

SECOND EDITION

ADVANCED ENGINEERING ELECTROMAGNETICS

Constantine A. Balanis

SOLUTIONS MANUAL



CHAPTER 1

1.1

$$\nabla \times \underline{\underline{H}} = \underline{\underline{J}}_{ic} + \frac{\partial \underline{\underline{B}}}{\partial t}$$

Taking the divergence of both sides

$$\nabla \cdot (\nabla \times \underline{\underline{H}}) = \nabla \cdot \underline{\underline{J}}_{ic} + \nabla \cdot \frac{\partial \underline{\underline{B}}}{\partial t} = \nabla \cdot \underline{\underline{J}}_{ic} + \frac{\partial}{\partial t} \nabla \cdot \underline{\underline{B}}$$

Using the vector identity of

$\nabla \cdot (\nabla \times \underline{\underline{A}} = 0)$ and (1-3) we can write that

$$0 = \nabla \cdot \underline{\underline{J}}_{ic} + \frac{\partial}{\partial t} (q_{ve}) \Rightarrow \boxed{\nabla \cdot \underline{\underline{J}}_{ic} = - \frac{\partial q_{ve}}{\partial t}}$$

1.2

$$\nabla \times \underline{\underline{E}} = - \underline{\underline{M}}_i - \frac{\partial \underline{\underline{B}}}{\partial t}$$

Taking a surface integral of both sides, we can write that

$$\iint_S (\nabla \times \underline{\underline{E}}) \cdot d\underline{s} = - \iint_S \underline{\underline{M}}_i \cdot d\underline{s} - \frac{\partial}{\partial t} \iint_S \underline{\underline{B}} \cdot d\underline{s}$$

Applying Stokes' theorem of (1-7) to the left side of the equation above leads to

$$\oint_C \underline{\underline{E}} \cdot d\underline{l} = - \iint_S \underline{\underline{M}}_i \cdot d\underline{s} - \frac{\partial}{\partial t} \iint_S \underline{\underline{B}} \cdot d\underline{s}$$

Using the same procedure, we can write that

$$\oint_C \underline{\underline{H}} \cdot d\underline{l} = \iint_S \underline{\underline{J}}_i \cdot d\underline{s} + \iint_S \underline{\underline{J}}_c \cdot d\underline{s} + \frac{\partial}{\partial t} \iint_S \underline{\underline{D}} \cdot d\underline{s}$$

For the remaining three equations of Table 1-1 we proceed as follows:

$$\nabla \cdot \underline{\underline{D}} = q_{ve}$$

Taking a volume integral of both sides, we can write that

$$\iiint_V \nabla \cdot \underline{\underline{D}} dv = \iiint_V q_{ve} ds = Q_e$$

1.2 Cont'd

Applying the divergence theorem of (1-8) on the left side of the equation above leads to

$$\oint_S \underline{D} \cdot d\underline{s} = Q_e$$

Using the same procedure, we can write that

$$\oint_S \underline{B} \cdot d\underline{s} = Q_m$$

and

$$\oint_S \underline{J}_{ic} \cdot d\underline{s} = -\frac{\partial}{\partial t} \iiint_V q_{ve} dv = -\frac{\partial Q_e}{\partial t}$$

1.3

(a) $\underline{D} = \hat{a}_x(3+x)$

$$Q_e = \oint_S \underline{D} \cdot d\underline{s} = \int_0^1 \int_0^1 \hat{a}_x(3+x) | \cdot (\hat{a}_x dy dz) + \int_0^1 \int_0^1 \hat{a}_x(3+x) | \cdot dy dz = -3 + 4 = 1$$

(b) $\underline{D} = \hat{a}_y(4+y^2)$

$$Q_e = \oint_S \underline{D} \cdot d\underline{s} = \int_0^1 \int_0^1 \hat{a}_y(4+y^2) | \cdot (-\hat{a}_y dx dz) + \int_0^1 \int_0^1 \hat{a}_y(4+y^2) | \cdot \hat{a}_y dx dz = -4 + 5 = 1$$

1.4 $D_2 = 6\hat{a}_x + 3\hat{a}_z, \chi_{e2} = \epsilon_{sr2} - 1 = 2.56 - 1 = 1.56$

(a) $E_2 = \frac{D_2}{\epsilon_{e2}} = \frac{1}{2.56\epsilon_0} (6\hat{a}_x + 3\hat{a}_z) = \frac{1}{\epsilon_0} \left(\frac{6}{2.56} \hat{a}_x + \frac{3}{2.56} \hat{a}_z \right)$

$$E_2 = \frac{1}{\epsilon_0} (2.34\hat{a}_x + 1.1718\hat{a}_z)$$

(b) $P_2 = \epsilon_0 \chi_e E_2 = \epsilon_0 \left[1.56 \frac{1}{\epsilon_0} (2.34\hat{a}_x + 1.1718\hat{a}_z) \right]$
 $P_2 = 3.65\hat{a}_x + 1.828\hat{a}_z$

(c) $E_{1x} = E_{2x}$ Continuity of tangential components of E -field

$$\frac{D_{1x}}{\epsilon_0} = \frac{D_{2x}}{\epsilon_{e2}\epsilon_0} \Rightarrow D_{1x} = \frac{D_{2x}}{\epsilon_{e2}} = \frac{6}{2.56} = 2.344$$

$$\hat{n} \cdot (D_2 - D_1) = q_{es} \text{ Discontinuity of normal components of } D \text{ density}$$

$$\hat{n} = \hat{a}_z; D_{2z} - D_{1z} = q_{es} = 0.2 \Rightarrow D_{1z} = D_{2z} - q_{es} = 3 - 0.2 = 2.8$$

$$D_{1z} = 2.8$$

$$\underline{D}_1 = 2.344\hat{a}_x + 2.8\hat{a}_z$$

Cont'd

[1.4 cont'd]

$$(d) \quad \underline{\epsilon}_1 \underline{E}_1 = \underline{D}_1 \Rightarrow \epsilon_0 \underline{E}_1 = \underline{D}_1 \Rightarrow \underline{E}_1 = \frac{1}{\epsilon_0} \underline{D}_1 = \frac{1}{\epsilon_0} (2.34 \hat{a}_x + 2.8 \hat{a}_z)$$

$$\underline{E}_1 = \frac{1}{\epsilon_0} (2.34 \hat{a}_x + 2.8 \hat{a}_z)$$

$$(e) \quad \chi_{e1} = \epsilon_{sr1} - 1 = 1 - 1 = 0$$

$$\underline{P}_1 = \epsilon_0 \gamma_{e1} \underline{E}_1 = 0$$

[1.5]

$$\underline{H}_1 = 3 \hat{a}_x + \hat{a}_z q, \quad \mu_2 = 4 \mu_0$$

$$(a) \quad \underline{B}_1 = \mu_0 \underline{H}_1 = \mu_0 (3 \hat{a}_x + q \hat{a}_z)$$

(b)

$$\underline{M}_1 = \chi_{m1} \underline{H}_1, \quad \chi_{m1} = \mu_{sr} - 1 = 1 - 1 = 0$$

$$\underline{M}_1 = 0$$

(c)

$$H_{2x} = H_{1x} = 3 \quad \text{Continuity of tangential } \underline{H}-\text{field}$$

Continuity of normal \underline{B} density

$$B_{2z} = B_{1z} = q \mu_0 \Rightarrow \mu_2 H_{2z} = q \mu_0 \Rightarrow 4 \mu_0 H_{2z} = q \mu_0$$

$$H_{2z} = \frac{q}{4} = 2.25$$

$$\underline{H}_2 = 3 \hat{a}_x + 2.25 \hat{a}_z$$

(d)

$$\underline{B}_2 = \mu_2 \underline{H}_2 = 4 \mu_0 (3 \hat{a}_x + \frac{q}{4} \hat{a}_z) = \mu_0 (12 \hat{a}_x + q \hat{a}_z)$$

(e)

$$\underline{M}_2 = \chi_{m2} \underline{H}_2 \quad \chi_{m2} = \mu_{sr} - 1 = 4 - 1 = 3$$

$$\underline{M}_2 = 3 (3 \hat{a}_x + 2.25 \hat{a}_z) = 9 \hat{a}_x + 6.75 \hat{a}_z$$

1.6) $D_0 = \epsilon_0 E_0$

$$(a) E_{on} = E_0 \cos 30^\circ = 0.866 E_0$$

$$E_{ot} = E_0 \sin 30^\circ = 0.5 E_0$$

$$D_{on} = \epsilon_0 E_{on} = 0.866 \epsilon_0 E_0$$

$$D_{ot} = \epsilon_0 E_{ot} = 0.5 \epsilon_0 E_0$$

From B.C.s.

$$E_{1t} = E_{ot} = 0.5 E_0$$

$$D_{1t} = \epsilon_1 E_{1t} = 0.5(4) \epsilon_0 E_0$$

$$D_{1t} = 2 \epsilon_0 E_0$$

$$D_{in} = D_{on} = 0.866 \epsilon_0 E_0$$

$$E_{in} = \frac{D_{in}}{\epsilon_1} = \frac{0.866 \epsilon_0 E_0}{4 \epsilon_0}$$

$$E_m = 0.2165 E_0$$

$$E_1 = (0.5 \hat{a}_t + 0.2165 \hat{a}_n) E_0$$

$$E_1 = \sqrt{(E_{1t})^2 + (E_{in})^2} = \sqrt{(0.5)^2 + (0.2165)^2} E_0 = \sqrt{0.25 + 0.046875} E_0$$

$$E_1 = \sqrt{0.296875} E_0 = 0.54486 E_0$$

$$E_1 = 0.54486 E_0$$

$$D_1 = (2 \hat{a}_t + 0.866 \hat{a}_n) \epsilon_0 E_0$$

$$D_1 = \sqrt{(D_{1t})^2 + (D_{in})^2} = \sqrt{(2)^2 + (0.866)^2} \epsilon_0 E_0 = \sqrt{4 + 0.749956} \epsilon_0 E_0$$

$$D_1 = \sqrt{4.749956} \epsilon_0 E_0 = 2.17944 \epsilon_0 E_0 = 0.54486(4\epsilon_0) E_0$$

$$D_1 = 2.17944 E_0 = 0.54486(4\epsilon_0) E_0$$

$$(b) \theta_1 = \tan^{-1} \left(\frac{E_{1t}}{E_{in}} \right) = \tan^{-1} \left(\frac{0.5 E_0}{0.2165 E_0} \right) = \tan^{-1} (2.309) = 66.587^\circ$$

$$\theta_1 = \tan^{-1} \left(\frac{D_{1t}}{D_{in}} \right) = \tan^{-1} \left(\frac{2 \epsilon_0 E_0}{0.866 \epsilon_0 E_0} \right) = \tan^{-1} \left(\frac{2}{0.866} \right) = \tan^{-1} (2.308) = 66.581^\circ$$

$$\theta_1 = 66.571^\circ$$

$$[1.7] B_0 = \mu_0 H_0$$

$$H_{0x} = H_0 \cos 30^\circ = 0.866 H_0$$

$$H_{0y} = H_0 \sin 30^\circ = 0.5 H_0$$

$$B_{0x} = \mu_0 H_{0x} = 0.866 \mu_0 H_0$$

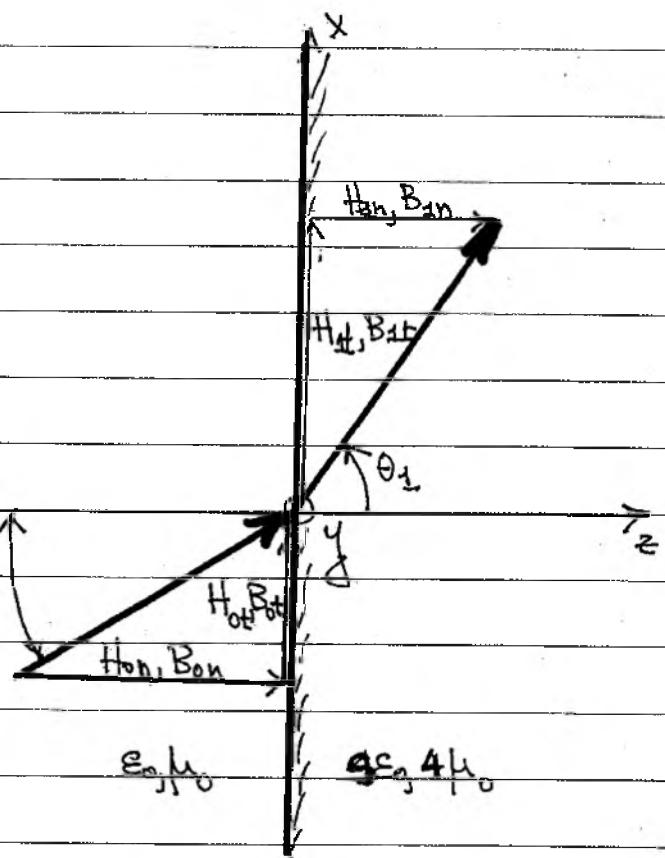
$$B_{0y} = \mu_0 H_{0y} = 0.5 \mu_0 H_0$$

From B.C.s.

$$H_{1x} = H_{0x} = 0.5 H_0$$

$$B_{1x} = \mu_0 H_{1x} = 0.5 \mu_0 H_0$$

$$B_{1y} = 2 \mu_0 H_0$$



$$B_{0x} = B_{0y} = 0.866 \mu_0 H_0$$

$$H_{1x} = B_{1x} = 0.866 \mu_0 H_0$$

$$H_{1x} = 0.2165 H_0$$

$$H_1 = (\sqrt{0.5^2 + 0.2165^2}) H_0$$

$$H_1 = \sqrt{(H_{1x})^2 + (H_{1y})^2} = \sqrt{(0.5)^2 + (0.2165)^2} H_0 = \sqrt{0.296875} H_0$$

$$H_1 = 0.54486 H_0$$

$$B_1 = (\sqrt{2^2 + 0.866^2}) \mu_0 H_0$$

$$B_1 = \sqrt{(B_{1x})^2 + (B_{1y})^2} = \sqrt{(2)^2 + (0.866)^2} \mu_0 H_0 = \sqrt{4.751} \mu_0 H_0$$

$$B_1 = 2.18 \mu_0 H_0 = 0.545(4\mu_0) H_0$$

$$b. \quad \theta_1 = \tan^{-1} \left(\frac{H_{1x}}{H_{1y}} \right) = \tan^{-1} \left(\frac{0.5 H_0}{0.2165 H_0} \right) = \tan^{-1} (2.309) = 66.587^\circ$$

$$\Theta_1 = 66.587^\circ$$

$$\theta_1 = \tan^{-1} \left(\frac{B_{1x}}{B_{1y}} \right) = \tan^{-1} \left(\frac{2 \mu_0 H_0}{0.866 \mu_0 H_0} \right) = \tan^{-1} (2.308) = 66.571^\circ$$

$$\Theta_1 = 66.571^\circ$$

1.8 Snell's Law of Refraction $\epsilon_1 = 1$

$$(a) \beta_1 \sin\theta_1 = \beta_2 \sin\theta_2$$

$$w\sqrt{\epsilon_1} \sin\theta_1 = w\sqrt{\epsilon_2} \sin\theta_2$$

Since $\mu_1 = \mu_2 = \mu_0$

$$\sqrt{\epsilon_1} \sin\theta_1 = \sqrt{\epsilon_2} \sin\theta_2$$

$$\sin\theta_2 = \sqrt{\frac{\epsilon_1}{\epsilon_2}} \sin\theta_1$$

$$= \sqrt{\frac{1}{\epsilon_2/\epsilon_1}} \sin\theta_1$$

$$\sin\theta_2 = \sqrt{\frac{1}{4}} \sin\theta_1 = \frac{1}{2} \sin(30^\circ) = \frac{1}{2} \left(\frac{1}{2}\right) = \frac{1}{4}$$

$$\theta_2 = \sin^{-1}\left(\frac{1}{4}\right) = 14.4775^\circ = \theta_4$$

$$\beta_2 \sin\theta_4 = \beta_3 \sin\alpha \Rightarrow w\sqrt{\epsilon_2} \sin\theta_4 = w\sqrt{\epsilon_3} \sin\alpha$$

Since $\mu_2 = \mu_3 = \mu_0$:

$$\sqrt{\epsilon_3} \sin\alpha = \sqrt{\epsilon_2} \sin\theta_4$$

$$\alpha = \sin^{-1}\left[\sqrt{\frac{\epsilon_2}{\epsilon_3}} \sin\theta_4\right] = \sin^{-1}\left(\sqrt{\frac{4}{9}} \sin\theta_4\right) = \sin^{-1}\left(\frac{2}{3} \sin(14.4775^\circ)\right)$$

$$\alpha = \sin^{-1}\left(\frac{2}{3}(0.25)\right) = \sin^{-1}(0.1667)$$

$$\alpha = 9.594^\circ$$

$$b) \tan\theta_2 = \frac{(h-3)}{6} = \tan(14.4775^\circ)$$

$$\frac{h-3}{6} = (0.2582) \Rightarrow h-3 = 6(0.2582)$$

$$h = 3 + 6(0.2582) = 4.5492 \text{ cm}$$

$$\boxed{h = 4.5492 \text{ cm}}$$

1.9

$$\underline{D} = \epsilon_0 \underline{E}, \quad Q_e = \oint_S \underline{D} \cdot d\underline{s}$$

$$Q_e = \iint_S \underline{D} \cdot d\underline{s} = \epsilon_0 \iint_{z=0}^h [\hat{a}_x E_x] \cdot (-\hat{a}_z ds) + \epsilon_0 \iint_{z=h} [\hat{a}_x E_x] \cdot \hat{a}_z ds$$

$$Q_e = -\epsilon_0 \iint_S \left[-\frac{c}{h} - \frac{bh^2}{6\epsilon_0} \right] ds + \epsilon_0 \iint_S \left[-\frac{c}{h} + 2\frac{bh^2}{6\epsilon_0} \right] ds = \frac{3h^2 b}{6} (\pi a^2) = \frac{\pi}{2} b (ha)^2$$

$$[1.10] \quad \epsilon_r = 4, \mu_r = 9, a = 4 \text{ cm}$$

$$\underline{H} = 3\hat{a}_p + 6\hat{a}_\phi + 8\hat{a}_z$$

$$(a) \quad \underline{B} = \mu \underline{H} = \mu_r \mu_0 \underline{H} = 9 \mu_0 (3\hat{a}_p + 6\hat{a}_\phi + 8\hat{a}_z) = \mu_0 (27\hat{a}_p + 54\hat{a}_\phi + 72\hat{a}_z)$$

$$\underline{B} = \mu_0 (27\hat{a}_p + 54\hat{a}_\phi + 72\hat{a}_z)$$

$$(b) \quad \underline{H}_o = \hat{a}_p H_{po} + \hat{a}_\phi H_{\phi o} + \hat{a}_z H_{zo}$$

$$H_{\phi o} = H_\phi = 6$$

$$H_{zo} = H_z = 8$$

$$B_{po} = B_p = \mu_0 H_{po} = \mu H_p = \mu_r \mu_0 H_p$$

$$H_{po} = \mu_r H_p = 9(3) = 27$$

$$\underline{H}_o = (27\hat{a}_p + 6\hat{a}_\phi + 8\hat{a}_z)$$

(c)

$$\underline{B}_o = \mu_0 \underline{H}_o = \mu_0 (27\hat{a}_p + 6\hat{a}_\phi + 8\hat{a}_z)$$

[1.11] $\nabla \cdot \underline{\mathcal{E}} = 0$ for a source-free and homogeneous medium.

Thus

$$\nabla \cdot \underline{\mathcal{E}} = \left[\hat{a}_x \frac{\partial}{\partial x} + \hat{a}_y \frac{\partial}{\partial y} + \hat{a}_z \frac{\partial}{\partial z} \right] \cdot [\hat{a}_x A(x+y) + \hat{a}_y B(x-y)] \cos(\omega t) = 0$$

$$A \frac{\partial^2}{\partial x^2} (x+y) + B \frac{\partial^2}{\partial y^2} (x-y) = A(1) + B(-1) = 0 \Rightarrow A = B$$

Also

$$\underline{\mathcal{E}} = \hat{a}_x A(x+y) + \hat{a}_y B(x-y)$$

$$\nabla \times \underline{\mathcal{E}} = -j\omega \mu \underline{H} \Rightarrow \underline{H} = -\frac{1}{j\omega \mu} \begin{vmatrix} \hat{a}_x & \hat{a}_y & \hat{a}_z \\ \frac{\partial}{\partial x} & \frac{\partial}{\partial y} & \frac{\partial}{\partial z} \\ A(x+y) & B(x-y) & 0 \end{vmatrix}$$

$$\underline{H} = -\frac{1}{j\omega \mu} [\hat{a}_x(0) + \hat{a}_y(0) + \hat{a}_z(B-A)]$$

$$\nabla \times \underline{H} = j\omega \epsilon \underline{\mathcal{E}} \Rightarrow \underline{\mathcal{E}} = 0 \Rightarrow A = B = 0$$

$$[1.12] \quad \underline{B} = \hat{a}_z \frac{10^{-12}}{1+25\rho} \cos(1500\pi t)$$

$$\text{a. } \Psi_m = \iint_{S_0} \underline{B} \cdot d\underline{s} = \int_0^{2\pi} \int_0^a \hat{a}_z B_z \cdot \hat{a}_z \rho d\rho d\phi = \int_0^{2\pi} \int_0^a B_z \rho d\rho d\phi \\ = 2\pi \int_0^a B_z \rho d\rho = 2\pi \times 10^{-12} \cos(1500\pi t) \int_0^a \frac{\rho}{1+25\rho} d\rho$$

Using the integral

$$\int \frac{u}{a+bu} du = \frac{1}{b^2} [a + bu - a \ln(a+bu)]$$

we can write the flux as

$$\Psi_m = 2\pi \times 10^{-12} \cos(1500\pi t) \left\{ \frac{1}{(25)^2} \left[1 + 25\rho - \ln(1+25\rho) \right]_0^a \right\}$$

$$= 2\pi \times 10^{-12} \cos(1500\pi t) \left\{ \frac{1}{625} [25a - \ln(1+25a)] \right\}$$

$$\Psi_m = 2\pi \times 10^{-12} \cos(1500\pi t) \left\{ \frac{1}{625} [2.5 - \ln(2.5)] \right\} = 1.2539 \times 10^{-14} \cos(1500\pi t)$$

$$\text{b. } \oint \underline{\phi} \cdot d\underline{l} = - \frac{\partial \Psi}{\partial t}$$

$$\oint_{C} (\hat{a}_\phi \underline{\mathcal{E}}_\phi) \cdot \hat{a}_\phi \rho d\phi = - \frac{\partial}{\partial t} \int_0^{2\pi} \int_0^a \underline{B} \cdot d\underline{s} = - \frac{\partial}{\partial t} \int_0^{2\pi} \int_0^a (\hat{a}_z \underline{B}_z) \cdot \hat{a}_z \rho d\rho d\phi$$

$$2\pi \rho \underline{\mathcal{E}}_\phi = -2\pi \frac{\partial}{\partial t} \int_0^a B_z \rho d\rho = 2\pi \times 10^{-12} (1500\pi) \sin(1500\pi t) \int_0^a \frac{\rho}{1+25\rho} d\rho$$

$$\rho \underline{\mathcal{E}}_\phi = 1500\pi \times 10^{-12} \sin(1500\pi t) \left\{ \frac{1}{(25)^2} \left[1 + 25\rho - \ln(1+25\rho) \right]_0^a \right\}$$

$$\underline{\mathcal{E}}_\phi = \frac{7.5398 \times 10^{-12}}{a} [25a - \ln(1+25a)] \sin(1500\pi t)$$

To check: Use Maxwell's equation $\nabla \times \underline{\mathcal{E}} = - \frac{\partial \underline{B}}{\partial t}$

$$\hat{a} \cdot \frac{1}{\rho} \frac{\partial}{\partial \rho} (\rho \underline{\mathcal{E}}_\phi) = - \frac{\partial}{\partial t} \left[\hat{a}_z \frac{10^{-12}}{1+25\rho} \cos(1500\pi t) \right]$$

$$\frac{1}{\rho} \frac{\partial}{\partial \rho} \left\{ 7.5398 \times 10^{-12} [25\rho - \ln(1+25\rho)] \sin(1500\pi t) \right\} = 1500\pi \times 10^{-12} \sin(1500\pi t) \frac{1}{1+25\rho} \\ \frac{4.712 \times 10^{-9}}{1+25\rho} \sin(1500\pi t) = \frac{4.712 \times 10^{-9}}{1+25\rho} \sin(1500\pi t) \quad \text{QED}$$

$$[1.13] \quad \underline{B} = \hat{a}_x B_x \cos(2y) \sin(wt - \pi z) + \hat{a}_y B_y \cos(2x) \cos(wt - \pi z)$$

$$\nabla \times \underline{H} = \hat{x} \frac{\partial}{\partial z} + \hat{y} \frac{\partial}{\partial x} + \hat{z} \cdot \underline{J}_d = \underline{J}_d = \frac{\partial \underline{B}}{\partial t}$$

$$\underline{J}_d = \nabla \times \underline{H} = \frac{1}{\mu_0} \nabla \times \underline{B} = \frac{1}{\mu_0} \left\{ -\hat{a}_x \frac{\partial B_y}{\partial z} + \hat{a}_y \frac{\partial B_x}{\partial z} + \hat{a}_z \left[\frac{\partial B_y}{\partial x} - \frac{\partial B_x}{\partial y} \right] \right\}$$

$$= \frac{1}{\mu_0} \left\{ -\hat{a}_x \pi B_y \cos(2x) \sin(wt - \pi z) - \hat{a}_y \pi B_x \cos(2y) \cos(wt - \pi z) + \hat{a}_z \left[-2B_y \sin(2x) \cos(wt - \pi z) + 2B_x \sin(2y) \sin(wt - \pi z) \right] \right\}$$

$$\underline{J}_d = -\hat{a}_x 2.5 \times 10^6 B_y \cos(2x) \sin(wt - \pi z) - \hat{a}_y 2.5 \times 10^6 B_x \cos(2y) \cos(wt - \pi z) + \hat{a}_z \left[-1.59 \times 10^6 B_y \sin(2x) \cos(wt - \pi z) + 1.59 \times 10^6 B_x \sin(2y) \sin(wt - \pi z) \right]$$

$$[1.14] \quad \underline{J}_d = \hat{a}_x y z + \hat{a}_y y^2 + \hat{a}_z x y z, \quad I_d = \iint_S \underline{J}_d \cdot d\underline{s}$$

$$\begin{aligned} I_d &= \int_{-1/2}^{1/2} \int_{-1/2}^{1/2} \left| (\hat{a}_x y z) \right| \cdot (-\hat{a}_x dy dz) + \int_{-1/2}^{1/2} \int_{-1/2}^{1/2} \left| (\hat{a}_x y z) \right| \cdot (\hat{a}_x dy dz) \\ &\quad + \int_{-1/2}^{1/2} \int_{-1/2}^{1/2} \left| (\hat{a}_y y^2) \right| \cdot (-\hat{a}_y dx dz) + \int_{-1/2}^{1/2} \int_{-1/2}^{1/2} \left| (\hat{a}_y y^2) \right| \cdot (\hat{a}_y dx dz) \\ &\quad + \int_{-1/2}^{1/2} \int_{-1/2}^{1/2} \left| (\hat{a}_z x y z) \right| \cdot (-\hat{a}_z dx dy) + \int_{-1/2}^{1/2} \int_{-1/2}^{1/2} \left| (\hat{a}_z x y z) \right| \cdot (\hat{a}_z dx dy) \end{aligned}$$

$$I_d = 0 + 0 + 0 = 0$$

$$[1.15] \quad \underline{D} = \hat{a}_r \frac{10^{-9}}{4\pi} \frac{1}{r^2} \cos(wt - \beta r) = \operatorname{Re} \left[\hat{a}_r \frac{10^{-9}}{4\pi r^2} e^{j(wt - \beta r)} \right] = \operatorname{Re} \left[\underline{D} e^{jwt} \right]$$

$$\text{where } \underline{D} = \hat{a}_r \frac{10^{-9}}{4\pi r^2} e^{-j\beta r}, \quad \beta = \omega \sqrt{\mu_0 \epsilon_0}$$

$$Q_e = \iint_S \underline{D} \cdot d\underline{s} = \int_0^{2\pi} \int_0^\pi \int_0^r \left(\hat{a}_r D_r \right) \cdot \hat{a}_r r^2 \sin\theta d\theta d\phi = \frac{10^{-9}}{4\pi} e^{-j\beta r} \int_0^{2\pi} \int_0^\pi \sin\theta d\theta d\phi$$

$$Q_e = \frac{10^{-9}}{4\pi} e^{-j\beta r} 2\pi (-\cos\theta)_0^\pi = 10^{-9} e^{-j\beta r}$$

$$|Q_e| = 10^{-9} \text{ Coulombs}$$

$$1.16 \quad \underline{\underline{E}} = \operatorname{Re} [\underline{\underline{E}}(r, \theta) e^{j\omega t}] = \operatorname{Re} [\hat{a}_\phi E_0 \sin \theta \frac{e^{-j\beta_0 r}}{r} e^{j\omega t}] = \hat{a}_\phi E_0 \sin \theta \frac{\cos(\omega t - \beta_0 r)}{r}$$

where $\underline{\underline{E}}(r, \theta) = \hat{a}_\phi E_0 \sin \theta \frac{e^{-j\beta_0 r}}{r}$

Using Maxwell's equation

$$\underline{\underline{H}} = -\frac{1}{j\omega\mu_0} \nabla \times \underline{\underline{E}} = -\frac{1}{j\omega\mu_0} \left[\hat{a}_y \frac{1}{r \sin \theta} \frac{\partial}{\partial \theta} (\underline{\underline{E}} \cdot \hat{a}_\phi \sin \theta) - \hat{a}_\phi \frac{1}{r} \frac{\partial}{\partial r} (r \underline{\underline{E}} \cdot \hat{a}_y) \right]$$

$$= -\hat{a}_y \frac{2 E_0}{j\omega\mu_0} \cos \theta \frac{e^{-j\beta_0 r}}{r^2} - \hat{a}_\phi \sqrt{\frac{\epsilon_0}{\mu_0}} E_0 \sin \theta \frac{e^{-j\beta_0 r}}{r}$$

$$\underline{\underline{H}} \approx -\hat{a}_\phi \sqrt{\frac{\epsilon_0}{\mu_0}} E_0 \sin \theta \frac{e^{-j\beta_0 r}}{r} \Rightarrow \underline{\underline{H}} = \operatorname{Re} [\underline{\underline{H}} e^{j\omega t}] \approx -\hat{a}_\phi \sqrt{\frac{\epsilon_0}{\mu_0}} E_0 \sin \theta \frac{\cos(\omega t - \beta_0 r)}{r}$$

$$1.17 \quad V(t) = 10 \cos(\omega t), \quad \underline{\underline{J}}_c = \sigma \underline{\underline{E}}, \quad \underline{\underline{J}}_d = \epsilon \frac{\partial \underline{\underline{E}}}{\partial t} = \epsilon \frac{\partial \underline{\underline{E}}}{\partial t}$$

$$\underline{\underline{E}}(t) = \frac{10}{2 \times 10^{-2}} \cos(\omega t) = 500 \cos(\omega t), \quad \frac{\partial \underline{\underline{E}}}{\partial t} = -500 \omega \sin(\omega t)$$

a. $f = 1 \text{ MHz}$

$$|\underline{\underline{J}}_c|_{\max} = |\sigma \underline{\underline{E}}|_{\max} = (3.7 \times 10^{-4}) 500 = 0.185$$

$$|\underline{\underline{J}}_d|_{\max} = \left| \epsilon \frac{\partial \underline{\underline{E}}}{\partial t} \right|_{\max} = \left| -500 \omega \epsilon \sin(\omega t) \right|_{\max} = \frac{2.56}{36} = 0.07111$$

b. $f = 100 \text{ MHz}$

$$|\underline{\underline{J}}_c|_{\max} = |\sigma \underline{\underline{E}}|_{\max} = (3.7 \times 10^{-4}) 500 = 0.185$$

$$|\underline{\underline{J}}_d|_{\max} = \left| \epsilon \frac{\partial \underline{\underline{E}}}{\partial t} \right|_{\max} = \frac{2.56 (100)}{36} = 7.111$$

$$1.18 \quad \underline{\underline{E}} = [\hat{a}_y 5 + \hat{a}_z 10] \cos(\omega t - \beta x) = \operatorname{Re} [(\hat{a}_y 5 + \hat{a}_z 10) e^{j(\omega t - \beta x)}] = \operatorname{Re} [\underline{\underline{E}} e^{j\omega t}]$$

where $\underline{\underline{E}} = (\hat{a}_y 5 + \hat{a}_z 10) e^{-j\beta x}$

a. $\underline{\underline{H}} = -\frac{1}{j\omega\mu_0} \nabla \times \underline{\underline{E}} = \hat{a}_y \left(-\frac{\partial E_3}{\partial x} \right) + \hat{a}_z \left(\frac{\partial E_4}{\partial x} \right) = 4.244 \times 10^{-3} (-\hat{a}_y 10 + \hat{a}_z 5) e^{-j\beta x}$

$$\underline{\underline{H}} = \operatorname{Re} [\underline{\underline{H}} e^{j\omega t}] = 4.244 \times 10^{-3} (-\hat{a}_y 10 + \hat{a}_z 5) \cos(\omega t - \beta x)$$

(continued)

1.18 cont'd. The tangential components of the electric and magnetic fields must be continuous across the boundaries.

The normal components of the electric and magnetic fields must be discontinuous across the boundaries.

b. $E_x^o(y=h^+) = 10 \cos(\omega t - \beta x); E_y^o(y=h^+) = \frac{\epsilon}{\epsilon_0}(5) \cos(\omega t - \beta x) = 12.8 \cos(\omega t - \beta x)$
 $\underline{E}^o(y=h^+) = (\hat{a}_y 12.8 + \hat{a}_z 10) \cos(\omega t - \beta x)$

$$\underline{H}_x^o(y=h^+) = 4.244 \times 10^{-3}(5) \cos(\omega t - \beta x); \underline{H}_y^o(y=h^+) = 4.244 \times 10^{-3}(-10) \cos(\omega t - \beta x)$$

$$\underline{H}^o(y=h^+) = 4.244 \times 10^{-3}(-\hat{a}_y 10 + \hat{a}_z 5) \cos(\omega t - \beta x)$$

In a similar manner

$$\underline{E}^o(y=-h^-) = (\hat{a}_y 12.8 + \hat{a}_z 10) \cos(\omega t - \beta x)$$

$$\underline{H}^o(y=-h^-) = 4.244 \times 10^{-3}(-\hat{a}_y 10 + \hat{a}_z 5) \cos(\omega t - \beta x)$$

1.19

$$\underline{J} = \hat{a}_z 10^4 e^{-10^6 y} \cos(2\pi \times 10^9 t)$$

$$\underline{J}(y=0.25 \times 10^{-3}) = \hat{a}_z 10^4 e^{-10^6 (2.5 \times 10^{-4})} \cos(2\pi \times 10^9 t) = \hat{a}_z 10^4 e^{-250} \cos(2\pi \times 10^9 t) \approx 0$$

$$I = \iint_S \underline{J} \cdot d\underline{s} \approx 2 \int_0^{2.5 \times 10^{-4}} \int_0^{5 \times 10^{-3}} [\hat{a}_z 10^4 e^{-10^6 y} \cos(2\pi \times 10^9 t)] \cdot \hat{a}_z dy dx$$

$$I \approx 2(5 \times 10^{-3})(10^4) \cos(2\pi \times 10^9 t) \int_0^{2.5 \times 10^{-4}} e^{-10^6 y} dy = 10^{-4} \cos(2\pi \times 10^9 t)$$

1.20 a. $\underline{E} = \hat{a}_y E_0 \sin\left(\frac{\pi}{\alpha} x\right) \cos(\omega t - \beta z) = \operatorname{Re}[\hat{a}_y E_0 \sin\left(\frac{\pi}{\alpha} x\right) e^{j(\omega t - \beta z)}] = \operatorname{Re}[E e^{j\omega t}]$

where $\underline{E} = \hat{a}_y E_0 \sin\left(\frac{\pi}{\alpha} x\right) e^{-j\beta z}$

$$\underline{H} = -\frac{1}{j\omega\mu_0} \nabla \times \underline{E} = \hat{a}_x \frac{1}{j\omega\mu_0} \frac{\partial E_y}{\partial z} - \hat{a}_z \frac{1}{j\omega\mu_0} \frac{\partial E_y}{\partial x}$$

$$= -\hat{a}_x \frac{\beta_z}{\omega\mu_0} E_0 \sin\left(\frac{\pi}{\alpha} x\right) e^{-j\beta z} + \hat{a}_z \frac{E_0}{j\omega\mu_0} \left(\frac{\pi}{\alpha}\right) \cos\left(\frac{\pi}{\alpha} x\right) e^{-j\beta z}$$

$$\underline{H} = \operatorname{Re}[\underline{H} e^{j\omega t}] = -\hat{a}_x \frac{\beta_z}{\omega\mu_0} E_0 \sin\left(\frac{\pi}{\alpha} x\right) \cos(\omega t - \beta z) + \hat{a}_z \frac{E_0}{\omega\mu_0} \left(\frac{\pi}{\alpha}\right) \cos\left(\frac{\pi}{\alpha} x\right) \cos(\omega t + \frac{\pi}{2} - \beta z)$$

b. Using $\nabla \times \underline{H} = j\omega\epsilon_0 \underline{E} \Rightarrow \hat{a}_y \left(\frac{\partial H_x}{\partial z} - \frac{\partial H_z}{\partial x} \right) = j\omega\epsilon_0 E_0 \sin\left(\frac{\pi}{\alpha} x\right) e^{-j\beta z}$

$$\hat{a}_y j \frac{E_0}{\omega\mu_0} \left[\beta_z^2 + \left(\frac{\pi}{\alpha}\right)^2 \right] \sin\left(\frac{\pi}{\alpha} x\right) e^{-j\beta z} = j\omega\epsilon_0 E_0 \sin\left(\frac{\pi}{\alpha} x\right) e^{-j\beta z}$$

$$\omega\epsilon_0 = \frac{1}{\omega\mu_0} \left[\beta_z^2 + \left(\frac{\pi}{\alpha}\right)^2 \right] \Rightarrow \beta_z = \pm \sqrt{\omega^2\mu_0\epsilon_0 - \left(\frac{\pi}{\alpha}\right)^2}$$

$$1.21 \quad \underline{E} = \hat{a}_p \left(\frac{100}{\rho} \right) \cos(10^8 t - \beta z) = \operatorname{Re} \left[\hat{a}_p \frac{100}{\rho} e^{j(10^8 t - \beta z)} \right] = \operatorname{Re} \left[\underline{E} e^{j\omega t} \right]$$

where $\underline{E} = \hat{a}_p \frac{100}{\rho} e^{-j\beta z}$

a. $\underline{H} = -\frac{1}{j\omega\mu_0} \nabla \times \underline{E} = -\frac{1}{j\omega\mu_0} \left(\hat{a}_p \frac{\partial E_p}{\partial z} \right) = \hat{a}_p \frac{\beta}{\omega\mu_0} \frac{100}{\rho} e^{-j\beta z}$

$$\underline{H} = \operatorname{Re} \left[\underline{H} e^{j\omega t} \right] = \operatorname{Re} \left[\underline{H} e^{j10^8 t} \right] = \hat{a}_p \frac{\beta}{\omega\mu_0} \left(\frac{100}{\rho} \right) \cos(10^8 t - \beta z)$$

b. $\nabla \times \underline{H} = j\omega \epsilon \underline{E} \Rightarrow -\hat{a}_p \frac{\partial H_p}{\partial z} = \hat{a}_p j\omega \epsilon \frac{100}{\rho} e^{-j\beta z}$

$$\hat{a}_p j \frac{\beta^2}{\omega\mu_0} \left(\frac{100}{\rho} \right) e^{-j\beta z} = \hat{a}_p j\omega \epsilon \left(\frac{100}{\rho} \right) e^{-j\beta z}$$

$$\omega \epsilon = \frac{\beta^2}{\omega\mu_0} \Rightarrow \beta^2 = \omega^2 \mu_0 \epsilon$$

c. $\underline{J}_d = \epsilon_0 \frac{\partial \underline{E}}{\partial t} = \epsilon_0 \frac{\partial}{\partial t} \left[\hat{a}_p \frac{100}{\rho} \cos(10^8 t - \beta z) \right] = -\hat{a}_p \frac{10^8 \epsilon_0}{\rho} \sin(10^8 t - \beta z)$
 $= -\hat{a}_p \frac{8.854 \times 10^{-2}}{\rho} \sin(10^8 t - \beta z) = -\hat{a}_p \frac{2.25 (8.854 \times 10^{-2})}{\rho} \sin(10^8 t - \beta z)$
 $\underline{J}_d = -\hat{a}_p \frac{0.1992}{\rho} \sin(10^8 t - \beta z)$

$$1.22 \quad \underline{H} = \hat{a}_p \frac{2}{\rho} \cos\left(\frac{\pi}{L} z\right) \cos(4\pi \times 10^8 t) = \operatorname{Re} \left[\hat{a}_p \frac{2}{\rho} \cos\left(\frac{\pi}{L} z\right) e^{j4\pi \times 10^8 t} \right] = \operatorname{Re} \left[\underline{H} e^{j4\pi \times 10^8 t} \right]$$

where $\underline{H} = \hat{a}_p \frac{2}{\rho} \cos\left(\frac{\pi}{L} z\right)$, $\omega = 4\pi \times 10^8$

a. $\underline{E} = \frac{1}{j\omega\epsilon_0} \nabla \times \underline{H} = -\hat{a}_p \frac{1}{j\omega\epsilon_0} \frac{\partial H_p}{\partial z} = -\hat{a}_p j \frac{1}{\omega\epsilon_0} \left(\frac{\pi}{L} \right) \frac{2}{\rho} \sin\left(\frac{\pi}{L} z\right) \sin(4\pi \times 10^8 t)$

$$\underline{E} = \operatorname{Re} \left[\underline{E} e^{j\omega t} \right] = \hat{a}_p \frac{1}{\omega\epsilon_0} \left(\frac{\pi}{L} \right) \frac{2}{\rho} \sin\left(\frac{\pi}{L} z\right) \sin(4\pi \times 10^8 t)$$

b. $\underline{J}_s = \hat{n} \times (\underline{H}_2 - \underline{H}_1)$: At $\rho = a$: $\underline{J}_s = \hat{a}_p \times (\hat{a}_p \underline{H}_p) \Big|_{\rho=a} = \hat{a}_p \frac{2}{a} \cos\left(\frac{\pi}{L} z\right) \cos(4\pi \times 10^8 t)$

At $\rho = b$: $\underline{J}_s = -\hat{a}_p \times (\hat{a}_p \underline{H}_p) \Big|_{\rho=b} = -\hat{a}_p \frac{2}{b} \cos\left(\frac{\pi}{L} z\right) \cos(4\pi \times 10^8 t)$

c. $\underline{J}_d = \epsilon_0 \frac{\partial \underline{E}_0}{\partial t} = \hat{a}_p \frac{4\pi \times 10^8}{\omega} \left(\frac{\pi}{L} \right) \frac{2}{\rho} \sin\left(\frac{\pi}{L} z\right) \cos(4\pi \times 10^8 t) = \hat{a}_p \frac{\pi}{L} \left(\frac{2}{\rho} \right) \sin\left(\frac{\pi}{L} z\right) \cos(4\pi \times 10^8 t)$

d. $I_d = \int_0^L \int_0^{\frac{\pi}{L}} \underline{J}_d \cdot d\underline{z} = \int_0^L \int_0^{\frac{\pi}{L}} \left(\hat{a}_p \underline{J}_d \right) \cdot \hat{a}_p \rho d\phi dz = \frac{\pi}{L} (4\pi) \left[-\frac{2}{\pi} \cos\left(\frac{\pi}{L} z\right) \right]_0^L \cos(4\pi \times 10^8 t)$

$$I_d = 8\pi \cos(4\pi \times 10^8 t)$$

$$1.23 \quad \nabla \times \underline{E} = -\underline{M}_i - \frac{\partial \underline{B}}{\partial t}$$

Defining $\underline{E} = \operatorname{Re}[\underline{E} e^{j\omega t}]$
 $\underline{M}_i = \operatorname{Re}[\underline{M}_i e^{j\omega t}]$
 $\underline{B} = \operatorname{Re}[\underline{B} e^{j\omega t}]$

Thus

$$\nabla \times (\operatorname{Re}[\underline{E} e^{j\omega t}]) = -\operatorname{Re}[\underline{M}_i e^{j\omega t}] - \frac{\partial}{\partial t}(\operatorname{Re}[\underline{B} e^{j\omega t}])$$

Interchanging differentiation with the Real part leads to

or $\operatorname{Re}[\nabla \times (\underline{E} e^{j\omega t})] = \operatorname{Re}[-\underline{M}_i e^{j\omega t}] + \operatorname{Re}\left[-\frac{\partial}{\partial t}(\underline{B} e^{j\omega t})\right]$
 $\operatorname{Re}[(\nabla \times \underline{E}) e^{j\omega t}] = \operatorname{Re}[-\underline{M}_i e^{j\omega t}] + \operatorname{Re}[-j\omega \underline{B} e^{j\omega t}]$

Lemma: If \underline{A} and \underline{B} are complex quantities and

then $\operatorname{Re}[\underline{A} e^{j\omega t}] = \operatorname{Re}[\underline{B} e^{j\omega t}] \text{ for all } t$

$$\underline{A} = \underline{B}$$

Using this lemma, we can write that

$$\nabla \times \underline{E} = -\underline{M}_i - j\omega \underline{B}$$

The same procedure can be used for all the other differential form equations.

For the integral form

$$C \oint \underline{E} \cdot d\underline{l} = - \iint_S \underline{M}_i \cdot d\underline{s} - \frac{\partial}{\partial t} \iint_S \underline{B} \cdot d\underline{s}$$

Using the above definitions, we can write the integral form as

$$C \oint \operatorname{Re}[\underline{E} e^{j\omega t}] \cdot d\underline{l} = - \iint_S \operatorname{Re}[\underline{M}_i e^{j\omega t}] \cdot d\underline{s} - \frac{\partial}{\partial t} \iint_S \operatorname{Re}[\underline{B} e^{j\omega t}] \cdot d\underline{s}$$

$$\operatorname{Re}\left\{ \left[\oint \underline{E} \cdot d\underline{l} \right] e^{j\omega t} \right\} = \operatorname{Re}\left\{ \left[- \iint_S \underline{M}_i \cdot d\underline{s} \right] e^{j\omega t} \right\} + \operatorname{Re}\left\{ \left[-j\omega \iint_S \underline{B} \cdot d\underline{s} \right] e^{j\omega t} \right\}$$

Using the above lemma leads to

$$C \oint \underline{E} \cdot d\underline{l} = - \iint_S \underline{M}_i \cdot d\underline{s} - j\omega \iint_S \underline{B} \cdot d\underline{s}$$

The same procedure can be used for all the other integral form equations.

$$\begin{aligned}
 1.24 \quad \underline{E} &= \operatorname{Re}[\underline{E} e^{j\omega t}] = \operatorname{Re}[(\underline{E}_R + j\underline{E}_x)(\cos\omega t + j\sin\omega t)] \\
 &= \operatorname{Re}[(\underline{E}_R \cos\omega t - \underline{E}_x \sin\omega t) + j(\underline{E}_R \sin\omega t + \underline{E}_x \cos\omega t)] \\
 \underline{E} &= (\underline{E}_R \cos\omega t - \underline{E}_x \sin\omega t)
 \end{aligned}$$

Similarly

$$\begin{aligned}
 \underline{E} &= \frac{1}{2} [\underline{E} e^{j\omega t} + (\underline{E} e^{j\omega t})^*] = \frac{1}{2} [\underline{E} e^{j\omega t} + \underline{E}^* e^{-j\omega t}] \\
 &= \frac{1}{2} [(\underline{E}_R + j\underline{E}_x)(\cos\omega t + j\sin\omega t) + (\underline{E}_R - j\underline{E}_x)(\cos\omega t - j\sin\omega t)] \\
 \underline{E} &= \frac{1}{2} [2(\underline{E}_R \cos\omega t - \underline{E}_x \sin\omega t)] = (\underline{E}_R \cos\omega t - \underline{E}_x \sin\omega t)
 \end{aligned}$$

1.25

$$\begin{aligned}
 (a) \quad \underline{E}(z, t) &= \hat{a}_x \underline{E}_0 \sin\left[(\omega t - \beta_0 z) - \frac{\pi}{2}\right] \\
 &= \hat{a}_x \underline{E}_0 \left[\sin(\omega t - \beta_0 z) \cos\left(-\frac{\pi}{2}\right) + \cos(\omega t - \beta_0 z) \sin\left(-\frac{\pi}{2}\right) \right] \\
 &= -\hat{a}_x \underline{E}_0 \cos(\omega t - \beta_0 z) = -\hat{a}_x \underline{E}_0 \operatorname{Re}[e^{j(\omega t - \beta_0 z)}] \\
 \underline{E}(z, t) &= -\hat{a}_x \underline{E}_0 \operatorname{Re}[e^{j\omega t} e^{-j\beta_0 z}] = \operatorname{Re}[\underbrace{-\hat{a}_x \underline{E}_0 e^{-j\beta_0 z}}_{E_x(z)} e^{j\omega t}]
 \end{aligned}$$

$$\underline{E}(z) = -\hat{a}_x \underline{E}_0 e^{-j\beta_0 z}$$

$$(b) \quad \nabla \times \underline{E} = -j\omega \mu_0 \underline{H} \Rightarrow \underline{H} = -\frac{1}{j\omega \mu_0} \nabla \times \underline{E} = -\frac{1}{j\omega \mu_0} \nabla \times [-\hat{a}_x \underline{E}_0 e^{-j\beta_0 z}]$$

$$\underline{H} = -\frac{1}{j\omega \mu_0} \begin{vmatrix} \hat{a}_x & \hat{a}_y & \hat{a}_z \\ \frac{\partial}{\partial x} & \frac{\partial}{\partial y} & \frac{\partial}{\partial z} \\ \underline{E}_x & 0 & 0 \end{vmatrix} = -\frac{1}{j\omega \mu_0} \left[\hat{a}_x (0) + \hat{a}_y \left(\frac{\partial \underline{E}_x}{\partial z} \right) + \hat{a}_z \left(-\frac{\partial \underline{E}_x}{\partial y} \right) \right]$$

$$\underline{H} = -\frac{1}{j\omega \mu_0} \left[\hat{a}_y \left(\frac{\partial \underline{E}_x}{\partial z} \right) \right] = -\hat{a}_y \frac{1}{j\omega \mu_0} \frac{\partial}{\partial z} \left[-\underline{E}_0 e^{-j\beta_0 z} \right] = -\hat{a}_y \underline{E}_0 \frac{\beta_0}{\omega \mu_0} e^{-j\beta_0 z}$$

$$\underline{H} = -\hat{a}_y \underline{E}_0 \frac{\omega \sqrt{\mu_0 \epsilon_0}}{\omega \mu_0} e^{-j\beta_0 z} = -\hat{a}_y \underline{E}_0 \sqrt{\frac{\epsilon_0}{\mu_0}} e^{-j\beta_0 z} = -\hat{a}_y \underline{E}_0 \frac{1}{\eta_0} e^{-j\beta_0 z}, \eta_0 = \sqrt{\frac{\mu_0}{\epsilon_0}}$$

$$(c) \quad S_{ave} = \frac{1}{2} \operatorname{Re}(\underline{E} \times \underline{H}^*) = \frac{1}{2} \operatorname{Re} \left[(-\hat{a}_x \underline{E}_0 e^{-j\beta_0 z}) \times (-\hat{a}_y \underline{E}_0 \sqrt{\frac{\epsilon_0}{\mu_0}} e^{-j\beta_0 z})^* \right]$$

$$S_{ave} = \hat{a}_z \frac{1}{2} |\underline{E}_0|^2 \sqrt{\frac{\epsilon_0}{\mu_0}} = \hat{a}_z \frac{1}{2} \eta_0 |\underline{E}_0|^2$$

$$1.26 \quad \underline{H} = \hat{a}_\phi H_0 \frac{e^{-j\beta_0 \rho}}{\sqrt{\rho}}$$

$$\begin{aligned}\underline{E} &= \frac{1}{j\omega\epsilon_0} \nabla \times \underline{H} = \frac{1}{j\omega\epsilon_0} \left[\hat{a}_z \frac{1}{\rho} \frac{\partial}{\partial \rho} (\rho H_\phi) \right] = \frac{H_0}{j\omega\epsilon_0} \frac{\hat{a}_z}{\rho} \frac{\partial}{\partial \rho} \left[\rho^{1/2} e^{-j\beta_0 \rho} \right] \\ &= \hat{a}_z \frac{H_0}{j\omega\epsilon_0} \frac{1}{\rho} \left[\rho^{1/2} (-j\beta_0 e^{-j\beta_0 \rho}) + \frac{1}{2} \frac{1}{\rho^{1/2}} e^{-j\beta_0 \rho} \right] \\ &= \hat{a}_z H_0 \left[-\frac{\beta_0}{\omega\epsilon_0} \frac{e^{-j\beta_0 \rho}}{\sqrt{\rho}} + \frac{e^{-j\beta_0 \rho}}{j^2 \omega \epsilon_0 (\rho)^{3/2}} \right] \xrightarrow{\rho \rightarrow \text{large}} -\hat{a}_z H_0 \frac{\beta_0}{\omega\epsilon_0} \frac{e^{-j\beta_0 \rho}}{\sqrt{\rho}} \\ \underline{E} &= -\hat{a}_z H_0 \frac{\omega \sqrt{\mu_0 \epsilon_0}}{\omega \epsilon_0} \frac{e^{-j\beta_0 \rho}}{\sqrt{\rho}} = -\hat{a}_z H_0 \sqrt{\frac{\mu_0}{\epsilon_0}} \frac{e^{-j\beta_0 \rho}}{\sqrt{\rho}}\end{aligned}$$

$$1.27 \quad \underline{E} = \hat{a}_r E_r(r, \theta) + \hat{a}_\theta E_\theta(r, \theta), \quad E_r = E_0 \frac{\cos \theta}{r^2} \left(1 + \frac{1}{j\beta_0 r} \right) e^{j\beta_0 r}, \quad E_\theta = j \frac{\epsilon_0}{r} \frac{\beta_0 \sin \theta}{2r} \left[1 + \frac{1}{j\beta_0 r} - \frac{1}{(\beta_0 r)^2} \right] e^{j\beta_0 r}$$

$$\underline{H} = -\frac{1}{j\omega\mu_0} \nabla \times \underline{E} = -\frac{1}{j\omega\mu_0} \left\{ \frac{\hat{a}_\phi}{r} \left[\frac{\partial}{\partial r} (r E_\theta) - \frac{\partial E_r}{\partial \theta} \right] \right\}$$

Expanding using the above electric field components leads to

$$\underline{H} = \hat{a}_\phi j \sqrt{\frac{\epsilon_0}{\mu_0}} \frac{\beta_0 \sin \theta}{2r} \left(1 + \frac{1}{j\beta_0 r} \right) e^{-j\beta_0 r}$$

$$\text{or } H_r = H_\theta = 0$$

$$H_\phi = j \sqrt{\frac{\epsilon_0}{\mu_0}} \frac{\beta_0 \sin \theta}{2r} \left(1 + \frac{1}{j\beta_0 r} \right) e^{-j\beta_0 r}$$

1.28 $\underline{E} = \hat{a}_\phi E_0 \frac{\sin \theta}{r} \left(1 + \frac{1}{j\beta_0 r}\right) e^{-j\beta_0 r} = \hat{a}_\phi E_\phi(r, \theta)$

$$\underline{H} = -\frac{1}{j\omega\mu_0} \nabla \times \underline{E} = -\frac{1}{j\omega\mu_0} \left\{ \hat{a}_r \frac{2}{r \sin \theta} \frac{\partial}{\partial \theta} (E_\phi \sin \theta) + \frac{\hat{a}_\theta}{r} \left[\frac{\partial}{\partial r} (r E_\phi) \right] \right\}$$

Using the above electric field component leads to

$$\underline{H} = \hat{a}_r H_r + \hat{a}_\theta H_\theta \text{ where } H_r = j \frac{2 E_0 \cos \theta}{\omega \mu_0 r^2} \left(1 + \frac{1}{j\beta_0 r}\right) e^{-j\beta_0 r}$$

$$H_\theta = -\frac{E_0 \sin \theta}{\eta r} \left[1 + \frac{1}{j\beta_0 r} - \frac{1}{(\beta_0 r)^2} \right] e^{-j\beta_0 r}$$

1.29 $\underline{E} = \hat{a}_z (1+j) \sin\left(\frac{\pi}{a}x\right) \sin\left(\frac{\pi}{b}y\right)$

Using Maxwell's equation of $\nabla \times \underline{E} = -j\omega\mu_0 \underline{H}$ leads to

$$\underline{H} = -\frac{1}{j\omega\mu_0} \left[\hat{a}_x \frac{\partial E_z}{\partial y} - \hat{a}_y \frac{\partial E_z}{\partial x} \right] = -\frac{(1+j)}{j\omega\mu_0} \left[\hat{a}_x \left(\frac{\pi}{b}\right) \cos\left(\frac{\pi}{b}y\right) \sin\left(\frac{\pi}{a}x\right) - \hat{a}_y \left(\frac{\pi}{a}\right) \sin\left(\frac{\pi}{b}y\right) \cos\left(\frac{\pi}{a}x\right) \right]$$

Now using Maxwell's equation of $\nabla \times \underline{H} = \underline{J}_c + j\omega \epsilon \underline{E} = \sigma \underline{E} + j\omega \epsilon \underline{E}$ leads to

$$\nabla \times \underline{H} = -\hat{a}_z \frac{(1+j)}{j\omega\mu_0} \left[\left(\frac{\pi}{a}\right)^2 + \left(\frac{\pi}{b}\right)^2 \right] \sin\left(\frac{\pi}{b}y\right) \sin\left(\frac{\pi}{a}x\right) = \hat{a}_z (\sigma + j\omega\epsilon) (1+j) \sin\left(\frac{\pi}{b}y\right) \sin\left(\frac{\pi}{a}x\right)$$

Equating both sides, we can write that

$$(\sigma + j\omega\epsilon) = -\frac{1}{j\omega\mu_0} \left[\left(\frac{\pi}{a}\right)^2 + \left(\frac{\pi}{b}\right)^2 \right] \Rightarrow \sigma = 0$$

$$\epsilon_r = \frac{\epsilon}{\epsilon_0} = \frac{1}{\omega^2 \mu_0 \epsilon_0} \left[\left(\frac{\pi}{a}\right)^2 + \left(\frac{\pi}{b}\right)^2 \right]$$

1.30 $\underline{E}^i = \hat{a}_x e^{-j\beta_0 z} \Rightarrow \underline{H}^i = -\frac{1}{j\omega\mu_0} \nabla \times \underline{E}^i = -\hat{a}_y \frac{1}{j\omega\mu_0} \frac{\partial E_x^i}{\partial z} = \hat{a}_y \sqrt{\frac{\epsilon_0}{\mu_0}} e^{-j\beta_0 z}$

$$\underline{E}^r = -\hat{a}_x e^{+j\beta_0 z} \Rightarrow \underline{H}^r = -\hat{a}_y \frac{1}{j\omega\mu_0} \frac{\partial E_x^r}{\partial z} = \hat{a}_y \sqrt{\frac{\epsilon_0}{\mu_0}} e^{+j\beta_0 z}$$

$$\underline{J}_s = \hat{n} \times (\underline{H}_2 - \underline{H}_1) = \hat{n} \times \underline{H}_2 = -\hat{a}_z \times \hat{a}_y (H_x^i + H_x^r) = \hat{a}_x (H_x^i + H_x^r) = \hat{a}_x \sqrt{\frac{\epsilon_0}{\mu_0}} (e^{-j\beta_0 z} + e^{+j\beta_0 z})$$

$$= \hat{a}_x 2\sqrt{\frac{\epsilon_0}{\mu_0}} = \hat{a}_x \frac{2}{377} = \hat{a}_x 5.3 \times 10^{-3} \text{ A/m}$$

1.31 $\underline{E}^i = \hat{a}_y E_0 e^{-j\beta_o(x \sin \theta_i + z \cos \theta_i)}$, $\underline{E}^r = \hat{a}_y E_0 \Gamma_h e^{-j\beta_o(x \sin \theta_i - z \cos \theta_i)}$

Along the interface ($z=0$) $(\underline{E}^i + \underline{E}^r)_{z=0}^{\tan} = 0 = \hat{a}_y E_0 (1 + \Gamma_h) e^{-j\beta_o x \sin \theta_i}$

which is satisfied provided $(1 + \Gamma_h) = 0 \Rightarrow \Gamma_h = -1$

1.32 Using Maxwell's equation of $\nabla \times \underline{E} = -j\omega \mu_0 \underline{H}$, the magnetic field components corresponding to the electric fields of Problem 1.31 can be written as

$$\underline{H}^i = \frac{E_0}{\sqrt{\mu_0/E_0}} (-\hat{a}_x \cos \theta_i + \hat{a}_z \sin \theta_i) e^{-j\beta_o(x \sin \theta_i + z \cos \theta_i)}$$

$$\underline{H}^r = -\frac{E_0}{\sqrt{\mu_0/E_0}} (\hat{a}_x \cos \theta_i + \hat{a}_z \sin \theta_i) e^{-j\beta_o(x \sin \theta_i - z \cos \theta_i)}$$

b.

$$\underline{J}_s = \hat{n} \times (\underline{H}_2 - \underline{H}_1) \Big|_{z=0} = -\hat{a}_z \times (\underline{H}^i + \underline{H}^r) \Big|_{z=0} = -\hat{a}_z \times [\hat{a}_x (H^i + H^r) + \hat{a}_z (H^i + H^r)] \Big|_{z=0}$$

$$\underline{J}_s = \hat{a}_z \times \hat{a}_x (H^i + H^r) = \hat{a}_y \frac{2E_0}{\sqrt{\mu_0/E_0}} \cos \theta_i e^{-j\beta_o x \sin \theta_i}$$

1.33 $\underline{E}^i = (\hat{a}_x \cos \theta_i - \hat{a}_z \sin \theta_i) E_0 e^{-j\beta_o(x \sin \theta_i + z \cos \theta_i)}$

$$\underline{E}^r = (\hat{a}_x \cos \theta_i + \hat{a}_z \sin \theta_i) \Gamma_e e^{-j\beta_o(x \sin \theta_i - z \cos \theta_i)}$$

Along the interface ($z=0$) $(\underline{E}^i + \underline{E}^r)_{z=0}^{\tan} = 0 = \hat{a}_x \cos \theta_i E_0 (1 + \Gamma_e) e^{-j\beta_o x \sin \theta_i}$

which is satisfied provided $(1 + \Gamma_e) = 0 \Rightarrow \Gamma_e = -1$

1.34 Along the interface the normal components of the electric flux density must be continuous; that is

$$\hat{n} \cdot (\underline{D}_2 - \underline{D}_1) = \hat{n} \cdot \underline{D}_2 = -\hat{n} \cdot \hat{a}_z \epsilon_0 E_0 \sin \theta_i (1 + \Gamma_e) e^{-j\beta_o x \sin \theta_i} = 0$$

$$= -\hat{a}_z \cdot \hat{a}_z \epsilon_0 E_0 \sin \theta_i (1 + \Gamma_e) e^{-j\beta_o x \sin \theta_i} = 0$$

which is satisfied provided

$$(1 + \Gamma_e) = 0 \Rightarrow \Gamma_e = -1$$

1.35 Given the electric fields of Problem 1.33, the corresponding magnetic field components can be found using Maxwell's equation of

$$\nabla \times \underline{E} = -j\omega \mu_0 \underline{H} \Rightarrow \underline{H} = -\frac{1}{j\omega \mu_0} \nabla \times \underline{E} = -\frac{1}{j\omega \mu_0} \left[\hat{a}_x \frac{\partial E_z}{\partial y} + \hat{a}_y \left(\frac{\partial E_x}{\partial z} - \frac{\partial E_z}{\partial x} \right) + \hat{a}_z \frac{\partial E_y}{\partial x} \right]$$

For the incident field

$$E_x^i = \cos \theta_i E_0 e^{-j\beta_0 (x \sin \theta_i + z \cos \theta_i)}$$

$$E_y^i = 0$$

$$E_z^i = -\sin \theta_i E_0 e^{-j\beta_0 (x \sin \theta_i + z \cos \theta_i)}$$

Using these

$$\frac{\partial E_x^i}{\partial y} = 0, \quad \frac{\partial E_z^i}{\partial y} = 0$$

However

$$\frac{\partial E_x^i}{\partial z} = -j\beta_0 \cos^2 \theta_i E_0 e^{-j\beta_0 (x \sin \theta_i + z \cos \theta_i)}$$

$$\frac{\partial E_z^i}{\partial x} = +j\beta_0 \sin^2 \theta_i E_0 e^{-j\beta_0 (x \sin \theta_i + z \cos \theta_i)}$$

Thus we can write the incident magnetic field as

$$\underline{H}^i = -\frac{E_0}{j\omega \mu_0} \hat{a}_y (j\beta_0) (\cos^2 \theta_i + \sin^2 \theta_i) e^{-j\beta_0 (x \sin \theta_i + z \cos \theta_i)}$$

$$= \hat{a}_y \frac{\beta_0}{\omega \mu_0} E_0 e^{-j\beta_0 (x \sin \theta_i + z \cos \theta_i)} = \hat{a}_y \sqrt{\frac{\epsilon_0}{\mu_0}} E_0 e^{-j\beta_0 (x \sin \theta_i + z \cos \theta_i)}$$

Using the same procedure we can write the reflected magnetic field as

$$\underline{H}^r = -\hat{a}_y \sqrt{\frac{\epsilon_0}{\mu_0}} E_0 \Gamma_r e^{-j\beta_0 (x \sin \theta_i - z \cos \theta_i)} = \hat{a}_y \sqrt{\frac{\epsilon_0}{\mu_0}} E_0 e^{-j\beta_0 (x \sin \theta_i - z \cos \theta_i)}$$

b. $\underline{J}_s = \hat{n} \times (\underline{H}_2 - \underline{H}_1) = \hat{n} \times \underline{H}_2 = \hat{a}_z \times \hat{a}_y (\underline{H}^i + \underline{H}^r) = \hat{a}_x 2 E_0 \sqrt{\frac{\epsilon_0}{\mu_0}} e^{-j\beta_0 x \sin \theta_i}$

1.36 To determine the coefficients Γ_0 and T_0 we apply the boundary conditions along the interface at $z=0$. To do this we first find the corresponding magnetic field components. This is accomplished using Maxwell's equation of $\nabla \times \underline{E} = -j\omega \mu_0 \underline{H} \Rightarrow \underline{H}_0 = -\frac{1}{j\omega \mu_0} \nabla \times \underline{E}$. Doing (continued)

1.36 cont'd this for each component leads to

$$\underline{H}^i = \hat{a}_y \sqrt{\frac{\epsilon_0}{\mu_0}} E_0 e^{-j\beta_0 z}, \quad \underline{H}^r = -\hat{a}_y \sqrt{\frac{\epsilon_0}{\mu_0}} E_0 \Gamma_0 e^{+j\beta_0 z}, \quad \underline{H}^t = \hat{a}_y \sqrt{\frac{\epsilon}{\mu_0}} E_0 T_0 e^{-j\beta_0 z}$$

Applying the boundary conditions on the continuity of the tangential electric and magnetic fields along the interface at $z=0$ leads to

$$1 + \Gamma_0 = T_0 \quad \text{from continuity of the electric fields}$$

$$1 - \Gamma_0 = \sqrt{\epsilon_r} T_0 = \sqrt{81} T_0 \quad \text{from continuity of the magnetic fields}$$

Solving these two equations, we find that

$$T_0 = \frac{2}{1 + \sqrt{\epsilon_r}} = \frac{2}{1 + \sqrt{81}} = \frac{2}{1 + 9} = \frac{1}{5} = 0.2$$

$$\Gamma_0 = T_0 - 1 = \frac{1}{5} - 1 = -\frac{4}{5} = -0.8$$

1.37 The boundary conditions require continuity of the tangential components of the electric and magnetic fields.

$$\text{Electric Fields: } (\underline{E}^i + \underline{E}^r)_{z=0}^{\tan} = (\underline{E}^t)_{z=0}^{\tan}$$

$$1 + \Gamma_h = T_h$$

$$\text{Magnetic Fields: } (\underline{H}^i + \underline{H}^r)_{z=0}^{\tan} = (\underline{H}^t)_{z=0}^{\tan}$$

$$\cos\theta_i (-1 + \Gamma_h) \sqrt{\epsilon_0} = -\sqrt{1 - \frac{\epsilon_0}{\epsilon} \sin^2\theta_i} \sqrt{\epsilon} T_h$$

Solving these two equations leads to

$$\Gamma_h = \frac{\cos\theta_i - \sqrt{\frac{\epsilon}{\epsilon_0}} \sqrt{1 - \frac{\epsilon_0}{\epsilon} \sin^2\theta_i}}{\cos\theta_i + \sqrt{\frac{\epsilon}{\epsilon_0}} \sqrt{1 - \frac{\epsilon_0}{\epsilon} \sin^2\theta_i}}$$

$$T_h = \frac{2 \cos\theta_i}{\cos\theta_i + \sqrt{\frac{\epsilon}{\epsilon_0}} \sqrt{1 - \frac{\epsilon_0}{\epsilon} \sin^2\theta_i}}$$

1.38 The boundary conditions require continuity of the normal components of the electric flux density and magnetic flux density.

$$\text{Electric Flux Density: } (\underline{D}^i + \underline{D}^r)_{z=0}^{\text{nor}} = (\underline{D}^t)_{z=0}^{\text{nor}}$$

Since there are no normal components, this boundary condition is automatically satisfied.

$$\text{Magnetic Flux Density: } (\underline{B}^i + \underline{B}^r)_{z=0}^{\text{nor}} = (\underline{B}^t)_{z=0}^{\text{nor}}$$

$$\text{or } \sin\theta_i \mu_0 \sqrt{\epsilon_0} (1 + \Gamma_h) = \sqrt{\frac{\epsilon_0}{\epsilon}} \sin\theta_i \mu_0 \sqrt{\epsilon} T_h$$

$$1 + \Gamma_h = T_h$$

This is identical to one of the equations for the solution of Problem 1.37. However we do not have another equation from the normal components of the electric field. Therefore we can not solve for Γ_h and T_h using only the normal components of Problem 1.37.

1.39 The boundary conditions require continuity of the tangential components of the electric and magnetic fields.

$$\text{Electric Fields: } (\underline{E}^i + \underline{E}^r)_{z=0}^{\text{tan}} = (\underline{E}^t)_{z=0}^{\text{tan}}$$

$$\cos\theta_i (1 + \Gamma_e) = \sqrt{1 - \frac{\epsilon_0}{\epsilon} \sin^2 \theta_i} T_e$$

$$\text{Magnetic Fields: } (\underline{H}^i + \underline{H}^r)_{z=0}^{\text{tan}} = (\underline{H}^t)_{z=0}^{\text{tan}}$$

$$\sqrt{\epsilon_0} (1 - \Gamma_e) = \sqrt{\epsilon} T_e$$

Solving these two equations leads to

$$\Gamma_e = \frac{-\cos\theta_i + \sqrt{\frac{\epsilon_0}{\epsilon}} \sqrt{1 - \frac{\epsilon_0}{\epsilon} \sin^2 \theta_i}}{\cos\theta_i + \sqrt{\frac{\epsilon_0}{\epsilon}} \sqrt{1 - \frac{\epsilon_0}{\epsilon} \sin^2 \theta_i}}$$

$$T_e = \frac{2 \sqrt{\frac{\epsilon_0}{\epsilon}} \cos\theta_i}{\cos\theta_i + \sqrt{\frac{\epsilon_0}{\epsilon}} \sqrt{1 - \frac{\epsilon_0}{\epsilon} \sin^2 \theta_i}}$$

1.40 The boundary conditions require continuity of the normal components of the electric flux density and magnetic flux density.

$$\text{Electric Flux Density: } (\underline{D}^i + \underline{D}^r)_{z=0}^{\text{nor}} = (\underline{D}^t)_{z=0}^{\text{nor}}$$

$$\sin\theta_i E_0 (-1 + \Gamma_e) = -\sqrt{\frac{\epsilon_0}{\mu_0}} \epsilon \sin\theta_i T_e$$

or

$$\sqrt{\epsilon_0} (1 - \Gamma_e) = \sqrt{\epsilon} T_e : \text{This equation is identical to one of the equations for the solution of Problem 1.37.}$$

$$\text{Magnetic Flux Density: } (\underline{B}^i + \underline{B}^r)_{z=0}^{\text{nor}} = (\underline{D}^t)_{z=0}^{\text{nor}}$$

Since there are no normal components, this boundary condition is automatically satisfied. However we only have one equation and two unknowns; therefore we can not solve for Γ_e and T_e using only the normal components of Problem 1.37.

1.41 From Problem 1.16 and at large distances

$$\underline{E} = \hat{a}_\phi E_0 \sin\theta \frac{\cos(\omega t - \beta r)}{r}, \underline{E} = \text{Re}[E e^{j\omega t}] \Rightarrow \underline{E} = E_0 \sin\theta \frac{e^{-j\beta r}}{r} \hat{a}_\phi$$

$$\underline{H} = -\hat{a}_\theta \sqrt{\frac{\epsilon_0}{\mu_0}} E_0 \sin\theta \frac{\cos(\omega t - \beta r)}{r}, \underline{H} = \text{Re}[H e^{j\omega t}] \Rightarrow \underline{H} = -\hat{a}_\theta \sqrt{\frac{\epsilon_0}{\mu_0}} E_0 \sin\theta \frac{e^{-j\beta r}}{r}$$

$$\begin{aligned} a. \quad \underline{S}_{av} &= \underline{S} = \frac{1}{2} \text{Re}(\underline{E} \times \underline{H}^*) = \frac{1}{2} \text{Re}(\hat{a}_\phi E_0 \times \hat{a}_\theta H_0^*) = \frac{1}{2} \text{Re}(\hat{a}_r E_0 H_0^*) \\ &= \hat{a}_r \frac{|E_0|^2}{2} \sqrt{\frac{\epsilon_0}{\mu_0}} \frac{\sin^2\theta}{r^2} \end{aligned}$$

$$\begin{aligned} b. \quad P_{av} &= \iint_S \underline{S}_{av} \cdot d\underline{s} = \iint_S \underline{S}_{av} \cdot \hat{a}_r r^2 \sin\theta d\theta d\phi = \frac{|E_0|^2}{2} \sqrt{\frac{\epsilon_0}{\mu_0}} \int_0^{2\pi} \int_0^\pi \sin^3\theta d\theta d\phi = \frac{|E_0|^2}{2} \sqrt{\frac{\epsilon_0}{\mu_0}} (2\pi) \left(\frac{4}{3}\right) \\ &= |E_0|^2 \frac{4\pi}{3} \sqrt{\frac{\epsilon_0}{\mu_0}} \end{aligned}$$

$$\begin{aligned} \boxed{1.42} \quad \underline{H}^{inc} &= \frac{1}{377} (-\hat{a}_x \cos\theta_i + \hat{a}_z \sin\theta_i), \underline{H}^{ref} = \frac{1}{377} (-\hat{a}_x \cos\theta_i - \hat{a}_z \sin\theta_i) \\ \underline{H}^{total} &= \underline{H}^{inc} + \underline{H}^{ref} = -\hat{a}_x \frac{2}{377} \cos\theta_i = -\hat{a}_x 5.31 \times 10^{-3} \cos\theta_i \\ \underline{T} &= \hat{a}_y \times \underline{H}^{total} = \hat{a}_y \times (-\hat{a}_x 5.31 \times 10^{-3} \cos\theta_i) = \hat{a}_z 5.31 \times 10^{-3} \cos\theta_i \end{aligned}$$

$$\boxed{1.43} \quad \underline{E} = \hat{a}_y E_0 \sin\left(\frac{\pi}{a} x\right) e^{-j\beta z}$$

$$a. \quad \underline{H} = -\frac{1}{j\omega\mu_0} \nabla \times \underline{E} = -\frac{1}{j\omega\mu_0} \left(-\hat{a}_x \frac{\partial E_y}{\partial z} + \hat{a}_z \frac{\partial E_y}{\partial x} \right) = -\hat{a}_x \frac{\beta_z}{\omega\mu_0} E_0 \sin\left(\frac{\pi}{a} x\right) e^{-j\beta z} - \hat{a}_z \frac{E_0 (\pi/a)}{j\omega\mu_0} \cos\left(\frac{\pi}{a} x\right) e^{-j\beta z}$$

$$b. \quad P_s = -\frac{1}{2} \iiint_V (\underline{H}^* \cdot \underline{M}_i + \underline{E} \cdot \underline{A}_i^*) dV = 0$$

$$c. \quad P_e = \iint_S \left(\frac{1}{2} \underline{E} \times \underline{H}^* \right) \cdot d\underline{s}$$

cont'd

$$1.43 \text{ cont'd} \quad \frac{1}{2} \underline{\underline{E}} \times \underline{\underline{H}}^* = \frac{1}{2} \hat{a}_y E_y \times (-\hat{a}_x H_x^* - \hat{a}_z H_z^*) = \frac{1}{2} (\hat{a}_z E_y H_x^* - \hat{a}_x E_y H_z^*)$$

$$\underline{\underline{S}} = \frac{1}{2} \underline{\underline{E}} \times \underline{\underline{H}}^* = \hat{a}_z \frac{\beta_2}{2\omega\mu_0} |E_0|^2 \sin^2\left(\frac{\pi}{a}x\right) + \hat{a}_x \frac{|E_0|^2}{j2\omega\mu_0} \left(\frac{\pi}{a}\right) \sin\left(\frac{\pi}{a}x\right) \cos\left(\frac{\pi}{a}x\right)$$

From the two side walls at $x=0$ and $x=a$:

$$\int_0^1 \int_{x=0}^b (\hat{a}_x S_x) | \cdot (-\hat{a}_x dy dz) + \int_0^1 \int_{x=a}^b (\hat{a}_x S_x) | \cdot (\hat{a}_x dy dz) = 0 + 0 = 0$$

From the front and back cross sections at $z=0$ and $z=a$:

$$\int_0^b \int_{z=0}^a (\hat{a}_z S_z) | \cdot (\hat{a}_z dx dy) = \frac{\beta_2}{2\omega\mu_0} |E_0|^2 \int_0^b \int_0^a \sin^2\left(\frac{\pi}{a}x\right) dx dy = \frac{\beta_2}{2\omega\mu_0} \frac{ab}{2} |E_0|^2$$

$$\int_0^b \int_{z=a}^c (\hat{a}_z S_z) | \cdot (-\hat{a}_z dx dy) = -\frac{\beta_2}{2\omega\mu_0} |E_0|^2 \int_0^b \int_a^c \sin^2\left(\frac{\pi}{a}x\right) dx dy = -\frac{\beta_2}{2\omega\mu_0} \frac{ab}{2} |E_0|^2$$

From the top and bottom walls at $y=0$ and $y=b$:

Since there are no y components of the power density, there is no contribution from the top and bottom walls.

Therefore

$$P_e = 0 + \frac{\beta_2}{2\omega\mu_0} \frac{ab}{2} |E_0|^2 - \frac{\beta_2}{2\omega\mu_0} \frac{ab}{2} |E_0|^2 = 0$$

d. $P_d = \frac{1}{2} \iiint \sigma |\underline{\underline{E}}|^2 dv = 0$

e. $\bar{W}_m = \iiint \frac{1}{4} \epsilon_0 |\underline{\underline{H}}|^2 dv = \frac{1}{4} \epsilon_0 |E_0|^2 \left\{ \int_0^1 \int_0^b \int_0^a \left(\frac{\beta_2}{\omega\mu_0} \right)^2 \sin^2\left(\frac{\pi}{a}x\right) dx dy dz + \left(\frac{1}{\omega\mu_0} \right)^2 \left(\frac{\pi}{a} \right)^2 \int_0^1 \int_0^b \int_0^a \cos^2\left(\frac{\pi}{a}x\right) dx dy dz \right\}$
 $= \frac{1}{4} \epsilon_0 |E_0|^2 \left\{ \left(\frac{\beta_2}{\omega\mu_0} \right)^2 \frac{ab}{2} + \left(\frac{1}{\omega\mu_0} \right)^2 \left(\frac{\pi}{a} \right)^2 \left(\frac{ab}{2} \right) \right\} = \frac{1}{4} \epsilon_0 |E_0|^2 \left\{ \frac{\epsilon_0 ab}{\mu_0 2} - \left(\frac{1}{\omega\mu_0} \right)^2 \left(\frac{\pi}{a} \right)^2 \frac{ab}{2} + \left(\frac{1}{\omega\mu_0} \right)^2 \left(\frac{\pi}{a} \right)^2 \frac{ab}{2} \right\}$

$$\bar{W}_m = \frac{\epsilon_0}{8} ab |E_0|^2$$

f. $\bar{W}_e = \iiint \frac{1}{4} \epsilon_0 |\underline{\underline{E}}|^2 dv = \frac{\epsilon_0}{4} |E_0|^2 \int_0^1 \int_0^b \int_0^a \sin^2\left(\frac{\pi}{a}x\right) dx dy dz = \frac{\epsilon_0}{4} |E_0|^2 \frac{ab}{2} = \frac{\epsilon_0}{8} ab |E_0|^2$

$$\bar{W}_e = \frac{\epsilon_0}{8} ab |E_0|^2$$

Ultimately $P_s = P_e + P_d + j2\omega(\bar{W}_m - \bar{W}_e)$

$$0 = 0 + 0 + j2\omega \left(\frac{\epsilon_0}{8} ab |E_0|^2 - \frac{\epsilon_0}{8} ab |E_0|^2 \right) = 0 + 0 + 0 = 0$$

QED

$$1.44 - \nabla \cdot (\frac{1}{2} \underline{\underline{\mathbf{E}}} \times \underline{\underline{\mathbf{H}}}^*) = \frac{1}{2} \underline{\underline{\mathbf{H}}}^* \cdot \underline{\underline{\mathbf{M}}}_0 + \frac{3}{2} \underline{\underline{\mathbf{E}}} \cdot \underline{\underline{\mathbf{A}}}_0^* + j 2\omega \left(\frac{1}{4} \mu_0 |\underline{\underline{\mathbf{H}}}^*|^2 - \frac{1}{4} \epsilon_0 |\underline{\underline{\mathbf{E}}}^*|^2 \right)$$

$$\frac{1}{2} (\underline{\underline{\mathbf{E}}} \times \underline{\underline{\mathbf{H}}}^*) = \hat{a}_x \frac{|\underline{\underline{\mathbf{E}}}_0|^2}{j 2\omega \mu_0} \left(\frac{\pi}{a} x \right) \sin \left(\frac{\pi}{a} x \right) \cos \left(\frac{\pi}{a} x \right) + \hat{a}_z \frac{\beta_2}{2\omega \mu_0} |\underline{\underline{\mathbf{E}}}_0|^2 \sin^2 \left(\frac{\pi}{a} x \right)$$

$$= \hat{a}_x \frac{|\underline{\underline{\mathbf{E}}}_0|^2}{j 4\omega \mu_0} \left(\frac{\pi}{a} \right) \sin \left(\frac{2\pi}{a} x \right) + \hat{a}_z \frac{\beta_2}{2\omega \mu_0} |\underline{\underline{\mathbf{E}}}_0|^2 \sin^2 \left(\frac{\pi}{a} x \right)$$

$$- \nabla \cdot \left(\frac{1}{2} \underline{\underline{\mathbf{E}}} \times \underline{\underline{\mathbf{H}}}^* \right) = - \frac{|\underline{\underline{\mathbf{E}}}_0|^2}{j 4\omega \mu_0} \left(\frac{\pi}{a} \right) \left(\frac{2\pi}{a} \right) \cos \left(\frac{2\pi}{a} x \right) = j \frac{|\underline{\underline{\mathbf{E}}}_0|^2}{2\omega \mu_0} \left(\frac{\pi}{a} \right)^2 \cos \left(\frac{2\pi}{a} x \right)$$

$$\frac{1}{4} |\underline{\underline{\mathbf{H}}}|^2 = \frac{\mu_0 |\underline{\underline{\mathbf{E}}}_0|^2}{4 (\omega \mu_0)} \left[\beta_0^2 \sin^2 \left(\frac{\pi}{a} x \right) + \left(\frac{\pi}{a} \right)^2 \cos^2 \left(\frac{\pi}{a} x \right) \right]$$

$$= \frac{\mu_0 |\underline{\underline{\mathbf{E}}}_0|^2}{4 (\omega \mu_0)} \left[\beta_0^2 \sin^2 \left(\frac{\pi}{a} x \right) + \left(\frac{\pi}{a} \right)^2 \left(\cos^2 \frac{\pi}{a} x - \sin^2 \frac{\pi}{a} x \right) \right]$$

$$= \frac{\mu_0 |\underline{\underline{\mathbf{E}}}_0|^2}{4 (\omega \mu_0)} \left[\beta_0^2 \sin^2 \left(\frac{\pi}{a} x \right) + \left(\frac{\pi}{a} \right)^2 \left(\frac{1 + \cos \left(\frac{2\pi}{a} x \right)}{2} - \frac{1 - \cos \left(\frac{2\pi}{a} x \right)}{2} \right) \right]$$

$$\frac{1}{4} |\underline{\underline{\mathbf{H}}}|^2 = \frac{|\underline{\underline{\mathbf{E}}}_0|^2}{4} \epsilon_0 \sin^2 \left(\frac{\pi}{a} x \right) + \frac{|\underline{\underline{\mathbf{E}}}_0|^2}{4} \frac{1}{\omega^2 \mu_0} \left(\frac{\pi}{a} \right)^2 \cos \left(\frac{2\pi}{a} x \right)$$

$$\frac{1}{4} \epsilon_0 |\underline{\underline{\mathbf{E}}}^*|^2 = \frac{|\underline{\underline{\mathbf{E}}}_0|^2}{4} \epsilon_0 \sin^2 \left(\frac{\pi}{a} x \right)$$

Therefore the conservation of energy equation in differential form can be written as

$$j \frac{|\underline{\underline{\mathbf{E}}}_0|^2}{2\omega \mu_0} \left(\frac{\pi}{a} \right)^2 \cos \left(\frac{2\pi}{a} x \right) = 0 + 0 + j 2\omega \left[\frac{|\underline{\underline{\mathbf{E}}}_0|^2}{4} \epsilon_0 \sin^2 \left(\frac{\pi}{a} x \right) + \frac{|\underline{\underline{\mathbf{E}}}_0|^2}{4} \frac{1}{\omega^2 \mu_0} \left(\frac{\pi}{a} \right)^2 \cos \left(\frac{2\pi}{a} x \right) - \frac{|\underline{\underline{\mathbf{E}}}_0|^2}{4} \epsilon_0 \sin^2 \left(\frac{\pi}{a} x \right) \right]$$

$$j \frac{|\underline{\underline{\mathbf{E}}}_0|^2}{2\omega \mu_0} \left(\frac{\pi}{a} \right)^2 \cos \left(\frac{2\pi}{a} x \right) = j \frac{|\underline{\underline{\mathbf{E}}}_0|^2}{2\omega \mu_0} \left(\frac{\pi}{a} \right)^2 \cos \left(\frac{2\pi}{a} x \right) \quad QED$$

1.45

Boundary Conditions on PEC:

$$(a) \underline{\underline{\mathbf{E}}}_x (0 \leq x \leq a, y=0, 0 \leq z \leq c) = \underline{\underline{\mathbf{E}}}_x (\text{bottom wall}) = 0$$

$$\underline{\underline{\mathbf{E}}}_x = (\cos(\beta_x x) \sin(\beta_y y) \sin(\beta_z z))|_{y=0} = 0$$

$$\underline{\underline{\mathbf{E}}}_x (0 \leq x \leq a, y=b, 0 \leq z \leq c) = \underline{\underline{\mathbf{E}}}_x (\text{top wall}) = 0$$

$$\underline{\underline{\mathbf{E}}}_x = \cos(\beta_x x) \sin(\beta_y b) \sin(\beta_z z) = 0$$

$$\sin(\beta_y b) = 0 \Rightarrow \beta_y b = \sin^{-1}(0) = n\pi, n = 0, \pm 1, \pm 2, \dots$$

$$\beta_y = \frac{(n\pi)}{b}, n = 0, \pm 1, \pm 2, \dots$$

Cont'd

1.45 (cont'd)

$$(b) E_y(x=0, 0 \leq y \leq b, 0 \leq z \leq c) = E_y(\text{left wall}) = 0$$

$$E_y = \sin(\beta_x 0) \cos(\beta_y y) \sin(\beta_z z) = 0$$

$$E_y(x=a, 0 \leq y \leq b, 0 \leq z \leq c) = E_y(\text{right wall}) = 0$$

$$E_y = \sin(\beta_x a) \cos(\beta_y y) \sin(\beta_z z) = 0$$

$$\sin(\beta_x a) = 0 \Rightarrow \beta_x a = \sin(0) = m\pi, m=0, \pm 1, \pm 2, \dots$$

$$\boxed{\beta_x = (m\pi/a), m=0, \pm 1, \pm 2, \dots}$$

$$c. E_y(0 \leq x \leq a, 0 \leq y \leq b, z=0) = E_y(\text{front wall}) = 0$$

$$E_y = \sin(\beta_x x) \cos(\beta_y y) \sin(0) = 0$$

$$E_y(0 \leq x \leq a, 0 \leq y \leq b, z=c) = E_y(\text{back wall}) = 0$$

$$E_y = \sin(\beta_x x) \cos(\beta_y y) \sin(\beta_z c) = 0$$

$$\sin(\beta_z c) = 0 \Rightarrow \beta_z c = \sin(0) = p\pi, p=\pm 1, \pm 2, \dots$$

$$\boxed{\beta_z = (p\pi/c) = (p\pi/c), p=\pm 1, \pm 2, \dots}$$

$\rightarrow p \neq 0$, because $p=0$ leads to trivial solution; E-fields vanish

1.46 $\underline{E} = \hat{a}_y E_0 \sin\left(\frac{\pi}{a}x\right) \sin\left(\frac{\pi}{c}z\right), \omega_r = \omega = \frac{1}{\sqrt{\mu_0 \epsilon_0}} \sqrt{\left(\frac{\pi}{a}\right)^2 + \left(\frac{\pi}{c}\right)^2}$

a. $\underline{H} = -\frac{1}{j\omega_r \mu_0} \nabla \times \underline{E} = -\frac{1}{j\omega_r \mu_0} \left[-\hat{a}_x \frac{\partial E_y}{\partial z} + \hat{a}_z \frac{\partial E_y}{\partial x} \right]$

$$= \hat{a}_x \frac{E_0}{j\omega_r \mu_0} \left(\frac{\pi}{c} \right) \sin\left(\frac{\pi}{a}x\right) \cos\left(\frac{\pi}{c}z\right) - \hat{a}_z \frac{E_0}{j\omega_r \mu_0} \left(\frac{\pi}{a} \right) \cos\left(\frac{\pi}{a}x\right) \sin\left(\frac{\pi}{c}z\right)$$

b. $P_s = -\frac{1}{2} \iint_V [\underline{H}^* \cdot \underline{M}_i + \underline{E}^* \cdot \underline{J}_i^*] dV = 0$

(continued)

1.46 cont'd.

$$c. P_e = \iint_{S_a} (\frac{1}{2} \underline{E} \times \underline{H}^*) \cdot d\underline{s} = \iint_{S_a} \underline{S} \cdot d\underline{s}, \quad \underline{S} = \frac{1}{2} \underline{E} \times \underline{H}^*$$

$$\underline{S} = \frac{1}{2} \underline{E} \times \underline{H}^* = \frac{1}{2} \hat{a}_y E_y \times (\hat{a}_x H_x^* - \hat{a}_z H_z^*) = \frac{1}{2} (-\hat{a}_x E_y H_x^* - \hat{a}_x E_y H_z^*)$$

$$= \frac{1}{2} \left[\hat{a}_x \frac{E_0}{j\omega\mu_0} \left(\frac{\pi}{a} \right) \sin \left(\frac{\pi}{a} x \right) \cos \left(\frac{\pi}{a} x \right) \sin^2 \left(\frac{\pi}{c} z \right) + \hat{a}_x \frac{|E_0|^2}{j\omega\mu_0} \left(\frac{\pi}{c} \right) \sin^2 \left(\frac{\pi}{a} x \right) \sin \left(\frac{\pi}{c} z \right) \cos \left(\frac{\pi}{c} z \right) \right]$$

Contributions to P_e from the different walls:

Left and right walls:

$$\iint_{\text{left}}^{\text{right}} (\hat{a}_x S_x) \cdot (-\hat{a}_x dy dz) = 0; \quad \iint_{\text{left}}^{\text{right}} (\hat{a}_x S_x) \cdot (\hat{a}_x dy dz) = 0$$

Back and front walls:

$$\iint_{\text{back}}^{\text{front}} (\hat{a}_z S_z) \cdot (-\hat{a}_z dx dy) = 0; \quad \iint_{\text{back}}^{\text{front}} (\hat{a}_z S_z) \cdot (\hat{a}_z dx dy) = 0$$

Top and bottom walls:

Since there are no u components of the power density, there are no contributions from the top and bottom walls.

Therefore $P_e = 0 + 0 + 0 + 0 + 0 = 0$

$$d. P_d = \frac{1}{2} \iiint \sigma |\underline{E}|^2 dv = 0$$

$$e. \bar{W}_m = \frac{1}{4} \iint \iint |\underline{H}|^2 dv = |E_0|^2 \frac{1}{4} \frac{1}{(\omega\mu_0)^2} \left[\left(\frac{\pi}{c} \right)^2 \iint_{\text{top}}^{\text{bottom}} \iint_{\text{left}}^{\text{right}} \sin^2 \left(\frac{\pi}{a} x \right) \cos^2 \left(\frac{\pi}{c} z \right) dx dy dz \right. \\ \left. + \left(\frac{\pi}{a} \right)^2 \iint_{\text{top}}^{\text{bottom}} \iint_{\text{left}}^{\text{right}} \cos^2 \left(\frac{\pi}{a} x \right) \sin^2 \left(\frac{\pi}{c} z \right) dx dy dz \right]$$

$$\bar{W}_m = |E_0|^2 \frac{1}{4} \frac{1}{(\omega\mu_0)^2} \left[\left(\frac{\pi}{c} \right)^2 \frac{abc}{4} + \left(\frac{\pi}{a} \right)^2 \frac{abc}{4} \right] = |E_0|^2 \frac{abc}{16} \frac{\epsilon_0}{\omega\mu_0} \left[\left(\frac{\pi}{a} \right)^2 + \left(\frac{\pi}{c} \right)^2 \right] = |E_0|^2 \frac{abc}{16} \epsilon_0$$

$$f. \bar{W}_e = \frac{\epsilon_0}{4} \iint \iint |\underline{E}|^2 dv = |E_0|^2 \frac{\epsilon_0}{4} \iint_{\text{top}}^{\text{bottom}} \iint_{\text{left}}^{\text{right}} \sin^2 \left(\frac{\pi}{a} x \right) \sin^2 \left(\frac{\pi}{c} z \right) dx dy dz = |E_0|^2 \frac{\epsilon_0}{4} \frac{abc}{4} = |E_0|^2 \frac{abc}{16} \epsilon_0$$

$$\text{Ultimately } P_s = P_e + P_d + j2\omega(\bar{W}_m - \bar{W}_e)$$

$$0 = 0 + 0 + j2\omega \left(|E_0|^2 \frac{abc}{16} \epsilon_0 - |E_0|^2 \frac{abc}{16} \epsilon_0 \right)$$

QED

CHAPTER 2

2.1 $\underline{P} = \hat{a}_y 2.762 \times 10^{-11} \text{ C/m}^2, \underline{\Xi} = \hat{a}_y 2 \text{ V/m}$

a. $q_{sb} = |\underline{P}| = 2.762 \times 10^{-11} \text{ C/m}^2$

$q_{sb} = +2.762 \times 10^{-11} \text{ C/m}^2 \text{ at } y=2 \text{ cm face}$

$q_{sb} = -2.762 \times 10^{-11} \text{ C/m}^2 \text{ at } y=0 \text{ cm face}$

b. $Q_p = \iint_S \underline{P} \cdot d\underline{s} = \underbrace{\iint_0^{0.04} \hat{a}_y P_y \cdot \hat{a}_y dx dz}_{y=2 \text{ cm face}} + \underbrace{\iint_0^{0.04} \hat{a}_y P_y \cdot (-\hat{a}_y dx dz)}_{y=0 \text{ cm face}} = \iint_0^{0.04} P_y dx dz - \iint_0^{0.04} P_y dx dz = 0$

c. $\iint_S \underline{P} \cdot d\underline{s} = \iiint_V \nabla \cdot \underline{P} dv = \iiint_V q_{ub} dv = Q_b$

$$q_{ub} = \nabla \cdot \underline{P} = \frac{\partial P_y}{\partial y} = 0$$

d. $\epsilon_r = \frac{\epsilon}{\epsilon_0} = \frac{(1+\chi_e)\epsilon_0}{\epsilon_0} = 1+\chi_e; \quad D = \epsilon_0 \epsilon_a + \underline{P} = (1+\chi_e) \epsilon_0 \epsilon_a$

$$\chi_e = \frac{P}{\epsilon_0 \epsilon_a} = \frac{2.762 \times 10^{-11}}{8.854 \times 10^{-12} (2)} = \frac{2.762}{1.7708} = 1.5597$$

$$\epsilon_r = 1 + 1.5597 = 2.5597 \approx 2.56$$

2.2 $\underline{P} = \hat{a}_p \frac{2}{p} \times 10^{-10} \text{ C/m}^2, \underline{\Xi} = \hat{a}_p \frac{7.53}{p} \text{ V/m}; \quad \text{as } p \leq b$

a. $q_{sb} \Big|_{p=a} = -|\underline{P}| = -\frac{2 \times 10^{-10}}{2 \times 10^{-2}} = -10^{-8} \text{ C/m}^2$

$$q_{sb} \Big|_{p=b} = +|\underline{P}| = \frac{2 \times 10^{-10}}{6 \times 10^{-2}} = \frac{1}{3} \times 10^{-8} \text{ C/m}^2$$

$$q_{sb} = 0 \text{ elsewhere}$$

b. $Q_p \Big|_{p=a} = q_{sb} A_a = -\frac{2}{p} \times 10^{-10} (2\pi p l) = -4\pi l \times 10^{-10} = -1.256 \times 10^{-10}$

$$Q_p \Big|_{p=b} = q_{sb} A_b = \frac{2}{p} \times 10^{-10} (2\pi p l) = 4\pi l \times 10^{-10} = +1.256 \times 10^{-10}$$

$$Q_{total} = Q_p \Big|_{p=a} + Q_p \Big|_{p=b} = 0$$

(continued)

2.2 cont'd.

c. $Q_p = \iiint_S \underline{P} \cdot d\underline{s} = \iiint_V \nabla \cdot \underline{P} dv = \iiint_V q_{vb} dv$

$$q_{vb} = \nabla \cdot \underline{P} = \frac{1}{\rho} \frac{\partial}{\partial \rho} (\rho P_p) = \frac{1}{\rho} \frac{\partial}{\partial \rho} \left(\rho \frac{2}{\rho} \times 10^{-10} \right) = 0$$

d. $\underline{D} = \epsilon \underline{E}_a = \epsilon_0 \underline{E}_a + \underline{P} = \epsilon_0 \underline{E}_a + \chi_e \epsilon_0 \underline{E}_a = \epsilon_0 (1 + \chi_e) \underline{E}_a$

$$\epsilon_r = \frac{\epsilon}{\epsilon_0} = (1 + \chi_e)$$

$$\chi_e = \frac{P}{\epsilon_0 E_a} = \frac{\frac{2}{\rho} \times 10^{-10}}{\epsilon_0 \frac{7.53}{\rho}} = \frac{2 \times 10^{-10}}{7.53 \epsilon_0} = 2.9998$$

$$\epsilon_r = 1 + \chi_e = 1 + 2.9998 = 3.9998 \approx 4$$

2.3 $\underline{P} = \hat{a}_r \frac{31.87}{r^2} \times 10^{-12} \text{ C/m}^2, \underline{E} = \hat{a}_r \frac{0.45}{r^2} \text{ V/m}; a \leq r \leq b$

a. $q_{sb}|_{r=a} = -|\underline{P}| = -\frac{31.87 \times 10^{-12}}{(2 \times 10^{-2})^2} = -\frac{31.87 \times 10^{-12}}{4 \times 10^{-4}} = -7.9675 \times 10^{-8} \text{ C/m}^2$

$$q_{sb}|_{r=b} = +|\underline{P}| = \frac{31.87 \times 10^{-12}}{(4 \times 10^{-2})^2} = \frac{31.87 \times 10^{-12}}{16 \times 10^{-4}} = 1.9919 \times 10^{-8} \text{ C/m}^2$$

b. $Q_p|_{r=a} = q_{sb} A_a = -\frac{31.87 \times 10^{-12}}{r^2} (4\pi r^2) = -31.87 (4\pi \times 10^{-12}) = -4.005 \times 10^{-10}$

$$Q_p|_{r=b} = q_{sb} A_b = \frac{31.87 \times 10^{-12}}{r^2} (4\pi r^2) = 31.87 (4\pi \times 10^{-12}) = 4.005 \times 10^{-10}$$

$$Q_{total} = Q_p|_{r=a} + Q_p|_{r=b} = 0$$

c. $Q_p = \iint_S \underline{P} \cdot d\underline{s} = \iiint_V \nabla \cdot \underline{P} dv = \iiint_V q_{vb} dv$

$$q_{vb} = \nabla \cdot \underline{P} = \frac{1}{r^2} \frac{\partial}{\partial r} (r^2 P_r) = \frac{1}{r^2} \frac{\partial}{\partial r} \left(r^2 \frac{31.87}{r^2} \times 10^{-12} \right) = 0$$

d. $\chi_e = \frac{P}{\epsilon_0 E_a} = \frac{\frac{31.87 \times 10^{-12}}{r^2}}{\epsilon_0 \frac{0.45}{r^2}} = \frac{31.87 \times 10^{-12}}{0.45 \epsilon_0} = 7.9989$

$$\epsilon_r = 1 + \chi_e = 1 + 7.9989 = 8.9989 \approx 9$$

2.4

$$a. E_0(0.25 \times 10^{-3}) + E_1(1 \times 10^{-3}) = 100 \text{ Volts}$$

$$\text{Also } \epsilon_0 E_0 = \epsilon_1 E_1 \Rightarrow E_1 = \frac{\epsilon_0}{\epsilon_1} E_0$$

$$\text{Thus } E_0(0.25 \times 10^{-3}) + E_0\left(\frac{10^{-3}}{5}\right) = E_0(0.25 + 0.2) \times 10^{-3} = 0.45 \times 10^{-3} E_0 = 100$$

$$E_0 = \frac{100 \times 10^3}{0.45} = 2.222 \times 10^5 \text{ V/m}$$

$$E_1 = \frac{\epsilon_0}{\epsilon_1} E_0 = \frac{2.222 \times 10^5}{5} = 0.4444 \times 10^5 \text{ V/m}$$

$$b. D_0 = \epsilon_0 E_0 = 8.854 \times 10^{-12} (2.222 \times 10^5) = 19.6755 \times 10^{-7} = 1.96755 \times 10^{-6} \text{ C/m}^2$$

$$D_1 = \epsilon_1 E_1 = 5(8.854 \times 10^{-12})(0.4444 \times 10^5) = 1.96755 \times 10^{-6} = D_0 \text{ C/m}^2$$

$$c. q_{bs} = 1.96755 \times 10^{-6} \text{ C/m}^2 \text{ in upper plate where voltage is positive}$$

$$q_{bs} = -1.96755 \times 10^{-6} \text{ C/m}^2 \text{ in lower plate where voltage is negative}$$

$$d. Q = q_{bs} A = 1.96755 \times 10^{-6} (2 \times 10^{-2}) = 3.9351 \times 10^{-8} \text{ C in upper plate}$$

$$Q = q_{bs} A = -1.96755 \times 10^{-6} (2 \times 10^{-2}) = -3.9351 \times 10^{-8} \text{ C in lower plate}$$

$$e. C_0 = \frac{Q}{V_0}, V_0 = E_0(0.25 \times 10^{-3}) = 2.222 \times 10^5 (0.25 \times 10^{-3}) = 55.5556 \text{ Volts}$$

$$C_0 = \frac{3.9351 \times 10^{-8}}{55.5556} = 7.0832 \times 10^{-10}$$

$$C_1 = \frac{Q}{V_1}, V_1 = E_1(1 \times 10^{-3}) = 44.4444 \text{ Volts}$$

$$C_1 = \frac{3.9351 \times 10^{-8}}{44.4444} = 8.854 \times 10^{-10}$$

$$C_t = \frac{Q}{V} = \frac{3.9351 \times 10^{-8}}{100} = 3.9351 \times 10^{-10} = \frac{C_0 C_1}{C_0 + C_1} = \frac{7.0832 (8.854)}{7.0832 + 8.854} \times 10^{-10}$$

$$f. W_{eo} = \frac{1}{2} C_0 V_0^2 = \frac{1}{2} (7.0832 \times 10^{-10}) (55.5556)^2 = 1.0931 \times 10^{-6} \text{ Joules}$$

$$W_{e1} = \frac{1}{2} C_1 V_1^2 = \frac{1}{2} (8.854 \times 10^{-10}) (44.4444)^2 = 0.87445 \times 10^{-6} \text{ Joules}$$

$$W_{et} = W_{eo} + W_{e1} = 1.96755 \times 10^{-6} \text{ Joules}$$

$$\text{Also } W_{et} = \frac{1}{2} C_t V^2 = \frac{1}{2} (3.9351 \times 10^{-10}) (100)^2 = 1.96755 \times 10^{-6} \text{ Joules}$$

2.5 Before the insertion of the slab the electric flux density between the plates was equal to

$$D = \epsilon_0 E_0 = 8.854 \times 10^{-12} \left(\frac{100}{1.25 \times 10^{-3}} \right) = 7.0832 \times 10^{-7} \text{ C/m}^2$$

Thus the surface charge density was equal to $|q_s| = D = 7.0832 \times 10^{-7} \text{ C/m}^2$
 c,b. After the removal of the battery and the insertion of the dielectric sheet, the charge density on the plates remains the same as before the insertion of the dielectric sheet. Thus

$q_s = +7.0832 \times 10^{-7} \text{ C/m}^2$ on the upper plate where the voltage is positive.

$$Q = q_s A = 7.0832 \times 10^{-7} (2 \times 10^{-2}) = 14.1664 \times 10^{-9} \text{ C}$$

$q_s = -7.0832 \times 10^{-7} \text{ C/m}^2$ on the lower plate where the voltage is negative.

$$Q = q_s A = -7.0832 \times 10^{-7} (2 \times 10^{-2}) = -14.1664 \times 10^{-9} \text{ C}$$

c. Using the boundary conditions on the normal components of the electric flux density

$$\hat{n} \cdot (D_2 - D_1) = q_s \text{ or } D_{2n} - D_{1n} = D_{2n} = q_s$$

Thus the electric flux density in the dielectric slab and free space is equal to

$$D_0 = D_1 = 7.0832 \times 10^{-7} \text{ C/m}^2$$

$$d. E_0 = \frac{D_0}{\epsilon_0} = \frac{7.0832 \times 10^{-7}}{8.854 \times 10^{-12}} = 80 \times 10^3 \text{ V/m}$$

$$E_1 = \frac{D_1}{\epsilon_1} = \frac{7.0832 \times 10^{-7}}{5 \epsilon_0} = 16 \times 10^3 \text{ V/m}$$

$$e. V_0 = E_0 (0.25 \times 10^{-3}) = 20 \text{ Volts}$$

$$V_1 = E_1 (1 \times 10^{-3}) = 16 \text{ Volts}$$

$$V_t = V_0 + V_1 = 20 + 16 = 36 \text{ Volts}$$

$$f. C_0 = \frac{Q}{V_0} = \frac{14.1664 \times 10^{-9}}{20} = 0.70832 \times 10^{-9} = 7.0832 \times 10^{-10}$$

$$C_1 = \frac{Q}{V_1} = \frac{14.1664 \times 10^{-9}}{16} = 0.8854 \times 10^{-9} = 8.854 \times 10^{-10}$$

$$C_t = \frac{Q}{V_t} = \frac{14.1664 \times 10^{-9}}{36} = 0.39351 \times 10^{-9} = 3.9351 \times 10^{-10} = \frac{C_0 C_1}{C_0 + C_1} = \frac{7.0832 (8.854) \times 10^{-10}}{7.0832 + 8.854}$$

$$g. W_{e0} = \frac{1}{2} C_0 V_0^2 = \frac{1}{2} (7.0832 \times 10^{-10}) (20)^2 = 0.14166 \times 10^{-6} \text{ Joules}$$

$$W_{e1} = \frac{1}{2} C_1 V_1^2 = \frac{1}{2} (8.854 \times 10^{-10}) (16)^2 = 0.11333 \times 10^{-6} \text{ Joules}$$

$$W_{et} = \frac{1}{2} C_t V_t^2 = \frac{1}{2} (3.9351 \times 10^{-10}) (36)^2 = 0.25499 \times 10^{-6} = W_{e0} + W_{e1} \text{ Joules}$$

2.6 Before the insertion of the slab

$$E = \frac{V}{4 \times 10^{-2}} = \frac{8}{4 \times 10^{-2}} = 200 \text{ V/m}$$

$$D = \epsilon_0 E = 8.854 \times 10^{-12} (200) = 17.708 \times 10^{-10} = 1.7708 \times 10^{-9} \text{ C/m}^2$$

a. $q_s = \pm D = \pm 1.7708 \times 10^{-9} \text{ C/m}^2$

$$Q = \pm q_s A = \pm 1.7708 \times 10^{-9} (64 \times 10^{-4}) = \pm 113.3312 \times 10^{-13} = \pm 1.133312 \times 10^{-11} \text{ C}$$

b. $E = 200 \text{ V/m}$

c. $D = 1.7708 \times 10^{-9} \text{ C/m}^2$

d. $C = \epsilon_0 \frac{A}{d} = 8.854 \times 10^{-12} \frac{64 \times 10^{-4}}{4 \times 10^{-2}} = 141.664 \times 10^{-14} = 1.41664 \times 10^{-12}$

e. $W_e = \frac{1}{2} CV^2 = \frac{1}{2} (1.41664 \times 10^{-12})(8)^2 = 45.3326 \times 10^{-12} \text{ Joules}$

After the insertion of the slab the electric field intensity remained the same; that is $E_0 = E_1 = 200 \text{ Volts/m}$

$$D_0 = \epsilon_0 E_0 = 8.854 \times 10^{-12} (200) = 1.7708 \times 10^{-9} \Rightarrow q_{s0} = 1.7708 \times 10^{-9}$$

$$D_1 = \epsilon_1 E_1 = 2.56 D_0 = 2.56 (1.7708 \times 10^{-9}) = 4.5332 \times 10^{-9} \Rightarrow q_{s1} = 4.5332 \times 10^{-9}$$

f. $Q_0 = \pm q_{s0} A = \pm 1.7708 \times 10^{-9} (32 \times 10^{-4}) = \pm 5.66656 \times 10^{-12} \text{ C (in free space)}$

$$Q_1 = \pm q_{s1} A = \pm 4.5332 \times 10^{-9} (32 \times 10^{-4}) = \pm 14.50639 \times 10^{-12} \text{ C (in dielectric)}$$

g. $E_0 = E_1 = 200 \text{ V/m}$

h. $D_0 = 1.7708 \times 10^{-9} \text{ C/m}^2 \text{ (in free space)}$

$$D_1 = 4.5332 \times 10^{-9} \text{ C/m}^2 \text{ (in dielectric)}$$

i. $C_0 = \frac{Q_0}{V} = \frac{5.66656 \times 10^{-12}}{8} = 0.70832 \times 10^{-12} \text{ (in free space)}$

$$C_1 = \frac{Q_1}{V} = \frac{14.50639 \times 10^{-12}}{8} = 1.813299 \times 10^{-12} \text{ (in dielectric)}$$

j. $C_t = \frac{Q_t}{V} = \frac{(5.66656 + 14.50639) \times 10^{-12}}{8} = \frac{20.17295 \times 10^{-12}}{8} = 2.521619 \times 10^{-12} = C_0 + C_1$

k. $W_{e0} = \frac{1}{2} C_0 V^2 = \frac{1}{2} (0.70832 \times 10^{-12})(8)^2 = 22.66624 \times 10^{-12} \text{ (in free space)}$

$$W_{e1} = \frac{1}{2} C_1 V^2 = \frac{1}{2} (1.813299 \times 10^{-12})(8)^2 = 58.02528 \times 10^{-12} \text{ (in dielectric)}$$

l. $W_{et} = \frac{1}{2} C_t V^2 = \frac{1}{2} (2.521619 \times 10^{-12})(8)^2 = 80.6918 \times 10^{-12} \text{ Joules}$

$$W_{et} = W_{e0} + W_{e1}$$

2.7 Parts (a)-(e) are the same as in the solution of Problem 2.6.

f. After the removal of the voltage source and the insertion of the slab, the total charge on the plates stays the same as before the removal of the source. Therefore from part a of Problem 2.6

$$Q_0 + Q_1 = Q_t = 1.133312 \times 10^{-11}$$

Also along the interface between the free space and the dielectric slab the tangential components of the electric field must be continuous. Therefore

$$E_0 = E_1 \Rightarrow \frac{D_0}{\epsilon_0} = \frac{D_1}{\epsilon_1} \Rightarrow \frac{q_{s0}}{\epsilon_0} = \frac{q_{s1}}{\epsilon_1} \Rightarrow \frac{Q_0/A_0}{\epsilon_0} = \frac{Q_1/A_1}{\epsilon_1} \Rightarrow Q_1 = \frac{\epsilon_1 A_1}{\epsilon_0 A_0} Q_0$$

Since the two areas are the same $A_0 = A_1$. Thus

$$Q_1 = \frac{\epsilon_1}{\epsilon_0} Q_0 = \epsilon_r Q_0 \quad \text{where } \epsilon_r = \frac{\epsilon_1}{\epsilon_0}$$

Using these two equations

$$Q_0 + Q_1 = Q_t$$

$$Q_1 = \epsilon_r Q_0$$

we can write that

$$Q_0 = \frac{Q_t}{1+\epsilon_r} = \frac{1.133312 \times 10^{-11}}{1+2.56} = 0.31835 \times 10^{-11} = 3.1835 \times 10^{-12} \text{ C} \quad (\text{in free space})$$

$$Q_1 = \epsilon_r Q_0 = 2.56 (3.1835 \times 10^{-11}) = 8.14966 \times 10^{-12} \text{ C} \quad (\text{in dielectric})$$

$$g. q_{s0} = \frac{Q_0}{A_0} = \frac{3.1835 \times 10^{-12}}{32 \times 10^{-4}} = 0.9948 \times 10^{-9} = D_0 \Rightarrow E_0 = \frac{0.9948 \times 10^{-9}}{8.854 \times 10^{-12}} = 112.36 \text{ V/m} \quad (\text{in free space})$$

$$q_{s1} = \frac{Q_1}{A_1} = \frac{8.14966 \times 10^{-12}}{32 \times 10^{-4}} = 2.5468 \times 10^{-9} = D_1 \Rightarrow E_1 = \frac{2.5468 \times 10^{-9}}{2.56(8.854 \times 10^{-12})} = 112.36 \text{ V/m} \quad (\text{in dielectric})$$

$$h. D_0 = 0.9948 \times 10^{-9} \text{ C/m}^2 \quad (\text{in free space}); D_1 = 2.5468 \times 10^{-9} \text{ C/m}^2 \quad (\text{in dielectric})$$

$$i. C_0 = \frac{Q_0}{V_0} = \frac{3.1835 \times 10^{-12}}{112.36(4 \times 10^{-2})} = 0.708326 \times 10^{-12} \text{ farads} = \epsilon_0 \frac{A_0}{d}$$

$$C_1 = \frac{Q_1}{V_1} = \frac{8.14966 \times 10^{-12}}{112.36(4 \times 10^{-2})} = 1.81329 \times 10^{-12} \text{ farads} = \epsilon_1 \frac{A_1}{d}$$

$$j. C_t = C_0 + C_1 = 2.521616 \times 10^{-12} \text{ farads}$$

$$k. W_0 = \frac{1}{2} C_0 V_0^2 = \frac{1}{2} (0.708326 \times 10^{-12}) [112.36(4 \times 10^{-2})]^2 = 7.15396 \times 10^{-12} \text{ Joules}$$

$$l. W_1 = \frac{1}{2} C_1 V_1^2 = \frac{1}{2} (1.81329 \times 10^{-12}) [112.36(4 \times 10^{-2})]^2 = 18.31389 \text{ Joules}$$

$$W_t = W_0 + W_1 = 25.46785 \times 10^{-12} \text{ Joules}$$

$$[2.8] V = 100 \text{ Volts}, A = 2 \text{ cm}^2 = 2 \text{ cm}^2 \left(\frac{\text{m}}{100 \text{ cm}} \right)^2 = 2 \times 10^{-4} \text{ m}^2$$

(a) Before the insertion of the slab, the electric flux density between the plates is equal to

$$D = \epsilon_0 E_0 = 8.854 \times 10^{-12} \left(\frac{100}{1.25 \times 10^{-3}} \right) = 7.0832 \times 10^{-7} \text{ C/m}^2$$

thus the surface charge density is equal to $|q_s| = D = 7.0832 \times 10^{-7} \text{ C/m}^2$

After removal of the battery and the insertion of the dielectric slab, the charge density remains the same. Thus,

$$Q_u = Q(\text{upper}) = q_s(\text{upper}) A = +7.0832 \times 10^{-7} (2 \times 10^{-4}) = +14.1664 \times 10^{-11} \text{ C}$$

$$Q_l = Q(\text{lower}) = q_s(\text{lower}) A = -7.0832 \times 10^{-7} (2 \times 10^{-4}) = -14.1664 \times 10^{-11} \text{ C}$$

$$(b) q_s(\text{upper}) = +7.0832 \times 10^{-7} \text{ C/m}^2$$

$$q_s(\text{lower}) = -7.0832 \times 10^{-7} \text{ C/m}^2$$

(c) The electric flux density in the dielectric slab and free space is

$$\hat{n} \cdot (D_2 - D_1) = q_s \Rightarrow D_2 - D_1 = q_s$$

$$D_1 = 7.0832 \times 10^{-7} \text{ C/m}^2$$

$$D_0 = 7.0832 \times 10^{-7} \text{ C/m}^2$$

$$(d) \epsilon_0 E_d = D_d \Rightarrow E_d = \frac{D_d}{\epsilon_0} = \frac{7.0832 \times 10^{-7}}{8.854 \times 10^{-12}} = 0.8 \times 10^5 = 16 \times 10^3 \text{ V/m}$$

$$\epsilon_0 E_0 = D_0 \Rightarrow E_0 = \frac{D_0}{\epsilon_0} = \frac{7.0832 \times 10^{-7}}{8.854 \times 10^{-12}} = 0.8 \times 10^5 = 80 \times 10^3 \text{ V/m}$$

$$(e) V_d = E_d d_d = 16 \times 10^3 (1 \times 10^{-3}) = 16 \text{ Volts}$$

$$V_o = E_o d_o = 80 \times 10^3 (0.25 \times 10^{-3}) = 20 \text{ Volts}$$

$$V_t = V_o + V_d = 20 + 16 = 36 \text{ Volts}$$

$$(f) C_d = \frac{Q}{V_d} = \frac{14.1664 \times 10^{-11}}{16} = 0.8854 \times 10^{-11} = 8.854 \times 10^{-12} \text{ farads}$$

$$C_o = \frac{Q}{V_o} = \frac{14.1664 \times 10^{-11}}{20} = 0.70832 \times 10^{-11} = 7.0832 \times 10^{-12} \text{ farads}$$

$$C_t = \frac{Q}{V_t} = \frac{14.1664 \times 10^{-11}}{36} = 3.9351 \times 10^{-12} = \frac{C_0 C_1}{C_0 + C_1} = \frac{7.0832 (8.854) \times 10^{-24}}{(7.0832 + 8.854) \times 10^{-12}}$$

$$2.9 \quad A = 1 \text{ cm}^2, V = 10 \text{ volt}, \epsilon_{r1} = 2, \epsilon_{r2} = 6, d_1 = d_2 = 1 \text{ cm}$$

$$(a) \quad E_1 d_1 + E_2 d_2 = 10 \Rightarrow E_1 + E_2 = \frac{10}{10^2} = 10^3 \text{ V/m}$$

$$D_1 = D_2 \Rightarrow \epsilon_1 E_1 = \epsilon_2 E_2 \Rightarrow E_1 = \frac{\epsilon_2}{\epsilon_1} (E_2) = \frac{\epsilon_2}{\epsilon_1} E_2 = \frac{6}{2} E_2 = 3 E_2$$

$$E_1 + E_2 = 3E_2 + E_2 = 4E_2 = 10^3 \Rightarrow E_2 = 0.25 \times 10^3 \text{ V/m}, E_1 = 3E_2 = 0.75 \times 10^3 \text{ V/m}$$

$$E_1 = 0.75 \times 10^3 \text{ V/m}, E_2 = 0.25 \times 10^3 \text{ V/m}$$

$$(b) \quad D_1 = \epsilon_1 E_1 = 2\epsilon_0 (0.75 \times 10^3) = 1.5\epsilon_0 \times 10^3 \text{ C/m}^2$$

$$D_2 = \epsilon_2 E_2 = 6\epsilon_0 (0.25 \times 10^3) = 1.5\epsilon_0 \times 10^3 \text{ C/m}^2$$

$$(c) \quad Q_{\text{upper}} = A_1 q_{s1} = 10^4 \epsilon_0 (1.5 \times 10^3) = 0.15\epsilon_0 \text{ C}$$

$$Q_{\text{lower}} = A_2 q_{s2} = 10^4 \epsilon_0 (1.5 \times 10^3) = 0.15\epsilon_0 \text{ C}$$

$$(d) \quad C_1 = \frac{Q}{V} = \frac{Q_1}{E_1 d_1} = \frac{0.15\epsilon_0}{(0.75 \times 10^3)(10^2)} = 0.02\epsilon_0 \text{ F}$$

$$C_2 = \frac{Q_2}{V_2} = \frac{0.15\epsilon_0}{E_2 d_2} = \frac{0.15\epsilon_0}{(0.25 \times 10^3)(10^2)} = 0.06\epsilon_0 \text{ F}$$

$$C_T = \frac{Q}{V} = \frac{0.15\epsilon_0}{10} = 0.015\epsilon_0 \text{ F}$$

(e)

$$C_1 = \epsilon_1 \frac{A_1}{d_1} = 2\epsilon_0 \frac{10^{-4}}{10^2} = 0.02\epsilon_0$$

$$C_2 = \epsilon_2 \frac{A_2}{d_2} = 6\epsilon_0 \frac{10^{-4}}{10^2} = 0.06\epsilon_0$$

$$C_T = \frac{C_1 C_2}{C_1 + C_2} = \frac{(0.02\epsilon_0)(0.06\epsilon_0)}{(0.02\epsilon_0 + 0.06\epsilon_0)} = \frac{0.02(0.06)\epsilon_0}{0.08}$$

$$C_T = \frac{2(6)}{8} \times 10^{-2} \epsilon_0 = 0.015\epsilon_0$$

(f) They are the same, as they should be.

2.10

(a) $E_1 = \frac{10}{1 \text{ cm}} \left(\frac{100 \text{ cm}}{1 \text{ m}} \right) = 10^{+3} \text{ V/m}$

$$E_2 = \frac{10}{1 \text{ cm}} \left(\frac{100 \text{ cm}}{1 \text{ m}} \right) = 10^{+3} \text{ V/m}$$

(b) $D_1 = E_1 \epsilon_1 = 10^{+3} \left(2 \times 8.854 \times 10^{-12} \right) = 17.708 \times 10^{-9} \text{ C/m}^2$

$$D_2 = E_2 \epsilon_2 = 10^{+3} \left(6 \times 8.854 \times 10^{-12} \right) = 53.124 \times 10^{-9} \text{ C/m}^2$$

(c) $Q_1 = 17.708 \times 10^{-9} \text{ C/m}^2$

$$Q_2 = 53.124 \times 10^{-9} \text{ C/m}^2$$

(d) $Q_1 = q_1 A = 17.708 \times 10^{-9} \frac{\text{C}}{\text{m}^2} \left(1 \times 10^{-4} \text{ m}^2 \right) = 17.708 \times 10^{-13} = 1.7708 \times 10^{-12} \text{ C}$

$$Q_2 = q_2 A = 53.124 \times 10^{-9} \frac{\text{C}}{\text{m}^2} \left(1 \times 10^{-4} \text{ m}^2 \right) = 53.124 \times 10^{-13} = 5.3124 \times 10^{-12} \text{ C}$$

(e) $C_1 = \frac{Q_1}{V_1} = \frac{1.7708 \times 10^{-12}}{10} = 1.7708 \times 10^{-13} \text{ farads}$

$$C_2 = \frac{Q_2}{V_2} = \frac{5.3124 \times 10^{-12}}{10} = 5.3124 \times 10^{-13} \text{ farads}$$

(f) $C_T = C_1 + C_2 = (1.7708 + 5.3124) \times 10^{-13} = 7.0832 \times 10^{-13} = 7.0832 \times 10^{-12} \text{ C}$

(g) $C_1 = \epsilon_1 \frac{A_1}{S_1} = 2(8.854 \times 10^{-12}) \frac{10^{-4}}{10^{-2}} = 17.708 \times 10^{-14} = 1.7708 \times 10^{-13} \text{ farads}$

$$C_2 = \epsilon_2 \frac{A_2}{S_2} = 6(8.854 \times 10^{-12}) \frac{10^{-4}}{10^{-2}} = 53.124 \times 10^{-14} = 5.3124 \times 10^{-13} \text{ farad}$$

(h) $C_T = C_1 + C_2 = (1.7708 + 5.3124) \times 10^{-13} = 7.0832 \times 10^{-13} = 7.0832 \times 10^{-12} \text{ farads}$

(i) They are the same as they should. Both are derived based on basic and related principles.

2.11

Before the insertion of the slabs and before the power supply is disconnected

$$E_1 = E_2 = \frac{10}{1 \times 10^{-2}} = 1,000 \text{ V/m} = 10^3 \text{ V/m}$$

$$D_1 = D_2 = \epsilon_0 E_1 = \epsilon_0 E_2 = 8.854 \times 10^{-12} (10^3) = 8.854 \times 10^{-9} \text{ C/m}^2$$

$$q_{s1} = q_{s2} = D_1 = D_2 = 8.854 \times 10^{-9} \text{ C/m}^2$$

$$Q_1 = Q_2 = q_{s1} A_1 = q_{s2} A_2 = 8.854 \times 10^{-9} (10^{-4}) = 8.854 \times 10^{-13} \text{ C}$$

After the removal of the power supply, the charge density (C/m^2) in each plate remains the same ^{in the presence of the slabs} before the insertion of the slabs.

$$(a) q_{s1} = q_{s2} = 8.854 \times 10^{-9} \text{ C/m}^2$$

$$(b) D_1 = D_2 = |q_{s1}| = |q_{s2}| = 8.854 \times 10^{-9} \text{ C/m}^2$$

$$(c) Q_1 = Q_2 = q_{s1} A_1 = q_{s2} A_2 = 8.854 \times 10^{-13} \text{ C}$$

$$(d) D_1 = \epsilon_1 E_1 \Rightarrow E_1 = \frac{D_1}{\epsilon_1} = \frac{8.854 \times 10^{-9}}{2(8.854 \times 10^{-12})} = \frac{10^3}{2} = 500 \text{ V/m}$$

$$D_2 = \epsilon_2 E_2 \Rightarrow E_2 = \frac{D_2}{\epsilon_2} = \frac{8.854 \times 10^{-9}}{6(8.854 \times 10^{-12})} = \frac{10^3}{6} = 166.67 \text{ V/m}$$

$$(e) V_1 = E_1 d_1 = 500 (10^{-2}) = 5 \text{ Volts}$$

$$V_2 = E_2 d_2 = 166.67 (10^{-2}) = 1.6667 \text{ Volts}$$

$$(f) C_1 = \epsilon_1 \frac{A_1}{d_1} = 2(8.854 \times 10^{-12}) \frac{10^{-4}}{10^{-2}} = 17.708 \times 10^{-14} \text{ F}$$

$$C_2 = \epsilon_2 \frac{A_2}{d_2} = 6(8.854 \times 10^{-12}) \frac{10^{-4}}{10^{-2}} = 53.124 \times 10^{-14} \text{ F}$$

$$(g) C_1 = \frac{Q_1}{V_1} = \frac{8.854 \times 10^{-13}}{5} = 1.7708 \times 10^{-13} = 17.708 \times 10^{-14} \text{ F}$$

$$C_2 = \frac{Q_2}{V_2} = \frac{8.854 \times 10^{-13}}{1.6667} = 5.3124 \times 10^{-13} = 53.124 \times 10^{-14} \text{ F}$$

(h) They should be the same. Same definition.

2.12 Using Coulomb's Law

a. $\oint_S D \cdot d\hat{s} = \int_0^{\frac{1}{2}\pi} \int_0^{2\pi} (\hat{a}_p D_p) \cdot \hat{a}_p \rho d\phi dz = \int_0^{\frac{1}{2}\pi} \int_0^{2\pi} \epsilon_0 E_p \rho d\phi dz = 2\pi \epsilon_0 \ell_p E_p = Q$

or

$$E_p = \frac{Q}{2\pi \epsilon_0 \ell} \frac{1}{\rho} = \frac{Q}{\rho} (3 \times 10^{11}) = 3 \times 10^{11} \frac{Q}{\rho}$$

Since the voltage is 10 Volts, then

$$V = \int_a^b E_p \cdot d\ell = \int_a^b (\hat{a}_p E_p) \cdot \hat{a}_p d\rho = 3 \times 10^{11} Q \int_a^b \frac{d\rho}{\rho} = 3 \times 10^{11} Q \ln(\frac{b}{a})$$

$$V = 10 = 3 \times 10^{11} Q \ln(\frac{b}{a}) = 3 \times 10^{11} Q \ln(2) = 3 \times 10^{11} Q (0.69315)$$

$$Q = \frac{10}{3 \times 10^{11} (0.69315)} = 48.0898 \times 10^{-12} \text{ C}$$

$$E_p = 3 \times 10^{11} \frac{Q}{\rho} = 3 \times 10^{11} \frac{48.0898 \times 10^{-12}}{\rho} = \frac{14.42695}{\rho}$$

b. $Q = 48.0898 \times 10^{-12} \text{ C}$

c. $q_{sa} = -\frac{Q}{A_a} = -\frac{Q}{2\pi a \ell} = -\frac{48.0898 \times 10^{-12}}{2\pi (2 \times 10^{-2})(6 \times 10^{-2})} = -6.378 \times 10^{-9} \text{ C/m}^2$

$$q_{sb} = \frac{Q}{A_b} = \frac{Q}{2\pi b \ell} = \frac{48.0898 \times 10^{-12}}{2\pi (4 \times 10^{-2})(6 \times 10^{-2})} = 3.189 \times 10^{-9} \text{ C/m}^2$$

d. $D = \epsilon_0 E_p = 8.854 \times 10^{-12} \left(\frac{14.42695}{\rho} \right) = \frac{1.27736}{\rho} \times 10^{-10} \text{ C/m}^2$

e. $C = \frac{Q}{V} = \frac{48.0898 \times 10^{-12}}{10} = 4.80898 \times 10^{-12} \text{ farads}$

f. $W_e = \frac{1}{2} CV^2 = \frac{1}{2} (4.80898 \times 10^{-12})(100) = 2.4045 \times 10^{-10} \text{ Joules}$

2.13 After the removal of the voltage source, the total charge stays the same. Using this we can answer the questions.

a. Same as in part b of Problem 2.12

$$Q = 48.0898 \times 10^{-12} \text{ C}$$

b. Same as in part c of Problem 2.12

$$q_{sa} = -6.378 \times 10^{-9} \text{ C/m}^2$$

$$q_{sb} = 3.189 \times 10^{-9} \text{ C/m}^2 \quad (\text{continued})$$

2.13 cont'd

Same as in part d of Problem 2.12 (in free space and dielectric)
c. $D_{p0} = D_{p1} = \frac{1.27736 \times 10^{-10}}{\rho}$

d. $E_{p0} = \frac{D_{p0}}{\epsilon_0} = \frac{1.27736 \times 10^{-10}}{8.854 \times 10^{-12} \rho} = \frac{14.4269}{\rho}$

$$E_{p1} = \frac{D_{p1}}{\epsilon_1} = \frac{1.27736 \times 10^{-10}}{2.56(8.854 \times 10^{-12}) \rho} = \frac{5.6355}{\rho}$$

e. $V = \int_a^c E_{p2} dl + \int_c^b E_{p0} \cdot dl = 5.6355 \int_a^c \frac{dp}{\rho} + 14.4269 \int_c^b \frac{dp}{\rho} = 5.6355 \ln\left(\frac{c}{a}\right) + 14.4269 \ln\left(\frac{b}{c}\right)$
 $= 5.6355 \ln\left(\frac{3}{2}\right) + 14.4269 \ln\left(\frac{4}{3}\right) = 5.6355(0.405465) + 14.4269(0.28768)$

$$V = 2.285 + 4.1504 = 6.4354 \text{ Volts}$$

f. $C = \frac{Q}{V} = \frac{48.0898 \times 10^{-12}}{6.4354} = 7.472698 \times 10^{-12} \text{ farads}$

g. $W_e = \frac{1}{2} CV^2 = \frac{1}{2}(7.472698)(6.4354)^2 \times 10^{-12} = 1.54738 \times 10^{-10} \text{ Joules}$

2.14 Since the voltage source is connected at all times, the total voltage is maintained at 10 Volts. Also the normal components of the electric flux density is continuous along the interface.

a. $V = \int_a^c E_{p1} \cdot dl + \int_c^b E_{p0} \cdot dl = 10 \text{ Volts} = \int_a^c \hat{a}_p E_{p1} \cdot \hat{a}_p dp + \int_c^b \hat{a}_p E_{p0} \cdot \hat{a}_p dp$
 $V = \int_a^c E_{p1} dp + \int_c^b E_{p0} dp = \int_a^c \frac{D_{p1}}{\epsilon_1} dp + \int_c^b \frac{D_{p0}}{\epsilon_0} dp = \frac{1}{\epsilon_1} \int_a^c D_{p1} dp + \int_c^b D_{p0} dp$

However $D_{p1} = D_{p0} = \frac{Q}{2\pi L} \frac{1}{\rho}$

Thus

$$V = \frac{1}{\epsilon_1} \frac{Q}{2\pi L} \ln\left(\frac{c}{a}\right) + \frac{1}{\epsilon_0} \frac{Q}{2\pi L} \ln\left(\frac{b}{c}\right) = \frac{Q}{\epsilon_0(2\pi L)} \left[\frac{\ln(c/a)}{\epsilon_1} + \ln(b/c) \right] = 10$$

$$Q = \frac{10(2\pi \epsilon_0 L)}{\frac{\ln(3/2)}{2.56} + \ln(4/3)} = \frac{10^{-10}}{3 \left[\frac{0.405465}{2.56} + 0.28768 \right]} = 7.47276 \times 10^{-9} \text{ C}$$

$$E_{p0} = \frac{Q}{2\pi \epsilon_0 L} \frac{1}{\rho} = \frac{7.47276 \times 10^{-9}}{2\pi \epsilon_0 (6 \times 10^{-2}) \rho} = \frac{2.241.827}{\rho} \text{ (in free space)}$$

(continued)

2.14 cont'd. $E_{p1} = \frac{Q}{2\pi\epsilon_0 l} \frac{1}{\rho} = \frac{E_{p0}}{\epsilon_r} = \frac{875.7136}{\rho}$ (in dielectric)

b. $D_{p0} = \epsilon_0 E_{p0} = \frac{19.849 \times 10^{-9}}{\rho} \text{ C/m}^2$

$D_{p1} = \epsilon_1 E_{p1} = \frac{19.849 \times 10^{-9}}{\rho} \text{ C/m}^2$

c. $q_{bsa} = D_{p1} \Big|_{\rho=2 \times 10^{-2}} = \frac{19.849 \times 10^{-9}}{2 \times 10^{-2}} = 0.99246 \times 10^{-6} \text{ C/m}^2$

$q_{bsb} = D_{p0} \Big|_{\rho=4 \times 10^{-2}} = \frac{q_{bsa}}{2} = 0.49623 \times 10^{-6} \text{ C/m}^2$

d. $Q_a = q_{bsa} (2\pi a l) = 0.99246 \times 10^{-6} (2\pi \times 2 \times 10^{-2} \times 6 \times 10^{-2}) = 74.8295 \times 10^{-10} \text{ C}$

$Q_b = q_{bsb} (2\pi b l) = 0.49623 \times 10^{-6} (2\pi \times 4 \times 10^{-2} \times 6 \times 10^{-2}) = 74.8295 \times 10^{-10} \text{ C}$

e. $C = \frac{Q}{V} = \frac{74.8295 \times 10^{-10}}{10} = 7.48295 \times 10^{-10} = 0.748295 \times 10^{-9} \text{ farads}$

f. $W_e = \frac{1}{2} C V^2 = \frac{1}{2} (0.748295 \times 10^{-9})(100) = 37.41475 \times 10^{-9} \text{ Joules}$

2.15 The voltage source is connected across at all times.

a. $E_0 = \frac{100}{4 \times 10^{-2}} = 2,500 \text{ V/m} = E_1$

b. $D_0 = \epsilon_0 E_0 = 22.135 \times 10^{-9} \text{ C/m}^2$ (in free space)

$D_1 = \epsilon_1 E_1 = 2.56 D_0 = 56.6656 \times 10^{-9} \text{ C/m}^2$ (in dielectric)

c. $Q_0 = 22.135 \times 10^{-9} (2 \times 10^{-2}) = 0.4427 \times 10^{-9} \text{ C}$ (in freespace)

$Q_1 = 56.6656 \times 10^{-9} (2 \times 10^{-2}) = 1.13331 \times 10^{-9} \text{ C}$ (in dielectric)

d. $W_{e0} = \frac{1}{2} C_0 V_0^2 = \frac{1}{2} \left(\frac{Q_0}{V_0} \right) V_0^2 = \frac{1}{2} Q_0 V_0 = \frac{1}{2} (0.4427 \times 10^{-9})(100) = 22.135 \times 10^{-9} \text{ J (freespace)}$

$W_{e1} = \frac{1}{2} C_1 V_1^2 = \frac{1}{2} \left(\frac{Q_1}{V_1} \right) V_1^2 = \frac{1}{2} Q_1 V_1 = \frac{1}{2} (1.13331 \times 10^{-9})(100) = 56.6655 \times 10^{-9} \text{ J (dielectric)}$

e. $Q_t = Q_0 + Q_1 = 1.57601 \times 10^{-9}$

f. $C_t = \frac{Q_t}{V} = \frac{1.57601 \times 10^{-9}}{100} = 1.57601 \times 10^{-11} = 15.7601 \times 10^{-12} = C_0 + C_1 \text{ farads}$

g. $W_{et} = \frac{1}{2} C_t V^2 = \frac{1}{2} (15.7601 \times 10^{-12})(100)^2 = 78.8005 \times 10^{-9} \text{ J} = W_{e0} + W_{e1}$

2.16

$$(a) \oint_S \underline{D} \cdot d\underline{s} = \int_0^l \int_0^{\pi} \hat{a}_\rho D_{\rho 0} \cdot \hat{a}_\rho dz \rho d\phi + \int_0^l \int_0^{\pi} \hat{a}_\rho D_{\rho 1} \cdot \hat{a}_\rho dz \rho d\phi$$

$$\oint_S \underline{D} \cdot d\underline{s} = \pi l (D_{\rho 0} + D_{\rho 1}) \rho = Q_t \Rightarrow D_{\rho 0} + D_{\rho 1} = \frac{Q_t}{2\pi l} \left(\frac{1}{\rho}\right)$$

Also, because the tangential electric field components are continuous along the radial interfaces between the dielectric free space

$$E_{\rho 0} = E_{\rho 1} \Rightarrow \frac{D_{\rho 0}}{\epsilon_0} = \frac{D_{\rho 1}}{\epsilon_1} \Rightarrow D_{\rho 1} = \frac{\epsilon_1}{\epsilon_0} D_{\rho 0} = \epsilon_r D_{\rho 0}, \epsilon_r = \frac{\epsilon_1}{\epsilon_0}$$

Thus

$$D_{\rho 0} (1 + \epsilon_r) = \frac{Q_t}{2\pi l} \frac{1}{\rho} \Rightarrow D_{\rho 0} = \frac{Q_t}{2\pi l (1 + \epsilon_r)} \frac{1}{\rho}, D_{\rho 1} = \frac{Q_t}{2\pi l} \frac{\epsilon_r}{1 + \epsilon_r} \frac{1}{\rho}$$

$$E_{\rho 0} = \frac{D_{\rho 0}}{\epsilon_0} = \frac{Q_t}{\pi \epsilon_0 l (1 + \epsilon_r)} \frac{1}{\rho}, \text{ in free space}$$

$$E_{\rho 1} = \frac{D_{\rho 1}}{\epsilon_1} = \frac{Q_t}{\pi \epsilon_1 l} \frac{\epsilon_r}{1 + \epsilon_r} \frac{1}{\rho}, \text{ in dielectric}$$

$$(b) \int_a^b \underline{E} \cdot d\underline{l} = \int_a^b \hat{a}_\rho E_{\rho 0} \cdot \hat{a}_\rho d\rho = \int_a^b E_{\rho 0} d\rho = \frac{Q_t \ln(b/a)}{2\pi \epsilon_0 l (1 + \epsilon_r)} = V = 10$$

$$Q_t = \frac{10 (\pi \epsilon_0 l) (1 + \epsilon_r)}{\ln(2)} = \frac{(1 + 2.56) \times 10^{-10}}{6 \ln(2)} = 0.856 \times 10^{-10} \text{ C}$$

$$(c,d) D_{\rho 0} = \epsilon_0 E_{\rho 0} = \frac{Q_t}{\pi l (1 + \epsilon_r)} \frac{1}{\rho} = \frac{0.856 \times 10^{-10}}{2\pi (6 \times 10^{-2}) (1 + 2.56)} \frac{1}{\rho} = \frac{1.2756 \times 10^{-10}}{\rho} \text{ C/m}^2$$

$$D_{\rho 0}|_{\rho=a} = \frac{1.2756 \times 10^{-10}}{2 \times 10^{-2}} = 0.6378 \times 10^{-8} \text{ C/m}^2 \Rightarrow q_{\rho 0}|_{\rho=a} = 0.6378 \times 10^{-8} \text{ C/m}^2$$

$$D_{\rho 0}|_{\rho=b} = \frac{1.2756 \times 10^{-10}}{4 \times 10^{-2}} = 0.3189 \times 10^{-8} \text{ C/m}^2 \Rightarrow q_{\rho 0}|_{\rho=b} = 0.3189 \times 10^{-8} \text{ C/m}^2$$

$$D_{\rho 1} = \epsilon_1 E_{\rho 1} = \frac{Q_t}{\pi l} \frac{\epsilon_r}{1 + \epsilon_r} \frac{1}{\rho} = \epsilon_r D_{\rho 0} = \frac{3.2655 \times 10^{-10}}{\rho} \text{ C/m}^2$$

$$D_{\rho 1}|_{\rho=a} = \frac{3.2655 \times 10^{-10}}{2 \times 10^{-2}} = 1.63277 \times 10^{-8} \text{ C/m}^2 \Rightarrow q_{\rho 1}|_{\rho=a} = 1.63277 \times 10^{-8} \text{ C/m}^2$$

$$D_{\rho 1}|_{\rho=b} = \frac{3.2655 \times 10^{-10}}{4 \times 10^{-2}} = 0.81638 \times 10^{-8} \text{ C/m}^2 \Rightarrow q_{\rho 1}|_{\rho=b} = 0.81638 \times 10^{-8} \text{ C/m}^2$$

$$(e) C_0 = \frac{Q_0}{V} = \frac{0.6378 \times 10^{-8} (\pi \times 2 \times 10^{-2} \times 6 \times 10^{-2})}{10} = 2.40445 \times 10^{-12} \text{ farads (in free space)}$$

$$C_1 = \frac{Q_1}{V} = \frac{1.63277 \times 10^{-8} (\pi \times 2 \times 10^{-2} \times 6 \times 10^{-2})}{10} = 6.155398 \times 10^{-12} \text{ farads (in dielectric)}$$

$$C_t = C_0 + C_1 = 8.559847 \times 10^{-12} \text{ farads}$$

$$(f) W_{e0} = \frac{1}{2} C_0 V^2 = \frac{1}{2} (2.40445 \times 10^{-12})(100) = 1.2022 \times 10^{-10} \text{ Joules (in free space)}$$

$$W_{e1} = \frac{1}{2} C_1 V^2 = \frac{1}{2} (6.155398 \times 10^{-12})(100) = 3.077699 \times 10^{-10} \text{ Joules (in dielectric)}$$

$$W_t = W_{e0} + W_{e1} = 4.279949 \times 10^{-10} \text{ Joules (total)}$$

2.17 From Problem 2.16 $E_p = \frac{Q}{2\pi\epsilon_0 p}$ where Q is the total charge.

- a. Since the charge stays the same at all times and is the same as that of Problem 2.16, or $Q_t = 85.5985 \times 10^{-12} \text{ C}$. Also to satisfy the boundary conditions along the interface between the two media

$$E_{p0} = E_{p1}$$

b. $D_{p0} = \epsilon_0 E_{p0} = \frac{1.2756 \times 10^{-10}}{p} \text{ C/m}^2$

$$D_{p1} = \epsilon_1 E_{p1} = \epsilon_r E_{p0} = \frac{3.2655 \times 10^{-10}}{p} \text{ C/m}^2$$

c. $q_{sol|p=a} = -D_{p0}|_{p=a} = -\frac{1.2756 \times 10^{-10}}{2 \times 10^{-2}} = -0.6378 \times 10^{-8} \text{ C/m}^2$

$$q_{sol|p=b} = D_{p1}|_{p=b} = \frac{1.2756 \times 10^{-10}}{4 \times 10^{-2}} = 0.3189 \times 10^{-8} \text{ C/m}^2$$

$$q_{si|p=a} = -D_{p1}|_{p=a} = -\frac{3.2655 \times 10^{-10}}{2 \times 10^{-2}} = -1.63277 \times 10^{-8} \text{ C/m}^2$$

$$q_{si|p=b} = D_{p1}|_{p=b} = \frac{3.2655 \times 10^{-10}}{4 \times 10^{-2}} = 0.81638 \times 10^{-8} \text{ C/m}^2$$

d. $Q_0|_{p=a} = q_{sol|p=a}(\pi a b) = -0.6378 \times 10^{-8} (\pi \times 2 \times 10^{-2} \times 6 \times 10^{-2}) = -24.0445 \times 10^{-12} \text{ C}$

$$Q_0|_{p=b} = q_{sol|p=b}(\pi b a) = 0.3189 \times 10^{-8} (\pi \times 4 \times 10^{-2} \times 6 \times 10^{-2}) = 24.0445 \times 10^{-12} \text{ C}$$

$$Q_1|_{p=a} = q_{si|p=a}(\pi a b) = -1.63277 \times 10^{-8} (\pi \times 2 \times 10^{-2} \times 6 \times 10^{-2}) = -61.5540 \times 10^{-12} \text{ C}$$

$$Q_1|_{p=b} = q_{si|p=b}(\pi b a) = 1.63277 \times 10^{-8} (\pi \times 4 \times 10^{-2} \times 6 \times 10^{-2}) = 61.5540 \times 10^{-12} \text{ C}$$

e. $C_0 = \frac{Q_0}{V_0} = \frac{24.0445 \times 10^{-12}}{10} = 2.40445 \times 10^{-12} \text{ farads} \quad \text{in free space}$

$$C_1 = \frac{Q_1}{V_1} = \frac{61.5540 \times 10^{-12}}{10} = 6.15540 \times 10^{-12} \text{ farads} \quad \text{in polystyrene}$$

$$C_t = C_0 + C_1 = (2.40445 + 6.15540) \times 10^{-12} = 8.55985 \times 10^{-12}$$

f. $W_{eo} = \frac{1}{2} C_0 V_0^2 = \frac{1}{2} (2.40445 \times 10^{-12})(100) = 1.2022 \times 10^{-10} \text{ Joules (in free space)}$

$$W_{e1} = \frac{1}{2} C_1 V_1^2 = \frac{1}{2} (6.15540 \times 10^{-12})(100) = 3.0777 \times 10^{-10} \text{ Joules (in polystyrene)}$$

$$W_{et} = \frac{1}{2} C_t V_t^2 = \frac{1}{2} (8.55985 \times 10^{-12})(100) = 4.2799 \times 10^{-10} \text{ Joules}$$

$$= W_{eo} + W_{e1}$$

$$\underline{\epsilon} = \hat{a}_2 10^{-3} \sin(2\pi \times 10^7 t), \quad \epsilon_{sy} = 3.56$$

$$(a) 1 + \chi_e = \frac{\epsilon_s}{\epsilon_0} = \epsilon_{sy} \Rightarrow \chi_e = \epsilon_{sy} - 1 = 3.56 - 1 = 1.56$$

$$(b) \underline{\sigma} = \epsilon_s \underline{\epsilon} = \hat{a}_2 2.56 (\epsilon_0) 10^{-3} \sin(2\pi \times 10^7 t) = \hat{a}_2 2.56 \times 10^{-3} \epsilon_0 \sin(2\pi \times 10^7 t)$$

$$\underline{\sigma} = \hat{a}_2 2.56 \times 10^{-3} \epsilon_0 \sin(2\pi \times 10^7 t)$$

$$(c) \underline{P} = \epsilon_0 \chi_e \underline{\epsilon} = \hat{a}_2 \epsilon_0 (1.56 \times 10^{-3} \epsilon_0) \sin(2\pi \times 10^7 t) = \hat{a}_2 1.56 \times 10^{-3} \epsilon_0 \sin(2\pi \times 10^7 t)$$

$$\underline{P} = \hat{a}_2 1.56 \times 10^{-3} \epsilon_0 \sin(2\pi \times 10^7 t)$$

$$(d) \underline{J}_d = \frac{\partial \underline{\sigma}}{\partial t} = \hat{a}_2 2.56 \times 10^{-3} \epsilon_0 \frac{\partial}{\partial t} \sin(2\pi \times 10^7 t)$$

$$= \hat{a}_2 (2.56 \times 10^{-3} \times 2\pi \times 10^7) \epsilon_0 \cos(2\pi \times 10^7 t)$$

$$\underline{J}_d = \hat{a}_2 16.085 \times 10^4 \epsilon_0 \cos(2\pi \times 10^7 t)$$

$$(e) \underline{J}_p = \frac{\partial \underline{P}}{\partial t} = \hat{a}_2 \frac{\partial}{\partial t} [1.56 \times 10^{-3} \epsilon_0 \sin(2\pi \times 10^7 t)]$$

$$= \hat{a}_2 1.56 \times 10^{-3} (2\pi \times 10^7) \epsilon_0 \cos(2\pi \times 10^7 t)$$

$$\underline{J}_p = \hat{a}_2 9.802 \times 10^4 \epsilon_0 \cos(2\pi \times 10^7 t)$$

$$[2.19] \quad \underline{M} = \hat{a}_z 1.245 \times 10^6 \text{ A/m}, \underline{H} = \hat{a}_z 5 \times 10^3 \text{ A/m}$$

a. $\underline{J}_{ms} = \underline{M} \times \hat{n}$

$$y=0: \quad \underline{J}_{ms} = \hat{a}_z M_z \times (-\hat{a}_y) = \hat{a}_x M_z = \hat{a}_x 1.245 \times 10^6$$

$$x=4\text{cm}: \quad \underline{J}_{ms} = \hat{a}_z M_z \times \hat{a}_x = \hat{a}_y M_z = \hat{a}_y 1.245 \times 10^6$$

$$y=6\text{cm}: \quad \underline{J}_{ms} = \hat{a}_z M_z \times \hat{a}_y = -\hat{a}_x M_z = -\hat{a}_x 1.245 \times 10^6$$

$$x=0: \quad \underline{J}_{ms} = \hat{a}_z M_z \times (-\hat{a}_x) = -\hat{a}_y M_z = -\hat{a}_y 1.245 \times 10^6$$

$$z=0: \quad \underline{J}_{ms} = \hat{a}_z M_z \times (-\hat{a}_z) = 0$$

$$z=1\text{cm}: \quad \underline{J}_{ms} = \hat{a}_z M_z \times \hat{a}_z = 0$$

b. $\underline{J}_m = \nabla \times \underline{M} = 0$

c. $I_m = \iint_S \underline{J}_m \cdot d\underline{s} = \iiint_V (\nabla \cdot \underline{J}_m) dv = 0$

d. $\chi_m = \frac{M}{H} = \frac{1.245 \times 10^6}{5 \times 10^3} = 249$

$$\mu_r = 1 + \chi_m = 1 + 249 = 250$$

$$[2.20] \quad \underline{H} = \hat{a}_\phi \frac{0.3183}{\rho} \text{ A/m}, \quad \underline{M} = \hat{a}_\phi \frac{190.67}{\rho} \text{ A/m}$$

a. $\underline{J}_{ms} = \underline{M} \times \hat{n}$

$$\rho=a: \quad \underline{J}_{ms} = \hat{a}_\phi M_\phi \times (-\hat{a}_\rho) \Big|_{\rho=a} = \hat{a}_z M_\phi \Big|_{\rho=a} = \hat{a}_z \frac{190.67}{1 \times 10^{-2}} = \hat{a}_z 19.067 \times 10^3$$

$$\rho=b: \quad \underline{J}_{ms} = \hat{a}_\phi M_\phi \times \hat{a}_\rho \Big|_{\rho=b} = -\hat{a}_z M_\phi \Big|_{\rho=b} = -\hat{a}_z \frac{190.67}{3 \times 10^{-2}} = \hat{a}_z 6.3557 \times 10^3$$

b. $\underline{J}_m = \nabla \times \underline{M} = \hat{a}_z \frac{1}{\rho} \frac{\partial}{\partial \rho} (\rho M_\phi) = \hat{a}_z \frac{1}{\rho} \frac{\partial}{\partial \rho} (190.67) = 0$

c. $I_m = \iint_S \underline{J}_m \cdot d\underline{s} = \iiint_V (\nabla \cdot \underline{J}_m) dv = 0$

d. $\chi_m = \frac{M}{H} = \frac{190.67/\rho}{0.3183/\rho} = \frac{190.67}{0.3183} = 599.03 \approx 599$

$$\mu_r = 1 + \chi_m = 1 + 599 = 600$$

$$\left. \begin{aligned} z=0: \quad \underline{J}_{ms} &= \hat{a}_\phi M_\phi \times \hat{a}_z = \hat{a}_\rho M_\phi \\ \underline{J}_{ms} &= \hat{a}_\rho \frac{160.67}{\rho} \end{aligned} \right\}$$

$$\left. \begin{aligned} z=-l: \quad \underline{J}_{ms} &= \hat{a}_\phi M_\phi \times (-\hat{a}_z) \\ &= -\hat{a}_\rho M_\phi \\ \underline{J}_{ms} &= -\hat{a}_\rho \frac{160.67}{\rho} \end{aligned} \right\}$$

$$[2.21] \quad \underline{M} = \hat{a}_\phi 10 \frac{\Delta}{m}$$

$$a. \quad \underline{J}_{ms} = \sum_{p=1}^{\infty} \underline{M} \times \hat{a}_p = (\hat{a}_\phi M_\phi \times \hat{a}_p) = -\hat{a}_z M_\phi |_{p=1} = -\hat{a}_z 10 \text{ A/m}$$

$$b. \quad \underline{J}_m = \nabla \times \underline{M} = \hat{a}_z \frac{1}{p} \frac{\partial}{\partial p} (p M_\phi) = \hat{a}_z \frac{1}{p} \frac{\partial}{\partial p} (10_p) = \hat{a}_z \frac{10}{p} \text{ A/m}^2$$

$$c. \quad I_m = \iint_S \underline{J}_m \cdot d\underline{s} = \int_0^{2\pi} \int_0^1 \hat{a}_z J_m \cdot \hat{a}_z p dp d\phi = \int_0^{2\pi} \int_0^1 10 dp d\phi = 2\pi(10) = 62.83185 \text{ A}$$

$$[2.22] \quad \underline{J} \approx \hat{a}_z J_0 e^{-10^2[(a-x)+(\alpha-y)]} \text{ A/m}^2$$

$$\begin{aligned} I &= \iint_S \underline{J} \cdot d\underline{s} = 4 \int_0^a \int_0^a \hat{a}_z J_z \cdot \hat{a}_z dx dy = 4 J_0 \left[\int_0^a e^{-100(a-x)} dx \int_0^a e^{-100(a-y)} dy \right] \\ &= 4 J_0 \left[\int_0^a e^{-100(a-x)} dx \right]^2 = 4 J_0 \left\{ \left[\frac{1}{100} e^{-100(a-x)} \right]_0^a \right\}^2 = 4 J_0 \left\{ \frac{1}{100} [1 - e^{-100a}] \right\}^2 \end{aligned}$$

$$I = 4 J_0 \left\{ \frac{1}{100} [1 - e^{-1}] \right\}^2 = 4 J_0 \left[\frac{1}{100} [1 - 0.36788] \right]^2 = 2(7.9915 \times 10)^{-5} J_0 = 15,983 \times 10^{-5} J_0 \text{ A}$$

$$[2.23] \quad \underline{J} = \hat{a}_z J_0 e^{-10^4(a-p)} \text{ A/m}^2$$

$$\begin{aligned} a. \quad I &= \iint_S \underline{J} \cdot \hat{a}_z d\underline{s} = \int_0^{2\pi} \int_0^a \hat{a}_z J_z \cdot \hat{a}_z p dp d\phi = \int_0^{2\pi} \int_0^a J_0 e^{-10^4(a-p)} p dp d\phi \\ &= 2\pi J_0 \int_0^a e^{-10^4(a-p)} p dp = 2\pi J_0 \int_0^a e^{-10^4a} e^{10^4p} p dp \end{aligned}$$

$$\text{Let } u = 10^4 p \Rightarrow du = 10^4 dp \Rightarrow dp = \frac{1}{10^4} du, \quad p = 10^{-4} u$$

$$I = 2\pi \times 10^{-8} J_0 \int_0^{10^4 a} e^{-10^4 a} e^u u du = 2\pi \times 10^{-8} J_0 e^{-10^4 a} \int_0^{10^4 a} e^u u du$$

Using the integral $\int e^u u du = e^u (u-1)$, we can write that

$$I = 2\pi J_0 \times 10^{-8} e^{-10^4 a} \left[e^{10^4 a} (10^4 a - 1) - e^0 (0 - 1) \right]_{a=0,01} = 0.622 \times 10^{-5} J_0 = 10 \text{ A}$$

$$I = 10 \approx 0.622 \times 10^{-5} J_0 \Rightarrow J_0 = 10 / 0.622 \times 10^{-5} = 1.6077 \times 10^6 \text{ A/m}^2$$

$$b. \quad J = J_0 e^{-10^4(a-p)} = e^{-1} J_0 = 0.368 J_0 \Rightarrow 10^4(a-p) = 1$$

$$a-p = 10^{-4} \Rightarrow p = a - 10^{-4}$$

Distance from surface = 10^{-4} m (very small; almost on the surface)

2.24 $\sigma = 5.76 \times 10^7 \text{ S/m}$, $E = 8.854 \times 10^{-12} \text{ N/C}$

$$T_r = \frac{E}{\sigma} = \frac{8.854 \times 10^{-12}}{5.76 \times 10^7} = 1.5372 \times 10^{-19} \text{ sec.}$$

In the microwave region ($f = 1-10 \text{ GHz}$) the period is

$$f = 1 \text{ GHz}: T = \frac{1}{f} = \frac{1}{10^9} = 10^{-9} \gg T_r = 1.5372 \times 10^{-19}$$

$$f = 2 \text{ GHz}: T = \frac{1}{f} = \frac{1}{10^9} = 10^{-10} \gg T_r = 1.5372 \times 10^{-19}$$

For x-rays [$\lambda = (1-10) \times 10^{-8} \text{ cm}$] the period is

$$\lambda = 1 \times 10^{-8} \text{ cm}: f = \frac{c}{\lambda} = \frac{3 \times 10^{10}}{10^{-8}} = 3 \times 10^{18}$$

$$T = \frac{1}{f} = \frac{1}{3 \times 10^{18}} = 0.333 \times 10^{-18} = 3.33 \times 10^{-19}$$

$$\lambda = 10 \times 10^{-8} \text{ cm}: f = \frac{c}{\lambda} = \frac{3 \times 10^{10}}{10 \times 10^{-8}} = 3 \times 10^{17}$$

$$T = \frac{1}{f} = \frac{1}{3 \times 10^{17}} = 0.333 \times 10^{-17} = 3.33 \times 10^{-18}$$

which are comparable to the relaxation time constant.

