

EE 523: TAKE HOME MIDTERM

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PART 1

A gyroelectric medium at frequency ω has the constitutive relations: $\vec{D} = \vec{\bar{\epsilon}} \cdot \vec{E}$, $\vec{B} = \mu_0 \vec{H}$. The permeability μ is a scalar, whereas the dyadic permittivity $\vec{\bar{\epsilon}}$ is represented by the matrix:

$$\vec{\bar{\epsilon}} = \begin{bmatrix} \varepsilon_1 & j\varepsilon_2 & 0 \\ -j\varepsilon_2 & \varepsilon_1 & 0 \\ 0 & 0 & \varepsilon_3 \end{bmatrix}$$

where $\varepsilon_1 > 0$, $|\varepsilon_2| < \varepsilon_1$, and $\varepsilon_3 > 0$.

a. Our main aim is to characterize circularly polarized plane waves propagating in z -direction within this medium. For this purpose, we consider the two (complex-valued) unit vectors denoting the circular directions $\hat{e}_+ = \frac{1}{2}(\hat{a}_x - j\hat{a}_y)$ and $\hat{e}_- = \frac{1}{2}(\hat{a}_x + j\hat{a}_y)$ for right and left polarizations. Show that any \vec{E} field in the form $\vec{E} = E_x \hat{a}_x + E_y \hat{a}_y$ can be converted to the form $\vec{E} = E_+ \hat{e}_+ + E_- \hat{e}_-$, where $E_+ = E_x + jE_y$, and $E_- = E_x - jE_y$.

Remark: Note that this operation is a basis change from the linear basis $\{\hat{a}_x, \hat{a}_y\}$ to the circular basis $\{\hat{e}_+, \hat{e}_-\}$

Solution:

The circular coordinate transformation defined in the problem can be given as:

$$(1) \quad \begin{bmatrix} \frac{1}{2} & -\frac{j}{2} \\ \frac{1}{2} & \frac{j}{2} \end{bmatrix} \begin{bmatrix} \hat{a}_x \\ \hat{a}_y \end{bmatrix} = \begin{bmatrix} \hat{e}_+ \\ \hat{e}_- \end{bmatrix}$$

And also,

$$(2) \quad \begin{bmatrix} E_x & E_y \end{bmatrix} \begin{bmatrix} \hat{a}_x \\ \hat{a}_y \end{bmatrix} = \begin{bmatrix} E_+ & E_- \end{bmatrix} \begin{bmatrix} \hat{e}_+ \\ \hat{e}_- \end{bmatrix}$$

After replacing equation 1 into equation 2, the following relation can be obtained:

$$(3) \quad \begin{bmatrix} E_x & E_y \end{bmatrix} \begin{bmatrix} \hat{a}_x \\ \hat{a}_y \end{bmatrix} = \begin{bmatrix} E_+ & E_- \end{bmatrix} \begin{bmatrix} \frac{1}{2} & -\frac{j}{2} \\ \frac{1}{2} & \frac{j}{2} \end{bmatrix} \begin{bmatrix} \hat{a}_x \\ \hat{a}_y \end{bmatrix}$$

Therefore,

$$(4) \quad \begin{bmatrix} E_x & E_y \end{bmatrix} = \begin{bmatrix} E_+ & E_- \end{bmatrix} \begin{bmatrix} \frac{1}{2} & -\frac{j}{2} \\ \frac{j}{2} & \frac{1}{2} \end{bmatrix}$$

After inverting the matrix in equation 4 and multiplying both side of equation 4, the following system can be obtained:

$$(5) \quad \begin{bmatrix} E_x & E_y \end{bmatrix} \begin{bmatrix} 1 & 1 \\ j & -j \end{bmatrix} = \begin{bmatrix} E_+ & E_- \end{bmatrix}$$

As a result,

$$(6) \quad E_+ = E_x + jE_y,$$

$$(7) \quad E_- = E_x - jE_y.$$

b. Show that under this transformation the permittivity matrix is diagonalized. That is:

$$\begin{bmatrix} D_+ \\ D_- \\ D_z \end{bmatrix} = \begin{bmatrix} \varepsilon_1 + \varepsilon_2 & 0 & 0 \\ 0 & \varepsilon_1 - \varepsilon_2 & 0 \\ 0 & 0 & \varepsilon_3 \end{bmatrix} \begin{bmatrix} E_+ \\ E_- \\ E_z \end{bmatrix}$$

where $D_{\pm} = D_x \pm jD_y$.

