



Xi'an Jiaotong-Liverpool University  
西交利物浦大學

# EEE220 Instrumentation and Control System

*2018-19 Semester 2*

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*29 April, 2019*

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# Lecture 19

# Outline

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## **The Root Locus Method**

- ☐ The Root Locus Concept
- ☐ The Root Locus Procedure
- ☐ **The Root Locus Using Matlab**
- ☐ **Parameter Design by the Root Locus Method**
- ☐ PID Controllers
  - Concept
  - PID Tuning
- ☐ Design Examples

# A Complete Example 19.1

Obtain the root locus for the following characteristic equation of a system as  $K$  varies for  $0 \leq K < \infty$ .

$$1 + \frac{K}{s^4 + 12s^3 + 64s^2 + 128s} = 0.$$

*Step 1. prepare the sketch.*

$$1 + \frac{K}{s(s + 4)(s + 4 + j4)(s + 4 - j4)} = 0$$

- The system has no finite zeros. There are four poles.
- Because  $n = 4, M = 0$ , we have four separate loci.
- The root loci are symmetrical with respect to the real axis.

*Step 2. determine the segments on the real axis that are root loci.*

- A segment of the root locus exists on the real axis between  $s = 0$  and  $s = -4$ .

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*Step 3. determine the asymptotes.*

Angles:

$$\phi_A = \frac{(2k + 1)}{4} 180^\circ, \quad k = 0, 1, 2, 3;$$

$$\phi_A = +45^\circ, 135^\circ, 225^\circ, 315^\circ.$$

Centroid:

$$\sigma_A = \frac{-4 - 4 - 4j - 4 + 4j}{4} = -3$$

*Step 4. determine the point where the locus crosses the imaginary axis (if any).*

The characteristic equation is rewritten as

$$s(s + 4)(s^2 + 8s + 32) + K = s^4 + 12s^3 + 64s^2 + 128s + K = 0.$$

Therefore, the Routh array is

$$\begin{array}{c|ccc} s^4 & 1 & 64 & K \\ s^3 & 12 & 128 & \\ s^2 & b_1 & K & \\ s^1 & c_1 & & \\ s^0 & K & & \end{array},$$

where 
$$b_1 = \frac{12(64) - 128}{12} = 53.33 \quad \text{and} \quad c_1 = \frac{53.33(128) - 12K}{53.33}.$$

The gain  $K$  for marginally stability is  $K = 568.89$ , and the roots for the auxiliary equation are

$$53.33s^2 + 568.89 = 53.33(s^2 + 10.67) = 53.33(s + j3.266)(s - j3.266).$$

Therefore, the root locus crosses the  $j\omega$ -axis at  $s = \pm j3.266$  when  $K = 568.89$ .

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*Step 5. determine the breakaway point (if any).*

The breakaway point is estimated by evaluating

$$K = p(s) = -s(s + 4)(s + 4 + j4)(s + 4 - j4)$$

between  $s = -4$  and  $s = 0$ .

We set  $\frac{dK}{ds} = \frac{dp(s)}{ds} = 0$ , and find  $s = -1.577$ .

