

Essentials of MOSFETs

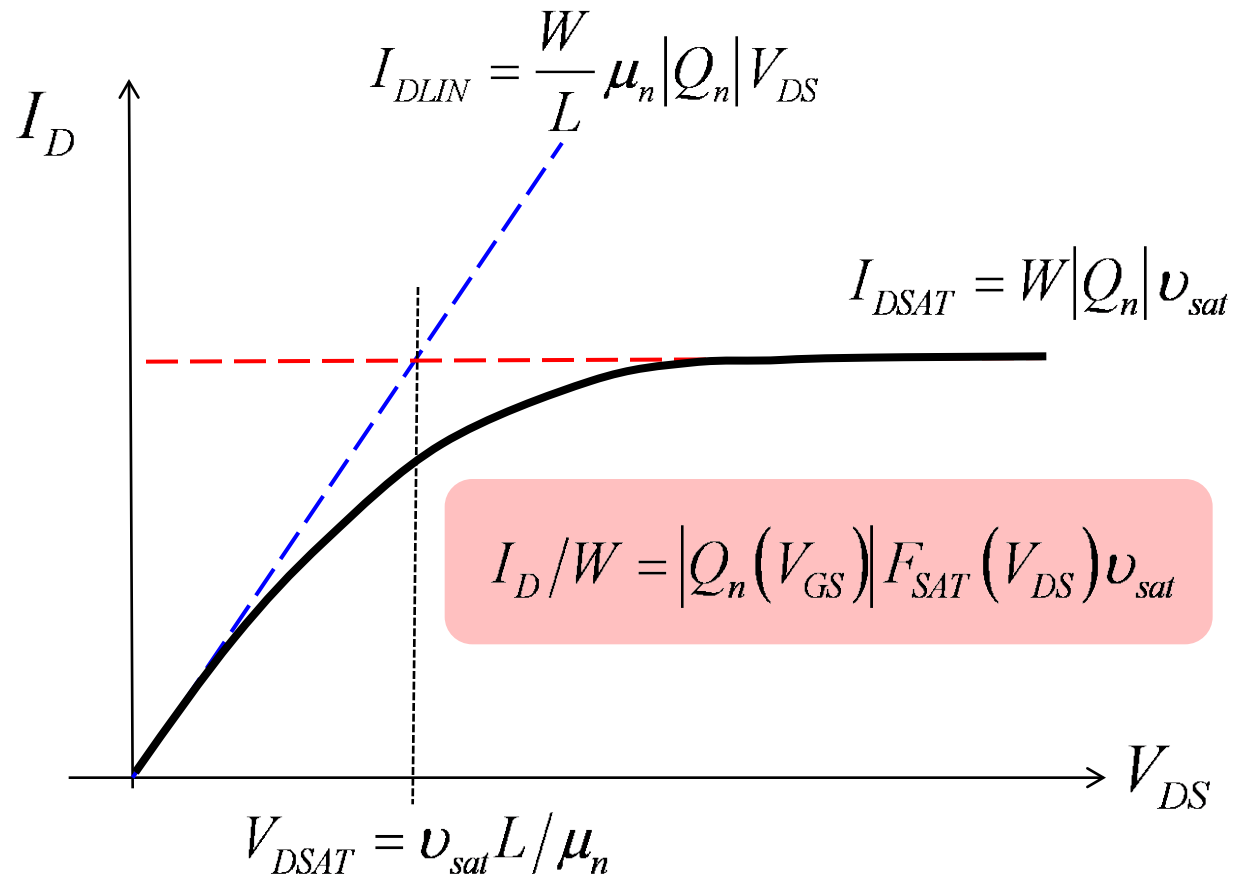
Unit 3: MOS Electrostatics

Lecture 3.9: The VS Model Revisited

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Level 0 VS Model



Level 0 VS model

$$1) \quad I_D/W = |Q_n(V_{GS})| \langle v_x(V_{DS}) \rangle$$

$$2) \quad Q_n(V_{GS}) = -C_{ox}(V_{GS} - V_T) \quad (V_{GS} > V_T)$$

$$V_T = V_{T0} - \delta V_{DS}$$