2.25 $\sigma = 3.96 \times 10^7 \text{ S/m}$, $\mu_e = 2.2 \times 10^{-3} \text{ m}^2/(\text{V-sec})$, $E = \hat{a}_x 2 \text{ V/m}$

a. $\sigma = -\mu_e g_{ve} \Rightarrow g_{ve} = -\frac{\sigma}{\mu_e} = -\frac{3.96 \times 10^7}{2.2 \times 10^{-3}} = -1.8 \times 10^{10} \text{ C/m}^3$

b. $V_e = -\mu_e E = -2.2 \times 10^{-3} (\hat{a}_x E_x) = -\hat{a}_x (2.2 \times 2 \times 10^{-3}) = -\hat{a}_x 4.4 \times 10^{-3} \text{ m/sec.}$

c. $J_e = \sigma E = 3.96 \times 10^7 (\hat{a}_x 2) = \hat{a}_x 7.92 \times 10^7 \text{ A/m}^2$

d. $I = \iint_{S_o} J_e \cdot d\mathbf{s} = \iint_{S_o} \hat{a}_x J_{ex} \cdot \hat{a}_x dS = \iint_{S_o} J_{ex} dS = 7.92 \times 10^7 (S_o) = 7.92 \times 10^7 (10^{-3})$

$$I = 7.92 \times 10^4 \text{ A}$$

e. $N_e = \frac{g_{ve}}{q_e} = \frac{-1.8 \times 10^{10}}{-1.602 \times 10^{-19}} = 1.1236 \times 10^{29} \frac{\text{electrons}}{\text{m}^3}$

CHAPTER 3

3.1 $\nabla \times \underline{E} = -\underline{M}_i - j\omega\mu\underline{H}$, $\nabla \times \underline{H} = \underline{J}_i + \sigma \underline{E} + j\omega\epsilon \underline{E}$

$$\nabla \times \nabla \times \underline{E} = -\nabla \times \underline{M}_i - j\omega\mu \nabla \times \underline{H}$$

$$\nabla \times \nabla \times \underline{H} = \nabla \times \underline{J}_i + \sigma \nabla \times \underline{E} + j\omega\epsilon \nabla \times \underline{E}$$

Using Maxwell's equations from above and the vector identity of

$$\nabla \times \nabla \times \underline{E} = \nabla(\nabla \cdot \underline{E}) - \nabla^2 \underline{E}$$

we can write

$$\begin{aligned} \nabla(\nabla \cdot \underline{E}) - \nabla^2 \underline{E} &= -\nabla \times \underline{M}_i - j\omega\mu[\underline{J}_i + \sigma \underline{E} + j\omega\epsilon \underline{E}] \\ &= -\nabla \times \underline{M}_i - j\omega\mu \underline{J}_i - j\omega\mu\sigma \underline{E} + \omega^2\mu\epsilon \underline{E} \end{aligned}$$

$$\text{Since } \nabla \cdot \underline{D} = \nabla \cdot (\epsilon \underline{E}) = \epsilon \nabla \cdot \underline{E} = q_{ve} \Rightarrow \nabla \cdot \underline{E} = \frac{q_{ve}}{\epsilon}$$

Now we can write that

$$\nabla^2 \underline{E} = \nabla \times \underline{M}_i + j\omega\mu \underline{J}_i + \frac{1}{\epsilon} \nabla q_{ve} + j\omega\mu\sigma \underline{E} - \omega^2\mu\epsilon \underline{E}$$

which is an uncoupled second-order differential equation.

Using the equation for the magnetic field from above along with Maxwell's equations and the vector identity, we can write

$$\nabla(\nabla \cdot \underline{H}) - \nabla^2 \underline{H} = \nabla \times \underline{J}_i + (\sigma + j\omega\epsilon) \nabla \times \underline{E} = \nabla \times \underline{J}_i + (\sigma + j\omega\epsilon)(-\underline{M}_i - j\omega\mu \underline{H})$$

$$\text{Since } \nabla \cdot \underline{B} = \nabla \cdot (\mu \underline{H}) = \mu \nabla \cdot \underline{H} = q_{vm} \Rightarrow \nabla \cdot \underline{H} = \frac{1}{\mu} q_{vm}$$

then

$$\nabla^2 \underline{H} = -\nabla \times \underline{J}_i + \sigma \underline{M}_i + \frac{1}{\mu} \nabla q_{vm} + j\omega\epsilon \underline{M}_i + j\omega\mu\sigma \underline{H} - \omega^2\mu\epsilon \underline{H}$$

which also is an uncoupled second order differential equation.

3.2 $\frac{d^2 f}{dx^2} = -\beta_x^2 f$, $f = f_1 = A_1 e^{-j\beta_x x} + B_1 e^{+j\beta_x x}$

Using $f = f_1 = A_1 e^{-j\beta_x x}$, then

$$(-j\beta_x)^2 A_1 e^{-j\beta_x x} = -\beta_x^2 A_1 e^{-j\beta_x x} = -\beta_x A_1 e^{-j\beta_x x} \quad \text{(continued)} \quad \text{Q.E.D.}$$

3.2 cont'd. The same can be shown by letting $f = f_1 = B_1 e^{+j\beta_x x}$

Now let us try the sinusoidal solutions.

$$\text{Let } f = f_2 = C_1 \cos(\beta_x x)$$

Substituting this into the differential equation leads to

$$-\beta_x^2 C_1 \cos(\beta_x x) = -\beta_x^2 C_1 \cos(\beta_x x) \quad \text{Q.E.D.}$$

The same can be shown by letting $f = f_2 = D_1 \sin(\beta_x x)$

3.3 $E_x(x, y, z, t) = [C_1 \cos(\beta_x x) + D_1 \sin(\beta_x x)] [C_2 \cos(\beta_y y) + D_2 \sin(\beta_y y)] B_3 \cos(\omega t + \beta_z z)$

To follow a point z_p for different values of t we must maintain constant the amplitude of the cosine term. This is accomplished by letting

$$\omega t + \beta_z z = C_1 = \text{constant}$$

Taking a derivative of both sides with respect to time, we can write

$$\omega(1) + \beta_z \frac{dz}{dt} = 0 \Rightarrow \omega + \beta_z v_p = 0 \Rightarrow v_p = -\frac{\omega}{\beta_z}$$

which indicates that the wave is moving in the $-z$ direction.

3.4 $\nabla^2 E_x - \gamma^2 E_x = 0 = \frac{d^2 E_x}{dx^2} + \frac{d^2 E_x}{dy^2} + \frac{d^2 E_x}{dz^2} - \gamma^2 E_x$

Letting $E_x(x, y, z) = f(x) g(y) h(z)$ and substituting above leads to

$$gh \frac{d^2 f}{dx^2} + fh \frac{d^2 g}{dy^2} + fg \frac{d^2 h}{dz^2} - \gamma^2 fgh = 0$$

Dividing both sides by fgh , we can write

$$\frac{1}{f} \frac{d^2 f}{dx^2} + \frac{1}{g} \frac{d^2 g}{dy^2} + \frac{1}{h} \frac{d^2 h}{dz^2} = \gamma^2$$

By letting each term on the left equal to a constant leads to

$$\frac{1}{f} \frac{d^2 f}{dx^2} = \gamma_x^2 \Rightarrow f = f_1 = A_1 e^{-\gamma_x x} + B_1 e^{+\gamma_x x}; f = f_2 = C_1 \cosh(\gamma_x x) + D_1 \sinh(\gamma_x x)$$

$$\frac{1}{g} \frac{d^2 g}{dy^2} = \gamma_y^2 \Rightarrow g = g_1 = A_2 e^{-\gamma_y y} + B_2 e^{+\gamma_y y}; g = g_2 = C_2 \cosh(\gamma_y y) + D_2 \sinh(\gamma_y y)$$

$$\frac{1}{h} \frac{d^2 h}{dz^2} = -\gamma_z^2 \Rightarrow h = h_1 = A_3 e^{-\gamma_z z} + B_3 e^{+\gamma_z z}; h = h_2 = C_3 \cosh(\gamma_z z) + D_3 \sinh(\gamma_z z)$$

provided that $\gamma_x^2 + \gamma_y^2 + \gamma_z^2 = \gamma^2$

$$3.5 \quad \nabla(\nabla \cdot \mathbf{E}) - \nabla \times \nabla \times \mathbf{E} = -\beta^2 \mathbf{E}, \quad \mathbf{E} = \hat{a}_r E_r + \hat{a}_\phi E_\phi + \hat{a}_z E_z$$

Using cylindrical coordinates

$$\nabla \cdot \mathbf{E} = \frac{1}{\rho} \frac{\partial}{\partial \rho} (\rho E_r) + \frac{1}{\rho} \frac{\partial E_\phi}{\partial \phi} + \frac{\partial E_z}{\partial z}$$

$$\begin{aligned} \nabla(\nabla \cdot \mathbf{E}) &= \nabla \left[\frac{1}{\rho} \frac{\partial}{\partial \rho} (\rho E_r) + \frac{1}{\rho} \frac{\partial E_\phi}{\partial \phi} + \frac{\partial E_z}{\partial z} \right] \\ &= \hat{a}_r \left\{ \frac{\partial}{\partial \rho} \left[\frac{1}{\rho} \frac{\partial}{\partial \rho} (\rho E_r) + \frac{1}{\rho} \frac{\partial E_\phi}{\partial \phi} + \frac{\partial E_z}{\partial z} \right] \right\} + \hat{a}_\phi \left\{ \frac{1}{\rho} \left[\frac{\partial}{\partial \rho} \left[\frac{1}{\rho} \frac{\partial}{\partial \rho} (\rho E_r) + \frac{1}{\rho} \frac{\partial E_\phi}{\partial \phi} + \frac{\partial E_z}{\partial z} \right] \right] \right\} \\ &\quad + \hat{a}_z \left\{ \frac{\partial}{\partial z} \left[\frac{1}{\rho} \frac{\partial}{\partial \rho} (\rho E_r) + \frac{1}{\rho} \frac{\partial E_\phi}{\partial \phi} + \frac{\partial E_z}{\partial z} \right] \right\} \end{aligned}$$

$$\begin{aligned} \nabla(\nabla \cdot \mathbf{E}) &= \hat{a}_r \left[\frac{\partial^2 E_r}{\partial \rho^2} + \frac{1}{\rho} \frac{\partial E_r}{\partial \rho} - \frac{E_r}{\rho^2} + \frac{1}{\rho} \frac{\partial^2 E_\phi}{\partial \rho \partial \phi} - \frac{1}{\rho^2} \frac{\partial E_\phi}{\partial \phi} + \frac{\partial^2 E_z}{\partial \rho \partial z} \right] \\ &\quad + \hat{a}_\phi \left[\frac{1}{\rho} \frac{\partial^2 E_\phi}{\partial \rho \partial \phi} + \frac{1}{\rho^2} \frac{\partial E_\phi}{\partial \phi} + \frac{1}{\rho^2} \frac{\partial^2 E_r}{\partial \phi^2} + \frac{1}{\rho} \frac{\partial^2 E_z}{\partial \phi \partial z} \right] + \hat{a}_z \left[\frac{\partial^2 E_z}{\partial \rho \partial z} + \frac{1}{\rho} \frac{\partial E_z}{\partial z} + \frac{1}{\rho^2} \frac{\partial^2 E_r}{\partial z^2} + \frac{\partial^2 E_\phi}{\partial z^2} \right] \end{aligned}$$

$$\nabla \times \mathbf{E} = \hat{a}_r \left[\frac{1}{\rho} \frac{\partial E_\phi}{\partial z} - \frac{\partial E_z}{\partial \phi} \right] + \hat{a}_\phi \left[\frac{\partial E_z}{\partial r} - \frac{\partial E_r}{\partial z} \right] + \hat{a}_z \left[\frac{\partial E_r}{\partial \phi} + \frac{E_\phi}{\rho} - \frac{1}{\rho} \frac{\partial E_\phi}{\partial r} \right]$$

$$\begin{aligned} \nabla \times \nabla \times \mathbf{E} &= \hat{a}_r \left\{ \frac{1}{\rho} \frac{\partial^2 E_\phi}{\partial z \partial \phi} + \frac{1}{\rho^2} \frac{\partial^2 E_\phi}{\partial \phi^2} - \frac{1}{\rho^2} \frac{\partial^2 E_r}{\partial \phi^2} - \frac{\partial^2 E_z}{\partial z^2} + \frac{\partial^2 E_r}{\partial z \partial r} \right\} \\ &\quad + \hat{a}_\phi \left\{ \frac{1}{\rho} \frac{\partial^2 E_z}{\partial z \partial \phi} - \frac{\partial^2 E_\phi}{\partial z^2} - \frac{1}{\rho^2} \frac{\partial^2 E_\phi}{\partial \phi^2} - \frac{1}{\rho} \frac{\partial^2 E_r}{\partial \phi \partial r} + \frac{E_\phi}{\rho^2} + \frac{1}{\rho} \frac{\partial^2 E_r}{\partial \phi \partial r} - \frac{1}{\rho^2} \frac{\partial^2 E_\phi}{\partial r \partial \phi} \right\} \\ &\quad + \hat{a}_z \left\{ \frac{\partial^2 E_r}{\partial z \partial r} + \frac{1}{\rho} \frac{\partial E_r}{\partial z} - \frac{\partial^2 E_\phi}{\partial z^2} - \frac{1}{\rho} \frac{\partial E_\phi}{\partial z} - \frac{1}{\rho^2} \frac{\partial^2 E_r}{\partial r^2} + \frac{1}{\rho} \frac{\partial^2 E_\phi}{\partial r \partial \phi} \right\} \end{aligned}$$

Substituting all these into the wave equation and equating identical components from the left and right sides, it leads to

r component:

$$\left[\frac{\partial^2 E_r}{\partial \rho^2} + \frac{1}{\rho} \frac{\partial E_r}{\partial \rho} + \frac{1}{\rho^2} \frac{\partial^2 E_r}{\partial \phi^2} + \frac{\partial^2 E_r}{\partial z^2} \right] - \frac{E_r}{\rho^2} - \frac{2}{\rho^2} \frac{\partial E_\phi}{\partial \phi} = -\beta^2 E_r$$

$$\text{or } \nabla^2 E_r + \left(-\frac{E_r}{\rho^2} - \frac{2}{\rho^2} \frac{\partial E_\phi}{\partial \phi} \right) = -\beta^2 E_r$$

φ component:

$$\left[\frac{\partial^2 E_\phi}{\partial \rho^2} + \frac{1}{\rho} \frac{\partial E_\phi}{\partial \rho} + \frac{1}{\rho^2} \frac{\partial^2 E_\phi}{\partial \phi^2} + \frac{\partial^2 E_\phi}{\partial z^2} \right] - \frac{1}{\rho^2} + \frac{2}{\rho^2} \frac{\partial E_r}{\partial \phi} = -\beta^2 E_\phi$$

$$\text{or } \nabla^2 E_\phi + \left[-\frac{E_\phi}{\rho^2} + \frac{2}{\rho^2} \frac{\partial E_r}{\partial \phi} \right] = -\beta^2 E_\phi$$

z component:

$$\left[\frac{\partial^2 E_z}{\partial \rho^2} + \frac{1}{\rho} \frac{\partial E_z}{\partial \rho} + \frac{1}{\rho^2} \frac{\partial^2 E_z}{\partial \phi^2} + \frac{\partial^2 E_z}{\partial z^2} \right] = -\beta^2 E_z$$

$$\text{or } \nabla^2 E_z = -\beta^2 E_z$$

3.6

$$\underline{E} = \hat{a}_\rho E_\rho(\rho, \phi, z) + \hat{a}_\phi E_\phi(\rho, \phi, z) + \hat{a}_z E_z$$

Using the rectangular to cylindrical coordinate transformation, we can write

$$\hat{a}_\rho = \hat{a}_x \cos \phi + \hat{a}_y \sin \phi$$

$$\hat{a}_\phi = -\hat{a}_x \sin \phi + \hat{a}_y \cos \phi$$

Therefore

$$\frac{\partial \hat{a}_\rho}{\partial \rho} = 0 ; \quad \frac{\partial \hat{a}_\phi}{\partial \rho} = 0 ; \quad \frac{\partial \hat{a}_z}{\partial \rho} = 0$$

$$\frac{\partial \hat{a}_\rho}{\partial \phi} = -\hat{a}_x \sin \phi + \hat{a}_y \cos \phi = \hat{a}_\phi ; \quad \frac{\partial^2 \hat{a}_\rho}{\partial \phi^2} = -\hat{a}_x \cos \phi - \hat{a}_y \sin \phi = -\hat{a}_\rho$$

$$\frac{\partial \hat{a}_\phi}{\partial \phi} = -\hat{a}_x \cos \phi - \hat{a}_y \sin \phi = -\hat{a}_\rho ; \quad \frac{\partial^2 \hat{a}_\phi}{\partial \phi^2} = \hat{a}_x \sin \phi - \hat{a}_y \cos \phi = -\hat{a}_\phi$$

$$\frac{\partial \hat{a}_z}{\partial \phi} = 0$$

$$\frac{\partial \hat{a}_\rho}{\partial z} = 0 ; \quad \frac{\partial \hat{a}_\phi}{\partial z} = 0 ; \quad \frac{\partial \hat{a}_z}{\partial z} = 0$$

$$\nabla^2 \underline{E} = \nabla^2(\hat{a}_\rho E_\rho) + \nabla^2(\hat{a}_\phi E_\phi) + \nabla^2(\hat{a}_z E_z) = \nabla^2(\hat{a}_\rho E_\rho) + \nabla^2(\hat{a}_\phi E_\phi) + \hat{a}_z \nabla^2 E_z$$

$$\nabla^2(\hat{a}_\rho E_\rho) = \frac{1}{\rho} \frac{2}{\rho} \left[\rho \frac{2}{\rho} (\hat{a}_\rho E_\rho) \right] + \frac{1}{\rho^2} \frac{\partial^2}{\partial \phi^2} (\hat{a}_\rho E_\rho) + \frac{\partial^2}{\partial z^2} (\hat{a}_\rho E_\rho)$$

$$\frac{\partial^2}{\partial \phi^2} (\hat{a}_\rho E_\rho) = \frac{2}{\partial \phi} \left[\frac{2}{\partial \phi} (\hat{a}_\rho E_\rho) \right] = \frac{2}{\partial \phi} \left[\hat{a}_\rho \frac{\partial E_\rho}{\partial \phi} + E_\rho \frac{\partial^2 E_\rho}{\partial \phi^2} \right] = \frac{2}{\partial \phi} \left[\hat{a}_\rho \frac{\partial E_\rho}{\partial \phi} + \hat{a}_\phi E_\phi \right]$$

$$= \hat{a}_\rho \frac{\partial^2 E_\rho}{\partial \phi^2} + \frac{\partial E_\rho}{\partial \phi} \frac{\partial \hat{a}_\rho}{\partial \phi} + \hat{a}_\phi \frac{\partial E_\rho}{\partial \phi} + E_\rho \frac{\partial \hat{a}_\phi}{\partial \phi}$$

$$= \hat{a}_\rho \frac{\partial^2 E_\rho}{\partial \phi^2} + \hat{a}_\phi \frac{\partial E_\rho}{\partial \phi} + \hat{a}_\phi \frac{\partial E_\rho}{\partial \phi} - \hat{a}_\rho E_\rho = -\hat{a}_\rho E_\rho + \hat{a}_\rho \frac{\partial^2 E_\rho}{\partial \phi^2} + \hat{a}_\phi 2 \frac{\partial E_\rho}{\partial \phi}$$

$$\frac{\partial^2}{\partial z^2} (\hat{a}_\rho E_\rho) = \frac{2}{\partial z} \left[\frac{2}{\partial z} (\hat{a}_\rho E_\rho) \right] = \frac{2}{\partial z} \left[\hat{a}_\rho \frac{\partial E_\rho}{\partial z} + E_\rho \frac{\partial^2 E_\rho}{\partial z^2} \right] = \hat{a}_\rho \frac{\partial^2 E_\rho}{\partial z^2} + \frac{\partial E_\rho}{\partial z} \frac{\partial \hat{a}_\rho}{\partial z}$$

Therefore

$$\nabla^2(\hat{a}_\rho E_\rho) = \hat{a}_\rho \left[\frac{\partial^2 E_\rho}{\partial \rho^2} + \frac{1}{\rho} \frac{\partial E_\rho}{\partial \rho} + \frac{1}{\rho^2} \left(\frac{\partial^2 E_\rho}{\partial \phi^2} - E_\rho \right) + \frac{\partial^2 E_\rho}{\partial z^2} \right] + \hat{a}_\phi \frac{2}{\rho^2} \frac{\partial E_\rho}{\partial \phi}$$

Similarly

$$\nabla^2(\hat{a}_\phi E_\phi) = \frac{\partial^2}{\partial \phi^2} (\hat{a}_\phi E_\phi) + \frac{1}{\rho} \frac{2}{\rho} (\hat{a}_\phi E_\phi) + \frac{1}{\rho^2} \frac{\partial^2}{\partial \phi^2} (\hat{a}_\phi E_\phi) + \hat{a}_\phi \frac{\partial^2 E_\phi}{\partial \phi^2}$$

$$= -\hat{a}_\phi \frac{2}{\rho^2} \frac{\partial E_\phi}{\partial \phi} + \hat{a}_\phi \left[\frac{\partial^2 E_\phi}{\partial \phi^2} + \frac{1}{\rho} \frac{\partial E_\phi}{\partial \rho} + \frac{1}{\rho^2} \frac{\partial^2 E_\phi}{\partial \phi^2} - \frac{E_\phi}{\rho^2} + \frac{\partial^2 E_\phi}{\partial z^2} \right]$$

$$\nabla^2(\hat{a}_z E_z) = \hat{a}_z \nabla^2 E_z = \hat{a}_z \left[\frac{\partial^2 E_z}{\partial \rho^2} + \frac{1}{\rho} \frac{\partial E_z}{\partial \rho} + \frac{1}{\rho^2} \frac{\partial^2 E_z}{\partial \phi^2} + \frac{\partial^2 E_z}{\partial z^2} \right] \quad (\text{continued})$$

3.6 cont'd. Substituting all these terms into the vector wave equation and collecting terms with the same direction vectors leads to

$$\left[\frac{\partial^2 E_\rho}{\partial \rho^2} + \frac{1}{\rho} \frac{\partial E_\rho}{\partial \rho} + \frac{1}{\rho^2} \frac{\partial^2 E_\rho}{\partial \phi^2} + \frac{\partial^2 E_\rho}{\partial z^2} \right] - \frac{E_\rho}{\rho^2} - \frac{2}{\rho^2} \frac{\partial E_\phi}{\partial \phi} = -\beta^2 E_\rho$$

$$\text{or } \nabla^2 E_\rho + \left[-\frac{E_\rho}{\rho^2} - \frac{2}{\rho^2} \frac{\partial E_\phi}{\partial \phi} \right] = -\beta^2 E_\rho$$

ϕ component:

$$\left[\frac{\partial^2 E_\phi}{\partial \rho^2} + \frac{1}{\rho} \frac{\partial E_\phi}{\partial \rho} + \frac{1}{\rho^2} \frac{\partial^2 E_\phi}{\partial \phi^2} + \frac{\partial^2 E_\phi}{\partial z^2} \right] - \frac{E_\phi}{\rho^2} + \frac{2}{\rho^2} \frac{\partial E_\rho}{\partial \phi} = -\beta^2 E_\phi$$

$$\text{or } \nabla^2 E_\phi + \left[-\frac{E_\phi}{\rho^2} + \frac{2}{\rho^2} \frac{\partial E_\rho}{\partial \phi} \right] = -\beta^2 E_\phi$$

z component:

$$\left[\frac{\partial^2 E_z}{\partial \rho^2} + \frac{1}{\rho} \frac{\partial E_z}{\partial \rho} + \frac{1}{\rho^2} \frac{\partial^2 E_z}{\partial \phi^2} + \frac{\partial^2 E_z}{\partial z^2} \right] = -\beta^2 E_z$$

$$\text{or } \nabla^2 E_z = -\beta^2 E_z$$

3.7 For large arguments the Bessel function of the first kind and the Hankel function of the second kind can be written according to (IV-13) and (IV-17) as

$$J_m(\beta\rho) \xrightarrow{\beta\rho \rightarrow \text{large}} \sqrt{\frac{2}{\pi\beta\rho}} \cos\left(\beta\rho - \frac{\pi}{4} - \frac{m\pi}{2}\right)$$

$$H_m^{(2)}(\beta\rho) \xrightarrow{\beta\rho \rightarrow \text{large}} \sqrt{\frac{2}{\pi\beta\rho}} e^{-j[\beta\rho - \frac{m\pi}{2} - \frac{\pi}{4}]}$$

which are recognized as decaying cosine and complex exponential functions. Thus they represent, respectively, standing and traveling waves.

3.8 For large argument the Hankel functions of the first and second kind can be written according to (IV-16) and (IV-17) as

$$H_m^{(1)}(\beta\rho) \xrightarrow{\beta\rho \rightarrow \text{large}} \sqrt{\frac{2}{\pi\beta\rho}} e^{j[\beta\rho - \frac{m\pi}{2} - \frac{\pi}{4}]}$$

$$H_m^{(2)}(\beta\rho) \xrightarrow{\beta\rho \rightarrow \text{large}} \sqrt{\frac{2}{\pi\beta\rho}} e^{-j[\beta\rho - \frac{m\pi}{2} - \frac{\pi}{4}]}$$

(continued)

3.8 cont'd. For an $e^{j\omega t}$ time convention the time varying field represented by the Hankel functions can be written as

$$H_m^{(1)}(\beta p; t) = \operatorname{Re}[H_m^{(0)}(\beta p) e^{j\omega t}] = \sqrt{\frac{2}{\pi \beta p}} \cos[\omega t + \beta p - \frac{m\pi}{2} - \frac{\pi}{4}]$$

$$H_m^{(2)}(\beta p; t) = \operatorname{Re}[H_m^{(2)}(\beta p) e^{j\omega t}] = \sqrt{\frac{2}{\pi \beta p}} \cos[\omega t - \beta p + \frac{m\pi}{2} + \frac{\pi}{4}]$$

For a constant wavefront

$$\omega t + \beta p - \frac{m\pi}{2} - \frac{\pi}{4} = C_1 \Rightarrow \omega(z) + \beta \frac{dz}{dt} = 0 \Rightarrow \omega + \beta v_p = 0 \Rightarrow v_p = -\frac{\omega}{\beta}$$

$$\omega t - \beta p + \frac{m\pi}{2} + \frac{\pi}{4} = C_2 \Rightarrow \omega(z) - \beta \frac{dz}{dt} = 0 \Rightarrow \omega - \beta v_p = 0 \Rightarrow v_p = +\frac{\omega}{\beta}$$

Therefore $H_m^{(1)}(\beta p)$ represents for an $e^{j\omega t}$ time convention waves traveling in the $-p$ direction while $H_m^{(2)}(\beta p)$ represents waves traveling in the $+p$ direction. It is obvious, by following a similar procedure, that the opposite would be true for an $e^{-j\omega t}$ time convention.

3.9 For large argument the Bessel function of the first kind can be written according to (IV-13) as

$$J_m(\gamma p) \xrightarrow{\gamma p \rightarrow \text{large}} \sqrt{\frac{2}{\pi \gamma p}} \cos\left[\gamma p - \frac{\pi}{4} - \frac{m\pi}{2}\right] = \sqrt{\frac{2}{\pi \gamma p}} \cos\left[\alpha_e p - \frac{\pi}{4} - \frac{m\pi}{2} + j\beta p\right]$$

$$\xrightarrow{\gamma p \rightarrow \text{large}} \sqrt{\frac{2}{\pi \gamma p}} \cos(\alpha_e p + j\beta p) = \sqrt{\frac{2}{\pi \gamma p}} \left\{ \cos(\alpha_e p) \cos(j\beta p) - \sin(\alpha_e p) \sin(j\beta p) \right\}$$

$$\xrightarrow{\gamma p \rightarrow \text{large}} \sqrt{\frac{2}{\pi \gamma p}} \left\{ \cos(\alpha_e p) \cosh(\beta p) - j \sin(\alpha_e p) \sinh(\beta p) \right\}$$

$$\xrightarrow{\gamma p \rightarrow \text{large}} \sqrt{\frac{2}{\pi \gamma p}} \left\{ \cos(\alpha_e p) \left(\frac{e^{\beta p} + e^{-\beta p}}{2} \right) - j \sin(\alpha_e p) \left(\frac{e^{\beta p} - e^{-\beta p}}{2} \right) \right\}$$

$$J_m(\gamma p) \xrightarrow{\gamma p \rightarrow \text{large}} \sqrt{\frac{2}{\pi \gamma p}} \left\{ \frac{e^{\beta p}}{2} \left[\cos(\alpha_e p) - j \sin(\alpha_e p) \right] + \frac{e^{-\beta p}}{2} \left[\cos(\alpha_e p) + j \sin(\alpha_e p) \right] \right\}$$

where $\gamma = \alpha + j\beta$ and $\alpha_e p = \alpha p - \frac{\pi}{4} - \frac{m\pi}{2}$

It is apparent from the above that the Bessel functions of complex argument represent attenuating standing waves.

3.10 For large argument the Hankel functions of the first and second kind can be written according to (IV-16) and (IV-17) as

$$H_m^{(1)}(\gamma p) \xrightarrow{\gamma p \rightarrow \text{large}} \sqrt{\frac{2}{\pi \gamma p}} e^{j[-j\gamma p - \frac{m\pi}{2} - \frac{\pi}{4}]} = \sqrt{\frac{2}{\pi \gamma p}} e^{j[-j(\alpha + j\beta)p - \frac{m\pi}{2} - \frac{\pi}{4}]}$$

$$\xrightarrow{\gamma p \rightarrow \text{large}} \sqrt{\frac{2}{\pi \gamma p}} e^{+\alpha p} e^{-j(\beta p - \frac{m\pi}{2} - \frac{\pi}{4})} = \sqrt{\frac{2}{\pi \gamma p}} e^{\alpha p} e^{+j(\beta p - \frac{m\pi}{2})}$$

which represents an attenuating traveling wave in the $-p$ direction because of the $e^{\alpha p}$ and $e^{j\beta p}$ factors.

$$H_m^{(2)}(\gamma p) \xrightarrow{\gamma p \rightarrow \text{large}} \sqrt{\frac{2}{\pi \gamma p}} e^{-j[-j\gamma p - \frac{m\pi}{2} - \frac{\pi}{4}]} = \sqrt{\frac{2}{\pi \gamma p}} e^{+j[j\gamma p + \frac{m\pi}{2} + \frac{\pi}{4}]}$$

$$\xrightarrow{\gamma p \rightarrow \text{large}} \sqrt{\frac{2}{\pi \gamma p}} e^{-\alpha p} e^{-j(\beta p - \frac{m\pi}{2} - \frac{\pi}{4})}$$

which represents an attenuating traveling wave in the $+p$ direction because of the $e^{-\alpha p}$ and $e^{-j\beta p}$ factors.

3.11 $\underline{E} = \hat{a}_r E_r(r, \theta, \phi) + \hat{a}_\theta E_\theta(r, \theta, \phi) + \hat{a}_\phi E_\phi(r, \theta, \phi)$

$$\nabla(\nabla \cdot \underline{E}) - \nabla \times \nabla \times \underline{E} = -\beta^2 \underline{E}$$

Using spherical coordinates

$$\nabla \cdot \underline{E} = \frac{1}{r^2} \frac{\partial}{\partial r} (r^2 E_r) + \frac{1}{r \sin \theta} \frac{\partial}{\partial \theta} (\sin \theta E_\theta) + \frac{1}{r \sin \theta} \frac{\partial E_\phi}{\partial \phi}$$

$$\begin{aligned} \nabla(\nabla \cdot \underline{E}) &= \hat{a}_r \frac{\partial}{\partial r} [\nabla \cdot \underline{E}] + \hat{a}_\theta \frac{1}{r} \frac{\partial}{\partial \theta} [\nabla \cdot \underline{E}] + \hat{a}_\phi \frac{1}{r \sin \theta} \frac{\partial}{\partial \phi} [\nabla \cdot \underline{E}] \\ &= \hat{a}_r \frac{\partial}{\partial r} \left[\frac{1}{r^2} \frac{\partial}{\partial r} (r^2 E_r) + \frac{1}{r \sin \theta} \frac{\partial}{\partial \theta} (\sin \theta E_\theta) + \frac{1}{r \sin \theta} \frac{\partial E_\phi}{\partial \phi} \right] \\ &\quad + \hat{a}_\theta \frac{1}{r} \frac{\partial}{\partial \theta} \left[\frac{1}{r^2} \frac{\partial}{\partial r} (r^2 E_r) + \frac{1}{r \sin \theta} \frac{\partial}{\partial \theta} (\sin \theta E_\theta) + \frac{1}{r \sin \theta} \frac{\partial E_\phi}{\partial \phi} \right] \\ &\quad + \hat{a}_\phi \frac{1}{r \sin \theta} \frac{\partial}{\partial \phi} \left[\frac{1}{r^2} \frac{\partial}{\partial r} (r^2 E_r) + \frac{1}{r \sin \theta} \frac{\partial}{\partial \theta} (\sin \theta E_\theta) + \frac{1}{r \sin \theta} \frac{\partial E_\phi}{\partial \phi} \right] \end{aligned}$$

(continued)

3.11 cont'd.

$$\nabla \times \underline{E} = \hat{a}_r \left[\frac{1}{r \sin \theta} \left(E_\phi \sin \theta - \frac{\partial E_r}{\partial \phi} \right) + \hat{a}_\theta \left[\frac{1}{r \sin \theta} \frac{\partial E_r}{\partial \phi} - \frac{2}{r} (r E_\phi) \right] + \hat{a}_\phi \left[\frac{2}{r} (r E_\phi) - \frac{\partial E_r}{\partial \theta} \right] \right]$$

$$= \hat{a}_r E_R + \hat{a}_\theta E_T + \hat{a}_\phi E_P$$

$$\text{where } E_R = \frac{1}{r \sin \theta} \left[\frac{2}{\partial \theta} (E_\phi \sin \theta) - \frac{\partial E_\phi}{\partial \theta} \right]$$

$$E_T = \frac{1}{r} \left[\frac{1}{\sin \theta} \frac{\partial E_r}{\partial \phi} - \frac{2}{r} (r E_\phi) \right]$$

$$E_P = \frac{1}{r} \left[\frac{2}{r} (r E_\phi) - \frac{\partial E_r}{\partial \theta} \right]$$

$$\nabla \times \nabla \times \underline{E} = \hat{a}_r \frac{1}{r \sin \theta} \left[\frac{2}{\partial \theta} (E_P \sin \theta) - \frac{\partial}{\partial \theta} (E_T) \right]$$

$$+ \hat{a}_\theta \frac{1}{r} \left[\frac{1}{\sin \theta} \frac{\partial}{\partial \phi} (E_R) - \frac{2}{r} (r E_P) \right]$$

$$+ \hat{a}_\phi \frac{1}{r} \left[\frac{2}{r} (r E_P) - \frac{\partial}{\partial \theta} (E_R) \right]$$

Using all these it can be shown that $\nabla(\nabla \cdot \underline{E}) - \nabla \times \nabla \times \underline{E} = -\beta^2 \underline{E}$ reduces by equating identical components from the left and right sides to

r component:

$$\left[\frac{1}{r^2} \frac{\partial}{\partial r} (r^2 E_r) + \frac{1}{r^2 \sin \theta} \frac{\partial}{\partial \theta} \left(\sin \theta \frac{\partial E_r}{\partial \theta} \right) + \frac{1}{r^2 \sin^2 \theta} \frac{\partial^2 E_r}{\partial \phi^2} \right] - \frac{2}{r^2} \left[E_r + E_\theta \cot \theta + \csc \theta \frac{\partial E_\phi}{\partial \theta} + \frac{\partial E_\theta}{\partial \phi} \right] = -\beta^2 E_r$$

$$\text{or } \nabla^2 E_r - \frac{2}{r^2} \left[E_r + E_\theta \cot \theta + \csc \theta \frac{\partial E_\phi}{\partial \theta} + \frac{\partial E_\theta}{\partial \phi} \right] = -\beta^2 E_r$$

θ component:

$$\left[\frac{1}{r^2} \frac{\partial}{\partial r} (r^2 E_\theta) + \frac{1}{r^2 \sin \theta} \frac{\partial}{\partial \theta} \left(\sin \theta \frac{\partial E_\theta}{\partial \theta} \right) + \frac{1}{r^2 \sin^2 \theta} \frac{\partial^2 E_\theta}{\partial \phi^2} \right] - \frac{2}{r^2} \left[E_\theta \csc^2 \theta - 2 \frac{\partial E_r}{\partial \theta} + 2 \cot \theta \csc \theta \frac{\partial E_\phi}{\partial \theta} \right] = -\beta^2 E_\theta$$

$$\text{or } \nabla^2 E_\theta - \frac{1}{r^2} \left[E_\theta \csc^2 \theta - 2 \frac{\partial E_r}{\partial \theta} + 2 \cot \theta \csc \theta \frac{\partial E_\phi}{\partial \theta} \right] = -\beta^2 E_\theta$$

φ component:

$$\left[\frac{1}{r^2} \frac{\partial}{\partial r} (r^2 E_\phi) + \frac{1}{r^2 \sin \theta} \frac{\partial}{\partial \theta} \left(\sin \theta \frac{\partial E_\phi}{\partial \theta} \right) + \frac{1}{r^2 \sin^2 \theta} \frac{\partial^2 E_\phi}{\partial \phi^2} \right] - \frac{2}{r^2} \left[E_\phi \csc^2 \theta - 2 \csc \theta \frac{\partial E_r}{\partial \phi} - 2 \cot \theta \csc \theta \frac{\partial E_\theta}{\partial \phi} \right] = -\beta^2 E_\phi$$

$$\text{or } \nabla^2 E_\phi - \frac{1}{r^2} \left[E_\phi \csc^2 \theta - 2 \csc \theta \frac{\partial E_r}{\partial \phi} - 2 \cot \theta \csc \theta \frac{\partial E_\theta}{\partial \phi} \right] = -\beta^2 E_\phi$$

$$3.12 \quad \underline{E} = \hat{a}_r E_r(r, \theta, \phi) + \hat{a}_\theta E_\theta(r, \theta, \phi) + \hat{a}_\phi E_\phi(r, \theta, \phi)$$

Using the rectangular to spherical coordinate transformation, we can write

$$\hat{a}_r = \hat{a}_x \sin \theta \cos \phi + \hat{a}_y \sin \theta \sin \phi + \hat{a}_z \cos \theta$$

$$\hat{a}_\theta = \hat{a}_x \cos \theta \cos \phi + \hat{a}_y \cos \theta \sin \phi - \hat{a}_z \sin \theta$$

$$\hat{a}_\phi = -\hat{a}_x \sin \phi + \hat{a}_y \cos \phi$$

Therefore

$$\frac{\partial \hat{a}_r}{\partial r} = 0 ; \quad \frac{\partial \hat{a}_\theta}{\partial r} = 0 ; \quad \frac{\partial \hat{a}_\phi}{\partial r} = 0$$

$$\frac{\partial \hat{a}_r}{\partial \theta} = \hat{a}_x \cos \theta \cos \phi + \hat{a}_y \cos \theta \sin \phi - \hat{a}_z \sin \theta = +\hat{a}_\theta$$

$$\frac{\partial \hat{a}_\theta}{\partial \theta} = -\hat{a}_x \sin \theta \cos \phi - \hat{a}_y \sin \theta \sin \phi - \hat{a}_z \cos \theta = -\hat{a}_r$$

$$\frac{\partial \hat{a}_\phi}{\partial \theta} = 0$$

$$\frac{\partial \hat{a}_r}{\partial \phi} = -\hat{a}_x \sin \theta \sin \phi + \hat{a}_y \sin \theta \cos \phi = \sin \theta [-\hat{a}_x \sin \phi + \hat{a}_y \cos \phi] = \sin \theta \hat{a}_\phi$$

$$\frac{\partial \hat{a}_\theta}{\partial \phi} = -\hat{a}_x \cos \theta \sin \phi + \hat{a}_y \cos \theta \cos \phi = \cos \theta [-\hat{a}_x \sin \phi + \hat{a}_y \cos \phi] = \cos \theta \hat{a}_\phi$$

$$\begin{aligned} \frac{\partial \hat{a}_\phi}{\partial \phi} &= -\hat{a}_x \cos \phi - \hat{a}_y \sin \phi = -\sin \theta \hat{a}_r - \cos \theta \hat{a}_\theta = -\hat{a}_x \sin^2 \theta \cos \phi - \hat{a}_y \sin^2 \theta \sin \phi - \hat{a}_z \sin \theta \cos \theta \\ &= -\hat{a}_x \cos \phi (\sin^2 \theta + \cos^2 \theta) - \hat{a}_y \sin \phi (\sin^2 \theta + \cos^2 \theta) - \hat{a}_x \cos^2 \theta \cos \phi - \hat{a}_y \cos^2 \theta \sin \phi + \hat{a}_z \sin \theta \cos \theta \end{aligned}$$

$$\frac{\partial^2 \underline{E}}{\partial \phi^2} = -\hat{a}_x \cos \phi - \hat{a}_y \sin \phi$$

$$\nabla^2 \underline{E} = \underbrace{\frac{1}{r^2} \frac{\partial}{\partial r} \left(r^2 \frac{\partial \underline{E}}{\partial r} \right)}_{I} + \underbrace{\frac{1}{r^2 \sin \theta} \frac{\partial}{\partial \theta} \left(\sin \theta \frac{\partial \underline{E}}{\partial \theta} \right)}_{II} + \underbrace{\frac{1}{r^2 \sin^2 \theta} \frac{\partial^2 \underline{E}}{\partial \phi^2}}_{III}$$

$$I. \quad \frac{1}{r^2} \frac{\partial}{\partial r} \left[r^2 \frac{\partial (\hat{a}_r E_r)}{\partial r} \right] + \frac{1}{r^2} \frac{\partial}{\partial r} \left[r^2 \frac{\partial (\hat{a}_\theta E_\theta)}{\partial r} \right] + \frac{1}{r^2} \frac{\partial}{\partial r} \left[r^2 \frac{\partial (\hat{a}_\phi E_\phi)}{\partial r} \right] = \frac{1}{r^2} \frac{\partial}{\partial r} \left(r^2 \frac{\partial \underline{E}}{\partial r} \right)$$

$$\hat{a}_r \frac{1}{r^2} \frac{\partial}{\partial r} \left[r^2 \frac{\partial E_r}{\partial r} \right] + \hat{a}_\theta \frac{1}{r^2} \frac{\partial}{\partial r} \left[r^2 \frac{\partial E_\theta}{\partial r} \right] + \hat{a}_\phi \frac{1}{r^2} \frac{\partial}{\partial r} \left[r^2 \frac{\partial E_\phi}{\partial r} \right] = \frac{1}{r^2} \frac{\partial}{\partial r} \left(r^2 \frac{\partial \underline{E}}{\partial r} \right)$$

$$II. \quad \frac{1}{r^2 \sin \theta} \frac{\partial}{\partial \theta} \left(\sin \theta \frac{\partial \underline{E}}{\partial \theta} \right) = \frac{1}{r^2 \sin \theta} \frac{\partial}{\partial \theta} \left[\sin \theta \frac{\partial (\hat{a}_r E_r)}{\partial \theta} \right] + \frac{1}{r^2 \sin \theta} \frac{\partial}{\partial \theta} \left[\sin \theta \frac{\partial (\hat{a}_\theta E_\theta)}{\partial \theta} \right] + \frac{1}{r^2 \sin \theta} \frac{\partial}{\partial \theta} \left[\sin \theta \frac{\partial (\hat{a}_\phi E_\phi)}{\partial \theta} \right]$$

$$\frac{\partial}{\partial \theta} (\underline{E}) = \frac{2}{\partial \theta} \left(\hat{a}_r E_r + \hat{a}_\theta E_\theta + \hat{a}_\phi E_\phi \right) = \hat{a}_r \left(\frac{\partial E_r}{\partial \theta} - E_\theta \right) + \hat{a}_\theta \left(\frac{\partial E_\theta}{\partial \theta} + E_r \right) + \hat{a}_\phi \frac{\partial E_\phi}{\partial \theta}$$

(continued)

3.12 cont'd.

$$\begin{aligned}\frac{\partial^2 \vec{E}}{\partial \theta^2} &= \hat{a}_r \left[\frac{\partial^2 E_r}{\partial \theta^2} - \frac{\partial E_\theta}{\partial \theta} - \left(\frac{\partial^2 E_\theta}{\partial \theta^2} + E_r \right) \right] + \hat{a}_\theta \left[\frac{\partial^2 E_\theta}{\partial \theta^2} + \frac{\partial E_r}{\partial \theta} + \left(\frac{\partial^2 E_r}{\partial \theta^2} - E_\theta \right) \right] + \hat{a}_\phi \frac{\partial^2 E_\phi}{\partial \theta^2} \\ &= \hat{a}_r \left(\frac{\partial^2 E_r}{\partial \theta^2} - 2 \frac{\partial E_\theta}{\partial \theta} - E_r \right) + \hat{a}_\theta \left(\frac{\partial^2 E_\theta}{\partial \theta^2} + 2 \frac{\partial E_r}{\partial \theta} - E_\theta \right) + \hat{a}_\phi \frac{\partial^2 E_\phi}{\partial \theta^2}\end{aligned}$$

Therefore

$$\begin{aligned}\frac{1}{r^2 \sin \theta} \frac{\partial}{\partial \theta} \left(\sin \theta \frac{\partial \vec{E}}{\partial \theta} \right) &= \frac{\cos \theta}{r^2 \sin \theta} \frac{\partial \vec{E}}{\partial \theta} + \frac{1}{r^2} \frac{\partial^2 \vec{E}}{\partial \theta^2} \\ &= \hat{a}_r \frac{1}{r^2} \left[\frac{\partial^2 E_r}{\partial \theta^2} + \frac{\cos \theta}{\sin \theta} \frac{\partial E_r}{\partial \theta} - 2 \frac{\partial E_\theta}{\partial \theta} - E_r - E_\theta \frac{\cos \theta}{\sin \theta} \right] \\ &\quad + \hat{a}_\theta \frac{1}{r^2} \left[\frac{\partial^2 E_\theta}{\partial \theta^2} + \frac{\cos \theta}{\sin \theta} \frac{\partial E_\theta}{\partial \theta} + 2 \frac{\partial E_r}{\partial \theta} + E_r \frac{\cos \theta}{\sin \theta} - E_\theta \right] \\ &\quad + \hat{a}_\phi \frac{1}{r^2} \left[\frac{\partial^2 E_\phi}{\partial \theta^2} + \frac{\cos \theta}{\sin \theta} \frac{\partial E_\phi}{\partial \theta} \right]\end{aligned}$$

or

$$\begin{aligned}\frac{1}{r^2 \sin \theta} \frac{\partial}{\partial \theta} \left(\sin \theta \frac{\partial \vec{E}}{\partial \theta} \right) &= \frac{1}{r^2} \left\{ \hat{a}_r \left[\frac{1}{\sin \theta} \frac{\partial}{\partial \theta} \left(\sin \theta \frac{\partial E_r}{\partial \theta} \right) - 2 \frac{\partial E_\theta}{\partial \theta} - E_\theta \frac{\cos \theta}{\sin \theta} - E_r \right] \right. \\ &\quad + \hat{a}_\theta \left[\frac{1}{\sin \theta} \frac{\partial}{\partial \theta} \left(\sin \theta \frac{\partial E_\theta}{\partial \theta} \right) + 2 \frac{\partial E_r}{\partial \theta} + E_r \frac{\cos \theta}{\sin \theta} - E_\theta \right] \\ &\quad \left. + \hat{a}_\phi \left[\frac{1}{\sin \theta} \frac{\partial}{\partial \theta} \left(\sin \theta \frac{\partial E_\phi}{\partial \theta} \right) \right] \right\}\end{aligned}$$

$$\text{III. } \frac{1}{r^2 \sin^2 \theta} \frac{\partial^2 \vec{E}}{\partial \phi^2} = \frac{1}{r^2 \sin^2 \theta} \frac{\partial^2}{\partial \phi^2} \left[\hat{a}_r E_r + \hat{a}_\theta E_\theta + \hat{a}_\phi E_\phi \right]$$

$$\frac{\partial \vec{E}}{\partial \phi} = \hat{a}_r \left(\frac{\partial E_r}{\partial \phi} - E_\phi \sin \theta \right) + \hat{a}_\theta \left(\frac{\partial E_\theta}{\partial \phi} - E_\phi \cos \theta \right) + \hat{a}_\phi \left(\frac{\partial E_\phi}{\partial \phi} + E_r \sin \theta + E_\theta \cos \theta \right)$$

$$\begin{aligned}\frac{\partial^2 \vec{E}}{\partial \phi^2} &= \hat{a}_r \left[\frac{\partial^2 E_r}{\partial \phi^2} - \frac{\partial E_\phi}{\partial \phi} \sin \theta - \left(\frac{\partial^2 E_\phi}{\partial \phi^2} + E_r \sin \theta + E_\theta \cos \theta \right) \sin \theta \right] \\ &\quad + \hat{a}_\theta \left[\frac{\partial^2 E_\theta}{\partial \phi^2} - \frac{\partial E_\phi}{\partial \phi} \cos \theta - \left(\frac{\partial^2 E_\phi}{\partial \phi^2} + E_r \sin \theta + E_\theta \cos \theta \right) \cos \theta \right] \\ &\quad + \hat{a}_\phi \left[\frac{\partial^2 E_\phi}{\partial \phi^2} + \frac{\partial E_r}{\partial \phi} \sin \theta + \frac{\partial E_\theta}{\partial \phi} \cos \theta + \left(\frac{\partial^2 E_r}{\partial \phi^2} - E_\phi \sin \theta \right) \sin \theta + \left(\frac{\partial^2 E_\theta}{\partial \phi^2} - E_\phi \cos \theta \right) \cos \theta \right]\end{aligned}$$

Therefore

$$\begin{aligned}\frac{1}{r^2 \sin^2 \theta} \frac{\partial^2 \vec{E}}{\partial \phi^2} &= \frac{1}{r^2 \sin^2 \theta} \left\{ \hat{a}_r \left[\frac{\partial^2 E_r}{\partial \phi^2} - 2 \frac{\partial E_\phi}{\partial \phi} \sin \theta - E_r \sin^2 \theta - E_\phi \sin \theta \cos \theta \right] \right. \\ &\quad + \hat{a}_\theta \left[\frac{\partial^2 E_\theta}{\partial \phi^2} - 2 \frac{\partial E_\phi}{\partial \phi} \cos \theta - E_r \sin \theta \cos \theta - E_\theta \cos^2 \theta \right] \\ &\quad \left. + \hat{a}_\phi \left[\frac{\partial^2 E_\phi}{\partial \phi^2} + 2 \frac{\partial E_r}{\partial \phi} \sin \theta + 2 \frac{\partial E_\theta}{\partial \phi} \cos \theta - E_\phi \right] \right\} \\ &\quad (\text{continued})\end{aligned}$$

3.12 cont'd. Finally using I, II and III, it can be shown that

$$\nabla^2 E = \nabla^2 (\hat{a}_r E_r + \hat{a}_\theta E_\theta + \hat{a}_\phi E_\phi) = -\beta^2 E = -\beta^2 (\hat{a}_r E_r + \hat{a}_\theta E_\theta + \hat{a}_\phi E_\phi)$$

reduces by equating identical components from the left and right sides to

$$\nabla^2 E_r - \frac{2}{r^2} \left[E_r + E_\theta \cot \theta + \csc \theta \frac{\partial E_\theta}{\partial \theta} + \frac{\partial E_\phi}{\partial \theta} \right] = -\beta^2 E_r$$

$$\nabla^2 E_\theta - \frac{4}{r^2} \left[E_\theta \csc^2 \theta - 2 \frac{\partial E_r}{\partial \theta} + 2 \cot \theta \csc \theta \frac{\partial E_\phi}{\partial \theta} \right] = -\beta^2 E_\theta$$

$$\nabla^2 E_\phi - \frac{1}{r^2} \left[E_\phi \csc^2 \theta - 2 \csc \theta \frac{\partial E_r}{\partial \phi} - 2 \cot \theta \csc \theta \frac{\partial E_\theta}{\partial \phi} \right] = -\beta^2 E_\phi$$

where

$$\nabla^2 \psi = \frac{1}{r^2} \frac{\partial^2}{\partial r^2} (r^2 \psi) + \frac{1}{r^2 \sin \theta} \frac{\partial^2}{\partial \theta^2} (\sin \theta \frac{\partial \psi}{\partial \theta}) + \frac{1}{r^2 \sin^2 \theta} \frac{\partial^2 \psi}{\partial \phi^2}$$

3.13 According to (3-90a)

$$j_n(\beta r) = \sqrt{\frac{\pi}{2\beta r}} J_{n+\frac{1}{2}}(\beta r) \xrightarrow{\beta r \rightarrow \text{large}} \sqrt{\frac{\pi}{2\beta r}} \sqrt{\frac{2}{\pi \beta r}} \cos(\beta r - \frac{\pi}{2} - \frac{n\pi}{2})$$

$$j_n(\beta r) \xrightarrow{\beta r \rightarrow \text{large}} \frac{1}{\beta r} \cos(\beta r - \frac{\pi}{2} - \frac{n\pi}{2})$$

which represents a standing wave in the r direction. The same can be shown for the spherical Bessel function of (3-90b).

3.14 According to (3-91)

$$h_n^{(2)}(\beta r) = \sqrt{\frac{\pi}{2\beta r}} H_{n+\frac{1}{2}}^{(2)}(\beta r) \xrightarrow{\beta r \rightarrow \text{large}} \sqrt{\frac{\pi}{2\beta r}} \sqrt{\frac{2}{\pi \beta r}} e^{-j[\beta r - \frac{n\pi}{2} - \frac{\pi}{2}]}$$

$$h_n^{(2)}(\beta r) \xrightarrow{\beta r \rightarrow \text{large}} \frac{1}{\beta r} e^{-j[\beta r - \frac{n\pi}{2} - \frac{\pi}{2}]}$$

which represents a traveling wave in the $+r$ direction (for $e^{i\omega t}$).

Similarly

$$h_n^{(1)}(\beta r) \xrightarrow{\beta r \rightarrow \text{large}} \frac{1}{\beta r} e^{j[\beta r - \frac{n\pi}{2} - \frac{\pi}{2}]}$$

which represents a traveling wave in the $-r$ direction (for $e^{+i\omega t}$).

3.15 According to (V-9) the Legendre polynomials are expressed in terms of cosine functions of the angle θ . Since cosine functions are used to represent standing wave functions, then the Legendre polynomials, and in turn the Legendre functions of (V-19a), are used to represent standing waves.

3.16 According to (3-90a)

$$j_n(\beta r) \equiv \sqrt{\frac{\pi}{2\beta r}} J_{n+\frac{1}{2}}(\beta r)$$

Since

$$\hat{J}_n(\beta r) \equiv \beta r j_n(\beta r)$$

then

$$\hat{J}_n(\beta r) = \beta r j_n(\beta r) = \beta r \sqrt{\frac{\pi}{2\beta r}} J_{n+\frac{1}{2}}(\beta r) = \sqrt{\frac{\pi \beta r}{2}} J_{n+\frac{1}{2}}(\beta r)$$

According to

$$h_n^{(a)}(\beta r) \equiv \sqrt{\frac{\pi}{2\beta r}} H_{n+\frac{1}{2}}^{(a)}(\beta r)$$

Since

$$\hat{H}_n^{(a)}(\beta r) \equiv \beta r h_n^{(a)}(\beta r)$$

then

$$\hat{H}_n^{(a)}(\beta r) = \beta r h_n^{(a)}(\beta r) = \beta r \sqrt{\frac{\pi}{2\beta r}} H_{n+\frac{1}{2}}^{(a)}(\beta r) = \sqrt{\frac{\pi \beta r}{2}} H_{n+\frac{1}{2}}^{(a)}(\beta r)$$

CHAPTER 4

4.1 According to Maxwell's equation of $\nabla \times \underline{H} = j\omega \epsilon \underline{E}$

$$\hat{a}_x \left(\frac{\partial H_z}{\partial y} - \frac{\partial H_y}{\partial z} \right) + \hat{a}_y \left(\frac{\partial H_x}{\partial z} - \frac{\partial H_z}{\partial x} \right) + \hat{a}_z \left(\frac{\partial H_y}{\partial x} - \frac{\partial H_x}{\partial y} \right) = j\omega \epsilon (\hat{a}_x E_x + \hat{a}_y E_y + \hat{a}_z E_z)$$

For a uniform plane wave traveling in the z direction the only component of \underline{H} that contributes assuming an x component of \underline{E} , is

$$-\hat{a}_x \frac{\partial H_y}{\partial z} = \hat{a}_x j\omega \epsilon E_x \Rightarrow \frac{\partial H_y}{\partial z} = -j\omega \epsilon E_x$$

Similarly using Maxwell's equation of $\nabla \times \underline{E} = -j\omega \mu \underline{H}$

$$\hat{a}_x \left(\frac{\partial E_z}{\partial y} - \frac{\partial E_y}{\partial z} \right) + \hat{a}_y \left(\frac{\partial E_x}{\partial z} - \frac{\partial E_z}{\partial x} \right) + \hat{a}_z \left(\frac{\partial E_y}{\partial x} - \frac{\partial E_x}{\partial y} \right) = -j\omega \mu (\hat{a}_x H_x + \hat{a}_y H_y + \hat{a}_z H_z)$$

the only components that contribute to the $-z$ plane wave traveling in the x direction are

$$\hat{a}_y \frac{\partial E_x}{\partial z} = -\hat{a}_y j\omega \mu H_y \Rightarrow \frac{\partial E_x}{\partial z} = -j\omega \mu H_y \Rightarrow H_y = -\frac{1}{j\omega \mu} \frac{\partial E_x}{\partial z}$$

Substituting this equation into the one above, we can write

$$\frac{\partial H_y}{\partial z} = \frac{\partial}{\partial z} \left[-\frac{1}{j\omega \mu} \frac{\partial E_x}{\partial z} \right] = -\frac{1}{j\omega \mu} \frac{\partial^2 E_x}{\partial z^2} = -j\omega \epsilon E_x \Rightarrow \frac{\partial^2 E_x}{\partial z^2} = -\omega^2 \mu \epsilon E_x$$

Since the field is only a function of z , then

$$\frac{d^2 E_x}{dz^2} = -\omega^2 \mu \epsilon E_x = -\beta^2 E_x \Rightarrow E_x = E_0^+ e^{-j\beta z} + E_0^- e^{+j\beta z}, E_0^+ = E_0^- e^{-j\beta z}$$

Then

$$\begin{aligned} H_y^+ &= -\frac{1}{j\omega \mu} \frac{\partial E_x^+}{\partial z} = -\frac{1}{j\omega \mu} \frac{\partial}{\partial z} [E_0^+ e^{-j\beta z}] = \frac{1}{\omega \mu} \beta E_0^+ e^{-j\beta z} \\ &= \frac{\beta}{\omega \mu} E_0^+ e^{-j\beta z} = \frac{\omega \sqrt{\mu \epsilon}}{\omega \mu} E_0^+ e^{-j\beta z} = \frac{1}{\sqrt{\mu/\epsilon}} E_0^+ e^{-j\beta z} \end{aligned}$$

$$H_y^+ = \frac{1}{\sqrt{\mu/\epsilon}} E_0^+$$

$$4.2 \quad \underline{E} = \hat{a}_y (E_0^+ e^{-j\beta_0 z} + E_0^- e^{+j\beta_0 z})$$

Using Maxwell's equation of

$$\nabla \times \underline{E} = -j\omega \mu \underline{H} \Rightarrow \underline{H} = -\frac{1}{j\omega \mu} \nabla \times \underline{E} = -\frac{1}{j\omega \mu} \left[\hat{a}_x \left(-\frac{\partial E_y}{\partial z} \right) \right] = \hat{a}_x \frac{1}{j\omega \mu} \frac{\partial E_y}{\partial z}$$

$$H_x = \frac{1}{j\omega \mu} \frac{\partial E_y}{\partial z} = \frac{\beta_0}{\omega \mu} \left(-E_0^+ e^{-j\beta_0 z} + E_0^- e^{+j\beta_0 z} \right) = \frac{\omega \mu \epsilon_0}{\omega \mu} \left(-E_0^+ e^{-j\beta_0 z} + E_0^- e^{+j\beta_0 z} \right)$$

$$H_x = \sqrt{\frac{\epsilon_0}{\mu}} \left(-E_0^+ e^{-j\beta_0 z} + E_0^- e^{+j\beta_0 z} \right) = \frac{1}{\eta} \left(-E_0^+ e^{-j\beta_0 z} + E_0^- e^{+j\beta_0 z} \right)$$

or

$$\underline{H} = \hat{a}_x \frac{1}{\eta} \left(-E_0^+ e^{-j\beta_0 z} + E_0^- e^{+j\beta_0 z} \right)$$

$$4.3 \quad \underline{H} = \frac{1}{120\pi} (\hat{a}_x - 2\hat{a}_y) e^{-j\beta_0 z}$$

$$a. \quad \underline{E} = \frac{120\pi}{120\pi} (-\hat{a}_y - 2\hat{a}_x) e^{-j\beta_0 z} = -(2\hat{a}_x + \hat{a}_y) e^{-j\beta_0 z}$$

$$b. \quad \underline{E} = \operatorname{Re} [\underline{E} e^{j\omega t}] = -(2\hat{a}_x + \hat{a}_y) \cos(\omega t - \beta_0 z)$$

$$\underline{H} = \operatorname{Re} [\underline{H} e^{j\omega t}] = \frac{1}{120\pi} (\hat{a}_x - 2\hat{a}_y) \cos(\omega t - \beta_0 z)$$

$$\underline{S} = \underline{E} \times \underline{H} = -\frac{1}{120\pi} (2\hat{a}_x + \hat{a}_y) \times (\hat{a}_x - 2\hat{a}_y) \cos^2(\omega t - \beta_0 z) = \hat{a}_y \frac{5}{120\pi} \cos^2(\omega t - \beta_0 z)$$

$$= \hat{a}_y 1.326 \times 10^{-2} \cos^2(\omega t - \beta_0 z)$$

$$c. \quad S_{ave} = \frac{1}{2} \operatorname{Re} (\underline{E} \times \underline{H}^*) = \frac{1}{2} \operatorname{Re} \left[-(2\hat{a}_x + \hat{a}_y) e^{-j\beta_0 z} \times \frac{1}{120\pi} (\hat{a}_x - 2\hat{a}_y) e^{+j\beta_0 z} \right]$$

$$= \hat{a}_y \frac{5}{240\pi} = \hat{a}_y 6.63 \times 10^{-3} \text{ W/m}^2$$

$$4.4 \quad \underline{E} = (\hat{a}_x + j\hat{a}_z) e^{-j\beta_0 y} + (2\hat{a}_x - j\hat{a}_z) e^{+j\beta_0 y}$$

$$a. \quad \underline{H} = (j\hat{a}_x - \hat{a}_z) \frac{1}{\eta_0} e^{-j\beta_0 y} + (j\hat{a}_x + 2\hat{a}_z) \frac{1}{\eta_0} e^{+j\beta_0 y}$$

$$b. \quad \underline{S}^+ = \frac{1}{2} \operatorname{Re} [\underline{E}^+ \times \underline{H}^{+*}] = \frac{1}{2} \operatorname{Re} [(\hat{a}_x + j\hat{a}_z) e^{-j\beta_0 y} \times (-j\hat{a}_x - \hat{a}_z) \frac{1}{\eta_0} e^{+j\beta_0 y}] = \hat{a}_y \frac{1}{\eta_0} = \hat{a}_y 2.6525 \times 10^{-3}$$

$$c. \quad \underline{S}_w = \frac{1}{2} \operatorname{Re} [\underline{E} \times \underline{H}^*] = \frac{1}{2} \operatorname{Re} [(2\hat{a}_x - j\hat{a}_z) e^{+j\beta_0 y} \times (j\hat{a}_x + 2\hat{a}_z) \frac{1}{\eta_0} e^{-j\beta_0 y}] = -\hat{a}_y \frac{5}{2\eta_0} = -\hat{a}_y 1.326 \times 10^{-2}$$

$$\underline{S}_{av} = -\hat{a}_y 0.6625 \times 10^{-2}$$

$$4.5 \quad \underline{H} = 10^{-6} [-\hat{a}_x(2+j) + \hat{a}_y(1+jz)] e^{+j\beta y}$$

$$a. \quad \underline{E} = \eta_0 \times 10^{-6} [\hat{a}_x(1+jz) + \hat{a}_y(2+j)] e^{+j\beta y}$$

$$b. \quad S_{av} = \frac{1}{2} \operatorname{Re} [\underline{E} \times \underline{H}^*] = \frac{\eta_0}{2} \times 10^{-12} \operatorname{Re} \left[[\hat{a}_x(1+jz) + \hat{a}_y(2+j)] e^{+j\beta y} \times [-\hat{a}_x(2-j) + \hat{a}_y(1-jz)] e^{-j\beta y} \right]$$

$$= -\frac{15}{2} \eta_0 \times 10^{-12} \hat{a}_y = -\hat{a}_y 2.8275 \times 10^{-9}$$

$$4.6 \quad \underline{E} = 10^{-3} (\hat{a}_x + j\hat{a}_y) \sin(\beta_0 z)$$

a. Standing wave because of $\sin(\beta_0 z)$ wave function.

$$b. \quad \underline{E} = 10^{-3} (\hat{a}_x + j\hat{a}_y) \sin(\beta_0 z) = 10^{-3} (\hat{a}_x + j\hat{a}_y) \left(\frac{e^{j\beta_0 z} - e^{-j\beta_0 z}}{2j} \right) = \frac{10^{-3}}{2} (-j\hat{a}_x + \hat{a}_y) e^{+j\beta_0 z} - \frac{10^{-3}}{2} (-j\hat{a}_x + \hat{a}_y) e^{-j\beta_0 z}$$

$$= \underbrace{\frac{10^{-3}}{2} (-j\hat{a}_x + \hat{a}_y)}_{-z \text{ direction}} e^{+j\beta_0 z} - \underbrace{\frac{10^{-3}}{2} (-j\hat{a}_x + \hat{a}_y)}_{+z \text{ direction}} e^{-j\beta_0 z}$$

$$c. \quad \underline{H} = -\frac{1}{j\omega\mu_0} \nabla \times \underline{E} = -\frac{1}{j\omega\mu_0} \left[-\hat{a}_x \frac{\partial E_y}{\partial z} + \hat{a}_y \frac{\partial E_x}{\partial z} \right] = 2.6525 \times 10^{-6} (\hat{a}_x + j\hat{a}_y) \cos(\beta_0 z) = \frac{10^{-3}}{377} (\hat{a}_x + j\hat{a}_y) \cos(\beta_0 z)$$

$$d. \quad S_{av} = \frac{1}{2} \operatorname{Re} (\underline{E} \times \underline{H}^*) = \frac{1}{2} 2.6525 \times 10^{-9} \operatorname{Re} [(\hat{a}_x + j\hat{a}_y) \sin(\beta_0 z) \times (\hat{a}_x + j\hat{a}_y) \cos(\beta_0 z)] = 0$$

$$4.7 \quad \underline{H} = 10^{-6} (\hat{a}_y + j\hat{a}_z) \cos(\beta_0 x)$$

a. Standing wave because of the $\cos(\beta_0 x)$ function.

$$b. \quad \underline{H} = 10^{-6} (\hat{a}_y + j\hat{a}_z) \cos(\beta_0 x) = 10^{-6} (\hat{a}_y + j\hat{a}_z) \left(\frac{e^{j\beta_0 x} + e^{-j\beta_0 x}}{2} \right) = \frac{10^{-6}}{2} (\hat{a}_y + j\hat{a}_z) e^{+j\beta_0 x} + \frac{10^{-6}}{2} (\hat{a}_y + j\hat{a}_z) e^{-j\beta_0 x}$$

$$= \underbrace{\frac{10^{-6}}{2} (\hat{a}_y + j\hat{a}_z)}_{-x \text{ direction}} e^{+j\beta_0 x} + \underbrace{\frac{10^{-6}}{2} (\hat{a}_y + j\hat{a}_z)}_{+x \text{ direction}} e^{-j\beta_0 x}$$

$$c. \quad \underline{E} = \frac{1}{j\omega\epsilon} \nabla \times \underline{H} = \frac{\beta_0 \times 10^{-6}}{j\omega\epsilon} [\hat{a}_y \sin \beta_0 x - \hat{a}_z \sin \beta_0 x] = \left\{ \eta [\hat{a}_y + j\hat{a}_z] \sin \beta_0 x \right\} 10^{-6}$$

$$d. \quad S_{av} = \frac{1}{2} \operatorname{Re} (\underline{E} \times \underline{H}^*) = \frac{10^{-12}}{2} \operatorname{Re} [\eta (\hat{a}_y + j\hat{a}_z) \sin \beta_0 x \times (\hat{a}_y - j\hat{a}_z) \cos \beta_0 x] = 0$$

4.8 $\underline{E} = \hat{a}_x 4 \times 10^{-3} e^{+j\beta_0 z}$, $\beta_0 = \omega \sqrt{\mu_0 \epsilon_0} = 2\pi \times 3 \times 10^8 \frac{1}{3 \times 10^5} = 2\pi \text{ rad/m}$

a. $\underline{H} = -\hat{a}_y \frac{4 \times 10^{-3}}{\eta_0} e^{+j\beta_0 z} = -\hat{a}_y 10.61 \times 10^{-6} e^{+j\beta_0 z}$

b. $\underline{E} = \operatorname{Re}[\underline{E} e^{j\omega t}] = \hat{a}_x 4 \times 10^{-3} \cos(\omega t + \beta_0 z)$

$\underline{H} = \operatorname{Re}[\underline{H} e^{j\omega t}] = -\hat{a}_y 10.61 \times 10^{-6} \cos(\omega t + \beta_0 z)$

c. $\underline{S}_{av} = \frac{1}{2} \operatorname{Re}(\underline{E} \times \underline{H}^*) = -\hat{a}_z 21.22 \times 10^{-9}$

$\underline{S} = \underline{E} \times \underline{H} = -\hat{a}_z 42.44 \times 10^{-9} \cos^2(\omega t + \beta_0 z)$

d. $\bar{W}_e = \frac{1}{2} \epsilon_0 |\underline{E}|^2 = \frac{1}{2} (8.854 \times 10^{-12}) (16 \times 10^{-6})^2 = 35.416 \times 10^{-18} \text{ J/m}^3$

$\bar{W}_m = \frac{1}{2} \mu_0 |\underline{H}|^2 = \frac{1}{2} (4\pi \times 10^{-7}) (10.61 \times 10^{-6})^2 = 353.656 \times 10^{-18} = 35.3656 \times 10^{-18} \text{ J/m}^3$

$w_e = \frac{1}{2} \epsilon_0 \underline{E}^2 = \frac{1}{2} (8.854 \times 10^{-12}) (16 \times 10^{-6}) \omega_r^2 (\omega t + \beta_0 z) = 70.832 \times 10^{-18} \cos^2(\omega t + \beta_0 z)$

$w_m = \frac{1}{2} \mu_0 \underline{H}^2 = \frac{1}{2} (4\pi \times 10^{-7}) (10.61 \times 10^{-6})^2 = 70.731 \times 10^{-18} \cos^2(\omega t - \beta_0 z)$

4.9

a. $H_0 = \frac{E_0}{\eta_0} = \frac{1 \times 10^{-3}}{377} = 2.6525 \times 10^{-6} \text{ A/m}$

b. $S_{av} = \frac{1}{2} \operatorname{Re}[\underline{E} \times \underline{H}^*] = \hat{a}_z \frac{|\underline{E}|^2}{2\eta_0} = \hat{a}_z \frac{(10^{-3})^2}{2(377)} = \hat{a}_z 1.3263 \times 10^{-9}$

c. $\underline{E} = \operatorname{Re}[\underline{E} e^{j\omega t}] = \hat{a}_x E_0 \cos(\omega t - \beta_0 z) \quad \left\{ w_e = \frac{1}{2} \epsilon_0 \underline{E}^2 = 4.427 \times 10^{-18} \right.$

$\underline{H} = \operatorname{Re}[\underline{H} e^{j\omega t}] = \hat{a}_y H_0 \cos(\omega t - \beta_0 z) \quad \left\{ w_m = \frac{1}{2} \mu_0 \underline{H}^2 = \frac{1}{2} (4\pi \times 10^{-7}) (2.6525 \times 10^{-6})^2 = 4.4207 \times 10^{-18} \right.$

4.10 a. $\underline{E} = \hat{a}_y 10^{-3} e^{-j2\pi z} \Rightarrow \beta = \frac{2\pi}{\lambda} \Rightarrow \lambda = \frac{\lambda}{\beta} = \frac{2\pi \times 10^{-6}}{2\pi} = 10^{-6} \text{ m/sec.}$

b. $v = \frac{1}{\sqrt{\mu_0 \epsilon_r \epsilon_0}} = \frac{1}{\sqrt{\mu_0 \epsilon_r \epsilon_0}} = \frac{1}{\sqrt{\epsilon_r \sqrt{\mu_0 \epsilon_0}}} = \frac{v_0}{\sqrt{\epsilon_r}} \Rightarrow \epsilon_r = \left(\frac{v_0}{v}\right)^2 = \left(\frac{3 \times 10^8}{10^6}\right)^2 = 9$

c. $\beta = \frac{2\pi}{\lambda} \Rightarrow \lambda = \frac{2\pi}{\beta} = \frac{2\pi}{2\pi} = 1 \text{ meters}$

d. $\underline{E} = \hat{a}_y 10^{-3} e^{-j2\pi z}, \underline{H} = -\hat{a}_x [10^3 / (377)] e^{-j2\pi z} = -\hat{a}_x 7.9576 \times 10^{-6} e^{-j2\pi z}$

$S_{av} = \frac{1}{2} \operatorname{Re}(\underline{E} \times \underline{H}^*) = \frac{\hat{a}_z |\underline{E}|^2}{2} = \frac{\hat{a}_z 10^{-6}}{2 \times 125.667} = \hat{a}_z 3.910 \times 10^{-9}$

e. $\bar{w}_e = \frac{1}{2} \epsilon_0 |\underline{E}|^2 = \frac{1}{2} (8.854 \times 10^{-12}) (10^{-3})^2 = 19.9215 \times 10^{-18} \quad \left\{ \bar{w} = \bar{w}_e + w_m = 39.8251 \times 10^{-18} \right.$

$w_m = \frac{1}{2} \mu_0 |\underline{H}|^2 = \frac{1}{2} (4\pi \times 10^{-7}) (7.9576 \times 10^{-6})^2 = 19.8936 \times 10^{-18} \quad \left\{ \bar{w} = \bar{w}_e + w_m = 39.8251 \times 10^{-18} \right.$

$$4.11 \quad \underline{E} = \hat{a}_x 10^{-3} (1+j) e^{-j(2/3)\pi x}$$

a. $\beta = \frac{2\pi}{\lambda} = \frac{2\pi}{3} \Rightarrow \lambda = 3 \text{ meters}$

b. $f = \frac{v}{\lambda} = \frac{3 \times 10^8}{3} = 10^8 \text{ m/sec}$

c. $H = -\hat{a}_y \frac{10^{-3}}{377} (1+j) e^{-j(2/3)\pi x} = -\hat{a}_y 2.6525 \times 10^{-6} (1+j) e^{-j(2/3)\pi x}$

4.12

a. $v = \frac{1}{\sqrt{\mu_0 \epsilon_0}} = \frac{1}{\sqrt{\mu_0 \epsilon_0 \epsilon_r}} = \frac{3 \times 10^8}{3} = 10^8 \text{ m/sec.}$

b. $Z_0 = \sqrt{\frac{\mu_0}{\epsilon_0}} = \sqrt{\frac{\mu_0}{\epsilon_0 \epsilon_r}} = \frac{377}{3} = 125.67$

c. $\eta_0 = Z_0 = \frac{377}{3} = 125.67$

d. $\lambda = \frac{v}{f} = \frac{10^8}{10^6} = 100 \text{ meters}$

4.13 $P_r = 50 \times 10^{-3}$

a. $(S_o)_{av} = \frac{P_r}{4\pi R^2} = \frac{50 \times 10^{-3}}{4\pi (3 \times 10^3)^2} = \frac{50 \times 10^{-3}}{36\pi \times 10^6} = 0.4421 \times 10^{-9} \text{ W/m}^2$

b. $(S_o)_{av} = \frac{1}{2} \frac{|E_{peak}|^2}{\eta_0} \Rightarrow |E_{peak}| = \sqrt{2\eta_0 (S_o)_{av}} = \sqrt{2(377)(4.421 \times 10^{-9})} = 5.7736 \times 10^{-4}$

$$|E_{rms}| = \frac{1}{\sqrt{2}} |E_{peak}| = \frac{5.7736 \times 10^{-4}}{\sqrt{2}} \text{ V/m} = 4.0826 \times 10^{-4} \text{ V/m}$$

$$|H_{rms}| = \frac{|E_{rms}|}{377} = \frac{4.0826 \times 10^{-4}}{377} \text{ A/m} = 1.0829 \times 10^{-6} \text{ V/m}$$

c. $\bar{W}_e = \frac{1}{4} \epsilon_0 |E_{peak}|^2 = \frac{1}{2} \epsilon_0 |E_{rms}|^2 = \frac{1}{2} (8.854 \times 10^{-12}) |4.0826 \times 10^{-4}|^2 = 0.7379 \times 10^{-18} \text{ J/m}^3$

$$\bar{W}_m = \frac{1}{4} \mu_0 |H_{peak}|^2 = \frac{1}{2} \mu_0 |H_{rms}|^2 = \frac{1}{2} (4\pi \times 10^{-7}) |1.0829 \times 10^{-6}|^2 = 0.7369 \times 10^{-18} \text{ J/m}^3$$

$$\bar{W} = \bar{W}_e + \bar{W}_m = 1.4748 \times 10^{-18} \text{ J/m}^3$$

4.14 $\underline{E} = \hat{a}_x 10^{-4} (1+j) e^{-j\beta_0 x} \Rightarrow S_{ave} = \frac{1}{2} \operatorname{Re}(\underline{E} \times \underline{H}^*) = \hat{a}_x \frac{|\underline{E}|^2}{2\eta_0} = \hat{a}_x \frac{10^{-8}}{2 \cdot 377} = 2.6525 \times 10^{-11} \hat{a}_x$

$$P_{av} = S_{ave} \cdot \hat{a}_x A = 2.6525 \times 10^{-11} (20 \times 10^{-4}) = 53.05 \times 10^{-15} = 5.305 \times 10^{-14} \text{ W}$$

$$4.15 \quad \underline{E} = 5 \times 10^{-3} (4 \hat{a}_y + 3 \hat{a}_z) e^{j(6y - \beta z)}$$

$$\underline{E} = 5 \times 10^{-3} (5) \left(\frac{4}{5} \hat{a}_y + \frac{3}{5} \hat{a}_z \right) e^{j10(0.6y - 0.8z)} = 25 \times 10^{-3} (0.8 \hat{a}_y + 0.6 \hat{a}_z) e^{j10(0.6y - 0.8z)}$$

a. $\underline{E} = 25 \times 10^{-3} (\cos \theta \hat{a}_y + \sin \theta \hat{a}_z) e^{j\beta (x \sin \theta y - z \cos \theta)}$

$$\cos \theta = 0.8 \Rightarrow \theta = \cos^{-1}(0.8) = 36.87^\circ$$

$$\sin \theta = 0.6 \Rightarrow \theta = \sin^{-1}(0.6) = 36.87^\circ$$

b. $\underline{E} = 25 \times 10^{-3} (\cos \theta \hat{a}_y + \sin \theta \hat{a}_z) e^{j(\beta_y y - \beta_z z)}$

$$\beta_y = 10 \text{ rad/m}$$

$$\beta_y = 6 \text{ rad/m}$$

$$\beta_z = 8 \text{ rad/m}$$

c. $\lambda_y = \frac{2\pi}{\beta_y} = \frac{2\pi}{10} = 0.6283 \text{ m} ; \lambda_y = \frac{2\pi}{\beta_y} = \frac{2\pi}{6} = 1.047 \text{ m} ; \lambda_z = \frac{2\pi}{\beta_z} = \frac{2\pi}{8} = 0.7854 \text{ m}$

d. $v_{fr} = \frac{1}{\sqrt{\mu_0 \epsilon_0}} = 3 \times 10^8 \text{ m/sec.}$

$$\cos \theta = \frac{v_r}{v_z} \Rightarrow v_z = \frac{v_r}{\cos \theta} = \frac{3 \times 10^8}{0.8} = 3.75 \times 10^8 \text{ m/sec.}$$

$$\cos\left(\frac{\pi}{2} - \theta\right) = \frac{v_r}{v_y} \Rightarrow v_y = \frac{v_r}{\cos\left(\frac{\pi}{2} - \theta\right)} = \frac{v_r}{\sin \theta} = \frac{3 \times 10^8}{0.6} = 5 \times 10^8 \text{ m/sec.}$$

e. $v_{ez} = 3 \times 10^8 \text{ m/sec}$

$$v_{ez} = v \cos \theta = 3 \times 10^8 (0.8) = 2.4 \times 10^8 \text{ m/sec.}$$

$$v_{ey} = v \sin \theta = 3 \times 10^8 (0.6) = 1.8 \times 10^8 \text{ m/sec.}$$

f. $\beta_0 = \omega \sqrt{\mu_0 \epsilon_0} = 2\pi f \sqrt{\mu_0 \epsilon_0} = 10 \Rightarrow f = \frac{10}{2\pi \sqrt{\mu_0 \epsilon_0}} = \frac{15 \times 10^8}{\pi} = 4.7746 \times 10^8 = 0.47746 \times 10^9$

g. $\underline{H} = -\hat{a}_x \frac{25 \times 10^{-3}}{377} e^{j10(0.6y - 0.8z)} = -\hat{a}_x 6.33 \times 10^{-6} e^{j(6y - 8z)}$

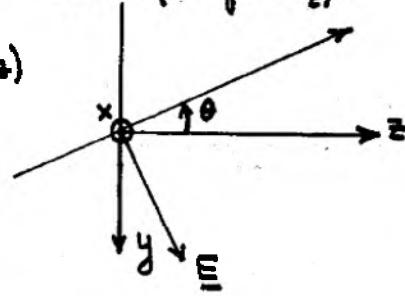
4.16 $\underline{E} = \underline{E}^+ + \underline{E}^- = E_0^+ (\hat{a}_x \cos \theta_i - \hat{a}_z \sin \theta_i) e^{-j\beta(x \sin \theta_i + z \cos \theta_i)} + E_0^- (\hat{a}_x \cos \theta_i - \hat{a}_z \sin \theta_i) e^{+j\beta(x \sin \theta_i + z \cos \theta_i)}$

$$\underline{H}^+ = \frac{1}{j\omega \mu} \nabla \times \underline{E}^+ = \frac{1}{j\omega \mu} \left[\hat{a}_x \frac{\partial E_x^+}{\partial y} + \hat{a}_y \left(\frac{\partial E_x^+}{\partial z} - \frac{\partial E_z^+}{\partial x} \right) + \hat{a}_z \frac{\partial E_z^+}{\partial y} \right] = \hat{a}_y \frac{1}{j\omega \mu} \left(\frac{\partial E_x^+}{\partial z} + \frac{\partial E_z^+}{\partial x} \right)$$

$$\frac{\partial E_x^+}{\partial z} = -j\beta \cos^2 \theta_i; E_0^+ e^{-j\beta(x \sin \theta_i + z \cos \theta_i)}$$

$$\frac{\partial E_z^+}{\partial x} = +j\beta \sin^2 \theta_i; E_0^+ e^{-j\beta(x \sin \theta_i + z \cos \theta_i)}$$

$$\underline{H}^+ = \hat{a}_y \frac{\beta E_0^+}{\omega \mu} (\cos^2 \theta_i + \sin^2 \theta_i) E_0^+ e^{-j\beta(x \sin \theta_i + z \cos \theta_i)} = \hat{a}_y \frac{E_0^+}{\eta} E_0^+ e^{-j\beta(x \sin \theta_i + z \cos \theta_i)}. \text{ Similarly for } \underline{H}^-.$$



$$4.17 \quad \underline{E} = \underline{E}^+ + \underline{E}^- = \hat{a}_y [E_0^+ e^{-j\beta(x \sin \theta_i + z \cos \theta_i)} + E_0^- e^{+j\beta(x \sin \theta_i + z \cos \theta_i)}]$$

$$\underline{H}^+ = -\frac{1}{j\omega\mu} \nabla \times \underline{E}^+ = -\frac{1}{j\omega\mu} \left[\hat{a}_x \left(-\frac{\partial E_0^+}{\partial z} \right) + \hat{a}_z \frac{\partial E_0^+}{\partial x} \right]$$

$$\frac{\partial E_0^+}{\partial z} = -j\beta E_0^+ e^{-j\beta(x \sin \theta_i + z \cos \theta_i)} \cos \theta_i$$

$$\frac{\partial E_0^+}{\partial x} = -j\beta E_0^+ e^{-j\beta(x \sin \theta_i + z \cos \theta_i)} \sin \theta_i$$

$$\underline{H}^+ = \frac{jE_0^+}{\omega\mu} \left[-\hat{a}_x \cos \theta_i + \hat{a}_z \sin \theta_i \right] e^{-j\beta(x \sin \theta_i + z \cos \theta_i)} = \frac{E_0^+}{\eta} \left(\hat{a}_x \cos \theta_i + \hat{a}_z \sin \theta_i \right) e^{j\beta(x \sin \theta_i + z \cos \theta_i)}$$

Similarly for $\underline{H}^- = \frac{E_0^-}{\eta} \left(\hat{a}_x \cos \theta_i - \hat{a}_z \sin \theta_i \right) e^{+j\beta(x \sin \theta_i + z \cos \theta_i)}$

$$4.18 \quad \underline{B}^+ \cdot \underline{r} - \omega t = \beta(x \sin \theta_i + z \cos \theta_i) - \omega t = C_0^+$$

$$\frac{\partial}{\partial t} \left[\underline{B}^+ \cdot \underline{r} - \omega t \right] = \beta \cos \theta_i \frac{\partial \underline{r}}{\partial t} - \omega(1) = 0 \Rightarrow \frac{\partial \underline{r}}{\partial t} = \underline{v}_{pz} = \frac{\omega}{\beta \cos \theta_i} = \frac{\omega}{C_0^+} = \frac{1}{\sqrt{\mu\epsilon} \cos \theta_i}$$

$$\underline{B}^+ \cdot \underline{r} - \omega t = \hat{a}_y \cdot \hat{a}_y \underline{r} - \omega t = \beta \underline{r} - \omega t = C_0^+$$

$$\frac{\partial}{\partial t} \left[\beta \underline{r} - \omega t \right] = \beta \frac{\partial \underline{r}}{\partial t} - \omega(1) = 0 \Rightarrow \frac{\partial \underline{r}}{\partial t} = \underline{v}_{pr} = \frac{\omega}{\beta} = \frac{1}{\sqrt{\mu\epsilon}}$$

$$4.19 \quad \underline{E}^+ = E_0^+ (\hat{a}_x \cos \theta_i - \hat{a}_z \sin \theta_i) e^{-j\beta(x \sin \theta_i + z \cos \theta_i)}$$

$$\underline{H}^+ = \hat{a}_y \frac{E_0^+}{\eta} e^{-j\beta(x \sin \theta_i + z \cos \theta_i)}$$

$$\underline{E}^+ = \text{Re}[\underline{E}^+ e^{j\omega t}] = E_0^+ (\hat{a}_x \cos \theta_i - \hat{a}_z \sin \theta_i) \cos[\omega t - \beta(x \sin \theta_i + z \cos \theta_i)]$$

$$\underline{H}^+ = \text{Re}[\underline{H}^+ e^{j\omega t}] = \hat{a}_y \frac{E_0^+}{\eta} \cos[\omega t - \beta(x \sin \theta_i + z \cos \theta_i)]$$

$$\underline{S}^+ = \underline{E}^+ \times \underline{H}^+ = \frac{(E_0^+)^2}{\eta} (\hat{a}_x \cos \theta_i + \hat{a}_z \sin \theta_i) \cos^2[\omega t - \beta(x \sin \theta_i + z \cos \theta_i)]$$

$$S_z^+ = \left[\frac{E_0^+}{\eta} \right]^2 \omega^2 [\omega t - \beta(x \sin \theta_i + z \cos \theta_i)] \cos \theta_i = \left(\frac{E_0^+}{\eta} \right)^2 \omega \cos \theta_i \cos^2[\omega t - \beta(x \sin \theta_i + z \cos \theta_i)]$$

$$w_e = \frac{1}{2} \epsilon |\underline{E}|^2 = \left(\frac{E_0^+}{2} \right)^2 \epsilon \cos^2[\omega t - \beta(x \sin \theta_i + z \cos \theta_i)]$$

$$w_m = \frac{1}{2} \mu \left| \underline{H}^+ \right|^2 = \frac{\mu}{2\eta^2} (E_0^+)^2 \cos^2[\omega t - \beta(x \sin \theta_i + z \cos \theta_i)] = \left(\frac{E_0^+}{2} \right)^2 \mu \cos^2[\omega t - \beta(x \sin \theta_i + z \cos \theta_i)]$$

$$V_{ea}^+ = \frac{S_z^+}{w_e + w_m} = \frac{\left[\left(\frac{E_0^+}{\eta} \right)^2 / \eta \right] \cos \theta_i}{\left(\frac{E_0^+}{2} \right)^2 \epsilon + \left(\frac{E_0^+}{2} \right)^2 \mu} = \frac{\cos \theta_i}{\eta \epsilon} = \frac{1}{\sqrt{\mu\epsilon}} \cos \theta_i \leq V \cos \theta_i \leq V$$

4.20 $f = 3 \text{ GHz}, \sigma = 5.76 \times 10^7 \text{ S/m}, \epsilon = \epsilon_0, \mu = \mu_0$

a. Since $\frac{\sigma}{\omega\epsilon_0} = \frac{5.76 \times 10^7}{2\pi(3 \times 10^9)(\frac{10^{-9}}{36\pi})} = 6(5.76) \times 10^7 \gg 1$

then $\eta = \sqrt{\frac{\omega\mu}{\sigma}}(1+j) = \sqrt{\frac{\omega\mu}{\sigma}} e^{j45^\circ} = \sqrt{\frac{2\pi(3 \times 10^9)(4\pi \times 10^{-7})}{2(5.76 \times 10^7)}}(1+j) = 2\pi\sqrt{\frac{30}{576}} \times 10^3(1+j)$
 $\approx 14.3393(1+j) \times 10^{-3} = 14.3393 \times 10^{-3}(1+j)$

b. $\delta = \frac{1}{\alpha} \approx \sqrt{\frac{2}{\omega\mu\sigma}} = \sqrt{\frac{2}{2\pi(3 \times 10^9)(4\pi \times 10^{-7})(5.76 \times 10^7)}} = \frac{1 \times 10^{-4}}{2\pi\sqrt{30(576)}} = 1.2107 \times 10^{-6} \text{ meters}$

4.21 $H = (\hat{a}_y + j2\hat{a}_z)H_0 e^{-\alpha x} e^{-j\beta x}, H_0 = 1 \text{ A/m}, \sigma = 10^{-4} \text{ S/m}, \epsilon_r = 9, f = 10^9 \text{ Hz}$

a. $E = \eta H_0 (j2\hat{a}_y - \hat{a}_z) e^{-\alpha x} e^{-j\beta x}, \frac{\sigma}{\omega\epsilon} = \frac{10^{-4}}{2\pi \times 10^9 \frac{9 \times 10^{-9}}{36\pi}} = 2 \times 10^{-4} \ll 1$

$\eta = \sqrt{\frac{\mu}{\epsilon}} = \sqrt{\frac{\mu_0}{\epsilon_r\epsilon_0}} = \frac{377}{3} = 125.67, \alpha \approx \frac{\sigma}{2} \sqrt{\frac{\mu_0}{\epsilon}} = \frac{10^{-4}}{2} \sqrt{\frac{\mu_0}{9\epsilon_0}} = \frac{377 \times 10^{-4}}{2(3)} = 62.83 \times 10^{-4} \text{ N/m}$

b. $S_{av} = \frac{1}{2} \operatorname{Re}[E \times H^*] \approx \frac{1}{2\eta} |E|^2 \approx \frac{\eta}{2} |H_0|^2 e^{-2\alpha x} = \hat{a}_x 314.167 e^{-2\alpha x} \times 10^{-12}$

c. $\beta \approx \omega\sqrt{\mu\epsilon} = \frac{2\pi \times 10^9(3) \times 10^{-6}}{3} = 2\pi \times 10 = 62.83 \text{ rad/m}$

d. $\beta = \frac{\omega}{\lambda} \Rightarrow \lambda = \frac{\omega}{\beta} \approx \frac{1}{\sqrt{\mu\epsilon}} = 10^8 \text{ m/sec}$

e. $\lambda = \frac{2\pi}{\beta} = \frac{2\pi}{20\pi} = \frac{1}{10} = 0.1 \text{ meters}$

f. $\alpha = \frac{\sigma}{2} \sqrt{\frac{\mu_0}{\epsilon}} = \frac{10^{-4}}{2} \sqrt{\frac{\mu_0}{9\epsilon_0}} = 62.83 \times 10^{-4} \text{ N/m}$

g. $\delta = \frac{1}{\alpha} \approx \frac{1}{62.83 \times 10^{-4}} = 159.1596 \text{ m}$

4.22 $\sigma = 4 \text{ S/m}, \epsilon_r = 81, \mu_r = 1, f = 10^4 \text{ Hz}$

$\frac{\sigma}{\omega\epsilon} = \frac{4}{2\pi \times 10^4 \frac{81 \times 10^{-9}}{36\pi}} = \frac{4(16)}{81} \times 10^5 = \frac{8}{9} \times 10^5 \gg 1$

a. $\alpha \approx \beta \approx \sqrt{\frac{\omega\mu}{2}} = \sqrt{\frac{2\pi \times 10^4(4\pi \times 10^{-7})}{2}} = 4\pi \times 10^{-2}\sqrt{10} = 0.39738 \Rightarrow Y = \alpha + j\beta = 0.39738(1+j)$

b. $\beta = \frac{\omega}{\lambda} \Rightarrow \lambda = \frac{\omega}{\beta} = \frac{2\pi \times 10^4}{0.39738} = 15.811 \times 10^4 = 1.5811 \times 10^5 = \sqrt{\frac{\omega\mu}{\epsilon}}$

c. $\lambda = \frac{2\pi}{\beta} \approx \frac{2\pi}{0.39738} = 15.811 \text{ meters}$

d. $\alpha \approx 0.39738 \text{ N/m}$

e. $\delta = \frac{1}{\alpha} \approx \frac{1}{0.39738} = 2.5165 \text{ meters}$

4.23 $\sigma = 10^4 \text{ S/m}$, $E_r = 4$, $\mu_r = 1$, $f = 10^6 \text{ Hz}$

$$\frac{f}{\omega} = \frac{10^6}{2\pi \times 10^6 \left(\frac{4 \times 10^{-7}}{2\pi} \right)} = \frac{1}{2} \times 10^2 = 450 > 1$$

a. $\frac{1.104 \times 10^{-2}}{3 \times 10^{-2}} = 0.368$

$$\delta = \sqrt{\frac{2}{\omega \mu \sigma}} = \sqrt{\frac{2}{2\pi \times 10^6 (4 \times 10^{-7}) 10^4}} = \frac{1}{2\pi} \times 10 = 1.5915 \text{ m}$$

b. $\alpha = \frac{1}{\delta} = \frac{\pi}{5} = 0.6283 \text{ N/m} = 0.6283 (0.68) \frac{dB}{m} = 5.4538 \frac{dB}{m}$

Total attenuation = $\alpha \delta = 1 = 8.68 \text{ dB}$

c. $2 = \frac{2\pi}{\beta} = \frac{2\pi}{\alpha} = \frac{2\pi}{0.6283} = \frac{2\pi}{\pi/5} = 10 \text{ meters, since } \beta \approx \alpha \text{ when } \frac{f}{\omega} > 1$

d. $v = \frac{\omega}{\beta} = \lambda f = 10 \times 10^6 = 10^7 \text{ m/sec.}$

e. $\eta \approx \sqrt{\frac{\omega \kappa}{2\sigma}} (1+j) = \sqrt{\frac{2\pi \times 10^6 (4 \times 10^{-7})}{2 \times 10^{-1}}} (1+j) = 2\pi (1+j) = 6.283 (1+j)$

4.24 $\underline{E} = 10^{-2} [\hat{a}_x \sqrt{2} + \hat{a}_z (1+j)] e^{j\pi/4} e^{-j\beta y} = 10^{-2} (\hat{a}_x \sqrt{2} + \hat{a}_z \sqrt{2} e^{j\pi/2}) e^{-j\beta y}$

a. Circular polarization.

b. Clockwise because the z component leads the x component.

c. $E_x = \sqrt{2} \times 10^{-2} \cos(\omega t - \beta y) \stackrel{y=0}{=} \Rightarrow E_x = \sqrt{2} \times 10^{-2} \cos(\omega t)$

$$E_z = \sqrt{2} \times 10^{-2} \cos(\omega t + \frac{\pi}{2} - \beta y) \stackrel{y=0}{=} \Rightarrow E_z = \sqrt{2} \times 10^{-2} \cos(\omega t + \frac{\pi}{2}) = -\sqrt{2} \times 10^{-2} \sin(\omega t)$$

$$|E| = \sqrt{E_x^2 + E_z^2} = \sqrt{2} \times 10^{-2} \sqrt{\cos^2 \omega t + \sin^2 \omega t} = \sqrt{2} \times 10^{-2}$$

$$\psi = \tan^{-1}\left(\frac{E_z}{E_x}\right) = \tan^{-1}\left[\frac{-\sin \omega t}{\cos \omega t}\right]$$

$$\psi = \tan^{-1}(-\tan \omega t) = -\omega t$$

4.25 $\underline{H} = \frac{10^{-3}}{120\pi} (\hat{a}_x - j\hat{a}_z) e^{+j\beta y}$

a. Circular polarization.

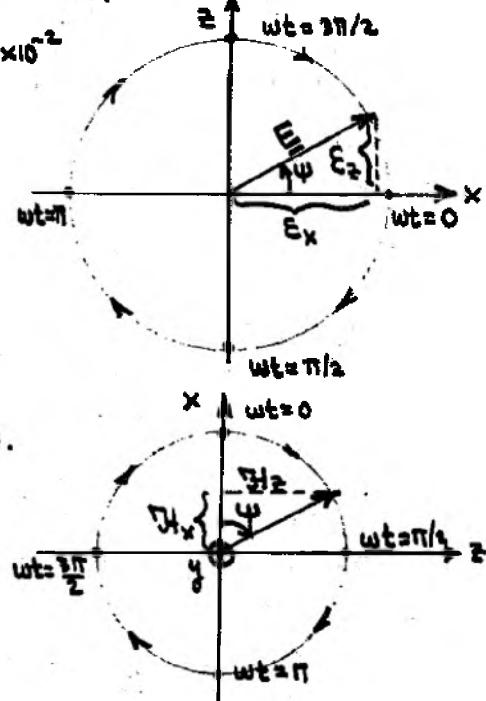
b. Clockwise because the x component leads the z .

c. $\Im H_x = \frac{10^{-3}}{120\pi} \cos(\omega t + \beta y) \stackrel{y=0}{=} \Rightarrow \Im H_x = \frac{10^{-3}}{120\pi} \cos \omega t$

$$\Im H_z = \frac{10^{-3}}{120\pi} \cos(\omega t - \frac{\pi}{2} + \beta y) \stackrel{y=0}{=} \Rightarrow \Im H_z = \frac{10^{-3}}{120\pi} \sin \omega t$$

$$|\Im H| = \sqrt{\Im H_x^2 + \Im H_z^2} = 10^{-3}/120\pi$$

$$\psi = \tan^{-1}(\Im H_z / \Im H_x) = \tan^{-1}(\tan \omega t) = \omega t$$



$$4.26 \quad H = [\hat{a}_x(1+j) + \hat{a}_z\sqrt{2} e^{j\pi/4}] \frac{E_0}{\eta_0} e^{-j\beta_0 y} = [\hat{a}_x\sqrt{2} e^{j\pi/4} + \hat{a}_z\sqrt{2} e^{j\pi/4}] \frac{E_0}{\eta_0} e^{-j\beta_0 y}$$

(a) Linear polarization because two components of the same time phase.

(b) No rotation.

$$(c) \quad E = \sqrt{2} e^{j\pi/4} (-\hat{a}_x + \hat{a}_z) E_0 e^{-j\beta_0 y} = \sqrt{2} (-\hat{a}_x + \hat{a}_z) E_0 e^{j(\frac{\pi}{4} - \beta_0 y)}$$

4.27 Let $E = \hat{a}_x E_0 e^{-j\beta z}$ which is linearly polarized. Then

$$\begin{aligned} E &= \hat{a}_x E_0 e^{-j\beta z} = \left[\hat{a}_x \frac{E_0}{2} e^{-j\beta z} + j \hat{a}_y \frac{E_0}{2} e^{-j\beta z} \right] + \left[\hat{a}_x \frac{E_0}{2} e^{-j\beta z} - j \hat{a}_y \frac{E_0}{2} e^{-j\beta z} \right] \\ &= \underbrace{\frac{E_0}{2} (\hat{a}_x + j \hat{a}_y) e^{-j\beta z}}_{\text{CIRCULAR: CCW}} + \underbrace{\frac{E_0}{2} (\hat{a}_x - j \hat{a}_y) e^{-j\beta z}}_{\text{CIRCULAR: CW}} \end{aligned}$$

The same can be shown for any other linearly polarized field.

4.28

$$E = (\hat{a}_x + j 2 \hat{a}_y) e^{-j 600\pi z} = (\hat{a}_x + j 2 \hat{a}_y) e^{-j \beta z}$$

$$(a) \quad \beta = \frac{2\pi}{\lambda} = 600\pi \Rightarrow \lambda = \frac{2\pi}{600\pi} = \frac{1}{300} = 3.333 \times 10^{-3} \text{ meters}$$

$$(b) \quad v = \lambda f = \frac{1}{300} (10 \times 10^9) = 3.333 \times 10^7 \text{ m/sec.}$$

$$(c) \quad \beta = \omega \sqrt{\mu_r \epsilon_r} = 2\pi f \sqrt{\mu_0 \epsilon_r \epsilon_0} = 2\pi f \sqrt{\mu_0 \epsilon_0 \sqrt{\epsilon_r}} = 2\pi f \sqrt{\epsilon_r} = \beta$$

$$\sqrt{\epsilon_r} = \frac{\beta v_0}{2\pi f} = \frac{600\pi (3 \times 10^8)}{2\pi (10 \times 10^9)} = 9$$

$$\epsilon_r = (9)^2 = 81$$

$$(d) \quad n = \sqrt{\frac{\mu_r}{\epsilon_r}} = \sqrt{\frac{\mu_0}{\epsilon_r \epsilon_0}} = \frac{1}{\sqrt{\epsilon_r}} \sqrt{\frac{\mu_0}{\epsilon_0}} = \frac{377}{\sqrt{81}} = \frac{377}{9} = 41.89$$

(e) For a uniform plane wave

$$Z_{\text{in}} = n \sqrt{\frac{\mu_r}{\epsilon_r}} = 41.89$$

cont'd

4.28 cont'd

$$(f) \underline{H} = \frac{1}{\eta} (\hat{a}_y - j 2 \hat{a}_x) e^{-j \frac{\beta}{600\pi} z} = \frac{1}{41.89} (-j 2 \hat{a}_x + \hat{a}_y) e^{-j \frac{\beta}{600\pi} z}$$

$$\underline{H} = (-j 47.74 \hat{a}_x + 23.87 \hat{a}_y) e^{-j \frac{\beta}{600\pi} z} \quad (\text{Using Right Rule between } \underline{E} \text{ and } \underline{H} \text{ fields})$$

Alternate: You can also use Maxwell's equation to find the \underline{H} -field.

$$\underline{H} = +j \frac{1}{\omega \mu} \nabla \times \underline{E} = j \frac{1}{\omega \mu} \begin{vmatrix} \hat{a}_x & \hat{a}_y & \hat{a}_z \\ \frac{\partial}{\partial x} & \frac{\partial}{\partial y} & \frac{\partial}{\partial z} \\ E_x & E_y & 0 \end{vmatrix} = j \frac{1}{\omega \mu} \left[\hat{a}_x \left(-\frac{\partial E_y}{\partial z} \right) + \hat{a}_y \left(\frac{\partial E_x}{\partial z} \right) + \hat{a}_z (0) \right]$$

$$= j \frac{1}{\omega \mu} \left[\hat{a}_x \left(-\frac{\partial E_y}{\partial z} \right) + \hat{a}_y \left(\frac{\partial E_x}{\partial z} \right) \right] = j \frac{1}{\omega \mu} \left[-\hat{a}_x (2j) (-j \frac{\beta}{600\pi}) + \hat{a}_y (-j \frac{\beta}{600\pi}) \right] e^{-j \frac{\beta}{600\pi} z}$$

$$\underline{H} = j \frac{(-j\beta)}{\omega \mu} \left[-\hat{a}_x (2j) + \hat{a}_y \right] e^{-j \frac{\beta}{600\pi} z} = \sqrt{\frac{\epsilon}{\mu}} (-j 2 \hat{a}_x + \hat{a}_y) e^{-j \frac{\beta}{600\pi} z}$$

$$\underline{H} = \frac{1}{\eta} (-j 2 \hat{a}_x + \hat{a}_y) e^{j \frac{\beta}{600\pi} z} = (-j 47.74 \hat{a}_x + 23.87 \hat{a}_y) e^{-j \frac{\beta}{600\pi} z}$$

which is the same as above (and as it should be).

4.29

$$\underline{E} = \hat{a}_y e^{-j(\beta_0 x - \pi/4)} + \hat{a}_z e^{-j(\beta_0 x - \pi/2)}$$

$$\underline{E} = \hat{a}_y e^{+j\pi/4} e^{-j\beta_0 x} + \hat{a}_z e^{+j\pi/2} e^{-j\beta_0 x} = (\hat{a}_y e^{+j\pi/4} + \hat{a}_z e^{+j\pi/2}) e^{-j\beta_0 x}$$

(a) Wave is traveling in the $+x$ direction



(b) Polarization is elliptical because:

- 2 orthogonal components not of the same magnitude
- Phase difference is 45° .

(c) The rotation is CCW(LH) because:

- z component is leading the y component in phase.
- Rotate the leading component (z component) toward the lagging component (y component) thru smaller angular rotation; thus CCW(LH).

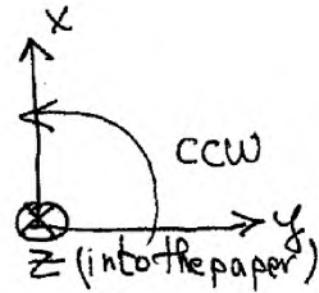
4.30

$$\underline{E} = \hat{a}_x 2 e^{-j(\beta_0 z - \frac{\pi}{4})} + \hat{a}_y e^{-j(\beta_0 z - \frac{3\pi}{4})}$$

$$\underline{E} = \hat{a}_x 2 e^{+j\frac{\pi}{4}} e^{-j\beta_0 z} + \hat{a}_y e^{+j\frac{3\pi}{4}} e^{-j\beta_0 z}$$

$$|E_{x0}| = 2, |E_{y0}| = 1, \phi_x = \frac{\pi}{4}, \phi_y = \frac{3\pi}{4}$$

Wave travels in +z direction



Using (f-50a) as a reference

$$\begin{aligned}\underline{E} &= \hat{a}_x E_x^+ e^{j(wt-\beta z)} + \hat{a}_y E_y^+ e^{j(wt-\beta z)} \\ \underline{E} &= [\hat{a}_x |E_{x0}| e^{j\phi_x - j\beta z} + \hat{a}_y |E_{y0}| e^{j\phi_y - j\beta z}] e^{jwt} \\ \underline{E} &= [\hat{a}_x |E_{x0}| e^{j\phi_x} + \hat{a}_y |E_{y0}| e^{j\phi_y}] e^{j\beta z}\end{aligned}$$

And using (4-58a)

$$\left\{ \begin{array}{l} \gamma = \tan^{-1} \left(\frac{|E_{y0}|}{|E_{x0}|} \right) = \tan^{-1} \left(\frac{1}{2} \right) = 26.565^\circ \\ \delta = \phi_y - \phi_x = \frac{3\pi}{4} - \frac{\pi}{4} = \frac{\pi}{2} = 90^\circ \end{array} \right\} \begin{array}{l} \gamma = 26.565^\circ \\ \delta = 90^\circ \end{array}$$

$$2\varepsilon = \sin^{-1} \left[\sin(2\gamma) \sin(\delta) \right] = \sin^{-1} \left[\sin(2\gamma)(1) \right] = 2\gamma$$

$$\varepsilon = \gamma = 26.565^\circ$$

$$2T = \tan^{-1} \left[\tan(2\gamma) \cos \delta \right] = \tan^{-1}(0) = 0^\circ \quad \left. \begin{array}{l} \varepsilon_r = 26.565^\circ \\ T = 0^\circ \end{array} \right\}$$

• $2\gamma = 53.13^\circ, \delta = 90^\circ \Rightarrow$ elliptical (along principal xy-plane)

• $2\varepsilon = 2\gamma = 53.13^\circ, T = 0^\circ \Rightarrow$ elliptical (along principal xy-plane)

• CCW (LH) because upper hemisphere of Poincaré sphere.

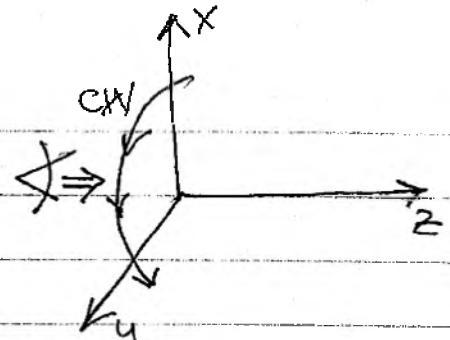
$$\cdot AR = \cot(\varepsilon) = \frac{1}{\tan \varepsilon} = \frac{1}{0.5} = 2$$

4.31

$$E(z) = (j2\hat{a}_x + 5\hat{a}_y)e^{-j\beta z}$$

(a) Elliptical because

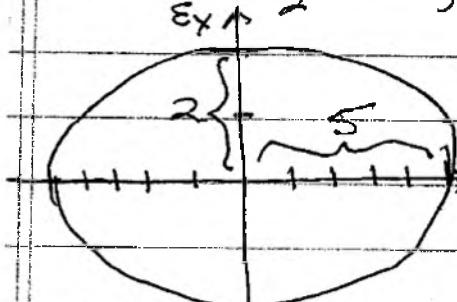
- 2 components, not of same magnitude and 90° time-phase difference



(b) CW. x-component leads y-component by $+90^\circ$

Rotate x-component toward y-component.

(c) $AR = -\frac{5}{2} = -2.5$, Since it is elliptical with 90° phase



difference between the 2 components, the major and minor axes of the ellipse align with the principal axes.

$$(d) \gamma = \tan^{-1}\left(\frac{E_{y0}}{E_{x0}}\right) = \tan^{-1}\left(\frac{5}{2}\right) = 68.2^\circ \quad (\text{or}) \quad \gamma = \tan^{-1}\left(\frac{E_{x0}}{E_{y0}}\right) = \tan^{-1}\left(\frac{2}{5}\right) = 21.8^\circ$$

$$\delta = \phi_y - \phi_x = 0 - 90^\circ = -90^\circ$$

Option I

$$\chi = 68.2^\circ, \delta = -90^\circ$$

$$2\varepsilon = \sin^{-1} [\sin(2\chi) \sin \delta] = \sin^{-1} [\sin(2\chi)(-1)] \\ = \sin^{-1} [-\sin 2\chi] = -\sin^{-1} (\sin 2\chi)$$

$$2\varepsilon = -136.4 \Rightarrow \varepsilon = -68.2^\circ$$

$$2\tau = \tan^{-1} [\tan 2\chi \cos \delta] = \tan^{-1} (\tan 2\chi \cdot 0)$$

$$2\tau = \tan^{-1}(0) = 0 \Rightarrow \tau = 0$$

Option II

$$\chi = 21.8^\circ, \delta = -90^\circ$$

$$2\varepsilon = \sin^{-1} [\sin(2\chi) \sin \delta] = \sin^{-1} [\sin 2\chi(-1)] \\ = \sin^{-1} (-\sin 2\chi) = -\sin^{-1} (\sin 2\chi)$$

$$2\varepsilon = -2\chi = -43.6 \Rightarrow \varepsilon = -21.8^\circ$$

$$2\tau = \tan^{-1} [\tan 2\chi \cos \delta] = \tan^{-1} (\tan 2\chi \cdot 0)$$

$$2\tau = \tan^{-1}(0) = 0 \Rightarrow \tau = 0$$

$\chi = 68.2^\circ, \delta = -90^\circ$

$\varepsilon = -68.2^\circ, \tau = 0^\circ$

$AR = \cot(\varepsilon) = \cot(-68.2)$

$AR = -0.4$

Cont'd.

$\chi = 21.8^\circ, \delta = -90^\circ$

$\varepsilon = -21.8^\circ, \tau = 0^\circ$

$AR = \cot(\varepsilon) = \cot(-21.8)$

$AR = -2.5$

431 cont'd

Option III

$$\gamma = 68.2^\circ, \delta = -90^\circ$$

$$2\epsilon = \sin^{-1} \left[\frac{\sin(2\gamma) \sin(\delta)}{1} \right] = \sin^{-1} \left(-\sin 2\gamma \right) = \begin{cases} \sin^{-1} \left[\frac{-\sin(2\gamma)}{1} \right] \\ \sin^{-1} \left[\frac{\sin(2\gamma)}{-1} \right] \end{cases}$$

Taking the first one:

$$2\epsilon = \sin^{-1} \left[\frac{-\sin(2\gamma)}{1} \right] = \sin^{-1} \left[\frac{-0.690}{1} \right] = -43.6^\circ \text{ or } 223.6^\circ$$

$$\epsilon = -21.8^\circ \text{ or } 111.8^\circ$$

$$\text{Since } -45^\circ \leq \epsilon \leq +45^\circ \Rightarrow \epsilon = -21.8^\circ$$

$$2\tau = \tan^{-1} \left[\tan(2\gamma) \cot \delta \right] = \tan^{-1}(0) = 0^\circ \text{ or } 180^\circ$$

$$\tau = -90^\circ \Rightarrow \tau = 0^\circ \text{ or } 90^\circ$$

$$\gamma = 68.2^\circ, \delta = -90^\circ, \epsilon = -21.8^\circ, \tau = 0^\circ$$

$$\gamma = 68.2^\circ, \delta = -90^\circ, \epsilon = 21.8^\circ, \tau = 90^\circ$$

Taking the second one:

This is not valid because

$$2\epsilon = \sin^{-1} \left[\frac{\sin(2\gamma)}{-1} \right]$$

it indicates a negative hypotenuse!

So, In Summary, the options are:

Options

1 st	X	$\gamma = 68.2^\circ, \delta = -90^\circ, \epsilon = 68.2^\circ, \tau = 0^\circ \Rightarrow AR = -0.4$
2 nd	X	$\gamma = 21.8^\circ, \delta = -90^\circ, \epsilon = -21.8^\circ, \tau = 0^\circ \Rightarrow AR = -2.5$
3 rd	X	$\gamma = 68.2^\circ, \delta = -90^\circ, \epsilon = -21.8^\circ, \tau = 0^\circ \Rightarrow AR = -2.5$
4 th	✓	$\gamma = 68.2^\circ, \delta = -90^\circ, \epsilon = 21.8^\circ, \tau = 90^\circ \Rightarrow AR = -2.5$

$$(e) AR = \cot(\epsilon) = \cot(-21.8^\circ) = -2.5 \quad AR = -2.5$$

Using Figure 4-19, τ is the tilt angle from the x-component toward the major axis. For our expression, $E = (j2\hat{x} + 5\hat{y})e^{jB_2}$, the y-component represents the major axis. Therefore, according to Figure 4-19, $\tau = 90^\circ$. Therefore the 4th option is the correct one since it gives the correct AR and tilt angle τ .

$$4.32 \quad \underline{H} = j(\hat{a}_y - j\hat{a}_z) \frac{E_0}{\eta_0} e^{+j\beta_0 x}$$

- a. Circular polarization because of two equal components and 90° time phase difference.
- b. CCW because the y component leads the z component.
- c. $\underline{E} = E_0(-\hat{a}_y + j\hat{a}_z) e^{+j\beta_0 x}$

$$\underline{S}_{av} = \frac{1}{2} \operatorname{Re}(\underline{E} \times \underline{H}^*) = -\hat{a}_x \frac{1}{2\eta_0} |\underline{E}|^2 = -\hat{a}_x \frac{2|E_0|^2}{2(377)} = -\hat{a}_x 2.6525 \times 10^{-3}$$

d. $\gamma = \tan^{-1}\left[\frac{1}{1}\right] = \tan^{-1}(1) = \pi/4 \Rightarrow 2\gamma = 90^\circ \quad \left. \begin{array}{l} \\ \end{array} \right\} \text{North pole}$
 $\delta = \phi_y - \phi_z = -180^\circ - 90^\circ = -270^\circ = +90^\circ$

$$4.33 \quad \underline{E} = 2 \times 10^{-3} (\hat{a}_x + \hat{a}_y) e^{-j2z}$$

- a. Linear polarization because of 2 components with no time phase difference.
- b. $\gamma = \tan^{-1}\left(\frac{1}{1}\right) = \frac{\pi}{4}, \delta = \phi_y - \phi_x = 0^\circ$
 $2\epsilon = \sin^{-1}[\sin(2\gamma)\sin\delta] = \sin^{-1}(0) = 0^\circ$
 $2\tau = \tan^{-1}[\tan(2\gamma)\cos\delta] = \tan^{-1}[\cos(0)] = 90^\circ \Rightarrow \tau = 45^\circ$
- c. The polarization point is located on the equator at the y axis.
 It represents linear polarization at a tilt angle of 45° .

$$4.34 \quad \underline{E} = E_0(\hat{a}_x - j2\hat{a}_y) e^{-j\beta_0 z}, AR = -2$$

a. $\delta = \phi_y - \phi_x = -\frac{\pi}{2} - 0 = -90^\circ$

$$\epsilon = \cot^{-1}(AR) = \cot^{-1}(-2) = -26.56^\circ$$

$$\gamma = \frac{1}{2} \sin^{-1} \left[\frac{\sin(2\epsilon)}{\sin\delta} \right] = \frac{1}{2} \sin^{-1} \left[\frac{\sin(-53.13^\circ)}{\sin(-90^\circ)} \right] = 26.56^\circ$$

b. $\tau = \frac{1}{2} \tan^{-1} [\tan(2\gamma) \cos\delta] = \frac{1}{2} [\tan(53.13^\circ) \cos(-90^\circ)] = 0^\circ$

Elliptical polarization, CW with an axial ratio of -2.

c. $2\gamma = 53.13^\circ$

Angle between $P_w P_a = 90^\circ - 53.13^\circ = 36.8699^\circ$

$$V = \cos\left(\frac{P_w P_a}{2}\right) = \cos\left(\frac{36.8699}{2}\right) = 0.9487 = 20 \log_{10}(0.9487) = -0.4575 \text{ dB}$$

$$4.35 \quad AR = \infty, 2\gamma = 109.47^\circ \Rightarrow \gamma = 54.735^\circ$$

a. $\gamma = \tan^{-1}\left(\frac{E_{y0}}{E_{x0}}\right) = 54.735^\circ \Rightarrow \frac{E_{y0}}{E_{x0}} = \tan(54.735^\circ) = \sqrt{2} \Rightarrow E_{y0} = \sqrt{2} E_{x0}$

E_y is more dominant ($E_y = \sqrt{2} E_x$)

b. $\epsilon = \cot^{-1}(\infty) = 0^\circ$

$$\cos(2\gamma) = \cos(2\epsilon) \cos(2\tau) \Big|_{\epsilon=0} = \cos(2\tau) \Rightarrow 2\tau = 2\gamma \Rightarrow \tau = 54.735^\circ \quad \left. \begin{array}{l} \gamma = 54.735^\circ \\ \epsilon = 0^\circ \\ \tau = 54.735^\circ \end{array} \right\} \tau = 54.735^\circ$$

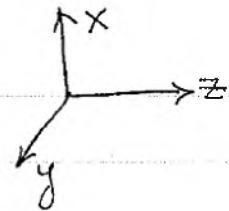
$$\tan\delta = \frac{\tan(2\epsilon)}{\sin(2\tau)} \Big|_{\epsilon=0, \tau=54.735^\circ} = \frac{0}{0.9428} = 0 \Rightarrow \delta = 0^\circ \quad \left. \begin{array}{l} \gamma = 54.735^\circ \\ \epsilon = 0^\circ \\ \tau = 54.735^\circ \end{array} \right\} \delta = 0^\circ$$

c. Linear because $\delta = 0^\circ$; also because $AR = \infty$

$$4.36 \quad E_W = (\hat{a}_x + j\hat{a}_y) e^{-j\beta z} E_0$$

$$(a) \quad E_{aa} = (\hat{a}_x + j\hat{a}_y) e^{+j\beta z} E_0$$

$$(b) \quad E_{ab} = (\hat{a}_x - j\hat{a}_y) e^{+j\beta z} E_0$$



1. Wave: Circular, CCW, AR=1 (North pole: $\delta=90^\circ, 2\gamma=90^\circ$)

2. Antenna(aa): Circular, CW, AR=-1 (South pole: $\delta=-90^\circ, 2\gamma=90^\circ$)

3. Antenna(ab): Circular, CCW, AR=+1 (North pole: $\delta=90^\circ, 2\gamma=90^\circ$)

4. w:aa

$$V = \cos\left(\frac{P_w P_a}{2}\right) = \cos\left(\frac{180^\circ}{2}\right) = \cos(90^\circ) = 0 = -\infty \text{ dB}$$

5. w:ab

$$V = \cos\left(\frac{P_w P_a}{2}\right) = \cos\left(\frac{0^\circ}{2}\right) = \cos(0^\circ) = 1 = 0 \text{ dB}$$

$$4.37 \quad E_{x0}^+ = E_{y0}^+, \quad AR = \infty$$

$$a. \quad \gamma = \tan^{-1}\left(\frac{E_{y0}^+}{E_{x0}^+}\right) = 45^\circ$$

$$\epsilon = \cot^{-1}(-\infty) = 0^\circ$$

$$\cos(2\gamma) = \cos(2\epsilon) \cos(2\pi) \Big|_{\epsilon=0} = \cos(2\pi) \Rightarrow 2\gamma = 2\pi \Rightarrow \gamma = \epsilon = 45^\circ$$

$$\tan \delta = \frac{\tan(2\epsilon)}{\sin(2\pi)} = 0 \Rightarrow \delta = 0^\circ$$

$\epsilon = 0^\circ; \pi = 45^\circ$

$$b. \quad V = \cos\left[\frac{2\gamma - 2\delta_a}{2}\right] = \cos(\gamma - \delta_a) = \cos(54.735^\circ - 45^\circ) = \cos(9.735^\circ) = 0.9856$$

$$= 20 \log_{10}(0.9856) = -0.1259 \text{ dB}$$

CHAPTER 5

5.1 $\underline{E}^i = \hat{a}_y 2 \times 10^{-3} e^{-j\beta z}$

a. $\underline{H}^i = -\hat{a}_x \frac{2 \times 10^{-3}}{\eta} e^{-j\beta z} = -\hat{a}_x \frac{2 \times 10^{-3}}{376.73/2} e^{-j\beta z} = -\hat{a}_x \frac{2 \times 10^{-3}}{188.37} e^{-j\beta z} = -\hat{a}_x 1.06 \times 10^{-5} e^{-j\beta z}$

b. $\eta_1 = \eta_0 = \sqrt{\frac{\mu_0}{\epsilon_0}} = 376.73 \quad , \quad \eta = \sqrt{\frac{\mu_0}{\epsilon}} = \sqrt{\frac{\mu_0}{4\epsilon_0}} = \frac{376.73}{2} = 188.37$

$$\Gamma^b = \frac{\eta_0 - \eta}{\eta_0 + \eta} = \frac{\sqrt{\frac{\mu_0}{\epsilon_0}} - \frac{1}{2}\sqrt{\frac{\mu_0}{\epsilon_0}}}{\sqrt{\frac{\mu_0}{\epsilon_0}} + \frac{1}{2}\sqrt{\frac{\mu_0}{\epsilon_0}}} = \frac{2-1}{2+1} = \frac{1}{3}$$

$$T^b = \frac{2\eta_0}{\eta_0 + \eta} = \frac{2\sqrt{\frac{\mu_0}{\epsilon_0}}}{\sqrt{\frac{\mu_0}{\epsilon_0}} + \frac{1}{2}\sqrt{\frac{\mu_0}{\epsilon_0}}} = \frac{2}{3} = \frac{4}{3} = 1.333$$

c. $\underline{E}^r = \hat{a}_y \Gamma^b \underline{E}_0 e^{+j\beta z} = \hat{a}_y \frac{2 \times 10^{-3}}{3} e^{+j\beta z} = \hat{a}_y 0.667 \times 10^{-3} e^{+j\beta z}$

$$\underline{H}^r = \hat{a}_x \frac{\Gamma^b \underline{E}_0}{\eta} e^{+j\beta z} = \hat{a}_x \frac{2 \times 10^{-3}}{3(188.37)} e^{+j\beta z} = \hat{a}_x 3.54 \times 10^{-6} e^{+j\beta z}$$

$$\underline{E}^t = \hat{a}_y T^b \underline{E}_0 e^{-j\beta_0 z} = \hat{a}_y \frac{4}{3}(2 \times 10^{-3}) e^{-j\beta_0 z} = \hat{a}_y 2.667 \times 10^{-3} e^{-j\beta_0 z}$$

$$\underline{H}^t = -\hat{a}_x T^b \underline{E}_0 e^{-j\beta_0 z} = -\hat{a}_x \frac{2.667 \times 10^{-3}}{376.73} e^{-j\beta_0 z} = -\hat{a}_x 7.08 \times 10^{-6} e^{-j\beta_0 z}$$

d. $S_{av}^i = \frac{1}{2} \operatorname{Re}(\underline{E}^i \times \underline{H}^{i*}) = \hat{a}_z \frac{|\underline{E}^i|^2}{2\eta} = \hat{a}_z \frac{|2 \times 10^{-3}|^2}{2(188.37)} = \hat{a}_z 1.06 \times 10^{-8} \text{ W/m}^2$

$$S_{av}^r = \frac{1}{2} \operatorname{Re}(\underline{E}^r \times \underline{H}^{r*}) = -\hat{a}_z |\Gamma^b|^2 S_{av}^i = -\hat{a}_z \left(\frac{1}{3}\right)^2 (1.06 \times 10^{-8}) = -\hat{a}_z 1.18 \times 10^{-9} \text{ W/m}^2$$

$$S_{av}^t = \frac{1}{2} \operatorname{Re}(\underline{E}^t \times \underline{H}^{t*}) = \hat{a}_z (S^i - S_{av}^r) = \hat{a}_z (1 - |\Gamma^b|^2) S_{av}^i = \hat{a}_z |\Gamma^b|^2 \eta \frac{\underline{S}_{av}^i}{\eta_0}$$

$$= \hat{a}_z 9.44 \times 10^{-9} \text{ W/m}^2$$

5.2 $\Gamma^b = \frac{\eta_1 - \eta_0}{\eta_1 + \eta_0} = \frac{\sqrt{\frac{\mu_0}{\epsilon_0}} - \sqrt{\frac{\mu_0}{\epsilon_0}}}{\sqrt{\frac{\mu_0}{\epsilon_0}} + \sqrt{\frac{\mu_0}{\epsilon_0}}} = \frac{1-1}{1+1} = \frac{0}{2} = 0.0 = -0.8$

$$S_{av}^{\text{ref}} = |\Gamma^b|^2 S_{av}^i$$

$$\frac{S_{av}^{\text{ref}}}{S_{av}^i} = |\Gamma^b|^2 = |-0.8|^2 = 0.64 \text{ or } 64\%$$

$$S_{av}^t = (1 - |\Gamma^b|^2) S_{av}^i$$

$$\frac{S_{av}^t}{S_{av}^i} = (1 - |\Gamma^b|^2) = 1 - 0.64 = 0.36 \text{ or } 36\%$$

$$5.3 \text{ a. } \Gamma^b = \frac{\eta_2 - \eta_1}{\eta_2 + \eta_1} = \frac{\frac{1}{3}\sqrt{k/\epsilon_0} - \frac{1}{2}\sqrt{k/\epsilon_0}}{\frac{1}{3}\sqrt{k/\epsilon_0} + \frac{1}{2}\sqrt{k/\epsilon_0}} = \frac{\frac{1}{3} - \frac{1}{2}}{\frac{1}{3} + \frac{1}{2}} = \frac{2-3}{2+3} = -\frac{1}{5} = -0.2$$

$$T^b = \frac{2\eta_2}{\eta_2 + \eta_1} = \frac{\frac{2}{3}\sqrt{k/\epsilon_0}}{\frac{1}{3}\sqrt{k/\epsilon_0} + \frac{1}{2}\sqrt{k/\epsilon_0}} = \frac{\frac{2}{3}}{\frac{1}{3} + \frac{1}{2}} = \frac{\frac{4}{6}}{\frac{5}{6}} = \frac{4}{5} = 0.8$$

$$\text{b. } \frac{S_{av}^{ref}}{S_{av}^t} = |\Gamma^b|^2 = |-0.2|^2 = 0.04 \text{ or } 4\%$$

$$\frac{S_{av}^t}{S_{av}^{ref}} = (1 - |\Gamma^b|^2) = (1 - |0.2|^2) = 0.96 \text{ or } 96\%$$

$$5.4 \quad f = 3 \text{ GHz}, \eta_1 = \eta_0, \eta_2 = \frac{1}{2}\eta_0$$

$$\text{a. } \Gamma^b = \frac{\eta_2 - \eta_1}{\eta_2 + \eta_1} = \frac{\frac{1}{2}\eta_0 - \eta_0}{\frac{1}{2}\eta_0 + \eta_0} = \frac{\frac{1}{2} - 1}{\frac{1}{2} + 1} = \frac{-\frac{1}{2}}{\frac{3}{2}} = -\frac{1}{3} = -0.3333$$

$$\text{b. } \text{SWR} = \frac{1 + |\Gamma^b|}{1 - |\Gamma^b|} = \frac{1 + \frac{1}{3}}{1 - \frac{1}{3}} = \frac{\frac{4}{3}}{\frac{2}{3}} = 2$$

$$\text{c. } E^{\text{total}} = E_0 e^{-j\beta_0 z} + \Gamma^b E_0 e^{+j\beta_0 z} = E_0 e^{-j\beta_0 z} [1 + \Gamma^b e^{-j2\beta_0 z}] = 2 \times 10^{-3} e^{-j\beta_0 z} [1 + \Gamma^b e^{-j2\beta_0 z}]$$

$$|E^{\text{total}}| = 2 \times 10^{-3} |1 - \frac{1}{3} e^{-j2\beta_0 z}|$$

$$|E^{\text{total}}|_{\max} = 2 \times 10^{-3} |1 + \frac{1}{3}| = 2 \times 10^{-3} \left(\frac{4}{3}\right) = \frac{8 \times 10^{-3}}{3} = 2.667 \times 10^{-3} \text{ when } 2\beta_0 z_{\max} = \pi, 3\pi, 5\pi, \dots$$

$$z_{\max} = \frac{m\pi}{2\beta_0} = \frac{m\pi}{2\left(\frac{2\pi}{\lambda_0}\right)} = \frac{m}{2} \lambda_0 = \frac{m}{2} \left(\frac{3 \times 10^8}{3 \times 10^9}\right) = m \left(\frac{0.1}{2}\right) = 0.025m, m=1, 3, 5, \dots$$

$$z_{\max} = 0.025m \text{ meters where } m=1, 3, 5, \dots$$

$$|E^{\text{total}}|_{\min} = 2 \times 10^{-3} |1 - \frac{1}{3}| = 2 \times 10^{-3} \left(\frac{2}{3}\right) = \frac{4 \times 10^{-3}}{3} = 1.333 \times 10^{-3} \text{ when } 2\beta_0 z_{\min} = 0, 2\pi, 4\pi, \dots$$

$$z_{\min} = \frac{2n\pi}{2\beta_0} = \frac{2n\pi}{2\left(\frac{2\pi}{\lambda_0}\right)} = \frac{n}{2} \lambda_0 = \frac{n}{2} \left(\frac{3 \times 10^8}{3 \times 10^9}\right) = n \left(\frac{0.1}{2}\right) = 0.05n, n=0, 1, 2, \dots$$

$$z_{\min} = 0.05n \text{ meters where } n=0, 1, 2, \dots$$

$$\text{d. } |E^{\text{total}}|_{\max} = 2 \times 10^{-3} |1 + \frac{1}{3}| = 2 \times 10^{-3} \left(\frac{4}{3}\right) = 2.667 \times 10^{-3} \text{ V/m}$$

$$|E^{\text{total}}|_{\min} = 2 \times 10^{-3} |1 - \frac{1}{3}| = 2 \times 10^{-3} \left(\frac{2}{3}\right) = 1.333 \times 10^{-3} \text{ V/m}$$

5.5

$$(a) \Gamma^b = \frac{\eta - \eta_0}{\eta + \eta_0} = \frac{\sqrt{\frac{\eta_0}{4\varepsilon_0}} - \sqrt{\frac{\eta_0}{\varepsilon_0}}}{\sqrt{\frac{\eta_0}{4\varepsilon_0}} + \sqrt{\frac{\eta_0}{\varepsilon_0}}} = \frac{\frac{1}{2} - 1}{\frac{3}{2} + 1} = \frac{-\frac{1}{2}}{\frac{5}{2}} = -\frac{1}{5}$$

$$\begin{array}{l} \varepsilon = 4\varepsilon_0 \\ \mu_2 = \mu_0 \\ \sigma_2 = 0 \end{array}$$

$$|E(\text{standing wave})|_{\max} = |1 + \Gamma^b e^{j2\beta_0 l}|_{z=-l} = \left|1 - \frac{1}{3} e^{-j2\beta_0 l}\right|_{\max}$$

$$|E(\text{sw})|_{\max} = \frac{4}{3} \quad \text{when } -2\beta_0 l = m\pi, m = \pi, 3\pi, 5\pi, \dots$$

$$(b) 2\beta_0 l = \pi \Rightarrow l = \frac{\pi}{2\beta_0} = \frac{\pi}{2(2\pi/\lambda_0)} = \frac{\lambda_0}{4} = \lambda_0/4$$

$$l = \lambda/4$$

$$(c) |E(\text{sw})|_{\min} = \left|1 + \Gamma^b e^{j2\beta_0 l}\right|_{z=l} = \left|1 - \frac{1}{3} e^{-j2\beta_0 l}\right|_{\min}$$

$$2\beta_0 l = 0 \Rightarrow l = 0 \quad |E|_{\min} = 1 - \frac{1}{3} = \frac{2}{3}$$

$$(d) 2\beta_0 l = \pi \Rightarrow l = \frac{\pi}{2\beta_0} \quad \text{Also } 2\beta_0 l = 2\pi \Rightarrow l = \frac{2\pi}{2\beta_0} = \frac{\lambda_0}{2}$$

$$(e) \text{SWR} = \frac{E_{\max}}{E_{\min}} = \frac{4/3}{2/3} = 2$$

$$(f) Z_{in}(l = \lambda_0/4) = \eta_0 \frac{\left[1 + \Gamma_{in}(l_0 = \lambda/4)\right]}{\left[1 - \Gamma_{in}(l_0 = \lambda/4)\right]} = \eta_0 \frac{\left[1 + \left(-\frac{1}{3}\right) e^{-j2\pi(l/\lambda_0)2\pi/4}\right]}{\left[1 - \left(-\frac{1}{3}\right) e^{-j2\pi(l/\lambda_0)2\pi/4}\right]} \\ = \eta_0 \left[\frac{1 + \frac{1}{3}(-1)}{1 - \frac{1}{3}(-1)} \right] = \eta_0 \left[\frac{4/3}{2/3} \right] = \eta_0 \left(\frac{2}{1} \right) = 2\eta_0$$

$$Z_{in}(l = \lambda_0/4)_{\max} = 2(\eta_0) = 2(377) = 754$$

$$(g) Z_{in}(l = 0) = \eta_0 \frac{\left[1 + \Gamma_{in}(l = 0)\right]}{\left[1 - \Gamma_{in}(l = 0)\right]} = \eta_0 \left[\frac{1 - \frac{1}{3}}{1 - (-\frac{1}{3})} \right] = \eta_0 \left[\frac{\frac{2}{3}}{\frac{4}{3}} \right] = \frac{\eta_0}{2}$$

$$Z_{in}(l = 0)_{\min} = \frac{\eta_0}{2} = \frac{377}{2} = 188.5$$

5.6

$$\underline{E} = (\hat{a}_x - j \hat{a}_y) e^{-j 6\pi z}, \quad z = \text{meters}, f = 10^8 \text{ Hz} = 100 \text{ MHz}$$

$$(a) \beta = \omega \sqrt{\mu \epsilon} = 2\pi f \sqrt{\mu_0 \epsilon_0 \epsilon_r} = 2\pi (10^8) \sqrt{\mu_0 \epsilon_0 \epsilon_r} = 6\pi$$

$$\frac{10^8 \sqrt{\epsilon_r}}{3 \times 10^8} = 3 \Rightarrow \frac{\sqrt{\epsilon_r}}{3} = 3 \Rightarrow \sqrt{\epsilon_r} = 9 \Rightarrow \epsilon_r = 81$$

$$(b) \Gamma_x = \Gamma_y = \frac{\eta_0 - \eta}{\eta_0 + \eta} = \frac{\sqrt{\frac{\mu_0}{\epsilon_0}} - \sqrt{\frac{\mu_0}{\epsilon_r \epsilon_0}}}{\sqrt{\frac{\mu_0}{\epsilon_0}} + \sqrt{\frac{\mu_0}{\epsilon_r \epsilon_0}}} = \frac{1 - \frac{1}{\sqrt{\epsilon_r}}}{1 + \frac{1}{\sqrt{\epsilon_r}}} = \frac{\sqrt{\epsilon_r} - 1}{\sqrt{\epsilon_r} + 1} = \frac{9 - 1}{9 + 1} = \frac{8}{10}$$

$$\Gamma_x = \Gamma_y = 0.8$$

$$(c) T_x = T_y = \frac{2\eta_0}{\eta_1 + \eta_0} = 2 \frac{\sqrt{\frac{\mu_0}{\epsilon_0}}}{\sqrt{\frac{\mu_0}{\epsilon_r \epsilon_0}} + \sqrt{\frac{\mu_0}{\epsilon_0}}} = 2 \frac{1}{\frac{1}{\sqrt{\epsilon_r}} + 1} = 2 \frac{\sqrt{\epsilon_r}}{1 + \sqrt{\epsilon_r}}$$

$$T_x = T_y = 2 \frac{9}{1+9} = \frac{18}{10} = 1.8 = 1 + \Gamma$$

$$(d) \underline{E}_x^r = \Gamma_x \underline{E}_x^i = 0.8 e^{+j 6\pi z}$$

$$\underline{E}_y^r = \Gamma_y \underline{E}_y^i = -j 0.8 e^{+j 6\pi z}$$

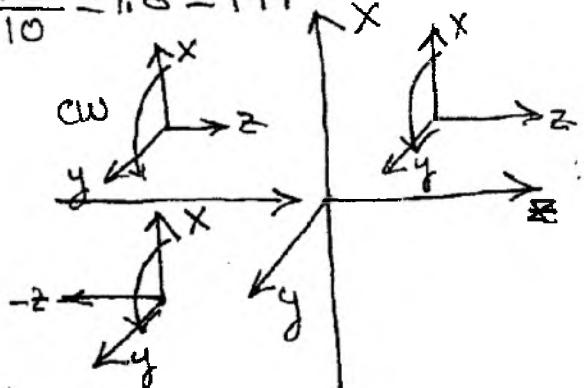
Circular

(e) CCW

$$(f) \underline{E}_x^t = T_x \underline{E}_x^i = 1.8 e^{-j \beta_0 z}$$

$$\underline{E}_y^t = T_y \underline{E}_y^i = -j 1.8 e^{-j \beta_0 z}$$

Circular



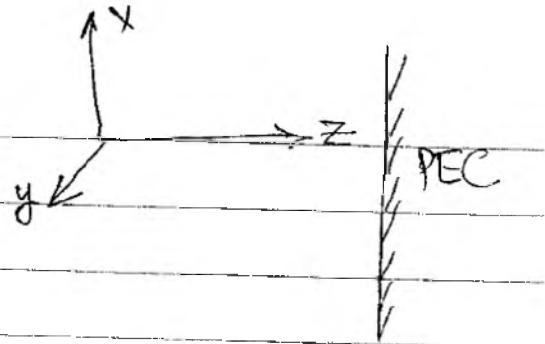
(g) cw

5.7

$$\underline{E}^i = (j\hat{x} + \hat{y}) E_0 e^{-j\beta z}$$

$$(a) \quad \underline{E}^r = (-j\hat{x} - \hat{y}) e^{+j\beta z}$$

$$= -(j\hat{x} + \hat{y}) e^{+j\beta z}$$



2 components, same magnitude, 90° out of phase \Rightarrow CP

Since x component leads y component \Rightarrow CCW(LH)

\therefore Reflected field is LH

(b)

On Poincaré sphere incident field is on South pole.

On Poincaré sphere reflected field is on North pole.

Therefore the great circle angle between the two is

$$P_w P_a = 180^\circ$$

The normalized voltage output is

$$V = C \cos\left(\frac{P_w P_a}{2}\right) = C \cos\left(\frac{180^\circ}{2}\right) = C \cos(90^\circ) = 0$$

$$V = 0 = -\infty \text{ dB}$$

The polarization of the antenna and the incident wave are cross-polarized (mismatched in polarization).

Therefore output is $0 = -\infty \text{ dB}$.

5.8 $\underline{E}^i = \hat{a}_x E_0 e^{-j\beta_0 z}$, $\Gamma^b = \frac{\eta_1 - \eta_0}{\eta_1 + \eta_0} = \frac{0 - \eta_0}{0 + \eta_0} = -1$

a. $\underline{E}^r = \hat{a}_x \Gamma^b E_0 e^{+j\beta_0 z} = -\hat{a}_x E_0 e^{+j\beta_0 z}$

b. $\underline{H}^i = \hat{a}_y \frac{E_0}{\eta_0} e^{-j\beta_0 z} = \hat{a}_y \frac{E_0}{3\pi} e^{-j\beta_0 z} = \hat{a}_y 2.6525 \times 10^{-3} E_0 e^{-j\beta_0 z}$

$$\underline{H}^r = -\hat{a}_y \frac{\Gamma^b E_0}{\eta_0} e^{+j\beta_0 z} = \hat{a}_y \frac{E_0}{3\pi} e^{+j\beta_0 z} = \hat{a}_y 2.6525 \times 10^{-3} E_0 e^{+j\beta_0 z}$$

c. $\underline{J}_s = \hat{n} \times (\underline{H}^i + \underline{H}^r) \Big|_{z=0} = \hat{n} \times (2 \underline{H}^i) \Big|_{z=0} = -2 \hat{a}_x \times \hat{a}_y 2.6525 \times 10^{-3} E_0 = \hat{a}_x 5.3 \times 10^{-3} A/m$

5.9 $\eta_1 = \sqrt{\eta_0 \eta_2} = \left\{ \sqrt{\frac{\mu_0}{\epsilon_0}} \frac{1}{2} \sqrt{\frac{\mu_0}{\epsilon_0}} \right\}^{1/2} = \frac{1}{\sqrt{2}} \sqrt{\frac{\mu_0}{\epsilon_0}} = \frac{1}{2} \eta_0 = \sqrt{\frac{\mu_0}{2\epsilon_0}} = \sqrt{\frac{\mu_0}{\epsilon_r \epsilon_0}} \Rightarrow \epsilon_r = 2$

5.10 In the three different regions, we can write the corresponding fields as:

$$E_1 = E_1^+ e^{-j\beta_0 z} + E_1^- e^{+j\beta_0 z} \quad \left. \begin{array}{l} \\ \end{array} \right\} z \leq 0$$

$$H_1 = \frac{E_1^+}{\eta_0} e^{-j\beta_0 z} - \frac{E_1^-}{\eta_0} e^{+j\beta_0 z} \quad \left. \begin{array}{l} \\ \end{array} \right\} x\text{-polarized while the magnetic fields are } y\text{-polarized.}$$

$$E_2 = E_2^+ e^{-j\beta_1 z} + E_2^- e^{+j\beta_1 z} \quad \left. \begin{array}{l} \\ \end{array} \right\} 0 \leq z \leq t$$

$$H_2 = \frac{E_2^+}{\eta_1} e^{-j\beta_1 z} - \frac{E_2^-}{\eta_1} e^{+j\beta_1 z} \quad \left. \begin{array}{l} \\ \end{array} \right\} 0 \leq z \leq t$$

$$E_3 = E_3^+ e^{-j\beta_0(z-t)} \quad \left. \begin{array}{l} \\ \end{array} \right\} t \leq z$$

$$H_3 = \frac{E_3^+}{\eta_0} e^{-j\beta_0(z-t)} \quad \left. \begin{array}{l} \\ \end{array} \right\} t \leq z$$

Applying the continuity of the tangential electric and magnetic fields at the interface at $z=0$, we have

$$E_1^+ + E_1^- = E_2^+ + E_2^- \Rightarrow E_1^+ + E_1^- = E_2^+ + E_2^- \quad (1)$$

$$\frac{E_1^+}{\eta_0} - \frac{E_1^-}{\eta_0} = \frac{E_2^+}{\eta_1} - \frac{E_2^-}{\eta_1} \Rightarrow E_1^+ - E_1^- = \frac{\eta_0}{\eta_1} (E_2^+ - E_2^-) \quad (2)$$

Solving (1) and (2), we get

$$E_1^+ = \frac{E_2^+}{2} \left(1 + \frac{\eta_0}{\eta_1} \right) + \frac{E_2^-}{2} \left(1 - \frac{\eta_0}{\eta_1} \right) = E_2^+ \left(\frac{\eta_1 + \eta_0}{2\eta_1} \right) + E_2^- \left(\frac{\eta_1 - \eta_0}{2\eta_1} \right) \quad \text{cont'd.} \quad (3)$$

5.10 cont'd.

$$E_1^- = E_2^+ \left(\frac{\eta_1 - \eta_0}{2\eta_1} \right) + E_2^- \left(\frac{\eta_1 + \eta_0}{2\eta_1} \right) \quad (4)$$

Now apply the boundary conditions at $z=t$, and we get the following:

$$E_2^+ e^{-j\beta_1 t} + E_2^- e^{+j\beta_1 t} = E_3^+ \Rightarrow E_2^+ e^{-j\beta_1 t} + E_2^- e^{+j\beta_1 t} = E_3^+ \quad (5)$$

$$\frac{E_2^+ e^{-j\beta_1 t}}{\eta_1} - \frac{E_2^- e^{+j\beta_1 t}}{\eta_1} = \frac{E_3^+}{\eta_0} \Rightarrow E_2^+ e^{-j\beta_1 t} - E_2^- e^{+j\beta_1 t} = \frac{\eta_1}{\eta_0} E_3^+ \quad (6)$$

Solving (5) and (6) for E_2^+ and E_2^- leads to

$$E_2^+ = \frac{E_3^+ e^{+j\beta_1 t}}{2} \left(\frac{\eta_1 + \eta_0}{\eta_0} \right) = E_3^+ e^{+j\beta_1 t} \left(\frac{\eta_1 + \eta_0}{2\eta_0} \right) = E_3^+ e^{+j\beta_1 t} \left(\frac{\eta_0 + \eta_1}{2\eta_0} \right) \quad (7)$$

$$E_2^- = E_3^+ e^{-j\beta_1 t} \left(\frac{\eta_0 - \eta_1}{2\eta_0} \right) = -E_3^+ e^{-j\beta_1 t} \left(\frac{\eta_1 - \eta_0}{2\eta_0} \right) \quad (8)$$

Substituting (7) and (8) into (3) we can write that

$$E_1^+ = E_3^+ \frac{e^{+j\beta_1 t}}{\eta_0 \eta_1} \left(\frac{\eta_0 + \eta_1}{2} \right)^2 - E_3^+ \frac{e^{-j\beta_1 t}}{\eta_0 \eta_1} \left(\frac{\eta_1 - \eta_0}{2} \right)^2 = \frac{E_3^+}{4\eta_0 \eta_1} \left[e^{+j\beta_1 t} (\eta_0 + \eta_1)^2 - e^{-j\beta_1 t} (\eta_1 - \eta_0)^2 \right]$$

or

$$\frac{E_3^+}{E_1^+} \equiv T^b = \frac{4\eta_0 \eta_1}{(\eta_1 + \eta_0)^2 e^{+j\beta_1 t} - (\eta_1 - \eta_0)^2 e^{-j\beta_1 t}} = \frac{4\eta_0 \eta_1 e^{-j\beta_1 t}}{(\eta_1 + \eta_0)^2 - (\eta_1 - \eta_0)^2 e^{-j2\beta_1 t}}$$

To find the reflection coefficient, we substitute (7) and (8) into (3) and (4) and then take the ratio of (4) and (3). This leads to

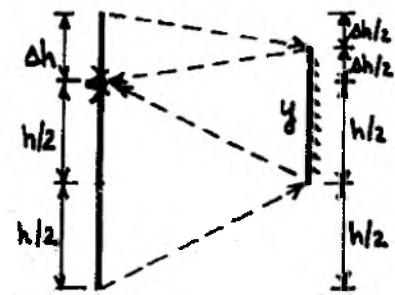
$$\frac{E_1^-}{E_1^+} \equiv R^b = \frac{\frac{e^{+j\beta_1 t}}{4\eta_0 \eta_1} (\eta_1 - \eta_0)(\eta_1 + \eta_0) - \frac{e^{-j\beta_1 t}}{4\eta_0 \eta_1} (\eta_1 + \eta_0)(\eta_1 - \eta_0)}{\frac{e^{+j\beta_1 t}}{4\eta_0 \eta_1} (\eta_1 + \eta_0)(\eta_1 + \eta_0) - \frac{e^{-j\beta_1 t}}{4\eta_0 \eta_1} (\eta_1 - \eta_0)(\eta_1 - \eta_0)}$$

$$\frac{E_1^-}{E_1^+} \equiv R = \frac{(\eta_1 + \eta_0)(\eta_1 - \eta_0)(1 - e^{-j2\beta_1 t})}{(\eta_1 + \eta_0)^2 - (\eta_1 - \eta_0)^2 e^{-j2\beta_1 t}}$$

5.11

As can be seen from the figure the mirror dimensions are equal to

$$y = \frac{h}{2} + \frac{\Delta h}{2} = \frac{1}{2}(h + \Delta h)$$



5.12

For the prism with a dielectric constant of 2.25, the critical angle is

$$\theta_c = \sin^{-1} \sqrt{\frac{\epsilon_0}{\epsilon_r \epsilon_0}} = \sin^{-1} \sqrt{\frac{1}{2.25}} = \sin^{-1} \sqrt{0.4444} = \sin^{-1}(0.6666) \approx 41.81^\circ$$

Therefore at the hypotenuse the reflection coefficient $|\Gamma| = 1$ since the incident angle of 45° is greater than the critical angle of 41.81° .

$$\Gamma_{20} = \frac{\eta_2 - \eta_0}{\eta_2 + \eta_0} = \frac{\frac{1}{2.5}\eta_0 - \eta_0}{\frac{1}{2.5}\eta_0 + \eta_0} = \frac{1 - 1.5}{1 + 1.5} = -\frac{0.5}{2.5} = -\frac{1}{5} = -0.2$$

$$\Gamma_{02} = -\Gamma_{20} = 0.2$$

$$\frac{S_{av}^{ex}}{S_{av}^c} = (1 - |\Gamma_{20}|^2)(1^2)(1 - |\Gamma_{02}|^2) = (1 - |\Gamma_{20}|^2)^2 = [1 - (0.2)^2]^2 = (0.96)^2 = 0.9216 \text{ or } 92.16\%$$

5.13

$$\beta_0 \sin \theta_i = \beta_2 \sin \theta_2 \Rightarrow \theta_2 = \sin^{-1} \left(\frac{\beta_0}{\beta_2} \sin \theta_i \right) = \sin^{-1} \left(\frac{\beta_0}{2\beta_0} \sin 30^\circ \right) = \sin^{-1} \left(\frac{1}{4} \right) = 14.48^\circ$$

$$\beta_0 \sin(30^\circ) = \beta_2 \sin \theta_2 = \beta_3 \sin \theta_3 \Rightarrow \theta_3 = 30^\circ$$

5.14

$$|\Gamma_1|^2 = \left| \frac{\cos \theta_i - \sqrt{\frac{\epsilon_2}{\epsilon_1}} \sqrt{1 - \left(\frac{\epsilon_1}{\epsilon_2}\right) \sin^2 \theta_i}}{\cos \theta_i + \sqrt{\frac{\epsilon_2}{\epsilon_1}} \sqrt{1 - \frac{\epsilon_1}{\epsilon_2} \sin^2 \theta_i}} \right|^2 = \left| \frac{\cos \theta_i - 2 \sqrt{1 - \frac{1}{4} \sin^2 \theta_i}}{\cos \theta_i + 2 \sqrt{1 - \frac{1}{4} \sin^2 \theta_i}} \right|^2 = \frac{1}{4}$$

When this equation is solved iteratively, it is found that $\theta_i \approx 52^\circ$

5.15

$$|\Gamma_{11}|^2 = \left| \frac{-\cos \theta_i + \sqrt{\frac{\epsilon_1}{\epsilon_2}} \sqrt{1 - \frac{\epsilon_1}{\epsilon_2} \sin^2 \theta_i}}{\cos \theta_i + \sqrt{\frac{\epsilon_1}{\epsilon_2}} \sqrt{1 - \frac{\epsilon_1}{\epsilon_2} \sin^2 \theta_i}} \right|^2 = \left| \frac{-\cos \theta_i + \frac{1}{2} \sqrt{1 - \frac{1}{4} \sin^2 \theta_i}}{\cos \theta_i + \frac{1}{2} \sqrt{1 - \frac{1}{4} \sin^2 \theta_i}} \right|^2 = \frac{1}{4}$$

When this equation is solved iteratively, it is found that $\theta_i \approx 81.65^\circ$

5.16

$$\theta_g = \tan^{-1} \sqrt{\frac{\epsilon_2}{\epsilon_1}}$$

$$\epsilon_r = 2.56 : \theta_g = \tan^{-1}(1.6) = 57.99^\circ$$

$$\epsilon_r = 4 : \theta_g = \tan^{-1}(2.0) = 63.435^\circ$$

$$\epsilon_r = 9 : \theta_g = \tan^{-1}(3.0) = 71.565^\circ$$

$$\epsilon_r = 16 : \theta_g = \tan^{-1}(4) = 75.964^\circ$$

$$\epsilon_r = 25 : \theta_g = \tan^{-1}(5) = 78.64^\circ$$

$$\epsilon_r = 81 : \theta_g = \tan^{-1}(9) = 83.66^\circ$$

5.17] The easiest way to solve this problem is to use the same procedure as used for the solution of Problem 5.10. The end result for the reflection and transmission coefficients is the same as that of Problem 5.10 except that instead of using the intrinsic impedances we must use the wave impedances in the z-direction, and we must use the phase constant along the z-direction.

Thus $\Gamma_{||}^b = \frac{(Z_{1z} + Z_{0z})(Z_{1z} - Z_{0z})[1 - e^{-j2\beta_{1z}t}]}{(Z_{1z} + Z_{0z})^2 - (Z_{1z} - Z_{0z})^2 e^{-j2\beta_{1z}t}}$

 $T_{||}^b = \frac{4Z_{0z}Z_{1z}e^{-j\beta_{1z}t}}{(Z_{1z} + Z_{0z})^2 - (Z_{1z} - Z_{0z})^2 e^{-j2\beta_{1z}t}}$

where according to (4-20b) and Figure 4-6

$$\left. \begin{array}{l} Z_{0z} = \eta_0 \cos \theta_i \\ Z_{1z} = \eta_1 \cos \theta_1 \\ \beta_{1z} = \beta_1 \cos \theta_1 \\ \eta_0 = \sqrt{\frac{\mu_0}{\epsilon_0}}, \eta_1 = \sqrt{\frac{\mu_1}{\epsilon_1}} \\ \beta_1 = \omega \sqrt{\mu_1 \epsilon_1} \end{array} \right\}$$

$$\left. \begin{array}{l} \beta_0 \sin \theta_i = \beta_1 \sin \theta_1 \\ \sin \theta_1 = \frac{\beta_0}{\beta_1} \sin \theta_i = \sqrt{\frac{\epsilon_0}{\epsilon_1}} \sin \theta_i \\ \cos \theta_1 = \sqrt{1 - \sin^2 \theta_1} = \sqrt{1 - \frac{\epsilon_0}{\epsilon_1} \sin^2 \theta_i} \end{array} \right.$$

5.18] As for Problem 5.17, the easiest way to solve this problem is to use the same procedure as used for the solution of Problem 5.10. The end result for the reflection and transmission coefficients is the same as that of Problem 5.10 except that instead of using the intrinsic impedances we must use the wave impedances in the z-direction, and we must use the phase constant along the z-direction.

Thus

$$\Gamma_{\perp}^b = \frac{(Z_{1z} + Z_{0z})(Z_{1z} - Z_{0z})[1 - e^{-j2\beta_{1z}t}]}{(Z_{1z} + Z_{0z})^2 - (Z_{1z} - Z_{0z})^2 e^{-j2\beta_{1z}t}}$$

$$T_{\perp} = \frac{4Z_{0z}Z_{1z}e^{-j\beta_{1z}t}}{(Z_{1z} + Z_{0z})^2 - (Z_{1z} - Z_{0z})^2 e^{-j2\beta_{1z}t}}$$

where according to (4-21b) and Figure 4-6

$$\left. \begin{array}{l} Z_{0z} = \frac{\eta_0}{\cos \theta_i} \quad \beta_{1z} = \beta_1 \cos \theta_1 \\ Z_{1z} = \frac{\eta_1}{\cos \theta_1} \quad \beta_1 = \omega \sqrt{\mu_1 \epsilon_1} \end{array} \right\} \left. \begin{array}{l} \beta_0 \sin \theta_i = \beta_1 \sin \theta_1 \\ \sin \theta_1 = \frac{\beta_0}{\beta_1} \sin \theta_i = \sqrt{\frac{\epsilon_0}{\epsilon_1}} \sin \theta_i \\ \cos \theta_1 = \sqrt{1 - \sin^2 \theta_1} = \sqrt{1 - \frac{\epsilon_0}{\epsilon_1} \sin^2 \theta_i} \end{array} \right.$$

5.19 $\theta_c = \sin^{-1} \sqrt{\frac{\epsilon_2}{\epsilon_1}} = \sin^{-1} \left(\sqrt{\frac{4}{3}} \right) = \sin^{-1} \left(\frac{2}{\sqrt{3}} \right) = \sin^{-1} (0.667) = 41.8^\circ$

5.20 $\theta_B = \tan^{-1} \left(\sqrt{\frac{\epsilon_2}{\epsilon_1}} \right), \quad \theta_c = \sin^{-1} \left(\sqrt{\frac{\epsilon_2}{\epsilon_1}} \right)$

(a) $\theta_B = \tan^{-1} \left(\sqrt{\frac{1}{3}} \right) = \tan^{-1} \left(\frac{1}{\sqrt{3}} \right) = 6.34^\circ \quad \theta_c = \sin^{-1} \left(\sqrt{\frac{1}{3}} \right) = \sin^{-1} \left(\frac{1}{\sqrt{3}} \right) = 6.379^\circ$

(b) $\theta_B = \tan^{-1} \left(\sqrt{3} \right) = \tan^{-1}(a) = 83.64^\circ \quad \theta_c = \sin^{-1} \left(\sqrt{3} \right) = \sin^{-1}(a) \approx \text{does not exist}$

(c) $\theta_B = \tan^{-1} \left(\sqrt{\frac{1}{9}} \right) = \tan^{-1} \left(\frac{1}{3} \right) = 18.436^\circ \quad \theta_c = \sin^{-1} \left(\sqrt{\frac{1}{9}} \right) = \sin^{-1} \left(\frac{1}{3} \right) = 19.47^\circ$

5.21 $\theta_c = \sin^{-1} \left(\sqrt{\frac{\epsilon_2}{\epsilon_1}} \right)$

a. $\theta_c = \sin^{-1} \left(\frac{1}{1.6} \right) = 38.68^\circ \quad \theta_c = \sin^{-1} \left(\frac{1}{4} \right) = 14.478^\circ$

$\theta_c = \sin^{-1} \left(\frac{1}{2} \right) = 30^\circ \quad \theta_c = \sin^{-1} \left(\frac{1}{3} \right) = 11.54^\circ$

$\theta_c = \sin^{-1} \left(\frac{1}{3} \right) = 19.47^\circ \quad \theta_c = \sin^{-1} \left(\frac{1}{9} \right) = 6.379^\circ$

b. $\theta_B = \tan^{-1} \left(\sqrt{\frac{\epsilon_2}{\epsilon_1}} \right)$

$\theta_B = \tan^{-1} \left(\frac{1}{1.6} \right) = 32.0^\circ \quad \theta_B = \tan^{-1} \left(\frac{1}{4} \right) = 14.036^\circ$

$\theta_B = \tan^{-1} \left(\frac{1}{2} \right) = 26.565^\circ \quad \theta_B = \tan^{-1} \left(\frac{1}{3} \right) = 11.31^\circ$

$\theta_B = \tan^{-1} \left(\frac{1}{3} \right) = 18.435^\circ \quad \theta_B = \tan^{-1} \left(\frac{1}{9} \right) = 6.34^\circ$

c. As the dielectric constant gets larger or the difference between the two media gets greater, the critical angle and Brewster angle are approaching the same value.

d. The plots are shown in the pages to follow.

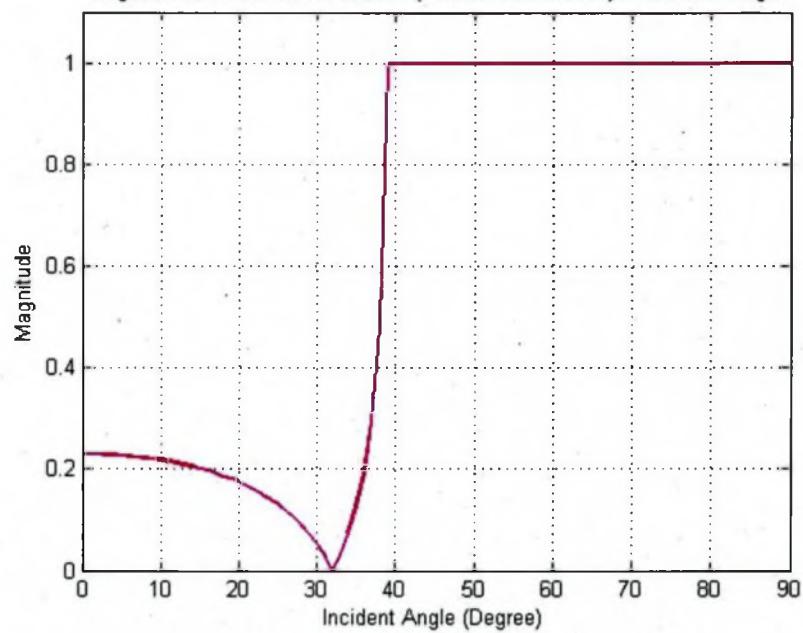
e. The plots are shown in the pages that follow.

Cont'd.

