



Xi'an Jiaotong-Liverpool University  
西交利物浦大學

# EEE220 Instrumentation and Control System

*2018-19 Semester 2*

Dr. Qing Liu

Email: [qing.liu@xjtlu.edu.cn](mailto:qing.liu@xjtlu.edu.cn)

Office: EE516

Department of Electrical and Electronic Engineering

*21 March, 2019*

---

# Lecture 10

# Outline

---

## **Control Systems:**

### **Mathematical Models of Systems: part 2/2**

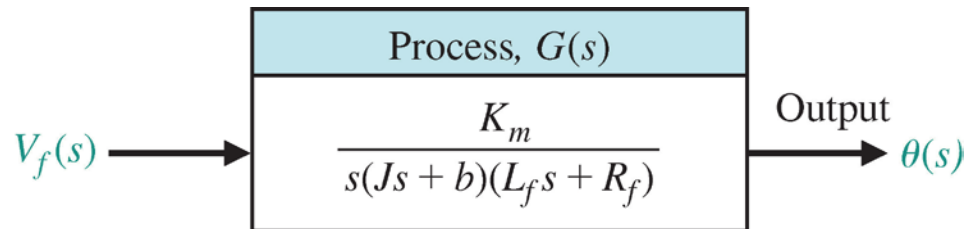
- ☐ Differential Equations of Physical Systems
- ☐ The Transfer Function of Linear Systems
- ☐ Block Diagram Models
- ☐ Signal-Flow Graph Models
- ☐ Simulation Tool

# Block Diagram Models

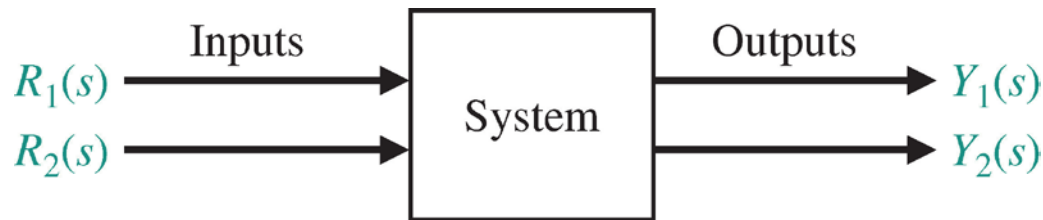
---

- The dynamic systems are typically represented mathematically by a set of simultaneous **differential equations**; the **Laplace transform** reduces the problem to the solution of a set of linear algebraic equations;
- **Transfer functions** are introduced to represent the relationship between the input and output variables of the system; it is an important relation in control engineering;
- **Block diagram** is a graphic way to represent this important cause-and-effect relationship. It consists of **unidirectional**, operational blocks that represent the transfer function of the systems of interest.

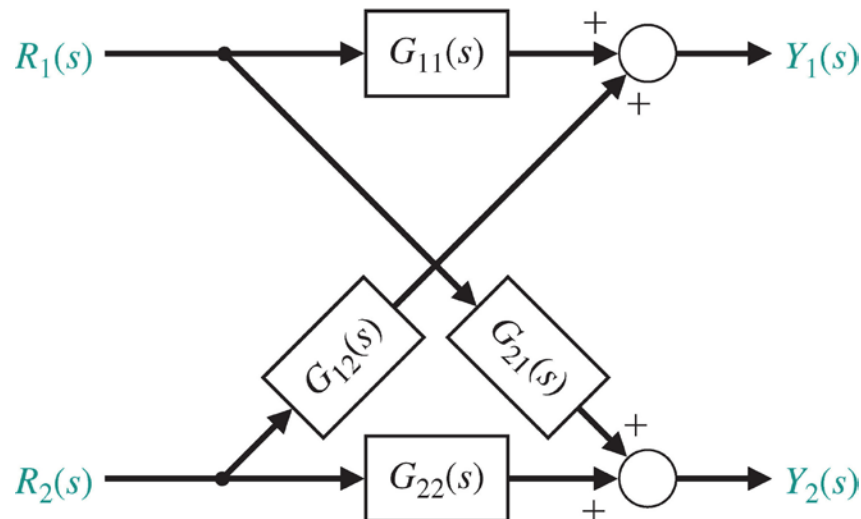
# Examples



(a) Block diagram of a DC motor.

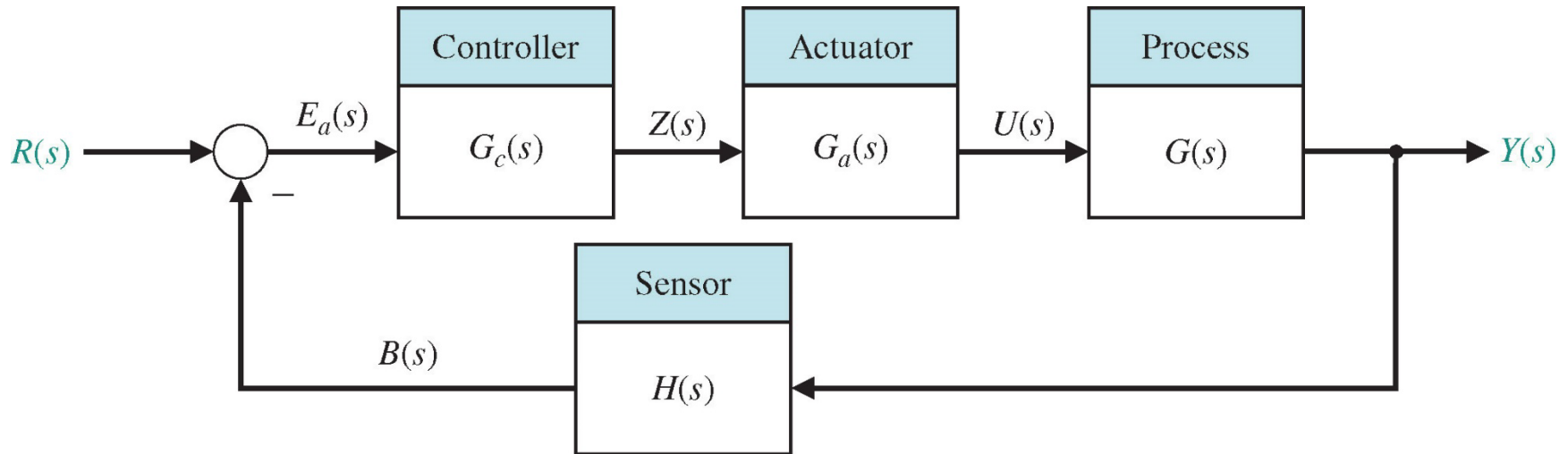


(b) General block representation of two-input, two-output system.



(c) Block diagram of a two-input, two-output *interconnected* system.

# Transfer Function of A Negative Feedback Control System



The error signal or actuating signal  $E_a(s)$  is

$$E_a(s) = R(s) - H(s)Y(s)$$

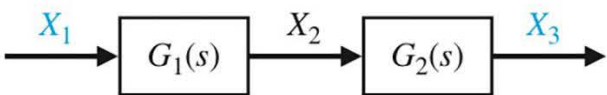
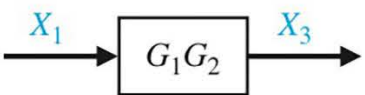
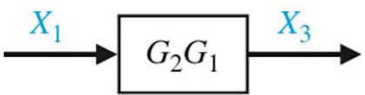
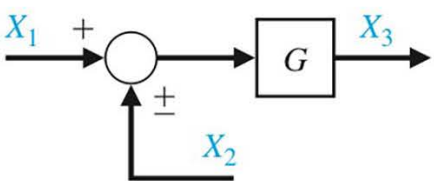
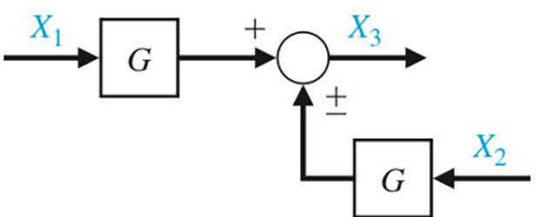
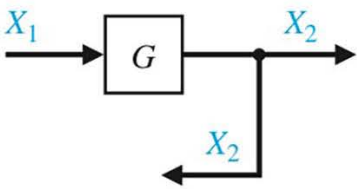
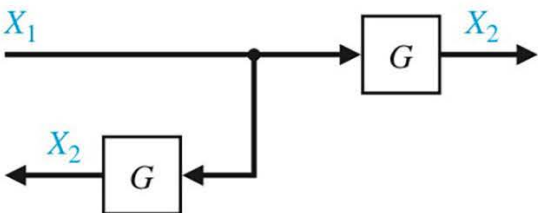
The output signal  $Y(s)$  can be represented as

$$Y(s) = E_a(s)G_c(s)G_a(s)G(s)$$

Combining the above two equations, the transfer function of the system is

$$\frac{Y(s)}{R(s)} = \frac{G_c(s)G_a(s)G(s)}{1 + G_c(s)G_a(s)G(s)H(s)}$$

# Block Diagram Transformations

Transformation	Original Diagram	Equivalent Diagram
1. Combining blocks in cascade		 or 
2. Moving a summing point behind a block		
3. Moving a pickoff point ahead of a block		