$$Q_n(V_{GS}) = 0 \quad (V_{GS} \leq V_T)$$

$$3) \quad \langle v(V_{DS}) \rangle = F_{SAT}(V_{DS}) v_{sat}$$

$$4) \quad F_{SAT}(V_{DS}) = \frac{V_{DS}/V_{DSAT}}{\left[1 + (V_{DS}/V_{DSAT})^\beta\right]^{1/\beta}}$$

$$5) \quad V_{DSAT} = v_{sat} L / \mu_n$$

There are only 8 device-specific parameters in this model:

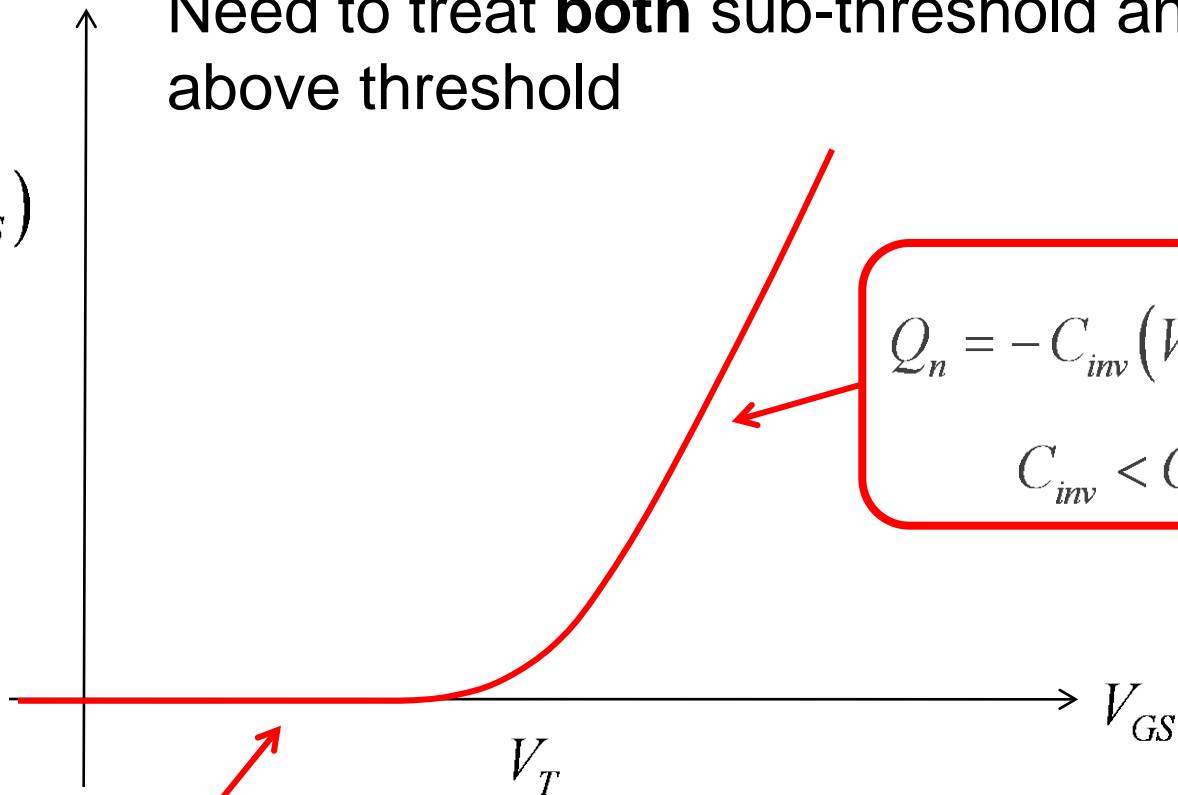
$$C_{ox}, V_{T0}, \delta, v_{sat}, \mu_n, L$$

