

## Studio 9: Root Locus with MATLAB

### Goal

The goal of this studio session is to use MATLAB to analyze and design control systems based on the root locus method. Everything in **red** is intended to be typed at a MATLAB command prompt. Results that should be displayed from MATLAB are listed in **blue**.

### MATLAB commands used in this session:

In this studio, you will practice some root locus-specific commands in MATLAB. The following MATLAB commands will be used. You can use "help command" to get information on the usage of a specific command.

#### Basic Functions:

`help` - MATLAB help function

#### Root Locus Functions:

`rlocus(sys)` - Calculates and plots the locus of the roots of `sys`  
`rlocfind` - Finds the root locus gains for a given set of roots  
`pzmap` - Pole-zero map of a linear system  
`sgrid` - Plot lines of constant damping ratio

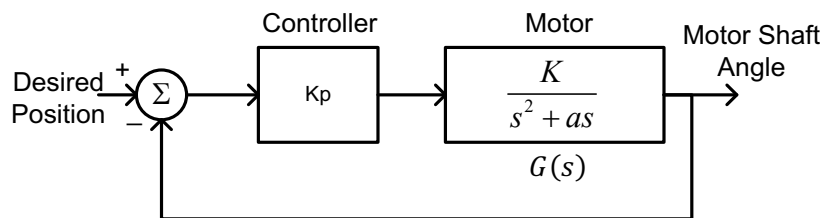
#### Root Locus Design GUI Tools:

`rltool` - Opens the SISO Design Tool for root locus design.

### 1.1 Motivation for Root Locus Plots

The root locus of a closed-loop system  $T(s)$  is the locus of all possible locations of the closed-loop system poles as a specified parameter of the system is varied. A common application of the root locus technique is to observe the variation in closed-loop poles as a linear gain element,  $K$ , is varied from zero to infinity. For example, this could help us determine a gain that will meet certain transient response specifications.

For example, consider the closed-loop system:



The closed-loop transfer function is

$$T(s) = \frac{K_p G(s)}{1 + K_p G(s)} = \frac{\frac{K_p K}{s(s+a)}}{1 + \frac{K_p K}{s(s+a)}} = \frac{K_p K}{s(s+a) + K_p K} = \frac{K_p K}{s^2 + as + K_p K}$$

The poles of the closed-loop system are those values of  $s$  which satisfy

$$1 + K_p G(s) = 0$$

or  $s^2 + as + K_p K = 0$

We would like to see how the behavior of this system changes as  $K_p$  changes – let's try some different values of  $K_p$ . We could obviously play around with different values of  $K_p$  (trial and error approach) to get the desired step response for this system, but this can take a while. Another approach would be to analytically calculate the poles of this transfer function and solve for a value of  $K_p$  that will place the poles at the point on the  $s$ -plane for a desired transient response.

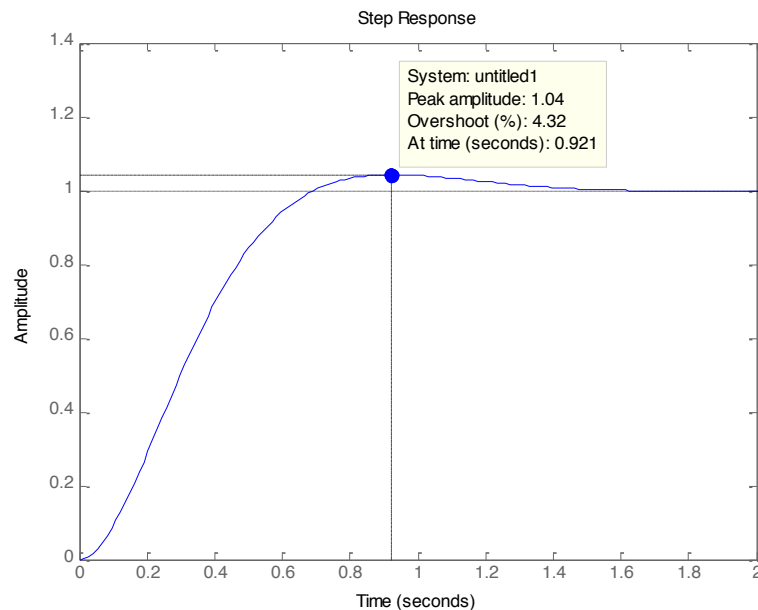
For example:

$$s = -\frac{a}{2} \pm \frac{j\sqrt{4K_p K - a^2}}{2}$$

For an underdamped response with damping ratio  $\zeta = 1/\sqrt{2}$ , we can set the magnitude of the real and imaginary parts of these roots equal to each other.

$$a = \sqrt{4K_p K - a^2} \rightarrow K_p = \frac{a^2}{2K} \rightarrow T(s) = \frac{a^2/2}{s^2 + as + a^2/2}$$

The step response with this  $K_p$  value provides a relatively fast response with a small overshoot as is expected from this  $\zeta$  value. However, this approach becomes more difficult if we have more complex systems (i.e., more complex transfer functions).

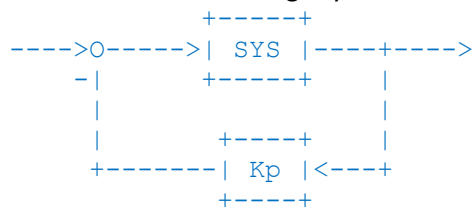


A better approach for finding good values of  $K_p$  is to use the root locus method.

## 1.2 Basics of Root Locus Using MATLAB

The root locus plots the path of the closed-loop (CL) system poles, i.e., roots of the denominator polynomial of the CL transfer function, as a parameter, usually a gain like  $K_p$  above, changes. In the lecture, we already discussed how to manually sketch the root locus diagram – a valuable skill since it allows the control system designer to develop the intuition necessary to see how changes to the loop transfer function  $K_p G(s)$  will affect the closed-loop system dynamics. However, as you have seen, this is a time consuming task, and MATLAB offers several powerful tools for plotting and analyzing the root locus. You will learn these tools in this studio.