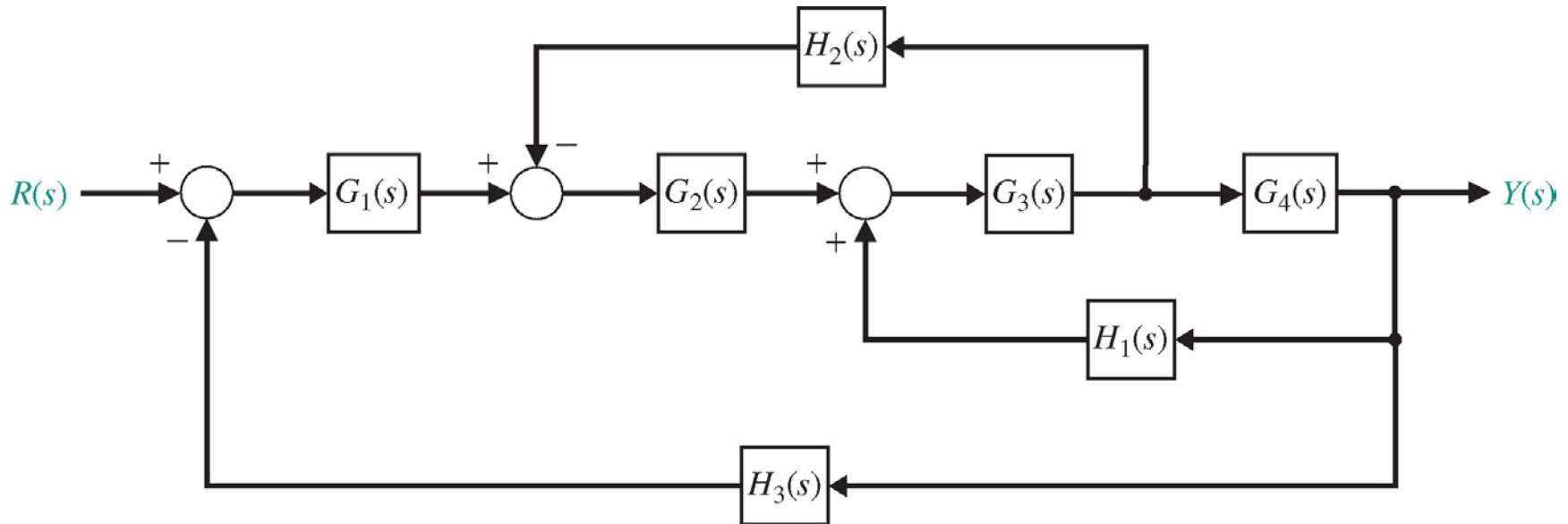
# Block Diagram Transformations (cont'd)

Transformation	Original Diagram	Equivalent Diagram
4. Moving a pickoff point behind a block		
5. Moving a summing point ahead of a block		
6. Eliminating a feedback loop		

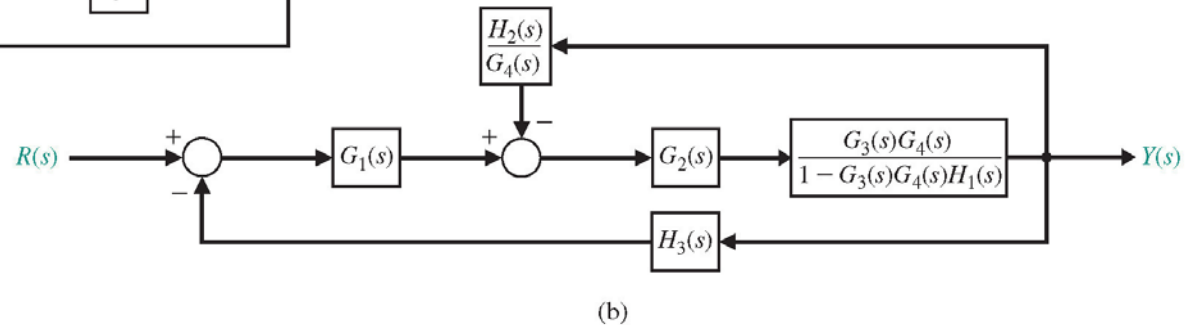
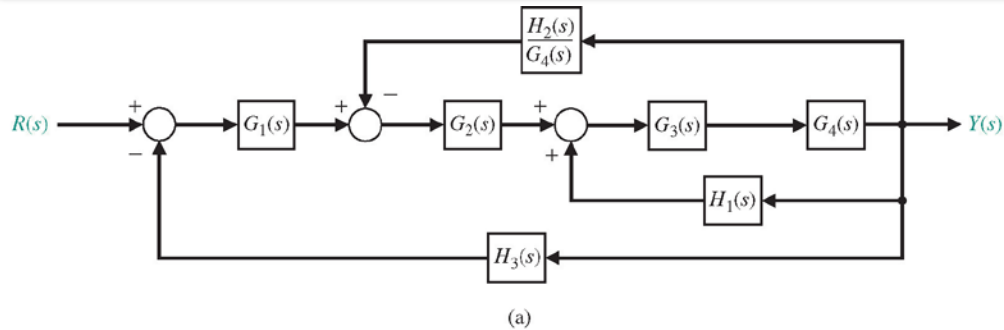


# Reduction of the Block Diagram

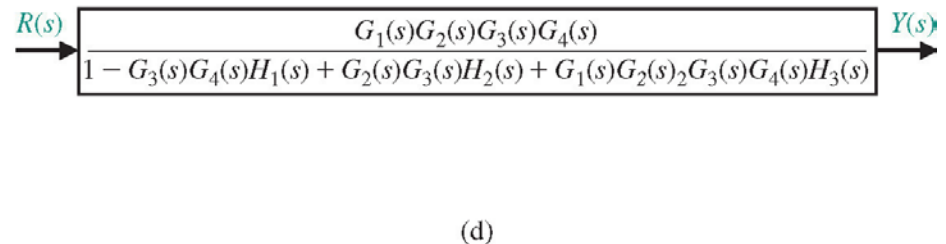
## A multi-loop feedback control system



Please note the feedback  $H_1G_3G_4$  is a **positive** feedback; while  $G_2G_3H_2$  and  $G_1G_2G_3G_4H_3$  are **negative** feedback loops.



- Step 1: move the pickoff point between G3 & G4 after G4 (from original diagram to (a));
- Step 2. eliminate loop 1 ( $G_3G_4H_1$ ) (from (a) to (b));
- Step 3. eliminate loop 2 (from (b) to (c));
- Step 4. eliminate loop 3 (from (c) to (d)).