$$R_{SD} = R_S + R_D, \beta$$

Improving the VS model: inversion charge

Need to treat **both** sub-threshold and above threshold

$Q_n(V_{GS})$

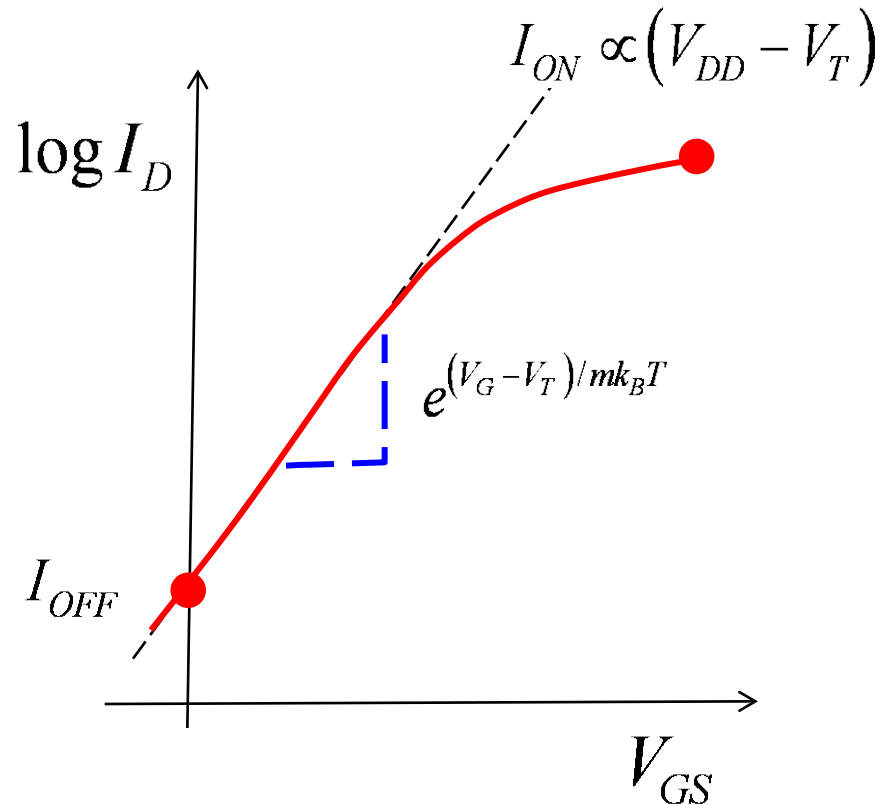


$$Q_n = -C_{inv}(V_{GS} - V_T)$$

$$C_{inv} < C_{ox}$$

$$Q_n(V_{GS}) = -(m-1)C_{ox}\left(\frac{k_B T}{q}\right)e^{q(V_{GS}-V_T)/mk_B T}$$

But first: the subthreshold current



- 1) subthreshold swing
- 2) off-current
- 3) on-current

$$Q_n(V_{GS}) \propto e^{q(V_{GS} - V_T)/mk_B T}$$

$$I_D \propto Q_n(V_{GS})$$

$$I_D \propto e^{q(V_{GS} - V_T)/mk_B T}$$

Subthreshold swing

$$I_D \propto e^{q\psi_S/k_B T}$$

$$\psi_S = V_{GS}/m$$

$$m \geq 1$$

$$\ln I_D = \frac{\psi_S}{(k_B T/q)} + c$$

$$\log_{10} I_D = \frac{\psi_S}{2.3(k_B T/q)} + \frac{c}{2.3}$$

$$\frac{\partial(\log_{10} I_D)}{\partial V_{GS}} = \frac{\partial(\log_{10} I_D)}{\partial \psi_S} \times \frac{\partial \psi_S}{\partial V_{GS}} = \frac{1}{2.3(k_B T/q)} \times \frac{1}{m} \quad \frac{\text{Decades of } I_D}{\text{Volts of } V_{GS}}$$

$$SS = \left(\frac{\partial(\log_{10} I_D)}{\partial V_{GS}} \right)^{-1} = 2.3m(k_B T/q) \frac{\text{mV}}{\text{dec}}$$

