Conversions Between S, Z, Y, h, ABCD, and T Parameters which are Valid for Complex Source and Load Impedances

Dean A. Frickey, Member, IEEE

Abstract—This paper provides tables which contain the conversion between the various common two-port parameters, Z, Y, h, ABCD, S, and T. The conversion are valid for complex normalizing impedances. An example is provided which verifies the conversions to and from S parameters.

I. INTRODUCTION

OST microwave textbooks these days seem to provide a table of the conversion between the various 2-port parameters. These 2-port parameters often include Z(impedance), Y (admittance), h (hybrid), ABCD (chain), S (scattering), and T (chain scattering or chain transfer). While the scattering parameters have been shown [1] to be valid for complex normalizing impedances (with positive real parts), the tables in [2]-[15] are not valid for complex source and load impedances. Often, the tables only provide conversions for the cases where port 1 and port 2 normalizing impedances are equal, i.e., $Z_{01} = Z_{02} = Z_0$. Some have results in which Z_{01} and Z_{02} are normalized to 1. Others provide equations for port 1 and port 2 impedances Z_{01} and Z_{02} to be unique. However, in all of these cases, the results are not valid when the impedances, Z_{01} and Z_{02} , or just Z_0 , are complex.

Of the two-port parameters mentioned, only the S and T parameters are dependent upon the source and load impedances. In this paper, the derivations of the conversions from the S and T parameters to the other 2-port parameters includes complex source and load impedances. The equations developed in this work are valid with port 1 and port 2 normalizing impedances complex and unique. When the normalizing impedances are real, the results simplify to those shown in other references. To make the list complete, the conversions between the Z, Y, h, and ABCD parameters as well as between S and T parameters are included.

II. DERIVATION

Two-port parameters are defined for a general 2-port network as shown in Fig. 1. Using the voltages and currents defined in this figure, the various 2-port parameters are written

Manuscript received December 2, 1992; revised April 13, 1993. The author is with EG&G Idaho, Idaho Falls, ID 83415. IEEE Log Number 9214525.



Fig. 1. A general two-port network with voltages and currents defined.

Z parameters

$$V_1 = Z_{11} \cdot I_1 + Z_{12} \cdot I_2 \tag{1a}$$

$$V_2 = Z_{21} \cdot I_1 + Z_{22} \cdot I_2, \tag{1b}$$

Y parameters

$$I_1 = Y_{11} \cdot V_1 + Y_{12} \cdot V_2 \tag{2a}$$

$$I_2 = Y_{21} \cdot V_1 + Y_{22} \cdot V_2, \tag{2b}$$

h parameters

$$V_1 = h_{11} \cdot I_1 + h_{12} \cdot V_2 \tag{3a}$$

$$I_2 = h_{21} \cdot I_1 + h_{22} \cdot V_2, \tag{3b}$$

ABCD parameters

$$V_1 = A \cdot V_2 - B \cdot I_2 \tag{4a}$$

$$I_1 = C \cdot V_2 - D \cdot I_2, \tag{4b}$$

S parameters

$$b_1 = S_{11} \cdot a_1 + S_{12} \cdot a_2 \tag{5a}$$

$$b_2 = S_{21} \cdot a_1 + S_{22} \cdot a_2, \tag{5b}$$

0018-9480/94\$04.00 © 1994 IEEE

TABLE I

Equations for the Conversion Between S Parameter	RS AND Z, Y, h , AND $ABCD$ PARAMETERS WITH	TH A SOURCE IMPEDANCE Z_{01} AND LOAD IMPEDANCE Z_{02}	9

$S_{11} = \frac{(Z_{11} - Z_{01}^*)(Z_{22} + Z_{02}) - Z_{12} Z_{21}}{(Z_{11} + Z_{01})(Z_{22} + Z_{02}) - Z_{12} Z_{21}}$	$Z_{11} = \frac{(Z_{01}^* + S_{11} Z_{01})(1 - S_{22}) + S_{12} S_{21} Z_{01}}{(1 - S_{11})(1 - S_{22}) - S_{12} S_{21}}$
$S_{12} = \frac{2Z_{12}(R_{01}R_{02})^{1/2}}{(Z_{11} + Z_{01})(Z_{22} + Z_{02}) - Z_{12}Z_{21}}$	$Z_{12} = \frac{2S_{12}(R_{01}R_{02})^{1/2}}{(1-S_{11})(1-S_{22}) - S_{12}S_{21}}$
$S_{21} = \frac{2Z_{21}(R_{01}R_{02})^{1/2}}{(Z_{11} + Z_{01})(Z_{22} + Z_{02}) - Z_{12}Z_{21}}$	$Z_{21} = \frac{2S_{21}(R_{01}R_{02})^{1/2}}{(1-S_{11})(1-S_{22})-S_{12}S_{21}}$
$S_{22} = \frac{(Z_{11} + Z_{01})(Z_{22} - Z_{02}^*) - Z_{12} Z_{21}}{(Z_{11} + Z_{01})(Z_{22} + Z_{02}) - Z_{12} Z_{21}}$	$Z_{22} = \frac{(1 - S_{11})(Z_{02}^* + S_{22} Z_{02}) + S_{12} S_{21} Z_{02}}{(1 - S_{11})(1 - S_{22}) - S_{12} S_{21}}$
$S_{11} = \frac{(1 - Y_{11} Z_{01}^*)(1 + Y_{22} Z_{02}) + Y_{12} Y_{21} Z_{01}^* Z_{02}}{(1 + Y_{11} Z_{01})(1 + Y_{22} Z_{02}) - Y_{12} Y_{21} Z_{01} Z_{02}}$	$Y_{11} = \frac{(1 - S_{11})(Z_{02}^* + S_{22} Z_{02}) + S_{12} S_{21} Z_{02}}{(Z_{01}^* + S_{11} Z_{01})(Z_{02}^* + S_{22} Z_{02}) - S_{12} S_{21} Z_{01} Z_{02}}$
$S_{12} = \frac{-2Y_{12}(R_{01}R_{02})^{1/2}}{(1+Y_{11}Z_{01})(1+Y_{22}Z_{02}) - Y_{12}Y_{21}Z_{01}Z_{02}}$	$Y_{12} = \frac{-2S_{12}(R_{01}R_{02})^{1/2}}{(Z_{01}^2 + S_{11}Z_{01})(Z_{02}^2 + S_{22}Z_{02}) - S_{12}S_{21}Z_{01}Z_{02}}$
$S_{21} = \frac{-2Y_{21}(R_{01}R_{02})^{1/2}}{(1+Y_{11}Z_{01})(1+Y_{22}Z_{02}) - Y_{12}Y_{21}Z_{01}Z_{02}}$	$Y_{21} = \frac{-2S_{21}(R_{01}R_{02})^{1/2}}{(Z_{01}^* + S_{11}Z_{01})(Z_{02}^* + S_{22}Z_{02}) - S_{12}S_{21}Z_{01}Z_{02}}$
$S_{22} = \frac{(1+Y_{11}Z_{01})(1-Y_{22}Z_{02}^*)+Y_{12}Y_{21}Z_{01}Z_{02}^*}{(1+Y_{11}Z_{01})(1+Y_{22}Z_{02})-Y_{12}Y_{21}Z_{01}Z_{02}^*}$	$Y_{22} = \frac{(Z_{01}^{2} + S_{11} Z_{01})(1 - S_{22}) + S_{12} S_{21} Z_{01}}{(Z_{01}^{2} + S_{11} Z_{01})(Z_{02}^{2} + S_{22} Z_{02}) - S_{12} S_{21} Z_{01}} Z_{02}}$
	(**************************************
$(h_{11}-Z_{01}^*)(1+h_{22}Z_{02})-h_{12}h_{21}Z_{02}$	(7* ±5:.7c.)(7* ±5:.7c.) -5:.57c.7c.
$S_{11} = \frac{(h_{11} - Z_{01}^{*})(1 + h_{22}Z_{02}) - h_{12}h_{21}Z_{02}}{(Z_{01} + h_{11})(1 + h_{22}Z_{02}) - h_{12}h_{21}Z_{12}Z_{02}}$	$h_{11} = \frac{(Z_{01}^* + S_{11} Z_{01})(Z_{02}^* + S_{22} Z_{02}) - S_{12} S_{21} Z_{01}}{(1 - S_{11})(Z_{02}^* + S_{22} Z_{02}) + S_{12} S_{21} Z_{02}}$
$S_{12} = \frac{2h_{12}(R_{01}R_{02})^{1/2}}{(Z_{01} + h_{11})(1 + h_{22}Z_{02}) - h_{12}h_{21}Z_{02}}$	$h_{12} = \frac{2S_{12}(R_{01}R_{02})^{1/2}}{(1-S_{11})(Z_{02}^* + S_{22}Z_{02}) + S_{12}S_{21}Z_{02}}$
$S_{21} = \frac{-2h_{21}(R_{01}R_{02})^{1/2}}{(Z_{01} + h_{11})(1 + h_{22}Z_{02}) - h_{12}h_{21}Z_{02}}$	$h_{21} = \frac{-2S_{21}(R_{01}R_{02})^{1/2}}{(1-S_{11})(Z_{02}^* + S_{22}Z_{02}) + S_{12}S_{21}Z_{02}}$
$S_{22} = \frac{(Z_{01} + h_{11})(1 - h_{22}Z_{02}^*) + h_{12}h_{21}Z_{02}^*}{(Z_{01} + h_{11})(1 + h_{22}Z_{02}) - h_{12}h_{21}Z_{02}}$	$h_{22} = \frac{(1 - S_{11})(1 - S_{22}) - S_{12}S_{21}}{(1 - S_{11})(Z_{02}^* + S_{22}Z_{02}) + S_{12}S_{21}Z_{02}}$
$S_{11} = \frac{AZ_{02} + B - CZ_{01}^* Z_{02} - DZ_{01}^*}{AZ_{02} + B + CZ_{01}^* Z_{02} + DZ_{01}^*}$	$A = \frac{(Z_{01}^* + S_{11}Z_{01})(1 - S_{22}) + S_{12}S_{21}Z_{01}}{2S_{21}(R_{01}R_{02})^{1/2}}$
$S_{12} = \frac{2(AD - BC)(R_{01}R_{02})^{1/2}}{AZ_{02} + B + CZ_{01}Z_{02} + DZ_{01}}$	$B = \frac{(Z_{01}^* + S_{11}Z_{01})(Z_{02}^* + S_{22}Z_{02}) - S_{12}S_{21}Z_{01}Z_{02}}{2S_{21}(R_{01}R_{02})^{1/2}}$
$S_{21} = \frac{\frac{AZ_{02} + B + CZ_{01}Z_{02} + DZ_{01}}{2(R_{01}R_{02})^{1/2}}}{\frac{Z_{02} + B + CZ_{01}Z_{02} + DZ_{01}}{2(R_{01}R_{02})^{1/2}}}$	$C = \frac{(1-S_{11})(1-S_{22})-S_{12}S_{21}}{2S_{21}(R_{01}R_{02})^{1/2}}$
	$-2S_{21}(R_{01}R_{02})^{1/2}$
$S_{22} = \frac{-AZ_{02}^* + B - CZ_{01}Z_{02}^* + DZ_{01}}{AZ_{02} + B + CZ_{01}Z_{02} + DZ_{01}}$	$D = \frac{(1 - S_{11})(Z_{02}^* + S_{22}Z_{02}) + S_{12}S_{21}Z_{02}}{2S_{21}(R_{01}R_{02})^{1/2}}$