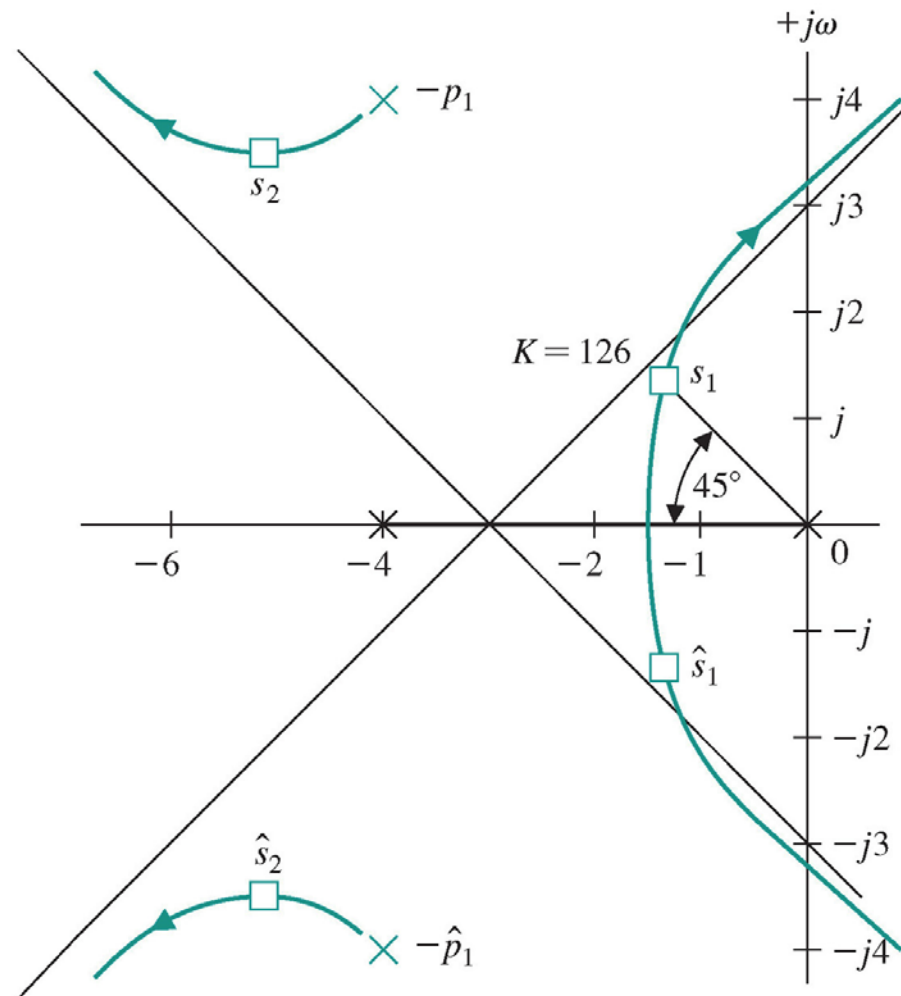
*Step 6. determine the departure angles.*

For angle of departure at complex pole  $-p_1$ , utilize angle criterion as follows