Subthreshold swing

$$S = 2.3m(k_B T / q) \frac{\text{mV}}{\text{dec}} \quad m \geq 1$$

$m \approx 1.1 - 1.4$ typically

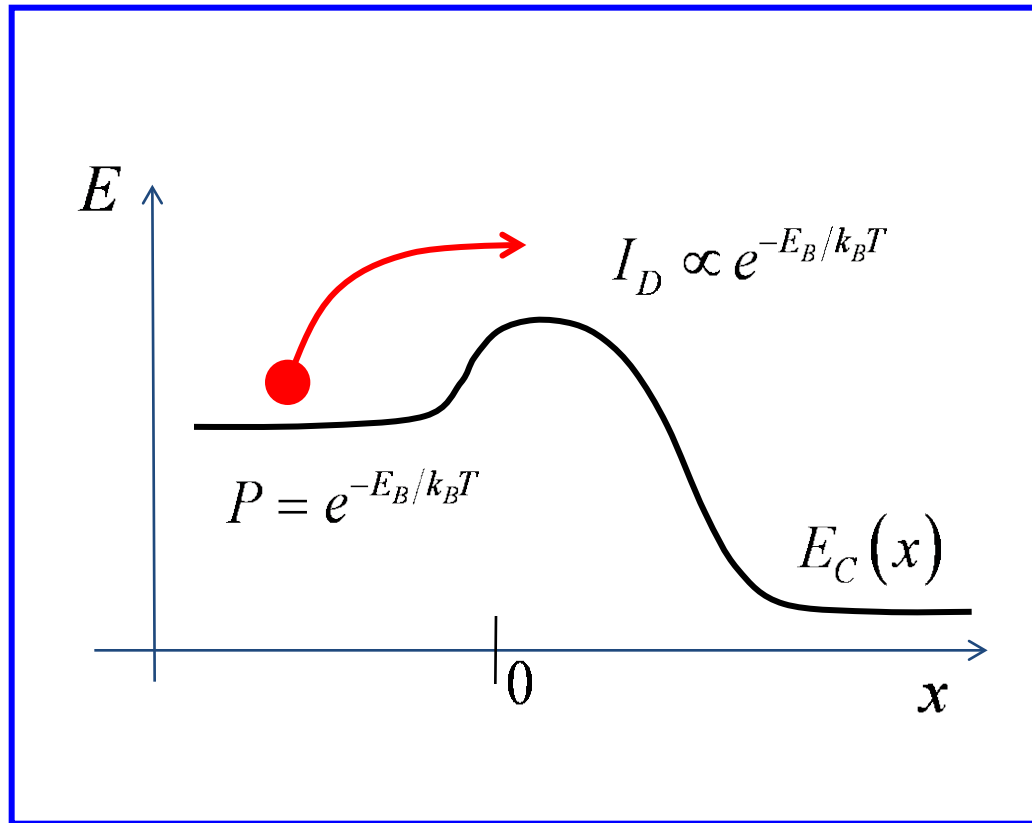
$$SS > 60 \frac{\text{mV}}{\text{dec}} \quad (T = 300\text{K})$$

$$SS < 100 \frac{\text{mV}}{\text{dec}} \quad (\text{typically})$$

Why is a small SS important?