Fig. 2. A general two port network with a's and b's defined.

T parameters¹

$$a_1 = T_{11} \cdot b_2 + T_{12} \cdot a_2 \tag{6a}$$

$$b_1 = T_{21} \cdot b_2 + T_{22} \cdot a_2 \tag{6b}$$

where the a's and b's are shown in Fig. 2 and defined below.

$$a_j = \left[\frac{Z_{0j} + Z_{0j}^*}{2}\right]^{1/2} \cdot I_{ji}$$
 (7a)

$$b_j = \left[\frac{Z_{0j} + Z_{0j}^*}{2} \right]^{1/2} \cdot I_{jr} \tag{7b}$$

 1 Some authors, (e.g. Rizzi [16]) define the T parameters as $b_1=T_{11}\cdot a_2+T_{12}\cdot b_2,$ and $a_1=T_{21}\cdot a_2+T_{22}b_2.$ In this case, the parameters can just be switched from what is derived in this paper. T_{11} and T_{22} are switched, T_{12} and T_{21} are switched.

where * indicates complex conjugate and Z_{0j} is the normalizing impedance for the jth port. For two-port networks, Z_{01} and Z_{02} are the source and load impedances of the system in which the S parameters of the two-port are measured or calculated. I_{ji} and I_{jr} are the incident and reflected currents for the jth port. Knowing that,

$$I_j = I_{ji} - I_{jr} \tag{8}$$

we can solve (7a) and (7b) for I_{ji} and I_{jr} and substitute them into (8) to get,

$$I_j = \left[\frac{2}{Z_{0j} + Z_{0j}^*}\right]^{1/2} \cdot (a_j - b_j). \tag{9}$$

Knowing also that,

$$V_j = V_{ji} + V_{jr} (10)$$

where V_{ji} and V_{jr} are the incident and reflected voltage at the jth port, we can substitute the expressions for I_{ji} and I_{jr} along with

$$V_{ji} = I_{ji} \cdot Z_{0j}^* \qquad V_{jr} = I_{jr} \cdot Z_{0j}$$

into (10) to get,

$$V_{j} = \left[\frac{2}{Z_{0j} + Z_{0j}^{*}}\right]^{1/2} \cdot (a_{j} \cdot Z_{0j}^{*} + b_{j} \cdot Z_{0j}).$$
 (11)

TABLE II
EQUATIONS FOR THE CONVERSION BETWEEN T PARAMETERS and Z, Y, h, and ABCD Parameters with a Source Impedance Z_{01} and Load Impedance Z_{02}