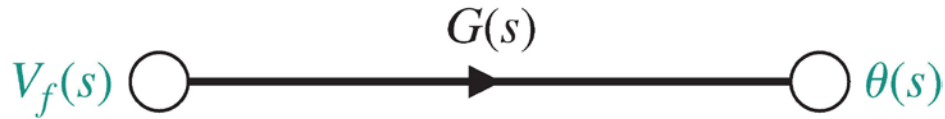


# Signal-Flow Graph Models

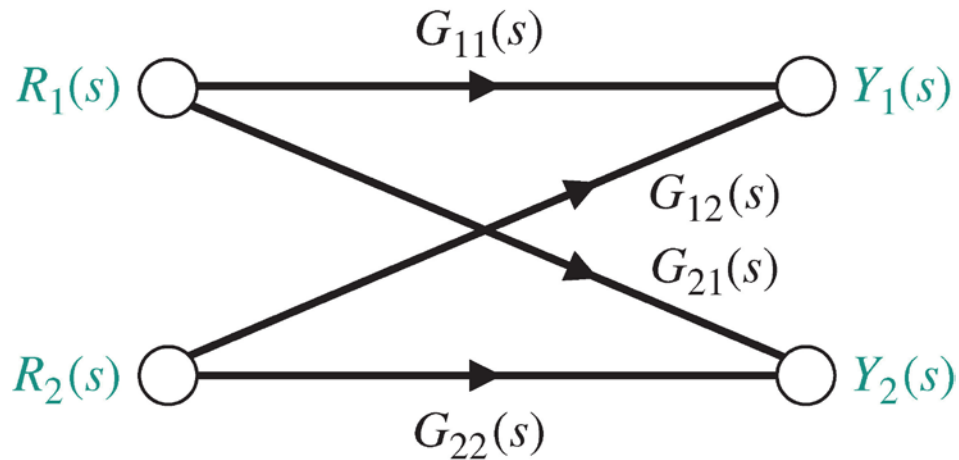
---

- A **signal-flow graph** is a diagram consisting of **nodes** that are connected by several **directed branches** and is a graphic representation of a set of linear relations;
- Signal-flow graph is particularly useful for feedback control systems because feedback theory is primarily concerned with the flow and processing of signals in the system;
- The basic element of a signal-flow graph is a unidirectional path segment called a **branch**, which relates the dependency of input and an output variable in a manner equivalent to a **block** of a block diagram.
- For complex systems, the block diagram method can become difficult to complete. By using the signal-flow graph model, the reduction procedure (used in the block diagram method) is not necessary to determine the relationship between system variables.

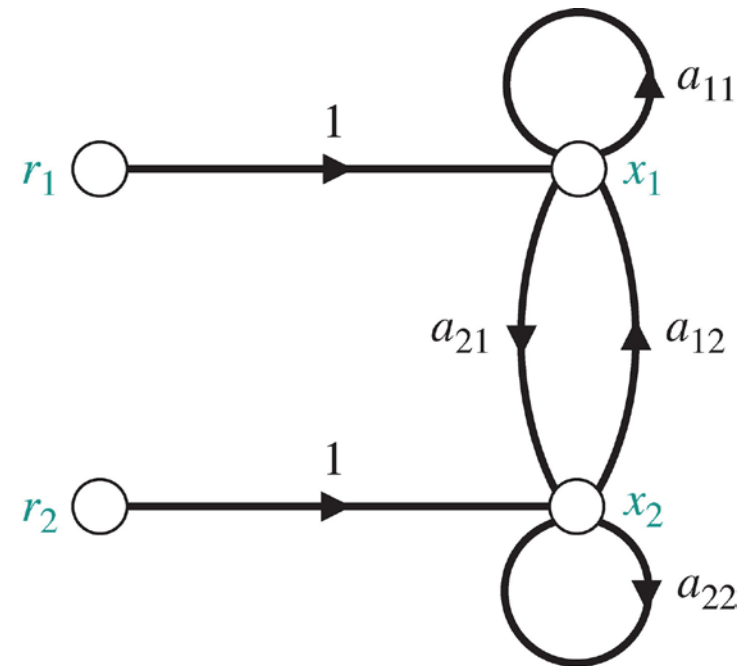
# Examples



(a) Signal-flow graph of a DC motor.



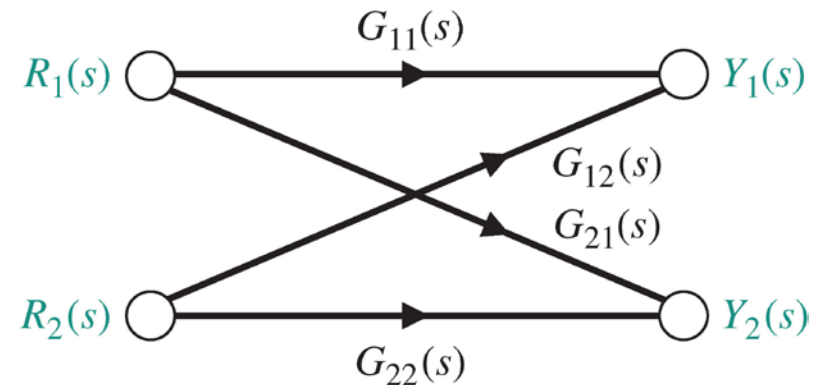
(b) Signal-flow graph of two-input, two-output interconnected system.



(c) Signal-flow graph of two algebraic equations.

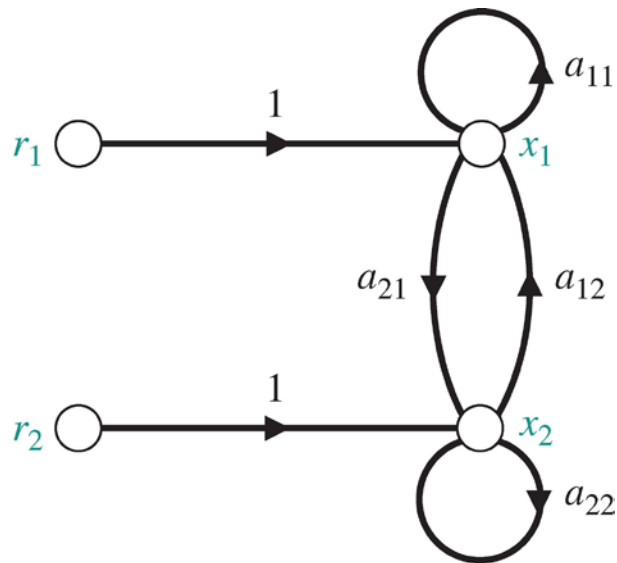
# Basic Concepts

- **Nodes**: are the input and output points or junctions; all branches leaving a node will pass the nodal signal to the output node of each branch (unidirectional); the summation of all signals entering a node is equal to the node variable;
- A **Path**: is a branch or a continuous sequence of branches that can be traversed from one node (signal) to another node (signal);
- A **Loop**: is a closed path that originates and terminates on the same node, with no node been met twice along the path;
- Two loops are said to be **nontouching** if they don't have a common node; two touching loops share one or more common nodes;



*Signal-flow graph of two-input, two-output interconnected system.*

# Determine Transfer Function from the Graph



$$a_{11}x_1 + a_{12}x_2 + r_1 = x_1$$

$$a_{21}x_1 + a_{22}x_2 + r_2 = x_2.$$



$$x_1(1 - a_{11}) + x_2(-a_{12}) = r_1,$$



$$x_1 = \frac{(1 - a_{22})r_1 + a_{12}r_2}{(1 - a_{11})(1 - a_{22}) - a_{12}a_{21}} = \frac{1 - a_{22}}{\Delta}r_1 + \frac{a_{12}}{\Delta}r_2,$$

$$x_2 = \frac{(1 - a_{11})r_2 + a_{21}r_1}{(1 - a_{11})(1 - a_{22}) - a_{12}a_{21}} = \frac{1 - a_{11}}{\Delta}r_2 + \frac{a_{21}}{\Delta}r_1.$$

Where:  $\Delta = (1 - a_{11})(1 - a_{22}) - a_{12}a_{21} = 1 - a_{11} - a_{22} + a_{11}a_{22} - a_{12}a_{21}.$

# Mason's Signal-flow Gain Formula

In general, the linear dependence  $T_{ij}(s)$  between the independent variable  $x_i$  (often called the input variable) and a dependent variable  $x_j$  is given by:

$$T_{ij} = \frac{\sum_k P_{ijk} \Delta_{ijk}}{\Delta}$$

The summation is taken over all possible  $k$  paths from  $x_i$  to  $x_j$ .