Solution:

By indicial notation, the relation $\bar{D} = \bar{\varepsilon} \cdot \bar{E}$ can be given as:

$$(8) \quad D_i = \varepsilon_{ij} E_j.$$

For given basis transformation, the linear coordinate transformation matrix is found as:

$$(9) \quad a_{ij} = \begin{bmatrix} 1 & j & 0 \\ 1 & -j & 0 \\ 0 & 0 & 1 \end{bmatrix}$$

After applying the coordinate transformation to the equation 8, it can be given as:

$$(10) \quad a_{ki} D_i = a_{ki} a_{lj} \varepsilon_{ij} a_{lj} E_j.$$

Note that

$$(11) \quad a_{ki} D_i = \begin{bmatrix} D_+ \\ D_- \\ D_z \end{bmatrix},$$

$$(12) \quad a_{lj} E_j = \begin{bmatrix} E_+ \\ E_- \\ E_z \end{bmatrix},$$

As a result,

(13)

$$a_{ki}a_{lj}\varepsilon_{ij} = \begin{bmatrix} 1 & j & 0 \\ 1 & -j & 0 \\ 0 & 0 & 1 \end{bmatrix} \begin{bmatrix} \varepsilon_1 & j\varepsilon_2 & 0 \\ -j\varepsilon_2 & \varepsilon_1 & 0 \\ 0 & 0 & \varepsilon_3 \end{bmatrix} \begin{bmatrix} 1 & 1 & 0 \\ j & -j & 0 \\ 0 & 0 & 1 \end{bmatrix} = \begin{bmatrix} \varepsilon_1 + \varepsilon_2 & 0 & 0 \\ 0 & \varepsilon_1 - \varepsilon_2 & 0 \\ 0 & 0 & \varepsilon_3 \end{bmatrix}$$

c. Show that the wave vectors for right and left circularly polarized plane waves propagating in z-direction are given as: $k_+ = \omega\sqrt{\mu\varepsilon_+}$ and $k_- = \omega\sqrt{\mu\varepsilon_-}$, where $\varepsilon_{\pm} = \varepsilon_1 \pm \varepsilon_2$. Note that the permittivity ε_3 has no role in this formulation.

Solution:

In a source free region, the governing equations for the electric field can be reduced to the following single differential equation:

$$(14) \quad (\mathbf{k} \cdot \mathbf{E})\mathbf{k} - (\mathbf{k} \cdot \mathbf{k})\mathbf{E} = -\omega^2\mu_0\bar{\varepsilon} \cdot \mathbf{E},$$

and for the wave field $\bar{E} = E_0e^{-jk_+z}\hat{e}_+$, the equation 14 can be given as:

$$(15) \quad -k_+^2 E_0 e^{-jk_+z} = -\omega^2\mu_0\varepsilon_+ E_0 e^{-jk_+z}.$$

For non-trivial electric field, the following relation must hold:

$$(16) \quad k_+ = \omega\sqrt{\mu\varepsilon_+}$$

Similarly, for the wave field $\bar{E} = E_0e^{-jk_-z}\hat{e}_-$, the equation 14 can be given as:

$$(17) \quad -k_-^2 E_0 e^{-jk_-z} = -\omega^2\mu_0\varepsilon_- E_0 e^{-jk_-z}.$$

For non-trivial electric field, the following relation must hold:

$$(18) \quad k_- = \omega\sqrt{\mu\varepsilon_-}$$

d. Assume that the region $0 < z < d$ is filled with this gyroelectric medium, whereas the regions $z < 0$ and $z > d$ are free space. Assume that a linearly polarized plane wave $\bar{E}^i = E_0e^{-jk_0z}\hat{a}_x$ in the free space region $z < 0$ is incident to the slab. Show that the transmitted wave in the region $z > d$ will be a linearly polarized plane wave $\bar{E}^t = \alpha E_0 [\cos\psi\hat{a}_x + \sin\psi\hat{a}_y] e^{-jk_0z}$, where α is a constant of proportionality (i.e. we are not concerned with the reflected waves from the boundaries of the slab), and ψ is the angle of rotation of the polarization direction which depends on k_+ , k_- and d . Evaluate ψ .

This phenomenon is known as Faraday rotation, and it was experimentally discovered by Michael Faraday in 1845.

