

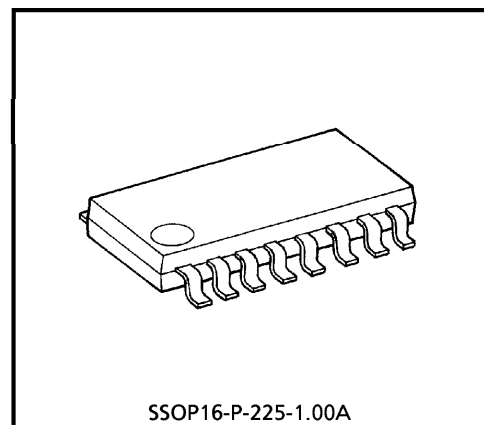
TA7688F

STEREO HEADPHONE AMPLIFIER (3V USE)

The TA7688F is a stereo headphone power amplifier IC designed for portable cassette player applications.

FEATURES

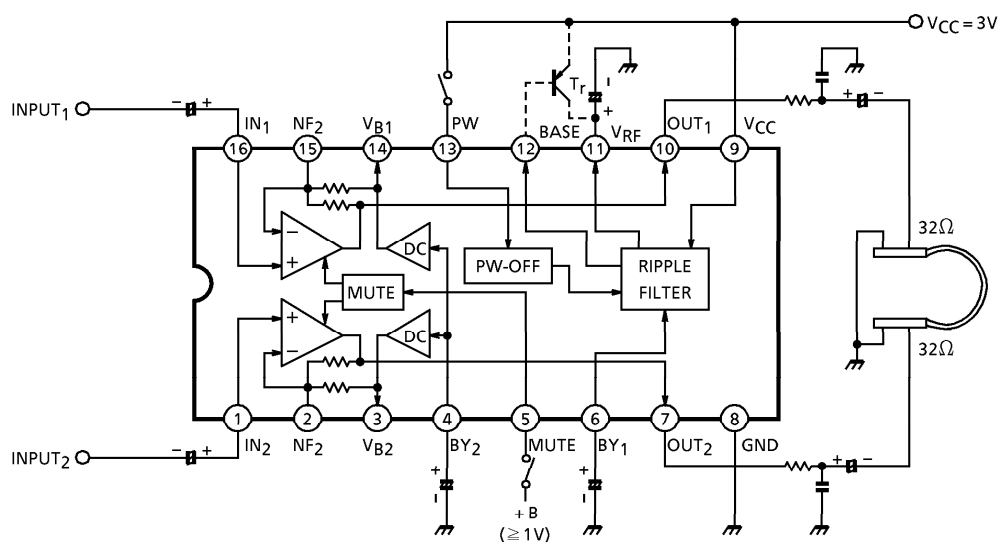
- Small installed area and few external parts
- Low supply current : $I_{CCQ} = 7\text{mA}$ (Typ.) at 3V
- Built-in a ripple filter
- Built-in a power amplifier mute
- Built-in a power off circuit
- Operating supply voltage range : $V_{CC}(\text{opr}) = 1.8\sim 5\text{V}$
- Recommended supply voltage : $V_{CC} = 3\text{V}$
- The standard model is TA7688F (SO)



SSOP16-P-225-1.00A

Weight : 0.14g (Typ.)

BLOCK DIAGRAM



Dotted Line is an additional circuit to boost the stabilized current. (Option)

ELECTRICAL CHARACTERISTICS

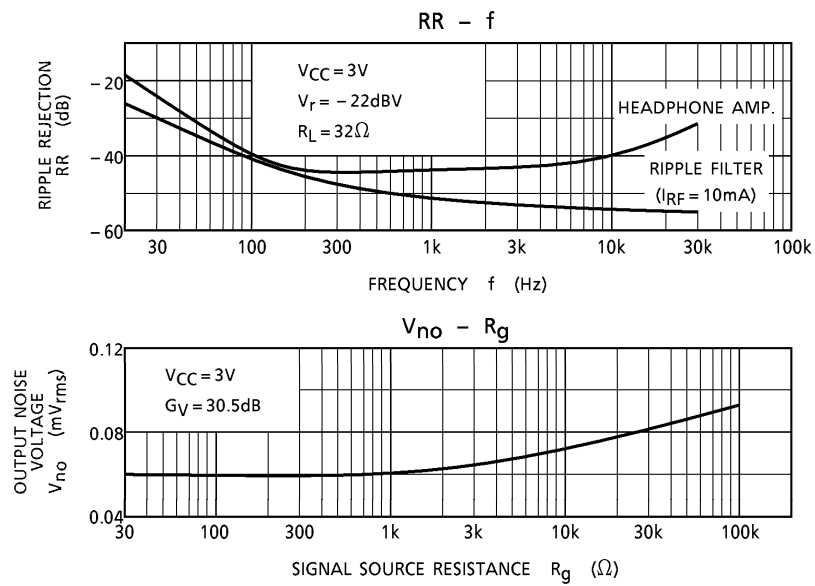
1. AC characteristics (Unless otherwise specified, $T_a = 25^\circ\text{C}$, $V_{CC} = 3\text{V}$, $R_g = 600\Omega$, $f = 1\text{kHz}$)
 $R_H = 3.9\Omega$, $R_L = 32\Omega$