MATLAB's rlocus command uses a slightly different system architecture as below.



In this case, if  $\text{sys}$  represents  $G(s)$ , the closed-loop transfer function  $T(s)$  is represented by:

$$T(s) = \frac{G(s)}{1 + K_p G(s)}$$

and thus the poles of the closed loop system are still those values of  $s$  which satisfy

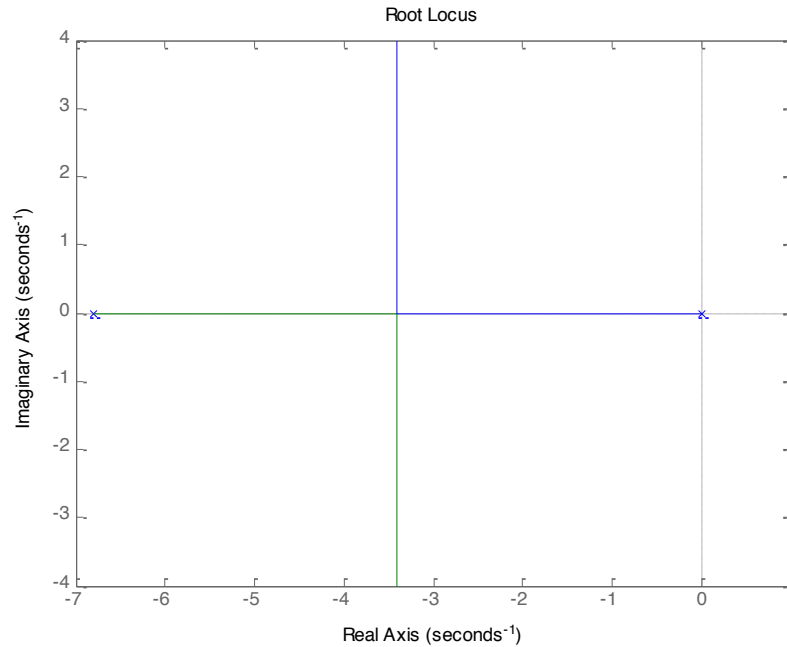
$$1 + K_p G(s) = 0$$

Using MATLAB's rlocus command, plot the root locus for the system above using  $K=1005$  and  $a=6.8$ .

```
K = 1005
a = 6.8
G = tf(K,[1 a 0])
    Transfer function:
         K
    -----
    s^2 + a s

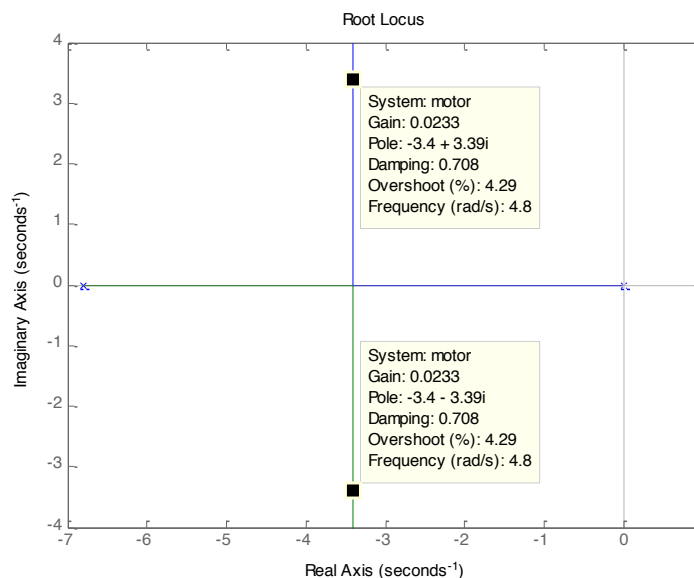
rlocus(G)    % plots the root locus of the system
```

A representative result is shown below. This plot shows all possible closed-loop pole locations for our pure gain  $K_p$ . The plot, however, does not give directions of the loci (what are they?). In this case, the starting points for the roots are shown by the two small 'x' marks at  $s = 0$  and  $s = -6.8$ . These starting points are given when  $K_p=0$  (\*\* note that  $K_p=0$  breaks the closed-loop transfer function and gives the roots of  $G(s)$  instead). As  $K_p$  increases, the two roots move horizontally toward each other, meeting at  $s = -3.4$ , and then move vertically away (break-away point) from each other.



Root locus of previous system.

The  $K_p$  value corresponding to points on this graph isn't plotted directly, but is easily found. Right click on the green line in the rlocus plot above. You should get a black square on the plot that you can then drag to different locations. You can do the same thing on the blue line. For example, find the gain  $K_p$  at pole locations where the real and imaginary parts are equivalent ( $s = -3.4 \pm 3.4j$  on this example plot). By clicking on this location, we find that gain  $K_p = 0.0233$  (which =  $a^2/2K$ ). This is the same  $K_p$  value that we calculated above!



You can also use the MATLAB command `rlocfind` to do the same thing. This function takes the system  $G$  and a vector of the desired pole locations as arguments and returns the  $K$  value.

```
rlocfind(G, [-3.4+3.4j, -3.4-3.4j])
```

ans =

0.0234      0.0234

This tells us that we want a  $K_p$  value of 0.0234 for a pole at  $-3.4+3.4j$  and a gain of 0.0234 for a pole value of  $-3.4-3.4j$ . Since a second order function will always have complex conjugates, we could only use one of these poles to find  $K_p$ .

### 1.3 Design Example

We'll start with a generic root locus design example and then you can go through the same analysis for your motor exercise in the assignment below.