$$\theta_1 + 90^\circ + 90^\circ + \theta_3 = 180^\circ + k360^\circ.$$

Since  $\theta_3 = 135^\circ \rightarrow \theta_1 = -135^\circ \equiv 225^\circ$

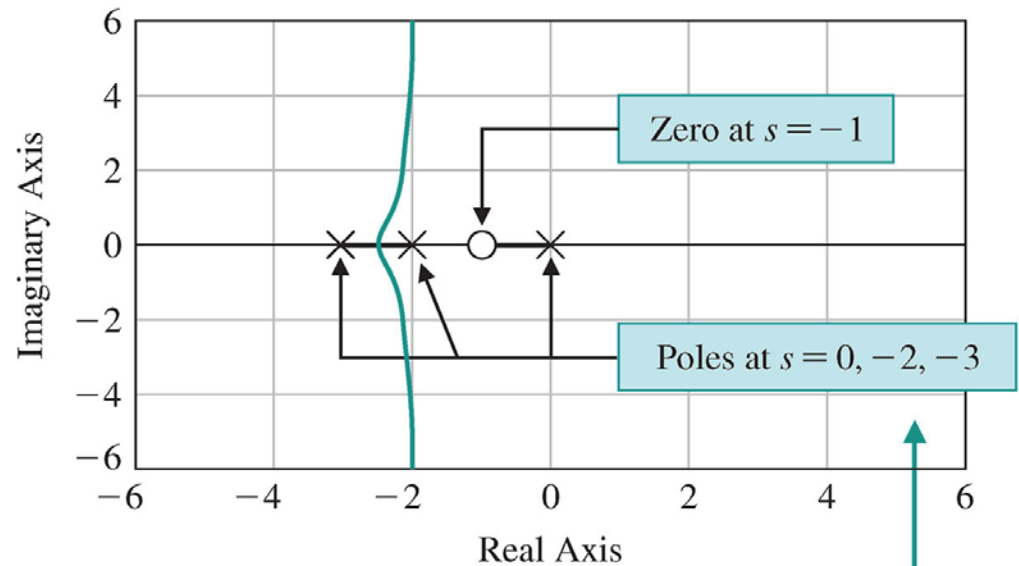
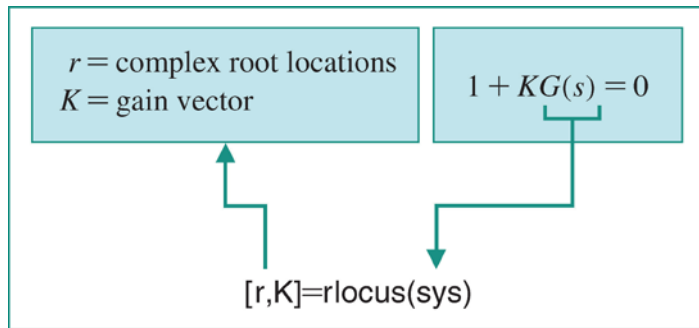
Step 7. complete the sketch.





# Root Locus Using Matlab

The **rlocus** function.



```
>>p=[1 1]; q=[1 5 6 0]; sys=tf(p,q); rlocus(sys)
```

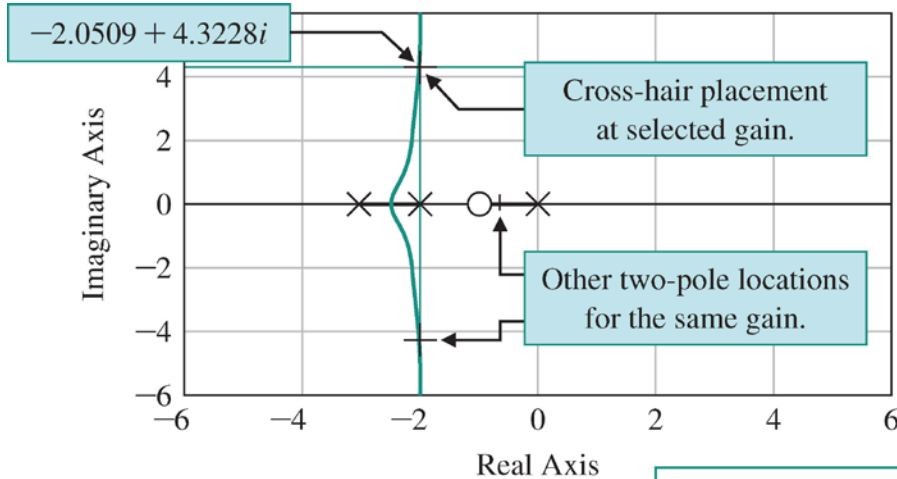
Generating a root locus plot.

```
>>p=[1 1]; q=[1 5 6 0]; sys=tf(p,q); [r,K]=rlocus(sys);
```

Obtaining root locations  $r$  associated with various values of the gain  $K$ .

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## Using the **rlocfind** function.