Hint: Decompose the wave propagating in the slab into its right and left circular components, and relate the rotation angle to the phase difference between the circular components at $z = d$, where the wave leaves the medium.

Solution:

The interface conditions at $z = 0$ can be given as:

$$(19) \quad E_{1t} = E_{2t},$$

$$(20) \quad \mathbf{a}_n \times (\mathbf{H}_1 - \mathbf{H}_2) = 0.$$

where n and t subscripts stand for normal and tangential components of the vector field.

$$(21) \quad \bar{E}_{tra} = E_- e^{-jk_- z} \hat{e}_- + E_+ e^{-jk_+ z} \hat{e}_+ = T_\perp^1 E_0 e^{-jk_0 z} \hat{a}_x.$$

which has to be satisfied at $z = 0$ where T_\perp^1 is the transmission coefficients, therefore,

$$(22) \quad E_- = E_+ = T_\perp^1 E_0.$$

At $z = d$, the electric field can be given as:

$$(23) \quad T_\perp^1 E_0 (e^{-jk_- d} \hat{e}_- + e^{-jk_+ d} \hat{e}_+) = \frac{T_\perp^1 E_0}{2} (\hat{a}_x (e^{-jk_- d} + e^{-jk_+ d}) + j \hat{a}_y (e^{-jk_+ d} - e^{-jk_- d})).$$

In the equation above, the sums and the differences of complex exponentials can be expanded as:

$$(24) \quad e^{-jk_- d} + e^{-jk_+ d} = e^{-j \frac{k_+ + k_-}{2} d} (e^{-j \frac{k_+ - k_-}{2} d} + e^{-j \frac{k_- - k_+}{2} d}) = 2 \cos\left(\frac{k_+ - k_-}{2} d\right) e^{-j \frac{k_+ + k_-}{2} d}$$

and

$$(25) \quad e^{-jk_+ d} - e^{-jk_- d} = e^{-j \frac{k_+ + k_-}{2} d} (e^{-j \frac{k_+ - k_-}{2} d} - e^{-j \frac{k_- - k_+}{2} d}) = -2j \sin\left(\frac{k_+ - k_-}{2} d\right) e^{-j \frac{k_+ + k_-}{2} d}$$

As a result, the field can be expressed as:

$$(26) \quad T_\perp^1 E_0 e^{-j \frac{k_+ + k_-}{2} d} (\cos\left(\frac{k_+ - k_-}{2} d\right) \hat{a}_x + \sin\left(\frac{k_+ - k_-}{2} d\right) \hat{a}_y).$$

Since at the second transmission, the transmitted wave field after $z > d$ has to match the electric field above due to interface conditions. The rotation angle of the polarization direction can be found as:

$$(27) \quad \psi = \frac{k_+ - k_-}{2} d.$$

PART 2

Using the guidelines given below, show that a plasma medium acts as a gyroelectric medium when a constant external magnetic field $\vec{B} = B_0 \hat{a}_z$ is applied.

a. The equation of motion of a free electron in this medium is governed by the following differential equation $m_e \frac{d\hat{v}}{dt} = e(\hat{E} + \hat{v} \times \hat{B})$, where m_e and e are the electron mass and charge respectively. Show that the components of the velocity vector \hat{v} satisfy:

$$\begin{aligned}\frac{dv_x}{dt} &= \frac{e}{m_e} E_x + \omega_b v_y, \\ \frac{dv_y}{dt} &= \frac{e}{m_e} E_y - \omega_b v_x.\end{aligned}$$

where $\omega_b = \frac{eB_0}{m_e}$ is called the cyclotron frequency.

Solution:

The forcing due to magnetic field can be expanded as:

$$(28) \quad \hat{v} \times \hat{B} = \det \begin{pmatrix} \hat{a}_x & \hat{a}_y & \hat{a}_z \\ v_x & v_y & 0 \\ 0 & 0 & B_0 \end{pmatrix} = B_0 v_y \hat{a}_x - B_0 v_x \hat{a}_y.$$

After replacing this term into equation of motion, the set of governing equations can be obtained as:

$$(29) \quad m_e \frac{dv_x}{dt} = e(E_x + B_0 v_y),$$

$$(30) \quad m_e \frac{dv_y}{dt} = e(E_y - B_0 v_x).$$

Dividing both sides of equations 29 and 30 results in the following form:

$$(31) \quad \frac{dv_x}{dt} = \frac{e}{m_e} E_x + \omega_b v_y,$$

$$(32) \quad \frac{dv_y}{dt} = \frac{e}{m_e} E_y - \omega_b v_x.$$

where $\omega_b = \frac{eB_0}{m_e}$ is called the cyclotron frequency.