5.21 Cont'd

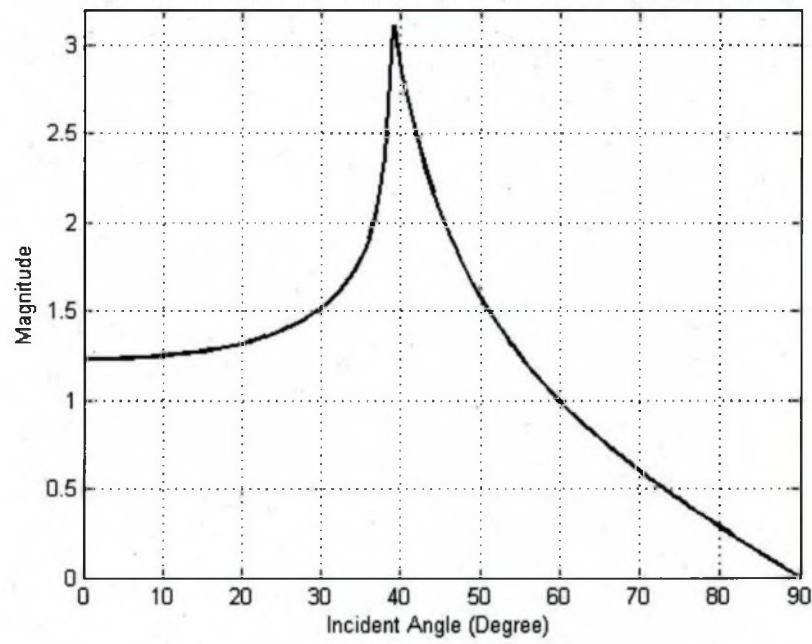
$$|\Gamma_{\parallel}|$$

Magnitude of Reflection coefficient (Parallel Polarization) vs Incident Angle



$$|T_{\parallel}|$$

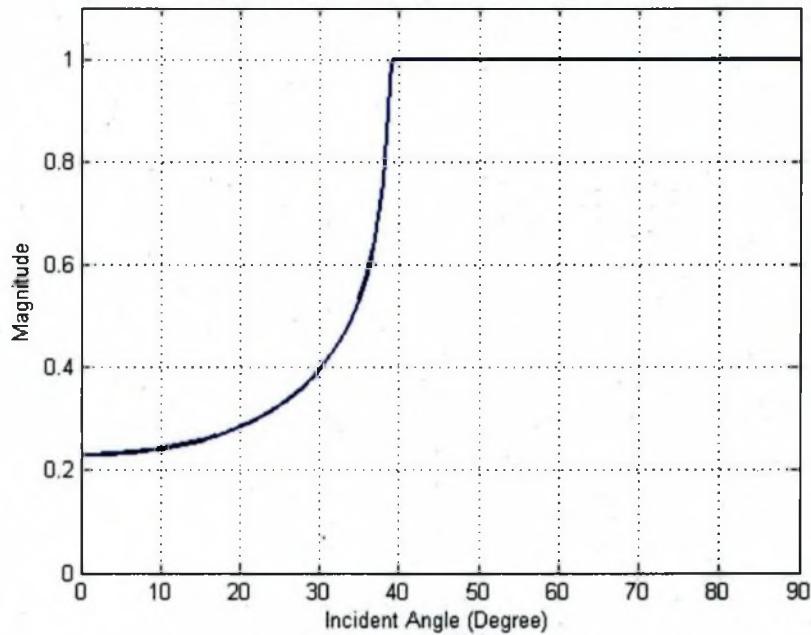
Magnitude of Transmission coefficient (Parallel Polarization) vs Incident Angle



5.21

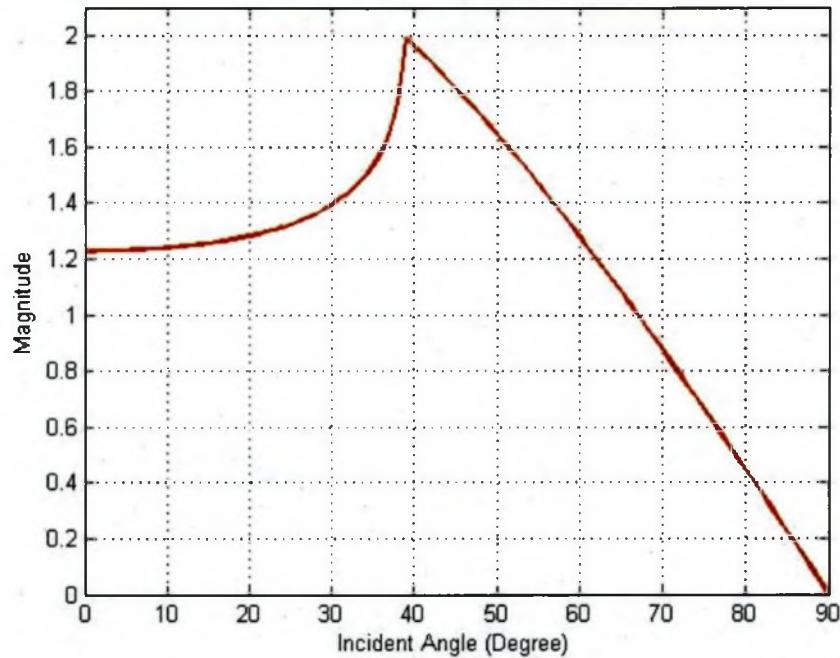
$$|\Gamma_{\perp}|$$

Magnitude of Reflection coefficient (Perpendicular Polarization) vs Incident Angle

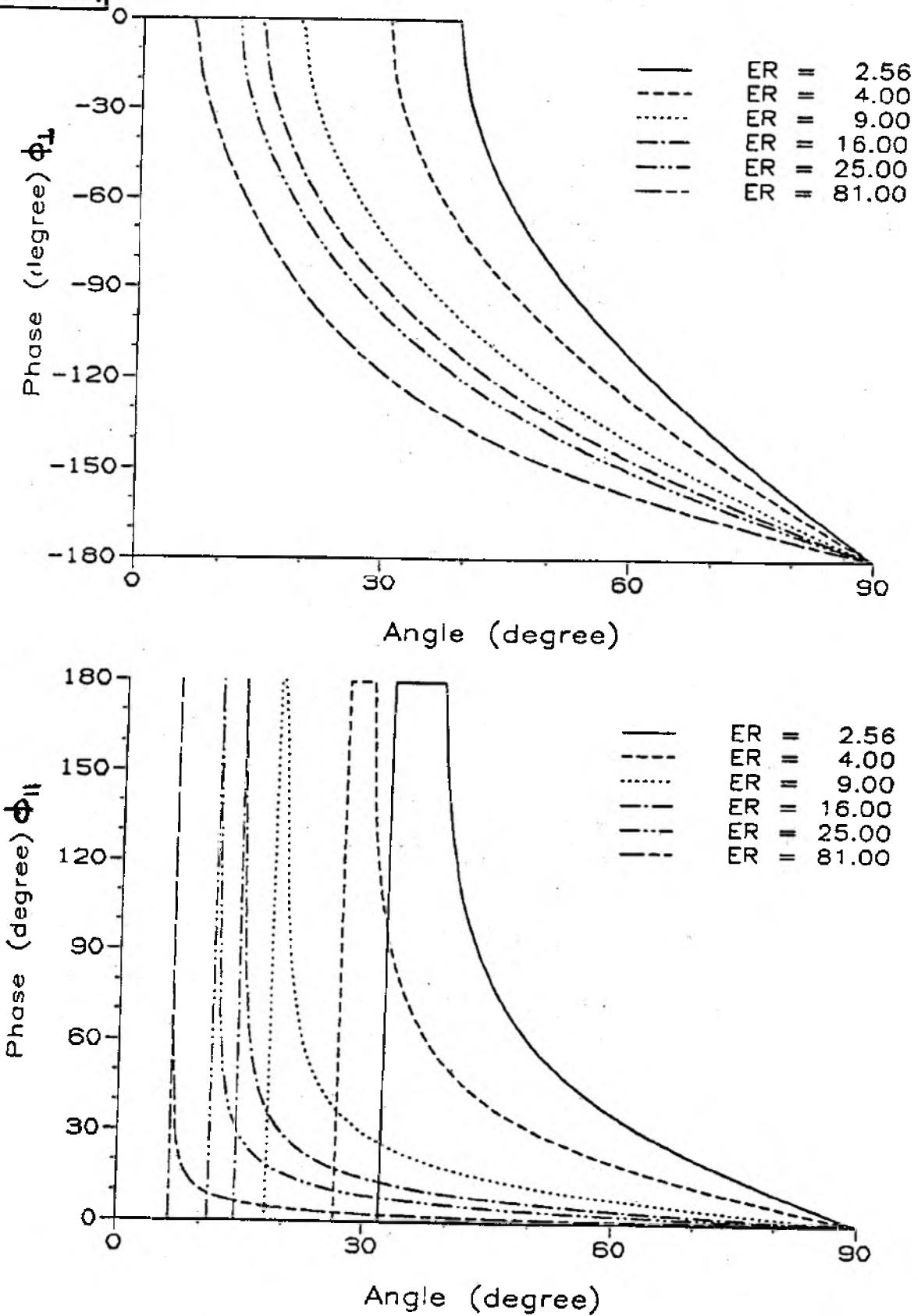


$$|T_{\perp}|$$

Magnitude of Transmission coefficient (Perpendicular Polarization) vs Incident Angle

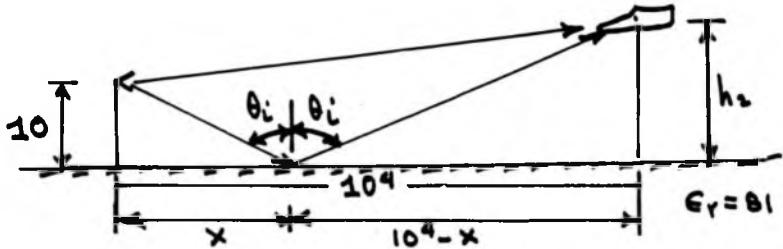


5.21 Cont'd.



5.22

$$\theta_B = \tan^{-1}(\sqrt{\frac{\epsilon_2}{\epsilon_1}})$$



For the reflected wave

not to possess a parallel polarized component, then the incidence angle must be equal to the Brewster angle. Thus

$$\theta_i = \theta_B = \tan^{-1}\left(\frac{x}{10}\right) = \tan^{-1}\left(\sqrt{\frac{\epsilon_2}{\epsilon_1}}\right) = \tan^{-1}\left(\sqrt{81}\right) = \tan^{-1}(9) = 83.66^\circ$$

$x = 90 \text{ meter}$

$$\text{Thus } \tan^{-1}\left(\frac{10^4 - 90}{h_2}\right) = \tan^{-1}\left(\sqrt{81}\right) = \tan^{-1}(9) \Rightarrow \frac{10^4 - 90}{h_2} = 9 \Rightarrow h_2 = \frac{10^4 - 90}{9} = 1,101.11 \text{ m}$$

5.23

For a height of 1,101.11 m the incident angle is equal to 83.66° which is equal to the Brewster angle (see the solution to Problem 5.22). The refraction angle θ_t measured from the vertical is equal to

$$\beta_0 \sin \theta_i = \beta \sin \theta_t \Rightarrow \theta_t = \sin^{-1}\left(\frac{\beta_0}{\beta} \sin \theta_i\right) = \sin^{-1}\left[\frac{1}{9} \sin(83.66^\circ)\right] = 6.34^\circ$$

Therefore

$$\Gamma_{||} = \frac{\eta \cos \theta_t - \eta_0 \cos \theta_i}{\eta \cos \theta_t + \eta_0 \cos \theta_i} = \frac{\cos \theta_t - \sqrt{\epsilon_r} \cos \theta_i}{\cos \theta_t + \sqrt{\epsilon_r} \cos \theta_i} = \frac{\cos(6.34^\circ) - 9 \cos(83.66^\circ)}{\cos(6.34^\circ) + 9 \cos(83.66^\circ)} = 0 = |\Gamma_{||}| |\xi_{||}^t|$$

$|\Gamma_{||}| = 0$
 $|\xi_{||}^t| = 0^\circ$

$$T_{||} = \frac{\omega \theta_i}{\cos \theta_t} (1 + \Gamma_{||}) = \frac{\omega \theta_i}{\cos \theta_t} = \frac{\omega \sin(83.66^\circ)}{\cos(6.34^\circ)} = 0.111 \text{ L}^0 = |\Gamma_{||}| |\xi_{||}^t| \Rightarrow |\Gamma_{||}| = 0.111$$

$\xi_{||}^t = 0^\circ$

$$\Gamma_{\perp} = \frac{\eta \cos \theta_i - \eta_0 \cos \theta_t}{\eta \cos \theta_i + \eta_0 \cos \theta_t} = \frac{\cos \theta_i - \sqrt{\epsilon_r} \cos \theta_t}{\cos \theta_i + \sqrt{\epsilon_r} \cos \theta_t} = -0.9756 = |\Gamma_{\perp}| |\xi_{\perp}^r| \Rightarrow |\Gamma_{\perp}| = 0.9756$$

$\xi_{\perp}^r = 180^\circ$

$$T_{\perp} = 1 + \Gamma_{\perp} = 1 - 0.9756 = 0.0244 = |\Gamma_{\perp}| |\xi_{\perp}^r| \Rightarrow |\Gamma_{\perp}| = 0.0244$$

$|\xi_{\perp}^r| = 0^\circ$

- a. $\delta_r = \delta_i - \pi + (\gamma_{\perp}^r - \gamma_{||}^t) = -90^\circ - 180^\circ + (180^\circ - 0) = -90^\circ$ } The reflected field
 $\gamma_r = \tan^{-1}\left(\frac{|\Gamma_{||}|}{|\Gamma_{\perp}|} \tan \gamma_i\right) = \tan^{-1}\left(\frac{0}{0.9756}(1)\right) = 0^\circ$ } is linearly polarized.

cont'd.

5.23 cont'd.

$$b. \delta_t = \delta_i + \left(\frac{t}{\epsilon_{\perp}} - \frac{t}{\epsilon_{||}} \right) = -90 + (0 - 0) = -90^\circ$$

$$\gamma_t = \tan^{-1} \left(\frac{1/T_{\perp}}{1/T_{||}} \tan \gamma_i \right) = \tan^{-1} \left[\frac{0.0244}{0.111} (1) \right] = \tan^{-1}(0.2198) = 12.398^\circ$$

Therefore the transmitted wave is elliptically, right-hand polarized.

5.24

$$\tan(90 - \theta_i) = \frac{100}{x} \quad (1)$$

$$\tan(90 - \theta_i) = \frac{10}{s-x} \quad (2)$$

Therefore

$$\frac{100}{x} = \frac{10}{s-x} \Rightarrow x = \frac{100}{110}s = \frac{10}{11}s \Rightarrow s = \frac{11}{10}x$$

To eliminate reflections for the parallel-polarized component the incidence angle must be equal to the Brewster angle. Thus

$$\theta_i = \tan^{-1} \left(\sqrt{\frac{\epsilon_2}{\epsilon_0}} \right) = \tan^{-1} \left(\sqrt{16} \right) = 75.96^\circ$$

Therefore from (1)

$$x = \frac{100}{\tan(90^\circ - 75.96^\circ)} = \frac{100}{\tan(14.04^\circ)} = \frac{100}{0.25} = 400 \text{ m} \Rightarrow s = \frac{11}{10}x = \frac{11}{10}(400) = 440 \text{ m}$$

5.25 $\theta_c = \sin^{-1} \left(\sqrt{\frac{1}{\epsilon_1}} \right) = \sin^{-1} \left(\frac{1}{\sqrt{81}} \right) = 6.379^\circ$

$$\tan \theta_c = \frac{x_1}{d} \Rightarrow x_1 = d \tan \theta_c = d \tan(6.379^\circ) = 0.1118d$$

Therefore $|x| \leq 0.1118d$

5.26 a. $\theta_r = \theta_c = \sin^{-1} \left(\sqrt{\frac{\epsilon_0}{\epsilon_r}} \right) = \sin^{-1} \left(\frac{1}{\sqrt{6}} \right) = 30^\circ \Rightarrow \frac{1}{\sqrt{\epsilon_r}} = \sin(30^\circ) = \frac{1}{2} \Rightarrow \epsilon_r = 4$

Since $\theta_c \geq \theta_r \Rightarrow \epsilon_r \leq 4$

b. $\beta \sin \theta_i = \beta_0 \sin \theta_t = \beta_0 \sin \theta_e$

$$\theta_e = \sin^{-1} \left(\frac{\beta}{\beta_0} \sin \theta_i \right) = \sin^{-1} \left(\sqrt{\epsilon_r} \sin 30^\circ \right) = \sin^{-1} \left(2 \frac{1}{2} \right) = \sin^{-1}(1) = 90^\circ$$

5.27

$$a. \theta_B = \tan^{-1}\left(\sqrt{\frac{\epsilon_0}{4\epsilon_0}}\right) = \tan^{-1}\left(\frac{1}{2}\right) = 26.565^\circ$$

$$b. \theta_c = \sin^{-1}\left(\sqrt{\frac{\epsilon_0}{4\epsilon_0}}\right) = \sin^{-1}\left(\frac{1}{2}\right) = 30^\circ$$

5.28

$$a. \beta_0 \sin \theta_i = \beta \sin \theta_c \Rightarrow \sin \theta_c = \frac{\beta_0}{\beta} \sin \theta_i$$

$$\theta_c = \sin^{-1}\left[\frac{1}{\sqrt{\epsilon_0}} \sin(90^\circ)\right] = \sin^{-1}\left(\frac{1}{2}\right) = 6.379^\circ \Rightarrow 2\theta_c = 12.758^\circ$$

$$b. \tan \theta_c = \frac{D/2}{h} \Rightarrow h = \frac{D/2}{\tan \theta_c} = \frac{D/2}{\tan(6.379^\circ)} = 4.472 D$$

5.29 To make the brightness temperature equal to the thermal temperature,

a. the reflection coefficient must be zero. Therefore the incidence angle must be the Brewster angle and the polarization must be parallel.

$$b. |\Gamma_{||}| = 0 \Rightarrow \theta_i = \theta_B = \tan^{-1}\left(\sqrt{\frac{\epsilon_0}{\epsilon_0}}\right) = \tan^{-1}(\sqrt{81}) = \tan^{-1}(9) = 83.66^\circ$$

5.30 Since the electric field must be reduced to 0.368 of its original value, it must travel a distance equal to its skin depth. Since

$$\frac{\sigma}{\omega \epsilon} = \frac{3}{2\pi(10^4)(\frac{10^{-9}}{36\pi})81} = \frac{3(10)}{81} \times 10^5 = \frac{2}{3} \times 10^5 \gg 1$$

then

$$\delta = \sqrt{\frac{2}{\omega \mu \sigma}} = \sqrt{\frac{1}{4\pi \mu \sigma}} = \sqrt{\frac{1}{\pi(4\pi \times 10^{-7})(10^4)3}} = \frac{1}{2\pi} \sqrt{\frac{1000}{3}} = 2.9057 \text{ meters}$$

5.31

$$a. \frac{\sigma}{\omega \epsilon} = \frac{1}{2\pi(10^4)(\frac{81 \times 10^{-9}}{36\pi})} = \frac{10}{81} \times 10^3 = \frac{2}{9} \times 10^3 \gg 1 \Rightarrow \text{Good conductor}$$

$$\eta = \sqrt{\frac{\omega \mu}{2\sigma}} (1+j) = \sqrt{\frac{2\pi \times 10^4 (4\pi \times 10^{-7})}{2(1)}} (1+j) = 1.987(1+j) = 2.810 \angle 45^\circ$$

$$\Gamma = \frac{\eta - \eta_0}{\eta + \eta_0} = \frac{1.987(1+j) - 377}{1.987(1+j) + 377} = \frac{-375.013 + j1.987}{378.987 + j1.987} = 0.9895 \angle 179.4^\circ$$

$$E^r = |\Gamma| E^i = 0.9895 (1 \times 10^{-3}) = 0.9895 \times 10^{-3} \text{ V/m}$$

$$b. SWR = \frac{1+|\Gamma|}{1-|\Gamma|} = \frac{1+0.9895}{1-0.9895} = 189.476$$

$$c. S^i = \frac{1}{2\eta} |E^i|^2 = \frac{1}{2(377)} (10^{-3})^2 = 1.326 \times 10^{-9} \text{ W/m}^2$$

$$S^r = \frac{1}{2\eta_0} |\Gamma E^i|^2 = \frac{1}{2\eta_0} |\Gamma E^i|^2 = |\Gamma|^2 S^i = (0.9895)^2 (1.326 \times 10^{-9}) = 1.2983 \times 10^{-9} \text{ W/m}^2$$

cont'd.

5.31 cont'd.

$$d. \quad T = \frac{2\eta}{\eta + \eta_0} = \frac{2(2.810^{45^\circ})}{378.992 + j1.981} = \frac{2(2.810^{45^\circ})}{378.992 \text{ rad}} = 0.0148 |44.7^\circ|$$

$$E^t = |T| E^L = 0.0148 \times 10^{-3} \text{ V/m}$$

$$e. \quad S_L^t = S^L - S^R = S^L - |\eta|^2 S^L = (1 - |\eta|^2) S^L = (1 - |0.9895|^2) 1.326 \times 10^{-9} = 0.0277 \times 10^{-9} \text{ W/m}^2$$

$$f. \quad \delta = \sqrt{\frac{2}{\omega \mu \sigma}} = \sqrt{\frac{2}{\pi f \mu \sigma}} = \sqrt{\frac{2}{\pi (10^6) (4\pi \times 10^{-7}) (1)}} = \frac{1}{2\pi} \sqrt{10} = 0.5033 \text{ m}$$

$$g. \quad \lambda = 2\pi \sqrt{\frac{2}{\omega \mu \sigma}} = 2\pi \delta = 2\pi (0.5033) = 3.1623 \text{ m}$$

$$d = 20\lambda = 2(3.1623) = 6.32455 \text{ m}$$

$$h. \quad v = \sqrt{\frac{2\omega}{\mu \sigma}} = \sqrt{\frac{2(2\pi \times 10^6)}{4\pi \times 10^{-7} (1)}} = \sqrt{10} (10^6) = 3.1623 \times 10^6 \text{ m/sec.}$$

$$t = \frac{d}{v} = \frac{100}{3.1623 \times 10^6} = 31.623 \times 10^{-6} \text{ sec.}$$

$$i. \quad \frac{v}{v_0} = \frac{3.1623 \times 10^6}{3 \times 10^8} = 1.054 \times 10^{-2}$$

5.32 $\eta_2 \approx \sqrt{\frac{\omega \mu_0}{2\sigma_2}} (1+j), \quad \eta_1 \approx \sqrt{\frac{\omega \mu_0}{2\sigma_1}} (1+j)$

$$|\Gamma| = \left| \frac{\eta_2 - \eta_1}{\eta_2 + \eta_1} \right| = \left| \frac{\sqrt{\sigma_1} - \sqrt{\sigma_2}}{\sqrt{\sigma_1} + \sqrt{\sigma_2}} \right| = \frac{|\sqrt{\sigma_1} - \sqrt{\sigma_2}|}{|\sqrt{\sigma_1} + \sqrt{\sigma_2}|} = \frac{\sqrt{\sigma_1} - \sqrt{\sigma_2}}{\sqrt{\sigma_1} + \sqrt{\sigma_2}}$$

$$\text{SWR} = \frac{1+|\Gamma|}{1-|\Gamma|} \Rightarrow |\Gamma| = \frac{\text{SWR} - 1}{\text{SWR} + 1} = \frac{1.5 - 1}{1.5 + 1} = \frac{0.5}{2.5} = \frac{1}{5}$$

$$|\Gamma| = \frac{1}{5} = \frac{\sqrt{\sigma_1} - \sqrt{\sigma_2}}{\sqrt{\sigma_1} + \sqrt{\sigma_2}} \Rightarrow \sqrt{\sigma_1} + \sqrt{\sigma_2} = 5\sqrt{\sigma_1} - 5\sqrt{\sigma_2} \Rightarrow 4\sqrt{\sigma_1} = 6\sqrt{\sigma_2} \Rightarrow \frac{\sqrt{\sigma_1}}{\sqrt{\sigma_2}} = 1.5 \Rightarrow \frac{\sigma_1}{\sigma_2} = (1.5)^2 = 2.25$$

5.33 a. For right hand circular $\psi = 90^\circ$

$$b. \quad H^L = (j\hat{a}_y - \hat{a}_z) \frac{E_0}{\eta_0} e^{+j\beta_0 x}$$

$$c. \quad \frac{\sigma}{\omega \epsilon} = \frac{10^{-1}}{2\pi(10^6) \left(\frac{8(2\pi 10^6)}{36\pi} \right)} = \frac{10}{81} \times 10^{-1} \ll 1 \Rightarrow \text{Good dielectric}$$

$$\Gamma = \frac{\eta - \eta_0}{\eta + \eta_0} \approx \frac{\sqrt{\mu_0} - \sqrt{\mu_0 \epsilon_0}}{\sqrt{\mu_0} + \sqrt{\mu_0 \epsilon_0}} = \frac{\sqrt{\epsilon_0} - \sqrt{\epsilon}}{\sqrt{\epsilon_0} + \sqrt{\epsilon}} = \frac{1 - \sqrt{\epsilon_r}}{1 + \sqrt{\epsilon_r}} = \frac{1 - \sqrt{81}}{1 + \sqrt{81}} = \frac{-9}{10} = -0.9$$

$$E^r = -0.9(\hat{a}_y + j\hat{a}_z) E_0 e^{-j\beta_0 x}$$

$$H^r = -\frac{0.9}{377}(\hat{a}_z - j\hat{a}_y) E_0 e^{-j\beta_0 x} = 2.122 \times 10^{-3} (j\hat{a}_y - \hat{a}_z) E_0 e^{-j\beta_0 x}$$

cont'd.

5.33 Cont'd.

d. Circular, CCW

$$e. T = \frac{2\eta}{\eta + \eta_0} \approx \frac{2\sqrt{\frac{\epsilon_0}{\epsilon}}}{\sqrt{\frac{\epsilon_0}{\epsilon}} - \sqrt{\frac{\epsilon_0}{\epsilon_y}}} = 2 \frac{\frac{1}{\sqrt{\epsilon}}}{\frac{1}{\sqrt{\epsilon}} - \frac{1}{\sqrt{\epsilon_y}}} = 2 \frac{\sqrt{\epsilon_0}}{\sqrt{\epsilon_0} - \sqrt{\epsilon_y}} = \frac{2}{1 + \sqrt{\epsilon_y}} = \frac{2}{1 + 9} = 0.2$$

$$\underline{E}^t = 0.2 (\hat{a}_y + j \hat{a}_z) e^{+j\beta x} = 0.2 (\hat{a}_y + j \hat{a}_z) e^{+(\alpha + j\beta)x}$$

$$\underline{H}^t = \frac{0.2}{377} \frac{(j \hat{a}_y - \hat{a}_z) \epsilon_0}{\sqrt{81}} e^{+(\alpha + j\beta)x} = (9) 5.3 \times 10^{-4} (j \hat{a}_y - \hat{a}_z) \epsilon_0 e^{+(\alpha + j\beta)x} = 47.7454 \times 10^{-4} (j \hat{a}_y - \hat{a}_z) \epsilon_0 e^{+(\alpha + j\beta)x}$$

$$\alpha \approx \frac{\sigma}{2} \sqrt{\frac{\epsilon_0}{\epsilon}} = \frac{10^{-1}}{2} \left(\frac{377}{9} \right) = 2.0944 \text{ N/m}$$

$$\beta \approx \omega \sqrt{\mu_0 \epsilon} = \beta_0 \sqrt{81} = 9 \beta_0 = 9 \omega \sqrt{\mu_0 \epsilon_0} = \frac{9(2\pi \times 10^9)}{3 \times 10^8} = 9(20.944) = 188.5 \text{ rad/m}$$

f. Circular, CW

$$g. \frac{S^r}{S^t} = |\Gamma|^2 = |-0.8|^2 = 0.64 = 64\%$$

$$\frac{S^t}{S^r} = 1 - |\Gamma|^2 = (1 - |-0.8|^2) = 0.36 = 36\%$$

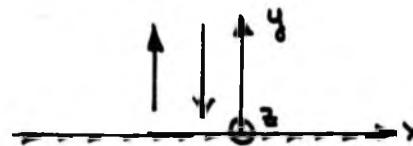
- 5.34** a. Because the reflection coefficient is -1 , the polarization is also left-hand circularly polarized.

b. $\underline{E}^i = (\hat{a}_x - j \hat{a}_z) e^{+j\beta y} \text{ (CW/RH)}$

$\underline{E}^r = -(\hat{a}_x - j \hat{a}_z) e^{-j\beta y} \text{ (CCW/LH)}$

$$V = \cos\left(\frac{P_w P_a}{2}\right) = \cos\left(\frac{90}{2}\right) = \cos(90^\circ) = 0 \quad \begin{cases} \text{One @ north pole, the} \\ \text{other @ south pole} \end{cases} \Rightarrow 2y = 180^\circ$$

c. $V = \cos\left(\frac{0}{2}\right) = \cos(0^\circ) = \cos(0^\circ) = 1 \quad \text{Both @ north pole} \Rightarrow 2y = 0^\circ$



5.35 $\frac{\sigma}{\omega \epsilon} = \frac{4}{2\pi(10^9)} \left(\frac{81 \times 10^{-9}}{\frac{36\pi}{25}} \right) = \frac{4(10)}{81} \times 10^2 = \frac{8}{9} \times 10^2 > 1 \Rightarrow \text{Good conductor}$

$$\eta \approx \sqrt{\frac{\omega \mu}{2\sigma}} (1+j) = \sqrt{\frac{\pi \times 10^9 (4\pi \times 10^{-7})}{2}} (1+j) = \pi(1+j) = \sqrt{2}\pi 145^\circ$$

$$\Gamma = \frac{\eta - \eta_0}{\eta + \eta_0} = \frac{\pi + \pi j}{\pi + j\pi + 377} = \frac{-373.858 + j\pi}{380.142 + j\pi} = \frac{373.871 179.5185}{380.165 0.4725} = 0.9835 179.045^\circ$$

Because the intrinsic impedance is very small and the reflection coefficient is almost -1 , the results of this problem are almost identical to those of problem 5.34.

5.35 According to (5-20a)-(5-22d)

$$\underline{E}_{\parallel}^l = (\hat{a}_x \cos \theta_i - \hat{a}_z \sin \theta_i) E_0 e^{-j\beta_i(x \sin \theta_i + z \cos \theta_i)}$$

$$\underline{H}_{\parallel}^l = \hat{a}_y \frac{E_0}{\eta_1} e^{-j\beta_i(x \sin \theta_i + z \cos \theta_i)}$$

$$\underline{E}_{\parallel}^r = (\hat{a}_x \cos \theta_r + \hat{a}_z \sin \theta_r) \Gamma_{\parallel}^b E_0 e^{-j\beta_i(x \sin \theta_r - z \cos \theta_r)}$$

$$\underline{H}_{\parallel}^r = -\hat{a}_y \frac{\Gamma_{\parallel}^b E_0}{\eta_2} e^{-j\beta_i(x \sin \theta_r - z \cos \theta_r)}$$

$$\underline{E}_{\parallel}^t = (\hat{a}_x \cos \theta_t - \hat{a}_z \sin \theta_t) T_{\parallel}^b E_0 e^{-j\beta_t(x \sin \theta_t + z \cos \theta_t)}$$

$$\underline{H}_{\parallel}^t = \hat{a}_y \frac{T_{\parallel}^b E_0}{\eta_2} e^{-j\beta_t(x \sin \theta_t + z \cos \theta_t)}$$

When the incidence angle is greater than the critical angle, then the exponentials for the transmitted fields can be written as those given by (5-43a)-(5-43b) for the perpendicular polarization. Thus the transmitted fields take the form of

$$\underline{E}_{\parallel}^t = (\hat{a}_x \cos \theta_t - \hat{a}_z \sin \theta_t) T_{\parallel}^b E_0 e^{-\alpha_e z} e^{-j\beta_e x}$$

$$\underline{H}_{\parallel}^t = \hat{a}_y \frac{T_{\parallel}^b E_0}{\eta_2} e^{-\alpha_e z} e^{-j\beta_e x}$$

where $\alpha_e = \omega \sqrt{\mu_1 \epsilon_1 \sin^2 \theta_i - \mu_2 \epsilon_2} \Big|_{\theta_i > \theta_c}$

$$\beta_e = \omega \sqrt{\mu_1 \epsilon_1 \sin \theta_i} \Big|_{\theta_i > \theta_c}$$

The reflection and transmission coefficients (Γ_{\parallel}^b and T_{\parallel}^b) of (5-24c) and (5-24d) reduce, respectively, to

$$\left| \Gamma_{\parallel}^b \right| = \frac{-\sqrt{\frac{\mu_1}{\epsilon_1}} \cos \theta_i + j \sqrt{\frac{\mu_2}{\epsilon_2}} \cos \theta_t}{\sqrt{\frac{\mu_1}{\epsilon_1}} \cos \theta_i + j \sqrt{\frac{\mu_2}{\epsilon_2}} \cos \theta_t} = \frac{-\sqrt{\frac{\mu_1}{\epsilon_1}} \cos \theta_i + j \sqrt{\frac{\mu_2}{\epsilon_2}} \sqrt{\frac{\mu_1 \epsilon_1}{\mu_2 \epsilon_2} \sin^2 \theta_i - 1}}{\sqrt{\frac{\mu_1}{\epsilon_1}} \cos \theta_i + j \sqrt{\frac{\mu_2}{\epsilon_2}} \sqrt{\frac{\mu_1 \epsilon_1}{\mu_2 \epsilon_2} \sin^2 \theta_i - 1}}$$

$$= |\Gamma_{\parallel}^b| e^{j\psi_{\parallel}}$$

where $|\Gamma_{\parallel}^b| = 1$

$$\psi_{\parallel} = \pi - \Psi_0 = \pi - \Psi_0, \Psi_0 = \tan^{-1} \left(\frac{x_{\parallel}}{R_{\parallel}} \right), x_{\parallel} = \sqrt{\frac{\mu_2}{\epsilon_2}} \sqrt{\frac{\mu_1 \epsilon_1}{\mu_2 \epsilon_2} \sin^2 \theta_i - 1}, R_{\parallel} = \sqrt{\frac{\mu_1}{\epsilon_1}} \cos \theta_i$$

cont'd.

5.35 cont'd.

$$T_{11}^b = \frac{2\sqrt{\frac{\mu_2}{\epsilon_2}} \cos \theta_i}{\sqrt{\frac{\mu_1}{\epsilon_1}} \cos \theta_i + \sqrt{\frac{\mu_2}{\epsilon_2}} \cos \theta_t} = \frac{2\sqrt{\frac{\mu_2}{\epsilon_2}} \cos \theta_i}{\sqrt{\frac{\mu_1}{\epsilon_1}} \cos \theta_i + j\sqrt{\frac{\mu_2}{\epsilon_2}} \sqrt{\frac{\mu_1 \epsilon_1}{\mu_2 \epsilon_2} \sin^2 \theta_i - 1}} = |T_{11}^b| e^{-j\psi_0}$$

$$|T_{11}^b| = \frac{2 R_{11}}{\sqrt{R_{11}^2 + x_i^2}}$$

$$\begin{aligned} S_{av}^l &= \frac{1}{2} \operatorname{Re}(\underline{E}^l \times \underline{H}^{l*}) = \frac{1}{2} \operatorname{Re}[(\hat{a}_x E_x^l + \hat{a}_z E_z^l) \times \hat{a}_y H_y^{l*}] = \frac{1}{2} \operatorname{Re}[\hat{a}_z E_x^l H_y^{l*} - \hat{a}_x E_z^l H_y^{l*}] \\ &= \frac{1}{2} \left\{ \hat{a}_z \left| \frac{E_0}{\eta_1} \right|^2 \cos \theta_i + \hat{a}_x \left| \frac{E_0}{\eta_1} \right|^2 \sin \theta_i \right\} = \frac{|E_0|^2}{2\eta_1} \left\{ \hat{a}_z \cos \theta_i + \hat{a}_x \sin \theta_i \right\} \end{aligned}$$

$$\begin{aligned} S_{av}^r &= \frac{1}{2} \operatorname{Re}(\underline{E}^r \times \underline{H}^r) = \frac{1}{2} \operatorname{Re}[(\hat{a}_x E_x^r + \hat{a}_z E_z^r) \times (\hat{a}_y H_y^r)] = \frac{1}{2} \operatorname{Re}[\hat{a}_z E_x^r H_y^r - \hat{a}_x E_z^r H_y^r] \\ &= \frac{1}{2} \left\{ \hat{a}_z \left| \frac{E_0}{\eta_1} \right|^2 \cos \theta_r + \hat{a}_x \left| \frac{E_0}{\eta_1} \right|^2 \sin \theta_r \right\} = \frac{|E_0|^2}{2\eta_1} \left\{ -\hat{a}_z \cos \theta_r + \hat{a}_x \sin \theta_r \right\} \end{aligned}$$

$$\begin{aligned} S_{av}^t &= \frac{1}{2} \operatorname{Re}\{ \underline{E}^t \times \underline{H}^{t*} \} = \frac{1}{2} \operatorname{Re}\{ (\hat{a}_x E_x^t + \hat{a}_z E_z^t) \times \hat{a}_y H_y^{t*} \} = \frac{1}{2} \operatorname{Re}[\hat{a}_z E_x^t H_y^{t*} - \hat{a}_x E_z^t H_y^{t*}] \\ &= \frac{\operatorname{Re}\{ \hat{a}_z \left| T_{11}^b \right|^2 \frac{|E_0|^2}{\eta_2} \cos \theta_t + \hat{a}_x \left| T_{11}^b \right|^2 \frac{|E_0|^2}{\eta_2} \sin \theta_t \}}{2\eta_2} = \frac{|T_{11}^b|^2 |E_0|^2}{2\eta_2} \operatorname{Re}[\hat{a}_z \cos \theta_t + \hat{a}_x \sin \theta_t] e^{-2d_e z} \end{aligned}$$

Since $\cos \theta_t = \pm j \sqrt{\frac{\mu_1 \epsilon_1}{\mu_2 \epsilon_2} \sin^2 \theta_i - 1}$ $|_{\theta_i > \theta_c}$ = imaginary [see (5-42b)]

$$\sin \theta_t = \sqrt{\frac{\mu_1 \epsilon_1}{\mu_2 \epsilon_2} \sin^2 \theta_i} |_{\theta_i > \theta_c}$$

then

$$S_{av}^t = \hat{a}_x \sqrt{\frac{\mu_1 \epsilon_1}{\mu_2 \epsilon_2} \sin^2 \theta_i} \left| \frac{|T_{11}^b|^2 |E_0|^2}{2\eta_2} \right| e^{-2d_e z} |_{\theta_i > \theta_c}$$

5.37

$$E_{r2} = 4, \theta_i = 60^\circ$$

(a) $Z_w = \eta_0 = 377 \text{ ohms}$ because plane wave

$Z_r = \eta_0 = 377 \text{ ohms}$ because plane wave

$$Z_t^t = \eta_0 \frac{1}{2} = \frac{\eta_0}{\sqrt{8\pi}} = \frac{377}{2} = 188.5 \text{ because plane wave}$$

$$(b) Z_{z0}^t(\text{directional}) = \frac{\eta_0}{\cos \theta_i} = \frac{377}{\cos(60^\circ)} = \frac{377}{0.5} = 754 \text{ ohms}$$

$$Z_{z2}^t(\text{directional}) = \frac{\eta_0}{\cos \theta_t}$$

$$\beta_0 \sin \theta_i = \beta_2 \sin \theta_t \Rightarrow \theta_t = \sin^{-1} \left[\frac{\beta_0 \sin \theta_i}{\beta_2} \right] = \sin^{-1} \left[\frac{1}{\sqrt{8\pi}} \sin \theta_i \right]$$

$$= \sin^{-1} \left[\frac{1}{2} \sin(60^\circ) \right] = \sin^{-1} \left[\frac{0.866}{2} \right]$$

$$\theta_t = \sin^{-1}(0.433) = 25.669^\circ$$

$$Z_{z2}^t = \frac{\eta_0}{\cos \theta_t} = \frac{\eta_0 / \sqrt{8\pi}}{\cos(25.669^\circ)} = \frac{188.5}{0.9014} = 209.12 \text{ ohms}$$

$$(c) \Gamma_{in} = \frac{Z_{z2}^t - Z_{z0}^t}{Z_{z2}^t + Z_{z0}^t} = \frac{209.12 - 754}{209.12 + 754} = \frac{544.88}{963.12} = -0.5657$$

$$|\Gamma_{in}| = 0.5657, \quad \angle \Gamma_{in} = 180^\circ$$

OR

$$\begin{aligned} \Gamma_L &= \frac{\cos \theta_i - \sqrt{8\pi} \sqrt{1 - \frac{1}{8\pi} \sin^2 \theta_i}}{\cos \theta_i + \sqrt{8\pi} \sqrt{1 - \frac{1}{8\pi} \sin^2 \theta_i}} = \frac{\cos(60^\circ) - \sqrt{4} \sqrt{1 - \frac{1}{4} \sin^2(60^\circ)}}{\cos(60^\circ) + \sqrt{4} \sqrt{1 - \frac{1}{4} \sin^2(60^\circ)}} \\ &= \frac{0.5 - 2\sqrt{1 - 0.75/4}}{0.5 + 2\sqrt{1 - 0.75/4}} = \frac{0.5 - 2(0.9011388)}{0.5 + 2(0.9011388)} = \frac{0.5 - 1.802776}{0.5 + 1.802776} \end{aligned}$$

$$\Gamma_L = -\frac{1.302776}{2.302776} = -0.5657 \Rightarrow |\Gamma_L| = 0.5657, \quad \angle \Gamma_L = 180^\circ$$

$$(d) SWR = \frac{1 + |\Gamma_{in}|}{1 - |\Gamma_{in}|} = \frac{1 + 0.5657}{1 - 0.5657} = \frac{1.5657}{0.4343} = 3.605$$

5.38

(a) Snell's Law of Refraction

$$\beta_0 \sin\theta_i = \beta \sin\theta_t$$

 ϵ_0, μ_0 ϵ_r, μ_r η_0 η_r

$$\sin\theta_t = \frac{\beta_0}{\beta} \sin\theta_i = \frac{w\sqrt{\epsilon_0}}{w\sqrt{\epsilon_r}} \sin\theta_i$$

$$= \frac{\sqrt{\epsilon_0}}{\sqrt{\epsilon_0 \epsilon_r}} \sin\theta_i$$

$$\sin\theta_t = \frac{1}{\sqrt{\epsilon_r}} \sin\theta_i \quad \text{for } \epsilon_r \gg 1 \approx 0$$



$\theta_t \approx \sin^{-1}(0) \approx 0^\circ$ for both I and II polarizations

(b)

$$\Gamma_I = \frac{\eta \cos\theta_i - \eta_0 \cos\theta_t}{\eta \cos\theta_i + \eta_0 \cos\theta_t} \quad \theta_t = 0 \quad \eta \cos\theta_i - \eta_0 = 0 \Rightarrow \eta \cos\theta_i - \eta_0 = 0$$

$$\theta_i = \cos^{-1}\left(\frac{\eta_0}{\eta}\right) = \cos^{-1}\left(\frac{\eta_0}{\eta_0 \sqrt{\frac{\epsilon_r}{\epsilon_r}}}\right) = \cos^{-1}\left(\sqrt{\frac{\epsilon_r}{\epsilon_r}}\right)$$

exists if $\epsilon_r > \epsilon_r$

$$\Gamma_{II} = \frac{-\eta_0 \cos\theta_i + \eta \cos\theta_t}{\eta_0 \cos\theta_i + \eta \cos\theta_t} \quad \theta_t = 0 \quad -\eta_0 \cos\theta_i + \eta = 0 \quad \eta_0 \cos\theta_i + \eta$$

$$-\eta_0 \cos\theta_i + \eta = 0 \Rightarrow \cos\theta_i = \frac{\eta}{\eta_0}$$

$$\theta_i = \cos^{-1}\left(\frac{\eta}{\eta_0}\right) = \cos^{-1}\left(\frac{\eta_0 \sqrt{\frac{\epsilon_r}{\epsilon_r}}}{\eta_0}\right) = \cos^{-1}\left(\sqrt{\frac{\epsilon_r}{\epsilon_r}}\right)$$

exists if $\epsilon_r > \epsilon_r$

5.39 Due to the symmetry of the problem, both dielectric slabs have the same thickness and dielectric constant.

(a) To match the slab on each side,

we need $\lambda/4$ impedance transformers. (air)

$$\text{Thus } d = \lambda/4$$

(b) The intrinsic impedance of the $\lambda/4$ slabs is

$$\eta_i = \sqrt{\eta_0 \epsilon_{ri}} = \sqrt{\frac{\eta_0}{\epsilon_{ri}} \epsilon_0} = \frac{\eta_0}{\sqrt{1.6}} = 0.7906 \eta_0$$

$$\eta_i = 298 \text{ ohms}$$

$$(c) \eta_i = \sqrt{\frac{\eta_0}{\epsilon_{ri} \epsilon_0}} = \frac{1}{\sqrt{\epsilon_{ri}}} = \frac{377}{\sqrt{1.6}} = 298$$

$$\sqrt{\epsilon_{ri}} = 377/298 = 1.2651$$

$$\epsilon_{ri} = 1.6$$

(d) The wavelength in the slabs is

$$\lambda = \frac{\lambda_0}{\sqrt{\epsilon_{ri}}} = \frac{30 \times 10^9 / 10 \times 10^9}{\sqrt{1.6}} = \frac{3}{\sqrt{1.6}} = \frac{3}{1.2649} = 2.3717 \text{ cm}$$

Therefore the thickness of each slab is

$$d = \frac{\lambda}{4} = \frac{2.3717}{4} = 0.5929 \text{ cm}$$

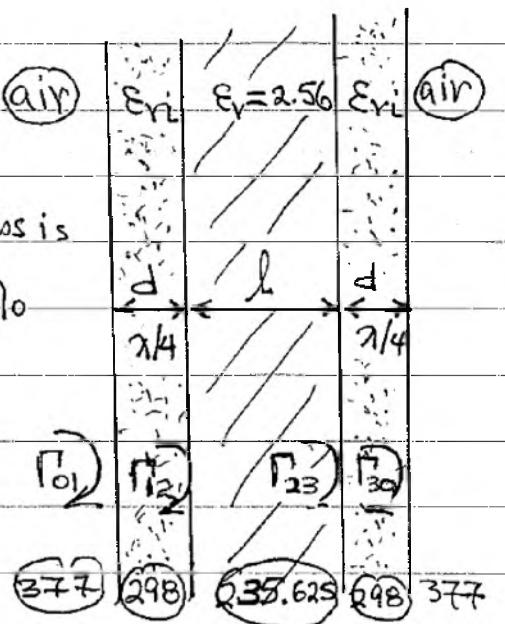
(e) From above

$$\eta_i = 298 \text{ ohms}$$

(f) Since the $\lambda/4$, the total input reflection coefficient is zero; thus the VSWR = 1

$$\Gamma_{01} = \frac{\eta_i - \eta_0}{\eta_i + \eta_0} = \frac{298 - 377}{298 + 377} = -0.117; \quad \Gamma_{12} = \frac{\eta - \eta_i}{\eta + \eta_i} = \frac{235.625 - 298}{235.625 + 298} = -0.1169$$

$$\Gamma_{23} = -\Gamma_{12} = +0.1169, \quad \Gamma_{30} = -\Gamma_{01} = 0.117$$



$$\left| \frac{\Gamma_{12}(1+e^{-j\pi x})}{1+(\Gamma_{12})^2 e^{-j\pi x}} \right| = \rho_m$$

$$x = f / f_o; \quad f_o = 10 \text{ GHz}$$

$$\Gamma_{12} = \frac{1-\sqrt{2}}{1+\sqrt{2}} = \frac{-0.41421}{2.41421} = -0.1716$$

Using an iterative program or solving it otherwise,
the following solutions are obtained.

.....

1. $\rho_m = 0.05, \Gamma_{12} = -0.1726,$

The lower (f_l) and upper (f_u) frequencies, and bandwidth (Δf),
due to frequency response symmetry about f_o , are:

$$f_l = 9.095 \text{ GHz}; \quad f_u = 10.905 \text{ GHz}$$

$$\Delta f = (f_u - f_l) = (10.905 - 9.095) \text{ GHz} = 1.81 \text{ GHz}$$

.....

2. $\rho_m = 0.10, \Gamma_{12} = -0.1726$

The lower (f_l) and upper (f_u) frequencies, and bandwidth (Δf),
due to frequency response symmetry about f_o , are:

$$f_l = 8.165 \text{ GHz}; \quad f_u = 11.835 \text{ GHz}$$

$$\Delta f = (f_u - f_l) = (11.835 - 8.165) \text{ GHz} = 3.67 \text{ GHz}$$

5.41

Exact Transmission Line Model:

$$\Gamma_{in}(z = -d^-) = \frac{Z_{in}(z = -d^+) - \eta_1}{Z_{in}(z = -d^+) + \eta_1} = \frac{N}{D}$$

$$N = \eta_2 \left[(\eta_3 + \eta_2) + (\eta_3 - \eta_2) e^{-j2\beta_2 d} \right] \\ - \eta_1 \left[(\eta_3 + \eta_2) - (\eta_3 - \eta_2) e^{-j2\beta_2 d} \right]$$

$$D = \eta_2 \left[(\eta_3 + \eta_2) + (\eta_3 - \eta_2) e^{-j2\beta_2 d} \right] \\ + \eta_1 \left[(\eta_3 + \eta_2) - (\eta_3 - \eta_2) e^{-j2\beta_2 d} \right]$$

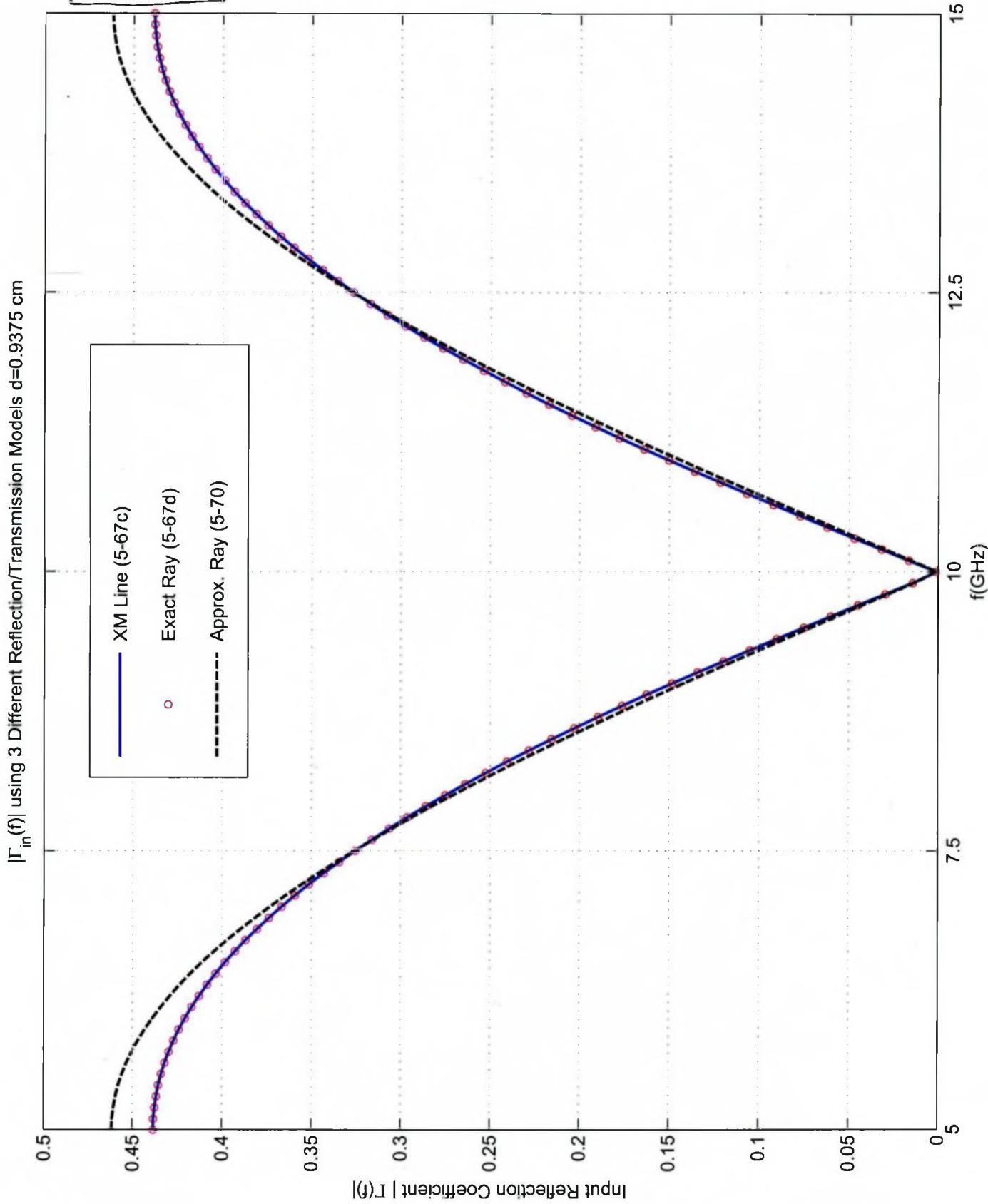
Exact Ray-Tracing Model:

$$\Gamma_{in}(z = -d^-) = \frac{\Gamma_{12} + \Gamma_{23} e^{-j2\beta_2 d}}{1 + \Gamma_{12} \Gamma_{23} e^{-j2\beta_2 d}}$$

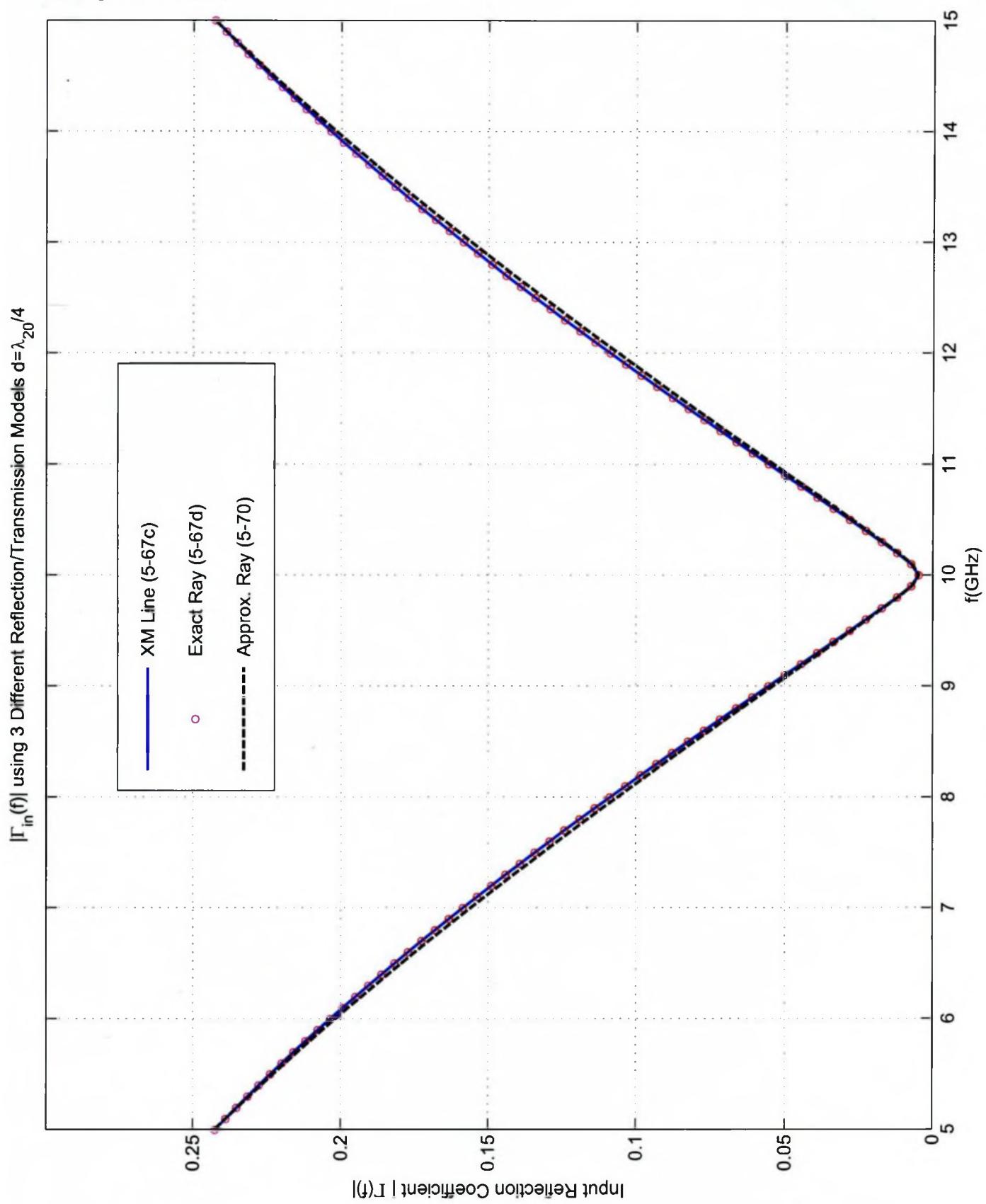
Approximate Ray-Tracing Model:

$$\Gamma_{in}(z = -d^-) \underset{|\Gamma_{23}| \ll 1}{\approx} \Gamma_{12} + \Gamma_{23} e^{-j2\beta_2 d}$$

5.41 cont'd



5.41 cont'd



5.42

$$(a) d = \lambda_1 / 4$$

$$(b) \eta_1 = \sqrt{\eta_0 \eta_2} = \left\{ \sqrt{\frac{\mu_0}{\epsilon_0}} \left(\frac{1}{4} \sqrt{\frac{\mu_0}{\epsilon_0}} \right) \right\}^{1/2}$$

$$\eta_0 = \frac{\mu_0}{\epsilon_0}$$

$$\epsilon_r \epsilon_0 \mu_0$$

$$16 \epsilon_0 \mu_0$$

$$\eta_1 = \sqrt{\frac{\mu_0}{\epsilon_r \epsilon_0}}$$

$$\eta_2 = \sqrt{\frac{\mu_0}{16 \epsilon_0}} = \frac{1}{4}$$

$$\eta_1 = \frac{1}{2} \sqrt{\frac{\mu_0}{\epsilon_0}} = \sqrt{\frac{\mu_0}{\epsilon_0}} \frac{1}{\sqrt{\epsilon_r}}$$

$$\eta_0 = \sqrt{\frac{\mu_0}{\epsilon_0}}$$

$$\eta_1 = \frac{1}{\sqrt{\epsilon_r}} \sqrt{\frac{\mu_0}{\epsilon_0}}$$

$$\eta_2 = \frac{1}{4} \sqrt{\frac{\mu_0}{\epsilon_0}}$$

$$\frac{1}{\sqrt{\epsilon_r}} = \frac{1}{2} \Rightarrow \boxed{\epsilon_r = 4}$$

$$(c) \lambda_1 = \frac{v}{f} = \frac{v_0 \sqrt{\epsilon_r}}{1 \times 10^9} = \frac{3 \times 10^10 / \sqrt{4}}{1 \times 10^9}$$

$$\lambda_1 = \frac{3}{2} \times 10 = \frac{30}{2} = 15 \text{ cm}$$

$$d = \frac{\lambda_1}{4} = \frac{15}{4} = \boxed{3.75 \text{ cm}}$$

5.43

$$a. \Gamma_{01} = \frac{\eta_1 - \eta_0}{\eta_1 + \eta_0} = \frac{\frac{\eta_0}{2} - \eta_0}{\frac{\eta_0}{2} + \eta_0} = \frac{\frac{1}{2} - 1}{\frac{1}{2} + 1} = -\frac{1}{3}$$

$$\boxed{\Gamma_{01} = -\frac{1}{3}}$$

$$\begin{array}{c|c|c|c} & \lambda_1/4 & \lambda_2/4 & \\ \epsilon_0 = 1 & \epsilon_1 = 4 & \epsilon_2 = 4 & \epsilon_3 = 1 \end{array}$$

$$\Gamma_{12} = \frac{\eta_2 - \eta_1}{\eta_2 + \eta_1} = \frac{\frac{\eta_0}{2} - \frac{\eta_0}{2}}{\frac{\eta_0}{2} + \frac{\eta_0}{2}} = 0$$

$$\boxed{\Gamma_{12} = 0}$$

$$\begin{array}{c|c|c|c} \Gamma_{01} & \Gamma_{12} & \Gamma_{23} & \\ \#0 & \#1 & \#2 & \#3 \end{array}$$

$$\Gamma_{23} = \frac{\eta_3 - \eta_2}{\eta_3 + \eta_2} = \frac{\eta_0 - \eta_0/2}{\eta_0 + \eta_0/2} = \frac{1 - 1/2}{1 + 1/2} = \frac{1}{3}$$

$$\boxed{\Gamma_{23} = \frac{1}{3}}$$

(b)

$$\Gamma_{in} = \Gamma_{01} + \Gamma_{12} e^{-j2\beta d_1} + \Gamma_{23} e^{-j(2\beta d_1 + 2\beta d_2)}$$

$$= \Gamma_{01} + \Gamma_{12} e^{-j\pi} + \Gamma_{23} e^{-j2(\frac{\pi}{2} + \frac{\pi}{2})} = \Gamma_{01} + \Gamma_{12} e^{-j\pi} + \Gamma_{23} e^{-j2\pi}$$

$$\Gamma_{in} = -\frac{1}{3} + 0 + \frac{1}{3} = 0$$

$$\boxed{\Gamma_{in} = 0}$$

5.44

Using one of the transmission line model

$$(a) Z_{in}(z=0^+) = 0 \text{ PEC}$$

$$\Gamma_{in}(z=0^-) = \frac{Z_m(z=0^+) - \eta}{Z_m(z=0^+) + \eta} = \frac{0 - \eta}{0 + \eta} = -1$$

$$\begin{aligned}\Gamma_{in}(z=-d^+) &= \Gamma_{in}(z=0^-) e^{-j2\beta d} = -1 e^{-j2(\frac{\pi}{4})(\frac{d}{\lambda})} \\ &= -1 e^{-j\pi/2} = -1(j) = +j \quad (\Gamma_{in}(z=-d^+))\end{aligned}$$

$$Z_{in}(z=-d^+) = \eta \left[\frac{1 + \Gamma_{in}(z=-d^+)}{1 - \Gamma_{in}(z=-d^+)} \right]$$

$$= \eta \left[\frac{1+j}{1-j} \right] = \eta \left[\frac{\sqrt{2} [45^\circ]}{\sqrt{2} [-45^\circ]} \right] = \eta |1+0| = \eta e^{j\pi/2} = j\eta$$

$$Z_{in}(z=-d^+) = j\eta = j \frac{1}{2} \eta_0$$

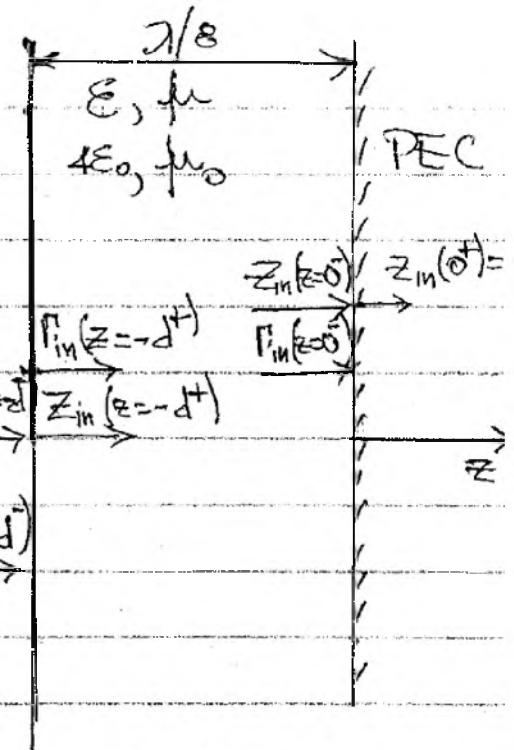
$$\Gamma_{in}(z=-d^+) = \Gamma_{in}(\text{total}) = \frac{Z_m(z=-d^+) - \eta_0}{Z_{in}(z=-d^+) + \eta_0} = \frac{j\frac{1}{2}\eta_0 - \eta_0}{j\frac{1}{2}\eta_0 + \eta_0} = \frac{j\frac{1}{2}-1}{j\frac{1}{2}+1}$$

$$= -\frac{1-j\frac{1}{2}}{1+j\frac{1}{2}} = -\frac{2-j}{2+j} = -\frac{\sqrt{5} [-26.565^\circ]}{\sqrt{5} [+26.565^\circ]} = -1 + 53.13$$

$$= -1(0.6 - j0.8) = -0.6 + j0.8 = 1 [180^\circ - 53.13^\circ]$$

$$\Gamma_{in}(\text{total}) = 1 [126.87^\circ] \Rightarrow |\Gamma_{in}(\text{total})| = 1$$

$$(b) SWR = \frac{1 + |\Gamma_{in}(\text{total})|}{1 - |\Gamma_{in}(\text{total})|} = \frac{1+1}{1-1} = \infty$$



alternate

(a) Using the other transmission line model or ray-tracing model

$$\Gamma(z=-d^-) = \frac{\Gamma_{12} + \Gamma_{23} e^{-j2\beta_2 d}}{1 + \Gamma_{12}\Gamma_{23} e^{-j2\beta_2 d}}$$

$$\Gamma_{12} = \frac{\eta_2 - \eta_1}{\eta_2 + \eta_1} = \frac{\frac{1}{2}\eta_0 - \eta_0}{\frac{1}{2}\eta_0 + \eta_0} = \frac{-\frac{1}{2}}{\frac{1}{2} + 1}$$

$$\Gamma_{12} = -\frac{1}{3} = -\frac{1}{3}$$

$$\Gamma_{23} = \frac{\eta_3 - \eta_2}{\eta_3 + \eta_2} = \frac{0 - \eta_2}{0 + \eta_2} = -1$$

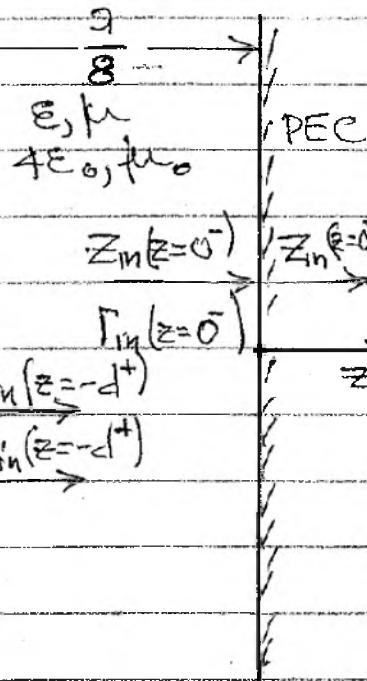
$$2\beta_2 d = 2 \left(\frac{2\pi}{\lambda_2} \right) \left(\frac{\eta_2}{8} \right) = \frac{\pi}{2}$$

$$\begin{aligned} \Gamma_{in}(z=-d^-) &= \frac{-\frac{1}{3} - 1}{1 + \left(-\frac{1}{3}\right)(-1)e^{-j\pi/2}} e^{j\pi/2} = \frac{\frac{1}{3} + 1}{1 + \frac{1}{3}e^{-j\pi/2}} e^{-j\pi/2} = \frac{\frac{4}{3}}{\frac{4}{3} - j} \\ &= \frac{1-j\sqrt{3}}{3-j} = \frac{\sqrt{10}[-71.565]}{\sqrt{10}[-18.43495]} = -11-53.13^\circ \end{aligned}$$

$$\Gamma_{in}(z=-d^-) = -(0.6 - j0.8) = -0.6 + j0.8 = 1 [180^\circ - 53.13^\circ]$$

$$\Gamma_{in}(\text{total}) = 1 [126.87^\circ] \Rightarrow |\Gamma_{in}(\text{total})| = 1, |\Gamma_{in}(\text{total})| = 126.87^\circ$$

$$(b) \quad \text{SWR} = \frac{1 + |\Gamma_{in}(\text{total})|}{1 - |\Gamma_{in}(\text{total})|} = \frac{1+1}{1-1} = \infty$$



5.45

$$f_0 = 2 \text{ GHz} \quad N=2$$

$$\frac{\Delta f}{f_0} = 0.5 \quad \eta_L = 377$$

$$\eta_0 = 251.33$$

$$\begin{array}{c|c|c|c} & \Gamma_0 & \Gamma_1 & \Gamma_2 \\ \epsilon_{r_0} = 2.25 & \epsilon_1 = ? & \epsilon_2 = ? & \epsilon_L = 1 \\ \eta_0 = 251.33 & \eta_1 = ? & \eta_2 = ? & \eta_L = 377 \end{array}$$

b. $\frac{\Delta f}{f_0} = 0.5 = 2 - \frac{4}{\pi} \cos^{-1} \left[\frac{\Gamma_m}{(\eta_L - \eta_0) / (\eta_L + \eta_0)} \right]^{1/2}$

$$\Gamma_m = \frac{\eta_L - \eta_0}{\eta_L + \eta_0} \left\{ \cos \left[\frac{\pi}{4} \left(2 - \frac{\Delta f}{f_0} \right) \right] \right\}^2 = \frac{377 - 251.33}{377 + 251.33} \left\{ \cos \left[\frac{\pi}{4} \left(2 - 0.5 \right) \right] \right\}^2 = 0.0293$$

b. $\Gamma_n = 2^{-N} \frac{\eta_L - \eta_0}{\eta_L + \eta_0} \frac{N!}{(N-n)!n!}$

$$n=0: \Gamma_0 = \frac{1}{4} \left(\frac{377 - 251.33}{377 + 251.33} \right) \frac{2!}{2!0!} = \frac{1}{4}(0.2) = 0.05$$

$$n=1: \Gamma_1 = \frac{1}{4}(0.2) \frac{2!}{1!1!} = 2(0.05) = 0.1$$

$$n=2: \Gamma_2 = \frac{1}{4}(0.2) \frac{2!}{(2-2)!2!} = 0.05$$

c. $\Gamma_0 = 0.05 = \frac{\eta_1 - \eta_0}{\eta_1 + \eta_0} \Rightarrow \eta_1 = \eta_0 \frac{1 + \Gamma_0}{1 - \Gamma_0} = 251.33 \frac{1 + 0.05}{1 - 0.05} = 277.786 \Rightarrow \epsilon_{r_1} = 1.842$

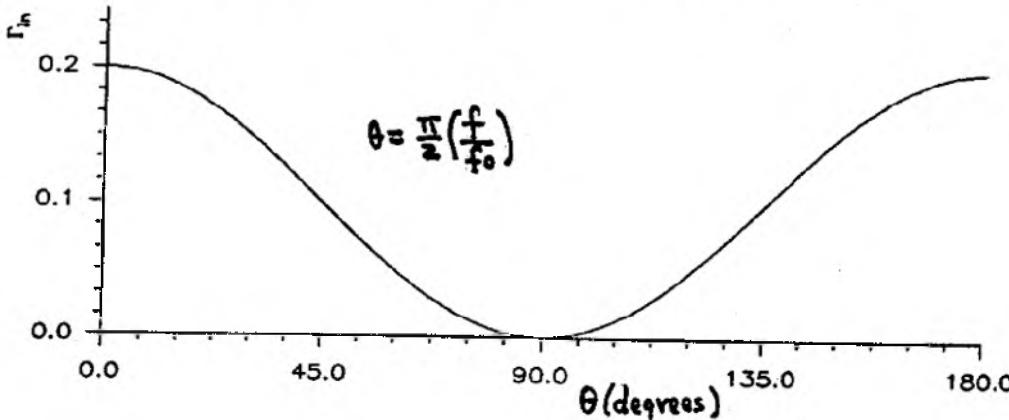
$$d_{12} = \frac{2\lambda_0}{4} = \frac{30 \times 10^9}{4(2 \times 10^9)\sqrt{1.842}} = 2.763 \text{ cm}$$

$$\Gamma_1 = 0.1 = \frac{\eta_2 - \eta_1}{\eta_2 + \eta_1} \Rightarrow \eta_2 = \eta_1 \frac{1 + \Gamma_1}{1 - \Gamma_1} = 277.786 \frac{1 + 0.1}{1 - 0.1} = 339.516 \Rightarrow \epsilon_{r_2} = 1.233$$

$$d_{20} = \frac{2\lambda_0}{4} = \frac{30 \times 10^9}{4(2 \times 10^9)\sqrt{1.233}} = 3.377 \text{ cm}$$

d. $\frac{\Delta f}{f_0} = \frac{\Delta f}{2 \text{ GHz}} = 0.5 \Rightarrow \Delta f = 1 \text{ GHz} \Rightarrow f_L = f_0 - \frac{\Delta f}{2} = 1.5 \text{ GHz}, f_M = f_0 + \frac{\Delta f}{2} = 2.5 \text{ GHz}$

e. $|\Gamma_{in}(f)| = \frac{\eta_L - \eta_0}{\eta_L + \eta_0} \cos^2(\theta) = \frac{377 - 251.33}{377 + 251.33} \cos^2 \left[\frac{\pi}{2} \left(\frac{f}{f_0} \right) \right] = 0.2 \cos^2 \left(\frac{\pi}{2} \left(\frac{f}{f_0} \right) \right)$



5.46 $f_0 = 3 \text{ GHz}$ $N = 3$
 $\text{SWR}_{\max} = 1.1$ $\eta_L = 251.33$
 $\eta_0 = 125.67$

Γ_0	Γ_1	Γ_2	Γ_3
$\epsilon_{r0} = 9$	$\epsilon_{r1} = ?$	$\epsilon_{r2} = ?$	$\epsilon_{r3} = ?$
$\eta_0 = 125.67$	$\eta_1 = ?$	$\eta_2 = ?$	$\eta_3 = ?$

$\eta_0 = 125.67$ $\eta_1 = ?$ $\eta_2 = ?$ $\eta_3 = ?$ $\eta_L = 251.33$

a. $\text{SWR}_m = \frac{1+\Gamma_m}{1-\Gamma_m} \Rightarrow \Gamma_m = \frac{\text{SWR}_m - 1}{\text{SWR}_m + 1} = \frac{1.1 - 1}{1.1 + 1} = 0.0476$

$$\Delta f = 2 - \frac{4}{\pi} \cos^{-1} \left[\frac{\Gamma_m}{(\eta_L - \eta_0)(\eta_L + \eta_0)} \right]^{1/2} = 2 - \frac{4}{\pi} \cos^{-1} \left[\frac{0.0476}{(251.33 - 125.67)/(251.33 + 125.67)} \right]^{1/2}$$

$$\frac{\Delta f}{f_0} = 2 - \frac{4}{\pi} \cos^{-1}(0.1428)^{1/2} = 2 - \frac{4}{\pi} \cos^{-1}(0.5227) = 2 - 1.3 = 0.7 \Rightarrow \Delta f = 2.1 \text{ GHz}$$

$$f_L = f_0 - \frac{\Delta f}{2} = (3 - 1.05) \text{ GHz} = 1.95 \text{ GHz}, f_H = f_0 + \frac{\Delta f}{2} = (3 + 1.05) \text{ GHz} = 4.05 \text{ GHz}$$

b. $\Gamma_n = 2^{-n} \frac{\eta_L - \eta_0}{\eta_L + \eta_0} \frac{N!}{(N-n)!n!} = 2^{-3} \frac{251.33 - 125.67}{251.33 + 125.67} \frac{N!}{(N-n)!n!} = \frac{1}{24} \frac{N!}{(N-n)!n!}$

$$n=0: \Gamma_0 = \frac{1}{24} \frac{3!}{(3-0)!0!} = \frac{1}{24}$$

$$n=1: \Gamma_1 = \frac{1}{24} \frac{3!}{(3-1)!1!} = \frac{1}{24} (3) = \frac{1}{8}$$

$$n=2: \Gamma_2 = \frac{1}{24} \frac{3!}{(3-2)!2!} = \frac{1}{24} (3) = \frac{1}{8}$$

$$n=3: \Gamma_3 = \frac{1}{24} \frac{3!}{(3-3)!3!} = \frac{1}{24}$$

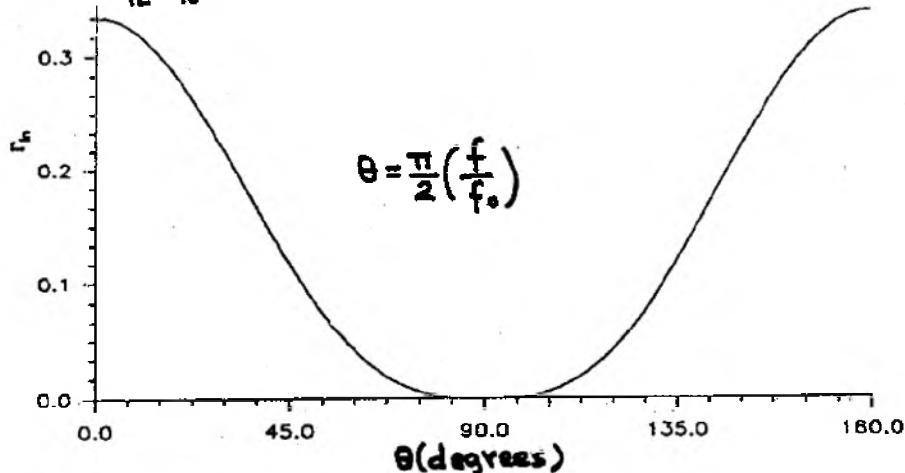
c. From Part a, $\Gamma_m = 0.0476$

d. $\Gamma_0 = \frac{1}{24} = \frac{\eta_L - \eta_0}{\eta_L + \eta_0} \Rightarrow \eta_1 = \eta_0 \frac{1 + \Gamma_0}{1 - \Gamma_0} = 125.67 \frac{1 + \frac{1}{24}}{1 - \frac{1}{24}} = 136.598, \epsilon_{r1} = 7.617, d_1 = 0.906 \text{ cm}$

$$\Gamma_1 = \frac{1}{8} = \frac{\eta_2 - \eta_1}{\eta_2 + \eta_1} \Rightarrow \eta_2 = \eta_1 \frac{1 + \Gamma_1}{1 - \Gamma_1} = 136.598 \frac{1 + \frac{1}{8}}{1 - \frac{1}{8}} = 175.626, \epsilon_{r2} = 4.608, d_2 = 1.165 \text{ cm}$$

$$\Gamma_2 = \frac{1}{8} = \frac{\eta_3 - \eta_2}{\eta_3 + \eta_2} \Rightarrow \eta_3 = \eta_2 \frac{1 + \Gamma_2}{1 - \Gamma_2} = 175.626 \frac{1 + \frac{1}{8}}{1 - \frac{1}{8}} = 225.804, \epsilon_{r3} = 2.788, d_3 = 1.497 \text{ cm}$$

e. $|\Gamma_{in}(f)| = \frac{\eta_L - \eta_0}{\eta_L + \eta_0} \cos^3(\theta) = \frac{251.33 - 125.67}{251.33 + 125.67} \cos^3\left(\frac{\pi}{2} \frac{f}{f_0}\right) = \frac{1}{3} \cos^3\left(\frac{\pi}{2} \frac{f}{f_0}\right)$



5.47

$$\begin{aligned} f_0 &= 10 \text{ GHz} & N &= 2 \\ \frac{\Delta f}{f_0} &= 0.375 & \eta_L &= 188.5 \\ & & \eta_o &= 377 \end{aligned}$$

a. Using (5-75) for $N=2$

$$\Gamma_{in}(f) = 2e^{-j2\theta} \left[\Gamma_0 \cos(2\theta) + \frac{1}{2} \Gamma_1 \right] \quad (1)$$

which when equated to (5-81) leads to

$$\text{or } 2e^{-j2\theta} \left[\Gamma_0 \cos(2\theta) + \frac{1}{2} \Gamma_1 \right] = e^{-j2\theta} \frac{\eta_L - \eta_o}{\eta_L + \eta_o} \frac{T_2(\sec \theta_m \cos \theta)}{T_2(\sec \theta_m)} \quad (2)$$

$$2\Gamma_0 \cos(2\theta) + \Gamma_1 = \frac{\eta_L - \eta_o}{\eta_L + \eta_o} \frac{1}{T_2(\sec \theta_m)} T_2(\sec \theta_m \cos \theta) \quad (3)$$

Since the fractional bandwidth is 0.375, then

$$\Delta f = 2(f_o - f_m)$$

$$\theta_m = \beta_m l = \frac{2\pi}{\lambda_m} \left(\frac{\lambda_o}{4} \right) = \frac{\pi}{2} \left(\frac{\lambda_o}{\lambda_m} \right) = \frac{\pi}{2} \left(\frac{f_m}{f_o} \right) \Rightarrow f_m = \frac{2}{\pi} f_o \theta_m$$

$$\Delta f = 2 \left[f_o - \frac{2}{\pi} f_o \theta_m \right] = 2 f_o \left(1 - \frac{2}{\pi} \theta_m \right) \Rightarrow \frac{\Delta f}{f_o} = 2 \left(1 - \frac{2}{\pi} \theta_m \right) = 2 - \frac{4}{\pi} \theta_m$$

$$\theta_m = \frac{\pi}{4} \left(2 - \frac{\Delta f}{f_o} \right) = \frac{\pi}{4} \left(2 - 0.375 \right) = \frac{13}{32} \pi = 0.40625\pi = 1.2763 \approx 73.125^\circ$$

$$\text{Thus } \sec \theta_m = \sec(73.125^\circ) = 3.445$$

$$T_2(\sec \theta_m) = 2 \sec^2 \theta_m - 1 = 2(3.445)^2 - 1 = 22.736 \quad (4)$$

Therefore we can write (3) using (5-89) for $T_2(\sec \theta_m \cos \theta)$ as

$$\begin{aligned} 2\Gamma_0 \cos(2\theta) + \Gamma_1 &= \frac{-377 + 188.5}{377 + 188.5} \frac{1}{22.736} \left[\sec^2 \theta_m \cos(2\theta) + (\sec^2 \theta_m - 1) \right] \\ &= \frac{-1}{3} \left(\frac{1}{22.736} \right) \left[\sec^2 \theta_m \cos(2\theta) + (\sec^2 \theta_m - 1) \right] \\ &= -0.01466 \left[\sec^2 \theta_m \cos(2\theta) + (\sec^2 \theta_m - 1) \right] \\ &= -0.01466 (\sec^2 \theta_m) \cos 2\theta - 0.01466 (\sec^2 \theta_m - 1) \\ &= -0.01466 (3.445)^2 \cos(2\theta) - 0.01466 [(3.445)^2 - 1] \end{aligned}$$

$$2\Gamma_0 \cos(2\theta) + \Gamma_1 = -0.174 \cos(2\theta) - 0.1593$$

$$\text{From these } \Gamma_1 = -0.1593$$

$$2\Gamma_0 = -0.174 \Rightarrow \Gamma_0 = -0.087$$

Cont'd.

5.47 cont'd.

$$n=0: \Gamma_0 = 0.087 = \frac{\eta_1 - \eta_0}{\eta_1 + \eta_0} \Rightarrow \eta_1 = \eta_0 \frac{1 + \Gamma_0}{1 - \Gamma_0} = 377 \frac{1 + 0.087}{1 - 0.087} = 316.65$$

$$n=1: \Gamma_1 = -0.1593 = \frac{\eta_2 - \eta_1}{\eta_2 + \eta_1} \Rightarrow \eta_2 = \eta_1 \frac{1 + \Gamma_1}{1 - \Gamma_1} = 316.65 \frac{1 - 0.1593}{1 + 0.1593} = 229.63$$

$$n=2: \Gamma_2 = \Gamma_0 = -0.087 = \frac{\eta_3 - \eta_2}{\eta_3 + \eta_2}$$

$$\epsilon_{r2} = \left(\frac{377}{316.65} \right)^2 = 1.4175 \Rightarrow d_1 = \frac{377}{4} = \frac{30 \times 10^9}{4(10 \times 10^9) \sqrt{1.4175}} = 0.63 \text{ cm}$$

$$\epsilon_{r2} = \left(\frac{377}{229.63} \right)^2 = 2.695 \Rightarrow d_2 = \frac{377}{4} = \frac{30 \times 10^9}{4(10 \times 10^9) \sqrt{2.695}} = 0.457 \text{ cm}$$

- b. Within the bandwidth, the maximum value of (5-81) occurs when $\theta = \theta_m$. Thus

$$|\Gamma_{in}|_{max} = \Gamma_m = \left| \frac{\eta_L - \eta_0}{\eta_L + \eta_0} \right| \frac{T_2(\sec \theta_m \cos \theta_m)}{T_2(\sec \theta_m)} = \left| \frac{\eta_L - \eta_0}{\eta_L + \eta_0} \right| \frac{T_2(1)}{T_2(\sec \theta_m)} = \left| \frac{\eta_L - \eta_0}{\eta_L + \eta_0} \right| \frac{1}{T_2(\sec \theta_m)}$$

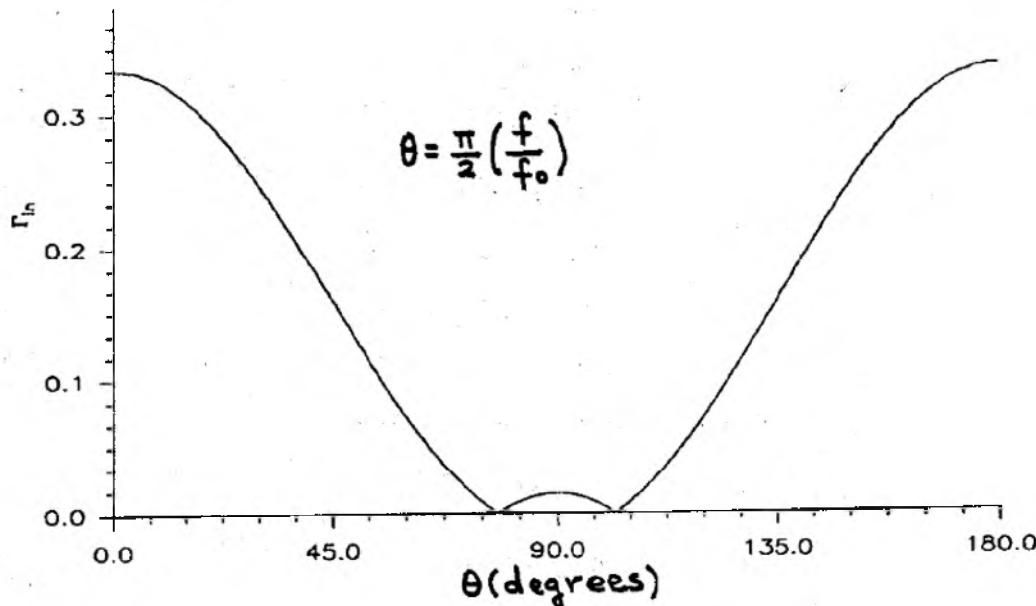
Since $T_2(1) = 1$

$$\text{Thus } \Gamma_m = \frac{1}{3} \left(\frac{1}{22.736} \right) = 0.01466$$

$$\text{SWR}_m = \frac{1 + \Gamma_m}{1 - \Gamma_m} = \frac{1 + 0.01466}{1 - 0.01466} = 1.03$$

$$c. |\Gamma_{in}(f)| = \left| \frac{\eta_L - \eta_0}{\eta_L + \eta_0} \right| \frac{1}{T_2(\sec \theta_m)} T_2(\omega s \theta \sec \theta_m) = \Gamma_m T_2(\sec \theta_m \cos \theta) = 0.174 \cos(2\theta) + 0.1593$$

$$|\Gamma_{in}(f)| = 0.174 \cos(2\theta) + 0.1593, \quad \theta = \frac{\pi}{2} \left(\frac{f}{f_0} \right)$$



5.48

$$\begin{aligned} f_0 &= 2 \text{ GHz} & N &= 2 \\ \frac{\Delta f}{f_0} &= 0.5 & \eta_L &= 377 \\ & & \eta_0 &= 251.33 \end{aligned}$$

a. Using (5-78) and (5-79)

$$\frac{\Delta f}{f_0} = 2 - \frac{4}{\pi} \theta_m \Rightarrow \theta_m = \frac{\pi}{4} \left(2 - \frac{\Delta f}{f_0} \right) = \frac{\pi}{4} \left(2 - 0.5 \right) = \frac{3\pi}{8} = 67.5^\circ \Rightarrow \sec \theta_m = 2.613$$

$$T_2(\sec \theta_m) = 2 \sec^2 \theta_m - 1 = 2(6.828) - 1 = 12.656$$

Within the bandwidth the maximum value of (5-81) occurs when $\theta = \theta_m$. Then $T_2(\sec \theta_m \cos \theta_m) = T_2(1) = 1$. Thus

$$|\Gamma_{in}|_{max} = \Gamma_m = \left| \frac{\eta_L - \eta_0}{\eta_L + \eta_0} \right| \frac{1}{T_2(\sec \theta_m)} = \frac{0.2}{12.656} = 0.0158$$

b. Using (5-75) for $N=2$ and equating it to (5-81) leads to

$$\Gamma_{in}(f) = 2 e^{-j2\theta} \left[\Gamma_0 \cos(2\theta) + \frac{1}{2} \Gamma_1 \right] = e^{-j2\theta} \frac{\eta_L - \eta_0}{\eta_L + \eta_0} \frac{1}{T_2(\sec \theta_m \cos \theta)} T_2(\sec \theta_m \cos \theta)$$

$$2\Gamma_0 \cos(2\theta) + \Gamma_1 = \Gamma_m T_2(\sec \theta_m \cos \theta)$$

By using (5-83) for $T_2(\sec \theta_m \cos \theta)$

$$\begin{aligned} 2\Gamma_0 \cos(2\theta) + \Gamma_1 &= \Gamma_m \left[\sec^2 \theta_m \cos(2\theta) + (\sec^2 \theta_m - 1) \right] \\ &= 0.0158 \left[6.828 \cos(2\theta) + 5.828 \right] \end{aligned}$$

$$2\Gamma_0 \cos(2\theta) + \Gamma_1 = 0.108 \cos(2\theta) + 0.092$$

From these

$$2\Gamma_0 = 0.108 \Rightarrow \Gamma_0 = 0.054$$

$$\Gamma_1 = 0.092$$

$$c. n=0: \Gamma_0 = 0.054 = \frac{\eta_1 - \eta_0}{\eta_1 + \eta_0} \Rightarrow \eta_1 = \eta_0 \frac{1 + \Gamma_0}{1 - \Gamma_0} = 251.33 \frac{1 + 0.054}{1 - 0.054} = 280.023$$

$$\epsilon_{r1} = \left(\frac{377.0}{280.023} \right)^2 = 1.812, d_3 = \frac{\lambda_{10}}{4} = \frac{30 \times 10^9}{4(2 \times 10^9) \sqrt{1.812}} = 2.785 \text{ cm}$$

$$n=1: \Gamma_1 = 0.092 = \frac{\eta_2 - \eta_1}{\eta_2 + \eta_1} \Rightarrow \eta_2 = \eta_1 \frac{1 + \Gamma_1}{1 - \Gamma_1} = 280.023 \frac{1 + 0.092}{1 - 0.092} = 336.768$$

$$\epsilon_{r2} = \left(\frac{377}{336.768} \right)^2 = 1.253 \Rightarrow d_2 = \frac{\lambda_{20}}{4} = \frac{30 \times 10^9}{4(2 \times 10^9) \sqrt{1.253}} = 3.35 \text{ cm}$$

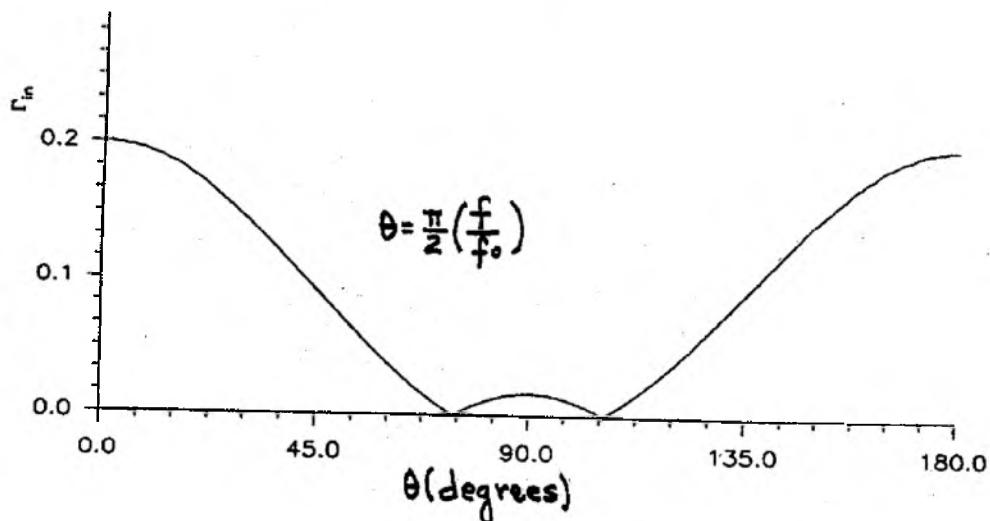
d. From the solution of Problem 5.45

$$f_2 = f_0 - \frac{\Delta f}{2} = (2 - \frac{1}{2}) \text{ GHz} = 1.5 \text{ GHz}, f_4 = f_0 + \frac{\Delta f}{2} = (2 + \frac{1}{2}) \text{ GHz} = 2.5 \text{ GHz}$$

Cont'd.

548 Cont'd.

$$\begin{aligned}
 e. \quad |\Gamma_{in}(f)| &= \left| \frac{\eta_L - \eta_0}{\eta_L + \eta_0} \right| \frac{1}{T_2(\sec \theta_m)} T_2(\cos \theta \sec \theta_m) = \Gamma_m T_2(\sec \theta_m \cos \theta) \\
 &= 0.0158 T_2(\sec \theta_m \cos \theta) = 0.108 \cos(2\theta) + 0.092 \\
 |\Gamma_{in}(f)| &= 0.108 \cos(2\theta) + 0.092, \quad \theta = \frac{\pi}{2} \left(\frac{f}{f_0} \right)
 \end{aligned}$$



5.49

$$f_0 = 3 \text{ GHz} \quad N = 3 \\ \eta_L = 251.33 \\ SWR_{max} = 1.1 \quad \eta_o = 125.67$$

$$a. SWR_m = \frac{1+\Gamma_m}{1-\Gamma_m} \Rightarrow \Gamma_m = \frac{SWR_m - 1}{SWR_m + 1} = \frac{1.1 - 1}{1.1 + 1} = 0.0476$$

Within the bandwidth the maximum value of (5-81) occurs when $\theta = \theta_m$.

Then $T_3(\sec \theta_m \cos \theta_m) = T_3(1) = 1$. Thus

$$|\Gamma_{in}(f)|_{max} = \Gamma_m = \left| \frac{\eta_L - \eta_o}{\eta_L + \eta_o} \right| \frac{1}{T_3(\sec \theta_m)} = \frac{1}{3} \frac{1}{T_3(\sec \theta_m)} \Rightarrow T_3(\sec \theta_m) = \frac{1}{3\Gamma_m} = 7$$

From (5-83)

$$T_3(\sec \theta_m) = 4(\sec \theta_m)^3 - 3 \sec \theta_m = 7 = \sec \theta_m [4 \sec^2 \theta_m - 3]$$

$$\text{By iteration } \theta_m = 44.87^\circ \Rightarrow \sec \theta_m = 1.411$$

$$\theta_m = 0.783 \text{ rad.} \quad \begin{cases} \sec^2 \theta_m = 1.991 \\ \sec^3 \theta_m = 2.809 \end{cases}$$

Therefore

$$\frac{\Delta f}{f_0} = 2 \left(1 - \frac{2}{\pi} \theta_m \right) = 1.003 \approx 1$$

b. Using (5-75) for $N = 3$ and equating it to (5-81) for $N = 3$, lead to

$$\Gamma_{in}(f) = 2e^{-j3\theta} [\Gamma_0 \cos(3\theta) + \Gamma_1 \cos(\theta)] = e^{-j3\theta} \frac{\eta_L - \eta_o}{\eta_L + \eta_o} \frac{T_3(\sec \theta_m \cos \theta)}{T_3(\sec \theta_m)}$$

$$2 [\Gamma_0 \cos(3\theta) + \Gamma_1 \cos(\theta)] = \frac{\eta_L - \eta_o}{\eta_L + \eta_o} \frac{1}{T_3(\sec \theta_m)} T_3(\sec \theta_m \cos \theta) = \Gamma_m T_3(\sec \theta_m \cos \theta)$$

$$2 [\Gamma_0 \cos(3\theta) + \Gamma_1 \cos(\theta)] = \Gamma_m T_3(\sec \theta_m \cos \theta)$$

Using the expansion of (5-89) to represent $T_3(\sec \theta_m \cos \theta)$, we can write that

$$2 [\Gamma_0 \cos(3\theta) + \Gamma_1 \cos(\theta)] = 0.0476 [\sec^3 \theta_m \cos(3\theta) + 3(\sec^2 \theta_m - \sec \theta_m) \cos \theta] \\ = 0.0476 (2.809) \cos(3\theta) + 0.0476 (3)(2.809 - 1.411) \cos \theta$$

$$2 [\Gamma_0 \cos(3\theta) + \Gamma_1 \cos(\theta)] = 0.1337 \cos(3\theta) + 0.1996 \cos(\theta)$$

Matching terms

$$2\Gamma_0 = 0.1337 \Rightarrow \Gamma_0 = 0.06685$$

$$2\Gamma_1 = 0.1996 \Rightarrow \Gamma_1 = 0.0998$$

From symmetry

$$\Gamma_2 = \Gamma_1 = 0.0998$$

$$\Gamma_3 = \Gamma_0 = 0.06685$$

Cont'd.

5.49 cont'd.

c. From Part a, $\Gamma_m = 0.0476$

$$d. \Gamma_0 = 0.06685 = \frac{\eta_1 - \eta_0}{\eta_1 + \eta_0} \Rightarrow \eta_1 = \eta_0 \frac{1 + \Gamma_0}{1 - \Gamma_0} = 125.67 \frac{1 + 0.06685}{1 - 0.06685} = 143.67$$

$$\Gamma_1 = 0.0998 = \frac{\eta_2 - \eta_1}{\eta_2 + \eta_1} \Rightarrow \eta_2 = \eta_1 \frac{1 + \Gamma_1}{1 - \Gamma_1} = 143.67 \frac{1 + 0.0998}{1 - 0.0998} = 175.533$$

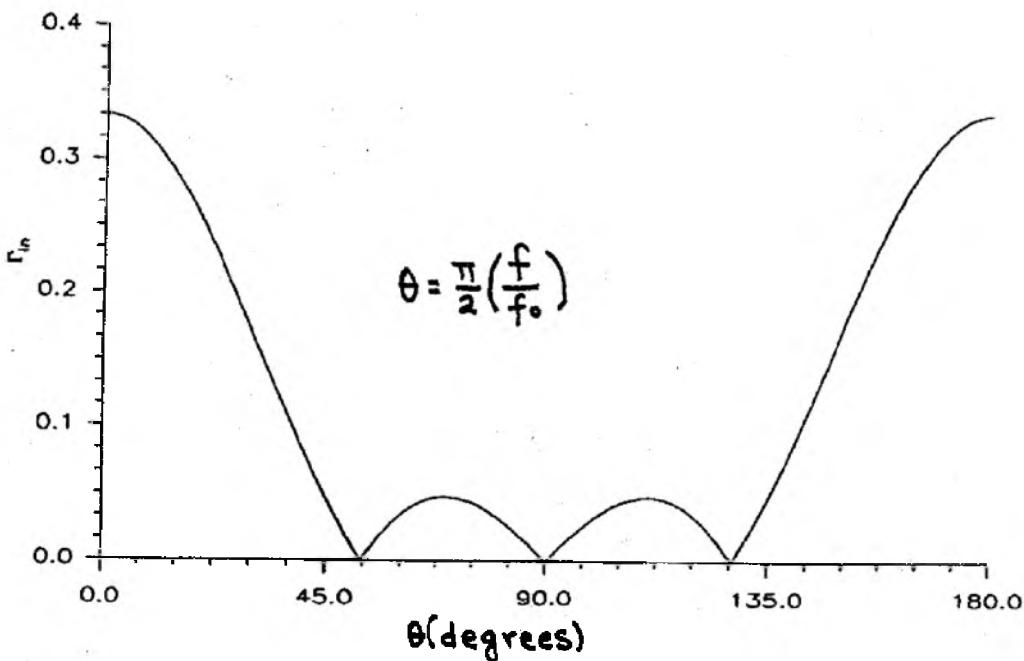
$$\Gamma_2 = 0.0998 = \frac{\eta_3 - \eta_2}{\eta_3 + \eta_2} \Rightarrow \eta_3 = \eta_2 \frac{1 + \Gamma_2}{1 - \Gamma_2} = 175.533 \frac{1 + 0.0998}{1 - 0.0998} = 214.453$$

$$\epsilon_{r1} = \left(\frac{377}{143.67} \right)^2 = 6.886 \Rightarrow d_1 = \frac{30 \times 10^9}{4(3 \times 10^9) \sqrt{6.886}} = 0.953 \text{ cm}$$

$$\epsilon_{r2} = \left(\frac{377}{175.533} \right)^2 = 4.613 \Rightarrow d_2 = \frac{30 \times 10^9}{4(3 \times 10^9) \sqrt{4.613}} = 1.164 \text{ cm}$$

$$\epsilon_{r3} = \left(\frac{377}{214.453} \right)^2 = 3.090 \Rightarrow d_3 = \frac{30 \times 10^9}{4(3 \times 10^9) \sqrt{3.090}} = 1.422 \text{ cm}$$

$$e. \Gamma_{in}(f) = 0.1337 \cos(3\theta) + 0.1996 \cos(\theta) \quad \text{where } \theta = \frac{\pi}{2} \left(\frac{f}{f_0} \right)$$



$$5.50 \quad \theta_i^i = 30^\circ, AR = -2$$

$$\gamma_i^i = \tan^{-1}\left(\frac{E_{\perp}^i}{E_{\parallel}^i}\right) = \tan^{-1}\left(\frac{1}{2}\right) = 26.565^\circ$$

$$\delta_i^i = \phi_{\perp}^i - \phi_{\parallel}^i = -90^\circ$$

$$\epsilon^i = \cot^{-1}(AR^i) = \cot^{-1}(-2) = -26.565^\circ$$

$$\gamma^i = \frac{1}{2} \sin^{-1} \left[\frac{\sin(2\epsilon^i)}{\sin \delta^i} \right] = \frac{1}{2} \sin^{-1} \left[\frac{\sin(-53.13^\circ)}{\sin(-90^\circ)} \right]$$

$$= 26.565^\circ = 0.4636 \text{ rad.}$$

$$\tan \gamma^i = \tan(26.565^\circ) = 0.5$$

$$\Gamma_{\parallel}^b = -1 \Rightarrow |\Gamma_{\parallel}^b| = 1, \gamma_{\parallel}^r = 180^\circ$$

$$\Gamma_{\perp}^b = -1 \Rightarrow |\Gamma_{\perp}^b| = 1, \gamma_{\perp}^r = 180^\circ$$

$$\gamma^r = \tan^{-1} \left[\frac{|\Gamma_{\perp}^b| \tan \gamma^i}{|\Gamma_{\parallel}^b|} \right] = \tan^{-1} \left(\frac{1}{2} \right) = 26.565^\circ$$

$$\delta^r = (\delta^i - \pi) + (\gamma_{\perp}^r - \gamma_{\parallel}^r) = -\frac{\pi}{2} - \pi = -\frac{3\pi}{2} = \frac{\pi}{2}$$

$$2\epsilon^r = \sin^{-1} \left[\sin(2\gamma^r) \sin(\delta^r) \right]$$

$$= \sin^{-1} \left[\sin(53.13^\circ) \sin(90^\circ) \right] = \sin^{-1}(0.5)$$

$$2\epsilon^r = 53.13^\circ$$

$$\epsilon^r = 26.565^\circ$$

$$AR^r = \cot(26.565^\circ) = 2$$

Therefore the reflected field is elliptically polarized with CCW rotation and an axial ratio of AR = 2

$$5.51 \quad \theta_i^i = 30^\circ, AR = -2$$

$$\delta_i^i = \phi_{\perp}^i - \phi_{\parallel}^i = -90^\circ$$

$$\epsilon^i = \cot^{-1}(AR^i) = \cot^{-1}(-2) = -26.565^\circ$$

$$\gamma^i = \frac{1}{2} \sin^{-1} \left[\frac{\sin 2\epsilon^i}{\sin \delta^i} \right] = \frac{1}{2} \sin^{-1} \left[\frac{\sin(-53.13^\circ)}{\sin(-90^\circ)} \right] = 26.565^\circ$$

$$\tan \gamma^i = 0.5$$

From Example 5.13

$$\Gamma_{\perp}^b = -0.824 \Rightarrow |\Gamma_{\perp}^b| = 0.824, \gamma_{\perp}^r = 180^\circ; \Gamma_{\parallel}^b = -0.773 \Rightarrow |\Gamma_{\parallel}^b| = 0.773, \gamma_{\parallel}^r = 180^\circ$$

$$\gamma^r = \tan^{-1} \left[\frac{|\Gamma_{\perp}^b| \tan \gamma^i}{|\Gamma_{\parallel}^b|} \right] = \tan^{-1} \left(\frac{0.824 \cdot 1}{0.773 \cdot 2} \right) = 28.06^\circ$$

$$2\epsilon^r = \sin^{-1} \left[\sin(2\gamma^r) \sin \delta^r \right]$$

$$= \sin^{-1} \left[\sin(56.114^\circ) \sin(+90^\circ) \right] = 56.114^\circ$$

$$\delta^r = (\delta^i - \pi) + (\gamma_{\perp}^r - \gamma_{\parallel}^r) = -90^\circ - 180^\circ + (180^\circ - 180^\circ) = -270^\circ$$

$$\epsilon^r = 28.057^\circ = 0.4897 \text{ rad.}$$

$$\delta^r = +90^\circ \text{ so that } -180^\circ < \delta^r < 180^\circ \quad \epsilon^r = +90^\circ \quad AR^r = \cot(\epsilon^r) = 1.8762$$

Therefore the reflected field is elliptically polarized with counter-clockwise rotation (CCW) and with an axial ratio of AR = 1.8762.

From Example 5.13

$$\Gamma_{\perp}^b = 0.1758 \Rightarrow |\Gamma_{\perp}^b| = 0.1758, \gamma_{\perp}^t = 0^\circ$$

$$\Gamma_{\parallel}^b = 0.197 \Rightarrow |\Gamma_{\parallel}^b| = 0.197, \gamma_{\parallel}^t = 0^\circ$$

$$\gamma^t = \tan^{-1} \left[\frac{|\Gamma_{\perp}^b| \tan \gamma^i}{|\Gamma_{\parallel}^b|} \right] = \tan^{-1} \left[\frac{0.1758 \cdot 1}{0.197 \cdot 2} \right]$$

$$= \tan^{-1}(0.4462) = 24.0464^\circ$$

$$\delta^t = \delta^i + (\gamma_{\perp}^t - \gamma_{\parallel}^t) = -90^\circ + (0^\circ - 0^\circ) = -90^\circ$$

$$2\epsilon^t = \sin^{-1} \left[\sin(2\gamma^t) \sin(\delta^t) \right]$$

$$= \sin^{-1} \left[\sin(48.093^\circ) \sin(-90^\circ) \right]$$

$$2\epsilon^t = \sin^{-1}(-0.7442) = -48.093^\circ$$

$$\epsilon^t = -24.0464^\circ = -0.4197 \text{ rad.}$$

$$AR^t = \cot(-24.0464^\circ) = -2.2411$$

Therefore the transmitted field is elliptically polarized with clockwise (CW) rotation and with an axial ratio of AR = -2.2411.

$$5.52 \quad E^i(z) = (\hat{a}_z + j 2\hat{a}_y) e^{-j\beta_0 x}$$

$$(a) \quad E^r(z) = -(\hat{a}_z + j 2\hat{a}_y) e^{+j\beta_0 x}$$

- (b)
- Elliptical: 2 components; not of the same magnitude; 90° phase diff.
 - CW: y component is leading the z component
 - Rotate y component to z component thru smaller angular range (CW)
 - $AR = \frac{2}{1} = -2$ since it is elliptical with 90° phase difference;
the major and minor axes of the ellipse align with principal axes.

$$(c) \quad E^r(z) = -(\hat{a}_z + j 2\hat{a}_y) e^{+j\beta_0 x}$$

- Elliptical: 2 components; not of the same magnitude; 90° phase diff.

• CCW: y component is leading the z component
Rotate y component to z component
thru smaller angular range (CCW)

- $AR = \frac{2}{1} = 2$ since it is elliptical with 90° phase difference;
the major and minor axes of the ellipse align with principal axes

$$5.53 \quad E^i = (2\hat{a}_x - j\hat{a}_z) E_0 e^{-j\beta_0 y}$$

- (a) Since reflecting surface is PMC $\Rightarrow \Gamma = +1$. Therefore

$$E^r = (2\hat{a}_x - j\hat{a}_y) E_0 e^{+j\beta_0 y}$$

$$(b) \quad E^i = (2\hat{a}_x - j\hat{a}_z) E_0 e^{-j\beta_0 y}$$

- Elliptical: 2 components; not of the same magnitude; 90° phase diff.

- CCW: x component is leading the z component

Rotate the x -component to z component
thru smaller angular range (CCW)

- $AR = \frac{2}{1} = 2$ (major and minor axes of ellipse align with principal axes)

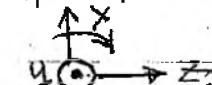
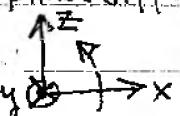
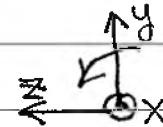
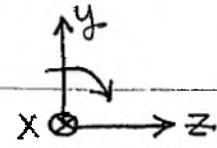
$$(c) \quad E^r = (2\hat{a}_x - j\hat{a}_z) E_0 e^{+j\beta_0 y}$$

- Elliptical: 2 components; not of the same magnitude; 90° phase diff.

- CW: x component is leading the z component

Rotate the x -component to the z component thru smaller range

- $AR = -\frac{2}{1} = -2$ (major and minor axes of ellipse align with principal axes)



5.54

$$\epsilon = \epsilon_0, \mu = \mu_0, \sigma = 0; \text{CCW(LH) CP}, \theta_i = 18.43495^\circ$$

$$\theta_c = \sin^{-1}\left(\sqrt{\frac{\epsilon_2}{\epsilon_1}}\right) = \sin^{-1}\left(\sqrt{\frac{1}{\epsilon_{ri}}}\right) = \sin^{-1}\left(\sqrt{\frac{1}{q}}\right) = 19.4712^\circ$$

$$\theta_B = \tan^{-1}\left(\sqrt{\frac{\epsilon_2}{\epsilon_1}}\right) = \tan^{-1}\left(\sqrt{\frac{1}{\epsilon_{ri}}}\right) = \tan^{-1}\left(\sqrt{\frac{1}{q}}\right) = 18.4349^\circ$$

(a) Since one of the components is totally transmitted, because the incident angle is equal to the Brewster angle, the reflected field has only one component. Therefore the reflected field is linearly polarized.

(b) No rotation because it is linearly polarized.

$$(c, d) T_1^b = \frac{2 \cos \theta_i}{\cos \theta_i + \sqrt{\frac{\epsilon_2}{\epsilon_1}} \sqrt{1 - \frac{\epsilon_1}{\epsilon_2} \sin^2 \theta_i}} = \frac{2 \cos(18.43495)}{\cos(18.43495) + \sqrt{q} \sqrt{1 - q \sin^2(18.43495)}} \\ = \frac{2(0.9487)}{0.9487 + \frac{1}{\sqrt{q}}(0.3162)} = 1.8$$

$$T_{||}^b = \frac{2 \sqrt{\frac{\epsilon_1}{\epsilon_2}} \cos \theta_i}{\cos \theta_i + \sqrt{\frac{\epsilon_1}{\epsilon_2}} \sqrt{1 - \frac{\epsilon_1}{\epsilon_2} \sin^2 \theta_i}} = \frac{2(3) \cos(18.43495)}{\cos(18.43495) + \sqrt{q} \sqrt{1 - q \sin^2(18.43495)}} \\ = \frac{2(3) 0.9487}{0.9487 + 3(0.3162)} = 3$$

1. Since the transmission coefficients are not identical, the transmitted field components will not be identical.
2. Also both transmission coefficients are real positive (zero phase angle); the transmitted field components will maintain the same phase as the incident field components; i.e. will have the same sense of rotation.
Thus the polarization of the transmitted wave is:

(c) Elliptically polarized because the 2 field components are not equal.

(d) The rotation is CCW(LH) because the transmitted field components maintain the same phase as the incident components.

5.55 $\epsilon = \epsilon_0, \mu = \mu_0, \sigma = 0; \text{CW(RH) CP}, \theta_i = 18.43495^\circ$

$$\theta_c = \sin^{-1}(\sqrt{\frac{\epsilon_2}{\epsilon_1}}) = \sin^{-1}(\sqrt{\frac{1}{\epsilon_1}}) = \sin^{-1}(\sqrt{\frac{1}{\sigma}}) = 19.4712^\circ$$

$$\theta_B = \tan^{-1}(\sqrt{\frac{\epsilon_2}{\epsilon_1}}) = \tan^{-1}(\sqrt{\frac{1}{\epsilon_1}}) = \tan^{-1}(\sqrt{\frac{1}{\sigma}}) = 18.4349^\circ$$

Based on the solution of Problem 5.54 :

- (a) The reflected field is linearly polarized.
- (b) No rotation because the reflected wave is linearly polarized.
- (c) T_{\perp}^b (based on the solution of Problem 5.54) = 1.8
 T_{\parallel}^b (based on the solution of Problem 5.54) = 3
 - 1. Since the transmission coefficients are not identical, the transmitted field components will not be identical.
 - 2. Also both transmission coefficients are real positive (zero phase angle); therefore the transmitted field components will maintain the same phase as the incident field components; i.e., will have the same sense of rotation.
- Thus the polarization of the transmitted wave is :
- (d) Elliptically polarized because the 2 field components are not equal.
- (e) The rotation is CW(RH) because the transmitted field components maintain the same phase as the incident field components; thus have the same sense of rotation.

5.56 Based on the geometry and ray-tracing method illustrated in Figure 5-11, we can write the transmitted wave as a summation of the following terms:

$$T = T_{32} (T_{21} e^{-j\theta}) + T_{32} \Gamma_{21} \Gamma_{23} T_{21} e^{-j\frac{3\theta}{2}} + T_{32} \Gamma_{21}^2 \Gamma_{23}^2 T_{21} e^{-j\frac{5\theta}{2}} + \dots$$

Factoring the common terms $T_{32} T_{21} e^{-j\theta}$ that appear in each of the terms, we can write the above equation as

$$T = T_{32} T_{21} e^{-j\theta} \left[1 + (\Gamma_{21} \Gamma_{23} e^{-j\frac{2\theta}{2}}) + (\Gamma_{21} \Gamma_{23} e^{-j\frac{2\theta}{2}})^2 + \dots \right]$$

which can be written in compact form, since the terms within the bracket form a geometric series, as

$$T = T_{32} T_{21} e^{-j\theta} \left[\frac{1}{1 - \Gamma_{21} \Gamma_{23} e^{-j\frac{2\theta}{2}}} \right] = \frac{T_{32} T_{21} e^{-j\theta}}{\Gamma_{21} - \Gamma_{12} + \Gamma_{12} \Gamma_{23} e^{-j\frac{2\theta}{2}}}$$

Since

$$T_{32} = \frac{2\eta_3}{\eta_3 + \eta_2}, T_{21} = \frac{2\eta_2}{\eta_2 + \eta_1} \text{ and } \eta_3 = \eta_1,$$

we can reduce the above to

$$T = \frac{4\eta_1 \eta_2 e^{-j\theta}}{(\eta_1 + \eta_2)^2} \left[\frac{1}{1 + \Gamma_{12} \Gamma_{23} e^{-j\frac{2\theta}{2}}} \right] = \frac{4\eta_1 \eta_2 e^{-j\theta}}{(\eta_1 + \eta_2)^2} \left[\frac{1}{1 - (\Gamma_{12})^2 e^{-j\frac{2\theta}{2}}} \right]$$

$$\Gamma_{23} = -\Gamma_{12}$$

Since $\theta = |B_2|d$, the transmission coefficient can be written as

$$T = \frac{4\eta_1 \eta_2 e^{-j|B_2|d}}{(\eta_1 + \eta_2)^2} \left[\frac{1}{1 - (\Gamma_{12})^2 e^{-j2|B_2|d}} \right]$$

*In the first printing of the book, there is an extra factor of 2 in the exponential at the numerator; this is a typographical error.

5-57 For parallel polarization, the wavenumbers are the same as those of the perpendicular polarization and given by (5-110a)–(5-110c), or

$$\beta_i = \eta_1 \frac{w}{v_0} (\hat{a}_x \sin \theta_i + \hat{a}_z \cos \theta_i) \quad (1)$$

$$\beta_r = \eta_1 \frac{w}{v_0} (\hat{a}_x \sin \theta_i - \hat{a}_z \cos \theta_i) \quad (2)$$

$$\beta_t = \eta_2 \frac{w}{v_0} (\hat{a}_x \sin \theta_t + \hat{a}_z \cos \theta_t) \quad (3)$$

The electric and magnetic fields for parallel polarization, based on the geometries of Figure 5-4 and 5-29 (but with the transmitted fields of Figure 5-4 in the 4th quadrant), we can write them as

$$\underline{E}_{||}^i = (\hat{a}_x \cos \theta_i - \hat{a}_z \sin \theta_i) E_0 e^{-j\beta_i(x \sin \theta_i + z \cos \theta_i)}$$

$$\underline{H}_{||}^i = \hat{a}_y \frac{E_0}{\eta_1} e^{j\beta_i(x \sin \theta_i + z \cos \theta_i)}$$

$$\underline{E}_{||}^r = (\hat{a}_x \cos \theta_r + \hat{a}_z \sin \theta_r) \Gamma_{||}^b e^{-j\beta_r(x \sin \theta_r - z \cos \theta_r)}$$

$$\underline{H}_{||}^r = -\hat{a}_y \frac{\Gamma_{||}^b}{\eta_1} E_0 e^{-j\beta_r(x \sin \theta_r - z \cos \theta_r)}$$

$$\underline{E}_{||}^t = (\hat{a}_x \cos \theta_t + \hat{a}_z \sin \theta_t) T_{||}^b E_0 e^{-j\beta_t(x \sin \theta_t + z \cos \theta_t)}$$

$$\underline{H}_{||}^t = \hat{a}_y \frac{T_{||}^b}{\eta_2} E_0 e^{-j\beta_t(x \sin \theta_t + z \cos \theta_t)}$$

The corresponding Poynting vectors can be written, using the above equations for the incident, reflected and transmitted electric and magnetic fields, as

$$S_{||}^i = \frac{1}{2} \operatorname{Re}(\underline{E}_{||}^i \times \underline{H}_{||}^i) = \frac{|E_0|^2}{2\eta_1} (\hat{a}_x \sin \theta_i + \hat{a}_z \cos \theta_i) \quad (4)$$

$$S_{||}^r = \frac{1}{2} \operatorname{Re}(\underline{E}_{||}^r \times \underline{H}_{||}^r) = \frac{|E_0 \Gamma_{||}^b|^2}{2\eta_1} (\hat{a}_x \sin \theta_r - \hat{a}_z \cos \theta_r) \quad (5)$$

$$S_{||}^t = \frac{1}{2} \operatorname{Re}(\underline{E}_{||}^t \times \underline{H}_{||}^t) = \frac{|E_0 T_{||}^b|^2}{2\eta_2} (\hat{a}_x \sin \theta_t + \hat{a}_z \cos \theta_t) \quad (6)$$

Cont'd

5-57 (cont'd)

It is apparent by examining (1)-(3) and (4)-(6) that the wavenumbers and Poynting vector for parallel polarization:

- Incident Fields

The wavenumber and Poynting vector are both parallel to each other and both point inward toward the origin.

- Reflected Fields

The wave number and Poynting vector are both parallel to each other and both point outward from the origin.

- Transmitted Fields

The wave number and Poynting vector are antiparallel to each other; with the wave number pointing inward toward the origin while the Poynting vector pointing outward from the origin.

These are the same observations made for the perpendicular polarization as represented, respectively, by (5-110a)-(5-110c) and (5-111a)-(5-111c) and illustrated graphically in Figure 5-29.

CHAPTER 6

6.1 If $\underline{H}_e = j\omega \epsilon \nabla \times \underline{\Pi}_e$

Maxwell's curl equation $\nabla \times \underline{E}_e = -j\omega \mu \underline{H}_e$ can be written as

$$\nabla \times \underline{E}_e = -j\omega \mu \underline{H}_e = -j\omega \mu (j\omega \epsilon \nabla \times \underline{\Pi}_e) = \omega^2 \mu \epsilon \nabla \times \underline{\Pi}_e \quad (1)$$

or

$$\nabla \times (\underline{E}_e - \omega^2 \mu \epsilon \underline{\Pi}_e) = \nabla \times (\underline{E}_e - \beta^2 \underline{\Pi}_e) = 0 \quad \text{where } \beta^2 = \omega^2 \mu \epsilon.$$

Letting

$$\underline{E}_e - \beta^2 \underline{\Pi}_e = -\nabla \phi_e \Rightarrow \underline{E}_e = -\nabla \phi_e + \beta^2 \underline{\Pi}_e \quad (2)$$

Taking the curl of (1) and using the vector identity of

$$\nabla \times \nabla \times \underline{A} = \nabla(\nabla \cdot \underline{A}) - \nabla^2 \underline{A}$$

leads to

$$\nabla \times \underline{H}_e = j\omega \epsilon \nabla \times \nabla \times \underline{\Pi}_e = j\omega \epsilon [\nabla(\nabla \cdot \underline{\Pi}_e) - \nabla^2 \underline{\Pi}_e]$$

Using Maxwell's equation of

$$\nabla \times \underline{H}_e = \underline{J} + j\omega \epsilon \underline{E}_e \quad (3)$$

reduces (3) to

$$\underline{J} + j\omega \epsilon \underline{E}_e = j\omega \epsilon [\nabla(\nabla \cdot \underline{\Pi}_e) - \nabla^2 \underline{\Pi}_e] \quad (4)$$

Substituting (2) into (4) reduces to

$$\nabla^2 \underline{\Pi}_e + \beta^2 \underline{\Pi}_e = j \frac{\underline{J}}{\omega \epsilon} + [\nabla(\nabla \cdot \underline{\Pi}_e) + \nabla \phi_e] \quad (5)$$

Letting $\phi_e = -\nabla \cdot \underline{\Pi}_e$ simplifies (5) to

$$\nabla^2 \underline{\Pi}_e + \beta^2 \underline{\Pi}_e = j \frac{\underline{J}}{\omega \epsilon} \quad (6)$$

and (2) to

$$\underline{E}_e = \nabla(\nabla \cdot \underline{\Pi}_e) + \beta^2 \underline{\Pi}_e \quad (7)$$

Comparing (6) with (6-16) leads to the relation

$$\underline{\Pi}_e = -j \frac{1}{\omega \mu \epsilon} \underline{A} \quad (8)$$

6.2 If $\underline{E}_m = -j\omega \mu \nabla \times \underline{H}_m$

Maxwell's curl equation $\nabla \times \underline{H}_m = j\omega \epsilon \underline{E}_m$ can be written as

$$\nabla \times \underline{H}_m = j\omega \epsilon (-j\omega \mu \nabla \times \underline{H}_m) = +\omega^2 \mu \epsilon \nabla \times \underline{H}_m \quad (1)$$

or

$$\nabla \times (\underline{H}_m - \omega^2 \mu \epsilon \underline{H}_m) = \nabla \times (\underline{H}_m - \beta^2 \underline{H}_m) = 0 \text{ where } \beta^2 = \omega^2 \mu \epsilon.$$

Letting

$$\underline{H}_m - \beta^2 \underline{H}_m = -\nabla \phi_m \Rightarrow \underline{H}_m = -\nabla \phi_m + \beta^2 \underline{H}_m \quad (2)$$

Taking the curl of (1) and using the vector identity of

$$\nabla \times \nabla \times \underline{E} = \nabla (\nabla \cdot \underline{E}) - \nabla^2 \underline{E}$$

leads to

$$\nabla \times \underline{E}_m = -j\omega \mu \nabla \times \nabla \times \underline{H}_m = -j\omega \mu [\nabla (\nabla \cdot \underline{H}_m) - \nabla^2 \underline{H}_m] \quad (3)$$

Using Maxwell's equation of

$$\nabla \times \underline{E}_m = -\underline{M} = -j\omega \mu \underline{H}_m$$

reduces (3) to

$$-\underline{M} - j\omega \mu \underline{H}_m = -j\omega \mu [\nabla (\nabla \cdot \underline{H}_m) - \nabla^2 \underline{H}_m] \quad (4)$$

Substituting (2) into (4) reduces to

$$\nabla^2 \underline{H}_m + \beta^2 \underline{H}_m = j \frac{\underline{M}}{\omega \mu} + [\nabla (\nabla \cdot \underline{H}_m) + \nabla \phi_m] \quad (5)$$

Letting

$$\phi_m = -\nabla \cdot \underline{H}_m$$

simplifies (5) to

$$\nabla^2 \underline{H}_m + \beta^2 \underline{H}_m = j \frac{\underline{M}}{\omega \mu} \quad (6)$$

and (2) to

$$\underline{H}_m = \nabla (\nabla \cdot \underline{H}_m) + \beta^2 \underline{H}_m \quad (7)$$

Comparing (6) with (6-28) leads to the relation

$$\underline{H}_m = -j \frac{1}{\omega \mu \epsilon} \underline{E} \quad (8)$$

$$6.3 \quad H_A = \frac{1}{\mu} \nabla \times A \quad , \quad A = \frac{k}{4\pi} \iiint J(r') \frac{e^{-j\beta R}}{R} dv' \quad , \quad E_A = \frac{1}{j\omega\epsilon} \nabla \times H_A$$

$$H_A = \frac{1}{\mu} \nabla \times A = \frac{1}{\mu} \nabla \times \left[\frac{k}{4\pi} \iiint J(r') \frac{e^{-j\beta R}}{R} dv' \right] = \frac{1}{4\pi} \nabla \times \iiint J(r') \frac{e^{-j\beta R}}{R} dv'$$

$$H_A = \frac{1}{4\pi} \iiint \nabla \times \left[J(r') \frac{e^{-j\beta R}}{R} \right] dv' \quad (1)$$

Using the identity

$$\nabla \times (\alpha B) = \alpha \nabla \times B - B \times \nabla \alpha \quad (2)$$

we can write (1), remembering that $J(r')$ is not a function of r , as

$$H_A = -\frac{1}{4\pi} \iiint \left[J(r') \times \nabla \left(\frac{e^{-j\beta R}}{R} \right) \right] dv' \quad (3)$$

Now using (3) we can write the electric field as

$$E_A = \frac{1}{j\omega\epsilon} \nabla \times H_A = -\frac{1}{j\omega\epsilon} \nabla \times \left\{ \frac{1}{4\pi} \iiint \left[J(r') \times \nabla \left(\frac{e^{-j\beta R}}{R} \right) \right] dv' \right\}$$

$$E_A = -\frac{1}{j4\pi\omega\epsilon} \iiint \nabla \times \left[J(r') \times \nabla \left(\frac{e^{-j\beta R}}{R} \right) \right] dv' \quad (4)$$

Now using the vector identity

$$\nabla \times (B \times C) = B \nabla \cdot C - C \nabla \cdot B + (C \cdot \nabla) B - (B \cdot \nabla) C \quad (5)$$

we can write (4) as

$$E_A = -\frac{1}{j4\pi\omega\epsilon} \iiint \left\{ \left[J \nabla \cdot \nabla \left(\frac{e^{-j\beta R}}{R} \right) - \nabla \left(\frac{e^{-j\beta R}}{R} \right) [\nabla \cdot J] + (\nabla \left(\frac{e^{-j\beta R}}{R} \right) \cdot \nabla) J - (J \cdot \nabla) \nabla \left(\frac{e^{-j\beta R}}{R} \right) \right] \right\} dv' \quad (6)$$

Again, since $J(r')$ is not a function of r , the second and third terms are equal to zero. Therefore (6) reduces to

$$E_A = -\frac{1}{j4\pi\omega\epsilon} \iiint \left\{ J(r') \nabla^2 \left(\frac{e^{-j\beta R}}{R} \right) - [J(r') \cdot \nabla] \nabla \left(\frac{e^{-j\beta R}}{R} \right) \right\} dv' \quad (7)$$

Since $\nabla^2 \left(\frac{e^{-j\beta R}}{R} \right) = \frac{1}{R^2} \frac{\partial}{\partial R} \left[R^2 \frac{\partial}{\partial R} \left(\frac{e^{-j\beta R}}{R} \right) \right] = -\beta^2 \frac{e^{-j\beta R}}{R^2}$, then (7) reduces to

$$E_A = -j \frac{1}{4\pi\omega\epsilon} \iiint \left\{ \left[\beta^2 J(r') + (J(r') \cdot \nabla) \nabla \right] \frac{e^{-j\beta R}}{R^2} \right\} dv' \quad (8)$$

$$\begin{aligned}
 6.4 \quad & \underline{\underline{E}}_F = -\frac{1}{\epsilon} \nabla \times \underline{\underline{E}}, \quad \underline{\underline{E}} = \frac{\epsilon}{4\pi} \iiint \underline{\underline{M}}(r') \frac{e^{-j\beta R}}{R} dv', \quad \underline{\underline{H}}_F = -\frac{1}{j\omega\mu} \nabla \times \underline{\underline{E}}_F \\
 & \underline{\underline{E}}_F = -\frac{1}{\epsilon} \nabla \times \underline{\underline{E}} = -\frac{1}{\epsilon} \nabla \times \left[\frac{\epsilon}{4\pi} \iiint \underline{\underline{M}}(r') \frac{e^{-j\beta R}}{R} dv' \right] = -\frac{1}{4\pi} \nabla \times \iiint \underline{\underline{M}} \frac{e^{-j\beta R}}{R} dv' \\
 & \underline{\underline{E}}_F = -\frac{1}{4\pi} \iiint \nabla \times \left[\underline{\underline{M}}(r') \frac{e^{-j\beta R}}{R} \right] dv' \tag{1}
 \end{aligned}$$

Using the identity $\nabla \times (\alpha \underline{\underline{B}}) = \alpha \nabla \times \underline{\underline{B}} - \underline{\underline{B}} \times \nabla \alpha$ (2)

we can write (1), remembering that $\underline{\underline{M}}(r')$ is not a function of r , as

$$\underline{\underline{E}}_F = +\frac{1}{4\pi} \iiint \left[\underline{\underline{M}}(r') \times \nabla \left(\frac{e^{-j\beta R}}{R} \right) \right] dv' \tag{3}$$

Now using (3) we can write the magnetic field as

$$\begin{aligned}
 \underline{\underline{H}}_F &= -\frac{1}{j\omega\mu} \nabla \times \underline{\underline{E}}_F = -\frac{1}{j\omega\mu} \nabla \times \left\{ \frac{1}{4\pi} \iiint \left[\underline{\underline{M}}(r') \times \nabla \left(\frac{e^{-j\beta R}}{R} \right) \right] dv' \right\} \\
 \underline{\underline{H}}_F &= -\frac{1}{j4\pi\omega\mu} \iiint \nabla \times \left[\underline{\underline{M}}(r') \times \nabla \left(\frac{e^{-j\beta R}}{R} \right) \right] dv' \tag{4}
 \end{aligned}$$

Now using the vector identity

$$\nabla \times (\underline{\underline{B}} \times \underline{\underline{C}}) = \underline{\underline{B}} \cdot \nabla \underline{\underline{C}} - \underline{\underline{C}} \cdot \nabla \underline{\underline{B}} + (\underline{\underline{C}} \cdot \nabla) \underline{\underline{B}} - (\underline{\underline{B}} \cdot \nabla) \underline{\underline{C}} \tag{5}$$

we can write (4) as

$$\underline{\underline{H}}_F = -\frac{1}{j4\pi\omega\mu} \iiint \left\{ \underline{\underline{M}} \nabla \cdot \nabla \left(\frac{e^{-j\beta R}}{R} \right) - \nabla \left(\frac{e^{-j\beta R}}{R} \right) [\nabla \cdot \underline{\underline{M}}] + \left[\left(\nabla \frac{e^{-j\beta R}}{R} \right) \cdot \nabla \right] \underline{\underline{M}} - (\underline{\underline{M}} \cdot \nabla) \nabla \left(\frac{e^{-j\beta R}}{R} \right) \right\} dv' \tag{6}$$

Again, since $\underline{\underline{M}}(r')$ is not a function of r , the second and third terms are equal to zero. Therefore (6) reduces to

$$\underline{\underline{H}}_F = -\frac{1}{j4\pi\omega\mu} \iiint \left\{ \underline{\underline{M}}(r') \nabla^2 \left(\frac{e^{-j\beta R}}{R} \right) - [\underline{\underline{M}}(r') \cdot \nabla] \nabla \left(\frac{e^{-j\beta R}}{R} \right) \right\} dv' \tag{7}$$

Since

$$\nabla^2 \left(\frac{e^{-j\beta R}}{R} \right) = \frac{1}{R^2} \frac{\partial^2}{\partial R^2} \left[R^2 \frac{\partial}{\partial R} \left(\frac{e^{-j\beta R}}{R} \right) \right] = -\beta^2 \frac{e^{-j\beta R}}{R}$$

then (7) reduces to

$$\underline{\underline{H}}_F = -j \frac{1}{4\pi\omega\mu} \iiint \left\{ \left[\beta^2 \underline{\underline{M}}(r') + (\underline{\underline{M}}(r') \cdot \nabla) \nabla \right] \frac{e^{-j\beta R}}{R} \right\} dv' \tag{8}$$

6.5 A. TEM^x $\Rightarrow E_x = H_x = 0$

It is apparent that by examining (6-41) and (6-43) that $E_x = H_x = 0$ can be obtained by any of the following three combinations.

1. $A_y = A_z = F_y = F_z = 0, A_x \neq 0, F_x \neq 0, \frac{\partial}{\partial y} \neq 0, \frac{\partial}{\partial z} \neq 0$

For this combination, according to (6-41)

$$E_x = -j\omega A_x - j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 A_x}{\partial x^2} = -j \frac{1}{\omega \mu \epsilon} \left(\frac{\partial^2}{\partial x^2} + \omega^2 \mu \epsilon \right) A_x = 0$$

provided that

$$A_x(x, y, z) = A_x^+(y, z) e^{-j\beta x} + A_x^-(y, z) e^{+j\beta x}$$

Similarly according to (6-43)

$$H_x = -j\omega F_x - j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 F_x}{\partial x^2} = -j \frac{1}{\omega \mu \epsilon} \left(\frac{\partial^2}{\partial x^2} + \omega^2 \mu \epsilon \right) F_x = 0$$

provided that

$$F_x(x, y, z) = F_x^+(y, z) e^{-j\beta x} + F_x^-(y, z) e^{+j\beta x}$$

Also according to (6-41) and (6-43)

$$E_y = \left(-\frac{1}{\sqrt{\mu \epsilon}} \frac{\partial A_x^+}{\partial y} - \frac{1}{\epsilon} \frac{\partial F_x^+}{\partial z} \right) e^{-j\beta x} + \left(\frac{1}{\sqrt{\mu \epsilon}} \frac{\partial A_x^-}{\partial y} - \frac{1}{\epsilon} \frac{\partial F_x^-}{\partial z} \right) e^{+j\beta x} = E_y^+ + E_y^-$$

$$E_z = \left(-\frac{1}{\sqrt{\mu \epsilon}} \frac{\partial A_x^+}{\partial z} + \frac{1}{\epsilon} \frac{\partial F_x^+}{\partial y} \right) e^{-j\beta x} + \left(\frac{1}{\sqrt{\mu \epsilon}} \frac{\partial A_x^-}{\partial z} + \frac{1}{\epsilon} \frac{\partial F_x^-}{\partial y} \right) e^{+j\beta x} = E_z^+ + E_z^-$$

$$\begin{aligned} H_y &= \left(\frac{1}{\mu} \frac{\partial A_x^+}{\partial z} - \frac{1}{\sqrt{\mu \epsilon}} \frac{\partial F_x^+}{\partial y} \right) e^{-j\beta x} + \left(\frac{1}{\mu} \frac{\partial A_x^-}{\partial z} + \frac{1}{\sqrt{\mu \epsilon}} \frac{\partial F_x^-}{\partial y} \right) e^{+j\beta x} \\ &= \sqrt{\frac{\epsilon}{\mu}} \left(\frac{1}{\sqrt{\mu \epsilon}} \frac{\partial A_x^+}{\partial z} - \frac{1}{\epsilon} \frac{\partial F_x^+}{\partial y} \right) e^{-j\beta x} + \sqrt{\frac{\epsilon}{\mu}} \left(\frac{1}{\sqrt{\mu \epsilon}} \frac{\partial A_x^-}{\partial z} + \frac{1}{\epsilon} \frac{\partial F_x^-}{\partial y} \right) e^{+j\beta x} \end{aligned}$$

$$H_y = \sqrt{\frac{\epsilon}{\mu}} (-E_z^+) + \sqrt{\frac{\epsilon}{\mu}} (E_z^-) = H_y^+ + H_y^-$$

$$\begin{aligned} H_z &= \left(-\frac{1}{\mu} \frac{\partial A_x^+}{\partial y} - \frac{1}{\sqrt{\mu \epsilon}} \frac{\partial F_x^+}{\partial z} \right) e^{-j\beta x} + \left(\frac{1}{\mu} \frac{\partial A_x^-}{\partial y} + \frac{1}{\sqrt{\mu \epsilon}} \frac{\partial F_x^-}{\partial z} \right) e^{+j\beta x} \\ &= \sqrt{\frac{\epsilon}{\mu}} \left(-\frac{1}{\sqrt{\mu \epsilon}} \frac{\partial A_x^+}{\partial y} - \frac{1}{\epsilon} \frac{\partial F_x^+}{\partial z} \right) e^{-j\beta x} + \sqrt{\frac{\epsilon}{\mu}} \left(\frac{1}{\sqrt{\mu \epsilon}} \frac{\partial A_x^-}{\partial y} + \frac{1}{\epsilon} \frac{\partial F_x^-}{\partial z} \right) e^{+j\beta x} \end{aligned}$$

$$H_z = \sqrt{\frac{\epsilon}{\mu}} (E_y^+) + \sqrt{\frac{\epsilon}{\mu}} (-E_y^-) = H_z^+ + H_z^-$$

cont'd.

6.5 Cont'd.

2. $A_x = A_y = A_z = F_y = F_z = 0, F_x \neq 0, \partial/\partial y \neq 0, \partial/\partial z \neq 0$

For this combination

$$E_x = 0$$

$$H_x = -j\omega F_x - j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 F_x}{\partial x^2} = - \frac{1}{\omega \mu \epsilon} \left(\frac{\partial^2 F_x}{\partial x^2} + \omega^2 \mu \epsilon F_x \right) = -j \frac{1}{\omega \mu \epsilon} \left(\frac{\partial^2}{\partial x^2} + \omega^2 \mu \epsilon \right) F_x = 0$$

provided that

$$F_x(x, y, z) = F_x^+(y, z) e^{-j\beta x} + F_x^-(y, z) e^{+j\beta x}$$

Also according to (6-41) and (6-43)

$$E_y = -\frac{1}{\epsilon} \frac{\partial F_x^+}{\partial z} e^{-j\beta x} - \frac{1}{\epsilon} \frac{\partial F_x^-}{\partial z} e^{+j\beta x} = E_y^+ + E_y^-$$

$$E_z = \frac{1}{\epsilon} \frac{\partial F_x^+}{\partial y} e^{-j\beta x} + \frac{1}{\epsilon} \frac{\partial F_x^-}{\partial y} e^{+j\beta x} = E_z^+ + E_z^-$$

$$H_y = \sqrt{\frac{\epsilon}{\mu}} \left(-\frac{1}{\epsilon} \frac{\partial F_x^+}{\partial y} \right) e^{-j\beta x} + \sqrt{\frac{\epsilon}{\mu}} \left(\frac{1}{\epsilon} \frac{\partial F_x^-}{\partial y} \right) e^{+j\beta x} = \sqrt{\frac{\epsilon}{\mu}} (-E_z^+) + \sqrt{\frac{\epsilon}{\mu}} (E_z^-) = H_y^+ + H_y^-$$

$$H_z = \sqrt{\frac{\epsilon}{\mu}} \left(-\frac{1}{\epsilon} \frac{\partial F_x^+}{\partial z} \right) e^{-j\beta x} - \sqrt{\frac{\epsilon}{\mu}} \left(\frac{1}{\epsilon} \frac{\partial F_x^-}{\partial z} \right) e^{+j\beta x} = \sqrt{\frac{\epsilon}{\mu}} (E_y^+) + \sqrt{\frac{\epsilon}{\mu}} (-E_y^-) = H_z^+ + H_z^-$$

3. $A_y = A_z = F_x = F_z = 0, A_x \neq 0, \partial/\partial y \neq 0, \partial/\partial z \neq 0$

For this combination

$$E_x = -j\omega A_x - j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 A_x}{\partial x^2} = -j \frac{1}{\omega \mu \epsilon} \left(\frac{\partial^2}{\partial x^2} + \omega^2 \mu \epsilon \right) A_x = 0$$

$$H_x = 0$$

provided that

$$A_x(x, y, z) = A_x^+(y, z) e^{-j\beta x} + A_x^-(y, z) e^{+j\beta x}$$

Also according to (6-41) and (6-43)

$$E_y = \left(-\frac{1}{\sqrt{\mu \epsilon}} \frac{\partial A_x^+}{\partial y} \right) e^{-j\beta x} + \left(\frac{1}{\sqrt{\mu \epsilon}} \frac{\partial A_x^-}{\partial y} \right) e^{+j\beta x} = E_y^+ + E_y^-$$

$$E_z = \left(-\frac{1}{\sqrt{\mu \epsilon}} \frac{\partial A_x^+}{\partial z} \right) e^{-j\beta x} + \left(\frac{1}{\sqrt{\mu \epsilon}} \frac{\partial A_x^-}{\partial z} \right) e^{+j\beta x} = E_z^+ + E_z^-$$

$$H_y = \sqrt{\frac{\epsilon}{\mu}} \left(\frac{1}{\sqrt{\mu \epsilon}} \frac{\partial A_x^+}{\partial z} \right) e^{-j\beta x} + \sqrt{\frac{\epsilon}{\mu}} \left(\frac{1}{\sqrt{\mu \epsilon}} \frac{\partial A_x^-}{\partial z} \right) e^{+j\beta x} = \sqrt{\frac{\epsilon}{\mu}} (-E_z^+) + \sqrt{\frac{\epsilon}{\mu}} (E_z^-) = H_y^+ + H_y^-$$

$$H_z = \sqrt{\frac{\epsilon}{\mu}} \left(-\frac{1}{\sqrt{\mu \epsilon}} \frac{\partial A_x^+}{\partial y} \right) e^{-j\beta x} - \sqrt{\frac{\epsilon}{\mu}} \left(\frac{1}{\sqrt{\mu \epsilon}} \frac{\partial A_x^-}{\partial y} \right) e^{+j\beta x} = \sqrt{\frac{\epsilon}{\mu}} (E_y^+) + \sqrt{\frac{\epsilon}{\mu}} (-E_y^-) = H_z^+ + H_z^-$$

Cont'd.

6.5 cont'd. B. TEM⁰ $\Rightarrow E_y = H_y = 0$

It is apparent that by examining (6-41) and (6-43) that $E_y = H_y = 0$ can be obtained by any of the following three combinations.

1. $A_x = A_z = F_x = F_z = 0, A_y \neq 0, F_y \neq 0, \frac{\partial}{\partial x} \neq 0, \frac{\partial}{\partial z} \neq 0$

For this combination, according to (6-41)

$$E_y = -j\omega A_y - j \frac{1}{\mu_0 \epsilon_0} \frac{\partial^2 A_y}{\partial y^2} = -j \frac{1}{\mu_0 \epsilon_0} \left(\frac{\partial^2}{\partial y^2} + \omega^2 \mu_0 \epsilon_0 \right) A_y = 0$$

provided that

$$A_y(x, y, z) = A_y^+(x, z) e^{-j\beta y} + A_y^-(x, z) e^{+j\beta y}$$

Similarly, according to (6-43)

$$H_y = -j\omega F_y - j \frac{1}{\mu_0 \epsilon_0} \frac{\partial^2 F_y}{\partial y^2} = -j \frac{1}{\mu_0 \epsilon_0} \left(\frac{\partial^2}{\partial y^2} + \omega^2 \mu_0 \epsilon_0 \right) F_y = 0$$

provided that

$$F_y(x, y, z) = F_y^+(x, z) e^{-j\beta y} + F_y^-(x, z) e^{+j\beta y}$$

Also according to (6-41) and (6-43)

$$E_x = \left(-\frac{1}{\sqrt{\mu_0 \epsilon_0}} \frac{\partial A_y^+}{\partial x} + \frac{1}{\epsilon_0} \frac{\partial F_y^+}{\partial z} \right) e^{-j\beta y} + \left(\frac{1}{\sqrt{\mu_0 \epsilon_0}} \frac{\partial A_y^-}{\partial x} + \frac{1}{\epsilon_0} \frac{\partial F_y^-}{\partial z} \right) e^{+j\beta y} = E_x^+ + E_x^-$$

$$E_z = \left(-\frac{1}{\sqrt{\mu_0 \epsilon_0}} \frac{\partial A_y^+}{\partial z} - \frac{1}{\epsilon_0} \frac{\partial F_y^+}{\partial x} \right) e^{-j\beta y} + \left(\frac{1}{\sqrt{\mu_0 \epsilon_0}} \frac{\partial A_y^-}{\partial z} - \frac{1}{\epsilon_0} \frac{\partial F_y^-}{\partial x} \right) e^{+j\beta y} = E_z^+ + E_z^-$$

$$\begin{aligned} H_x &= \left(-\frac{1}{\mu_0} \frac{\partial A_y^+}{\partial z} - \frac{1}{\sqrt{\mu_0 \epsilon_0}} \frac{\partial F_y^+}{\partial x} \right) e^{-j\beta y} + \left(-\frac{1}{\mu_0} \frac{\partial A_y^-}{\partial z} + \frac{1}{\sqrt{\mu_0 \epsilon_0}} \frac{\partial F_y^-}{\partial x} \right) e^{-j\beta y} \\ &= \sqrt{\frac{\epsilon_0}{\mu_0}} \left(\frac{1}{\sqrt{\mu_0 \epsilon_0}} \frac{\partial A_y^+}{\partial z} - \frac{1}{\epsilon_0} \frac{\partial F_y^+}{\partial x} \right) e^{-j\beta y} + \sqrt{\frac{\epsilon_0}{\mu_0}} \left(\frac{1}{\sqrt{\mu_0 \epsilon_0}} \frac{\partial A_y^-}{\partial z} + \frac{1}{\epsilon_0} \frac{\partial F_y^-}{\partial x} \right) e^{-j\beta y} \end{aligned}$$

$$H_x = \sqrt{\frac{\epsilon_0}{\mu_0}} (E_z^+ + E_z^-) = H_x^+ + H_x^-$$

$$\begin{aligned} H_z &= \left(\frac{1}{\mu_0} \frac{\partial A_y^+}{\partial x} - \frac{1}{\sqrt{\mu_0 \epsilon_0}} \frac{\partial F_y^+}{\partial z} \right) e^{-j\beta y} + \left(\frac{1}{\mu_0} \frac{\partial A_y^-}{\partial x} + \frac{1}{\sqrt{\mu_0 \epsilon_0}} \frac{\partial F_y^-}{\partial z} \right) e^{+j\beta y} \\ &= \sqrt{\frac{\epsilon_0}{\mu_0}} \left(\frac{1}{\sqrt{\mu_0 \epsilon_0}} \frac{\partial A_y^+}{\partial x} - \frac{1}{\epsilon_0} \frac{\partial F_y^+}{\partial z} \right) e^{-j\beta y} + \sqrt{\frac{\epsilon_0}{\mu_0}} \left(\frac{1}{\sqrt{\mu_0 \epsilon_0}} \frac{\partial A_y^-}{\partial x} + \frac{1}{\epsilon_0} \frac{\partial F_y^-}{\partial z} \right) e^{+j\beta y} \end{aligned}$$

$$H_z = \sqrt{\frac{\epsilon_0}{\mu_0}} (-E_x^+ + E_x^-) = H_z^+ + H_z^-$$

cont'd.

6.5 cont'd

2. $A_x = A_y = A_z = F_x = F_z = 0, F_y \neq 0, \partial/\partial x \neq 0, \partial/\partial z \neq 0$

For this combination

$$E_y = 0$$

$$H_y = -j\omega F_y - j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 F_y}{\partial y^2} = -j \frac{1}{\omega \mu \epsilon} \left(\frac{\partial^2}{\partial y^2} + \omega^2 \mu \epsilon \right) F_y = 0$$

provided that

$$F_y(x, y, z) = F_y^+(x, z) e^{-j\beta y} + F_y^-(x, z) e^{+j\beta y}$$

Also according to (6-41) and (6-43)

$$E_x = \frac{1}{\epsilon} \frac{\partial F_y^+}{\partial z} e^{-j\beta y} + \frac{1}{\epsilon} \frac{\partial F_y^-}{\partial z} e^{+j\beta y} = E_x^+ + E_x^-$$

$$E_z = -\frac{1}{\epsilon} \frac{\partial F_y^+}{\partial x} e^{-j\beta y} - \frac{1}{\epsilon} \frac{\partial F_y^-}{\partial x} e^{+j\beta y} = E_z^+ + E_z^-$$

$$H_x = \sqrt{\frac{\epsilon}{\mu}} \left(-\frac{1}{\epsilon} \frac{\partial F_y^+}{\partial z} \right) e^{-j\beta y} + \sqrt{\frac{\epsilon}{\mu}} \left(\frac{1}{\epsilon} \frac{\partial F_y^-}{\partial z} \right) e^{+j\beta y} = \sqrt{\frac{\epsilon}{\mu}} (E_x^+) + \sqrt{\frac{\epsilon}{\mu}} (-E_x^-) = H_x^+ + H_x^-$$

$$H_z = \sqrt{\frac{\epsilon}{\mu}} \left(-\frac{1}{\epsilon} \frac{\partial F_y^+}{\partial x} \right) e^{-j\beta y} + \sqrt{\frac{\epsilon}{\mu}} \left(\frac{1}{\epsilon} \frac{\partial F_y^-}{\partial x} \right) e^{+j\beta y} = \sqrt{\frac{\epsilon}{\mu}} (-E_z^+) + \sqrt{\frac{\epsilon}{\mu}} (E_z^-) = H_z^+ + H_z^-$$

3. $A_x = A_z = F_x = F_y = F_z = 0, A_y \neq 0, \partial/\partial x \neq 0, \partial/\partial z \neq 0$

For this combination

$$E_y = -j\omega A_y - j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 A_y}{\partial y^2} = -j \frac{1}{\omega \mu \epsilon} \left(\frac{\partial^2}{\partial y^2} + \omega^2 \mu \epsilon \right) A_y = 0$$

$$H_y = 0$$

provided that

$$A_y(x, y, z) = A_y^+(x, z) e^{-j\beta y} + A_y^-(x, z) e^{+j\beta y}$$

Also according to (6-41) and (6-43)

$$E_x = \left(-\frac{1}{\sqrt{\mu \epsilon}} \frac{\partial A_y^+}{\partial z} \right) e^{-j\beta y} + \left(\frac{1}{\sqrt{\mu \epsilon}} \frac{\partial A_y^-}{\partial z} \right) e^{+j\beta y} = E_x^+ + E_x^-$$

$$E_z = \left(-\frac{1}{\sqrt{\mu \epsilon}} \frac{\partial A_y^+}{\partial x} \right) e^{-j\beta y} + \left(\frac{1}{\sqrt{\mu \epsilon}} \frac{\partial A_y^-}{\partial x} \right) e^{+j\beta y} = E_z^+ + E_z^-$$

$$H_x = \sqrt{\frac{\epsilon}{\mu}} \left(-\frac{1}{\sqrt{\mu \epsilon}} \frac{\partial A_y^+}{\partial z} \right) e^{-j\beta y} + \sqrt{\frac{\epsilon}{\mu}} \left(\frac{1}{\sqrt{\mu \epsilon}} \frac{\partial A_y^-}{\partial z} \right) e^{+j\beta y} = \sqrt{\frac{\epsilon}{\mu}} (E_x^+) + \sqrt{\frac{\epsilon}{\mu}} (-E_x^-) = H_x^+ + H_x^-$$

$$H_z = \sqrt{\frac{\epsilon}{\mu}} \left(\frac{1}{\sqrt{\mu \epsilon}} \frac{\partial A_y^+}{\partial x} \right) e^{-j\beta y} + \sqrt{\frac{\epsilon}{\mu}} \left(\frac{1}{\sqrt{\mu \epsilon}} \frac{\partial A_y^-}{\partial x} \right) e^{+j\beta y} = \sqrt{\frac{\epsilon}{\mu}} (-E_z^+) + \sqrt{\frac{\epsilon}{\mu}} (E_z^-) = H_z^+ + H_z^-$$

6.6

A. TEM⁴ $\Rightarrow E_\phi = H_\phi = 0$

It is apparent that by examining (6-50) and (6-51) that $E_\phi = H_\phi = 0$ can be obtained by any of the following three combinations.

1. $A_\rho = A_z = F_\rho = F_z = 0, A_\phi \neq 0, F_\phi \neq 0, \partial/\partial\rho \neq 0, \partial/\partial z \neq 0$

For this combination, according to (6-50) and (6-51)

$$E_\phi = -j\omega A_\phi - j \frac{1}{\omega\mu\epsilon} \frac{1}{\rho} \frac{\partial}{\partial\rho} \left(\frac{1}{\rho} \frac{\partial A_\phi}{\partial\phi} \right) = -j\omega A_\phi - j \frac{1}{\omega\mu\epsilon} \frac{1}{\rho^2} \frac{\partial^2 A_\phi}{\partial\phi^2} = 0$$

$$\frac{\partial^2 A_\phi}{\partial\phi^2} + \omega^2 \mu\epsilon \rho^2 A_\phi = 0$$

$$H_\phi = -j\omega F_\phi - j \frac{1}{\omega\mu\epsilon} \frac{1}{\rho} \frac{\partial}{\partial\rho} \left(\frac{1}{\rho} \frac{\partial F_\phi}{\partial\phi} \right) = -j\omega F_\phi - j \frac{1}{\omega\mu\epsilon} \frac{1}{\rho^2} \frac{\partial^2 F_\phi}{\partial\phi^2} = 0$$

$$\frac{\partial^2 F_\phi}{\partial\phi^2} + \omega^2 \mu\epsilon \rho^2 F_\phi = 0$$

$$E_\rho = -j \frac{1}{\omega\mu\epsilon} \frac{\partial}{\partial\rho} \left(\frac{1}{\rho} \frac{\partial A_\phi}{} \right) + \frac{1}{\epsilon} \frac{\partial F_\phi}{\partial z}$$

$$E_z = -j \frac{1}{\omega\mu\epsilon} \frac{\partial}{\partial z} \left(\frac{1}{\rho} \frac{\partial A_\phi}{} \right) - \frac{1}{\epsilon} \frac{1}{\rho} \left[\frac{\partial}{\partial\rho} (\rho F_\phi) \right] = -j \frac{1}{\omega\mu\epsilon} \frac{1}{\rho} \frac{\partial^2 A_\phi}{\partial\phi\partial z} - \frac{1}{\epsilon} \frac{1}{\rho} \frac{\partial}{\partial\rho} (\rho F_\phi)$$

$$H_\rho = -j \frac{1}{\omega\mu\epsilon} \frac{\partial}{\partial\rho} \left(\frac{1}{\rho} \frac{\partial F_\phi}{} \right) - \frac{1}{\mu} \frac{\partial A_\phi}{\partial z}$$

$$H_z = -j \frac{1}{\omega\mu\epsilon} \frac{\partial}{\partial z} \left(\frac{1}{\rho} \frac{\partial F_\phi}{} \right) + \frac{1}{\mu} \frac{1}{\rho} \left[\frac{\partial}{\partial\rho} (\rho A_\phi) \right] = -j \frac{1}{\omega\mu\epsilon} \frac{1}{\rho} \frac{\partial^2 F_\phi}{\partial\phi\partial z} + \frac{1}{\mu} \frac{1}{\rho} \frac{\partial}{\partial\rho} (\rho A_\phi)$$

2. $A_\rho = A_\phi = A_z = F_\rho = F_z = 0, F_\phi \neq 0, \partial/\partial\rho \neq 0, \partial/\partial z \neq 0$

For this combination, according to (6-50) and (6-51)

$$E_\phi = 0$$

$$H_\phi = -j\omega F_\phi - j \frac{1}{\omega\mu\epsilon} \frac{1}{\rho^2} \frac{\partial^2 F_\phi}{\partial\phi^2} = 0 \Rightarrow \frac{\partial^2 F_\phi}{\partial\phi^2} + \omega^2 \mu\epsilon \rho^2 F_\phi = 0$$

$$E_\rho = \frac{1}{\epsilon} \frac{\partial F_\phi}{\partial z}$$

$$H_\rho = -j \frac{1}{\omega\mu\epsilon} \frac{\partial}{\partial z} \left(\frac{1}{\rho} \frac{\partial F_\phi}{} \right)$$

$$E_z = -\frac{1}{\epsilon} \frac{1}{\rho} \frac{\partial}{\partial\rho} (\rho F_\phi)$$

$$H_z = -j \frac{1}{\omega\mu\epsilon} \frac{1}{\rho} \frac{\partial^2 F_\phi}{\partial\phi\partial z}$$

3. $A_\rho = A_z = F_\rho = F_z = 0, A_\phi \neq 0, \partial/\partial\rho \neq 0, \partial/\partial z \neq 0$

For this combination, according to (6-50) and (6-51)

$$E_\phi = -j\omega A_\phi - j \frac{1}{\omega\mu\epsilon} \frac{1}{\rho^2} \frac{\partial^2 A_\phi}{\partial\phi^2} = 0 \Rightarrow \frac{\partial^2 A_\phi}{\partial\phi^2} + \omega^2 \mu\epsilon \rho^2 A_\phi = 0 \quad H_\phi = 0$$

$$E_\rho = -j \frac{1}{\omega\mu\epsilon} \frac{2}{\rho} \left(\frac{1}{\rho} \frac{\partial A_\phi}{\partial\phi} \right)$$

$$H_\rho = -\frac{1}{\mu} \frac{\partial A_\phi}{\partial z}$$

$$E_z = -j \frac{1}{\omega\mu\epsilon} \frac{1}{\rho} \frac{\partial^2 A_\phi}{\partial\phi\partial z}$$

Cont'd.

$$H_z = \frac{1}{\mu} \frac{1}{\rho} \frac{2}{\rho} \left(\rho A_\phi \right)$$

6.6 Cont'd. B. TEM^z $\Rightarrow E_z = H_z = 0$

1. $A_p = A_\phi = F_p = F_\phi = 0, A_z \neq 0, F_z \neq 0, \frac{\partial}{\partial p} \neq 0, \frac{\partial}{\partial \phi} \neq 0$

For this combination, according to (6-50) and (6-51)

$$E_z = -j\omega A_z - j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 A_z}{\partial z^2} = 0 \Rightarrow \frac{\partial^2 A_z}{\partial z^2} + \omega^2 \mu \epsilon A_z = 0$$

$$H_z = -j\omega F_z - j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 F_z}{\partial z^2} = 0 \Rightarrow \frac{\partial^2 F_z}{\partial z^2} + \omega^2 \mu \epsilon F_z = 0$$

$$E_p = -j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 A_z}{\partial p \partial z} - \frac{1}{\epsilon p} \frac{\partial F_z}{\partial \phi}$$

$$E_\phi = -j \frac{1}{\omega \mu \epsilon} \frac{1}{p} \frac{\partial^2 A_z}{\partial \phi \partial z} + \frac{1}{\epsilon} \frac{\partial F_z}{\partial p}$$

$$H_p = -j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 F_z}{\partial p \partial z} + \frac{1}{\epsilon p} \frac{\partial A_z}{\partial \phi}$$

$$H_\phi = -j \frac{1}{\omega \mu \epsilon} \frac{1}{p} \frac{\partial^2 F_z}{\partial \phi \partial z} - \frac{1}{\epsilon} \frac{\partial A_z}{\partial p}$$

2. $A_p = A_\phi = A_z = F_p = F_\phi = 0, F_z \neq 0, \frac{\partial}{\partial p} \neq 0, \frac{\partial}{\partial \phi} \neq 0$

For this combination, according to (6-50) and (6-51)

$$E_z = 0$$

$$H_z = -j\omega F_z - j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 F_z}{\partial z^2} = 0 \Rightarrow \frac{\partial^2 F_z}{\partial z^2} + \omega^2 \mu \epsilon F_z = 0$$

$$E_p = - \frac{1}{\epsilon p} \frac{\partial F_z}{\partial \phi}$$

$$E_\phi = \frac{1}{\epsilon} \frac{\partial F_z}{\partial p}$$

$$H_p = -j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 F_z}{\partial p \partial z}$$

$$H_\phi = -j \frac{1}{\omega \mu \epsilon} \frac{1}{p} \frac{\partial^2 F_z}{\partial \phi \partial z}$$

3. $A_p = A_\phi = F_p = F_\phi = 0, A_z \neq 0, \frac{\partial}{\partial p} \neq 0, \frac{\partial}{\partial \phi} \neq 0$

For this combination, according to (6-50) and (6-51)

$$E_z = -j\omega A_z - j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 A_z}{\partial z^2} = 0 \Rightarrow \frac{\partial^2 A_z}{\partial z^2} + \omega^2 \mu \epsilon A_z = 0$$

$$H_z = 0$$

$$E_p = -j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 A_z}{\partial p \partial z}$$

$$E_\phi = -j \frac{1}{\omega \mu \epsilon} \frac{1}{p} \frac{\partial^2 A_z}{\partial \phi \partial z}$$

$$H_p = \frac{1}{\epsilon p} \frac{\partial A_z}{\partial \phi}$$

$$H_\phi = -\frac{1}{\epsilon} \frac{\partial A_z}{\partial p}$$

6.7

A. TM^x $\Rightarrow H_x = 0$

It is apparent by examining (6-41) and (6-43) that $H_x = 0$ can be obtained by letting

$$A_y = A_z = F_x = F_y = F_z = 0 \text{ and } A_x \neq 0.$$

When these are substituted into (6-41) and (6-43), we get (6-61).

B. TM^y $\Rightarrow H_y = 0$

It is apparent by examining (6-41) and (6-43) that $H_y = 0$ can be obtained by letting

$$A_x = A_z = F_x = F_y = F_z = 0$$

When these are substituted into (6-41) and (6-43), we get (6-64).

6.8

A. TM^x $\Rightarrow H_x = 0$

Let $E_x \neq 0$ so that $\nabla^2 E_x + \beta^2 E_x = 0$. Solve this for E_x .

