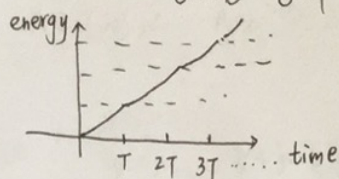


$$1. a. \text{energy} = \frac{V^2}{R} t = \frac{1 + e^{-\frac{2t}{T}} - 2e^{-\frac{t}{T}}}{R} t$$

energy is accumulated in each period



$$b. \frac{1}{T} \int_0^T \frac{(1 - e^{-\frac{t}{T}})^2}{R} dt = \frac{1}{TR} \left[t - \frac{T}{2} e^{-\frac{2t}{T}} + 2T e^{-\frac{t}{T}} \right] \Big|_0^T = \frac{1 - \frac{1}{2}e^{-2} + 2e^{-1}}{R}$$

$$4. \quad C V_C' + \frac{V_C}{10} + \frac{V_2 + V_C}{10} = 12A = \frac{V_2}{5} + 6 + \frac{1}{L} \int_0^{0^+} V_2 dt + \frac{V_2 + 60}{10} = 12A \quad V_2(0^+) = 0V$$

$V_2(0^-)$ comes from previous state when switch is ~~open~~^{closed}, and $V_2(0^-) = 0V$ because

the inductor acts like a short. the current in circuit, $i = \frac{V_2}{5} + \frac{1}{L} \int_0^t V_2 dt + i_L(0^-) =$

$$\frac{120V - V_C(0^-) - V_C - V_2}{10\Omega}, \text{ thus } \frac{di}{dt} = \frac{1}{5} V_2' + \frac{1}{L} V_2 = -\frac{1}{10} V_C' - \frac{1}{10} V_2 = -\frac{1}{10} \frac{di}{dt} - \frac{1}{10} V_2 = -\frac{1}{10C} \left(\frac{V_2}{5} + \right.$$

$$i_L(0^-) \Big) - \frac{1}{10} V_2 \text{ at } t=0^+, \frac{1}{5} V_2'(0^+) = -\frac{1}{10} \cdot 6A = -\frac{3}{5}, \text{ so } V_2'(0^+) = -3$$

$$5. a. \quad \text{Circuit diagram: a resistor } R=1\Omega \text{ in series with a parallel combination of a resistor } R=1\Omega \text{ and a capacitor } C. \text{ The voltage across the capacitor is } V_C(t). \quad V_C(t) = k e^{-2t} \text{ at } t=0 \quad k = V_C(0), \text{ thus } V_C = V_C(0) e^{-2t}$$

$$b. \quad V_C' + 2V_C = A \cos(\omega t + \varphi) \quad \text{let } V_C = k_1 \sin(\omega t + \varphi) + k_2 \cos(\omega t + \varphi) \text{ and plug } V_C \text{ back into the differential equation, we get } \begin{cases} k_1 = \frac{A\omega}{\omega^2 + 4} \\ k_2 = \frac{2A}{\omega^2 + 4} \end{cases}, \text{ at } t=0 \quad V_C =$$

$$C e^{-2t} + k_1 \sin \varphi + k_2 \cos \varphi = 0, \text{ thus } C = 0, \text{ thus } V_C = (-k_1 \sin \varphi - k_2 \cos \varphi) e^{-2t} + \frac{A}{\sqrt{\omega^2 + 4}} \cos(\varphi + \arctan(-\frac{\omega}{2}))$$

$$c. \text{ in steady state } e^{-2t} \text{ die out, } V_C = \frac{A}{\sqrt{\omega^2 + 4}} \cos(\varphi + \arctan(-\frac{\omega}{2}))$$

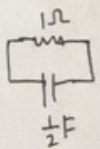
$$d. \quad V_C = A e^{j\varphi} \frac{\frac{1}{j\omega + 1}}{1 + \frac{1}{j\omega + 1}} = A e^{j\varphi} \frac{1}{j\omega + 2} \quad |V_C| = A \frac{1}{\sqrt{\omega^2 + 4}} \quad \varphi = \arctan(-\frac{\omega}{2})$$

$$\text{thus } V_C = \frac{A}{\sqrt{\omega^2 + 4}} \cos(\varphi + \arctan(-\frac{\omega}{2}))$$

$$e. \text{ plug in the numbers, we get } V_C = \frac{3}{5} e^{-2t} + \frac{1}{\sqrt{5}} \cos(t + \arctan(-\frac{1}{2}))$$