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```
>>p=[1 1]; q=[1 5 6 0]; sys=tf(p,q); rlocus(sys)  
>>rlocfind(sys)
```

rlocfind follows the rlocus function.

Select a point in the graphics window

```
selected_point =  
    -2.0509 + 4.3228i
```

```
ans =
```

```
    20.5775
```

Value of  $K$  at selected point

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# Parameter Design by the Root Locus Method

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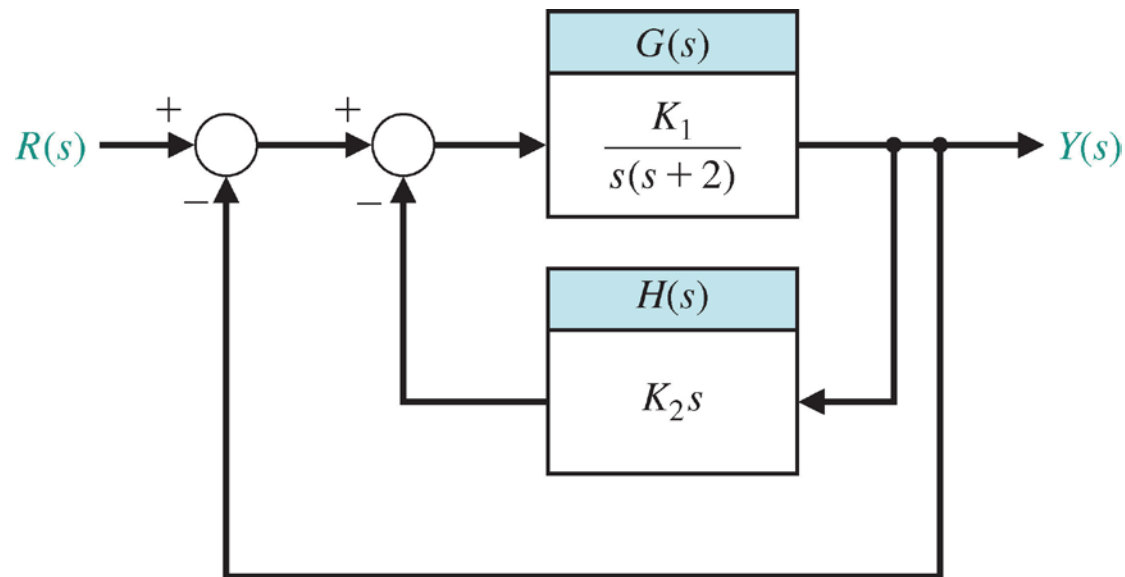
- ❑ Originally, the root locus method was developed to determine the locus of roots of the characteristic equation as the system gain  $K$  is varied from zero to infinity; while the effect of other system parameters can be readily investigated by rearranging the characteristic equation;
- ❑ It appears that the root locus method is a single-parameters. The question arises: **How do we investigate the effect of two or more parameters?**

--- Fortunately, this method can be extended to more than one parameters, based on which **parameter design** for the system is enabled.

# Example 19.2: Welding Head Control

A welding head for an auto body requires an accurate control system for positioning the welding head. The feedback control system is to be designed (i.e., values of  $K_1$  and  $K_2$  are to be determined) to satisfy the following specifications:

1. Steady-state error for a ramp input is  $e_{ss} \leq 35\%$  of the input slope
2. Damping ratio of dominant roots is  $\zeta \geq 0.707$
3. Settling time to within 2% of the final value is  $T_s \leq 3s$



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## Solutions:

Step 1. determine root locations in the s-plane to satisfy the design specifications.

- For steady-state error requirement:

$$E(s) = R(s) - Y(s) = \frac{s^2 + (K_1 K_2 + 2)s}{s^2 + (K_1 K_2 + 2)s + K_1} R(s)$$

$$\text{Ramp input} \rightarrow R(s) = \frac{A}{s^2}$$

$$e_{ss} = \lim_{s \rightarrow 0} sE(s) = \lim_{s \rightarrow 0} s \frac{s^2 + (K_1 K_2 + 2)s}{s^2 + (K_1 K_2 + 2)s + K_1} \frac{A}{s^2} = \frac{K_1 K_2 + 2}{K_1} A \leq 0.35A$$

$$K_2 + \frac{2}{K_1} \leq 0.35 \quad \rightarrow \text{we need small value of } K_2.$$

- For damping ratio requirement:

$$\zeta \geq 0.707$$

$$\text{as } \theta = \cos^{-1}\zeta$$

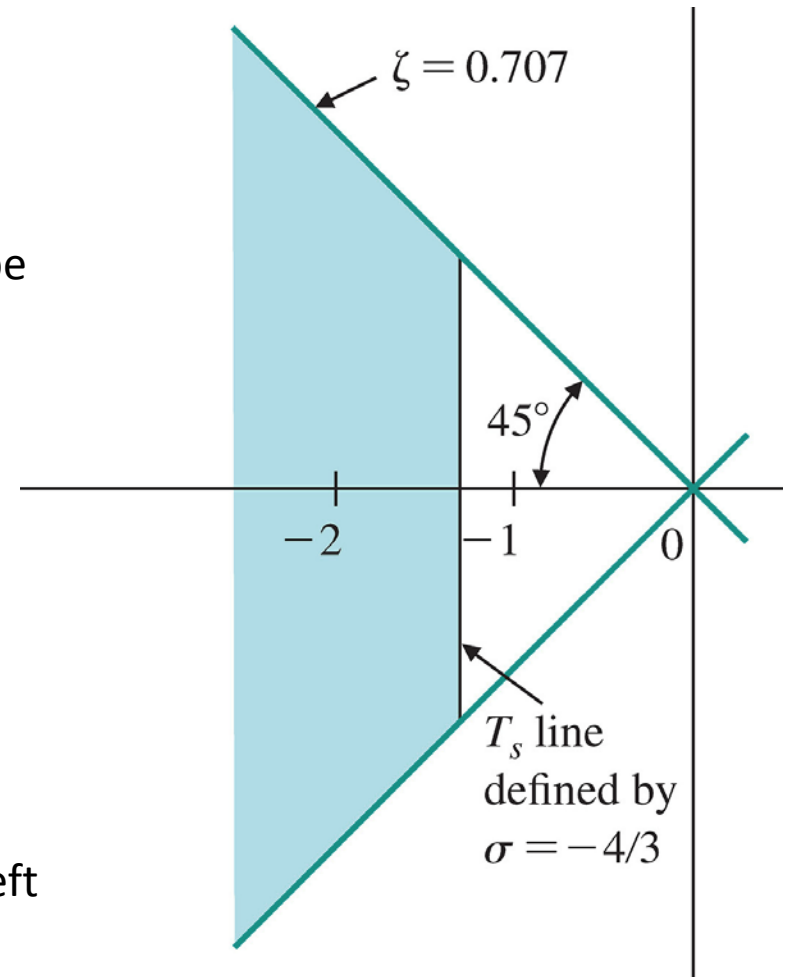
--> The roots of the closed-loop system must be below the line at  $45^\circ$  in the left-hand s-plane.

- For settling time requirement:

$$T_s = \frac{4}{\zeta\omega_n} = \frac{4}{-\sigma} \leq 3$$

$$-\sigma \geq \frac{4}{3}$$

--> We want the dominant roots to lie to the left of the line defined by  $\sigma = -\frac{4}{3}$ .