EQUATIONS FOR THE CONVERSION BETWEEN 1 PARAMETERS AND 2, 1, n, AND 2	$ABCD$ Parameters with a Source impedance z_{01} and Luad impedance z_{02}
$T_{11} = \frac{(Z_{11} + Z_{01})(Z_{22} + Z_{02}) - Z_{12}Z_{21}}{2Z_{21}(R_{01}R_{02})^{1/2}}$	$Z_{11} = \frac{Z_{01}^*(T_{11} + T_{12}) + Z_{01}(T_{21} + T_{22})}{T_{11} + T_{12} - T_{21} - T_{22}}$
$T_{12} = \frac{(Z_{11} + Z_{01})(Z_{02}^* - Z_{22}) + Z_{12}Z_{21}}{2Z_{21}(R_{01}R_{02})^{1/2}}$	$Z_{12} = \frac{2(R_{01}R_{02})^{1/2}(T_{11}T_{22} - T_{12}T_{21})}{T_{11} + T_{12} - T_{21} - T_{22}}$
$T_{21} = \frac{(Z_{11} - Z_{01}^*)(Z_{22} + Z_{02}) - Z_{12} Z_{21}}{2Z_{21}(R_{01}R_{02})^{1/2}}$	$Z_{21} = \frac{2(R_{01}R_{02})^{1/2}}{T_{11} + T_{12} - T_{21} - T_{22}}$
$T_{22} = \frac{(Z_{01}^{\bullet} - Z_{11})(Z_{22} - Z_{02}^{\bullet}) + Z_{12}Z_{21}}{2Z_{21}(R_{01}R_{02})^{1/2}}$	$Z_{22} = \frac{Z_{02}^*(T_{11} - T_{21}) - Z_{02}(T_{12} - T_{22})}{T_{11} + T_{12} - T_{21} - T_{22}}$
$T_{11} = \frac{(-1 - Y_{11} Z_{01})(1 + Y_{22} Z_{02}) + Y_{12} Y_{21} Z_{01} Z_{02}}{2Y_{21} (R_{01} R_{02})^{1/2}}$	$Y_{11} = \frac{Z_{02}^*(T_{11} - T_{21}) - Z_{02}(T_{12} - T_{22})}{T_{11}Z_{01}^*Z_{02}^* - T_{12}Z_{01}^*Z_{02} + T_{21}Z_{01}Z_{02} - T_{22}Z_{01}Z_{02}}$
$T_{12} = \frac{(1 + Y_{11} Z_{01})(1 - Y_{22} Z_{02}^*) + Y_{12} Y_{21} Z_{01} Z_{02}^*}{2Y_{21} (R_{01} R_{02})^{1/2}}$	$Y_{12} = \frac{-2(R_{01}R_{02})^{1/2}(T_{11}T_{22} - T_{12}T_{21})}{T_{11}Z_{01}^*Z_{02}^* - T_{12}Z_{01}^*Z_{02} + T_{21}Z_{01}Z_{02}^* - T_{22}Z_{01}Z_{02}}$
$T_{21} = \frac{(Y_{11}Z_{01}^{\bullet}-1)(1+Y_{22}Z_{02})-Y_{12}Y_{21}Z_{01}^{\bullet}Z_{02}}{2Y_{21}(R_{01}R_{02})^{1/2}}$	$Y_{21} = \frac{-2(R_{01}R_{02})^{1/2}}{T_{11}Z_{01}^*Z_{02}^* - T_{12}Z_{01}^*Z_{02} + T_{21}Z_{01}Z_{02}^* - T_{22}Z_{01}Z_{02}}$
$T_{22} = \frac{(1 - Y_{11} Z_{01}^*) (1 - Y_{22} Z_{02}^*) - Y_{12} Y_{21} Z_{01}^* Z_{02}^*}{2 Y_{21} (R_{01} R_{02})^{1/2}}$	$Y_{22} = \frac{Z_{01}^{*}(T_{11} + T_{12}) + Z_{01}(T_{21} + T_{22})}{T_{11}Z_{01}^{*}Z_{02}^{*} - T_{12}Z_{01}^{*}Z_{02} + T_{21}Z_{01}Z_{02} - T_{22}Z_{01}Z_{02}}$
$T_{11} = \frac{(-h_{11} - Z_{01})(1 + h_{22}Z_{02}) + h_{12}h_{21}Z_{02}}{2h_{21}(R_{01}R_{02})^{1/2}}$	$h_{11} = \frac{Z_{02}^{\star}(T_{11}Z_{01}^{\star} + T_{21}Z_{01}) - Z_{02}(T_{12}Z_{01}^{\star} + T_{22}Z_{01})}{Z_{02}^{\star}(T_{11} - T_{21}) - Z_{02}(T_{12} + T_{22})}$
$T_{12} = \frac{(h_{11} + Z_{\underline{0}1})(1 - h_{22}Z_{\underline{0}2}^{\star}) + h_{12}h_{21}Z_{\underline{0}2}^{\star}}{2h_{21}(R_{01}R_{02})^{1/2}}$	$h_{12} = \frac{2(R_{01}R_{02})^{1/2}(T_{11}T_{22} - T_{12}T_{21})}{Z_{02}^*(T_{11} - T_{21}) - Z_{02}(T_{12} + T_{22})}$
$T_{21} = \frac{(Z_{01}^{\bullet} - h_{11})(1 + h_{22}Z_{02}) + h_{12}h_{21}Z_{02}}{2h_{21}(R_{01}R_{02})^{1/2}}$	$h_{21} = \frac{-2(R_{01}R_{02})^{1/2}}{Z_{02}^*(T_{11} - T_{21}) - Z_{02}(T_{12} + T_{22})}$
$T_{22} = \frac{(h_{11} - Z_{01}^*)(1 - h_{22} Z_{02}^*) + h_{12} h_{21} Z_{02}^*}{2h_{21} (R_{01} R_{02})^{1/2}}$	$h_{22} = \frac{T_{11} + T_{12} - T_{21} - T_{22}}{Z_{02}^* (T_{11} - T_{21}) - Z_{02} (T_{12} + T_{22})}$
$T_{11} = \frac{AZ_{02} + B + CZ_{01}Z_{02} + DZ_{01}}{2(R_{01}R_{02})^{1/2}}$	$A = \frac{Z_{01}^{\star}(T_{11} + T_{12}) + Z_{01}(T_{21} + T_{22})}{2(R_{01}R_{02})^{1/2}}$
$T_{12} = \frac{AZ_{02}^* - B + CZ_{01}Z_{02}^* - DZ_{01}}{2(R_{01}R_{02})^{1/2}}$	$B = \frac{Z_{02}^{\star}(T_{11}Z_{01}^{\star} + T_{21}Z_{01}) - Z_{02}(T_{12}Z_{01}^{\star} + T_{22}Z_{01})}{2(R_{01}R_{02})^{1/2}}$
$T_{21} = \frac{AZ_{02} + B - CZ_{01}^* Z_{02} - DZ_{01}^*}{2(R_{01}R_{02})^{1/2}}$	$C = \frac{T_{11} + T_{12} - T_{21} - T_{22}}{2(R_{01}R_{02})^{1/2}}$
$T_{22} = \frac{AZ_{02}^{\bullet} - B - CZ_{01}^{\bullet} Z_{02}^{\bullet} + DZ_{01}^{\bullet}}{2(R_{01}R_{02})^{1/2}}$	$D = \frac{Z_{02}^{*}(T_{11} - T_{21}) - Z_{02}(T_{12} - T_{22})}{2(R_{01}R_{02})^{1/2}}$

Solving (9) and (11) for a_j and b_j gives

$$a_j = \frac{V_j + Z_{0j}I_j}{[2(Z_{0j} + Z_{0j}^*)]^{1/2}}$$
 (12)

$$b_j = \frac{V_j - Z_{0j}^* I_j}{[2(Z_{0j} + Z_{0j}^*)]^{1/2}}. (13)$$

Equations (12) and (13) are (3) and (4) in [1] and served as the starting point.

The notation, $S \leftrightarrow Z$, indicates the conversion from S parameters to Z parameters and Z parameters to S parameters. Since S and T parameters are defined in terms of a's and b's, they will contain the source and load normalizing impedances Z_{01} and Z_{02} . The other 2-port parameters are defined independent of the source and load impedances.

To derive the conversions, $S \leftrightarrow Z$, $S \leftrightarrow Y$, $S \leftrightarrow h$, $S \leftrightarrow ABCD$, $T \leftrightarrow Z$, $T \leftrightarrow Y$, $T \leftrightarrow h$, and $T \leftrightarrow ABCD$, it is necessary to use (9), (11)–(13). For example, to derive the expressions for S parameters in terms of the Z parameters, first substitute (9) and (11) into (1a) and (1b) and solve for b_1

and b_2 to get in the form of (5a) and (5b). Likewise, to get the expressions for the Z parameters in terms of the S parameters, substitute (12) and (13) into (5a) and (5b) and solve for V_1 and V_2 to get in the form of (1a) and (1b).