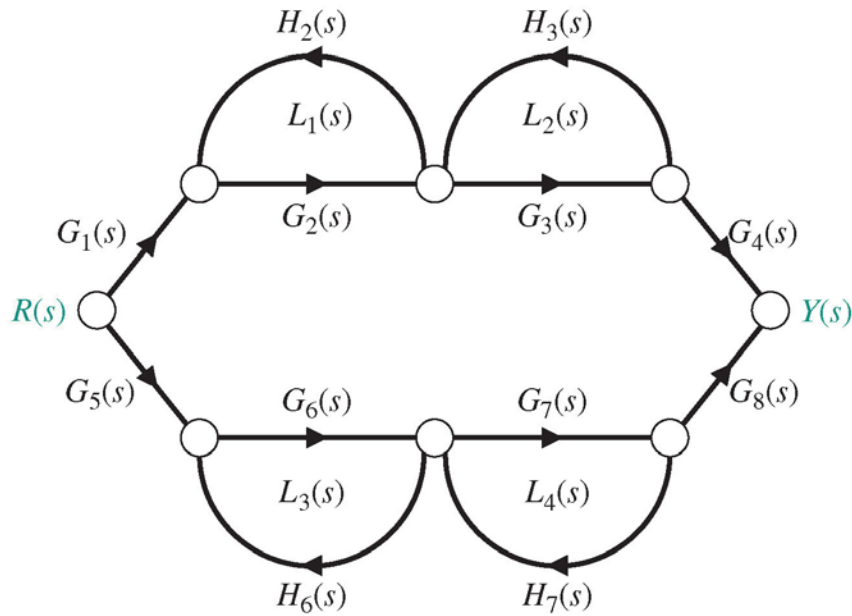
- $P_{ijk}$ : is the path gain defined as the product of gains of the branches of the path, traversed in the direction of the arrows **with no node encountered more than once**;
- $\Delta_{ijk}$ : cofactor, is the determinant **with the loops touching the  $k$ th path removed**.
- $\Delta$ : the determinant, is:

$$\Delta = 1 - \sum_{n=1}^N L_n + \sum_{\substack{n, m \\ \text{nontouching}}} L_n L_m - \sum_{\substack{n, m, p \\ \text{nontouching}}} L_n L_m L_p + \dots,$$

Where  $L_q$  equals the value of  $q$ th loop transmittance, therefore:

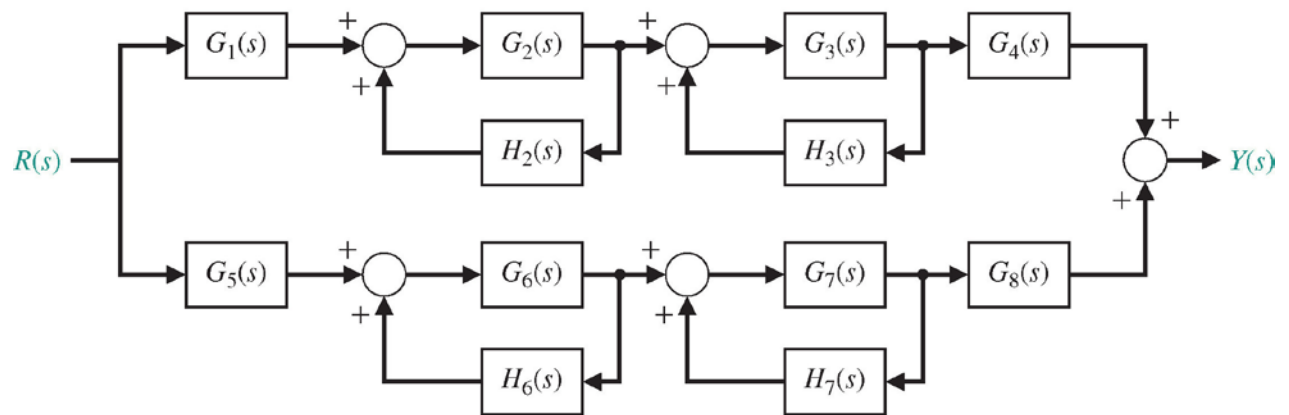
$$\begin{aligned} \Delta = & 1 - (\text{sum of all different loop gains}) \\ & + (\text{sum of the gain products of all combinations of two nontouching loops}) \\ & - (\text{sum of the gain products of all combinations of three nontouching loops}) \\ & + \dots \end{aligned}$$

# Mason's Rule: Application



**How to obtain  
Transfer Function between  
Input  $R(s)$  and Output  $Y(s)$ ?**

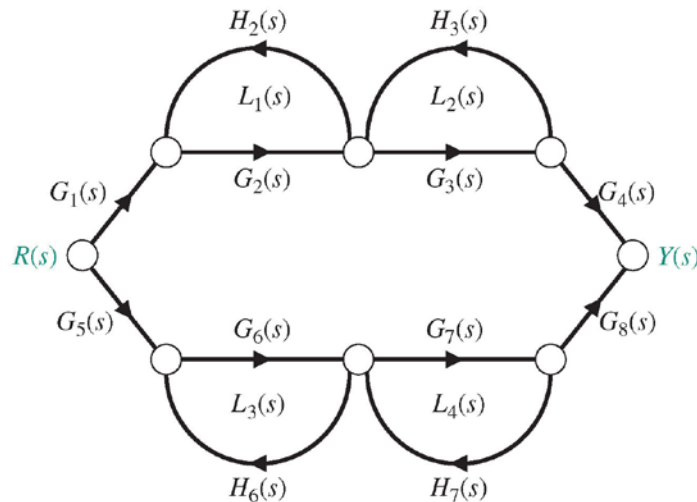
(a)



(b)



# Example 10.1



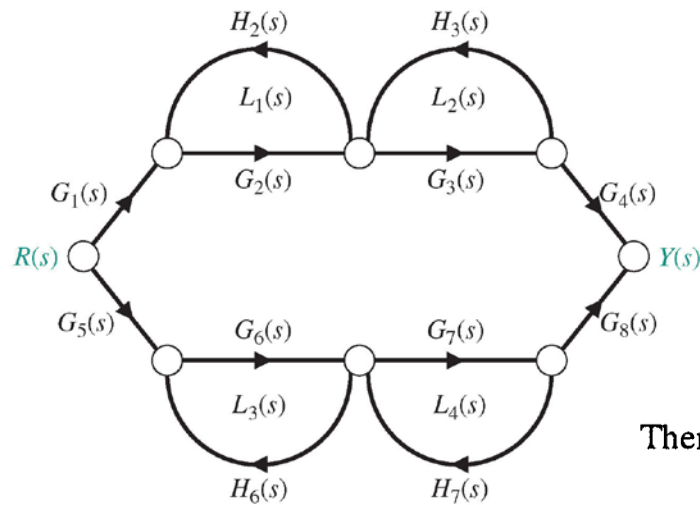
$$T_{ij} = \frac{\sum_k P_{ijk} \Delta_{ijk}}{\Delta}$$

1. How many possible paths?  $P_1 = G_1 G_2 G_3 G_4$  (path 1) and  $P_2 = G_5 G_6 G_7 G_8$  (path 2).

2. How many loops? There are four self-loops:  
 $L_1 = G_2 H_2$ ,  $L_2 = H_3 G_3$ ,  $L_3 = G_6 H_6$ , and  $L_4 = G_7 H_7$ .

3. How many groups of 2 nontouching loops? Loops  $L_1$  and  $L_2$  do not touch  $L_3$  and  $L_4$ .

4. How many groups of 3 nontouching loops? None.



$$T_{ij} = \frac{\sum_k P_{ijk} \Delta_{ijk}}{\Delta}$$

There are four self-loops:

$$L_1 = G_2 H_2, \quad L_2 = H_3 G_3, \quad L_3 = G_6 H_6, \quad \text{and} \quad L_4 = G_7 H_7.$$

Loops  $L_1$  and  $L_2$  do not touch  $L_3$  and  $L_4$ . Therefore, the determinant is

$$\Delta = 1 - (L_1 + L_2 + L_3 + L_4) + (L_1 L_3 + L_1 L_4 + L_2 L_3 + L_2 L_4).$$