According to (6-61) E_x is proportional to A_x . Therefore

$$A_x = \frac{E_x}{-j \frac{1}{\omega \mu \epsilon} \left(\frac{\partial^2}{\partial x^2} + \beta^2 \right) E_x} E_x = C_x E_x \text{ where } C_x = j \frac{E_x}{\frac{1}{\omega \mu \epsilon} \left(\frac{\partial^2}{\partial x^2} + \beta^2 \right) E_x}$$

Thus we can write the remaining E-and H-field components as

$$E_y = -j \frac{C_x}{\omega \mu \epsilon} \frac{\partial^2 E_x}{\partial x \partial y} \quad H_y = \frac{C_x}{\mu} \frac{\partial E_x}{\partial z}$$

$$E_z = -j \frac{C_x}{\omega \mu \epsilon} \frac{\partial^2 E_x}{\partial x \partial z} \quad H_z = -\frac{C_x}{\mu} \frac{\partial E_x}{\partial y}$$

B. TM^y $\Rightarrow H_y = 0$

Let $E_y \neq 0$ so that $\nabla^2 E_y + \beta^2 E_y = 0$. Solve this for E_y .

According to (6-64) E_y is proportional to A_y . Therefore

$$A_y = \frac{E_y}{-j \frac{1}{\omega \mu \epsilon} \left(\frac{\partial^2}{\partial y^2} + \beta^2 \right) E_y} E_y = C_y E_y \text{ where } C_y = j \frac{E_y}{\frac{1}{\omega \mu \epsilon} \left(\frac{\partial^2}{\partial y^2} + \beta^2 \right) E_y}$$

Thus we can write the remaining E-and H-field components as

$$E_x = -j \frac{C_y}{\omega \mu \epsilon} \frac{\partial^2 E_y}{\partial x \partial y} \quad H_x = -\frac{C_y}{\mu} \frac{\partial E_y}{\partial z}$$

$$E_z = -j \frac{C_y}{\omega \mu \epsilon} \frac{\partial^2 E_y}{\partial y \partial z} \quad H_z = \frac{C_y}{\mu} \frac{\partial E_y}{\partial x}$$

cont'd.

6.8 Cont'd. C. TM² $\Rightarrow H_z = 0$

Let $E_z \neq 0$ so that $\nabla^2 E_z + \beta^2 E_z = 0$. Solve this for E_z .

According to (6-59) E_z is proportional to A_z . Therefore

$$A_z = \frac{E_z}{-j \frac{1}{\omega \mu \epsilon} \left(\frac{\partial^2}{\partial z^2} + \beta^2 \right) E_z} E_z = C_z E_z \text{ where } C_z = -j \frac{1}{\omega \mu \epsilon} \left(\frac{\partial^2}{\partial z^2} + \beta^2 \right) E_z$$

Thus we can write the remaining E - and H -field components as

$$E_x = -j \frac{C_z}{\omega \mu \epsilon} \frac{\partial^2 E_z}{\partial x \partial z} \quad H_x = \frac{C_z}{\mu} \frac{\partial E_z}{\partial y}$$

$$E_y = -j \frac{C_z}{\omega \mu \epsilon} \frac{\partial^2 E_z}{\partial y \partial z} \quad H_y = -\frac{C_z}{\mu} \frac{\partial E_z}{\partial x}$$

6.9

$$TE^2: E_z = 0, H_z(x, y, z) = f(x)g(y)h(z) = f(x)g(y)e^{-j\beta_z z}$$

$$\nabla \times \underline{E} = -j\omega \underline{H} \Rightarrow \underline{H} = -\frac{1}{j\omega \mu} \nabla \times \underline{E} = -\frac{1}{j\omega \mu} \begin{vmatrix} \hat{a}_x & \hat{a}_y & \hat{a}_z \\ \frac{\partial}{\partial x} & \frac{\partial}{\partial y} & \frac{\partial}{\partial z} \\ E_x & E_y & 0 \end{vmatrix}$$

$$\underline{H} = -\frac{1}{j\omega \mu} \left[\hat{a}_x \frac{\partial E_y}{\partial z} + \hat{a}_y \frac{\partial E_x}{\partial z} + \hat{a}_z \left(\frac{\partial E_y}{\partial x} - \frac{\partial E_x}{\partial y} \right) \right]$$

$$H_x = \frac{1}{j\omega \mu} \frac{\partial E_y}{\partial z} - \frac{-j\beta_z}{j\omega \mu} E_y = \frac{-\beta_z}{\omega \mu} E_y \quad (1)$$

$$H_y = -\frac{1}{j\omega \mu} \frac{\partial E_x}{\partial z} = -\frac{j\beta_z}{j\omega \mu} E_x = +\frac{\beta_z}{\omega \mu} E_x \quad (2)$$

$$H_z = \left(\frac{\partial E_y}{\partial x} - \frac{\partial E_x}{\partial y} \right) \frac{1}{j\omega \mu} \quad (3)$$

$$\nabla \times \underline{H} = +j\omega \epsilon \underline{E} \Rightarrow \underline{E} = \frac{1}{j\omega \epsilon} \nabla \times \underline{H} = \frac{1}{j\omega \epsilon} \begin{vmatrix} \hat{a}_x & \hat{a}_y & \hat{a}_z \\ \frac{\partial}{\partial x} & \frac{\partial}{\partial y} & \frac{\partial}{\partial z} \\ H_x & H_y & H_z \end{vmatrix}$$

$$\underline{E} = \frac{1}{j\omega \epsilon} \left[\hat{a}_x \left(\frac{\partial H_z}{\partial y} - \frac{\partial H_y}{\partial z} \right) + \hat{a}_y \left(\frac{\partial H_x}{\partial z} - \frac{\partial H_z}{\partial x} \right) + \hat{a}_z \left(\frac{\partial H_y}{\partial x} - \frac{\partial H_x}{\partial y} \right) \right]$$

$$E_x = \frac{1}{j\omega \epsilon} \left(\frac{\partial H_z}{\partial y} - \frac{\partial H_y}{\partial z} \right) = \frac{1}{j\omega \epsilon} \left(\frac{\partial H_z}{\partial y} + j\beta_z \frac{H_y}{\beta_z} \right) \quad (4)$$

$$E_y = \frac{1}{j\omega \epsilon} \left(\frac{\partial H_x}{\partial z} - \frac{\partial H_z}{\partial x} \right) = \frac{1}{j\omega \epsilon} \left(-j\beta_z \frac{H_x}{\beta_z} - \frac{\partial H_z}{\partial x} \right) \quad (5)$$

$$E_z = \frac{1}{j\omega \epsilon} \left(\frac{\partial H_y}{\partial x} - \frac{\partial H_x}{\partial y} \right) = 0 \quad (6)$$

Cont'd.

6.9 cont'd

Substitute (4) into (2) and using that $\beta^2 = \omega^2 \mu \epsilon = \beta_c^2 + \beta_z^2 \Rightarrow \beta - \beta_z^2 = \beta_c^2$

$$H_y = + \frac{\beta_z}{\omega \mu} E_x = \frac{\beta_z}{\omega \mu} \left[\frac{1}{j \omega \epsilon} \left(\frac{\partial H_z}{\partial y} + j \beta_z H_y \right) \right] = -j \frac{\beta_z}{\omega^2 \mu \epsilon} \left(\frac{\partial H_z}{\partial y} + j \beta_z H_y \right)$$

$$\Rightarrow -j \frac{\beta_z}{\omega^2 \mu \epsilon} \frac{\partial H_z}{\partial y} + \frac{\beta_z^2}{\omega^2 \mu \epsilon} H_y = -j \frac{\beta_z}{\beta_c^2} \frac{\partial H_z}{\partial y} + \frac{\beta_z^2}{\beta_c^2} H_y$$

$$H_y \left(1 - \frac{\beta_z^2}{\beta_c^2} \right) = H_y \left(\frac{\beta_c^2 - \beta_z^2}{\beta_c^2} \right) = H_y \left(\frac{\beta_c^2}{\beta_c^2} \right) = -j \frac{\beta_z}{\beta_c^2} \frac{\partial H_z}{\partial y}$$

$$H_y = \frac{\beta_c^2}{\beta_c^2} \left(-j \frac{\beta_z}{\beta_c^2} \right) \frac{\partial H_z}{\partial y} = -j \frac{\beta_z}{\beta_c^2} \frac{\partial H_z}{\partial y}$$

$$\boxed{H_y = -j \frac{\beta_z}{\beta_c^2} \frac{\partial H_z}{\partial y}} \quad (7)$$

Substitute (5) into (1), and using that $\beta^2 = \omega^2 \mu \epsilon = \beta_c^2 + \beta_z^2 \Rightarrow \beta - \beta_z^2 = \beta_c^2$

$$H_x = - \frac{\beta_z}{\omega \mu} E_y = - \frac{\beta_z}{\omega \mu} \left[\frac{1}{j \omega \epsilon} \left(-j \beta_z H_x - \frac{\partial H_z}{\partial x} \right) \right] = +j \frac{\beta_z}{\omega^2 \mu \epsilon} \left(-j \beta_z H_x - \frac{\partial H_z}{\partial x} \right)$$

$$= \frac{\beta_z^2}{\omega^2 \mu \epsilon} H_x - j \frac{\beta_z}{\omega^2 \mu \epsilon} \frac{\partial H_z}{\partial x} - \frac{\beta_z^2}{\beta_c^2} H_x - j \frac{\beta_z}{\beta_c^2} \frac{\partial H_z}{\partial x}$$

$$H_x \left(1 - \frac{\beta_z^2}{\beta_c^2} \right) = H_x \left(\frac{\beta_c^2 - \beta_z^2}{\beta_c^2} \right) = H_x \left(\frac{\beta_c^2}{\beta_c^2} \right) = -j \frac{\beta_z}{\beta_c^2} \frac{\partial H_z}{\partial x}$$

$$H_x = \frac{\beta_c^2}{\beta_c^2} \left(-j \frac{\beta_z}{\beta_c^2} \right) \frac{\partial H_z}{\partial x} = -j \frac{\beta_z}{\beta_c^2} \frac{\partial H_z}{\partial x}$$

$$\boxed{H_x = -j \frac{\beta_z}{\beta_c^2} \frac{\partial H_z}{\partial x}} \quad (8)$$

Cont'd.

6.9 cont'd

Substitute (7) into (4)

$$F_x = \frac{1}{j\omega\epsilon} \left(\frac{\partial H_z}{\partial y} + j\beta_z H_y \right) = \frac{1}{j\omega\epsilon} \left[\frac{\partial H_z}{\partial y} + j\beta_z \left(-j \frac{\beta_z}{\beta_c^2} \right) \frac{\partial H_z}{\partial y} \right]$$

$$= \frac{1}{j\omega\epsilon} \left[\frac{\partial H_z}{\partial y} + \frac{\beta_z^2}{\beta_c^2} \frac{\partial H_z}{\partial y} \right] = \frac{1}{j\omega\epsilon} \left(1 + \frac{\beta_z^2}{\beta_c^2} \right) \frac{\partial H_z}{\partial y} = \frac{1}{j\omega\epsilon} \left(\frac{\beta_z^2 + \beta_c^2}{\beta_c^2} \right) \frac{\partial H_z}{\partial y}$$

$$F_x = \frac{1}{j\omega\epsilon} \frac{\beta_z^2}{\beta_c^2} \frac{\partial H_z}{\partial y} = \frac{\omega^2 \mu \epsilon}{j\omega\epsilon} \frac{1}{\beta_c^2} \frac{\partial H_z}{\partial y} = -j \frac{\omega \mu}{\beta_c^2} \frac{\partial H_z}{\partial y}$$

$F_x = -j \frac{\omega \mu}{\beta_c^2} \frac{\partial H_z}{\partial y}$

(9)

Substitute (8) into (5)

$$F_y = \frac{1}{j\omega\epsilon} \left(-j\beta_z H_x - \frac{\partial H_z}{\partial x} \right) = \frac{1}{j\omega\epsilon} \left[-j\beta_z \left(-j \frac{\beta_z}{\beta_c^2} \frac{\partial H_z}{\partial x} \right) - \frac{\partial H_z}{\partial x} \right]$$

$$= \frac{1}{j\omega\epsilon} \left[-\frac{\beta_z^2}{\beta_c^2} \frac{\partial H_z}{\partial x} - \frac{\partial H_z}{\partial x} \right] = -\frac{1}{j\omega\epsilon} \left(\frac{\beta_z^2 + 1}{\beta_c^2} \right) \frac{\partial H_z}{\partial x} = -\frac{1}{j\omega\epsilon} \left(\frac{\beta_z^2 + \beta_c^2}{\beta_c^2} \right) \frac{\partial H_z}{\partial x}$$

$$F_y = -\frac{1}{j\omega\epsilon} \frac{\beta_z^2}{\beta_c^2} \frac{\partial H_z}{\partial x} = \frac{1}{j\omega\epsilon} \frac{\omega^2 \mu \epsilon}{\beta_c^2} \frac{\partial H_z}{\partial x} = +j \frac{\omega \mu}{\beta_c^2} \frac{\partial H_z}{\partial x}$$

$F_y = +j \frac{\omega \mu}{\beta_c^2} \frac{\partial H_z}{\partial x}$

(10)

cont'd

6.9 cont'd

SUMMARY

TE^z:

$$E_x = -j \frac{w\epsilon}{\beta_c^2} \frac{\partial H_z}{\partial y}$$

$$E_y = +j \frac{w\epsilon}{\beta_c^2} \frac{\partial H_z}{\partial x}$$

$$E_{z=0} = \frac{1}{jw\epsilon} \left(\frac{\partial H_y}{\partial x} - \frac{\partial H_x}{\partial y} \right) \Rightarrow \frac{\partial H_x}{\partial y} = \frac{\partial H_y}{\partial x}$$

$$H_x = -j \frac{\beta_z}{\beta_c^2} \frac{\partial H_z}{\partial x}$$

$$H_y = -j \frac{\beta_z}{\beta_c^2} \frac{\partial H_z}{\partial y}$$

$$H_x = h_x(x, y) e^{-j\beta_z z} = f(x) g(y) e^{-j\beta_z z}$$

$$\boxed{\nabla^2 H_z + \beta^2 H_z = (\nabla^2 + \beta^2) H_z(x, y, z) = 0}$$

6.10

TM^z: It can be shown that for TM^z $[H_z = 0, E_z(x, y, z) = e(x, y) e^{-j\beta_z z}]$

$$E_x = -j \frac{\beta_z}{\beta_c^2} \frac{\partial E_z}{\partial x}, \quad E_y = -j \frac{\beta_z}{\beta_c^2} \frac{\partial E_z}{\partial y}$$

$$H_x = j \frac{w\epsilon}{\beta_c^2} \frac{\partial E_z}{\partial y}, \quad H_y = -j \frac{w\epsilon}{\beta_c^2} \frac{\partial E_z}{\partial x}$$

$$E_z(x, y, z) = e(x, y) e^{-j\beta_z z} = f(x) g(y) e^{-j\beta_z z}$$

$$\boxed{\nabla^2 E_z + \beta^2 E_z = (\nabla^2 + \beta^2) E_z(x, y, z) = 0}$$

6.11 $TM^2 \Rightarrow H_z = 0$

Let $E_z \neq 0$ so that $\nabla^2 E_z + \beta^2 E_z = 0$. Solve this for E_z . Then according to (6-70) E_z is proportional to A_z . Therefore

$$A_z = \frac{E_z}{-j \frac{1}{\omega \mu \epsilon} \left(\frac{\partial^2}{\partial z^2} + \beta^2 \right) E_z} E_z = C_z E_z \text{ where } C_z = j \frac{1}{\omega \mu \epsilon} \frac{E_z}{\left(\frac{\partial^2}{\partial z^2} + \beta^2 \right) E_z}$$

Thus we can write the remaining E - and H -field components in terms of E_z as

$$E_y = -j \frac{C_z}{\omega \mu \epsilon} \frac{\partial^2 E_z}{\partial y \partial z} \quad H_y = \frac{1}{\mu} \frac{C_z}{\beta} \frac{\partial E_z}{\partial y}$$

$$E_\phi = -j \frac{C_z}{\omega \mu \epsilon} \frac{1}{\beta} \frac{\partial^2 E_z}{\partial \phi \partial z} \quad H_\phi = -\frac{C_z}{\mu} \frac{\partial E_z}{\partial \phi}$$

6.12 $TE^2 \Rightarrow E_z = 0$

It is apparent by examining (6-41) and (6-43) that $E_z = 0$ can be obtained by letting

$$A_x = A_y = A_z = F_x = F_y = 0, \quad F_z \neq 0$$

When these are substituted into (6-41) and (6-43), we get (6-72).

B. $TE^2 \Rightarrow E_x = 0$

It is apparent by examining (6-41) and (6-43) that $E_x = 0$ can be obtained by letting

$$A_x = A_y = A_z = F_y = F_z = 0, \quad F_x \neq 0$$

When these are substituted into (6-41) and (6-43), we get (6-74).

C. $TE^2 \Rightarrow E_y = 0$

It is apparent by examining (6-41) and (6-43) that $E_y = 0$ can be obtained by letting

$$A_x = A_y = A_z = F_x = F_z = 0, \quad F_y \neq 0$$

When these are substituted into (6-41) and (6-43), we get (6-77).

6.13 $TE^z \Rightarrow E_z = 0$

It is apparent by examining (6-50) and (6-51) that $E_z = 0$ can be obtained by letting

$$A_x = A_y = A_z = F_x = F_y = 0, \quad F_z \neq 0$$

When these are substituted into (6-50) and (6-51), we get (6-80).

6.14 A. $TE^x \Rightarrow E_x = 0$

Let $H_x \neq 0$ so that $\nabla^2 H_x + \beta^2 H_x = 0$. Solve this for H_x . Then according to (6-74) H_x is proportional to F_x . Therefore

$$F_x = \frac{H_x}{-j \frac{1}{\omega \mu \epsilon} \left(\frac{\partial^2}{\partial x^2} + \beta^2 \right) H_x} \quad H_x = D_x H_x \text{ where } D_x = j \frac{H_x}{\frac{1}{\omega \mu \epsilon} \left(\frac{\partial^2}{\partial x^2} + \beta^2 \right) H_x}$$

Thus we can write the remaining E - and H -field components in terms of H_x as

$$E_y = -\frac{D_x}{\epsilon} \frac{\partial H_x}{\partial z} \quad H_y = -j \frac{D_x}{\omega \mu \epsilon} \frac{\partial^2 H_x}{\partial x \partial y}$$

$$E_z = \frac{D_x}{\epsilon} \frac{\partial H_x}{\partial y} \quad H_z = -j \frac{D_x}{\omega \mu \epsilon} \frac{\partial^2 H_x}{\partial x \partial z}$$

B. $TE^y \Rightarrow E_y = 0$

Let $H_y \neq 0$ so that $\nabla^2 H_y + \beta^2 H_y = 0$. Solve this for H_y . Then according to (6-77) H_y is proportional to F_y . Therefore

$$F_y = \frac{H_y}{-j \frac{1}{\omega \mu \epsilon} \left(\frac{\partial^2}{\partial y^2} + \beta^2 \right) H_y} \quad H_y = D_y H_y \text{ where } D_y = j \frac{H_y}{\frac{1}{\omega \mu \epsilon} \left(\frac{\partial^2}{\partial y^2} + \beta^2 \right) H_y}$$

Thus we can write the remaining E - and H -field components in terms of H_y as

$$E_x = \frac{D_y}{\epsilon} \frac{\partial H_y}{\partial z} \quad H_x = -j \frac{D_y}{\omega \mu \epsilon} \frac{\partial^2 H_y}{\partial x \partial y}$$

$$E_z = -\frac{D_y}{\epsilon} \frac{\partial H_y}{\partial x} \quad H_z = -j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 H_y}{\partial y \partial z}$$

Cont'd.

6.14 Cont'd.

C. $\text{TE}^2 \Rightarrow E_z = 0$

Let $H_z \neq 0$ so that $\nabla^2 H_z + \beta^2 H_z = 0$. Solve this for H_z . Then according to (6-72) H_z is proportional to F_z . Therefore

$$F_z = \frac{H_z}{-j \frac{1}{\omega \mu \epsilon} \left(\frac{\partial^2}{\partial z^2} + \beta^2 \right) H_z} H_z = D_z H_z \text{ where } D_z = j \frac{H_z}{\frac{1}{\omega \mu \epsilon} \left(\frac{\partial^2}{\partial z^2} + \beta^2 \right) H_z}$$

Thus we can write the remaining E - and H -field components in terms of H_z as

$$E_x = -\frac{D_z}{\epsilon} \frac{\partial H_z}{\partial y} \quad H_x = -j \frac{D_z}{\omega \mu \epsilon} \frac{\partial^2 H_z}{\partial x \partial z}$$

$$E_y = \frac{D_z}{\epsilon} \frac{\partial H_z}{\partial x} \quad H_y = -j \frac{D_z}{\omega \mu \epsilon} \frac{\partial^2 H_z}{\partial y \partial z}$$

6.15

$\text{TE}^2 \Rightarrow E_z = 0$

Let $H_z \neq 0$ so that $\nabla^2 H_z + \beta^2 H_z = 0$. Solve this for H_z . Then according to (6-80) H_z is proportional to F_z . Therefore

$$F_z = \frac{H_z}{-j \frac{1}{\omega \mu \epsilon} \left(\frac{\partial^2}{\partial z^2} + \beta^2 \right) H_z} H_z = D_z H_z \text{ where } D_z = j \frac{H_z}{\frac{1}{\omega \mu \epsilon} \left(\frac{\partial^2}{\partial z^2} + \beta^2 \right) H_z}$$

Thus we can write the remaining E - and H -field components in terms of H_z as

$$E_\phi = -\frac{D_z}{\epsilon \rho} \frac{\partial H_z}{\partial \phi} \quad H_\phi = -j \frac{D_z}{\omega \mu \epsilon} \frac{\partial^2 H_z}{\partial \rho \partial z}$$

$$E_\theta = \frac{D_z}{\epsilon} \frac{\partial H_z}{\partial \rho} \quad H_\theta = -j \frac{D_z}{\omega \mu \epsilon} \frac{1}{\rho} \frac{\partial^2 H_z}{\partial \theta \partial z}$$

6.16

$$A_{z1} = C_1 \frac{e^{-j\beta r}}{r}, A_{z2} = C_2 \frac{e^{+j\beta r}}{r}$$

$$\frac{dA_{z1}}{dr} = C_1 \left[-j\beta \frac{e^{-j\beta r}}{r} - \frac{e^{-j\beta r}}{r^2} \right], \quad \frac{dA_{z2}}{dr} = C_2 \left[j\beta \frac{e^{+j\beta r}}{r} - \frac{e^{+j\beta r}}{r^2} \right]$$

$$\frac{d^2 A_{z1}}{dr^2} = C_1 \left[-\beta^2 \frac{e^{-j\beta r}}{r} + j\beta \frac{e^{-j\beta r}}{r^2} + j\beta \frac{e^{-j\beta r}}{r^2} + 2 \frac{e^{-j\beta r}}{r^3} \right]$$

$$= C_1 \left[-\beta^2 \frac{e^{-j\beta r}}{r} + j2\beta \frac{e^{-j\beta r}}{r^2} + 2 \frac{e^{-j\beta r}}{r^3} \right]$$

$$\frac{d^2 A_{z2}}{dr^2} = C_2 \left[-\beta^2 \frac{e^{+j\beta r}}{r} - j\beta \frac{e^{+j\beta r}}{r^2} - j\beta \frac{e^{+j\beta r}}{r^2} + 2 \frac{e^{+j\beta r}}{r^3} \right]$$

$$= C_2 \left[-\beta^2 \frac{e^{+j\beta r}}{r} - j2\beta \frac{e^{+j\beta r}}{r^2} + 2 \frac{e^{+j\beta r}}{r^3} \right]$$

Substituting each of these into (6-84a) balances the equation.

6.17

a. $A_z = \frac{k}{4\pi} \iiint \frac{J_z}{r} dv'$, $\nabla^2 A_z = \frac{1}{r^2} \frac{\partial}{\partial r} \left(r^2 \frac{\partial A_z}{\partial r} \right)$

When $r \neq 0$ and $\underline{r} \notin V'$, then

$$\begin{cases} \frac{\partial A_z}{\partial r} = -\frac{k}{4\pi} \iiint \frac{J_z}{r^2} dv' \\ r^2 \frac{\partial A_z}{\partial r} = -\frac{k}{4\pi} \iiint J_z dv' \\ \frac{\partial}{\partial r} \left(r^2 \frac{\partial A_z}{\partial r} \right) = 0 \end{cases}$$

Therefore when $\underline{r} \neq \underline{r}'$ ($\underline{r} \notin V'$) $\nabla^2 A_z = 0$. However when $\underline{r} \rightarrow \underline{r}'$

$$\nabla^2 A_z(x, y, z) = \frac{k}{4\pi} \iiint \nabla \cdot \left(\nabla \frac{1}{r} \right) J_z(x', y', z') dv' = \frac{k}{4\pi} \iiint \nabla \cdot \left(-\frac{r}{r^3} \hat{r} \right) J_z(x', y', z') dx' dy' dz'$$

$\underline{r} \in S_e$
 $\epsilon \rightarrow 0$

where $S_\epsilon = \{ \underline{r}' : |\underline{r}' - \underline{r}| \leq \epsilon \}$

Thus

$$\nabla^2 A_z = -\frac{k}{4\pi} \iint_S \frac{r}{r^3} \cdot \hat{r} J_z(x', y', z') r^2 \sin\theta d\theta d\phi = -\frac{k}{4\pi} \int_0^{2\pi} \int_0^\pi J_z(x, y, z) \sin\theta d\theta d\phi$$

$$\nabla^2 A_z = -\frac{k}{4\pi} J_z(2)(2\pi) = -k J_z \Rightarrow \nabla^2 A_z = -k J_z$$

cont'd.

6.17 cont'd.

b. $A_z = \frac{\mu}{4\pi} \iiint J_z \frac{e^{-j\beta r}}{r} dv' , \nabla^2 A_z + \beta^2 A_z = -\mu J_z$

When $r \neq 0$ and $v \neq v'$.

$$(\nabla^2 + \beta^2) \frac{e^{-j\beta r}}{r} = \beta^2 \frac{e^{-j\beta r}}{r} + \frac{1}{r^2} \frac{\partial}{\partial r} \left[r^2 \frac{\partial}{\partial r} \left(\frac{e^{-j\beta r}}{r} \right) \right]$$

$$= \beta^2 \frac{e^{-j\beta r}}{r} + \frac{1}{r^2} \frac{\partial}{\partial r} \left[-(1+j\beta r) e^{-j\beta r} \right]$$

$$(\nabla^2 + \beta^2) \frac{e^{-j\beta r}}{r} = \beta^2 \frac{e^{-j\beta r}}{r} + \frac{1}{r^2} \left[-j\beta e^{-j\beta r} + j\beta e^{-j\beta r} - \beta^2 r e^{-j\beta r} \right] = 0$$

Thus when $r \neq 0$ and $v \neq v'$

$$\nabla^2 A_z + \beta^2 A_z = \frac{\mu}{4\pi} (\nabla^2 + \beta^2) \iiint J_z \frac{e^{-j\beta r}}{r} dv' = \frac{\mu}{4\pi} \iiint J_z \left[(\nabla^2 + \beta^2) \frac{e^{-j\beta r}}{r} \right] dv' = 0$$

When $r \rightarrow r'$

$$\begin{aligned} \nabla^2 A_z + \beta^2 A_z &= \lim_{r \rightarrow 0} \left\{ \frac{\mu}{4\pi} \iiint_{S_r} ((\nabla^2 + \beta^2) \frac{e^{-j\beta r}}{r}) J_z(x, y, z) dv' \right\} \\ &= \lim_{r \rightarrow 0} \int_0^{2\pi} \int_0^\pi \int_0^\pi \nabla \cdot \left(-\frac{r}{r^3} \right) e^{-j\beta r} J_z(x, y, z) r^2 \sin\theta d\theta d\phi \\ &\quad \times \frac{\mu}{4\pi} J_z(x, y, z) (z)(2\pi) \end{aligned}$$

$$\nabla^2 A_z + \beta^2 A_z = -\frac{\mu}{4\pi} \int_0^{2\pi} \int_0^\pi J_z(x, y, z) \sin\theta d\theta d\phi = -\frac{\mu}{4\pi} J_z(x, y, z) (z)(2\pi)$$

$$\nabla^2 A_z + \beta^2 A_z = -\frac{\mu}{4\pi} J_z (4\pi) = -\mu J_z$$

Thus

$$\nabla^2 A_z + \beta^2 A_z = -\mu J_z$$

$$6.18 \quad H_r = H_\theta = 0, \quad H_\phi = j \frac{\beta I_0 l \sin\theta}{4\pi r} \left(1 + \frac{1}{j\beta r} \right) e^{-j\beta r}$$

$$\nabla \times \underline{H} = \hat{a}_r \left[\frac{2}{r \sin\theta} (H_\phi \sin\theta) \right] - \frac{\hat{a}_\theta}{r} \left[\frac{2}{\sin\theta} (r H_\phi) \right]$$

$$\frac{2}{\sin\theta} (H_\phi \sin\theta) = 2 \sin\theta \cos\theta H_\phi$$

$$\frac{1}{r \sin\theta} \frac{2}{\sin\theta} (H_\phi \sin\theta) = j \frac{\beta I_0 l \cos\theta}{2\pi r^2} \left(1 + \frac{1}{j\beta r} \right) e^{-j\beta r}$$

$$r H_\phi = j \frac{\beta I_0 l \sin\theta}{4\pi} \left(1 + \frac{1}{j\beta r} \right) e^{-j\beta r}$$

$$\begin{aligned} \frac{2}{\sin\theta} (r H_\phi) &= j \frac{\beta I_0 l \sin\theta}{4\pi} \left\{ -j\beta \left(1 + \frac{1}{j\beta r} \right) e^{-j\beta r} + \left(-\frac{1}{j\beta r^2} \right) e^{-j\beta r} \right\} \\ &= j \frac{\beta I_0 l \sin\theta}{4\pi} \left\{ -j\beta \left(1 + \frac{1}{j\beta r} - \frac{1}{(\beta r)^2} \right) e^{-j\beta r} \right\} \end{aligned}$$

$$\frac{1}{r} \frac{2}{\sin\theta} (r H_\phi) = \frac{\beta^2 I_0 l \sin\theta}{4\pi r} \left[1 + \frac{1}{j\beta r} - \frac{1}{(\beta r)^2} \right] e^{-j\beta r}$$

Using the above equations we can write that

$$\underline{E} = \frac{1}{j\omega\epsilon} \nabla \times \underline{H} = \frac{1}{j\omega\epsilon} \left\{ \frac{\hat{a}_r}{r \sin\theta} \left[\frac{2}{\sin\theta} (H_\phi \sin\theta) \right] - \frac{\hat{a}_\theta}{r} \left[\frac{2}{\sin\theta} (r H_\phi) \right] \right\}$$

is equal to

$$E_r = \eta \frac{\beta I_0 l \cos\theta}{2\pi r^2} \left[1 + \frac{1}{j\beta r} \right] e^{-j\beta r}$$

$$E_\theta = j\eta \frac{\beta I_0 l \sin\theta}{4\pi r} \left[1 + \frac{1}{j\beta r} - \frac{1}{(\beta r)^2} \right] e^{-j\beta r}, \quad E_\phi = 0$$

6.19 When $\beta r \gg 1$ the most dominant term in each of the above expressions is the first one. Also since the most dominant term for E_r varies as $1/r^2$ while for E_θ varies as $1/r$, the E_r component is very small compared to E_θ so that is set approximately equal to zero. Therefore

$$E_r \approx 0, \quad E_\theta = H_r = H_\theta = 0$$

$$E_\theta \approx j\eta \frac{\beta I_0 l e^{-j\beta r}}{4\pi r} \sin\theta, \quad H_\phi \approx E_\theta / \eta$$

6.20

$$\begin{aligned} \bullet S_{ave} &= \frac{1}{2} \operatorname{Re}(E \times H^*) = \frac{1}{2} \operatorname{Re} \left[\hat{a}_\theta E_\theta \times (\hat{a}_\theta E_\theta)^* \right] = \hat{a}_r \frac{1}{2\eta} |E_\theta|^2 \\ &= \frac{\hat{a}}{2\eta} \int_0^{2\pi} \int_{-\pi/2}^{\pi/2} \beta I_0 l e^{-j\beta r} \sin\theta \left[\frac{\eta \beta^2 |I_0 l|^2}{32\pi^2 r^2} \sin^2\theta \right] d\theta = \hat{a}_r \eta \frac{(4\pi^2/3)}{32\pi^2 r^2} |I_0 l|^2 \sin^2\theta \end{aligned}$$

$$S_{ave} = \hat{a}_r \frac{\eta}{8} \frac{|I_0 l|^2}{r^2} \sin^2\theta$$

$$\bullet U = r^2 S_{ave} = \frac{\eta}{8} \frac{|I_0 l|^2}{r^2} \sin^2\theta$$

$$\bullet P_{rad} = \int_0^{2\pi} \int_0^{\pi} U(\theta, \phi) \sin\theta d\theta d\phi = \frac{\eta}{8} \frac{|I_0 l|^2}{r^2} \int_0^{2\pi} d\phi \int_0^{\pi} \sin^3\theta d\theta = \frac{\eta}{8} \frac{|I_0 l|^2}{r^2} (2\pi) \left(\frac{4}{3} \right)$$

$$P_{rad} = \eta \left(\frac{\pi}{3} \right) \frac{|I_0 l|^2}{r^2}$$

$$\bullet U = \frac{\eta}{8} \frac{|I_0 l|^2}{r^2} \sin^2\theta \Rightarrow U_{max} \Big|_{\theta=90^\circ} = \frac{\eta}{8} \frac{|I_0 l|^2}{r^2}$$

$$D_0 = \frac{4\pi U_{max}}{P_{rad}} = \frac{4\pi \left[\frac{\eta}{8} \frac{|I_0 l|^2}{r^2} \right]}{\eta \left(\frac{\pi}{3} \right) \frac{|I_0 l|^2}{r^2}} = \frac{3}{2}$$

$$D_0 = \frac{3}{2} \text{ (dimensionless)} = 20 \log_{10} \left(\frac{3}{2} \right) = 1.761 \text{ dB}$$

$$R_r = \frac{2 P_{rad}}{|I_0 l|^2} = 2 \frac{\eta \left(\frac{\pi}{3} \right) \frac{|I_0 l|^2}{r^2}}{|I_0 l|^2} = \eta \left(\frac{2\pi}{3} \right) \frac{|I_0 l|^2}{r^2 |I_0 l|^2} = \eta \left(\frac{2\pi}{3} \right)$$

$$R_r = 120\pi \left(\frac{2\pi}{3} \right) \left(\frac{l}{r} \right)^2 = 80\pi^2 \left(\frac{l}{r} \right)^2$$

$$R_r = 80\pi^2 \left(\frac{l}{r} \right)^2$$

6.21

$$A = \frac{\mu}{4\pi} \int \frac{I_e e^{-j\beta R}}{R} dl' = \frac{\mu}{4\pi} \int_{-l/2}^{+l/2} \hat{a}_x I_0 \frac{e^{-j\beta r}}{r} dx' = \hat{a}_x \frac{I_0 \mu}{4\pi} \frac{e^{-jkr}}{r} \int_{-l/2}^{+l/2} dx'$$

$$A = \hat{a}_x \frac{l \mu I_0 e^{-j\beta r}}{4\pi r} = \hat{a}_x A_x \Rightarrow A_x = \frac{\mu I_0 l}{4\pi r} e^{-j\beta r}$$

$$\begin{pmatrix} A_r \\ A_\theta \\ A_\phi \end{pmatrix} = \begin{pmatrix} \sin \theta \cos \phi & \sin \theta \sin \phi & \cos \theta \\ \cos \theta \cos \phi & \cos \theta \sin \phi & -\sin \theta \\ -\sin \phi & \cos \phi & 0 \end{pmatrix} \begin{pmatrix} A_x \\ 0 \\ 0 \end{pmatrix}$$

$$A_r = A_x \sin \theta \cos \phi = \frac{\mu I_0 l e^{-j\beta r}}{4\pi r} \sin \theta \cos \phi$$

$$A_\theta = A_x \cos \theta \cos \phi = \frac{\mu I_0 l e^{-j\beta r}}{4\pi r} \cos \theta \cos \phi$$

$$A_\phi = -A_x \sin \phi = -\frac{\mu I_0 l e^{-j\beta r}}{4\pi r} \sin \phi$$

- In far-field:

$$\left. \begin{array}{l} E_r \simeq 0 \\ E_\theta \simeq -j\omega A_\theta \\ E_\phi \simeq -j\omega A_\phi \end{array} \right\} (3-58a) \Rightarrow \left. \begin{array}{l} E_r \simeq 0 \\ E_\theta = -j \frac{\omega \mu I_0 l e^{-j\beta r}}{4\pi r} \cos \theta \cos \phi \\ E_\phi = -j \frac{\omega \mu I_0 l e^{-j\beta r}}{4\pi r} \sin \phi \end{array} \right\} \left. \begin{array}{l} H_r \simeq 0 \\ H_\phi = \frac{E_\theta}{\eta} \\ H_\theta = -\frac{E_\phi}{\eta} \end{array} \right.$$

$$\left. \begin{array}{l} \underline{S}_{ave} = \frac{1}{2} \operatorname{Re} [E_r \times H_r^*] = \frac{1}{2} \operatorname{Re} [\hat{a}_\theta E_\theta + \hat{a}_\phi E_\phi] \times [\hat{a}_\theta H_\theta + \hat{a}_\phi H_\phi]^* \\ \underline{S}_{ave} = \frac{1}{2\eta} \left\{ |E_\theta|^2 + |E_\phi|^2 \right\} = \hat{a}_\theta \frac{\eta}{2} \left| \frac{I_0 l}{r} \right|^2 \frac{\sin^2 \theta}{r^2} = \hat{a}_\theta W_{ave} = \hat{a}_\theta W_r \end{array} \right\}$$

$$U = \frac{r^2}{2\eta} [|E_\theta|^2 + |E_\phi|^2] \quad (2-12a)$$

$$U = \left(\frac{\omega \mu I_0 l}{4\pi} \right)^2 \frac{1}{2\eta} [\cos^2 \theta \cos^2 \phi + \sin^2 \phi]$$

$$= B_0 [\cos^2 \theta \cos^2 \phi + \sin^2 \phi]$$

$$B_0 = \frac{1}{2\eta} \left(\frac{\omega \mu I_0 l}{4\pi} \right)^2 = \frac{1}{2\eta} \left(\frac{\eta \omega \mu I_0 l}{4\pi} \right)^2 = \frac{1}{2\eta} \left[\frac{\eta \omega \mu I_0 l}{4\pi \sqrt{\mu/\epsilon}} \right]^2$$

$$= \frac{1}{2\eta} \left[\frac{\eta \omega \sqrt{\mu\epsilon}}{4\pi} I_0 l \right] = \frac{1}{2\eta} \left[\frac{\eta \beta I_0 l}{4\pi} \right] = \frac{\eta^2}{2\eta} \left(\frac{\beta I_0 l}{4\pi} \right)^2 = \frac{\eta}{2} \left(\frac{\beta I_0 l}{4\pi} \right)^2$$

$$B_0 = \frac{\eta}{2} \left(\frac{\beta I_0 l}{4\pi} \right)^2$$

$$U = B_0 (\cos^2 \theta \cos^2 \phi + \sin^2 \phi) \Rightarrow U_{max} = B_0 \text{ when } \phi = 90^\circ, 270^\circ$$

$$0 \leq \theta \leq 180^\circ$$

Cont'd

6.21 cont'd

$$\bullet \quad P_{\text{rad}} = \int_0^{2\pi} \int_0^{\pi} U \sin \theta \, d\theta \, d\phi \\ = B_0 \left\{ \int_0^{2\pi} \int_0^{\pi} \underbrace{\cos^2 \theta \cos^2 \phi \sin \theta \, d\theta \, d\phi}_{I_1} + \int_0^{2\pi} \int_0^{\pi} \underbrace{\sin^2 \phi \sin \theta \, d\theta \, d\phi}_{I_2} \right\}$$

$$I_1 = \int_0^{2\pi} \cos^2 \phi \, d\phi \int_0^{\pi} \cos^2 \theta \sin \theta \, d\theta = \int_0^{2\pi} \cos^2 \phi \, d\phi \int_0^{\pi} \cos^2 \theta \, d(-\cos \theta) \\ = - \int_0^{2\pi} \left(\frac{1 + \cos(2\phi)}{2} \right) d\phi \int_0^{\pi} (\cos \theta)^2 \, d(\cos \theta) \\ = -\frac{1}{2} \left[(\phi + \frac{1}{2} \sin 2\phi) \Big|_0^{2\pi} \right] \left[\frac{\cos^3 \theta}{3} \Big|_0^{\pi} \right]$$

$$I_1 = -\frac{1}{2}[(2\pi)] \left(-\frac{1}{3} - \frac{1}{3} \right) = \frac{1}{2}(2\pi) \left(\frac{2}{3} \right) = \frac{2\pi}{3}$$

$$I_2 = \int_0^{2\pi} \int_0^{\pi} \sin^2 \phi \sin \theta \, d\theta \, d\phi = \int_0^{2\pi} \sin^2 \phi \, d\phi \int_0^{\pi} \sin \theta \, d\theta \\ = \int_0^{2\pi} \left(\frac{1 - \cos(2\phi)}{2} \right) d\phi \int_0^{\pi} \sin \theta \, d\theta$$

$$I_2 = \frac{1}{2} [\phi + \frac{1}{2} \sin 2\phi] \Big|_0^{2\pi} (-\cos \theta) \Big|_0^{\pi} = \frac{1}{2}(2\pi)[-(-1) + 1] = 2\pi$$

$$I_1 + I_2 = \frac{2\pi}{3} + 2\pi = \frac{8\pi}{3}$$

$$P_{\text{rad}} = B_0(I_1 + I_2) = B_0 \left(\frac{8\pi}{3} \right) = \left(\frac{8\pi}{3} \right) \left[\frac{1}{2} \left(\frac{\beta I_o l}{4\pi} \right)^2 \right] = \frac{8\pi}{3} \left[\frac{1}{2} \left(\frac{\beta l}{4\pi} \right)^2 I_o^2 \right]$$

$$D_0 = \frac{4\pi U_{\text{max}}}{P_{\text{rad}}} = \frac{4\pi(B_0)}{\frac{8\pi}{3}(B_0)} = \frac{3}{2} = 1.761 \text{ dB}$$

$$D_0 = 1.5 = +1.761 \text{ dB}$$

$$\bullet \quad R_V = \frac{2P_{\text{rad}}}{|I_o|^2} = 2 \left(\frac{8\pi}{3} \right) \left[\frac{1}{2} \left(\frac{\beta l}{4\pi} \right)^2 \right] = 8\pi^2 \left(\frac{l}{\pi} \right)^2$$

6.22 $I_e = \hat{a}_y I_o e^{-j\beta r}$

- $A = \frac{k}{4\pi} \int_{-\infty}^{\infty} I_e \frac{e^{-j\beta r}}{r} dy' = \hat{a}_y I_o \frac{\mu e^{-j\beta r}}{4\pi Y} \int_{-\ell/2}^{\ell/2} dy' = \hat{a}_y I_o \frac{\mu e^{-j\beta r}}{4\pi Y}, A_y = \frac{\mu I_o \ell}{4\pi Y} e^{-j\beta r}$

$$A_r = A_y \sin \theta \sin \phi, A_\theta = A_y \cos \theta \sin \phi, A_\phi = A_y \cos \phi$$

$$A_r = \frac{l \mu I_o e^{-j\beta r}}{4\pi Y} \sin \theta \sin \phi, A_\theta = \frac{l \mu I_o e^{-j\beta r}}{4\pi Y} \cos \theta \sin \phi, A_\phi = \frac{l \mu I_o e^{-j\beta r}}{4\pi Y} \cos \phi$$

- $E_r \approx 0$

$$E_\theta \approx -j\omega A_\theta = -j \frac{\omega \mu I_0 l e^{-j\beta r}}{4\pi Y} \cos \theta \sin \phi$$

$$E_\phi \approx -j\omega A_\phi = -j \frac{\omega \mu I_0 l e^{-j\beta r}}{4\pi Y} \cos \phi$$

$$H_r = 0$$

$$H_\theta = -E_\phi / \eta$$

$$H_\phi = E_\theta / \eta$$

- $S_{ave} = \frac{1}{2} \operatorname{Re} [E \times H^*] = \frac{1}{2} \operatorname{Re} [\hat{a}_y E_\theta + \hat{a}_\phi E_\phi] \times (\hat{a}_\theta H_\theta + \hat{a}_\phi H_\phi)^*$

$$S_{ave} = \frac{1}{2\eta} \left\{ |E_\theta|^2 + |E_\phi|^2 \right\} = \hat{a}_y \frac{\eta}{2} \left| \frac{I_o l}{a} \right|^2 \frac{\sin^2 \theta}{r^2} = \hat{a}_y W_{ave} = \hat{a}_y W_r$$

- $U(\theta, \phi) = \frac{r^2}{2\eta} (|E_\theta|^2 + |E_\phi|^2) = \frac{1}{2\eta} \left(\frac{\omega \mu I_0 l}{4\pi} \right)^2 [\cos^2 \theta \sin^2 \phi + \cos^2 \phi]$

$$U(\theta, \phi) = \frac{1}{2\eta} \left(\frac{\eta \omega \mu I_0 l}{4\pi \sqrt{\mu/\epsilon}} \right)^2 [\cos^2 \theta \sin^2 \phi + \cos^2 \phi] = B_0 (\cos^2 \theta \sin^2 \phi + \cos^2 \phi)$$

$$B_0 = \frac{1}{2\eta} \left(\frac{\eta \omega \sqrt{\mu \epsilon} I_0 l}{4\pi} \right)^2 = \frac{\eta}{2} \left(\frac{\beta I_0 l}{4\pi} \right)^2$$

$$U(\theta, \phi) = B_0 (\cos^2 \theta \sin^2 \phi + \cos^2 \phi) \Rightarrow U_{max} = B_0 @ \phi = 0^\circ, 180^\circ$$

$$P_{rad} = \int_0^{2\pi} \int_0^\pi U(\theta, \phi) \sin \theta d\theta d\phi = B_0 \int_0^{2\pi} \int_0^\pi (\cos^2 \theta \sin^2 \phi + \cos^2 \phi) \sin \theta d\theta d\phi$$

$$= B_0 \left\{ \underbrace{\int_0^{2\pi} \int_0^\pi \cos^2 \theta \sin^2 \phi \sin \theta d\theta d\phi}_{I_1} + \underbrace{\int_0^{2\pi} \int_0^\pi \cos^2 \phi \sin \theta d\theta d\phi}_{I_2} \right\}$$

$$I_1 = \int_0^{2\pi} \sin^2 \phi d\phi \int_0^\pi \cos^2 \theta \sin \theta d\theta$$

$$= \int_0^{2\pi} \left(\frac{1 - \cos 2\phi}{2} \right) d\phi \int_0^\pi \cos^2 \theta d(-\cos \theta)$$

$$= -\frac{1}{2} [\phi - \frac{1}{2} \sin 2\phi]_0^{2\pi} \left(\frac{\cos^3 \theta}{3} \right)_0^\pi = -\frac{1}{2} (2\pi) \left(\frac{-1 - 1}{3} \right) = \frac{2\pi}{3}$$

$$I_2 = \int_0^{2\pi} \int_0^\pi \cos^2 \phi \sin \theta d\theta d\phi = \int_0^{2\pi} \left(\frac{1 + \cos 2\phi}{2} \right) d\phi \int_0^\pi \sin \theta d\theta$$

$$= \frac{1}{2} [\phi + \frac{1}{2} \sin 2\phi]_0^{2\pi} (-\cos \theta)_0^\pi = \frac{1}{2} (2\pi)(2) = 2\pi$$

$$P_{rad} = B_0 (I_1 + I_2) = B_0 \left(\frac{2\pi}{3} + 2\pi \right) = B_0 \left(\frac{8\pi}{3} \right), \bar{B}_o = \frac{\eta}{2} \left(\frac{\beta I_0 l}{4\pi} \right)^2 = \frac{\eta}{2} \left(\frac{\beta l}{4\pi} \right)^2 F_o^2$$

- $D_0 = \frac{4\pi U_{max}}{P_{rad}} = \frac{4\pi B_0}{\frac{8\pi}{3} B_0} = \frac{3}{2}$ (same as in Problem 6.21)

- $R_Y = \frac{2P_{rad}}{|F_o|^2} = 2 \left(\frac{8\pi}{3} \right) \left[\frac{\eta}{2} \left(\frac{\beta l}{4\pi} \right)^2 \right] = 80\pi^2 \left(\frac{l}{\lambda} \right)^2$

$$6.23 \quad \underline{E}_A = -j\omega \underline{A} - j \frac{1}{\omega \mu \epsilon} \nabla (\nabla \cdot \underline{A})$$

$$\underline{A} \approx [\hat{a}_r A'_r(\theta, \phi) + \hat{a}_\theta A'_\theta(\theta, \phi) + \hat{a}_\phi A'_\phi(\theta, \phi)] \frac{e^{-j\beta r}}{r}$$

$$\begin{aligned} \Psi = \nabla \cdot \underline{A} &= \frac{1}{r^2} \frac{\partial}{\partial r} (r^2 A_r) - \frac{1}{r \sin \theta} \frac{\partial}{\partial \theta} (A_\theta \sin \theta) + \frac{1}{r \sin \theta} \frac{\partial A_\phi}{\partial \phi} \\ &= -j\beta \frac{e^{-j\beta r}}{r} + \frac{e^{-j\beta r}}{r^2} [\dots] + \frac{e^{-j\beta r}}{r^3} [\dots] + \dots \end{aligned}$$

$$\begin{aligned} \nabla (\nabla \cdot \underline{A}) &= \nabla \Psi = \hat{a}_r \frac{\partial \Psi}{\partial r} + \hat{a}_\theta \frac{1}{r} \frac{\partial \Psi}{\partial \theta} + \hat{a}_\phi \frac{1}{r \sin \theta} \frac{\partial \Psi}{\partial \phi} \\ &= \hat{a}_r \left\{ \frac{1}{r} \left[-\omega^2 \mu \epsilon e^{-j\beta r} A'_r(\theta, \phi) \right] + \frac{1}{r^2} [\dots] + \frac{1}{r^3} [\dots] + \dots \right\} \\ &\quad + \hat{a}_\theta \left\{ \frac{1}{r} (0) + \frac{1}{r^2} [\dots] + \frac{1}{r^3} [\dots] + \dots \right\} \\ &\quad + \hat{a}_\phi \left\{ \frac{1}{r} (0) + \frac{1}{r^2} [\dots] + \frac{1}{r^3} [\dots] + \dots \right\} \end{aligned}$$

Therefore

$$\underline{E}_A = -j\omega \underline{A} - j \frac{1}{\omega \mu \epsilon} \nabla (\nabla \cdot \underline{A})$$

$$\begin{aligned} \underline{E}_A &\approx -j\omega \left[\hat{a}_r A'_r + \hat{a}_\theta A'_\theta + \hat{a}_\phi A'_\phi \right] \frac{e^{-j\beta r}}{r} \\ &\quad - j \frac{1}{\omega \mu \epsilon} \left\{ \hat{a}_r \left[\omega^2 \mu \epsilon \frac{e^{-j\beta r}}{r} + \frac{1}{r^2} (\dots) + \frac{1}{r^3} (\dots) + \dots \right] \right. \\ &\quad \left. + \hat{a}_\theta \left[\frac{1}{r} (0) + \frac{1}{r^2} (\dots) + \frac{1}{r^3} (\dots) + \dots \right] \right. \\ &\quad \left. + \hat{a}_\phi \left[\frac{1}{r} (0) + \frac{1}{r^2} (\dots) + \frac{1}{r^3} (\dots) + \dots \right] \right\} \end{aligned}$$

or

$$\underline{E}_A \approx \frac{1}{r} \left\{ -j\omega e^{-j\beta r} \left[\hat{a}_r(0) + \hat{a}_\theta A'_\theta + \hat{a}_\phi A'_\phi \right] + \frac{1}{r^2} [\dots] + \dots \right\}$$

In a similar manner, it can be shown that

$$\underline{H} = \frac{1}{\mu} \nabla \times \underline{A}$$

$$\approx \frac{1}{r} \left\{ j \frac{\omega}{\eta} e^{-j\beta r} \left[\hat{a}_r(0) + \hat{a}_\theta A'_\phi - \hat{a}_\phi A'_\theta \right] \right\} + \frac{1}{r^2} [\dots] + \frac{1}{r^3} [\dots] + \dots$$

$$[6.24] \quad R = \left[r^2 + (r')^2 - 2rr' \cos\psi \right]^{1/2} = r \left[1 + \left(\frac{r'}{r} \right)^2 - 2 \frac{r'}{r} \cos\psi \right]^{1/2}$$

Using the binomial expansion of

$$(a+b)^n = \frac{a^n b^0}{0!} + \frac{n a^{n-1} b^1}{1!} + \frac{n(n-1) a^{n-2} b^2}{2!} + \frac{n(n-1)(n-2) a^{n-3} b^3}{3!} + \dots$$

it can be shown by letting

$$a = r^2$$

$$b = (r')^2 - 2rr' \cos\psi$$

$$n = 1/2$$

that

$$R = r - r' \cos\psi + \frac{1}{r} \left(\frac{r'^2}{2} \sin^2\psi \right) + \frac{1}{r^2} \left(\frac{r'}{2} \cos\psi \sin\psi \right) + \frac{1}{r^3} \left[\frac{(r')^3}{8} (-1 + 6 \cos^2\psi - 5 \cos^4\psi) \right] + \dots$$

By keeping the first two terms, the most significant neglected term is the third whose maximum value is

$$\frac{1}{r} \left(\frac{r'^2}{2} \sin\psi \right) \Big|_{\max} = \frac{(r')^2}{2r} \text{ when } \psi = \frac{\pi}{2}$$

Using the $\pi/8$ rad (or 22.5°) as the maximum phase error, then

$$\frac{\beta(r')^2}{2r} \leq \frac{\pi}{8}$$

If the maximum value of r' is $D/2$, then

$$\frac{\pi}{8} \frac{1}{r} \left(\frac{D^2}{4} \right) \leq \frac{\pi}{8} \Rightarrow r \geq \frac{2D^2}{3}$$

[6.25]

Using (6-112a) and (6-112b) we can reduce (6-97a) to

$$A \approx \frac{i}{4\pi} \frac{e^{-j\beta r}}{r} \int_{-\theta/2}^{+\theta/2} I_z e^{+j\beta r' \cos\psi} dz' \quad \begin{cases} \sin[\beta(\frac{\ell}{2} - z')], & 0 \leq z' \leq \ell/2 \\ \sin[\beta(\frac{\ell}{2} + z')], & -\ell/2 \leq z' \leq 0 \end{cases}$$

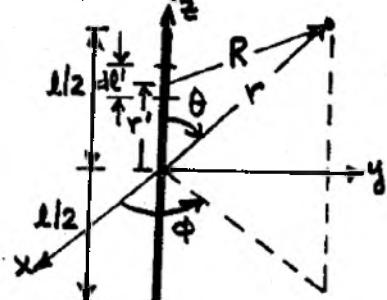
$$I_z = \hat{a}_z I_z(z) = \hat{a}_z I_0 \cdot \begin{cases} \sin[\beta(\frac{\ell}{2} - z')], & 0 \leq z' \leq \ell/2 \\ \sin[\beta(\frac{\ell}{2} + z')], & -\ell/2 \leq z' \leq 0 \end{cases}$$

$$A \approx \hat{a}_z \frac{i}{4\pi} \frac{e^{-j\beta r}}{r} \int_{-\theta/2}^{+\theta/2} I_z(z) e^{+j\beta z' \cos\theta} dz' = \hat{a}_z A_z$$

$$E_z \approx -j\omega A_z$$

$$E_\theta \approx -j\omega A_\theta$$

$$E_\phi \approx -j\omega A_\phi$$



cont'd.

6.25 cont'd.

$$A_\theta \approx -\frac{\mu I_0 \sin \theta}{2\pi r} e^{-j\beta r} \int_{-\ell/2}^{\ell/2} I_z(z') e^{j\beta z' \cos \theta} dz'$$

$$E_\theta = +j \frac{\omega \mu I_0 \sin \theta e^{-j\beta r}}{2\pi r} \int_{-\ell/2}^{\ell/2} I_z(z') e^{j\beta z' \cos \theta} dz'$$

Using the sinusoidal current distribution and the integral of

$$\int e^{\alpha x} \sin(\beta x + \gamma) dx = \frac{e^{\alpha x}}{\alpha^2 + \beta^2} [\alpha \sin(\beta x + \gamma) - \beta \cos(\beta x + \gamma)]$$

we can write that

$$E_\theta = j\eta \frac{\beta I_0 \sin \theta e^{-j\beta r}}{2\pi r} \left\{ \int_{-\ell/2}^0 \sin[\beta(\frac{r}{2} + z')] e^{j\beta z' \cos \theta} dz' + \int_0^{\ell/2} \sin[\beta(\frac{r}{2} - z')] e^{j\beta z' \cos \theta} dz' \right\}$$

After some mathematical manipulations, we can reduce the above equation to

$$E_\theta \approx j\eta \frac{I_0 e^{-j\beta r}}{2\pi r} \left[\frac{\cos(\frac{\beta \ell}{2} \cos \theta) - \cos(\frac{\beta \ell}{2})}{\sin \theta} \right], H_\phi \approx +\frac{E_\theta}{\eta}$$

6.26 When $\ell = \lambda/2$ we can write the far-zone electric and magnetic fields of Problem 6.25 when $\beta \ell / 2 = \frac{2\pi}{\lambda} \left(\frac{\lambda}{4} \right) = \frac{\pi}{2}$

$$E_\theta \approx j\eta \frac{I_0 e^{-j\beta r}}{2\pi r} \left[\frac{\cos(\frac{\pi}{2} \cos \theta) - \cos(\frac{\pi}{2})}{\sin(\frac{\pi}{2})} \right]$$

$$E_\theta \approx j\eta \frac{I_0 e^{-j\beta r}}{2\pi r} \left[\frac{\cos(\frac{\pi}{2} \cos \theta)}{\sin \theta} \right]$$

$$H_\phi \approx \frac{E_\theta}{\eta}$$

$$E_r \approx E_\phi \approx H_r \approx H_\theta \approx 0$$

6.27 Since the operator ∇ in the solution of Problem 6.3 acts on the coordinates of the sources while the observation point is maintained fixed, then

$$\nabla \left(\frac{e^{-j\beta R}}{R} \right) = -\hat{R} \left(j\beta + \frac{1}{R} \right) \frac{e^{-j\beta R}}{R} \quad (1)$$

$$(\underline{J} \cdot \nabla) \nabla \left(\frac{e^{-j\beta R}}{R} \right) = \left[\beta^2 (\underline{J} \cdot \hat{R}) \hat{R} + \frac{3}{R} (j\beta + \frac{1}{R}) (\underline{J} \cdot \hat{R}) \hat{R} - \frac{1}{R} (j\beta + \frac{1}{R}) \right] \frac{e^{-j\beta R}}{R} \quad (2)$$

Thus for first-order variations in R , (1) and (2) reduce to

$$\nabla \left(\frac{e^{-j\beta R}}{R} \right) \approx -\hat{R} \left(j\beta \frac{e^{-j\beta R}}{R} \right) \quad (3)$$

$$(\underline{J} \cdot \nabla) \nabla \left(\frac{e^{-j\beta R}}{R} \right) \approx -\hat{R} \beta^2 (\underline{J} \cdot \hat{R}) \frac{e^{-j\beta R}}{R} \quad (4)$$

Using (3) and (4) we can reduce (3) and (6) of the solution of Problem 6.3 to

$$\underline{H}_A \approx -\frac{1}{4\pi} \int \int \int \int \left[\underline{J} \times \left(-j\beta \hat{R} \frac{e^{-j\beta R}}{R} \right) \right] d\underline{v}' = +j \frac{\beta}{4\pi} \int \int \int \int \left[(\underline{J} \times \hat{R}) \frac{e^{-j\beta R}}{R} \right] d\underline{v}' \quad (5)$$

$$\underline{E}_A \approx -j \frac{1}{4\pi \omega \epsilon} \int \int \int \int \left[\beta^2 \underline{J} - \hat{R} \beta^2 (\underline{J} \cdot \hat{R}) \right] \frac{e^{-j\beta R}}{R} d\underline{v}' = -j \eta \frac{\beta}{4\pi} \int \int \int \int \left[\underline{J} - \hat{R} (\underline{J} \cdot \hat{R}) \right] \frac{e^{-j\beta R}}{R} d\underline{v}' \quad (6)$$

Using the geometry of Figure 6-3 we can write for far-field observations

$$R \approx r - \underline{r}' \cdot \hat{a}_r \quad \text{for phase variations} \quad (7a)$$

$$R \approx r \quad \text{for amplitude variations} \quad (7b)$$

Thus (5) and (6) of above can be reduced to

$$\underline{H}_A \approx j \frac{\beta e^{-j\beta}}{4\pi r} \int \int \int (\underline{J} \times \hat{a}_r) e^{j\beta \underline{r}' \cdot \hat{a}_r} d\underline{v}' \quad (8)$$

$$\underline{E}_A \approx -j \eta \frac{\beta e^{-j\beta R}}{4\pi r} \int \int \int \left[\underline{J} - \hat{a}_r (\underline{J} \cdot \hat{a}_r) \right] e^{j\beta \underline{r}' \cdot \hat{a}_r} d\underline{v}' \quad (9)$$

6.28 Using the same procedure as in Problem 6.27 solution, we can reduce (3) and (4) of the solution of Problem 6.4 for far-field observations to

$$\underline{E}_F \approx -j \frac{\beta e^{-j\beta R}}{4\pi r} \int \int \int (\underline{M} \times \hat{a}_r) e^{+j\beta \underline{r}' \cdot \hat{a}_r} d\underline{v}'$$

$$\underline{H}_F \approx -j \frac{\beta e^{-j\beta R}}{4\pi \eta r} \int \int \int \left[\underline{M} - \hat{a}_r (\underline{M} \cdot \hat{a}_r) \right] e^{j\beta \underline{r}' \cdot \hat{a}_r} d\underline{v}'$$

$$6.29 \quad \underline{E}_a = \hat{a}_z E_0, \quad \underline{M}_s = -2\hat{n} \times \underline{E}_a = -2\hat{a}_x \times \hat{a}_z E_0 = \hat{a}_y 2E_0$$

$$\text{Thus } M_y = 2E_0, \quad M_x = M_z = J_x = J_y = J_z = 0$$

$$N_\theta = N_\phi = 0$$

$$L_\theta = \iiint [M_x^0 \cos\theta \cos\phi + M_y \cos\theta \sin\phi - M_z \sin\theta] e^{j\beta(y \sin\theta \sin\phi + z \cos\theta)} dy' dz'$$

$$= 2E_0 \cos\theta \sin\phi \int_{-a/2}^{a/2} e^{j\beta y' \sin\theta \sin\phi} dy' \int_{-b/2}^{b/2} e^{j\beta z' \cos\theta} dz'$$

$$L_\theta = 2E_0 ab \left[\cos\theta \sin\phi \frac{\sin Y}{Y} \frac{\sin Z}{Z} \right], \quad Y = \frac{pa}{2} \sin\theta \sin\phi, \quad Z = \frac{pb}{2} \cos\theta$$

$$L_\phi = \iiint [-M_x^0 \sin\phi + M_y \cos\phi] e^{j\beta(y \sin\theta \sin\phi + z \cos\theta)} dy' dz'$$

$$= 2E_0 ab \left[\cos\phi \frac{\sin Y}{Y} \frac{\sin Z}{Z} \right]$$

Using (6-122a) - (6-122f), we have $E_r \approx H_r \approx 0$ and

$$E_\theta \approx -\frac{j\beta e^{-j\beta r}}{4\pi r} [L_\phi + \eta N_\phi^0] = -j \frac{ab\beta E_0 e^{-j\beta r}}{2\pi r} \left[\cos\phi \frac{\sin Y}{Y} \frac{\sin Z}{Z} \right]$$

$$E_\phi \approx \frac{j\beta e^{-j\beta r}}{4\pi r} [L_\theta - \eta N_\theta^0] = -j \frac{ab\beta E_0 e^{-j\beta r}}{2\pi r} \left[-\cos\theta \sin\phi \frac{\sin Y}{Y} \frac{\sin Z}{Z} \right]$$

$$H_\theta \approx -\frac{E_\theta}{\eta}, \quad H_\phi \approx +\frac{E_\phi}{\eta}$$

6.30

$$\underline{E}_a = \hat{a}_z E_0, \quad \underline{H}_a = -\hat{a}_y \frac{E_0}{\eta}$$

$$\underline{M}_s = -\hat{n} \times \underline{E}_a, \quad \underline{J}_s = \hat{n} \times \underline{H}_a$$

$$\underline{M}_s = -\hat{a}_x \times (\hat{a}_z E_0) = \hat{a}_y E_0 \Rightarrow M_y = E_0$$

$$\underline{J}_s = \hat{a}_x \times (-\hat{a}_y \frac{E_0}{\eta}) = -\hat{a}_z \frac{E_0}{\eta} \Rightarrow J_z = -\frac{E_0}{\eta}$$

$$N_\theta = \iint (\hat{J}_x^0 \cos\theta \cos\phi + \hat{J}_y^0 \cos\theta \sin\phi - \hat{J}_z^0 \sin\theta) e^{j\beta r' \cos\psi} ds'$$

$$= - \iint \hat{J}_z^0 \sin\theta e^{j\beta r' \cos\psi} ds' = + \frac{E_0}{\eta} \sin\theta \int_{-a/2}^{a/2} e^{j\beta y' \sin\theta \sin\phi} dy' \int_{-b/2}^{b/2} e^{j\beta z' \cos\theta} dz'$$

$$N_\theta = \frac{E_0 ab}{\eta} \sin\theta \left[\frac{\sin Y}{Y} \left(\frac{\sin Z}{Z} \right) \right]_{-a/2}^{a/2}, \quad Y = \frac{pa}{2} \sin\theta \sin\phi$$

$$N_\phi = \iint (-\hat{J}_x^0 \sin\phi + \hat{J}_y^0 \cos\phi) e^{j\beta r' \cos\psi} ds' = 0, \quad Z = \frac{pb}{2} \cos\theta$$

cont'd

6.30 cont'd]

$$\begin{aligned}
 L_\theta &= \iiint_S \left(M_x^0 \cos\theta \cos\phi + M_y \cos\theta \sin\phi - M_z \sin\theta \right) e^{j\beta r' \cos\psi} ds' = \iint_S M_y \cos\theta \sin\phi e^{j\beta r' \cos\psi} ds' \\
 &= E_0 \cos\theta \sin\phi \int_{-a/2}^{a/2} e^{j\beta y' \sin\theta \sin\phi} dy' \int_{-b/2}^{b/2} e^{j\beta z' \cos\theta} dz' = E_0 ab \cos\theta \sin\phi \left[\frac{\sin Y}{Y} \frac{\sin Z}{Z} \right] \\
 L_\phi &= \iiint_S \left(-M_x^0 \sin\phi + M_y \cos\phi \right) e^{j\beta r' \cos\psi} dx = \iint_S M_y \cos\phi e^{j\beta r' \cos\psi} ds' \\
 &= E_0 \cos\phi \int_{-a/2}^{a/2} e^{j\beta y' \sin\theta \sin\phi} dy' \int_{-b/2}^{b/2} e^{j\beta z' \cos\theta} dz' = E_0 (ab) \cos\phi \left[\frac{\sin Y}{Y} \frac{\sin Z}{Z} \right] \\
 E_r &\approx 0, E_\theta \approx -j \frac{\beta e^{-j\beta r}}{4\pi r} (L_\theta + \eta N_\theta) = -j E_0 \frac{ab \beta e^{-j\beta r}}{4\pi r} \left\{ (\cos\phi + \sin\phi) \frac{\sin Y}{Y} \frac{\sin Z}{Z} \right\} \\
 E_\phi &\approx +j \frac{\beta e^{-j\beta r}}{4\pi r} (L_\phi - \eta N_\phi) = +j E_0 \frac{ab \beta e^{-j\beta r}}{4\pi r} \left\{ \cos\theta \sin\phi \frac{\sin Y}{Y} \frac{\sin Z}{Z} \right\}
 \end{aligned}$$

$$H_r \approx 0, H_\theta = -E_\theta/\eta, H_\phi = E_\phi/\eta$$

6.31 $\underline{E}_0 = \hat{a}_x E_0, \underline{M}_S = -2\hat{n} \times \underline{E}_0 = -2\hat{a}_y \times \hat{a}_x E_0 = \hat{a}_z 2E_0$

$$\text{Thus } M_z = 2E_0, M_x = M_y = J_x = J_y = J_z = 0$$

$$N_\theta = N_\phi = 0$$

$$\begin{aligned}
 L_\theta &= \iint_{S_a} \left[M_x^0 \cos\theta \cos\phi + M_y \cos\theta \sin\phi - M_z \sin\theta \right] e^{j\beta (x' \sin\theta \cos\phi + z' \cos\theta)} dx' dz' \\
 &= -2E_0 \sin\theta \int_{-a/2}^{a/2} e^{j\beta z' \cos\theta} dz' \int_{-b/2}^{b/2} e^{j\beta x' \sin\theta \cos\phi} dx' \\
 L_\theta &= -2E_0 ab \left[\sin\theta \frac{\sin X}{X} \frac{\sin Z}{Z} \right], X = \frac{\beta b}{2} \sin\theta \cos\phi, Z = \frac{\beta a}{2} \cos\theta
 \end{aligned}$$

(cont'd)

6.31 Cont'd.

$$L_\phi = \iint_{S_a} \left[-M_x \overset{\circ}{\sin \phi} + M_y \overset{\circ}{\cos \phi} \right] e^{j\beta(x' \sin \theta \cos \phi + z' \cos \theta)} dx' dz' = 0$$

Using (6-122a)-(6-122f), we have

$$E_r \approx H_r \approx 0$$

$$E_\theta \approx -\frac{j\beta e^{-j\beta r}}{4\pi r} \left[L_\theta + \eta N_\theta \right] = 0$$

$$E_\phi \approx \frac{j\beta e^{-j\beta r}}{4\pi r} \left[L_\phi - \eta N_\phi \right] = -j \frac{ab\beta e^{-j\beta r}}{2\pi r} E_0 \left[\sin \theta \frac{\sin X}{X} \frac{\sin Y}{Y} \right]$$

$$H_\theta \approx -\frac{E_\phi}{\eta}, \quad H_\phi \approx \frac{E_\theta}{\eta} = 0$$

6.32] $E_a = \hat{a}_y E_0, M_s = -2\hat{n} \times E_a = -2\hat{a}_z \times \hat{a}_y E_0 = \hat{a}_x 2E_0$

Thus $M_x = 2E_0, M_y = M_z = J_x = J_y = J_z = 0$

$$N_\theta = N_\phi = 0$$

$$L_\theta = \iint_{S_a} \left[M_x \cos \theta \cos \phi + M_y \overset{\circ}{\cos \theta \sin \phi} - M_z \overset{\circ}{\sin \theta} \right] e^{j\beta(x' \sin \theta \cos \phi + y' \sin \theta \sin \phi)} dx' dy'$$

$$= 2E_0 \cos \theta \cos \phi \int_{-b/2}^{b/2} e^{j\beta y' \sin \theta \sin \phi} dy' \int_{-a/2}^{a/2} e^{j\beta x' \sin \theta \cos \phi} dx'$$

$$L_\phi = 2abE_0 \left[\cos \theta \cos \phi \frac{\sin X}{X} \frac{\sin Y}{Y} \right], \quad X = \frac{\beta a}{2} \sin \theta \cos \phi, \quad Y = \frac{\beta b}{2} \sin \theta \sin \phi$$

$$L_\phi = \iint_{S_a} \left[-M_x \overset{\circ}{\sin \phi} + M_y \overset{\circ}{\cos \phi} \right] e^{j\beta(x' \sin \theta \cos \phi + y' \sin \theta \sin \phi)} dx' dy'$$

$$= -2abE_0 \left[\sin \phi \frac{\sin X}{X} \frac{\sin Y}{Y} \right]$$

Using (6-122a)-(6-122f), we have $E_r \approx H_r \approx 0$

$$E_\theta \approx -\frac{j\beta e^{-j\beta r}}{4\pi r} \left[L_\phi + \eta N_\phi \right] = j \frac{ab\beta e^{-j\beta r}}{2\pi r} E_0 \left[\sin \phi \frac{\sin X}{X} \frac{\sin Y}{Y} \right]$$

$$E_\phi \approx \frac{j\beta e^{-j\beta r}}{4\pi r} \left[L_\theta - \eta N_\theta \right] = j \frac{ab\beta e^{-j\beta r}}{2\pi r} E_0 \left[\cos \theta \cos \phi \frac{\sin X}{X} \frac{\sin Y}{Y} \right]$$

$$H_\theta \approx -\frac{E_\phi}{\eta}, \quad H_\phi \approx \frac{E_\theta}{\eta}$$

$$6.33 \quad E_a = \hat{a}_z E_0 \omega \left(\frac{\pi}{a} y' \right)$$

The only difference between this problem and 6.29 is that for the y variations the integral reduces to

$$\int_{-a/2}^{a/2} \cos\left(\frac{\pi}{a} y'\right) e^{j\beta y' \sin\theta \sin\phi} dy' = -\left(\frac{\pi a}{2}\right) \frac{\cos\left(\frac{\beta a}{2} \sin\theta \sin\phi\right)}{\left(\frac{\beta a}{2} \sin\theta \sin\phi\right)^2 - \left(\frac{\pi}{2}\right)^2}$$

Thus $E_r \approx H_r \approx 0$

$$E_\theta \approx -\frac{j\beta e^{-j\beta r}}{4\pi r} L_\theta = j \frac{ab\beta E_0 e^{-j\beta r}}{4r} \left[\cos\phi \frac{\cos Y}{(Y)^2 - (\pi/2)^2} \frac{\sin Z}{Z} \right], \quad Y = \frac{\beta a}{2} \sin\theta \sin\phi$$

$$E_\phi \approx \frac{j\beta e^{-j\beta r}}{4\pi r} L_\phi = j \frac{ab\beta E_0 e^{-j\beta r}}{4r} \left[-\cos\theta \sin\phi \frac{\cos Y}{(Y)^2 - (\pi/2)^2} \frac{\sin Z}{Z} \right], \quad Z = \frac{\beta b}{2} \cos\theta$$

$$H_\theta \approx -\frac{E_\phi}{\eta}, \quad H_\phi \approx +\frac{E_\theta}{\eta}$$

$$6.34 \quad E_a = \hat{a}_x E_0 \omega \left(\frac{\pi}{a} z' \right)$$

The only difference between this problem and 6.31 is that for the z variations the integral reduces to

$$\int_{-a/2}^{a/2} \cos\left(\frac{\pi}{a} z'\right) e^{j\beta z' \cos\theta} dz' = -\left(\frac{\pi a}{2}\right)^2 \frac{\cos\left(\frac{\beta a}{2} \cos\theta\right)}{\left(\frac{\beta a}{2} \cos\theta\right)^2 - \left(\frac{\pi}{2}\right)^2}$$

Thus $E_r \approx H_r \approx 0 \approx E_\theta$

$$E_\phi \approx j \frac{\beta e^{-j\beta r}}{4\pi r} L_\phi = j \frac{ab\beta E_0 e^{-j\beta r}}{4r} \left[\sin\theta \frac{\sin X}{X} \frac{\cos Z}{(Z)^2 - (\pi/2)^2} \right], \quad X = \frac{\beta b}{2} \sin\theta \cos\phi$$

$$H_\phi \approx -\frac{E_\phi}{\eta}, \quad H_\theta \approx \frac{E_\theta}{\eta} = 0$$

$$6.35 \quad E_a = \hat{a}_y E_0 \omega \left(\frac{\pi}{a} x' \right)$$

The only difference between this problem and 6.32 is that for the x variations the integral reduces to

$$\int_{-a/2}^{a/2} \cos\left(\frac{\pi}{a} x'\right) e^{j\beta x' \sin\theta \cos\phi} dx' = -\left(\frac{\pi a}{2}\right)^2 \frac{\cos\left(\frac{\beta a}{2} \sin\theta \cos\phi\right)}{\left(\frac{\beta a}{2} \sin\theta \cos\phi\right)^2 - \left(\frac{\pi}{2}\right)^2}$$

Thus $E_r \approx H_r \approx 0$

$$E_\theta \approx -\frac{j\beta e^{-j\beta r}}{4\pi r} L_\phi = -j \frac{ab\beta E_0 e^{-j\beta r}}{4r} \left[\sin\phi \frac{\cos X}{(X)^2 - (\pi/2)^2} \frac{\sin Y}{Y} \right], \quad X = \frac{\beta b}{2} \sin\theta \cos\phi$$

$$E_\phi \approx \frac{j\beta e^{-j\beta r}}{4\pi r} L_\theta \approx -j \frac{ab\beta E_0 e^{-j\beta r}}{4r} \left[\cos\theta \cos\phi \frac{\cos X}{(X)^2 - (\pi/2)^2} \frac{\sin Y}{Y} \right], \quad Y = \frac{\beta b}{2} \sin\theta \sin\phi$$

$$H_\theta = -\frac{E_\phi}{\eta}, \quad H_\phi = \frac{E_\theta}{\eta}$$

6.36

(a) In the E-plane $E_r \approx E_d \approx 0$ and the normalized E_θ

$$E_{\theta n} = \left(\frac{1+ic_1\theta}{2}\right) \frac{\sin\left(\frac{\beta b}{2} \sin\theta\right)}{\frac{\beta b}{2} \sin\theta} \xrightarrow{b \rightarrow \text{large}} \frac{\sin\left(\frac{\beta b}{2} \sin\theta\right)}{\frac{\beta b}{2} \sin\theta}$$

The 0.707 point of the sinc function occurs when its argument is 1.391. Thus

$$\frac{\beta b}{2} \sin\theta_h = 1.391 \Rightarrow \theta_h = \sin^{-1}\left(\frac{2.782}{\beta b}\right) = \sin^{-1}\left(\frac{0.4432}{b}\right)$$

Therefore the half-power beamwidth Θ_h is given by

$$\Theta_h = 2\theta_h = 2\sin^{-1}\left(\frac{0.4432}{b}\right) \Big|_{b=3\lambda} = 2\sin^{-1}\left(\frac{0.443}{3}\right) = 2(6.49) = 16.98^\circ$$

(b) In the H-plane $E_r \approx E_d \approx 0$ and the normalized E_ϕ

$$E_{\phi n} = \left(\frac{1+ic_1\theta}{2}\right) \frac{\sin\left(\frac{\beta a}{2} \sin\theta\right)}{\frac{\beta a}{2} \sin\theta} \xrightarrow{a \rightarrow \text{large}} \frac{\sin\left(\frac{\beta a}{2} \sin\theta\right)}{\frac{\beta a}{2} \sin\theta}$$

Therefore

$$\frac{\beta a}{2} \sin\theta_h = 1.391 \Rightarrow \theta_h = \sin^{-1}\left(\frac{0.4432}{a}\right)$$

$$\Theta_h = 2\theta_h = 2\sin^{-1}\left(\frac{0.4432}{a}\right) \Big|_{a=4\lambda} = 2\sin^{-1}\left(\frac{0.443}{4}\right) = 2(6.36) = 12.72^\circ$$

6.37.

(a) $\underline{\underline{E}}_a = \hat{a}_y E_0 \left[1 - \left(\frac{\rho'}{a} \right)^2 \right], \rho' \leq a$

(b) $\underline{\underline{E}}_a = \hat{a}_y E_0 \left[1 - \left(\frac{\rho'}{a} \right)^2 \right]^2, \rho' \leq a$

(c) $\underline{\underline{E}}_a = \hat{a}_y E_0 \left[1 - \left(\frac{\rho'}{a} \right)^2 \right], \rho' \leq a$

We start by writing the magnetic current from the expression of the electric field given in the exercise

$$\mathbf{M}_s = -2\hat{n} \times \mathbf{E}_a = \hat{a}_x 2E_0 [1 - (\rho'/a)^2].$$

Since the electric currents are zero then N_θ and N_ϕ will be equal to zero. We note that the aperture fields are given in terms of their cartesian components, so we use

$$\begin{aligned} L_\theta &= \int_S M_x \cos \theta \cos \phi e^{j\beta r' \cos \psi} ds' \\ &= 2E_0 \cos \theta \cos \phi \int_0^a \int_0^{2\pi} \rho' [1 - (\rho'/a)^2] e^{j\beta \rho' \sin \theta \cos(\phi - \phi')} d\rho' d\phi'. \end{aligned}$$

We observe that

$$\int_0^{2\pi} e^{j\beta \rho' \sin \theta \cos(\phi - \phi')} d\phi' = 2\pi J_0(\beta \rho' \sin \theta),$$

which yields

$$L_\theta = 4\pi E_0 \cos \theta \cos \phi \int_0^a J_0(\beta \rho' \sin \theta) [1 - (\rho'/a)^2] \rho' d\rho'.$$

By making the substitution

$$t = \beta \rho' \sin \theta, \rho' = \frac{t}{\beta \sin \theta}, \Rightarrow dt = \beta \rho' \sin \theta d\rho', d\rho' = \frac{dt}{\beta \sin \theta}$$

the previous integral becomes

$$L_\theta = 4\pi E_0 \cos \theta \cos \phi \left[\frac{1}{(\beta \sin \theta)^2} \int_0^{\beta a \sin \theta} t J_0(t) dt - \frac{1}{a^2 (\beta \sin \theta)^4} \int_0^{\beta a \sin \theta} t^3 J_0(t) dt \right].$$

Then we use the Bessel function indefinite integrals

$$\begin{aligned} \int x^{p+1} J_p(ax) dx &= \frac{1}{\alpha} x^{p+1} J_{p+1}(ax) + C \Rightarrow \int t J_0(t) dt = t J_1(t) + C \\ \int t^3 J_0(t) dt &= t^3 J_1(t) - 2t^2 J_2(t) + C \end{aligned}$$

and we obtain

$$\begin{aligned} L_\theta &= 4\pi E_0 \cos \theta \cos \phi \left[\frac{a}{\beta \sin \theta} J_1(\beta a \sin \theta) - \frac{a}{\beta \sin \theta} J_1(\beta a \sin \theta) + \frac{2}{(\beta \sin \theta)^2} J_2(\beta a \sin \theta) \right] \\ &= 8\pi E_0 \frac{\cos \theta \cos \phi}{(\beta \sin \theta)^2} J_2(\beta a \sin \theta). \end{aligned}$$

cont'd

The procedure to find the expression for L_ϕ is very similar and yields

$$\begin{aligned}
 L_\phi &= \int_s -M_x \sin \phi e^{j\beta r' \cos \psi} ds' \\
 &= -2E_0 \sin \phi \int_0^a \rho' \left[1 - (\rho'/a)^2 \right] \int^2 \pi_0 e^{j\beta \rho' \sin \theta \cos(\phi-\phi')} d\phi' d\rho' \\
 &= -2E_0 \sin \phi \int_0^a \rho' \left[1 - (\rho'/a)^2 \right] 2\pi J_0(\beta \rho' \sin \theta) d\rho' \\
 &= -4\pi E_0 \sin \phi \int_0^a \left[\rho' J_0(\beta \rho' \sin \theta) - \frac{\rho'^3}{a^2} J_0(\beta \rho' \sin \theta) \right] d\rho' \\
 &= -8\pi E_0 \sin \phi \frac{J_2(\beta a \sin \theta)}{(\beta \sin \theta)^2}
 \end{aligned}$$

It is now possible to write the expressions of the electric and magnetic fields. The result is

$$E_r \simeq 0$$

$$E_\theta \simeq -\frac{j\beta e^{-j\beta r}}{4\pi r} L_\phi = j2\beta E_0 \frac{e^{-j\beta r}}{r} \sin \phi \frac{J_2(\beta a \sin \theta)}{(\beta \sin \theta)^2}$$

$$E_\phi \simeq +\frac{j\beta e^{-j\beta r}}{4\pi r} L_\theta = j2\beta E_0 \frac{e^{-j\beta r}}{r} \cos \theta \cos \phi \frac{J_2(\beta a \sin \theta)}{(\beta \sin \theta)^2}$$

$$H_r \simeq 0$$

$$H_\theta \simeq -\frac{1}{\eta} \frac{j\beta e^{-j\beta r}}{4\pi r} L_\theta = -E_\phi / \eta = j2 \frac{\beta E_0}{\eta} \frac{e^{-j\beta r}}{r} \cos \theta \cos \phi \frac{J_2(\beta a \sin \theta)}{(\beta \sin \theta)^2}$$

$$H_\phi \simeq -\frac{1}{\eta} \frac{j\beta e^{-j\beta r}}{4\pi r} L_\phi = E_\theta / \eta = j2 \frac{\beta E_0}{\eta} \frac{e^{-j\beta r}}{r} \sin \phi \frac{J_2(\beta a \sin \theta)}{(\beta \sin \theta)^2}$$

$$(b) \quad \underline{E}_a = \hat{a}_y \underline{E}_0 \left[1 - \left(\frac{\rho'}{a} \right)^2 \right]^2, \quad \rho' \leq a$$

This case is very similar to previous one. The magnetic current is

$$\mathbf{M}_s = -2\hat{n} \times \mathbf{E}_a = \hat{a}_x 2E_0 \left[1 - (\rho'/a)^2 \right]^2.$$

and

$$N_\theta = 0, N_\phi = 0.$$

We have

$$\begin{aligned}
 L_\theta &= \int_s M_x \cos \theta \cos \phi e^{j\beta r' \cos \psi} ds' \\
 &= \int_s \left(2E_0 \left[1 - (\rho'/a)^2 \right]^2 \cos \theta \cos \phi \right) e^{j\beta \rho' \sin \theta \cos(\phi-\phi')} \rho' d\phi' d\rho' \\
 &= 2E_0 \cos \theta \cos \phi \int_0^a \rho' \left[1 - (\rho'/a)^2 \right]^2 \int_0^{2\pi} e^{j\beta \rho' \sin \theta \cos(\phi-\phi')} d\phi' d\rho' \\
 &= 2E_0 \cos \theta \cos \phi \int_0^a \rho' \left[1 - (\rho'/a)^2 \right]^2 2\pi J_0(\beta \rho' \sin \theta) d\rho' \\
 &= 4\pi E_0 \cos \theta \cos \phi \int_0^a \left[\rho' J_0(\beta \rho' \sin \theta) - 2 \frac{\rho'^3}{a^2} J_0(\beta \rho' \sin \theta) + \frac{\rho'^5}{a^4} J_0(\beta \rho' \sin \theta) \right] d\rho'
 \end{aligned}$$

Cont'd

Using the substitution, we obtain

$$L_\theta = 4\pi E_0 \cos \theta \cos \phi \int_0^{\beta a \sin \theta} \left[\frac{t}{(\beta \sin \theta)^2} J_0(t) - 2 \frac{t^3}{(\beta \sin \theta)^4 a^2} J_0(t) + \frac{t^5}{(\beta \sin \theta)^6 a^4} J_0(t) \right] dt.$$

We use again the Bessel indefinite integrals, and

$$\begin{aligned} \int t^m J_n(t) dt &= t^m J_{n+1}(t) - (m-n-1) \int t^{m-1} J_{n+1}(t) dt \\ \Rightarrow \int t^5 J_0(t) dt &= t^5 J_1(t) - 4 \left[\int t^4 J_1(t) dt \right] = t^5 J_1(t) - 4 \left[t^4 J_2(t) - 2 \int t^3 J_2(t) dt \right] \\ &= t^5 J_1(t) - 4 [t^4 J_2(t) - 2t^3 J_3(t)] + C \end{aligned}$$

Hence,

$$\begin{aligned} L_\theta &= 4\pi E_0 \cos \theta \cos \phi \left[\frac{\beta a \sin \theta}{(\beta \sin \theta)^2} J_1(\beta a \sin \theta) - \frac{2(\beta a \sin \theta)^3}{(\beta \sin \theta)^4 a^2} J_1(\beta a \sin \theta) + \frac{4(\beta a \sin \theta)^2}{(\beta \sin \theta)^4 a^2} J_2(\beta a \sin \theta) \right] \\ &\quad + 4\pi E_0 \cos \theta \cos \phi \left[\frac{(\beta a \sin \theta)^5}{(\beta \sin \theta)^6 a^4} J_1(\beta a \sin \theta) - 4 \frac{(\beta a \sin \theta)^4}{(\beta \sin \theta)^6 a^4} J_2(\beta a \sin \theta) + 8 \frac{(\beta a \sin \theta)^3}{(\beta \sin \theta)^6 a^4} J_3(\beta a \sin \theta) \right] \\ &= 32\pi E_0 \cos \theta \cos \phi \frac{J_3(\beta a \sin \theta)}{a(\beta \sin \theta)^3} \end{aligned}$$

and, similarly,

$$L_\phi = -32\pi E_0 \sin \phi \frac{1}{a(\beta \sin \theta)^3} J_3(a\beta \sin \theta)$$

The expressions of the fields are again obtained from

$$E_r \simeq 0$$

$$E_\theta \simeq -\frac{j\beta e^{-j\beta r}}{4\pi r} L_\phi = j32\beta E_0 \frac{e^{-j\beta r}}{r} \sin \phi \frac{J_3(\beta a \sin \theta)}{a(\beta \sin \theta)^3}$$

$$E_\phi \simeq \frac{j\beta e^{-j\beta r}}{4\pi r} L_\theta = j32\beta E_0 \frac{e^{-j\beta r}}{r} \cos \theta \cos \phi \frac{J_3(\beta a \sin \theta)}{a(\beta \sin \theta)^3}$$

$$H_r \simeq 0$$

$$H_\theta \simeq -\frac{1}{\eta} \frac{j\beta e^{-j\beta r}}{4\pi r} L_\theta = -E_\phi/\eta = j32 \frac{\beta E_0}{\eta} \frac{e^{-j\beta r}}{r} \cos \theta \cos \phi \frac{J_3(\beta a \sin \theta)}{a(\beta \sin \theta)^3}$$

$$H_\phi \simeq -\frac{1}{\eta} \frac{j\beta e^{-j\beta r}}{4\pi r} L_\phi = -E_\theta/\eta = j32 \frac{\beta E_0}{\eta} \frac{e^{-j\beta r}}{r} \sin \phi \frac{J_3(\beta a \sin \theta)}{a(\beta \sin \theta)^3}$$

$$6.38 \quad \underline{E}_a = -\hat{a}_p \frac{\nabla}{2\ln(b/a)} \frac{1}{p'} = -\hat{a}_p C \frac{1}{p'} \quad \text{where } C = \frac{\nabla}{2\ln(b/a)}$$

$$\underline{M}_s = -2\hat{n} \times \underline{E}_a = -2\hat{a}_p \times \left(-\hat{a}_p \frac{C}{p'} \right) = \hat{a}_p 2 \frac{C}{p'} \Rightarrow M_\phi = \frac{2C}{p'}$$

Thus using (6-131c)

$$L_\theta = \int_a^b \int_0^{2\pi} \left[M_p \cos\theta \cos(\phi-\phi') + M_\phi \cos\theta \sin(\phi-\phi') - M_z \sin\theta \right] e^{jpr' \cos\theta} d\phi' ds'$$

$$= 2C \cos\theta \int_a^b \left[\int_0^{2\pi} \sin(\phi-\phi') e^{j\beta p' \sin\theta \cos(\phi-\phi')} d\phi' \right] dp'$$

Because of azimuthal symmetry, the field is not a function of ϕ . Choosing $\phi=0$

$$L_\theta = 2C \cos\theta \int_a^b \left[- \int_0^{2\pi} \sin\phi' e^{j\beta p' \sin\theta \cos\phi'} d\phi' \right] dp' = 2C \cos\theta \int_a^b [0] dp' = 0$$

$$L_\phi = \int_a^b \int_0^{2\pi} \left[-M_p \sin(\phi-\phi') + M_\phi \cos(\phi-\phi') \right] e^{jpr' \cos\theta} d\phi' ds' = 2C \int_a^b \left[\int_0^{2\pi} \cos\phi' e^{j\beta p' \sin\theta \cos\phi'} d\phi' \right] dp'$$

$$= 2C \int_a^b \left[\int_0^\pi \cos\phi' e^{j\beta p' \sin\theta \cos\phi'} d\phi' + \int_0^{2\pi} \cos\phi' e^{j\beta p' \sin\theta \cos\phi'} d\phi' \right] dp'$$

$$= 2C \int_a^b \left[\int_0^\pi \cos\phi' e^{j\beta p' \sin\theta \cos\phi'} d\phi' - \underbrace{\int_0^\pi \cos\phi'' e^{-j\beta p'' \sin\theta \cos\phi''} d\phi''}_{\text{Let } \phi'' = \phi' - \pi \Rightarrow d\phi'' = d\phi'} \right] dp'$$

$$L_\phi = 2C \int_a^b \left[J_1(\beta p' \sin\theta) - J_1(-\beta p' \sin\theta) \right] dp' = 4C \int_a^b J_1(\beta p' \sin\theta) dp'$$

$$\text{Using } \int x^{1-p} J_p(ax) dx = -\frac{1}{a} x^{1-p} J_{p-1}(ax)$$

$$L_\phi = -4C \left[\frac{1}{\beta \sin\theta} (J_0(\beta b \sin\theta) - J_0(\beta a \sin\theta)) \right]$$

If however the slot is very thin, L_ϕ can be approximated by

$$L_\phi = 4C \int_a^b J_1(\beta p' \sin\theta) dp' \approx 4C J_1(\beta a \sin\theta) \int_a^b dp' = 4C (b-a) J_1(\beta a \sin\theta), a = \frac{a+b}{2}$$

Using (6-122a)-(6-122f), we can write the radiated fields as

$$\underline{E}_r \approx \underline{H}_r \approx 0$$

$$\underline{E}_\theta \approx -j \frac{\beta e^{-j\beta r}}{4\pi r} [L_\phi + \eta N_\phi] = j \frac{C e^{-j\beta r}}{\pi r} \left[\frac{J_0(\beta b \sin\theta) - J_0(\beta a \sin\theta)}{\sin\theta} \right] = j \frac{(b-a)C \beta e^{-j\beta r}}{\pi r} \frac{J_1(\beta a \sin\theta)}{\sin\theta}$$

$$\underline{E}_\phi \approx j \frac{\beta e^{-j\beta r}}{4\pi r} [L_\phi - \eta N_\phi] = 0, H_\theta = -\frac{\underline{E}_\phi}{\eta}, H_\phi = \frac{\underline{E}_\phi}{\eta}$$

6.39

(a) For the E-plane the normalized electric field is

$$E_{\phi n} = \frac{J_1(\beta a \sin \theta)}{\beta a \sin \theta}$$

whose maximum value is 0.5 when $\beta a \sin \theta = 0$. Its 0.707 value of its maximum value, or its 0.3535 value, occurs according to the table on page 943 when $\beta a \sin \theta_h \approx 1.616 \Rightarrow \theta_h = \sin^{-1}\left(\frac{1.616}{\beta a}\right) = \sin^{-1}\left(\frac{0.25722}{a}\right)$

Thus the half-power beamwidth is equal to

$$\Theta_h = 2\theta_h = 2 \sin^{-1}\left(\frac{0.25722}{a}\right)$$

When $a = 3\lambda$

$$\Theta_h = 2 \sin^{-1}\left(\frac{0.2572}{3}\right) = 2 \sin^{-1}(0.08573) = 2(4.918^\circ) = 9.836^\circ$$

(b) In the H-plane the normalized electric field is

$$E_{\phi n} = \cos \theta \frac{J_1(\beta a \sin \theta)}{\beta a \sin \theta} \xrightarrow[a \rightarrow \text{large}]{\sim} \frac{J_1(\beta a \sin \theta)}{\beta a \sin \theta}$$

since the $J_1(x)/x$ term varies much faster than the $\cos \theta$ term.

Therefore the half-power beamwidth in the H-plane is approximately the same as that in the E-plane, or

$$\Theta_h \approx 9.836^\circ$$

CHAPTER 7

7.1 a. $H_\phi = \frac{E_\theta}{\eta} = j \frac{\beta I_0 l e^{-j\beta r}}{4\pi r} \sin\theta [2 \cos(\beta h \cos\theta)]$

b. $S_{av} = \frac{1}{2} \operatorname{Re}[E \times H^*] = \frac{1}{2} \operatorname{Re}[\hat{a}_\theta E_\theta \times \hat{a}_\phi H_\phi^*] = \hat{a}_r \operatorname{Re}[E_\theta H_\phi^*]$
 $= \hat{a}_r \frac{1}{2} \operatorname{Re}[E_\theta \frac{E_\theta^*}{\eta}] = \hat{a}_r \frac{1}{2} \frac{|E_\theta|^2}{\eta}$

$$S_{av} = \hat{a}_r \frac{n}{2} \left| \frac{\beta I_0 l}{4\pi r} \right|^2 \sin^2\theta [2 \cos(\beta h \cos\theta)]^2 = \hat{a}_r 2n \left(\frac{\beta I_0 l}{4\pi r} \right)^2 \sin^2\theta \cos^2(\beta h \cos\theta)$$

c. $P_{av} = \iint_S S_{av} \cdot d\bar{s} = \int_0^{2\pi} \int_0^{\pi/2} \hat{a}_r S_{av} \hat{a}_r r^2 \sin\theta d\theta dr = \int_0^{2\pi} \int_0^{\pi/2} S_{av} r^2 \sin\theta d\theta dr$

$$P_{av} = 4\pi n \left| \frac{\beta I_0 l}{4\pi} \right|^2 \int_0^{\pi/2} \sin^3\theta \cos^2(\beta h \cos\theta) d\theta = \pi n \left| \frac{I_0 l}{2} \right|^2 \int_0^{\pi/2} \sin^3\theta \cos^2(\beta h \cos\theta) d\theta$$

$$P_{rad} = P_{av} = \pi n \left| \frac{I_0 l}{2} \right|^2 I$$

where $I = I_1 + I_2 = \int_0^{\pi/2} \sin^3\theta \cos^2(\beta h \cos\theta) d\theta = \int_0^{\pi/2} \sin^3\theta \left[\frac{1 + \cos(2\beta h \cos\theta)}{2} \right] d\theta$

$$I = I_1 + I_2 = \frac{1}{2} \int_0^{\pi/2} \sin^3\theta d\theta + \frac{1}{2} \int_0^{\pi/2} \sin^3\theta \cos(2\beta h \cos\theta) d\theta$$

$$I_1 = \frac{1}{2} \int_0^{\pi/2} \sin^3\theta d\theta = -\frac{1}{6} \cos\theta (\sin^2\theta + 2) \Big|_0^{\pi/2} = \frac{1}{3}$$

$$I_2 = \frac{1}{2} \int_0^{\pi/2} \sin^3\theta \cos(2\beta h \cos\theta) d\theta = \frac{1}{2} \int_0^{\pi/2} \sin^2\theta \cos(2\beta h \cos\theta) \sin\theta d\theta$$

Let $u = \sin^2\theta \quad du = 2 \sin\theta \cos\theta d\theta$ $d\theta = -\frac{\cos(2\beta h \cos\theta)}{2\beta h} d(2\beta h \cos\theta)$, $v = -\frac{1}{2\beta h} \sin(2\beta h \cos\theta)$

Thus

$$I_2 = -\underbrace{\frac{\sin^2\theta \sin(2\beta h \cos\theta)}{4\beta h} \Big|_0^{\pi/2}}_{=0} + \frac{1}{2\beta h} \int_0^{\pi/2} \cos\theta \sin(2\beta h \cos\theta) \sin\theta d\theta$$

Let $u = \cos\theta \quad du = -\frac{1}{2\beta h} \sin(2\beta h \cos\theta) d(2\beta h \cos\theta)$
 $du = -\sin\theta d\theta \quad v = \frac{1}{2\beta h} \cos(2\beta h \cos\theta)$

$$I_2 = 0 + \frac{1}{2\beta h} \left\{ \frac{\cos\theta}{2\beta h} \cos(2\beta h \cos\theta) \Big|_0^{\pi/2} + \frac{1}{2\beta h} \int_0^{\pi/2} \cos(2\beta h \cos\theta) \sin\theta d\theta \right\}$$

$$I_2 = \frac{1}{2\beta h} \left\{ -\frac{1}{2\beta h} \cos(2\beta h) - \frac{1}{(2\beta h)^2} \sin(2\beta h \cos\theta) \Big|_0^{\pi/2} \right\} = \left\{ -\frac{\cos(2\beta h)}{(2\beta h)^2} + \frac{\sin(2\beta h)}{(2\beta h)^3} \right\}$$

Therefore

$$P_{rad} = \pi n \left| \frac{I_0 l}{2} \right|^2 \left[\frac{1}{3} - \frac{\cos(2\beta h)}{(2\beta h)^2} + \frac{\sin(2\beta h)}{(2\beta h)^3} \right]$$

7.2 a. From Problem 7.1, part b

$$S_{av} = 2\eta \left| \frac{8I_0 l}{4\pi r} \right|^2 \sin^2 \theta \cos^2(\beta h \cos \theta) = \frac{\eta}{2} \left| \frac{I_0 l}{2r} \right|^2 \sin^2 \theta \cos^2(\beta h \cos \theta)$$

$$U = r^2 S_{av} = \frac{\eta}{2} \left| \frac{I_0 l}{2r} \right|^2 \sin^2 \theta \cos^2(\beta h \cos \theta)$$

b. $D_o = \frac{4\pi U_{max}}{P_{rad}}$, $U_{max} = U_{|r \rightarrow \infty} = \frac{\eta}{2} \left| \frac{I_0 l}{2r} \right|^2$

From Problem 7.1, part c

$$P_{rad} = \pi \eta \left| \frac{I_0 l}{2r} \right|^2 \left[\frac{1}{3} - \frac{\cos(2\beta h)}{(2\beta h)^2} + \frac{\sin(2\beta h)}{(2\beta h)^3} \right] = \pi \eta \left| \frac{I_0 l}{2r} \right|^2 F(\beta h)$$

where $F(\beta h) = \left[\frac{1}{3} - \frac{\cos(2\beta h)}{(2\beta h)^2} + \frac{\sin(2\beta h)}{(2\beta h)^3} \right]$

Thus

$$D_o = \frac{4\pi \frac{\eta}{2} \left| \frac{I_0 l}{2r} \right|^2}{\pi \eta \left| \frac{I_0 l}{2r} \right|^2 F(\beta h)} = \frac{2}{F(\beta h)}$$

c. $R_r = \frac{2 P_{rad}}{|I_0|^2} = 2\pi \eta \left| \frac{l}{2r} \right|^2 F(\beta h)$

7.3 The radiation characteristics of a vertical magnetic dipole can be obtained easily by referring to the solution of the radiation characteristics of a vertical electric dipole and using duality.

Based on the solution of Example 6-3, and duality, we can write:

- $F_r = \frac{Im l e^{-j\beta r}}{4\pi r} \cos \theta$, $F_\theta = -\frac{Im l e^{-j\beta r}}{4\pi r} \sin \theta$, $F_\phi = 0$

- $H_r = \frac{Im l \cos \theta}{2\pi r^2} \left(1 + \frac{1}{j\beta r} \right) e^{-j\beta r}$

$$H_\theta = j \frac{\beta Im l \sin \theta}{4\pi r^2} \left[1 + \frac{1}{j\beta r} - \frac{1}{(\beta r)^2} \right] e^{-j\beta r}$$

$$H_\phi = 0$$

- $E_r = E_\theta = 0$

- $E_\phi = -j \frac{\beta Im l \sin \theta}{4\pi r^2} \left(1 + \frac{1}{j\beta r} \right) e^{-j\beta r}$

Cont'd

7.3 cont'd

- $S_{ave} = \frac{1}{2} \operatorname{Re} [E_x H^*] = \hat{a}_r \frac{1}{2\eta} \left(\frac{\beta l}{4\pi} \right) \frac{|I_m|^2}{r^2} \sin^2 \theta = \hat{a}_r S_r = \hat{a}_r S_{ave}$

- $M \approx r^2 S_{ave} = \frac{1}{2\eta} \left(\frac{\beta l}{4\pi} \right)^2 |I_m|^2 \sin^2 \theta \Rightarrow M_{max} = \frac{1}{2\eta} \left(\frac{\beta l}{4\pi} \right)^2 |I_m|^2$

- $P_{rad} = \int_0^{2\pi} \int_0^\pi U(\theta, \phi) \sin \theta d\theta d\phi = \frac{1}{2\eta} \left(\frac{\beta l}{4\pi} \right)^2 \frac{4}{3} (2\pi) = \frac{8\pi}{2(3\eta)} \left(\frac{\beta l}{4\pi} \right)^2 |I_m|^2$

- $D_o = \frac{4\pi M_{max}}{P_{rad}} = \frac{4\pi}{\frac{8\pi}{3\eta} \left(\frac{\beta l}{4\pi} \right)^2} = \frac{3}{2}$

- $R_r = \frac{2 P_{rad}}{|I_m|^2} = \frac{2}{2} \left[\frac{8\pi}{3\eta} \left(\frac{\beta l}{4\pi} \right)^2 \right] = \frac{8\pi}{12\eta} \left(\frac{l}{2} \right)^2 = \frac{8}{12(120)} \left(\frac{l}{2} \right)^2$

$$R_r = \frac{1}{180} \left(\frac{l}{2} \right)^2$$

7.4

$$E_r = E_\theta = H_\phi = 0, \quad F_r = \frac{I_m l e^{-j\beta r}}{4\pi r} \cos \theta$$

$$E_\phi = -j \frac{B I_m l e^{-j\beta r}}{4\pi r} \sin \theta$$

$$H_\theta = j \frac{B I_m l e^{-j\beta r}}{4\pi r} \sin \theta, \quad F_\phi = 0$$

7.5

$$F = \frac{\epsilon}{4\pi} \int_{-l/2}^{+l/2} I_m \frac{e^{-j\beta r}}{4\pi r} dl \approx \hat{a}_x \frac{\epsilon I_m e^{-j\beta r}}{4\pi r} \int_{-\delta/2}^{+\delta/2} dx' = \hat{a}_x \frac{\epsilon l I_m e^{-j\beta r}}{4\pi r}$$

- Using (II-12a)

$$F_r = F_x \sin \theta \cos \phi, \quad F_\theta = F_x \cos \theta \sin \phi, \quad F_\phi = F_x \sin \phi$$

$$F_y = \frac{\epsilon l I_m e^{-j\beta r}}{4\pi r} \sin \theta \cos \phi, \quad F_\theta = \frac{\epsilon l I_m e^{-j\beta r}}{4\pi r} \cos \theta \sin \phi, \quad F_\phi = -\frac{\epsilon l I_m e^{-j\beta r}}{4\pi r} \sin \phi$$

- Using (6-119a)-(6-119d)

$$E_\theta = -j w \eta F_\phi = j \frac{w \eta \epsilon l I_m e^{-j\beta r}}{4\pi r} \sin \phi = j \frac{\beta l I_m e^{-j\beta r}}{4\pi r} \sin \phi$$

$$E_\phi = +j w \eta F_\theta = j \frac{w \eta \epsilon l I_m e^{-j\beta r}}{4\pi r} \cos \theta \cos \phi = j \frac{\beta l I_m e^{-j\beta r}}{4\pi r} \cos \theta \cos \phi$$

cont'd

7.5 cont'd

$$H_\theta = -j\omega F_\theta = -j\omega \ell I_m e^{-j\beta r} \cos\theta \cos\phi = -E_\theta / \eta$$

$$H_\phi = -j\omega F_\phi = j\omega \ell I_m e^{-j\beta r} \sin\theta = E_\phi / \eta$$

- $S_{ave} = \frac{1}{2} \operatorname{Re} [E_x H^*] = \hat{a}_r \frac{1}{2\pi} \left[|E_0|^2 + |E_\phi|^2 \right] = \hat{a}_r \frac{1}{2\pi} \left[\frac{\beta \ell I_m}{4\pi r} \right]^2 \left[\sin^2 \phi + \cos^2 \theta \cos^2 \phi \right]$
- $S_{ave} = \hat{a}_r S_r = \hat{a}_r \frac{I_m^2}{2\pi r} \left[\frac{\beta \ell}{4\pi} \right]^2 \left[\sin^2 \phi + \cos^2 \theta \cos^2 \phi \right]$

- $U = r^2 S_r = \left| \frac{I_m}{2\pi} \left(\frac{\beta \ell}{4\pi} \right)^2 \right|^2 \left[\sin^2 \phi + \cos^2 \theta \cos^2 \phi \right]$

- $P_{rad} = \int_0^{2\pi} U \sin\theta d\theta d\phi = \left| \frac{I_m}{2\pi} \left(\frac{\beta \ell}{4\pi} \right)^2 \right|^2 \left[\sin^2 \phi + \cos^2 \theta \cos^2 \phi \right] \sin\theta d\theta d\phi$

Using a similar integration procedure as in solution to Problem 6.21:

$$P_{rad} = \frac{|I_m|^2}{2\pi} (2\pi) \left(\frac{4}{3} \right) \left(\frac{\beta \ell}{4\pi} \right)^2 = \frac{|I_m|^2}{2\pi} \left(\frac{8\pi}{3} \right) \left(\frac{\beta \ell}{4\pi} \right)^2$$

$$U_{max} = \left| \frac{I_m}{2\pi} \left(\frac{\beta \ell}{4\pi} \right)^2 \right|^2 \left[\sin^2 \phi + \cos^2 \theta \cos^2 \phi \right] = \frac{|I_m|^2}{2\pi} \left(\frac{\beta \ell}{4\pi} \right)^2$$

or $\theta = 0, \phi = 0$
 $\theta = 90^\circ, \phi = 90^\circ$

$$D_0 = \frac{4\pi U_{max}}{P_{rad}} = \frac{4\pi}{(8\pi)/3} = \left(\frac{3}{2} \right)$$

- $R_r = 2P_{rad} - 2 \frac{1}{2\pi} \left(\frac{8\pi}{3} \right) \left(\frac{\beta \ell}{4\pi} \right)^2 = \frac{1}{\pi} \left(\frac{8\pi}{3} \right) \left[\frac{2\pi \ell}{2(4\pi)} \right]^2 = \frac{1}{\pi} \left(\frac{8\pi}{3} \right) \left(\frac{\ell}{2} \right)^2$

$$R_r = \frac{8\pi}{120\pi(4/3)} \left(\frac{\ell}{2} \right)^2 = \frac{1}{3(60)} \left(\frac{\ell}{2} \right)^2 = \frac{1}{120} \left(\frac{\ell}{2} \right)^2 + l/2$$

7.6

$$F = \frac{\epsilon}{4\pi} \int_{-l/2}^{l/2} T_m \frac{e^{-j\beta R}}{R} dR \xrightarrow{R \approx r} \text{by } \epsilon I_m e^{-j\beta r} \frac{l_y}{4\pi r} \quad \text{by } \epsilon I_m e^{-j\beta r}$$

$$F = \hat{a}_y \frac{\epsilon l I_m e^{-j\beta r}}{4\pi r} = \hat{a}_y F_y, \quad F_y = \frac{\epsilon l I_m e^{-j\beta r}}{4\pi r}$$

$$F_r = F_y \sin\theta \sin\phi = \frac{\epsilon l I_m e^{-j\beta r}}{4\pi r} \sin\theta \sin\phi$$

$$F_\theta = F_y \cos\theta \sin\phi = \frac{\epsilon l I_m e^{-j\beta r}}{4\pi r} \cos\theta \sin\phi$$

$$F_\phi = F_y \cos\phi = \frac{\epsilon l I_m e^{-j\beta r}}{4\pi r} \cos\phi$$

cont'd

7.6 | (cont'd) Using (6-119a)-(6-119d)

$$E_\theta = -j\omega\eta F_\theta = -j \frac{\omega\epsilon I_m e^{-j\beta r}}{4\pi r} \cos\phi = -j \frac{\beta I_m e^{-j\beta r}}{4\pi r} \cos\phi$$

$$E_\phi = j\omega\eta F_\phi = j \frac{\omega\epsilon I_m e^{-j\beta r}}{4\pi r} \sin\phi = j \frac{\beta I_m e^{-j\beta r}}{4\pi r} \cos\phi \sin\phi$$

$$H_\theta = -j\omega F_\theta = -j \frac{\omega\epsilon I_m e^{-j\beta r}}{4\pi r} \cos\phi \sin\phi = -j \frac{\beta I_m e^{-j\beta r}}{4\pi r} \cos\phi \sin\phi = -\frac{E_\phi}{\eta}$$

$$H_\phi = -j\omega F_\phi = -j \frac{\omega\epsilon I_m e^{-j\beta r}}{4\pi r} \cos\phi = -j \frac{\beta I_m e^{-j\beta r}}{4\pi r} \cos\phi = \frac{E_\theta}{\eta}$$

7.7 | For a horizontal electric dipole, directed along the y -direction, the far-zone electric and magnetic fields are given by (see solution of Problem 6.22) by

$$A_r = \frac{l\mu I_0 e^{-j\beta r}}{4\pi r} \sin\phi \sin\theta, A_\theta = \frac{l\mu I_0 e^{-j\beta r}}{4\pi r} \cos\phi \sin\theta, A_\phi = \frac{l\mu I_0 e^{-j\beta r}}{4\pi r} \cos\phi$$

$$E_r \approx H_r \approx 0, E_\theta \approx$$

$$E_\phi \approx -j \frac{\omega l I_0 e^{-j\beta r}}{4\pi r} \cos\phi \sin\theta, H_\phi \approx -E_\phi / \eta$$

$$E_\phi \approx -j \frac{\omega l I_0 e^{-j\beta r}}{4\pi r} \cos\phi, H_\phi \approx E_\phi / \eta$$

When the infinitesimal electric dipole is placed a height h above a PEC ground plane, its image is, according to Figures 7-2 and 7-5, in the $-y$ direction (same magnitude, 180° phase difference). Therefore, according to (7-17) its array factor is

$$AF = 2j \sin(\beta h \cos\theta)$$

and the total fields are now given by

$$E_\theta \approx -j \frac{\omega l I_0 e^{-j\beta r}}{4\pi r} [2j \sin(\beta h \cos\theta)] \cos\phi \sin\theta$$

$$E_\phi \approx -j \frac{\omega l I_0 e^{-j\beta r}}{4\pi r} [2j \sin(\beta h \cos\theta)] \cos\phi$$

Cont'd

7.7 [cont'd.] The radiation intensity is given

$$\bullet \text{II} = \frac{4}{2\pi r^2} \left[|E_\theta|^2 + |E_\phi|^2 \right] = \frac{|I_0|^2 (\omega \mu)^2}{2\pi} \left\{ \cos^2 \theta \sin^2 \phi + \cos^2 \phi \right\} \left[2 \sin(\beta h \cos \theta) \right]^2$$

which reduces to

$$\text{II} = \frac{\eta}{2} |I_0|^2 \left(\frac{l}{\lambda} \right)^2 \left[\cos^2 \theta \sin^2 \phi + \cos^2 \phi \right] \sin^2(\beta h \cos \theta)$$

- The power radiated is

$$P_{\text{rad}} = \int_{0}^{2\pi} \int_{\pi/2}^{\pi} II(\theta, \phi) \sin \theta d\theta d\phi = \frac{\eta}{2} |I_0|^2 \left(\frac{l}{\lambda} \right)^2 \int_{0}^{2\pi} \left[\cos^2 \theta \sin^2 \phi + \cos^2 \phi \right] \sin^2(\beta h \cos \theta) \sin \theta d\theta d\phi$$

Using an integration procedure similar to that indicated in the solution of Problem 7.1, it can be shown that the above reduces to

$$P_{\text{rad}} = \eta \left(\frac{\pi}{2} \right) |I_0|^2 \left(\frac{l}{\lambda} \right)^2 \left[\frac{2}{3} - \frac{\sin(2\beta h)}{2\beta h} - \frac{\cos(2\beta h)}{(2\beta h)^2} + \frac{\sin(2\beta h)}{(2\beta h)^3} \right]$$

- Using the maximum value of $\text{II}_{\max} = \text{II}(\theta, \phi)$ from above, it can be shown that the directivity can be written as

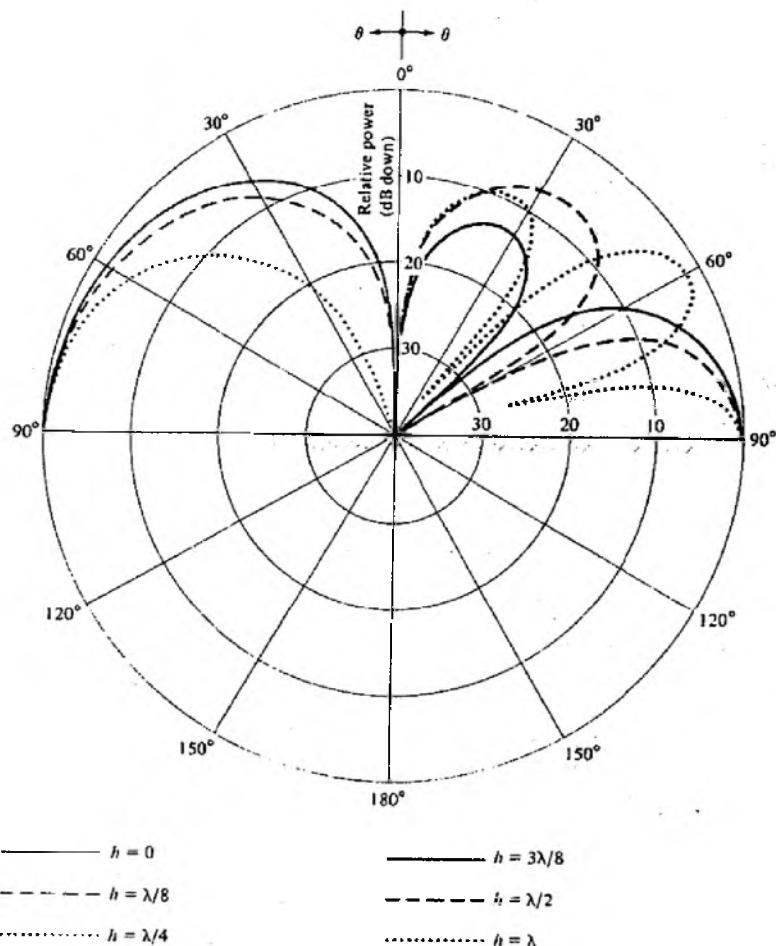
$$D_0 = \frac{4\pi \text{II}_{\max}}{P_{\text{rad}}} = \begin{cases} \frac{4 \sin^2(\beta h)}{R(\beta h)} & \beta h \leq \pi/2 \quad (h \leq \lambda/4) \\ \frac{4}{R(\beta h)} & \beta h > \pi/2 \quad (h > \lambda/4) \end{cases}$$

$$R(\beta h) = \left[\frac{2}{3} - \frac{\sin(2\beta h)}{2\beta h} - \frac{\cos(2\beta h)}{(2\beta h)^2} + \frac{\sin(2\beta h)}{(2\beta h)^3} \right]$$

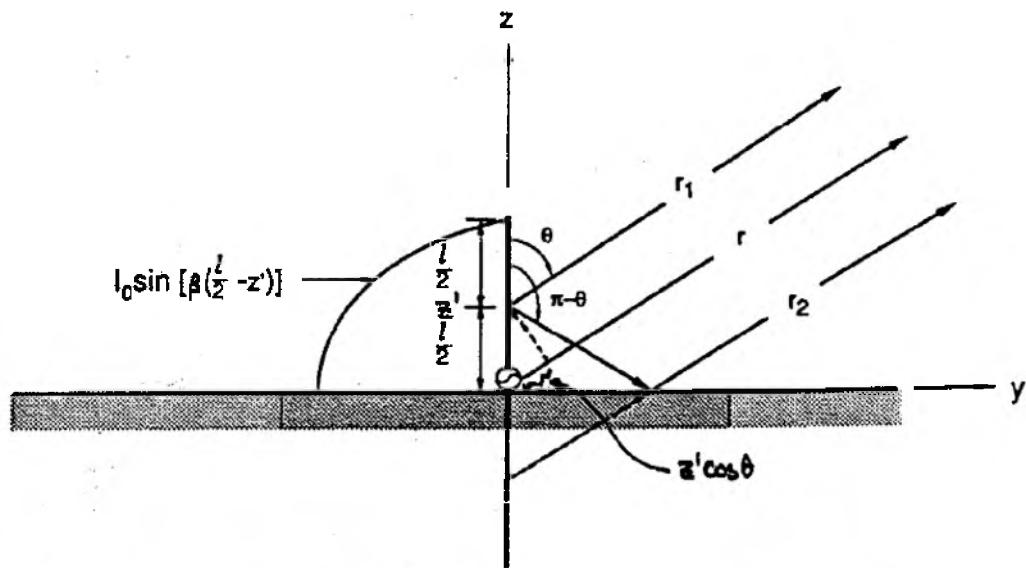
- The radiation resistance, according to the definition of

$$\begin{aligned} R_r &= 2 P_{\text{rad}} = 2\eta \left(\frac{\pi}{2} \right) \left(\frac{l}{\lambda} \right)^2 \left[\frac{2}{3} - \frac{\sin(2\beta h)}{2\beta h} - \frac{\cos(2\beta h)}{(2\beta h)^2} + \frac{\sin(2\beta h)}{(2\beta h)^3} \right] \\ &= \eta \pi \left(\frac{l}{\lambda} \right)^2 \left[\frac{2}{3} - \frac{\sin(2\beta h)}{2\beta h} - \frac{\cos(2\beta h)}{(2\beta h)^2} + \frac{\sin(2\beta h)}{(2\beta h)^3} \right] \end{aligned}$$

7.8



7.9



The direct far-zone field is obtained using (7.9) modified as

$$E_\theta^i = j\eta \frac{\beta e^{-j\beta r_1}}{4\pi r_1} \sin\theta \int_{-\ell/2}^{\ell/2} I_2(z') e^{j\beta z' \sin\theta} dz' = j\eta \frac{\beta e^{-j\beta r_1}}{4\pi r_1} \sin\theta \int_{-\ell/2}^{\ell/2} \sin[\beta(\frac{l}{2} - z')] e^{j\beta z' \cos\theta} dz' \quad (1)$$

The integral can be integrated as

$$\begin{aligned} \int_{-\ell/2}^{\ell/2} \sin[\beta(\frac{l}{2} - z')] e^{j\beta z' \cos\theta} dz' &= \frac{e^{j\beta z' \cos\theta}}{(j\beta \cos\theta)^2 + \beta^2} \left\{ j\beta \cos\theta \sin[\beta(\frac{l}{2} - z')] + \beta \cos[\beta(\frac{l}{2} - z')] \right\} \Big|_{-\ell/2}^{\ell/2} \\ &= \frac{1}{\beta^2 \sin^2\theta} \left[\beta e^{j\frac{\beta l}{2} \cos\theta} - e^{-j\frac{\beta l}{2} \cos\theta} \right] \left[j\beta \cos\theta \sin(\beta l) + \beta \cos(\beta l) \right] \\ \int_{-\ell/2}^{\ell/2} \sin[\beta(\frac{l}{2} - z')] e^{j\beta z' \cos\theta} dz' &= \frac{1}{\beta \sin^2\theta} \left\{ e^{j\frac{\beta l}{2} \cos\theta} - e^{-j\frac{\beta l}{2} \cos\theta} \right\} \left[j \cos\theta \sin(\beta l) + \cos(\beta l) \right] \end{aligned} \quad (2)$$

Substituting (2) into (1) we can write the direct far-zone field as

$$E_\theta^i = j\eta \frac{I_0 e^{-j\beta r_1}}{4\pi r_1} \left\{ \frac{e^{j\frac{\beta l}{2} \cos\theta} - e^{-j\frac{\beta l}{2} \cos\theta}}{\sin\theta} \left[j \cos\theta \sin(\beta l) + \cos(\beta l) \right] \right\} \quad (3)$$

Using the far zone approximations of

$$\begin{aligned} r_1 &\approx r - \frac{l}{2} \cos\theta && \text{for phase terms} \\ r_1 &\approx r && \text{for amplitude terms} \end{aligned} \quad (4)$$

cont'd.

7.9 Cont'd. we can reduce (3) to

$$E_\theta^L = j\eta \frac{I_0 e^{-j\beta r}}{4\pi r} \left\{ \frac{e^{j\beta l \cos\theta} - j \cos\theta \sin(\beta l) - \cos(\beta l)}{\sin\theta} \right\} \quad (5)$$

By referring to the geometry of the figure it is evident that the reflected field can be obtained from the direct field by replacing θ by $\pi-\theta$ in (3).

Doing this, the reflected field can be written by referring to (3) as

$$\begin{aligned} E_\theta^R &= j\eta \frac{I_0 e^{-j\beta r_2}}{4\pi r_2} \left\{ \frac{e^{j\frac{\beta l}{2} \cos(\pi-\theta)} - e^{-j\frac{\beta l}{2} \cos(\pi-\theta)} [j \cos(\pi-\theta) \sin(\beta l) + \cos(\beta l)]}{\sin(\pi-\theta)} \right\} \\ &= j\eta \frac{I_0 e^{-j\beta r_2}}{4\pi r_2} \left\{ \frac{e^{-j\frac{\beta l}{2} \cos\theta} - e^{j\frac{\beta l}{2} \cos\theta} [-j \cos\theta \sin(\beta l) + \cos(\beta l)]}{\sin\theta} \right\} \end{aligned} \quad (6)$$

which by using the far zone approximations of

$$\begin{aligned} r_2 &\approx r + \frac{l}{2} \cos\theta && \text{for phase terms} \\ r_2 &\approx r && \text{for amplitude terms} \end{aligned} \quad (7)$$

reduces to

$$E_\theta^R = j\eta \frac{I_0 e^{-j\beta r}}{4\pi r} \left\{ \frac{e^{-j\beta l \cos\theta} + j \cos\theta \sin(\beta l) - \cos(\beta l)}{\sin\theta} \right\} \quad (8)$$

Combining (5) and (8) leads to

$$E_\theta^T = E_\theta^L + E_\theta^R = j\eta \frac{I_0 e^{-j\beta r}}{4\pi r} \left\{ \frac{\cos(\beta l \cos\theta) - \cos(\beta l)}{\sin\theta} \right\} \quad (9)$$

When $l = \pi/4$, (9) reduces to

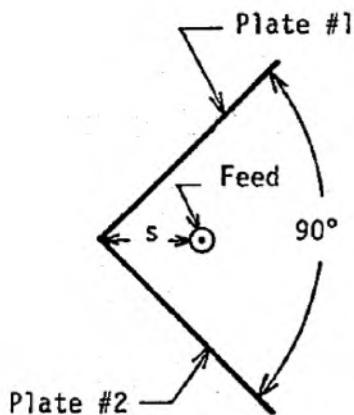
$$E_\theta^T = j\eta \frac{I_0 e^{-j\beta r}}{4\pi r} \left\{ \frac{\cos(\frac{\pi}{2} \cos\theta)}{\sin\theta} \right\} \quad (10)$$

The other electric and magnetic field components are

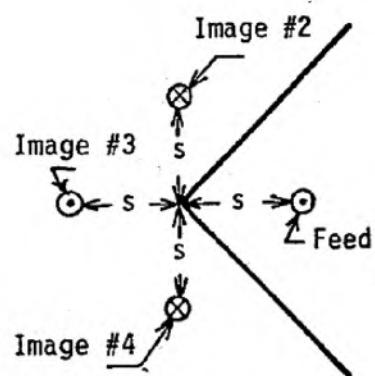
$$E_r \approx E_\phi \approx H_r \approx H_\phi \approx 0$$

$$H_\phi^T = \frac{E_\theta^T}{\eta}$$

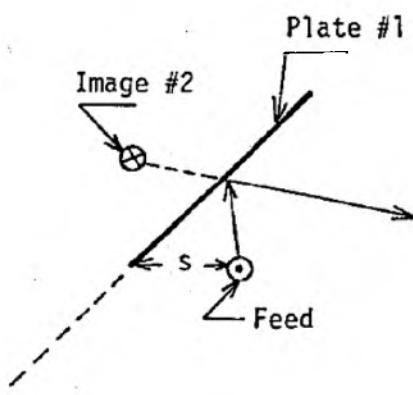
- 7.10** a. The number of images (3 of them), their polarizations, and their positions are shown below.



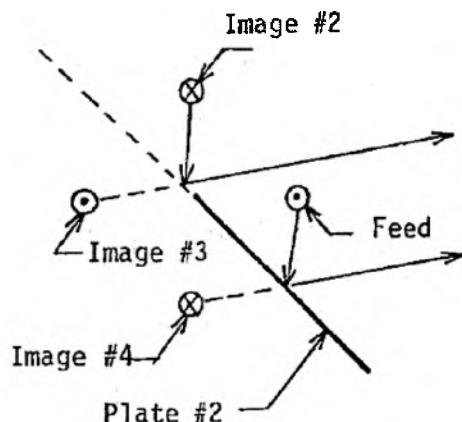
(a) 90° corner reflector



(b) images for 90° corner reflector



(c) placement of image #2
(due to feed)



(d) placement of images #4
and #3 (due to feed and
image #2)

Cont'd.

7.10 cont'd. b. The total field is derived by summing the contributions from the feed and its images. Thus

$$E^t(r, \theta, \phi) = E_1(r_1, \theta, \phi) + E_2(r_2, \theta, \phi) + E_3(r_3, \theta, \phi) + E_4(r_4, \theta, \phi) \quad (1)$$

In the far zone, the normalized field can be written as

$$\begin{aligned} E^t(r, \theta, \phi) &= f(\theta, \phi) \frac{e^{-j\beta r_1}}{r_1} - f(\theta, \phi) \frac{e^{-j\beta r_2}}{r_2} + f(\theta, \phi) \frac{e^{-j\beta r_3}}{r_3} - f(\theta, \phi) \frac{e^{-j\beta r_4}}{r_4} \\ &= [e^{j\beta s \cos \psi_1} - e^{j\beta s \cos \psi_2} + e^{j\beta s \cos \psi_3} - e^{j\beta s \cos \psi_4}] f(\theta, \phi) \frac{e^{-j\beta r}}{r} \end{aligned} \quad (2)$$

where $\cos \psi_1 = \hat{a}_x \cdot \hat{a}_r = \sin \theta \cos \phi$

$$\cos \psi_2 = \hat{a}_y \cdot \hat{a}_r = \sin \theta \sin \phi$$

$$\cos \psi_3 = -\hat{a}_x \cdot \hat{a}_r = -\sin \theta \cos \phi \quad (3)$$

$$\cos \psi_4 = -\hat{a}_y \cdot \hat{a}_r = -\sin \theta \sin \phi$$

since

$$\hat{a}_r = \hat{a}_x \sin \theta \cos \phi + \hat{a}_y \sin \theta \sin \phi + \hat{a}_z \cos \theta \quad (3a)$$

Substituting (3) into (2) we can write it as

$$E^t(r, \theta, \phi) = 2 [\cos(\beta s \sin \theta \cos \phi) - \cos(\beta s \sin \theta \sin \phi)] f(\theta, \phi) \frac{e^{-j\beta r}}{r} \quad (4)$$

By letting the field of the isolated element (in this case the dipole) to be

$$E_g^0 = f(\theta, \phi) \frac{e^{-j\beta r}}{r} = j \eta \frac{\beta I_0 e^{-j\beta r}}{4\pi r} \sin \theta \quad (5)$$

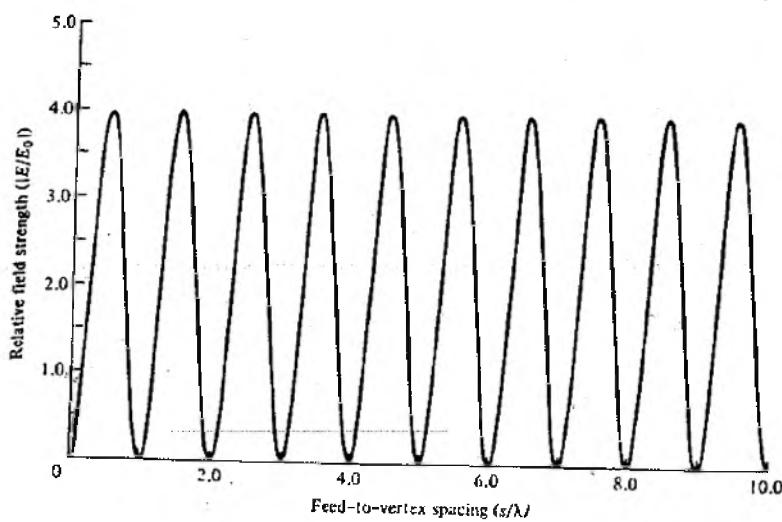
then (4) can be written as

$$E^t(r, \theta, \phi) = E_g^0 F(\beta s)$$

where

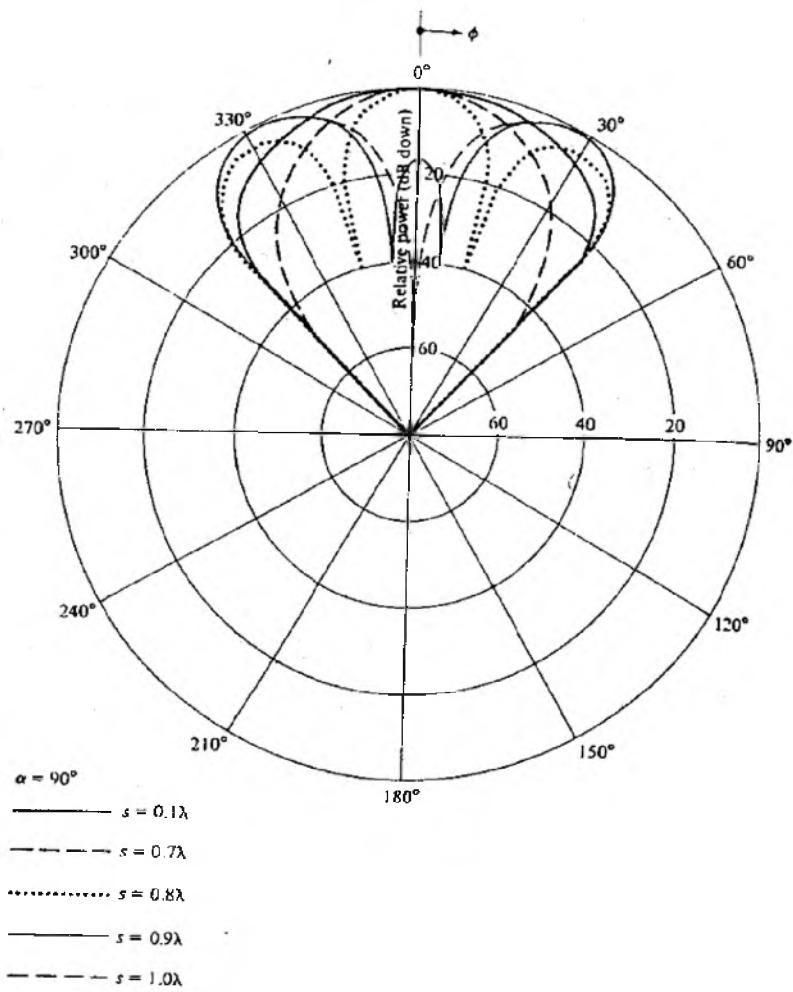
$$F(\beta s) = 2 [\cos(\beta s \sin \theta \cos \phi) - \cos(\beta s \sin \theta \sin \phi)]$$

7.11

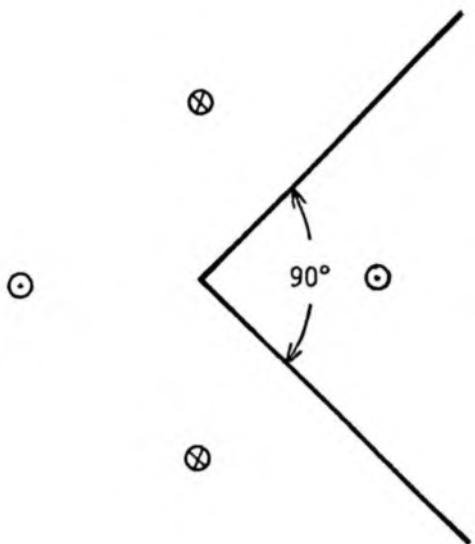


The maximum value of $F(s)$ is 4. The function is periodic with a period of π .

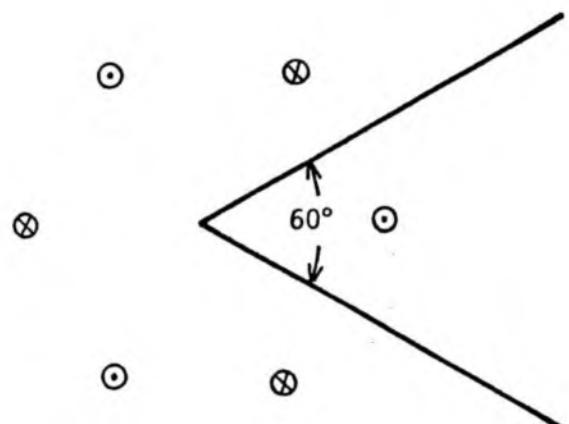
7.12



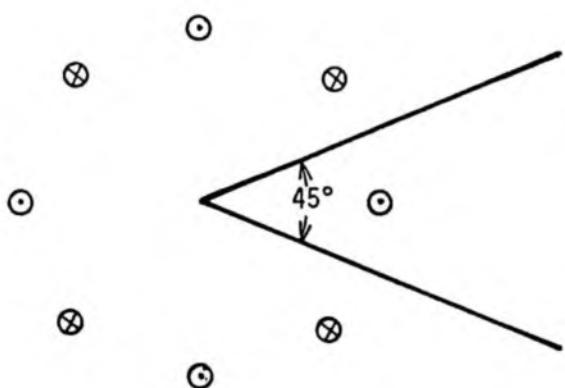
7.13 For all three corner reflectors ($\alpha = 60^\circ, 45^\circ, 30^\circ$), the geometrical coordinate system is that shown in Problem 7.10. The sources will be numbered so that the feed will be #1. The images will be designated as #2, #3, (in a clockwise rotation) as shown in the figure of the solution of Problem 7.10 for the 90° corner reflector. The image arrangement for each geometry is shown below.



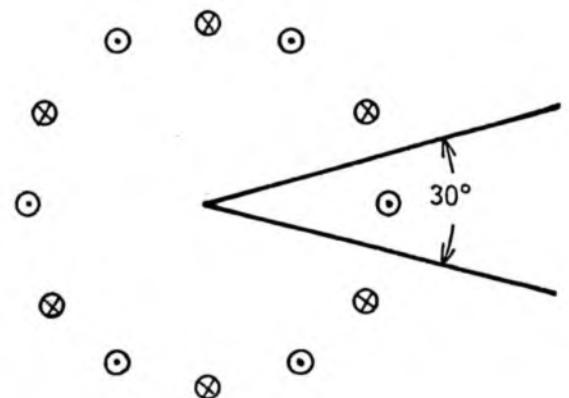
(a) 90°



(b) 60°



(c) 45°



(d) 30°

cont'd.

7.13 Cont'd.

$$\alpha = 60^\circ$$

Using the arrangement of Figure b of the previous page

$$E(r, \theta, \phi) = \left(\frac{e^{-j\beta r_1}}{r_1} + \frac{e^{-j\beta r_2}}{r_2} + \frac{e^{-j\beta r_3}}{r_3} + \frac{e^{-j\beta r_4}}{r_4} + \frac{e^{-j\beta r_5}}{r_5} + \frac{e^{-j\beta r_6}}{r_6} \right) f(\theta, \phi)$$

for the main source (#1) and the five images. For far-field observations

For Phase Terms

$$r_1 \approx r - s \cos \psi_1 = r - s (\hat{a}_x \cdot \hat{a}_r) = r - s \sin \theta \cos \phi$$

$$r_2 \approx r - s \cos \psi_2 = r - s [(0.5 \hat{a}_x + 0.866 \hat{a}_y) \cdot \hat{a}_r] = r - s (0.5 \sin \theta \cos \phi + 0.866 \sin \theta \sin \phi)$$

$$r_3 \approx r - s \cos \psi_3 = r - s [(-0.5 \hat{a}_x + 0.866 \hat{a}_y) \cdot \hat{a}_r] = r - s (-0.5 \sin \theta \cos \phi + 0.866 \sin \theta \sin \phi)$$

$$r_4 \approx r - s \cos \psi_4 = r - s (-\hat{a}_x \cdot \hat{a}_r) = r + s \sin \theta \cos \phi$$

$$r_5 \approx r - s \cos \psi_5 = r - s [(-0.5 \hat{a}_x - 0.866 \hat{a}_y) \cdot \hat{a}_r] = r + s (0.5 \sin \theta \cos \phi + 0.866 \sin \theta \sin \phi)$$

$$r_6 \approx r - s \cos \psi_6 = r - s [(0.5 \hat{a}_x - 0.866 \hat{a}_y) \cdot \hat{a}_r] = r + s (-0.5 \sin \theta \cos \phi + 0.866 \sin \theta \sin \phi)$$

$$\text{where } \hat{a}_r = \hat{a}_x \sin \theta \cos \phi + \hat{a}_y \sin \theta \sin \phi + \hat{a}_z \cos \theta$$

For Amplitude Terms

$$r_1 \approx r_2 \approx r_3 \approx r_4 \approx r_5 \approx r_6 \approx r$$

Making these substitutions and combining terms (first with fourth, second with fifth, third with sixth), we can write that

$$E(r, \theta, \phi) = f(\theta, \phi) \frac{e^{-j\beta r}}{r} 2 \left\{ \sin(\beta s \sin \theta \cos \phi) - \sin[\beta s (0.5 \sin \theta \cos \phi + 0.866 \sin \theta \sin \phi)] \right. \\ \left. + \sin[\beta s (-0.5 \sin \theta \cos \phi + 0.866 \sin \theta \sin \phi)] \right\}$$

Using the identities of

$$\sin(x \pm y) = \sin(x) \cos(y) \pm \cos(x) \sin(y)$$

reduces the field to

$$\frac{E}{E_0} = 2 \left[\sin(\beta s \sin \theta \cos \phi) - 2 \sin\left(\frac{\beta s}{2} \sin \theta \cos \phi\right) \cos\left(\frac{\sqrt{3}}{2} \beta s \sin \theta \sin \phi\right) \right] \text{ where } E_0 = f(\theta, \phi) \frac{e^{-j\beta r}}{r}$$

Utilizing the identity of

$$\sin(2x) = 2 \sin(x) \cos(x)$$

we can write in the final form that

$$\frac{E}{E_0} = 4 \sin\left(\frac{\beta s}{2} \sin \theta \cos \phi\right) \left[\cos\left(\frac{\beta s}{2} \sin \theta \cos \phi\right) - \cos\left(\frac{\sqrt{3}}{2} \beta s \sin \theta \sin \phi\right) \right]$$

Cont'd.

7.13 cont'd.

$\alpha = 45^\circ$

Using the source arrangement of Figure C

$$E(r, \theta, \phi) = \left(\frac{e^{-j\beta r_1}}{r_1} + \frac{e^{-j\beta r_2}}{r_2} + \frac{e^{-j\beta r_3}}{r_3} + \frac{e^{-j\beta r_4}}{r_4} + \frac{e^{-j\beta r_5}}{r_5} + \frac{e^{-j\beta r_6}}{r_6} + \frac{e^{-j\beta r_7}}{r_7} + \frac{e^{-j\beta r_8}}{r_8} \right) f(\theta, \phi)$$

For far-field observations

For Phase Terms

$$r_1 \approx r - s \cos \psi_1 = r - s (\hat{a}_x \cdot \hat{a}_r) = r - s \sin \theta \cos \phi$$

$$r_2 \approx r - s \cos \psi_2 = r - s \left[\frac{1}{\sqrt{2}} (\hat{a}_x + \hat{a}_y) \cdot \hat{a}_r \right] = r - \frac{s}{\sqrt{2}} (\sin \theta \cos \phi + \sin \theta \sin \phi)$$

$$r_3 \approx r - s \cos \psi_3 = r - s (\hat{a}_y \cdot \hat{a}_r) = r - s \sin \theta \sin \phi$$

$$r_4 \approx r - s \cos \psi_4 = r - s \left[\frac{1}{\sqrt{2}} (-\hat{a}_x + \hat{a}_y) \cdot \hat{a}_r \right] = r + \frac{s}{\sqrt{2}} (\sin \theta \cos \phi - \sin \theta \sin \phi)$$

$$r_5 \approx r - s \cos \psi_5 = r - s (-\hat{a}_x \cdot \hat{a}_r) = r + s \sin \theta \cos \phi$$

$$r_6 \approx r - s \cos \psi_6 = r - s \left[\frac{1}{\sqrt{2}} (-\hat{a}_x - \hat{a}_y) \cdot \hat{a}_r \right] = r + \frac{s}{\sqrt{2}} (\sin \theta \cos \phi + \sin \theta \sin \phi)$$

$$r_7 \approx r - s \cos \psi_7 = r - s (-\hat{a}_y \cdot \hat{a}_r) = r + s \sin \theta \sin \phi$$

$$r_8 \approx r - s \cos \psi_8 = r - s \left[\frac{1}{\sqrt{2}} (\hat{a}_x - \hat{a}_y) \cdot \hat{a}_r \right] = r - \frac{s}{\sqrt{2}} (\sin \theta \cos \phi - \sin \theta \sin \phi)$$

For Amplitude Terms

$$r_1 \approx r_2 \approx r_3 \approx r_4 \approx r_5 \approx r_6 \approx r_7 \approx r_8 \approx r$$

Making these substitutions and combining terms (first with fifth, second with sixth, third with seventh, fourth with eighth), we can write

$$\frac{E}{E_0} = 2 \left[\cos(\beta s \sin \theta \cos \phi) + \cos(\beta s \sin \theta \sin \phi) - 2 \cos\left(\frac{\beta s}{\sqrt{2}} \sin \theta \cos \phi\right) \cos\left(\frac{\beta s}{\sqrt{2}} \sin \theta \sin \phi\right) \right]$$

where $E_0 = f(\theta, \phi) \frac{e^{-j\beta r}}{r}$

$\alpha = 30^\circ$

The procedure for this reflector follows those of the others. Using the geometry of Figure d

$$E = \left(\frac{e^{-j\beta r_1}}{r_1} + \frac{e^{-j\beta r_2}}{r_2} + \frac{e^{-j\beta r_3}}{r_3} + \frac{e^{-j\beta r_4}}{r_4} + \frac{e^{-j\beta r_5}}{r_5} + \frac{e^{-j\beta r_6}}{r_6} + \frac{e^{-j\beta r_7}}{r_7} + \frac{e^{-j\beta r_8}}{r_8} + \frac{e^{-j\beta r_9}}{r_9} + \frac{e^{-j\beta r_{10}}}{r_{10}} + \frac{e^{-j\beta r_{11}}}{r_{11}} + \frac{e^{-j\beta r_{12}}}{r_{12}} \right) f(\theta, \phi)$$

For far-field observations

For Phase Terms

$$r_1 \approx r - s \cos \psi_1 = r - s (\hat{a}_x \cdot \hat{a}_r) = r - s \sin \theta \cos \phi$$

$$r_2 \approx r - s \cos \psi_2 = r - s \left[\frac{1}{\sqrt{3}} (\sqrt{2} \hat{a}_x + \hat{a}_y) \cdot \hat{a}_r \right] = r - \frac{s}{\sqrt{3}} (\sqrt{2} \sin \theta \cos \phi + \sin \theta \sin \phi)$$

$$r_3 \approx r - s \cos \psi_3 = r - s \left[\frac{1}{\sqrt{3}} (\hat{a}_x + \sqrt{2} \hat{a}_y) \cdot \hat{a}_r \right] = r - \frac{s}{\sqrt{3}} (\sin \theta \cos \phi + \sqrt{2} \sin \theta \sin \phi)$$

Cont'd.

7.13 Cont'd.

$$\begin{aligned}
 r_4 &\approx r - s \cos \psi_4 = r - s (\hat{a}_y \cdot \hat{a}_r) = r - s \sin \theta \sin \phi \\
 r_5 &\approx r - s \cos \psi_5 = r - s \left[\frac{1}{2} (-\hat{a}_x + \sqrt{3} \hat{a}_y) \cdot \hat{a}_r \right] = r - \frac{s}{2} (-\sin \theta \cos \phi + \sqrt{3} \sin \theta \sin \phi) \\
 r_6 &\approx r - s \cos \psi_6 = r - s \left[\frac{1}{2} (-\sqrt{3} \hat{a}_x + \hat{a}_y) \cdot \hat{a}_r \right] = r - \frac{s}{2} (-\sqrt{3} \sin \theta \cos \phi + \sin \theta \sin \phi) \\
 r_7 &\approx r - s \cos \psi_7 = r - s (-\hat{a}_x \cdot \hat{a}_r) = r + s \sin \theta \cos \phi \\
 r_8 &\approx r - s \cos \psi_8 = r - s \left[\frac{1}{2} (-\sqrt{3} \hat{a}_x - \hat{a}_y) \cdot \hat{a}_r \right] = r + \frac{s}{2} (\sqrt{3} \sin \theta \cos \phi + \sin \theta \sin \phi) \\
 r_9 &\approx r - s \cos \psi_9 = r - s \left[\frac{1}{2} (\hat{a}_x - \sqrt{3} \hat{a}_y) \cdot \hat{a}_r \right] = r + \frac{s}{2} (\sin \theta \cos \phi + \sqrt{3} \sin \theta \sin \phi) \\
 r_{10} &\approx r - s \cos \psi_{10} = r - s (-\hat{a}_y \cdot \hat{a}_r) = r + s \sin \theta \sin \phi \\
 r_{11} &\approx r - s \cos \psi_{11} = r - s \left[\frac{1}{2} (\hat{a}_x - \sqrt{3} \hat{a}_y) \cdot \hat{a}_r \right] = r + \frac{s}{2} (-\sin \theta \cos \phi + \sqrt{3} \sin \theta \sin \phi) \\
 r_{12} &\approx r - s \cos \psi_{12} = r - s \left[\frac{1}{2} (\sqrt{3} \hat{a}_x - \hat{a}_y) \cdot \hat{a}_r \right] = r + \frac{s}{2} (-\sqrt{3} \sin \theta \cos \phi + \sin \theta \sin \phi)
 \end{aligned}$$

For Amplitude Terms

$$r_1 \approx r_2 \approx r_3 \approx r_4 \approx r_5 \approx r_6 \approx r_7 \approx r_8 \approx r_9 \approx r_{10} \approx r_{11} \approx r_{12} \approx r$$

Making these substitutions and combining terms (first with seventh, second with eighth, third with ninth, fourth with tenth, fifth with eleventh, sixth with twelfth), we can write

$$\begin{aligned}
 \frac{E}{E_0} &= 2 \left[\cos(\beta s \sin \theta \cos \phi) - 2 \cos\left(\frac{\sqrt{3}}{2} \beta s \sin \theta \cos \phi\right) \cos\left(\frac{\beta s}{2} \sin \theta \sin \phi\right) \right. \\
 &\quad \left. - \cos(\beta s \sin \theta \sin \phi) + 2 \cos\left(\frac{\beta s}{2} \sin \theta \cos \phi\right) \cos\left(\frac{\sqrt{3}}{2} \beta s \sin \theta \sin \phi\right) \right]
 \end{aligned}$$

$$\text{where } E_0 = f(\theta, \phi) \frac{e^{-j\beta r}}{r}$$

7.14 The period of each is

$$\alpha = 60^\circ : \Delta s = 2.02$$

$$\alpha = 45^\circ : \Delta s = 16.692$$

$$\alpha = 30^\circ : \Delta s = 30.02$$

while the peak values are

$$\alpha = 60^\circ : |E/E_0|_{\max} = 5.2$$

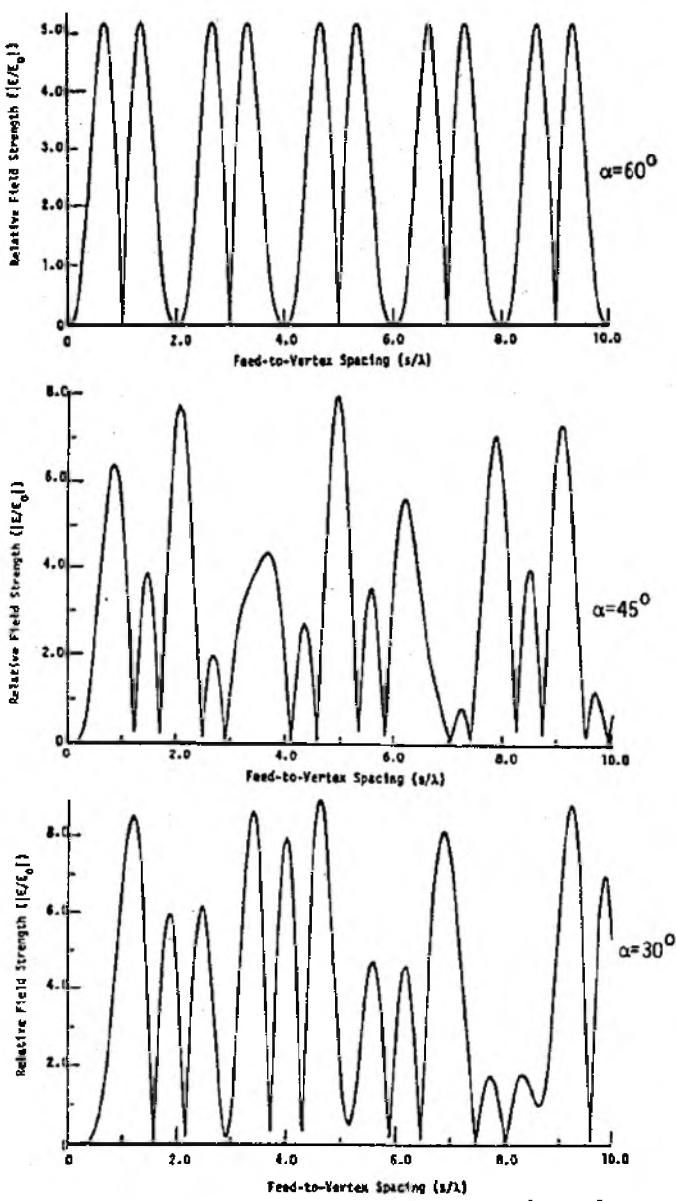
$$\alpha = 45^\circ : |E/E_0|_{\max} = 8.0$$

$$\alpha = 30^\circ : |E/E_0|_{\max} = 9.0$$

The plots of $|E/E_0|$ as a function of s in the range $0 \leq s \leq 102$ are shown on the next page

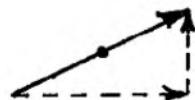
Cont'd.

7.14 cont'd.

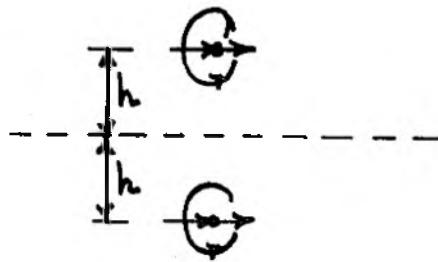


Relative field strengths along the axis ($\theta=90^\circ, \phi=0^\circ$) for $\alpha=60^\circ, 45^\circ, 30^\circ$ corner reflectors as a function of feed-to-vertex spacing.

7.15



7.16 Since a small loop can be modeled as a linear magnetic dipole, as shown in the figure, then using the sources and images of Figure 7-2(a) we choose the image to be another magnetic horizontal linear dipole as shown in the figure. The corresponding small loop to represent the image is also shown.



7.17

- The equivalent of a small circular loop is a linear magnetic dipole perpendicular to the plane of the loop (\perp to the x-y plane) and pointed in the $+z$ direction.

The image of the magnetic dipole above a PMC is a linear magnetic dipole in the same direction ($+z$) as the actual source.

Therefore the image is also a small electric loop with the same magnitude and phase as the actual source.

- In the same direction as the actual source; CCW.

7.18 To maintain the maximum along the z -axis, the field must add (in phase) with that of the actual dipole. Therefore the smallest phase (excluding zero) of the additional round trip of the reflected field must be 2π rads or 360° . Thus

$$2\beta h = 2\pi \Rightarrow h = \frac{2\pi}{2\beta} = \frac{2\pi}{2(2\pi/\lambda)} = \frac{\lambda}{2}$$

h = $\lambda/2$

- $\lambda = 30 \times 10^9 / 10 \times 10^9 = 3 \text{ cm} \Rightarrow [h = 1.5 \text{ cm}]$

- 7.19 • To maintain the maximum along the z -axis, the field must add (in phase) with that of the actual dipole. Since the image is in the opposite direction of the actual source (dipole), the smallest phase of the round trip of the reflected field must be π rads or 180° . Thus

$$2\beta h = \pi \Rightarrow h = \pi/(2\beta) = \pi/[2(2\pi/3)] = \boxed{2/4}$$

$$\bullet \lambda = 30 \times 10^9 / (10 \times 10^9) = 3 \text{ cm} \Rightarrow \boxed{h = 0.75 \text{ cm}}$$

- 7.20 • Since the source is an electric dipole, the image is in the opposite direction. Therefore smallest phase of the round trip of the reflected field must be π rads or 180° . Thus

$$2\beta h = \pi \Rightarrow h = \pi/(2\beta) = \pi/[2(2\pi/3)] = \boxed{2/4}$$

$$\bullet \lambda = 30 \times 10^9 / (10 \times 10^9) = 3 \text{ cm} \Rightarrow \boxed{h = 0.75 \text{ cm}}$$

- 7.21 • Since the source is an electric dipole and the ground plane is a PMC, the image is in the same direction as the actual element. Therefore the smallest phase of the round trip of the reflected field must (excluding zero) 2π rads (360°). Thus

$$2\beta h = 2\pi \Rightarrow h = 2\pi/(2\beta) = 2\pi/[2(2\pi/3)] = \boxed{2/2}$$

$$\bullet \lambda = 30 \times 10^9 / (10 \times 10^9) = 3 \text{ cm} \Rightarrow \boxed{h = 1.5 \text{ cm}}$$

$$\begin{aligned} \underline{\underline{E}}^L &= \hat{x} \underline{\underline{E}}_0 e^{-j\beta z} & \underline{\underline{E}}^R &= \hat{x} \Gamma \underline{\underline{E}}_0 e^{+j\beta z} & \underline{\underline{E}}^T &= \hat{x} T \underline{\underline{E}}_0 e^{j\beta z} \\ \underline{\underline{H}}^L &= \hat{y} \frac{\underline{\underline{\Gamma}}}{\eta} \underline{\underline{E}}_0 e^{-j\beta z} & \underline{\underline{H}}^R &= -\hat{y} \frac{\underline{\underline{\Gamma}}}{\eta} \underline{\underline{E}}_0 e^{+j\beta z} & \underline{\underline{H}}^T &= \hat{y} \frac{\underline{\underline{T}}}{\eta} \underline{\underline{E}}_0 e^{-j\beta z} \end{aligned}$$

$$\underline{\underline{J}}_{eq} = j\omega(\epsilon - \epsilon_0) \underline{\underline{E}}^+ = \hat{x} j\omega(\epsilon - \epsilon_0) T \underline{\underline{E}}_0 e^{-j\beta z}$$

$$\underline{\underline{M}}_{eq} = j\omega(\mu - \mu_0) \underline{\underline{H}}^+ = \hat{y} j\omega(\mu - \mu_0) \frac{T \underline{\underline{E}}_0}{\eta} e^{-j\beta z}$$

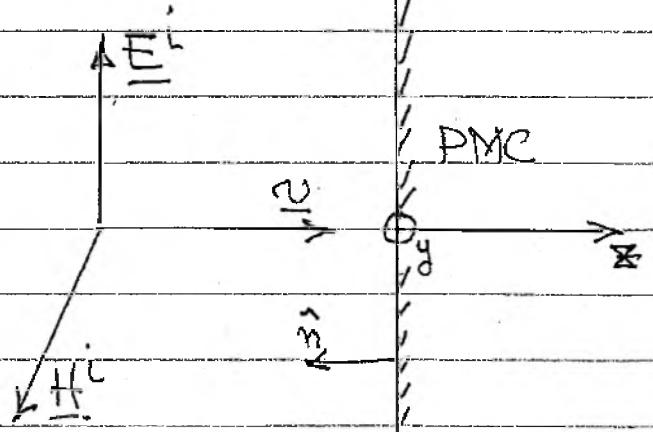
$$\Gamma = \frac{\eta - \eta_0}{\eta + \eta_0} \quad T = \frac{2\eta}{\eta + \eta_0}$$

7.23

$$\underline{E}_1 = \hat{a}_x E_0 e^{-jB_0 z}$$

a. $\underline{H}_1^L = \hat{a}_y \frac{\underline{E}_1}{\eta_0} e^{-jB_0 z} = \hat{a}_y \frac{E_0}{\eta_0} e^{-jB_0 z}$

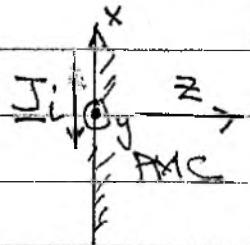
$$\underline{H}_1^L = \hat{a}_y 2.6525 \times 10^{-3} E_0 e^{-jB_0 z}$$



b. Using Figure 7-12(a), but with PMC instead of PEC, the current densities that you need are, according to Figure 7-12(b) with PMC instead of PEC, are

$$\underline{J}_i = \hat{n} \times \underline{H}_1 \Big|_{z=0} = (\hat{a}_z) \times \left(\hat{a}_y \frac{E_0}{\eta_0} e^{-jB_0 z} \right) \Big|_{z=0} = \hat{a}_z \times \hat{a}_y \frac{E_0}{\eta_0} = -\hat{a}_x \frac{E_0}{\eta_0}$$

$$\underline{M}_i = \hat{n} \times \underline{E}_1 \Big|_{z=0} = -\hat{a}_x \times \hat{a}_x \frac{E_0}{\eta_0} e^{-jB_0 z} \Big|_{z=0} = -\hat{a}_y E_0$$

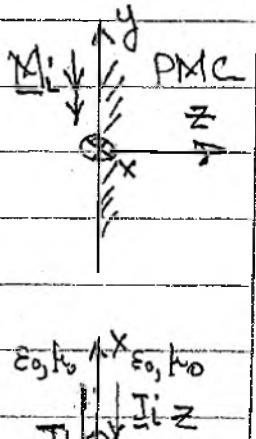


However since both of them are next (parallel) to a PMC, then M_i is shorted out by the PMC.