Since Z, Y, h, and ABCD parameters do not require normalizing impedances, the conversions, $Z \leftrightarrow Y$, $Z \leftrightarrow h$, $Z \leftrightarrow ABCD$, $Y \leftrightarrow h$, $Y \leftrightarrow ABCD$, and $h \leftrightarrow ABCD$, as well as $S \leftrightarrow T$, are straight forward. These conversions are accomplished by rearranging one set of equations into the form of the other. These conversions appear in many of the references cited and are included here for completeness.

III. RESULTS

The results are given in the following tables. In these tables, Z_{01} and Z_{02} are the source and load impedances of the system to which the S and T parameters pertain. Complex conjugate is indicated by *, and R_{01} and R_{02} are the real parts of Z_{01} and Z_{02} .

Table I gives the conversions between S parameters and Z, Y, h, and ABCD parameters. Table II gives the conversions

TABLE III Equations for the Conversion Between S Parameters and Normalized $Z,\,Y,\,h,$ and ABCD Parameters with a Source Impedance Z_{01} and Load Impedance Z_{02}

	CE IMPEDANCE Z_{01} AND LOAD IMPEDANCE Z_{02}
$S_{11} = \frac{\left[Z_{11n} - \frac{Z_{01}^*}{Z_{01}} \right] (Z_{22n} + 1) - Z_{12n} Z_{21n}}{(Z_{11n} + 1)(Z_{22n} + 1) - Z_{12n} Z_{21n}}$	$Z_{11n} = \frac{\left[\frac{Z_{01}^*}{Z_{01}} + S_{11}\right](1 - S_{22}) + S_{12}S_{21}}{(1 - S_{11})(1 - S_{22}) - S_{12}S_{21}}$
$S_{12} = \frac{2Z_{12n} \left[\frac{R_{01}R_{02}}{Z_{01}Z_{02}} \right]^{1/2}}{(Z_{11n}+1)(Z_{22n}+1) - Z_{12n}Z_{21n}}$	$Z_{12n} = \frac{2S_{12} \left[\frac{R_{01} R_{02}}{Z_{01} Z_{02}} \right]^{1/2}}{(1 - S_{11})(1 - S_{22}) - S_{12} S_{21}}$
$S_{21} = \frac{2Z_{21n} \left[\frac{R_{01}R_{02}}{Z_{01}Z_{02}} \right]^{1/2}}{(Z_{11n}+1)(Z_{22n}+1) - Z_{12n}Z_{21n}}$	$Z_{21n} = \frac{2S_{21} \left[\frac{R_{01} R_{02}}{Z_{01} Z_{02}} \right]^{1/2}}{(1 - S_{11})(1 - S_{22}) - S_{12} S_{21}}$
$S_{22} = \frac{(Z_{11n}+1) \left[Z_{22n} - \frac{Z_{02}^*}{Z_{02}} \right] - Z_{12n} Z_{21n}}{(Z_{11n}+1)(Z_{22n}+1) - Z_{12n} Z_{21n}}$	$Z_{22n} = \frac{(1 - S_{11}) \left[\frac{Z_{02}^2}{Z_{02}} + S_{22} \right] + S_{12} S_{21}}{(1 - S_{11})(1 - S_{22}) - S_{12} S_{21}}$
$Z_{11n} = \frac{Z_{11}}{Z_{01}}$ $Z_{12n} = \frac{Z_{12}}{(Z_{01}Z_{02})^{1/2}}$	$Z_{21n} = \frac{Z_{21}}{(Z_{01}Z_{02})^{1/2}} \qquad \qquad Z_{22n} = \frac{Z_{22}}{Z_{02}}$
$S_{11} = \frac{\left[\frac{1-Y_{11n}\left[\frac{Z_{01}^{\delta}}{Z_{01}}\right]}{(1+Y_{11n})(1+Y_{22n})-Y_{12n}Y_{21n}}\left[\frac{Z_{01}^{\delta}}{Z_{01}}\right]}{(1+Y_{11n})(1+Y_{22n})-Y_{12n}Y_{21n}}\right]}$	$Y_{11n} = \frac{\frac{(1-S_{11})}{\frac{Z_{02}^{n}}{Z_{02}} + S_{22}} + S_{12}S_{21}}{\frac{Z_{01}^{n}}{Z_{01}} + S_{11}} \frac{\frac{Z_{02}^{n}}{Z_{02}} + S_{22}}{\frac{Z_{02}^{n}}{Z_{02}} + S_{22}} - S_{12}S_{21}}$
$S_{12} = \frac{-2Y_{12n} \left[\frac{R_{01}R_{02}}{Z_{01}Z_{02}} \right]^{1/2}}{(1+Y_{11n})(1+Y_{22n}) - Y_{12n}Y_{21n}}$	$Y_{12n} = \frac{-2S_{12} \left[\frac{R_{01}R_{02}}{Z_{01}}\right]^{1/2}}{\left[\frac{Z_{01}^*}{Z_{01}} + S_{11}\right] \left[\frac{Z_{02}^*}{Z_{02}} + S_{22}\right] - S_{12}S_{21}}$
$S_{21} = \frac{-2Y_{21n} \left[\frac{R_{01} R_{02}}{Z_{01} Z_{02}} \right]^{1/2}}{(1+Y_{11n})(1+Y_{22n}) - Y_{12n} Y_{21n}}$	$Y_{21n} = \frac{-2S_{21} \left[\frac{R_{01} R_{02}}{Z_{01}} \right]^{1/2}}{\left[\frac{Z_{01}^{*}}{Z_{01}} + S_{11} \right] \left[\frac{Z_{02}^{*}}{Z_{02}} + S_{22} - S_{12} S_{21}} \right]}$
$S_{22} = \frac{(1+Y_{11n}) \left[1-Y_{22n} \left[\frac{Z_{02}^{\bullet 2}}{Z_{02}}\right]\right] + Y_{12n}Y_{21n} \left[\frac{Z_{02}^{\bullet 2}}{Z_{02}}\right]}{(1+Y_{11n})(1+Y_{22n}) - Y_{12n}Y_{21n}}$	$Y_{22n} = \frac{\left[\frac{Z_{01}^{*}}{Z_{01}^{*}} + S_{11}\right](1 - S_{22}) + S_{12}S_{21}}{\left[\frac{Z_{01}^{*}}{Z_{01}^{*}} + S_{11}\right]\left[\frac{Z_{02}^{*}}{Z_{02}^{*}} + S_{22}\right] - S_{12}S_{21}}$
$Y_{11n} = Y_{11}Z_{01}$ $Y_{12n} = Y_{12}(Z_{01}Z_{02})^{1/2}$	$Y_{21n} = Y_{21}(Z_{01}Z_{02})^{1/2} Y_{22n} = Y_{22}Z_{02}$
$S_{11} = \frac{\left[\frac{h_{11n} - \frac{Z_{01}^*}{Z_{01}}}{(1 + h_{11n})(1 + h_{22n}) - h_{12n}h_{21n}}}{(1 + h_{11n})(1 + h_{22n}) - h_{12n}h_{21n}}\right]}$	$h_{11n} = \frac{\left[\frac{z_{01}^*}{z_{01}} + S_{11}\right] \left[\frac{z_{02}^*}{z_{02}} + S_{22}\right] - S_{12}S_{21}}{(1 - S_{11}) \left[\frac{z_{02}^*}{z_{02}} + S_{22}\right] + S_{12}S_{21}}$
$S_{12} = \frac{2h_{12n} \left[\frac{R_{01}R_{02}}{Z_{01}Z_{02}} \right]^{1/2}}{(1+h_{11n})(1+h_{22n}) - h_{12n}h_{21n}}$	$h_{12n} = \frac{2S_{12} \left[\frac{R_{01}R_{02}}{Z_{01}Z_{02}} \right]^{1/2}}{(1-S_{11}) \left[\frac{Z_{02}^2}{Z_{02}} + S_{22} \right] + S_{12}S_{21}}$
$S_{21} = \frac{-2h_{21n} \left[\frac{R_{01}R_{02}}{Z_{01}Z_{02}} \right]^{1/2}}{(1+h_{11n})(1+h_{22n}) - h_{12n}h_{21n}}$	$h_{21n} = \frac{-2S_{21} \left[\frac{R_{01}R_{02}}{Z_{01}Z_{02}} \right]^{1/2}}{(1 - S_{11}) \left[\frac{Z_{02}}{Z_{02}} + S_{22} \right] + S_{12}S_{21}}$
$S_{22} = \frac{(1+h_{11n})\left[1-h_{22n}\left[\frac{Z_{02}^{\bullet}}{Z_{02}}\right]\right] + h_{12n}h_{21n}\left[\frac{Z_{02}^{\bullet}}{Z_{02}}\right]}{(1+h_{11n})(1+h_{22n}) - h_{12n}h_{21n}}$	$h_{22n} = \frac{(1-S_{11})(1-S_{22}) - S_{12}S_{21}}{(1-S_{11})\left[\frac{Z_{12}}{Z_{02}} + S_{22}\right] + S_{12}S_{21}}$
$h_{11n} = \frac{h_{11}}{Z_{01}}$ $h_{12n} = h_{12} \left[\frac{Z_{02}}{Z_{01}} \right]^{1/2}$	$h_{21n} = h_{21} \left[\frac{Z_{02}}{Z_{01}} \right]^{1/2}$ $h_{22n} = h_{22} Z_{02}$
$S_{11} = \frac{A_n + B_n - C_n \left[\frac{Z_{01}^*}{Z_{01}} \right] - D_n \left[\frac{Z_{01}^*}{Z_{01}} \right]}{A_n + B_n + C_n + D_n}$	$A_n = \frac{\left[\frac{Z_{01}^* + S_{11}}{Z_{01}} + S_{11}\right] (1 - S_{22}) + S_{12}S_{21}}{2S_{21}\left[\frac{R_{01}R_{02}}{Z_{01}Z_{02}}\right]}$
$S_{12} = \frac{2\left[\frac{R_{01}R_{02}}{Z_{01}Z_{02}}\right](A_nD_n - B_nC_n)}{A_n + B_n + C_n + D_n} = \frac{2(AD - BC)}{A_n + B_n + C_n + D_n}$	$B_n = \frac{\begin{bmatrix} \frac{z_{01}^n}{z_{01}} + S_{11} \end{bmatrix} \begin{bmatrix} \frac{z_{02}^n}{z_{02}} + S_{22} \end{bmatrix} - S_{12}S_{21}}{2S_{21} \begin{bmatrix} \frac{R_{01}R_{02}}{z_{01}z_{02}} \end{bmatrix}}$
$S_{21} = \frac{2}{A_n + B_n + C_n + D_n}$	$C_n = \frac{(1 - S_{11})(1 - S_{22}) - S_{12}S_{21}}{2S_{21} \left[\frac{R_{01}R_{02}}{Z_{01}R_{02}}\right]}$
$S_{22} = rac{-A_n \left[rac{Z_{02}^*}{Z_{02}} \right] + B_n - C_n \left[rac{Z_{02}^*}{Z_{02}} \right] + D_n}{A_n + B_n + C_n + D_n}$	
$S_{22} = \frac{1}{A_n + B_n + C_n + D_n}$	$D_n = \frac{(1 - S_{11}) \left[\frac{Z_{02}^n + S_{22}}{Z_{02}} \right] + S_{12} S_{21}}{2S_{21} \left[\frac{R_0 1 R_{02}}{Z_{01} Z_{02}} \right]}$