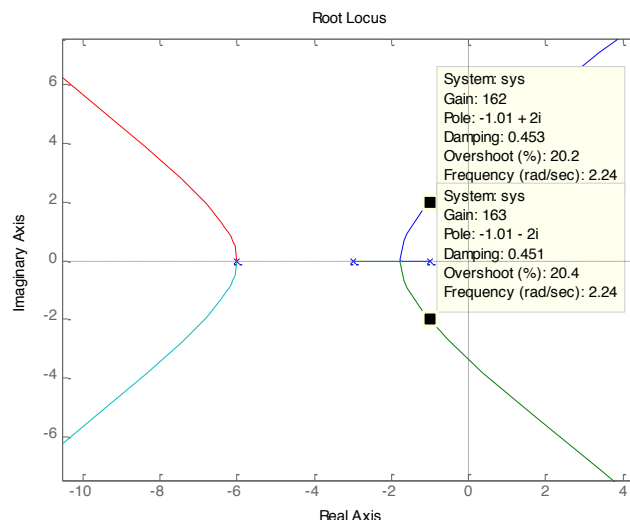
Given the following transfer function  $G(s)$ , pick a value of  $K_p$  that will yield second-order response of the closed-loop transfer function  $T(s)$  with a settling time of  $T_s = 4$  sec (assume negative feedback is used).

$$G(s) = \frac{1}{(s+1)(s+3)(s+6)^2}$$

First, recall that settling time for a second-order system is defined by  $T_s = 4/\sigma$ , where  $\sigma$  is the absolute value of the real part of the dominant second-order poles ( $\sigma = \omega_n \zeta$  for an underdamped stable 2<sup>nd</sup>-order system). Thus, to achieve a settling time of 4 sec,  $\sigma = 1$ , i.e. the dominant poles must have real parts equal to -1. To find suitable closed-loop pole locations, we will use `rlocus()` to graphically select the poles and find the value of  $K_p$  which will put the poles at this point of the root locus:

```
sys = zpk([], [-1 -3 -6 -6], 1)
rlocus(sys)
```

When you graphically pick your poles with real parts of -1, the result should look like this:



You may need to zoom in on your plot to find the exact value of  $K_p$  that results in a real part of -1. This plot tells us that  $K_p \approx 163$  will yield the desired settling time of 4s.

Note that you should also check that the 'approximate 2<sup>nd</sup>-order' assumption is valid. You can use `rlocfind` without a desired pole location to display all of the pole locations at a selected point.

```
[kp,poles] = rlocfind(sys)
```

Using `rlocfind` in this way will ask you to select a point in your rlocus plot. Select a point so that the real value of the dominant poles is approximately -1.

```
Select a point in the graphics window
```

```
selected_point =
```

```
-1.0545 - 2.0031i
```

```
kp =
```

```
159.3061
```

```
poles =
```

```
-6.9659 + 2.1526i
```

```
-6.9659 - 2.1526i
```

```
-1.0341 + 1.9898i
```

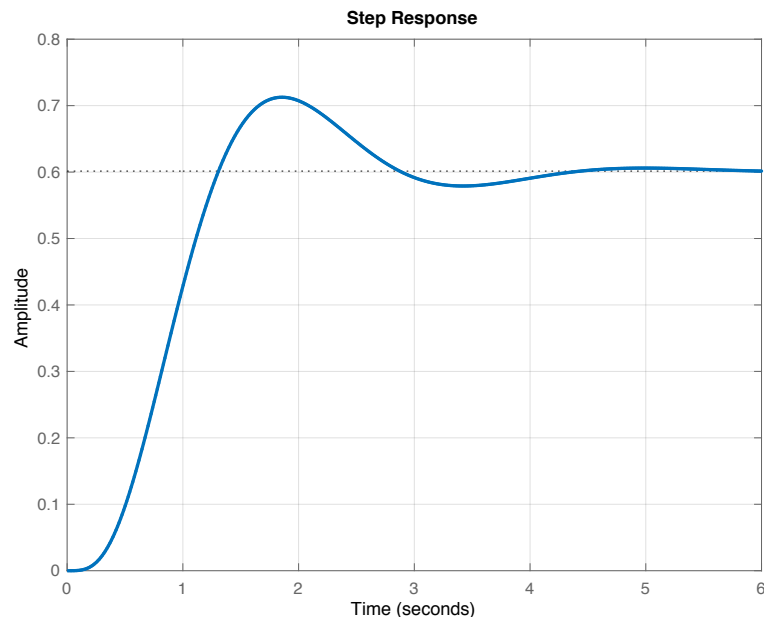
```
-1.0341 - 1.9898i
```

Are the non-dominant poles on the leftmost loci at least 5x further into the left half plane than the two dominant poles? If so, you can assume the system will behave approximately like a true 2<sup>nd</sup>-order system.

To validate your newly designed system, let's plot the time-domain response of the closed-loop system. First build the closed-loop system  $T(s)$ . To do this, assume that the feedback is unity gain and  $G(s)$  is in the forward path along with our newly selected controller gain  $K_p$ .

```
G = zpk([], [-1 -3 -6 -6], 1)
Kp=163
T = feedback(Kp*G,1)
step(T)
```

This code should result in the following step response plot:



The settling time  $T_s$  is defined as the time to reach and stay within 2% of the final steady-state value. Estimate your settling time (you can also right click, select Characteristics → Settling Time and move the mouse over the dot that appears on the plot). Did you meet the performance specification with your controller?

It is also possible to graphically plot the poles and zeros of a closed-loop system without drawing the entire root locus plot by using the `pzmap()` command. For example, if you want to see the location of the poles and zeros of a closed-loop system  $T(s)$ , simply enter to display the result.