The J_i , which is also parallel to the PMC (flat end infinite in extent), using image theory of Figure 7-2(b), the image is the same as the actual source (magnitude and phase). Thus

$$\underline{J}_i \Big|_{z=0} = -2 \hat{a}_x \frac{E_0}{\eta_0} = -\hat{a}_x 2 (2.6525 \times 10^{-3} E_0)$$

$$\underline{J}_i \Big|_{z=0} = -\hat{a}_x 5.305 \times 10^{-3} E_0$$

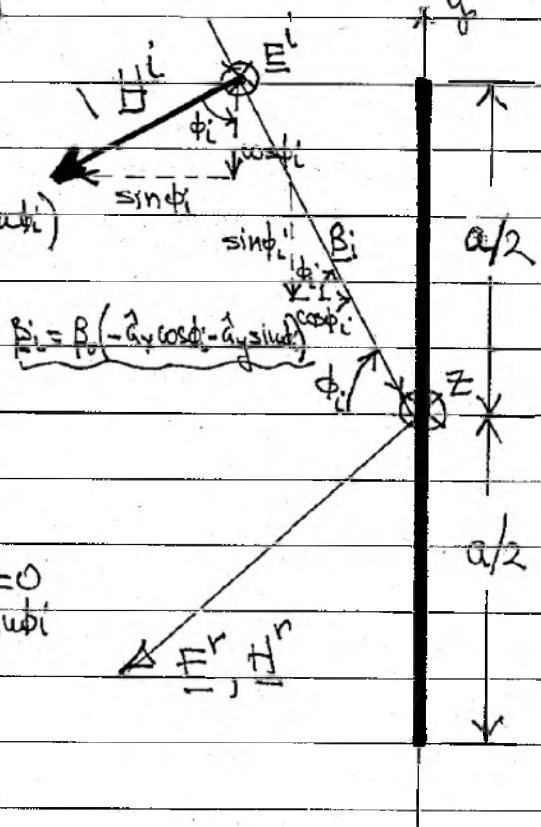


$$[7.24] \underline{E}^l = \hat{a}_z E_0 e^{-j\beta L z} \\ = \hat{a}_z E_0 e^{-j\beta (-x \cos \phi_i - y \sin \phi_i)} (\hat{a}_x x + \hat{a}_y y + \hat{a}_z z) \\ \underline{E}^l = \hat{a}_z E_0 e^{+j\beta (x \cos \phi_i + y \sin \phi_i)}$$

$$\underline{H}^l = H_0 (\hat{a}_x \sin \phi_i - \hat{a}_y \cos \phi_i) e^{+j\beta (x \cos \phi_i + y \sin \phi_i)}$$

$$H_0 = E_0 / n$$

(a) Induction Equivalent



$$M_x^l = 2 \hat{n} \times \underline{E}^l \Big|_{x=0} = 2 \hat{a}_x \times (\hat{a}_z E^l) \Big|_{x=0} = +2 \hat{a}_y E^l$$

$$M_y^l = +\hat{a}_y 2 E_0 e^{j\beta y \sin \phi_i} \Rightarrow M_y^l = +2 E_0 e^{j\beta y \sin \phi_i}$$

$$M_x = M_z = T_x = T_z = 0$$

$$M_y = +2 E_0 e^{j\beta y \sin \phi_i}$$

According to (6-125a)-(6-128c)

$$N_\theta = N_\phi = 0$$

$$L_\phi = \int_S \left[M_x \sin \phi_s \cos \phi_s + M_y \cos \phi_s \sin \phi_s - M_z \sin \phi_s \right] e^{j\beta r' \cos \phi} ds'$$

$$L_\phi = \iint_S \left[-M_x \sin \phi_s + M_y \cos \phi_s \right] e^{j\beta r' \cos \phi} ds'$$

According to (6-125a)-(6-128d)

$$N_f = N_d = 0$$

$$\begin{aligned}
 L_d &= \iint_{\Omega} \left(M_x \cos \theta_s \cos \phi_s + M_y \cos \theta_s \sin \phi_s - M_z \sin \theta_s \right) e^{j p' \cos \psi} ds \\
 &= \int_{-h/2}^{h/2} \int_{-b/2}^{b/2} M_y \cos \theta_s \sin \phi_s e^{j p' (\sin \theta_s \sin \phi_s + z' \cos \phi_s)} dy' dz' \\
 &\quad \times \int_{-b/2}^{b/2} \int_{-c/2}^{c/2} e^{j \beta z' \cos \theta_s} e^{j \beta (\sin \theta_s \sin \phi_s + \sin \phi_i') y'} dy' dz' \\
 L_d &= +2E_0 \cos \theta_s \sin \phi_s
 \end{aligned}$$

According to

$$\int_{-c/2}^{c/2} e^{j \alpha z} dz = c \left[\frac{\sin(\frac{\alpha}{2}c)}{\frac{\alpha}{2}} \right]$$

Thus

$$L_d = +2E_0 \cos \theta_s \sin \phi_s \left[b \frac{\sin\left(\frac{\beta b}{2} \cos \theta_s\right)}{\frac{\beta b}{2} \cos \theta_s} \right] \left[a \frac{\sin\left[\frac{\beta a}{2} (\sin \theta_s \sin \phi_s + \sin \phi_i')\right]}{\frac{\beta a}{2} (\sin \theta_s \sin \phi_s + \sin \phi_i')} \right]$$

$$L_d = +2E_0 \cos \theta_s \sin \phi_s \left[\frac{\sin(Y)}{Y} \right] \left[\frac{\sin(Z)}{Z} \right]$$

$$Y = \frac{\beta R}{2} (\sin \theta_s \sin \phi_s + \sin \phi_i')$$

$$Z = \frac{\beta b}{2} \cos \theta_s$$

$$L_\phi = \int \int \int (-M_x \sin\phi_s + M_y \cos\phi_s) e^{j\beta r \cos\psi} ds$$

$$L_\phi = +2E_0 ab \cos\phi_s \left[\frac{\sin Y}{Y} \right] \left[\frac{\sin Z}{Z} \right]$$

$$E_y \approx -j \frac{\beta e^{-j\beta Y}}{4\pi r} (L_\phi + \eta N_\phi) = -j \frac{\beta e^{-j\beta Y}}{4\pi r} L_\phi$$

$$E_\phi \approx +j \frac{\beta e^{-j\beta Y}}{4\pi r} (1 + \eta N_\phi) = +j \frac{\beta e^{-j\beta Y}}{2\pi r} L_\phi$$

$$E_\phi \approx -j \frac{\beta e^{-j\beta Y}}{4\pi r} [+2E_0 ab \cos\phi_s] \left[\frac{\sin(Y)}{Y} \right] \left[\frac{\sin(Z)}{Z} \right]$$

$$E_\phi \approx -j \frac{ab E_0 \beta e^{-j\beta Y}}{2\pi r} \left[\cos\phi_s \frac{\sin Y}{Y} \frac{\sin Z}{Z} \right]$$

$$E_y \approx +j \frac{\beta e^{-j\beta Y}}{4\pi r} L_\phi = +j \frac{\beta e^{-j\beta Y}}{4\pi r} [+2E_0 ab \cos\phi_s \sin\phi_s] \left[\frac{\sin Y}{Y} \right] \left[\frac{\sin Z}{Z} \right]$$

$$E_\phi \approx +j \frac{ab E_0 \beta e^{-j\beta Y}}{2\pi r} \left[\cos\phi_s \sin\phi_s \frac{\sin Y}{Y} \frac{\sin Z}{Z} \right]$$

$$Y = \frac{\beta a}{2} \left[\sin\phi_s \sin\phi_s + \sin\phi_i \right]$$

$$Z = \frac{\beta b}{2} \cos\phi_s$$

For backscattering: $\theta_s = \pi/2, \phi_s = \phi_i$:

$$E_y \approx j \frac{ab E_0 \beta e^{-j\beta Y}}{2\pi r} \left[\cos\phi_i \frac{\sin(\beta a \sin\phi_i)}{\beta a \sin\phi_i} \right]$$

$$E_\phi \approx j \frac{ab E_0 \beta e^{-j\beta Y}}{2\pi r} [0] = 0$$

(b) Physical Equivalent:

$$\underline{J}_P = 2 \hat{\eta} \times \underline{H}^l |_{x=0}$$

$$\underline{H}^l = \frac{E_0}{\eta} \left(\hat{a}_x \sin \phi_i - \hat{a}_y \cos \phi_i \right) e^{j \beta (x \cos \phi_i + y \sin \phi_i)}$$

$$\underline{J}_P = 2 \hat{a}_x \times \left(\frac{E_0}{\eta} \left(\hat{a}_x \sin \phi_i - \hat{a}_y \cos \phi_i \right) e^{j \beta (x \cos \phi_i + y \sin \phi_i)} \right) |_{x=0}$$

$$\underline{J}_P = -\hat{a}_z 2 \frac{E_0}{\eta} \cos \phi_i e^{j \beta y' \sin \phi_i}$$

$$\underline{J}_z = -2 \frac{E_0}{\eta} \cos \phi_i e^{j \beta y' \sin \phi_i}, \quad J_x = J_y = 0, \quad M_x = M_y = M_z = 0$$

Using $(6-125a) - (6-128c)$

$$L_\theta = L_\phi = 0$$

$$N_\phi = \iiint \left[J_x \cos \theta \cos \phi_s + J_y \cos \theta \sin \phi_s - J_z \sin \theta \right] e^{j \beta r \cos \psi} ds'$$

$$N_\phi = \iiint \left[-J_x \sin \phi_s + J_y \cos \phi_s \right] e^{j \beta r \cos \psi} ds' = 0$$

$$N_\theta = \int_{-b/2}^{b/2} \int_{-a/2}^{a/2} -J_z \sin \theta_s e^{j \beta r \cos \psi} ds' = +2 \frac{E_0 \cos \phi_i \sin \theta_s}{\eta} \int_{-b/2}^{b/2} e^{j \beta r \cos \psi} dz \int_{-a/2}^{a/2} e^{j \beta z \sin \phi_i} dy$$

$$N_\theta = +2 \frac{E_0 \cos \phi_i \sin \theta_s}{\eta} \left[b \frac{\sin \left(\frac{\beta b}{2} \cos \theta_s \right)}{\frac{\beta b}{2} \cos \theta_s} \right] \left[a \frac{\sin \left(\frac{\beta a}{2} (\sin \theta_s \sin \phi_i + \sin \phi_i) \right)}{\frac{\beta a}{2} (\sin \theta_s \sin \phi_i + \sin \phi_i)} \right]$$

$$N_\theta = +2 \frac{E_0 ab}{\eta} \cos \phi_i \sin \theta_s \left[\frac{\sin Y}{Y} \right] \left[\frac{\sin Z}{Z} \right]$$

$$E_\phi \approx -j \frac{\beta e^{-j\beta r}}{4\pi r} \left[\frac{1}{b+\eta N_0} \right] = -j \frac{\beta e^{-j\beta r}}{4\pi r} N_0$$

$$\approx -j \frac{\beta e^{-j\beta r}}{4\pi r} \left[+2 \frac{E_{0ab} \cos\phi_i \sin\theta_i}{\eta} \left[\frac{\sin Y}{Y} \right] \left[\frac{\sin Z}{Z} \right] \right]$$

$$E_\phi \approx -j \frac{ab E_{0\beta}}{2\pi r} \left[\cos\phi_i \sin\theta_i \left[\frac{\sin Y}{Y} \right] \left[\frac{\sin Z}{Z} \right] \right]$$

$$E_\phi \approx +j \frac{\beta b}{4\pi r} \left[\frac{1}{b-\eta N_0} \right] = 0$$

$$Y = \frac{\beta a}{2} \left[\sin\theta_i \sin\phi_i + \sin\theta_i \right]$$

$$Z = \frac{\beta b}{2} \cos\theta_i$$

For backscattering: $\theta_i = \pi/2, \phi_i = \phi$

$$E_\phi \approx -j \frac{ab E_{0\beta}}{2\pi r} \left[\cos\phi_i \frac{\sin(\beta a \sin\phi_i)}{\beta a \sin\phi_i} \right]$$

(which, for backscattering is the same as for the Induction)
 Equivalent. The solutions are not the same in other directions.

7.25

$$\underline{E}_a = \hat{a}_z E_0, \underline{M}_s = -2\hat{n} \times \underline{E}_a = -2\hat{a}_x \times \hat{a}_z E_0 = \hat{a}_y 2E_0 \quad \text{Thus}$$

$$M_y = 2E_0, M_x = M_z = J_x = J_y = J_z = 0$$

$$N_\theta = N_\phi = 0$$

$$L_\theta = \iint_{S_a} [\mathcal{M}_x^0 \cos \theta \cos \phi + M_y \cos \theta \sin \phi - \mathcal{M}_z^0 \sin \theta] e^{j\beta(y' \sin \theta \sin \phi + z' \cos \theta)} dy' dz'$$

$$= 2E_0 \cos \theta \cdot \sin \phi \int_{-a/2}^{a/2} e^{j\beta y' \sin \theta \sin \phi} dy' \int_{-b/2}^{b/2} e^{j\beta z' \cos \theta} dz'$$

$$L_\theta = 2E_0 ab \left[\cos \theta \sin \phi \frac{\sin Y \sin Z}{Y \cdot Z} \right], Y = \frac{\beta a}{2} \sin \theta \cdot \sin \phi, Z = \frac{\beta b}{2} \cos \theta$$

$$L_\phi = \iint_{S_a} [-\mathcal{M}_x^0 \sin \phi + M_y \cos \phi] e^{j\beta(y' \sin \theta \cdot \sin \phi + z' \cos \theta)} dy' dz'$$

$$= 2E_0 \cdot a \cdot b \left[\cos \phi \cdot \frac{\sin Y}{Y} \cdot \frac{\sin Z}{Z} \right]$$

$$E_r \approx 0$$

$$E_\theta \approx -j \frac{\beta e^{-j\beta r}}{4\pi r} [L_\phi + \eta \mathcal{M}_\phi^0] = -j \frac{ab\beta E_0 e^{-j\beta r}}{2\pi r} \left[\cos \phi \cdot \frac{\sin Y \sin Z}{Y \cdot Z} \right]$$

$$E_\phi \approx j \frac{\beta e^{-j\beta r}}{4\pi r} [L_\theta - \eta \mathcal{M}_\theta^0] = +j \frac{ab\beta E_0 e^{-j\beta r}}{2\pi r} \left[+\cos \theta \cdot \sin \phi \frac{\sin Y}{Y} \cdot \frac{\sin Z}{Z} \right]$$

$$H_r \approx 0, \quad H_\theta \approx -\frac{E_\phi}{\eta}, \quad H_\phi \approx +\frac{E_\theta}{\eta}$$

7.26

Assuming the magnetic field at the aperture is related to the electric field by the wave impedance \underline{Z}_a .

$\underline{H}_a = -\hat{a}_y \frac{E_0}{\underline{Z}_a}$, and using the equivalent model of 7.8(c) and an equivalent procedure as that outlined in Figure 7.10 but using a PMC ground plane (instead of a PEC), the equivalent current densities to represent the aperture and PMC ground plane are

$$\underline{J}_s = 2\hat{n} \times \underline{H}_a = 2\hat{a}_x \times \left(-\hat{a}_y \frac{E_0}{\underline{Z}_a} \right) = -\hat{a}_z \frac{2E_0}{\underline{Z}_a} \Rightarrow J_z = -\frac{2E_0}{\underline{Z}_a}, \quad J_x = J_y = 0$$

$$M_s = 0$$

$$\underline{N}_\phi = \iint_S -J_z \sin \theta e^{j\beta r' \cos \phi} ds' = -\frac{2E_0 ab}{\underline{Z}_a} \left[\sin \theta \frac{\sin Y}{Y} \frac{\sin Z}{Z} \right] \quad Y = \frac{\beta a}{2} \sin \theta \sin \phi \\ Z = \frac{\beta b}{2} \cos \theta$$

$$E_\phi = j \frac{ab\beta e^{-j\beta r}}{2\pi r} \left(\frac{\eta}{\underline{Z}_a} \right) \left[\sin \theta \frac{\sin Y}{Y} \frac{\sin Z}{Z} \right], \quad E_\phi = 0$$

$$H_\theta = -\frac{E_\phi}{\eta} = 0, \quad H_\phi = \frac{E_\theta}{\eta}$$

7.27

$$\underline{E}_a = \hat{a}_z E_0, \underline{M}_s = -\hat{n} \times \underline{E}_a = \hat{a}_y E_0 \Rightarrow M_x = M_z = 0, M_y = E_0$$

$$\underline{H}_a = -\hat{a}_y \frac{E_0}{Z_a}, J_s = \hat{n} \times \underline{H}_a = \hat{a}_x \times \left(-\hat{a}_y \frac{E_0}{Z_a} \right) = -\hat{a}_z \frac{E_0}{Z_a}$$

$$\Rightarrow J_x = J_y = 0, J_z = -E_0 / Z_a$$

From Problem 7.25, without the factor of 2

$$L_\theta = E_0 \cdot a \cdot b \left[\cos \theta \sin \phi \cdot \frac{\sin Y}{Y} \cdot \frac{\sin Z}{Z} \right], Y = \frac{\beta a}{2} \sin \theta \sin \phi, Z = \frac{\beta b}{2} \cos \theta$$

$$L_\phi = E_0 \cdot ab \left[\cos \phi \cdot \frac{\sin Y}{Y} \cdot \frac{\sin Z}{Z} \right]$$

$$N_\theta = \iint_{S_a} [J_x^0 \cos \theta \cos \phi + J_y^0 \cos \theta \sin \phi - J_z \sin \theta] e^{j\beta(y' \sin \theta \sin \phi + z' \cos \theta)} dy' dz'$$

$$= + \frac{E_0 ab}{Z_a} \left[\sin \theta \cdot \frac{\sin Y}{Y} \cdot \frac{\sin Z}{Z} \right]$$

$$N_\phi = \iint_{S_a} [-J_x^0 \sin \phi + J_y^0 \cos \phi] e^{j\beta(y' \sin \theta \sin \phi + z' \cos \theta)} dy' dz' = 0$$

$$E_r \simeq 0$$

$$E_\theta \simeq -j \frac{\beta e^{-j\beta r}}{4\pi r} [L_\phi + \eta N_\theta] = -j \frac{\beta ab E_0 e^{-j\beta r}}{4\pi r} \left[(\cos \phi + \sin \theta) \frac{\sin Y}{Y} \cdot \frac{\sin Z}{Z} \right]$$

$$E_\phi \simeq j \frac{\beta e^{-j\beta r}}{4\pi r} [L_\theta - \eta N_\phi^0] = -j \frac{\beta ab E_0 e^{-j\beta r}}{4\pi r} \left[-\cos \theta \sin \phi \cdot \frac{\sin Y}{Y} \cdot \frac{\sin Z}{Z} \right]$$

$$H_r \simeq 0$$

$$H_\theta \simeq -E_\phi / \eta$$

$$H_\phi \simeq E_\theta / \eta$$

7.28

The only difference between this problem and 7.25 is that for the y variations the integral reduces to

$$\int_{-a/2}^{a/2} \cos \left(\frac{\pi}{a} y' \right) e^{j\beta y' \sin \theta \sin \phi} dy' = - \left(\frac{\pi a}{2} \right) \frac{\cos \left(\frac{\beta a}{2} \sin \theta \sin \phi \right)}{\left(\frac{\beta a}{2} \sin \theta \sin \phi \right)^2 - \left(\frac{\pi}{2} \right)^2}$$

Thus

$$E_r \simeq 0$$

$$E_\theta \simeq -j \frac{\beta e^{-j\beta r}}{4\pi r} L_\phi = j \frac{ab \beta E_0 e^{-j\beta r}}{4r} \left[\cos \phi \cdot \frac{\cos Y}{(Y)^2 - (\pi/2)^2} \cdot \frac{\sin Z}{Z} \right], Y = \frac{\beta a}{2} \sin \theta \sin \phi$$

$$E_\phi \simeq j \frac{\beta e^{-j\beta r}}{4\pi r} \cdot L_\theta = +j \frac{ab \beta E_0 \cdot e^{-j\beta r}}{4r} \left[-\cos \theta \cdot \sin \phi \frac{\cos Y}{(Y)^2 - (\pi/2)^2} \cdot \frac{\sin Z}{Z} \right],$$

$$Z = \frac{\beta b}{2} \cos \theta$$

$$H_r \simeq 0, \quad H_\theta \simeq -E_\phi / \eta, \quad H_\phi \simeq E_\theta / \eta$$

7.29 Assuming the magnetic field at the aperture is related to the electric field by the wave impedance Z_a

$$\underline{H}_a = -\hat{a}_y \frac{E_0}{Z_a} \cos\left(\frac{\pi}{a} y'\right)$$

and using the equivalent model of 7.8(c) and an equivalent procedure as that outlined in Figure 7.10 but using a PMC ground plane (instead of a PEC), the equivalent current densities to represent the aperture and PMC ground plane are

$$\underline{J}_S = 2\hat{n} \times \underline{H}_a = 2\hat{e}_x \times \left[-\hat{a}_y \frac{E_0}{Z_a} \cos\left(\frac{\pi}{a} y'\right) \right] = -\hat{a}_z \frac{2E_0}{Z_a} \cos\left(\frac{\pi}{a} y'\right)$$

$$\underline{J}_Z = -\frac{2E_0}{Z_a} \cos\left(\frac{\pi}{a} y'\right), \quad \underline{J}_x = \underline{J}_y = \underline{M}_S = 0$$

Using the integration with respect to y' from the solution of Problem 7.28

$$\int_{-\alpha/2}^{\alpha/2} \cos\left(\frac{\pi}{a} y'\right) e^{j\beta y' \sin\theta \sin\phi} dy' = -\left(\frac{\pi a}{2}\right) \frac{\cos\left(\frac{\beta a}{2} \sin\theta \sin\phi\right)}{\left(\frac{\beta a}{2} \sin\theta \sin\phi\right)^2 - \left(\frac{\pi}{2}\right)^2}$$

$$N_\phi = \iint_S J_Z \sin\theta e^{j\beta r \cos\phi} ds' = \frac{2E_0 abr}{2Z_a} \left[\sin\theta \frac{\cos(Y)}{(Y)^2 - (\pi/2)^2} \frac{\sin Z}{Z} \right]$$

$$N_\phi = L_\theta = L_\phi = 0, \quad Y = \frac{\beta a}{2} \sin\theta \sin\phi, \quad Z = \frac{\beta b}{2} \cos\theta$$

$$E_\phi = -j \frac{ab\beta E_0 e^{-j\beta r}}{2r} \left(\frac{\eta}{Z_a} \right) \left[\sin\theta \frac{\cos(Y)}{(Y)^2 - (\pi/2)^2} \frac{\sin(Z)}{Z} \right], \quad E_\phi = 0, H_\phi = E_0/\eta$$

7.30 Since the aperture is not mounted on a ground plane, then both electric (J_S) and magnetic (M_S) current densities must be used, as in Problem 7.27. Using the solution of Problem 7.27 and integration of the solutions of Problems 7.28 and 7.29, we can write (with $Z_a = \eta$)

$$E_\theta \approx j \frac{ab\beta E_0 e^{-j\beta r}}{8r} \left[(\cos\phi + \sin\theta) \frac{\cos Y}{(Y)^2 - (\pi/2)^2} \frac{\sin Z}{Z} \right]$$

$$E_\phi \approx -j \frac{ab\beta E_0 e^{-j\beta r}}{8r} \left[+ \cos\theta \cdot \sin\phi \cdot \frac{\cos Y}{(Y)^2 - (\pi/2)^2} \right]$$

$$Y = \frac{\beta a}{2} \sin\theta \sin\phi, \quad Z = \frac{\beta b}{2} \cos\theta, \quad H_r \approx 0, \quad H_\theta \approx -E_\phi/\eta, \quad H_\phi \approx E_\theta/\eta$$

F.31

$$\underline{E}_a = \hat{a}_x E_0, M_s = -2\hat{n} \times \underline{E}_a = -2\hat{a}_y \times \hat{a}_x E_0 = \hat{a}_z 2E_0$$

$$\text{Thus } M_z = 2E_0, M_x = M_y = J_x = J_y = J_z = 0$$

$$N_\theta = N_\phi = 0$$

$$L_\theta = \iint_{S_a} [M_x^0 \cos \theta \cdot \cos \phi + M_y^0 \cos \theta \sin \phi - M_z \cdot \sin \theta] e^{j\beta(x' \sin \theta \cos \phi + z' \cos \theta)} dx' dz'$$

$$= -2E_0 \sin \theta \int_{-a/2}^{a/2} e^{j\beta z' \cos \theta} dz' \int_{-b/2}^{b/2} e^{j\beta x' \sin \theta \cos \phi} dx'$$

$$L_\theta = -2E_0 ab \left[\sin \theta \cdot \frac{\sin x}{x} \cdot \frac{\sin Z}{Z} \right], X = \frac{Bb}{2} \sin \theta \cos \phi, Z = \frac{Ba}{2} \cos \theta$$

$$L_\phi = \iint_{S_a} [-M_x^0 \sin \phi + M_y^0 \cos \phi] e^{j\beta(x' \sin \theta \cos \phi + z' \cos \theta)} dx' dz' = 0$$

$$E_r \simeq 0, \quad E_\theta \simeq -\frac{jBe^{-j\beta r}}{4\pi r} [L_\phi^0 + \eta M_\theta^0] = 0$$

$$E_\phi \simeq j \frac{Be^{-j\beta r}}{4\pi r} [L_\theta - \eta M_\phi^0] = -j \frac{abBE_0 e^{-j\beta r}}{2\pi r} \left[\sin \theta \cdot \frac{\sin X}{X} \cdot \frac{\sin Z}{Z} \right]$$

$$H_r \simeq 0, \quad H_\theta \simeq -\frac{E_\phi}{\eta}, \quad H_\phi \simeq \frac{E_\theta}{\eta} = 0$$

F.32

Assuming the magnetic field at the aperture is related to the electric field by the wave impedance $\frac{Z_0}{2}$,

$\underline{H}_a = -\hat{a}_z \frac{\underline{E}_0}{Z_0}$, and using the equivalent model of F.8(c) and an equivalent procedure as that outlined in Figure F.10 but using a PMC ground plane (instead of a PEC), the equivalent current densities to represent the aperture and PMC ground plane are

$$\underline{J}_s = 2\hat{n} \times \underline{H}_a = 2\hat{a}_y \times \left(-\hat{a}_z \frac{\underline{E}_0}{Z_0} \right) = -\hat{a}_x \frac{2E_0}{Z_0}, \quad J_y = J_z = 0$$

$$M_s = 0$$

$$N_\theta = \iint_S J_x \cos \theta \sin \phi e^{j\beta r' \cos \phi} ds' = -\frac{2E_0 \cos \theta \sin \phi}{Z_0} \frac{\sin X}{X} \frac{\sin Z}{Z}$$

$$N_\phi = \iint_S -J_x \sin \phi e^{j\beta r' \cos \phi} ds' = +\frac{2E_0 \sin \phi}{Z_0} \frac{\sin X}{X} \frac{\sin Z}{Z}$$

$$E_\phi = -j \frac{Be^{-j\beta r}}{2\pi r} \left(\frac{\eta}{Z_0} \right) \left[\cos \theta \cos \phi \frac{\sin X}{X} \frac{\sin Z}{Z} \right], \quad X = \frac{Bb}{2} \sin \theta \cos \phi \\ Z = \frac{Ba}{2} \cos \theta$$

$$H_\phi = -E_\phi / \eta, \quad H_\theta = E_\theta / \eta$$

7.33

$$E_a = \hat{a}_x E_0, \quad M_s = -\hat{n} \times \underline{E}_a = \hat{a}_z E_0 \Rightarrow M_x = M_y = 0, \quad M_z = E_0$$

$$\underline{H}_a = -\hat{a}_z \frac{E_0}{Z}, \quad J_s = \hat{n} \times \underline{H}_a = \hat{a}_y \times \left(-\hat{a}_z \frac{E_0}{Z} \right) = -\hat{a}_x \frac{E_0}{Z} \Rightarrow J_y = J_z = 0,$$

$$J_x = -\frac{E_0}{Z}$$

From Problem 7.31, without the factor of 2

$$L_\theta = -E_0 ab \left[\sin \theta \cdot \frac{\sin X}{X} \cdot \frac{\sin Z}{Z} \right]$$

$$L_\phi = 0$$

$$X = \frac{\beta b}{2} \sin \theta \cos \phi, \quad Z = \frac{\beta a}{2} \cos \theta$$

$$N_\theta = \iint_{Sa} [J_x \cos \theta \cos \phi + J_y^0 \cos \theta \sin \phi - J_z^0 \sin \theta] e^{j\beta(x' \sin \theta \cos \phi + z' \cos \theta)} dx' dz'$$

$$= -\frac{E_0}{Z} ab \left[\cos \theta \cos \phi \cdot \frac{\sin X}{X} \cdot \frac{\sin Z}{Z} \right]$$

$$N_\phi = \iint_{Sa} [-J_x \sin \phi + J_y \cos \phi] e^{j\beta(x' \sin \theta \cos \phi + z' \cos \theta)} dx' dz'$$

$$= \frac{E_0}{Z} ab \left[\sin \phi \cdot \frac{\sin X}{X} \cdot \frac{\sin Z}{Z} \right]$$

$$E_r \simeq 0$$

$$E_\theta \simeq -j \frac{\beta e^{-j\beta r}}{4\pi r} [L_\phi^0 + \eta N_\theta] = j \frac{\beta ab E_0 e^{-j\beta r}}{4\pi r} \left[\cos \theta \cos \phi \frac{\sin X}{X} \frac{\sin Z}{Z} \right] \left(\frac{\eta}{Z} \right)$$

$$E_\phi \simeq j \frac{\beta e^{-j\beta r}}{4\pi r} [L_\theta - \eta N_\phi] = j \frac{\beta ab E_0 e^{-j\beta r}}{4\pi r} \left[-(\sin \theta + \frac{1}{Z}) \frac{\sin X}{X} \frac{\sin Z}{Z} \right]$$

$$H_r \simeq 0, \quad H_\theta \simeq -E_\phi / \eta, \quad H_\phi \simeq E_\theta / \eta$$

7.34

The only difference between this problem and 7.31 is that for the z variations the integral reduces to

$$E_r \simeq 0 \quad \int_{-a/2}^{a/2} \cos \left(\frac{\pi}{a} z' \right) e^{j\beta z' \cos \theta} dz' = - \left(\frac{\pi a}{2} \right) \frac{\cos \left(\frac{\beta a}{2} \cos \theta \right)}{\left(\frac{\beta a}{2} \cos \theta \right)^2 - \left(\frac{\pi}{2} \right)^2}$$

$$E_\theta \simeq 0$$

$$E_\phi \simeq j \frac{\beta e^{-j\beta r}}{4\pi r} L_\theta = j \frac{\beta ab E_0 e^{-j\beta r}}{4\pi r} \left[\sin \theta \frac{\sin X}{X} \frac{\cos Z}{(Z)^2 - \left(\frac{\pi}{2} \right)^2} \right]$$

$$X = \frac{\beta b}{2} \sin \theta \cos \phi, \quad Z = \frac{\beta a}{2} \cos \theta$$

$$H_r \simeq 0, \quad H_\theta \simeq -E_\phi / \eta, \quad H_\phi \simeq E_\theta / \eta = 0$$

7.35 Assuming the magnetic field at the aperture is related to the electric field by the wave impedance Z_a

$$\underline{H}_a = -\hat{a}_x \frac{E_0}{Z_a} \cos\left(\frac{\pi}{a} z'\right)$$

and using the equivalent model of 7.8(c) and an equivalent procedure as that outlined in Figure 7.10 but using a PMC ground plane (instead of a PEC), the equivalent current densities to represent the aperture and the ground plane are

$$\underline{J}_S = 2\hat{n} \times \underline{H}_a = 2\hat{a}_y \times \left[-\hat{a}_x \frac{E_0}{Z_a} \cos\left(\frac{\pi}{a} z'\right) \right] = \hat{a}_z \frac{2E_0}{Z_a} \cos\left(\frac{\pi}{a} z'\right)$$

$$J_z = \frac{2E_0}{Z_a} \cos\left(\frac{\pi}{a} z'\right), J_x = J_y = M_S = 0$$

Using the integration with respect to z' from the solution of

Problem 7.34 $\int_{-a/2}^{+a/2} \cos\left(\frac{\pi}{a} z'\right) e^{jBz' \cos\theta} dz' = -\left(\frac{\pi a}{2}\right) \frac{\cos\left(\frac{B a}{2} \cos\theta\right)}{\left(\frac{B a}{2} \cos\theta\right)^2 - \left(\frac{\pi}{2}\right)^2}$

$$N_\phi = \iint_{-a/2}^{+a/2} -J_z \sin\theta e^{jBz' \cos\theta} ds' = \frac{2ab\pi E_0}{2Z_a} \left[\sin\theta \frac{\sin X}{X} \frac{\cos Z}{(Z)^2 - (\pi/2)^2} \right]$$

$$N_\phi = L_\phi = L_\theta = 0, \quad X = \left(\frac{Ba}{2} \sin\theta \cos\phi\right), \quad Z = \frac{Ba}{2} \cos\theta$$

$$E_\phi = -j \frac{BabE_0 e^{-jBr}}{4r} \left[\sin\theta \frac{\sin X}{X} \frac{\cos(Z)}{(Z)^2 - (\pi/2)^2} \right]$$

$$E_\theta = 0, \quad H_\theta = -E_\phi/\eta = 0, \quad H_\phi = E_\theta/\eta$$

7.36 Since the aperture is not mounted on a ground plane, then both the electric (\underline{J}_S) and magnetic (\underline{M}_S) current densities must be used, as in Problem 7.33. Using the solution of Problem 7.33 and integration of the solutions of Problems 7.33 and 7.34, we can write (with $Z_a = \eta$)

$$E_\theta \approx j \frac{kabE_0 e^{-jkr}}{8r} \left[\cos\theta \cos\phi \cdot \frac{\sin X}{X} \cdot \frac{\cos Z}{(Z)^2 - (\pi/2)^2} \right]$$

$$E_\phi \approx j \frac{kabE_0 e^{-jkr}}{8r} \left[-(\sin\theta + \sin\phi) \cdot \frac{\sin X}{X} \cdot \frac{\cos Z}{(Z)^2 - (\pi/2)^2} \right]$$

$$X = \frac{kb}{2} \sin\theta \cdot \cos\phi, \quad Z = \frac{ka}{2} \cos\theta$$

$$H_\theta \approx -E_\phi/\eta$$

$$H_\phi \approx E_\theta/\eta$$

7.37 $\underline{E}_a = \hat{a}_y E_0$

$$(a) M_s = \hat{n} \times \underline{E}_a = 2\hat{a}_z \times (\hat{a}_y E_0) = +\hat{a}_x 2E_0 \Rightarrow M_x = +2E_0$$

$$\underline{J}_s = 0$$

$$(b) N_\theta = N_\phi = 0$$

$$L_\theta = \iint_S M_x \cos\theta \cos\phi e^{j\beta r' \cos\phi} ds' = +2E_0 \cos\theta \cos\phi \int_{-a/2}^{+a/2} e^{j\beta x' \sin\theta \cos\phi} dx' \int_{-b/2}^{+b/2} e^{j\beta y' \sin\theta \cos\phi} dy'$$

$$L_\theta = +2E_0 ab \left[\cos\theta \cos\phi \frac{\sin(x)}{x} \frac{\sin(y)}{y} \right], X = \frac{\beta x}{2} \sin\theta \cos\phi, Y = \frac{\beta y}{2} \sin\theta \cos\phi$$

$$L_\phi = \iint_S -M_x \sin\phi = -2E_0 ab \left[\sin\phi \frac{\sin x}{X} \frac{\sin y}{Y} \right]$$

$$\underline{E}_\theta = +j \frac{\beta ab E_0 e^{-j\beta r}}{2\pi r} \left[\sin\phi \frac{\sin x}{X} \frac{\sin y}{Y} \right]$$

$$\underline{E}_\phi = +j \frac{\beta ab E_0 e^{-j\beta r}}{2\pi r} \left[\cos\theta \cos\phi \frac{\sin x}{X} \frac{\sin y}{Y} \right]$$

7.38 Assuming the magnetic field at the aperture is related to the electric field by the wave impedance Z_a , $\underline{H}_a = \hat{a}_x \frac{\underline{E}_0}{Z_a}$, and using the equivalent model of 7.8(c) and an equivalent procedure as that outlined in Figure 7.10 but using a PMC ground plane (instead of a PEC), the equivalent current densities to represent the aperture and PMC ground plane are

$$\underline{J}_s = 2\hat{n} \times \underline{H}_a = 2\hat{a}_z \times \left(-\hat{a}_x \frac{E_0}{Z_a} \right) = -\hat{a}_y \frac{2E_0}{Z_a} \Rightarrow J_y = -\frac{2E_0}{Z_a}, J_x = J_z = 0$$

$$M_s = 0$$

$$N_\theta = \iint_S J_y \cos\theta \sin\phi e^{j\beta r \cos\phi} ds' = -\frac{2E_0 \cos\theta \sin\phi}{Z_a} \int_{-a/2}^{+a/2} e^{j\beta x' \sin\theta \cos\phi} dx' \int_{-b/2}^{+b/2} e^{j\beta y' \sin\theta \sin\phi} dy'$$

$$= -\frac{2abE_0}{Z_a} \left[\cos\theta \sin\phi \frac{\sin x}{X} \frac{\sin y}{Y} \right], X = \frac{\beta x}{2} \sin\theta \cos\phi, Y = \frac{\beta y}{2} \sin\theta \sin\phi$$

$$N_\phi = \iint_S J_y \cos\phi e^{j\beta r \cos\phi} ds' = \frac{(ab)2E_0}{Z_a} \left[\cos\phi \frac{\sin x}{X} \frac{\sin y}{Y} \right], L_\theta = L_\phi = 0$$

$$\underline{E}_\theta = +j \frac{\beta ab E_0 e^{-j\beta r}}{2\pi r} \frac{m}{Z_a} \left[\cos\phi \frac{\sin x}{X} \frac{\sin y}{Y} \right]$$

$$\underline{E}_\phi = j \frac{\beta ab E_0 e^{-j\beta r}}{2\pi r} \frac{m}{Z_a} \left[\cos\phi \frac{\sin x}{X} \frac{\sin y}{Y} \right]$$

7.39 When the aperture is not mounted on a ground plane, we need both \underline{J}_s and \underline{M}_s ($\underline{J}_s = \hat{n} \times \underline{H}_a$, $\underline{M}_s = -\hat{n} \times \underline{E}_a$).

From the solutions of 7.37 and 7.38 (with $z_a = \eta$)

$$\underline{M}_s = \hat{a}_X \underline{E}_o, \underline{J}_s = -\hat{a}_Y \frac{\underline{E}_o}{\eta}$$

$$L_\theta = E_o ab \left[\cos \theta \sin \phi \frac{\sin X}{X} \frac{\sin Y}{Y} \right], L_\phi = -E_o ab \left[\sin \phi \frac{\sin X}{X} \frac{\sin Y}{Y} \right]$$

$$N_\theta = -\frac{E_o ab}{\eta} \left[\cos \theta \sin \phi \frac{\sin X}{X} \frac{\sin Y}{Y} \right], N_\phi = -\frac{E_o ab}{\eta} \left[\cos \phi \left(\frac{\sin X}{X} \right) \frac{\sin Y}{Y} \right]$$

$$E_\theta = j \frac{\beta ab E_o e^{-j\beta r}}{4\pi r} \left[\sin \phi (1 + \cos \theta) \frac{\sin X}{X} \frac{\sin Y}{Y} \right], X = \frac{\beta a}{2} \sin \theta \cos \phi,$$

$$Y = \frac{\beta b}{2} \sin \theta \sin \phi$$

$$E_\phi = j \frac{\beta ab E_o e^{-j\beta r}}{4\pi r} \left[\cos \phi (1 + \cos \theta) \frac{\sin X}{X} \frac{\sin Y}{Y} \right]$$

7.40

The only difference between this problem and 7.37 is that for the x variations the integral reduces to

$$\int_{-\alpha_2}^{\alpha_2} \cos(\frac{\pi}{a} x') dx' e^{j\beta x' \sin \theta \cos \phi} dx' = -\left(\frac{\pi a}{2}\right) \frac{\sin\left(\frac{\beta a}{2} \sin \theta \cos \phi\right)}{\left(\frac{\beta a}{2} \sin \theta \cos \phi\right)^2 - \left(\frac{\pi}{2}\right)^2}$$

Thus, from the solution of 7.37 $[X = \frac{\beta a}{2} \sin \theta \cos \phi, Y = \frac{\beta b}{2} \sin \theta \sin \phi]$

$$L_\theta = -E_o \pi ab \left[\cos \theta \cos \phi \frac{\cos(X)}{(X)^2 - (\pi/2)^2} \frac{\sin Y}{Y} \right], L_\phi = E_o \pi ab \left[\sin \phi \frac{\cos X}{(X)^2 - (\pi/2)^2} \frac{\sin Y}{Y} \right]$$

$$E_\theta = -j \frac{\beta ab E_o e^{-j\beta r}}{4r} \left[\sin \phi \frac{\cos X}{(X)^2 - (\pi/2)^2} \frac{\sin Y}{Y} \right]$$

$$E_\phi = -j \frac{\beta ab E_o e^{-j\beta r}}{4r} \left[\cos \theta \cos \phi \frac{\cos(X)}{(X)^2 - (\pi/2)^2} \frac{\sin Y}{Y} \right]$$

$$H_\theta = -E_\phi / \eta, H_\phi = E_\theta / \eta$$

7.41

Assuming the magnetic field at the aperture is related to the electric field by the wave impedance Z_a

$$\underline{H}_a = -\hat{a}_x \frac{E_0 \cos(\frac{\pi}{a}x')}{Z_a}$$

and using the equivalent model of 7.8(c) and an equivalent procedure as that outlined in Figure 7.10 but using a PMC ground plane (instead of a PEC) the equivalent current densities to represent the aperture and PMC ground plane are

$$\underline{J}_S = -\hat{a}_y \frac{2E_0 \cos(\frac{\pi}{a}x')}{Z_a} \Rightarrow J_y = -\frac{2E_0}{Z_a} \cos(\frac{\pi}{a}x'), J_x = J_z = 0$$

$$M_S = 0$$

Using the integral of $\int_{-a/2}^{a/2} \cos(\frac{\pi}{a}x) dx' = -\left(\frac{\pi a}{2}\right) \frac{\cos(X)}{(X)^2 - (\frac{\pi}{2})^2}$, $X = \frac{\beta a}{2} \sin \theta \cos \phi$
 $Y = \frac{\beta b}{2} \sin \theta \sin \phi$

$$N_\theta = -\frac{2E_0 ab}{Z_a} \left(-\frac{\pi}{2} \right) \left[\cos \theta \sin \phi \frac{\cos X}{(X)^2 - (\frac{\pi}{2})^2} \frac{\sin Y}{Y} \right], N_\phi = -\frac{2E_0 ab}{Z_a} \left(\frac{\pi}{2} \right) \left[\cos \phi \cos X \frac{\cos Y}{(X)^2 - (\frac{\pi}{2})^2} \frac{\sin Y}{Y} \right]$$

$$E_\theta = -j \frac{\beta ab E_0 (\eta)}{4\pi} \left[\cos \theta \sin \phi \frac{\cos X}{(X)^2 - (\frac{\pi}{2})^2} \frac{\sin Y}{Y} \right], E_\phi = -j \frac{\beta ab E_0 (\eta)}{4\pi} \left[\cos \phi \cos X \frac{\cos Y}{(X)^2 - (\frac{\pi}{2})^2} \frac{\sin Y}{Y} \right]$$

$$H_\theta = -E_\phi / \eta, H_\phi = E_\theta / \eta$$

7.42

$$\underline{E}_a = \hat{a}_y E_0 \cos\left(\frac{\pi}{a}x'\right) \Rightarrow \underline{M}_s = -\hat{n} \times \underline{E}_a = \hat{a}_x E_0 \cos\left(\frac{\pi}{a}x'\right)$$

$$\text{Thus } M_x = E_0 \cos\left(\frac{\pi}{a}x'\right), M_y = M_z = 0$$

$$\begin{aligned} L_\theta &= \int_{-b/2}^{b/2} \int_{-a/2}^{a/2} M_x \cos \theta \cos \phi e^{j\frac{\beta}{a}(x' \sin \theta \cos \phi + y' \sin \theta \sin \phi)} dx' dy' \\ &= E_0 \cos \theta \cos \phi \int_{-a/2}^{a/2} \cos\left(\frac{\pi}{a}x'\right) e^{j\frac{\beta}{a}x' \sin \theta \cos \phi} dx' \int_{-b/2}^{b/2} e^{j\frac{\beta}{a}y' \sin \theta \sin \phi} dy' \end{aligned}$$

Since

$$\int_{-a/2}^{a/2} \cos\left(\frac{\pi}{a}x'\right) e^{j\frac{\beta}{a}x' \sin \theta \cos \phi} dx' = -\left(\frac{\pi a}{2}\right) \frac{\cos(X)}{(X)^2 - \left(\frac{\pi}{2}\right)^2}$$

$$\int_{-b/2}^{b/2} e^{j\frac{\beta}{a}y' \sin \theta \sin \phi} dy' = b \frac{\sin Y}{Y}, \quad X = \frac{\beta a}{2} \sin \theta \cos \phi, Y = \frac{\beta b}{2} \sin \theta \sin \phi$$

then

$$L_\theta = -\frac{\pi ab}{2} E_0 \left[\cos \theta \cdot \cos \phi \frac{\cos X}{(X)^2 - \left(\frac{\pi}{2}\right)^2} \cdot \frac{\sin Y}{Y} \right]$$

Similarly

$$\begin{aligned}
 L_\phi &= \int_{-b/2}^{b/2} \int_{-a/2}^{a/2} -M_x \sin \phi e^{j\beta(x' \sin \theta \cos \phi + y' \sin \theta \sin \phi)} dx' dy' \\
 &= -E_0 \sin \phi \int_{-a/2}^{a/2} \cos\left(\frac{\pi}{a}x'\right) e^{j\beta x' \sin \theta \cos \phi} dx' \int_{-b/2}^{b/2} e^{j\beta y' \sin \theta \sin \phi} dy' \\
 L_\phi &= +\frac{\pi ab}{2} E_0 \left[\sin \phi \frac{\cos X}{(X)^2 - (\frac{\pi}{2})^2} \cdot \frac{\sin Y}{Y} \right] \\
 \underline{H}_a &\simeq -\hat{a}_x \frac{E_a}{\eta} \Rightarrow \underline{J}_s = \hat{n} \times \underline{H}_a = \hat{a}_z \times \left(-\hat{a}_x \frac{E_a}{\eta} \right) = -\hat{a}_y \frac{E_a}{\eta} = -\hat{a}_y \frac{E_0}{\eta} \cos\left(\frac{\pi}{a}x'\right)
 \end{aligned}$$

$$J_x = J_z = 0, J_y = -\frac{E_0}{\eta} \cos\left(\frac{\pi}{a}x'\right)$$

$$\begin{aligned}
 N_\theta &= \int_{-b/2}^{b/2} \int_{-a/2}^{a/2} J_y \cos \theta \sin \phi e^{j\beta(x' \sin \theta \cos \phi + y' \sin \theta \sin \phi)} dx' dy' \\
 &= -\frac{E_0}{\eta} \cos \theta \sin \phi \int_{-a/2}^{a/2} \cos\left(\frac{\pi}{a}x'\right) e^{j\beta x' \sin \theta \cos \phi} dx' \int_{-b/2}^{b/2} e^{j\beta y' \sin \theta \sin \phi} dy' \\
 &= +\frac{\pi ab E_0}{2\eta} \left[\cos \theta \sin \phi \cdot \frac{\cos X}{(X)^2 - (\pi/2)^2} \cdot \frac{\sin Y}{Y} \right] \\
 N_\phi &= \int_{-b/2}^{b/2} \int_{-a/2}^{a/2} J_y \cos \phi e^{j\beta(x' \cos \phi \sin \theta + y' \sin \theta \sin \phi)} dx' dy' \\
 &= +\frac{\pi ab E_0}{2\eta} \left[\cos \phi \cdot \frac{\cos X}{(X)^2 - (\frac{\pi}{2})^2} \cdot \frac{\sin Y}{Y} \right]
 \end{aligned}$$

$$E_r \simeq 0$$

$$\begin{aligned}
 E_\theta &\simeq -j \frac{\beta e^{-j\beta r}}{4\pi r} [L_\phi + \eta N_\theta] = -j \frac{ab\beta E_0 e^{-j\beta r}}{8r} \left[\sin \phi (1 + \cos \theta) \frac{\cos X}{(X)^2 - (\frac{\pi}{2})^2} \cdot \frac{\sin Y}{Y} \right] \\
 E_\phi &\simeq j \frac{\beta e^{-j\beta r}}{4\pi r} [L_\theta - \eta N_\phi] = -j \frac{ab\beta E_0 e^{-j\beta r}}{8r} \left[\cos \phi (1 + \cos \theta) \frac{\cos X}{(X)^2 - (\frac{\pi}{2})^2} \cdot \frac{\sin Y}{Y} \right]
 \end{aligned}$$

$$H_r \simeq 0, H_\theta = -E_\phi/\eta, H_\phi = E_\theta/\eta$$

7.43 Since $\underline{E}_a = \hat{a}_p E_p + \hat{a}_\phi E_\phi$ exists only over the circular aperture where $p \leq a$, and it is zero elsewhere, then over the aperture and outside it along a flat plane coincident with the circular aperture we can form the electric and magnetic equivalent current, as shown below

$$\underline{M}_a = \hat{n} \times \underline{E}_a \quad \underline{M}_s = 0, \underline{J}_s \neq 0 \quad \underline{J}_s, \underline{M}_a \uparrow^n \quad \underline{M}_s = 0, \underline{J}_s \neq 0 \quad \epsilon_0, \mu_0$$

$\xleftarrow[a + a]{}$

Now if we move a perfect electric conductor to coincide with the interface we have the following equivalent.

$$\underline{M}_s = \underline{J}_s = 0 \quad \underline{J}_s = 0, \underline{M}_a \uparrow^n \quad \underline{M}_s = \underline{J}_s = 0 \quad \epsilon_0, \mu_0$$

$\sigma = \infty$

which is equivalent to

$$\underline{M}_s = 2\hat{n} \times \underline{E}_a \quad \underline{J}_s = \underline{M}_s = 0 \quad \underline{J}_s = 0, \underline{M}_s = 2\underline{M}_a \quad \underline{J}_s = \underline{M}_s = 0 \quad \frac{\epsilon_0}{\epsilon_0, \mu_0}$$

$$\underline{M}_s = 2\hat{a}_z \times (\hat{a}_p E_p + \hat{a}_\phi E_\phi) = 2\hat{a}_p E_p - 2\hat{a}_p E_\phi \quad \xleftarrow[a + a]{}$$

$$7.44 \quad \underline{E}^i = \hat{a}_x E_0 e^{-j\beta_0(y \sin \theta_i - z \cos \theta_i)}$$

$$\underline{H}^i = \frac{E_0}{\eta} (-\hat{a}_y \cos \theta_i - \hat{a}_z \sin \theta_i) e^{-j\beta_0(y \sin \theta_i - z \cos \theta_i)}$$

a. Induction Equivalent

$$\underline{M}_i = 2\hat{n} \times \underline{E}^i \Big|_{z=0} = 2\hat{a}_z \times \hat{a}_x E_x \Big|_{z=0} = \hat{a}_y 2E_0 e^{-j\beta_0 y \sin \theta_i}$$

$$M_x = M_z = 0, \quad M_y = 2E_0 e^{-j\beta_0 y \sin \theta_i}$$

Using (6-125a)-(6-125d)

$$N_\theta = N_\phi = 0$$

$$L_\theta = \iint_{S_a} M_y \cos \theta_i \sin \phi_i e^{j\beta_0 r' \cos \psi} ds' = 2E_0 \cos \theta_i \sin \phi_i \iint_{S_a} e^{-j\beta_0 y' \sin \theta_i} e^{j\beta_0 r' \cos \psi} ds'$$

Using (6-130b), (6-132a) and (6-132b) we can write

$$r' \cos \psi = x' \sin \theta_i \cos \phi_i + y' \sin \theta_i \sin \phi_i$$

$$ds' = dx' dy' = p' dp' d\phi'$$

Thus

$$L_\theta = 2E_0 \cos \theta_i \sin \phi_i \iint_{S_a} e^{-j\beta_0 [x' \sin \theta_i \cos \phi_i + y' (\sin \theta_i \sin \phi_i - \sin \theta_i)]} p' dp' d\phi'$$

$$\text{Since } x' = p' \cos \phi' \\ y' = p' \sin \phi'$$

then the exponent of the exponential can be written as

$$\begin{aligned} x' \sin \theta_i \cos \phi_i + y' (\sin \theta_i \sin \phi_i - \sin \theta_i) &= p' \underbrace{[\sin \theta_i \cos \phi_i \cos \phi' + \sin \phi' (\sin \theta_i \sin \phi_i - \sin \theta_i)]}_{A} \\ &= p' (A \cos \phi' + B \sin \phi') \\ &= p' \sqrt{A^2 + B^2} \left(\frac{A}{\sqrt{A^2 + B^2}} \cos \phi' + \frac{B}{\sqrt{A^2 + B^2}} \sin \phi' \right) \\ &= p' \sqrt{A^2 + B^2} (\cos \phi_0 \cos \phi' + \sin \phi_0 \sin \phi') \end{aligned}$$

$$x' \sin \theta_i \cos \phi_i + y' (\sin \theta_i \sin \phi_i - \sin \theta_i) = \sqrt{A^2 + B^2} p' \cos(\phi' - \phi_0)$$

$$\text{where } A = \sin \theta_i \cos \phi_i$$

$$B = \sin \theta_i \sin \phi_i \sin \theta_i$$

$$\phi_0 = \tan^{-1} \left(\frac{B}{A} \right)$$

cont'd.

7.14 Cont'd. Thus we can write L_θ as

$$L_\theta = 2E_0 \cos\theta_s \sin\phi_s \int_0^a \left\{ \int_0^{2\pi} e^{j\beta_0 \sqrt{A^2+B^2} \rho' \cos(\phi'-\phi_0)} d\phi' \right\} \rho' d\rho'$$

$$= 2E_0 \cos\theta_s \sin\phi_s \int_0^a \left[2\pi J_0(\beta_0 \sqrt{A^2+B^2} \rho') \right] \rho' d\rho' = 4\pi E_0 \cos\theta_s \sin\phi_s \int_0^a J_0(\beta_0 \sqrt{A^2+B^2} \rho') \rho' d\rho'$$

$$L_\theta = 4\pi E_0 \cos\theta_s \sin\phi_s \int_0^a \frac{J_1(\beta_0 \sqrt{A^2+B^2} \rho')}{\beta_0 \sqrt{A^2+B^2}} \rho' d\rho' = 4\pi a^2 E_0 \cos\theta_s \sin\phi_s \frac{J_1(\beta_0 a \sqrt{A^2+B^2})}{\beta_0 a \sqrt{A^2+B^2}}$$

Using (6-125d)

$$L_\phi = \iint_{S_a} M_y \cos\phi_s e^{j\beta_0 r' \cos\phi_s} ds' = 4\pi a^2 E_0 \cos\phi_s \frac{J_1(\beta_0 a \sqrt{A^2+B^2})}{\beta_0 a \sqrt{A^2+B^2}}$$

The electric and magnetic scattered fields can be written using (6-122a) - (6-122f) as

$$E_r^s \approx H_y^s = 0$$

$$E_\theta^s \approx -j \frac{\beta_0 e^{-j\beta_0 r}}{4\pi r} (L_\theta + \eta H_\theta^s) = -j \frac{\beta_0 a^2 E_0 e^{-j\beta_0 r}}{r} \cos\phi_s \frac{J_1(\beta_0 a \sqrt{A^2+B^2})}{\beta_0 a \sqrt{A^2+B^2}}$$

$$E_\phi^s \approx j \frac{\beta_0 e^{-j\beta_0 r}}{4\pi r} (L_\phi - \eta H_\phi^s) = j \frac{\beta_0 a^2 E_0 e^{-j\beta_0 r}}{r} \cos\theta_s \sin\phi_s \frac{J_1(\beta_0 a \sqrt{A^2+B^2})}{\beta_0 a \sqrt{A^2+B^2}}$$

$$H_\theta^s \approx -\frac{E_\phi^s}{\eta}, \quad H_\phi^s = \frac{E_\theta^s}{\eta}$$

For backscatter observations

$$\theta_s = \theta_i, \quad \phi_s = 3\pi/2$$

$$A = \sin\theta_s \cos\phi_s = 0$$

$$B = \sin\theta_s \sin\phi_s - \sin\theta_i = -\sin\theta_i = -2 \sin\theta_i \quad \left\{ \sqrt{A^2+B^2} = 2 \sin\theta_i \right.$$

Thus

$$E_\theta^s \approx 0$$

$$E_\phi^s \approx -j \frac{\beta_0 a^2 E_0 e^{-j\beta_0 r}}{r} \cos\theta_i \frac{J_1(2\beta_0 a \sin\theta_i)}{2\beta_0 a \sin\theta_i}$$

$$H_\theta^s \approx -\frac{E_\phi^s}{\eta}$$

$$H_\phi^s \approx \frac{E_\theta^s}{\eta}$$

cont'd.

7.44 Cont'd.

b. Physical Equivalent

$$\underline{J}_p = 2 \hat{n} \times \underline{H}^i \Big|_{\substack{z=0 \\ y=y'}} = 2 \hat{a}_z \times (\hat{a}_y H_y^i + \hat{a}_z H_z^i) \Big|_{\substack{z=0 \\ y=y'}} = -\hat{a}_x 2 H_y^i \Big|_{\substack{z=0 \\ y=y'}} = \hat{a}_x 2 \frac{\epsilon_0}{\eta} \cos \theta_i e^{-j \beta_0 y' \sin \theta_i}$$

$$J_y = J_z = 0, \quad J_x = \frac{2 \epsilon_0}{\eta} \cos \theta_i e^{-j \beta_0 y' \sin \theta_i}$$

Using (6-125a)-(6-125d) and the results of the induction equivalent, we can write

$$N_\theta = \iint_{S_a} J_x \cos \theta_i \cos \phi_i e^{j \beta_0 r' \cos \psi} ds' = \frac{4 \pi a^2 \epsilon_0}{\eta} \cos \theta_i \cos \phi_i \cos \phi_s \frac{J_1(\beta_0 a \sqrt{A^2 + B^2})}{\beta_0 a \sqrt{A^2 + B^2}}$$

$$N_\phi = \iint_{S_a} -J_x \sin \phi_i e^{j \beta_0 r' \cos \psi} ds' = -\frac{4 \pi a^2 \epsilon_0}{\eta} \cos \theta_i \sin \phi_i \frac{J_1(\beta_0 a \sqrt{A^2 + B^2})}{\beta_0 a \sqrt{A^2 + B^2}}$$

$$A = \sin \theta_i \cos \phi_i, \quad B = \sin \theta_i \sin \phi_i - \sin \theta_i$$

Now using (6-122a)-(6-122f) we can write the scattered electric and magnetic fields as

$$E_r^s \approx H_r^s \approx 0$$

$$E_\theta^s \approx -j \frac{\beta_0 e^{-j \beta_0 r}}{4 \pi r} (k_\theta^s + \eta N_\theta) = -j \frac{\beta_0 a^2 \epsilon_0 e^{-j \beta_0 r}}{r} \cos \theta_i \cos \theta_i \cos \phi_i \frac{J_1(\beta_0 a \sqrt{A^2 + B^2})}{\beta_0 a \sqrt{A^2 + B^2}}$$

$$E_\phi^s \approx j \frac{\beta_0 e^{-j \beta_0 r}}{4 \pi r} (k_\phi^s - \eta N_\phi) = j \frac{\beta_0 a^2 \epsilon_0 e^{-j \beta_0 r}}{r} \cos \theta_i \sin \phi_i \frac{J_1(\beta_0 a \sqrt{A^2 + B^2})}{\beta_0 a \sqrt{A^2 + B^2}}$$

$$H_\theta^s \approx -\frac{E_\theta^s}{\eta}, \quad H_\phi^s \approx \frac{E_\phi^s}{\eta}$$

For backscatter observations

$$\theta_s = \theta_i, \quad \phi_s = 3\pi/2$$

$$\begin{aligned} A &= 0 \\ B &= -2 \sin \theta_i \end{aligned} \quad \left\{ \sqrt{A^2 + B^2} = 2 \sin \theta_i \right.$$

$$E_\theta^s \approx 0$$

$$E_\phi^s \approx -j \frac{\beta_0 a^2 \epsilon_0 e^{-j \beta_0 r}}{r} \cos \theta_i \frac{J_1(2 \beta_0 a \sin \theta_i)}{2 \beta_0 a \sin \theta_i}$$

$$H_\theta^s \approx -\frac{E_\phi^s}{\eta}$$

$$H_\phi^s \approx \frac{E_\phi^s}{\eta}$$

7.45

$$\underline{H}^L = \hat{a}_x H_0 e^{-j\beta_0(y \sin \theta_i - z \cos \theta_i)}$$

$$\underline{E}^L = \eta H_0 (\hat{a}_y \cos \theta_i + \hat{a}_z \sin \theta_i) e^{-j\beta_0(y \sin \theta_i - z \cos \theta_i)}$$

a. Induction Equivalent

$$M_z = 2 \hat{a}_x \times E^L|_{z=0} = 2 \hat{a}_x \times (\hat{a}_y E_y + \hat{a}_z E_z)|_{z=0} = -\hat{a}_x^2 E_y|_{z=0} = -\hat{a}_x^2 2 \eta H_0 \cos \theta_i e^{-j\beta_0 y \sin \theta_i}$$

$$M_y = M_z = 0, \quad M_x = -2 \eta H_0 \cos \theta_i e^{-j\beta_0 y \sin \theta_i}$$

Using (6-125a)-(6-125d) and the solution of Problem 7.14, we can write

$$L_\theta = \int \int_{S_a} M_x \cos \theta_i \cos \phi_s e^{i\beta_0 r \cos \psi} ds' = -4\pi a^2 \eta H_0 \cos \theta_i \cos \phi_s \frac{J_1(\beta_0 a \sqrt{A^2 + B^2})}{\beta_0 a \sqrt{A^2 + B^2}}$$

$$L_\phi = \int \int_{S_a} -M_x \sin \phi_s e^{i\beta_0 r \cos \psi} ds' = +4\pi a^2 \eta H_0 \cos \theta_i \sin \phi_s \frac{J_2(\beta_0 a \sqrt{A^2 + B^2})}{\beta_0 a \sqrt{A^2 + B^2}}$$

$$A = \sin \theta_i \cos \phi_s, \quad B = \sin \theta_i \sin \phi_s - \sin \theta_i$$

Now using (6-122a)-(6-122f) we can write the scattered electric and magnetic fields as

$$E_r^s \approx H_r^s \approx 0$$

$$E_\theta^s \approx -\frac{j\beta_0 e^{-j\beta_0 r}}{4\pi r} (L_\theta + jN_\theta) = -j \frac{\beta_0 a^2 \eta H_0 e^{-j\beta_0 r}}{r} \cos \theta_i \sin \phi_s \frac{J_1(\beta_0 a \sqrt{A^2 + B^2})}{\beta_0 a \sqrt{A^2 + B^2}}$$

$$E_\phi^s \approx j \frac{\beta_0 e^{-j\beta_0 r}}{4\pi r} (L_\phi - jN_\phi) = -j \frac{\beta_0 a^2 \eta H_0 e^{-j\beta_0 r}}{r} \cos \theta_i \cos \phi_s \cos \phi_i \frac{J_2(\beta_0 a \sqrt{A^2 + B^2})}{\beta_0 a \sqrt{A^2 + B^2}}$$

$$H_\theta^s \approx -\frac{E_\theta^s}{\eta}, \quad H_\phi^s \approx \frac{E_\phi^s}{\eta}$$

For backscatter observations

$$\theta_s = \theta_i, \quad \phi_s = 3\pi/2$$

$$\left. \begin{array}{l} A=0 \\ B=-2 \sin \theta_i \end{array} \right\} \sqrt{A^2 + B^2} = 2 \sin \theta_i$$

$$E_\theta^s \approx +j \frac{\beta_0 a^2 \eta H_0 e^{-j\beta_0 r}}{r} \cos \theta_i \frac{J_1(2\beta_0 a \sin \theta_i)}{2\beta_0 a \sin \theta_i}$$

$$E_\phi^s \approx 0$$

$$H_\theta^s \approx -\frac{E_\theta^s}{\eta}$$

$$H_\phi^s \approx \frac{E_\phi^s}{\eta}$$

cont'd.

b. Physical Equivalent

$$\underline{J}_P = 2 \hat{n} \times \underline{H}^L \Big|_{z=0} = 2 \hat{a}_z \times \hat{a}_x H_x = \hat{a}_y 2 H_0 e^{-j \beta_0 y' \sin \theta_i}$$

$y=y'$

$$J_x = J_z = 0, \quad J_y = 2 H_0 e^{-j \beta_0 y' \sin \theta_i}$$

Using (6-125a)-(6-125d) and the solution of Problem 7.44, we can write

$$N_\theta = \iint_{S_a} J_y \cos \theta_s \sin \phi_s e^{j \beta_0 r' \cos \psi} ds' = 4 \pi a^2 H_0 \cos \theta_s \sin \phi_s \frac{J_1(B_0 a \sqrt{A^2 + B^2})}{B_0 a \sqrt{A^2 + B^2}}$$

$$N_\phi = \iint_{S_a} J_y \cos \phi_s e^{j \beta_0 r' \cos \psi} ds' = 4 \pi a^2 H_0 \cos \phi_s \frac{J_1(B_0 a \sqrt{A^2 + B^2})}{B_0 a \sqrt{A^2 + B^2}}$$

$A = \sin \theta_s \cos \phi_s, B = \sin \theta_s \sin \phi_s - \sin \theta_i$

Now using (6-122a)-(6-122f) we can write the scattered electric and magnetic fields as

$$\underline{E}_\theta^s \approx -j \frac{\beta_0 e^{-j \beta_0 r}}{4 \pi r} (\underline{k}_\theta^\circ + \eta N_\theta) = -j \frac{\beta_0 a^2 \eta H_0 e^{-j \beta_0 r}}{r} \cos \theta_s \sin \phi_s \frac{J_1(B_0 a \sqrt{A^2 + B^2})}{B_0 a \sqrt{A^2 + B^2}}$$

$$\underline{E}_\phi^s \approx j \frac{\beta_0 e^{-j \beta_0 r}}{4 \pi r} (\underline{k}_\phi^\circ - \eta N_\phi) = -j \frac{\beta_0 a^2 \eta H_0 e^{-j \beta_0 r}}{r} \cos \phi_s \frac{J_1(B_0 a \sqrt{A^2 + B^2})}{B_0 a \sqrt{A^2 + B^2}}$$

$$\underline{H}_\theta^s \approx -\frac{\underline{E}_\theta^s}{\eta}$$

$$\underline{H}_\phi^s \approx \frac{\underline{E}_\phi^s}{\eta}$$

For backscatter directions

$$\theta_s = \theta_i, \phi_s = 3\pi/2$$

$$\begin{aligned} A &= 0 \\ B &= -2 \sin \theta_i \end{aligned} \quad \left\{ \sqrt{A^2 + B^2} = 2 \sin \theta_i \right.$$

$$\underline{E}_\theta^s \approx j \frac{\beta_0 a^2 \eta H_0 e^{-j \beta_0 r}}{r} \cos \theta_i \frac{J_1(2 \beta_0 a \sin \theta_i)}{2 \beta_0 a \sin \theta_i}$$

$$\underline{E}_\phi^s \approx 0$$

$$\underline{H}_\theta^s \approx -\frac{\underline{E}_\theta^s}{\eta}$$

$$\underline{H}_\phi^s \approx \frac{\underline{E}_\phi^s}{\eta}$$

CHAPTER 8

8.1 $v_g = v \cos \theta, \cos \theta = \frac{\beta_z}{\beta} = \sqrt{1 - (\frac{f_c}{f})^2}$

$$(f_c)_{10} = \frac{1}{2a\sqrt{\mu_0\epsilon_0}} = \frac{3 \times 10^{10}}{2(2.54)(0.9)} = 6.56 \times 10^9, \frac{(f_c)_{10}}{f} = \frac{6.56}{10} = 0.656$$

$$\cos \theta = \sqrt{1 - (0.656)^2} = \sqrt{1 - 0.43} = \sqrt{0.57} = 0.755$$

$$v_g = 3 \times 10^8 (0.755) = 2.265 \times 10^8 \text{ m/sec.}$$

$$d = \text{delay} \times \text{velocity} = (2 \times 10^{-6})(2.265 \times 10^8) = 4.53 \times 10^2 = 453 \text{ meters}$$

8.2 a. $(f_c)_{10} = \frac{1}{2a\sqrt{\mu_0\epsilon_0}} = \frac{1}{2a\sqrt{\mu_0\epsilon_0}\sqrt{\epsilon_r}} = \frac{v_0}{2a\sqrt{\epsilon_r}} = \frac{30 \times 10^9}{2(2.286)\sqrt{2.56}} = 4.101 \text{ GHz}$

b. $\lambda_g = \frac{\lambda_{10}}{\sqrt{1 - (\frac{f_c}{f})^2}} = \frac{\lambda_{10}/\sqrt{\epsilon_r}}{\sqrt{1 - (\frac{4.101}{10})^2}} = \frac{30 \times 10^9 / (10 \times 10^9 \sqrt{2.56})}{\sqrt{1 - (0.4101)^2}} = \frac{1.875}{0.912} = 2.056 \text{ cm}$

c. $Z_w = \frac{\eta}{\sqrt{1 - (\frac{f_c}{f})^2}} = \frac{377/\sqrt{2.56}}{0.912} = 258.36 \text{ Ohms}$

d. $v_p = \frac{v}{\sqrt{1 - (\frac{f_c}{f})^2}} = \frac{v_0/\sqrt{\epsilon_r}}{0.912} = \frac{3 \times 10^8 / \sqrt{2.56}}{0.912} = \frac{1.875 \times 10^8}{0.912} = 2.056 \times 10^8 \text{ m/sec.}$

e. $v_g = v \sqrt{1 - (\frac{f_c}{f})^2} = \frac{v_0}{\sqrt{\epsilon_r}} \sqrt{1 - (\frac{f_c}{f})^2} = \frac{3 \times 10^8}{\sqrt{2.56}} (0.912) = 1.71 \times 10^8 \text{ m/sec.}$

8.3 TE₁₀ mode; $a = 0.622 \text{ in} = 1.57988 \text{ cm}; b = 0.311 \text{ in} = 0.78944 \text{ cm}$

a. $f_c = \frac{1}{2\pi\sqrt{\mu_0\epsilon_0}} \left(\frac{\pi}{a}\right) = \frac{1}{2a\sqrt{\mu_0\epsilon_0}} = \frac{30 \times 10^9}{2(1.57988)} = 9.49439 \times 10^9$

$$\beta_z = \beta \sqrt{1 - \left(\frac{f_c}{f}\right)^2} = \omega \sqrt{\mu_0\epsilon_0} \sqrt{1 - \left(\frac{9.49439}{15}\right)^2} = \frac{2\pi(15 \times 10^9)}{30 \times 10^9} \sqrt{1 - (0.63296)^2}$$

$$\beta_z = \pi(0.77418) = 139.35329 \text{ degrees/cm}$$

$$139.35329(\Delta z) = 300^\circ \Rightarrow \Delta z = 300 / 139.35329 = 2.1528 \text{ cm}$$

| $\Delta z = 2.1528 \text{ cm}$ |

(Cont'd.)

8.3 cont'd

$$b. (f_c)_{10} = \frac{1}{2\pi\sqrt{\epsilon_0\epsilon_r}} \left(\frac{\pi}{a} \right) = \frac{1}{2} (f_c)_{\epsilon_r=1} = \frac{9.49439}{2} \times 10^9 = 4.7472 \times 10^9$$

$$\beta_z = 2\pi f \sqrt{\epsilon_0 \epsilon_r} \sqrt{1 - \left(\frac{f_c}{f}\right)^2} = \frac{2\pi (1.5 \times 10^9)}{30 \times 10^9} \sqrt{4} \sqrt{1 - \left(\frac{4.7472}{1.5}\right)^2} \\ = 2\pi \sqrt{1 - (0.31648)^2} = 2\pi \sqrt{1 - 0.1} = 2\pi (0.9486) = 5.96023 \text{ rad/cm}$$

$$\beta_z = 5.96023 \text{ rad/cm} = 341.496 \text{ degrees/cm}$$

$$\Delta\phi = \beta_z \Delta z = 341.496 (2.1528) = 735.173^\circ = 12.83118 \text{ rads}$$

$$\text{Additional phase: } \Delta\psi = (735.173 - 720) = 15.173^\circ$$

$$\Delta\psi = (12.83118 - 12.5664) = 0.26481 \text{ rads}$$

8.4

$$a = 2.286 \text{ cm}, b = 1.016 \text{ cm}$$

$$(f_c)_{10} = \frac{1}{2\pi\sqrt{\epsilon_0\epsilon_r}} = \frac{30 \times 10^9}{2(2.286)\sqrt{2.25}} = \frac{30 \times 10^9}{2(2.286)(1.5)} = 4.3745 \text{ GHz}$$

$$\cos\theta = \sqrt{1 - (f_c/f)^2} = \sqrt{1 - (4.3745/10)^2} = \sqrt{1 - (0.43745)^2} = 0.8992$$

$$v_g = v_0 \cos\theta = \frac{30 \times 10^9}{\sqrt{2.25}} (0.8992) = 17.9849 \times 10^9 \text{ cm/sec} \\ = 0.179849 \times 10^8 \text{ m/sec}$$

$$(V_g/d) = 1/2 \times 10^{-6} = 0.179849 \left(\frac{1}{d}\right) \times 10^8 = 0.5 \times 10^6$$

$$d = (2 \times 10^{-6}) \times (0.179849 \times 10^8) = 35.9698 \text{ meters}$$

$$8.5 \quad a. \text{ For the empty } (f_c)_{10} = 6.56 \times 10^9 \Rightarrow Z_{w0} = \frac{\eta_0}{\sqrt{1 - (f_c/f)^2}} = \frac{377}{\sqrt{1 - (6.56/10)^2}} = \frac{377}{0.7547} = 499.5$$

For the filled waveguide, from Problem 8.2

$$(f_c)_{10} = 4.101 \times 10^9 \Rightarrow Z_{w2} = 258.36$$

Therefore for the intermediate waveguide ($1/4$ section)

$$Z_{w1} = \sqrt{Z_{w0} Z_{w2}} = \sqrt{499.5 (258.36)} = 359.235 \text{ ohms}$$

b.

$$Z_{w1} = 359.235 = \frac{377/\sqrt{\epsilon_{r1}}}{\sqrt{1 - (\frac{f_c}{f})^2}} = \frac{377/\sqrt{\epsilon_{r1}}}{\sqrt{1 - (\frac{f_{c0}/\epsilon_{r1}}{f})^2}} = \frac{377/\sqrt{\epsilon_{r1}}}{\sqrt{1 - \epsilon_{r1}(\frac{6.56}{10})^2}} \sqrt{1 - 0.43745\epsilon_{r1}}$$

$$\epsilon_{r1} = 1.531$$

c.

$$l = \frac{\lambda_{g1}}{4} = \frac{2\pi/\sqrt{\epsilon_{r1}}}{4\sqrt{1 - (\frac{f_{c1}}{f})^2}} = \frac{30 \times 10^9 / (10 \times 10^9 \sqrt{1.531})}{4\sqrt{1 - (\frac{f_{c0}/\sqrt{\epsilon_{r1}}}{f})^2}} = \frac{3/\sqrt{1.531}}{4\sqrt{1 - (\frac{6.56}{10 \times 1.531})^2}} = 0.7149 \text{ cm}$$

8.6 a. For the empty waveguide, from Problem 8.5 $\Rightarrow Z_{w0} = 499.5 \text{ Ohms}$

For filled waveguide with polystyrene, from Problem 8.2 $\Rightarrow Z_{wL} = 258.36 \text{ Ohms}$

$N=2$

Using (5-76a) and (5-77)

$$\Gamma_n = 2^{-N} \frac{Z_{wL} - Z_{w0}}{Z_{wL} + Z_{w0}} C_n^N = 2^{-N} \frac{Z_{wL} - Z_{w0}}{Z_{wL} + Z_{w0}} \frac{N!}{(N-n)!n!}$$

$$\underline{n=0:} \quad \Gamma_0 = 2^{-2} \frac{258.36 - 499.5}{258.36 + 499.5} \frac{2!}{2!0!} = \frac{1}{4} (-0.318) = -0.0795 = \frac{Z_{w1} - Z_{w0}}{Z_{w1} + Z_{w0}}$$

$$\Rightarrow Z_{w1} = Z_{w0} \left(\frac{1 - 0.0795}{0 + 0.0795} \right) = 499.5 (0.8527) = \underline{425.928}$$

$$\underline{n=1:} \quad \Gamma_1 = 2^{-2} \frac{258.36 - 499.5}{258.36 + 499.5} \frac{2!}{1!1!} = -0.159 = \frac{Z_{w2} - Z_{w1}}{Z_{w2} + Z_{w1}}$$

$$\Rightarrow Z_{w2} = Z_{w1} \frac{1 - 0.159}{1 + 0.159} = 425.928 (0.7255) = \underline{309.03}$$

b. $Z_{w1} = \frac{377/\sqrt{\epsilon_{r1}}}{\sqrt{1 - \left(\frac{6.56}{10\sqrt{\epsilon_{r1}}}\right)^2}} \Rightarrow \epsilon_{r1} = 0.43 + \left(\frac{377}{Z_{w1}}\right)^2 = 1.213$

$$Z_{w2} = \frac{377/\sqrt{\epsilon_{r2}}}{\sqrt{1 - \left(\frac{6.56}{10\sqrt{\epsilon_{r2}}}\right)^2}} \Rightarrow \epsilon_{r2} = 0.43 + \left(\frac{377}{Z_{w2}}\right)^2 = 1.918$$

c. $L_1 = \frac{\lambda_{g1}}{4} = \frac{\lambda_0/\sqrt{\epsilon_{r1}}}{4\sqrt{1 - \left(\frac{6.56}{10\sqrt{\epsilon_{r1}}}\right)^2}} = \frac{3/\sqrt{1.213}}{4\sqrt{1 - 0.43/\epsilon_{r1}}} = 0.8476 \text{ cm}$

$$L_2 = \frac{\lambda_{g2}}{4} = \frac{\lambda_0/\sqrt{\epsilon_{r2}}}{4\sqrt{1 - \left(\frac{6.56}{10\sqrt{\epsilon_{r2}}}\right)^2}} = \frac{3/\sqrt{1.918}}{4\sqrt{1 - 0.43/\epsilon_{r2}}} = 0.6148 \text{ cm}$$

B.7 The expressions for α_c of the TE_{mn}^{\pm} and TM_{mn}^{\pm} modes are derived using (B-64a), or

$$\alpha_c = \frac{P_c/l}{2 P_{mn}} \quad (1)$$

The procedure follows the derivation of (B-69a) for $\alpha_c^{TE,10}$

a. TE_{mn}^{\pm} : $E_x^{\pm} = A_{mn} \frac{n\pi}{b\epsilon} \cos\left(\frac{m\pi}{a}x\right) \sin\left(\frac{n\pi}{b}y\right) e^{-j\beta_z z}$

$$E_y^{\pm} = -A_{mn} \frac{n\pi}{b\epsilon} \sin\left(\frac{m\pi}{a}x\right) \cos\left(\frac{n\pi}{b}y\right) e^{-j\beta_z z}$$

$$H_x^{\pm} = A_{mn} \frac{n\pi \beta_z}{\omega \mu \epsilon} \sin\left(\frac{m\pi}{a}x\right) \cos\left(\frac{n\pi}{b}y\right) e^{-j\beta_z z}$$

$$H_y^{\pm} = A_{mn} \frac{n\pi \beta_z}{b\omega \mu \epsilon} \cos\left(\frac{m\pi}{a}x\right) \sin\left(\frac{n\pi}{b}y\right) e^{-j\beta_z z}$$

$$H_z^{\pm} = -jA_{mn} \frac{\beta_z^2}{\omega \mu \epsilon} \cos\left(\frac{m\pi}{a}x\right) \cos\left(\frac{n\pi}{b}y\right) e^{-j\beta_z z}$$

$$P_{mn} = \int_0^b \int_0^a \frac{1}{2} \operatorname{Re}(E_x H_z^*) \cdot \hat{a}_z dx dy = \frac{1}{2} \operatorname{Re} \int_0^b \int_0^a \hat{a}_z (E_x H_y^* - E_y H_x^*) \cdot \hat{a}_z dx dy$$

$$P_{mn} = \frac{1}{2} \operatorname{Re} \int_0^b \int_0^a (E_x H_y^* - E_y H_x^*) dx dy$$

$$E_x H_y^* = |A_{mn}|^2 \frac{\beta_z}{\omega \mu \epsilon^2} \left(\frac{n\pi}{b}\right)^2 \cos^2\left(\frac{m\pi}{a}x\right) \sin^2\left(\frac{n\pi}{b}y\right)$$

$$E_y H_x^* = -|A_{mn}|^2 \frac{\beta_z}{\omega \mu \epsilon^2} \left(\frac{m\pi}{a}\right)^2 \sin^2\left(\frac{m\pi}{a}x\right) \cos^2\left(\frac{n\pi}{b}y\right)$$

$$P_{mn} = \frac{|A_{mn}|^2}{2} \left(\frac{\beta_z}{\omega \mu \epsilon^2} \right) \left\{ \left(\frac{n\pi}{b} \right)^2 \frac{ab}{4} + \left(\frac{m\pi}{a} \right)^2 \frac{ab}{4} \right\} = \frac{|A_{mn}|^2}{2} \frac{ab}{4} \left(\frac{\beta_z}{\omega \mu \epsilon^2} \right) \left[\left(\frac{n\pi}{b} \right)^2 + \left(\frac{m\pi}{a} \right)^2 \right] \epsilon_{mn}$$

$$P_c = \frac{R_s l}{2} \iint (J_z J_z) ds = 2 \left\{ \frac{R_s l}{2} \int_0^b \int_0^a (|H_x|^2 + |H_z|^2) dx dy + \frac{R_s l}{2} \int_0^b \int_0^a (|H_y|^2 + |H_z|^2) dy dx \right\}$$

$$P_c = R_s l |A_{mn}|^2 \left\{ \left(\frac{n\pi \beta_z}{\omega \mu \epsilon} \right)^2 \int_0^a \sin^2\left(\frac{n\pi}{b}y\right) dy + \left(\frac{\beta_z^2}{\omega \mu \epsilon} \right)^2 \int_0^a \cos^2\left(\frac{n\pi}{b}y\right) dy \right. \\ \left. + \left[\left(\frac{n\pi \beta_z}{\omega \mu \epsilon} \right)^2 \int_0^b \sin^2\left(\frac{n\pi}{b}y\right) dy + \left(\frac{\beta_z^2}{\omega \mu \epsilon} \right)^2 \int_0^b \cos^2\left(\frac{n\pi}{b}y\right) dy \right] \right\}$$

$$P_c = \frac{R_s l |A_{mn}|^2}{(\omega \mu \epsilon)^2} \left\{ (\beta_z)^2 \left[\left(\frac{m\pi}{a} \right)^2 \left(\frac{a}{2} \right) + \left(\frac{n\pi}{b} \right)^2 \frac{b}{2} \right] + (\beta_z^2)^2 \left[\frac{a}{2} \epsilon_m + \frac{b}{2} \epsilon_n \right] \right\}, \epsilon_p = \begin{cases} 2, p=0 \\ 1, p \neq 0 \end{cases}$$

$$\alpha_c = \frac{4R_s}{ab(\omega \mu \epsilon)^2} \left(\frac{\omega \mu \epsilon^2}{\beta_z} \right) \left\{ \frac{(\beta_z)^2 \left[\left(\frac{m\pi}{a} \right)^2 \frac{a}{2} + \left(\frac{n\pi}{b} \right)^2 \frac{b}{2} \right]}{\left(\frac{m\pi}{a} \right)^2 + \left(\frac{n\pi}{b} \right)^2} + \frac{\beta_z^4 \left(\frac{a}{2} \epsilon_m + \frac{b}{2} \epsilon_n \right)}{\left(\frac{m\pi}{a} \right)^2 + \left(\frac{n\pi}{b} \right)^2} \right\} \frac{1}{\epsilon_m \epsilon_n}$$

cont'd.

8.7 cont'd. Since $(\frac{m\pi}{a})^2 + (\frac{n\pi}{b})^2 = \beta_c^2$, then

$$\begin{aligned}\alpha_c^{TE} &= \frac{2Rs}{w\mu\epsilon_m\epsilon_n} \left\{ \frac{\beta_z \left(\frac{m^2}{a^2} + \frac{n^2}{b^2} \right)}{ab \left(\frac{m}{a} \right)^2 + \left(\frac{n}{b} \right)^2} + \frac{\beta_c^2 (a\epsilon_m + b\epsilon_n)}{ab\beta_z} \right\} \\ &= \frac{2Rs}{\eta \beta_c^2 \epsilon_m \epsilon_n} \left\{ \beta \sqrt{1 - \left(\frac{f_c}{f} \right)^2} \frac{(m^2 b + n^2 a)}{(mb)^2 + (na)^2} + \frac{\beta_c^2}{\beta \sqrt{1 - \left(\frac{f_c}{f} \right)^2}} \left(\frac{a\epsilon_m + b\epsilon_n}{ab} \right) \right\} \\ &= \frac{2Rs}{\eta \epsilon_m \epsilon_n} \left\{ \sqrt{1 - \left(\frac{f_c}{f} \right)^2} \frac{(m^2 b + n^2 a)}{(mb)^2 + (na)^2} + \frac{(\beta_c/\beta)^2}{\sqrt{1 - \left(\frac{f_c}{f} \right)^2}} \left(\frac{a\epsilon_m + b\epsilon_n}{ab} \right) \right\}\end{aligned}$$

$$\alpha_c^{TE} = \frac{2Rs}{\eta} \left\{ \sqrt{1 - \left(\frac{f_c}{f} \right)^2} \frac{m^2 b + n^2 a}{(mb)^2 + (na)^2} + \frac{\left(f_c/f \right)^2}{\sqrt{1 - \left(f_c/f \right)^2}} \left(\frac{a\epsilon_m + b\epsilon_n}{ab} \right) \right\}$$

which can also be written as

$$\begin{aligned}\alpha_c^{TE} &= \frac{2Rs}{\epsilon_m \epsilon_n b \eta} \sqrt{1 - \left(\frac{f_c}{f} \right)^2} \left\{ \left(\epsilon_m + \frac{b}{a} \epsilon_n \right) \left(\frac{f_c}{f} \right)^2 + \frac{b}{a} \left[1 - \left(\frac{f_c}{f} \right)^2 \right] \frac{m^2 ab + (na)^2}{(mb)^2 + (na)^2} \right\} \\ \epsilon_{-p} &= \begin{cases} 2, & p=0 \\ 1, & p \neq 0 \end{cases}\end{aligned}$$

b. $TM_{mn}^z (H_z^+ = 0)$: $E_x^+ = B_{mn} \frac{m\pi\beta_z}{aw\mu\epsilon} \cos\left(\frac{m\pi}{a}x\right) \sin\left(\frac{n\pi}{b}y\right) e^{-j\beta_z z}$

$$E_y^+ = -B_{mn} \frac{n\pi\beta_z}{bw\mu\epsilon} \sin\left(\frac{m\pi}{a}x\right) \cos\left(\frac{n\pi}{b}y\right) e^{-j\beta_z z}$$

$$E_z^+ = -j B_{mn} \frac{\beta_c^2}{w\mu\epsilon} \sin\left(\frac{m\pi}{a}x\right) \sin\left(\frac{n\pi}{b}y\right) e^{-j\beta_z z}$$

$$H_x^+ = B_{mn} \frac{n\pi}{b\mu\epsilon} \sin\left(\frac{m\pi}{a}x\right) \cos\left(\frac{n\pi}{b}y\right) e^{-j\beta_z z}$$

$$H_y^+ = -B_{mn} \frac{m\pi}{a\mu\epsilon} \cos\left(\frac{m\pi}{a}x\right) \sin\left(\frac{n\pi}{b}y\right) e^{-j\beta_z z}$$

$$P_{mn} = \int_0^b \int_0^a \frac{1}{2} \operatorname{Re} [E_x H_y^* - E_y H_x^*] \cdot \hat{a}_z dx dy = \frac{1}{2} \operatorname{Re} \int_0^b \int_0^a \hat{a}_z (E_x H_y^* - E_y H_x^*) dx dy$$

$$= \frac{1}{2} \operatorname{Re} \int_0^b \int_0^a (E_x H_y^* - E_y H_x^*) dx dy$$

cont'd.

8.7 cont'd.

$$E_x H_y^* = |B_{mn}|^2 \left(\frac{m\pi}{a}\right)^2 \frac{\beta_2}{w\mu^2 \epsilon} \cos^2\left(\frac{m\pi}{a}x\right) \sin^2\left(\frac{n\pi}{b}y\right)$$

$$E_y H_x^* = |B_{mn}|^2 \left(\frac{n\pi}{b}\right)^2 \frac{\beta_2}{w\mu^2 \epsilon} \sin^2\left(\frac{m\pi}{a}x\right) \cos^2\left(\frac{n\pi}{b}y\right)$$

$$P_{mn} = |B_{mn}|^2 \frac{\beta_2}{w\mu^2 \epsilon} \left\{ \left(\frac{m\pi}{a}\right)^2 \left(\frac{a}{2}\right)\left(\frac{b}{2}\right) + \left(\frac{n\pi}{b}\right)^2 \left(\frac{a}{2}\right)\left(\frac{b}{2}\right) \right\} = |B_{mn}|^2 \frac{ab}{4} \frac{\beta_2}{w\mu^2 \epsilon} \left[\left(\frac{m\pi}{a}\right)^2 + \left(\frac{n\pi}{b}\right)^2 \right]$$

$$P_c = \frac{R_s}{2} \iint \mathbf{J}_s \cdot \mathbf{J}_s ds = 2 \left\{ \frac{R_s}{2} \int_0^a |H_x^+|^2 dx dz + \frac{R_s}{2} \int_0^b |H_y^+|^2 dy dz \right\}$$

$$= R_s l \left\{ \int_0^a |H_x^+|_{y=0}^2 dx + \int_0^b |H_y^+|_{x=0}^2 dy \right\}$$

$$P_c = R_s l \frac{|B_{mn}|^2}{\mu^2} \left\{ \left(\frac{n\pi}{b}\right)^2 \frac{a}{2} + \left(\frac{m\pi}{a}\right)^2 \frac{b}{2} \right\}$$

Thus

$$\alpha_c^{TM} = \frac{R_s}{l^2} \frac{4}{ab} \frac{w\mu^2 \epsilon}{\beta_2} \frac{\left[\left(\frac{m\pi}{b}\right)^2 \frac{a}{2} + \left(\frac{m\pi}{a}\right)^2 \frac{b}{2}\right]}{\left(\frac{m\pi}{a}\right)^2 + \left(\frac{n\pi}{b}\right)^2} = \frac{2R_s w \epsilon}{\beta_2 ab} \frac{\left(\frac{n}{b}\right)^2 a + \left(\frac{m}{a}\right)^2 b}{\left(\frac{m}{a}\right)^2 + \left(\frac{n}{b}\right)^2}$$

$$= \frac{2R_s w \epsilon}{ab w \sqrt{\mu \epsilon} \sqrt{1 - \left(\frac{f_c}{f}\right)^2}} \frac{\frac{n^2 a^3 + m^2 b^3}{a^2 b^2}}{\frac{(mb)^2 + (na)^2}{a^2 b^2}} = \frac{2R_s}{ab \eta} \sqrt{1 - \left(\frac{f_c}{f}\right)^2} \frac{\frac{m^2 b^3 + n^2 a^3}{(mb)^2 + (na)^2}}{(mb)^2 + (na)^2}$$

$$\alpha_c^{TM} = \frac{2R_s}{ab \eta} \sqrt{1 - \left(\frac{f_c}{f}\right)^2} \frac{\frac{m^2 b^3 + n^2 a^3}{(mb)^2 + (na)^2}}{(mb)^2 + (na)^2}$$

$$8.8 \quad E_x = E_0 \sin(\beta_y y) e^{-\gamma z} \Rightarrow \beta_x = 0, \beta_y = \beta_y, \beta_z = -j\gamma$$

$$a. \quad \beta_x^2 + \beta_y^2 + \beta_z^2 = \beta^2 \Rightarrow \beta_y^2 - \gamma^2 = \beta^2 \Rightarrow \gamma^2 = \beta_y^2 - \beta^2$$

$$\nabla^2 E_x + \beta^2 E_x = \left(\frac{\partial^2}{\partial x^2} + \frac{\partial^2}{\partial y^2} + \frac{\partial^2}{\partial z^2} \right) E_x + \beta^2 E_x = -\beta_y^2 + \gamma^2 + \beta^2 = 0$$

$$b. \quad E_x(y=0) = E_0(0) e^{-\gamma z} = 0$$

$$E_x(y=b) = E_0 \sin(\beta_y b) e^{-\gamma z} = 0 \Rightarrow \sin(\beta_y b) = 0 \Rightarrow \beta_y b = n\pi, n=1,2,3,\dots$$

$$\beta_y = \left(\frac{n\pi}{b} \right), n=1,2,3,4,\dots$$

$$\gamma = \sqrt{\beta_y^2 - \beta^2} = \sqrt{\left(\frac{n\pi}{b} \right)^2 - \beta^2} = j\sqrt{\beta^2 - \left(\frac{n\pi}{b} \right)^2} \text{ for } f > f_c$$

$$\beta_c = \omega_c \sqrt{\mu \epsilon} = \beta_y = \frac{n\pi}{b} \Rightarrow \omega_c = \frac{1}{\sqrt{\mu \epsilon}} \left(\frac{n\pi}{b} \right) = 2\pi f_c \Rightarrow f_c = \frac{n}{2b\sqrt{\mu \epsilon}}, n=1,2,3,\dots$$

$$H = -\frac{1}{j\omega \mu} \nabla \times E = -\frac{1}{j\omega \mu} \left[\hat{a}_x(0) + \hat{a}_y \frac{\partial E_x}{\partial z} - \hat{a}_z \frac{\partial E_x}{\partial y} \right]$$

$$H_y = -\frac{1}{j\omega \mu} \frac{\partial E_x}{\partial z} = \frac{\gamma}{j\omega \mu} E_0 \sin(\beta_y y) e^{-\gamma z} \Big|_{y=j\beta_z} = \frac{\beta_z}{\omega \mu} E_0 \sin(\beta_y y) e^{-j\beta_z z}$$

$$\begin{aligned} P_n &= \frac{1}{2} \iint \operatorname{Re}(E \times H^*) \cdot dS = \frac{1}{2} \iint \operatorname{Re} \left[\hat{a}_x E_x \times (\hat{a}_y H_y^* + \hat{a}_z H_z^*) \right] \cdot \hat{a}_z dx dy \\ &= \frac{1}{2} \iint \operatorname{Re} \left(\hat{a}_z E_x H_y^* - \hat{a}_y E_x H_z^* \right) \hat{a}_z dx dy = \operatorname{Re} \left\{ \frac{1}{2} \iint E_x H_y dx dy \right\} \\ &= \operatorname{Re} \left\{ \frac{1}{2} \iint |E_0|^2 \sin^2(\beta_y y) \frac{\beta_z}{\omega \mu} dx dy \right\} = \frac{|E_0|^2 \beta_z l}{4 \omega \mu} \int_0^b [1 - \cos(2\beta_y y)] dy \end{aligned}$$

$$P_n = \frac{|E_0|^2 \beta_z l}{4 \omega \mu} \left[y - \frac{1}{2\beta_y} \sin(2\beta_y y) \right]_0^b = \frac{|E_0|^2}{4} \frac{bl\beta_z}{\omega \mu}$$

$$P_n = \frac{|E_0|^2}{4} \frac{bl}{\omega \mu} \sqrt{1 - \left(\frac{f_c}{f} \right)^2}$$

$$8.9 \quad a. \quad (f_c)_{10} = \frac{1}{2a\sqrt{\mu \epsilon} f_0} = \frac{30 \times 10^9}{2(2.25)} = 6.667 \times 10^9$$

$$b. \quad (f'_c)_{10} = \frac{1}{3} (f_c)_{10} = \frac{1}{3(2a\sqrt{\mu \epsilon} f_0)} = \frac{1}{2a\sqrt{\mu \epsilon} \epsilon_r} \Rightarrow \epsilon_r = 9$$

$$\epsilon_r = 9$$

8.10 a. $H_z(x=0) = H_z(x=a) = 0 \quad \left. \begin{array}{l} \\ H_z(y=0) = H_z(y=b) = 0 \end{array} \right\}$ These are sufficient to derive the appropriate expressions and evaluate the constants.

b. $F_z = [C_1 \cos(\beta_x x) + D_1 \sin(\beta_x x)] [C_2 \cos(\beta_y y) + D_2 \sin(\beta_y y)] e^{-j\beta_z z}$

$$E_x = -\frac{1}{\epsilon} \frac{\partial F_z}{\partial y} \quad H_x = -j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 F_z}{\partial x \partial z}$$

$$E_y = \frac{1}{\epsilon} \frac{\partial F_z}{\partial x} \quad H_y = -j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 F_z}{\partial y \partial z}$$

$$E_z = 0 \quad H_z = -j \frac{1}{\omega \mu \epsilon} (\frac{\partial^2}{\partial z^2} + \beta^2) F_z$$

$$H_z = -j \frac{1}{\omega \mu \epsilon} (-\beta_z^2 + \beta^2) F_z = -j \frac{\beta_z^2}{\omega \mu \epsilon} F_z = -j \frac{\beta_z^2}{\omega \mu \epsilon} [C_1 \cos(\beta_x x) + D_1 \sin(\beta_x x)] [C_2 \cos(\beta_y y) + D_2 \sin(\beta_y y)] e^{-j\beta_z z}$$

$$H_z(x=0) = -j \frac{\beta_z^2}{\omega \mu \epsilon} [C_1(1) + D_1(0)] [C_2(1) + D_2(0)] e^{-j\beta_z z} = 0 \Rightarrow C_1 = 0$$

$$H_z(x=a) = -j \frac{\beta_z^2}{\omega \mu \epsilon} [D_1 \sin(\beta_x a)] [C_2(1) + D_2(0)] e^{-j\beta_z z} = 0 \Rightarrow \beta_x a = \sin^{-1}(0) = n\pi \Rightarrow \beta_x = \frac{n\pi}{a}, n=1,2,\dots$$

$$H_z(y=0) = -j \frac{\beta_z^2}{\omega \mu \epsilon} D_1 \sin(\beta_x x) [C_2(1) + D_2(0)] e^{-j\beta_z z} = 0 \Rightarrow C_2 = 0$$

$$H_z(y=b) = -j \frac{\beta_z^2}{\omega \mu \epsilon} D_1 \sin(\beta_x x) D_2 \sin(\beta_y b) e^{-j\beta_z z} = 0 \Rightarrow \beta_y b = \sin^{-1}(0) = n\pi \Rightarrow \beta_y = \frac{n\pi}{b}, n=1,2,\dots$$

$$F_z = A_{mn} \sin(\beta_x x) \sin(\beta_y y) e^{-j\beta_z z}, \quad \beta_x = \frac{m\pi}{a}, m=1,2,\dots \quad \beta_y = \frac{n\pi}{b}, n=1,2,3,\dots$$

$$E_x = -A_{mn} \frac{\beta_y}{\epsilon} \sin(\beta_x x) \cos(\beta_y y) e^{-j\beta_z z}, \quad H_x = -A_{mn} \frac{\beta_x \beta_z}{\omega \mu \epsilon} \cos(\beta_x x) \sin(\beta_y y) e^{-j\beta_z z}$$

$$E_y = +A_{mn} \frac{\beta_x}{\epsilon} \cos(\beta_x x) \sin(\beta_y y) e^{-j\beta_z z}, \quad H_y = -A_{mn} \frac{\beta_y \beta_z}{\omega \mu \epsilon} \sin(\beta_x x) \cos(\beta_y y) e^{-j\beta_z z}$$

$$E_z = 0, \quad H_z = -j A_{mn} \frac{\beta_z^2}{\omega \mu \epsilon} \sin(\beta_x x) \sin(\beta_y y) e^{-j\beta_z z}$$

$$\beta_z = \sqrt{\beta_x^2 + \beta_y^2} = 2\pi f_c \sqrt{\mu \epsilon} = \sqrt{(\frac{m\pi}{a})^2 + (\frac{n\pi}{b})^2}$$

$$(f_c)_{mn} = \frac{1}{2\pi \sqrt{\mu \epsilon}} \sqrt{(\frac{m\pi}{a})^2 + (\frac{n\pi}{b})^2}, \quad m=1,2,3,\dots, \quad n=1,2,3,\dots$$

c. If $a > b$ the dominant mode is the TE₁₁. Thus

$$(f_c)_{11} = \frac{1}{2\sqrt{\mu \epsilon}} \sqrt{(\frac{1}{a})^2 + (\frac{1}{b})^2}$$

B.11 c. $H_x(y=0) = H_x(y=b) = 0$ These are sufficient to derive the appropriate
 $H_y(x=0) = H_y(x=a) = 0$ expressions and evaluate the constants.

b. $A_2 = [C_1 \cos \beta_x x + D_1 \sin \beta_x x] [C_2 \cos \beta_y y + D_2 \sin \beta_y y] e^{-j\beta_2 z}$

$$E_x = -j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 A_2}{\partial x \partial z}$$

$$E_y = -j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 A_2}{\partial y \partial z}$$

$$E_z = -j \frac{1}{\omega \mu \epsilon} \left(\frac{\partial^2}{\partial z^2} + \beta^2 \right) A_2 \quad H_z = 0$$

$$H_y(x=0) = \frac{\beta_x}{\mu} [C_1(0) + D_1(1)] [C_2 \cos \beta_y y + D_2 \sin \beta_y y] e^{-j\beta_2 z} = 0 \Rightarrow D_1 = 0$$

$$H_y(x=a) = -\frac{\beta_x}{\mu} [-C_1 \sin(\beta_x a)] [C_2 \cos \beta_y y + D_2 \sin \beta_y y] e^{-j\beta_2 z} = 0 \Rightarrow \beta_x a = \sin^{-1}(0) = m\pi \Rightarrow \beta_x = \left(\frac{m\pi}{a}\right), m=0,1,2,\dots$$

$$H_x(y=0) = \frac{\beta_y}{\mu} [-C_2(0) + D_2(1)] [C_1 \cos \beta_x x + D_1 \sin \beta_x x] e^{-j\beta_2 z} = 0 \Rightarrow D_2 = 0$$

$$H_x(y=b) = \frac{\beta_y}{\mu} [-C_2 \sin(\beta_y b)] [C_1 \cos \beta_x x + D_1 \sin \beta_x x] e^{-j\beta_2 z} = 0 \Rightarrow \beta_y b = \sin^{-1}(0) = n\pi \Rightarrow \beta_y = \left(\frac{n\pi}{b}\right), n=1,2,\dots$$

$$A_2 = B_{mn} \cos(\beta_x x) \cos(\beta_y y) e^{-j\beta_2 z}, \quad \beta_x = \frac{m\pi}{a}, m=0,1,2,\dots; \quad \beta_y = \frac{n\pi}{b}, n=0,1,2,\dots$$

$$E_x = B_{mn} \frac{\beta_x \beta_2}{\omega \mu \epsilon} \sin(\beta_x x) \cos(\beta_y y) e^{-j\beta_2 z}$$

$$E_y = B_{mn} \frac{\beta_y \beta_2}{\omega \mu \epsilon} \cos(\beta_x x) \sin(\beta_y y) e^{-j\beta_2 z}$$

$$E_z = -j A_{mn} \frac{\beta^2}{\omega \mu \epsilon} \cos(\beta_x x) \cos(\beta_y y) e^{-j\beta_2 z}$$

$$H_x = -A_{mn} \frac{\beta_y}{\mu} \cos(\beta_x x) \sin(\beta_y y) e^{-j\beta_2 z}$$

$$H_y = A_{mn} \frac{\beta_x}{\mu} \sin(\beta_x x) \cos(\beta_y y) e^{-j\beta_2 z}$$

$$H_z = 0$$

$$\beta_c = 2\pi f_c \sqrt{\mu \epsilon} = \sqrt{\beta_x^2 + \beta_y^2} = \sqrt{\left(\frac{m\pi}{a}\right)^2 + \left(\frac{n\pi}{b}\right)^2}$$

$$(f_c)_{mn} = \frac{1}{2\pi \sqrt{\mu \epsilon}} \sqrt{\left(\frac{m\pi}{a}\right)^2 + \left(\frac{n\pi}{b}\right)^2}, \quad m=0,1,2,\dots, \quad n=0,1,2,\dots \quad \left. \right\} m=n \neq 0$$

c. If $a > b$ the dominant mode is the TM_{10}^2 . Thus

$$(f_c)_{10}^{TM^2} = \frac{1}{2\pi \sqrt{\mu \epsilon}} \left(\frac{\pi}{a}\right) = \frac{1}{2a \sqrt{\mu \epsilon}}$$

8.12

 TE^2

$$\text{B.C.'s: } \left. \begin{array}{l} E_x(y=0) = E_x(y=b) = 0 \\ H_y(x=0) = H_y(x=a) = 0 \end{array} \right\} F_2 = [C_1 \cos \beta_x x + D_1 \sin \beta_x x] [C_2 \cos \beta_y y + D_2 \sin \beta_y y] e^{-j \beta_z z}$$

$$E_x = -\frac{1}{\epsilon} \frac{\partial F_2}{\partial y} = -\frac{\beta_y}{\epsilon} [C_1 \cos \beta_x x + D_1 \sin \beta_x x] [-C_2 \sin \beta_y y + D_2 \cos \beta_y y] e^{-j \beta_z z}$$

$$H_y = -j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 F_2}{\partial y^2} = -\frac{\beta_y \beta_z}{\omega \mu \epsilon} [C_1 \cos \beta_x x + D_1 \sin \beta_x x] [-C_2 \sin(\beta_y y) + D_2 \cos(\beta_y y)] e^{-j \beta_z z}$$

$$E_x(y=0) = 0 \Rightarrow D_2 = 0$$

$$E_x(y=b) = 0 \Rightarrow \sin(\beta_y b) = 0 \Rightarrow \beta_y b = \sin^{-1}(0) = n\pi \Rightarrow \beta_y = \frac{n\pi}{b}, n = 0, 1, 2, \dots$$

$$H_y(x=0) = 0 \Rightarrow C_1 = 0$$

$$H_y(x=a) = 0 \Rightarrow \sin(\beta_x a) = 0 \Rightarrow \beta_x a = \sin^{-1}(0) = m\pi \Rightarrow \beta_x = \frac{m\pi}{a}, m = 1, 2, 3, \dots$$

$$F_2 = A_{mn} \sin(\beta_x x) \cos(\beta_y y) e^{-j \beta_z z}, \beta_x = \frac{m\pi}{a}, \beta_y = \frac{n\pi}{b}, n = 0, 1, 2, \dots$$

$$E_x = -\frac{1}{\epsilon} \frac{\partial F_2}{\partial y} = A_{mn} \frac{\beta_y}{\epsilon} \sin(\beta_x x) \sin(\beta_y y) e^{-j \beta_z z}, H_x = -j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 F_2}{\partial y^2} = -A_{mn} \frac{\beta_x \beta_z}{\omega \mu \epsilon} \cos(\beta_x x) \cos(\beta_y y) e^{-j \beta_z z}$$

$$E_y = \frac{1}{\epsilon} \frac{\partial F_2}{\partial x} = A_{mn} \frac{\beta_x}{\epsilon} \cos(\beta_x x) \cos(\beta_y y) e^{-j \beta_z z}, H_y = -j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 F_2}{\partial x^2} = +A_{mn} \frac{\beta_y \beta_z}{\omega \mu \epsilon} \sin(\beta_x x) \sin(\beta_y y) e^{-j \beta_z z}$$

$$E_z = 0, H_z = -j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 F_2}{\partial z^2} = -j \frac{\beta_z^2}{\omega \mu \epsilon} \sin(\beta_x x) \cos(\beta_y y) e^{-j \beta_z z}$$

$$(f_c)_{mn} = \frac{1}{2\pi\sqrt{\mu\epsilon}} \sqrt{\left(\frac{m\pi}{a}\right)^2 + \left(\frac{n\pi}{b}\right)^2}, \lambda_g = \frac{2}{\sqrt{1 - \left(\frac{f_c}{c}\right)^2}}, Z_w = \frac{E_x}{H_y} = -\frac{E_y}{H_x} = \frac{\omega \mu}{\beta_z}$$

$$\beta_z = \sqrt{\beta^2 - \beta_c^2} = \begin{cases} \beta \sqrt{1 - \left(\frac{f_c}{c}\right)^2} & f > f_c \\ 0 & f = f_c \\ -j \beta \sqrt{\left(\frac{f_c}{c}\right)^2 - 1} & f < f_c \end{cases} \quad \text{Dominant Mode is the } TE_{10}$$

$$\underline{TM^2} \Rightarrow A_2 = [C_1 \cos(\beta_x x) + D_1 \sin(\beta_x x)] [C_2 \cos \beta_y y + D_2 \sin \beta_y y] e^{-j \beta_z z}$$

$$E_x = -j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 A_2}{\partial x \partial z} = -\frac{\beta_x \beta_z}{\omega \mu \epsilon} [C_1 \sin(\beta_x x) + D_1 \cos(\beta_x x)] [C_2 \cos \beta_y y + D_2 \sin \beta_y y] e^{-j \beta_z z}$$

$$H_y = -\frac{1}{\mu} \frac{\partial A_2}{\partial x} = -\frac{\beta_x}{\mu} [-C_1 \sin(\beta_x x) + D_1 \cos(\beta_x x)] [C_2 \cos \beta_y y + D_2 \sin \beta_y y] e^{-j \beta_z z}$$

$$E_x(y=0) = 0 \Rightarrow C_2 = 0; E_x(y=b) = \sin(\beta_y b) = 0 \Rightarrow \beta_y = \frac{(n\pi)}{b}, n = 1, 2, 3, \dots$$

$$H_y(x=0) = 0 \Rightarrow D_1 = 0; H_y(x=a) = \sin(\beta_x a) = 0 \Rightarrow \beta_x = \frac{(m\pi)}{a}, m = 0, 1, 2, \dots$$

$$A_2 = B_{mn} \cos(\beta_x x) \sin(\beta_y y) e^{-j \beta_z z}, \beta_x = \frac{m\pi}{a}, m = 0, 1, 2, \dots; \beta_y = \frac{n\pi}{b}, n = 1, 2, \dots$$

$$E_x = -j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 A_2}{\partial y \partial z} = B_{mn} \frac{\beta_y \beta_z}{\omega \mu \epsilon} \sin(\beta_x x) \sin(\beta_y y) e^{-j \beta_z z}, H_x = \frac{1}{\mu} \frac{\partial A_2}{\partial y} = B_{mn} \frac{\beta_y}{\mu} \cos(\beta_x x) \cos(\beta_y y) e^{-j \beta_z z}$$

$$E_y = -j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 A_2}{\partial x \partial z} = B_{mn} \frac{\beta_x \beta_z}{\omega \mu \epsilon} \cos(\beta_x x) \cos(\beta_y y) e^{-j \beta_z z}, H_y = -\frac{1}{\mu} \frac{\partial A_2}{\partial x} = B_{mn} \frac{\beta_x}{\mu} \sin(\beta_x x) \sin(\beta_y y) e^{-j \beta_z z}$$

$$E_z = j \frac{1}{\omega \mu \epsilon} \left(\frac{\beta^2}{\beta_z^2} + \beta^2 \right) A_2 = -j \frac{\beta_z^2}{\omega \mu \epsilon} \cos(\beta_x x) \sin(\beta_y y) e^{-j \beta_z z}, H_z = 0$$

$$Z_w = \frac{E_x}{H_y} = -\frac{E_y}{H_x} = \frac{\beta_z}{\omega \epsilon} \quad \text{Dominant Mode is the } TM_{01} \left(f_c = \frac{1}{2b\sqrt{\mu\epsilon}} \right)$$

If $a > b$: Lowest Order Mode is TE_{10} ; If $a < b$, Lowest Order Mode is TM_{01}

B.C.'s: $H_x(y=0) = H_x(y=b) = 0$ } Necessary and sufficient boundary
 $E_y(x=0) = E_y(x=a) = 0$ } conditions

TE₀₁: $F_2 = [C_1 \cos(\beta_x x) + D_1 \sin(\beta_x x)][C_2 \cos(\beta_y y) + D_2 \sin(\beta_y y)] e^{-j\beta z^2}$

$$H_x = -j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 F_2}{\partial x^2} = -\frac{\beta_x \beta_z}{\omega \mu \epsilon} [-C_1 \sin(\beta_x x) + D_1 \cos(\beta_x x)][C_2 \cos(\beta_y y) + D_2 \sin(\beta_y y)] e^{-j\beta z^2}$$

$$E_y = \frac{1}{\epsilon} \frac{\partial F_2}{\partial y} = \frac{\beta_x}{\epsilon} [-C_1 \sin(\beta_x x) + D_1 \cos(\beta_x x)][C_2 \cos(\beta_y y) + D_2 \sin(\beta_y y)] e^{-j\beta z^2}$$

$$H_x(y=0) = 0 \Rightarrow C_2 = 0; H_x(y=b) = 0 \Rightarrow \sin(\beta_y b) = 0 \Rightarrow \beta_y = \left(\frac{n\pi}{b}\right), n=1, 2, 3, \dots$$

$$E_y(x=0) = 0 \Rightarrow D_1 = 0; E_y(x=a) = 0 \Rightarrow \sin(\beta_x a) = 0 \Rightarrow \beta_x = \left(\frac{m\pi}{a}\right), m=0, 1, 2, \dots$$

$$F_2 = A_{mn} \cos(\beta_x x) \sin(\beta_y y) e^{-j\beta z^2}, \beta_x = \left(\frac{m\pi}{a}\right), m=0, 1, 2, \dots; \beta_y = \left(\frac{n\pi}{b}\right), n=1, 2, 3, \dots$$

$$E_x = -\frac{1}{\epsilon} \frac{\partial F_2}{\partial y} = -A_{mn} \frac{\beta_y}{\epsilon} \cos(\beta_x x) \cos(\beta_y y) e^{-j\beta z^2}, H_x = -j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 F_2}{\partial x^2} = A_{mn} \frac{\beta_x \beta_z}{\omega \mu \epsilon} \sin(\beta_x x) \sin(\beta_y y) e^{-j\beta z^2}$$

$$E_y = \frac{1}{\epsilon} \frac{\partial F_2}{\partial x} = -A_{mn} \frac{\beta_x}{\epsilon} \sin(\beta_x x) \sin(\beta_y y) e^{-j\beta z^2}, H_y = j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 F_2}{\partial y^2} = -A_{mn} \frac{\beta_y \beta_z}{\omega \mu \epsilon} \cos(\beta_x x) \cos(\beta_y y) e^{-j\beta z^2}$$

$$E_z = 0, H_z = -j \frac{1}{\omega \mu \epsilon} \left(\frac{\partial^2}{\partial z^2} + \beta_z^2 \right) F_2 = -j A_{mn} \frac{\beta_z}{\omega \mu \epsilon} \cos(\beta_x x) \sin(\beta_y y) e^{-j\beta z^2}$$

$$(f_c)_{mn} = \frac{1}{2\pi\sqrt{\mu\epsilon}} \sqrt{\left(\frac{m\pi}{a}\right)^2 + \left(\frac{n\pi}{b}\right)^2}, \beta_g = \frac{1}{\sqrt{1-(f_c/\epsilon)^2}}, Z_W = \frac{E_x}{H_y} = -\frac{E_y}{H_x} = \frac{\omega h}{\beta z}$$

$$\beta_z = \pm \sqrt{\beta^2 - \beta_g^2} = \begin{cases} \beta \sqrt{1 - \left(\frac{f_c}{\epsilon}\right)^2} & f > f_c \\ 0 & f = f_c \\ -j \beta \sqrt{\left(\frac{f_c}{\epsilon}\right)^2 - 1} & f < f_c \end{cases} \quad \text{Dominant Mode is the TE}_{01}$$

$$(f_c)_{01} = \frac{1}{2b\sqrt{\mu\epsilon}}$$

TM₀₁: $A_2 = [C_1 \cos(\beta_x x) + D_1 \sin(\beta_x x)][C_2 \cos(\beta_y y) + D_2 \sin(\beta_y y)] e^{-j\beta z^2}$

$$H_x = +j \frac{1}{\mu} \frac{\partial A_2}{\partial y} = \frac{\beta_y}{\mu} [C_1 \cos(\beta_x x) + D_1 \sin(\beta_x x)][-C_2 \sin(\beta_y y) + D_2 \cos(\beta_y y)] e^{-j\beta z^2}$$

$$E_y = -j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 A_2}{\partial y^2} = -\frac{\beta_x \beta_z}{\omega \mu \epsilon} [C_1 \cos(\beta_x x) + D_1 \sin(\beta_x x)][-C_2 \sin(\beta_y y) + D_2 \cos(\beta_y y)] e^{-j\beta z^2}$$

$$H_x(y=0) = 0 \Rightarrow D_2 = 0; H_x(y=b) = 0 \Rightarrow \sin(\beta_y b) = 0 \Rightarrow \beta_y = \left(\frac{n\pi}{b}\right), n=0, 1, 2, \dots$$

$$E_y(x=0) = 0 \Rightarrow C_1 = 0; E_y(x=a) = 0 \Rightarrow \sin(\beta_x a) = 0 \Rightarrow \beta_x = \left(\frac{m\pi}{a}\right), m=1, 2, 3, \dots$$

$$A_2 = B_{mn} \sin(\beta_x x) \cos(\beta_y y) e^{-j\beta z^2}, \beta_x = \frac{m\pi}{a}, m=1, 2, 3, \dots; \beta_y = \frac{n\pi}{b}, n=0, 1, 2, \dots$$

$$E_x = -j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 A_2}{\partial x^2} = -B_{mn} \frac{\beta_x \beta_z}{\omega \mu \epsilon} \cos(\beta_x x) \cos(\beta_y y) e^{-j\beta z^2}, H_x = \frac{1}{\mu} \frac{\partial A_2}{\partial y} = -B_{mn} \frac{\beta_y}{\mu} \sin(\beta_x x) \sin(\beta_y y) e^{-j\beta z^2}$$

$$E_y = -j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 A_2}{\partial y^2} = B_{mn} \frac{\beta_y \beta_z}{\omega \mu \epsilon} \sin(\beta_x x) \sin(\beta_y y) e^{-j\beta z^2}, H_y = -j \frac{1}{\omega \mu \epsilon} \frac{\partial A_2}{\partial x} = -B_{mn} \frac{\beta_x}{\omega \mu \epsilon} \cos(\beta_x x) \cos(\beta_y y) e^{-j\beta z^2}$$

$$E_z = -j \frac{1}{\omega \mu \epsilon} \left(\frac{\partial^2}{\partial z^2} + \beta_z^2 \right) A_2 = -j \frac{\beta_z^2}{\omega \mu \epsilon} \sin(\beta_x x) \cos(\beta_y y) e^{-j\beta z^2}, H_z = 0$$

$$Z_W = \frac{E_x}{H_y} = -\frac{E_y}{H_x} = \frac{\beta_z}{\omega \epsilon} \quad \left\{ (f_c)_{mn}, \beta_g \text{ and } \beta_z \text{ are the same as above.} \right.$$

The dominant mode is the TM₁₀ mode. $(f_c)_{10} = \frac{1}{2a\sqrt{\mu\epsilon}}$

If $a > b$: Lowest order mode is TM₁₀.

If $a < b$: Lowest order mode is TE₀₁.

8.14

$$\text{TE}^{\frac{1}{2}}: E_z^+ (x, y, z) = [C_1 \cos(\beta_x x) + D_1 \sin(\beta_x x)] [C_2 \cos(\beta_y y) + D_2 \sin(\beta_y y)] A_3 e^{-j\beta_z z} \quad (8.7)$$

$$\beta_x^2 + \beta_y^2 + \beta_z^2 = \beta^2 \quad (8.7a)$$

$$E_x = -\frac{1}{\epsilon} \frac{\partial F_z}{\partial y}$$

$$H_x = -j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 F_z}{\partial y \partial z}$$

$$E_y = \frac{1}{\epsilon} \frac{\partial F_z}{\partial x}$$

$$H_y = -j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 F_z}{\partial x \partial z}$$

$$E_z = 0$$

$$H_z = -j \frac{1}{\omega \mu \epsilon} (\frac{\partial^2}{\partial z^2} + \beta^2) F_z$$

$$\text{B.C.s: } E_x^+ (0 \leq x \leq a, y=0, z) = 0 \Rightarrow E_x^+ (0 \leq x \leq a, y=b, z)$$

$$H_y^+ (x=0, 0 \leq y \leq b, z) = 0 = H_y^+ (x=a, 0 \leq y \leq b, z)$$

$$E_x^+ = -\frac{1}{\epsilon} \frac{\partial F_y^+}{\partial y} = -\frac{\beta_y}{\epsilon} [C_1 \cos(\beta_x x) + D_1 \sin(\beta_x x)] [-C_2 \sin(\beta_y y) + D_2 \cos(\beta_y y)] A_3 e^{-j\beta_z z}$$

$$\text{B.C. } E_x^+ (y=0) = 0 \Rightarrow \frac{\beta_y}{\epsilon} [C_1(0) + D_1(1)] A_3 e^{-j\beta_z z} = 0 \Rightarrow D_1 = 0$$

$$E_x^+ (y=b) = -\frac{\beta_y}{\epsilon} [-C_2(\sin \beta_y b)] A_3 e^{-j\beta_z z} = 0$$

$$\sin(\beta_y b) = 0 \Rightarrow \beta_y b = n\pi, n=0, 1, 2, 3, \dots$$

$$\beta_y = \left(\frac{n\pi}{b}\right), n=0, 1, 2, 3, \dots$$

$$F_z^+ (x, y, z) = C_2 [C_1 \cos(\beta_x x) + D_1 \sin(\beta_x x)] \cos(\beta_y y) A_3 e^{-j\beta_z z}$$

$$H_y^+ = -j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 F_z^+}{\partial y \partial z} = +j \frac{\beta_y (-j\beta_z)}{\omega \mu \epsilon} C_2 [C_1 \cos(\beta_x x) + D_1 \sin(\beta_x x)] \sin(\beta_y y) A_3 e^{-j\beta_z z}$$

$$\text{B.C. } H_y^+ (x=0) = j \beta_y (-j\beta_z) C_2 [C_1(-1) + D_1(0)] \sin(\beta_y y) A_3 e^{-j\beta_z z} = 0 \Rightarrow C_1 = 0$$

$$H_y^+ (x=a) = j \beta_y (-j\beta_z) C_2 [D_1 \sin(\beta_x a)] \sin(\beta_y y) A_3 e^{-j\beta_z z} = 0$$

$$\sin(\beta_x a) = 0 \Rightarrow \beta_x a = n\pi, n=0, 1, 2, \dots$$

$$\beta_x = \left(\frac{m\pi}{a}\right), m=0, 1, 2, \dots$$

$$F_z^+ (x, y, z) = A_{mn} \sin(\beta_x x) \cos(\beta_y y) e^{-j\beta_z z} \quad \begin{cases} \beta_x = \left(\frac{m\pi}{a}\right), m=0, 1, 2, \dots \\ \beta_y = \left(\frac{n\pi}{b}\right), n=0, 1, 2, \dots \end{cases}$$

TE^z Dominant Mode:

$$(f_c)_{mn} = \frac{1}{2\pi f \mu \epsilon} \sqrt{\left(\frac{m\pi}{a}\right)^2 + \left(\frac{n\pi}{b}\right)^2}$$

$$(f_c)_{10} = \frac{1}{2\pi f \mu \epsilon} \sqrt{\left(\frac{\pi}{a}\right)^2} = \frac{1}{2\pi f \mu \epsilon} \sqrt{\frac{\pi^2}{a^2}} = \frac{30 \times 10^9}{2(2.286)\sqrt{1(81)}}$$

$$(f_c)_{10} = \frac{30 \times 10^9}{2(2.286)(9)} = 0.729 \text{ GHz}$$

8.15 Since dominant mode is TE₁₀:

$$a. (\alpha_c)^0 = \frac{R_s}{\eta b} \frac{1+2(\frac{b}{\lambda})(\frac{f_c}{f})^2}{\sqrt{1-(\frac{f_c}{f})^2}}, R_s = \sqrt{\frac{w}{2\sigma}}$$

$$(f_c)^0 = \frac{1}{2\pi\sqrt{\mu\varepsilon}} = \frac{30 \times 10^9}{2(2.286)\sqrt{2.56}} = 4.1 \times 10^9$$

$$(\alpha_c)^0 = \frac{1}{b\sqrt{\frac{\mu}{\varepsilon}}} \sqrt{\frac{w\varepsilon}{2\sigma}} \frac{1+2(\frac{b}{\lambda})(\frac{f_c}{f})^2}{\sqrt{1-(\frac{f_c}{f})^2}} = \frac{1}{b} \sqrt{\frac{w\varepsilon}{2\sigma}} \frac{1+2(\frac{b}{\lambda})(\frac{f_c}{f})^2}{\sqrt{1-(\frac{f_c}{f})^2}}$$

$$(\alpha_c)^0 = 1.6054 \times 10^{-2} \text{ Np/m} = 8.68 (1.6054 \times 10^{-2}) \\ = 1.393 \times 10^{-1} \text{ dB/m}$$

$$\left. \begin{array}{l} b = 1.016 \times 10^{-2} \text{ m} \\ f = 6.15 \times 10^9 \text{ Hz} \\ f_c = 4.1 \times 10^9 \text{ Hz} \\ \sigma = 5.76 \times 10^7 \text{ S/m} \end{array} \right.$$

$$b. \alpha_d = \frac{\epsilon_r''}{\epsilon_r'} \frac{\pi}{\lambda} \left(\frac{2g}{\lambda} \right) = \tan \delta_e \frac{\pi}{\lambda} \frac{2g}{\lambda}$$

$$\lambda = \frac{2g}{\lambda_{cv}} = \frac{30 \times 10^9 / 6.15 \times 10^9}{\sqrt{2.56}} = \frac{4.8747}{1.6} = 3.0467 \text{ cm} = 3.0467 \times 10^{-2} \text{ m}$$

$$2g = \frac{\lambda}{\sqrt{1-(\frac{f_c}{f})^2}} = \frac{3.0467 \times 10^{-2}}{\sqrt{1-(\frac{4.1}{6.15})^2}} = 4.0862 \times 10^{-2} \text{ m}$$

$$\alpha_d = 4 \times 10^{-4} \left(\frac{\pi}{3.0467 \times 10^{-2}} \right) \left(\frac{4.0862}{3.0467} \right) = 5.5318 \times 10^{-2} \text{ Np/m} = 4.802 \text{ dB/m}$$

8.16 a. From Problem 8.15 $(f_c)^0 = 4.1 \times 10^9$

$$b. \omega s \theta = \frac{\beta_z}{k} = \sqrt{1 - \left(\frac{f_c}{f} \right)^2} \Rightarrow \left(\frac{f_c}{f} \right)^2 = 1 - \omega s^2 \theta \Rightarrow f = \frac{f_c}{\sqrt{1 - \omega s^2 \theta}} = \frac{4.1 \times 10^9}{\sqrt{1 - 0.5}} = 5.798 \times 10^9$$

$$c. \lambda_g = \frac{\lambda}{\sqrt{1 - \left(\frac{f_c}{f} \right)^2}} = \frac{30 \times 10^9 / (\sqrt{2.56} \times 5.798 \times 10^9)}{\sqrt{1 - \left(\frac{4.1}{5.798} \right)^2}} = \frac{3.2338}{0.707} = 4.574 \text{ cm}$$

$$d. \beta_z = \beta \sqrt{1 - \left(\frac{f_c}{f} \right)^2} = \frac{2\pi}{\lambda_g}$$

$$\beta_z d = \frac{2\pi}{\lambda_g} d = 2\pi \Rightarrow d = \lambda_g = 4.574 \text{ cm}$$

$$8.17 \quad a = 0.9 \text{ in.} (2.286 \text{ cm}), b = 0.4 \text{ in.} (1.016 \text{ cm}), f = 10 \text{ GHz}$$

$$a. \quad Z_w = \frac{\eta}{\sqrt{1 - (\frac{f_c}{f})^2}} = \frac{377}{\sqrt{1 - (\frac{6.56}{10})^2}} = 499.5$$

$$\Gamma_L = \frac{\eta_0 - Z_w}{\eta_0 + Z_w} = \frac{377 - 499.5}{377 + 499.5} = -0.1398 = 0.1398 [180^\circ]$$

$$b. \quad SWR = \frac{1 + |\Gamma_L|}{1 - |\Gamma_L|} = \frac{1 + 0.1398}{1 - 0.1398} = 1.325$$

$$c. \quad \Gamma_{in} = \Gamma_L e^{-2\gamma z} = \Gamma_L e^{-2\alpha_c z} e^{-j2\beta z} \Rightarrow |\Gamma_{in}| = |\Gamma_L| e^{-2\alpha_c z}$$

$$\alpha_c = \frac{R_s}{b\eta} \frac{\left[1 + 2\left(\frac{b}{a}\right)\left(\frac{f_c}{f}\right)^2\right]}{\sqrt{1 - \left(\frac{f_c}{f}\right)^2}} = \frac{1}{b} \sqrt{\frac{\omega h}{2\sigma}} \frac{\left[1 + 2\left(\frac{b}{a}\right)\left(\frac{f_c}{f}\right)^2\right]}{\sqrt{1 - \left(\frac{f_c}{f}\right)^2}} = \frac{1}{b} \sqrt{\frac{\omega \epsilon}{2\sigma}} \sqrt{\left[1 + 2\left(\frac{b}{a}\right)\left(\frac{f_c}{f}\right)^2\right]}$$

$$(f_c)_{10} = 6.56 \times 10^9, \sigma = 5.76 \times 10^7 \text{ S/m}, f = 10 \text{ GHz}, \epsilon = \epsilon_0$$

$$\alpha_c = 1.25217 \times 10^{-2} \text{ Np/m} = 1.08813 \times 10^{-4} \text{ dB/m}$$

$$\gamma_g = \frac{2}{\sqrt{1 - (\frac{f_c}{f})^2}} = \frac{2\eta/\sqrt{\epsilon_r}}{\sqrt{1 - (\frac{f_c}{f})^2}} = \frac{30 \times 10^9 / (0.0813 \sqrt{1})}{\sqrt{1 - (\frac{6.56}{10})^2}} = \frac{3}{0.7548} = 3.975 \text{ cm} = 3.975 \times 10^{-2} \text{ m}$$

$$z=0: |\Gamma_{in}| = |\Gamma_L| = 0.1398, \text{ SWR} = 1.325$$

$$z=\gamma_g/4: |\Gamma_{in}| = |\Gamma_L| e^{-2\alpha_c z} = 0.1398 e^{-2(1.25217 \times 10^{-2})(3.975 \times 10^{-2}/4)} = 0.1398 e^{-0.000238}$$

$$|\Gamma_{in}| = 0.1398 (0.9998) = 0.1398 \Rightarrow \text{SWR} = \frac{1+0.1398}{1-0.1398} = 1.325$$

$$z=\gamma_g/2: |\Gamma_{in}| = |\Gamma_L| e^{-2\alpha_c z} = 0.1398 e^{-2(1.25217 \times 10^{-2})(3.975 \times 10^{-2}/2)} = 0.1398 e^{-0.000475}$$

$$|\Gamma_{in}| = 0.1398 (0.9995) = 0.1397 \Rightarrow \text{SWR} = \frac{1+0.1397}{1-0.1397} = 1.325$$

8.1B

$$\epsilon_r = 81, a = 2.286 \text{ cm}, b = 1.016 \text{ cm}, \mu_r = 1$$

a. Since the walls of the waveguide can be modeled as PMC, the expressions for the cutoff frequencies are the same as for the PEC except reversed. Therefore for the PMC, the expressions for the cutoff frequencies are:

$$TE^z(\text{PMC}) = TM^z(\text{PEC}) = \frac{1}{2\pi\sqrt{\mu\epsilon}} \sqrt{\left(\frac{m\pi}{a}\right)^2 + \left(\frac{n\pi}{b}\right)^2} \quad \begin{cases} m=1, 2, 3, \dots \\ n=1, 2, 3, \dots \end{cases}$$

$$TM^z(\text{PMC}) = TE^z(\text{PEC}) = \frac{1}{2\pi\sqrt{\mu\epsilon}} \sqrt{\left(\frac{m\pi}{a}\right)^2 + \left(\frac{n\pi}{b}\right)^2} \quad \begin{cases} m=0, 1, 2, \dots \\ n=0, 1, 2, \dots \end{cases}$$

b. For $a > b$, the dominant mode is the $TM_{10}^z(\text{PMC})$

$$\begin{aligned} c. f_c(TM_{10}^z) &= \frac{1}{2\pi\sqrt{\mu\epsilon}} \left(\frac{\pi}{a} \right) = \frac{1}{2a\sqrt{\mu\epsilon}} \quad \begin{cases} \epsilon = 81\epsilon_0, \mu = \mu_0 \\ \end{cases} \\ &= \frac{1}{2a} \frac{1}{\sqrt{\epsilon_r\mu_0}} - \frac{1}{2a} \frac{V_0}{\sqrt{\epsilon_r}} = \frac{30 \times 10^9}{2(2.286)\sqrt{81}} \end{aligned}$$

$$f_c(TM_{10}^z) = 0.7291 \times 10^9 = 0.7291 \text{ GHz}$$

$$\begin{aligned} d. \beta_z &= \beta \sqrt{1 - (\frac{f_c}{f})^2} = \omega \sqrt{\mu\epsilon} \sqrt{1 - \left(\frac{f_c}{2f_c}\right)^2} = 2\pi f \sqrt{\mu\epsilon} \sqrt{1 - \left(\frac{1}{2}\right)^2} \\ &= 2\pi f \frac{\sqrt{\epsilon_r}}{V_0} \sqrt{1 - \left(\frac{1}{2}\right)} = 2\pi \frac{(2 \times 0.7291 \times 10^9)}{30 \times 10^9} \sqrt{\frac{3}{4}} \end{aligned}$$

$$\beta_z = 2.3803 \text{ rad/cm} = 2.3803 \left(\frac{180}{\pi}\right) = 136.382 \text{ degrees/cm}$$

$$\beta_z d = 136.382(d) = 360 \Rightarrow d = 360 / 136.382 = 2.6396$$

$$d = 2.6396 \text{ cm}$$

$$8.19 \quad (f_r)_{101} = \frac{1}{2\pi\sqrt{\mu_0\epsilon_r}} \sqrt{\left(\frac{\pi}{a}\right)^2 + \left(\frac{\pi}{c}\right)^2} = \frac{1}{2\sqrt{\mu_0\epsilon_r}} \sqrt{\left(\frac{1}{a}\right)^2 + \left(\frac{1}{c}\right)^2} = \frac{v_0}{2} \sqrt{\left(\frac{1}{a}\right)^2 + \left(\frac{1}{c}\right)^2}$$

$$(f_r)^2 = \frac{v_0^2}{4} \left[\frac{1}{a^2} + \frac{1}{c^2} \right] \Rightarrow c = \frac{1}{\sqrt{\left(\frac{2f_r}{v_0}\right)^2 - \left(\frac{1}{a}\right)^2}} = \frac{1}{\sqrt{\left[\frac{2(10)}{30}\right]^2 - \left(\frac{1}{2}\right)^2}} = 2.2678 \text{ cm}$$

$$8.20 \quad (f_r)_{101} = \frac{1}{2\sqrt{\mu_0\epsilon_r}} \sqrt{\frac{1}{a^2} + \frac{1}{c^2}} \stackrel{c=a}{=} \frac{1}{\sqrt{2\sqrt{\mu_0\epsilon_r} a}} = \frac{v_0}{\sqrt{2}a} \Rightarrow a = \frac{v_0}{\sqrt{2}(f_r)_{101}}$$

a. $a = \frac{v_0/\sqrt{\epsilon_r}}{\sqrt{2}(10^9)} = \frac{30 \times 10^9}{\sqrt{2} \times 10^9} = 21.216 \text{ cm} = 0.21216 \text{ m} = c, b = a/2 = 0.10608 \text{ m}$

b. $a = \frac{v_0/\sqrt{\epsilon_r}}{\sqrt{2}(10^9)} = \frac{30 \times 10^9 / 1.6}{\sqrt{2}(10^9)} = 13.26 \text{ cm} = 0.1326 \text{ m} = c, b = a/2 = 0.0663 \text{ m}$

8.21 For a PMC resonator the:

TE^z modes are analogous to the TM^z modes of a PEC resonator

TM^z modes are analogous to the TE^z modes of a PEC resonator

c. TE^z PMC (Section 8.3.2): $\beta_x = \left(\frac{m\pi}{a}\right), \beta_y = \left(\frac{n\pi}{b}\right), \beta_z = \left(\frac{p\pi}{c}\right)$ $m=1, 2, 3, \dots$
 $n=1, 2, 3, \dots$
 $p=0, 1, 2, \dots$

b. TM^z PMC (Section 8.3.1): $\beta_x = \left(\frac{m\pi}{a}\right), \beta_y = \left(\frac{n\pi}{b}\right), \beta_z = \left(\frac{p\pi}{c}\right)$ $m=0, 1, 2, \dots$
 $n=0, 1, 2, \dots$
 $p=1, 2, 3, \dots$
 $m=n \neq 0$

c. TE^z PMC: $(f_r)_{mnp}^{\text{TE}} = \frac{1}{2\pi\sqrt{\mu_0\epsilon_r}} \sqrt{\left(\frac{m\pi}{a}\right)^2 + \left(\frac{n\pi}{b}\right)^2 + \left(\frac{p\pi}{c}\right)^2}$ $m=1, 2, 3, \dots$
 $n=1, 2, 3, \dots$
 $p=0, 1, 2, \dots$

TM^z PMC: $(f_r)_{mnp}^{\text{TM}} = \frac{1}{2\pi\sqrt{\mu_0\epsilon_r}} \sqrt{\left(\frac{m\pi}{a}\right)^2 + \left(\frac{n\pi}{b}\right)^2 + \left(\frac{p\pi}{c}\right)^2}$ $m=0, 1, 2, \dots$
 $n=0, 1, 2, \dots$
 $p=1, 2, 3, \dots$
 $m=n \neq 0$

d. For $a/b=2, c/b=4$:

Dominant Mode: TM_{101} $\Rightarrow (f_r)_{101}^{\text{TM}} = \frac{1}{2\pi\sqrt{\mu_0\epsilon_r}} \sqrt{\left(\frac{\pi}{a}\right)^2 + \left(\frac{\pi}{c}\right)^2}$

$$(f_r)_{101}^{\text{TM}} = \frac{30 \times 10^9}{2\sqrt{81}} \sqrt{\left(\frac{1}{2}\right)^2 + \left(\frac{1}{4}\right)^2} = \frac{30 \times 10^9}{2(9)} (1.118) = \boxed{1.86339 \text{ GHz}}$$

Next Mode: TM_{102}

$$(f_r)_{102}^{\text{TM}} = \frac{30 \times 10^9}{2\sqrt{81}} \sqrt{\left(\frac{1}{2}\right)^2 + \left(\frac{2}{4}\right)^2} = \frac{30 \times 10^9}{2(9)} \sqrt{\left(\frac{1}{2}\right)^2 + \left(\frac{1}{2}\right)^2} = \frac{30 \times 10^9}{2(9)} \sqrt{2}$$

$$\boxed{(f_r)_{102}^{\text{TM}} = 2.357 \text{ GHz}}$$

B.22 a. Since there are no x or y variations, then the electric field must be represented as a standing wave. Thus

$$\mathbf{E}_y = [C_1 \cos(\beta_z z) + D_1 \sin(\beta_z z)]$$

$$E_y(z=0) = 0 = C_1(1) + D_1(0) \Rightarrow C_1 = 0$$

$$E_y(z=d) = 0 = D_1 \sin(\beta_z d) = 0 \Rightarrow \beta_z d = \sin^{-1}(0) = n\pi, \beta_z = \frac{n\pi}{d}, n=1,2,3,\dots$$

$$\text{Thus } E_y = D_1 \sin(\beta_z z) = E_0 \sin(\beta_z z), \beta_x^2 + \beta_y^2 + \beta_z^2 = \beta^2 \Rightarrow \beta_z = \beta = \omega \sqrt{\mu \epsilon}$$

$$E_y = E_0 \sin(\beta z) \Rightarrow H = \frac{1}{j \omega \mu} \nabla \times \mathbf{E} = -\frac{1}{j \omega \mu} \left[\hat{a}_x \frac{\partial E_y}{\partial z} \right] = -j \frac{\omega \epsilon}{\omega \mu} \hat{a}_x \cos(\beta z)$$

$$H_x = -j \frac{\omega \sqrt{\mu \epsilon} E_0}{\omega \mu} \cos(\beta z) = -j \sqrt{\frac{\epsilon}{\mu}} \cos(\beta z)$$

b. From Part a

$$\beta_z = \frac{n\pi}{d} \Rightarrow d = \frac{n\pi}{\beta_z} = \frac{n\pi}{\beta} = \frac{n\pi}{\frac{2\pi}{\lambda}} = \frac{n\lambda}{2}, n=1,2,3,\dots$$

$$\begin{aligned} c. W = 2W_e &= 2 \left[\frac{\epsilon_0}{4} \iiint |\mathbf{E}|^2 dz \right] = \frac{\epsilon_0}{2} \int \int \int |\mathbf{E}_0|^2 \sin^2(\beta z) dz dy dx \\ &= \frac{\epsilon_0 ab}{2} \int \int \int |\mathbf{E}_0|^2 \sin^2(\beta z) dz = \frac{ab \epsilon_0 |\mathbf{E}_0|^2}{4} \int \int [1 - \cos(2\beta z)] dz \end{aligned}$$

$$W = 2W_e = \frac{ab \epsilon_0 |\mathbf{E}_0|^2}{4} \left[z - \frac{1}{2\beta} \sin(2\beta z) \right]_0^d = \frac{ab \epsilon_0 d |\mathbf{E}_0|^2}{4} = \frac{\epsilon_0 ab d |\mathbf{E}_0|^2}{4}$$

$$P_d = 2 \left\{ R_m \int \int \mathbf{J}_s \cdot \mathbf{J}_s^* dx dy \right\}, \mathbf{J}_s = \hat{a}_z \times \mathbf{H} = \hat{a}_z \times \left[\hat{a}_x j \frac{\epsilon_0}{\eta} \omega(\beta z) \right]$$

$$P_d = R_m \left| \frac{|\mathbf{E}_0|^2}{\eta^2} \right| \int \int dx dy = R_m ab \left| \frac{|\mathbf{E}_0|^2}{\eta^2} \right|$$

$$Q = \frac{W}{P_d} = \frac{W}{R_m ab \left| \frac{|\mathbf{E}_0|^2}{\eta^2} \right|} = \frac{W \epsilon_0 d \eta^2}{4 R_m}$$

$$d = 5\lambda = 5 \left(\frac{30 \times 10^9}{60 \times 10^9} \right) = 2.5 \text{ cm} = 2.5 \times 10^{-2} \text{ m}$$

$$Q = \frac{W \epsilon_0 d \eta^2}{4 R_m} = \frac{2\pi \times 60 \times 10^9 (10^9 / 36\pi) (2.5 \times 10^{-2}) (377)^2}{4 (6.4127 \times 10^{-2})} = 46,237.296$$

$$Q \approx 46,237$$

$$d = 10\lambda$$

$$Q = 2 (46,237.296) = 92,474.592 \approx 92,475$$

8.23 a. $\beta_{x0} = \beta_{xd} = \left(\frac{m\pi}{a}\right) \Big|_{m=1} = \frac{\pi}{a} = \frac{\pi}{2.286} = 1.3743 \text{ Rad/cm}$

b. Since $\epsilon_d \approx \epsilon_0$ and $\mu_d = \mu_0$, then according to (8-133)

$$\beta_{y0} \approx \beta_{yd} \approx 0$$

c. According to :

$$(f_c)_0^0 = \frac{1}{2a\sqrt{\mu_0\epsilon_0}} = \frac{30 \times 10^9}{2(2.286)} = 6.56 \times 10^9 \quad \text{If filled with air}$$

$$(f_c)_0^d = \frac{1}{2a\sqrt{\mu_d\epsilon_d}} = \frac{6.56}{\sqrt{1.1}} = 6.25 \times 10^9 \quad \text{If filled with styrofoam}$$

Therefore the cutoff frequency of the partially filled should be

$$6.25 \times 10^9 < (f_c)_0 < 6.56 \times 10^9$$

According to (8-135)

$$(f_c)_0 = \frac{1}{2a\sqrt{\mu_r\epsilon_r\epsilon_0}} \sqrt{\frac{h + \epsilon_r(b-h)}{(b-h) + \mu_r h}} = \frac{30 \times 10^9}{2(2.286)\sqrt{1.1}} \sqrt{\frac{0.254+1.1(1.016-0.254)}{1.016-0.254+0.254}}$$

$$(f_c)_0 = 6.2563 (1.0368) \times 10^9 = 6.487 \times 10^9$$

$$\text{d. } \beta_{x0}^2 + \beta_{y0}^2 + \beta_z^2 = \beta_0^2 \Rightarrow \beta_z = \sqrt{\beta_0^2 - \beta_{x0}^2} = \sqrt{\frac{[2\pi(1.25)(6.487 \times 10^9)]^2}{30 \times 10^9} - (1.3743)^2} \\ = \sqrt{(1.6982)^2 - (1.3743)^2} = 0.9976 \text{ Rad/cm}$$

8.24

TE^x(LSE^x)

$$F_x = [C_1 \cos(\beta_x x) + D_1 \sin(\beta_x x)][C_2 \cos(\beta_y y) + D_2 \sin(\beta_y y)] e^{-j\beta_z z}$$

$$\begin{aligned} E_x &= 0 & H_x &= -j \frac{1}{\omega \mu \epsilon} \left(\frac{\partial^2}{\partial x^2} + \beta^2 \right) F_x \\ E_y &= -\frac{1}{\epsilon} \frac{\partial F_x}{\partial z} & H_y &= -j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 F_x}{\partial x \partial y} \\ E_z &= \frac{1}{\epsilon} \frac{\partial F_x}{\partial y} & H_z &= -j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 F_x}{\partial x \partial z} \end{aligned} \quad \left. \begin{array}{l} \text{B.C.'s} \\ E_z(x=0) = E_z(x=a) = 0 \\ E_z(y=0) = E_z(y=b) = 0 \end{array} \right\}$$

$$E_z = \frac{\theta_4}{\epsilon} [C_1 \cos \beta_x x + D_1 \sin \beta_x x] [-C_2 \sin \beta_y y + D_2 \cos \beta_y y] e^{-j\beta_z z}$$

$$E_z(x=0) = 0 \Rightarrow C_1 = 0; \quad E_x(x=0) = 0 \Rightarrow \sin(\beta_x a) = 0 \Rightarrow \beta_x = \left(\frac{m\pi}{a}\right), m = 1, 2, 3, \dots$$

$$E_z(y=0) = 0 \Rightarrow D_2 = 0; \quad E_z(y=b) = 0 \Rightarrow \sin(\beta_y b) = 0 \Rightarrow \beta_y = \left(\frac{n\pi}{b}\right), n = 0, 1, 2, \dots$$

$$F_x = A_{mn} \sin(\beta_x x) \cos(\beta_y y) e^{-j\beta_z z}, \quad \beta_x = \frac{m\pi}{a}, m = 1, 2, 3, \dots; \quad \beta_y = \frac{n\pi}{b}, n = 0, 1, 2, \dots$$

cont'd.

$$8.24 \text{ Cont'd.} \quad (f_c)_{mn} = \frac{1}{2\pi\sqrt{\mu\epsilon}} \sqrt{\left(\frac{m\pi}{a}\right)^2 + \left(\frac{n\pi}{b}\right)^2}, \quad m=1,2,3,\dots \\ n=0,1,2,\dots$$

$$\begin{aligned} E_x &= 0 & H_x &= -j A_{mn} \frac{\beta^2 - \beta_x^2}{\omega\mu\epsilon} \sin(\beta_x x) \cos(\beta_y y) e^{-j\beta_z z} \\ E_y &= j A_{mn} \frac{\beta_x}{\epsilon} \sin(\beta_x x) \cos(\beta_y y) e^{-j\beta_z z} & H_y &= j A_{mn} \frac{\beta_x \beta_y}{\omega\mu\epsilon} \cos(\beta_x x) \sin(\beta_y y) e^{-j\beta_z z} \\ E_z &= -A_{mn} \frac{\beta_y}{\epsilon} \sin(\beta_x x) \sin(\beta_y y) e^{-j\beta_z z} & H_z &= -A_{mn} \frac{\beta_x \beta_y}{\omega\mu\epsilon} \cos(\beta_x x) \cos(\beta_y y) e^{-j\beta_z z} \end{aligned}$$

Dominant mode is the TE_{z0}^X ($f_{c10} = \frac{1}{2a\sqrt{\mu\epsilon}}$)

TM^X (LSM^X)

$$A_x = [C_1 \cos(\beta_x x) + D_1 \sin(\beta_x x)] [C_2 \cos(\beta_y y) + D_2 \sin(\beta_y y)] e^{-j\beta_z z}$$

$$E_x = -j \frac{1}{\omega\mu\epsilon} \left(\frac{\partial^2}{\partial x^2} + \beta^2 \right) A_x \quad H_x = 0 \quad \left. \begin{array}{l} \\ \end{array} \right\} \text{B.C.'s}$$

$$E_y = -j \frac{1}{\omega\mu\epsilon} \frac{\partial^2 A_x}{\partial x \partial y} \quad H_y = \frac{1}{\mu} \frac{\partial A_x}{\partial y} \quad \left. \begin{array}{l} \\ \end{array} \right\} E_z(x=0) = E_z(x=a) = 0$$

$$E_z = -j \frac{1}{\omega\mu\epsilon} \frac{\partial^2 A_x}{\partial x \partial z} \quad H_z = -\frac{1}{\mu} \frac{\partial A_x}{\partial z} \quad \left. \begin{array}{l} \\ \end{array} \right\} E_z(y=0) = E_z(y=b) = 0$$

$$E_z = -\frac{\beta_x \beta_z}{\omega\mu\epsilon} [-C_1 \sin(\beta_x x) + D_1 \cos(\beta_x x)] [C_2 \cos(\beta_y y) + D_2 \sin(\beta_y y)] e^{-j\beta_z z}$$

$$E_z(x=0) = 0 \Rightarrow D_1 = 0; \quad E_z(x=a) = 0 \Rightarrow \sin(\beta_x a) = 0 \Rightarrow \beta_x = \frac{m\pi}{a}, \quad m=0,1,2,\dots$$

$$E_z(y=0) = 0 \Rightarrow C_2 = 0; \quad E_z(y=b) = 0 \Rightarrow \sin(\beta_y b) = 0 \Rightarrow \beta_y = \frac{n\pi}{b}, \quad n=1,2,3,\dots$$

$$A_x = B_{mn} \cos(\beta_x x) \sin(\beta_y y) e^{-j\beta_z z}, \quad \beta_x = \frac{m\pi}{a}, \quad m=0,1,2,\dots; \quad \beta_y = \frac{n\pi}{b}, \quad n=1,2,3,\dots$$

$$(f_c)_{mn} = \frac{1}{2\pi\sqrt{\mu\epsilon}} \sqrt{\left(\frac{m\pi}{a}\right)^2 + \left(\frac{n\pi}{b}\right)^2}, \quad m=0,1,2,\dots \\ n=1,2,3,\dots$$

$$E_x = -j B_{mn} \frac{\beta^2 - \beta_x^2}{\omega\mu\epsilon} \cos(\beta_x x) \sin(\beta_y y) e^{-j\beta_z z}, \quad H_x = 0$$

$$E_y = j B_{mn} \frac{\beta_x \beta_y}{\omega\mu\epsilon} \sin(\beta_x x) \cos(\beta_y y) e^{-j\beta_z z}, \quad H_y = -j B_{mn} \frac{\beta_x}{\mu} \cos(\beta_x x) \sin(\beta_y y) e^{-j\beta_z z}$$

$$E_z = B_{mn} \frac{\beta_x \beta_y}{\omega\mu\epsilon} \sin(\beta_x x) \sin(\beta_y y) e^{-j\beta_z z}, \quad H_z = -B_{mn} \frac{\beta_y}{\mu} \cos(\beta_x x) \cos(\beta_y y) e^{-j\beta_z z}$$

The dominant mode is the TM_{01}^X ($f_{c01} = \frac{1}{2b\sqrt{\mu\epsilon}}$)

8.25

B.C.'s:

$$E_z^d(0 \leq x \leq w, y=0, z) = E_z^d(0 \leq x \leq w, y=b, z) = 0$$

$$E_z^o(x=a, 0 \leq y \leq b, z) = 0$$

$$E_z^d(x=w, 0 \leq y \leq b, z) = E_z^o(x=w, 0 \leq y \leq b, z)$$

$$E_z^o(w \leq x \leq a, y=0, z) = E_z^o(w \leq x \leq a, y=b, z) = 0$$

$$E_z^d(x=a, 0 \leq y \leq b, z) = 0$$

$$H_z^d(x=w, 0 \leq y \leq b, z) = H_z^o(x=w, 0 \leq y \leq b, z)$$

TE^x:

$$F_x^d = [C_1^d \cos(\beta_{xd}x) + D_1^d \sin(\beta_{xd}x)] [C_2^d \cos(\beta_{yd}y) + D_2^d \sin(\beta_{yd}y)] A_3^d e^{-j\beta_z z} \quad (1)$$

$$\beta_{xd}^2 + \beta_{yd}^2 + \beta_z^2 = \beta_d^2 = w^2 \mu_d \epsilon_d$$

$$F_x^o = \{C_1^o \cos[\beta_{xo}(a-x)] + D_1^o \sin[\beta_{xo}(a-x)]\} \{C_2^o \cos(\beta_{yo}y) + D_2^o \sin(\beta_{yo}y)\} A_3^o e^{-j\beta_z^o z}$$

By applying the boundary conditions we reduce (1) and (2) to

$$F_x^d = A_{mn}^d \sin(\beta_{xd}x) \cos(\beta_{yd}y) e^{-j\beta_z z}$$

$$\beta_{yd} = \frac{n\pi}{b}, n = 0, 1, 2, \dots$$

$$\beta_{xd}^2 + \beta_{yd}^2 + \beta_z^2 = \beta_{xd}^2 + \left(\frac{n\pi}{b}\right)^2 + \beta_z^2 = \beta_d^2 = w^2 \mu_d \epsilon_d$$

$$F_x^o = A_{mn}^o \sin[\beta_{xo}(a-x)] \cos(\beta_{yo}y) e^{-j\beta_z^o z}$$

$$\beta_{yo} = \frac{n\pi}{b}, n = 0, 1, 2, \dots$$

$$\beta_{xo}^2 + \beta_{yo}^2 + \beta_z^2 = \beta_{xo}^2 + \left(\frac{n\pi}{b}\right)^2 + \beta_z^2 = \beta_o^2 = w^2 \mu_o \epsilon_o$$

Now apply boundary conditions along the interface. Doing this leads to

$$\frac{1}{\epsilon_d} A_{mn}^d \sin[\beta_{xo}(a-w)] = \frac{1}{\epsilon_d} A_{mn}^d \sin(\beta_{xd}w) \quad (3)$$

$$\frac{\beta_{xo}}{\mu_o \epsilon_o} A_{mn}^o \cos[\beta_{xo}(a-w)] = -\frac{\beta_{xd}}{\mu_d \epsilon_d} A_{mn}^d \cos(\beta_{xd}w) \quad (4)$$

Dividing (4) by (3) leads to

$$\frac{\beta_{xo}}{\mu_o} \cot[\beta_{xo}(a-w)] = -\frac{\beta_{xd}}{\mu_d} \cot(\beta_{xd}w)$$

$$\begin{aligned} \beta_{xo}^2 + \beta_{yo}^2 + \beta_z^2 &= \beta_{xo}^2 + \left(\frac{n\pi}{b}\right)^2 + \beta_z^2 = \beta_o^2 = w^2 \mu_o \epsilon_o \\ \beta_{xd}^2 + \beta_{yd}^2 + \beta_z^2 &= \beta_{xd}^2 + \left(\frac{n\pi}{b}\right)^2 + \beta_z^2 = \beta_d^2 = w^2 \mu_d \epsilon_d \end{aligned} \quad \left. \right\} n = 0, 1, 2, \dots$$

The dominant mode is the TE₁₀^x with a cutoff frequency value of

$$\frac{1}{2a\sqrt{\mu_d \epsilon_d}} \leq (f_c)_{10} \leq \frac{1}{2a\sqrt{\mu_o \epsilon_o}}$$

cont'd.

8.25 cont'd.