TABLE IV Equations for the Conversion Between T Parameters and Normalized $Z,\,Y,\,h,$ and ABCD Parameters with a Source Impedance Z_{01} and Load Impedance Z_{02}

$T_{11} = \frac{(Z_{11n} + 1)(Z_{22n} + 1) - Z_{12n} Z_{21n}}{2Z_{21n} \left[\frac{R_{01}R_{02}}{Z_{01}Z_{02}}\right]^{1/2}}$	$Z_{11n} = \frac{\left[\frac{z_{01}^{n}}{z_{01}}\right]^{(T_{11} + T_{12}) + (T_{21} + T_{22})}}{T_{11} + T_{12} - T_{21} - T_{22}}$
$T_{12} = \frac{(Z_{11n} + 1) \left[\frac{Z_{02}^*}{Z_{02}} - Z_{22n} \right] + Z_{12n} Z_{21n}}{2Z_{21n} \left[\frac{Z_{01}}{Z_{01}} \frac{Z_{02}}{Z_{02}} \right]^{1/2}}$	$Z_{12n} = \frac{2\left[\frac{R_{01}R_{02}}{Z_{01}Z_{02}}\right]^{1/2} (T_{11}T_{22} - T_{12}T_{21})}{T_{11} + T_{12} - T_{21} - T_{22}}$
$T_{21} = \frac{\left[Z_{11n} - \frac{z_{01}^*}{Z_{01}}\right] (Z_{22n} + 1) - Z_{12n} Z_{21n}}{2Z_{21n} \left[\frac{R_{01} R_{02}}{Z_{01} Z_{02}}\right]^{1/2}}$	$Z_{21n} = \frac{2\left[\frac{R_{01}R_{02}}{Z_{01}Z_{02}}\right]^{1/2}}{T_{11} + T_{12} - T_{21} - T_{22}}$
$T_{22} = \frac{\left[\frac{Z_{01}^*}{Z_{01}} - Z_{11n}\right] \left[Z_{22n} - \frac{Z_{02}^*}{Z_{02}}\right] + Z_{12n}Z_{21n}}{2Z_{21n} \left[\frac{R_{01}R_{02}}{Z_{01}Z_{02}}\right]^{1/2}}$	$Z_{22n} = \frac{\left[\frac{Z_{02}^*}{Z_{02}}\right](T_{11} - T_{21}) - (T_{12} - T_{22})}{T_{11} + T_{12} - T_{21} - T_{22}}$
$Z_{11n} = \frac{Z_{11}}{Z_{01}}$ $Z_{12n} = \frac{Z_{12}}{(Z_{01}Z_{02})^{1/2}}$	$Z_{21n} = \frac{Z_{21}}{(Z_{01}Z_{02})^{1/2}} \qquad \qquad Z_{22n} = \frac{Z_{22}}{Z_{02}}$
$T_{11} = \frac{(-1 - Y_{11n})(1 + Y_{22n}) + Y_{12n}Y_{21n}}{2Y_{21n} \left[\frac{R_{01}R_{02}}{Z_{01}Z_{02}}\right]^{1/2}}$	$Y_{11n} = \frac{\left[\frac{Z_{02}^*}{Z_{02}}\right](T_{11} - T_{21}) - (T_{12} - T_{22})}{T_{11}\left[\frac{Z_{01}^*}{Z_{01}^*}\right] - T_{12}\left[\frac{Z_{01}^*}{Z_{01}}\right] + T_{21}\left[\frac{Z_{02}^*}{Z_{02}}\right] - T_{22}}$
$T_{12} = \frac{(1+Y_{11n})\left[1-Y_{22n}\left[\frac{Z_{02}^*}{Z_{02}}\right]\right] + Y_{12n}Y_{21n}\left[\frac{Z_{02}^*}{Z_{02}}\right]}{2Y_{21n}\left[\frac{R_{01}R_{02}}{Z_{01}Z_{02}}\right]^{1/2}}$	$Y_{12n} = \frac{-2\left[\frac{R_{01}R_{02}}{Z_{01}Z_{02}}\right]^{1/2}(T_{11}T_{22} - T_{12}T_{21})}{T_{11}\left[\frac{Z_{01}^{*}Z_{01}^{*}}{Z_{01}Z_{02}^{*}}\right] - T_{12}\left[\frac{Z_{01}^{*}}{Z_{01}}\right] + T_{21}\left[\frac{Z_{02}^{*}}{Z_{02}^{*}}\right] - T_{22}}$
$T_{21} = \frac{\left[Y_{11n}\left[\frac{z_{01}^*}{z_{01}}\right] - 1\right](1 + Y_{22n}) - Y_{12n}Y_{21n}\left[\frac{z_{01}^*}{z_{01}}\right]}{2Y_{21n}\left[\frac{R_{01}R_{02}}{z_{01}z_{02}}\right]^{1/2}}$	$Y_{21n} = \frac{-2\left[\frac{R_{01}R_{02}}{Z_{01}}\right]^{1/2}}{T_{11}\left[\frac{Z_{01}^{5}Z_{02}^{5}}{Z_{01}^{2}Z_{02}}\right] - T_{12}\left[\frac{Z_{01}^{5}}{Z_{01}}\right] + T_{21}\left[\frac{Z_{02}^{5}}{Z_{02}^{5}}\right] - T_{22}}$
$T_{22} = \frac{\left[\frac{1 - Y_{11n} \left[\frac{Z_{01}^*}{Z_{01}}\right]}{\left[\frac{Z_{01}^*}{Z_{01}}\right]}\right] \left[1 - Y_{22n} \left[\frac{Z_{02}^*}{Z_{02}}\right]\right] - Y_{12n} Y_{21n} \left[\frac{Z_{01}^* Z_{02}^*}{Z_{01} Z_{02}}\right]}{2Y_{21n} \left[\frac{Z_{01}^* Z_{02}}{Z_{01} Z_{02}}\right]^{1/2}}$	$Y_{22n} = \frac{\left\lceil \frac{Z_{01}^{\perp}}{Z_{01}} \right\rceil (T_{11} + T_{12}) + (T_{21} + T_{22})}{T_{11} \left[\frac{Z_{01}^{\perp}}{Z_{01}^{\perp} Z_{02}^{\perp}} \right] - T_{12} \left[\frac{Z_{01}^{\perp}}{Z_{01}} \right] + T_{21} \left[\frac{Z_{02}^{\perp}}{Z_{02}^{\perp}} \right] - T_{22}}$
