

EECS240 – Spring 2010

Lecture 7: Noise and Feedback



Elad Alon
Dept. of EECS

Noise and Feedback

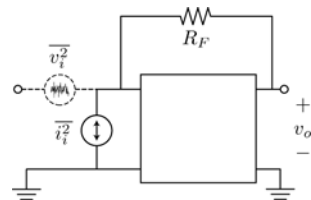
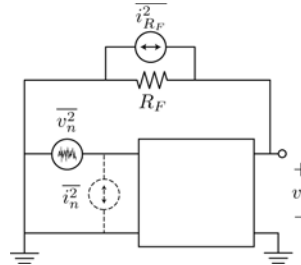
- **Ideal feedback:**
 - No increase of input referred noise
 - No decrease of SNR at output
- **Practical feedback: increased noise**
 - Noise from feedback network
 - Noise gain from elements outside feedback loop

Real Feedback

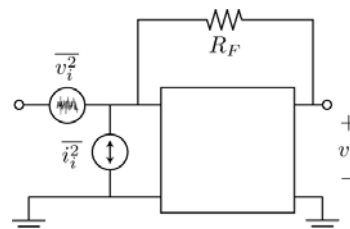
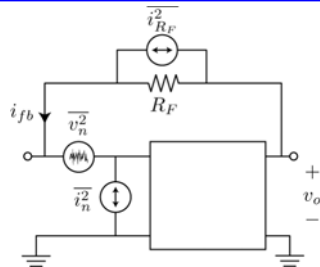
- Conceptually identical to standard two port calculations

- Use $R_s = 0$ to find $v_{i,eq}^2$
 - $R_s = \infty$ to find $i_{i,eq}^2$

- Calculations get tedious...

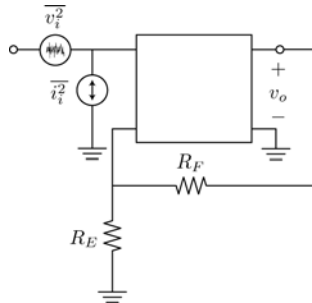


Practical Feedback Analysis



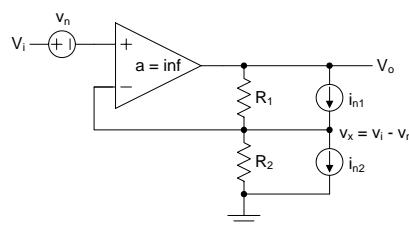
- Quick approximation method:**
 - Consider loading of feedback network on the input
 - Add a noise source associated with this element.
- Example: shunt feedback**
 - Loading at input is $R_F \rightarrow i_i^2 = i_n^2 + 4kT\Delta f/R_F$

Example #2: Series-Shunt Feedback



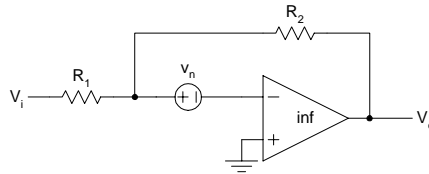
- Loading is $R_F || R_E$
- So, noise voltage becomes:
 - $v_i^2 = v_n^2 + 4kT(R_F || R_E)\Delta f$

Implications: Non-Inverting Amp



- Minimum power from feedback \rightarrow large $R_1 + R_2$
- Example:
 - $A_v = 10$, $R_2 = 100k\Omega$, $R_1 = R_2(A_{v0} - 1) = 900k\Omega$
 - $v_{nfb}^2 = 40nV/\sqrt{Hz}$ (very high)
- Only way to lower noise is increase power...

Example: Inverting Amplifier



- Ignoring noise from R_1 , R_2 :

$$v_o = -v_i \underbrace{\frac{R_2}{R_1}}_{-A_{v0}} + v_n \left(1 + \frac{R_2}{R_1} \right) = -v_i \underbrace{\frac{R_2}{R_1}}_{-A_{v0}} + v_n \frac{R_1 + R_2}{R_1}$$

$$\overline{v_{ieq}^2} = \overline{v_n^2} \left(\frac{R_1 + R_2}{R_1} \frac{R_1}{R_2} \right)^2 = \overline{v_n^2} \left(\frac{R_1 + R_2}{R_2} \right)^2 = \overline{v_n^2} \left(1 + \frac{1}{|A_{v0}|} \right)^2$$

- “Ideal” feedback, why is $v_{ieq}^2 > v_n^2$?

Example

Example