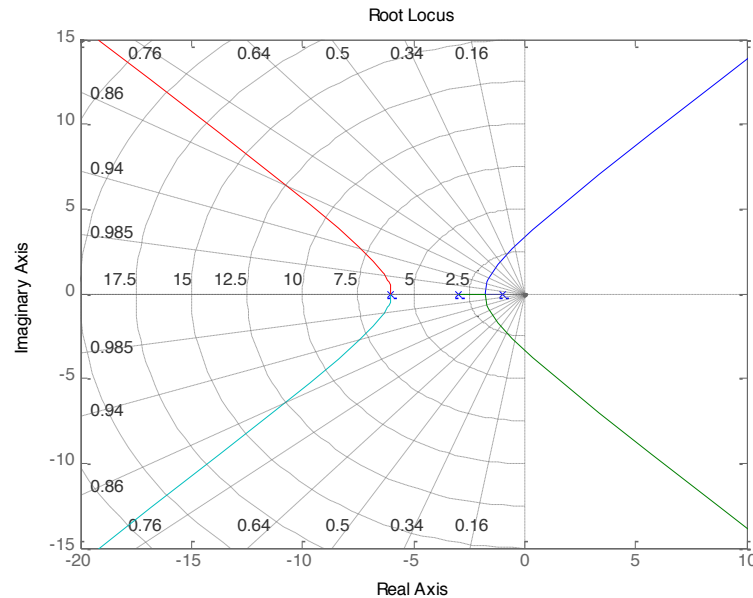
```
pzmap(T)
```

Of course in this case you need to use  $T(s)$  in the function `pzmap()`, unlike the root locus plot which only uses the open-loop transfer function  $G(s)$ . Note the locations of the non-dominant poles, which are far enough into the left-half plane compared to the dominant poles to significantly affect the dynamic response.

## 1.4 Plotting Constant Damping Ratio Lines

If the above design problem had required finding closed-loop poles with a particular damping ratio (or %OS), it would have been a bit more challenging to get the correct answer since the root locus plot does not show lines of constant damping. This is easily corrected using the `sgrid` command. To see how this works, re-plot the root locus for the previous example, and then apply the `sgrid` command:

```
rlocus(G)
sgrid
```



Note that both constant damping ( $\zeta$ ) and constant natural frequency ( $\omega_n$ , NOT damped frequency  $\omega_d$ ) lines are displayed in the left-half plane, i.e. both radial lines with constant slopes (constant  $\zeta$ ) and circular lines with constant radii (constant  $\omega_n$ ). The values of  $\zeta$  for each radial line are displayed around the periphery of the root locus plot.

## 1.5 Design Example Using rltool

MATLAB also has a special interactive design tool using the root locus. Let's go through the previous design example (settling time of 4s) using the `rltool`.

First, define the open-loop transfer function as before:

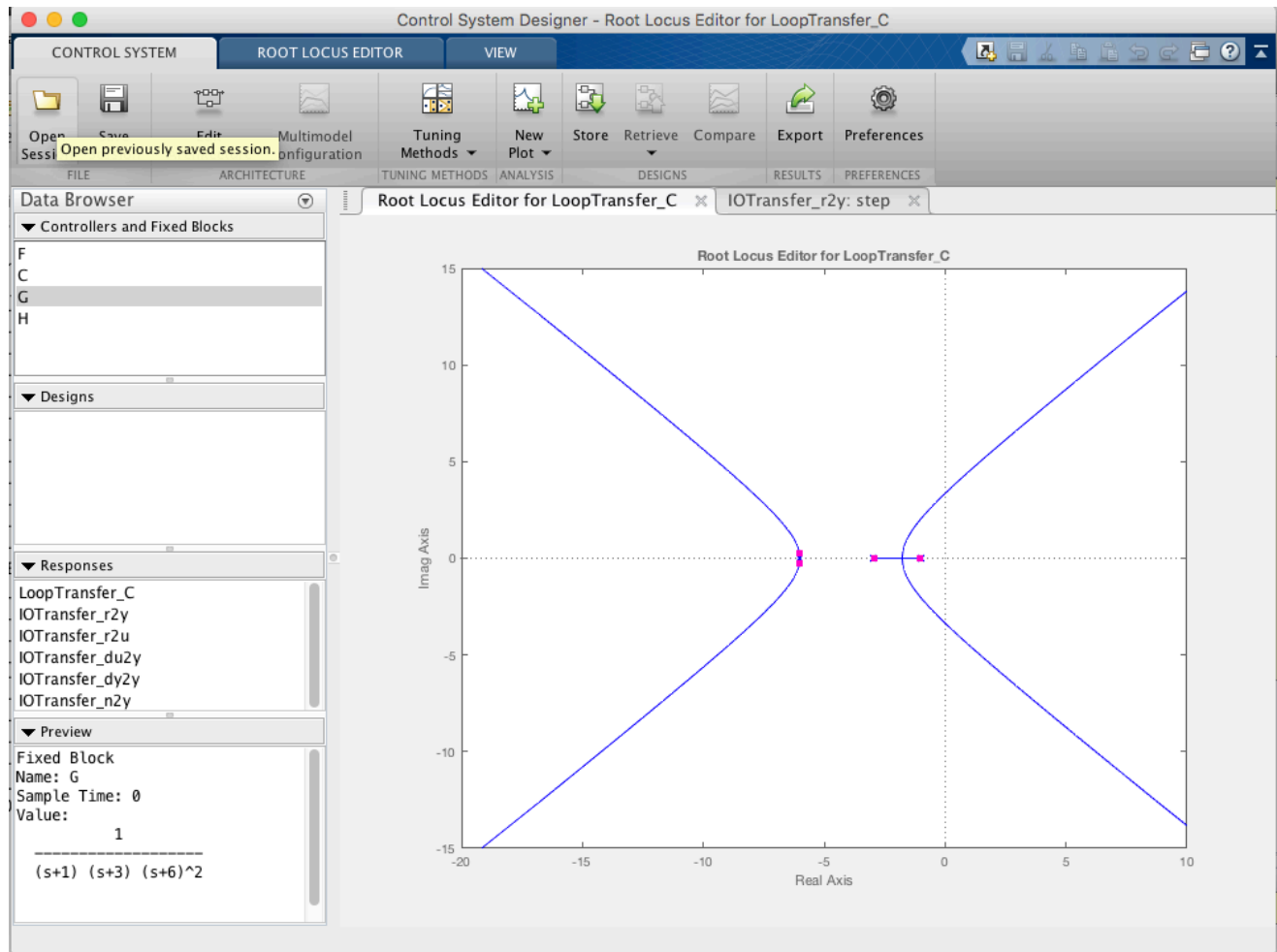
```
G = zpk([], [-1 -3 -6 -6], 1)
```

then, open `rltool` with `G` as an argument

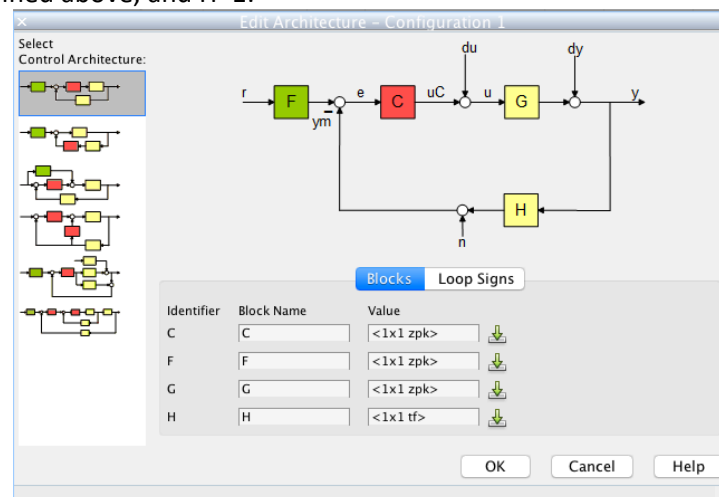
```
rltool(G)
```

A graphical user interface with the root locus plot and a "Control System Designer – Root Locus Editor for LoopTransfer\_C" window should pop up as shown below:



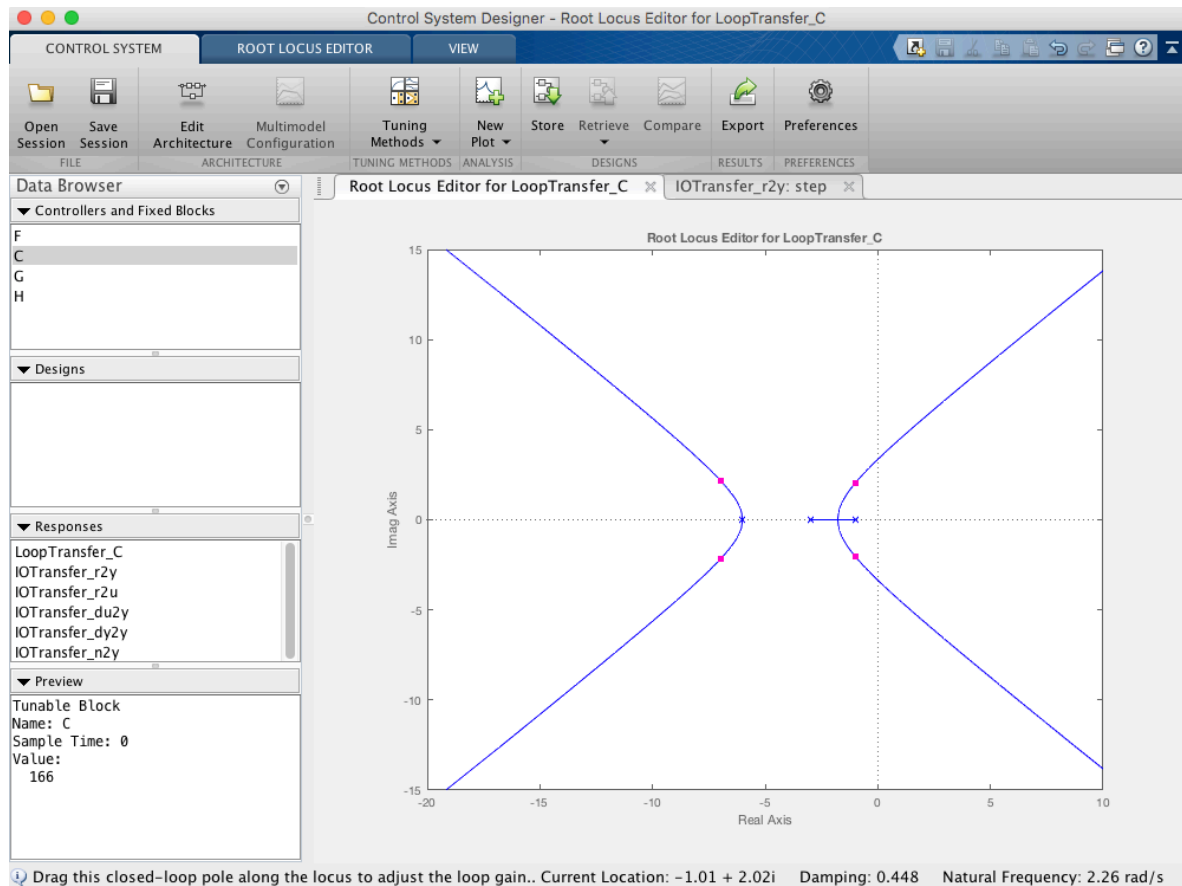


**rltool** defaults to a controller in the forward path as shown in the block diagram under Edit Architecture tab with  $F=1$ ,  $C=K_p$ ,  $G$  as defined above, and  $H=1$ :