$Y_{11n} = Y_{11}Z_{01} Y_{12n} = Y_{12}(Z_{01}Z_{02})^{1/2}$	$Y_{21n} = Y_{21}(Z_{01}Z_{02})^{1/2} Y_{22n} = Y_{22}Z_{02}$
$T_{11} = \frac{(-h_{11n} - 1)(1 + h_{22n}) + h_{12n}h_{21n}}{2h_{21n} \left[\frac{R_{01}R_{02}}{Z_{01}Z_{02}}\right]^{1/2}}$	$h_{11n} = \frac{T_{11} \begin{bmatrix} Z_{01}^{\bullet} Z_{02}^{\bullet} \\ Z_{01}^{\bullet} Z_{02} \end{bmatrix} - T_{12} \begin{bmatrix} Z_{01}^{\bullet} \\ Z_{01} \end{bmatrix} + T_{21} \begin{bmatrix} Z_{02}^{\bullet} \\ Z_{02} \end{bmatrix} - T_{22}}{\begin{bmatrix} Z_{02}^{\bullet} \\ Z_{02} \end{bmatrix} (T_{11} - T_{21}) - (T_{12} - T_{22})}$
$T_{12} = \frac{(h_{11n} + 1)\left[1 - h_{22n}\left[\frac{Z_{02}^{\bullet}}{Z_{02}}\right]\right] + h_{12n}h_{21n}\left[\frac{Z_{02}^{\bullet}}{Z_{02}}\right]}{2h_{21n}\left[\frac{R_{01}R_{02}}{R_{01}Z_{02}}\right]^{1/2}}$	$h_{12n} = \frac{2\left[\frac{R_{01}R_{02}}{Z_{01}Z_{02}}\right]^{1/2} (T_{11}T_{22} - T_{12}T_{21})}{\left[\frac{Z_{02}}{Z_{02}}\right] (T_{11} - T_{21}) - (T_{12} - T_{22})}$
$T_{21} = \frac{\left[\frac{Z_{01}^{*}}{Z_{01}} - h_{11n}\right] (1 + h_{22n}) + h_{12n}h_{21n}}{2h_{21n}\left[\frac{R_{01}R_{02}}{Z_{01}Z_{02}}\right]^{1/2}}$	$h_{21n} = \frac{-2\left[\frac{R_{01}R_{02}}{Z_{01}Z_{02}}\right]^{1/2}}{\left[\frac{Z_{02}^*}{Z_{02}}\right](T_{11} - T_{21}) - (T_{12} - T_{22})}$
$T_{22} = \frac{\left[\frac{h_{11n} - \frac{Z_{01}^*}{Z_{01}}}{Z_{01}}\right] \left[\frac{1 - h_{22n}}{Z_{02}}\right] + h_{12n}h_{21n} \left[\frac{Z_{02}^*}{Z_{02}}\right]}{2h_{21n} \left[\frac{R_{01}R_{02}}{Z_{01}Z_{02}}\right]^{1/2}}$	$h_{22n} = \frac{T_{11} + T_{12} - T_{21} - T_{22}}{\left[\frac{Z_{02}}{Z_{02}}\right] (T_{11} - T_{21}) - (T_{12} - T_{22})}$
$h_{11n} = \frac{h_{11}}{Z_{01}}$ $h_{12n} = h_{12} \left[\frac{Z_{02}}{Z_{01}} \right]^{1/2}$	$h_{21n} = h_{21} \left[\frac{Z_{02}}{Z_{01}} \right]^{1/2}$ $h_{22n} = h_{22} Z_{02}$
$T_{11} = \frac{A_n + B_n + C_n + D_n}{2}$	$A_n = \frac{\left[\frac{Z_{01}^*}{Z_{01}}\right](T_{11} + T_{12}) + (T_{21} + T_{22})}{2\left[\frac{R_{01}R_{02}}{Z_{01}Z_{02}}\right]}$
$T_{12} = \frac{A_n \left[\frac{Z_{02}^*}{Z_{02}} \right] - B_n + C_n \left[\frac{Z_{02}^*}{Z_{02}} \right] - D_n}{2}$	$B_n = \frac{T_{11} \left[\frac{Z_{01}^2 Z_{02}^2}{Z_{01}^2 Z_{02}} \right] - T_{12} \left[\frac{Z_{01}^2}{Z_{01}} \right] + T_{21} \left[\frac{Z_{02}^2}{Z_{02}} \right] - T_{22}}{2 \left[\frac{R_{01}}{Z_{01}^2 Z_{02}} \right]}$
$T_{21} = \frac{A_n + B_n - C_n \left[\frac{z_{01}^*}{Z_{01}} \right] - D_n \left[\frac{z_{01}^*}{Z_{01}} \right]}{2}$	$C_n = \frac{T_{11} + T_{12} - T_{21} - T_{22}}{2 \left[\frac{R_{01} R_{02}}{R_{01} R_{02}} \right]}$
$T_{22} = \frac{A_n \left[\frac{Z_{02}^*}{Z_{02}} \right] - B_n - C_n \left[\frac{Z_{01}^* Z_{02}^*}{Z_{01} Z_{02}} \right] + D_n \left[\frac{Z_{01}^*}{Z_{01}} \right]}{2}$	$D_n = \frac{\left[\frac{Z_{02}^*}{Z_{02}}\right](T_{11} - T_{21}) - (T_{12} - T_{22})}{2\left[\frac{R_{01}R_{02}}{Z_{01}Z_{02}}\right]}$
$A_n = \frac{AZ_{02}}{(R_{01}R_{02})^{1/2}} \qquad B_n = \frac{B}{(R_{01}R_{02})^{1/2}}$	$C_n = \frac{CZ_{01}Z_{02}}{(R_{01}R_{02})^{1/2}}$ $D_n = \frac{DZ_{01}}{(R_{01}R_{02})^{1/2}}$