CHARACTERISTIC	SYMBOL	TEST CIR- CUIT	TEST CONDITION	MIN.	TYP.	MAX.	UNIT
Quiescent Current	$I_{CCO}(1)$	—	$V_{in} = 0$	—	7	12	mA
	$I_{CCO}(2)$		$V_{in} = 0$, SW_2 : OFF	—	1	10	μA
Output Power	$P_O(1)$	—	THD = 10%	20	27	—	mW
	$P_O(2)$		$R_L = 16\Omega$, THD = 10%	—	38	—	
Total Harmonic Distortion	THD	—	$P_O = 10\text{mW/ch}$	—	0.12	1.0	%
Closed Loop Voltage Gain	G_V	—	$V_{in} = -42\text{dBV}$	28.5	30.5	32.5	dB
Channel Balance	ΔG_V	—	$V_{in} = -42\text{dBV}$	—	0	± 1	dB
Cross Talk	CT	—	$V_O = -2\text{dBV}$, $ch1 \leftrightarrow ch2$	45	65	—	dB
Ripple Rejection	Headphone AMP	RR (1)	$f_r = 1\text{kHz}$, $V_r = -22\text{dBV}$	30	45	—	dB
	Ripple Filter	RR (2)	$f_r = 100\text{Hz}$, $V_r = -22\text{dBV}$	—	40	—	dB
Output Noise Voltage	V_{no}	—	BPF = 20Hz~20kHz	—	0.06	0.2	mV_{rms}
Input Resistance	R_{IN}	—	$f = 1\text{kHz}$	15	20	25	$\text{k}\Omega$
Ripple Filter Output Voltage	$V_{RF}(1)$	—	$V_{CC} = 2\text{V}$, $I_{RF} = 10\text{mA}$	1.45	1.6	—	V
	$V_{RF}(2)$		$I_{RF} = 10\text{mA}$	2.1	2.3	2.5	
	$V_{RF}(3)$		$V_{CC} = 4.5\text{V}$, $I_{RF} = 10\text{mA}$	—	3.4	—	
Muting Attenuation	ATT	—	$V_{MUTE} = 3\text{V}$ (0dB = 240 mV_{rms})	60	80	—	dB
Muting Input Voltage	V_{MUTE}	—	ATT $\geq 50\text{dB}$ (0dB = 240 mV_{rms})	—	0.7	1.0	V
Muting Input Current	I_{MUTE}	—	ATT $\geq 50\text{dB}$ (0dB = 240 mV_{rms})	—	35	—	μA
Ripple Filter Current	I_B	—	—	—	0.05	—	mA