Xilai-Zhang-804796478

6. a. the circuit is second order because there are two energy storing elements.

b.  $i_R(0^+) = 10A$ $i_R(\infty) = 0A$ $i_R = 10e^{-2t}$

c. $10 - \int_0^t i dt = i + V$, $i = V'$, thus $10 - V = V' + V$, thus $V = Ce^{-2t} + 5$
at $t=0$, $V=0$, thus $C+5=0$, $C=-5$, thus $V(t) = -5e^{-2t} + 5$

d. $V(\infty) = 5V$. the final voltage is not 0 because when voltage on the two capacitors are the same, there will be no current in the circuit, and thus no more energy loss on the resistor, And at that state the circuit is steady.

8. a. $s^3 + 1 = 0 = (s+1)(s^2 - s + 1)$ $s_1 = -1$, $s_2 = \frac{1}{2} + \frac{\sqrt{3}}{2}j$, $s_3 = \frac{1}{2} - \frac{\sqrt{3}}{2}j$, thus the solution is of the form $k_1 e^{-t} + k_2 e^{\frac{t}{2}} \sin \frac{\sqrt{3}}{2} t + k_3 e^{\frac{t}{2}} \cos \frac{\sqrt{3}}{2} t + y_p(t)$. let $y_p(t) = C_1 \cos t + C_2 \sin t$,

plug in $y_p(t)$ into equation we get $\begin{cases} C_1 = \frac{1}{2} \\ C_2 = -\frac{1}{2} \end{cases}$ thus $y_p(t) = \frac{1}{2} \cos t - \frac{1}{2} \sin t$, plug in

and use three initial conditions

general solution we get $\begin{cases} k_1 = \frac{1}{6} \\ k_2 = \frac{\sqrt{3}}{3} \\ k_3 = \frac{1}{3} \end{cases}$ thus $y(t) = \frac{1}{6} e^{-t} + \frac{\sqrt{3}}{3} e^{\frac{t}{2}} \sin \frac{\sqrt{3}}{2} t + \frac{1}{3} e^{\frac{t}{2}} \cos \frac{\sqrt{3}}{2} t + \frac{1}{2} \cos t - \frac{1}{2} \sin t$

b. $s^2 + 2s + 1 = 0$ $(s+1)^2 = 0$ thus $y(t) = k_1 e^{-t} + k_2 t e^{-t}$ plug $y(t)$ back into the equation we get $\begin{cases} k_1 = 0 \\ k_2 = 1 \end{cases}$ thus $y(t) = t e^{-t} u(t)$

c. by superposition, $\begin{cases} y' + 2y = 1 & 0 \leq t < 1 \\ y' + 2y = -e^{-(t-1)} & 1 < t < \infty \end{cases}$ the $y' + 2y = 1$ gives us $y(t) = -\frac{1}{2} e^{-2t} + \frac{1}{2}$

$\frac{1}{2}$, the $y' + 2y = -e^{-(t-1)}$ has a solution of the form $k_1 e^{-(t-1)}$ and $k_1 = -1$, thus in the second part when $1 < t < \infty$, $y(t) = [C_1 e^{-2t} - e^{-(t-1)}] u(t-1)$, thus $y(t) =$

$$\begin{cases} -\frac{1}{2} e^{-2t} + \frac{1}{2} & 0 \leq t < 1 \\ [C_1 e^{-2t} - e^{-(t-1)}] u(t-1) & 1 < t < \infty \end{cases}$$

10. a. $\omega_0 = \frac{1}{\sqrt{LC}}$ $Q = \frac{1}{2} = \frac{1}{R} \sqrt{\frac{L}{C}}$ thus $L = \frac{1}{C}$, $R = \frac{2}{C}$. $L i_L' + V_C(0) + \frac{1}{C} \int_0^t i_L dt + i_L R = 0$,

thus $L i_L'' + R i_L' + \frac{i_L}{C} = 0$, thus $i_L'' + 2i_L' + i_L = 0$, let $i_L = k_1 e^{-t} + k_2 t e^{-t}$. $i_L(0) = 0$ thus