	EQUATIONS SHOWING THE CONVERSIONS I	SETWEEN Z , Y , h , AND $ABCD$ PARAMETERS	
$Z_{11} = \frac{Y_{22}}{Y_{11}Y_{22} - Y_{12}Y_{21}}$ $Z_{12} = \frac{Y_{21}}{Y_{11}Y_{22} - Y_{12}Y_{21}}$ $Z_{21} = \frac{Y_{22}}{Y_{11}Y_{22} - Y_{12}Y_{21}}$ $Z_{22} = \frac{Y_{22}}{Y_{21}Y_{22} - Y_{12}Y_{21}}$	$Y_{11} = \frac{Z_{22}}{Z_{11}Z_{22}Z_{12}Z_{21}}$ $Y_{12} = \frac{Z_{12}Z_{12}Z_{21}}{Z_{11}Z_{22}Z_{12}Z_{21}}$ $Y_{21} = \frac{Z_{11}Z_{22}Z_{12}Z_{21}}{Z_{11}Z_{22}Z_{12}Z_{21}}$ $Y_{22} = \frac{Z_{11}Z_{22}Z_{12}Z_{21}}{Z_{11}Z_{22}Z_{12}Z_{21}}$	$Z_{11} = \frac{h_{11}h_{22} - h_{12}h_{21}}{h_{22}}$ $Z_{12} = \frac{h_{12}}{h_{22}}$ $Z_{21} = \frac{-h_{21}}{h_{22}}$ $Z_{22} = \frac{1}{h_{22}}$	$h_{11} = \frac{Z_{11}Z_{22} - Z_{12}Z_{21}}{Z_{22}}$ $h_{12} = \frac{Z_{12}}{Z_{23}}$ $h_{21} = \frac{-Z_{21}}{Z_{22}}$ $h_{22} = \frac{1}{Z_{22}}$
$Z_{11} = \frac{A}{C}$ $Z_{12} = \frac{AD - BC}{C}$ $Z_{21} = \frac{1}{C}$ $Z_{22} = \frac{D}{C}$	$A = \frac{Z_{11}}{Z_{21}}$ $B = \frac{Z_{11}Z_{22} - Z_{12}Z_{21}}{Z_{21}}$ $C = \frac{1}{Z_{21}}$ $D = \frac{Z_{22}}{Z_{21}}$	$Y_{11} = \frac{1}{h_{11}}$ $Y_{12} = \frac{h_{12}}{h_{11}}$ $Y_{21} = \frac{h_{21}}{h_{21}}$ $Y_{22} = \frac{h_{11}h_{22} - h_{12}h_{21}}{h_{11}}$	$h_{11} = \frac{1}{Y_{11}}$ $h_{12} = \frac{Y_{12}}{Y_{11}}$ $h_{21} = \frac{Y_{21}}{Y_{21}}$ $h_{22} = \frac{Y_{11}Y_{22} - Y_{12}Y_{21}}{Y_{11}}$
$Y_{11} = \frac{D}{B}$ $Y_{12} = \frac{BC - AD}{B}$ $Y_{21} = \frac{-1}{B}$ $Y_{22} = \frac{A}{B}$	$A = \frac{-Y_{22}}{Y_{21}}$ $B = \frac{-1}{Y_{21}}$ $C = \frac{Y_{12}Y_{21} - Y_{11}Y_{22}}{Y_{21}}$ $D = \frac{-Y_{11}}{Y_{21}}$	$h_{11} = \frac{B}{D}$ $h_{12} = \frac{AD - BC}{D}$ $h_{21} = \frac{-1}{D}$ $h_{22} = \frac{C}{D}$	$A = \frac{h_{12}h_{21} - h_{11}h_{22}}{h_{21}}$ $B = \frac{-h_{11}}{h_{21}}$ $C = \frac{-h_{22}}{h_{21}}$ $D = \frac{1}{h_{21}}$

TABLE V

between T parameters and Z, Y, h, and ABCD parameters. Tables III and IV provide the conversions from \boldsymbol{S} and \boldsymbol{T} parameters to the normalized Z, Y, h, and ABCD parameters, respectively. From Tables III and IV, it is easy to see that if Z_{01} and Z_{02} are real, the conversions become those shown in many of the references cited, e.g., [2], [4], [7], [8], [11], [12], [14], [15]. Finally, Table V shows the conversions between Z, Y, h, and ABCD parameters while Table VI shows the conversions between \boldsymbol{S} and \boldsymbol{T} parameters. These are included to make the table of conversions in this paper complete.

IV. VERIFICATION

Using PSPICE, a SPICE based circuit analysis program, a lumped element model of an NE32000 HEMT was analyzed. The netlist was taken from the NEC databook and is shown below:

By properly configuring a source at first port 1 then port 2, and opening and shorting out the other port, PSPICE will provide the complex voltages and currents required to calculate the Z, Y, h, and ABCD parameters. Tables VII and VIII show the voltages and currents from PSPICE under the conditions listed in those tables. The Z, Y, h, and ABCD parameters are calculated from these using (1)-(4) and are shown in Table IX.

The NE32000 lumped element model was also analyzed using Super Compact. For no particular reason, I chose to

Equations Showing the Conversions Between ${\cal S}$ and ${\cal T}$ Parameters

$S_{11} = \frac{T_{21}}{T_{11}}$	$T_{11} = \frac{1}{S_{21}}$
$S_{12} = \frac{T_{11}T_{22} - T_{12}T_{21}}{T_{11}}$	$T_{12} = \frac{-S_{22}}{S_{21}}$
$S_{21} = \frac{1}{T_{11}}$	$T_{21} = \frac{S_{11}}{S_{21}}$
$S_{22} = \frac{-T_{12}}{T_{11}}$	$T_{22} = \frac{S_{12}S_{21} - S_{11}S_{22}}{S_{21}}$

calculate the S parameters for the NE32000 in a system with a source impedance, Z_{01} , equal to 70+j 30 and load impedance, Z_{02} , equal to 25-j 35 at the single frequency of 10 GHz. The results of the Super Compact analysis are shown in Table X.

If a person uses the Z, Y, h, or ABCD parameters of Table IX, in the equations of Table I, with $Z_{01} = 70 + j$ 30 and $Z_{02} = 25 - j$ 35, they will find that the calculated S parameters agree with those from Super Compact. In a like fashion, using the S parameters of Super Compact in the other equations in Table I will result in Z, Y, h, and ABCD parameters shown in Table IX.