The cofactor of the determinant along path 1 is evaluated by removing the loops that touch path 1 from  $\Delta$ . Hence, we have

$$L_1 = L_2 = 0 \quad \text{and} \quad \Delta_1 = 1 - (L_3 + L_4).$$

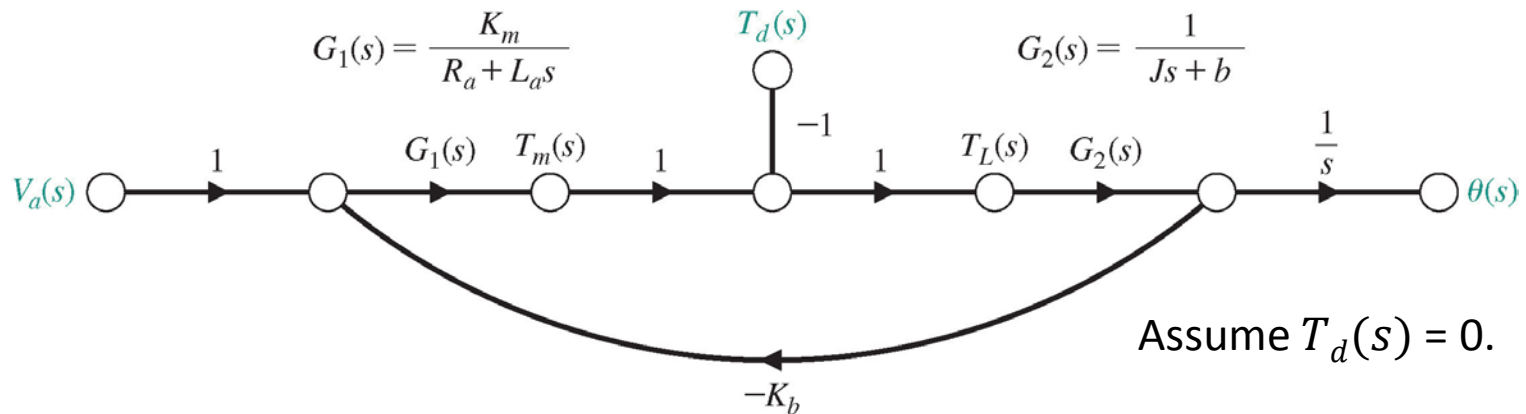
Similarly, the cofactor for path 2 is

$$\Delta_2 = 1 - (L_1 + L_2).$$

Therefore, the transfer function of the system is

$$\begin{aligned} \frac{Y(s)}{R(s)} = T(s) &= \frac{P_1 \Delta_1 + P_2 \Delta_2}{\Delta} \\ &= \frac{G_1 G_2 G_3 G_4 (1 - L_3 - L_4) + G_5 G_6 G_7 G_8 (1 - L_1 - L_2)}{1 - L_1 - L_2 - L_3 - L_4 + L_1 L_3 + L_1 L_4 + L_2 L_3 + L_2 L_4} \end{aligned}$$

# Example 10.2



There is only 1 forward path, which touches one loop:

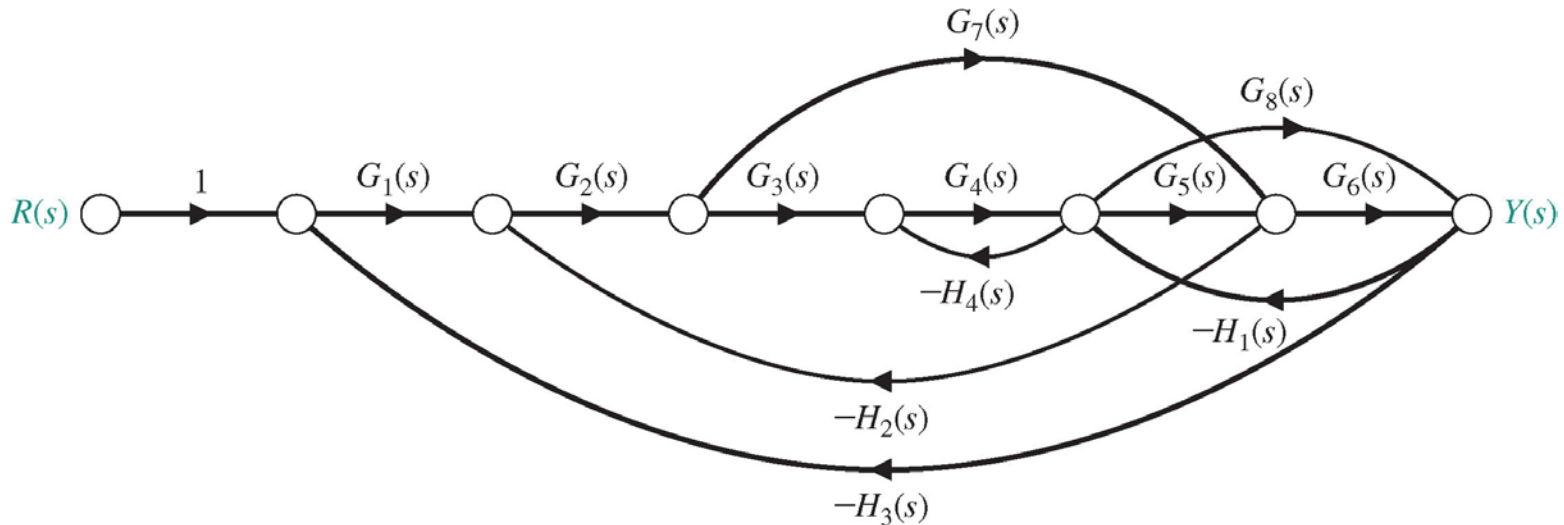
$$P_1(s) = \frac{1}{s} G_1(s) G_2(s) \quad \text{and} \quad L_1(s) = -K_b G_1(s) G_2(s).$$

Therefore, the transfer function is

$$T(s) = \frac{P_1(s)}{1 - L_1(s)} = \frac{(1/s) G_1(s) G_2(s)}{1 + K_b G_1(s) G_2(s)} = \frac{K_m}{s[(R_a + L_a s)(J s + b) + K_b K_m]}$$

# Example 10.3

Consider a reasonably complex system that would be difficult to reduce by block diagram techniques:



$$P_1 = G_1G_2G_3G_4G_5G_6, \quad P_2 = G_1G_2G_7G_6, \quad \text{and} \quad P_3 = G_1G_2G_3G_4G_8.$$

The feedback loops are

$$\begin{aligned} L_1 &= -G_2G_3G_4G_5H_2, & L_2 &= -G_5G_6H_1, & L_3 &= -G_8H_1, & L_4 &= -G_7H_2G_2, \\ L_5 &= -G_4H_4, & L_6 &= -G_1G_2G_3G_4G_5G_6H_3, & L_7 &= -G_1G_2G_7G_6H_3, & \text{and} \\ L_8 &= -G_1G_2G_3G_4G_8H_3. \end{aligned}$$

---

Loop  $L_5$  does not touch loop  $L_4$  or loop  $L_7$ , and loop  $L_3$  does not touch loop  $L_4$ ; but all other loops touch. Therefore, the determinant is

$$\Delta = 1 - (L_1 + L_2 + L_3 + L_4 + L_5 + L_6 + L_7 + L_8) + (L_5L_7 + L_5L_4 + L_3L_4).$$

The cofactors are

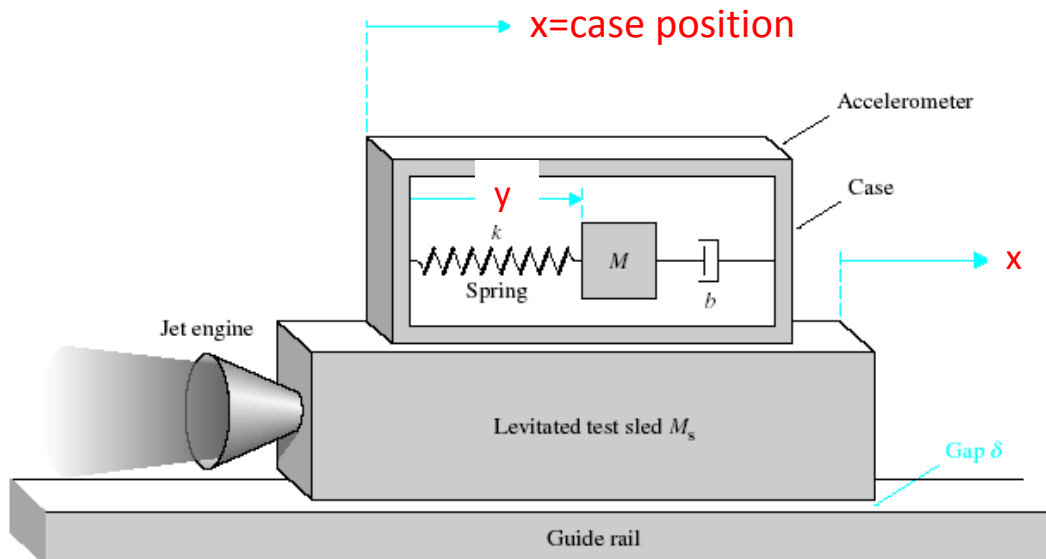
$$\Delta_1 = \Delta_3 = 1 \quad \text{and} \quad \Delta_2 = 1 - L_5 = 1 + G_4H_4.$$