b. In order to solve these coupled differential equations, search (in phasor domain) solutions in the form $v_{\pm} = v_x \pm jv_y$. Show that:

$$v_{\pm} = v_x \pm jv_y = \frac{\frac{e}{m_e}(E_x \pm jE_y)}{j(\omega \pm \omega_b)}.$$

Solution:

Fourier transforming equations 31 and 32 in time results in the following set of algebraic equations:

$$(33) \quad j\omega v_x = \frac{e}{m_e} E_x + \omega_b v_y,$$

$$(34) \quad j\omega v_y = \frac{e}{m_e} E_y - \omega_b v_x.$$

In matrix form, equations 33 and 34 can be given as:

$$(35) \quad \begin{bmatrix} j\omega & -\omega_b \\ \omega_b & j\omega \end{bmatrix} \begin{bmatrix} v_x \\ v_y \end{bmatrix} = \begin{bmatrix} \frac{e}{m_e} E_x \\ \frac{e}{m_e} E_y \end{bmatrix}$$

By inverting the matrix in equation 35, the velocity components can be obtained as:

$$(36) \quad \begin{bmatrix} v_x \\ v_y \end{bmatrix} = \frac{1}{\omega_b^2 - \omega^2} \begin{bmatrix} j\omega & \omega_b \\ -\omega_b & j\omega \end{bmatrix} \begin{bmatrix} \frac{e}{m_e} E_x \\ \frac{e}{m_e} E_y \end{bmatrix}$$

Therefore,

$$(37) \quad v_+ = v_x + jv_y = \frac{\frac{e}{m_e}(E_x + jE_y)j(\omega - \omega_b)}{\omega_b^2 - \omega^2} = \frac{\frac{e}{m_e}(E_x + jE_y)}{j(\omega_b + \omega)},$$

and also

$$(38) \quad v_- = v_x - jv_y = \frac{\frac{e}{m_e}(E_x - jE_y)j(\omega + \omega_b)}{\omega_b^2 - \omega^2} = \frac{\frac{e}{m_e}(E_x - jE_y)}{j(\omega - \omega_b)}.$$

c. Finally let N be the number of electrons per unit volume. Show that $\varepsilon_{\pm} = \varepsilon_1 \pm \varepsilon_2 = \varepsilon_0 \left[1 - \frac{\omega_p^2}{\omega(\omega \pm \omega_b)} \right]$, where $\omega_p = \sqrt{\frac{Ne^2}{m_e \varepsilon_0}}$ is the plasma frequency. Hint: Integrate the velocity to find the displacement of a single electron and obtain the susceptibility of the bulk material in phasor domain.

Solution:

The dipole moment is defined as:

$$(39) \quad \bar{p}(t) = q(-l(t))\hat{e}_{\pm}$$

where $q = e$ and $l(t)$ can be obtained by integrating the velocity field v_{\pm} as:

$$(40) \quad l(t) = \frac{v_{\pm}}{j\omega} = -\frac{\frac{e}{m_e}(E_x \pm jE_y)}{\omega(\omega \pm \omega_b)}$$

The macroscopic electric polarization vector $\bar{P}(t)$ is evaluated as:

$$(41) \quad \bar{P}(t) = Np(t)$$

therefore,

$$(42) \quad \bar{P}(t) = -\frac{\frac{Ne^2}{m_e}}{\omega(\omega \pm \omega_b)}(E_x \pm jE_y).$$

Also

$$(43) \quad \bar{D} = \varepsilon_0 \bar{E} + \bar{P}.$$

After replacing equation 42 into the relation 43, the following relation is obtained:

$$(44) \quad \bar{D} = \varepsilon_0 \bar{E} \left(1 - \frac{\frac{Ne^2}{m_e \varepsilon_0}}{\omega(\omega \pm \omega_b)} \right)$$

As a result,

$$(45) \quad \varepsilon_{\pm} = \varepsilon_1 \pm \varepsilon_2 = \varepsilon_0 \left[1 - \frac{\omega_p^2}{\omega(\omega \pm \omega_b)} \right].$$