TM^x:

$$A_x^d = [C_1^d \cos(\beta_{xd}x) + D_1^d \sin(\beta_{xd}x)] [C_2^d \cos(\beta_{yd}y) + D_2^d \sin(\beta_{yd}y)] A_3^d e^{-j\beta_z z} \quad (1)$$

$$\beta_{xd}^2 + \beta_{yd}^2 + \beta_z^2 = \beta_d^2 = \omega^2 \mu_d \epsilon_d$$

$$A_x^o = \{C_1^o \cos[\beta_{xo}(a-x)] + D_1^o \sin[\beta_{xo}(a-x)]\} \{C_2^o \cos(\beta_{yo}y) + D_2^o \sin(\beta_{yo}y)\} A_3^o e^{-j\beta_z z} \quad (2)$$

$$\beta_{xo}^2 + \beta_{yo}^2 + \beta_z^2 = \beta_o^2 = \omega^2 \mu_o \epsilon_o$$

By applying the boundary conditions along the walls of the waveguide we reduce (1) and (2) to

$$A_x^d = A_{mn} \cos(\beta_{xd}x) \sin(\beta_{yd}y) e^{-j\beta_z z}$$

$$\beta_{yd} = \frac{n\pi}{b}, n = 1, 2, 3, \dots$$

$$\beta_{xd}^2 + \beta_{yd}^2 + \beta_z^2 = \beta_{xd}^2 + \left(\frac{n\pi}{b}\right)^2 + \beta_z^2 = \beta_d^2 = \omega^2 \mu_d \epsilon_d$$

$$A_x^o = A_{mn} \cos[\beta_{xo}(a-x)] \sin(\beta_{yo}y) e^{-j\beta_z z}$$

$$\beta_{yo} = \frac{n\pi}{b}, n = 1, 2, 3, \dots$$

$$\beta_{xo}^2 + \beta_{yo}^2 + \beta_z^2 = \beta_{xo}^2 + \left(\frac{n\pi}{b}\right)^2 + \beta_z^2 = \beta_o^2 = \omega^2 \mu_o \epsilon_o$$

Now applying the boundary conditions along the interface leads to

$$\frac{\beta_{xd}}{\mu_d \epsilon_d} A_{mn}^d \sin(\beta_{xd}w) = - \frac{\beta_{xo}}{\mu_o \epsilon_o} A_{mn}^o \sin[\beta_{xo}(a-w)] \quad (3)$$

$$\frac{1}{\mu_d} A_{mn}^d \cos(\beta_{xd}w) = \frac{1}{\mu_o} A_{mn}^o \cos[\beta_{xo}(a-w)] \quad (4)$$

Dividing (3) by (4) leads to

$$\frac{\beta_{xo}}{\epsilon_o} \tan[\beta_{xo}(a-w)] = - \frac{\beta_{xd}}{\epsilon_d} \tan(\beta_{xd}w)$$

$$\left. \begin{aligned} \beta_{xo}^2 + \beta_{yo}^2 + \beta_z^2 &= \beta_{xo}^2 + \left(\frac{n\pi}{b}\right)^2 + \beta_z^2 = \beta_o^2 = \omega^2 \mu_o \epsilon_o \\ \beta_{xd}^2 + \beta_{yd}^2 + \beta_z^2 &= \beta_{xd}^2 + \left(\frac{n\pi}{b}\right)^2 + \beta_z^2 = \beta_d^2 = \omega^2 \mu_d \epsilon_d \end{aligned} \right\} n = 1, 2, 3, \dots$$

The dominant mode is the TM₀₁^x with a cutoff frequency value of

$$\frac{1}{2b\sqrt{\mu_d \epsilon_d}} \leq (f_c)_{01}^{TM^x} \leq \frac{1}{2b\sqrt{\mu_o \epsilon_o}}$$

8.26

$$1. (f_c)_{mn} = \frac{4}{2\pi\sqrt{\mu_0\epsilon_0}} \sqrt{\left(\frac{m\pi}{a}\right)^2 + \left(\frac{n\pi}{b}\right)^2} \quad \left\{ \begin{array}{l} TE^z \\ TM^z \end{array} \right\} \begin{cases} m=0,1,2,\dots \\ n=0,1,2,\dots \\ m=n \neq 0 \end{cases}$$

$$\left\{ \begin{array}{l} TE^z \\ TM^z \end{array} \right\} \begin{cases} m=1,2,3,\dots \\ n=1,2,3,\dots \end{cases}$$

a. For $b > a$ ($m=0, n=1$)

$$f_c(TE_{01}^z) = \frac{1}{2\pi\sqrt{\mu_0\epsilon_0}} \left(\frac{\pi}{b}\right) = \frac{1}{2b\sqrt{\mu_0\epsilon_0}} = \frac{30 \times 10^9}{2(2.286)} = 6.5617 \text{ GHz}$$

b. For $b > a$ ($m=0, n=1$)

$$f_c(TE_{01}^y) = \frac{1}{2\pi\sqrt{\mu_0\epsilon_0}} \left(\frac{\pi}{b}\right) = \frac{1}{2b\sqrt{\mu_0\epsilon_0}} \quad \left\{ \begin{array}{l} TE^y \\ TM^y \end{array} \right\} \begin{cases} m=0,1,2,\dots \\ n=1,2,3,\dots \\ m=n \neq 0 \end{cases}$$

$$f_c(TE_{01}^y) = \frac{30 \times 10^9}{2(2.286)} = 6.5617 \text{ GHz}$$

c. For $b > a$ ($m=0, n=2$)

$$f_c(TE_{02}^z) = \frac{1}{2\pi\sqrt{\mu_0\epsilon_0}} \left(\frac{2\pi}{b}\right) = \frac{1}{b\sqrt{\mu_0\epsilon_0}} = 2(6.5617) \text{ GHz}$$

$$f_c(TE_{02}^z) = 13.1234 \text{ GHz}$$

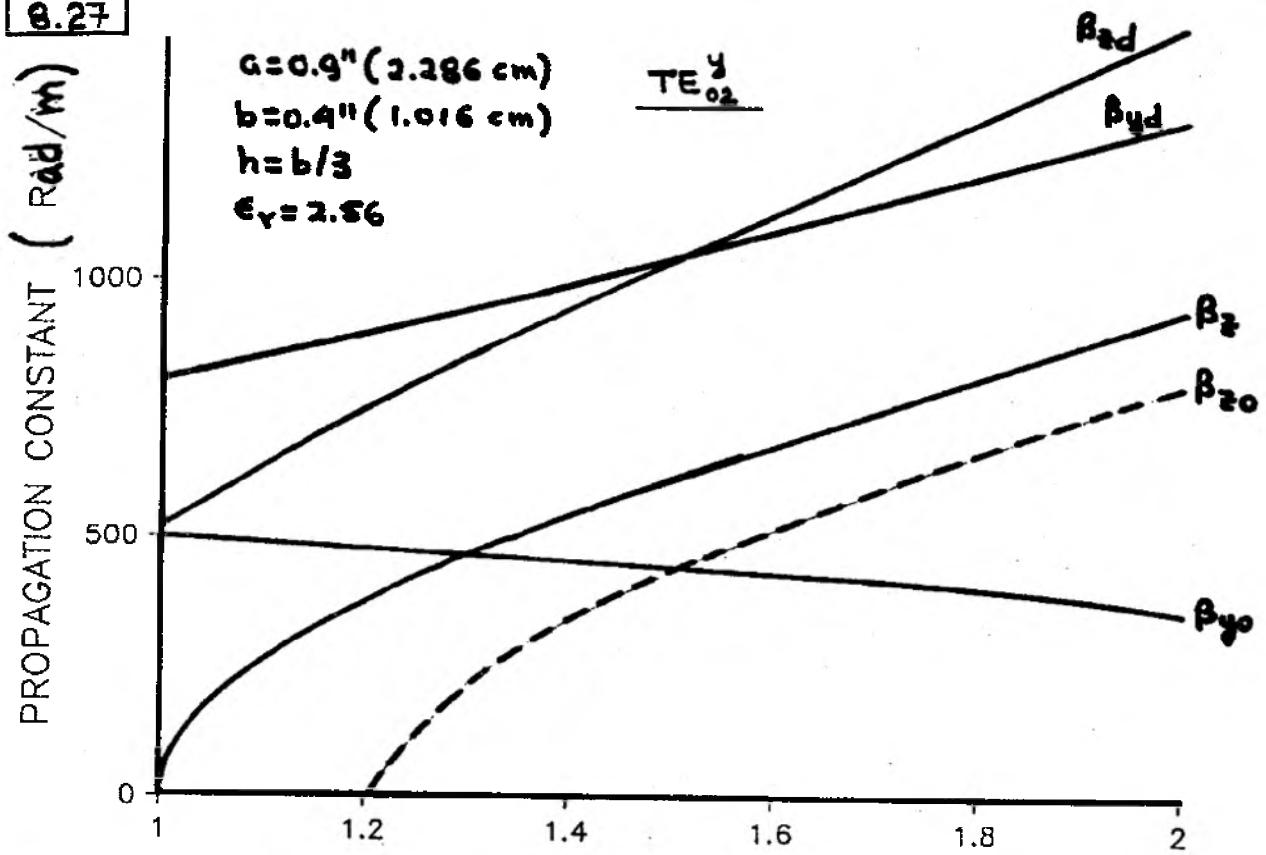
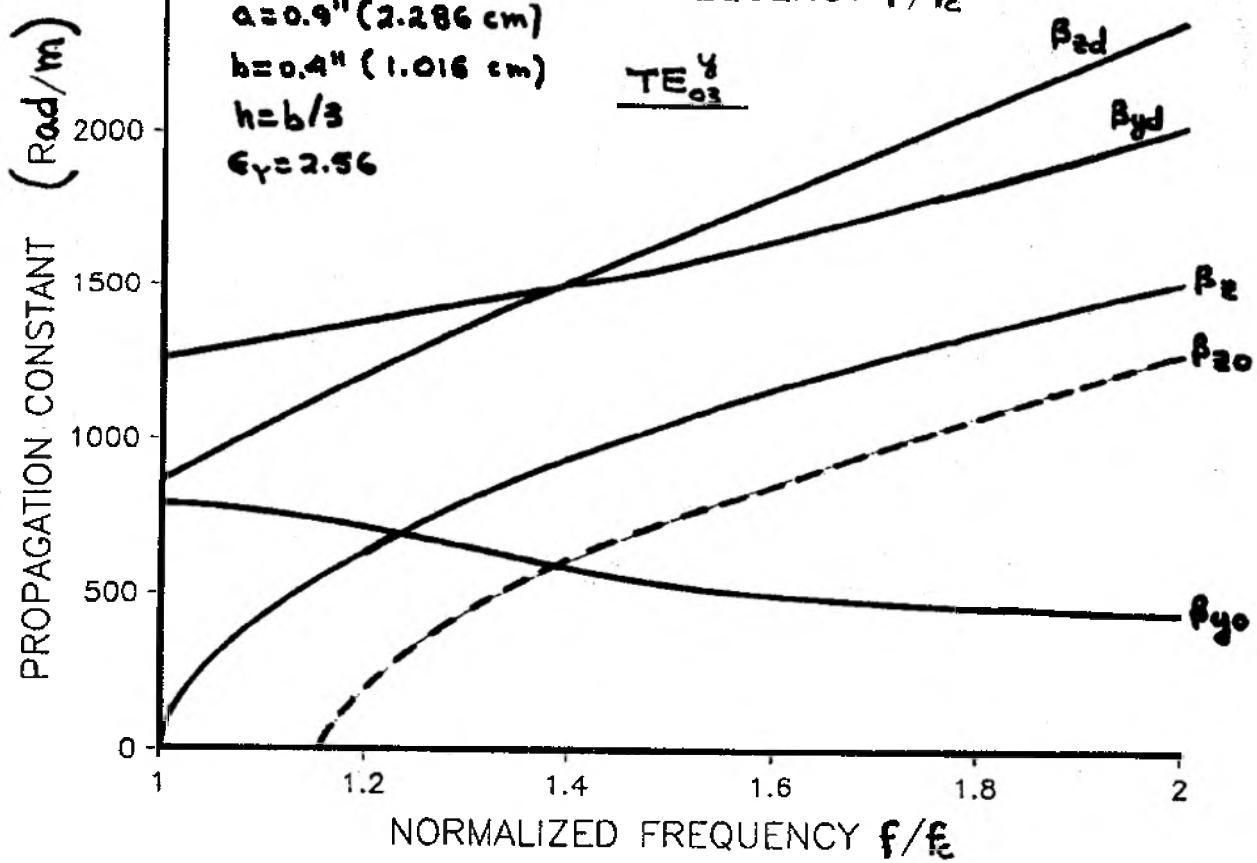
$$d. (f_c)_{02}^{TE^2} = \frac{1}{2\pi\sqrt{\mu_0\epsilon_r\epsilon_0}} \left(\frac{\pi}{b}\right) = \frac{1}{2b\sqrt{\mu_0\epsilon_0\epsilon_r}} = 4 \times 10^9$$

$$\frac{6.5617 \times 10^9}{\sqrt{\epsilon_r}} = 4 \times 10^9$$

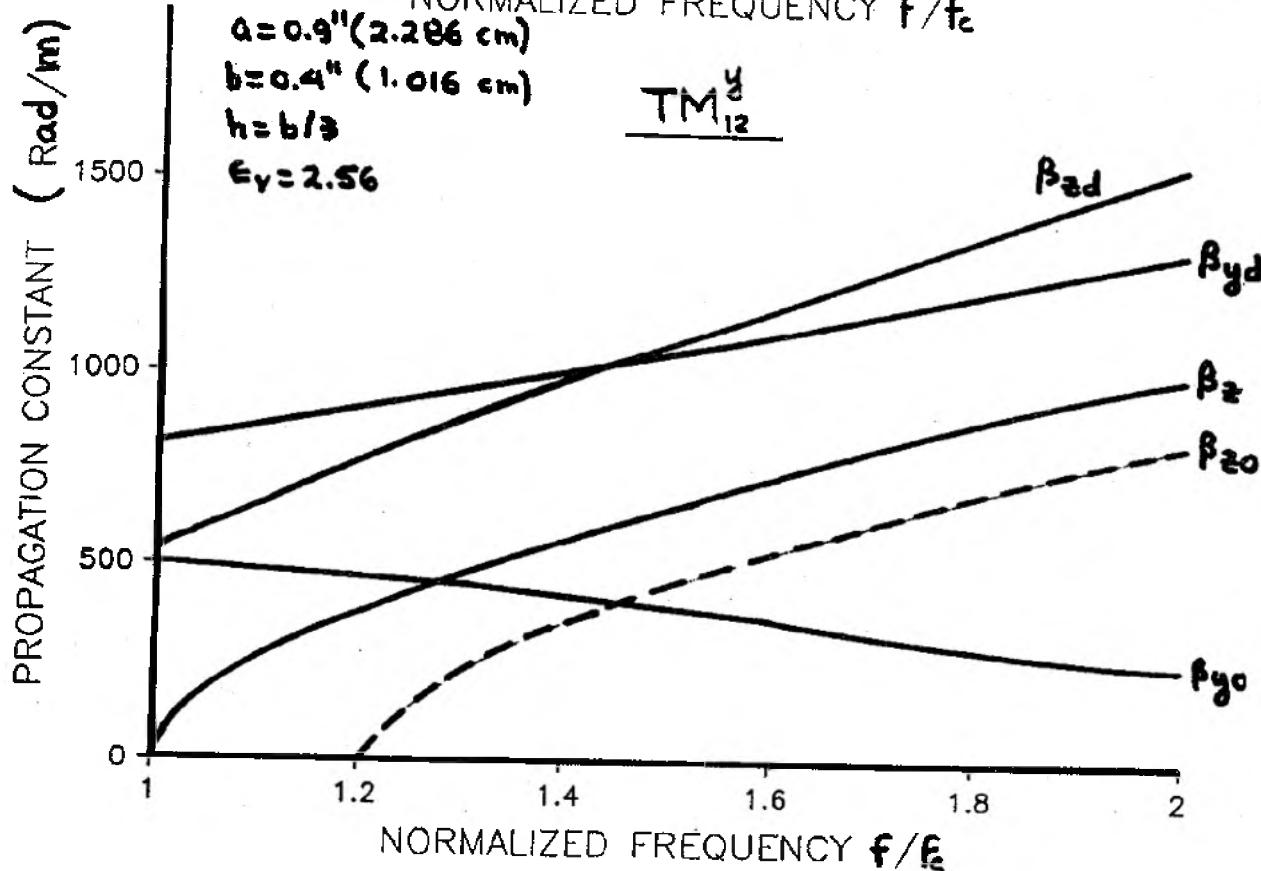
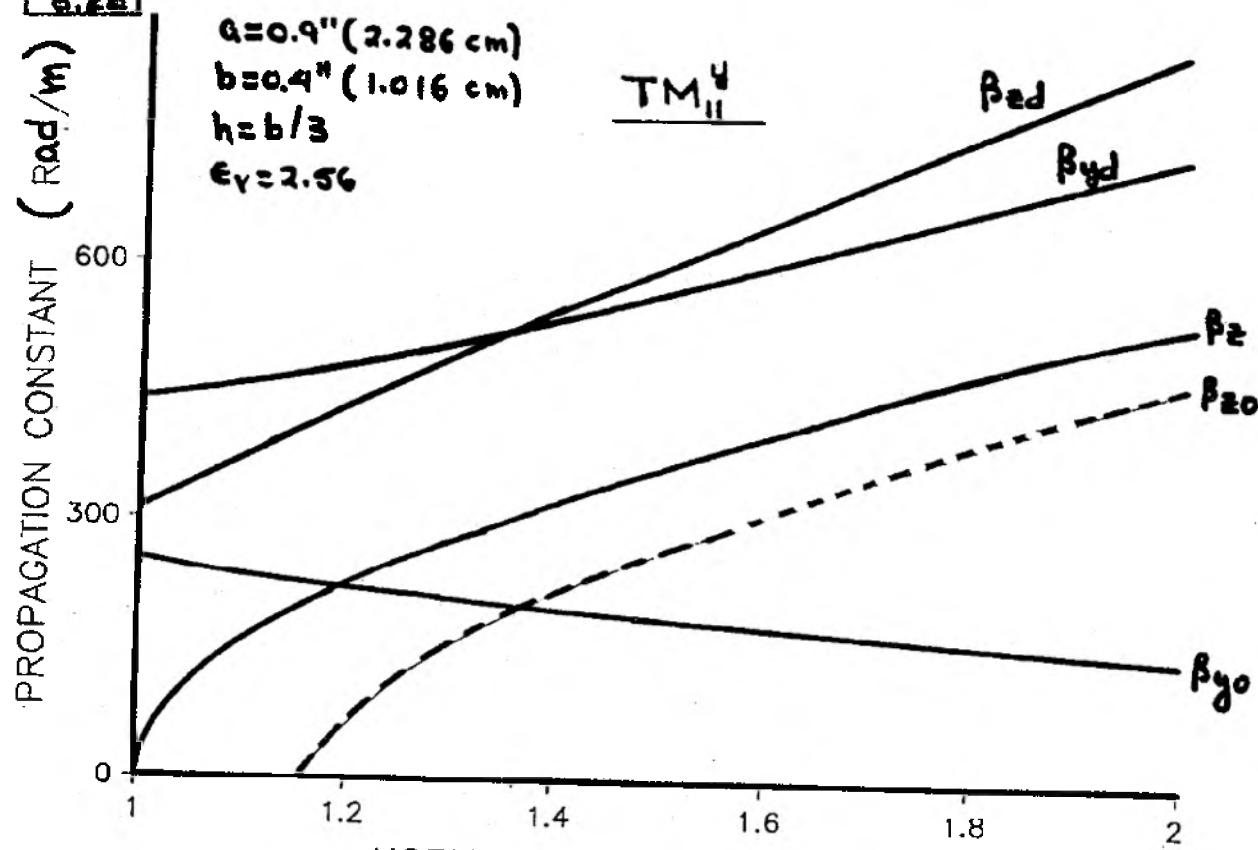
$$\sqrt{\epsilon_r} = \frac{6.5617}{4} = 1.64$$

$$\epsilon_r = 2.691$$

8.27

NORMALIZED FREQUENCY f/f_c NORMALIZED FREQUENCY f/f_c

0.28



8.29

TE^g(A) [Figure 8.14(b)]

Using the procedure outlined in Section 8.6.1 for Figure 8.13(a), we can write for Figure 8.14(b):

$$Z_{cd} = Z_d^e = \frac{w\mu_d}{\beta_{xd}}, \quad Z_{c0} = Z_0^e = \frac{w\mu_0}{\beta_{x0}}.$$

$$\beta_{td} = \beta_{xd}, \quad \beta_{to} = \beta_{x0}$$

$$\boxed{\frac{\beta_{x0}}{\mu_0} \cot[\beta_{x0}(a-w)] = - \frac{\beta_{xd}}{\mu_d} \cot(\beta_{xd}w)}$$

TM^g(E) [Figure 8.14(b)]

Using the procedure outlined in Section 8.6.2 for Figure 8.13(a), we can write for Figure 8.14(b):

$$Z_{cd} = Z_d^e = \frac{\beta_{xd}}{w\epsilon_d}, \quad Z_{c0} = Z_0^e = \frac{\beta_{x0}}{w\epsilon_0}.$$

$$\beta_{td} = \beta_{xd}, \quad \beta_{to} = \beta_{x0}$$

$$\boxed{\frac{\beta_{x0}}{\epsilon_0} \tan[\beta_{x0}(a-w)] = - \frac{\beta_{xd}}{\epsilon_d} \tan(\beta_{xd}w)}$$

Cont'd

8.29 cont'd

TE^{4(A)} [Figure 8.14(b)]

Using the procedure outlined in Section 8.6.1 for Figure 8.14(a), we can write for Figure 8.14(b):

$$Z_{cd} = Z_d^e = \frac{w\mu_d}{\beta_{xd}}, \quad Z_{c0} = Z_0^e = \frac{w\mu_0}{\beta_{x0}}.$$

$$\beta_{td} = \beta_{xd}, \quad \beta_{to} = \beta_{x0}$$

$$\boxed{\frac{\beta_{x0}}{\epsilon_0} \cot[\beta_{x0}(a-w)] = -\frac{\beta_{xd}}{\epsilon_d} \cot(\beta_{xd}w)}$$

TM^{4(E)} [Figure 8.14(b)]

Using the procedure outlined in Section 8.6.2 for Figure 8.14(a), we can write for Figure 8.14(b):

$$Z_{cd} = Z_d^e = \frac{\beta_{xd}}{w\epsilon_d}, \quad Z_{c0} = Z_0^e = \frac{\beta_{x0}}{w\epsilon_0}$$

$$\beta_{td} = \beta_{xd}, \quad \beta_{to} = \beta_{x0}$$

$$\boxed{\frac{\beta_{x0}}{\epsilon_0} \tan[\beta_{x0}(a-w)] = -\frac{\beta_{xd}}{\epsilon_d} \tan(\beta_{xd}w)}$$

8.30

$$(f_c)^{TE^y} = \frac{1}{2\pi\sqrt{\mu\epsilon}} \sqrt{\left(\frac{m\pi}{a}\right)^2 + \left(\frac{n\pi}{b}\right)^2} \quad \begin{cases} m=0,1,2,\dots \\ n=1,2,3,\dots \end{cases} \quad (8-100a)$$

a. TE_{01}^y Mode

b. 1. TE_{01}^y : Air-filled: $(f_c)^{TE^y} = \frac{1}{2\pi\sqrt{\mu_0\epsilon_0}} \left(\frac{\pi}{b}\right) = \frac{1}{2b\sqrt{\mu_0\epsilon_0}}$

2. TE_{01}^y : Dielectric-filled: $(f_c)^{TE^y} = \frac{1}{2\pi\sqrt{\mu_0\epsilon_0}} \left(\frac{\pi}{b}\right) = \frac{1}{2b\sqrt{\mu_0\epsilon_0}}$

3. TE_{01}^y : Partially-filled:

$$\sqrt{\frac{\epsilon_0}{\mu_0}} \cot \left[w_c \sqrt{\mu_0\epsilon_0} (b-h) \right] = - \sqrt{\frac{\epsilon_0}{\mu_0}} \cot \left[w_c \sqrt{\mu_0\epsilon_0} h \right] \quad (8-118)$$

$$\sqrt{\frac{\epsilon_0}{\mu_0}} \cot \left[w_c \sqrt{\mu_0\epsilon_0} (b-h) \right] = - \sqrt{\frac{4\epsilon_0}{4\mu_0}} \cot \left[w_c \sqrt{\mu_0\epsilon_0} h \right] = - \sqrt{\frac{\epsilon_0}{\mu_0}} \cot \left(w_c \sqrt{\mu_0\epsilon_0} h \right)$$

$$\cot \left[w_c \sqrt{\mu_0\epsilon_0} (b-h) \right] = - \cot \left(w_c \sqrt{4\mu_0\epsilon_0} h \right) = \cot \left[\pi - 4h w_c \sqrt{\mu_0\epsilon_0} \right]$$

$$w_c \sqrt{\mu_0\epsilon_0} (b-h) = \pi - w_c 4h \sqrt{\mu_0\epsilon_0}$$

$$w_c \sqrt{\mu_0\epsilon_0} (b-h+4h) = w_c \sqrt{\mu_0\epsilon_0} (b+3h) = \pi$$

$$2\pi f_c \sqrt{\mu_0\epsilon_0} (b+3h) = 2\pi f_c \sqrt{\mu_0\epsilon_0} (2b) = \pi$$

$$h = b/3$$

$$(f_c) = \frac{1}{4b\sqrt{\mu_0\epsilon_0}} = \frac{1}{2(b+3h)\sqrt{\mu_0\epsilon_0}}$$

c. 1. TE_{01}^y (air-filled): $f_c = \frac{1}{2b\sqrt{\mu_0\epsilon_0}} = \frac{30 \times 10^9}{2(1.016)} = 14.764 \text{ GHz}$

2. TE_{01}^y (dielectric-filled): $f_c = \frac{1}{2b\sqrt{\mu_0\epsilon_0}(4)} = \frac{14.764 \text{ GHz}}{4} = 3.691 \text{ GHz}$

3. TE_{01}^y (partially-filled): $f_c = \frac{1}{4b\sqrt{\mu_0\epsilon_0}} = \frac{30 \times 10^9}{4(1.016)} = 7.382 \text{ GHz}$

d. $3.691 \text{ GHz} < f_c (\text{partially-filled}) < 14.764 \text{ GHz} < f_c (\text{air-filled})$

$$3.691 \text{ GHz} < 7.382 \text{ GHz} < 14.764 \text{ GHz}$$

as it should be.

TM^x : $a=2.286 \text{ cm}$, $b=1.1016 \text{ cm}$, $\epsilon_r=1.1$, $\mu_r=1$, $w=a/2$

From Problem 8.24(b)

$$(f_c)_{mn} (\text{TM}^x - \text{LSM}^x) = \frac{1}{2\pi\sqrt{\mu\epsilon}} \sqrt{\left(\frac{m\pi}{a}\right)^2 + \left(\frac{n\pi}{b}\right)^2} \quad \begin{matrix} m=0, 1, 2, \dots \\ n=1, 2, 3, \dots \end{matrix}$$

$$\beta_{x0} = \left(\frac{m\pi}{a}\right), \quad \beta_{x1} = \left(\frac{n\pi}{a}\right) \quad m=0, 1, 2, \dots$$

The dominant mode is the TM_{01}^x ($m=0, n=1$) with

$$(f_c)_{01}^{\text{TM}^x} = \frac{1}{2\pi\sqrt{\mu\epsilon}} \left(\frac{\pi}{b}\right) = \frac{1}{2b\sqrt{\mu\epsilon}}$$

a. Free-Space: $f_{01}^{\text{TM}^x} = \frac{30 \times 10^9}{2(0.106)} = 13.617 \text{ GHz}$

b. Dielectric Filled: $f_{01}^{\text{TM}^x} = \frac{30 \times 10^9}{2(1.1016)\sqrt{1.1}} = 12.983 \text{ GHz}$

c. Based on Figure 8-15(b), the pertinent equations for TM^x are:

$$\frac{\beta_{x0}}{\epsilon} \tan[\beta_{x0}(a-w)] = - \frac{\beta_{xd}}{\epsilon_d} \tan(\beta_{xd}w)$$

$$\beta_{x0}^2 + \beta_{y0}^2 + \beta_z^2 = \beta_{x0}^2 + \left(\frac{n\pi}{b}\right)^2 + \beta_z^2 = \beta_0^2 = w\mu_0\epsilon_0 \quad \begin{matrix} n=1, 2, 3, \dots \\ \beta_{xd}^2 + \beta_{yd}^2 + \beta_z^2 = \beta_{xd}^2 + \left(\frac{n\pi}{b}\right)^2 + \beta_z^2 = \beta_d^2 = w\mu_d\epsilon_d \end{matrix}$$

The dominant mode is the TM_{01}^x ($m=0, n=1$) whose cutoff

frequency is $\frac{1}{2b\sqrt{\mu_d\epsilon_d}} < (f_c)_{01}^{\text{TM}^x} < \frac{1}{2b\sqrt{\mu_0\epsilon_0}}$

Because $\epsilon_r=1.1 \Rightarrow \epsilon \approx \epsilon_0 \Rightarrow \beta_{x0} \approx \beta_{xd} \approx \text{small} \Rightarrow 0$

Therefore the transcendental equation reduces, as shown similarly on page 401 of the book for the TM^{y0} , to (8-135)

which for TM^x mode can be written as

$$(f_c)_{01}^{\text{TM}^x} \approx \frac{\pi}{b\sqrt{\mu_d\epsilon_d}} \sqrt{\frac{w+\epsilon_r(a-w)}{(a-w)+\mu_r w}} \quad \begin{matrix} a=2.286 & \epsilon_r=1.1 \\ w=1.143 & \mu_r=1 \end{matrix}$$

$$(f_c)_{01}^{\text{TM}^x} = \frac{30 \times 10^9}{2(1.1016)\pi} \sqrt{\frac{1.143+1.1(1.143)}{1.143+1.143}} = 13.3035 \text{ GHz}$$

d.
$$12.983 \times 10^9 < 13.3035 \times 10^9 < 13.617 \times 10^9$$

8.32

$$2h = 1 \text{ cm}, \epsilon_r = 5, \mu_r = 1$$

a. $(f_c)_m = \frac{m}{4h\sqrt{\mu_d\epsilon_d - h^2\epsilon_0}}$ } $m = 0, 2, 4, \dots, TM_m^2(\text{odd}) \text{ and } TE_m^2(\text{odd})$
 } $m = 1, 3, 5, \dots, TM_m^2(\text{even}) \text{ and } TE_m^2(\text{even})$

$$m=0: (f_c)_0 = 0$$

$$m=1: (f_c)_1 = \frac{30 \times 10^9}{4(0.5 \times 10^{-2})\sqrt{5-1}} = 7.4949 \times 10^9$$

$$\left. \begin{array}{l} TM_0^2(\text{odd}) \\ TE_0^2(\text{odd}) \end{array} \right\} (f_c)_0 = 0$$

$$\left. \begin{array}{l} TM_1^2(\text{even}) \\ TE_1^2(\text{even}) \end{array} \right\} (f_c)_1 = 7.4949 \times 10^9$$

- b. Using a graphical solution similar to that of Figure 8-22 it leads to the following answers (radius of circle is $a = 1.6767$)

See graph next page

$$TM_0^2(\text{odd})$$

$$\alpha_{y_0}h \approx 1.04$$

$$\beta_{yd}h \approx 1.325$$

$$\alpha_{y_0} \approx 208 \text{ Np/m}$$

$$\beta_{yd} \approx 265 \text{ Rad/m}$$

$$TM_1^2(\text{even})$$

$$\alpha_{y_0}h \approx 0.04$$

$$\beta_{yd}h \approx 1.675$$

$$\alpha_{y_0} \approx 8 \text{ Np/m}$$

$$\beta_{yd} \approx 335 \text{ Rad/m}$$

$$TE_0^2(\text{odd})$$

See graph next page

$$TE_0^2(\text{odd})$$

$$\alpha_{y_0}h \approx 1.39$$

$$\beta_{yd}h \approx 0.956$$

$$\alpha_{y_0} \approx 278 \text{ Np/m}$$

$$\beta_{yd} \approx 191.2 \text{ Rad/m}$$

$$TE_1^2(\text{even})$$

$$\alpha_{y_0}h \approx 0.180$$

$$\beta_{yd}h \approx 1.663$$

$$\alpha_{y_0} \approx 36 \text{ Np/m}$$

$$\beta_{yd} \approx 332.6 \text{ Rad/m}$$

- c. Using a graphical solution similar to that of Figure 8-30 it leads to the following answers:

$$TM_0^2(\text{odd}): \theta_1 \approx 45.2^\circ$$

$$TE_0^2(\text{odd}): \theta_1 = 59.2^\circ$$

$$TM_1^2(\text{even}): \theta_1 \approx 26.7^\circ$$

$$TE_1^2(\text{even}): \theta_1 = 27.2^\circ$$

See Graph.

It follows.

- b. Alternate values obtained analytically using iteratively techniques:

$$TM_0^2(\text{odd})$$

$$TM_1^2(\text{even})$$

$$\alpha_{y_0} = 206.3335 \text{ Np/m} \quad \alpha_{y_0} = 7.0177 \text{ Np/m}$$

$$\beta_{yd} = 264.0467 \text{ Rad/m} \quad \beta_{yd} = 335.0297 \text{ Rad/m}$$

$$TE_0^2(\text{odd})$$

$$TE_1^2(\text{even})$$

$$y_0 = 274.59 \text{ Np/m} \quad \alpha_{y_0} = 32.4091 \text{ Np/m}$$

$$\beta_{yd} = 192.0794 \text{ Rad/m} \quad \beta_{yd} = 333.5323 \text{ Rad/m}$$

- c. Alternate values obtained analytically using iteratively techniques:

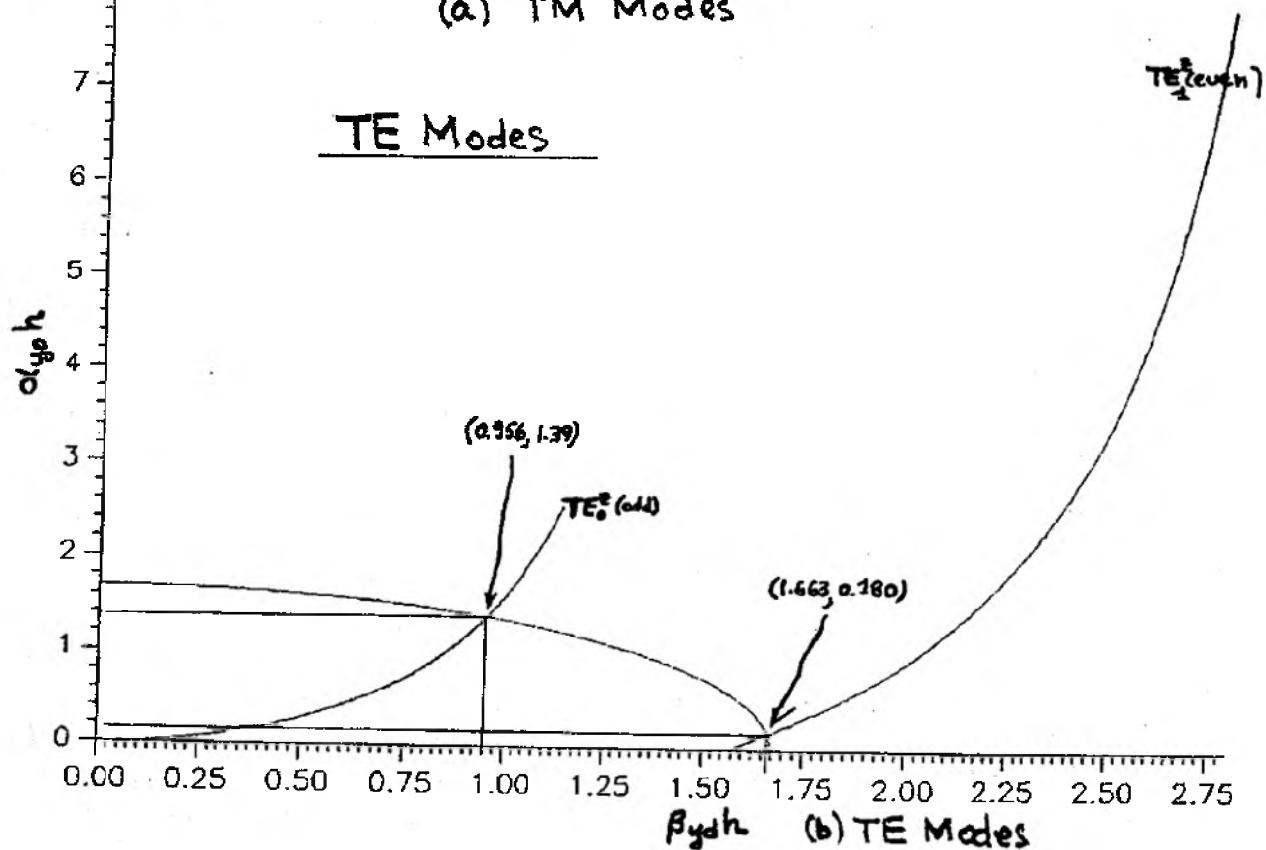
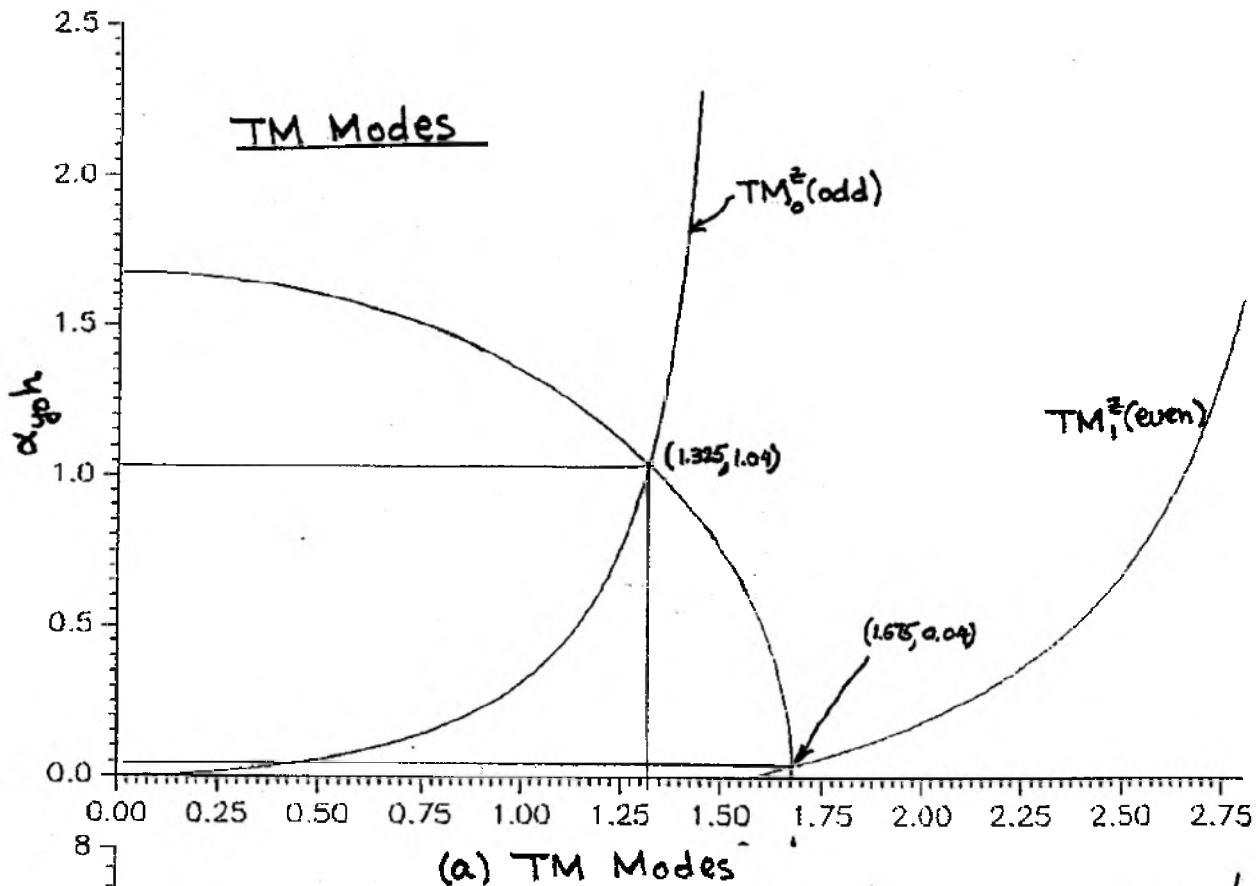
$$TM_0^2(\text{odd}): \theta_1 = 45.1877^\circ$$

$$TE_0^2(\text{odd}): \theta_1 = 59.1563^\circ$$

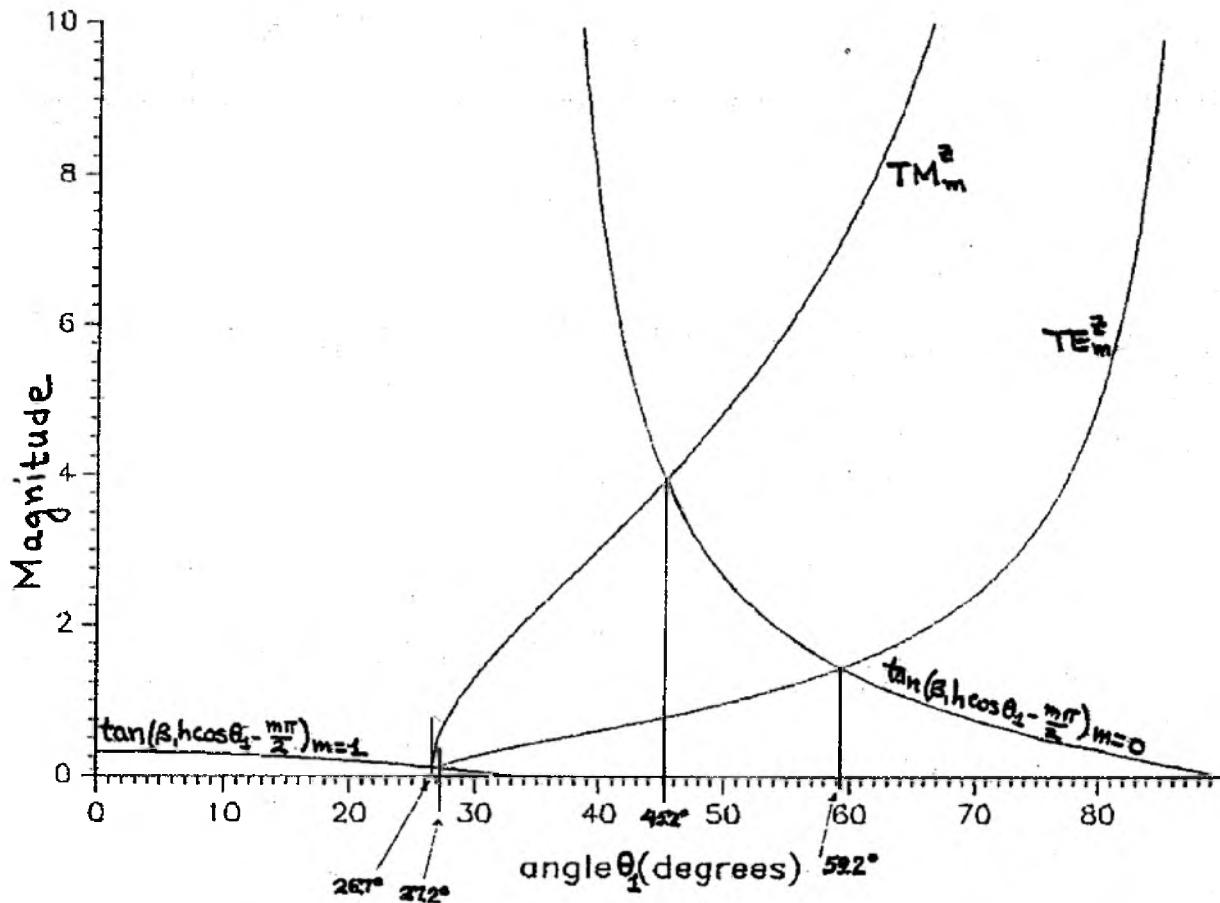
$$TM_1^2(\text{even}): \theta_1 = 26.5901^\circ$$

$$TE_1^2(\text{even}): \theta_1 = 27.0944^\circ$$

8.32 cont'd.



B.32 cont'd.



(c) Incidence angles for TM and TE modes

8.33

a. TM_0 (odd): $f_c = 0$

TE_1 (even):

$$f_c = \frac{1}{4h\sqrt{\mu_0}\sqrt{\epsilon_r - 1}} = \frac{30 \times 10^9}{4(1.25)\sqrt{4-1}} = 3.464 \times 10^9$$

b. $\theta_c = \sin^{-1} \sqrt{\frac{1}{4}} = \sin^{-1}(0.5) = 30^\circ \left\{ \begin{array}{l} \cos \theta_c = 0.867 \\ \sin \theta_c = 0.5 \end{array} \right.$

$\text{TM}_{(m=0)}$: $\tan(\beta_1 h \cos \theta_1 - \frac{m\pi}{2}) = \frac{\sqrt{\epsilon_r} \sqrt{\epsilon_r \sin^2 \theta_1 - 1}}{\cos \theta_1}, m = 0, 1, 2, \dots$

$m=0$: $\tan(\beta_1 h \cos \theta_1) = \frac{\sqrt{\epsilon_r} \sqrt{\epsilon_r \sin^2 \theta_1 - 1}}{\cos \theta_1} \left\{ \begin{array}{l} \theta_1 = 60^\circ \\ \sin \theta_1 = 0.867 \end{array} \right.$

$$\beta_1 = \frac{1}{h \cos \theta_1} \tan^{-1} \left[\frac{\sqrt{\epsilon_r} \sqrt{\epsilon_r \sin^2 \theta_1 - 1}}{\cos \theta_1} \right] = \frac{1}{1.25(0.5)} \tan^{-1} \left\{ \frac{\sqrt{4} \sqrt{4(0.867)^2 - 1}}{0.5} \right\}$$

$$\beta_1 = \frac{1}{0.625} \tan^{-1}(5.6664) = \frac{79.9916^\circ (1.3961 \text{ rads})}{0.625} = \frac{1.3961}{0.625}$$

$$\beta_1 = 2.2338 \text{ rads/cm}$$

$$\beta_{yd} = \beta_1 \cos \theta_1 = 2.2338(0.5) = 1.1169 \text{ rads/cm}$$

$\text{TE}_1(m=1)$: $\tan(\beta_1 h \cos \theta_1 - \frac{m\pi}{2}) = \frac{\sqrt{\epsilon_r \sin^2 \theta_1 - 1}}{\sqrt{\epsilon_r \cos \theta_1}}$

$$\beta_1 = \frac{1}{h \cos \theta_1} \left\{ \frac{\pi}{2} + \tan^{-1} \frac{\sqrt{\epsilon_r \sin^2 \theta_1 - 1}}{\sqrt{\epsilon_r \cos \theta_1}} \right\} = \frac{1}{1.25(0.5)} \left\{ \frac{\pi}{2} + \tan^{-1} \left[\frac{\sqrt{4(0.867)^2 - 1}}{\sqrt{4(0.5)}} \right] \right\}$$

$$\beta_1 = \frac{1}{0.625} \left\{ \frac{\pi}{2} + 0.9561 \right\} = \frac{2.5269}{0.625} = 4.043 \text{ rads/cm}$$

$$\beta_1 = 4.043 \text{ rads/cm}$$

$$\beta_{yd} = \beta_1 \cos \theta_1 = 4.043(0.5) = 2.0215 \text{ rads/cm}$$

$$\beta_{yd} = 2.0215 \text{ rads/cm}$$

$$[B.34] \quad (f_c)_{\text{m}} = \frac{m}{4h\sqrt{\mu_d \epsilon_d - h \beta_0}}$$

$$f = (f_c)_1 + 0.1 (f_c)_2 = 1.1 (f_c)_1 = \frac{1.1}{4h\sqrt{\mu_0 \epsilon_0} \sqrt{\mu_r \epsilon_r - 1}} = \frac{1.1 (30 \times 10^9)}{4(0.635) \sqrt{\epsilon_r - 1}} = 10 \times 10^9$$

$$\sqrt{\epsilon_r - 1} = \frac{1.1 (3)}{4(0.635)} = 1.299 \Rightarrow \epsilon_r - 1 = (1.299)^2 = 1.688$$

$$\epsilon_r = 2.688$$

At cutoff $\beta_z = \beta_0$.

$$\beta_{yd} = \sqrt{\beta_d^2 - \beta_z^2} \Big|_{\beta_z = \beta_0} = \sqrt{\beta_d^2 - \beta_0^2} = \beta_0 \sqrt{\epsilon_r - 1} = \beta_{co} \sqrt{\epsilon_r - 1}$$

$$\beta_{yd} = 2\pi f_c \sqrt{\mu_0 \epsilon_0} \sqrt{\epsilon_r - 1}$$

$$f_c = \frac{10 \times 10^9}{3.1} = 9.091 \text{ GHz}$$

$$\beta_{yd} = 2\pi \frac{9.091 \times 10^9}{3 \times 10^8} \sqrt{2.688 - 1} = 247.375 \text{ Rad/m} = 2.47375 \text{ Rad/cm}$$

$$\alpha_{yo} = \sqrt{\beta_z^2 - \beta_0^2} \Big|_{\beta_z = \beta_0} = 0$$

Given: $\epsilon_r = 5, \mu_r = 1$ (assumed), thickness = $h = 5.625 \text{ cm}$

$$(f_c)_m = \frac{m}{4h\sqrt{\mu_d\epsilon_d - \mu_0\epsilon_0}} = \frac{m}{4h\sqrt{\mu_0\epsilon_0}\sqrt{\mu_r\epsilon_r - 1}} = \frac{m \cdot c}{4h\sqrt{\mu_r\epsilon_r - 1}} \quad c = 30 \times 10^9 \text{ cm/s}$$

$$\begin{aligned} m &= 1, 3, 5, \dots, \text{ TE}^z \text{ (even)} \\ m &= 0, 2, 4, \dots, \text{ TE}^z \text{ (odd)} \end{aligned}$$

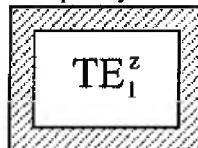
$$(f_c)_0 = \frac{(0) \cdot (30 \times 10^9 \text{ cm/s})}{(4) \cdot (5.625 \text{ cm})\sqrt{(1) \cdot (5) - 1}} = 0 \text{ Hz} \quad \Rightarrow \quad (f_c)_0 = 0 \text{ Hz } [\text{TM}_0^z \text{ (odd)}]$$

$$(f_c)_1 = \frac{(1) \cdot (30 \times 10^9 \text{ cm/s})}{(4) \cdot (5.625 \text{ cm})\sqrt{(1) \cdot (5) - 1}} = 6.67 \times 10^8 \text{ Hz} \quad \Rightarrow \quad (f_c)_1 = 0.667 \text{ GHz } [\text{TE}_1^z \text{ (even)}]$$

$$(f_c)_2 = \frac{(2) \cdot (30 \times 10^9 \text{ cm/s})}{(4) \cdot (5.625 \text{ cm})\sqrt{(1) \cdot (5) - 1}} = 1.33 \times 10^9 \text{ Hz} \quad \Rightarrow \quad (f_c)_2 = 1.33 \text{ GHz } [\text{TM}_2^z \text{ (odd)}]$$

$$(f_c)_3 = \frac{(3) \cdot (30 \times 10^9 \text{ cm/s})}{(4) \cdot (5.625 \text{ cm})\sqrt{(1) \cdot (5) - 1}} = 2.00 \times 10^9 \text{ Hz} \quad \Rightarrow \quad (f_c)_3 = 2.00 \text{ GHz } [\text{TE}_2^z \text{ (even)}]$$

For an operating frequency of 1 GHz, the ONLY the TE^z mode that can propagate inside the slab unattenuated is the TE_1^z mode, whose cutoff frequency is 0.667 GHz (< 1 GHz).



For the TE_1^z mode, find the propagation constant β_z .

Use the Graphical Method:

Plot:
$$-\frac{\mu_0}{\mu} (\beta_{yd} \cdot h) \cot(\beta_{yd} \cdot h) = \alpha_{y0} \cdot h$$

$$(\alpha_{y0} \cdot h)^2 + (\beta_{yd} \cdot h)^2 = a^2$$

$\alpha_{y0} \cdot h$ (ordinate) versus $\beta_{yd} \cdot h$ (abscissa)

Problem 8.35 Cont.

where,

and

$$a = \omega h \sqrt{\mu_0 \epsilon_0} \sqrt{\mu_r \epsilon_r - 1} = \frac{\omega h}{c} \sqrt{\mu_r \epsilon_r - 1} = \beta_0 h \sqrt{\mu_r \epsilon_r - 1}$$

$$\beta_{yd}^2 = \beta_d^2 - \beta_z^2 = \omega^2 \mu_d \epsilon_d - \beta_z^2$$

$$\alpha_{y0}^2 = \beta_z^2 - \beta_0^2 = \beta_z^2 - \omega^2 \mu_0 \epsilon_0$$

Solve for β_z ,

$$\beta_z^2 = \beta_d^2 - \beta_{yd}^2 = \omega^2 \mu_d \epsilon_d - \beta_{yd}^2$$

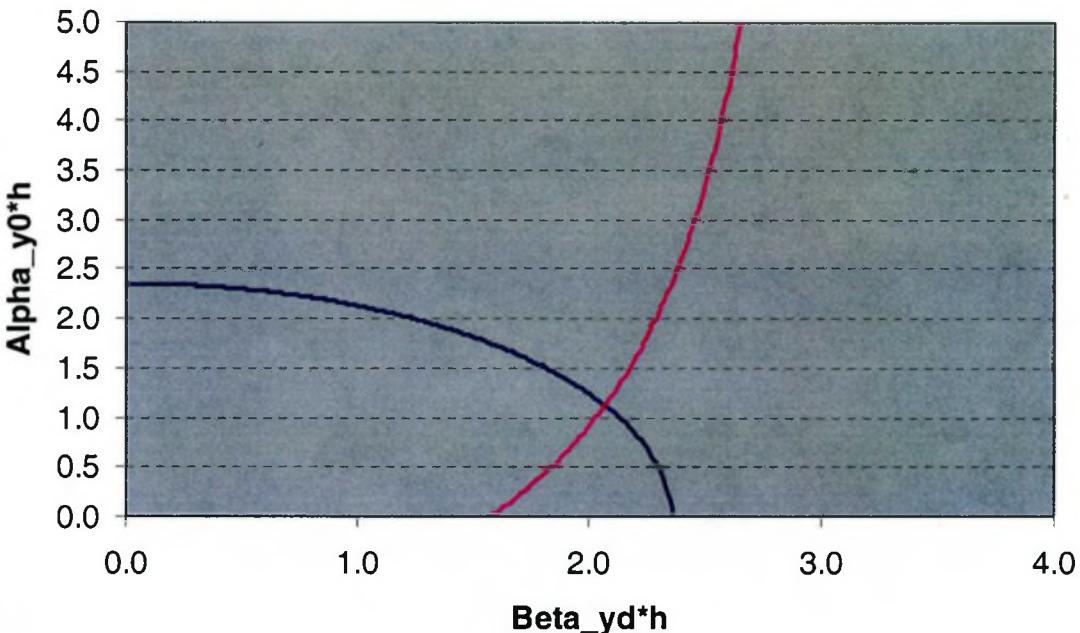
or

$$\beta_z^2 = \alpha_{y0}^2 + \beta_0^2 = \alpha_{y0}^2 + \omega^2 \mu_0 \epsilon_0$$

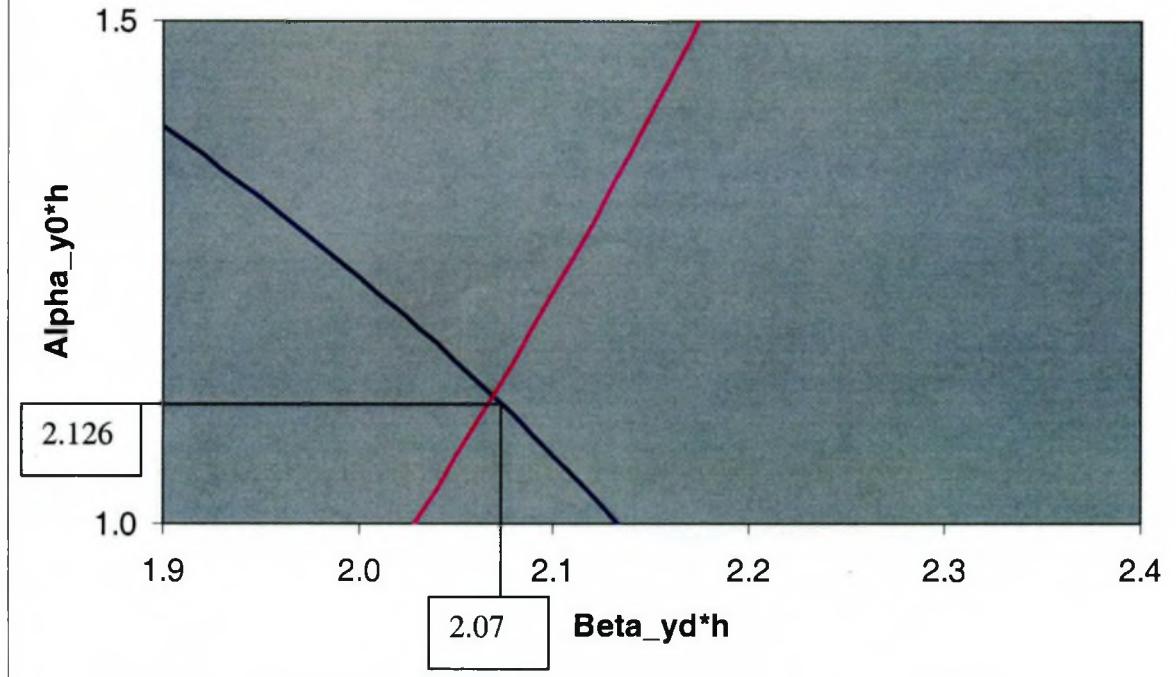
$$\beta_z = \pm \sqrt{\alpha_{y0}^2 + \beta_0^2} = \pm \sqrt{\alpha_{y0}^2 + \omega^2 \mu_0 \epsilon_0} = \pm \sqrt{\alpha_{y0}^2 + \frac{\omega^2}{c^2}}$$

TEz1 (even)

Graphical Solution for attenuation and phase constants of TEzm modes in a dielectric slab waveguide



TEz1 (even)
Graphical Solution for attenuation and phase constants of TEzm modes in a dielectric slab waveguide



$$\beta_z^2 = \alpha_{y0}^2 + \beta_0^2 = \alpha_{y0}^2 + \omega^2 \mu_0 \epsilon_0$$

$$\beta_z = \pm \sqrt{\alpha_{y0}^2 + \beta_d^2} = \pm \sqrt{\alpha_{y0}^2 + \omega^2 \mu_0 \epsilon_0} = \pm \sqrt{\alpha_{y0}^2 + \frac{\omega^2}{c^2}}$$

$$\beta_{yd} \cdot h = 2.07 \text{ rad}$$

$$\beta_{yd} = \frac{2.07 \text{ rad}}{5.625 \text{ cm}} = 0.368 \text{ rad/cm} \\ = 36.8 \text{ rad/m}$$

$$\alpha_{y0} \cdot h = 1.126 \text{ Np}$$

$$\alpha_{y0} = \frac{1.126 \text{ Np}}{5.625 \text{ cm}} = 0.200 \text{ Np/cm} = 20 \text{ Np/m}$$

$$\beta_z = \pm \sqrt{(0.200)^2 + \frac{(2 \cdot \pi \cdot 1 \times 10^9 \text{ Hz})^2}{(30 \times 10^9 \text{ cm/s})^2}} = 0.290 \text{ rad/cm}$$

$$\beta_z = 0.290 \text{ rad/cm}$$

β_z = 29 rad/m

$$8.36 \quad (f_c)_m = \frac{m}{4h\sqrt{\mu_0\epsilon_0 - \epsilon_r}} = \frac{m}{4h\sqrt{\mu_0}\sqrt{\epsilon_r - 1}} \quad TM_z^z(\text{odd}): m=0, 2, 4, \dots \\ TE_z^z(\text{even}): m=1, 3, 5, \dots$$

(a) $m=0, TM_z^z(\text{odd}) : f_c = 0$

$$(b) \quad m=1, TE_1^z(\text{even}) : (f_c)_1^{TE} = \frac{1}{4h\sqrt{\mu_0\epsilon_0}\sqrt{\epsilon_r - 1}} = \frac{30 \times 10^9}{4(0.113)\sqrt{\epsilon_r - 1}} = 20 \times 10^9$$

$$\sqrt{\epsilon_r - 1} = \frac{30}{4(0.113)(20)} = 3.3186 \Rightarrow \epsilon_r = 1 + (3.3186)^2 = 12.013$$

(c) $TE_1^z(\text{even}) : (f_c) = 20 \text{ GHz}$

$$(d) \quad \beta_{yd}^2 + \beta_z^2 = \beta_d^2 = \omega^2 \mu_0 \epsilon_0 \epsilon_r$$

At $f = 20 \text{ GHz}$, the TE_1^z mode is at cutoff $\Rightarrow \beta_z = \beta_0 = \omega \sqrt{\mu_0 \epsilon_0}$

$$\beta_{yd}^2 = \beta_d^2 - \beta_z^2 = \beta_d^2 - \beta_0^2 = (\omega \sqrt{\mu_0 \epsilon_0 \epsilon_r})^2 - (\omega \sqrt{\mu_0 \epsilon_0})^2 = (\omega \sqrt{\mu_0 \epsilon_0})^2 (\epsilon_r - 1)$$

$$\beta_{yd} = \sqrt{2\pi (20 \times 10^9)^2 (12.013 - 1)} = 193.2337$$

$$\beta_{yd} = 193.2337 \Rightarrow \beta_{yd} = 13.901 \text{ rad/cm} = 796.460 \text{ degrees/cm}$$

$$\beta_{yd} = 796.460 \text{ degrees/cm}$$

$$-\alpha_{yo}^2 + \beta_z^2 = \beta_0^2 \Rightarrow \alpha_{yo}^2 = \beta_z^2 - \beta_0^2 = 0$$

$\beta_z = \beta_0$
@ cutoff

At cutoff: $\alpha_{yo} = 0 \text{ dB/cm}$

(e) At 25 GHz \Rightarrow For $TE_1^z(\text{even})$, what is β_{yd} and α_{yo} ?

Using the computer program at the end of Chapter 8

Ground-TE-TM-graph

$$\beta_{yd} = 16.5909 \text{ rad/cm} = 950.5885 \text{ degrees/cm}$$

$$\alpha_{yo} = 5.2039 \text{ Nepers/cm} = 45.1699 \text{ dB/cm}$$

8.37

$$a. (f_c) = \frac{m}{4h\sqrt{\mu_1\varepsilon_1 - \mu_2\varepsilon_2}} = \frac{m}{4h\sqrt{\mu_1\varepsilon_1} \sqrt{1 - \frac{\mu_2\varepsilon_2}{\mu_1\varepsilon_1}}} = \frac{m}{4h\sqrt{\varepsilon_1} \sqrt{1 - \frac{\varepsilon_2}{\varepsilon_1}}}$$

$$(f_c)_m = \frac{m}{4h\sqrt{\varepsilon_1} \sqrt{1 - \frac{\varepsilon_2}{\varepsilon_1}}} = \frac{m}{4h\sqrt{\varepsilon_1} \sqrt{\varepsilon_0} \sqrt{1 - \frac{\varepsilon_2}{\varepsilon_1}}}, m=0, 2, 4, TM^2(\text{odd})$$

$$m=1, 3, 5, TE^2(\text{even})$$

a. $m=0: TM^2_0(\text{odd})$ b. $[m=1: TE^2_1(\text{even})]$

$$c. (f_c)_1 = (f_c)_0 = (3.1 - 0) \times 10^9 = 3.1 \times 10^9 = \frac{1}{4h\sqrt{\varepsilon_{r2}} \sqrt{\varepsilon_0} \sqrt{1 - \frac{\varepsilon_{r2}}{\varepsilon_{r1}}}}$$

$$3.1 \times 10^9 = \frac{30 \times 10^9}{4(2)\sqrt{2.56}} \frac{1}{\sqrt{1 - \frac{\varepsilon_{r2}}{\varepsilon_{r1}}}} \Rightarrow \sqrt{1 - \frac{\varepsilon_{r2}}{\varepsilon_{r1}}} = \frac{30}{4(2)(1.6)3.1}$$

$$\sqrt{1 - \frac{\varepsilon_{r2}}{\varepsilon_{r1}}} = 0.756 \Rightarrow 1 - \frac{\varepsilon_{r2}}{\varepsilon_{r1}} = (0.756)^2 = 0.5716$$

$$\frac{\varepsilon_{r2}}{\varepsilon_{r1}} = 1 - 0.5716 = 0.42839 \Rightarrow \varepsilon_{r2} = 0.42839 \varepsilon_{r1} = 0.42839(2.56)$$

$$\varepsilon_{r2} = 0.42839(2.56) = 1.09668 \approx 1.1$$

$$\boxed{\varepsilon_{r2} = 1.09668 \approx 1.1}$$

$$d. \beta_{yd}^2 + \beta_z^2 = \beta_{d1}^2 = \omega^2 \mu_1 \varepsilon_{d1} = \omega^2 \mu_1 \varepsilon_{r1} \varepsilon_0$$

At $f=3.1$ GHz the TE^2_1 mode is @ cutoff $\Rightarrow \beta_2 = \beta_z = \omega \sqrt{\mu_1 \varepsilon_0}$.

$$\beta_{yd}^2 = \beta_{d1}^2 - \beta_z^2 = \beta_{d1}^2 - \beta_{d2}^2 = \omega^2 \mu_1 \varepsilon_1 - \omega^2 \mu_2 \varepsilon_2 = \omega^2 \mu_1 \varepsilon_{r1} \varepsilon_0 - \omega^2 \mu_1 \varepsilon_{r2} \varepsilon_0$$

$$\beta_{yd}^2 = \omega^2 \mu_1 \varepsilon_0 (\varepsilon_{r1} - \varepsilon_{r2}) = \omega^2 \mu_1 \varepsilon_0 \varepsilon_{r1} \left(1 - \frac{\varepsilon_{r2}}{\varepsilon_{r1}}\right) = 0.756$$

$$\beta_{yd} = \omega \sqrt{\mu_1 \varepsilon_0} \sqrt{\varepsilon_{r1} \left(1 - \frac{\varepsilon_{r2}}{\varepsilon_{r1}}\right)^{1/2}} = 2\pi(3.1 \times 10^9) \sqrt{2.56} \sqrt{1 - \frac{\varepsilon_{r2}}{\varepsilon_{r1}}}$$

$$\boxed{\beta_{yd} = \frac{2\pi(3.1)(1.6)(0.756)}{30} = 0.4854 \text{ rad/cm} = 44.997 \text{ deg/cm}}$$

$$-\alpha_{y2}^2 + \beta_z^2 = \beta_z^2 \Rightarrow \alpha_{y2}^2 = \beta_z^2 - \beta_z^2 |_{\beta_2 = \beta_z = 0} = 0$$

$$\boxed{\alpha_{y2} = 0} \quad @ \text{cutoff}$$

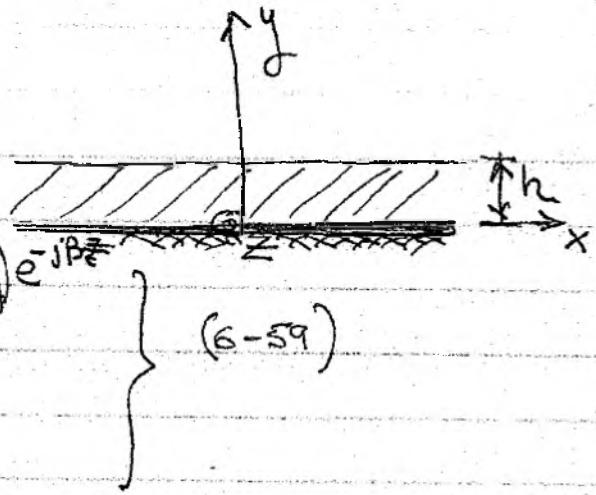
8.38

$$\underline{\text{TM}^z(\text{even})}: A_{ze}^d = A_{me}^d \cos(\beta_{yd}y) e^{-j\beta_e z}$$

$$H_{xe}^d = \frac{1}{\mu_d} \frac{\partial A_{ze}^d}{\partial y} = -A_{me} \frac{\beta_{yd}}{\mu_d} \sin(\beta_{yd}y) e^{-j\beta_e z}$$

$$H_{ye}^d = -\frac{1}{\mu_d} \frac{\partial A_{ze}^d}{\partial x} = 0$$

$$H_{ze}^d = 0$$



$$H_{xe}^d(y=0) = -A_{me} \frac{\beta_{yd}}{\mu_d} \sin(\beta_{yd}y)|_{y=0} e^{-j\beta_e z} = 0$$

$\therefore \text{TM}^z(\text{even})$ satisfy the additional B.C. [$H_{tan}^d(y=0)=0$].
Therefore they can propagate as guided waves.

$$\underline{\text{TM}^z(\text{odd})}: A_{mo}^d = A_{mo}^d \sin(\beta_{yd}y) e^{-j\beta_e z}$$

$$H_{xe}^d = \frac{1}{\mu_d} \frac{\partial A_{mo}^d}{\partial y} = +A_{mo} \frac{\beta_{yd}}{\mu_d} \cos(\beta_{yd}y) e^{-j\beta_e z}$$

$$H_{yo}^d = -\frac{1}{\mu_d} \frac{\partial A_{mo}^d}{\partial x} = 0$$

$$H_{zo}^d = 0$$

$$H_{xe}^d(y=0) = +A_{mo} \frac{\beta_{yd}}{\mu_d} \cos(\beta_{yd}y)|_{y=0} e^{-j\beta_e z} = A_{mo} \frac{\beta_{yd}}{\mu_d} (1) e^{-j\beta_e z} \neq 0$$

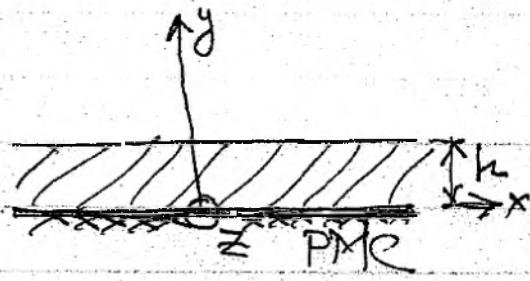
$\therefore \text{TM}^z(\text{odd})$ do NOT satisfy the additional B.C. [$H_{tan}^d(y=0)=0$].
Therefore they cannot propagate as guided waves.

$$(b) (f_c)_m(\text{even}) = \frac{m}{4h\sqrt{\mu_d\varepsilon_d - \beta_{eo}^2}}, \quad m = 1, 3, 5, \dots$$

$$(c) (f_c)_1(\text{even}) = \frac{1}{4h\sqrt{\mu_d\varepsilon_d - \beta_{eo}^2}} = \frac{1}{4h\beta_{eo}\sqrt{\mu_d\varepsilon_d - 1}} = \frac{30 \times 10^9}{4(0.125)\sqrt{4-1}} = \frac{30 \times 10^9}{0.5(\sqrt{3})} = 34.64 \text{ GHz}$$

8.39

$$(a) \underline{\text{TE}^z \text{ (even)}}: F_{ze}^d = B_{me}^d \cos(\beta_{yd} y) e^{-j\beta_z z}$$



$$H_{xe}^d = -j \frac{1}{\omega \mu_d \epsilon_d} \frac{\partial^2 F_{ze}^d}{\partial x \partial z} = 0$$

$$H_{ye}^d = -j \frac{1}{\omega \mu_d \epsilon_d} \frac{\partial^2 F_{ze}^d}{\partial y \partial z} = -j \frac{(-\beta_{yd})(-\beta_z)}{\omega \mu_d \epsilon_d} B_{me}^d \sin(\beta_{yd} y) e^{-j\beta_z z}$$

$$H_{ze}^d = -j \frac{1}{\omega \mu_d \epsilon_d} \left(\frac{\partial^2}{\partial z^2} + \beta_d^2 \right) F_{ze}^d = -j \frac{1}{\omega \mu_d \epsilon_d} \left[(-j\beta_z^2) + \beta_d^2 \right] B_{me}^d \cos(\beta_{yd} y) e^{-j\beta_z z}$$

$$H_{ze}^d(y=0) = -j \frac{B_{me}^d}{\omega \mu_d \epsilon_d} \left[-\beta_z^2 + \beta_d^2 \right] \cos(\beta_{yd} 0) \Big|_{y=0} e^{-j\beta_z z} \neq 0$$

$\therefore \text{TE}^z \text{ (even)}$ do NOT satisfy the additional B.C. $[H_{tan}^d(y=0) = 0]$.
Therefore they cannot propagate as guided waves.

$$\underline{\text{TE}^z \text{ (odd)}}: F_{ze0}^d = B_{mo}^d \sin(\beta_{yd} y) e^{-j\beta_z z}$$

$$H_{xe0}^d = -j \frac{1}{\omega \mu_d \epsilon_d} \frac{\partial^2 F_{ze0}^d}{\partial x \partial z} = 0$$

$$H_{ye}^d = -j \frac{1}{\omega \mu_d \epsilon_d} \frac{\partial^2 F_{ze0}^d}{\partial y \partial z} = -j \frac{(\beta_{yd})(-\beta_z)}{\omega \mu_d \epsilon_d} B_{mo}^d \cos(\beta_{yd} y) e^{-j\beta_z z}$$

$$H_{ze}^d = -j \frac{1}{\omega \mu_d \epsilon_d} \left(\frac{\partial^2}{\partial z^2} + \beta_d^2 \right) F_{ze0}^d = -j \frac{B_{mo}^d}{\omega \mu_d \epsilon_d} \left[(-j\beta_z^2) + \beta_d^2 \right] \sin(\beta_{yd} y) e^{-j\beta_z z}$$

$$H_{ze}^d(y=0) = -j \frac{B_{mo}^d}{\omega \mu_d \epsilon_d} \left(-\beta_z^2 + \beta_d^2 \right) \sin(\beta_{yd} 0) \Big|_{y=0} e^{-j\beta_z z} = 0$$

$\therefore \text{TE}^z \text{ (odd)}$ do satisfy the additional B.C. $[H_{tan}^d(y=0) = 0]$.

Therefore they can propagate as guided waves.

$$(b) (f_c)^{(odd)}_m = \frac{m}{4h\sqrt{\mu_d \epsilon_d - \beta_d^2}}, m=0, 2, 4, \dots \quad (\text{e-16ff})$$

$$(c) (f_c)^{(odd)}_{m=0} = 0$$

8.40

$$\epsilon_r = 2.20$$

a. $w/b = 1$: For zero thickness center conductor

$$\frac{C_f}{\epsilon} \approx \frac{1}{\pi} [2 \ln(2)] = 0.4413$$

$$Z_c \sqrt{\epsilon_r} = \frac{30\pi}{\frac{w/b}{1-t/b} + \frac{C_f}{\epsilon}} = \frac{30\pi}{w/b + \frac{C_f}{\epsilon}} = \frac{30\pi}{1+0.4413} = 65.39$$

$$Z_c = \frac{65.39}{\sqrt{2.2}} = 44.086 \text{ Vs. } 44.09 \text{ from Example 8-17.}$$

Good agreement between the two because large w/b .

b. $w/b = 0.1$:

$$Z_c \sqrt{\epsilon_r} = \frac{30\pi}{\frac{w/b}{1-t/b} + \frac{C_f}{\epsilon}} = \frac{30\pi}{w/b + \frac{C_f}{\epsilon}} = \frac{30\pi}{0.1+0.4413} = 174.114$$

$$Z_c = \frac{174.114}{\sqrt{2.2}} = 117.387 \text{ Vs. } 131.04 \text{ from Example 8-17.}$$

Poor agreement between the two because of small w/b which leads to greater fringing. Equations (8-200a) - (8-201a) are not as accurate for small values of w/b .

8.41

$$a. Z_c \sqrt{\epsilon_r} = \frac{30\pi}{\frac{w/b}{1-t/b} + \frac{C_f}{\epsilon}} \Big|_{t/b=0} = \frac{30\pi}{\frac{w}{b} + \frac{C_f}{\epsilon}} = \frac{30\pi}{\frac{w}{b} + 0.4413}$$

$$\frac{w}{b} = -0.4413 + \frac{30\pi}{Z_c \sqrt{\epsilon_r}} = -0.4413 + \frac{30\pi}{30\sqrt{4}} = -0.4413 + 1.570796$$

$$\boxed{\frac{w}{b} = 1.1295}$$

$$b. k = \tanh\left(\frac{\pi}{2} \frac{w}{b}\right) = \tanh\left(\frac{\pi}{2} (1.1295)\right) = \tanh(1.7742) = 0.944$$

$$k = 0.944 < 1$$

$$\frac{k(k)}{K(k)} \approx \frac{1}{\pi} \ln\left[2 \frac{1+\sqrt{k}}{1-\sqrt{k}}\right] \Big|_{k=0.944} = \frac{1}{\pi} \ln\left[2 \frac{1+\sqrt{0.944}}{1-\sqrt{0.944}}\right]$$

$$\approx \frac{1}{\pi} \ln\left[2 \frac{1.9716}{0.028365}\right] = \frac{1}{\pi} \ln(139.0159) = \frac{1}{\pi} (4.9346) = 1.5707$$

$$Z_c = \frac{1}{\sqrt{\epsilon_r}} \frac{30\pi}{K(k)/K(k')} = \frac{1}{\sqrt{4}} \frac{30\pi}{1.5707} = \frac{60}{2} = 30$$

$$\boxed{Z_c = 30 \text{ ohms !!}}$$

B.42

$$\underline{E} = \hat{a}_y E_0 e^{-j\beta z} \Rightarrow \underline{H} = -\hat{a}_x \frac{E_0}{\eta} e^{-j\beta z}$$

$$\alpha_c = \frac{P_c/l}{2P_d}$$

$$P_d = \int_0^h \int_0^w \frac{1}{2} \operatorname{Re}(\underline{E} \times \underline{H}^*) \cdot \hat{a}_z dx dy = \frac{1}{2} \operatorname{Re} \iint \left[\hat{a}_y E_y \times (-\hat{a}_x H_x^*) \right] \cdot \hat{a}_z dx dy$$

$$= \frac{1}{2} \operatorname{Re} \iint (\hat{a}_z E_y H_x^*) \cdot \hat{a}_z dx dy = \frac{1}{2} \operatorname{Re} \iint E_y H_x^* dx dy$$

$$P_d = \frac{1}{2} \int_0^h \int_0^w \frac{|E_0|^2}{\eta} dx dy = \frac{|E_0|^2}{2\eta} wh$$

$$P_c = 2 \left\{ R_s \int_0^w \int_0^l \underline{J}_s \cdot \underline{J}_s^* dz dx \right\} = R_s \int_0^w \int_0^l |\underline{J}_s|^2 dz dx$$

$$\underline{J}_s = \hat{n} \times \underline{H} \Big|_{y=0} = \hat{a}_y \times \left(-\hat{a}_x \frac{E_0}{\eta} e^{-j\beta z} \right) = \hat{a}_z \frac{E_0}{\eta} e^{-j\beta z}$$

$$P_c = R_s \frac{|E_0|^2}{\eta^2} \int_0^w \int_0^l dz dx = R_s \frac{|E_0|^2}{\eta^2} wl$$

$$\frac{P_c}{l} = R_s \frac{|E_0|^2}{\eta^2} w$$

Therefore

$$\alpha_c = \frac{P_c/l}{2P_d} = \frac{R_s \frac{|E_0|^2}{\eta^2} w}{2 \frac{|E_0|^2}{2\eta} wh} = R_s \frac{1}{\eta h} = \frac{R_s}{\eta h}$$

B.43

The capacitance of a wire of radius a_e and length l located parallel at a height h above a grounded conducting plane is given by

$$C = \frac{2\pi E_0 l}{\cosh^{-1}(h/a_e)} \approx \frac{2\pi E_0 l}{\ln(2h/a_e)} \Rightarrow \frac{C}{l} = \frac{2\pi E_0}{\cosh^{-1}(h/a_e)} \approx \frac{2\pi E_0}{\ln(2h/a_e)}$$

Using (B-198) we can write the characteristic impedance as

$$Z_c = \frac{\sqrt{h_0 \epsilon_0}}{C/l} = \frac{\sqrt{h_0 \epsilon_0}}{2\pi E_0} \cosh^{-1}(h/a_e) \approx \frac{1}{2\pi} \sqrt{\frac{F_0}{\epsilon_0}} \ln(2h/a_e) = \frac{\pi}{2\pi} \ln\left(\frac{2h}{0.25w}\right)$$

8.44

$$w_1 = 1.505 \text{ mm}, h = 1 \text{ mm}, w_3 = 0.549 \text{ mm}, \epsilon_{r1} = \epsilon_{r3} = 6$$

$$(a) \quad w_1/h = 1.505/1 = 1.505, \quad w_3/h = 0.549/1 = 0.549$$

$$\epsilon_{r1,\text{eff}} = \frac{\epsilon_{r1}+1}{2} + \frac{\epsilon_{r1}-1}{2} \left[1 + 12 \left(\frac{h}{w} \right) \right]^{-1/2} = \frac{6+1}{2} + \frac{6-1}{2} \left[\frac{1}{\sqrt{1 + \frac{12}{1.505}}} \right]$$

$$= 3.5 + \frac{2.5}{2.9956} = 3.5 + 0.8346 = 4.3346$$

$$\epsilon_{r2,\text{eff}} = 3.5 + 2.5 \left\{ \frac{1}{\sqrt{1 + \frac{12}{0.549}}} + 0.04 [1 - 0.549]^2 \right\} = 3.5 + 2.5 [0.2092 + 0.0081]$$

$$= 3.5 + 2.5 (0.2173) = 3.5 + 0.5433 = 4.0433$$

(b) Using the expressions of (8-205a) and (8-204a) for the characteristic impedances, respectively, for $w/h > 1$ and $w/h < 1$

$$Z_{c1} = \frac{120\pi/\sqrt{\epsilon_{r1,\text{eff}}}}{\frac{w}{h} + 1.393 + 0.667 \ln(\frac{w}{h} + 1.444)} = \frac{120\pi/\sqrt{4.3346}}{1.505 + 1.393 + 0.667 \ln(1.505 + 1.444)}$$

$$= \frac{181.0742}{1.505 + 1.393 + 0.667(1.0815)} = \frac{181.0742}{3.6193} = 50.03$$

$$Z_{c3} = \frac{60}{\sqrt{\epsilon_{r3}}} \ln \left[\frac{8}{w/h} + \frac{w/h}{4} \right] = \frac{60}{\sqrt{4.0433}} \ln \left[\frac{8}{0.549} + \frac{0.549}{4} \right]$$

$$= \frac{60}{2.0108} \ln(14.5719 + 0.1373) = \frac{60}{2.0108} \ln(14.7092) = \frac{60(2.6885)}{2.0108}$$

$$Z_{c3} = 80.22 \text{ ohms}$$

$$(c) \quad Z_{c2} = \sqrt{Z_{c1} Z_{c3}} = \sqrt{50.03(80.22)} = 63.35 \text{ ohms}$$

(d) Assuming the effective dielectric constant of the line between the input and output lines is approximately the average of the two. $\epsilon_{r2,\text{eff}} = \frac{4.3346 + 4.0433}{2} = 4.189$

the wavelength @ $2\frac{3}{4}\text{Hz}$ is

$$\lambda_2 = \frac{\lambda_0}{\sqrt{\epsilon_{r2,\text{eff}}}} = \frac{30 \times 10^9}{\sqrt{4.189}} = 14.6578 \text{ cm}$$

Therefore the length of the $2/4$ transformer is $\frac{14.6578}{4} = 3.6644 \text{ cm}$

TE^Z (H^Z) Modes

For the waveguide geometry of Figure 8-3, the TE^Z mode fields can be found using the potential functions

$$\underline{F}^i = \hat{a}_z F_z^i(x, y, z) = \hat{a}_z \psi_h^i(x, y) e^{-j\beta_z z} \quad (1a)$$

$$\underline{A}^i = 0 \quad (1b)$$

where i refers to the two media (air and dielectric; $i=0$ for air and $i=d$ for the dielectric). The normalized field components (normalized with respect to $-\beta_z/\omega_0\epsilon$)

$$E_{hx}^i = -\frac{\omega_0 \epsilon_i}{\beta_z} \left[-\frac{1}{\epsilon} \frac{\partial \psi_h^i(x, y) e^{-j\beta_z z}}{\partial y} \right] = \frac{\omega_0 \epsilon_i}{\beta_z} \frac{\partial \psi_h^i}{\partial y} e^{-j\beta_z z} \quad (2a)$$

$$E_{hy}^i = -\frac{\omega_0 \epsilon_i}{\beta_z} \left[\frac{1}{\epsilon} \frac{\partial \psi_h^i(x, y) e^{-j\beta_z z}}{\partial x} \right] = -\frac{\omega_0 \epsilon_i}{\beta_z} \frac{\partial \psi_h^i}{\partial x} e^{-j\beta_z z} \quad (2b)$$

$$E_{hz}^i = 0 \quad (2c)$$

$$H_{hx}^i = -\frac{\omega_0 \epsilon_i}{\beta_z} \left[-\frac{j}{\omega_0 \epsilon_i} \frac{\partial^2 \psi_h^i(x, y) e^{-j\beta_z z}}{\partial x \partial z} \right] = \frac{\partial \psi_h^i}{\partial x} e^{-j\beta_z z} \quad (2d)$$

8.45 Contd.

$$H_{hy}^i = - \frac{\omega \mu_i \epsilon_i}{\beta_z} \left[- \frac{j}{\omega \mu_i \epsilon_i} \frac{\partial^2 \psi_h^i(x, y) e^{-j\beta_z z}}{\partial y \partial z} \right] = \frac{\partial \psi_h^i}{\partial y} e^{-j\beta_z z} \quad (2e)$$

$$H_{hz}^i = - \frac{\omega \mu_i \epsilon_i}{\beta_z} \left[- \frac{j}{\omega \mu_i \epsilon_i} \left(\frac{\partial^2}{\partial z^2} + \beta_i^2 \right) \psi_h^i(x, y) e^{-j\beta_z z} \right] = j \left(\frac{\beta_i^2 - \beta_z^2}{\beta_z} \right) \psi_h^i e^{-j\beta_z z} \quad (2f)$$

where ψ_h^i satisfies the transverse wave equation of

$$\left[\frac{\partial^2 \psi_h^i}{\partial x^2} + \frac{\partial^2 \psi_h^i}{\partial y^2} \right] + (\beta_i^2 - \beta_z^2) \psi_h^i = \nabla_t^2 \psi_h^i + (\beta_i^2 - \beta_z^2) \psi_h^i = 0 \quad (3)$$

For the symmetrical modes which require the tangential magnetic field components to vanish along the symmetry plane of the waveguide ($x=0, 0 \leq y \leq b$), ψ_h^i takes the form of

$$\psi_h^d = \sum_{n=1}^{\infty} \left[A_{hn}^d \cos(\beta_{xn}^d x) + B_{hn}^d \sin(\beta_{xn}^d x) \right] \left[C_{hn}^d \cos(\beta_{yn}^d y) + D_{hn}^d \sin(\beta_{yn}^d y) \right] \quad (4a)$$

$$\psi_h^0 = \sum_{n=1}^{\infty} \left[A_{hn}^0 \cos(\beta_{xn}^0 x) + B_{hn}^0 \sin(\beta_{xn}^0 x) \right] \left[C_{hn}^0 \cos(\beta_{yn}^0 (b-y)) + D_{hn}^0 \sin(\beta_{yn}^0 (b-y)) \right] \quad (4b)$$

For the TE^Z modes, the independent boundary conditions on the metallic walls of the waveguide are

$$E_{hx}^d(-a/2 \leq x \leq a/2, y=0, z) = E_{hy}^0(-a/2 \leq x \leq a/2, y=b, z) = 0 \quad (5a)$$

$$E_{hy}^d(x=\pm a/2, 0 \leq y \leq b, z) = E_{hy}^0(x=\pm a/2, b \leq y \leq b, z) = 0 \quad (5b)$$

Enforcing the first boundary condition of (5a) using and (4a), leads to

$$E_{hx}^d(-a/2 \leq x \leq a/2, y=0, z) = \frac{\omega \mu_d \beta_{yn}^d}{\beta_z} \sum_{n=1}^{\infty} \left[A_{hn}^d \cos(\beta_{xn}^d x) + B_{hn}^d \sin(\beta_{xn}^d x) \right] \cdot \left[-C_{hn}^d(0) + D_{hn}^d(1) \right] e^{-j\beta_z z} = 0 \Rightarrow D_{hn}^d = 0 \quad (6)$$

Applying the first boundary condition of (5b) using (2b),

Contd.

8.45 Contd. (4a) and (6) leads to

$$E_{hy}^d(x=\pm a/2, 0 \leq y \leq h, z) = -\frac{\omega \mu_d \beta_{xn}^d}{\beta_z} \sum_{n=1}^{\infty} \left[-A_{hn}^d \sin(\pm \frac{\beta_{xn}^d a}{2}) + B_{hn}^d \cos(\pm \frac{\beta_{xn}^d a}{2}) \right] \\ \cdot [C_{hn}^d \cos(\beta_{yn}^d y)] e^{-j\beta_z z} = 0 \quad (7)$$

which is satisfied when

$$-A_{hn}^d \sin(\frac{\beta_{xn}^d a}{2}) + B_{hn}^d \cos(\frac{\beta_{xn}^d a}{2}) = 0 \quad (7a)$$

$$A_{hn}^d \sin(\frac{\beta_{xn}^d a}{2}) + B_{hn}^d \cos(\frac{\beta_{xn}^d a}{2}) = 0 \quad (7b)$$

The above two equations are satisfied when

$$A_{hn}^d = 0 \quad (7c)$$

and

$$\cos(\frac{\beta_{xn}^d a}{2}) = 0 \Rightarrow \frac{\beta_{xn}^d a}{2} = \cos^{-1}(0) = (n - \frac{1}{2})\pi$$

$$\boxed{\beta_{xn}^d = \frac{2\pi}{a}(n - \frac{1}{2}), \quad n=1, 2, 3, \dots} \quad (7d)$$

Thus (4a) reduces to

$$\Psi_h^d = \sum_{n=1}^{\infty} B_{hn}^d C_{hn}^d \sin(\beta_{xn}^d x) \cos(\beta_{yn}^d y) \quad (8)$$

where

$$(\beta_{xn}^d)^2 + (\beta_{yn}^d)^2 + \beta_z^2 = (\beta_d)^2 \Rightarrow (\beta_{yn}^d)^2 = (\beta_d)^2 - [(\beta_{xn}^d)^2 + \beta_z^2] \quad (8a)$$

or

$$\beta_{yn}^d = \sqrt{(\beta_d)^2 - [(\beta_{xn}^d)^2 + \beta_z^2]} = j\sqrt{[(\beta_{xn}^d)^2 + \beta_z^2] - (\beta_d)^2} = j\alpha_n^d \quad (8b)$$

with

cont'd.

3.4.5 cont'd.

$$\alpha_n^d = \sqrt{[(\beta_{xn}^d)^2 + \beta_z^2] - (\beta_d)^2}$$

(8c)

Therefore (8) can ultimately be written as

$$\psi_h^d = \sum_{n=1}^{\infty} B_{hn}^d C_{hn}^d \sin(\beta_{xn}^d x) \cos(\beta_{yn}^d y) = \sum_{n=1}^{\infty} B_{hn}^d C_{hn}^d \sin(\beta_{xn}^d x) \cos(j\alpha_n^d y)$$

$$\boxed{\psi_h^d = \sum_{n=1}^{\infty} A_{hn}^d \sin(\beta_{xn}^d x) \cosh(\alpha_n^d y)} \quad (9)$$

where β_{xn}^d and α_n^d are given, respectively, by (7d) and (8c). If the α_n^d is imaginary, then the hyperbolic cosine must be replaced by a trigonometric cosine.

Applying the second boundary condition of (5a) using (2a) and (4b) leads to

$$E_{hx}^0(-a/2 \leq x \leq a/2, y=b, z) = \frac{\omega \mu_0 \beta_{vn}^0}{\beta_z} \sum_{n=1}^{\infty} [A_{hn}^0 \cos(\beta_{xn}^0 x) + B_{hn}^0 \sin(\beta_{xn}^0 x)] \\ \cdot [-C_{hn}^0(0) + D_{hn}^0(1)] e^{-j\beta_z z} = 0 \Rightarrow D_{hn}^0 = 0 \quad (10)$$

Enforcing the second boundary condition of (5b) using (2b) (4b), and (10) leads to

$$E_{hy}^0(x=\pm a/2, h \leq y \leq b, z) = - \frac{\omega \mu_0 \beta_{xn}^0}{\beta_z} \sum_{n=1}^{\infty} [-A_{hn}^0 \sin(\pm \frac{\beta_{xn}^0 a}{2}) + B_{hn}^0 \cos(\pm \frac{\beta_{xn}^0 a}{2})] \\ \cdot [-C_{hn}^0 \cos(\beta_{yn}^0(h-y))] e^{-j\beta_z z} = 0 \quad (11)$$

which is satisfied when

$$-A_{hn}^0 \sin(\frac{\beta_{xn}^0 a}{2}) + B_{hn}^0 \cos(\frac{\beta_{xn}^0 a}{2}) = 0 \quad (11a)$$

cont'd.

B.45 Contd.

$$A_{hn}^0 \sin\left(\frac{\beta_{xn}^0 a}{2}\right) + B_{hn}^0 \cos\left(\frac{\beta_{xn}^0 a}{2}\right) = 0 \quad (11b)$$

The above two equations are satisfied when

$$A_{hn}^0 = 0 \quad (11c)$$

and

$$\cos\left(\frac{\beta_{xn}^0 a}{2}\right) = 0 \Rightarrow \frac{\beta_{xn}^0 a}{2} = \cos^{-1}(0) = (n - \frac{1}{2})\pi$$

$$\boxed{\beta_{xn}^0 = \frac{2\pi}{a}(n - \frac{1}{2}) = \beta_{xn}^d = \beta_{xn}, \quad n=1, 2, 3, \dots} \quad (11d)$$

Thus (4b) reduces to

$$\psi_h^0 = \sum_{n=1}^{\infty} B_{hn}^0 C_{hn}^0 \sin(\beta_{xn}^0 x) \cos[\beta_{yn}^0(b-y)] \quad (12)$$

where

$$\begin{aligned} (\beta_{xn}^0)^2 + (\beta_{yn}^0)^2 + \beta_z^2 &= (\beta_0)^2 \Rightarrow (\beta_{yn}^0)^2 = (\beta_0)^2 - [(\beta_{xn}^0)^2 + \beta_z^2] \\ \text{or} \end{aligned} \quad (12a)$$

$$\beta_{yn}^0 = \sqrt{(\beta_0)^2 - [(\beta_{xn}^0)^2 + \beta_z^2]} = j\sqrt{[(\beta_{xn}^0)^2 + \beta_z^2] - (\beta_0)^2} = j\alpha_n^0 \quad (12b)$$

with

$$\boxed{\alpha_n^0 = \sqrt{[(\beta_{xn}^0)^2 + \beta_z^2] - (\beta_0)^2}} \quad (12c)$$

Therefore (12) can ultimately be written as

$$\psi_h^0 = \sum_{n=1}^{\infty} B_{hn}^0 C_{hn}^0 \sin(\beta_{xn}^0 x) \cos[\beta_{yn}^0(b-y)] = \sum_{n=1}^{\infty} B_{hn}^0 C_{hn}^0 \sin(\beta_{xn}^0 x) \cos[j\alpha_n^0(b-y)]$$

$$\boxed{\psi_h^0 = \sum_{n=1}^{\infty} B_{hn}^0 \sin(\beta_{xn}^0 x) \cosh[\alpha_n^0(b-y)]} \quad (13)$$

Contd.

B45 cont'd.

where β_{xn}^0 and α_n^0 are given, respectively, by (11d) and (12c).

In summary then for the TE^Z modes

$$\psi_h^d = \sum_{n=1}^{\infty} A_{hn}^d \sin(\beta_{xn}x) \cosh(\alpha_n^d y) \quad (14a)$$

$$\psi_h^0 = \sum_{n=1}^{\infty} B_{hn}^0 \sin(\beta_{xn}x) \cosh[\alpha_n^0(b-y)] \quad (14b)$$

$$\beta_{xn} = \frac{2\pi}{a}(n - \frac{1}{2}), \quad n=1, 2, 3, \dots \quad (14c)$$

$$\alpha_n^d = \sqrt{[(\beta_{xn})^2 + \beta_z^2] - (\beta_d)^2} \quad (14d)$$

$$\alpha_n^0 = \sqrt{[(\beta_{xn})^2 + \beta_z^2] - (\beta_0)^2} \quad (14e)$$

$TM^Z (E^Z)$ Modes

Following a similar procedure as for the TE^Z modes it can be shown that the fields of the TM^Z modes can be derived using the potential functions

$$A^i = \hat{a}_z A_z^i(x, y, z) = \hat{a}_z \psi_e^i(x, y) e^{-j\beta_z z} \quad (15a)$$

$$F^i = 0 \quad (15b)$$

Using the normalized field components (normalized with respect to $-\beta_z/\omega \mu_e \epsilon_i$) the electric and magnetic fields can be written as

$$E_{ex}^i = \frac{\partial \psi_e^i}{\partial x} e^{-j\beta_z z} \quad (16a)$$

$$E_{ey}^i = \frac{\partial \psi_e^i}{\partial y} e^{-j\beta_z z} \quad (16b)$$

$$E_{ez}^i = j \frac{(\beta_1^2 - \beta_z^2)}{\beta_z} \psi_e^i e^{-j\beta_z z} \quad (16c)$$

cont'd.

8.45 cont'd.

$$H_{ex}^i = - \frac{\omega \epsilon_i}{\beta_z} \frac{\partial \psi_e^i}{\partial y} e^{-j\beta_z z} \quad (16d)$$

$$H_{ey}^i = \frac{\omega \epsilon_i}{\beta_z} \frac{\partial \psi_e^i}{\partial x} e^{-j\beta_z z} \quad (16e)$$

$$H_{ez}^i = 0 \quad (16f)$$

where ψ_e^i satisfies the transverse wave equation of

$$\left(\frac{\partial^2 \psi_e^i}{\partial x^2} + \frac{\partial^2 \psi_e^i}{\partial y^2} \right) + (\beta_i^2 - \beta_z^2) \psi_e^i = \nabla_t^2 \psi_e^i + (\beta_i^2 - \beta_z^2) \psi_e^i = 0 \quad (17)$$

For the symmetrical modes which require the tangential magnetic field components to vanish along the symmetry plane of the waveguide ($x=0, 0 \leq y \leq b$), ψ_e^i takes the form of

$$\psi_e^d = \sum_{n=1}^{\infty} [A_{en}^d \cos(\beta_{xn}^d x) + B_{en}^d \sin(\beta_{xn}^d x)] [C_{en}^d \cos(\beta_{yn}^d y) + D_{en}^d \sin(\beta_{yn}^d y)] \quad (18a)$$

$$\psi_e^0 = \sum_{n=1}^{\infty} [A_{en}^0 \cos(\beta_{xn}^0 x) + B_{en}^0 \sin(\beta_{xn}^0 x)] [C_{en}^0 \cos(\beta_{yn}^0 (h-y)) + D_{en}^0 \sin(\beta_{yn}^0 (h-y))] \quad (18b)$$

Applying the appropriate boundary conditions along the metallic walls of the waveguide reduces the above scalar potentials to

$$\psi_e^d = \sum_{n=1}^{\infty} A_{en}^d \cos(\beta_{xn}^d x) \sinh(\alpha_n^d y) \quad (19a)$$

$$\psi_e^0 = \sum_{n=1}^{\infty} B_{en}^0 \cos(\beta_{xn}^0 x) \sinh[\alpha_n^0 (b-y)] \quad (19b)$$

$$\beta_{xn} = \frac{2\pi}{a} \left(n - \frac{1}{2} \right), \quad n=1, 2, 3, \dots \quad (19c)$$

$$\alpha_n^d = \sqrt{[(\beta_{xn})^2 + \beta_z^2] - (\beta_d)^2} \quad (19d)$$

$$\alpha_n^0 = \sqrt{[(\beta_{xn})^2 + \beta_z^2] - (\beta_0)^2} \quad (19e)$$

cont'd.

845 cont'd.

The coefficients A_{hn}^d , B_{hn}^0 , A_{en}^d and B_{en}^0 are yet unknowns, and their relations will be found by applying the boundary conditions along the air-dielectric interface.

The next step is to apply the appropriate boundary conditions along the air-dielectric interface including the center conductor. It should be stated again that the superposition of the TE^z and TM^z modes must satisfy the boundary along this interface. The appropriate boundary condition along the interface for $x>0$ that must be enforced are the following (the same apply for $x<0$):

$$E_{ez}^d(w/2 \leq x \leq a/2, y=h, z) = E_{ez}^0(w/2 \leq x \leq a/2, y=h, z) \quad (20a)$$

$$\begin{aligned} E_{hx}^d(w/2 \leq x \leq a/2, y=h, z) &+ E_{ex}^d(w/2 \leq x \leq a/2, y=h, z) \\ &= E_{hx}^0(w/2 \leq x \leq a/2, y=h, z) + E_{ex}^0(w/2 \leq x \leq a/2, y=h, z) \end{aligned} \quad (20b)$$

$$E_{ez}^d(0 \leq x \leq w/2, y=h, z) = E_{ez}^0(0 \leq x \leq w/2, y=h, z) \quad (20c)$$

$$\begin{aligned} H_{hx}^d(w/2 \leq x \leq a/2, y=h, z) &+ H_{ex}^d(w/2 \leq x \leq a/2, y=h, z) \\ &= H_{hx}^0(w/2 \leq x \leq a/2, y=h, z) + H_{ex}^0(w/2 \leq x \leq a/2, y=h, z) \end{aligned} \quad (20d)$$

$$\begin{aligned} E_{hx}^d(0 \leq x \leq w/2, y=h, z) &+ E_{ex}^d(0 \leq x \leq w/2, y=h, z) \\ &= E_{hx}^0(0 \leq x \leq w/2, y=h, z) + E_{ex}^0(0 \leq x \leq w/2, y=h, z) \end{aligned} \quad (20e)$$

$$H_{hz}^d(w/2 \leq x \leq a/2, y=h, z) = H_{hz}^0(w/2 \leq x \leq a/2, y=h, z) \quad (20f)$$

The boundary conditions of (20a) and (20b) are used, in conjunction with (14a), (14b), (19a) and (19b), to express B_{en}^0 and B_{hn}^0 each in terms of A_{en}^d and A_{hn}^d .

Applying (20a) leads to

Cont'd.

8.45 cont'd.

$$B_{en}^0 = A_{en}^d \frac{\beta_d^2 - \beta_z^2}{\beta_0^2 - \beta_z^2} \frac{\sinh(\alpha_n^{dh})}{\sinh[\alpha_n^0(b-h)]}$$

(21)

while (20b) reduces for $\mu_d = \mu_0$ to

$$B_{hn}^0 = \left[A_{en}^d \frac{\beta_{xn}\beta_z}{\omega\mu_0\alpha_n^0} \left\{ 1 - \frac{\beta_d^2 - \beta_z^2}{\beta_0^2 - \beta_z^2} \right\} - A_{hn}^d \frac{\alpha_n^d}{\alpha_n^0} \right] \frac{\sinh(\alpha_n^{dh})}{\sinh[\alpha_n^0(b-h)]}$$

Applying (20c) leads to

$$\sum_{n=1}^{\infty} A_{en}^d \cos(\beta_{xn}x) = 0$$

when $0 \leq x \leq w/2$

(23)

where

$$\bar{A}_{en}^d = A_{en}^d \sinh(\alpha_n^{dh})$$

(23a)

while (20d) reduces to

$$\sum_{n=1}^{\infty} \bar{A}_{en}^d \beta_{xn} P_n(\beta_z) \cos(\beta_{xn}x) - \sum_{n=1}^{\infty} \bar{A}_{hn}^d \beta_{xn} T_n(\beta_z) \cos(\beta_{xn}x) = 0$$

when $w/2 \leq x \leq a/2$

(24)

where

$$\bar{A}_{hn}^d = A_{hn}^d \frac{\omega\mu_0}{\beta_z} \sinh(\alpha_n^{dh})$$

(24a)

Cont'd.

8.45 cont'd.

$$P_n(\beta_z) = \epsilon_r \frac{\alpha_n^d}{\beta_{xn}} \coth(\alpha_n^d h) + \frac{\epsilon_r - \bar{\beta}_z^2}{1 - \bar{\beta}_z^2} \frac{\alpha_n^0}{\beta_{xn}} \coth[\alpha_n^0(b-h)] \\ + \bar{\beta}_z^2 \frac{\beta_{xn}}{\alpha_n^d} \frac{1 - \epsilon_r}{1 - \bar{\beta}_z^2} \coth[\alpha_n^0(b-h)] \quad (24b)$$

$$T_n(\beta_z) = \bar{\beta}_z^2 \left[\frac{\beta_{xn}}{\alpha_n^d} \coth(\alpha_n^d h) + \frac{\beta_{xn}}{\alpha_n^0} \cot[\alpha_n^0(b-h)] \right] \quad (24c)$$

$$\bar{\beta}_z = \frac{\beta_z}{\beta_0} = \frac{\beta_z}{\omega \sqrt{\mu_0 \epsilon_0}} = \frac{\lambda_0}{\lambda_g} = \sqrt{\epsilon_r}_{eff} = \frac{\nu_p}{\nu_0} \quad (24d)$$

Applying (8-214e) leads to

$$\sum_{n=1}^{\infty} \bar{A}_{en}^d \beta_{xn} \sin(\beta_{xn} x) - \sum_{n=1}^{\infty} \bar{A}_{hn}^d \beta_{xn} \sin(\beta_{xn} x) = 0 \quad (25)$$

when $0 \leq x \leq w/2$

while (20f) reduces to

$$\sum_{n=1}^{\infty} \bar{A}_{en}^d Q_n(\beta_z) \sin(\beta_{xn} x) - \sum_{n=1}^{\infty} \bar{A}_{hn}^d W_n(\beta_z) \sin(\beta_{xn} x) = 0 \quad (26)$$

when $w/2 \leq x \leq a/2$

where

$$Q_n(\beta_z) = \frac{\beta_{xn}}{\alpha_n^0} \frac{1 - \epsilon_r}{1 - \bar{\beta}_z^2} \coth[\alpha_n^0(b-h)] \quad (26a)$$

$$W_n(\beta_z) = \frac{\epsilon_r - \bar{\beta}_z^2}{1 - \bar{\beta}_z^2} \frac{\beta_{xn}}{\alpha_n^d} \coth(\alpha_n^d h) + \frac{\beta_{xn}}{\alpha_n^0} \cot[\alpha_n^0(b-h)] \quad (26b)$$

The objective now is to use (23)-(26b) to solve for

cont'd.

8.45 cont'd.

$\beta_z = \beta_z / \beta_0 = \sqrt{\epsilon_{r_{\text{eff}}}}$ of (24d). This can be accomplished by transforming

(23)-(26) into an infinite set of homogeneous simultaneous

equations for A_{en}^d and A_{hn}^d by using a conventional procedure such as the

singular integral equation of taking a scalar product with

a complete set of functions which are appropriate for the various

regions. By solving for the zeros of the determinant associated

with the above matrix equation, the values of β_z can be determined.

8.46

Given: $\epsilon_r = 6.8$, $\mu_r = 1$ (assumed), $w/h = 1.5$, $t/h = 0.01$

(a) Effective width-to-height ratio at zero frequency.

Since $w/h = 1.5 > \frac{1}{2\pi} = 0.159$,

$$\frac{w_{\text{eff}}(f=0)}{h} = \frac{w}{h} + \frac{1.25}{\pi} \cdot \frac{t}{h} \left[1 + \ln\left(\frac{2h}{t}\right) \right]$$

$$\frac{w_{\text{eff}}(f=0)}{h} = 1.5 + \frac{1.25}{\pi} \cdot (0.01) \cdot \left[1 + \ln\left(2 \cdot \frac{1}{0.01}\right) \right] = 1.525 \Rightarrow$$

$$\frac{w_{\text{eff}}(f=0)}{h} = 1.525$$

(b) Effective dielectric constant at zero frequency.

$$\frac{w_{\text{eff}}(f=0)}{h} = 1.525 > 1$$

$$\epsilon_{r,\text{eff}}(f=0) = \frac{\epsilon_r + 1}{2} + \frac{\epsilon_r - 1}{2} \left[1 + 12 \frac{h}{\omega_{\text{eff}}(f=0)} \right]^{\frac{1}{2}}$$

$$\epsilon_{r,\text{eff}}(f=0) = \frac{6.8 + 1}{2} + \frac{6.8 - 1}{2} \left[1 + 12 \cdot \left(\frac{1}{1.525} \right) \right]^{\frac{1}{2}} = 4.874 \Rightarrow$$

$$\epsilon_{r,\text{eff}}(f=0) = 4.874$$

(c) Characteristic impedance at zero frequency.

$$Z_c(f=0) = \frac{\frac{120\pi}{\sqrt{\epsilon_{r,\text{eff}}(f=0)}}}{\frac{\omega_{\text{eff}}(f=0)}{h} + 1.393 + 0.667 \cdot \ln\left[\frac{\omega_{\text{eff}}(f=0)}{h} + 1.444\right]}$$

$$Z_c(f=0) = \frac{\frac{120\pi}{\sqrt{4.874}}}{1.525 + 1.393 + 0.667 \cdot \ln[1.525 + 1.444]} = 46.861 \text{ Ohms} \Rightarrow Z_c(f=0) = 46.861 \text{ Ohms}$$

(Problem 8.46 cont.)

(d) Approximate frequency where dispersion will begin when $h = 0.05$

$$f_c \geq 0.3 \sqrt{\frac{Z_c(0)}{h} \cdot \frac{1}{\sqrt{\epsilon_r - 1}}} \times 10^9 = 0.3 \sqrt{\frac{46.861 \text{ Ohms}}{0.05 \text{ cm}} \cdot \frac{1}{\sqrt{6.8 - 1}}} \times 10^9 = 5.918 \times 10^9 \text{ Hz}$$

Approximate frequency where dispersion will begin for $h = 0.05 \text{ cm}$ is \Rightarrow

$$f_c \geq 5.918 \text{ GHz}$$

(e) Effective dielectric constant at 15 GHz

$$\epsilon_{r,\text{eff}}(f) = \epsilon_r - \left[\frac{\epsilon_r - \epsilon_{r,\text{eff}}(0)}{1 + \frac{\epsilon_{r,\text{eff}}(0)}{\epsilon_r} \left(\frac{f}{f_t} \right)^2} \right] \Rightarrow f_t = \frac{Z_c(0)}{2\mu_0 h} = \frac{46.861 \text{ Ohms}}{2 \cdot (4\pi \times 10^{-7})(0.0005 \text{ cm})} = 37.29 \times 10^9 \text{ Hz}$$

$$\epsilon_{r,\text{eff}}(15 \text{ GHz}) = 6.8 - \left[\frac{6.8 - 4.874}{1 + \frac{4.874}{6.8} \left(\frac{15 \text{ GHz}}{37.29 \text{ GHz}} \right)^2} \right] = 5.074 \Rightarrow \epsilon_{r,\text{eff}}(15 \text{ GHz}) = 5.074$$

(f) Characteristic impedance at 15 GHz. Compare with the value if dispersion is neglected.

$$Z_c(f) = Z_c(0) \sqrt{\frac{\epsilon_{r,\text{eff}}(0)}{\epsilon_{r,\text{eff}}(f)}}$$

$$Z_c(f = 15 \text{ GHz}) = (46.861 \text{ Ohms}) \sqrt{\frac{4.874}{5.074}} = 45.928 \text{ Ohms} \Rightarrow Z_c(f = 15 \text{ GHz}) = 45.928 \text{ Ohms}$$

If dispersion is neglected, $Z_c(f = 15 \text{ GHz}) = Z_c(f = 0 \text{ Hz}) = 46.861 \text{ Ohms}$.

The characteristic impedance calculation shows no frequency dependence if dispersion neglected.

(g) Phase velocity at 15 GHz. Compare with the value if dispersion is neglected

$$v_p(f) = \frac{c}{\sqrt{\mu_r \cdot \epsilon_{r,\text{eff}}(f)}}$$

$$v_p(f = 15 \text{ GHz}) = \frac{3 \times 10^8 \text{ m/s}}{\sqrt{(1) \cdot (5.074)}} = 1.332 \times 10^8 \text{ m/s} \Rightarrow v_p(f = 15 \text{ GHz}) = 1.332 \times 10^8 \text{ m/s}$$

If dispersion is neglected, then $\epsilon_{r,\text{eff}}(f = 15 \text{ GHz}) = \epsilon_{r,\text{eff}}(f = 0 \text{ Hz}) = 4.874$

(Problem 8A6 cont.)

$$v_p(f = 0 \text{ GHz}) = \frac{3 \times 10^8 \text{ m/s}}{\sqrt{(1) \cdot (4.874)}} = 1.359 \times 10^8 \text{ m/s} \Rightarrow v_p(f = 15 \text{ GHz}) = 1.359 \times 10^8 \text{ m/s}$$

Dispersion Neglected

$$v_p = \frac{3 \times 10^8 \text{ m/s}}{\sqrt{(1) \cdot (6.8)}} = 1.15 \times 10^8 \text{ m/s} \Rightarrow v_p = 1.15 \times 10^8 \text{ m/s} \text{ (no fringing)}$$

(h) Guide wavelength at 15 GHz.

$$\lambda_g(f) = \frac{c}{f \sqrt{\mu_r \cdot \epsilon_{r,\text{eff}}(f)}}$$

$$\lambda_g(f = 15 \text{ GHz}) = \frac{3 \times 10^8 \text{ m/s}}{15 \times 10^9 \text{ GHz} \sqrt{(1) \cdot (5.074)}} = 8.879 \times 10^{-3} \text{ meters}$$

$$\lambda_g(f = 15 \text{ GHz}) = 0.8879 \text{ cm}$$

CHAPTER 9

Q.1

Given: $a = 1.12 \text{ cm}$, $\epsilon_r = ?$, $\mu_r = 1$, $BW = 1.5 \text{ GHz}$ (over a single dominant mode)

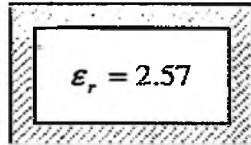
Dominant mode for a circular waveguide: TE_{11}^z with $\chi_{11} = 1.8412$

Next higher-order mode: TM_{01}^z with $\chi_{01} = 2.4049$

BW of the TE_{11}^z mode is: $(f_c)_{01}^{TM^z} - (f_c)_{11}^{TE^z}$

$$BW = (f_c)_{01}^{TM^z} - (f_c)_{11}^{TE^z} = \frac{\chi_{01}}{2\pi \cdot a \sqrt{\mu\epsilon}} - \frac{\chi_{11}}{2\pi \cdot a \sqrt{\mu\epsilon}} = \frac{\chi_{01} - \chi_{11}}{2\pi \cdot a \sqrt{\mu\epsilon}} = \frac{(\chi_{01} - \chi_{11}) \cdot c}{2\pi \cdot a \sqrt{\mu_r \epsilon_r}}$$

$$\epsilon_r = \left[\frac{(\chi_{01} - \chi_{11}) \cdot c}{2\pi \cdot a \cdot BW \sqrt{\mu_r}} \right]^2 = \left[\frac{(2.4049 - 1.8412) \cdot 30 \times 10^9 \text{ cm/s}}{2\pi \cdot (1.12 \text{ cm}) \cdot 1.5 \times 10^9 \text{ Hz} \sqrt{1}} \right]^2 = 2.57$$



Lower Frequency for the TE_{11}^z mode:

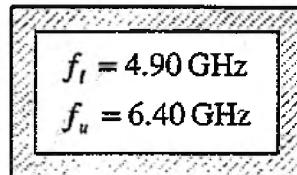
$$f_l = (f_c)_{11}^{TE^z} = \frac{\chi_{11}}{2\pi \cdot a \sqrt{\mu\epsilon}} = \frac{\chi_{11} \cdot c}{2\pi \cdot a \sqrt{\mu_r \epsilon_r}} = \frac{(1.8412) \cdot (30 \times 10^9 \text{ cm/s})}{2\pi \cdot (1.12 \text{ cm}) \sqrt{(1) \cdot (2.57)}} = 4.90 \text{ GHz}$$

Upper Frequency for the TE_{11}^z mode:

$$f_u = (f_c)_{01}^{TM^z} = \frac{\chi_{01} \cdot c}{2\pi \cdot a \sqrt{\mu_r \epsilon_r}} = \frac{(2.4049) \cdot (30 \times 10^9 \text{ cm/s})}{2\pi \cdot (1.12 \text{ cm}) \sqrt{(1) \cdot (2.57)}} = 6.40 \text{ GHz}$$

or

$$f_u = f_l + 1.5 \text{ GHz} = (6.21 + 1.5) \text{ GHz} = 6.40 \text{ GHz}$$



9.2

$$a = 2 \text{ cm}, L = 5 \text{ cm}, f = 6 \text{ GHz}$$

$$\beta_2 = \beta \sqrt{1 - \left(\frac{f_c}{f}\right)^2}, \quad f_c = \frac{\lambda_0'}{2\pi a \sqrt{\mu_0 \epsilon_0}} = \frac{1.8412}{2\pi a \sqrt{\mu_0 \epsilon_0}}$$

$$(f_{c1}) = \frac{1.8412}{2\pi a \sqrt{\mu_0 \epsilon_0} \epsilon_{r1}} = \frac{1.8412 \times 30 \times 10^9}{2\pi (2) \sqrt{2.56}} = \frac{1.8412 (30) \times 10^9}{4\pi (1.6)} = 2.7472 \times 10^9$$

$$(f_{c0}) = \frac{1.8412}{2\pi a \sqrt{\mu_0 \epsilon_0}} = \frac{1.8412 \times 30 \times 10^9}{4\pi} = 4.3955 \times 10^9$$

$$\beta_{21} = \beta_1 \sqrt{1 - \left(\frac{f_{c1}}{f}\right)^2} = \omega \sqrt{\mu_0 \epsilon_0 \epsilon_{r1}} \sqrt{1 - \left(\frac{f_{c1}}{f}\right)^2} = \beta_0 \sqrt{\epsilon_{r1}} \sqrt{1 - \left(\frac{f_{c1}}{f}\right)^2}$$

$$\beta_{21} = \frac{2\pi}{\lambda_0} \sqrt{\epsilon_{r1}} \sqrt{1 - \left(\frac{2.7472}{6}\right)^2} = \frac{2\pi}{\lambda_0} (1.6) \sqrt{1 - (0.45707)^2} = \frac{2\pi}{\lambda_0} (1.6) (0.88902)$$

$$\beta_{20} = \beta_0 \sqrt{1 - \left(\frac{f_{c0}}{f}\right)^2} = \frac{2\pi}{\lambda_0} \sqrt{1 - \left(\frac{4.3955}{6}\right)^2} = \frac{2\pi}{\lambda_0} \sqrt{1 - (0.73258)^2} = \frac{2\pi}{\lambda_0} (0.68068)$$

$$\Delta \beta = \beta_{21} - \beta_0 = \frac{2\pi}{\lambda_0} (1.6) (0.88902) - \frac{2\pi}{\lambda_0} (0.68068) = \frac{2\pi}{\lambda_0} [1.42243 - 0.68068] \\ = \frac{2\pi}{5} (1.42243 - 0.68068) = 1.78748 - 0.88537 = 0.93211 \text{ rad/cm}$$

$$\Delta \beta = \frac{2\pi}{\lambda_0} (0.74175), \quad \lambda_0 = \frac{30 \times 10^9}{6 \times 10^9} = 5 \text{ cm}$$

$$\Delta \phi = \Delta \beta = (5) \frac{2\pi}{\lambda_0} (0.74175) = (5) \frac{2\pi}{5} (0.74175) = 4.66058 \text{ radians} \\ \Delta \phi = \Delta \phi_1 - \Delta \phi_0 = (8.93739 - 4.27684) = 512.075^\circ - 245.0448^\circ$$

$$\Delta \phi = 267.03167^\circ = 4.66058 \text{ radians}$$

$$\Delta \phi = \Delta \phi - \Delta \phi_0 = 512.075^\circ - 245.0448^\circ$$

9.3

$a = 2 \text{ cm}$, $\epsilon_r = 2.25$, $f = 3.5 \text{ GHz}$, delay = 2 microseconds

Dominant mode: $TE_{11} \Rightarrow (f_c)_{11}^{TE} = \frac{\pi c}{2\pi a \sqrt{\mu_r \epsilon_r}} = \frac{\pi c}{2\pi a \sqrt{\mu_r \epsilon_r}}$

From Table 9-1 $\Rightarrow \gamma_1' = 1.8412$

$$(f_c)_{11}^{TE} = \frac{1.8412(30 \times 10^9)}{2\pi(2) \sqrt{2.25}} = 2.93084 \text{ GHz}$$

$$V_g U_p = V^2 \Rightarrow V_g = \frac{V^2}{U_p}$$

$$\beta_2 = \frac{2\pi}{\gamma_g} \Rightarrow \gamma_g = \frac{2\pi}{\beta_2} \Rightarrow U_p = \gamma_g f = \frac{\gamma f}{\sqrt{1 - \left(\frac{f_c}{f}\right)^2}} = \frac{v}{\sqrt{1 - \left(\frac{f_c}{f}\right)^2}}$$

$$V_g = \frac{V^2}{U_p} = \frac{V^2}{v} \sqrt{1 - \left(\frac{f_c}{f}\right)^2} = v \sqrt{1 - \left(\frac{f_c}{f}\right)^2}$$

$$v = \frac{V_0}{\sqrt{\epsilon_r}} = \frac{3 \times 10^8}{\sqrt{2.25}} = 2 \times 10^8 \text{ m/sec.}$$

$$V_g = 2 \times 10^8 \sqrt{1 - \left(\frac{2.93084}{3.5}\right)^2} = 2 \times 10^8 \sqrt{1 - (0.83738)^2} = 2 \times (0.54662 \times 10^8)$$

$$V_g = 1.09324 \times 10^8 \text{ m/sec}$$

$$d = V_g t_d = (1.09324 \times 10^8) 2 \times 10^{-6} = 2.18647 \times 10^2$$

$d = 218.647 \text{ meters}$

9.4

$$\Delta\phi = \beta_z \Delta l \Rightarrow \Delta l = \Delta\phi / \beta_z, \quad \beta_z = \beta \sqrt{1 - (\frac{f_c}{f})^2}$$

$$(f_c)_{TE}^{TE} = \frac{x_1}{2\pi a \sqrt{\epsilon_0 \epsilon_r}} = \frac{1.84(30 \times 10^9) / \sqrt{2.25}}{2\pi(2)} = 2.93 \text{ GHz}$$

$$\begin{aligned}\beta_z &= \beta \sqrt{1 - (\frac{f_c}{f})^2} = \frac{2\pi}{\lambda_0} \sqrt{1 - (\frac{f_c}{f})^2} = \frac{2\pi \sqrt{\epsilon_r}}{\lambda_0} \sqrt{1 - (\frac{f_c}{f})^2} = \frac{2\pi \sqrt{2.25}}{5} \sqrt{1 - (\frac{2.93}{6})^2} \\ &= \frac{2\pi(1.5)}{5} (0.8727) = 1.6449 \text{ rad/cm} = 94.247^\circ/\text{cm}\end{aligned}$$

$$\Delta l = \frac{\Delta\phi}{\beta_z} = \frac{180}{94.247} = 1.9099 \text{ cm}$$

9.5

Vector Potential Component:

For a Circular Waveguide,

$$F_z(\rho, \phi, z) = [A_1 J_m(\beta_\rho \rho) + B_1 Y_m(\beta_\rho \rho)] \cdot [C_2 \cos(m\phi) + D_2 \sin(m\phi)] \cdot [A_3 e^{-j\beta_z z} + B_3 e^{+j\beta_z z}]$$

$$\beta_\rho^2 + \beta_z^2 = \beta^2$$

$f_1(\rho) = A_1 J_m(\beta_\rho \rho) + B_1 Y_m(\beta_\rho \rho)$ represent standing waves in the $\pm \rho$ directions.

$g_2(\phi) = C_2 \cos(m\phi) + D_2 \sin(m\phi)$ represent periodic waves in the $\pm \phi$ directions.

$h_3(z) = A_3 e^{-j\beta_z z} + B_3 e^{+j\beta_z z}$ represent traveling waves in the $\pm z$ directions.

Find the constants, $A_1, B_1, C_2, D_2, A_3, B_3, m, \beta_\rho$, and β_z using the following conditions...

- 1) $E_\phi(\rho = a, \phi, z) = 0$: Tangential component of the Electric field at the waveguide wall must be zero. $\cancel{E_\phi(\phi = \pi, z)}$
 - 2) $E_\rho(\rho = 0, z) = 0$: Tangential component of the Electric field at the conducting baffle must be zero.
 - 3) The fields must be finite everywhere.
- ~~4) The fields must repeat every 2π radians in ϕ .~~

According to 3), $B_1 = 0$ since $Y_m[\rho = 0] = -\infty$.According to 4), $m = 1, 2, 3, \dots$ Consider only waves that propagate in the $+z$ direction. Then $B_3 = 0$.

Prob 9-5 Cont.

$$F_z^+(\rho, \phi, z) = A_1 A_3 J_m(\beta_\rho \rho) \cdot [C_2 \cos(m\phi) + D_2 \sin(m\phi)] \cdot e^{-j\beta_z z}, \text{ with } \beta_\rho^2 + \beta_z^2 = \beta^2$$

Apply BC # 1) $E_\phi(\rho = a, \phi, z) = 0$:

$$E_\phi^+(\rho, \phi, z) = \frac{1}{\epsilon} \frac{\partial}{\partial \rho} F_z^+(\rho, \phi, z) = \beta_\rho \frac{A_1 A_3}{\epsilon} J_m'(\beta_\rho \rho) \cdot [C_2 \cos(m\phi) + D_2 \sin(m\phi)] \cdot e^{-j\beta_z z}$$

$$E_\phi^+(\rho = a, \phi, z) = \beta_\rho \frac{A_1 A_3}{\epsilon} J_m'(\beta_\rho a) \cdot [C_2 \cos(m\phi) + D_2 \sin(m\phi)] \cdot e^{-j\beta_z z} = 0$$

This is true only for $J_m'(\beta_\rho a) = 0 \Rightarrow \beta_\rho a = \chi_{mn} \Rightarrow \boxed{\beta_\rho = \frac{\chi_{mn}}{a}}$

where, χ_{mn} represents the nth zero ($n = 1, 2, 3, \dots$) of the derivative of the Bessel function J_m of the first kind of order m ($m = 0, 1, 2, 3, \dots$).

Apply BC # 2) $E_\rho(\rho, \phi = 0, z) = 0$:

$$E_\rho^+(\rho, \phi, z) = -\frac{1}{\epsilon \rho} \frac{\partial}{\partial \phi} F_z^+(\rho, \phi, z) = -m \frac{A_1 A_3}{\epsilon \rho} J_m(\beta_\rho \rho) \cdot [-C_2 \sin(m\phi) + D_2 \cos(m\phi)] \cdot e^{-j\beta_z z}$$

$$E_\rho^+(\rho, \phi = 0, z) = -m \frac{A_1 A_3}{\epsilon \rho} J_m(\beta_\rho \rho) \cdot [-C_2 \sin(0) + D_2 \cos(0)] \cdot e^{-j\beta_z z} = 0$$

$$E_\rho^+(\rho, \phi = 0, z) = m \frac{A_1 A_3}{\epsilon \rho} J_m(\beta_\rho \rho) \cdot [D_2] \cdot e^{-j\beta_z z} = 0 \quad \therefore \quad D_2 = 0$$

$$E_\rho^+(\phi, \phi = 2\pi, z) = m \frac{A_1 A_3}{\epsilon \rho} J_m(\beta_\rho \rho) \left[-C_2 \sin(2m\pi) \right] = \sin(2m\pi) = 0 \Rightarrow 2m\pi = \sin^{-1}(0) = p\pi$$

Finally, the simplified expression for the Electric Vector Potential...

$$F_z^+(\rho, \phi, z) = A_1 A_3 C_2 \cdot J_m(\beta_\rho \rho) \cdot [\cos(m\phi)] \cdot e^{-j\beta_z z}$$

$$\boxed{\begin{aligned} m &= \frac{p\pi}{2}, \quad p = 0, \pm 1, 2, 3 \\ m &= 0, \frac{1}{2}, 1, \frac{3}{2}, \dots \\ n &= 1, 2, 3, 4, \dots \end{aligned}}$$

$$\boxed{F_z^+(\rho, \phi, z) = A_{mn} J_m(\beta_\rho \rho) \cdot [\cos(m\phi)] \cdot e^{-j\beta_z z}}$$

with $\beta_\rho^2 + \beta_z^2 = \beta^2$ and $\beta_\rho = \frac{\chi_{mn}}{a}$, $m = 0, \frac{1}{2}, 1, \frac{3}{2}, \dots$

Prob 9-5 Cont

$$F_z^+(\rho, \phi, z) = A_{mn} J_m(\beta_\rho \rho) \cdot [\cos(m\phi)] \cdot e^{-j\beta_z z}$$

$$E_\rho^+(\rho, \phi, z) = -\frac{1}{\epsilon \rho} \frac{\partial F_z^+}{\partial \phi} = A_{mn} \frac{m}{\epsilon \rho} J_m(\beta_\rho \rho) \cdot [\sin(m\phi)] \cdot e^{-j\beta_z z}$$

$$E_\phi^+(\rho, \phi, z) = \frac{1}{\epsilon} \frac{\partial F_z^+}{\partial \rho} = A_{mn} \frac{\beta_\rho}{\epsilon} J_m(\beta_\rho \rho) \cdot [\cos(m\phi)] \cdot e^{-j\beta_z z}$$

$$E_z^+(\rho, \phi, z) = 0$$

$$H_\rho^+(\rho, \phi, z) = -j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 F_z^+}{\partial \rho \cdot \partial z} = -A_{mn} \frac{\beta_\rho \beta_z}{\omega \mu \epsilon} J_m(\beta_\rho \rho) \cdot [\cos(m\phi)] \cdot e^{-j\beta_z z}$$

$$H_\phi^+(\rho, \phi, z) = -j \frac{1}{\omega \mu \epsilon} \frac{1}{\rho} \frac{\partial^2 F_z^+}{\partial \phi \cdot \partial z} = A_{mn} \frac{m \beta_z}{\omega \mu \epsilon} \frac{1}{\rho} J_m(\beta_\rho \rho) \cdot [\sin(m\phi)] \cdot e^{-j\beta_z z}$$

$$H_z^+(\rho, \phi, z) = -j \frac{1}{\omega \mu \epsilon} \left(\frac{\partial^2}{\partial z^2} + \beta^2 \right) F_z^+ = -j A_{mn} \frac{\beta_\rho^2}{\omega \mu \epsilon} J_m(\beta_\rho \rho) \cdot [\cos(m\phi)] \cdot e^{-j\beta_z z}$$

Cutoff Frequency

$$\beta_\rho^2 + \beta_z^2 = \beta^2 \quad \Leftarrow \quad \beta_\rho = \frac{\chi_{mn}}{a}$$

$$\beta_\rho^2 + 0 = \beta_c^2 \Rightarrow \beta_c = \omega_c \sqrt{\mu \epsilon} = 2\pi \cdot f_c \sqrt{\mu \epsilon} = \beta_\rho = \frac{\chi_{mn}}{a}$$

$$f_c = \frac{\chi_{mn}}{2\pi \cdot a \cdot \sqrt{\mu \epsilon}} = \frac{\chi_{mn}}{2\pi \cdot a \cdot \sqrt{\mu_0 \epsilon_0} \sqrt{\mu_r \epsilon_r}} = \frac{\chi_{mn} \cdot c}{2\pi \cdot a \cdot \sqrt{\mu_r \epsilon_r}}$$

- (a) The cutoff frequencies of the three lowest-order propagating modes in order of ascending cutoff frequency when the radius of the cylinder is 1 cm are:

$$(\text{TE}_1^z)_z : \quad (f_c)_{1z} = \frac{\chi_1^z \cdot c}{2\pi \cdot a \cdot \sqrt{\mu_r \epsilon_r}} = \frac{(1.1655) \cdot (30 \times 10^9 \text{ cm/s})}{2\pi \cdot (1 \text{ cm}) \cdot \sqrt{(1)} \cdot (1)} = 5.565 \text{ GHz}$$

$$(\text{TE}_2^z)_y : \quad (f_c)_{2y} = \frac{\chi_2^z \cdot c}{2\pi \cdot a \cdot \sqrt{\mu_r \epsilon_r}} = \frac{(1.8412) \cdot (30 \times 10^9 \text{ cm/s})}{2\pi \cdot (1 \text{ cm}) \cdot \sqrt{(1)} \cdot (1)} = 8.791 \text{ GHz}$$

$$(\text{TE}_3^z)_y : \quad (f_c)_{3y} = \frac{\chi_3^z \cdot c}{2\pi \cdot a \cdot \sqrt{\mu_r \epsilon_r}} = \frac{(2.4605) \cdot (30 \times 10^9 \text{ cm/s})}{2\pi \cdot (1 \text{ cm}) \cdot \sqrt{(1)} \cdot (1)} = 11.748 \text{ GHz}$$

Prob 9-5 Cont

$(f_c)_{\frac{1}{2^1}} = 5.565 \text{ GHz}$
$(f_c)_{\frac{1}{2^2}} = 8.791 \text{ GHz}$
$(f_c)_{\frac{1}{2^3}} = 11.748 \text{ GHz}$

Phase Constant along the axis of the guide:

$$\beta_\rho^2 + \beta_z^2 = \beta^2 \Rightarrow \beta_z^2 = \beta^2 - \beta_\rho^2 = \beta^2 - \left(\frac{\chi_{mn}}{a} \right)^2 = \beta^2 \left[1 - \left(\frac{\chi_{mn}}{a \cdot \beta} \right)^2 \right]$$

$$f_c = \frac{\chi_{mn}}{2\pi \cdot a \cdot \sqrt{\mu \epsilon}} = \frac{\chi_{mn} \cdot v}{2\pi \cdot a} = \frac{\chi_{mn}}{2\pi \cdot a} \cdot \lambda f = \frac{\chi_{mn}}{2\pi \cdot a} \cdot \left(\frac{2\pi}{\beta} \right) f \Rightarrow \frac{f_c}{f} = \frac{\chi_{mn}}{\beta \cdot a}$$

$$(\beta_z)_{mn} = \beta \sqrt{1 - \left(\frac{\chi_{mn}}{a \cdot \beta} \right)^2} \Rightarrow \boxed{\beta_z = \beta \sqrt{1 - \left(\frac{f_c}{f} \right)^2}} \quad f > f_c = (f_c)_{mn}$$

Guide Wavelength:

$$(\lambda_g)_{mn} = \frac{2\pi}{(\beta_z)_{mn}} = \frac{2\pi}{\beta \sqrt{1 - \left(\frac{f_c}{f} \right)^2}} \Rightarrow \boxed{(\lambda_g)_{mn} = \frac{\lambda}{\sqrt{1 - \left(\frac{f_c}{f} \right)^2}}} \quad f > f_c = (f_c)_{mn}$$

where, λ is the wavelength of the wave in an infinite medium which exists inside the waveguide.

Guide Wavelength:

$$(Z_w^{+z})_{mn}^{\text{TE}} = \frac{E_\rho^+}{H_\phi^+} = \frac{\omega \mu}{(\beta_z)_{mn}} = \frac{\omega \mu}{\beta \sqrt{1 - \left(\frac{f_c}{f} \right)^2}} = \frac{\omega \mu}{\omega \sqrt{\mu \epsilon} \sqrt{1 - \left(\frac{f_c}{f} \right)^2}} = \frac{\sqrt{\frac{\mu}{\epsilon}}}{\sqrt{1 - \left(\frac{f_c}{f} \right)^2}}$$

$$\boxed{(Z_w^{+z})_{mn}^{\text{TE}} = \frac{\eta_0 \sqrt{\frac{\mu_r}{\epsilon_r}}}{\sqrt{1 - \left(\frac{f_c}{f} \right)^2}}} \quad f > f_c = (f_c)_{mn}$$

Prob 9-5 Cont

(b) The wave impedance and guide wavelength (in cm) for the lowest-order mode at $f = 1.5 f_c$ where f is the cutoff frequency of the lowest-order mode.

$$(\lambda_g)_{mn} = \frac{\lambda}{\sqrt{1 - \left(\frac{f_c}{f}\right)^2}} \quad \lambda = \frac{v}{f} = \frac{1}{\sqrt{\mu\epsilon}} \cdot \frac{1}{1.5f_c} = \frac{c}{\sqrt{\mu_r\epsilon_r}} \cdot \frac{1}{1.5f_c}$$

$$(\lambda_g)_{\frac{1}{2},1} = \frac{\frac{c}{\sqrt{\mu_r\epsilon_r}} \cdot \frac{1}{1.5f_c}}{\sqrt{1 - \left(\frac{f_c}{f}\right)^2}} = \frac{30 \times 10^9 \text{ cm/s}}{(1.5) \cdot (5.565 \times 10^9 \text{ Hz}) \sqrt{1 - \left(\frac{1}{1.5}\right)^2}} = 4.822 \text{ cm}$$

$$(\lambda_g)_{\frac{1}{2},1} = 4.822 \text{ cm}$$

$$(Z_w^{+z})_{\frac{1}{2},1}^{\text{TE}} = \frac{(120 \cdot \pi) \cdot \sqrt{\frac{1}{1}}}{\sqrt{1 - \left(\frac{1}{1.5}\right)^2}} = 505.77 \Omega$$

$$(Z_w^{+z})_{\frac{1}{2},1}^{\text{TE}} = 505.77 \Omega$$

9.6

$$A_z = [C_1 J_m(\beta_p \rho) + D_1 Y_m(\beta_p \rho)] [C_2 \cos(m\phi) + D_2 \sin(m\phi)] e^{j\beta_z z}$$

$$A_z = C_1 J_m(\beta_p \rho) [C_2 \cos(m\phi) + D_2 \sin(m\phi)] e^{j\beta_z z}$$

$$\text{B.C.'s: } E_z(\phi=0) = E_z(\phi=2\pi) = 0$$

$$E_z(\rho=a) = 0$$

$$E_z = -j \frac{1}{\omega \mu \epsilon} \left(\frac{\partial^2}{\partial z^2} + \beta^2 \right) A_z = -j \frac{C_1}{\omega \mu \epsilon} (\beta^2 - \beta_z^2) J_m(\beta_p \rho) [C_2 \cos(m\phi) + D_2 \sin(m\phi)] e^{j\beta_z z}$$

$$E_z(\phi=0) = -j \frac{C_1}{\omega \mu \epsilon} (\beta^2 - \beta_z^2) J_m(\beta_p \rho) [C_2(1) + D_2(0)] e^{-j\beta_z z} = 0 \Rightarrow C_2 = 0$$

$$E_z(\phi=2\pi) = -j \frac{C_1}{\omega \mu \epsilon} (\beta^2 - \beta_z^2) J_m(\beta_p \rho) D_2 \sin(2m\pi) = 0 \Rightarrow 2m\pi = \sin^{-1}(0) = -p\pi, p=1, 2, \dots$$

$$m = \frac{p}{2}, p=1, 2, 3, \dots \Rightarrow m = \frac{1}{2}, 1, \frac{3}{2}, 2, \dots$$

$$A_z = B_{mn} J_m(\beta_p \rho) \sin(m\phi) e^{-j\beta_z z}, \quad m = \frac{1}{2}, 1, \frac{3}{2}, 2, \dots$$

cont'd.

$$9.6 \text{ cont'd.} \quad E_z(\rho=a) = -j B_{mn} \frac{\beta^2 - \beta_p^2}{w\mu\varepsilon} J_m(\beta_p\rho) e^{-j\beta_p z} \Big|_{\rho=a}$$

$$E_z(\rho=a) = -j B_{mn} \frac{\beta^2 - \beta_p^2}{w\mu\varepsilon} J_m(\beta_p a) \sin(m\phi) e^{-j\beta_p a^2} = 0 \Rightarrow J_m(\beta_p a) = 0$$

$$\beta_p a = \lambda_{mn} \Rightarrow \beta_p = \frac{\lambda_{mn}}{a}, \quad m = \frac{1}{2}, 1, \frac{3}{2}, 2, \dots$$

$$E_\theta = -j \frac{1}{w\mu\varepsilon} \frac{\partial^2 A_\theta}{\partial \rho \partial z} = -B_{mn} \frac{\beta_2 \beta_0}{w\mu\varepsilon} J_m'(\beta_p \rho) \sin(m\phi) e^{-j\beta_p z^2}, \quad H_\theta = \frac{1}{\rho \beta_p} \frac{\partial A_\theta}{\partial \rho} = B_{mn} \frac{m}{\beta_p} J_m(\beta_p \rho) \cos(m\phi) e^{-j\beta_p z^2}$$

$$E_\phi = -j \frac{1}{w\mu\varepsilon} \frac{1}{\rho} \frac{\partial^2 A_\phi}{\partial \theta \partial z} = -B_{mn} \frac{m \beta_2}{w\mu\varepsilon} \frac{1}{\rho} J_m(\beta_p \rho) \cos(m\phi) e^{-j\beta_p z^2}, \quad H_\phi = -\frac{1}{\rho} \frac{\partial A_\phi}{\partial \theta} = -B_{mn} \frac{\beta_0}{\rho} J_m'(\beta_p \rho) \sin(m\phi) e^{-j\beta_p z^2}$$

$$E_z = -j \frac{1}{w\mu\varepsilon} \left(\frac{\partial^2}{\partial \rho^2} + \rho^2 \right) A_z = -j B_{mn} \frac{\beta_p^2}{w\mu\varepsilon} J_m(\beta_p \rho) \sin(m\phi) e^{-j\beta_p z^2}, \quad H_z = 0$$

$$\beta_z = \begin{cases} \sqrt{\beta^2 - \beta_p^2} = \sqrt{\beta^2 - (\frac{\lambda_{mn}}{a})^2}, & \beta > \beta_p = \lambda_{mn}/a \\ 0, & \beta = \beta_p \\ -j \sqrt{\beta_p^2 - \beta^2} = -j \sqrt{(\frac{\lambda_{mn}}{a})^2 - \beta^2}, & \beta < \beta_p = \lambda_{mn}/a \end{cases}$$

$$(f_c)_{mn} = \frac{\lambda_{mn}}{2\pi\sqrt{\mu\varepsilon}}, \quad m = \frac{1}{2}, 1, \frac{3}{2}, 2, \dots$$

$$\beta_z = \frac{2\pi}{\lambda_p} \Rightarrow \lambda_p = \frac{2\pi}{\beta_z} = \frac{2\pi}{\sqrt{\beta^2 - \beta_p^2}} = \frac{2\pi}{\rho \sqrt{1 - (\frac{f_c}{f})^2}} = \frac{2}{\sqrt{1 - (\frac{f_c}{f})^2}}$$

$$Z_N = \frac{E_\theta}{H_\theta} = -\frac{E_\phi}{H_\theta} = \frac{\beta_p}{w\varepsilon} = \frac{w\sqrt{\mu\varepsilon}}{w\varepsilon} \sqrt{1 - (\frac{f_c}{f})^2} = \sqrt{\frac{\mu}{\varepsilon}} \sqrt{1 - (\frac{f_c}{f})^2}$$

a.

$$m = \frac{1}{2}, n = 1: \quad \lambda_{\frac{1}{2},1} = 3.1416 \Rightarrow (f_c)_{\frac{1}{2},1} = \frac{3.1416(30)}{2\pi(1)} \times 10^9 = 15 \times 10^9 = 15 \text{ GHz}$$

$$m = 1, n = 1: \quad \lambda_{1,1} = 3.8318 \Rightarrow (f_c)_{1,1} = \frac{3.8318(30)}{2\pi(1)} \times 10^9 = 18.295 \times 10^9 = 18.295 \text{ GHz}$$

$$m = \frac{3}{2}, n = 1: \quad \lambda_{\frac{3}{2},1} = 4.4934 \Rightarrow (f_c)_{\frac{3}{2},1} = \frac{4.4934(30)}{2\pi(1)} \times 10^9 = 21.4544 \times 10^9 = 21.4544 \text{ GHz}$$

$$b. \quad \sqrt{1 - (\frac{f_c}{f})^2} = \sqrt{1 - (\frac{f_c}{1.5 f_c})^2} = 0.7453$$

$$Z_N = \eta \sqrt{1 - (\frac{f_c}{f})^2} = 377 (0.7453) =$$

$$\lambda_p = \frac{2}{\sqrt{1 - (\frac{f_c}{f})^2}} = \frac{30 \times 10^9 / (1.5 \times 15 \times 10^9)}{0.7453} = 1.789 \text{ cm}$$

9.7

TE²

$$F_z = [A_1 J_m(\beta_p p) + \cancel{B_1 Y_m(\beta_p p)}] [C_2 \cos(m\phi) + D_2 \sin(m\phi)] e^{-j\beta_z z}$$

$$E_p = -\frac{1}{\epsilon} \frac{1}{p} \frac{\partial F_z}{\partial \phi} = -\frac{A_1 m}{\epsilon p} J_m'(\beta_p p) [-C_2 \sin(m\phi) + D_2 \cos(m\phi)] e^{-j\beta_z z}$$

$$E_p(\phi=0) = -\frac{A_1 m}{\epsilon p} J_m(\beta_p p) [-C_2(0) + D_2(1)] e^{-j\beta_z z} = 0 \Rightarrow D_2 = 0$$

$$E_p(\phi=\pi) = -\frac{A_1 m}{\epsilon p} J_m(\beta_p p) [-C_2 \sin(m\pi)] e^{-j\beta_z z} = 0 \Rightarrow \sin(m) = 0 \Rightarrow m\pi = \sin^{-1}(0) = p\pi$$

$$m = -p \quad -p = 0, 1, 2, 3, \dots \Rightarrow m = 0, 1, 2, 3, \dots$$

$$E_\phi = \frac{1}{\epsilon} \frac{\partial F_z}{\partial \phi} = A_{mn} \frac{p}{\epsilon} J_m'(\beta_p p) \cos(m\phi) e^{-j\beta_z z}$$

$$E_\phi(p=a) = A_{mn} \frac{p}{\epsilon} J_m'(\beta_p a) \cos(m\phi) e^{-j\beta_z z} = 0 \Rightarrow \beta_p a = \alpha_m \Rightarrow \beta_p = \left(\frac{\alpha_m}{a}\right)$$

$$F_z = A_{mn} J_m(\beta_p p) \cos(m\phi) e^{-j\beta_z z}, \quad \beta_p = \frac{\alpha_m}{a}, \quad m = 0, 1, 2, \dots, \quad n = 1, 2, 3, \dots$$

$$(f_c)_{mn} = \frac{\alpha_m}{2\pi a \sqrt{\mu\epsilon}},$$

$$E_p = -\frac{1}{\epsilon} \frac{1}{p} \frac{\partial F_z}{\partial \phi} = +A_{mn} \frac{1}{\epsilon p} J_m(\beta_p p) \sin(m\phi) e^{-j\beta_z z}, \quad H_p = j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 F_z}{\partial p^2} = -A_{mn} \frac{\beta_p \beta_z}{\omega \mu \epsilon} J_m'(\beta_p p) \cos(m\phi) e^{-j\beta_z z}$$

$$E_\phi = \frac{1}{\epsilon} \frac{\partial F_z}{\partial \phi} = A_{mn} \frac{p}{\epsilon} J_m'(\beta_p p) \cos(m\phi) e^{-j\beta_z z}, \quad H_\phi = -j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 F_z}{\partial p^2} = A_{mn} \frac{m \beta_z}{\omega \mu \epsilon} J_m(\beta_p p) \sin(m\phi) e^{-j\beta_z z}$$

$$E_z = 0, \quad H_z = -j \frac{1}{\omega \mu \epsilon} \left(\frac{\partial^2}{\partial p^2} + \beta_z^2 \right) F_z = -j \frac{A_{mn}}{\omega \mu \epsilon} \beta_p^2 J_m(\beta_p p) \cos(m\phi) e^{-j\beta_z z}$$

TM²

$$A_z = [A_1 J_m(\beta_p p) + \cancel{B_1 Y_m(\beta_p p)}] [C_2 \cos(m\phi) + D_2 \sin(m\phi)] e^{-j\beta_z z}$$

$$E_z = -j \frac{1}{\omega \mu \epsilon} \left(\frac{\partial^2}{\partial p^2} + \beta_z^2 \right) A_z = -j A_1 \frac{\beta_p^2}{\omega \mu \epsilon} J_m(\beta_p p) [C_2 \cos(m\phi) + D_2 \sin(m\phi)] e^{-j\beta_z z}$$

$$E_z(\phi=0) = -j A_1 \frac{\beta_p^2}{\omega \mu \epsilon} J_m(\beta_p p) [C_2(1) + D_2(0)] e^{-j\beta_z z} = 0 \Rightarrow C_2 = 0$$

$$E_z(\phi=\pi) = -j A_1 \frac{\beta_p^2}{\omega \mu \epsilon} J_m(\beta_p p) [D_2 \sin(m\pi)] e^{-j\beta_z z} = 0 \Rightarrow m\pi = \sin^{-1}(0) = p\pi, \quad m = -p = 1, 2, 3, \dots$$

$$E_z(p=a) = -j A_1 \frac{\beta_p^2}{\omega \mu \epsilon} J_m(\beta_p a) [D_2 \sin(m\phi)] e^{-j\beta_z z} = 0 \Rightarrow \beta_p a = \alpha_m \Rightarrow \beta_p = \alpha_m/a, \quad n = 1, 2, 3, \dots$$

$$A_z = B_{mn} J_m(\beta_p p) \sin(m\phi) e^{-j\beta_z z}, \quad \beta_p = \alpha_m/a, \quad m = 1, 2, 3, \dots, \quad n = 1, 2, 3, \dots$$

$$(f_c)_{mn} = \frac{\alpha_m}{2\pi a \sqrt{\mu\epsilon}}$$

$$E_p = -j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 A_z}{\partial p^2} = -B_{mn} \frac{\beta_p \beta_z}{\omega \mu \epsilon} J_m'(\beta_p p) \sin(m\phi) e^{-j\beta_z z}, \quad H_p = \frac{1}{\beta_p} \frac{\partial A_z}{\partial p} = B_{mn} \frac{m}{\beta_p} J_m(\beta_p p) \cos(m\phi) e^{-j\beta_z z}$$

$$E_\phi = -j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 A_z}{\partial p^2} = -B_{mn} \frac{m}{\omega \mu \epsilon} \frac{\beta_z}{p} J_m(\beta_p p) \cos(m\phi) e^{-j\beta_z z}, \quad H_\phi = -j \frac{1}{\mu} \frac{\partial A_z}{\partial p} = -B_{mn} \frac{\beta_p}{\beta_z} J_m'(\beta_p p) \sin(m\phi) e^{-j\beta_z z}$$

$$E_z = -j \frac{1}{\omega \mu \epsilon} \left(\frac{\partial^2}{\partial p^2} + \beta_z^2 \right) A_z = -j \frac{A_{mn}}{\omega \mu \epsilon} \beta_p^2 J_m(\beta_p p) \sin(m\phi) e^{-j\beta_z z}, \quad H_z = 0$$

9.8 The solution to this problem is identical to that of Problem 9.7 with the following exceptions:

$$\underline{\text{TE}^2}$$

$$E_p(\phi=\pi/2) = -\frac{A_1}{\epsilon} \frac{\beta_p^2}{w\mu\epsilon} J_m(\beta_p g) [C_2 \sin(m\phi)] e^{-j\beta_p z^2} = 0 \Rightarrow \frac{m\pi}{2} = \sin'(0) = +\pi \Rightarrow m=2p, p=0, 1, 2, \dots$$

$$m=0, 2, 4, \dots$$

$$\underline{\text{TM}^2}$$

$$E_z(\phi=\pi/2) = -j A_2 \frac{\beta_p^2}{w\mu\epsilon} J_m(\beta_p g) [D_2 \sin(m\phi/2)] e^{-j\beta_p z^2} = 0 \Rightarrow \frac{m\pi}{2} = \sin'(0) = +\pi, -p=1, 2, 3, \dots$$

$$m=2, 4, 6, \dots$$

9.9 The solution to this problem is identical to that of Problem 9.7 with the following exceptions:

$$\underline{\text{TE}^2}$$

$$E_p(\phi=\phi_0) = -\frac{A_1}{\epsilon} \frac{\beta_p^2}{w\mu\epsilon} J_m(\beta_p g) [C_2 \sin(m\phi_0)] e^{-j\beta_p z^2} = 0 \Rightarrow m\phi_0 = \sin'(0) = +\pi \Rightarrow m = \frac{p\pi}{\phi_0}, p=0, 1, 2, 3, \dots$$

$$\underline{\text{TM}^2}$$

$$E_z(\phi=\phi_0) = -j A_2 \frac{\beta_p^2}{w\mu\epsilon} J_m(\beta_p g) [D_2 \sin(m\phi_0)] e^{-j\beta_p z^2} = 0 \Rightarrow m\phi_0 = \sin'(0) = +\pi \Rightarrow m = \frac{p\pi}{\phi_0}, p=1, 2, 3, \dots$$

9.10 $F_2 = [A_1 J_m(\beta_p g) + B_1 Y_m(\beta_p g)] [C_2 \cos(m\phi) + D_2 \sin(m\phi)] e^{-j\beta_p z^2}, m=0, 1, 2, \dots$ for π periodicity in ϕ .

$$E_\phi = -\frac{1}{\epsilon} \frac{\partial F_2}{\partial \phi} = -\frac{A_1}{\epsilon} \beta_p [A_1 J'_m(\beta_p g) + B_1 Y'_m(\beta_p g)] [C_2 \cos(m\phi) + D_2 \sin(m\phi)] e^{-j\beta_p z^2}$$

$$E_\phi(p=a) = -\frac{A_1}{\epsilon} \beta_p [A_1 J'_m(\beta_p a) + B_1 Y'_m(\beta_p a)] [C_2 \cos(m\phi) + D_2 \sin(m\phi)] e^{-j\beta_p z^2} = 0$$

$$A_1 J'_m(\beta_p a) + B_1 Y'_m(\beta_p a) = 0$$

$$E_\phi(p=b) = -\frac{A_1}{\epsilon} \beta_p [A_1 J'_m(\beta_p b) + B_1 Y'_m(\beta_p b)] [C_2 \cos(m\phi) + D_2 \sin(m\phi)] e^{-j\beta_p z^2} = 0$$

$$A_1 J'_m(\beta_p b) + B_1 Y'_m(\beta_p b) = 0$$

$$\begin{bmatrix} J'_m(\beta_p a) & Y'_m(\beta_p a) \\ J'_m(\beta_p b) & Y'_m(\beta_p b) \end{bmatrix} \begin{bmatrix} A_1 \\ B_1 \end{bmatrix} = \begin{bmatrix} 0 \\ 0 \end{bmatrix} \Rightarrow J'_m(\beta_p a) Y'_m(\beta_p b) - J'_m(\beta_p b) Y'_m(\beta_p a) = 0$$

$$E_p = -\frac{1}{\epsilon} \frac{1}{\beta_p} \frac{\partial F_2}{\partial p} = -\frac{m}{\epsilon} \frac{1}{\beta_p} [A_1 J_m(\beta_p g) + B_1 Y_m(\beta_p g)] [-C_2 \sin(m\phi) + D_2 \cos(m\phi)] e^{-j\beta_p z^2}$$

$$E_\phi = \frac{1}{\epsilon} \frac{\partial F_2}{\partial p} = \frac{\beta_p}{\epsilon} [A_1 J'_m(\beta_p g) + B_1 Y'_m(\beta_p g)] [C_2 \cos(m\phi) + D_2 \sin(m\phi)] e^{-j\beta_p z^2}$$

$$\bar{E}_z = 0$$

$$H_p = -j \frac{1}{w\mu\epsilon} \frac{\partial^2 F_2}{\partial p \partial z^2} = -\frac{\beta_p \beta_t}{w\mu\epsilon} [A_1 J'_m(\beta_p g) + B_1 Y'_m(\beta_p g)] [C_2 \cos(m\phi) + D_2 \sin(m\phi)] e^{-j\beta_p z^2}$$

$$H_\phi = -j \frac{1}{w\mu\epsilon} \frac{\partial^2 F_2}{\partial \phi \partial z^2} = -\frac{m}{w\mu\epsilon} [A_1 J_m(\beta_p g) + B_1 Y_m(\beta_p g)] [C_2 \sin(m\phi) + D_2 \cos(m\phi)] e^{-j\beta_p z^2}$$

$$H_z = -j \frac{1}{w\mu\epsilon} (\frac{\partial^2}{\partial z^2} + \beta^2) F_2 = -j \frac{\beta_p^2}{w\mu\epsilon} [A_1 J_m(\beta_p g) + B_1 Y_m(\beta_p g)] [C_2 \cos(m\phi) + D_2 \sin(m\phi)] e^{-j\beta_p z^2}$$

9.11

$$A_2 = [A_1 J_m(\beta_p g) + B_1 Y_m(\beta_p g)] [C_2 \cos(m\phi) + D_2 \sin(m\phi)] e^{-j\beta_2 z^2}$$

$$E_z = -j \frac{1}{\omega \mu \epsilon} \left(\frac{\beta^2}{\beta_2^2} + \rho^2 \right) A_2 = -j \frac{\beta_p^2}{\omega \mu \epsilon} [A_1 J_m(\beta_p g) + B_1 Y_m(\beta_p g)] [C_2 \cos(m\phi) + D_2 \sin(m\phi)] e^{-j\beta_2 z^2}$$

$$E_z(g=a) = -j \frac{\beta_p^2}{\omega \mu \epsilon} [A_1 J_m(\beta_p a) + B_1 Y_m(\beta_p a)] [C_2 \cos(m\phi) + D_2 \sin(m\phi)] e^{-j\beta_2 z^2} = 0$$

$$A_1 J_m(\beta_p a) + B_1 Y_m(\beta_p a) = 0$$

$$E_z(g=b) = -j \frac{\beta_p^2}{\omega \mu \epsilon} [A_1 J_m(\beta_p b) + B_1 Y_m(\beta_p b)] [C_2 \cos(m\phi) + D_2 \sin(m\phi)] e^{-j\beta_2 z^2} = 0$$

$$A_1 J_m(\beta_p b) + B_1 Y_m(\beta_p b) = 0$$

$$\begin{bmatrix} J_m(\beta_p a) & Y_m(\beta_p a) \\ J_m(\beta_p b) & Y_m(\beta_p b) \end{bmatrix} \begin{bmatrix} A_1 \\ B_1 \end{bmatrix} = \begin{bmatrix} 0 \\ 0 \end{bmatrix} \Rightarrow J_m(\beta_p a) Y_m(\beta_p b) - J_m(\beta_p b) Y_m(\beta_p a) = 0$$

$$E_p = -j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 A_2}{\partial p \partial z} = -\frac{\beta_p \beta_2}{\omega \mu \epsilon} [A_1 J_m'(\beta_p g) + B_1 Y_m'(\beta_p g)] [C_2 \cos(m\phi) + D_2 \sin(m\phi)] e^{-j\beta_2 z^2}$$

$$E_\phi = -j \frac{1}{\omega \mu \epsilon} \frac{1}{\rho} \frac{\partial^2 A_2}{\partial p \partial z} = -\frac{m \beta_2}{\omega \mu \epsilon} [A_1 J_m(\beta_p g) + B_1 Y_m(\beta_p g)] [-C_2 \sin(m\phi) + D_2 \cos(m\phi)] e^{-j\beta_2 z^2}$$

$$E_z = -j \frac{1}{\omega \mu \epsilon} \left(\frac{\beta^2}{\beta_2^2} + \rho^2 \right) A_2 = -j \frac{\beta_p^2}{\omega \mu \epsilon} [A_1 J_m(\beta_p g) + B_1 Y_m(\beta_p g)] [C_2 \cos(m\phi) + D_2 \sin(m\phi)] e^{-j\beta_2 z^2}$$

$$H_p = \frac{1}{k_p^2} \frac{\partial A_2}{\partial p} = \frac{m}{k_p^2} [A_1 J_m(\beta_p g) + B_1 Y_m(\beta_p g)] [-C_2 \sin(m\phi) + D_2 \cos(m\phi)] e^{-j\beta_2 z^2}$$

$$H_\phi = -\frac{1}{\rho} \frac{\partial A_2}{\partial p} = -\frac{\beta_p}{\rho} [A_1 J_m'(\beta_p g) + B_1 Y_m'(\beta_p g)] [C_2 \cos(m\phi) + D_2 \sin(m\phi)] e^{-j\beta_2 z^2}$$

$$H_z = 0$$

9.12 The solution to this problem is identical to that of Problem 9.10 with the following exceptions:

$$E_p = -\frac{1}{\epsilon} \frac{1}{\rho} \frac{\partial F_2}{\partial \phi} = -\frac{m}{\epsilon} \frac{1}{\rho} [A_1 J_m(\beta_p g) + B_1 Y_m(\beta_p g)] [-C_2 \sin(m\phi) + D_2 \cos(m\phi)] e^{-j\beta_2 z^2}$$

$$E_p(\phi=0) = -\frac{m}{\epsilon} \frac{1}{\rho} [A_1 J_m(\beta_p g) + B_1 Y_m(\beta_p g)] [-C_2(0) + D_2(0)] e^{-j\beta_2 z^2} = 0 \Rightarrow D_2 = 0$$

$$E_p(\phi=\phi_0) = -\frac{m}{\epsilon} \frac{1}{\rho} [A_1 J_m(\beta_p g) + B_1 Y_m(\beta_p g)] [-C_2 \sin(m\phi_0)] e^{-j\beta_2 z^2} = 0$$

$$\sin(m\phi_0) = 0 \Rightarrow m\phi_0 = \sin^{-1}(0) = \pm \pi \Rightarrow m = \frac{\pm \pi}{\phi_0}, \quad \rho = 0, 1, 2, \dots$$

$$F_2 = A_{mn} [A_1 J_m(\beta_p g) + B_1 Y_m(\beta_p g)] \cos(m\phi) e^{-j\beta_2 z^2}$$

9.13 The solution to this problem is identical to that of Problem 9.11 with the following exceptions:

$$E_z = -j \frac{1}{\omega \mu \epsilon} \left(\frac{\beta^2}{\beta_2^2} + \rho^2 \right) A_2 = -j \frac{\beta_p^2}{\omega \mu \epsilon} [A_1 J_m(\beta_p g) + B_1 Y_m(\beta_p g)] [C_2 \cos(m\phi) + D_2 \sin(m\phi)] e^{-j\beta_2 z^2}$$

$$E_z(\phi=0) = -j \frac{\beta_p^2}{\omega \mu \epsilon} [A_1 J_m(\beta_p g) + B_1 Y_m(\beta_p g)] [C_2(0) + D_2(0)] e^{-j\beta_2 z^2} = 0 \Rightarrow C_2 = 0$$

$$E_z(\phi=\phi_0) = -j \frac{\beta_p^2}{\omega \mu \epsilon} [A_1 J_m(\beta_p g) + B_1 Y_m(\beta_p g)] [D_2 \sin(m\phi_0)] e^{-j\beta_2 z^2} = 0 \Rightarrow \sin(m\phi_0) = 0$$

$$m\phi_0 = \sin^{-1}(0) = \pm \pi \Rightarrow m = \frac{\pm \pi}{\phi_0}, \quad \rho = 1, 2, 3, \dots$$

$$A_2 = B_{mn} [A_1 J_m(\beta_p g) + B_1 Y_m(\beta_p g)] \sin(m\phi) e^{-j\beta_2 z^2}$$

9.14 $f = 7 \times 10^9 \text{ Hz}$, $\sigma = 5.76 \times 10^7 \text{ S/m}$, $\epsilon = \epsilon_0$, $a = 3 \times 10^{-2} \text{ m}$

TE_{11} Mode

$$f_c = \frac{\alpha'_{11}}{2\pi a \sqrt{\mu_0 \epsilon_0}} = \frac{1.8412 (3 \times 10^8)}{2\pi (3 \times 10^{-2})} \approx 2.93 \times 10^9$$

$$(\alpha_c)_{11}^{TE} = \frac{R_s}{a \eta_0 \sqrt{1 - \left(\frac{f_c}{f}\right)^2}} \left[\left(\frac{f_c}{f}\right)^2 + \frac{1}{(1.8412)^2 - 1} \right]$$

$$\left(\frac{f_c}{f}\right)^2 = \left(\frac{2.93}{7}\right)^2 = (0.4186)^2 = 0.1752$$

$$R_s = \sqrt{\frac{\omega \mu}{2\sigma}} = \sqrt{\frac{2\pi \times 10^9 \times 7 \times 4\pi \times 10^{-7}}{2(5.76 \times 10^7)}} = 2\pi \sqrt{\frac{70}{5.76}} \times 10^{-3} = 0.0219$$

$$(\alpha_c)_{11}^{TE} = \frac{0.0219}{3 \times 10^{-2} (377) \sqrt{1 - 0.1752}} \left[0.1752 + \frac{1}{(1.8412)^2 - 1} \right]$$

$$(\alpha_c)_{11}^{TE} = 2.132 \times 10^{-3} [0.5936] = 1.266 \times 10^{-3} \text{ Np/m} = 10.99 \times 10^{-3} \text{ dB/m}$$

TE_{01} Mode

$$f_c = \frac{\alpha'_{01}}{2\pi a \sqrt{\mu_0 \epsilon_0}} = \frac{3.8318 (3 \times 10^8)}{2\pi (3 \times 10^{-2})} = 6.0985 \times 10^9$$

$$(\alpha_c)_{01}^{TE} = \frac{R_s}{a \eta} \frac{\left(\frac{f_c}{f}\right)^2}{\sqrt{1 - \left(\frac{f_c}{f}\right)^2}}$$

$$\left(\frac{f_c}{f}\right)^2 = \left(\frac{6.0985}{7}\right)^2 = (0.8712)^2 = 0.759$$

$$(\alpha_c)_{01}^{TE} = \frac{0.0219}{3 \times 10^{-2} (377)} \frac{0.759}{\sqrt{1 - 0.759}} = 2.9938 \times 10^{-3} \text{ Np/m} = 26 \times 10^{-3} \text{ dB/m}$$

9.15 $\underline{J}_s = \hat{n} \times \underline{H} = -\hat{a}_p \times (\hat{a}_p H_p + \hat{a}_\phi H_\phi) = -\hat{a}_z H_\phi = -\hat{a}_z \left(-\frac{\partial A_z}{\partial p}\right) = \hat{a}_z \frac{\partial A_z}{\partial p}$

$$\underline{J}_s = \hat{a}_z B_{01} \frac{\beta_p}{\mu} \underline{J}'_0 (\beta_p p) = \hat{a}_z B_{01} \frac{\beta_{01}}{\mu a} \underline{J}'_0 \left(\frac{\beta_{01} p}{a}\right)$$

$$P_e = \frac{R_s}{2} \iint_{\rho=a} \left| \underline{J}_s \cdot \underline{J}_s^* \right| ds = \frac{R_s}{2} \left(\frac{\beta_{01}}{\mu} \right)^2 |B_{01}|^2 \iint_{\rho=a} \left| \underline{J}'_0 (\beta_{01}) \right|^2 ad\phi d\theta$$

$$P_e = \pi R_s l \left(\frac{\beta_{01}}{\mu} \right)^2 |B_{01}|^2 \left[\underline{J}'_0 (\beta_{01}) \right]^2 \quad \text{Cont'd.}$$

9.15 Cont'd.

$$\frac{P_c}{l} = \pi R_s \left(\frac{\beta_p}{\mu} \right)^2 |B_{01}|^2 [J_0'(\chi_{01})]^2$$

However

$$[J_0'(\chi_{01})]^2 = \left\{ \frac{d}{d(\beta_p p)} J_0(\beta_p p) \right\}_{p=0}^2 = \left\{ - J_1(\beta_p p) \right\}_{p=0}^2 = [J_1(\chi_{01})]^2$$

Thus

$$\frac{P_c}{l} = \pi R_s \left(\frac{\beta_p}{\mu} \right)^2 |B_{01}|^2 [J_1(\chi_{01})]^2$$

$$P_{01} = \frac{1}{2} \iint_{S_0} \operatorname{Re} [E \times H^*] \cdot dS = \frac{1}{2} \int_0^{\pi} \int_0^a \left\{ \hat{a}_p E_p + \hat{a}_\phi E_\phi + \hat{a}_z E_z \right\} \times \left[\hat{a}_p H_p^* + \hat{a}_\phi H_\phi^* + \hat{a}_z H_z^* \right] \cdot dS$$

$$P_{01} = \frac{1}{2} \int_0^{2\pi} \int_0^a \operatorname{Re} \hat{a}_z (E_p H_\phi^* - E_\phi H_p^*) \cdot \hat{a}_z p d\phi d\theta = \frac{1}{2} \int_0^{2\pi} \int_0^a \operatorname{Re} \{ [E_p H_\phi^* - E_\phi H_p^*] \} p d\phi d\theta$$

$$\left. \begin{aligned} E_p &= -B_{01} \frac{\beta_p \beta_z}{\omega \mu \epsilon} J_m'(\beta_p p) e^{-j\beta_z z} \\ H_\phi &= -B_{01} \frac{\beta_p}{\mu} J_m'(\beta_p p) e^{-j\beta_z z} \end{aligned} \right\} \quad \begin{aligned} E_p H_\phi^* - E_\phi H_p^* &= E_p H_\phi^* \\ &= |B_{01}|^2 \left(\frac{\beta_p}{\mu} \right)^2 \frac{\beta_z}{\omega \epsilon} [J_m'(\beta_p p)]^2 \end{aligned}$$

$$P_{01} = \frac{1}{2} |B_{01}|^2 \left(\frac{\beta_p}{\mu} \right)^2 \frac{\beta_z}{\omega \epsilon} \int_0^{\pi} \int_0^a [J_m'(\beta_p p)]^2 p d\phi d\theta = |B_{01}|^2 \left(\frac{\beta_p}{\mu} \right)^2 \frac{\pi \beta_z}{\omega \epsilon} \int_0^a [J_0'(\beta_p p)]^2 p dp$$

$$J_0'(\beta_p p) = -J_1(\beta_p p)$$

$$P_{01} = |B_{01}|^2 \left(\frac{\beta_p}{\mu} \right)^2 \frac{\pi \beta_z}{\omega \epsilon} \int_0^a J_1^2(\beta_p p) p dp = |B_{01}|^2 \left(\frac{\beta_p}{\mu} \right)^2 \frac{\pi \beta_z}{\omega \epsilon} \left\{ \frac{p^2}{2} [J_1^2(\beta_p p) - J_0(\beta_p p) J_2(\beta_p p)] \right\}_0^a$$

$$P_{01} = |B_{01}|^2 \left(\frac{\beta_p}{\mu} \right)^2 \frac{\pi \beta_z a^2}{2 \omega \epsilon} \left\{ J_1^2(\chi_{01}) - J_0(\chi_{01}) J_2(\chi_{01}) \right\}$$

However $J_0(\chi_{01}) = 0$ by definition. Thus

$$P_{01} = |B_{01}|^2 \left(\frac{\beta_p}{\mu} \right)^2 \frac{\pi a^2 \beta_z}{2 \omega \epsilon} J_1^2(\chi_{01})$$

Therefore

$$(\alpha_c)^{TM}_{01} = \frac{P_c/l}{2 P_{01}} = \frac{\alpha \pi R_s \left(\frac{\beta_p}{\mu} \right)^2 |B_{01}|^2 J_1^2(\chi_{01})}{|B_{01}|^2 \left(\frac{\beta_p}{\mu} \right)^2 \frac{\pi \beta_z a^2}{2 \omega \epsilon} J_1^2(\chi_{01})} = \frac{\alpha R_s \omega \epsilon}{a^2 \beta_z}$$

$$(\alpha_c)^{TM}_{01} = \frac{R_s}{a \beta_z} \frac{\beta}{\eta} = \frac{R_s}{a \eta} \frac{\beta}{\beta_z} = \frac{R_s}{a \eta} \frac{1}{\sqrt{1 - (\frac{f_c}{f})^2}}$$

9.16 $a = 6 \text{ cm}, h = 10 \text{ cm}, \epsilon_r = 4$

$$a. (f_r)^{TE}_{mn0} = \frac{1}{2\pi\sqrt{\mu\epsilon}} \sqrt{\left(\frac{x'_{mn}}{a}\right)^2 + \left(\frac{p\pi}{h}\right)^2} = \frac{1}{2\pi a\sqrt{\mu\epsilon}} \sqrt{\left(x'_{mn}\right)^2 + \left(\frac{p\pi a}{h}\right)^2}$$

$$(f_r)^{TM}_{mn-p} = \frac{1}{2\pi\sqrt{\mu\epsilon}} \sqrt{\left(\frac{x'_{mn}}{a}\right)^2 + \left(\frac{p\pi}{h}\right)^2} = \frac{1}{2\pi a\sqrt{\mu\epsilon}} \sqrt{\left(x'_{mn}\right)^2 + \left(\frac{p\pi a}{h}\right)^2}$$

$$(f_r)^{TM}_{010} = \frac{1}{2\pi a\sqrt{\mu\epsilon}} (x'_{01}) = \frac{2.4049 \times 3 \times 10^9}{2\pi (6 \times 10^{-2}) 2} = 0.9569 \times 10^9$$

$$(f_r)^{TE}_{111} = \frac{1}{2\pi a\sqrt{\mu\epsilon}} \sqrt{\left(x'_{11}\right)^2 + \left(\frac{\pi a}{h}\right)^2} = \frac{3 \times 10^9}{2\pi (6 \times 10^{-2}) 2} \sqrt{(1.8412)^2 + (0.6\pi)^2} = 1.0482 \times 10^9$$

$$(f_r)^{TM}_{011} = \frac{1}{2\pi a\sqrt{\mu\epsilon}} \sqrt{\left(x'_{01}\right)^2 + \left(\frac{\pi a}{h}\right)^2} = \frac{3 \times 10^9}{2\pi (6 \times 10^{-2}) 2} \sqrt{(2.4049)^2 + (0.6\pi)^2} = 1.2158 \times 10^9$$

$$(f_r)^{TE}_{211} = \frac{1}{2\pi a\sqrt{\mu\epsilon}} \sqrt{\left(x'_{21}\right)^2 + \left(\frac{\pi a}{h}\right)^2} = \frac{3 \times 10^9}{2\pi (6 \times 10^{-2}) 2} \sqrt{(3.0542)^2 + (0.6\pi)^2} = 1.428 \times 10^9$$

b. $Q = \frac{1.2025 \eta}{R_s \left(1 + \frac{a}{h}\right)}$

$$R_s = \sqrt{\frac{wh}{2\sigma}} = \sqrt{\frac{2\pi (0.9569 \times 10^9) (4\pi \times 10^{-7})}{2 (5.76 \times 10^7)}} = 2\pi \sqrt{\frac{9.569}{5.76} \times 10^{-3}} = 8.098 \times 10^{-3}$$

$$Q = \frac{1.2025 (377/2) \times 10^3}{8.098 (1 + 0.6)} = 17.4944 \times 10^3 = 17,494.4 \approx 17,495$$

9.17

From Table 9-4 it is seen that for $h/a = 1$ the dominant mode is the TM_{010} and that the next higher order mode is the TE_{111} such that $(f_r)^{TE}_{111} = 1.514 (f_r)^{TM}_{010} \approx 1.5 (f_r)^{TM}_{010}$

c. There $h/a = 1 \Rightarrow h = a = 4 \text{ cm}$

$$b. (f_r)^{TM}_{010} = \frac{2.4049}{2\pi a\sqrt{\mu\epsilon}} = \frac{2.4049 \times 3 \times 10^9}{2\pi (4 \times 10^{-2})} = 2.8706 \times 10^9$$

$$c. (f_r')^{TM}_{010} = \frac{1}{1.5} (f_r)^{TM}_{010} = \frac{1}{1.5} \frac{2.4049}{2\pi a\sqrt{\mu_0\epsilon_0}} = \frac{2.4049}{2\pi a\sqrt{\mu_0\epsilon_0} \sqrt{\epsilon_r}} \Rightarrow \sqrt{\epsilon_r} \approx 1.5$$

$$\sqrt{\epsilon_r} = 1.5 \Rightarrow \epsilon_r = (1.5)^2 = 2.25$$

9.18

$$(f_r)_{010}^{TM^2} = \frac{1}{2\pi\sqrt{\mu_0\epsilon_0}} \frac{2.4049}{a} = \frac{1}{2\pi a\sqrt{\mu_0\epsilon_0}} (2.4049)$$

$$(f_r)_{111}^{TE^2} = \frac{1}{2\pi\sqrt{\mu_0\epsilon_0}} \sqrt{\left(\frac{1.8412}{a}\right)^2 + \left(\frac{\pi}{h}\right)^2} = \frac{1}{2\pi a\sqrt{\mu_0\epsilon_0}} \sqrt{(1.8412)^2 + \left(\frac{a}{h}\pi\right)^2}$$

When $h/a < 2.03$ the dominant mode is the TM_{010} mode. Therefore for $h/a = 1.9$ the dominant mode is the TM_{010} mode.