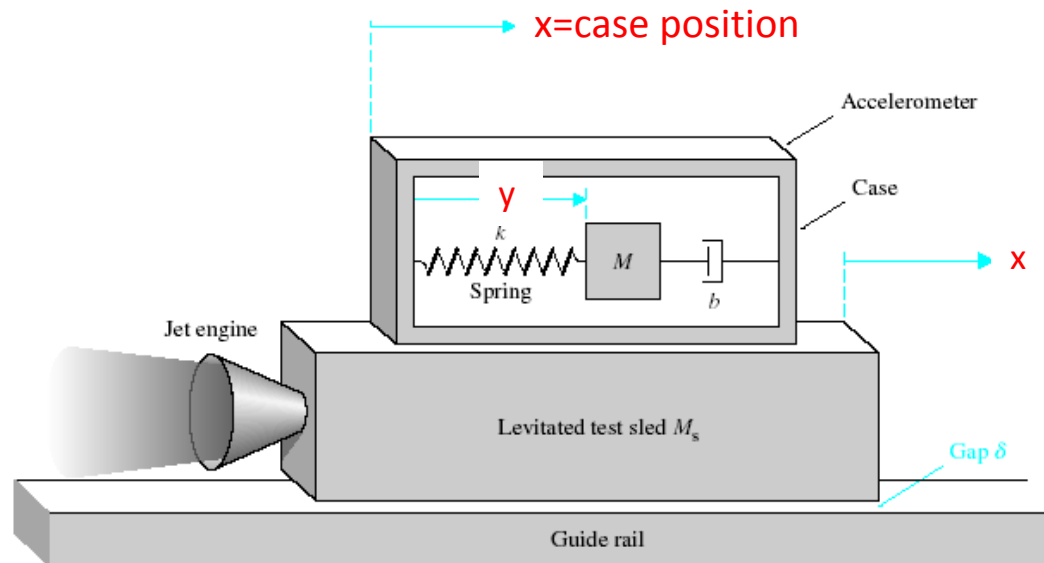
Finally, the transfer function is

$$T(s) = \frac{Y(s)}{R(s)} = \frac{P_1 + P_2\Delta_2 + P_3}{\Delta}.$$

# Real-world Example: Mechanical Accelerometer

- A mechanical accelerometer is used to measure the acceleration of a rocket test sled. The test sled maneuvers above a guide rail a small distance  $\delta$ .
- The accelerometer provides a measurement of the acceleration  $a(t)$  of the sled, since the position  $y$  of the mass  $M$ , with respect to the accelerometer case, is proportional to the acceleration of the case (and the sled).
- The goal is to design an accelerometer with an appropriate dynamic responsiveness. We wish to design an accelerometer with an acceptable time for the desired measurement characteristic.





The sum of the forces acting on the mass is

$$-b \frac{dy}{dt} - ky = M \frac{d^2}{dt^2}(y + x)$$

or

$$M \frac{d^2 y}{dt^2} + b \frac{dy}{dt} + ky = -M \frac{d^2 x}{dt^2}.$$

Since

$$M_s \frac{d^2 x}{dt^2} = F(t),$$

is the engine force, we have

$$M \ddot{y} + b \dot{y} + ky = -\frac{M}{M_s} F(t),$$

or

$$\ddot{y} + \frac{b}{M} \dot{y} + \frac{k}{M} y = -\frac{F(t)}{M_s}.$$

---

We select the coefficients where  $b/M = 3$ ,  $k/M = 2$ ,  $F(t)/M_s = Q(t)$ , and we consider the initial conditions  $y(0) = -1$  and  $\dot{y}(0) = 2$ . We then obtain the Laplace transform equation, when the force, and thus  $Q(t)$ , is a step function, as follows:

$$(s^2Y(s) - sy(0) - \dot{y}(0)) + 3(sY(s) - y(0)) + 2Y(s) = -Q(s).$$

Since  $Q(s) = P/s$ , where  $P$  is the magnitude of the step function, we obtain

$$(s^2Y(s) + s - 2) + 3(sY(s) + 1) + 2Y(s) = -\frac{P}{s},$$

or

$$(s^2 + 3s + 2)Y(s) = \frac{-(s^2 + s + P)}{s}.$$

Thus the output transform is

$$Y(s) = \frac{-(s^2 + s + P)}{s(s^2 + 3s + 2)} = \frac{-(s^2 + s + P)}{s(s + 1)(s + 2)}.$$

Expanding in partial fraction form yields

$$Y(s) = \frac{k_1}{s} + \frac{k_2}{s + 1} + \frac{k_3}{s + 2}.$$



---

We then have

$$k_1 = \frac{-(s^2 + s + P)}{(s + 1)(s + 2)} \Big|_{s=0} = -\frac{P}{2}.$$

Similarly,  $k_2 = +P$  and  $k_3 = \frac{-P - 2}{2}$ . Thus,

$$Y(s) = \frac{-P}{2s} + \frac{P}{s + 1} + \frac{-P - 2}{2(s + 2)}.$$

Therefore, the output measurement is

$$y(t) = \frac{1}{2}[-P + 2Pe^{-t} - (P + 2)e^{-2t}], \quad t \geq 0.$$

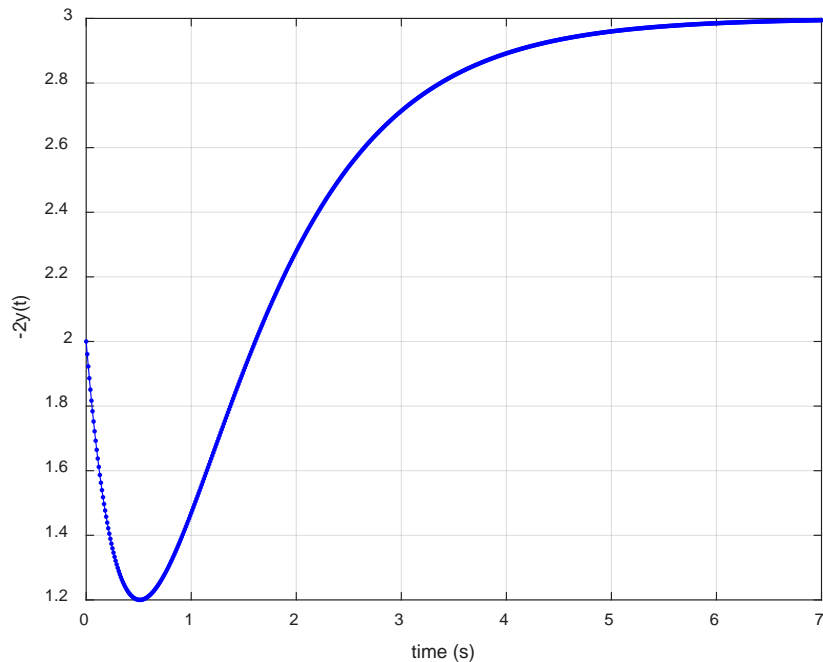
# Time-domain Response Simulation by Matlab

Command Window

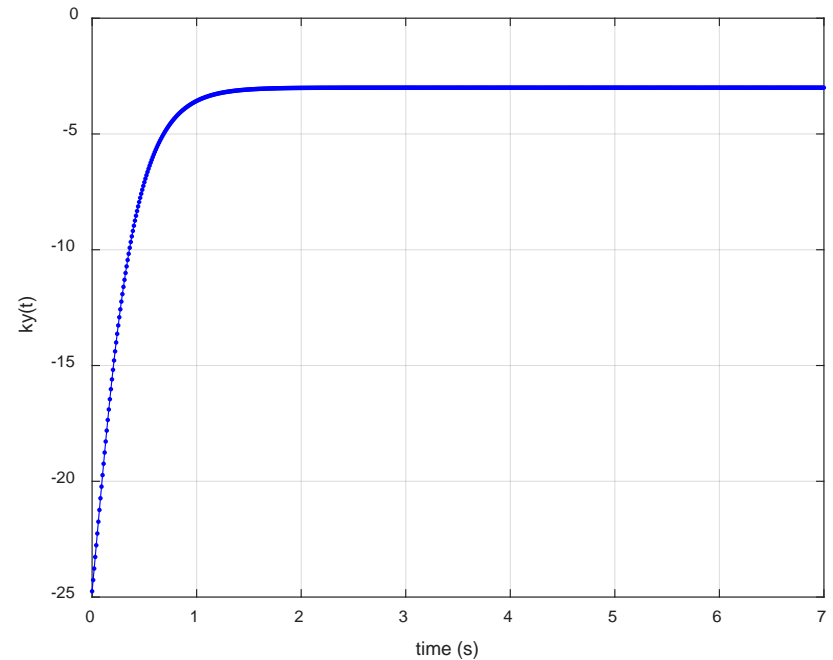
New to MATLAB? See resources for [Getting Started](#).

```
>> t = 0:0.01:7;  
>> y = 1/2*(-P+2*P*exp(-t)-(P+2)*exp(-2*t));  
>> figure,plot(t,-2*y,'b.-'); grid on; xlabel('time (s)'); ylabel('-2y(t)');  
fx >>
```