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Step 2. Look into the root locus with one varying parameter, while setting the other parameter to be zero.

Characteristic equation for the closed-loop system:

$$\Delta(s) = s^2 + (K_1K_2 + 2)s + K_1$$

Assume  $K_1 = \alpha$ ,  $K_1K_2 = \beta$ , then

$$\Delta(s) = s^2 + \beta s + 2s + \alpha$$

Set  $\beta = 0$ , sketch the root locus with varying  $\alpha$  from zero to infinity

$$1 + \alpha \frac{1}{s(s+2)} = 0$$

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Step 3. Select a fixed value of  $\alpha$ , investigate the effect of another parameter by sketching the corresponding root locus.

For example, choose a gain of  $K_1 = \alpha = 20$ , the roots are

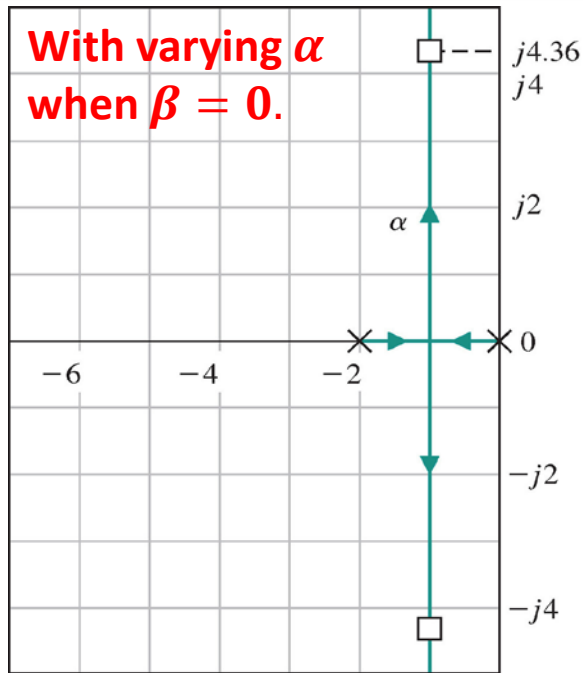
$$s = -1 \pm j4.36$$

Then the effect of varying  $\beta = 20K_2$  is determined from the locus equation

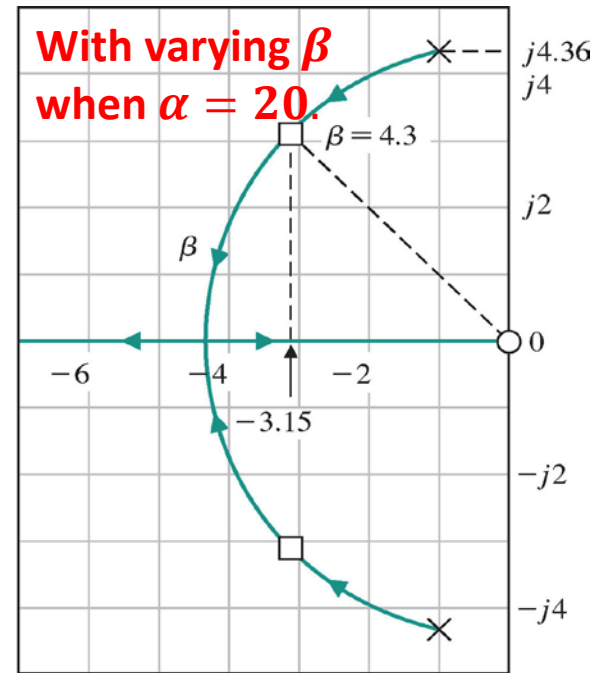
$$1 + \beta \frac{s}{s^2 + 2s + 20} = 0$$

The root locus for  $\alpha = 20$  and with varying  $\beta$  can be then obtained.





(a)



(b)

#### Step 4. determine the parameter values.

The root with  $\zeta = 0.707$  are obtained when  $\beta = 4.3$ , the real part of these roots is  $\sigma = -3.15$ , then  $T_s = 1.27s$ . Therefore, when  $K_1 = 20$ ,  $K_2 = 0.215$ , the design specifications can be met.