(b) The next higher order mode is the TE_{111} . The difference in the resonant frequencies is

$$\frac{1}{2\pi a\sqrt{\mu_0\epsilon_0}} \left\{ \sqrt{(1.8412)^2 + \left(\frac{a}{h}\pi\right)^2} - 2.4049 \right\} = 50 \times 10^6$$

$$\frac{1}{2\pi a\sqrt{\mu_0\epsilon_0} \sqrt{\epsilon_r}} \left\{ \sqrt{(1.8412)^2 + \left(\frac{\pi}{1.9}\right)^2} - 2.4049 \right\} = 50 \times 10^6$$

$$\sqrt{\epsilon_r} = \frac{1}{2\pi a (50 \times 10^6) \mu_0 \epsilon_0} \left\{ \sqrt{(1.8412)^2 + \left(\frac{\pi}{1.9}\right)^2} - 2.4049 \right\}$$

$$= \frac{30 \times 10^9}{2\pi (2)(50 \times 10^6)} \left\{ \sqrt{(1.8412)^2 + \left(\frac{\pi}{1.9}\right)^2} - 2.4049 \right\}$$

$$\sqrt{\epsilon_r} = \frac{3 \times 10^3}{20\pi} \left\{ 2.4747 - 2.4049 \right\} = \frac{300}{2\pi} (0.06977) = 3.331$$

$$\epsilon_r = 11.097 \approx 11$$

9.19

 $T \in \mathbb{N}$

$$(f_c)_n = \frac{n}{2h\sqrt{\mu_0\varepsilon_0}} \Big|_{n=2} = \frac{2}{2h\sqrt{\mu_0\varepsilon_0}} = \frac{2 \times 3 \times 10^8}{2h} = 300 \times 10^6 = 3 \times 10^8 \Rightarrow h = 1 \text{ m}$$

$$h = 1 \text{ m} = 100 \text{ cm} \Rightarrow h = 1 \text{ m}$$

 TM_n^2

$$(f_c)_n = \frac{n}{2h\sqrt{\mu_0\varepsilon_0}} \Big|_{n=1} = \frac{3 \times 10^8}{2h} = 3 \times 10^8 \Rightarrow h = 0.5 \text{ m}$$

9.20

$$F_z^+ (\rho, \phi, z) = D_1 H_m^{(2)}(\beta_p \rho) [C_2 \cos(m\phi) + D_2 \sin(m\phi)] [C_3 \cos(\beta_z z) + D_3 \sin(\beta_z z)]$$

$$\beta_p^2 + \beta_z^2 = \beta^2$$

$$E_g^+ = -\frac{1}{\epsilon} \frac{1}{z} \frac{\partial F_z^+}{\partial \phi} = -D_1 \frac{m}{\epsilon p} H_m^{(2)}(\beta_p \rho) [-C_2 \sin(m\phi) + D_2 \cos(m\phi)] [C_3 \cos(\beta_z z) + D_3 \sin(\beta_z z)]$$

$$E_g^+ (\phi = 0) = -D_1 \frac{m}{\epsilon p} H_m^{(2)}(\beta_p \rho) [-C_2(0) + D_2(0)] [C_3(0) + D_3(0)] = 0 \Rightarrow D_2 = 0$$

$$E_g^+ (\phi = \phi_0) = -D_1 \frac{m}{\epsilon p} H_m^{(2)}(\beta_p \rho) [-C_2 \sin(m\phi_0)] [C_3(0) + D_3(0)] = 0 \Rightarrow \sin(m\phi_0) = 0 \Rightarrow m = \frac{n\pi}{\phi_0}, n = 0, 1, 2, \dots$$

$$E_g^+ (z = 0) = -D_1 \frac{m}{\epsilon p} H_m^{(2)}(\beta_p \rho) [-C_2 \sin(m\phi)] [C_3(0) + D_3(0)] = 0 \Rightarrow C_3 = 0$$

$$E_g^+ (z = h) = -D_1 \frac{m}{\epsilon p} H_m^{(2)}(\beta_p \rho) [-C_2 \sin(m\phi)] [D_3 \sin(\beta_z h)] = 0 \Rightarrow \sin(\beta_z h) = 0 \Rightarrow \beta_z = \frac{n\pi}{h}, n = 1, 2, \dots$$

Therefore

$$F_z^+ (\rho, \phi, z) = A_{pn} H_m^{(2)}(\beta_p \rho) \cos(m\phi) \sin(\beta_z z)$$

$$9.21 \quad A_z^+ (\rho, \phi, z) = D_1 H_m^{(2)}(\beta_p \rho) [C_2 \cos(m\phi) + D_2 \sin(m\phi)] [C_3 \cos(\beta_z z) + D_3 \sin(\beta_z z)]$$

$$E_g^+ = -j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 A_z^+}{\partial \rho \partial z} = -j \frac{D_1 \beta_p \beta_z}{\omega \mu \epsilon} H_m^{(2)}(\beta_p \rho) [C_2 \cos(m\phi) + D_2 \sin(m\phi)] [-C_3 \sin(\beta_z z) + D_3 \cos(\beta_z z)]$$

$$E_g^+ (\phi = 0) = -j D_1 \frac{\beta_p \beta_z}{\omega \mu \epsilon} H_m^{(2)}(\beta_p \rho) [C_2(0) + D_2(0)] [-C_3(0) + D_3(0)] = 0 \Rightarrow C_2 = 0$$

$$E_g^+ (\phi = \phi_0) = -j D_1 \frac{\beta_p \beta_z}{\omega \mu \epsilon} H_m^{(2)}(\beta_p \rho) [D_2 \sin(m\phi_0)] [-C_3(0) + D_3(0)] = 0 \Rightarrow \sin(m\phi_0) = 0 \Rightarrow m = \frac{n\pi}{\phi_0}, n = 1, 2, \dots$$

$$E_g^+ (z = 0) = -j D_1 \frac{\beta_p \beta_z}{\omega \mu \epsilon} H_m^{(2)}(\beta_p \rho) [D_2 \sin(m\phi)] [-C_3(0) + D_3(0)] = 0 \Rightarrow D_3 = 0$$

$$E_g^+ (z = h) = -j D_1 \frac{\beta_p \beta_z}{\omega \mu \epsilon} H_m^{(2)}(\beta_p \rho) [D_2 \sin(m\phi)] [-C_3 \sin(\beta_z h)] = 0 \Rightarrow \sin(\beta_z h) = 0 \Rightarrow \beta_z = \frac{n\pi}{h}, n = 0, 1, 2, \dots$$

$$\text{Therefore } A_z^+ (\rho, \phi, z) = B_{pn} H_m^{(2)}(\beta_p \rho) \sin(m\phi) \cos(\beta_z z)$$

9.22

Given: $\epsilon_r = 130$, $\mu_r = 1$, surface of the dielectric can be approximated by a PMC
radius of the rod, $a = 3 \text{ cm}$

For a Circular Dielectric Waveguide, the TE^z modes can be constructed using the vector potential F_z , where,

$$F_z(\rho, \phi, z) = [A_1 J_m(\beta_\rho^d \rho) + B_1 Y_m(\beta_\rho^d \rho)] \cdot [C_2 \cos(m\phi) + D_2 \sin(m\phi)] \cdot [A_3 e^{-j\beta_z z} + B_3 e^{+j\beta_z z}]$$

$$(\beta_\rho^d)^2 + \beta_z^2 = \beta^2$$

$f_1(\rho) = A_1 J_m(\beta_\rho^d \rho) + B_1 Y_m(\beta_\rho^d \rho)$ represents standing waves in the $\pm \rho$ directions.

$g_2(\phi) = C_2 \cos(m\phi) + D_2 \sin(m\phi)$ represents periodic waves in the $\pm \phi$ directions.

$h_3(z) = A_3 e^{-j\beta_z z} + B_3 e^{+j\beta_z z}$ represents traveling waves in the $\pm z$ directions.

Since the surface can be modeled as a PMC, the following boundary conditions exist:

$$\boxed{BC \#1) H_\phi(\rho = a, 0 \leq \phi \leq 2\pi, z) = 0}$$

or

$$\boxed{BC \#2) H_z(\rho = a, 0 \leq \phi \leq 2\pi, z) = 0}$$

using the following conditions...

- 1) The fields must be finite everywhere.
- 2) The fields must repeat every 2π radians in ϕ .

According to 1), $B_1 = 0$ since $Y_m[\rho = 0] = -\infty$.

According to 2), $m = 1, 2, 3, \dots$

If we only consider only waves that propagate in the $+z$ direction. Then $B_3 = 0$.

$$\boxed{F_z^+(\rho, \phi, z) = A_{mn} J_m(\beta_\rho^d \rho) \cdot [C_2 \cos(m\phi) + D_2 \sin(m\phi)] \cdot e^{-j\beta_z z}}$$

with $(\beta_\rho^d)^2 + \beta_z^2 = \beta^2$

Problem 9-22 cont.

$$E_\rho^+ = -\frac{1}{\epsilon_d \rho} \frac{\partial F_z^+}{\partial \phi} = -A_{mn} \frac{m}{\epsilon_d \rho} J_m(\beta_\rho^d \rho) \cdot [-C_2 \sin(m\phi) + D_2 \cos(m\phi)] \cdot e^{-j\beta_z z}$$

$$E_\phi^+ = \frac{1}{\epsilon_d} \frac{\partial F_z^+}{\partial \rho} = A_{mn} \frac{\beta_\rho^d}{\epsilon_d} J_m(\beta_\rho^d \rho) \cdot [C_2 \cos(m\phi) + D_2 \sin(m\phi)] \cdot e^{-j\beta_z z}$$

$$E_z^+ = 0$$

$$H_\rho^+ = -j \frac{1}{\omega \mu_d \epsilon_d} \frac{\partial^2 F_z^+}{\partial \rho \partial z} = -A_{mn} \frac{\beta_\rho^d \beta_z}{\omega \mu_d \epsilon_d} J_m(\beta_\rho^d \rho) \cdot [C_2 \cos(m\phi) + D_2 \sin(m\phi)] \cdot e^{-j\beta_z z}$$

$$H_\phi^+ = -j \frac{1}{\omega \mu_d \epsilon_d} \frac{1}{\rho} \frac{\partial^2 F_z^+}{\partial \phi \partial z} = -A_{mn} \frac{m \beta_z}{\omega \mu_d \epsilon_d} \frac{1}{\rho} J_m(\beta_\rho^d \rho) \cdot [-C_2 \sin(m\phi) + D_2 \cos(m\phi)] \cdot e^{-j\beta_z z}$$

$$H_z^+ = -j \frac{1}{\omega \mu_d \epsilon_d} \left(\frac{\partial^2}{\partial z^2} + \beta^2 \right) F_z^+ = -j A_{mn} \frac{(\beta_\rho^d)^2}{\omega \mu_d \epsilon_d} J_m(\beta_\rho^d \rho) \cdot [C_2 \cos(m\phi) + D_2 \sin(m\phi)] \cdot e^{-j\beta_z z}$$

with, $' = \frac{\partial}{\partial(\beta_\rho^d \rho)}$

$$\text{Apply BC # 1 : } H_\phi(\rho = a, 0 \leq \phi \leq 2\pi, z) = 0$$

$$H_\phi(\rho = a, 0 \leq \phi \leq 2\pi, z) = -A_{mn} \frac{m \beta_z}{\omega \mu_d \epsilon_d} \frac{1}{a} J_m(\beta_\rho^d a) \cdot [-C_2 \sin(m\phi) + D_2 \cos(m\phi)] \cdot e^{-j\beta_z z} = 0$$

$$\Rightarrow J_m(\beta_\rho^d a) = 0 \Rightarrow \beta_\rho^d a = \chi_{mn} \Rightarrow \boxed{\beta_\rho^d = \frac{\chi_{mn}}{a}}$$

where χ_{mn} represents the nth zero ($n = 1, 2, 3, \dots$) of the Bessel function J_m of the first kind of order m ($m = 0, 1, 2, 3, \dots$).

Cutoff is defined when $(\beta_z)_{mn} = 0$. Therefore,

$$(\beta_\rho^d)^2 + \beta_z^2 = \beta^2 \Rightarrow \beta_z = \sqrt{\beta^2 - (\beta_\rho^d)^2} \quad \text{when } \beta > \beta_\rho^d = \frac{\chi_{mn}}{a}$$

$$\beta_c = \omega_c \sqrt{\mu_d \epsilon_d} = 2\pi f_c \sqrt{\mu_d \epsilon_d} = \beta_\rho^d = \frac{\chi_{mn}}{a}$$

Cutoff Frequency:

$$\boxed{(f_c)_{mn} = \frac{\chi_{mn}}{2\pi a \sqrt{\mu_d \epsilon_d}}}$$

$$\begin{aligned} m &= 0, 1, 2, 3, \dots \\ n &= 1, 2, 3, \dots \end{aligned}$$

Problem 9.22 cont.

Determine the cutoff frequencies of the lowest two modes when the radius of the rod is 3 cm.

For the TE^z modes, the lowest two modes are the TE_{01}^z and TE_{11}^z .

$$\text{TE}_{01}^z \Rightarrow \chi_{01} = 2.409$$

$$(f_c)_{01} = \frac{\chi_{01}}{2\pi a \sqrt{\mu \epsilon}} = \frac{\chi_{mn} \cdot c}{2\pi a \sqrt{\mu_d \epsilon_d}} = \frac{(2.409) \cdot (30 \times 10^9 \text{ cm/s})}{2\pi (3 \text{ cm}) \sqrt{(1) \cdot (130)}} = 0.336 \text{ GHz}$$

$$\text{TE}_{11}^z \Rightarrow \chi_{11} = 3.8318$$

$$(f_c)_{11} = \frac{\chi_{11} \cdot c}{2\pi a \sqrt{\mu_d \epsilon_d}} = \frac{(3.8318) \cdot (30 \times 10^9 \text{ cm/s})}{2\pi (3 \text{ cm}) \sqrt{(1) \cdot (130)}} = 0.535 \text{ GHz}$$

$\text{TE}_{01}^z \Rightarrow (f_c)_{01} = 0.336 \text{ GHz}$
 $\text{TE}_{11}^z \Rightarrow (f_c)_{11} = 0.535 \text{ GHz}$

9.23

Given: $\epsilon_r = 130$, $\mu_r = 1$, surface of the dielectric can be approximated by a PMC
radius of the rod, $a = 3 \text{ cm}$

For a Circular Dielectric Waveguide, the TM^z modes can be constructed using the vector potential A_z , where,

$$A_z(\rho, \phi, z) = [A_1 J_m(\beta_\rho^d \rho) + B_1 Y_m(\beta_\rho^d \rho)] \cdot [C_2 \cos(m\phi) + D_2 \sin(m\phi)] \cdot [A_3 e^{-j\beta_z z} + B_3 e^{+j\beta_z z}]$$

$$(\beta_\rho^d)^2 + \beta_z^2 = \beta^2$$

$f_1(\rho) = A_1 J_m(\beta_\rho^d \rho) + B_1 Y_m(\beta_\rho^d \rho)$ represents standing waves in the $\pm \rho$ directions.

$g_2(\phi) = C_2 \cos(m\phi) + D_2 \sin(m\phi)$ represents periodic waves in the $\pm \phi$ directions.

$h_3(z) = A_3 e^{-j\beta_z z} + B_3 e^{+j\beta_z z}$ represents traveling waves in the $\pm z$ directions.

Since the surface can be modeled as a PMC, the following boundary conditions exist:

$$\boxed{\text{BC #1)} H_\phi(\rho = a, 0 \leq \phi \leq 2\pi, z) = 0}$$

or

$$\boxed{\text{BC #2)} H_z(\rho = a, 0 \leq \phi \leq 2\pi, z) = 0}$$

using the following conditions...

- 1) The fields must be finite everywhere.
- 2) The fields must repeat every 2π radians in ϕ .

According to 1), $B_1 = 0$ since $Y_m[\rho = 0] = -\infty$.

According to 2), $m = 1, 2, 3, \dots$

If we only consider only waves that propagate in the $+z$ direction. Then $B_3 = 0$.

$$\boxed{A_z^+(\rho, \phi, z) = A_{mn} J_m(\beta_\rho^d \rho) \cdot [C_2 \cos(m\phi) + D_2 \sin(m\phi)] \cdot e^{-j\beta_z z}}$$

with $(\beta_\rho^d)^2 + \beta_z^2 = \beta^2$

Problem 9.23 cont.

$$\begin{aligned}
 E_\rho^+ &= -j \frac{1}{\omega_r \mu_d \epsilon_d} \frac{\partial^2 A_z^+}{\partial \rho \partial z} = -B_{mn} \frac{\beta_\rho^d \beta_z}{\omega \mu_d \epsilon_d} J_m(\beta_\rho^d \rho) \cdot [C_2 \cos(m\phi) + D_2 \sin(m\phi)] \cdot e^{-j\beta_z z} \\
 E_\phi^+ &= -j \frac{1}{\omega_r \mu_d \epsilon_d} \frac{1}{\rho} \frac{\partial^2 A_z^+}{\partial \phi \partial z} = -B_{mn} \frac{m \beta_z}{\omega \mu_d \epsilon_d} \frac{1}{\rho} J_m(\beta_\rho^d \rho) \cdot [-C_2 \sin(m\phi) + D_2 \cos(m\phi)] \cdot e^{-j\beta_z z} \\
 E_z^+ &= -j \frac{1}{\omega \mu_d \epsilon_d} \left(\frac{\partial^2}{\partial z^2} + \beta^2 \right) A_z^+ = -j B_{mn} \frac{(\beta_\rho^d)^2}{\omega \mu_d \epsilon_d} J_m(\beta_\rho^d \rho) \cdot [C_2 \cos(m\phi) + D_2 \sin(m\phi)] \cdot e^{-j\beta_z z} \\
 H_\rho^+ &= \frac{1}{\mu_d} \frac{1}{\rho} \frac{\partial A_z^+}{\partial \phi} = B_{mn} \frac{m}{\mu_d} \frac{1}{\rho} J_m(\beta_\rho^d \rho) \cdot [-C_2 \sin(m\phi) + D_2 \cos(m\phi)] \cdot e^{-j\beta_z z} \\
 H_\phi^+ &= -\frac{1}{\mu_d} \frac{\partial A_z^+}{\partial \rho} = -B_{mn} \frac{\beta_\rho^d}{\mu_d} J_m(\beta_\rho^d \rho) \cdot [C_2 \cos(m\phi) + D_2 \sin(m\phi)] \cdot e^{-j\beta_z z} \\
 H_z^+ &= 0
 \end{aligned}$$

with, $\dot{} = \frac{\partial}{\partial(\beta_\rho^d \rho)}$

Apply BC # 1) : $H_\phi(\rho = a, 0 \leq \phi \leq 2\pi, z) = 0$

$$\begin{aligned}
 H_\phi(\rho = a, 0 \leq \phi \leq 2\pi, z) &= -B_{mn} \frac{\beta_\rho^d}{\mu_d} J_m(\beta_\rho^d \rho) \cdot [C_2 \cos(m\phi) + D_2 \sin(m\phi)] \cdot e^{-j\beta_z z} \\
 \Rightarrow J_m(\beta_\rho^d a) &= 0 \quad \Rightarrow \quad \beta_\rho^d a = \chi_{mn} \quad \Rightarrow \quad \boxed{\beta_\rho^d = \frac{\chi_{mn}}{a}}
 \end{aligned}$$

where χ_{mn} represents the nth zero ($n = 1, 2, 3, \dots$) of the derivative of the Bessel function J_m of the first kind of order m ($m = 0, 1, 2, 3, \dots$).

Cutoff is defined when $(\beta_z)_{mn} = 0$. Therefore,

$$(\beta_\rho^d)^2 + \beta_z^2 = \beta^2 \quad \Rightarrow \quad \beta_z = \sqrt{\beta^2 - (\beta_\rho^d)^2} \quad \text{when } \beta > \beta_\rho^d = \frac{\chi_{mn}}{a}$$

$$\beta_c = \omega_c \sqrt{\mu_d \epsilon_d} = 2\pi f_c \sqrt{\mu_d \epsilon_d} = \beta_\rho^d = \frac{\chi_{mn}}{a}$$

Cutoff Frequency:

$$(f_c)_{mn} = \frac{\chi_{mn}}{2\pi a \sqrt{\mu_d \epsilon_d}}$$

$$m = 0, 1, 2, 3, \dots$$

$$n = 1, 2, 3, \dots$$

Problem 9-23 cont.

Determine the cutoff frequencies of the lowest two modes when the radius of the rod is 3 cm.

For the TM^z modes, the lowest two modes are the TM_{11}^z and TM_{21}^z .

$$\text{TM}_{11}^z \Rightarrow \chi_{11} = 1.8412$$

$$(f_c)_{11} = \frac{\chi_{11}}{2\pi a \sqrt{\mu \epsilon}} = \frac{\chi_{11} \cdot c}{2\pi a \sqrt{\mu_d \epsilon_d}} = \frac{(1.8412) \cdot (30 \times 10^9 \text{ cm/s})}{2\pi (3 \text{ cm}) \sqrt{(1) \cdot (130)}} = 0.257 \text{ GHz}$$

$$\text{TM}_{21}^z \Rightarrow \chi_{21} = 3.0542$$

$$(f_c)_{21} = \frac{\chi_{21}}{2\pi a \sqrt{\mu \epsilon}} = \frac{\chi_{21} \cdot c}{2\pi a \sqrt{\mu_d \epsilon_d}} = \frac{(3.0542) \cdot (30 \times 10^9 \text{ cm/s})}{2\pi (3 \text{ cm}) \sqrt{(1) \cdot (130)}} = 0.425 \text{ GHz}$$

$\text{TM}_{11}^z \Rightarrow (f_c)_{11} = 0.257 \text{ GHz}$
 $\text{TM}_{21}^z \Rightarrow (f_c)_{21} = 0.425 \text{ GHz}$

9.24

Given: $a = 3 \text{ cm}$, $\epsilon_r = 20, 38, \text{ and } 130$

Dominant Mode: HEM₁₁ (HE₁₁) with zero cutoff frequency. $(f_c)_{11}^{\text{HE}} = 0 \text{ Hz}$

$$\zeta = \beta_z a = \sqrt{(\beta_0 a)^2 \epsilon_r - \chi^2}$$

$$\xi = \sqrt{(\beta_0 a)^2 (\epsilon_r \mu_r - 1) - \chi^2}$$

$$\text{At Cutoff: } \beta_z = \beta_0$$

$$\zeta = \beta_0 a = \sqrt{(\beta_0 a)^2 \epsilon_r - \chi^2}$$

$$(\beta_0 a)^2 = (\beta_0 a)^2 \epsilon_r - \chi^2$$

$$\chi^2 = (\beta_0 a)^2 \epsilon_r - (\beta_0 a)^2$$

$$\chi = \beta_0 a \sqrt{\epsilon_r - 1}$$

$$\text{slope} = \frac{\chi_{mn}}{\beta_0 a} = \sqrt{\epsilon_r - 1}$$

$$\text{At Cutoff: } \xi = 0 \Rightarrow \beta_0 = \beta_{oc}$$

$$\xi|_{\beta_0=\beta_{oc}} = 0 = \sqrt{(\beta_{oc} a)^2 (\epsilon_r \mu_r - 1) - \chi_{mn}^2}$$

$$0 = (\beta_{oc} a)^2 (\epsilon_r \mu_r - 1) - \chi_{mn}^2$$

$$\chi_{mn}^2 = (\beta_{oc} a)^2 (\epsilon_r \mu_r - 1)$$

$$\chi_{mn} = \beta_{oc} a \sqrt{\epsilon_r \mu_r - 1}$$

$$\chi_{mn} = \frac{\omega_c}{v} a \sqrt{\epsilon_r \mu_r - 1} = \omega_c \sqrt{\mu_0 \epsilon_0} \cdot a \sqrt{\epsilon_r \mu_r - 1}$$

$$\chi_{mn} = 2\pi \cdot f_c \sqrt{\mu_0 \epsilon_0} \cdot a \sqrt{\epsilon_r \mu_r - 1}$$

$$(f_c)_{mn} = \frac{\chi_{mn}}{2\pi \cdot a \sqrt{\mu_0 \epsilon_0} \sqrt{\epsilon_r \mu_r - 1}} = \frac{\chi_{mn} \cdot c}{2\pi \cdot a \sqrt{\epsilon_r \mu_r - 1}}$$

$$1. \text{ HE}_{11} (\text{HEM}_{11}) \quad J_1(\chi_{10}) = 0 \quad \chi_{10} = 0$$

$$(f_c)_{11}^{\text{HE}} = 0 \text{ Hz}$$

$$2. \text{ TE}_{01}, \text{TM}_{01}, \text{HE}_{21} (\text{HEM}_{21}) \quad J_0(\chi_{01}) = 0 \quad \chi_{01} = 2.4049$$

$$\epsilon_r = 20 \Rightarrow (f_c)_{01} = \frac{(2.4049) \cdot (30 \times 10^9 \text{ cm/s})}{2\pi \cdot (3 \text{ cm}) \sqrt{(20) \cdot (1) - 1}} = 0.878 \text{ GHz}$$

$$\epsilon_r = 38 \Rightarrow (f_c)_{01} = \frac{(2.4049) \cdot (30 \times 10^9 \text{ cm/s})}{2\pi \cdot (3 \text{ cm}) \sqrt{38 \cdot (1) - 1}} = 0.629 \text{ GHz}$$

$$\epsilon_r = 130 \Rightarrow (f_c)_{01} = \frac{(2.4049) \cdot (30 \times 10^9 \text{ cm/s})}{2\pi \cdot (3 \text{ cm}) \sqrt{(130) \cdot (1) - 1}} = 0.337 \text{ GHz}$$

Problem 9-24 Cont.

$$3. \text{ HE}_{12}(\text{HEM}_{13}), \text{ EH}_{11}(\text{HEM}_{12}), \text{ HE}_{13}(\text{HEM}_{31}) \quad J_1(\chi_{11}) = 0 \quad \chi_{11} = 3.8318$$

$$\epsilon_r = 20 \Rightarrow (f_c)_{01} = \frac{(3.8318) \cdot (30 \times 10^9 \text{ cm/s})}{2\pi \cdot (3 \text{ cm}) \sqrt{(20) \cdot (1) - 1}} = 1.40 \text{ GHz}$$

$$\epsilon_r = 38 \Rightarrow (f_c)_{01} = \frac{(3.8318) \cdot (30 \times 10^9 \text{ cm/s})}{2\pi \cdot (3 \text{ cm}) \sqrt{38 \cdot (1) - 1}} = 1.00 \text{ GHz}$$

$$\epsilon_r = 130 \Rightarrow (f_c)_{01} = \frac{(3.8318) \cdot (30 \times 10^9 \text{ cm/s})}{2\pi \cdot (3 \text{ cm}) \sqrt{(130) \cdot (1) - 1}} = 0.537 \text{ GHz}$$

$$4. \text{ EH}_{21}(\text{HEM}_{22}), \text{ EH}_{11}, \text{ HE}_{41}(\text{HEM}_{41}) \quad J_2(\chi_{21}) = 0 \quad \chi_{21} = 5.1357$$

$$\epsilon_r = 20 \Rightarrow (f_c)_{01} = \frac{(5.1357) \cdot (30 \times 10^9 \text{ cm/s})}{2\pi \cdot (3 \text{ cm}) \sqrt{(20) \cdot (1) - 1}} = 1.88 \text{ GHz}$$

$$\epsilon_r = 38 \Rightarrow (f_c)_{01} = \frac{(5.1357) \cdot (30 \times 10^9 \text{ cm/s})}{2\pi \cdot (3 \text{ cm}) \sqrt{38 \cdot (1) - 1}} = 1.34 \text{ GHz}$$

$$\epsilon_r = 130 \Rightarrow (f_c)_{01} = \frac{(5.1357) \cdot (30 \times 10^9 \text{ cm/s})}{2\pi \cdot (3 \text{ cm}) \sqrt{(130) \cdot (1) - 1}} = 0.720 \text{ GHz}$$

9.25

$$f_c = \frac{\chi_{mn}}{2\pi \cdot a \sqrt{\mu_0 \epsilon_0 \sqrt{\epsilon_r - 1}}} = \frac{\chi_{mn} \cdot c}{2\pi \cdot a \sqrt{\epsilon_r - 1}} \Rightarrow a = \frac{\chi_{mn} \cdot c}{2\pi \cdot f_c \sqrt{\epsilon_r - 1}}$$

$$\epsilon_r = 2.56 \Rightarrow a = \frac{(2.4049) \cdot (30 \times 10^9 \text{ cm/s})}{2\pi \cdot (3 \times 10^9 \text{ Hz}) \sqrt{2.56 - 1}} = 3.064 \text{ cm}$$

$$\epsilon_r = 4 \Rightarrow a = \frac{(2.4049) \cdot (30 \times 10^9 \text{ cm/s})}{2\pi \cdot (3 \times 10^9 \text{ Hz}) \sqrt{4 - 1}} = 2.21 \text{ cm}$$

$$\epsilon_r = 9 \Rightarrow a = \frac{(2.4049) \cdot (30 \times 10^9 \text{ cm/s})}{2\pi \cdot (3 \times 10^9 \text{ Hz}) \sqrt{9 - 1}} = 1.35 \text{ cm}$$

$$\epsilon_r = 16 \Rightarrow a = \frac{(2.4049) \cdot (30 \times 10^9 \text{ cm/s})}{2\pi \cdot (3 \times 10^9 \text{ Hz}) \sqrt{16 - 1}} = 0.988 \text{ cm}$$

$$[9.26] \gamma = \beta_z a \Big|_{\beta_z = \beta_0} = \beta_0 a = \sqrt{(\beta_0 a)^2 \epsilon_r - \gamma^2} \Rightarrow \epsilon_r = \frac{(\beta_0 a)^2 + \gamma^2}{(\beta_0 a)^2}, \gamma = 2.4049$$

$$\text{At } f = 5 \text{ GHz} \Rightarrow \lambda_0 = \frac{\nu_0}{f} = \frac{30 \times 10^9}{5 \times 10^9} = 6 \text{ cm} \Rightarrow \beta_0 = 2\pi/\lambda_0$$

$$\text{a. } a = 1.315 \text{ cm: } \beta_0 a = \frac{2\pi}{\lambda_0} (1.315) = 1.377 \Rightarrow \epsilon_r = \frac{(1.377)^2 + (2.4049)^2}{(1.377)^2} = 4.05$$

$$\text{b. } a = 1.838 \text{ cm: } \beta_0 a = \frac{2\pi}{\lambda_0} (1.838) = 1.9247 \Rightarrow \epsilon_r = \frac{(1.9247)^2 + (2.4049)^2}{(1.9247)^2} = 2.56$$

$$[9.27] \epsilon_{\text{eff}} = (\beta_z / \beta_0)^2 = 2.78, \epsilon_r = 4, f = 3 \text{ GHz}, a/\lambda_0 = 0.3$$

$$(a) \beta_z / \beta_0 = \sqrt{2.78} = 1.667 \Rightarrow \beta_z = 1.667 \beta_0 = 1.667 \left(\frac{2\pi}{\lambda_0} \right) = 1.667 \left(\frac{2\pi}{0.1} \right) = 104.74 \text{ rad/m}$$

$$\beta_z = 104.74 \text{ rad/m} = 1.0474 \text{ rad/cm}$$

$$(\beta_p^d)^2 + (\beta_z)^2 = \beta_0^2 = \epsilon_r \beta_0^2 \Rightarrow (\beta_p^d)^2 = \epsilon_r \beta_0^2 - \beta_z^2 = \epsilon_r \beta_0^2 - \epsilon_{\text{eff}} \beta_0^2 = \beta_0^2 (\epsilon_r - \epsilon_{\text{eff}})$$

$$\beta_p^d = \beta_0 \sqrt{\epsilon_r - \epsilon_{\text{eff}}} = 20\pi \sqrt{4 - 2.78} = 20\pi (1.1045) = 69.4 \text{ rad/m}$$

$$\beta_p^d = 69.4 \text{ rad/m} = 0.694 \text{ rad/cm}$$

(b)

$$-(\alpha_p^o)^2 + \beta_z^2 = \beta_0^2 \Rightarrow (\alpha_p^o)^2 = \beta_0^2 - \beta_z^2 = (1.667 \beta_0)^2 - \beta_0^2 = \beta_0^2 [(1.667)^2 - 1] = \beta_0^2 (2.78 - 1)$$

$$\alpha_p^o = \beta_0 \sqrt{2.78 - 1} = 1.334 \beta_0 = 1.334 (20\pi) = 26.683\pi = 83.82 \text{ Nepers/m}$$

$$\alpha_p^o = 83.82 \text{ Nepers/m} = 0.8382 \text{ Nepers/cm}$$

9.28

$$\gamma = \beta_z a = \sqrt{(\beta_0 a)^2 \epsilon_r - \chi^2}$$

At cutoff $\gamma = \beta_z a = \beta_0 a$

$$\beta_0 a = \sqrt{(\beta_0 a)^2 \epsilon_r - \chi^2}$$

$$(\beta_0 a)^2 = (\beta_0 a)^2 \epsilon_r - \chi^2$$

$$(\beta_0 a)^2 (\epsilon_r - 1) = \chi^2$$

$$(\beta_0 a)^2 = \frac{\chi^2}{\epsilon_r - 1}$$

$$\beta_0 a = \frac{\chi}{\sqrt{\epsilon_r - 1}}$$

$$a = \frac{\chi}{\beta_0 \sqrt{\epsilon_r - 1}} = \frac{\lambda_0}{2\pi} \frac{\chi}{\sqrt{\epsilon_r - 1}}$$

$$\lambda_0 = \frac{\omega_0}{f} = \frac{30 \times 10^9}{4 \times 10^9} = 7.5 \text{ cm}$$

The next modes after the HE₁₁ mode are TE₀₂, TM₀₁, HE₂₁
 $\chi = 2.4049$.

$$a = \frac{7.5}{2\pi} \frac{2.4049}{\sqrt{4-1}} = \frac{7.5}{2\pi} \frac{2.4049}{\sqrt{3}} = 1.6574 \text{ cm}$$

$a = 1.6574 \text{ cm}$ ζ

9.29

$$(a) \frac{2a}{\lambda_0} \geq \frac{1}{\pi} \frac{2.4049}{\sqrt{\epsilon_r - 1}}$$

$$\sqrt{\epsilon_r - 1} \geq \frac{\lambda_0 (2.4049)}{2\pi a} = \frac{2.4049 \lambda_0}{2\pi a}$$

$$\epsilon_r - 1 \geq \left(\frac{2.4049 \lambda_0}{2\pi a} \right)^2$$

$$\epsilon_r \geq 1 + \left(\frac{2.4049 \lambda_0}{2\pi a} \right)^2$$

$$\lambda_0 = \frac{c}{f} = \frac{3 \times 10^8}{100 \times 10^9} = \frac{3 \times 10^8}{10^{11}} = 3 \times 10^{-3}$$

$$a = 1 \times 10^{-3}$$

$$\epsilon_r \geq 1 + \left(\frac{2.4049 \times 3 \times 10^{-3}}{2\pi \times 10^{-3}} \right)^2 = 1 + (1.148255)^2 = 1 + 1.3185$$

$$\epsilon_r \geq 2.3185$$

$$(b). \quad \lambda_{01} = 2.4049 \quad \xi = \sqrt{(\beta_0 a)^2 (\epsilon_r - 1) - x^2}$$

$$\text{At cutoff } \xi = 0, \quad \beta_{0c} = \omega_{co} \sqrt{\mu_0 \epsilon_0} = 2\pi f_c \sqrt{\mu_0 \epsilon_0}$$

$$(\beta_{0c})^2 (\epsilon_r - 1) = x^2 \Rightarrow \beta_{0c} = \frac{x}{\sqrt{\epsilon_r - 1}} \Rightarrow 2\pi f_{co} \sqrt{\mu_0 \epsilon_0} = \frac{x}{\sqrt{\epsilon_r - 1}}$$

$$f_{co} = \frac{1}{2\pi a} \cdot \frac{1}{\sqrt{\mu_0 \epsilon_0}} \cdot \frac{x}{\sqrt{\epsilon_r - 1}} = \frac{2.4049 (3 \times 10^8)}{2\pi (10^{-3}) / 2.3185 - 1} = \frac{2.4049 (3) \times 10^{11}}{2\pi \sqrt{1.3185}}$$

$$f_{co} = 10^{11} = 100 \times 10^9 = 100 \text{ GHz}$$

9.30

For TE^z modes

$$E_z = A_{mn} J_m(\beta_p p) [C_2 \cos(m\phi) + D_2 \sin(m\phi)] e^{-j\beta_p z} \quad (9-7)$$

(a). The B.C.s for PMC walls are that the tangential magnetic fields must vanish. Choosing the H_z component from (6-80), we can write that

$$H_z = -j \frac{1}{w\mu\epsilon} \left(\frac{\partial^2}{\partial z^2} + \beta_p^2 \right) E_z = -j \frac{1}{w\mu\epsilon} [-\beta_p^2 + \beta_p^2] A_{mn} J_m(\beta_p p) e^{-j\beta_p z}$$

$$H_z(p=a) = 0 = J_m(\beta_p a) \left[-j \frac{1}{w\mu\epsilon} (-\beta_p^2 + \beta_p^2) \right] A_{mn} e^{-j\beta_p z}$$

$$J_m(\beta_p a) = 0 \Rightarrow \beta_p a = \chi_{mn} \Rightarrow \beta_p = \frac{\chi_{mn}}{a} \quad n=1, 2, 3, \dots$$

$$H_z(\phi=0) = 0 \Rightarrow J_m(\beta_p p) \left[-j \frac{1}{w\mu\epsilon} (-\beta_p^2 + \beta_p^2) \right] A_{mn} [C_2(1) + D_2(0)] e^{-j\beta_p z} \Rightarrow C_2 = 0$$

$$H_z(\phi=\pi) = 0 = J_m(\beta_p p) \left[-j \frac{1}{w\mu\epsilon} (-\beta_p^2 + \beta_p^2) \right] A_{mn} [D_2 \sin(m\pi)] e^{-j\beta_p z} \Rightarrow \sin(m\pi) = 0$$

$$\sin(m\pi) = 0 \Rightarrow m\pi = \sin^{-1}(0) = \pm\pi \Rightarrow m = \pm = 1, 2, 3, \dots$$

$$\therefore E_z = B_{mn} J_m(\beta_p p) \sin(m\phi) e^{-j\beta_p z}$$

$$m = 1, 2, 3, \dots ; n = 1, 2, 3, \dots$$

$$\beta_p = w_c \sqrt{\mu\epsilon} = 2\pi f_c \sqrt{\mu\epsilon} = \beta_p = \chi_{mn}/a$$

$$(f_r) = \frac{1}{2\pi} \sqrt{\mu\epsilon} \left(\frac{\chi_{mn}}{a} \right) \quad \begin{cases} m = 1, 2, 3, \dots \\ n = 1, 2, 3, \dots \end{cases}$$

(b). From Table 9-2, for the TE_{11} dominant mode ($\chi_{11} = 3.8318$)

$$(f_r)^{TE}_{11} = \frac{20 \times 10^9 (3.8318)}{2\pi (1.5) \sqrt{81}} = \boxed{1.3552 \times 10^9 = 1.3552 \text{ GHz}}$$

9.31

For TM^2 Modes

$$A_z = B_{mn} J_m(\beta_p \rho) [C_2 \cos(m\phi) + D_2 \sin(m\phi)] e^{-j\beta_p z}$$

(a). The B.C.s for PMC walls are that the tangential magnetic fields must vanish. Choosing the H_ϕ and H_z components from (6-70), we can write that.

$$H_\phi = \frac{1}{\mu} \frac{1}{\rho} \frac{\partial A_z}{\partial \phi} = \frac{m}{\mu \rho} B_{mn} J_m'(\beta_p \rho) [-C_2 \sin(m\phi) + D_2 \cos(m\phi)] e^{-j\beta_p z}$$

$$H_\phi = -\frac{1}{\mu} \frac{\partial A_z}{\partial \phi} = -\frac{\beta_p B_{mn}}{\mu} J_m'(\beta_p \rho) [C_2 \cos(m\phi) + D_2 \sin(m\phi)] e^{-j\beta_p z}$$

$$H_\phi(\phi=0) = 0 = \frac{m}{\mu \rho} B_{mn} J_m'(\beta_p \rho) [-C_2(0) + D_2(\pm)] e^{-j\beta_p z} \Rightarrow D_2 = 0$$

$$H_\phi(\phi=\pi) = 0 = \frac{m}{\mu \rho} B_{mn} J_m'(\beta_p \rho) [C_2 \sin(m\pi)] e^{-j\beta_p z} \Rightarrow \sin(m\pi) = 0 \Rightarrow m\pi = \sin^{-1}(0)$$

$$m\pi = \sin^{-1}(0) = \pm\pi \Rightarrow m = \pm 1, \pm 2, \dots$$

$$H_\phi(\rho=a) = 0 = -\frac{\beta_p}{\mu} B_{mn} J_m'(\beta_p a) [C_2 \cos(m\phi)] e^{-j\beta_p z} \Rightarrow J_m'(\beta_p a) = 0$$

$$\beta_p a = \chi_{mn}^1 \Rightarrow \beta_p = \left(\frac{\chi_{mn}^1}{a} \right), \quad m=0,1,2,\dots, \quad n=1,2,3,\dots$$

$$\beta_p = \omega \sqrt{\mu \epsilon} = 2\pi f_c \sqrt{\mu \epsilon} = \beta_p = \left(\frac{\chi_{mn}^1}{a} \right) \Rightarrow (f_c)_{mn} = \frac{1}{2\pi \sqrt{\mu \epsilon} a} \chi_{mn}^1$$

$$(f_c)_{mn} = \frac{1}{2\pi \sqrt{\mu \epsilon}} \left(\frac{\chi_{mn}^1}{a} \right) \quad \begin{cases} m=0,1,2,\dots \\ n=1,2,3,\dots \end{cases}$$

(b). From Table 9-1, the smallest χ_{mn}^1 is $\chi_{11}^1 = 1.8412$. Therefore the lowest order mode is the TM_{11} or

$$(f_c)_{11}^{TM} = \frac{1}{2\pi \sqrt{\mu \epsilon} a} (\chi_{11}^1) = \frac{30 \times 10^9}{2\pi \sqrt{81}} \left(\frac{1.8412}{1.5} \right) = 0.6511 \times 10^9$$

$$(f_c)_{11}^{TM} = 0.6511 \times 10^9 = 0.6511 \text{ GHz}$$

$$F_z = [A_1 J_m(\beta_p \rho) + B_1 Y_m(\beta_p \rho)] [C_2 \cos(m\phi) + D_2 \sin(m\phi)] [C_3 \cos(\beta_2 z) + D_3 \sin(\beta_2 z)]$$

with $\beta_p^2 + \beta_2^2 = \beta^2$

$B_1 = 0$ because field must be finite at $\rho = 0$. Because of periodicity $m = 0, 1, 2, \dots$

$$\text{Thus } F_z(g, \phi, z) = A_1 J_m(\beta_p \rho) [C_2 \cos(m\phi) + D_2 \sin(m\phi)] [C_3 \cos(\beta_2 z) + D_3 \sin(\beta_2 z)]$$

$$H_\phi = -j \frac{1}{\omega \mu \epsilon \rho} \frac{\partial^2 F_z}{\partial \phi \partial z} = -j A_1 \frac{m \beta_2}{\omega \mu \epsilon \rho} J_m(\beta_p \rho) [-C_2 \sin(m\phi) + D_2 \cos(m\phi)] [-C_3 \sin(\beta_2 z) + D_3 \cos(\beta_2 z)]$$

$$H_\phi(\rho = a) = -j \frac{1}{\omega \mu \epsilon} J_m(\beta_p a) [-C_2 \sin(m\phi) + D_2 \cos(m\phi)] [-C_3 \sin(\beta_2 z) + D_3 \cos(\beta_2 z)] = 0$$

$$J_m(\beta_p a) = 0 \Rightarrow \beta_p a = \lambda_{mn} \Rightarrow \beta_p = \left(\frac{\lambda_{mn}}{a} \right)$$

$$H_\phi(z=0) = -j \frac{1}{\omega \mu \epsilon \rho} J_m(\beta_p \rho) [-C_2 \sin(m\phi) + D_2 \cos(m\phi)] [-C_3(0) + D_3(1)] = 0 \Rightarrow D_3 = 0$$

$$H_\phi(z=h) = -j \frac{1}{\omega \mu \epsilon \rho} J_m(\beta_p \rho) [-C_2 \sin(m\phi) + D_2 \cos(m\phi)] [-C_3 \sin(\beta_2 h)] = 0$$

$$\sin(\beta_2 h) = 0 \Rightarrow \beta_2 h = \sin^{-1}(0) = -p\pi \Rightarrow \beta_2 = \left(\frac{-p\pi}{h} \right), \quad p = 0, 1, 2, \dots$$

$$\beta_p^2 + \beta_2^2 = \left(\frac{\lambda_{mn}}{a} \right)^2 + \left(\frac{-p\pi}{h} \right)^2 = \beta_r^2 = \omega_r^2 \mu \epsilon \Rightarrow (f_r)_{mn\phi} = \frac{1}{2\pi\sqrt{\mu\epsilon}} \sqrt{\left(\frac{\lambda_{mn}}{a} \right)^2 + \left(\frac{-p\pi}{h} \right)^2}, \quad m = 0, 1, 2, 3, \dots \\ n = 1, 2, 3, \dots \quad p = 0, 1, 2, \dots$$

$$F_z = A_{mn\rho} J_m(\beta_p \rho) [C_2 \cos(m\phi) + D_2 \sin(m\phi)] \cos(\beta_2 z)$$

The electric and magnetic fields can be obtained using (6-80) and F_z from above.

$$A_2 = A_1 J_m(\beta_p \rho) [C_2 \cos(m\phi) + D_2 \sin(m\phi)] [C_3 \cos(\beta_2 z) + D_3 \sin(\beta_2 z)], \quad m = 0, 1, 2, \dots$$

$$H_\phi = -\frac{1}{\mu} \frac{\partial A_2}{\partial \rho} = -A_1 \frac{\beta_p}{\mu} J'_m(\beta_p \rho) [C_2 \cos(m\phi) + D_2 \sin(m\phi)] [C_3 \cos(\beta_2 z) + D_3 \sin(\beta_2 z)]$$

$$H_\phi(\rho = a) = -A_1 \frac{\beta_p}{\mu} J'_m(\beta_p a) [C_2 \cos(m\phi) + D_2 \sin(m\phi)] [C_3 \cos(\beta_2 z) + D_3 \sin(\beta_2 z)] = 0$$

$$J'_m(\beta_p a) = 0 \Rightarrow \beta_p a = \lambda'_{mn} \Rightarrow \beta_p = \left(\frac{\lambda'_{mn}}{a} \right)$$

$$H_\phi(z=0) = -A_1 \frac{\beta_p}{\mu} J'_m(\beta_p \rho) [C_2 \cos(m\phi) + D_2 \sin(m\phi)] [C_3(1) + D_3(0)] = 0 \Rightarrow C_3 = 0$$

$$H_\phi(z=h) = -A_1 \frac{\beta_p}{\mu} J'_m(\beta_p \rho) [C_2 \cos(m\phi) + D_2 \sin(m\phi)] [D_3 \sin(\beta_2 h)] = 0$$

$$\sin(\beta_2 h) = 0 \Rightarrow \beta_2 h = \sin^{-1}(0) = -p\pi \Rightarrow \beta_2 = \left(\frac{-p\pi}{h} \right), \quad p = 1, 2, 3, \dots$$

$$\beta_p^2 + \beta_2^2 = \left(\frac{\lambda'_{mn}}{a} \right)^2 + \left(\frac{-p\pi}{h} \right)^2 = \beta_r^2 = \omega_r^2 \mu \epsilon \Rightarrow (f_r)_{mn\rho} = \frac{1}{2\pi\sqrt{\mu\epsilon}} \sqrt{\left(\frac{\lambda'_{mn}}{a} \right)^2 + \left(\frac{-p\pi}{h} \right)^2}, \quad m = 0, 1, 2, \dots \\ n = 1, 2, 3, \dots \quad p = 1, 2, 3, \dots$$

$$A_2 = B_{mn\rho} J'_m(\beta_p \rho) [C_2 \cos(m\phi) + D_2 \sin(m\phi)] \sin(\beta_2 z)$$

The electric and magnetic fields can be obtained using (6-70) and A_2 from above.

- (a) PMC modeling predicts the lowest TE^z mode to be the TE_{010}^z mode provided that $h/a < 2.03$.

$$(b) Q = \omega \frac{2W_e}{P_d} = \omega \frac{2 \left[\frac{\epsilon'}{4} \iiint_V |\bar{E}|^2 dv \right]}{\frac{1}{2} \iiint_V \sigma |\bar{E}|^2 dv} = \frac{\omega \epsilon'}{\sigma}$$

In the above simplified expression, ϵ is actually $\epsilon(\omega)$ which can be further expanded into a real and imaginary part,

$$\epsilon(\omega) = \epsilon'(\omega) - j\epsilon''(\omega)$$

$$Q = \frac{\omega \epsilon'}{\sigma} = \frac{1}{\tan \delta_\epsilon}$$

- (c) The resonant frequency is calculated using equation (9-101) on p. 529 for the TE_{010}^z mode as follows,

$$(f_r)_{mnp}^{TE^z} = \frac{1}{2\pi\sqrt{\mu_d \epsilon_d}} \sqrt{\left(\frac{\chi_{mn}}{a}\right)^2 + \left(\frac{p\pi}{h}\right)^2} \quad m = 0, 1, 2, \dots \\ n = 1, 2, 3, \dots \\ p = 0, 1, 2, \dots$$

where $\chi_{01} = 2.4049$ and $m = 0, n = 1, p = 0$

$$(f_r)_{010}^{TE^z} = \frac{\chi_{01} c}{2\pi a \sqrt{\epsilon_d}} = \frac{(2.4049)(3 \times 10^8 \text{ m/s})}{2\pi(0.1148 \times 10^{-2} \text{ m})\sqrt{130}} = \boxed{8.773 \text{ GHz}}$$

- (d) Using the expression derived in part (b), the value of Q is,

$$Q = \frac{\omega \epsilon'}{\sigma} = \frac{1}{\tan \delta_\epsilon} = \frac{1}{(4 \times 10^{-4})} = \boxed{2,500}$$

9.34

For TE^2 : $A=0$, $F=\hat{a}_z F_z(\rho, \phi, z)$

$$(a). F_z = [A, J_m(\beta_{\rho}\rho) + B, Y_m(\beta_{\rho}\rho)] [C_2 \cos(m\phi) + D_2 \sin(m\phi)] [C_3 \cos(\beta_z z) + D_3 \sin(\beta_z z)]$$

$$\text{B.c's: } 1. E_p(0 \leq \rho \leq a, \phi = 0, z) = E_p(0 \leq \rho \leq a, \phi = \pi, z) = 0$$

$$2. H_\phi(\rho = a, 0 \leq \phi \leq \pi, z) = 0 \text{ or } H_z(\rho = a, 0 \leq \phi \leq \pi, z) = 0$$

$$3. H_\phi(0 \leq \rho \leq a, 0 \leq \phi \leq \pi, z = 0) = 0 \text{ or } H_p(0 \leq \rho \leq a, 0 \leq \phi \leq \pi, z = 0) = 0$$

$$H_\phi(0 \leq \rho \leq a, 0 \leq \phi \leq \pi, z = l) = 0 \text{ or } H_p(0 \leq \rho \leq a, 0 \leq \phi \leq \pi, z = l) = 0$$

4. E & H field components finite everywhere $\Rightarrow B_1 = 0$

$$F_z = A_{mn} J_m(\beta_{\rho}\rho) [C_2 \cos(m\phi) + D_2 \sin(m\phi)] [C_3 \cos(\beta_z z) + D_3 \sin(\beta_z z)]$$

$$(b). E_p = -\frac{1}{\epsilon_0} \frac{\partial F_z}{\partial \phi} = -\frac{\omega}{\epsilon_0} A_{mn} J_m(\beta_{\rho}\rho) [-C_2 \sin(m\phi) + D_2 \cos(m\phi)]$$

$$\text{Using (1): } E_p(\phi = 0) = -\frac{\omega}{\epsilon_0} A_{mn} J_m(\beta_{\rho}\rho) [-C_2(0) + D_2(1)] = 0 \Rightarrow D_2 = 0$$

$$\text{Using (2): } E_p(\phi = \pi) = -\frac{\omega}{\epsilon_0} A_{mn} J_m(\beta_{\rho}\rho) [-C_2 \sin(m\pi)] = 0$$

$$\Rightarrow \sin(m\pi) = 0 \Rightarrow m\pi = \sin^{-1}(0) = q\pi \Rightarrow m = q, q = m = 0, 1, 2, 3, \dots$$

$$m = q = 0, 1, 2, 3, \dots$$

$$H_\phi = -j \frac{1}{\omega \mu_0 \epsilon_0} \frac{\partial^2 F_z}{\partial \phi \partial z} = +j \frac{\omega \beta_z}{\omega \mu_0 \epsilon_0} A_{mn} C_2 J_m(\beta_{\rho}\rho) C_2 \sin(m\phi) [-C_3 \sin(\beta_z z) + D_3 \cos(\beta_z z)]$$

$$\text{Using (2): } H_\phi(\rho = a) = +j \frac{\omega \beta_z}{\omega \mu_0 \epsilon_0} A_{mn} C_2 J_m(\beta_{\rho}\rho) \sin(m\phi) = 0$$

$$\Rightarrow J_m(\beta_{\rho}a) = 0 \Rightarrow \beta_{\rho}a = \lambda_{mn} \Rightarrow \beta_{\rho} = \left(\frac{\lambda_{mn}}{a} \right)$$

$$\beta_{\rho} = \left(\frac{\lambda_{mn}}{a} \right), \quad m = 0, 1, 2, \dots, \quad n = 1, 2, 3, \dots$$

$$\text{Using (3): } H_\phi(z = 0) = +j \frac{\omega \beta_z}{\omega \mu_0 \epsilon_0} A_{mn} C_2 J_m(\beta_{\rho}\rho) \sin(m\phi) [-C_3(0) + D_3(1)] = 0 \Rightarrow D_3 = 0$$

$$\text{Using (3): } H_\phi(z = l) = +j \frac{\omega \beta_z}{\omega \mu_0 \epsilon_0} A_{mn} C_2 J_m(\beta_{\rho}\rho) \sin(m\phi) [-C_3 \sin(\beta_z l)] = 0$$

$$\sin(\beta_z l) = 0 \Rightarrow \beta_z l = \sin^{-1}(0) = p\pi$$

$$\beta_z = \left(\frac{p\pi}{l} \right), \quad p = 0, 1, 2, \dots$$

9.34 cont'd.

9.34 cont'd.

$$F_z(p, \phi, z) = A_{mn} J_m(\beta_p p) \cos(m\phi) \cos(\beta_2 z)$$

$$\beta_p = \frac{\alpha_{mn}}{a} \quad m=1, 2, \dots$$

$$\beta_2 = \frac{cp\pi}{l} \quad n=1, 2, \dots$$

$$(c). \quad (\beta_p)^2 + (\beta_2)^2 = \beta_r^2 = \alpha_r^2 \mu \epsilon = (2\pi f_r)^2 \mu \epsilon$$

$$(f_r)_{mn} = \frac{1}{2\pi\sqrt{\mu\epsilon}} \sqrt{(\beta_p)^2 + (\beta_2)^2} = \frac{1}{2\pi\sqrt{\mu\epsilon}} \sqrt{\left(\frac{\alpha_{mn}}{a}\right)^2 + \left(\frac{cp\pi}{l}\right)^2}$$

$$(f_r)_{010} = \frac{1}{2\pi\sqrt{\mu\epsilon}} \sqrt{\left(\frac{\alpha_{01}}{a}\right)^2 + \left(\frac{cp\pi}{l}\right)^2}$$

Dominant: From Table 9-2 $\Rightarrow \alpha_{mn}|_{m=0} = \alpha_{01}|_{p=1} = 2.4049$

$$p=0 \Rightarrow \beta_2=0$$

$$(f_r)_{010}^{TE} = \frac{1}{2\pi\sqrt{\mu\epsilon}} \sqrt{\left(\frac{\alpha_{01}}{a}\right)^2} = \frac{\alpha_{01}}{2\pi a \sqrt{\mu_0 \epsilon_0 \epsilon_r}}$$

$$= \frac{\alpha_{01}}{2\pi a \sqrt{\mu_0 \epsilon_0 \epsilon_r}} = \frac{(2.4049)(30 \times 10^9)}{2\pi(2)\sqrt{81}}$$

$$(f_r)_{010}^{TE} = 0.63792 \times 10^9$$

[Q.35] For TM^z : $A = \hat{A}_z A_z(\rho, \phi, z)$, $E = 0$

$$(a) A_z = [A_1 J_m(\beta_p \rho) + B_1 Y_m(\beta_p \rho)] [C_2 \cos(m\phi) + D_2 \sin(m\phi)] [C_3 \cos(\beta_z z) + D_3 \sin(\beta_z z)]$$

$$\text{B.C.s: } 1. E_\rho (0 \leq \rho \leq a, \phi = 0, z) = E_\rho (0 \leq \rho \leq a, \phi = \pi, z) = 0 \text{ or } E_z (\phi = 0) = E_z (\phi = \pi) = 0$$

$$2. H_\phi (\rho = a, 0 \leq \phi \leq \pi, z) = 0 \text{ or } H_z (\rho = a, 0 \leq \phi \leq \pi, z) = 0$$

$$3. H_\rho (0 \leq \rho \leq a, 0 \leq \phi \leq \pi, z = 0) = 0 \text{ or } H_\rho (0 \leq \rho \leq a, 0 \leq \phi \leq \pi, z = l) = 0$$

$$H_\phi (0 \leq \rho \leq a, 0 \leq \phi \leq \pi, z = l) = 0 \text{ or } H_\rho (0 \leq \rho \leq a, 0 \leq \phi \leq \pi, z = l) = 0$$

4. E & H field components are finite everywhere $\Rightarrow B_1 = 0$

$$\text{Thus } A_z = B_{mn} J_m(\beta_p \rho) [C_2 \cos(m\phi) + D_2 \sin(m\phi)] [C_3 \cos(\beta_z z) + D_3 \sin(\beta_z z)]$$

$$(b) E_z = -j \frac{1}{\omega \mu \epsilon} \left(\frac{\partial^2}{\partial z^2} + \beta_p^2 \right) A_z = -j \frac{\beta_z B_{mn}}{\omega \mu \epsilon} \underbrace{\left(-(\beta_z)^2 + \beta_p^2 \right)}_{\beta_p^2} J_m(\beta_p \rho) [C_2 \cos(m\phi) + D_2 \sin(m\phi)] [C_3 \cos(\beta_z z) + D_3 \sin(\beta_z z)]$$

$$(1) E_z (\phi = 0) = j B_{mn} \frac{\beta_z \beta_p^2}{\omega \mu \epsilon} J_m(\beta_p \rho) [C_2 (1) + D_2 (0)] [C_3 \cos(\beta_z z) + D_3 \sin(\beta_z z)] = 0 \Rightarrow C_2 = 0$$

$$(1) E_z (\phi = \pi) = -j B_{mn} \frac{\beta_z \beta_p^2}{\omega \mu \epsilon} J_m(\beta_p \rho) D_2 \sin(m\pi) [C_3 \cos(\beta_z z) + D_3 \sin(\beta_z z)] = 0$$

$$\sin(m\pi) = 0 \Rightarrow m\pi = \sin(0) = 0, q = m = 1, 2, \dots$$

$$m = q = 1, 2, 3, \dots$$

$$H_\phi = -\frac{1}{\mu} \frac{\partial A_z}{\partial \rho} = -B_{mn} D_2 \frac{1}{\mu} J_m'(\beta_p \rho) \sin(m\phi) [C_3 \cos(\beta_z z) + D_3 \sin(\beta_z z)]$$

$$(2) H_\phi (\rho = a) = -B_{mn} D_2 \frac{1}{\mu} J_m'(\beta_p a) \sin(m\phi) [C_3 \cos(\beta_z z) + D_3 \sin(\beta_z z)] = 0$$

$$J_m'(\beta_p a) = 0 \Rightarrow \beta_p a = x_m' \Rightarrow \beta_p = \left(\frac{x_m'}{a} \right)$$

$$\beta_p = \left(\frac{x_m'}{a} \right), m = 1, 2, 3, \dots, n = 1, 2, 3, \dots$$

(3)

$$H_\phi (z = 0) = -B_{mn} D_2 \left(\frac{1}{\mu} \right) J_m'(\beta_p \rho) \sin(m\phi) [C_3 (1) + D_3 (0)] = 0 \Rightarrow C_3 = 0$$

$$H_\phi (z = l) = -B_{mn} D_2 \left(\frac{1}{\mu} \right) J_m'(\beta_p \rho) \sin(m\phi) D_3 \sin(\beta_z l) = 0$$

$$\sin(\beta_z l) = 0 \Rightarrow \beta_z l = \sin(0) = p\pi, p = 1, 2, \dots$$

$$\beta_z = \left(\frac{p\pi}{l} \right), p = 1, 2, 3, \dots$$

$$A_z (\rho, \phi, z) = B_{mn} J_m(\beta_p \rho) \sin(m\phi) \sin(\beta_z z)$$

$$\beta_p = \left(\frac{x_m'}{a} \right), m = 1, 2, 3, \dots, n = 1, 2, 3, \dots$$

$$\beta_z = \left(\frac{p\pi}{l} \right), p = 1, 2, 3, \dots$$

Cont'd

9.35 cont'd

$$(c) (\beta_p)^2 + (\beta_z)^2 = \beta_r^2 = \omega_r \mu \epsilon = (2\pi f_r)^2 \mu \epsilon$$

$$(f_r)_{mnp} = \frac{1}{2\pi\sqrt{\mu\epsilon}} \sqrt{(\beta_p)^2 + \beta_z^2} = \frac{1}{2\pi\sqrt{\mu\epsilon}} \sqrt{\left(\frac{\lambda_{mn}'}{a}\right)^2 + \left(\frac{p\pi}{l}\right)^2}$$

$$(f_r)_{mnp} = \frac{1}{2\pi\sqrt{\mu\epsilon}} \sqrt{\left(\frac{\lambda_{mn}'}{a}\right)^2 + \left(\frac{p\pi}{l}\right)^2}$$

The dominant mode, based on the values of λ'_{mn} from Table 9-1, are: $m=1, n=1 \Rightarrow \lambda'_{11} = 1.841, p=1$

$$(f_r)_{111} = \frac{1}{2\pi\sqrt{\mu\epsilon}} \sqrt{\left(\frac{1.841}{a}\right)^2 + \left(\frac{\pi}{l}\right)^2}$$

$$a = 2 \text{ cm}, l = 4 \text{ cm}, \epsilon_r = 81$$

$$(f_r)_{111} = \frac{1}{2\pi\sqrt{\epsilon_r\mu_0\epsilon_0}} \sqrt{\left(\frac{1.841}{2}\right)^2 + \left(\frac{\pi}{4}\right)^2}$$

$$= \frac{30 \times 10^9}{2\pi\sqrt{81}} \sqrt{\left(\frac{1.841}{2}\right)^2 + \left(\frac{\pi}{4}\right)^2} = \frac{30 \times 10^9}{2\pi(9)} \sqrt{(0.9205)^2 + (0.7854)^2}$$

$$= 0.5305 \sqrt{0.8473 + 0.6169} \times 10^9 = (0.5305 \sqrt{1.4642}) \times 10^9$$

$$= (0.5305) 1.21 \times 10^9 = 0.6419 \times 10^9$$

$$(f_r)_{111}^{TM} = 0.6419 \times 10^9 = 0.6419 \text{ GHz}$$

9.36

(a), (b) For $T \in \mathbb{Z}$ ($E_z = 0$)

$$F_2(p, \phi, z) = [A_1 J_m(\beta_p^d p) + B_1 Y_m(\beta_p^d p)] [C_2 \cos(m\phi) + D_2 \sin(m\phi)] [C_3 \cos(\beta_z z) + D_3 \sin(\beta_z z)]$$

B.C.'s: Fields must be finite everywhere including $p=0$.

$$H_p(0 \leq p \leq a, 0 \leq \phi \leq \phi_0, z=0) = H_p(0 \leq p \leq a, 0 \leq \phi \leq \phi_0, z=h) = 0$$

$$H_p(0 \leq p \leq a, \phi = 0, 0 \leq z \leq h) = H_p(0 \leq p \leq a, \phi = \phi_0, 0 \leq z \leq h) = 0$$

$$H_\phi(p=a, 0 \leq \phi \leq \phi_0, 0 \leq z \leq h) = 0$$

For the fields to be finite everywhere, including $p=0$, $B_1 = 0$.

$$F_2(p, \phi, z) = A_1 J_m(\beta_p^d p) [C_2 \cos(m\phi) + D_2 \sin(m\phi)] [C_3 \cos(\beta_z z) + D_3 \sin(\beta_z z)]$$

$$H_p = -j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 F_2}{\partial p \partial z} \quad H_\phi = -j \frac{1}{\omega \mu \epsilon p} \frac{\partial^2 F_2}{\partial \phi \partial z}$$

$$H_p = -j A_1 \frac{\beta_p \beta_z}{\omega \mu \epsilon} J_m'(\beta_p p) [C_2 \cos(m\phi) + D_2 \sin(m\phi)] [C_3 \sin(\beta_z z) + D_3 \cos(\beta_z z)]$$

$$H_p(z=0) = -j A_1 \frac{\beta_p \beta_z}{\omega \mu \epsilon} J_m'(\beta_p p) [C_2(0) + D_2(0)] = 0 \Rightarrow D_2 = 0$$

$$H_p(z=h) = -j A_1 \frac{\beta_p \beta_z}{\omega \mu \epsilon} J_m'(\beta_p p) [C_2(h) + D_2(h)] = 0$$

$$\sin(\beta_z h) = 0 \Rightarrow \beta_z h = \sin^{-1}(0) = p\pi \Rightarrow \beta_z = \left(\frac{-p\pi}{h}\right), p=0, 1, 2, \dots$$

$$H_\phi(\phi=0) = -j A_1 \frac{\beta_p \beta_z}{\omega \mu \epsilon} J_m'(\beta_p p) [C_2(1) + D_2(0)] = 0 \Rightarrow C_2 = 0$$

$$H_\phi(\phi=\phi_0) = -j A_1 \frac{\beta_p \beta_z}{\omega \mu \epsilon} J_m'(\beta_p p) [D_2 \sin(m\phi_0)] = 0$$

$$\sin(m\phi_0) = 0 \Rightarrow m\phi_0 = \sin^{-1}(0) = q\pi \Rightarrow m = \frac{q\pi}{\phi_0}, q=1, 2, 3, \dots$$

$$F_2 = A_{mn} J_m(\beta_p p) \sin(m\phi) \cos(\beta_z z)$$

$$H_\phi = -j \frac{1}{\omega \mu \epsilon p} \frac{\partial^2 F_2}{\partial \phi \partial z} = j A_{mn} \frac{m \beta_z}{\omega \mu \epsilon p} J_m(\beta_p p) \cos(m\phi) \sin(\beta_z z)$$

$$H_\phi(p=a) = j A_{mn} \frac{m \beta_z}{\omega \mu \epsilon p} J_m(\beta_p a) \cos(m\phi) \sin(\beta_z z) = 0 \Rightarrow J_m(\beta_p a) = 0 \Rightarrow \beta_p a = \gamma_{mn}$$

$$\beta_p = \left(\frac{\gamma_{mn}}{a}\right), m = q\pi/\phi_0, q=1, 2, 3, \dots$$

$$F_2 = A_{mn} J_m(\beta_p p) \sin(m\phi) \cos(\beta_z z)$$

Since the circumferential surface is identical to that of Problem 9.32 the boundary conditions over this surface are identical to those of Problem 9.32. From the solution of 9.32 we have that

$$F_z = A_1 J_m(\beta_p p) [C_2 \cos(m\phi) + D_2 \sin(m\phi)] [C_3 \cos(\beta_z z) + D_3 \sin(\beta_z z)]$$

with

$$\beta_p = \left(\frac{\chi_{mn}}{a} \right), \quad m = 0, 1, 2, 3, \dots$$

$$E_p = -\frac{1}{\epsilon} \frac{1}{p} \frac{\partial F_z}{\partial \phi} = -A_1 \frac{m}{\epsilon} \frac{1}{p} J_m'(\beta_p p) [-C_2 \sin(m\phi) + D_2 \cos(m\phi)] [C_3 \cos(\beta_z z) + D_3 \sin(\beta_z z)]$$

Applying the boundary conditions at $z=0$ and $z=h$ leads to

$$E_p(z=0) = -A_1 \frac{m}{\epsilon} \frac{1}{p} J_m(\beta_p p) [-C_2 \sin(m\phi) + D_2 \cos(m\phi)] [C_3(1) + D_3(0)] = 0 \Rightarrow C_3 = 0$$

$$E_p(z=h) = -A_1 \frac{m}{\epsilon} \frac{1}{p} J_m(\beta_p p) [-C_2 \sin(m\phi) + D_2 \cos(m\phi)] [D_3 \sin(\beta_z h)] = 0$$

$$\sin(\beta_z h) = 0 \Rightarrow \beta_z h = \sin^{-1}(0) = -p\pi \Rightarrow \beta_z = \frac{-p\pi}{h}, \quad p = -1, 2, 3, \dots$$

Therefore

$$F_z = A_{mn} J_m(\beta_p p) [C_2 \cos(m\phi) + D_2 \sin(m\phi)] \sin(\beta_z z), \quad \begin{matrix} m=0, 1, 2, 3, \dots \\ n=1, 2, 3, \dots \\ p=-1, 2, 3, \dots \end{matrix}$$

$$(f_r)^{TE}_{mn\bar{p}} = \frac{1}{2\pi\sqrt{\mu\epsilon}} \sqrt{\left(\frac{\chi_{mn}}{a}\right)^2 + \left(\frac{-p\pi}{h}\right)^2}$$

Again from Problem 9.32 solution

$$A_z = A_1 J_m(\beta_p p) [C_2 \cos(m\phi) + D_2 \sin(m\phi)] [C_3 \cos(\beta_z z) + D_3 \sin(\beta_z z)]$$

$$\beta_p = \left(\frac{\chi_{mn}}{a} \right), \quad \begin{matrix} m=0, 1, 2, \dots \\ n=1, 2, 3, \dots \end{matrix}$$

Applying the boundary conditions at $z=0$ and $z=h$ leads to

$$E_p = -j \frac{1}{\omega\mu\epsilon} \frac{\partial^2 A_z}{\partial p \partial z} = -j A_1 \frac{\beta_p \beta_z}{\omega\mu\epsilon} J_m'(\beta_p p) [C_2 \cos(m\phi) + D_2 \sin(m\phi)] [-C_3 \sin(\beta_z z) + D_3 \cos(\beta_z z)]$$

$$E_p(z=0) = -j A_1 \frac{\beta_p \beta_z}{\omega\mu\epsilon} J_m'(\beta_p p) [C_2 \cos(m\phi) + D_2 \sin(m\phi)] [-C_3(0) + D_3(1)] = 0 \Rightarrow D_3 = 0$$

$$E_p(z=h) = -j A_1 \frac{\beta_p \beta_z}{\omega\mu\epsilon} J_m'(\beta_p p) [C_2 \cos(m\phi) + D_2 \sin(m\phi)] [-C_3 \sin(\beta_z h)] = 0$$

$$\sin(\beta_z h) = 0 \Rightarrow \beta_z h = \sin^{-1}(0) = -p\pi \Rightarrow \beta_z = \frac{-p\pi}{h}, \quad p = 0, 1, 2, \dots$$

Therefore

$$A_z = B_{mn} J_m(\beta_p p) [C_2 \cos(m\phi) + D_2 \sin(m\phi)] \cos(\beta_z z), \quad \begin{matrix} m=0, 1, 2, 3, \dots \\ n=1, 2, 3, \dots \\ p=0, 1, 2, \dots \end{matrix}$$

$$(f_r)^{TM}_{mn\bar{p}} = \frac{1}{2\pi\sqrt{\mu\epsilon}} \sqrt{\left(\frac{\chi_{mn}}{a}\right)^2 + \left(\frac{-p\pi}{h}\right)^2}$$

9.38 Since the only difference between Problems 9.37 and 9.38 is the angular extend of the cross section, then from Problems 9.32 and 9.37 solutions

$$H_p = -j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 F_z}{\partial \rho \partial z} = -j A_1 \frac{\beta_p \beta_z}{\omega \mu \epsilon} J_m'(\beta_p \rho) [C_2 \cos(m\phi) + D_2 \sin(m\phi)] [C_3 \sin(\beta_z z) + D_3 \cos(\beta_z z)]$$

$$H_p(t=0) = -j A_1 \frac{\beta_p \beta_z}{\omega \mu \epsilon} J_m'(\beta_p \rho) [C_3 \sin(\beta_z z) + D_3 \cos(\beta_z z)] [C_2(1) + D_2(0)] = 0 \Rightarrow C_2 = 0$$

$$H_p(\phi=\phi_0) = -j A_1 \frac{\beta_p \beta_z}{\omega \mu \epsilon} J_m'(\beta_p \rho) [C_3 \sin(\beta_z z) + D_3 \cos(\beta_z z)] D_2 \sin(m\phi_0) = 0$$

$$\sin(m\phi_0) = 0 \Rightarrow m\phi_0 = \sin^{-1}(0) = -p\pi \Rightarrow m = -p\left(\frac{\pi}{\phi_0}\right), p = 1, 2, 3, \dots$$

From Problem 9.32 solution

$$\beta_p = \left(\frac{\alpha_{mn}}{a}\right), n = 1, 2, 3, \dots$$

TM^z

$$H_p = \frac{1}{\hbar p} \frac{\partial A_z}{\partial \phi} = A_1 J_m(\beta_p \rho) \frac{m}{\hbar p} [C_2 \sin(m\phi) + D_2 \cos(m\phi)] [C_3 \cos(\beta_z z) + D_3 \sin(\beta_z z)]$$

$$H_p(\phi=0) = A_1 \frac{m}{\hbar p} J_m(\beta_p \rho) [-C_2(0) + D_2(1)] [C_3 \cos(\beta_z z) + D_3 \sin(\beta_z z)] = 0 \Rightarrow D_2 = 0$$

$$H_p(\phi=\phi_0) = A_1 \frac{m}{\hbar p} J_m(\beta_p \rho) [-C_2 \sin(m\phi_0)] [C_3 \cos(\beta_z z) + D_3 \sin(\beta_z z)] = 0$$

$$\sin(m\phi_0) = 0 \Rightarrow m\phi_0 = \sin^{-1}(0) = -p\pi \Rightarrow m = -p\left(\frac{\pi}{\phi_0}\right), p = 0, 1, 2, 3, \dots$$

From the solution of Problem 9.32

$$\beta_p = \left(\frac{\alpha_{mn}}{a}\right), n = 1, 2, 3, \dots$$

9.39 (a) TE^z

For the region: $(a \leq \rho \leq b, 0 \leq \phi \leq 2\pi, z)$

$$F_z(\rho, \phi, z) = [A_1 J_m(\beta_\rho^d \rho) + B_1 Y_m(\beta_\rho^d \rho)] \cdot [C_2 \cos(m\phi) + D_2 \sin(m\phi)] \cdot [C_3 \cos(\beta_z z) + D_3 \sin(\beta_z z)]$$

$$E_\phi = \frac{1}{\epsilon_d} \frac{\partial F_z}{\partial \rho} = \frac{\beta_\rho^d}{\epsilon_d} [A_1 J_m(\beta_\rho^d \rho) + B_1 Y_m(\beta_\rho^d \rho)] \cdot [C_2 \cos(m\phi) + D_2 \sin(m\phi)] \\ \times [C_3 \cos(\beta_z z) + D_3 \sin(\beta_z z)]$$

$$H_\phi = -j \frac{1}{\omega \mu_d \epsilon_d} \frac{1}{\rho} \frac{\partial^2 F_z}{\partial \phi \partial z} = -j \frac{m \beta_z}{\omega \mu_d \epsilon_d} \frac{1}{\rho} [A_1 J_m(\beta_\rho^d \rho) + B_1 Y_m(\beta_\rho^d \rho)] \\ \times [-C_2 \sin(m\phi) + D_2 \cos(m\phi)] \cdot [-C_3 \sin(\beta_z z) + D_3 \cos(\beta_z z)]$$

(Cont'd)

Prob. 9.33 cont.

Apply BC #1: $E_\phi(a \leq \rho \leq b, 0 \leq \phi \leq 2\pi, z = 0) = 0$

$$E_\phi(a \leq \rho \leq b, 0 \leq \phi \leq 2\pi, z = 0) = \frac{\beta_\rho^d}{\varepsilon_d} [A_1 J_m(\beta_\rho^d \rho) + B_1 Y_m(\beta_\rho^d \rho)] \cdot [C_2 \cos(m\phi) + D_2 \sin(m\phi)] \cdot [C_3 \cos(0) + D_3 \sin(0)] = 0$$

$$\frac{\beta_\rho^d}{\varepsilon_d} [A_1 J_m(\beta_\rho^d \rho) + B_1 Y_m(\beta_\rho^d \rho)] \cdot [C_2 \cos(m\phi) + D_2 \sin(m\phi)] \cdot [C_3(1) + D_3(0)] = 0 \Rightarrow C_3 = 0$$

Apply BC #2: $E_\phi(a \leq \rho \leq b, 0 \leq \phi \leq 2\pi, z = h) = 0$

$$E_\phi(a \leq \rho \leq b, 0 \leq \phi \leq 2\pi, z = h) = \frac{\beta_\rho^d}{\varepsilon_d} [A_1 J_m(\beta_\rho^d \rho) + B_1 Y_m(\beta_\rho^d \rho)] \cdot [C_2 \cos(m\phi) + D_2 \sin(m\phi)] \cdot [D_3 \sin(\beta_z h)] = 0$$

$$\sin(\beta_z h) = 0 \Rightarrow \beta_z h = p\pi \Rightarrow \beta_z = \frac{p\pi}{h} \quad p = 0, 1, 2, \dots$$

Apply BC #3: $H_\phi(\rho = b, 0 \leq \phi \leq 2\pi, 0 \leq z \leq h) = 0$

$$H_\phi(\rho = b, 0 \leq \phi \leq 2\pi, 0 \leq z \leq h) = -j \frac{m\beta_z}{\omega \mu_d \varepsilon_d} \frac{1}{b} [A_1 J_m(\beta_\rho^d \cdot b) + B_1 Y_m(\beta_\rho^d \cdot b)]$$

$$\times [-C_2 \sin(m\phi) + D_2 \cos(m\phi)] \cdot [D_3 \cos(\beta_z z)] = 0$$

$$A_1 J_m(\beta_\rho^d \cdot b) + B_1 Y_m(\beta_\rho^d \cdot b) = 0$$

Apply BC #4: $H_\phi(\rho = a, 0 \leq \phi \leq 2\pi, 0 \leq z \leq h) = 0$

$$H_\phi(\rho = a, 0 \leq \phi \leq 2\pi, 0 \leq z \leq h) = -j \frac{m\beta_z}{\omega \mu_d \varepsilon_d} \frac{1}{a} [A_1 J_m(\beta_\rho^d \cdot a) + B_1 Y_m(\beta_\rho^d \cdot a)]$$

$$\times [-C_2 \sin(m\phi) + D_2 \cos(m\phi)] \cdot [D_3 \cos(\beta_z z)] = 0$$

$$A_1 J_m(\beta_\rho^d \cdot a) + B_1 Y_m(\beta_\rho^d \cdot a) = 0$$

$$[A_1 J_m(\beta_\rho^d \cdot a) + B_1 Y_m(\beta_\rho^d \cdot a)] = 0$$

$$[A_1 J_m(\beta_\rho^d \cdot b) + B_1 Y_m(\beta_\rho^d \cdot b)] = 0$$

(Cont'd)

Prob. 9.39 cont.

$$Fg = \begin{bmatrix} J_m(\beta_\rho^d \cdot a) & Y_m(\beta_\rho^d \cdot a) \\ J_m(\beta_\rho^d \cdot b) & Y_m(\beta_\rho^d \cdot b) \end{bmatrix} \cdot \begin{bmatrix} A_1 \\ B_1 \end{bmatrix} = 0$$

The equation $Fg = 0$ has a nontrivial solution provided that the determinant of F is equal to zero. $\det(F) = 0$

$$\det(F) = J_m(\beta_\rho^d \cdot a) \cdot Y_m(\beta_\rho^d \cdot b) - J_m(\beta_\rho^d \cdot b) \cdot Y_m(\beta_\rho^d \cdot a) = 0$$

Therefore, the eigenvalues are obtained as solutions to:

$$J_m(\beta_\rho^d \cdot a) \cdot Y_m(\beta_\rho^d \cdot b) - J_m(\beta_\rho^d \cdot b) \cdot Y_m(\beta_\rho^d \cdot a) = 0, \quad m = 0, 1, 2, \dots$$

$$(\beta_\rho^d)^2 + \beta_z^2 = \beta_d^2$$

$$\beta_r = \omega_r \sqrt{\mu_d \epsilon_d} = 2\pi f_r \sqrt{\mu_d \epsilon_d} = \sqrt{(\beta_\rho^d)^2 + \left(\frac{p\pi}{h}\right)^2}$$

Expression for the resonant frequency for TE^z modes:

$$f_r = \frac{1}{2\pi \sqrt{\mu_d \epsilon_d}} \sqrt{(\beta_\rho^d)^2 + \left(\frac{p\pi}{h}\right)^2}$$

(b) **TM^z MODES**

For the region: $(a \leq \rho \leq b, 0 \leq \phi \leq 2\pi, z)$

$$A_z(\rho, \phi, z) = [A_1 J_m(\beta_\rho^d \rho) + B_1 Y_m(\beta_\rho^d \rho)] \cdot [C_2 \cos(m\phi) + D_2 \sin(m\phi)] \cdot [C_3 \cos(\beta_z z) + D_3 \sin(\beta_z z)]$$

$$E_\phi = -j \frac{1}{\omega_r \mu_d \epsilon_d} \frac{1}{\rho} \frac{\partial^2 A_z}{\partial \phi \partial z} = -j \frac{m \beta_z}{\omega_r \mu_d \epsilon_d} \frac{1}{\rho} [A_1 J_m(\beta_\rho^d \rho) + B_1 Y_m(\beta_\rho^d \rho)] \times [-C_2 \sin(m\phi) + D_2 \cos(m\phi)] \cdot [-C_3 \sin(\beta_z z) + D_3 \cos(\beta_z z)]$$

$$H_\phi = -\frac{1}{\mu_d} \frac{\partial A_z}{\partial \rho} = -\frac{\beta_\rho^d}{\mu_d} [A_1 J_m(\beta_\rho^d \rho) + B_1 Y_m(\beta_\rho^d \rho)] \times [C_2 \cos(m\phi) + D_2 \sin(m\phi)] \cdot [C_3 \cos(\beta_z z) + D_3 \sin(\beta_z z)]$$

(Cont'd)

Prob. 9.39 Cont.

Apply BC #1: $E_\phi(a \leq \rho \leq b, 0 \leq \phi \leq 2\pi, z = 0) = 0$

$$E_\phi(a \leq \rho \leq b, 0 \leq \phi \leq 2\pi, z = 0) = -j \frac{m\beta_z}{\omega_r \mu_d \epsilon_d} \frac{1}{\rho} [A_1 J_m(\beta_\rho^d \rho) + B_1 Y_m(\beta_\rho^d \rho)] \\ \times [-C_2 \sin(m\phi) + D_2 \cos(m\phi)] \cdot [-C_3 \sin(\beta_z 0) + D_3 \cos(\beta_z 0)] = 0 \\ -j \frac{m\beta_z}{\omega_r \mu_d \epsilon_d} \frac{1}{\rho} [A_1 J_m(\beta_\rho^d \rho) + B_1 Y_m(\beta_\rho^d \rho)] \cdot [-C_2 \sin(m\phi) + D_2 \cos(m\phi)] \cdot [-C_3(0) + D_3(1)] = 0 \Rightarrow D_3 = 0$$

Apply BC #2: $E_\phi(a \leq \rho \leq b, 0 \leq \phi \leq 2\pi, z = h) = 0$

$$E_\phi(a \leq \rho \leq b, 0 \leq \phi \leq 2\pi, z = h) = -j \frac{m\beta_z}{\omega_r \mu_d \epsilon_d} \frac{1}{\rho} [A_1 J_m(\beta_\rho^d \rho) + B_1 Y_m(\beta_\rho^d \rho)] \\ \times [-C_2 \sin(m\phi) + D_2 \cos(m\phi)] \cdot [-C_3 \sin(\beta_z h)] = 0 \\ \sin(\beta_z h) = 0 \Rightarrow \beta_z h = p\pi \Rightarrow \boxed{\beta_z = \frac{p\pi}{h} \quad p = 0, 1, 2, \dots}$$

Apply BC #3: $H_\phi(\rho = b, 0 \leq \phi \leq 2\pi, 0 \leq z \leq h) = 0$

$$H_\phi(\rho = b, 0 \leq \phi \leq 2\pi, 0 \leq z \leq h) = -\frac{\beta_\rho^d}{\mu_d} [A_1 J_m(\beta_\rho^d \cdot b) + B_1 Y_m(\beta_\rho^d \cdot b)] \\ \times [C_2 \cos(m\phi) + D_2 \sin(m\phi)] \cdot [C_2 \cos(\beta_z z) + D_2 \sin(\beta_z z)] = 0$$

$$A_1 J_m(\beta_\rho^d \cdot b) + B_1 Y_m(\beta_\rho^d \cdot b) = 0$$

Apply BC #4: $H_\phi(\rho = a, 0 \leq \phi \leq 2\pi, 0 \leq z \leq h) = 0$

$$H_\phi(\rho = a, 0 \leq \phi \leq 2\pi, 0 \leq z \leq h) = -\frac{\beta_\rho^d}{\mu_d} [A_1 J_m(\beta_\rho^d \cdot a) + B_1 Y_m(\beta_\rho^d \cdot a)] \\ \times [C_2 \cos(m\phi) + D_2 \sin(m\phi)] \cdot [C_2 \cos(\beta_z z) + D_2 \sin(\beta_z z)] = 0$$

$$A_1 J_m(\beta_\rho^d \cdot a) + B_1 Y_m(\beta_\rho^d \cdot a) = 0$$

(Cont'd)

$$[A_1 J_m'(\beta_\rho^d \cdot a) + B_1 Y_m'(\beta_\rho^d \cdot a)] = 0$$

$$[A_1 J_m'(\beta_\rho^d \cdot b) + B_1 Y_m'(\beta_\rho^d \cdot b)] = 0$$

$$Fg = \begin{bmatrix} J_m'(\beta_\rho^d \cdot a) & Y_m'(\beta_\rho^d \cdot a) \\ J_m'(\beta_\rho^d \cdot b) & Y_m'(\beta_\rho^d \cdot b) \end{bmatrix} \cdot \begin{bmatrix} A_1 \\ B_1 \end{bmatrix} = 0$$

The equation $Fg = 0$ has a nontrivial solution provided that the determinant of F is equal to zero. $\det(F) = 0$

$$\det(F) = J_m'(\beta_\rho^d \cdot a) \cdot Y_m'(\beta_\rho^d \cdot b) - J_m'(\beta_\rho^d \cdot b) \cdot Y_m'(\beta_\rho^d \cdot a) = 0$$

Therefore, the eigenvalues are obtained as solutions to:

$$J_m'(\beta_\rho a) \cdot Y_m'(\beta_\rho b) - Y_m'(\beta_\rho a) \cdot J_m'(\beta_\rho b) = 0, \quad m = 0, 1, 2, \dots$$

$$(\beta_\rho^d)^2 + \beta_z^2 = \beta^2$$

$$\beta_r = \omega_r \sqrt{\mu_d \epsilon_d} = 2\pi f_r \sqrt{\mu_d \epsilon_d} = \sqrt{(\beta_\rho^d)^2 + \left(\frac{p\pi}{h}\right)^2}$$

Expression for the resonant frequency for TM^z modes:

$$f_r = \frac{1}{2\pi\sqrt{\mu_d \epsilon_d}} \sqrt{(\beta_\rho^d)^2 + \left(\frac{p\pi}{h}\right)^2}$$

9.40

 TE^z

Since this structure is a combination of the structure for Problems 9.38 and 9.39, it must satisfy the boundary conditions of Problems 9.38 and 9.39. Therefore from the solutions of Problems 9.38 and 9.39, we can write that

$$J_m(\beta_\rho a) Y_m(\beta_\rho b) - Y_m(\beta_\rho a) J_m(\beta_\rho b) = 0$$

$$m = -p\left(\frac{\pi}{\phi_0}\right), \quad p = 1, 2, 3, \dots \dots$$

 TM^z

For the same reasons as for the TE^z modes, we can write from the solutions of Problems 9.38 and 9.39 that

$$J_m'(\beta_\rho a) Y_m'(\beta_\rho b) \cdot Y_m'(\beta_\rho a) J_m'(\beta_\rho b) = 0$$

$$m = -p\left(\frac{\pi}{\phi_0}\right), \quad p = 0, 1, 2, 3, \dots \dots$$

9.41

$$IE^z: F_z = [A_1 J_m(\beta_{pp}) + B_1 Y_m(\beta_{pp})] [C_2 \cos(\beta_2 z) + D_2 \sin(\beta_2 z)] [C_3 \cos(\beta_3 z) + D_3 \sin(\beta_3 z)]$$

$$H_p = -j \frac{1}{w\mu\varepsilon} \frac{\partial^2 F_z}{\partial p^2 z}$$

$$H_\phi = -j \frac{1}{w\mu\varepsilon} \frac{1}{\rho} \frac{\partial^2 F_z}{\partial \phi \partial z}$$

$$H_z = -j \frac{1}{w\mu\varepsilon} \left(\frac{\partial^2}{\partial z^2} + \beta^2 \right) F_z$$

$$H_z = -j \frac{1}{w\mu\varepsilon} \left(\beta_2^2 + \beta_3^2 \right) [A_1 J_m(\beta_{pp}) + B_1 Y_m(\beta_{pp})] [C_2 \cos(\beta_2 z) + D_2 \sin(\beta_2 z)] [C_3 \cos(\beta_3 z) + D_3 \sin(\beta_3 z)]$$

$$= -j \frac{\beta_p^2}{w\mu\varepsilon} [A_1 J_m(\beta_{pp}) + B_1 Y_m(\beta_{pp})] [C_2 \cos(\beta_2 z) + D_2 \sin(\beta_2 z)] [C_3 \cos(\beta_3 z) + D_3 \sin(\beta_3 z)]$$

$$\textcircled{1} H_z(\phi=0) = 0 = -j \frac{\beta_p^2}{w\mu\varepsilon} [C_2(1) + D_2(0)] [] = 0 \Rightarrow C_2 = 0$$

$$\textcircled{2} H_z(\phi=\phi_0) = 0 = -j \frac{\beta_p^2}{w\mu\varepsilon} [D_2 \sin(m\phi)] [] = 0 \Rightarrow \sin(m\phi_0) = 0$$

$$m\phi_0 = \sin^{-1}(0) = q\pi \Rightarrow m = \frac{q\pi}{\phi_0}, q = 1, 2, 3, \dots$$

$$H_p = -j \frac{\beta_p}{w\mu\varepsilon} [A_1] [D_2 \sin(m\phi)] [-C_3 \sin(\beta_3 z) + D_3 \cos(\beta_3 z)]$$

$$\textcircled{3} H_p(z=0) = -j \frac{\beta_p}{w\mu\varepsilon} [] [] [-C_3(0) + D_3(1)] = 0 \Rightarrow D_3 = 0$$

$$\textcircled{4} H_p(z=h) = -j \frac{\beta_p}{w\mu\varepsilon} [] [] [-C_3 \sin(\beta_3 h)] = 0 \Rightarrow \beta_3 h = \sin^{-1}(0) = p\pi$$

$$\textcircled{5} H_z(p=a) = -j \frac{\beta_p^2}{w\mu\varepsilon} B_2 C_3 [A_1 J_m(\beta_{pa}) + B_1 Y_m(\beta_{pa})] [\sin(m\phi) \sin(\beta_2 z)] = 0$$

$$\left. \begin{aligned} A_1 J_m(\beta_{pa}) + B_1 Y_m(\beta_{pa}) &= 0 \\ A_1 J_m(\beta_{pb}) + B_1 Y_m(\beta_{pb}) &= 0 \end{aligned} \right\} \left. \begin{aligned} J_m(\beta_{pa}) &= 0 \\ J_m(\beta_{pb}) &= 0 \end{aligned} \right\} \left. \begin{aligned} Y_m(\beta_{pa}) &= 0 \\ Y_m(\beta_{pb}) &= 0 \end{aligned} \right\} \left. \begin{aligned} A_1 &= 0 \\ B_1 &= 0 \end{aligned} \right\}$$

$$\textcircled{6} H_z(p=b) = 0 \Rightarrow A_1 J_m(\beta_{pb}) + B_1 Y_m(\beta_{pb}) = 0$$

$$\boxed{J_m(\beta_{pa}) Y_m(\beta_{pb}) - J_m(\beta_{pb}) Y_m(\beta_{pa}) = 0}$$

$$\boxed{\beta_p^2 + \beta_2^2 = \beta_r^2}$$

$$m = \frac{q\pi}{\phi_0}, q = 1, 2, \dots ; \beta_r = \frac{p\pi}{h}, p = 0, 1, 2, \dots$$

9.42

$$\underline{\text{TM}^2}: A_2 = [A_1 J_m(\beta_{pp}) + B_1 Y_m(\beta_{pp})] [C_2 \cos(m\phi) + D_2 \sin(m\phi)] [C_3 \cos(\beta_2 z) + D_3 \sin(\beta_2 z)]$$

$$H_\theta = \frac{1}{r} \frac{d}{dp} \frac{\partial A_2}{\partial p} = \frac{m}{rp} [A_1 J_m'(\beta_{pp}) + B_1 Y_m'(\beta_{pp})] [-C_2 \sin(m\phi) + D_2 \cos(m\phi)] [C_3 \cos(\beta_2 z) + D_3 \sin(\beta_2 z)]$$

$$H_\phi = -\frac{1}{r} \frac{\partial A_2}{\partial p} = -\frac{\beta_p}{r} [A_1 J_m'(\beta_{pp}) + B_1 Y_m'(\beta_{pp})] [C_2 \cos(m\phi) + D_2 \sin(m\phi)] [C_3 \cos(\beta_2 z) + D_3 \sin(\beta_2 z)]$$

$$H_\phi(\phi=0) = \frac{m}{rp} [\quad] [-C_2(0) + D_2(1)] [\quad] = 0 \Rightarrow \boxed{D_2 = 0}$$

$$H_\phi(\phi=\phi_0) = \frac{m}{rp} [\quad] [-C_2 \sin(m\phi_0)] [\quad] = 0 \Rightarrow \sin(m\phi_0) = 0$$

$$m\phi_0 = \sin^{-1}(0) = q\pi \Rightarrow \boxed{m = \frac{q\pi}{\phi_0}, \quad q = 0, 1, 2, \dots}$$

$$H_p(z=0) = \frac{m}{rp} [\quad] [\quad] [C_3(1) + D_3(0)] = 0 \Rightarrow \boxed{C_3 = 0}$$

$$H_p(z=h) = \frac{m}{rp} [\quad] [\quad] [D_3 \sin(\beta_2 h)] = 0 \Rightarrow \sin(\beta_2 h) = 0$$

$$\beta_2 h = \sin^{-1}(0) = p\pi \Rightarrow \boxed{\beta_2 = \left(\frac{p\pi}{h}\right), \quad p = 1, 2, 3, \dots}$$

$$H_\phi(p=a) = -\frac{\beta_p}{r} [A_1 J_m'(\beta_{pa}) + B_1 Y_m'(\beta_{pa})] [\quad] [\quad] = 0$$

$$\boxed{A_1 J_m'(\beta_{pa}) + B_1 Y_m'(\beta_{pa}) = 0}$$

$$H_\phi(p=b) = 0 \Rightarrow \boxed{A_1 J_m'(\beta_{pb}) + B_1 Y_m'(\beta_{pb}) = 0}$$

$$\begin{bmatrix} J_m'(\beta_{pa}) & Y_m'(\beta_{pa}) \\ J_m'(\beta_{pb}) & Y_m'(\beta_{pb}) \end{bmatrix} \begin{bmatrix} A_1 \\ B_1 \end{bmatrix} = 0 \Rightarrow \boxed{J_m'(\beta_{pa}) Y_m'(\beta_{pb}) - J_m'(\beta_{pb}) Y_m'(\beta_{pa}) = 0}$$

$$\boxed{\beta_p^2 + \beta_z^2 = \beta_r^2}$$

9.43

$$F_2^d = [A_0^d J_0(\beta_p^d \rho) + B_0^d Y_0(\beta_p^d \rho)] \quad , \quad (\beta_p^d)^2 + (\beta_z)^2 = (\beta_d)^2$$

$$F_2^o = [A_0^o K_0(\alpha_p^o \rho)] e^{-j\beta_z z} \quad , \quad -(\alpha_p^o)^2 + (\beta_z)^2 = (\beta_o)^2$$

$$E_p^d = -\frac{1}{\epsilon_0} \frac{1}{\rho} \frac{\partial F_2^d}{\partial \phi} = 0$$

$$E_\phi^d = \frac{1}{\epsilon_0} \frac{\partial F_2^d}{\partial \rho} = \frac{\beta_p^d}{\epsilon_0} [A_0^d J'_0(\beta_p^d \rho) + B_0^d Y'_0(\beta_p^d \rho)] e^{-j\beta_z z}$$

$$E_z^d = 0$$

$$H_p^d = -j \frac{1}{\omega \mu_0 \epsilon_0} \frac{\partial^2 F_2^d}{\partial \rho^2} = -\frac{\beta_p^d \beta_z}{\omega \mu_0 \epsilon_0} [A_0^d J'_0(\beta_p^d \rho) + B_0^d Y'_0(\beta_p^d \rho)] e^{-j\beta_z z}$$

$$H_\phi^d = -j \frac{1}{\omega \mu_0 \epsilon_0} \frac{1}{\rho} \frac{\partial^2 F_2^d}{\partial \phi^2} = 0$$

$$H_z^d = -j \frac{1}{\omega \mu_0 \epsilon_0} \left(\frac{\partial^2}{\partial z^2} + \beta_d^2 \right) F_2^d = -j \frac{(\beta_p^d)^2}{\omega \mu_0 \epsilon_0} [A_0^d J_0(\beta_p^d \rho) + B_0^d Y_0(\beta_p^d \rho)] e^{-j\beta_z z}$$

$$E_p^o = -\frac{1}{\epsilon_0} \frac{1}{\rho} \frac{\partial F_2^o}{\partial \phi} = 0$$

$$E_\phi^o = \frac{1}{\epsilon_0} \frac{\partial F_2^o}{\partial \rho} = \frac{\alpha_p^o}{\epsilon_0} A_0^o K'_0(\alpha_p^o \rho) e^{-j\beta_z z}$$

$$E_z^o = 0$$

$$H_p^o = -j \frac{1}{\omega \mu_0 \epsilon_0} \frac{\partial^2 F_2^o}{\partial \rho^2} = -\frac{\alpha_p^o \beta_z}{\omega \mu_0 \epsilon_0} A_0^o K'_0(\alpha_p^o \rho) e^{j\beta_z z}$$

$$H_\phi^o = -j \frac{1}{\omega \mu_0 \epsilon_0} \frac{1}{\rho} \frac{\partial^2 F_2^o}{\partial \phi^2} = 0$$

$$H_z^o = -j \frac{1}{\omega \mu_0 \epsilon_0} \left(\frac{\partial^2}{\partial z^2} + \beta_o^2 \right) F_2^o = j \frac{(\alpha_p^o)^2}{\omega \mu_0 \epsilon_0} A_0^o K_0(\alpha_p^o \rho) e^{-j\beta_z z}$$

The boundary conditions are the following:

$$E_\phi^d(\rho=a, 0 \leq \phi \leq 2\pi, z) = 0 \quad (1)$$

$$E_\phi^d(\rho=b, 0 \leq \phi \leq 2\pi, z) = E_\phi^o(\rho=b, 0 \leq \phi \leq 2\pi, z) \quad (2)$$

$$H_z^d(\rho=b, 0 \leq \phi \leq 2\pi, z) = H_z^o(\rho=b, 0 \leq \phi \leq 2\pi, z) \quad (3)$$

Applying boundary condition (1) leads to

$$\frac{\beta_p^d}{\epsilon_0} [A_0^d J'_0(\beta_p^d a) + B_0^d Y'_0(\beta_p^d a)] e^{-j\beta_z z} = 0 \Rightarrow B_0^d = -A_0^d \frac{J'_0(\beta_p^d a)}{Y'_0(\beta_p^d a)} \quad (4)$$

Applying boundary condition (2) leads to

$$\frac{\beta_p^d}{\epsilon_0} [A_0^d J'_0(\beta_p^d b) + B_0^d Y'_0(\beta_p^d b)] = \frac{\alpha_p^o}{\epsilon_0} A_0^o K'_0(\alpha_p^o b)$$

cont'd.

9.43 cont'd.

$$A_0^0 K_0'(\alpha_p^0 b) = \frac{\epsilon_0}{\epsilon_d} \frac{\beta_p^d}{\alpha_p^0} [A_0^d J_0'(\beta_p^d b) + B_0^d Y_0'(\beta_p^d b)] \quad (5)$$

which can be written using (4) as

$$A_0^0 K_0'(\alpha_p^0 b) = \frac{\epsilon_0}{\epsilon_d} \frac{\beta_p^d}{\alpha_p^0} A_0^d \left[J_0'(\beta_p^d b) - Y_0'(\beta_p^d b) \frac{J_0'(\beta_p^d a)}{Y_0'(\beta_p^d a)} \right] \quad (6)$$

Applying boundary condition (3) leads to

$$-j \frac{(\beta_p^d)^2}{\omega_p^2 \epsilon_d} [A_0^d J_0(\beta_p^d b) + B_0^d Y_0(\beta_p^d b)] e^{-j\beta_p^d z} = j \frac{(\alpha_p^0)^2}{\omega_p^2 \epsilon_0} A_0^0 K_0(\alpha_p^0 b) e^{-j\beta_p^d z}$$

$$A_0^0 K_0(\alpha_p^0 b) = - \frac{\mu_0 \epsilon_0}{\mu_d \epsilon_d} \left(\frac{\beta_p^d}{\alpha_p^0} \right)^2 [A_0^d J_0(\beta_p^d b) + B_0^d Y_0(\beta_p^d b)]$$

which can be written as

$$A_0^0 K_0(\alpha_p^0 b) = - A_0^d \frac{\mu_0 \epsilon_0}{\mu_d \epsilon_d} \left(\frac{\beta_p^d}{\alpha_p^0} \right)^2 \left[J_0(\beta_p^d b) - Y_0(\beta_p^d b) \frac{J_0'(\beta_p^d a)}{Y_0'(\beta_p^d a)} \right] \quad (7)$$

Dividing (6) by (7) leads to

$$\frac{K_0'(\alpha_p^0 b)}{K_0(\alpha_p^0 b)} = - \frac{\mu_d (\alpha_p^0)}{\mu_0 (\beta_p^d)} \frac{[J_0'(\beta_p^d b) Y_0'(\beta_p^d a) - J_0'(\beta_p^d a) Y_0'(\beta_p^d b)]}{[J_0(\beta_p^d b) Y_0'(\beta_p^d a) - J_0'(\beta_p^d a) Y_0(\beta_p^d b)]} \quad (8)$$

In addition

$$(\beta_p^d)^2 + \beta_z^2 = \beta_d^2$$

$$-(\alpha_p^0)^2 + \beta_z^2 = \beta_0^2$$

Subtracting these two leads to

$$(\beta_p^d)^2 + (\alpha_p^0)^2 = \beta_d^2 - \beta_0^2 = \beta_0^2 (\mu_r \epsilon_r - 1) = \beta_0^2 (\mu_r \epsilon_r - \frac{1}{\epsilon_0}) \quad (9)$$

Let us now examine the bracketed terms in the numerator and denominator of (8), using $J_0'(\beta_p^d a) = -J_1(\beta_p^d a)$, $Y_0'(\beta_p^d a) = -Y_1(\beta_p^d a)$, etc.

$$\text{NUM} = J_0'(\beta_p^d b) Y_0'(\beta_p^d a) - J_0'(\beta_p^d a) Y_0'(\beta_p^d b) \approx Y_1(\beta_p^d a) \left\{ J_1(\beta_p^d a) \left[1 - \frac{b-a}{a} \right] + \beta_p^d (b-a) J_0(\beta_p^d a) \right\}$$

$$- J_1(\beta_p^d a) \left\{ Y_1(\beta_p^d a) \left[1 - \frac{b-a}{a} \right] + \beta_p^d (b-a) Y_0(\beta_p^d a) \right\}$$

$$\text{NUM} \approx (\beta_p^d)(b-a) [J_1(\beta_p^d a) Y_0(\beta_p^d a) - Y_1(\beta_p^d a) J_0(\beta_p^d a)] \quad (10)$$

cont'd.

9.43 cont'd.

$$\text{DEN} = J_0(\beta_p^d b) Y'_0(\beta_p^d a) - J'_0(\beta_p^d a) Y_0(\beta_p^d b) \approx -Y_1(\beta_p^d a) [J_0(\beta_p^d a) - \beta_p^d(b-a) J_1(\beta_p^d a)] \\ + J_1(\beta_p^d a) [Y_0(\beta_p^d a) - \beta_p^d(b-a) Y_1(\beta_p^d a)]$$

$$\text{DEN} \approx [J_1(\beta_p^d a) Y_0(\beta_p^d a) - Y_1(\beta_p^d a) J_0(\beta_p^d a)] \quad (11)$$

Using (10) and (11) as well as (9-11a) and (9-11b), we can write (8) as

$$\frac{K'_0(\alpha_p^0 b)}{K_0(\alpha_p^0 b)} = -\frac{1/b\alpha_p^0}{-\ln(0.89\alpha_p^0 b)} = -\frac{k_d}{k_0} \frac{\alpha_p^0}{\beta_p^d} [-\beta_p^d(b-a)]$$

$$\frac{1}{b\alpha_p^0 \ln(0.89\alpha_p^0 b)} = k_r \alpha_p^0 (b-a)$$

$$1 = k_r b (\alpha_p^0)^2 (b-a) \ln(0.89\alpha_p^0 b)$$

CHAPTER 10

10.1 $A_z^t = A_z^{(1)}(x - \frac{z}{2}, y, z) - A_z^{(2)}(x + \frac{z}{2}, y, z) \xrightarrow{z \rightarrow 0} -s \frac{\partial A_z^{(0)}}{\partial x}$

$$A_z^t = j \frac{\sin \beta I e^{\Delta l}}{4\pi} \frac{\partial}{\partial x} h_0^{(2)}(br) = j \frac{\sin \beta I e^{\Delta l}}{4\pi} \frac{\partial r}{\partial x} \frac{\partial}{\partial r} h_0^{(2)}(br) = j \frac{\sin \beta^2 I e^{\Delta l}}{4\pi} \frac{\partial r}{\partial x} h_0^{(2)}(br)$$

$$r = \sqrt{x^2 + y^2 + z^2}, \quad x = r \sin \theta \cos \phi$$

$$\frac{\partial r}{\partial x} = \frac{1}{r} (x^2 + y^2 + z^2)^{-1/2} (2x) = \frac{x}{\sqrt{x^2 + y^2 + z^2}} = \frac{r \sin \theta \cos \phi}{r} = \sin \theta \cos \phi$$

$$A_z^t \approx j \frac{\sin \beta^2 I e^{\Delta l}}{4\pi} h_0^{(2)}(br) \sin \theta \cos \phi$$

Since according to (II-19a)

$$P_1^1(x) = -(1-x^2)^{1/2}$$

$$P_1^1(\cos \theta) = -(1-\cos^2 \theta)^{1/2} = -\sin \theta$$

then

$$A_z^t \approx -j \frac{\sin \beta^2 I e^{\Delta l}}{4\pi} h_0^{(2)}(br) P_1^1(\cos \theta) \cos \phi = +j \frac{\sin \beta^2 I e^{\Delta l}}{4\pi} h_0^{(2)}(br) P_1^1(\cos \theta) \cos \phi$$

10.2 $A_z^t = A_z^{(1)}(x, y - \frac{z}{2}, z) - A_z^{(2)}(x, y + \frac{z}{2}, z) \xrightarrow{z \rightarrow 0} -s \frac{\partial A_z^{(0)}}{\partial y}$

$$A_z^t \approx +j \frac{\sin \beta I e^{\Delta l}}{4\pi} \frac{\partial}{\partial y} h_0^{(2)}(br) = j \frac{\sin \beta I e^{\Delta l}}{4\pi} \frac{\partial r}{\partial y} \frac{\partial}{\partial r} h_0^{(2)}(br) = j \frac{\sin \beta^2 I e^{\Delta l}}{4\pi} \frac{\partial r}{\partial y} h_0^{(2)}(br)$$

$$r = \sqrt{x^2 + y^2 + z^2}, \quad y = r \sin \theta \sin \phi$$

$$\frac{\partial r}{\partial y} = \frac{1}{r} (x^2 + y^2 + z^2)^{-1/2} (2y) = \frac{y}{\sqrt{x^2 + y^2 + z^2}} = \frac{r \sin \theta \sin \phi}{r} = \sin \theta \sin \phi$$

$$A_z^t \approx j \frac{\sin \beta^2 I e^{\Delta l}}{4\pi} h_0^{(2)}(br) \sin \theta \sin \phi$$

then Since from solution of Problem 10.1 $P_1^1(\cos \theta) = -\sin \theta$

$$A_z^t \approx -j \frac{\sin \beta^2 I e^{\Delta l}}{4\pi} h_0^{(2)}(br) P_1^1(\cos \theta) \sin \phi = +j \frac{\sin \beta^2 I e^{\Delta l}}{4\pi} h_0^{(2)}(br) P_1^1(\cos \theta) \sin \phi$$

10.3 $A_z^t = A_z^{(1)}(x, y, z - \frac{z}{2}) - A_z^{(2)}(x, y, z + \frac{z}{2}) \xrightarrow{z \rightarrow 0} -s \frac{\partial A_z^{(0)}}{\partial z}$

$$A_z^t \approx j \frac{\sin \beta I e^{\Delta l}}{4\pi} \frac{\partial}{\partial z} h_0^{(2)}(br) = j \frac{\sin \beta I e^{\Delta l}}{4\pi} \frac{\partial r}{\partial z} \frac{\partial}{\partial r} h_0^{(2)}(br) = j \frac{\sin \beta^2 I e^{\Delta l}}{4\pi} \frac{\partial r}{\partial z} h_0^{(2)}(br)$$

$$r = \sqrt{x^2 + y^2 + z^2}, \quad z = r \cos \theta$$

$$\frac{\partial r}{\partial z} = \frac{1}{r} (x^2 + y^2 + z^2)^{-1/2} (2z) = \frac{z}{\sqrt{x^2 + y^2 + z^2}} = \frac{r \cos \theta}{r} = \cos \theta$$

Since according to (II-19a)

$$P_1^0(x) = P_1^1(x) = x \Rightarrow P_1(\cos \theta) = \cos \theta$$

then

$$A_z^t \approx j \frac{\sin \beta^2 I e^{\Delta l}}{4\pi} h_0^{(2)}(br) \cos \theta = j \frac{\sin \beta^2 I e^{\Delta l}}{4\pi} h_0^{(2)}(br) P_1(\cos \theta)$$

$$10.4 \quad \underline{E} = -\frac{1}{\epsilon} \nabla \times \underline{F}, \quad \underline{H} = -j\omega \underline{F} - j \frac{1}{\omega \mu \epsilon} \nabla (\nabla \cdot \underline{E})$$

For TE²: $\underline{F} = \hat{a}_z F_z = (\hat{a}_r \cos\theta - \hat{a}_\phi \sin\theta) F_z(r, \theta, \phi)$ using (II-15b)

$$F_r = \cos\theta F_z(r, \theta, \phi)$$

$$F_\phi = -\sin\theta F_z(r, \theta, \phi)$$

$$\nabla \cdot \underline{F} = \frac{1}{r^2} \frac{\partial}{\partial r} (r^2 F_r) + \frac{1}{r \sin\theta} \frac{\partial}{\partial \theta} [\sin\theta F_\phi] + \frac{1}{r \sin\theta} \frac{\partial F_z}{\partial \phi}$$

$$= \frac{1}{r^2} \frac{\partial}{\partial r} [r^2 \cos\theta F_z] + \frac{1}{r \sin\theta} \frac{\partial}{\partial \theta} [-\sin^2\theta F_z]$$

$$\nabla \cdot \underline{F} = \frac{\cos\theta}{r^2} \frac{\partial}{\partial r} (r^2 F_z) - \frac{1}{r \sin\theta} \frac{\partial}{\partial \theta} (\sin^2\theta F_z) = DF$$

$$\nabla(\nabla \cdot \underline{F}) = \hat{a}_r \frac{1}{r^2} \frac{\partial}{\partial r} (DF) + \hat{a}_\theta \frac{1}{r} \frac{\partial}{\partial \theta} (DF) + \hat{a}_\phi \frac{1}{r \sin\theta} \frac{\partial}{\partial \phi} (DF)$$

$$\nabla \times \underline{F} = \hat{a}_r \frac{1}{r \sin\theta} \left[\frac{\partial}{\partial \theta} \left(\overset{\circ}{F}_\phi \sin\theta \right) - \frac{\partial F_z}{\partial \phi} \right] + \hat{a}_\theta \left[\frac{1}{\sin\theta} \frac{\partial F_r}{\partial \phi} - \frac{\partial}{\partial r} \left(\overset{\circ}{F}_\phi \right) \right] + \hat{a}_\phi \left[\frac{2}{r \sin^2\theta} \left(\overset{\circ}{F}_\phi \right) \right]$$

$$\nabla \times \underline{F} = \hat{a}_r \frac{1}{r \sin\theta} \left[\frac{\partial}{\partial \phi} (\sin\theta F_z) \right] + \hat{a}_\theta \frac{1}{r} \left[\frac{1}{\sin\theta} \frac{\partial}{\partial \phi} (\cos\theta F_z) \right]$$

$$+ \hat{a}_\phi \frac{1}{r} \left\{ \frac{2}{\partial r} (-r \sin\theta F_z) - \frac{2}{\partial \theta} (\cos\theta F_z) \right\}$$

Expanding all these operations, it can be shown after many lengthy manipulations that

$$E_r = -\frac{1}{\epsilon r} \frac{\partial F_z}{\partial \phi}$$

$$E_\theta = -\frac{\cos\theta}{\epsilon} \frac{\partial F_z}{\partial \phi}$$

$$E_\phi = +\frac{1}{\epsilon r} \left[\sin\theta \frac{\partial}{\partial r} (r F_z) + \frac{2}{\partial \theta} (F_z \cos\theta) \right]$$

$$H_r = -j \omega F_z \cos\theta - j \frac{1}{\omega \mu \epsilon} \frac{\partial}{\partial r} \left[\cos\theta \frac{2}{\partial r} (r^2 F_z) - \frac{1}{r \sin\theta} \frac{\partial}{\partial \theta} (F_z \sin^2\theta) \right]$$

$$H_\theta = j \omega F_z \sin\theta + j \frac{1}{\omega \mu \epsilon r} \frac{\partial}{\partial \theta} \left[\cos\theta \frac{2}{\partial r} (r^2 F_z) - \frac{1}{r \sin\theta} \frac{\partial}{\partial \theta} (F_z \sin^2\theta) \right]$$

$$H_\phi = -j \frac{1}{\omega \mu \epsilon r \sin\theta} \frac{\partial}{\partial \phi} \left[\cos\theta \frac{2}{\partial r} (r^2 F_z) - \frac{1}{r \sin\theta} \frac{\partial}{\partial \theta} (F_z \sin^2\theta) \right]$$

with $\nabla^2 F_z + \beta^2 F_z = 0$

$$[10.5] \quad \underline{H} = \frac{1}{\mu} \nabla \times \underline{A}, \quad \underline{E} = -j\omega \underline{A} - j \frac{1}{\omega \mu \epsilon} \nabla (\nabla \cdot \underline{A})$$

For TM²: $\underline{A} = \hat{a}_r \underline{R}_2 = (\hat{a}_r \cos\theta - \hat{a}_\phi \sin\theta) \underline{R}_2(r, \theta, \phi)$ using (II-136)

$$A_r = \cos\theta A_2(r, \theta, \phi)$$

$$A_\phi = -\sin\theta A_2(r, \theta, \phi)$$

Using the duality theorem of Section 7.2, Chapter 7, we can write utilizing the solution of Problem 10.4 that

$$\underline{E}_r = -j\omega A_2 \cos\theta - j \frac{1}{\omega \mu \epsilon} \frac{\partial}{\partial r} \left[\frac{\cos\theta}{r^2} \frac{\partial}{\partial r} (r^2 A_2) - \frac{1}{r \sin\theta} \frac{\partial}{\partial \theta} (A_2 \sin^2\theta) \right]$$

$$\underline{E}_\theta = +j\omega A_2 \sin\theta - j \frac{1}{\omega \mu \epsilon r} \frac{\partial}{\partial \theta} \left[\frac{\cos\theta}{r^2} \frac{\partial}{\partial r} (r^2 A_2) - \frac{1}{r \sin\theta} \frac{\partial}{\partial \theta} (A_2 \sin^2\theta) \right]$$

$$\underline{E}_\phi = -j \frac{1}{\omega \mu \epsilon r \sin\theta} \frac{\partial}{\partial \phi} \left[\frac{\cos\theta}{r^2} \frac{\partial}{\partial r} (r^2 A_2) - \frac{1}{r \sin\theta} \frac{\partial}{\partial \theta} (A_2 \sin^2\theta) \right]$$

$$H_r = \frac{1}{\mu r} \frac{\partial A_2}{\partial \theta}$$

$$H_\theta = \frac{\cos\theta}{\mu r} \frac{\partial A_2}{\partial \phi}$$

$$H_\phi = -\frac{1}{\mu r} \left[\sin\theta \frac{\partial}{\partial r} (r A_2) + \frac{2}{\partial \theta} (A_2 \cos\theta) \right]$$

$$\text{with } \nabla^2 \underline{R}_2 + \beta^2 \underline{A}_2 = 0$$

[10.6] According to (6-4a)

$$\underline{H}_A = \frac{1}{\mu} \nabla \times \underline{A} \quad (1)$$

Away from the source \underline{J}

$$\nabla \times \underline{H}_A = j\omega \epsilon \underline{E}_A \Rightarrow \underline{E}_A = \frac{1}{j\omega \mu \epsilon} \nabla \times \underline{H}_A = \frac{1}{j\omega \mu \epsilon} \nabla \times \nabla \times \underline{A} \quad (2)$$

Taking the curl of both sides of (1) leads to

$$\nabla \times \underline{H}_A = \frac{1}{\mu} \nabla \times \nabla \times \underline{A} \quad (3)$$

Using Maxwell's equation of

$$\nabla \times \underline{H}_A = \underline{J} + j\omega \epsilon \underline{E}_A \quad (4)$$

and equating it to (3) leads to

$$\nabla \times \nabla \times \underline{A} = \mu \underline{J} + j\omega \epsilon \mu \underline{E}_A \quad (5)$$

The electric and magnetic fields are also related by Maxwell's equation

$$\nabla \times \underline{E}_A = -j\omega \mu \underline{H}_A \quad (6)$$

cont'd.

[10.6 cont'd.] Substituting (1) into (6) and regrouping leads to

$$\nabla \times \underline{\underline{E}}_A = -j\omega \mu \underline{\underline{H}}_A = -j\omega \mu \left(\frac{1}{\mu} \nabla \times \underline{\underline{A}} \right) = -j\omega \nabla \times \underline{\underline{A}} \quad (7)$$

or

$$\nabla \times (\underline{\underline{E}}_A + j\omega \underline{\underline{A}}) = 0 \quad (7a)$$

Using the vector identity of

$$\nabla \times (-\nabla \psi_e) = 0 \quad (8)$$

and equating (7a) and (8) leads to

$$\underline{\underline{E}}_A + j\omega \underline{\underline{A}} = -\nabla \psi_e \Rightarrow \underline{\underline{E}}_A = -j\omega \underline{\underline{A}} - \nabla \psi_e \quad (9)$$

Substituting (9) into (5) leads to

$$\nabla \times \nabla \times \underline{\underline{A}} = \mu \underline{\underline{J}} + j\omega \epsilon \mu (-j\omega \underline{\underline{A}} - \nabla \psi_e) = \mu \underline{\underline{J}} + \omega^2 \mu \epsilon \underline{\underline{A}} - j\omega \mu \epsilon \nabla \psi_e \quad (10)$$

or

$$\nabla \times \nabla \times \underline{\underline{A}} - \omega^2 \mu \epsilon \underline{\underline{A}} = \mu \underline{\underline{J}} - j\omega \mu \epsilon \nabla \psi_e \quad (10a)$$

For source-free region ($\underline{\underline{J}} = 0$) (10a) reduces to

$$\nabla \times \nabla \times \underline{\underline{A}} - \omega^2 \mu \epsilon \underline{\underline{A}} = -j\omega \mu \epsilon \nabla \psi_e \quad (11)$$

[10.7] The required solution can be obtained by following the procedure of (10-17a)–(10-22a) which is used to derive (10-22b). Another alternative procedure is to use the duality theorem of Section 7.2, and Tables 7-1 and 7-2. The results follow easily.

[10.8] Using (10-43) we can write (10-27) as

$$\underline{\underline{E}}_\phi = \frac{1}{j\omega \mu \epsilon} \frac{1}{r \sin \theta} \frac{\partial^2 A_r}{\partial r \partial \phi} = \frac{B_m}{j\omega \mu \epsilon} \frac{1}{r \sin \theta} D_1 H_n^{(2)'}(kr) [A_2 P_n^m(\cos \alpha_1) + B_2 P_n^m(-\cos \alpha_1)] \\ \times [C_3 \sin(m\phi) + D_3 \cos(m\phi)]$$

Applying (10-44b) we have that

$$\underline{\underline{E}}_\phi(\theta = \alpha_1) = \frac{B_m}{j\omega \mu \epsilon r \sin \alpha_1} D_1 H_n^{(2)'}(kr) [A_2 P_n^m(\cos \alpha_1) + B_2 P_n^m(-\cos \alpha_1)] [\quad] = 0$$

$$A_2 P_n^m(\cos \alpha_1) + B_2 P_n^m(-\cos \alpha_1) = 0$$

$$\underline{\underline{E}}_\phi(\theta = \alpha_2) = \frac{B_m}{j\omega \mu \epsilon r \sin \alpha_2} D_1 H_n^{(2)'}(kr) [A_2 P_n^m(\cos \alpha_2) + B_2 P_n^m(-\cos \alpha_2)] [\quad] = 0$$

$$\text{or } \begin{bmatrix} P_n^m(\cos \alpha_1) & A_2 P_n^m(\cos \alpha_2) + B_2 P_n^m(-\cos \alpha_2) \\ P_n^m(-\cos \alpha_1) & P_n^m(-\cos \alpha_2) \end{bmatrix} \begin{bmatrix} A_2 \\ B_2 \end{bmatrix} = 0 \Rightarrow P_n^m(\cos \alpha_1) P_n^m(-\cos \alpha_2) - P_n^m(-\cos \alpha_1) P_n^m(\cos \alpha_2) = 0$$

$$\boxed{10.9} \quad \nabla \times \underline{E} = -j\omega \mu H \quad (1)$$

$$\nabla \times \underline{H} = j\omega E \underline{E} \quad (2)$$

Expanding (1) in spherical coordinates and assuming that the E -field has only an E_ϕ component independent of ϕ , reduces to

$$\nabla \times \underline{E} = \hat{a}_\phi \frac{1}{r} \frac{\partial}{\partial r} (r E_\phi) = -j\omega \mu (\hat{a}_r H_r + \hat{a}_\theta H_\theta + \hat{a}_\phi H_\phi) \quad (3)$$

Since \underline{H} has only an H_ϕ component, necessary to form the TEM mode with E_ϕ , (3) can be written as

$$\frac{1}{r} \frac{\partial}{\partial r} (r E_\phi) = -j\omega \mu H_\phi \quad (4)$$

From Ampere's law of (2), we have when expanded in spherical coordinates, and assuming only E_ϕ and H_ϕ components independent of ϕ , that

$$\hat{a}_r \frac{1}{r^2 \sin \theta} \left[\frac{\partial}{\partial \theta} (r \sin \theta H_\phi) \right] - \hat{a}_\theta \frac{1}{r \sin \theta} \left[\frac{\partial}{\partial r} (r \sin \theta H_\phi) \right] = j\omega \epsilon (Z_0 E_\phi) \quad (5)$$

which can also be written as

$$\frac{\partial}{\partial \theta} (r \sin \theta H_\phi) = 0 \quad (6a)$$

$$\frac{1}{r \sin \theta} \frac{\partial}{\partial r} (r \sin \theta H_\phi) = -j\omega \epsilon E_\phi \quad (6b)$$

Rewriting (6b) as

$$\frac{1}{r} \frac{\partial}{\partial r} (r H_\phi) = -j\omega \epsilon E_\phi \quad (7)$$

and substituting it into (4) we form a differential equation for H_ϕ as

$$-\frac{1}{j\omega \epsilon r} \frac{\partial}{\partial r} \left[\frac{1}{r} \frac{\partial}{\partial r} (r H_\phi) \right] = -j\omega \mu H_\phi \quad (8)$$

or

$$\frac{\partial^2}{\partial r^2} (r H_\phi) = -\omega^2 \mu \epsilon (r H_\phi) = -\beta^2 (r H_\phi) \quad (8a)$$

A solution for (8a) must be obtained to satisfy (6a). To meet the condition of (6a), the θ variations of H_ϕ must be of the form

$$H_\phi = \frac{f(r)}{\sin \theta} \quad (9)$$

A solution of (8a), which also meets the requirements of (9) and represents an outward traveling wave, is

$$H_\phi = \frac{H_0}{\sin \theta} e^{-j\beta r} \quad \text{where } f(r) = H_0 \frac{e^{-j\beta r}}{r} \quad (10)$$

An inward traveling wave is also a solution but does not apply to the infinitely long structure.

Since the field is of TEM mode, the electric field is related to the magnetic field by the intrinsic impedance, and we can write it as

$$E_\phi = \eta H_\phi = \eta \frac{H_0}{\sin \theta} \frac{e^{-j\beta r}}{r} \quad (11)$$

[10.10] $E_\phi = \eta \frac{H_0}{\sin \theta} \frac{e^{-j\beta r}}{r}$, $H_\phi = \frac{H_0}{\sin \theta} \frac{e^{-j\beta r}}{r}$ from solution of Problem 10.9.

$$\underline{S}_{ave} = \frac{1}{2} \operatorname{Re} [\underline{E} \times \underline{H}^*] = \frac{1}{2} \operatorname{Re} [E_\phi H_\phi \times \hat{a}_\phi H_\phi^*] = \frac{\hat{a}_r}{2} \operatorname{Re} [E_\phi H_\phi^*] = \hat{a}_r \frac{|H_0|^2 \eta}{2r^2 \sin^2 \theta}$$

$$P_{rad} = \oint \underline{S}_{ave} \cdot d\underline{s} = \int_{\alpha/2}^{\pi - \alpha/2} \int_{0}^{2\pi} \hat{a}_r S_r \cdot \hat{a}_r r^2 \sin \theta d\theta d\phi = |H_0|^2 \pi \eta \int_{\alpha/2}^{\pi - \alpha/2} \frac{d\theta}{\sin \theta} = 2\pi \eta |H_0|^2 \ln [\cot(\frac{\alpha}{2})]$$

[10.11] From solution of Problem 10.9

$$H_\phi = \frac{H_0}{\sin \theta} \frac{e^{-j\beta r}}{r}$$

The current on the surface of the cones, a distance r from the origin, is

$$I(r) = \int_0^{2\pi} H_\phi r \sin \theta d\phi = H_0 e^{-j\beta r} \int_0^{2\pi} d\phi = 2\pi H_0 e^{-j\beta r}$$

At $r=0$

$$I(r=0) = 2\pi H_0$$

Using $I(r=0)$ from above and the P_{rad} from the solution of Problem 10.10, we can write the radiation resistance as

$$R_Y = \frac{2 P_{rad}}{|I(r=0)|^2} = \frac{4\pi \eta |H_0|^2 \ln [\cot(\frac{\alpha}{2})]}{(2\pi)^2 |H_0|^2} = \frac{\eta}{\pi} \ln [\cot(\frac{\alpha}{2})]$$

[10.12]

$$(a) R_Y = 300 = \frac{120\pi}{\pi} \ln [\cot(\frac{\alpha}{2})] = 120 \ln [\cot(\frac{\alpha}{2})] \Rightarrow \ln [\cot(\frac{\alpha}{2})] = \frac{300}{120} = 2.5$$

$$\cot(\frac{\alpha}{2}) = \ln^{-1}(2.5) = e^{2.5} = 12.18 \Rightarrow \frac{\alpha}{2} = \cot^{-1}(12.18) = 4.69^\circ$$

$$\alpha = 9.38^\circ \Rightarrow 2\alpha = 18.76^\circ \text{ total cone angle}$$

$$(b) R_Y = 50 = 120 \ln [\cot(\frac{\alpha}{2})] \Rightarrow \ln [\cot(\frac{\alpha}{2})] = \frac{50}{120} = 0.4166 \Rightarrow \cot(\frac{\alpha}{2}) = \ln^{-1}(0.4166)$$

$$\cot(\frac{\alpha}{2}) = 1.5169 \Rightarrow \frac{\alpha}{2} = \cot^{-1}(1.5169) = 33.39^\circ \Rightarrow \alpha = 66.789^\circ$$

$$2\alpha = 133.58^\circ \text{ total cone angle}$$

[10.13]

a. For $\frac{IEr}{r^2}$ $\Rightarrow (F_r)_{mnp} = [A_2 P_n^m(\cos \theta) + B_2 Q_n^m(\cos \theta)] [C_2 \hat{H}_n^{(c)}(kr) + D_2 \hat{H}_n^{(a)}(kr)] [C_3 \cos(m\phi) + D_3 \sin(m\phi)]$
Since the fields must be finite at $\theta = 0^\circ$ where $Q_n^m(\cos \theta)$ has a singularity, then $B_2 = 0$ and $(F_r)_{mnp}$ reduces to

$$(F_r)_{mnp} = A_2 P_n^m(\cos \theta) [C_2 \hat{H}_n^{(c)}(kr) + D_2 \hat{H}_n^{(a)}(kr)] [C_3 \cos(m\phi) + D_3 \sin(m\phi)]$$

b. The boundary condition is

$$E_\phi(0 \leq r \leq \infty, \theta = \alpha, 0 \leq \phi \leq \pi)$$

$$\text{Using (10-23c)} \Rightarrow E_\phi = \frac{1}{EY} \frac{1}{r} \frac{\partial F_r}{\partial \theta} = \frac{A_2}{EY} \frac{\partial P_n^m(\cos \theta)}{\partial \theta} [C_2 \hat{H}_n^{(c)}(kr) + D_2 \hat{H}_n^{(a)}(kr)] [C_3 \cos(m\phi) + D_3 \sin(m\phi)]$$

$$\text{Thus } E_\phi(\theta = \alpha) = \frac{A_2}{EY} \frac{\partial P_n^m(\cos \theta)}{\partial \theta} \Big|_{\theta=\alpha} = 0 \Rightarrow \frac{\partial P_n^m(\cos \theta)}{\partial \theta} \Big|_{\theta=\alpha} = 0$$

10.14

TM^r

- a. For TM^r $\Rightarrow (Ar)_{mnp} = [A_2 P_n^m(\cos\theta) + B_2 Q_n^m(\cos\theta)] [C_1 \hat{H}_n^{(1)}(ar) + D_1 \hat{H}_n^{(2)}(ar)] [C_3 \cos(m\phi) + D_3 \sin(m\phi)]$
Since the fields must be finite at $\theta=0^\circ$ where $Q_n^m(\cos\theta)$ has a singularity, then
 $B_2 = 0$ and $(Ar)_{mnp}$ reduces to

$$(Ar)_{mnp} = A_2 P_n^m(\cos\theta) [C_1 \hat{H}_n^{(1)}(ar) + D_1 \hat{H}_n^{(2)}(ar)] [C_3 \cos(m\phi) + D_3 \sin(m\phi)]$$

- b. The boundary condition is

$$E_\phi (0 \leq r \leq \infty, \theta = \alpha, 0 \leq \phi \leq 2\pi) = 0 \quad \text{or} \quad E_r (0 \leq r \leq \infty, \theta = \alpha, 0 \leq \phi \leq 2\pi) = 0$$

Using (10-27c) we can write that

$$\nabla_\phi = \frac{1}{j\omega \epsilon r \sin\theta} \frac{1}{r} \frac{\partial}{\partial \phi} Ar = \frac{A_2 m\phi}{j\omega \epsilon r \sin\theta} P_n^m(\cos\theta) [C_1 \hat{H}_n^{(1)}(ar) + D_1 \hat{H}_n^{(2)}(ar)] [C_3 \cos(m\phi) + D_3 \sin(m\phi)]$$

thus

$$E_\phi (\theta = \alpha) = \frac{A_2 m\phi}{j\omega \epsilon r \sin\theta} P_n^m(\cos\theta) [\quad] [\quad] \Big|_{\theta=\alpha} = 0 \Rightarrow P_n^m(\cos\theta) \Big|_{\theta=\alpha} = 0$$

10.15

TE^r

$$(Fr)_{mnp} = [A_2 P_n^m(\cos\theta) + B_2 Q_n^m(\cos\theta)] [C_1 \hat{H}_n^{(1)}(ar) + D_1 \hat{H}_n^{(2)}(ar)] [C_3 \cos(m\phi) + D_3 \sin(m\phi)]$$

$$n = 1, 2, 3, \dots$$

- Since the fields must be finite at $\theta=0$ and $\theta=\pi$ where $Q_n^m(\cos\theta)$ has a singularity, then $B_2 = 0$ and $(Fr)_{mnp}$ reduces to

$$(Fr)_{mnp} = A_2 P_n^m(\cos\theta) [C_1 \hat{H}_n^{(1)} + D_1 \hat{H}_n^{(2)}(ar)] [C_3 \cos(m\phi) + D_3 \sin(m\phi)]$$

- The boundary condition is

$$E_\phi (0 \leq r \leq \infty, 0 \leq \theta \leq \pi, \phi = 0) = E_\phi (0 \leq r \leq \infty, 0 \leq \theta \leq \pi, \phi = \alpha) = 0$$

Using (10-23b) we can write

$$E_\phi = -\frac{1}{Er} \frac{1}{\sin\theta} \frac{\partial Fr}{\partial \phi} = -\frac{m A_2}{Er \sin\theta} P_n^m(\cos\theta) [C_1 \hat{H}_n^{(1)}(ar) + D_1 \hat{H}_n^{(2)}(ar)] [-C_3 \sin(m\phi) + D_3 \cos(m\phi)]$$

$$E_\phi (\phi = 0) = -\frac{m A_2}{Er \sin\theta} P_n^m(\cos\theta) [\quad] [-C_3(0) + D_3(1)] = 0 \Rightarrow D_3 = 0$$

$$E_\phi (\phi = \alpha) = -\frac{m A_2}{Er \sin\theta} P_n^m(\cos\theta) [\quad] [-C_3 \sin(m\alpha)] = 0 \Rightarrow \sin(m\alpha) = 0$$

$$m\alpha = \sin^{-1}(0) = -p\pi \Rightarrow m = \frac{-p\pi}{\alpha}, \quad p = 0, 1, 2, \dots$$

thus $(Fr)_{mnp} = A_{mnp} P_n^m(\cos\theta) [C_1 \hat{H}_n^{(1)}(ar) + D_1 \hat{H}_n^{(2)}(ar)] \cos(m\phi)$

$$m = \frac{-p\pi}{\alpha}, \quad p = 0, 1, 2, \dots$$

$$n = 1, 2, 3, \dots$$

10.16

TM^r

$$(A_r)_{mnp} = [A_2 P_n^m(\cos\theta) + B_2 Q_n^m(\cos\theta)] [C_1 \hat{H}_n^{(1)}(pr) + D_1 \hat{H}_n^{(2)}(pr)] [C_3 \cos(m\phi) + D_3 \sin(m\phi)]$$

$n = 1, 2, 3, \dots$

Since the fields must be finite at $\theta=0$ and $\theta=\pi$ where $Q_n^m(\cos\theta)$ has singularities, then $B_2=0$ and $(A_r)_{mnp}$ reduces to

$$(A_r)_{mnp} = A_2 P_n^m(\cos\theta) [C_1 \hat{H}_n^{(1)}(pr) + D_1 \hat{H}_n^{(2)}(pr)] [C_3 \cos(m\phi) + D_3 \sin(m\phi)]$$

The boundary condition is

$$E_\theta(0 \leq r \leq a, 0 \leq \theta \leq \pi, \phi=0) = E_\theta(0 \leq r \leq a, 0 \leq \theta \leq \pi, \phi=\pi) = 0$$

Using (10-27b) we can write

$$E_\theta = \frac{1}{j\omega \mu \epsilon r} \frac{\partial^2 A_r}{\partial r \partial \theta} = \frac{\beta A_2}{j\omega \mu \epsilon r} \frac{\partial P_n^m(\cos\theta)}{\partial \theta} [C_1 \hat{H}_n^{(1)}(pr) + D_1 \hat{H}_n^{(2)}(pr)] [C_3 \cos(m\phi) + D_3 \sin(m\phi)]$$

$$E_\theta(\phi=0) = \frac{\beta A_2}{j\omega \mu \epsilon r} \frac{\partial P_n^m(\cos\theta)}{\partial \theta} [\quad] [C_3(1) + D_3(0)] = 0 \Rightarrow C_3 = 0$$

$$E_\theta(\phi=\pi) = \frac{\beta A_2}{j\omega \mu \epsilon r} \frac{\partial P_n^m(\cos\theta)}{\partial \theta} [\quad] [D_3 \sin(m\phi)] = 0 \Rightarrow \sin(m\alpha) = 0 \Rightarrow m\alpha = \sin^{-1}(0) = -\pi$$

Thus $m = \frac{-2\pi}{\alpha}, p = 1, 2, 3, \dots$

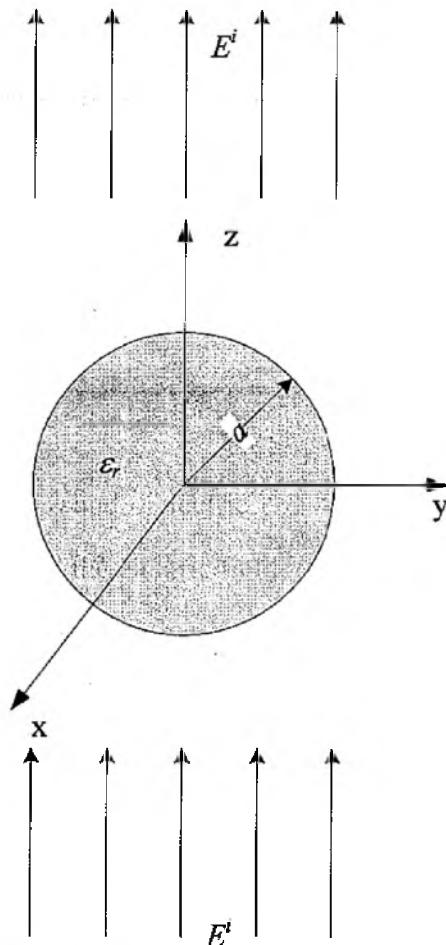
$$(A_r)_{mnp} = A_{mnp} P_n^m(\cos\theta) [C_1 \hat{H}_n^{(1)}(pr) + D_1 \hat{H}_n^{(2)}(pr)] \sin(m\phi)$$

$$m = \frac{-2\pi}{\alpha}, p = 1, 2, 3, \dots$$

$$n = 1, 2, 3, \dots$$

10.17

A dielectric sphere of radius a and dielectric constant ϵ_r is placed in a uniform static electric field $\vec{E}^i = \hat{a}_z E_0$. Find the electric field \vec{E} inside and outside the sphere.



The electric field \vec{E} can be found by using (6-9a) with $\omega = 0$.

$$\vec{E} = -\nabla \psi \quad (1)$$

where ϕ_e in (6-9a) is replaced by ψ .

To find ψ , we need to solve (3-75) with $\beta = 0$ which reduces to the Laplace equation

$$\nabla^2 \psi = 0 \quad (2)$$

For our problem, (2) needs to be solved twice: first for the potential inside the sphere $\psi_{in}(r, \theta)$, and then for the potential outside the sphere $\psi_{out}(r, \theta)$. These two solutions are connected by the boundary conditions (1-26a) and (1-30a) on the surface of the sphere.

Eq. (3-75) in spherical coordinates is given by (3.77). Assuming the solution of the form (3-76) and applying the separation of variables method, partial differential equation (3-77) can be separated into three ordinary differential equations (3-86).

Since our problem is fully symmetric (invariant) with respect to ϕ ,

$$\frac{\partial \psi}{\partial \phi} = fg \frac{\partial h}{\partial \phi} = \frac{\partial h}{\partial \phi} = 0 \quad (3)$$

Therefore, (3-86c) can only be satisfied if $m = 0$. From now on we can ignore (3-86c) and function $h(\phi)$ in (3-76).

Substituting $m = 0$ into (3-86b) and $\beta = 0$ into (3-86a), and expanding the derivative in (3-86a), we get two ordinary differential equation:

$$r^2 \frac{d^2 f}{dr^2} + 2r \frac{df}{dr} - n(n+1)f = 0 \quad (4)$$

$$\frac{1}{\sin \theta} \frac{d}{d\theta} \left(\sin \theta \frac{dg}{d\theta} \right) + n(n+1)g = 0 \quad (5)$$

Cont'd

Cont'd

Equation (4) is known as the Euler's differential equation [1]. It has a formal solution

$$f(r) = Ar^n + Br^{-(n+1)} \quad (6)$$

Using substitution $x = \cos\theta$, (5) can be transformed into (V-1) which is known as the Legendre's differential equation. Its formal solution is given by (V-4) or, equivalently, by

$$g(\theta) = CP_n(\cos\theta) + DQ_n(\cos\theta) \quad (7)$$

Thus, the general solution to (2) is given by

$$\psi(r, \theta) = \left(Ar^n + Br^{-(n+1)} \right) \times [CP_n(\cos\theta) + DQ_n(\cos\theta)] \quad (8)$$

As was already mentioned, (2) needs to be solved for the potentials ψ_{in} and ψ_{out} . Both of these solutions have the same general form (8), but they differ in the choice of the coefficients A , B , C , and D . The choice of the coefficients depends on the boundary condition for the electric field \bar{E} on the surface of the sphere and the asymptotic behavior of \bar{E} at infinity as well as at the point $r = 0$ inside the sphere.

Coefficient D is easily fixed for both potentials, ψ_{in} and ψ_{out} , by realizing that Legendre function $Q_n(\cos\theta)$ exhibits singularities at $\theta = 0, \pi$ or $x = \pm 1$ (see Figure V-2), therefore it should be removed from (8) by setting $D = 0$.

As far as potential $\psi_{in}(r, \theta)$ is concerned, it must be finite for all r , including $r = 0$. This means that $B_{in} = 0$, where subscript 'in' indicates that this coefficient is only for the potential $\psi_{in}(r, \theta)$. Thus, our two solutions to (2) are

$$\psi_{in}(r, \theta) = A_{in}r^n P_n(\cos\theta) \quad (9a)$$

$$\psi_{out}(r, \theta) = \left(A_{out}r^n + B_{out}r^{-(n+1)} \right) P_n(\cos\theta) \quad (9b)$$

Next, we use the knowledge of the asymptotic behavior of \bar{E} at infinity. Namely, for ($r = \infty$) the electric field \bar{E} should be equal to the original uniform field \bar{E}^i .

$$\bar{E}(r \rightarrow \infty, \theta) = -\nabla \psi_{out} = \bar{E}^i = \hat{a}_z E_0 \quad (10)$$

Cont'd

Cont'd

This condition requires that

$$\psi_{out}(r \rightarrow \infty, \theta) = -E_0 z = -E_0 r \cos \theta \quad (11)$$

Using (V-9), we make a substitution $\cos \theta = P_1(\cos \theta)$ and then equate (11) and (9b)

$$(A_{out} r^n + B_{out} r^{-(n+1)}) P_n(\cos \theta) = -E_0 r P_1(\cos \theta) \Big|_{r \rightarrow \infty} \quad (12)$$

Clearly, this equality is only possible if $n=1$ on the left hand side. Thus,

$$(A_{out} r + B_{out} r^{-2}) P_1(\cos \theta) = -E_0 r P_1(\cos \theta) \Big|_{r \rightarrow \infty} \quad (13)$$

$$A_{out} + \frac{B_{out}}{r^3} = -E_0 \Big|_{r \rightarrow \infty} \quad (14)$$

Therefore, letting ($r = \infty$) in (14), we get $A_{out} = -E_0$. Thus, potentials (9) become

$$\psi_{in}(r, \theta) = A_{in} r \cos \theta \quad (15a)$$

$$\psi_{out}(r, \theta) = \left(-E_0 r + B_{out} r^{-2} \right) \cos \theta \quad (15b)$$

The remaining two coefficient are fixed by using the boundary conditions (1-26a) and (1-30a). Using (1) and (II-24), the electric field \bar{E} can be written as

$$\bar{E}_{in} = A_{in} (-\hat{a}_r \cos \theta + \hat{a}_\theta \sin \theta) \quad (16a)$$

$$\bar{E}_{out}(r, \theta) = \hat{a}_r \left(E_0 + B_{out} \frac{2}{r^3} \right) \cos \theta - \hat{a}_\theta \left(E_0 - B_{out} \frac{1}{r^3} \right) \sin \theta \quad (16b)$$

Thus, conditions (1-26a) and (1-30a) on the surface of the sphere, ($r = a$), give

$$-\hat{a}_\theta \left(E_0 - B_{out} \frac{1}{a^3} \right) \sin \theta = \hat{a}_\theta A_{in} \sin \theta \quad (17a)$$

$$\hat{a}_r \varepsilon_1 \left(E_0 + B_{out} \frac{2}{a^3} \right) \cos \theta = -\hat{a}_r \varepsilon_2 A_{in} \cos \theta \quad (17b)$$

where $\varepsilon_1 = \varepsilon_0$, and $\varepsilon_2 = \varepsilon_0 \varepsilon_r$. From here, it is easy to show that

Cont'd

Cont'd

$$B_{out} = a^3 \left(\frac{\varepsilon_r - 1}{\varepsilon_r + 2} \right) E_0 \quad (19)$$

$$A_{in} = - \left(\frac{3}{\varepsilon_r + 2} \right) E_0$$

Thus, potentials (15) becomes

$$\psi_{in}(r, \theta) = - \left(\frac{3}{\varepsilon_r + 2} \right) E_0 r \cos \theta \quad (20a)$$

$$\psi_{out}(r, \theta) = -E_0 r \cos \theta + \left(\frac{\varepsilon_r - 1}{\varepsilon_r + 2} \right) E_0 \frac{a^3}{r^2} \cos \theta \quad (20b)$$

Finally, substituting (19) into (16), the electric field \bar{E} inside and outside the sphere is

$$\bar{E}_{in} = \left(\frac{3}{\varepsilon_r + 2} \right) E_0 (\hat{a}_r \cos \theta - \hat{a}_\theta \sin \theta) \quad (21a)$$

$$\bar{E}_{out}(r, \theta) = \hat{a}_r E_0 \left(1 + 2\varepsilon \frac{a^3}{r^3} \right) \cos \theta - \hat{a}_\theta E_0 \left(1 - \varepsilon \frac{a^3}{r^3} \right) \sin \theta \quad (21b)$$

where $\varepsilon = \left(\frac{\varepsilon_r - 1}{\varepsilon_r + 2} \right)$.