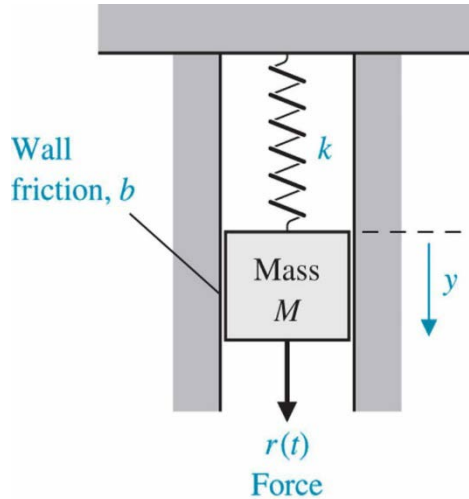
$$b/M = 3, k/M = 2,$$



$$b/M = 12 \quad k/M = 32$$



# Simulation of System Using Matlab



$$Y(s) = \frac{(M s^2 + b s + k) \cdot y_0}{M s^2 + b s + k} = \frac{p(s)}{q(s)}$$

$$y(s) = \frac{\left(s + \frac{b}{M}\right) \cdot (y_0)}{\left[s^2 + \left(\frac{b}{M}\right) \cdot s + \frac{k}{M}\right]} = \frac{(s + 2 \cdot \zeta \cdot \omega_n)}{s^2 + 2 \cdot \zeta \cdot \omega_n s + \omega_n^2}$$

$$s_1 = -(\zeta \cdot \omega_n) + \omega_n \cdot \sqrt{\zeta^2 - 1}$$

$$\omega_n = \sqrt{\frac{k}{M}} \quad \zeta = \frac{b}{(2 \cdot \sqrt{k \cdot M})}$$

$$s_2 = -(\zeta \cdot \omega_n) - \omega_n \cdot \sqrt{\zeta^2 - 1}$$

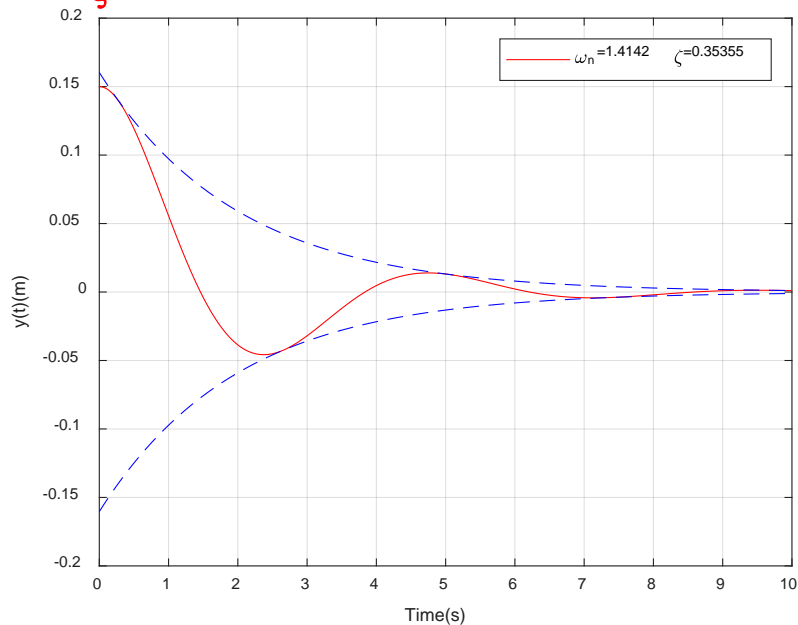
$$\square y(0) = 0.15 \text{ m}, \quad \omega_n = \sqrt{2} \frac{\text{rad}}{\text{sec}}, \quad \zeta = \frac{1}{2\sqrt{2}} \left( \frac{k}{M} = 2, \frac{b}{M} = 1 \right).$$

Command Window

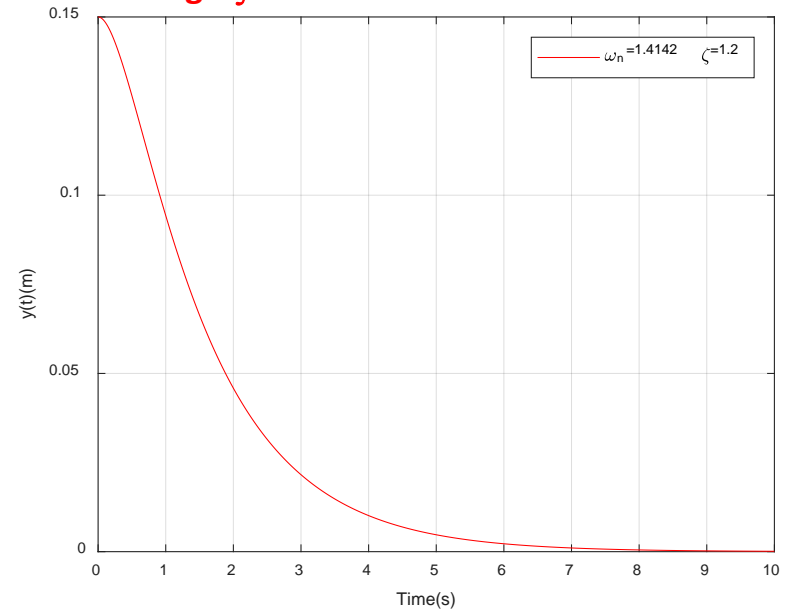
New to MATLAB? See resources for [Getting Started](#).

```
>> y0=0.15;
>> wn=sqrt(2);
>> zeta=1/(2*sqrt(2));
>> t=[0:0.01:10];
>> c=(y0/sqrt(1-zeta^2));
>> y=c*exp(-zeta*wn*t).*sin(wn*sqrt(1-zeta^2)*t+acos(zeta));
>> bu=c*exp(-zeta*wn*t);bl=-bu;
>> figure,plot(t,y,'r-',t,bu,'b--',t,bl,'b--'),grid; xlabel('Time(s)'), ylabel('y(t)(m)');
```

$\zeta < 1$ :



Change  $\zeta$  to be  $>1$ :



# Useful functions in Control Toolbox

- Calculate roots for polynomial (**roots, poly**)

$$p(s) = s^3 + 3s^2 + 4$$

```
>> p=[1 3 0 4]

p =

     1     3     0     4

>> r = roots(p)

r =

-3.3553 + 0.0000i
 0.1777 + 1.0773i
 0.1777 - 1.0773i

>> p = poly(r)

p =

     1.0000     3.0000    -0.0000     4.0000
```

- Multiply and evaluate polynomial (**conv, polyval**)

Using **conv** and **polyval** to multiply and evaluate the polynomials  $(3s^2 + 2s + 1)$   $(s + 4)$ .

```
>>p=[3 2 1]; q=[1 4];
```

```
>>n=conv(p,q)
```

```
n=
```

```
     3     14     9     4
```

```
>>value=polyval(n,-5)
```

```
value =
```

```
-66
```

Multiply  $p$  and  $q$ .

$n(s) = 3s^3 + 14s^2 + 9s + 4$

Evaluate  $n(s)$  at  $s = -5$ .

- Create transfer functions and add transfer functions (**tf**)

```
>> num1 = [10]; den1=[1 2 5];  
>> sys1 = tf(num1, den1)
```

sys1 =

$$\frac{10}{s^2 + 2s + 5}$$

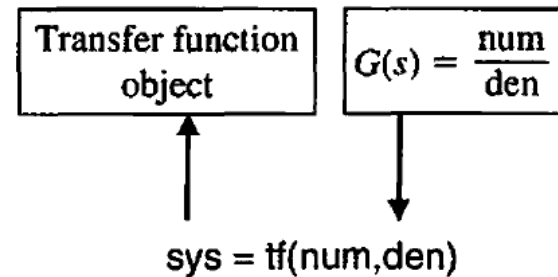
Continuous-time transfer function.

```
>> sys2 = tf([1], [1 1])
```

sys2 =

$$\frac{1}{s + 1}$$

Continuous-time transfer function.