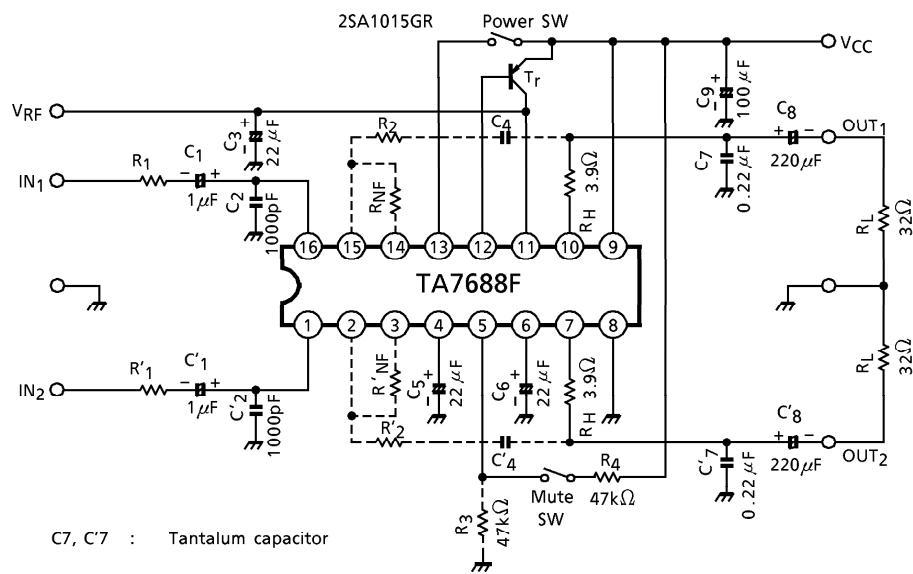
2. DC characteristics

($T_a = 25^\circ\text{C}$, $V_{CC} = 3\text{V}$, Terminal voltage at no signal)

ITEM	SYMBOL	RATING	UNIT
Terminal 1 (IN_2)	V_1	1.5	V
2 (NF_2)	V_2	1.5	V
3 (VB_2)	V_3	1.5	V
4 ($BYPASS_2$)	V_4	1.5	V
5 (MUTE)	V_5	0	V
6 ($BYPASS_1$)	V_6	2.2	V
7 (OUT_2)	V_7	1.5	V
8 (GND)	V_8	0	V
9 (V_{CC})	V_9	3.0	V
10 (OUT_1)	V_{10}	1.5	V
11 (V_{RF})	V_{11}	2.3	V
12 (BASE)	V_{12}	2.2	V
13 (PW ON / OFF)	V_{13}	3.0	V
14 (VB_1)	V_{14}	1.5	V
15 (NF_1)	V_{15}	1.5	V
16 (IN_1)	V_{16}	1.5	V



APPLICATION CIRCUIT



EXTERNAL PARTS TABLE (Mention only CH₁)

PARTS No.	TYPICAL	PURPOSE	INFLUENCE		NOTE
			SMALLER THAN TYP.	GREATER THAN TYP.	
C ₁	1 μ F	Coupling	Bad low frequency response	"P _{op} " noise is high.	Input
C ₂	1000pF	LPF	$f_{CH} = \frac{1}{2\pi C_2 (R_1 // Z_{in})}$ – 3dB (30kHz) at R ₁ = 5.6k Ω – 3dB (20kHz) at R ₁ = 12k Ω		Noise receiving protection
R ₁	—				Equivalent signal source impedance
C ₃	22 μ F	Decoupling for V _{RF}	Stability (OSC) decreases, V _{no} at V _{RF} increases	(It is better to connect to input side GND)	Use tantalum capacitor
R _{NF}	—	G _V Adjustment	Not available at G _V < 30dB If necessary deide at input level by resistors		—
R ₂	(15k Ω)	f-response control, THD improvement at high freq.	– 3dB point is 20kHz. Check ringing at clip by OSC margine down.		Low OSC margine at G _V < 40dB
C ₄	(180pF)				
C ₅	22 μ F	Bypass capacitor for bias	THD and V _{no} Degradation	—	It is better to connect to input side GND.
C ₆	22 μ F	Bypass capacitor for ripple filter	Ripple rejection ratio degradation	—	It is better to connect to output side GND.
R ₃	47k Ω	Pull down resistor at mute pin	I _{CC} increases at mute ON	Pull down effect down	Additional resistor at long pattern only
R ₄	47k Ω	I _{MUTE} limiter	I _{MUTE} increases (Unnecessary at V _{CC} = 3V)	I _{MUTE} decreases	I _{MUTE} < 150 μ A
R _H	3.9 Ω	Protection resistance. Phase compensation	Rush current increases. Phase compensation is out.	Output decreases. Phase compensation is out.	CR filter with C ₇
C ₇	0.22 μ F	Phase compensation	Oscillation	THD degradation by load capacitance	Recommended to use tantalum or film capacitor
C ₈	220 μ F	Coupling	Bad low frequency response	"P _{op} " noise is high.	Output
C ₉	100 μ F	V _{CC} decoupling	Oscillation margin decreases	—	Necessary to be near pin 9
T _r	2SA1015GR	Booster for V _{RF}	—	—	To be added at I _{RF} > 10mA