$k_1 = i_L(0)$. $L i_L'(0) + V_C(0) + R i_L(0) = 0$ $i_L'(0) = \frac{-V_C(0) - R i_L(0)}{L} = k_2 - k_1 = k_2 - i_L(0) = \frac{-V_C(0) - R i_L(0)}{L}$

thus $k_2 = \frac{-V_C(0) + (L-R) i_L(0)}{L}$, thus $i_L(t) = i_L(0) e^{-t} + \frac{-V_C(0) + (L-R) i_L(0)}{L} t e^{-t}$

b. $L i_L' + V_C(0) + \frac{1}{C} \int (i_L - \delta(t)) dt + (i_L - \delta(t)) R = 0$, thus $i_L'' + 2i_L' + i_L = \delta(t) + 2\delta'(t)$.

let $i_L = k_1 e^{-t} u(t) + k_2 t e^{-t} u(t)$ and plug i_L back into the equation, we get $\begin{cases} k_1 = 2 \\ k_2 = -1 \end{cases}$

thus $i_L(t) = 2e^{-t} u(t) - t e^{-t} u(t)$

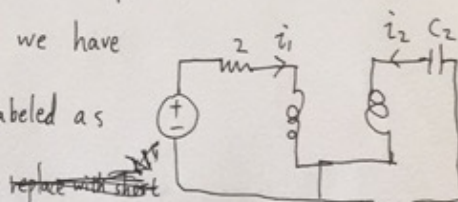
c. $i_L(0^+) = 2$ $i_L'(0^+) = -2e^{-0} - (e^{-0} - 0) = -3$

12. a. in the V_{TH} topology, $V_{TH} = \frac{2jM}{2+2j} V_s$ in phasor domain.

in the short circuit current I_{TH} topology, we have

$\begin{cases} (2+2j) i_1 + 2Mj I_2 = V_s \\ (8j + \frac{1}{2jc}) i_2 + 2Mj i_1 = 0 \end{cases}$

where i_1, i_2 is labeled as



the two equations are transformed into matrices: $\begin{bmatrix} 2+2j & 2Mj \\ 2Mj & 8j + \frac{1}{2jc} \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} V_s \\ 0 \end{bmatrix}$, and the

solution is $I_2 = \frac{-\frac{1}{2} Mj V_s}{4j + \frac{1}{4jc} - 4 + \frac{1}{4c} + M^2}$, thus $Z_{TH} = \frac{V_{TH}}{-i_2} = \frac{\frac{2jM}{2+2j} V_s}{\frac{-\frac{1}{2} Mj V_s}{4j + \frac{1}{4jc} - 4 + \frac{1}{4c} + M^2}} =$

$M^2 + j(8 - M^2 - \frac{1}{2C_2})$

b. maximum power is achieved when $Z_{TH} = Z_L^*$, thus $M^2 + (8 - M^2 - \frac{1}{2C_2})j = (1-j)^* = 1+j$ solving this equation we get $\begin{cases} M=1 \\ C_2 = \frac{1}{12} \end{cases}$

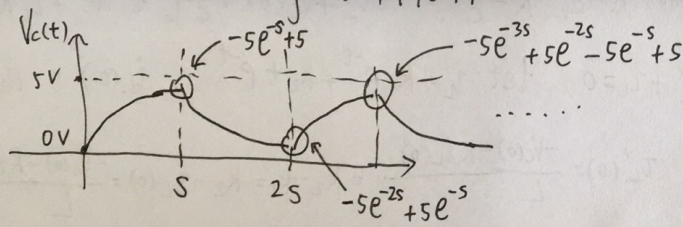
2. a. the circuit is transformed into where V is a square wave of amplitude 5V

when switch is off, $V' + V = 5$, thus $V(t) = (V_C(0) - 5)e^{-t} + 5$. when switch is on, $V' - V = 0$, thus

$V(t) = V_C(0)e^{-t}$ thus the plot $V_C(t)$ is on the next page,

$V_C(0)e^{-t}$ where $V_C(0)$ is the value at the beginning of period,

Xilai-Zhang-804796478



b. in steady state, $V_c(t)$ approaches 5V in the first ^{part of the} period, and becomes 0V in the second part of the period.

c. $\frac{1}{2S} \left[\int_0^S (V_c(t) - 5)e^{-t/S} dt + \int_0^S V_c(t) e^{-t/S} dt \right]$, where $V_c(t)$ is the voltage at the beginning of each S period. In steady state, $\frac{1}{2S} \left[\int_0^S (-5e^{-t/S} + 5) dt + \int_0^S 5e^{-t/S} dt \right] = \frac{1}{2S} \cdot 5S = \frac{5}{2}$