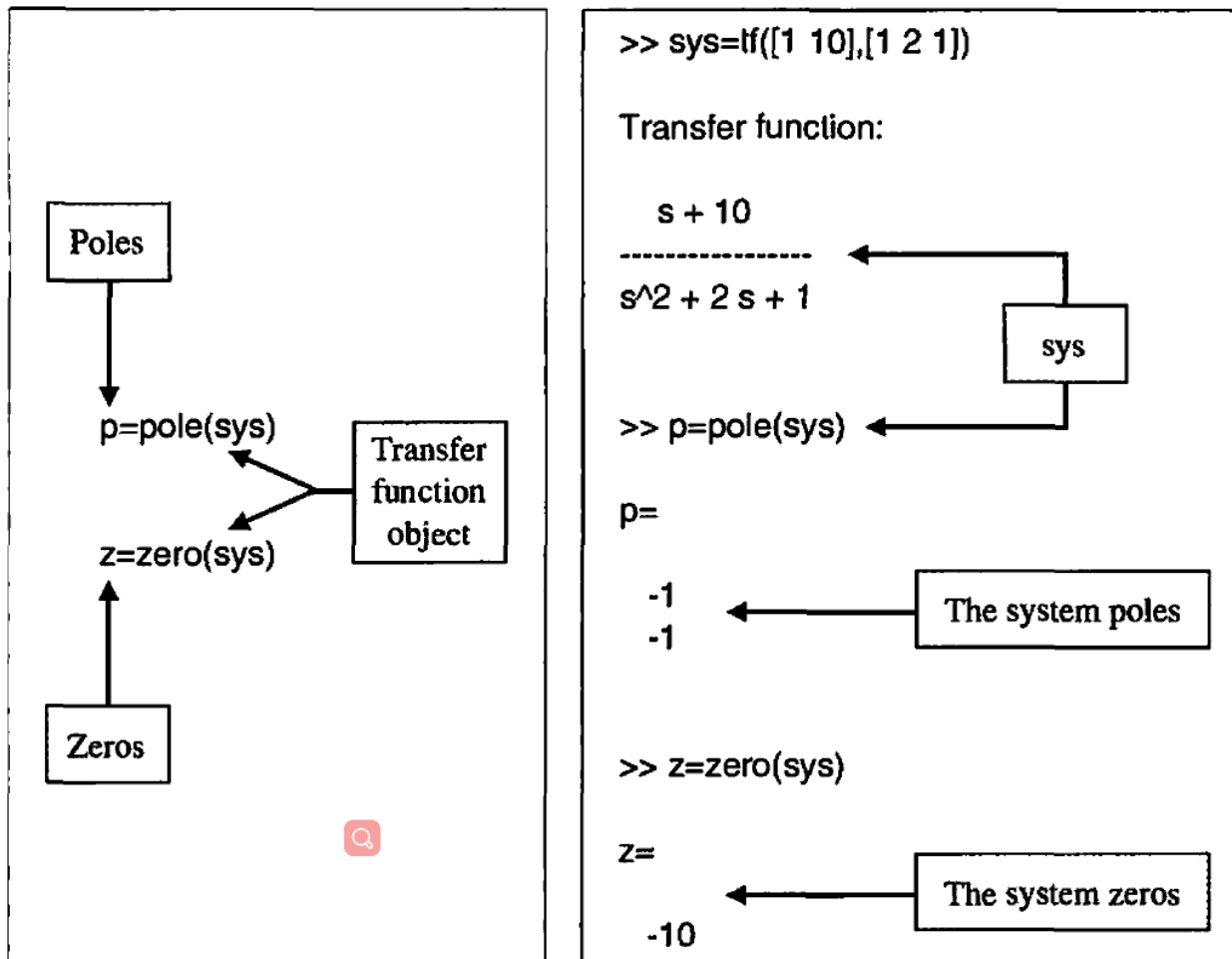
```
>> sys=sys1+sys2
```

sys =

$$\frac{s^2 + 12s + 15}{s^3 + 3s^2 + 7s + 5}$$

Continuous-time transfer function.

- Compute pole and zero locations of a linear system (**pole, zero**)

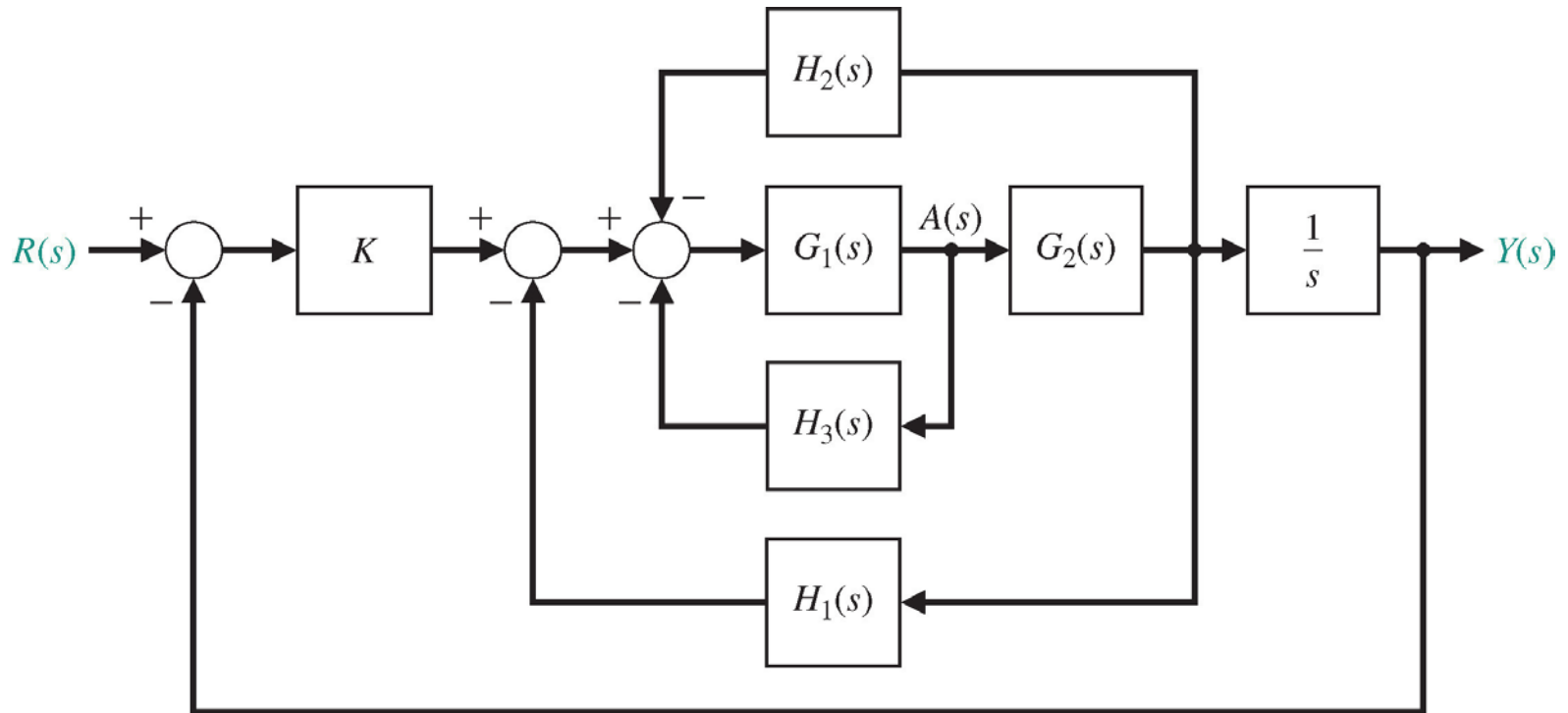


- 
- Pole-zero map (**pzmap**);
  - Block diagram (**series, parallel**);
  - Feedback system (**feedback**)
  - Time-domain response (**step, impulse**);
  - .... (Please refer to the manual of control system toolbox of Matlab)



# Quiz 10.1

Determine the transfer function for the following system:



---

# Thank You !