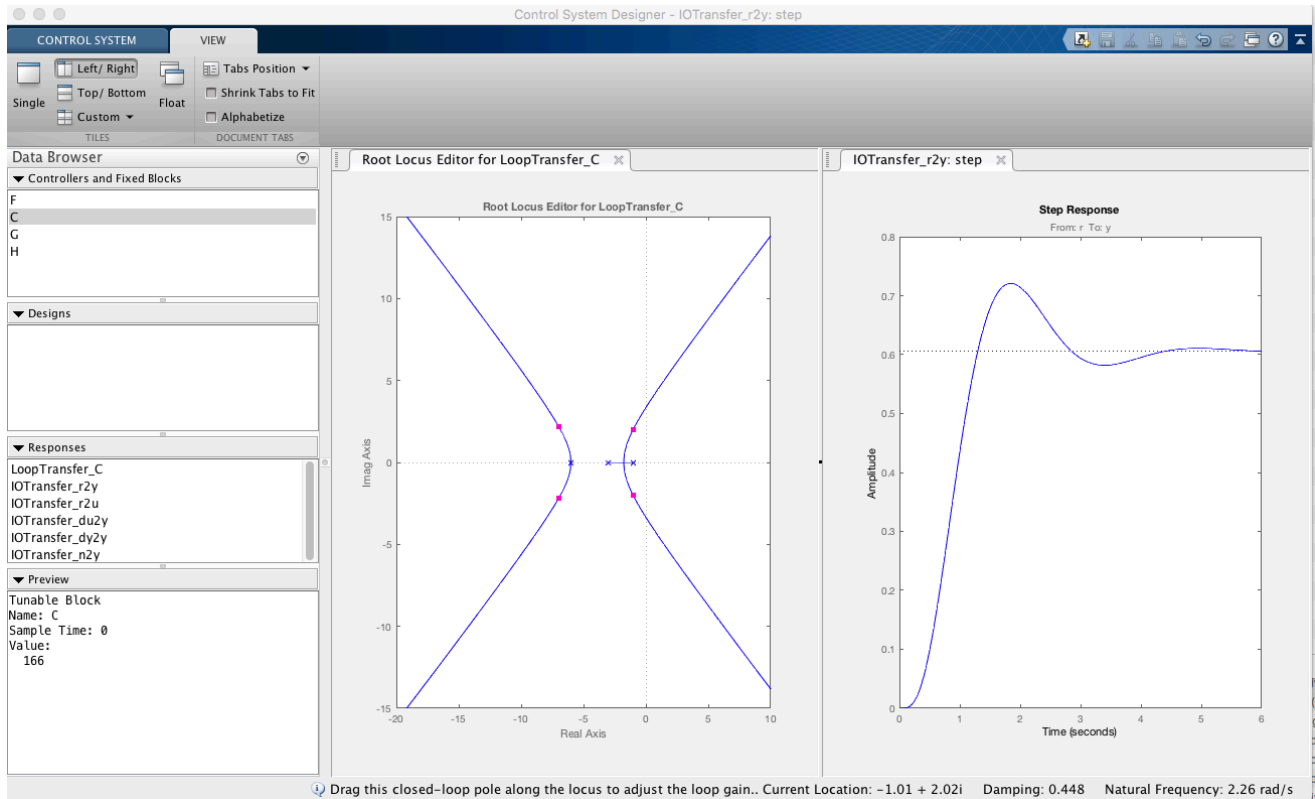
The pink squares on the root locus represent the closed-loop poles corresponding to the gain set point (note that the default  $K_p$  value is  $K_p = C = 1$ ).

To find a  $K_p$  value that results in a real pole location of  $\sigma = -1$  (for a settling time of 4s), drag one of the pink squares until you see a pole value of  $-1$ . The pole values are displayed on the bottom of the plot. The loop gain  $K_p$  is displayed in the Preview box in the lower left corner (make sure C is highlighted in the Controller and Fixed Blocks window).

