^\perp$$

where

$$\Pi_{u_d}^\perp := \left( I - \mathcal{H}_{n_{\max}}^\top(u_d) (\mathcal{H}_{n_{\max}}(u_d) \mathcal{H}_{n_{\max}}^\top(u_d))^{-1} \mathcal{H}_{n_{\max}}(u_d) \right)$$

Observe that  $\Pi_{u_d}^\perp$  is maximal rank and

$$\begin{bmatrix} \mathcal{H}_{n_{\max}}(u_d) \\ \mathcal{H}_{n_{\max}}(y_d) \end{bmatrix} \Pi_{u_d}^\perp = \begin{bmatrix} 0 \\ Y_0 \end{bmatrix}$$

$\Rightarrow$  **the orthogonal projection computes free responses**

## Comments

- $T - n_{\max} + 1$  free responses are computed via the orth. proj. while  $n_{\max}$  such responses suffice for the purpose of exact identification
- The orth. proj. is a **geometric operation**, whose system theoretic meaning is not revealed
- The **condition for  $\text{rank}(Y_0) = n$** , given in the MOESP literature,

$$\text{rank} \left( \begin{bmatrix} X_{\text{ini}} \\ \mathcal{H}_{n_{\max}}(u_d) \end{bmatrix} \right) = n + n_{\max}m$$

is **not verifiable from the data  $(u_d, y_d)$**   $\implies$  can not be checked whether the computation gives  $\mathcal{O}(A, C)$ , *cf.*, p.e. condition of FL

## N4SID-type algorithms

Consider the splitting of the data into “past” and “future”

$$\mathcal{H}_{2n_{\max}}(u_d) =: \begin{bmatrix} U_p \\ U_f \end{bmatrix}, \quad \mathcal{H}_{2n_{\max}}(y_d) =: \begin{bmatrix} Y_p \\ Y_f \end{bmatrix}$$

with  $\text{row dim}(U_p) = \text{row dim}(U_f) = \text{row dim}(Y_p) = \text{row dim}(Y_f) = n_{\max}$  and let

$$W_p := \begin{bmatrix} U_p \\ Y_p \end{bmatrix}$$

The key step of the N4SID algorithms is the **oblique projection** of the rows of  $Y_f$  along  $\text{row span}(U_f)$  onto  $\text{row span}(W_p)$

$$Y_0 := Y_f / U_f W_p := Y_f \underbrace{\begin{bmatrix} W_p^\top & U_f^\top \end{bmatrix} \begin{bmatrix} W_p W_p^\top & W_p U_f^\top \\ U_f W_p^\top & U_f U_f^\top \end{bmatrix}^+ \begin{bmatrix} W_p \\ 0 \end{bmatrix}}_{\Pi_{\text{obl}}}$$

## N4SID-type algorithms

Observe that

$$\begin{bmatrix} W_p \\ U_f \\ Y_f \end{bmatrix} \Pi_{\text{obl}} = \begin{bmatrix} W_p \\ 0 \\ Y_0 \end{bmatrix}$$

(in fact  $\Pi_{\text{obl}}$  is the least-norm, least-squares solution)

$\Rightarrow$  the oblique projection computes sequential free responses

## Comments

- $T - 2n_{\max} + 1$  sequential free responses are computed via the oblique projection while  $n_{\max} + m + 1$  such responses suffice for exact ident.
- The oblique proj. is a **geometric operation**, whose system theoretic meaning is not revealed
- The **conditions for  $\text{rank}(Y_0) = n$** , given in the N4SID literature,
  1.  $u_d$  persistently exciting of order  $2n_{\max}$  and
  2.  $\text{rowspan}(X_{\text{ini}}) \cap \text{rowspan}(U_f) = \{0\}$are **not verifiable from the data  $(u_d, y_d)$**



- **System theoretic interpretation** of the orth. and oblique proj.
- MOESP and N4SID alg. are computationally inefficient; more than what is necessary for exact ident. is computed  $\rightsquigarrow$  **cheaper algs**
- The FL gives **conditions for identifiability**, verifiable from the data
- We clarified the **role of the splitting**: the “past” assigns the initial conditions and in the “future” a desired response is computed  
 $\implies$  “past” should be chosen at least  $\ell$  samples long; the length of “future” is free as long as the p.e. condition is satisfied

# References

1. J. C. Willems.  
From time series to linear system—Part II. Exact modelling.  
*Automatica*, 22(6):675–694, 1986.
2. J. C. Willems, P. Rapisarda, I. Markovsky, and B. De Moor.  
A note on persistency of excitation.  
*Systems & Control Letters*, 54(4):325–329, 2005.
3. I. Markovsky, J. C. Willems, P. Rapisarda, and B. De Moor.  
Algorithms for deterministic balanced subspace identification.  
*Automatica*, 41(5):755–766, 2005.
4. I. Markovsky, J. C. Willems, S. Van Huffel, and B. De Moor.  
Exact and Approximate Modeling of Linear Systems  
SIAM, 2006

# Software

A MATLAB toolbox for exact SYSID is available from:

`ftp.esat.kuleuven.be/pub/SISTA/markovsky/  
abstracts/05-122.html`

In exercise 2 you will use the algorithms

- $w_d \mapsto R(\xi) \quad (w2r)$  and
- $w_d \mapsto (x_d(1), \dots, x_d(n_{\max} + m + 1)) \mapsto (A, B, C, D) \quad (uy2x2ss)$

in order to find the MPUM for given trajectory of an LTI system.