V. CONCLUSION

This paper developed the equations for converting between the various common 2-port parameters, Z, Y, h, ABCD, S, and T. The equations are derived from the definitions of the various 2-port parameters, the definition of a_i and b_i , and basic transmission line theory. As a result, the equations are completely general and are valid for complex and unique source and load impedances.

The validity of these results is shown by first calculating S parameters from Z, Y, h, and ABCD parameters for an NE32000 HEMT in a system with $Z_S = 70 + j$ 30 and $Z_L = 25 - j$ 35. These results agreed with the S parameters produced by Super Compact. Also, beginning with the Sparameters from Super Compact, the Z, Y, h, and ABCD parameters are calculated using the equations developed. The results are the same as those calculated from the voltages and currents produced by PSPICE.

TABLE VII

VOLTAGES AND CURRENTS FOR THE NE32000 HEMT AT 10 GHz with the Source at Port 1. The Voltages and Currents are Defined in Fig. 1

$V_1 = 1 + j \ 0$					
$I_2 = 0$ (Port	2 Open Circuited)	$V_2 = 0$ (Por	$V_2 = 0$ (Port 2 Short Circuited)		
I_1	V_2	I_1	I_2		
$8.844E - 03 + j \ 2.371E - 02$	$-8.181E + 00 + j \ 5.615E + 00$	$2.010E - 03 + j \ 1.292E - 02$	$4.018E - 02 - j \ 1.071E - 02$		

TABLE VIII

VOLTAGES AND CURRENTS FOR THE NE32000 HEMT AT 10 GHz with the Source at Port 2. The Voltages and Currents are Defined in Fig. 1

	V_2	=1+j0		
$I_1 = 0$ (Por	t 1 Open Circuited)	$V_1 = 0$ (Port 1 Short Circuited)		
I_2	V_1	$\overline{I_2}$	I_1	
$8.032E - 03 + j \ 1.119E - 03$	$9.661E - 02 + i \cdot 1.869E - 02$	$3.949E - 03 + i \cdot 1.402E - 03$	$4.741E - 05 - j \ 1.286E - 03$	

TABLE IX

 $Z,\,Y,\,h,\,{
m and}\,\,ABCD$ Parameters for the NE3200 HEMT at 10 GHz. These Parameters WERE CALCULATED FROM THE VOLTAGES AND CURRENTS IN TABLES VII AND VIII USING (1)-(4)

	11	12	21	22
\overline{z}	$1.380E + 01 - j \ 3.702E + 01$	1.212E + 01 + j 6.395E - 01	$9.518E + 01 + j \ 3.803E + 02$	$1.221E + 02 - j \ 1.701E + 01$
Y	$2.010E - 03 + j \ 1.292E - 02$	$4.741E - 05 - j \ 1.286E - 03$	$4.018E - 02 - j \ 1.071E - 02$	$3.949E - 03 + j \ 1.402E - 03$
h	$1.176E + 01 - j \ 7.557E + 01$	$9.661E - 02 + j \ 1.869E - 02$	$-3.370E - 01 - j \ 3.162E + 00$	$8.032E - 03 + j \ 1.119E - 03$
	A	В	C	D
\overline{ABCD}	-8.309E - 02 - i 5.703E - 02	-2.324E + 01 - i 6.194E + 00	$6.173E - 04 - i \ 2.474E - 03$	$3.332E - 02 - i \ 3.127E - 01$

TABLE X SUPER COMPACT RESULTS FOR THE NE32000 HEMT

$Z_S = 70 + j \ 30$					$Z_L = 25 - j 35$				
		MIC	ROWAVE HAR	MONICA PC V	/1.06 File: ne32	0-1.ckt 25-FEB-	92 21:42:46		
Freq	MS11	PS11	MS21	PS21	MS12	PS12	MS22	PS22	MS21
GHz	mag	deg	mag	deg	mag	deg	mag	deg	dΒ
	NE320L	NE320L	NE320L	NE320L	NE320L	NE320L	NE320L	NE320L	NE320L
10.000	0.665	121.4	2.104	110 2	0.069	45.2	0.706	12.4	6.82

ACKNOWLEDGMENT

The author would like to thank Dr. Ulrich Rohde and Dr. Ray Pengelly of Compact Software for the use of Microwave Harmonica. He also wishes to acknowledge as his Lord and Savior Jesus Christ, and thank Him for His love and guidance.

REFERENCES

- [1] D. C. Youla, "On scattering matrices normalized to complex port numbers," Proc. IRE, vol. 49, p. 1221, July 1961.
 [2] S. Y. Liao, Microwave Circuit Analysis & Amplifier Design. Engle-
- wood Cliffs, NJ: Prentice-Hall, 1987, pp. 471-472.
- [3] P. I. Somlo and J. D. Hunter, Microwave Impedance Measurement. (Electrical Measurement Ser.: No. 2), IEE 1985, p. 23.
 [4] R. S. Carson, High-Frequency Amplifiers, 2nd ed. New York: Wiley
- 1982, p. 200.
 [5] R. W. Beatty and D. M. Kerns, "Relationships between different kinds
- parameters, not assuming reciprocity or equality of the waveguide or transmission line characteristic impedances," *Proc. IEEE*, vol. 52, p.
- 426, Apr. 1964.
 [7] Hewlett-Packard, "S-Parameter techniques for faster, more accurate network design," Appl. Note 95-1, Palo Alto, CA, Feb. 1967.
 [8] G. D. Vendelin, A. M. Pavio, and U. L. Rohde, Microwave Circuit Design Using Linear and Nonlinear Techniques. New York: Wiley,
- 1990, pp. 16-17.
 [9] D. M. Kerns and R. W. Beatty, Basic Theory of Waveguide Junctions and Introductory Microwave Network Analysis. Elmsford, NJ: Pergamon, 1967, pp. 136-139.

- [10] D. M. Pozar, Microwave Engineering. Reading: MA: Addison-Wesley 1990, p. 235.
- Microwave Systems News, The Microwave System Designer's Hand-
- book, 5th ed., vol. 17, no. 8, July 1987, p. 229.

 [12] S. F. Adam, *Microwave Theory and Applications*. Englewood Cliffs, NJ: Prentice-Hall, 1969, p. 89.
- [13] L. J. Giaceletto, Ed., Electronics Designers' Handbook, 2nd Ed. New
- York: McGraw-Hill, 1977, pp. 5-77-5-79. G. Gonzalez, Microwave Transistor Amplifiers: Analysis and Design.
- [14] G. Gonzaiez, microwave transistor Amplifiers: Analysis and Design. Englewood Cliffs, NJ: Prentice-Hall, 1984, pp. 24-25.
 [15] R. Soares, Ed., GaAs MESFET Circuit Design. Butler, WI: Artech 1988, pp. 92-93.
 [16] P. A. Rizzi, Microwave Engineering: Passive Circuits. Englewood Cliffs, NJ: Prentice-Hall, 1988, p. 538.

Dean A. Frickey (S'76-M'82) was born in Sheridan, WY on March 10, 1958. He received the B.S. and M.S. degrees in electrical engineering from the South Dakota School of Mines and Technology, Rapid City in 1980 and 1981, respectively. He is pursuing, part-time, the Ph.D. degree in electrical engineering through the University of Idaho, Moscow.

He spent one year with the Boeing Military Airplane Company, Seattle, WA, in a systems analysis group before becoming involved in microwave technology at Raytheon Missile Systems Division, Tewksbury, MA, where he spent 6 years mostly involved in the analysis and design of hybrid microwave integrated circuits, but with some time spent working with W-Band IMPATT diodes. In November 1989, he joined EG&G Idaho, which operates the Idaho National Engineering Laboratory. His research interests are in microwave theory and applications.