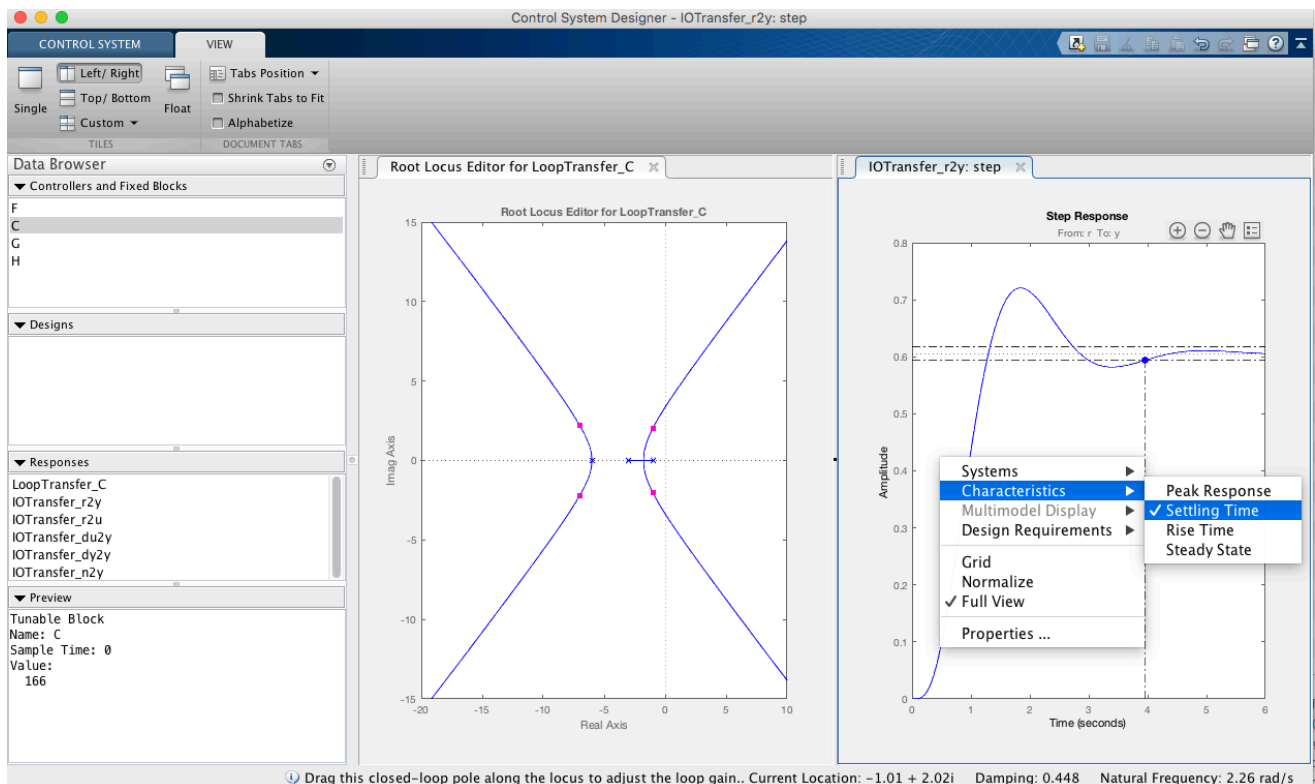


However, **rltool** provides an even easier way to find the  $K_p$  value that will result in a settling time of 4s.

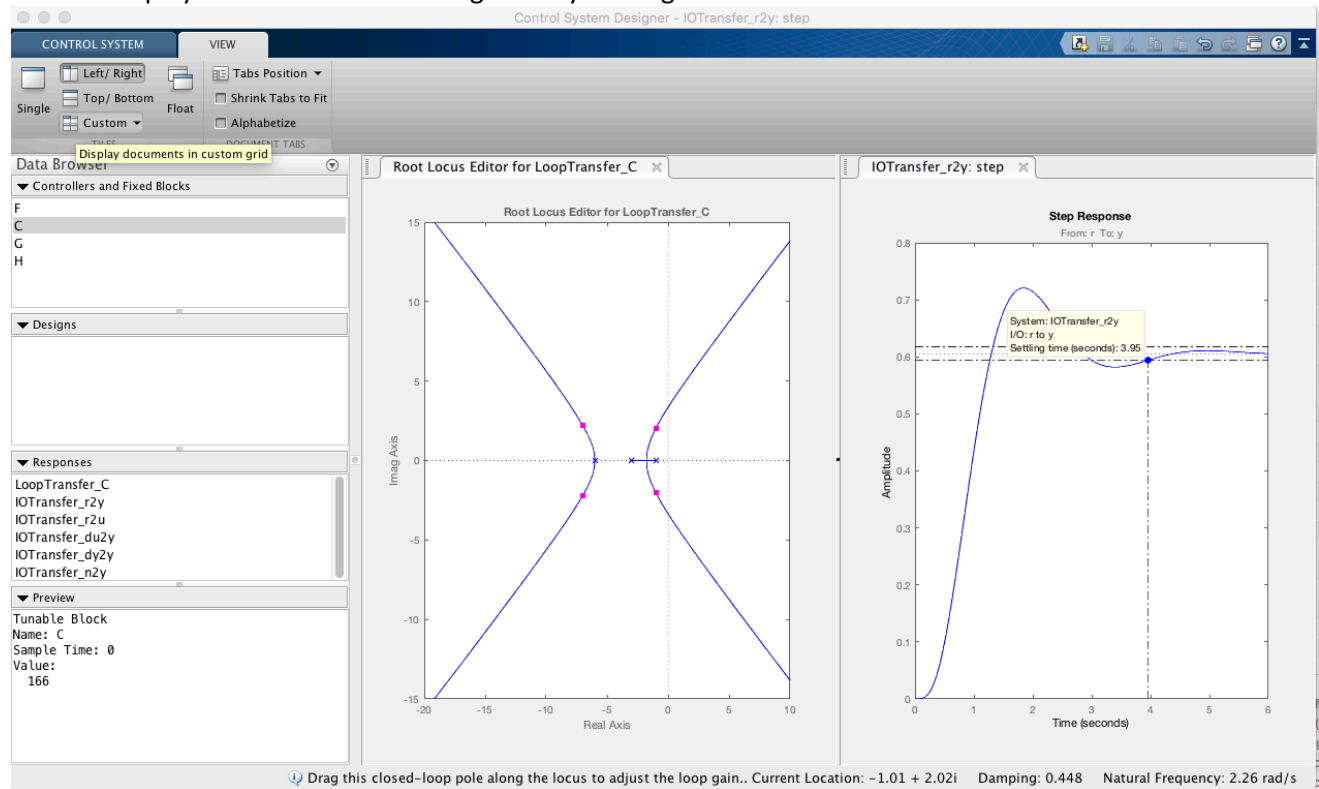
Click on the VIEW tab then click on Left/Right tab, this will result in a new subplot that shows the step response



To find the settling time from the step response, right click on the step response plot, select Characteristics > Settling Time. The settling time will be shown with a blue dot and a dotted line to the time-axis.