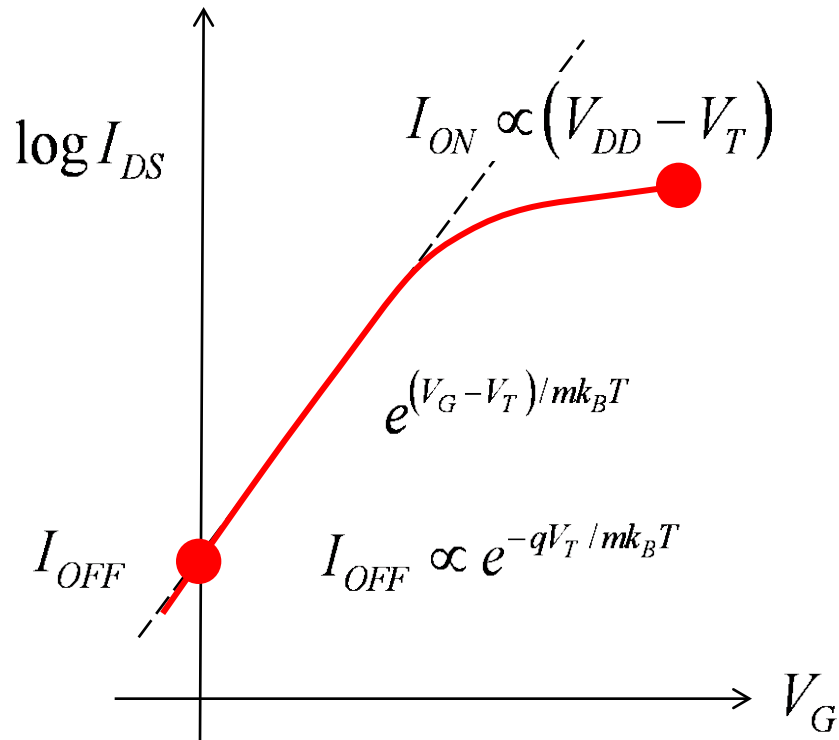
$$P_D \propto V_{DD}^2$$

Why is $S > 60$ mV/decade?

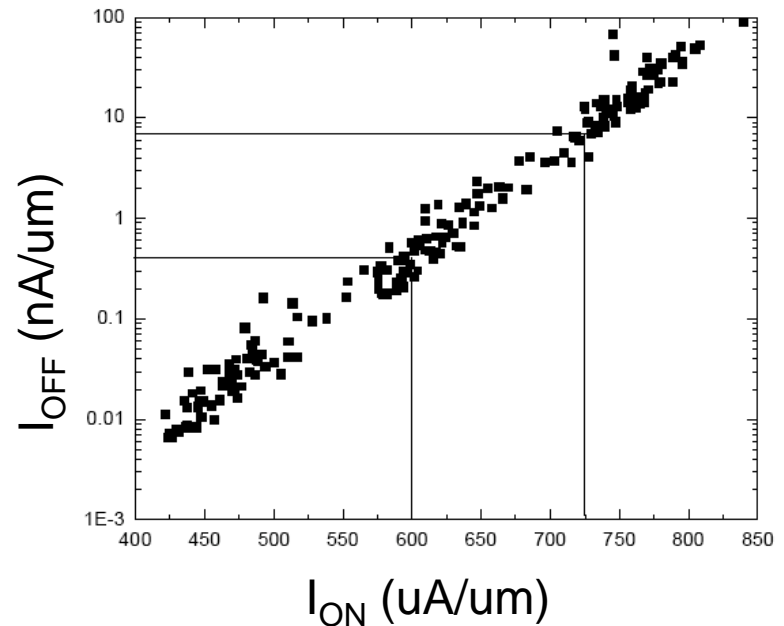


injection of thermal carriers over a barrier

Relation between I_{OFF} and I_{ON}



$$\log_{10} I_{OFF} = c_1 + c_2 I_{ON}$$



A. Steegen, et al., "65nm CMOS Technology for low power applications," Intern. Electron Dev. Meeting, Dec. 2005.

Back to the VS model

Now let us return to the question at hand:

“How do we describe $Q_n(V_{GS})$ **continuously** below and above threshold?”