*The root locus method can be extended to more than two parameters by extending the number of steps in the method.*

# Root Contours

Actually, a family of root loci can be generated for two parameters in order to determine the total effect of varying two parameters. For example, let us determine the effect of varying  $\alpha$  and  $\beta$  of the following characteristic equation:

$$s^3 + 3s^2 + 2s + \beta s + \alpha = 0$$

The root locus equation as a function of  $\alpha$  is (set  $\beta = 0$ )

$$1 + \frac{\alpha}{s(s+1)(s+2)} = 0 \quad (1)$$

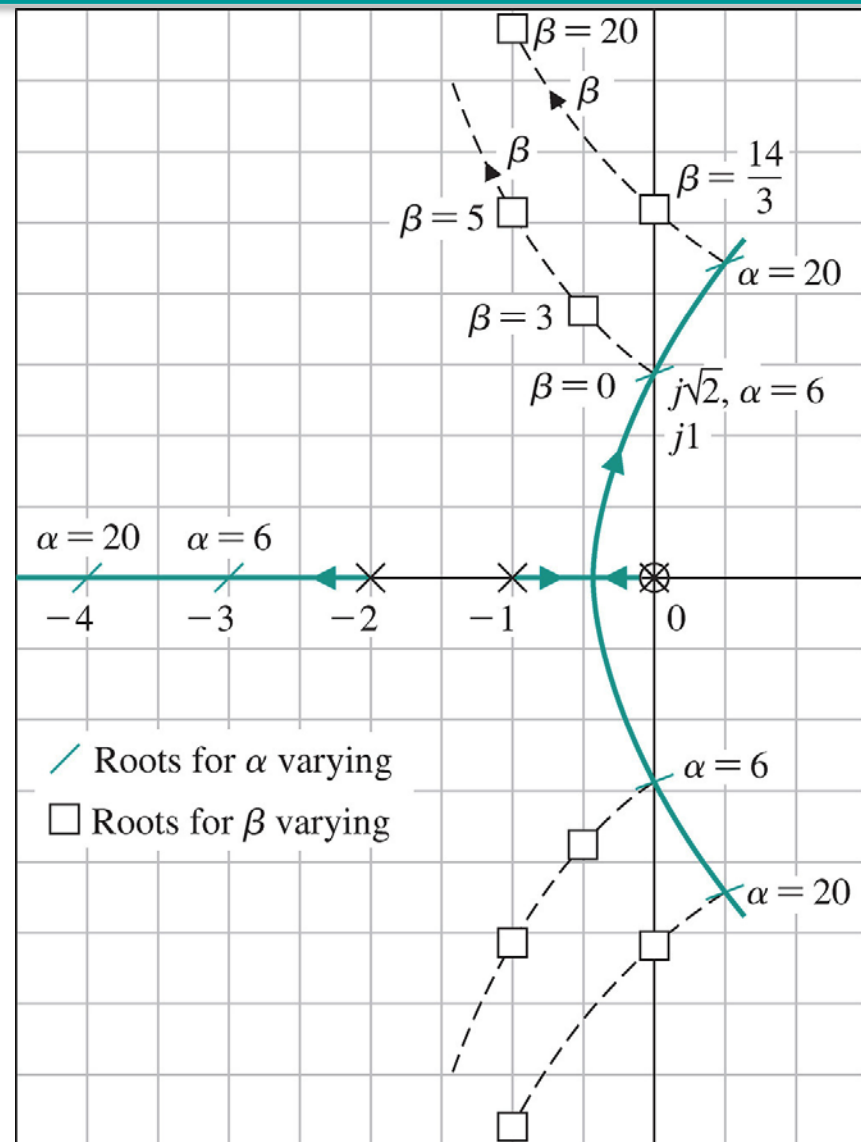
The root locus as a function of  $\beta$  is

$$1 + \frac{\beta s}{s^3 + 3s^2 + 2s + \alpha} = 0 \quad (2)$$

**Note:** the roots of eq.(1) become poles of eq.(2).

A family of loci, often called root contours can be sketched, which illustrates the effect of varying both  $\alpha$  and  $\beta$  on the roots of the system's characteristic equation.

Two-parameter root locus.  
The loci for  $\alpha$  varying are solid; the loci for  $\beta$  varying are dashed.



# Quiz 19.1

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Two unity feedback control systems have the loop transfer functions:

$$(a) \quad L(s) = G_c(s)G(s) = \frac{K}{s(s+2)(s^2+4s+5)}.$$

$$(b) \quad L(s) = G_c(s)G(s) = \frac{K(s^2+4s+8)}{s^2(s+4)}.$$

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# Thank You !