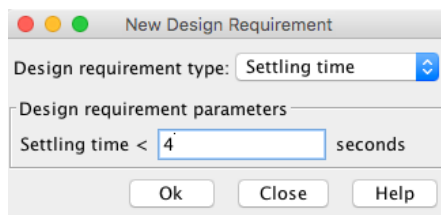


You can display the value of the settling time by clicking on the blue dot

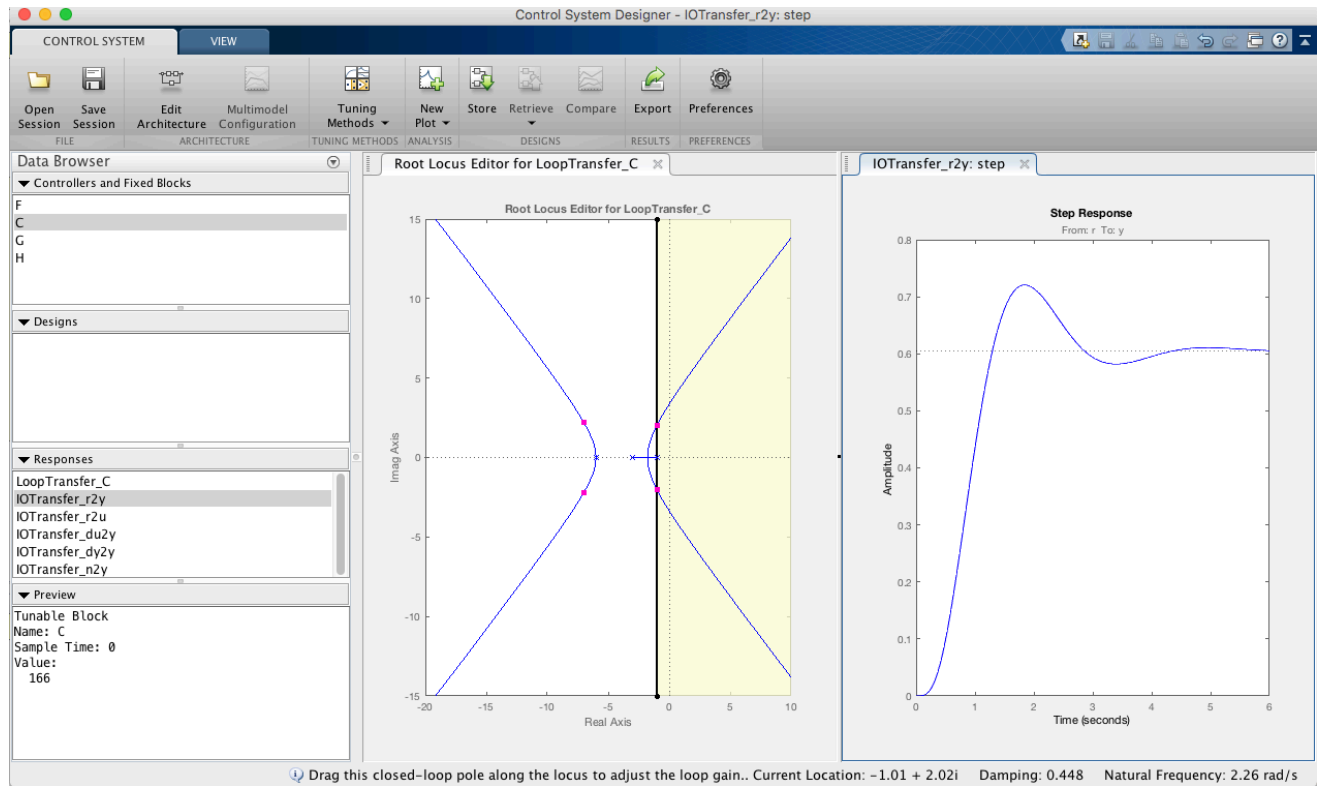


Try dragging one of the pink squares (closed-loop poles) in the root locus window to a new location and watch how the step response (and settling time) changes based on the location of the closed-loop poles. Find an approximate value of  $K_p$  that results in the shortest settling time possible. Does this pole location make sense?

Another way to find values of  $K_p$  that will result in a settling time less than 4s is to use the design feature in **rltool**. Right click in the root locus plot and select "Design Requirements" → New. A window will pop up that allows you to select several transient response parameters. Select settling time < 4 seconds as the design guideline, click Ok.



The root locus plot will now include a yellow area that defines gains that will NOT meet your maximum 4s settling time. Does this yellow area make sense based on the value you found for a settling time of 4s above?



## Exercise

Each student shall complete the exercise below and get their work checked off by the studio instructor.

**For remote students and students who do not finish within studio session:**

Compile your answers and outputs of Exercises 1 along with the MATLAB code you used to answer the questions in a word file and name it LastNameFirstNameStudio09.docx and upload your file to Canvas under the Studio09 link. Your file shall include the solutions to parts (a) through (e).

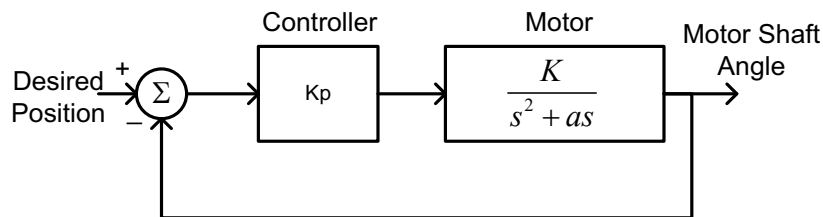
**Exercise 1. Design a control system for a (motor) transfer function using what you learned above.**

Let

$$G(s) = \frac{1005}{s(s + 6.8)}$$

A proportional controller, i.e., a pure gain, is added to the forward path of a unity-gain negative feedback loop placed around the plant as shown in the block diagram below. The transfer function of this controller is:

$$G_c(s) = K_p$$



Block diagram for Motor and proportional controller,  $K=1005$ ,  $a=6.8$ .

- (a)** Plot the root locus of this system using MATLAB's `rlocus` command.
- (b)** Find a value of  $K_p$  that will produce a closed-loop system with a damping ratio  $\zeta = 0.7$  (this can be approximate). Provide your chosen  $K_p$  and the location of all closed-loop poles for this gain.
- (c)** Calculate  $T_s$ ,  $T_p$ ,  $T_r$  and %OS for the value of  $K_p$  and closed-loop poles found in part (b).
- (d)** Plot the step response for the closed-loop system ( $T(s)$  NOT  $G_c(s)*G(s)$ ) for your chosen  $K_p$  to verify your results from (c).
- (e)** Is it possible to choose a gain that decreases the settling time in this system while maintaining a damping ratio  $< 0.7$ ? Why or why not?