Empirical treatment of inversion charge

$$Q_n(V_{GS}) = -C_{inv} m (k_B T / q) \ln \left(1 + e^{q(V_{GS} - V_T) / m k_B T} \right)$$

$$Q_n(V_{GS}) = -(m-1)C_{ox} (k_B T / q) e^{q(V_{GS} - V_T) / m k_B T}$$

----- correct result -----

G. T. Wright, "Threshold modelling of MOSFETs for CAD of CMOS VLSI," *Electron Lett.*, **21**, pp. 223–224, Mar. 1985.

Empirical treatment of inversion charge

$$Q_n(V_{GS}) = -C_{inv} m (k_B T / q) \ln \left(1 + e^{q(V_{GS} - V_T) / m k_B T} \right)$$

$$V_{GS} > V_T :$$

$$\ln(1 + x) \approx \ln(x)$$

$$Q_n(V_{GS}) \approx -C_{inv} (V_{GS} - V_T)$$

$$Q_n(V_{GS}) = -C_{inv} (V_{GS} - V_T)$$

correct

G. T. Wright, "Threshold modelling of MOSFETs for CAD of CMOS VLSI," *Electron Lett.*, **21**, pp. 223–224, Mar. 1985.

Empirical treatment of inversion charge

expression used in the MIT VS Model

$$Q_n(V_{GS}, V_{DS}) = -C_{inv} m (k_B T / q) \ln \left(1 + e^{q(V_{GS} - V_T + \alpha(k_B T_L / q) F_f) / m k_B T} \right)$$

$$V_T = V_{T0} - \delta V_{DS}$$

DIBL

different V_T 's in strong and weak inversion

Ali Khakifirooz, Osama M. Nayfeh, and Dimitri Antoniadis, "A Simple Semi-empirical Short-Channel MOSFET Current–Voltage Model Continuous Across All Regions of Operation and Employing Only Physical Parameters," *IEEE Trans. Electron Devices*, **56**, pp. 1674-1680, 2009.

Level 1 VS model

$$1) \quad I_D/W = |Q_n(V_{GS}, V_{DS})| \langle v_x(V_{DS}) \rangle$$

$$2) \quad Q_n(V_{GS}, V_{DS}) = -C_{inv} m (k_B T / q) \ln \left(1 + e^{q(V_{GS} - V_T + \alpha(k_B T_L / q) F_f) / m k_B T} \right)$$

$$V_T = V_{T0} - \delta V_{DS}$$

$$3) \quad \langle v_x(V_{DS}) \rangle = F_{SAT}(V_{DS}) v_{sat}$$

$$4) \quad F_{SAT}(V_{DS}) = \frac{V_{DS} / V_{DSAT}}{\left[1 + (V_{DS} / V_{DSAT})^\beta \right]^{1/\beta}}$$

$$5) \quad V_{DSAT} = \frac{v_{sat} L}{\mu_n}$$

Only 10 device-specific parameters in this model:

$$C_{inv}, V_{T0}, \delta, m, v_{sat}, \mu_n,$$

$$L, R_{SD} = R_S + R_D,$$

$$\alpha, \beta$$

Discussion

With this extension (subthreshold to above threshold conduction), the VS model accurately describes modern transistors providing:

- 1) The high-field saturation velocity is viewed as an empirical, fitting parameter.
- 2) The mobility of carriers in the inversion is viewed as an empirical, fitting parameter.
- 3) **But** we will see later, that these empirical parameters can be given a clear, physical interpretation.

Download the MVS model at: <https://nanohub.org/publications/15>

Summary

- 1) Using a semi-empirical expression for Q_n , we have extended the VS model to treat subthreshold to above threshold.
- 2) Excellent fits to measured transistor IV characteristics generally result.
- 3) But the physical understanding of the mobility and saturation velocity *at the nanoscale* needs to be clarified.