Using (II-13a), the fields in (21) can be transformed into rectangular coordinate system

$E_{(x)in}(x, y, z) = E_{(y)in}(x, y, z) = 0$
$E_{(z)in}(x, y, z) = E_0 \left(\frac{3}{\varepsilon_r + 2} \right)$

(22a)

Cont'd

cont'd

$$\begin{aligned}
 E_{(x)out}(x, y, z) &= E_0 3 \left(\frac{\epsilon_r - 1}{\epsilon_r + 2} \right) \frac{a^3}{r^5} xz \\
 E_{(y)out}(x, y, z) &= E_0 3 \left(\frac{\epsilon_r - 1}{\epsilon_r + 2} \right) \frac{a^3}{r^5} yz \\
 E_{(z)out}(x, y, z) &= E_0 + E_0 \frac{a^3}{r^5} (2z^2 - x^2 - y^2)
 \end{aligned} \tag{22b}$$

$r = \sqrt{x^2 + y^2 + z^2}$

[1] N.H. Asmar, "Applied Complex Analysis with Partial Differential Equations", Prentice-Hall, Inc., New Jersey, 2002

10.18 Using duality as outlined in Table 7-2, we can write the magnetic fields both internal and external to the magnetic sphere, as

$$H_{(x)in}(x, y, z) = H_{(y)in}(x, y, z) = 0$$

$$H_{(z)in}(x, y, z) = H_0 \left(\frac{3}{\mu_r + 2} \right)$$

$$H_{(x)out}(x, y, z) = H_0 3 \left(\frac{\mu_r - 1}{\mu_r + 2} \right) \frac{a^3}{r^5} xz$$

$$H_{(y)out}(x, y, z) = H_0 3 \left(\frac{\mu_r - 1}{\mu_r + 2} \right) \frac{a^3}{r^5} yz$$

$$H_{(z)out}(x, y, z) = H_0 + H_0 \frac{a^3}{r^5} (2z^2 - x^2 - y^2)$$

$$r = \sqrt{x^2 + y^2 + z^2}$$

$$10.19 \quad (f_r)_{011}^{TM} = \frac{2.744}{2\pi a \sqrt{\mu\epsilon}} \quad (f_r)_{021}^{TM} = \frac{3.870}{2\pi a \sqrt{\mu\epsilon}}$$

$$\frac{(f_r)_{021}^{TM}}{(f_r)_{011}^{TM}} = \frac{3.870/(2\pi a \sqrt{\mu\epsilon})}{2.744/(2\pi a \sqrt{\mu\epsilon})} = \frac{3.870}{2.744} = 1.41$$

$$(f_r)_{011}^{TM} = \frac{2.744(3 \times 10^9)}{2\pi a \sqrt{2.56}} = 1 \times 10^9 \Rightarrow a = \frac{2.744(30)}{2\pi(1.6)} = 8.168 \text{ cm}$$

10.20

a) Dominant degenerate modes:

TM_{001}^r (even), TM_{111}^r (even), TM_{111}^r (odd)

$$(f_r)_{mnp}^{TM^r} = \frac{\zeta'_{np}}{2\pi a \sqrt{\mu\epsilon}} \quad \zeta'_{11} = 2.744 \quad m = 0, 1 (\text{even, odd}) - 3 \text{ modes}$$

$$(f_r)_{011}^{TM^r} (\text{even}) = (f_r)_{111}^{TM^r} (\text{even}) = (f_r)_{111}^{TM^r} (\text{odd}) = \frac{\zeta'_{11}}{2\pi a \sqrt{\mu\epsilon}}$$

$$\frac{\zeta'_{11}}{2\pi a \sqrt{\mu\epsilon}} = \frac{\zeta'_{11} \cdot c}{2\pi a \sqrt{\mu_r \epsilon_r}} = \frac{2.744 \cdot (30 \times 10^9 \text{ cm/s})}{2\pi(2 \text{ cm})\sqrt{(1)(1)}} = 6.551 \times 10^9 \text{ Hz}$$

$$(f_r)_{011}^{TM^r} (\text{even}) = (f_r)_{111}^{TM^r} (\text{even}) = (f_r)_{111}^{TM^r} (\text{odd}) = 6.551 \text{ GHz}$$

b) Next higher order modes:

TM_{021}^r (even), TM_{121}^r (even), TM_{121}^r (odd), TM_{221}^r (even), TM_{221}^r (odd)

$$\zeta'_{21} = 3.870 \quad m = 0, 1, 2 (\text{even, odd}) - 8 \text{ modes}$$

$$\frac{\zeta'_{11}}{2\pi a \sqrt{\mu\epsilon}} = \frac{3.870 \cdot (30 \times 10^9 \text{ cm/s})}{2\pi(2 \text{ cm})\sqrt{(1)(1)}} = 9.239 \times 10^9 \text{ Hz}$$

$$BW = 9.239 \text{ GHz} - 6.551 \text{ GHz} = 2.688 \text{ GHz}$$

$$BW = 2.668 \text{ GHz}$$

c) The dominant degenerate modes can operate over a bandwidth of 2.688 GHz before the next higher order degenerate modes begin to operate.

$$\frac{(f_r)_{001}^{TM}}{(f_r)_{011}^{TM}} = \frac{1}{\sqrt{\epsilon_r}} = \frac{1}{2} \Rightarrow \boxed{\epsilon_r = 4}$$

$$10.21 \quad Q = 1.004 \frac{\eta_0}{R_s} = 1.004 \frac{\sqrt{\mu_0 \epsilon_0}}{R_s} = \frac{1.004 \eta_0}{\sqrt{\epsilon_r} R_s}$$

$$R_s = \sqrt{\frac{\omega \mu_0}{2\pi}} = \sqrt{\frac{2\pi f (4\pi \times 10^7)}{2\pi}} = 2\pi \sqrt{\frac{f (10^{-7})}{\sigma}} = \frac{2\pi}{3} \sqrt{\frac{10 \times 10^9 (10^{-7})}{5.7 \times 10^7}}$$

$$R_s = \frac{2\pi \times 10^{-2}}{\sqrt{5.7}} = 2.6317 \times 10^{-2}$$

$$Q = \frac{1.004}{\sqrt{\epsilon_r}} \frac{377}{2.6317 \times 10^{-2}} = 10,000 \Rightarrow \sqrt{\epsilon_r} = \frac{1.004(377)}{2.63(10,000)(10^{-2})}$$

$$\sqrt{\epsilon_r} = 1.4383 \Rightarrow \epsilon_r = 2.0686$$

$$10.22 \quad \sigma = 5.76 \times 10^7 \text{ S/m}, Q = 10,000$$

(a) The resonant frequency of the dominant TM₀₁₁ mode is

$$(f_r)_{011}^{TM} = \frac{J_{11}}{2\pi a \sqrt{\mu_0 \epsilon_0}} = \frac{2.744}{2\pi a \sqrt{\mu_0 \epsilon_0}}$$

$$Q = 1.004 \frac{\eta_0}{R_s} = 1.004 \frac{377}{\sqrt{2\pi f (4\pi \times 10^7)}} = \frac{1.004(377)}{2\pi \times 10^{-7} \sqrt{\frac{1}{5.76}}} = 10^4$$

$$\sqrt{f} = \frac{1.004(377) \times 10^7}{2\pi (10^4)} \sqrt{5.76} = 1.4458 \times 10^5$$

$$f_r = 20.903 \text{ GHz} = 20.903 \times 10^9$$

$$\text{Thus } \frac{2.744}{2\pi a \sqrt{\mu_0 \epsilon_0}} = \frac{2.744(30 \times 10^9)}{2\pi a} = 20.903 \times 10^9$$

$$a = \frac{2.744(30)}{2\pi (20.903)} = 0.6268 \text{ cm}$$

(b) From the solution of Problem 10.21

$$R_s = \frac{2\pi \times 10^{-2}}{\sqrt{5.76}} = 2.618 \times 10^{-2}, Q = \frac{1.004(377)}{\sqrt{\epsilon_r} (2.618 \times 10^{-2})} = \frac{10000}{3}$$

$$\sqrt{\epsilon_r} = \frac{1.004(377)(3)}{2.618 \times 10^{-2}} = 4.3374 \Rightarrow \epsilon_r = 18.813$$

$$(f_r)_{011}^{TM} = \frac{2.744}{2\pi a \sqrt{\mu_0 \epsilon_r}} = \frac{2.744 \times 10^9 (30)}{2\pi (0.6268) \sqrt{18.813}} = 4.8191 \times 10^9$$

$$(f_r)_{011}^{TM} = 4.8191 \times 10^9 = 4.8191 \text{ GHz}$$

$$10.23 \quad (A_r)_{111}(\text{even}) = -B_{111}C_3 \hat{J}_1(2.744 \frac{r}{a}) \sin\theta \cos\phi = -A_{111} \hat{J}_1(2.744 \frac{r}{a}) \sin\theta \cos\phi$$

$$H_r = 0$$

$$H_\theta = \frac{1}{\mu r} \frac{1}{\sin\theta} \frac{\partial A_r}{\partial \phi} = + \frac{A_{111}}{\mu r \sin\theta} \hat{J}_1(2.744 \frac{r}{a}) \sin\phi \sin\theta = \frac{A_{111}}{\mu r} \hat{J}_1(2.744 \frac{r}{a}) \sin\phi$$

$$H_\phi = -\frac{1}{\mu r} \frac{\partial A_r}{\partial \theta} = \frac{A_{111}}{\mu r} \hat{J}_1(2.744 \frac{r}{a}) \cos\theta \cos\phi$$

$$|H| = \sqrt{H_\theta^2 + H_\phi^2} = \frac{A_{111}}{\mu r} \hat{J}_1(2.744 \frac{r}{a}) \sqrt{\sin^2\phi + \cos^2\theta \cos^2\phi} = \frac{A_{111}}{\mu r} \hat{J}_1(2.744 \frac{r}{a}) H_o(\theta, \phi)$$

$$H_o(\theta, \phi) = \sqrt{\sin^2\phi + \cos^2\theta \cos^2\phi}$$

$$W = 2W_e = 2W_m = 2 \left\{ \frac{1}{4} \iint \iint |H|^2 dV \right\} = \frac{1}{2} \frac{|A_{111}|^2}{\mu^2} \int_a^a \hat{J}_1^2(2.744 \frac{r}{a}) r^2 dr \left\{ \int_0^{2\pi} \int_0^\pi |H_o(\theta, \phi)|^2 \sin\theta d\theta d\phi \right\}$$

$$W = \frac{|A_{111}|^2}{\mu^2} \int_a^a \hat{J}_1^2(2.744 \frac{r}{a}) dr \left[\int_0^\pi \int_0^\pi |H_o(\theta, \phi)|^2 \sin\theta d\theta d\phi \right] \text{ where } |H_o(\theta, \phi)|^2 = \sin^2\phi + \cos^2\theta \cos^2\phi$$

$$P_d = \frac{R_s}{2} \int_0^{2\pi} \int_0^\pi \underline{J}_s \cdot \underline{J}_s^* r^2 \sin\theta d\theta d\phi = \frac{R_s}{2} \int_0^\pi \int_0^\pi \underline{J}_s \cdot \underline{J}_s^* a^2 \sin\theta d\theta d\phi$$

$$\underline{J}_s = \hat{n} \times \underline{H} = -\hat{a}_r \times (\hat{a}_\theta H_\theta + \hat{a}_\phi H_\phi) = (\hat{a}_\theta H_\phi + \hat{a}_\phi H_\theta) \\ = \left[\hat{a}_\theta \frac{|A_{111}|}{\mu r} \hat{J}_1(2.744 \frac{r}{a}) \cos\theta \cos\phi + \hat{a}_\phi \frac{|A_{111}|}{\mu r} \hat{J}_1(2.744 \frac{r}{a}) \sin\phi \right]_{r=a}$$

$$\underline{J}_s = \hat{a}_\theta \frac{1}{\mu a} \hat{J}_1(2.744) \cos\theta \cos\phi + \hat{a}_\phi \frac{|A_{111}|}{\mu a} \hat{J}_1(2.744) \sin\phi$$

$$\underline{J}_s \cdot \underline{J}_s^* = \frac{|A_{111}|^2}{\mu^2 a^2} \hat{J}_1^2(2.744) [\cos^2\theta \cos^2\phi + \sin^2\phi] = \frac{|A_{111}|^2}{\mu^2 a^2} \hat{J}_1^2(2.744) |H_o(\theta, \phi)|^2$$

$$P_d = \frac{R_s}{2} \frac{|A_{111}|^2}{\mu^2} \hat{J}_1^2(2.744) \int_0^\pi \int_0^\pi |H_o(\theta, \phi)|^2 d\theta d\phi$$

Therefore

$$Q = \omega_r \frac{W}{P_d} = \omega_r \frac{\frac{1}{2} \frac{|A_{111}|^2}{\mu^2} \int_a^a \hat{J}_1^2(2.744 \frac{r}{a}) dr \left[\int_0^\pi \int_0^\pi |H_o(\theta, \phi)|^2 \sin\theta d\theta d\phi \right]}{\frac{R_s}{2} \frac{|A_{111}|^2}{\mu^2} \hat{J}_1^2(2.744) \left[\int_0^\pi \int_0^\pi |H_o(\theta, \phi)|^2 \sin\theta d\theta d\phi \right]}$$

$$Q = \frac{\mu \omega_r}{R_s} \frac{\int_a^a \hat{J}_1^2(2.744 \frac{r}{a}) dr}{\hat{J}_1^2(2.744)}$$

$$\text{From Pg. 567} \Rightarrow \int_0^a \hat{J}_1^2(2.744 \frac{r}{a}) dr = \frac{1.137}{B_r}, \quad \hat{J}_1^2(2.744) = (1.0640)^2 = 1.132$$

$$Q = \frac{\mu \omega_r}{R_s} \frac{1.137}{B_r 1.132} = 1.004 \frac{\mu \omega_r}{B_r R_s} = 1.004 \frac{\sqrt{\mu/\epsilon}}{R_s} = 1.004 \frac{\eta}{R_s} \quad \begin{matrix} \text{Same as} \\ (10-73) \end{matrix}$$

$$10.24 \quad (A_r)_{111}(\text{odd}) = B_{111} D_3 \hat{J}_1(2.744 \frac{r}{a} \sin\theta \sin\phi) = -C_{111} \hat{J}_1(2.744 \frac{r}{a} \sin\theta \sin\phi)$$

By comparing the solution of this to that of Problem 10.23, we see that the two have the same solution which is also the same as that of (10.73). Therefore

$$Q = 1.004 \frac{\eta}{R_s}$$

10.25

$$Q = 1.004 \frac{\eta}{R_s} \Rightarrow \eta = \sqrt{\frac{\mu}{\epsilon}} = \eta_0 \sqrt{\frac{\mu_r}{\epsilon_r}} = 120\pi \sqrt{\frac{\mu_r}{\epsilon_r}} \quad \text{and} \quad R_s = \sqrt{\frac{\omega\mu}{2\sigma}}$$

$$Q = 1.004 \frac{120\pi \sqrt{\frac{\mu_r}{\epsilon_r}}}{\sqrt{\frac{\omega\mu}{2\sigma}}} = 1.004 \frac{120\pi}{\sqrt{\frac{\omega\mu_0\epsilon_r}{2\sigma}}}$$

(a) The medium within the cavity is air: $\mu = \mu_0, \epsilon = \epsilon_0$

The resonant frequency for the dominant mode of a spherical cavity is:

$$(f_r)_{011}^{\text{TM}^r}(\text{even}) = (f_r)_{111}^{\text{TM}^r}(\text{even}) = \\ = (f_r)_{111}^{\text{TM}^r}(\text{odd}) = \frac{\zeta_{11}}{2\pi a \sqrt{\mu\epsilon}} \frac{\zeta_{11} \cdot c}{2\pi a \sqrt{\mu_r \epsilon_r}} = \frac{2.744 \cdot (30 \times 10^9 \text{ cm/s})}{2\pi (2 \text{ cm}) \sqrt{(1)(1)}} = 6.551 \times 10^9 \text{ Hz}$$

$$Q = 1.004 \frac{120\pi}{\sqrt{\frac{\omega\mu_0\epsilon_r}{2\sigma}}} = 1.004 \frac{120\pi}{\sqrt{\frac{2\pi(6.551 \times 10^9 \text{ Hz})(4\pi \times 10^{-7})(1)}{2(5.76 \times 10^7 \text{ S/m})}}} = 17,862$$

(b) The medium within the cavity is polystyrene: $\mu = \mu_0, \epsilon_r = 2.56$

The resonant frequency for the dominant mode of a spherical cavity is:

$$(f_r)_{011}^{\text{TM}^r}(\text{even}) = (f_r)_{111}^{\text{TM}^r}(\text{even}) = \\ = (f_r)_{111}^{\text{TM}^r}(\text{odd}) = \frac{\zeta_{11}}{2\pi a \sqrt{\mu\epsilon}} \frac{\zeta_{11} \cdot c}{2\pi a \sqrt{\mu_r \epsilon_r}} = \frac{2.744 \cdot (30 \times 10^9 \text{ cm/s})}{2\pi (2 \text{ cm}) \sqrt{(1)(2.56)}} = 4.094 \times 10^9 \text{ Hz}$$

$$Q = 1.004 \frac{120\pi}{\sqrt{\frac{\omega\mu_0\epsilon_r}{2\sigma}}} = 1.004 \frac{120\pi}{\sqrt{\frac{2\pi(4.094 \times 10^9 \text{ Hz})(4\pi \times 10^{-7})(2.56)}{2(5.76 \times 10^7 \text{ S/m})}}} = 14,122$$

air	$\Rightarrow Q = 17,862$
polystyrene	$\Rightarrow Q = 14,122$

$$10.26 \quad Q = 1.004 \frac{\eta}{R_s} = 1.004 \frac{377}{\sqrt{\pi f_r \mu_0}} = 19000 \approx 10^4 = \frac{1.004(377)\sqrt{\sigma}}{\sqrt{\pi f_r \mu_0}}$$

$$f_r = \frac{1.004(377)\sqrt{\sigma}}{\sqrt{\pi \mu_0}} \times 10^4 \Rightarrow f_r = [1.004(377) \times 10^4]^2 \frac{\sigma}{\pi \mu_0} = 20.89 \text{ GHz}$$

$$f_r = 20.89 \times 10^9 = 20.89 \text{ GHz}$$

$$f_r = \frac{2.744}{2\pi a \sqrt{\mu_0 \epsilon_r}} = 20.89 \times 10^9 \Rightarrow a = \frac{2.744}{2\pi \sqrt{\mu_0 \epsilon_r} (20.89 \times 10^9)} = \frac{2.744(30 \times 10^9)}{2\pi (20.89 \times 10^9)}$$

$$a = 0.6272 \text{ cm}$$

$$Q = 1.004 \frac{\sqrt{\mu/\epsilon}}{\sqrt{\pi f_r \mu_0}} = \frac{1.004 \sqrt{\mu/\epsilon_0}}{\sqrt{\epsilon_r}} \sqrt{\frac{\sigma}{\pi \mu_0} \frac{1}{f_r}} = 1.004 \sqrt{\frac{\sigma}{\pi \mu_0}} \frac{1}{f_r} \left\{ \frac{2\pi a \sqrt{\mu_0 \epsilon_r}}{2.744} \right\}^2$$

$$Q \approx (\epsilon_r)^{1/4} \frac{\sigma}{(\epsilon_r)^{1/2}} = \frac{1}{(\epsilon_r)^{1/4}} \Rightarrow (\epsilon_r)^{1/4} = 3 \Rightarrow \epsilon_r = (3)^4 = 81$$

$$10.27 \quad \text{TE}^r \text{ Modes: } (E_r)_{mnp} = A_{mnp} \hat{J}_n(br) P_n^m(\cos\theta) [C_3 \cos(m\phi) + D_3 \sin(m\phi)]$$

(a) Boundary Conditions: $H_\phi(r=a) = H_\phi(r=a) = 0$

Because the B.C.s are on the tangential magnetic fields, then through duality, the TE^r modes for the PMC are the same as the TM^r modes for the PEC. Thus

$$(f_r)_{mnp}^{\text{TE}^r} = \frac{J_{np}}{2\pi a \sqrt{\mu \epsilon}}, \quad \begin{matrix} m=0, 1, 2, \dots \leq n \\ n=1, 2, 3, \dots \\ p=1, 2, 3, \dots \end{matrix}$$

(b) The resonant frequency of the dominant TE_{011}^r mode is

$$(f_r)_{011}^{\text{TE}^r} = \frac{2.744}{2\pi a \sqrt{\mu \epsilon}} \quad \text{based on Table 10-2.}$$

$$10.28 \quad \text{TM}^r \text{ Modes: } (H_r)_{mnp} = B_{mnp} \hat{J}_n(br) P_n^m(\cos\theta) [C_3 \cos(m\phi) + D_3 \sin(m\phi)]$$

(a) Boundary Conditions: $H_\phi(r=a) = H_\phi(r=a) = 0$

Because the B.C.s are on the tangential magnetic fields, then thru duality the TM^r modes for the PMC are the same as the TE^r modes for the PEC. Thus

$$(f_r)_{mnp}^{\text{TM}^r} = \frac{J_{np}}{2\pi a \sqrt{\mu \epsilon}}, \quad \begin{matrix} m=0, 1, 2, \dots \leq n \\ n=1, 2, 3, \dots \\ p=1, 2, 3, \dots \end{matrix}$$

(b) The resonant frequency of the dominant TM_{011}^r mode is

$$(f_r)_{011}^{\text{TM}^r} = \frac{4.493}{2\pi a \sqrt{\mu \epsilon}} \quad \text{based on Table 10-1.}$$

10.29 A spherical cavity with $\epsilon_r \gg 1$ the boundary conditions
 (a) are similar to those of a PEC spherical cavity. Therefore, through duality, the TE_{mnp}^r modes for the PMC ($\epsilon_r \gg 1$) are the same as the TM_{mnp}^r of the PEC. Therefore the resonant frequency is given by (10-64) or

$$(f_r)_{mnp}^{TE^r}(\text{PMC}) = \frac{J_{n,p}}{2\pi a \sqrt{\mu \epsilon}} \quad \begin{matrix} m=0,1,2,\dots & \leq n \\ n=1,2,3,\dots & \\ p=1,2,3,\dots & \end{matrix}$$

The dominant mode is, the TE_{011}^r or

$$(f_r)_{011}^{TE^r}(\text{PMC}) = \frac{J_{0,1}}{2\pi a \sqrt{\mu \epsilon}} = \frac{2.744}{2\pi a \sqrt{\mu \epsilon}} \quad \text{based on Table 10-2}$$

(b) For $a = 3 \text{ cm}$, $\epsilon_r = 81$, $\mu_r = 1$

$$(f_r)_{011}^{TE^r}(\text{PMC}) = \frac{2.744(30 \times 10^9)}{2\pi(3)\sqrt{81}} = 0.4852 \times 10^9 = 0.4852 \text{ GHz}$$

10.30 A spherical cavity with $\epsilon_r \gg 1$ the boundary conditions are

(a) similar to those of a PEC spherical cavity. Therefore, through duality, the TM_{mnp}^r modes for the PMC ($\epsilon_r \gg 1$) are the same as the TE_{mnp}^r of the PEC. Therefore the resonant frequency is given by (10-59a), or

$$(f_r)_{mnp}^{TM^r}(\text{PMC}) = \frac{J_{n,p}}{2\pi a \sqrt{\mu \epsilon}} \quad \begin{matrix} m=0,1,2,\dots & \leq n \\ n=1,2,3,\dots & \\ p=1,2,3,\dots & \end{matrix}$$

The dominant mode is the TM_{011}^r or

$$(f_r)_{011}^{TM^r}(\text{PMC}) = \frac{J_{0,1}}{2\pi a \sqrt{\mu \epsilon}} = \frac{4.493}{2\pi a \sqrt{\mu \epsilon}} \quad \text{based on Table 10-1}$$

(b) For $a = 3 \text{ cm}$, $\epsilon_r = 81$, $\mu_r = 1$

$$(f_r)_{011}^{TM^r}(\text{PMC}) = \frac{4.493(30 \times 10^9)}{2\pi(3)\sqrt{81}} = 0.7945 \times 10^9 = 0.7945 \text{ GHz}$$

a. From Sections 10.4.1 and 10.4.2 we saw that the dominant mode for the complete spherical cavity is the TM_{011}^r with the reduced vector potential given by (10-65a), or

$$(A_r)_{011}^{\text{TM}} = B_{011} C_3 \hat{J}_1(2.744 \frac{r}{a}) \cos\theta = A_{011} \hat{J}_1(2.744 \frac{r}{a}) \cos\theta$$

The additional boundary conditions that the hemispherical cavity introduces are

$$E_r(0 \leq r \leq a, \theta = \pi/2, 0 \leq \phi \leq 2\pi) = 0$$

$$E_\phi(0 \leq r \leq a, \theta = \pi/2, 0 \leq \phi \leq 2\pi) = 0$$

Since the E_r and E_ϕ components for the TM^r modes are given by (10-27a) and (10-27c), or

$$E_r = \frac{1}{j\omega\mu\epsilon} \left(\frac{\partial^2}{\partial r^2} + \beta^2 \right) A_r \sim \cos\theta$$

$$E_\phi = \frac{1}{j\omega\mu\epsilon} \frac{1}{r \sin\theta} \frac{\partial^2 A_r}{\partial r \partial \phi} \sim \frac{\cos\theta}{\sin\theta} = \cot\theta$$

we see that both satisfy the additional boundary conditions introduced by the bottom of the hemispherical cavity. There the TM_{011}^r mode is also the dominant mode of the hemispherical cavity. Its resonant frequency is given by (10-64), or

$$(f_r)_{011}^{\text{TM}} = \frac{2.744}{2\pi a \sqrt{\mu\epsilon}}$$

b. The stored energy of the hemispherical cap is one half of the entire spherical cap, or one half of (10-70). In addition the power dissipated on the upper curved surface of the hemispherical cap is one half of that for the entire spherical cavity, or one half of (10-72). The power dissipated on the bottom flat surface is equal to

$$P_{df} = \frac{R_s}{2} \int_0^a \int_0^{\pi/2} \underline{\underline{J}}_s \cdot \underline{\underline{J}}_s^* | r dr d\phi$$

$$\underline{\underline{J}}_s = \hat{n} \times \underline{H} |_{\theta=\pi/2} = \hat{a}_\theta \times \hat{a}_\phi H_\phi |_{\theta=\pi/2} = -\hat{a}_r H_\phi |_{\theta=\pi/2} = -\hat{a}_r \frac{B'_{011}}{hr} \hat{J}_1(2.744 \frac{r}{a})$$

$$P_{df} = \frac{R_s}{2} \frac{|B'_{011}|^2}{\mu^2} \int_0^a \int_0^{2\pi} \frac{1}{r} \hat{J}_1^2(2.744 \frac{r}{a}) dr d\phi = \frac{R_s}{2} \frac{|B'_{011}|^2}{\mu^2} (2\pi) \int_0^a \frac{1}{r} \hat{J}_1^2(2.744 \frac{r}{a}) dr$$

$$\begin{aligned} \int_0^a \frac{1}{r} \hat{J}_1^2(2.744 \frac{r}{a}) dr &= \int_0^a \frac{1}{r} \left(\frac{\pi \beta r}{2} \right) \hat{J}_{3/2}^2(2.744 \frac{r}{a}) dr, \quad \beta = \frac{2.744}{a} \\ &= \frac{\pi R}{2} \int_0^a \hat{J}_{3/2}^2(2.744 \frac{r}{a}) dr \end{aligned}$$

cont'd

10.31 cont'd.

Let $r' = \frac{r}{a} \Rightarrow \alpha dr' = dr \Rightarrow dr = \alpha dr'$

$$\int_0^a \frac{1}{r} J_1^2 \left(2.744 \frac{r}{a} \right) dr = \frac{\pi \beta a}{2} \int_0^1 J_{3/2}^2 (2.744 r') dr'$$

Using numerical integration to approximate the integral (see page that follows)

$$\int_0^a \frac{1}{r} J_1^2 \left(2.744 \frac{r}{a} \right) dr = \frac{\pi \beta a}{2} \int_0^1 J_{3/2}^2 (2.744 r') dr' = \frac{\pi \beta a}{4} J_{3/2}^2 (2.744)$$

$$\int_0^a \frac{1}{r} J_1^2 \left(2.744 \frac{r}{a} \right) dr = \frac{2.744}{4} \frac{\pi \beta}{\mu} \left(\frac{2.744}{\beta} \right) J_{3/2}^2 = \frac{2.744 \pi}{4} J_{3/2}^2 (2.744)$$

$$\therefore P_{df} = \frac{R_s}{2} \frac{|B'_{0||}|^2}{\mu^2} (2\pi) \left(\frac{2.744 \pi}{4} \right) J_{3/2}^2 (2.744)$$

The power dissipated on the curved surface:

$$P_{dc} = \frac{|B'_{0||}|^2}{\mu^2} \frac{1.132(2\pi)R_s}{3}$$

Total Power Dissipated (P_d):

$$P_d = P_{dc} + P_{df} = \frac{|B'_{0||}|^2}{\mu^2} R_s \left[\frac{1.132(2\pi)}{3} + \frac{2.744\pi^2}{4} J_{3/2}^2 (2.744) \right]$$

$$J_{3/2}^2 (2.744) = 0.2627$$

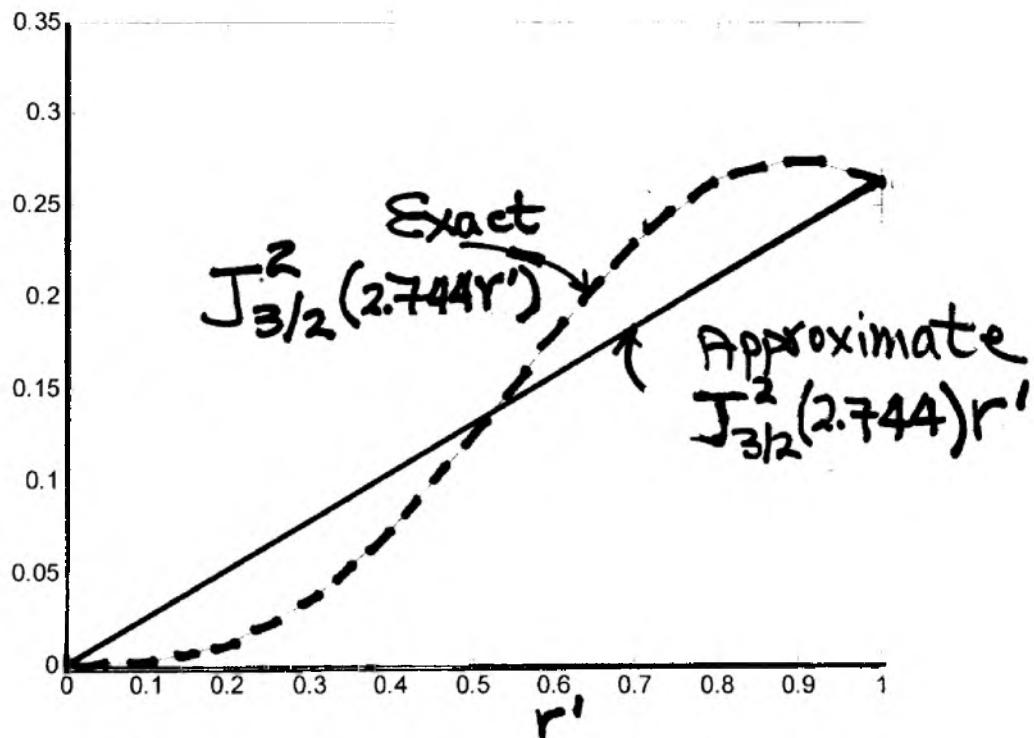
$$P_d = \frac{|B'_{0||}|^2}{\mu^2} R_s \left[2.3704 + 1.7786 \right] = \frac{|B'_{0||}|^2}{\mu^2} R_s (4.1495)$$

$$W = \frac{|B'_{0||}|^2}{\mu} \left(\frac{2\pi}{3} \right) \left(\frac{1.132}{\beta_r} \right) = \frac{|B'_{0||}|^2}{\mu} \left(\frac{2.3813}{\beta_r} \right)$$

$$\therefore Q_h = w_r \frac{W}{P_d} = w_r \frac{\frac{|B'_{0||}|^2}{\mu} \left(\frac{2.3813}{w_r \sqrt{\mu E}} \right)}{\frac{|B'_{0||}|^2}{\mu^2} R_s (4.1495)} = 0.5739 \left(\frac{n}{R_s} \right)$$

cont'd.

10.31 Cont'd.



$J_{3/2}^2(2.744r')$ was approximated by $J_{3/2}^2(2.744)r'$ so that the area of each from $0 \leq r' \leq 1$ is equal to that of the other; i.e.,

$$\int_0^1 J_{3/2}^2(2.744r') dr' = \int_0^1 J_{3/2}^2(2.744)r' dr'$$

$$Q_h = 0.5739 \frac{\sqrt{h/\epsilon}}{R_s} = 0.574 \frac{\eta}{R_s} \quad (\text{Hemispherical Cavity})$$

c. For a spherical cavity

$$Q_s = 1.004 \frac{\eta}{R_s}$$

For a cylindrical cavity with $h=d$ by (9-57) we have

$$Q_c = 0.8017 \frac{\eta}{R_s}$$

For a rectangular cavity with $a=b=c$ (cube) by (8-88a) we have

$$Q_r = 0.7405 \frac{\eta}{R_s}$$

Thus

$$\frac{Q_h}{Q_s} = \frac{0.574}{1.004} \times 100 = 57.17 \%$$

$$\frac{Q_h}{Q_c} = \frac{0.574}{0.8017} \times 100 = 71.6 \%$$

$$\frac{Q_h}{Q_r} = \frac{0.574}{0.7405} \times 100 = 77.52 \%$$

10.32 For a complete spherical cavity, the TM_{mp} mode, with the lowest resonant mode, is given in Section 10.4.2 by (10-65a) or

$$(A_r)_{011} (\text{even}) = B_{011} C_3 \hat{T}_1(\beta r) P_1^0(\cos \theta)$$

$$= B_{011} C_3 \hat{T}_1(\beta r) \cos \theta$$

The TM_{mp} electric and magnetic field components are given, respectively, by (10-27a)-(10-27c) and (10-28a)-(10-28c).

Now by placing a PEC ^{plate} right through the center in a horizontal plane reduces the complete spherical cap to 2 independent hemi-spherical PEC cavities, as shown in the Figure P10-3 for one of them.

In order to find the dominant TM_{mp} mode, we check to see if the TM_{mp} dominant mode [TM_{001}^r (even)] holds. The additional boundary condition is $[A_r]_{011} (\text{even}) = B_{011} C_3 \hat{T}_1(\beta r) \cos \theta$

$$E_r (0 \leq r \leq a, \theta = \pi/2, 0 \leq \phi \leq 2\pi) = 0 = \frac{1}{j\omega \mu \epsilon} \left(\frac{\partial^2}{\partial r^2} + \beta^2 \right) A_r \sim \cos \theta = 0$$

$$E_\phi (0 \leq r \leq a, \theta = \pi/2, 0 \leq \phi \leq 2\pi) = 0 = \frac{1}{j\omega \mu \epsilon r \sin \theta} \frac{\partial^2 A_r}{\partial \phi^2} \sim \cos \theta \Big|_{\theta=\pi/2} = 0$$

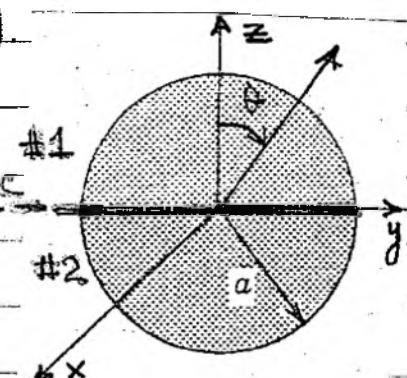
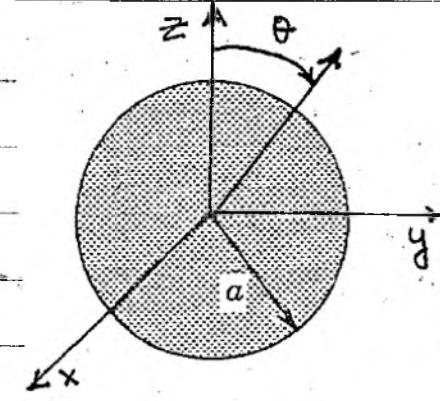
Looking at E_r , we have, from (10-27a)

$$E_r = \frac{1}{j\omega \mu \epsilon} \left(\frac{\partial^2}{\partial r^2} + \beta^2 \right) A_r = \frac{1}{j\omega \mu \epsilon} \left(\frac{\partial^2}{\partial r^2} + \beta^2 \right) \left[B_{011} C_3 \hat{T}_1(\beta r) \right] \cos \theta$$

which is satisfied for $\theta = \pi/2$. The same applies if we use E_ϕ from (10-27c) or

$$E_\phi = \frac{1}{j\omega \mu \epsilon} \frac{1}{r \sin \theta} \frac{\partial^2 A_r}{\partial \phi^2} = \frac{1}{j\omega \mu \epsilon} \frac{1}{r \sin \theta} \frac{\partial^2}{\partial \phi^2} \left[B_{011} C_3 \hat{T}_1(\beta r) \cos \theta \right]$$

which is also satisfied for $\theta = \pi/2$ ($E_\phi(\theta=\pi/2) = 0$).



the other 2 modes, TM_{011}^r (even) and TM_{111}^r (odd) do not satisfy the additional BCs on E_x and E_ϕ because their potential is proportional to $\sin\theta$, which is not equal to zero when $\theta = \pi/2$.

(a) therefore the dominant TM_{011}^r mode for the hemi-spherical cavity is the same as for the complete spherical cavity or TM_{011}^r (even) TM_{011}^r (even)

(b) The reduced vector potential is that given by (10-65a) or

$$TM_{011}^r \text{ (even)} = B_{011} C_3 \hat{T}_1(\beta r) \cos\theta$$

(c) the lowest resonant frequency is given by (10-64) and Table 10-2, or

$$(f_r)^{TM^r} = \frac{\Gamma_{011}^r}{2\pi a \sqrt{\mu_r}} \Big|_{\substack{n=0 \\ n=1 \\ p=1}} = \frac{\Gamma_{011}^r}{2\pi a \sqrt{\mu_0 \epsilon_0}} = \frac{2.744}{2\pi a \sqrt{\mu_0 \epsilon_0}} \Big|_{a=3 \text{ cm}}$$

$$(f_r)^{TM^r}_{011} = \frac{2.744 (3 \times 10^{10})}{2\pi (3)} = \frac{2.744}{2\pi} \times 10^{10} = 0.43672 \times 10^{10}$$

$$(f_r)^{TM^r}_{011} = 4.3672 \times 10^9 = 4.3672 \text{ GHz}$$

10.33 For a complete spherical cavity the TE_{011}^r mode, with the lowest resonant mode, is given in Section 10.4.1 by (10-60a) or $(F_r)_{011}^r = A_{011} C_3 \hat{T}_1(4.493 \frac{r}{a}) \cos\theta$. The corresponding to the TE_{011}^r mode, electric fields are given by (10-56a)-(10-56c), or

$$E_x = 0$$

$$E_\theta = -\frac{1}{\epsilon} \frac{1}{r} \frac{1}{rs \sin\theta} \frac{\partial F_r}{\partial \phi} = 0$$

$$E_\phi = \frac{1}{\epsilon} \frac{1}{r} \frac{1}{s} \frac{\partial F_r}{\partial \theta} = -\frac{A_{011} C_3}{\epsilon r} \hat{T}_1(4.493 \frac{r}{a}) \sin\theta$$

By placing a conductor along the xy -plane, as shown in

cont'd

10.33 cont'd

the second figure in the solution of Problem 10.32, to reduce the sphere to a hemi-sphere as shown in Figure P10-32, does not satisfy the additional boundary condition on the newly created boundary; i.e., tangential electric fields to vanish at $\theta = \pi/2$.

$$E_r = 0$$

$$E_\phi(\theta = \pi/2) = -\frac{A_{01}C_3}{\epsilon_r} \hat{J}_1(4.493 \frac{r}{a}) \sin\theta \neq 0$$

* However, the other 2 modes, TE_{111}^r (even) and TE_{111}^r (odd) do satisfy the BCs.

Therefore the dominant TE_{011}^r mode cannot exist, because it does not satisfy the boundary conditions, for the geometry of Figure P10-32. However, when the sphere is split by placing the PEC plate on the xz-plane (vertically instead of horizontally) as shown in this figure, then the dominant TE_{011}^r mode can exist because the additional boundary condition on the newly created boundary is satisfied; i.e.,

$$E_\phi(\theta = 9\pi) = -\frac{A_{01}C_3}{\epsilon_r} \hat{J}_1(4.493 \frac{r}{a}) \sin\theta = 0$$

Therefore, for this modified geometry $\theta = 9\pi$

- (a) The dominant TE_{011}^r mode does exist, but those of the TE_{111}^r (even) and TE_{111}^r (odd) do not exist.

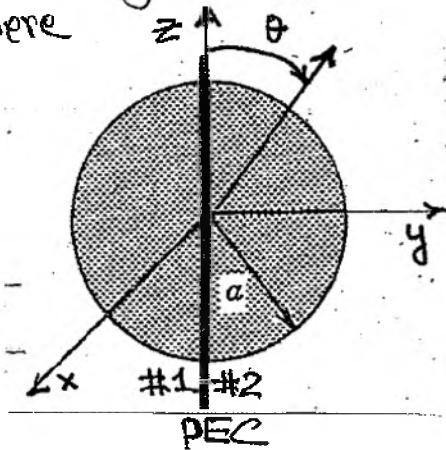
(b) The reduced vector potential for the modified geometry is

$$(F_r)_{011}^{TE} = A_{01}C_3 \hat{J}_1(4.493 \frac{r}{a}) \sin\theta$$

- (c) The corresponding resonant frequency ω is given by (10-59a),

$$\text{or } (f_r)^{TE} \Big|_{m=0, n=p=1} = (f_r)_{011}^{TE} = \frac{J_{np}}{2\pi a V \mu \epsilon_0} \Big|_{\begin{subarray}{l} J_1 = 4.493 \\ a = 3 \text{ cm} \end{subarray}} = \frac{4.493 (30 \times 10^9)}{2\pi (3)} = 7.151 \times 10^9 \text{ Hz}$$

$$(f_r)_{011}^{TE} = 7.151 \times 10^9 = 7.151 \text{ GHz}$$



10.34

For a PMC spherical cavity with $\epsilon_r \gg 1$ the modes with the lowest resonant frequency are, with vector potentials of

$\text{TE}^r(\text{PMC}) \sim \text{TM}^r(\text{PEC})$ [When $\epsilon_r \gg 1$, BCs are those of PMC (See Sol'n to Problem 10.27)]

$$(\mathbf{E}_r)_{\text{OH}} (\text{even}) \approx B_{\text{OH}} C_3 \hat{\mathbf{J}}_1 (2.744 \frac{r}{a}) \cos \theta \quad (10-65a)$$

$$(\mathbf{E}_r)_{\text{III}} (\text{even}) \approx -B_{\text{OH}} C_3 \hat{\mathbf{J}}_1 (2.744 \frac{r}{a}) \sin \theta \cos \phi \quad (10-65b)$$

$$(\mathbf{E}_r)_{\text{III}} (\text{odd}) \approx -B_{\text{III}} D_3 \hat{\mathbf{J}}_1 (2.744 \frac{r}{a}) \sin \theta \sin \phi \quad (10-65c)$$

By placing a PEC in the middle, we introduce an additional boundary condition of $E_\phi(\theta = \pi/2, 0 \leq \phi \leq 2\pi) = E_\phi(\theta = \pi/2, 0 \leq \phi \leq 2\pi) = 0$ where

$$E_\phi = 0 \quad (10-23a)$$

$$E_\phi = \frac{1}{\epsilon_r} \frac{1}{r} \frac{\partial F_r}{\partial \theta} \quad (10-23c)$$

Since $E_\phi = 0$, then:

$\text{TE}_{\text{OH}}^r (\text{even})$: $E_\phi = -B_{\text{OH}} C_3 \hat{\mathbf{J}}_1 (2.744 \frac{r}{a}) \sin \theta \neq 0$ which is not satisfied.
Thus it does not exist.

$\text{TE}_{\text{III}}^r (\text{even})$: $E_\phi = -B_{\text{III}} C_3 \hat{\mathbf{J}}_1 (2.744 \frac{r}{a}) \cos \theta \sin \phi \Big|_{\theta=\pi/2} = 0$ which is satisfied.

$\text{TE}_{\text{III}}^r (\text{odd})$: $E_\phi = -B_{\text{III}} D_3 \hat{\mathbf{J}}_1 (2.744 \frac{r}{a}) \cos \theta \sin \phi \Big|_{\theta=\pi/2} = 0$ which is satisfied.

Thus the TE_{III}^r modes with the lowest resonant frequency are:

$\left. \begin{array}{l} \text{TE}_{\text{III}}^r (\text{even}) \\ \text{TE}_{\text{III}}^r (\text{odd}) \end{array} \right\}$

with resonant frequency of $(f_r)^{\text{TE}} = \frac{J_{\text{mp}}^1}{2\pi r a \sqrt{\epsilon_r}} = \frac{J_{\text{II}}^1}{2\pi r a \sqrt{\epsilon_r}}$

$$(f_r)_{\text{III}}^{\text{TE}} = \frac{2.744 (30 \times 10^9)}{2\pi (3)(9)} = 0.48525 \times 10^9$$

$$(f_r)_{\text{III}}^{\text{TE}} = 0.48525 \text{ GHz}$$

[10.35]

$$\text{TM}^r(\text{PMC}) \sim \text{TE}^r(\text{PEC}) \quad [\text{When } \epsilon_r > 1, \text{ BCs are those of PMC; see solution to Problem 10.28}]$$

For complete sphere with $\epsilon_r > 1$

$$(A_r)_{011} (\text{even}) \approx A_{011} C_3 \hat{T}_1 (4.493 \frac{r}{a}) \cos\theta \quad (10-60a)$$

$$(A_r)_{111} (\text{even}) \approx -A_{111} C_3 \hat{T}_1 (4.493 \frac{r}{a}) \sin\theta \cos\phi \quad (10-60b)$$

$$(A_r)_{111} (\text{odd}) \approx -A_{111} D_3 \hat{T}_1 (4.493 \frac{r}{a}) \sin\theta \sin\phi \quad (10-60c)$$

By placing a PEC in the middle of the sphere, we introduce an additional boundary condition of $E_\phi (\theta = \pi/2, 0 \leq \phi \leq 2\pi) = E_\phi (\theta = \pi/2, 0 \leq \phi \leq 2\pi) = 0$ where

$$E_r = \frac{1}{j\omega \epsilon} \left(\frac{\partial^2}{\partial r^2} + \beta^2 \right) A_r \quad (10-27a)$$

$$E_\phi = \frac{1}{j\omega \epsilon} \frac{1}{r \sin\theta} \frac{\partial^2 A_r}{\partial \phi^2} \quad (10-27c)$$

TM₀₁₁ (even):

$$E_r \sim \cos\theta \Big|_{\substack{\theta = \pi/2 \\ 0 \leq \phi \leq 2\pi}} = 0 \quad \text{which is satisfied}$$

$$E_\phi = 0 \quad \text{which is satisfied}$$

TM₁₁₁ (even):

$$E_r \sim \sin\theta \cos\phi \Big|_{\substack{\theta = \pi/2 \\ 0 \leq \phi \leq 2\pi}} \neq 0 \quad \text{which is not satisfied}$$

$$E_\phi \sim \sin\theta \sin\phi \Big|_{\substack{\theta = \pi/2 \\ 0 \leq \phi \leq 2\pi}} \neq 0 \quad \text{which is not satisfied}$$

TM₁₁₁ (odd):

$$E_r \sim \sin\theta \sin\phi \Big|_{\substack{\theta = \pi/2 \\ 0 \leq \phi \leq 2\pi}} \neq 0 \quad \text{which is not satisfied}$$

$$E_\phi \sim \sin\theta \cos\phi \Big|_{\substack{\theta = \pi/2 \\ 0 \leq \phi \leq 2\pi}} \neq 0 \quad \text{which is not satisfied}$$

Thus the TM₀₁₁^r mode with the lowest resonant frequency is

$\boxed{\text{TM}_{011} (\text{even})}$	$\text{with } (f_r)_{011}^{\text{TM}^r} = \frac{\beta_{011}}{2\pi a \sqrt{\epsilon_r \mu_r}} = \frac{4.493 (30 \times 10^9)}{2\pi (3)(9)} = 0.79454 \times 10^9$
	$(f_r)_{011}^{\text{TM}^r} = 0.79454 \text{ GHz}$

[10.36] From the solution of Problem 10.27, the dominant TE_{mnp}^{r} modes for an entire sphere with PMC surface are the TM_{mnp}^{r} of a PEC, or $(TE_{011}^{r})_{\text{even}} = (TE_{111}^{r})_{\text{even}} = (TE_{111}^{r})_{\text{odd}}$

with a resonant frequency of

$$(f_r)_{011}^{TE} = (f_r)_{111}^{TE} (\text{even}) = (f_r)_{111}^{TE} (\text{odd}) = \frac{2.744}{2\pi a \sqrt{\epsilon}}$$

(c) The corresponding vector potential is analogous to (10-65a)-(10-65c), while the electric and magnetic fields are given by (10-23)-(10-24c).

Now by placing a PMC plate right through the center in a horizontal plane reduces the entire spherical cavity to two independent hemi-spherical PMC cavities.

In order to find the dominant TE_{mnp}^{r} mode, we check to see if the TE_{011}^{r} dominant mode [$TE_{011}^{r}(\text{even})$] holds. The additional BC is

$$H_r(0 \leq r \leq a, \theta = \pi/2, 0 \leq \phi \leq 2\pi) = 0$$

$$H_\phi(0 \leq r \leq a, \theta = \pi/2, 0 \leq \phi \leq 2\pi) = 0$$

$$(F_r)_{mnp} = (F_r)_{011}^{r(\text{even})} = [A_{011} C_3 \hat{T}_1 (4.493 \frac{r}{a})] \cos \theta \quad (10-60a)$$

$$H_r = \frac{1}{jw\mu\epsilon} \left(\frac{\partial^2}{\partial r^2} + \beta^2 \right) F_r \sim \left[\cos \theta \right] \quad (10-24a)$$

$$H_\phi = \frac{1}{jw\mu\epsilon} \frac{1}{r \sin \theta} \frac{\partial^2}{\partial \phi^2} F_r \sim \left[\frac{\cos \theta}{\sin \theta} \right] \quad (10-24e)$$

$$H_r(0 \leq r \leq a, \theta = \pi/2, 0 \leq \phi \leq 2\pi) = \left[\cos \theta \right] = 0$$

$$H_\phi(0 \leq r \leq a, \theta = \pi/2, 0 \leq \phi \leq 2\pi) = \left[\frac{\cos \theta}{\sin \theta} \right]_{\theta=\pi/2} = 0$$

however for the $TE_{111}^{r}(\text{even})$ mode, where F_r is

$$(F_r)_{111}^{r(\text{even})} = A_{111} C_3 \hat{T}_1 (4.493 \frac{r}{a}) \sin \theta \cos \phi \quad (10-60b)$$

and for the $TE_{111}^{r}(\text{odd})$ mode, where F_r is

$$(F_r)_{111}^{r(\text{odd})} = -A_{111} D_3 \hat{T}_1 (4.493 \frac{r}{a}) \sin \theta \sin \phi \quad (10-60c)$$

Cont'd

10.36 cont'd

The two boundary conditions on H_r and H_ϕ for the TE_{11}^r (even) and TE_{11}^r (odd) are not satisfied. Thus these two modes cannot be supported by the geometry of Fig. P10-36; only the TE_{01}^r (even) can be supported, and it is the dominant mode.

(b) The reduced vector potential for the dominant TE_{01}^r (even) mode is

$$(F_r)_{011}(\text{even}) = A_{011} C_3 \hat{J}_1(4.493 \frac{r}{a}) \cos\theta$$

(c) The resonant frequency of the dominant TE_{011}^r (even) mode is

$$(f_r)_{011}^{TE^r} = \frac{J_{11}}{2\pi a \sqrt{\mu\epsilon}} = \frac{2.744(30 \times 10^9)}{2\pi(3)} = 4.3672 \times 10^9 = 4.3672 \text{ GHz.}$$

10.37 From the solution of Problem 10.28, the dominant TM_{mnp}^r modes for an entire sphere with PMC surface are the TE_{mnp}^r of a PEC surface, or

$$(TM_{011}^r)^{(\text{even})} = (TM_{111}^r)^{(\text{even})} = (TM_{111}^r)^{(\text{odd})}$$

with a resonant frequency of

$$(f_r)_{011}^{TM^r} = (f_r)_{111}^{TM^r}(\text{even}) = (f_r)_{111}^{TM^r}(\text{odd}) = \frac{4.493}{2\pi a \sqrt{\mu\epsilon}}$$

(a) The corresponding vector potential is analogous to (10-60a)-(10-60c), while the electric and magnetic fields are given by (10-27a)-(10-28c). Now placing a PMC plate right through the center in the horizontal plane reduces the entire spherical cavity to two independent hemi-spherical PMC cavities.

In order to find the dominant TM_{mnp}^r mode, we check to see if the TM_{011}^r dominant mode [TM_{011}^r (even)] holds. The additional BCs are

$$H_r(0 \leq r \leq a, \theta = \pi/2, 0 \leq \phi \leq 2\pi) = 0$$

$$H_\phi(0 \leq r \leq a, \theta = \pi/2, 0 \leq \phi \leq 2\pi) = 0$$

$$(A_r)_{mnp} = (A_r)_{011}(\text{even}) = B_{011} C_3 \hat{J}_1(4.493 \frac{r}{a}) \cos\theta \quad (10-65a)$$

(cont'd)

10.37 Cont'd

Checking the additional boundary condition created by the PMC plate placed horizontally on the xy-plane, using (10-28a)-(10-28c)

$$H_r = 0$$

$$H_\theta = \frac{1}{\mu} - \frac{1}{r \sin \theta} \frac{\partial A_r}{\partial \phi} = 0$$

$$H_\theta = -\frac{1}{\mu} \frac{1}{r} \frac{\partial A_r}{\partial \theta} \Big|_{\theta=\pi/2} \approx [] \sin \theta \Big|_{\theta=\pi/2} \neq 0$$

which is not satisfied. However for the TM_{111} (even) mode, where

$$A_r \text{ is } (A_r)_{111} \text{ (even)} = B_{111} C_3 J_1(2.744 \frac{r}{a}) \sin \theta \cos \phi \quad (10-65b)$$

and for the TM_{111} (odd), where A_r is

$$(A_r)_{111} \text{ (odd)} = -B_{111} D_3 J_1(2.744 \frac{r}{a}) \sin \theta \sin \phi \quad (10-65c)$$

the additional boundary condition on $H_\theta (\theta = \pi/2)$ is satisfied by both the TM_{111} (even) and TM_{111} (odd) modes.

- (b) Therefore the reduced vector potential for the dominant TM_{111} (even) and TM_{111} (odd) is respectively

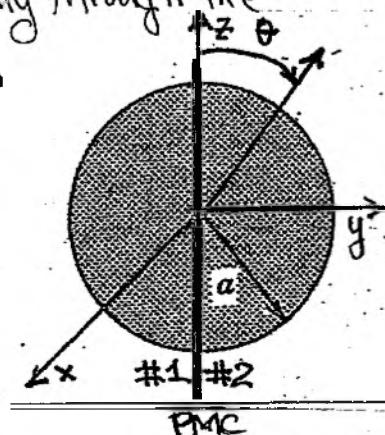
$$(A_r)_{111} \text{ (even)} = -B_{111} C_3 J_1(2.744 \frac{r}{a}) \sin \theta \cos \phi$$

$$(A_r)_{111} \text{ (odd)} = -B_{111} D_3 J_1(2.744 \frac{r}{a}) \sin \theta \sin \phi$$

- (c) The resonant frequency of the dominant TM_{111} (even) and TM_{111} (odd) modes is

$$(f_r)_{111}^{TM} \text{ (even, odd)} = \frac{J_{11}}{2\pi a V \epsilon_0} = \frac{4.493 (30 \times 10^9)}{2\pi (3)} = 7.151 \times 10^9 = 7.151 \text{ GHz}$$

P.S. It should be pointed out that if the ^{hemi-spherical} cavity was formed by placing the PMC ground plane vertically through the center of the sphere, as shown graphically, then the TM_{011} (even) mode would exist and the TM_{111} (even) and TM_{111} (odd) would not satisfy the newly created boundary conditions.



CHAPTER 11

11.1 For a displaced line source, the solution for an electric source is displayed in detail in Example 11-1. The solution for a magnetic line source follows the same procedure, and it does not have to be detailed as it can be obtained easier by using duality. Using Table 7-2 as a guide, the electric fields of the electric line source are replaced by the magnetic, the magnetic by the negative electric, ϵ by μ , μ by ϵ , n by $1/n$, I_e by I_m , etc. Therefore, from Example 11-1, we can write for the magnetic line source displaced from the origin that

$$E_p = -j \frac{\beta I_m}{4} \frac{p' \sin(\phi - \phi')}{\sqrt{|p-p'|^2}} H_1^{(2)}(\beta |p-p'|)$$

$$E_\phi = j \frac{\beta I_m}{4} \frac{[p-p' \cos(\phi - \phi')]}{\sqrt{|p-p'|^2}} H_1^{(2)}(\beta |p-p'|)$$

$$E_z = H_p = H_\phi = 0, \quad |p-p'| = \sqrt{p^2 + (p')^2 - 2pp' \cos(\phi - \phi')}$$

$$H_z = -\frac{\omega \epsilon I_m}{4} H_0^{(2)}(\beta |p-p'|)$$

11.2 Using the geometry of Figure 11-2, but with the electric

(a) line source replaced with a magnetic line source, we can write

$$\begin{aligned} H_z^t &= H_z^i + H_z^r = -\frac{\omega \epsilon I_m}{4} [H_0^{(2)}(\beta p_i) + H_0^{(2)}(\beta p_r)] \\ &= -I_m \frac{\beta}{4\eta} [H_0^{(2)}(\beta p_i) + H_0^{(2)}(\beta p_r)] \end{aligned}$$

because the image, to represent the reflected fields, has the same magnitude and phase as the actual source.

(b) For far-field observations the Hankel functions can be replaced by their asymptotic expansions of (IV-17). Therefore we can write the total magnetic field as

$$H_z^t = -I_m \frac{\beta}{4\eta} \sqrt{\frac{j2}{\pi \beta p}} \left[\frac{e^{-j\beta p_i}}{\sqrt{p_i}} + \frac{e^{-j\beta p_r}}{\sqrt{p_r}} \right] = -\frac{I_m}{\eta} \sqrt{\frac{j\beta}{8\pi}} \left[\frac{e^{-j\beta p_i}}{\sqrt{p_i}} + \frac{e^{-j\beta p_r}}{\sqrt{p_r}} \right]$$

For far-field observations

$$p_i \approx p - h \cos\left(\frac{\pi}{2} - \phi\right) = p - h \sin\phi \quad \left. \begin{array}{l} \text{for phase terms} \\ \text{for amplitude terms} \end{array} \right\}$$

$$p_r \approx p + h \cos\left(\frac{\pi}{2} - \phi\right) = p + h \sin\phi$$

$$p_i \approx p_r \approx p \quad \left. \begin{array}{l} \text{for amplitude terms} \end{array} \right.$$

Therefore the total field can be written in a more compact form,

with the radial distances referenced to the origin; as

$$H_z^t = -\frac{Im}{\eta} \sqrt{\frac{j\beta}{8\pi}} \left\{ \frac{e^{-j\beta p}}{\sqrt{p}} \left[\frac{e^{j\beta h \sin\phi} + e^{-j\beta h \sin\phi}}{2} \right] \right\}$$

$$H_z^t = -\frac{Im}{\eta} \sqrt{\frac{j\beta}{8\pi}} \frac{e^{-j\beta p}}{\sqrt{p}} \left\{ 2 \cos(\beta h \sin\phi) \right\}$$

(c)

To find the smallest height where the field will attain nulls at certain observation angles, we set the cosine function (usually referred to as the array factor) to zero. That is

$$\cos(\beta h \sin\phi_n) = 0 \Rightarrow \beta h \sin\phi_n = \cos^{-1}(0) = \frac{n\pi}{2}, n=1, 3, 5, \dots$$

For the smallest height h , we choose $n=1$, or

$$\beta h \sin\phi_n = \frac{\pi}{2} \Rightarrow h = \frac{\pi/2}{\beta \sin\phi_n} = \frac{\pi}{2} \frac{1}{2\pi \frac{1}{4\sin\phi_n}} = \frac{1}{4\sin\phi_n}$$

$$h = \frac{1}{4\sin\phi_n}$$

- For $\phi = 30^\circ$: $h = \frac{1}{4} \frac{1}{\sin 30^\circ} = \frac{1}{2}$

- For $\phi = 90^\circ$: $h = \frac{1}{4}$

11.3 For an electric line source placed above a PMC, the magnitude and phase ^{of the image} give the same as the actual source. Therefore the solution of this problem is the same as that of 11.2 except with the magnetic field replaced by the electric field using duality.

(a) $E_z^t = -I_e \eta \frac{\beta}{4} [H_0^{(2)}(\beta p_i) + H_0^{(2)}(\beta p_r)]$

(b) $H_z^t = -I_e \eta \sqrt{\frac{j\beta}{8\pi}} \left\{ \frac{e^{-j\beta p}}{\sqrt{p}} \left[2 \cos(\beta h \sin\phi) \right] \right\}$

(c) • $\phi = 30^\circ$: $h = \lambda/2$ • $\phi = 90^\circ$: $h = \lambda/4$

11.4 For a magnetic line source placed a height h above a PMC, the magnitude of the image is the same as that of the actual source while its phase is 180° (out of phase). Therefore, using the solution of Problem 11.2, we can write the solution of this problem as:

$$(a) H_z^t = -I_m \frac{\beta}{4\eta} [H_0^{(z)}(\beta p_i) - H_0^{(z)}(\beta p_r)]$$

$$(b) H_z^t = -\frac{I_m \sqrt{j\beta}}{\eta \sqrt{8\pi} \sqrt{\rho}} \left\{ 2j \left[e^{+j\beta h \sin\phi} - e^{-j\beta h \sin\phi} \right] \right\}$$

$$= -\frac{I_m \sqrt{j\beta}}{\eta \sqrt{8\pi} \sqrt{\rho}} [2j \sin(\beta h \sin\phi)]$$

(c) For the smallest height h for which the field will vanish

$$\sin(\beta h \sin\phi_n) = 0 \Rightarrow \beta h \sin\phi_n = \sin^{-1}(0) = n\pi, \quad n=1, 2, \dots$$

The smallest nontrivial ($h \neq 0$) height will be for $n=1$. Therefore

$$h = \frac{\pi}{\beta \sin\phi_n} = \frac{\pi}{2\pi \sin\phi_n} = \frac{1}{2 \sin\phi_n}$$

- $\phi = 30^\circ: h = \frac{1}{2(0.5)} = 1$

- $\phi = 90^\circ: h = \frac{1}{2}$

11.5 For an electric current element with current I_e located at

(a): $x=0, y=0$, the potential is given by

$$A_z^{(0)} = -j \frac{\mu I_e}{4} H_0^{(2)}(\beta p)$$

For a current element located at $x=x', y=y'$, the potential is given by

$$A_z = -j \frac{\mu I_e}{4} H_0^{(2)}(\beta |p - p'|) = -j \frac{\mu I_e}{4} H_0^{(2)}(\beta \sqrt{(x-x')^2 + (y-y')^2})$$

Thus for $x'=0, y'=h$ and current I_e

$$A_z^{(1)} = -j \frac{\mu I_e}{4} H_0^{(2)}(\beta \sqrt{x^2 + (y-h)^2}) = A_z^{(1)}(x, y-h)$$

(b) For the image at $x=0, y=-h$ and current $-I_e$

$$A_z^{(2)} = +j \frac{\mu I_e}{4} H_0^{(2)}(\beta \sqrt{x^2 + (y+h)^2}) = A_z^{(2)}(x, y+h)$$

Therefore the total field is equal to

$$A_z^t = A_z^{(1)}(x, y-h) - A_z^{(2)}(x, y+h) = -[A_z^{(0)}(x, y+h) - A_z^{(0)}(x, y-h)]$$

(cont'd)

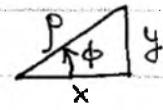
11.5 cont'd

$$\lim_{h \rightarrow 0} A_z^t = - \lim_{h \rightarrow 0} [A_z^{(e)}(x, y+h) - A_z^{(e)}(x, y-h)] = -2h \lim_{h \rightarrow 0} \frac{[A_z^{(e)}(x, y+h) - A_z^{(e)}(x, y-h)]}{2h}$$

$$= -2h \frac{\partial A_z^{(e)}}{\partial y} = -2h \frac{\partial}{\partial y} \left[-j \frac{\mu h I_e}{4} H_0^{(2)}(\beta p) \right] = j \frac{\mu h I_e}{4} \frac{\partial}{\partial y} H_0^{(2)}(\beta p)$$

where $p = \sqrt{x^2+y^2}$

$$\frac{\partial}{\partial y} = \frac{\partial}{\partial p} \frac{\partial p}{\partial y} = \frac{\partial p}{\partial y} \frac{\partial}{\partial p}$$



$$\frac{\partial p}{\partial y} = \frac{\partial}{\partial y} (x^2+y^2)^{1/2} = \frac{1}{2} (x^2+y^2)^{-1/2} (2y) = \frac{y}{\sqrt{x^2+y^2}} = \frac{y}{p} = \sin \phi$$

$$\frac{\partial}{\partial p} H_0^{(2)}(\beta p) = -\beta H_1^{(2)}(\beta p)$$

Therefore

$$\lim_{h \rightarrow 0} A_z^t = j \frac{\mu h I_e}{2} \sin \phi [-\beta H_1^{(2)}(\beta p)] = -j \frac{\beta \mu h I_e}{2} H_1^{(2)}(\beta p) \sin \phi$$

(c) For TM^Z

$$E_\phi = -j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 A_z^t}{\partial p \partial z} = 0$$

$$H_\phi = \frac{1}{h p} \frac{\partial A_z^t}{\partial p} = -j \frac{\beta h I_e}{2} H_1^{(2)}(\beta p) \cos \phi$$

$$E_\phi = -j \frac{1}{\omega \mu \epsilon} \frac{1}{p} \frac{\partial^2 A_z^t}{\partial z \partial z} = 0$$

$$H_\phi = -\frac{1}{p} \frac{\partial A_z^t}{\partial p} = -\frac{1}{p} \frac{\partial}{\partial p} \left[j \frac{\beta h I_e}{2} H_1^{(2)}(\beta p) \sin \phi \right]$$

$$E_z = -j \frac{1}{\omega \mu \epsilon} \frac{(\beta^2 + p^2)}{p} A_z^t = -j \frac{\beta^2}{\omega \mu \epsilon} A_z^t$$

$$= j \frac{\beta^2}{\omega \mu \epsilon} \left[-j \frac{\beta h I_e}{2} \right] H_1^{(2)}(\beta p) \sin \phi$$

$$= j \frac{\beta^2}{\omega \mu \epsilon} \left[-j \frac{\beta h I_e}{2} \right] H_1^{(2)}(\beta p) \sin \phi$$

$$E_z = -\frac{\beta^2 h I_e}{2} H_1^{(2)}(\beta p) \sin \phi$$

$$H_\phi = j \frac{\beta h I_e}{2} \sin \phi \left[\beta H_0^{(2)}(\beta p) - \frac{1}{p} H_1^{(2)}(\beta p) \right]$$

$$H_z = 0$$

$$E_z = 0 = -\frac{\beta^2 h I_e}{2} H_1^{(2)}(\beta p) \sin \phi = 0 \Rightarrow \phi = 90^\circ, 180^\circ.$$

11.6 Since a magnetic line source above a PMC is the dual of the solution of Problem 11.5 (electric line source above PEC), using duality

$$(a) F_z^{(1)} = -j \frac{\epsilon I_m}{4} H_0^{(2)} (\beta \sqrt{x^2+(y-h)^2}) = F_z^{(4)}(x, y-h)$$

$$(b) \lim_{h \rightarrow 0} F_z^t = -j \frac{\epsilon B^2 h I_m}{2} H_1^{(2)}(\beta p) \sin \phi$$

$$(c) H_\phi = H_\phi = 0$$

$$H_z = -\frac{\beta^2 h I_m}{2} H_1^{(2)}(\beta p) \sin \phi \quad H_z = 0 = \sin \phi = 0 \Rightarrow \phi = 0^\circ, 180^\circ$$

11.7

a. For a current element located at $x=0, y=0$, the potential is given by

$$A_z^{(0)} = -j \frac{\mu I}{4} H_0^{(2)}(\beta p)$$

For a current element located at $x=x', y=y'$, the potential is given by

$$A_z = -j \frac{\mu I}{4} H_0^{(2)}(\beta \sqrt{(x-x')^2 + (y-y')^2})$$

Thus for $x' = \pm s/2, y' = 0$ and with $+I$ current

$$A_z^{(1)} = -j \frac{\mu I}{4} H_0^{(2)}\left(\beta \sqrt{(x-x')^2 + y^2}\right) = A_z^{(0)}(x - \frac{s}{2}, y)$$

while for $x' = -s/2, y' = 0$ and with $-I$ current

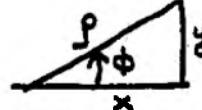
$$A_z^{(2)} = +j \frac{\mu I}{4} H_0^{(2)}\left(\beta \sqrt{(x+x')^2 + y^2}\right) = -A_z^{(0)}(x + \frac{s}{2}, y)$$

Therefore the total is equal to

$$A_z^t = A_z^{(0)}(x - \frac{s}{2}, y) - A_z^{(2)}(x + \frac{s}{2}, y) = -[A_z^{(0)}(x + \frac{s}{2}, y) - A_z^{(0)}(x - \frac{s}{2}, y)]$$

$$\begin{aligned} b. \lim_{s \rightarrow 0} A_z^t &= - \lim_{s \rightarrow 0} [A_z^{(0)}(x + \frac{s}{2}, y) - A_z^{(0)}(x - \frac{s}{2}, y)] = -s \lim_{s \rightarrow 0} \left[\frac{A_z^{(0)}(x + \frac{s}{2}, y) - A_z^{(0)}(x - \frac{s}{2}, y)}{s} \right] \\ &= -s \frac{\partial A_z^{(0)}}{\partial x} = -s \frac{\partial}{\partial x} \left\{ -j \frac{\mu I}{4} H_0^{(2)}(\beta p) \right\} = +j \frac{\mu s I}{4} \frac{\partial}{\partial x} H_0^{(2)}(\beta p) \end{aligned}$$

$$\frac{\partial}{\partial x} = \frac{\partial}{\partial p} \frac{\partial p}{\partial x} = \frac{\partial p}{\partial x} \frac{\partial}{\partial p} \quad \text{where } p = \sqrt{x^2 + y^2}$$



$$\frac{\partial p}{\partial x} = \frac{1}{2} (x^2 + y^2)^{-1/2} (2x) = \frac{x}{\sqrt{x^2 + y^2}} = \frac{x}{p} = \cos \phi$$

$$\frac{\partial}{\partial p} H_0^{(2)}(\beta p) = -\beta H_1^{(2)}(\beta p)$$

Therefore

$$\lim_{s \rightarrow 0} A_z^t = j \frac{\mu s I}{4} \cos \phi [-\beta H_1^{(2)}(\beta p)] = -j \frac{\mu \beta s I}{4} H_1^{(2)}(\beta p) \cos \phi$$

c. For TM²

$$E_p = -j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 A_z^t}{\partial p \partial z} = 0$$

$$E_\phi = -j \frac{1}{\omega \mu \epsilon} \frac{1}{p} \frac{\partial^2 A_z^t}{\partial p \partial t} = 0$$

$$\begin{aligned} E_z &= -j \frac{1}{\omega \mu \epsilon} \left(\frac{\partial^2}{\partial z^2} + p^2 \right) A_z^t = -j \frac{\beta^2}{\omega \mu \epsilon} A_z^t \\ &= -\frac{\beta^2 s I}{4 \omega \epsilon} H_1^{(2)}(\beta p) \cos \phi = -\frac{\beta^2 s I}{4} H_1^{(2)}(\beta p) \cos \phi \end{aligned}$$

$$H_p = \frac{1}{\mu p} \frac{\partial A_z^t}{\partial p} = +j \frac{\beta s I}{4 p} H_1^{(2)}(\beta p) \sin \phi$$

$$\begin{aligned} H_\phi &= -\frac{1}{\mu} \frac{\partial A_z^t}{\partial p} = +j \frac{\beta s I}{4} \cos \phi \frac{\partial}{\partial p} H_1^{(2)}(\beta p) \\ &= j \frac{\beta s I}{4} \cos \phi \left[\beta H_0^{(2)}(\beta p) - \frac{1}{p} H_1^{(2)}(\beta p) \right] \\ &= j \frac{\beta s I}{4 p} \cos \phi \left[(\beta p) H_0^{(2)}(\beta p) - H_1^{(2)}(\beta p) \right] \end{aligned}$$

$$H_z = 0$$

11.8 Two magnetic line sources displaced symmetrically along the x-axis a distance s apart, as shown in Figure P11-7, is the dual of the solution of Problem 11.7 (2 electric line sources displaced symmetrically along the x-axis a distance s apart). Therefore using the solution of Problem 11.7 and duality, we can write:

$$(a) F_z^t = F_z^{(1)}(x - \frac{s}{2}, y) - F_z^{(2)}(x + \frac{s}{2}, y) = -[F_z^{(1)}(x + \frac{s}{2}, y) - F_z^{(2)}(x - \frac{s}{2}, y)]$$

$$(b) \lim_{s \rightarrow 0} F_z^t = -j \frac{\beta s I_m}{4} H_1^{(2)}(\beta p) \cos \phi$$

$$(c) E_p = -j \frac{\beta^2 s I_m}{4p} H_1^{(2)}(\beta p) \sin \phi$$

$$E_\phi = -j \frac{\beta s I_m \cos \phi}{4p} [H_0^{(2)}(\beta p) - H_1^{(2)}(\beta p)]$$

$$E_z = 0$$

$$H_p = 0$$

$$H_\phi = 0$$

$$H_z = -\frac{\beta^2 s I_m}{4\pi} H_1^{(2)}(\beta p) \cos \phi$$

11.9

The solution of this problem is identical to the solution of the equivalent problem of Problem 11.5 (an electric line source placed a height h above a PEC ground plane). therefore using the solution of Problem 11.5 with $h \rightarrow s/2$, we can write:

$$(a) A_z^{(1)} = -j \frac{h I e}{4} H_0^{(2)}(\beta \sqrt{x^2 + (y - s/2)^2}) = A_z^{(1)}(x, y - s/2)$$

$$A_z^t = A_z^{(1)}(x, y - s/2) - A_z^{(2)}(x, y + s/2) = -[A_z^{(1)}(x, y + s/2) - A_z^{(1)}(x, y - s/2)]$$

$$(b) A_z^t = -j \frac{\beta h s I e}{4} H_1^{(2)}(\beta p) \sin \phi$$

$$(c) E_p = 0$$

$$H_p = -j \frac{\beta s I e}{4p} H_1^{(2)}(\beta p) \omega s \phi$$

$$E_\phi = 0$$

$$H_\phi = +j \frac{\beta s I e \sin \phi}{4} \frac{2}{\beta p} [H_2^{(2)}(\beta p)]$$

$$E_z = -\frac{\beta^2 \eta s I e}{4} H_1^{(2)}(\beta p) \sin \phi$$

$$= j \frac{\beta s I e \sin \phi}{4} [\beta H_0^{(2)}(\beta p) - \frac{1}{\beta} H_1^{(2)}(\beta p)]$$

$$H_z = 0$$

11.10 The solution of this problem can be obtained from the solution of Problem 11.9 by using duality (since the two problems are the duals of each other). Therefore, using duality as outlined in Table 7-2 and the solution of Problem 11.9, we can write that:

$$(a) F_z^{(1)} = -j \frac{\varepsilon \text{Im}}{4} H_0^{(2)}(\beta \sqrt{x^2 + (y-s/2)^2}) = z^{(1)}(x, y-s/2)$$

$$F_z^t = F_z^{(1)}(x, y-s/2) - F_z^{(2)}(x, y+s/2) = -[F_z^{(1)}(x, y+s/2) - F_z^{(2)}(x, y-s/2)]$$

$$(b) F_z^t = -j \frac{\beta \varepsilon \text{Im}}{4} H_1^{(2)}(\beta p) \sin\phi$$

$$(c) E_\phi = j \frac{\beta \varepsilon \text{Im}}{4p} H_1^{(2)}(\beta p) \cos\phi$$

$$H_\phi = 0$$

$$H_\phi = 0$$

$$E_\phi = -j \frac{\beta \varepsilon \text{Im}}{4} \sin\phi [\beta H_0^{(2)}(\beta p) - \frac{1}{p} H_1^{(2)}(\beta p)]$$

$$E_z = 0$$

$$H_z = -\frac{\beta^2 \varepsilon \text{Im}}{4p} H_1^{(2)}(\beta p) \sin\phi$$

11.11 a. Using the results from the solution of Problem 11.7, we can consider the two line sources along the negative x axis as one set similar to the one of Problem 11.7 and the two line sources along the positive x axis as another set similar to the one of Problem 11.7. Thus for each set

$$A_2^t(\text{left set}) = A_2^{tr} = \frac{\mu \beta I s_1}{4j} H_1^{(2)}(\beta p) \cos\phi = -A_2^o(s_1)$$

$$A_2^t(\text{right set}) = A_2^{tr} = \frac{\mu \beta I s_1}{4j} H_1^{(2)}(\beta p) \cos\phi = A_2^o(s_1)$$

Since the two sets are displaced relative to each other along the x axis by a separation s_2 , we can write that as $s_2 \rightarrow 0$ the total potential for all four sources (2 sets) can be written by treating each set as a source and using the results of Problem 11.7 as

$$A_2^t = A_2^{tr} + A_2^{tr} = A_2^{(o)}\left(x - \frac{s_2}{2}, y\right) - A_2^{(o)}\left(x + \frac{s_2}{2}, y\right) = -\left[A_2^{(o)}\left(x + \frac{s_2}{2}, y\right) - A_2^{(o)}\left(x - \frac{s_2}{2}, y\right)\right]$$

$$\lim_{s_2 \rightarrow 0} A_2^t = s_2 \lim_{s_2 \rightarrow 0} \left\{ \frac{A_2^t}{s_2} \right\} = -s_2 \lim_{s_2 \rightarrow 0} \left\{ \frac{A_2^{(o)}\left(x + \frac{s_2}{2}, y\right) - A_2^{(o)}\left(x - \frac{s_2}{2}, y\right)}{s_2} \right\} = -s_2 \frac{\partial A_2^{(o)}}{\partial x}(s_1)$$

$$= -s_2 \frac{\partial}{\partial x} \left\{ \frac{\mu \beta I s_1}{4j} H_1^{(2)}(\beta p) \cos(\phi) \right\} = -\frac{\mu \beta s_1 s_2 I}{4j} \frac{\partial}{\partial x} \left\{ H_1^{(2)}(\beta p) \cos\phi \right\}$$

$$\lim_{s_2 \rightarrow 0} A_2^t = -\frac{\mu \beta s_1 s_2 I}{4j} \left\{ H_1^{(2)}(\beta p) \frac{\partial}{\partial x} \cos\phi + \cos\phi \frac{\partial}{\partial x} H_1^{(2)}(\beta p) \right\}$$

$$\frac{\partial}{\partial x} \cos\phi = \frac{\partial}{\partial x} \left\{ \frac{x}{\sqrt{x^2+y^2}} \right\} = \frac{\partial}{\partial x} \left\{ x(x^2+y^2)^{-1/2} \right\} = (x^2+y^2)^{-1/2} + x(-\frac{1}{2})(x^2+y^2)^{-3/2}(2x)$$

$$= \frac{1}{\sqrt{x^2+y^2}} - \frac{x^2}{(x^2+y^2)^{3/2}} = \frac{1}{p} - \frac{(\rho \cos\phi)^2}{p^3} = \frac{1}{p} - \frac{\cos^2\phi}{p} = \frac{1-\cos^2\phi}{p} = \frac{\sin^2\phi}{p}$$

$$\frac{\partial}{\partial x} H_1^{(2)}(\beta p) = \frac{\partial p}{\partial x} \frac{\partial}{\partial p} H_1^{(2)}(\beta p) = \cos\phi \frac{\partial}{\partial p} H_1^{(2)}(\beta p) = \cos\phi \left[-\beta H_2^{(2)}(\beta p) + \frac{1}{p} H_1^{(2)}(\beta p) \right]$$

$$\lim_{s_2 \rightarrow 0} A_2^t = -\frac{\mu \beta s_1 s_2 I}{4j} \left\{ H_1^{(2)}(\beta p) \frac{\sin^2\phi}{p} + \cos^2\phi \left[-\beta H_2^{(2)}(\beta p) + \frac{1}{p} H_1^{(2)}(\beta p) \right] \right\}$$

$$= -\frac{\mu \beta s_1 s_2 I}{4j} \left\{ \frac{H_1^{(2)}(\beta p)}{p} - \beta \cos^2\phi H_2^{(2)}(\beta p) \right\}$$

b. For TM²

$$E_p = -j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 A_2^t}{\partial p \partial z}$$

$$E_\phi = -j \frac{1}{\omega \mu \epsilon} \frac{1}{p} \frac{\partial^2 A_2^t}{\partial p \partial z}$$

$$E_z = -j \frac{1}{\omega \mu \epsilon} \left(\frac{\partial^2}{\partial z^2} + p^2 \right) A_2^t$$

$$H_p = \frac{1}{\omega \mu} \frac{\partial A_2^t}{\partial z}$$

$$H_\phi = -\frac{1}{j} \frac{\partial A_2^t}{\partial p}$$

$$H_z = 0$$

11.12 a. Using the results from the solution of Problem 11.7, we can consider the two line sources along the negative y axis as one set similar to the one of Problem 11.7 and the two line sources along the positive y axis as another set similar to the one of Problem 11.7. Thus for each set

$$A_z^t(\text{upper}) = A_z^{tu} = \frac{\mu \beta I s}{4j} H_1^{(2)}(\beta p) \cos\phi = A_z^{(0)}(s)$$

$$A_z^t(\text{lower}) = A_z^{tl} = -\frac{\mu \beta I s}{4j} H_1^{(2)}(\beta p) \cos\phi = -A_z^{(0)}(s)$$

Since the two sets are displaced relative to each other along the y axis by a separation h, we can write that as $h \rightarrow 0$ the total potential for all four sources(2 sets) can be written by treating each set as a source and use a procedure similar to Problem 11.7. Therefore

$$A_z^t = A_z^{tu} + A_z^{tl} = -A_z^{(0)}(x, y + \frac{h}{2}) + A_z^{(0)}(x, y - \frac{h}{2}) = -[A_z^{(0)}(x, y + \frac{h}{2}) - A_z^{(0)}(x, y - \frac{h}{2})]$$

$$\lim_{h \rightarrow 0} A_z^t = h \lim_{h \rightarrow 0} \left(\frac{A_z^t}{h} \right) = h \lim_{h \rightarrow 0} \left\{ \frac{A_z^{(0)}(x, y + \frac{h}{2}) - A_z^{(0)}(x, y - \frac{h}{2})}{h} \right\} = -h \frac{\partial A_z^{(0)}(s)}{\partial y}$$

$$= -h \frac{\partial}{\partial y} \left\{ \frac{\mu \beta I s}{4j} H_1^{(2)}(\beta p) \cos\phi \right\} = -\frac{\mu \beta h s I}{4j} \frac{\partial}{\partial y} \left\{ H_1^{(2)}(\beta p) \cos\phi \right\}$$

$$\lim_{h \rightarrow 0} A_z^t = -\frac{\mu \beta h s I}{4j} \left\{ H_1^{(2)}(\beta p) \frac{\partial}{\partial y} \cos\phi + \cos\phi \frac{\partial}{\partial y} H_1^{(2)}(\beta p) \right\}$$

$$\begin{aligned} \frac{\partial}{\partial y} \cos\phi &= \frac{\partial}{\partial y} \left[\frac{x}{(x^2+y^2)^{1/2}} \right] = \frac{\partial}{\partial y} \left[x(x^2+y^2)^{-1/2} \right] = x(-\frac{1}{2})(x^2+y^2)^{-3/2} (2y) = -\frac{x^2 y}{(x^2+y^2)^{3/2}} \\ &= -\frac{1}{\sqrt{x^2+y^2}} \frac{x}{\sqrt{x^2+y^2}} \frac{y}{\sqrt{x^2+y^2}} = -\frac{1}{p} \cos\phi \sin\phi = -\frac{\cos\phi \sin\phi}{p} \end{aligned}$$

$$\frac{\partial}{\partial y} H_1^{(2)}(\beta p) = \frac{\partial p}{\partial y} \frac{\partial}{\partial p} H_1^{(2)}(\beta p) = \sin\phi \left[-\beta H_2^{(2)}(\beta p) + \frac{1}{p} H_1^{(2)}(\beta p) \right]$$

$$\begin{aligned} \lim_{h \rightarrow 0} A_z^t &= -\frac{\mu \beta h s I}{4j} \left\{ -\frac{H_1^{(2)}(\beta p)}{p} \cos\phi \sin\phi + \cos\phi \sin\phi \left[-\beta H_2^{(2)}(\beta p) + \frac{1}{p} H_1^{(2)}(\beta p) \right] \right\} \\ &= \frac{\mu \beta^2 h s I}{4j} H_2^{(2)}(\beta p) \cos\phi \sin\phi = \frac{\mu \beta^2 h s I}{8j} H_2^{(2)}(\beta p) \sin(2\phi) \end{aligned}$$

b. For TM^2

$$E_p = -j \frac{1}{\omega \mu \epsilon} \frac{\partial^2 A_z^t}{\partial p \partial z}$$

$$E_\phi = -j \frac{1}{\omega \mu \epsilon} \frac{1}{p} \frac{\partial^2 A_z^t}{\partial \phi \partial z}$$

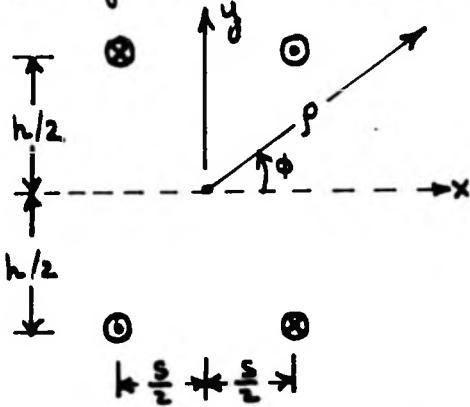
$$E_z = -j \frac{1}{\omega \mu \epsilon} \left(\frac{\partial^2}{\partial z^2} + \beta^2 \right) A_z^t$$

$$H_p = \frac{1}{\mu \beta} \frac{\partial A_z^t}{\partial p}$$

$$H_\phi = -\frac{1}{\mu} \frac{\partial A_z^t}{\partial \phi}$$

$$H_z = 0$$

11.13 The solution to this problem can be obtained using image theory. Since below the interface the fields are zero, the solution is only valid above the interface. The equivalent model for the geometry of Problem 11.13 is shown here where we replace the conductor with two image sources. Then the solution to this equivalent is identical to that of Problem 11.12 and it is valid on and above the interface; below the interface the fields are zero.



11.14 The solution to this problem can be obtained using image theory. Since the fields within the conductor are zero, the solution using image theory is only valid within the 90° sector. The equivalent [Three Images are indicated outside the 90° corner reflector] model for the geometry of Problem 11.14 is shown here where we replace the conductor with three image sources. Then the solution to this equivalent is identical to that of Problem 11.12, and it is valid within the 90° sector; within the conductor the fields are zero. The only substitution that needs to be made is to replace s and h in the solution of Problem 11.12 by $1.414 s$. Thus we can write the total potential as

$$A_z^t = \frac{(j\zeta)^2 \beta^2 s^2 I}{j\beta} H_2^{(2)}(\beta\rho) \sin(2\phi) = \frac{(\beta s)^2 I}{j^4} H_2^{(2)}(\beta\rho) \sin(2\phi)$$

b. For TM²

$$\left. \begin{aligned} E_p &= \frac{1}{j\omega \epsilon_0 \mu_0} \frac{\partial A_z^t}{\partial z} = 0 & H_p &= \frac{1}{\mu_0} \frac{\partial A_z^t}{\partial \phi} = \frac{(\beta s)^2 I}{j^2 \mu_0} H_2^{(2)}(\beta\rho) \cos(2\phi) \\ E_\phi &= \frac{1}{j\omega \epsilon_0 \mu_0} \frac{\partial A_z^t}{\partial \phi} = 0 & H_z &= 0 \\ E_z &= \frac{1}{j\omega \epsilon_0 \mu_0} \left(\frac{\partial^2 A_z^t}{\partial z^2} + \beta^2 \right) = -j \frac{\beta^2 (\beta s)^2 I}{4\omega \mu_0} H_2^{(2)}(\beta\rho) \sin(2\phi) \\ H_\phi &= -\frac{1}{\mu_0} \frac{\partial A_z^t}{\partial \rho} = \frac{(\beta s)^2 I}{j^4 \mu_0} \sin(2\phi) \left[\beta H_1^{(2)}(\beta\rho) - \frac{2}{\beta} H_2^{(2)}(\beta\rho) \right] \end{aligned} \right\} 0 \leq \phi \leq 90^\circ$$

c. $E_z = 0 = -j \frac{\beta^2 (\beta s)^2 I}{4\omega \mu_0} H_2^{(2)}(\beta\rho) \sin(2\phi) \Rightarrow \sin(2\phi) = 0 \Rightarrow 2\phi = \sin^{-1}(0) = 0, \pi$

$$\phi = 0, \frac{\pi}{2}$$

[11.15] The parallel slots can be replaced by equivalent magnetic current densities. Using the geometry of Figure P11-15, we represent each of the slots with magnetic current densities of

$$x=s/2: M_1 = -2\hat{n} \times E_{az} = -2\hat{a}_y \times (-\hat{a}_x E_0) = -\hat{a}_{22} E_0$$

$$x=-s/2: M_2 = -2\hat{n} \times E_{az} = -2\hat{a}_y \times (\hat{a}_x E_0) = +\hat{a}_{12} 2E_0$$

Each slot can then be modeled with magnetic line sources (with TE^z modes) with equivalent magnetic currents of

$$x=s/2: I_{m1} = -2E_0$$

$$x=-s/2: I_{m2} = 2E_0$$

- (a) The equivalent problem is then identical to Problem 11.8. Therefore, based on the solution of Problem 11.8, we can write

$$F_z^t = F_z^{(1)}(x-s/2, y) - F_z^{(2)}(x+s/2, y) = -[F_z^{(1)}(x+s/2, y) - F_z^{(2)}(x-s/2, y)]$$

$$(b) \lim_{s \rightarrow 0} F_z^t = -j \frac{\varepsilon \beta s}{4} (-2E_0) H_1^{(2)}(\beta p) \cos \phi = j \frac{\varepsilon \beta s E_0}{2} H_1^{(2)}(\beta p) \cos \phi$$

$$(c) E_p = +j \frac{\beta^2 s E_0}{2p} H_1^{(2)}(\beta p) \cos \phi \quad H_p = 0$$

$$E_\phi = j \frac{\beta s E_0}{2p} \cos \phi [H_0^{(2)}(\beta p) - H_1^{(2)}(\beta p)] \quad H_\phi = 0$$

$$E_z = 0 \quad H_z = \frac{\beta^2 s E_0}{2\eta} H_1^{(2)}(\beta p) \cos \phi$$

11.16

Based on the geometry of Figure 11-2 and (11-20a), we can write the total field above the ground plane as (for $p=p_0$)

$$E_z^t = jn I_e \sqrt{\frac{\beta}{2\pi}} \sin(\beta h \sin \phi) \frac{e^{-j\beta p_0}}{\sqrt{p_0}}$$

which vanishes when (for $p=p_0$)

$$\sin(\beta h \sin \phi) = 0 \Rightarrow \beta h \sin \phi = \sin^{-1}(0) = n\pi, \quad n=0, \pm 1, \dots$$

$$h = \frac{n\pi}{\beta \sin \phi} = \frac{n\pi}{(2\pi/\lambda) \sin \phi}$$

The smallest height for $h \neq 0$, $n=1$. Thus the smallest height

$$h = \frac{\pi \lambda}{2\pi \sin \phi} \Big|_{\phi=60^\circ} = \frac{\lambda}{2 \sin 60^\circ} \Big|_{\phi=60^\circ} = \frac{\lambda}{2(\sqrt{3}/2)} = \frac{\lambda}{\sqrt{3}}$$

$$h = 0.5774 \lambda$$

11.17 The normalized field for Problem 11.16 is represented by

$$E_{zn}^t = \sin(\beta h \sin\phi)$$

whose maximum value is 1 (or 0 dB). For the field to be -3dB from the maximum, its dimensionless value at that point is $1/\sqrt{2} = 0.70711$.

To determine the 2 smallest height that will result for the field to be 0.70711 (or -3 dB) at $\phi=60^\circ$, we set the normalized field to

$$\sin(\beta h \sin\phi) = 0.70711 \Rightarrow \beta h \sin\phi = \sin^{-1}(0.70711) = \frac{\pi}{4}, \frac{3\pi}{4}, \dots$$

$$\text{Therefore } h = \frac{n\pi/4}{\beta \sin\phi} \Big|_{\phi=60^\circ} = \frac{n\pi/4}{(2\pi/3)\sin\phi} \Big|_{\phi=60^\circ} = \frac{n\pi}{4 \sin\phi} \Big|_{\phi=60^\circ} = \frac{n\pi}{4 \left(\frac{\sqrt{3}}{2}\right)} = \frac{n\pi}{2\sqrt{3}}, n=1, 3, 5, \dots$$

$$n=1: h_1 = \frac{\pi}{2\sqrt{3}} = 0.57741$$

$$n=3: h_3 = \frac{3\pi}{2\sqrt{3}} = 1.73211$$

11.18

(a) From Example 11-2

$$\underline{H}^i = \hat{a}_z H_0 e^{j\beta(x \cos\phi + y \sin\phi)}$$

(b) The reflected magnetic field, in terms of the incident

magnetic field H_0 , can be written as [using Figure 11-4(c)]

$$\underline{H}^r = -\hat{a}_z \Gamma H_0 e^{-j\beta(x \cos\phi_r + y \sin\phi_r)}$$

(c) Using Snell's law of reflection ($\phi_i = \pi - \phi_r$), and that the tangential components of the magnetic field vanish on the PMC surface ($y=0$), we have that

$$\begin{aligned} H_0 e^{j\beta x \cos(\pi - \phi_r)} - \Gamma H_0 e^{j\beta \cos\phi_r} &= 0 \\ e^{-j\beta x \cos\phi_r} - \Gamma e^{j\beta \cos\phi_r} &= 0 \end{aligned}$$

$$1 - \Gamma = 0 \Rightarrow \boxed{\Gamma = +1} \text{ for a PMC surface.}$$

11.19 For normal incidence, the 3-D RCS of a flat plate is equal to
 $\sigma_{3-D} = 4\pi \left(\frac{A}{\lambda}\right)^2$ where A = area of the flat plate

(a) TE^X: The bistatic 3-D RCS at any observation angle is given by (11-44a), or

$$\sigma_{3-D} = 4\pi \left(\frac{A}{\lambda}\right)^2 \left\{ \cos^2 \theta_s \frac{\sin \left[\frac{\beta b}{2} (\sin \theta_s - \sin \theta_i) \right]}{\frac{\beta b}{2} (\sin \theta_s - \sin \theta_i)} \right\}$$

To find its maximum, we will assume the sinc function varies much faster than the $\cos^2 \theta_s$ function, which is usually the case, so that the $\cos^2 \theta_s$ is basically a constant at the maximum of the sinc.

The maximum value of the sinc function occurs when its argument is zero; in this case $\sin \theta_s = \sin \theta_i \Rightarrow \theta_s = \theta_i$ (specular direction). Therefore for the 3-D bistatic RCS to be reduced by 10 dB along the specular direction ($\theta_s = \theta_i$), then the cosine squared function must be equal to

$$\cos^2 \theta_s \Big|_{\theta_s = \theta_i} = \cos^2 \theta_i = 0.1 \quad (-10 \text{dB})$$

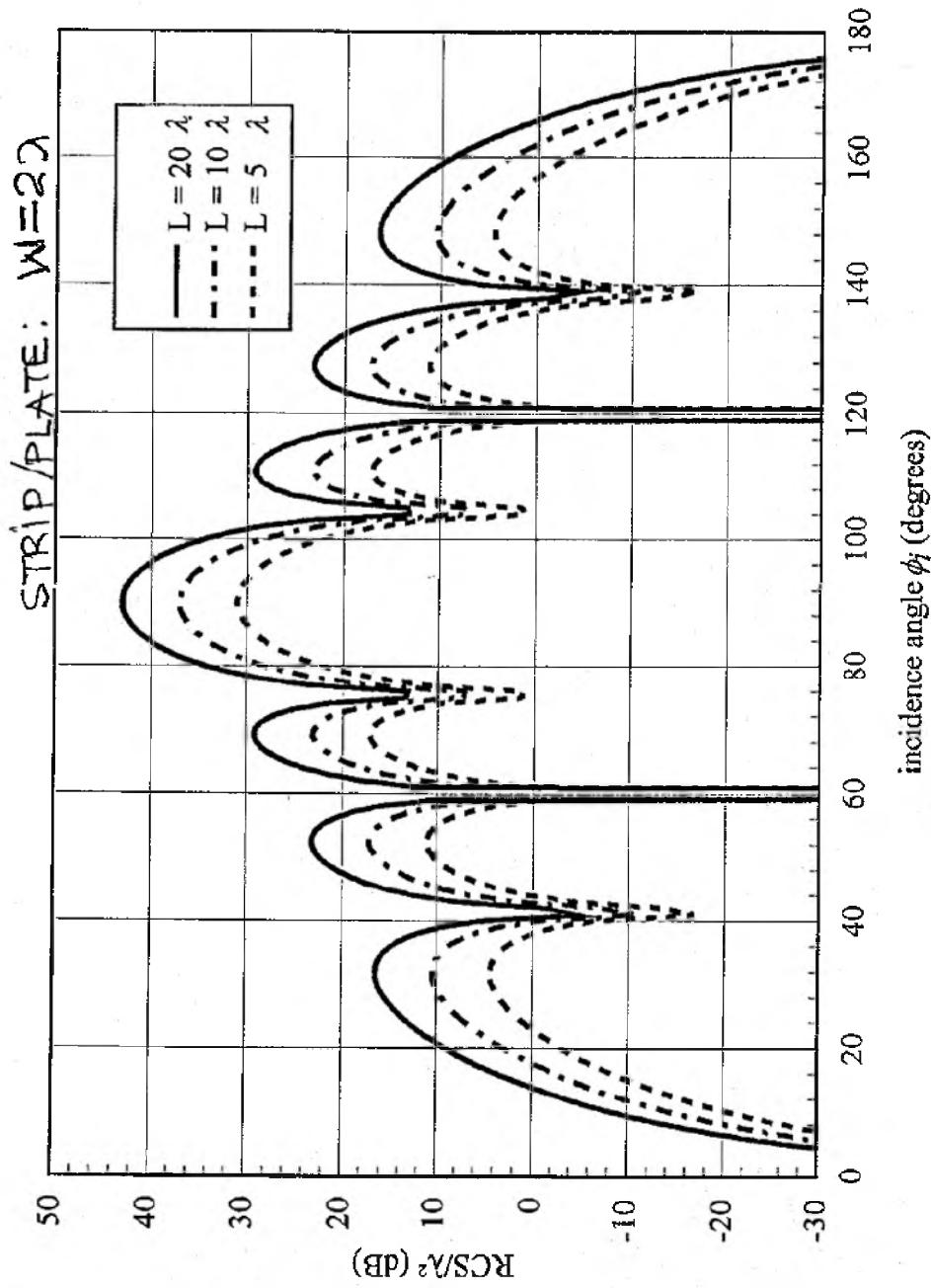
$$\theta_i = \cos^{-1}(0.1)^{1/2} = \cos^{-1}(0.3162) = \mp 1.565^\circ$$

(b) Similarly for the TM^X polarization, the 3-D RCS TM^X: is given (from Example 11-3) by (selecting $\phi_s = 90^\circ$)

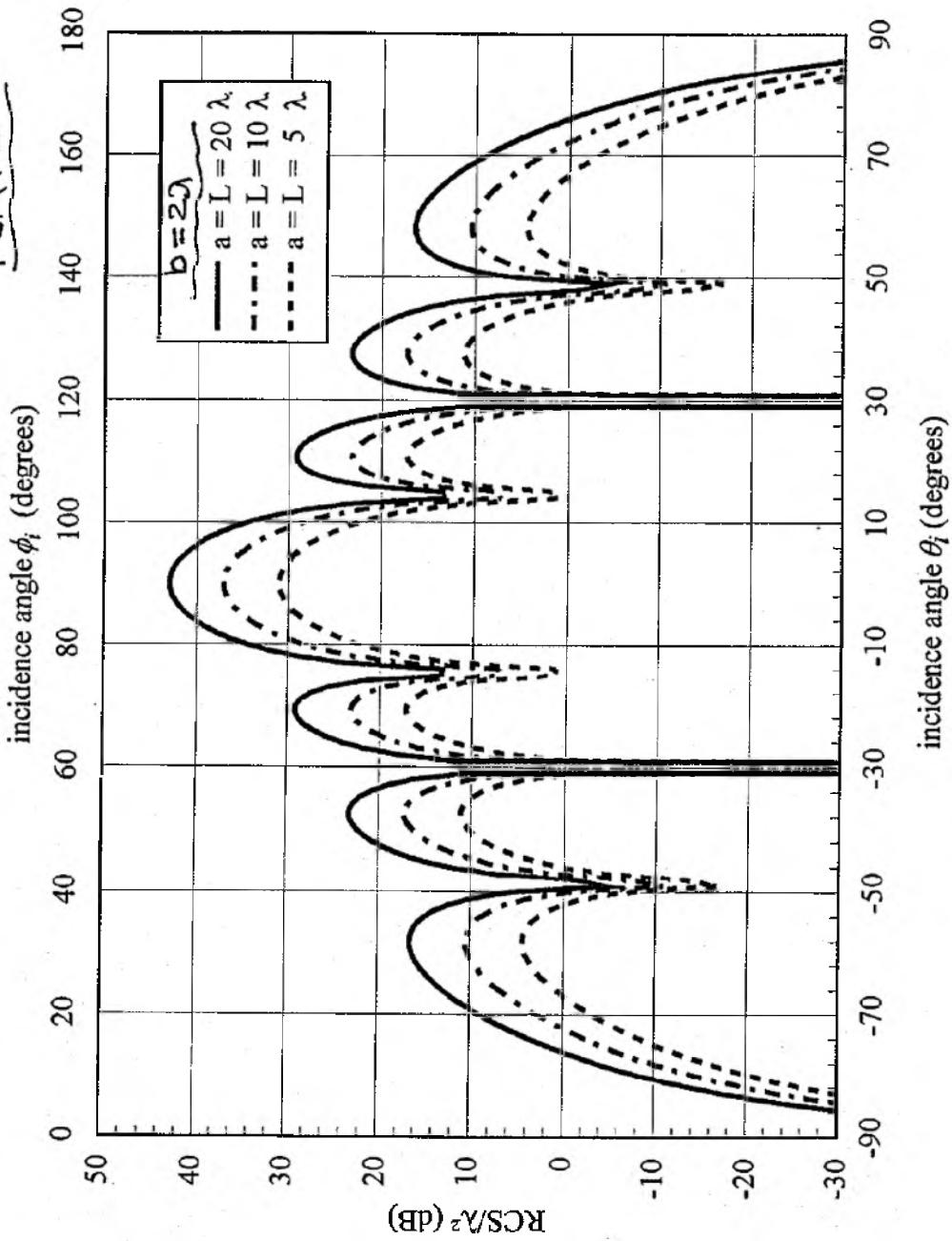
$$\sigma_{3-D} = 4\pi \left(\frac{A}{\lambda}\right)^2 \left[\cos^2 \theta_i \frac{\sin \left[\frac{\beta b}{2} (\sin \theta_s - \sin \theta_i) \right]}{\frac{\beta b}{2} (\sin \theta_s - \sin \theta_i)} \right]$$

which is identical to the one for the TE^X polarization, except that the cosine squared function is $\cos^2 \theta_i$ (instead of $\cos^2 \theta_s$ for TE^X). However, again, to maximize the sinc function, $\theta_s = \theta_i$. Therefore to reduce the maximum 3-D RCS toward the specular direction ($\theta_s = \theta_i$) by -10 dB, the incidence angle $\theta_i = \mp 1.565^\circ$, which is the same as for TE^X.

11.20



Normalized, monostatic radar cross section RCS/λ^2 versus incidence angle ϕ_i

PLATE

Normalized, monostatic radar cross section RCS/λ^2 versus incidence angle for strip ψ_i and plate θ_i

The maximum monostatic 3-D RCS of a flat plate is given by

$$\sigma_{3-D} = 4\pi \left(\frac{A}{\lambda}\right)^2 = 4\pi \frac{A^2}{\lambda^2}, \text{ where } A = \text{area of plate}$$

If the area of the plate is represented as

$$A = n\lambda^2$$

then the σ_{3-D} can be written as

$$\sigma_{3-D} = 4\pi \left(\frac{n^2 \lambda^4}{\lambda^2}\right) = 4\pi n^2 \lambda^2$$

so that its normalized expression be represented by

$$\frac{\sigma_{3-D}}{\lambda^2} = 4\pi n^2$$

- If it is desired to have the normalized 3-D RCS to be +20 dB (a factor of 100), then

$$\frac{\sigma_{3-D}}{\lambda^2} = 4\pi n^2 = 100 \Rightarrow n^2 = \frac{100}{4\pi} \Rightarrow n = \frac{10}{2\sqrt{\pi}} = 2.8209$$

Thus the flat plate must be $2.8209\lambda^2$ in order for its normalized maximum monostatic RCS to be +20 dB.

- At $f = 10 \text{ GHz} \Rightarrow \lambda = 3 \times 10^8 / 10 \times 10^9 = 0.03 \text{ meters}$

thus σ_{3-D} is equal to

$$\sigma_{3-D} = 4\pi \left(\frac{A}{\lambda}\right)^2 = 4\pi \left(7.9575\lambda^2\right)^2 = 4\pi (7.9575)(0.03)^2$$

$A = 2.8209\lambda^2$
 $\lambda = 0.03 \text{ m}$

$$\sigma_{3-D} = 0.09 \text{ m}^2$$

which normalized to 1 m² is equal to

$$\sigma_{3-D} = 0.09 = 10 \log_{10}(0.09) = -10.458 \text{ dBsm}$$

11.23 The field scattered by a two-dimensional scatterer when it is illuminated by a plane wave at normal incidence can be written as

$$\underline{E}^s(\rho, \phi, \psi) = \sqrt{\frac{2}{\pi \beta \rho}} e^{-j(\beta \rho - \frac{\pi}{4})} \underline{E}_2^s(\phi, \psi) \quad (1)$$

with the corresponding 2-D RCS expressed as

$$\sigma_{2-D} = \lim_{\rho \rightarrow \infty} \left\{ 2\pi \rho \frac{|\underline{E}^s|^2}{|\underline{E}^s|^2} \right\} = \frac{4}{\beta} |\underline{E}_2^s(\phi, \psi)|^2 = \frac{2}{\pi} |\underline{E}_2^s(\phi, \psi)|^2 \quad (2)$$

In a similar manner for a three-dimensional scatterer, the scattered field is in the same plane as the 2-D fields

$$\underline{E}^s(r, \phi, \psi) \sim \frac{e^{-j\beta r}}{r} \underline{E}_3^s(\phi, \psi) \quad (3)$$

with its corresponding RCS written as

$$\sigma_{3-D} = \lim_{r \rightarrow \infty} \left\{ 4\pi r^2 \frac{|\underline{E}^s|^2}{|\underline{E}^s|^2} \right\} = \frac{4\pi}{\beta^2} |\underline{E}_3^s(\phi, \psi)|^2 = \frac{2^2}{\pi} |\underline{E}_3^s(\phi, \psi)|^2 \quad (4)$$

Now let us assume that the 3-D scatterer is segmented into 2-D strips and the currents in the 3-D scatterer are identical to those of the 2-D when observations are made in a plane perpendicular to the scatterer. Then the scattered field can be written as a summation or integral; thus

$$\underline{E}^s = C \iint \underline{J}(r') \frac{e^{-j\beta r'}}{r'} ds' \approx C \frac{e^{-j\beta r}}{r} \iint \underline{J}(r') e^{j\beta \hat{a}_z r'} ds' \Rightarrow \underline{E}^s = \beta C \iint \underline{J}(r') e^{j\beta \hat{a}_z r'} ds' \quad (5)$$

If \underline{J} is independent of z (the axial coordinate), the integration with respect to z can be carried out and we can write for a scatterer of length L as

$$\underline{E}_3^s = \beta C \iint \underline{J}(r') e^{j\beta \hat{a}_z r'} ds' = \beta L C \int \underline{J}(p') e^{j\beta \hat{a}_z p'} dp' \quad (6)$$

If the p integration is carried out in

$$\begin{aligned} \underline{E}^s &= C \iint \underline{J}(r') \frac{e^{-j\beta r'}}{r'} ds' = -j\pi C \int \underline{J}(p') H_0^{(2)}(\beta(p-p')) dp' \\ &\approx -j\pi C e^{-j(\beta r - \frac{\pi}{4})} \sqrt{\frac{2}{\pi \beta r}} \underline{J}(p') e^{+j\beta \hat{a}_z p' L} dp' \end{aligned} \quad (7)$$

which when compared with (1) we can write

$$\underline{E}_2^s(\phi, \psi) = -j\pi C \int \underline{J}(p') e^{+j\beta \hat{a}_z p' L} dp' \quad (8)$$

By comparing (6) and (8)

$$\underline{E}_3^s = j \frac{\beta L}{\pi} \underline{E}_2^s \quad (9)$$

Now using (9) we can write (4) as

$$\sigma_{3-D} = \frac{2^2}{\pi} \frac{(\beta L)^2}{\pi^2} |\underline{E}_2^s|^2 = \frac{4L^2}{\pi} |\underline{E}_2^s|^2 = \frac{4L^2}{\pi} \left[\frac{2}{\pi} |\underline{E}_2^s|^2 \right] = \frac{2L^2}{\pi} \sigma_{2-D}$$

11.24

$$\text{a. } \underline{E}^i = E_0 (\hat{a}_y \cos \theta_i + \hat{a}_z \sin \theta_i) e^{-j\beta r_i}$$

$$\underline{E}^i = E_0 (\hat{a}_y \cos \theta_i + \hat{a}_z \sin \theta_i) = \\ \cdot e^{j\beta (\hat{a}_y \sin \theta_i - \hat{a}_z \cos \theta_i) \cdot (\hat{a}_x x + \hat{a}_y y + \hat{a}_z z)}$$

$$\underline{E}^i = E_0 (\hat{a}_y \cos \theta_i + \hat{a}_z \sin \theta_i) e^{-j\beta (y \sin \theta_i - z \cos \theta_i)}$$

$$\underline{H}^i = \hat{a}_x \frac{E_0}{\eta} e^{-j\beta (y \sin \theta_i - z \cos \theta_i)}$$

$$\underline{J}_s = 2\hat{n} \times \underline{H}^i \Big|_{\substack{x=0 \\ y=y}} = 2\hat{a}_z \times \hat{a}_x \frac{E_0}{\eta} e^{-j\beta y' \sin \theta_i} = \hat{a}_y \frac{2E_0}{\eta} e^{-j\beta y' \sin \theta_i}$$

$$J_x = J_z = 0, \quad J_y = \frac{2E_0}{\eta} e^{-j\beta y' \sin \theta_i}$$

$$\text{b. } N_\theta = \iint_S [J_x \cos \theta_s \cos \phi_s + J_y \cos \theta_s \sin \phi_s - J_z \sin \theta_s] e^{j\beta r' \cos \psi} ds$$

$$N_\phi = \iint_S [-J_x \sin \theta_s + J_y \cos \phi_s] e^{j\beta r' \cos \psi} ds'$$

$$r' \cos \psi = p' \sin \theta_s \cos(\phi_s - \phi) = x' \sin \theta_s \cos \phi_s + y' \sin \theta_s \sin \phi_s \quad [\text{see (6-132a)}]$$

$$ds' = p' dp' d\phi' \quad [\text{see (6-132b)}]$$

Therefore

$$N_\theta = \iint_S J_y \cos \theta_s \sin \phi_s e^{j\beta [x' \sin \theta_s \cos \phi_s + y' \sin \theta_s \sin \phi_s]} p' dp' d\phi'$$

$$N_\phi = \frac{2E_0 \cos \theta_s \sin \phi_s}{\eta} \int_0^{\pi} \int_0^{\alpha} e^{j\beta [x' \sin \theta_s \cos \phi_s + y' (\sin \theta_s \sin \phi_s - \sin \theta_i)]} p' dp' d\phi'$$

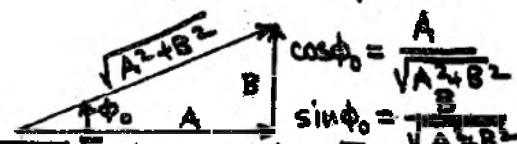
$$\begin{cases} x' = p' \cos \phi' \\ y' = p' \sin \phi' \end{cases}$$

$$\text{Thus } x' \sin \theta_s \cos \phi_s + y' (\sin \theta_s \sin \phi_s - \sin \theta_i) = p' \underbrace{\{\sin \theta_s \cos \phi_s \cos \phi' + (\sin \theta_s \sin \phi_s - \sin \theta_i) \sin \phi'\}}_B \\ = p' [A \cos \phi' + B \sin \phi']$$

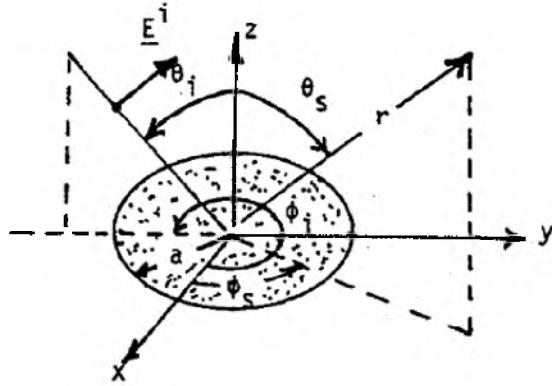
$$x' \sin \theta_s \cos \phi_s + y' (\sin \theta_s \sin \phi_s - \sin \theta_i) = p' \sqrt{A^2 + B^2} \left[\frac{A}{\sqrt{A^2 + B^2}} \cos \phi' + \frac{B}{\sqrt{A^2 + B^2}} \sin \phi' \right]$$

$$A = \sin \theta_s \cos \phi_s$$

$$B = \sin \theta_s \sin \phi_s - \sin \theta_i$$



$$x' \sin \theta_s \cos \phi_s + y' (\sin \theta_s \sin \phi_s - \sin \theta_i) = p' \sqrt{A^2 + B^2} [\cos \phi' \cos \phi_0 + \sin \phi' \sin \phi_0] \\ = p' \sqrt{A^2 + B^2} \cos(\phi' - \phi_0) \quad \text{cont'd.}$$



$$11.24 \text{ cont'd.} \quad N_\theta = \frac{2E_0}{\eta} \cos \theta_s \sin \phi_s \int_0^a \left\{ \int_{-\pi}^{\pi} e^{j\beta p' \sqrt{A^2+B^2} \cos(\phi'-\phi_s)} d\phi' \right\} p' dp'$$

$$N_\phi = \frac{2E_0}{\eta} \cos \theta_s \sin \phi_s \int_0^a [I] p' dp'$$

where $I = \int_{-\pi}^{\pi} e^{-j\beta \sqrt{A^2+B^2} p' \cos(\phi'-\phi_s)} d\phi'$

The function $e^{j\beta \sqrt{A^2+B^2} p' \cos(\phi'-\phi_s)}$ is a periodic function of ϕ' with a period of 2π . Therefore the integration over one period is the same no matter where the integration begins. That is

$$\text{Thus } \int_{-\pi}^{2\pi+\phi_s} e^{j\beta \sqrt{A^2+B^2} p' \cos(\phi'-\phi_s)} d\phi' = \int_{-\pi}^{2\pi} e^{j\beta \sqrt{A^2+B^2} p' \cos(\phi'-\phi_s)} d\phi' = 2\pi J_0(\beta \sqrt{A^2+B^2} p')$$

$$N_\theta = \frac{2E_0}{\eta} \cos \theta_s \sin \phi_s \int_0^a 2\pi J_0(\beta \sqrt{A^2+B^2} p') p' dp' = \frac{2\pi E_0}{\eta} \cos \theta_s \sin \phi_s \int_0^a J_0(\beta \sqrt{A^2+B^2} p') p' dp'$$

$$\text{Let } z = \beta \sqrt{A^2+B^2} p' \Rightarrow dz = \beta \sqrt{A^2+B^2} dp'$$

$$p' = z/\beta \sqrt{A^2+B^2} \Rightarrow dp' = dz/\beta \sqrt{A^2+B^2}$$

$$N_\theta = \frac{4\pi E_0}{\eta} \cos \theta_s \sin \phi_s \int_0^{\beta \sqrt{A^2+B^2} a} \frac{J_0(z) z dz}{(\beta \sqrt{A^2+B^2})^2} = \frac{4\pi E_0}{\eta (\beta \sqrt{A^2+B^2})^2} \cos \theta_s \sin \phi_s \int_0^{\beta \sqrt{A^2+B^2} a} z J_0(z) dz$$

$$\text{Since } \int_0^\gamma z J_0(z) dz = \gamma J_1(\gamma)$$

then

$$N_\theta = \frac{4\pi E_0}{\eta} \cos \theta_s \sin \phi_s \frac{\beta \sqrt{A^2+B^2} a}{(\beta \sqrt{A^2+B^2})} J_1(\beta a \sqrt{A^2+B^2}) = \frac{4\pi a^2}{\eta} \cos \theta_s \sin \phi_s \left[\frac{J_1(\beta a \sqrt{A^2+B^2})}{\beta a \sqrt{A^2+B^2}} \right]$$

$$\text{Similarly } N_\phi = \iint_S J_y \cos \phi_s e^{j\beta p' \sin \theta_s \cos(\phi_s - \phi)} p' dp' d\phi' = \frac{4\pi a^2}{\eta} E_0 \cos \phi_s \left[\frac{J_1(\beta a \sqrt{A^2+B^2})}{\beta a \sqrt{A^2+B^2}} \right]$$

$$\text{Thus } E_\theta^s = -j \frac{\rho e^{-j\beta r}}{4\pi r} \left[k_\theta + \eta N_\theta \right] = -j \eta \frac{\rho e^{-j\beta r}}{4\pi r} N_\theta = -j \frac{\beta a^2 E_0 e^{-j\beta r}}{r} \cos \theta_s \sin \phi_s J_1(\beta a \sqrt{A^2+B^2})$$

$$E_\phi^s = j \frac{\rho e^{-j\beta r}}{4\pi r} \left[k_\phi - \eta N_\phi \right] = -j \eta \frac{\rho e^{-j\beta r}}{4\pi r} N_\phi = -j \frac{\beta a^2 E_0 e^{-j\beta r}}{r} \cos \phi_s J_1(\beta a \sqrt{A^2+B^2})$$

$$\text{c. } |E|^2 = \sqrt{|E_\theta^s|^2 + |E_\phi^s|^2} = E_0 \frac{\rho a^2}{r} \sqrt{\cos^2 \theta_s \sin^2 \phi_s + \cos^2 \phi_s} \left| \frac{J_1(\beta a \sqrt{A^2+B^2})}{\beta a \sqrt{A^2+B^2}} \right|^2 = \frac{\beta a^2}{r} |F|^2$$

$$\xi = \lim_{r \rightarrow \infty} \left\{ 4\pi r^2 \frac{|E|^2}{|E|^2} \right\} = 4\pi r^2 \left(\frac{\beta a^2}{r} \right)^2 |F|^2 = 4\pi \left(\frac{\pi a^2}{\lambda} \right)^2 |F|^2$$

$$\text{where } |F| = \sqrt{\cos^2 \theta_s \sin^2 \phi_s + \cos^2 \phi_s} \left| \frac{J_1(\beta a \sqrt{A^2+B^2})}{\beta a \sqrt{A^2+B^2}} \right|$$

For monostatic observations $\phi_s = \pi/2, \theta_s = \theta_i$.

$$A = \sin \theta_s \cos \phi_s = 0$$

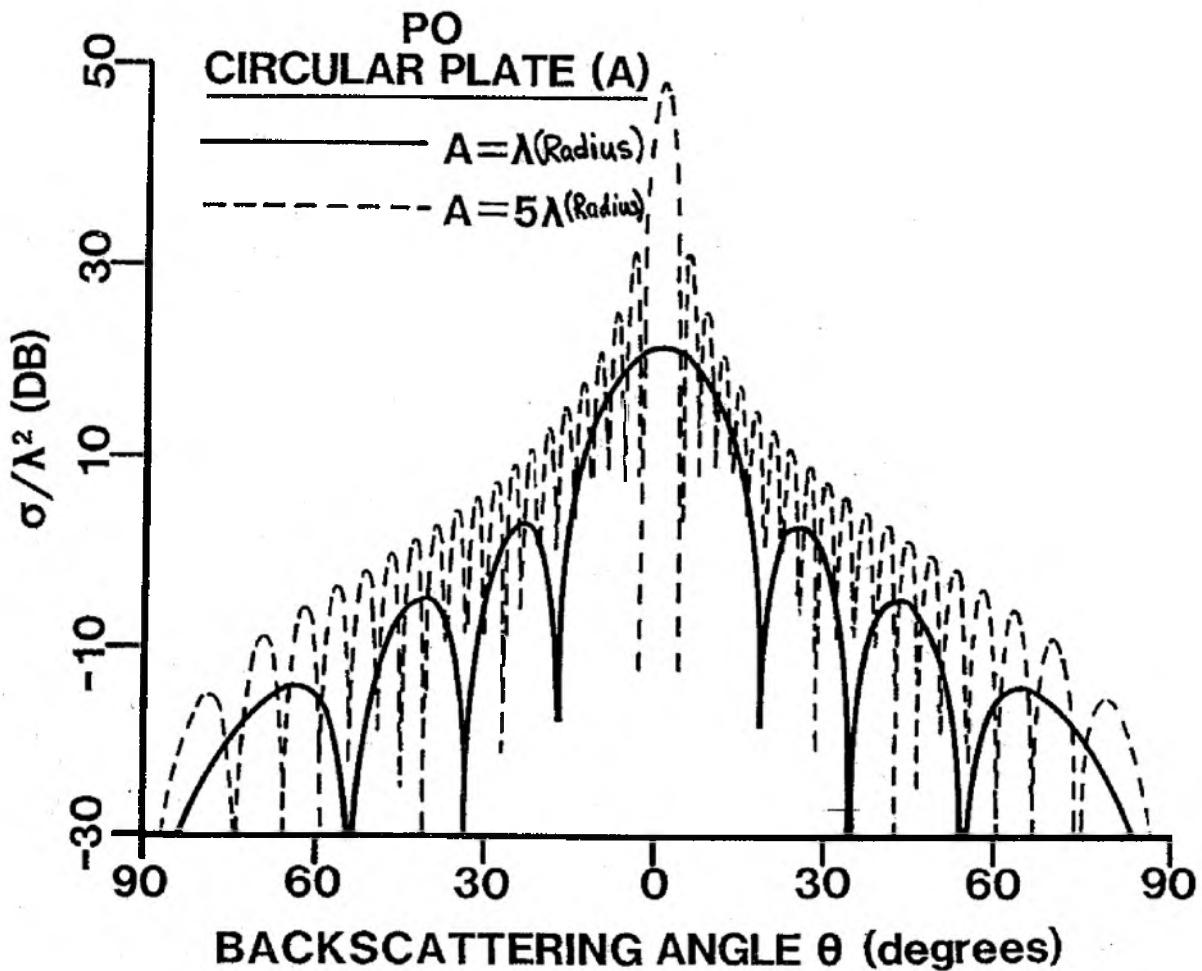
$$B = \sin \theta_s \sin \phi_s - \sin \theta_i = \sin \theta_i (-1) - \sin \theta_i = -2 \sin \theta_i \quad \left\{ \sqrt{A^2+B^2} = 2 \sin \theta_i \right\}$$

cont'd.

11.24 cont'd.

Thus $E_\theta^s = j \frac{\beta a^2 E_0 e^{-j\beta r}}{r} \cos \theta_i \frac{J_1(2\beta a \sin \theta_i)}{2\beta a \sin \theta_i}$, $E_\phi^s = 0$

$$\Gamma_{3-d} = \lim_{r \rightarrow \infty} \left\{ 4\pi r^2 \frac{|E_\theta^s|^2}{|E|^2} \right\} = 4\pi (\beta a^2)^2 \left[\cos^2 \theta_i \frac{J_1^2(2\beta a \sin \theta_i)}{2\beta a \sin \theta_i} \right] = 4\pi \left(\frac{\pi a^2}{\lambda} \right)^2 \left[2 \cos^2 \theta_i \frac{J_1^2(2\beta a \sin \theta_i)}{2\beta a \sin \theta_i} \right]^2$$



11.25 Using the geometry for the solution of Problem 11.24 but for TM^x plane

a. wave incidence, we can write

$$\underline{E}^i = \hat{a}_x E_0 e^{-j\beta(y \sin \theta_i - z \cos \theta_i)}, \underline{H}^i = \frac{E_0}{\eta} (-\hat{a}_y \cos \theta_i - \hat{a}_z \sin \theta_i) e^{-j\beta(y \sin \theta_i - z \cos \theta_i)}$$

$$\underline{J}_s = 2\hat{n} \times \underline{H}^i \Big|_{z=0, y=y'} = 2\hat{a}_x \times (-\hat{a}_y H_y - \hat{a}_z H_z) = \hat{a}_x \frac{2E_0}{\eta} \cos \theta_i e^{-j\beta y' \sin \theta_i}$$

$$J_y = J_z = 0, J_x = \frac{2E_0}{\eta} \cos \theta_i e^{-j\beta y' \sin \theta_i}$$

b. From the solution of Problem 11.24

$$N_\theta = \iint_S J_x \cos \theta_s \cos \phi_s e^{j\beta r' \cos \psi} ds' = \frac{2E_0}{\eta} \cos \theta_i \cos \theta_s \cos \phi_s \int_0^{2\pi} \int_0^a e^{-j\beta y'} e^{j\beta(x' \sin \theta_s \cos \phi_s + y' \sin \theta_s \sin \phi_s)} r' dr' d\phi'$$

$$N_\phi = \iint_S -J_x \sin \phi_s e^{j\beta r' \cos \psi} ds' = -\frac{2E_0}{\eta} \cos \theta_i \sin \phi_s \int_0^{2\pi} \int_0^a e^{-j\beta y'} e^{j\beta(x' \sin \theta_s \cos \phi_s + y' \sin \theta_s \sin \phi_s)} r' dr' d\phi'$$

Using the integration technique for the solution of Problem 11.24 we can write

$$N_\theta = \frac{4\pi a^2 E_0 \cos \theta_i \cos \theta_s \cos \phi_s}{\eta} \frac{J_1(\beta a \sqrt{A^2 + B^2})}{\beta a \sqrt{A^2 + B^2}}, N_\phi = -\frac{4\pi a^2 E_0 \cos \theta_i \sin \phi_s}{\eta} \frac{J_1(\beta a \sqrt{A^2 + B^2})}{\beta a \sqrt{A^2 + B^2}}$$

$$\underline{E}_\theta^s = -j \frac{\beta a^2 E_0 e^{-j\beta r}}{r} \cos \theta_i \cos \theta_s \cos \phi_s \frac{J_1(\beta a \sqrt{A^2 + B^2})}{\beta a \sqrt{A^2 + B^2}}, \underline{E}_\phi^s = j \frac{\beta a^2 E_0 e^{-j\beta r}}{r} \cos \theta_i \sin \phi_s \frac{J_1(\beta a \sqrt{A^2 + B^2})}{\beta a \sqrt{A^2 + B^2}}$$

$$A = \sin \theta_s \cos \phi_s$$

$$B = \sin \theta_s \sin \phi_s - \sin \theta_i$$

$$c. |\underline{E}^s| = \sqrt{|\underline{E}_\theta^s|^2 + |\underline{E}_\phi^s|^2} = E_0 \frac{\beta a^2}{r} \left\{ \sqrt{\cos^2 \theta_i \cos^2 \theta_s \cos^2 \phi_s + \cos^2 \theta_i \sin^2 \theta_s} \left| \frac{J_1(\beta a \sqrt{A^2 + B^2})}{\beta a \sqrt{A^2 + B^2}} \right| \right\}$$

$$|\underline{E}|^2 = \sqrt{|\underline{E}_\theta^s|^2 + |\underline{E}_\phi^s|^2} = E_0 \frac{\beta a^2}{r} \left\{ \cos \theta_i \sqrt{\cos^2 \theta_i \cos^2 \theta_s + \sin^2 \theta_s} \left| \frac{J_1(\beta a \sqrt{A^2 + B^2})}{\beta a \sqrt{A^2 + B^2}} \right| \right\} = E_0 \frac{\beta a^2}{r} |G|$$

$$G = \cos \theta_i \sqrt{\cos^2 \theta_i \cos^2 \theta_s + \sin^2 \theta_s} \left| \frac{J_1(\beta a \sqrt{A^2 + B^2})}{\beta a \sqrt{A^2 + B^2}} \right|$$

$$\sigma_{3-D} = \lim_{r \rightarrow \infty} \left\{ 4\pi r^2 \frac{|\underline{E}^s|^2}{|\underline{E}|^2} \right\} = 4\pi r^2 \left(\frac{\beta a^2}{r} \right)^2 |G|^2 = 4\pi \left(\frac{\pi a^2}{\lambda} \right)^2 |G|^2$$

For monostatic observations $\phi_s = \pi/2, \theta_s = \theta_i$.

$$A = \sin \theta_i \cos \phi_i = 0$$

$$B = \sin \theta_i \sin \phi_i - \sin \theta_i = \sin \theta_i (-1) - \sin \theta_i = -2 \sin \theta_i \quad \left\{ \sqrt{A^2 + B^2} = 2 \sin \theta_i \right\}$$

$$\underline{E}_\theta^s = 0, \underline{E}_\phi^s = -j \frac{\beta a^2 E_0 e^{-j\beta r}}{r} \cos \theta_i \frac{J_1(2\beta a \sin \theta_i)}{2\beta a \sin \theta_i}$$

$$\sigma_{3-D} = \lim_{r \rightarrow \infty} \left\{ 4\pi r^2 \frac{|\underline{E}_\phi^s|^2}{|\underline{E}|^2} \right\} = 4\pi \left(\frac{\pi a^2}{\lambda} \right)^2 \left[2 \cos \theta_i \frac{J_1(2\beta a \sin \theta_i)}{2\beta a \sin \theta_i} \right]^2$$

The monostatic for this polarization (TM^x) is the same as that of the TE^x polarization of Problem 11.24. Therefore the monostatic RCS plots for this polarization for $\alpha = 2, 5\lambda$ are those shown for the solution of Problem 11.24.

11.26) To solve this problem let us assume that the incident electric field is polarized in the x direction. The same answer is obtained using any other polarization.

$$\underline{E}^i = \hat{a}_x E_0 e^{j\beta z}, \quad \underline{H}^i = -\hat{a}_y \frac{E_0}{\eta} e^{j\beta z}$$

$$\underline{J}_s \approx 2\hat{n} \times \underline{H}^i \Big|_{z=0} = 2\hat{a}_z \times \left(-\hat{a}_y \frac{E_0}{\eta} e^{j\beta z} \right) = \hat{a}_x \frac{2E_0}{\eta}$$

$$J_x = \frac{2E_0}{\eta}, \quad J_y = J_z = 0, \quad \theta_s = 0$$

$$N_\theta = \iint_S [J_x \cos\phi_s \cos\theta_s + J_y \cos\phi_s \sin\theta_s - J_z \sin\phi_s] e^{j\beta(x' \sin\theta_s \cos\phi_s + y' \sin\theta_s \sin\phi_s)} dx dy,$$

$$= \frac{2E_0 \cos\phi_s}{\eta} \iint_S dx dy = \frac{2E_0 A \cos\phi_s}{\eta}$$

$$N_\phi = \iint_S [-J_x \sin\phi_s + J_y \cos\phi_s] e^{j\beta(x' \sin\theta_s \cos\phi_s + y' \sin\theta_s \sin\phi_s)} dx dy = -\frac{2E_0 \sin\phi_s}{\eta} \iint_S dx dy = -\frac{2E_0 A \sin\phi_s}{\eta}$$

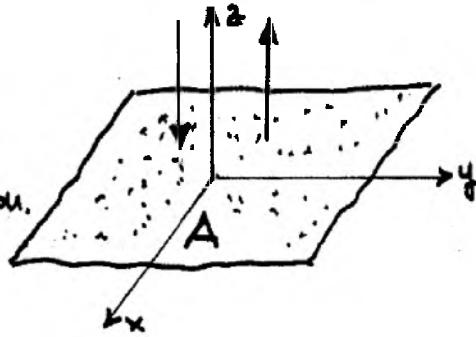
$$E_\theta^s \approx -j \frac{\beta e^{-j\beta r}}{4\pi r} [N_\theta + \eta N_\phi] = -j \frac{\beta e^{-j\beta r}}{4\pi r} (2E_0 A \cos\phi_s) = -j \frac{E_0 A \beta e^{-j\beta r}}{2\pi r} \cos\phi_s$$

$$E_\phi^s \approx j \frac{\beta e^{-j\beta r}}{4\pi r} [N_\theta - \eta N_\phi] = j \frac{\beta e^{-j\beta r}}{4\pi r} (2E_0 A \sin\phi_s) = j \frac{E_0 A \beta e^{-j\beta r}}{2\pi r} \sin\phi_s$$

$$|\underline{E}^s| = \sqrt{|E_\theta^s|^2 + |E_\phi^s|^2} = \left| \frac{E_0 A \beta}{2\pi r} \sqrt{\cos^2\phi_s + \sin^2\phi_s} \right| = \left| \frac{E_0 A \beta}{2\pi r} \right|$$

$$\sigma = \lim_{r \rightarrow \infty} \left[4\pi r^2 \frac{|\underline{E}^s|^2}{|\underline{E}^i|^2} \right] = \lim_{r \rightarrow \infty} \left[4\pi r^2 \frac{|\underline{E}^s|^2 A^2 \beta^2}{4\pi^2 r^2 |\underline{E}^i|^2} \right] = \left(\frac{A \beta}{\pi} \right)^2 = 4\pi \left(\frac{A}{\lambda} \right)^2$$

$$\sigma = 4\pi \left(\frac{A}{\lambda} \right)^2$$



11.24

- a. With the PO approximation, the current on the surface of the perfectly conducting object can be expressed as

$$\underline{J}_S = 2 H_0 \hat{n} \times \hat{a}_h e^{-j\beta R_1} \quad (1)$$

where \hat{a}_h represents the polarization of the H field and R_1 is the distance from the source to the scattering point. Then

$$\underline{A} = \frac{i}{4\pi} \iint_S \underline{J}_S \frac{e^{-j\beta R_2}}{R_2} d\underline{s}' = \frac{i_0 H_0}{2\pi} \iint_S (\hat{n} \times \hat{a}_h) \frac{e^{-j\beta(R_1+R_2)}}{R_2} d\underline{s}', \quad (2)$$

$$\underline{H}^S = \frac{1}{\mu_0} \nabla \times \underline{A} = \frac{H_0}{2\pi} \iint_S \nabla \times \left[(\hat{n} \times \hat{a}_h) \frac{e^{-j\beta(R_1+R_2)}}{R_2} \right] d\underline{s}' \quad (3)$$

where the integration is over the surface of the object. For far field and monostatic observations

$$R_1 = R_2 = R - z, \quad \nabla \leftrightarrow -j\beta \hat{a}_z \quad (4)$$

where R is the distance from the origin to the observation point, and z is the z axis projection of the position of the scattering point. With the above approximation, the scattered field of (3) can be reduced to

$$\underline{H}^S = -j \frac{\beta H_0}{2\pi R} \iint_S \hat{a}_z \times (\hat{n} \times \hat{a}_h) e^{-j\beta(R-z)} d\underline{s}' \quad (5)$$

Since

$$\text{then } \hat{a}_z \times (\hat{n} \times \hat{a}_h) = (\hat{a}_z \cdot \hat{a}_h) \hat{n} - (\hat{a}_z \cdot \hat{n}) \hat{a}_h, \quad \hat{a}_z \cdot \hat{a}_h = 0 \quad (6)$$

$$\underline{H}^S = j \frac{\beta H_0 e^{-j\beta R}}{2\pi R} \iint_S (\hat{a}_z \cdot \hat{n}) \hat{a}_h e^{+j2\beta z} d\underline{s}' \quad (7)$$

Because

$$\text{then } (\hat{a}_z \cdot \hat{n}) d\underline{s}' = dA \quad (8)$$

$$\underline{H}^S = j \frac{\beta H_0 e^{-j\beta R}}{2\pi R} \iint_A \hat{a}_h e^{+j2\beta z} dA \quad (9)$$

where the integration is performed in the xy -plane.

- b. The monostatic RCS can now be written as

$$\sigma_{3-D} = \lim_{R \rightarrow \infty} \left[4\pi R^2 \frac{|H^S|^2}{|H^S|^2} \right] = \left(\frac{\beta H_0}{\pi} \right)^2 \left| \iint_A \hat{a}_h e^{+j2\beta z} dA \right|^2$$

$$\sigma_{3-D} = \frac{4\pi}{\lambda^2} \left| \iint_A e^{+j2\beta z} dA \right|^2$$

cont'd.

11.27 cont'd.

c. It is sometimes more convenient to write the integral in Part b. in parameter form since many regular geometries have pretty simple expressions in terms of some special parameters. Thus by a change of variables we have

$$dA = dx dy = \begin{vmatrix} \frac{\partial x}{\partial u} & \frac{\partial x}{\partial v} \\ \frac{\partial y}{\partial u} & \frac{\partial y}{\partial v} \end{vmatrix} du dv = J(x, y; u, v) du dv \text{ where } J \text{ is the Jacobian.}$$

In many cases it is also possible to write the integral as

$$dA = J[u(\alpha), v] \frac{du}{d\alpha} d\alpha dv$$

which allows an independent integral

$$\sigma_{3-2} = \frac{4\pi}{\lambda^2} \left| \iint e^{j2\beta z} J[u(\alpha), v] \frac{du}{d\alpha} d\alpha dv \right|^2$$

Taking the cylindrical coordinates as a special case,

$$\begin{cases} u = \rho \\ v = \phi \end{cases} \quad J(x, y; u, v) = \rho$$

$$dA = \rho d\rho d\phi = \frac{1}{2} \frac{d\rho^2}{dz} dz d\phi$$

thus

$$\sigma_{3-2} = \frac{4\pi}{\lambda^2} \frac{1}{2} \left| \iint e^{j2\beta z} \frac{d\rho^2}{dz} dz d\phi \right|^2 = \frac{\pi}{\lambda^2} \left| \int_0^{2\pi} \int_0^{z(\phi)} e^{j2\beta z} \frac{d\rho^2}{dz} dz d\phi \right|^2$$

$$\boxed{11.28} \quad \left(\frac{x}{a} \right)^2 + \left(\frac{y}{b} \right)^2 = -\frac{z}{c}, \quad z < h$$

$$a. \quad z = -c \left[\left(\frac{x}{a} \right)^2 + \left(\frac{y}{b} \right)^2 \right]$$

For $y=0$ (i.e. xz -plane)

$$z = -c \left(\frac{x}{a} \right)^2$$

$$z' = \frac{dz}{dx} = -2c \left(\frac{x}{a} \right) \left(\frac{1}{a} \right) = -2 \frac{c}{a^2} x$$

$$z'' = \frac{d^2 z}{dx^2} = -\frac{2c}{a^2}$$

$$R_1 \Big|_{x=0} = \text{radius of curvature} = \left[\frac{1+(z')^2}{|z''|} \right]^{3/2} \Big|_{x=0} = \left[\frac{1+\left(\frac{2c}{a^2} x \right)^2}{\frac{2c}{a^2}} \right]^{3/2} \Big|_{x=0} = \frac{a^2}{2c} = a_1$$

$$R_1(x=0) = \frac{a^2}{2c} = a_1$$

For $x=0$ (i.e. yz -plane)

$$z = -c \left(\frac{y}{b} \right)^2$$

$$z' = \frac{dz}{dy} = -2c \left(\frac{y}{b} \right) \left(\frac{1}{b} \right) = -2 \frac{c}{b^2} y$$

$$z'' = \frac{d^2 z}{dy^2} = -\frac{2c}{b^2}$$

$$R_2 \Big|_{y=0} = \left[\frac{1+(z')^2}{|z''|} \right]^{3/2} \Big|_{y=0} = \left[\frac{1+\left(\frac{2c}{b^2} y \right)^2}{\frac{2c}{b^2}} \right]^{3/2} \Big|_{y=0} = \frac{b^2}{2c} = a_2$$

11.28 cont'd. $x = \rho \cos\phi, y = \rho \sin\phi$

b.

$$\begin{aligned} z &= -c \left[\frac{x^2}{a^2} + \frac{y^2}{b^2} \right] = -c \rho^2 \left[\left(\frac{\cos\phi}{a} \right)^2 + \left(\frac{\sin\phi}{b} \right)^2 \right] = -c \rho^2/a^2 \left[\omega^2 \phi + \left(\frac{a}{b} \right)^2 \sin^2\phi \right] \\ &= -\frac{c}{a^2} \rho^2 \left[1 - \sin^2\phi + \left(\frac{a}{b} \right)^2 \sin^2\phi \right] = -\frac{2c}{a^2} \left(\frac{\rho^2}{2} \right) \left[1 - \left[\sin^2\phi - \left(\frac{a}{b} \right)^2 \sin^2\phi \right] \right] \\ z &= -\frac{\rho^2}{2a_1} \left\{ 1 - \left[1 - \frac{a_1}{a_2} \right] \sin^2\phi \right\} \text{ where } a_1 = \frac{a^2}{2c}, a_2 = \frac{b^2}{2c} \end{aligned}$$

c. From Part b

$$dz = -\frac{d\rho^2}{2a_1} \left\{ 1 - \left[1 - \frac{a_1}{a_2} \right] \sin^2\phi \right\} \Rightarrow \frac{d\rho^2}{dz} = -\frac{2a_1}{1 - \left[1 - \frac{a_1}{a_2} \right] \sin^2\phi}$$

Then

$$\sigma_{3-D} = \frac{\pi}{\lambda^2} \left| \int_0^{2\pi} \int_0^h e^{j2\beta z} \frac{2a_1}{1 - \left[1 - \frac{a_1}{a_2} \right] \sin^2\phi} dz d\phi \right|^2$$

Using the integral

$$\int_0^{2\pi} \frac{d\phi}{1 - \alpha \sin^2\phi} = \frac{2\pi}{\sqrt{1 - \alpha}}$$

we can write the RCS as

$$\begin{aligned} \sigma_{3-D} &= \frac{\pi}{\lambda^2} \left| \left(\frac{e^{j2\beta h}}{j2\beta} \right) \int_0^{2\pi} \frac{2a_1(2\pi)}{\sqrt{1 - (1 - \frac{a_1}{a_2}) \sin^2\phi}} d\phi \right|^2 = \frac{\pi}{(\beta)^2} \left(\frac{e^{j2\beta h}}{j2} \right) 4\pi \sqrt{a_1 a_2} \end{aligned}$$

$$\sigma_{3-D} = 4\pi a_1 a_2 \sin^2(\beta h)$$

d. From Part c of the solution it is apparent that elimination of the exponential reduces the RCS to

$$\sigma_{3-D} = \pi a_1 a_2 \sin^2(\beta h)$$

$$11.29 \quad \left(\frac{x}{a}\right)^2 + \left(\frac{y}{b}\right)^2 + \left(\frac{z}{c}\right)^2 = 1, \quad x = \rho \cos\phi, \quad y = \rho \sin\phi$$

$$a. \quad \left(\frac{z}{c}\right)^2 = 1 - \left\{ \left(\frac{x}{a}\right)^2 + \left(\frac{y}{b}\right)^2 \right\} = 1 - \frac{1}{a^2} [x^2 + \left(\frac{y}{b}\right)^2] = 1 - \frac{\rho^2}{a^2} \left\{ 1 - \left[1 - \frac{a_1}{a_2}\right] \sin^2\phi \right\}$$

$$z^2 = c^2 - \frac{c^2 \rho^2}{a^2} \left\{ 1 - \left[1 - \frac{a_1}{a_2}\right] \sin^2\phi \right\} = c^2 - \frac{c \rho^2}{a^2} \left\{ 1 - \left[1 - \frac{a_1}{a_2}\right] \sin^2\phi \right\} = c^2 - \rho^2 \frac{c}{a_1} \left\{ 1 - \left(1 - \frac{a_1}{a_2}\right) \sin^2\phi \right\}$$

b. To find the RCS we perform the integration using parameter integral by assuming that

$$x = at \cos\phi$$

$$y = bt \sin\phi$$

then the ellipsoid equation can be expressed as

$$\left(\frac{z}{c}\right)^2 = 1 - \left[\left(\frac{x}{a}\right)^2 + \left(\frac{y}{b}\right)^2 \right] = 1 - \left[t^2 \cos^2\phi + t^2 \sin^2\phi \right] = 1 - t^2$$

$$dA = J(x, y; t, \phi) dt d\phi = - \frac{abz}{c^2} dz d\phi$$

Now from Part c of Problem 11.27 we evaluate the integral from $z = h$ to $z = c$. Thus

$$\sigma_{3-D} = \frac{4\pi}{\lambda^2} \left| \int_0^{2\pi} \int_h^c e^{j2\beta z} J[u(z), v] \frac{du}{dz} dz dv \right|^2 = \frac{4\pi}{\lambda^2} \left| \int_0^{2\pi} \int_h^c e^{j2\beta z} \frac{abz}{c^2} dz d\phi \right|^2$$

$$\sigma_{3-D} = \frac{4\pi}{\lambda^2} \left| \frac{ab(2\pi)}{c^2} \int_h^c e^{j2\beta z} z dz \right|^2$$

Neglecting the term with $(2\beta c)^{-1}$, the above equation for the RCS reduces to

$$\sigma_{3-D} = \frac{\pi(ab)^2}{c^4} \left| c e^{+j2\beta c} - h e^{+j2\beta h} \right|^2 = \frac{\pi a^2 b^2}{c^4} \left\{ c^2 + h^2 - 2hc \cos[2\beta(c-h)] \right\}$$

$$\sigma_{3-D} = \pi \frac{a^2 b^2}{c^2} \left\{ 1 + \left(\frac{h}{c}\right)^2 - 2 \frac{h}{c} \cos[2\beta(c-h)] \right\} = \pi a_1 a_2 \left\{ 1 + \left(\frac{h}{c}\right)^2 - 2 \frac{h}{c} \cos[2\beta(c-h)] \right\}$$

If $h = 0$ or if h is very irregular around the periphery of the ellipsoid, then the above expression for σ_{3-D} reduces to

$$\sigma_{3-D} = \pi a_1 a_2 \stackrel{a_1 = a_2 = a}{=} \pi a^2$$

$$11.30 \quad H_n^{(2)} = J_n(x) - j Y_n(x) \Rightarrow H_{-n}^{(2)} = J_{-n}(x) - j Y_{-n}(x)$$

For integer n : $J_{-n}(x) = (-1)^n J_n(x)$, $Y_{-n}(x) = (-1)^n Y_n(x)$

Therefore

$$H_{-n}^{(2)} = J_{-n}(x) - j Y_{-n}(x) = (-1)^n J_n(x) - j (-1)^n Y_n(x)$$

$$H_{-n}^{(2)} = (-1)^n [J_n(x) - j Y_n(x)] = (-1)^n H_n^{(2)}(x)$$

* In the first printing, there was an ^{typographical} error in the statement of the problem; an extra +1 appeared on the exponent of the (-1) .

$$11.31 \quad E_2^+ = e^{-j\beta x} = e^{-j\beta p \cos\phi} = \sum_{n=-\infty}^{+\infty} a_n J_n(\beta p) e^{jn\phi}$$

Using (11-54a) or $a_n = j^{-n}$, we can write the above expression as

$$E_2^+ = e^{-j\beta x} = e^{-j\beta p \cos\phi} = \sum_{n=-\infty}^{+\infty} j^{-n} J_n(\beta p) e^{jn\phi}$$

11.32 Since the only difference between $H_0^{(2)}$ and $H_0^{(1)}$ is the superscript, the expansion for $H_0^{(1)}$ in terms of cylindrical wave function originating at the origin proceeds along the same lines as the expansion $H_0^{(2)}$ with the replacement of the $H_0^{(2)}$ type of wave functions by $H_0^{(1)}$. Therefore (11-71a) is easily obtained by referring to (11-69a) and (11-71b) is obtained by referring to (11-69b).

$$11.33 \quad \underline{E}^i = \hat{a}_2 E_2^i = \hat{a}_2 E_0 \sum_{n=-\infty}^{+\infty} j^{-n} J_n(\beta p) e^{jn\phi}$$

$$\underline{E}^i = \hat{a}_2 E_0 \left\{ \underbrace{J_0(\beta p)}_{n=0} + \underbrace{j^{-1} J_1(\beta p) e^{j\phi}}_{n=1} + \underbrace{j^{-2} J_2(\beta p) e^{j2\phi}}_{n=2} + \dots \right. \\ \left. + j^1 \underbrace{J_{-1}(\beta p) e^{-j\phi}}_{n=-1} + j^2 \underbrace{J_{-2}(\beta p) e^{-j2\phi}}_{n=-2} + \dots \right\}$$

$$\text{Since } J_{-n}(x) = (-1)^n J_n(x) \Rightarrow J_{-1}(\beta p) = -J_1(\beta p), J_{-2}(\beta p) = -J_2(\beta p) \\ J_{-3}(\beta p) = J_3(\beta p), J_{-4}(\beta p) = J_4(\beta p)$$

Thus

$$\underline{E}^i = \hat{a}_2 E_0 \left\{ J_0(\beta p) + j^{-1} 2 J_1(\beta p) \left[e^{j\phi} - \frac{1}{2} e^{-j\phi} \right] + j^{-2} 2 J_2(\beta p) \left[e^{j2\phi} + \frac{1}{2} e^{-j2\phi} \right] + \dots \right\} \\ = \hat{a}_2 E_0 \left\{ J_0(\beta p) - j 2 J_1(\beta p) \cos(\phi) - 2 J_2(\beta p) \cos(2\phi) + \dots \right\}$$

$$\underline{E}^i = \hat{a}_2 E_0 \sum_{n=0}^{\infty} (-j)^n \epsilon_n J_n(\beta p) \cos(n\phi), \quad \epsilon_n = \begin{cases} 1 & n=0 \\ 2 & n \neq 0 \end{cases}$$

$$11.34 \quad \underline{J}_S = \hat{a}_2 \frac{2 E_0}{\pi a \omega \mu} \sum_{n=-\infty}^{+\infty} j^{-n} \frac{e^{jn\phi}}{H_n^{(2)}(\beta a)}$$

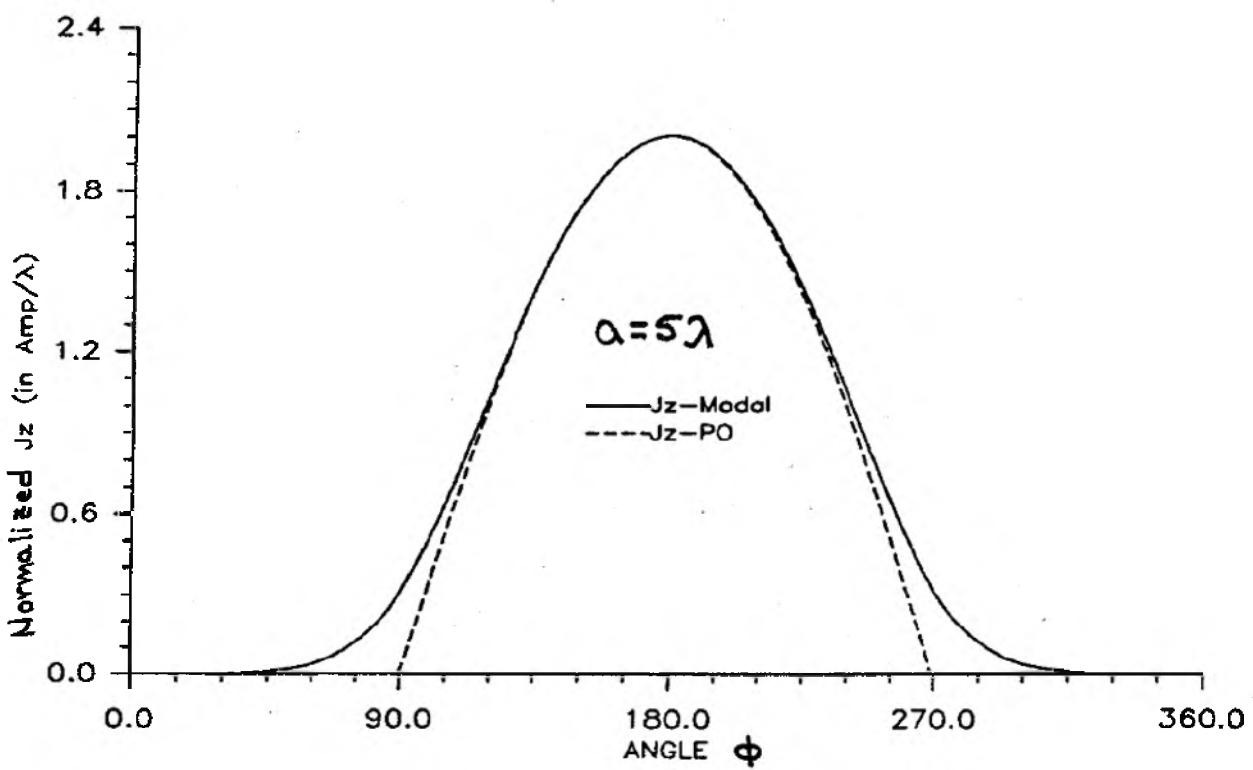
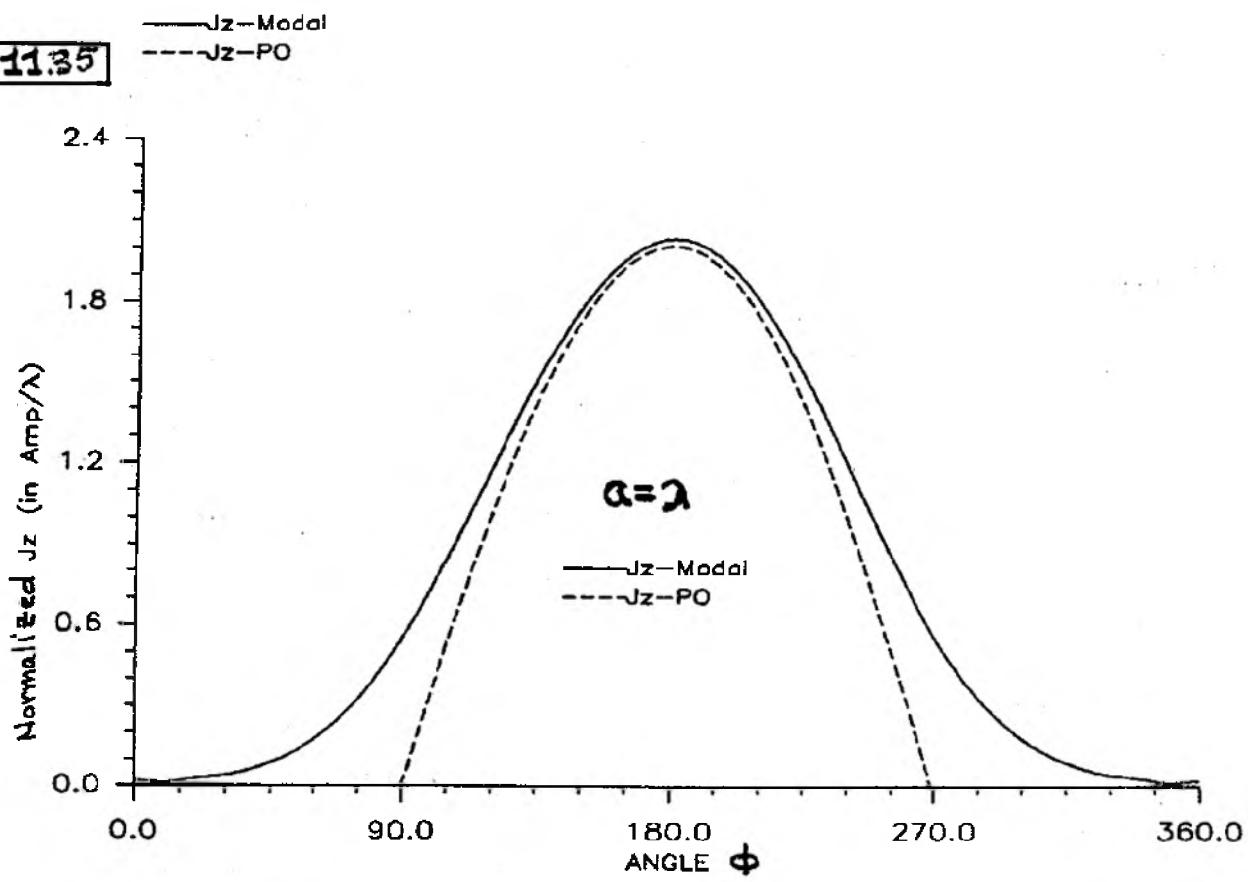
Using a procedure similar to that for the solution of Problem 11.33 we can write the above expression as

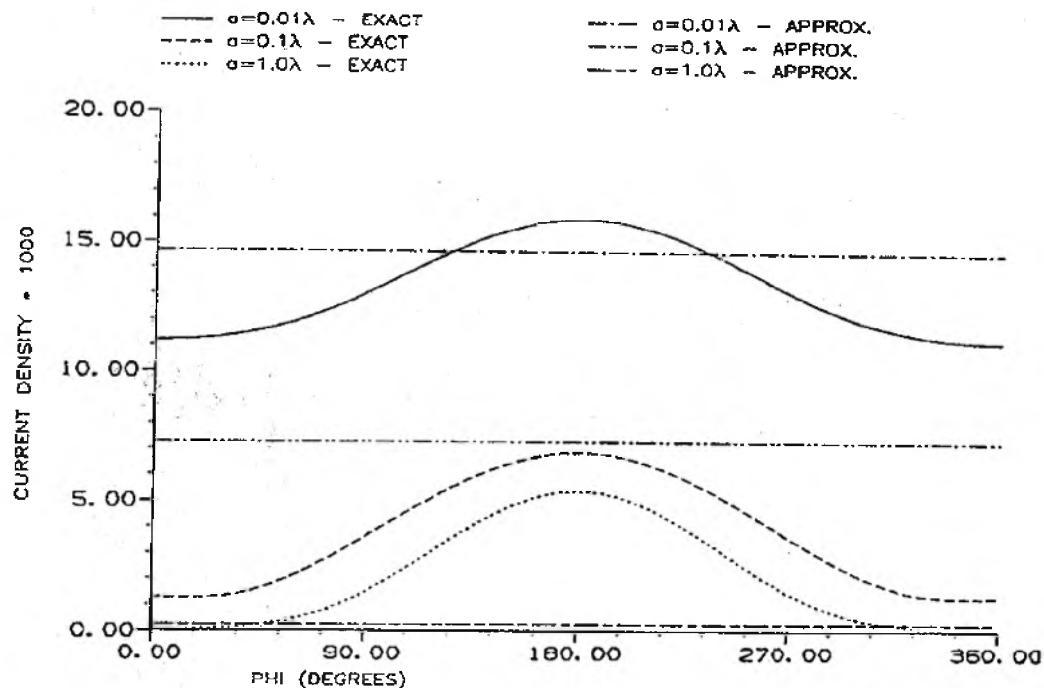
$$\underline{J}_S = \hat{a}_2 \frac{2 E_0}{\pi a \omega \mu} \sum_{n=0}^{\infty} (-j)^n \epsilon_n \frac{\cos(n\phi)}{H_n^{(2)}(\beta a)}$$

where

$$\epsilon_n = \begin{cases} 1 & n=0 \\ 2 & n \neq 0 \end{cases}$$

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INDUCED CURRENT DENSITY
CIRCULAR CYLINDERTM² POLARIZATION

$$11.37 \quad \sigma_{2-D} = \frac{4}{\beta} \left| \sum_{n=-\infty}^{+\infty} \frac{J_n(\beta a)}{H_n^{(2)}(\beta a)} e^{jn\phi} \right|^2 = \frac{4}{\beta} \left| \underbrace{\frac{J_0(\beta a)}{H_0^{(2)}(\beta a)}}_{n=0} + \underbrace{\frac{J_1(\beta a)}{H_1^{(2)}(\beta a)} e^{j\phi}}_{n=1} + \underbrace{\frac{J_2(\beta a)}{H_2^{(2)}(\beta a)} e^{j2\phi}}_{n=2} + \dots + \underbrace{\frac{J_{-1}(\beta a)}{H_{-1}^{(2)}(\beta a)} e^{-j\phi}}_{n=-1} + \underbrace{\frac{J_{-2}(\beta a)}{H_{-2}^{(2)}(\beta a)} e^{-j2\phi}}_{n=-2} + \dots \right|^2$$

$$\text{Since } J_{-n}(x) = (-1)^n J_n(x) \quad \left\{ \begin{array}{l} J_{-1}(x) = -J_1(x), J_{-2}(x) = J_2(x), J_{-3}(x) = -J_3(x), \dots \\ H_{-n}^{(2)}(x) = (-1)^n H_n^{(2)}(x) \end{array} \right. \quad \left\{ \begin{array}{l} H_{-1}^{(2)}(x) = -H_1^{(2)}(x), H_{-2}^{(2)}(x) = H_2^{(2)}(x), H_{-3}^{(2)}(x) = -H_3^{(2)}(x), \dots \end{array} \right.$$

then

$$\sigma_{2-D} = \frac{4}{\beta} \left| \frac{J_0(\beta a)}{H_0^{(2)}(\beta a)} + 2 \underbrace{\frac{J_1(\beta a)}{H_1^{(2)}(\beta a)} \left(\frac{e^{j\phi} + e^{-j\phi}}{2} \right)}_{\cos\phi} + 2 \underbrace{\frac{J_2(\beta a)}{H_2^{(2)}(\beta a)} \left(\frac{e^{j2\phi} + e^{-j2\phi}}{2} \right)}_{\cos 2\phi} + \dots \right|^2$$

$$\sigma_{2-D} = \frac{2}{\pi} \left| \sum_{n=0}^{\infty} \epsilon_n \frac{J_n(\beta a)}{H_n^{(2)}(\beta a)} \cos(n\phi) \right|^2$$

11.38 According to (11-92a) and (11-92b) we see that the ρ component of the magnetic field has a $1/\rho$ variation outside the summation while the ϕ component has no ρ variations outside the summation. Within the summation both have similar ρ variations when $\rho \gg \lambda$ (far field). Therefore we can approximate the far zone scattered magnetic field by

$$\underline{H}^S = \hat{a}_\phi H_\rho^S + \hat{a}_\phi H_\phi^S \xrightarrow{\rho \gg \lambda} \hat{a}_\phi H_\phi^S = \hat{a}_\phi \frac{\beta E_0}{j\omega\mu} \sum_{n=-\infty}^{+\infty} \frac{-n J_n(\beta a)}{H_n^{(2)}(\beta a)} H_n^{(2)'}(\beta \rho) e^{jn\phi}$$

$$H_n^{(2)'}(\beta \rho) = \frac{d}{d(\beta \rho)} H_n^{(2)}(\beta \rho) \xrightarrow{\rho \gg \lambda} \frac{d}{d(\beta \rho)} \left\{ \sqrt{\frac{2j}{\pi}} (-j)^{n+1} \beta^n j^n \right\}$$

$$\xrightarrow{\rho \gg \lambda} j^n \sqrt{\frac{2j}{\pi}} \left\{ \beta^{-\frac{1}{2}} (-j)^{-\frac{n+1}{2}} (\beta \rho)^{-\frac{n+1}{2}} e^{-j\beta \rho} \right\} = j^n \sqrt{\frac{2j}{\pi}} \left\{ -j \frac{e^{-j\beta \rho}}{(\beta \rho)^{\frac{n+1}{2}}} - \frac{e^{-j\beta \rho}}{a(\beta \rho)^{\frac{n+1}{2}}} \right\}$$

$$H_n^{(2)}(\beta \rho) \xrightarrow{\rho \gg \lambda} -j^n \sqrt{\frac{2j}{\pi}} \frac{e^{-j\beta \rho}}{\sqrt{\beta \rho}} = -j^{n+1} \sqrt{\frac{2j}{\pi}} \frac{e^{-j\beta \rho}}{\sqrt{\beta \rho}} = -j^{n+1} \sqrt{\frac{2j}{\pi \beta \rho}} e^{-j\beta \rho}$$

Therefore the scattered magnetic field can be approximated by

$$\underline{H}^S \approx \hat{a}_\phi \frac{\beta E_0}{j\omega\mu} \sqrt{\frac{2j}{\pi \beta \rho}} e^{-j\beta \rho} \sum_{n=-\infty}^{+\infty} \frac{J_n(\beta a)}{H_n^{(2)}(\beta a)} e^{jn\phi} = \hat{a}_\phi \frac{E_0}{\eta} \sqrt{\frac{2j}{\pi \beta \rho}} e^{-j\beta \rho} \sum_{n=-\infty}^{+\infty} \frac{J_n(\beta a)}{H_n^{(2)}(\beta a)} e^{jn\phi}$$

The RCS can now be written as

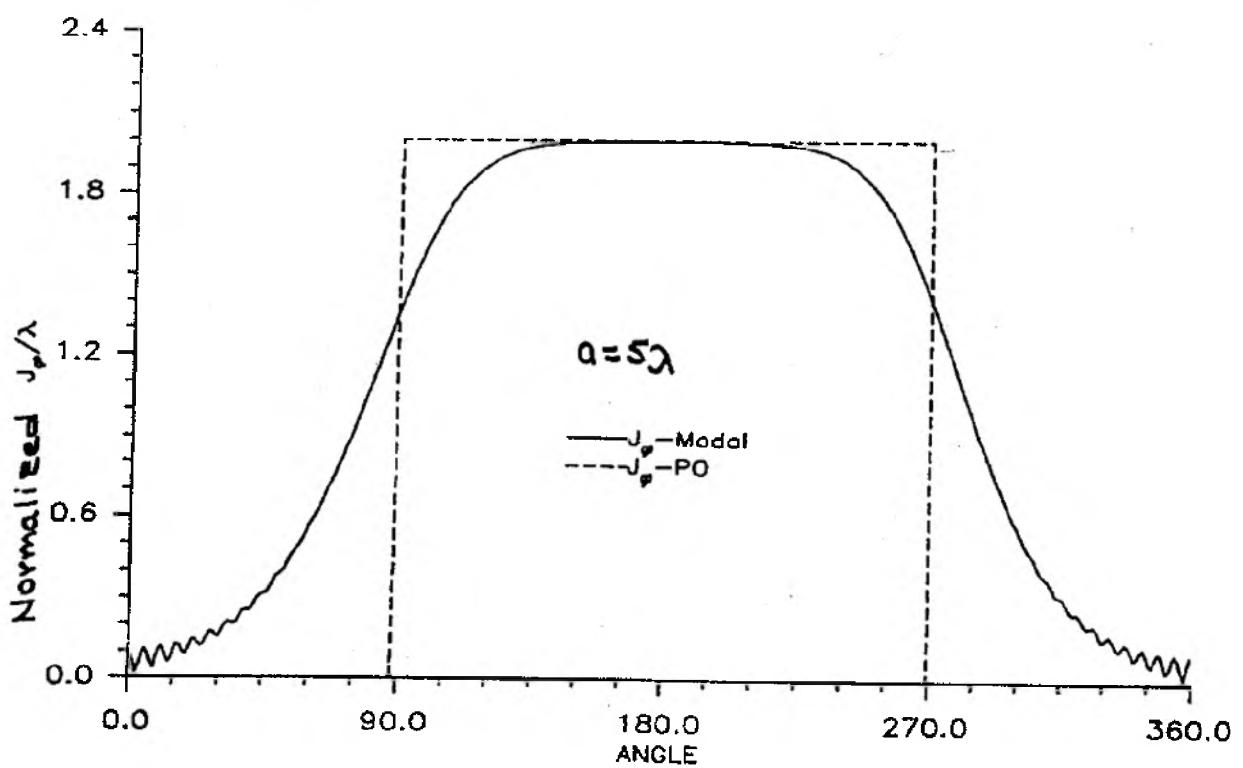
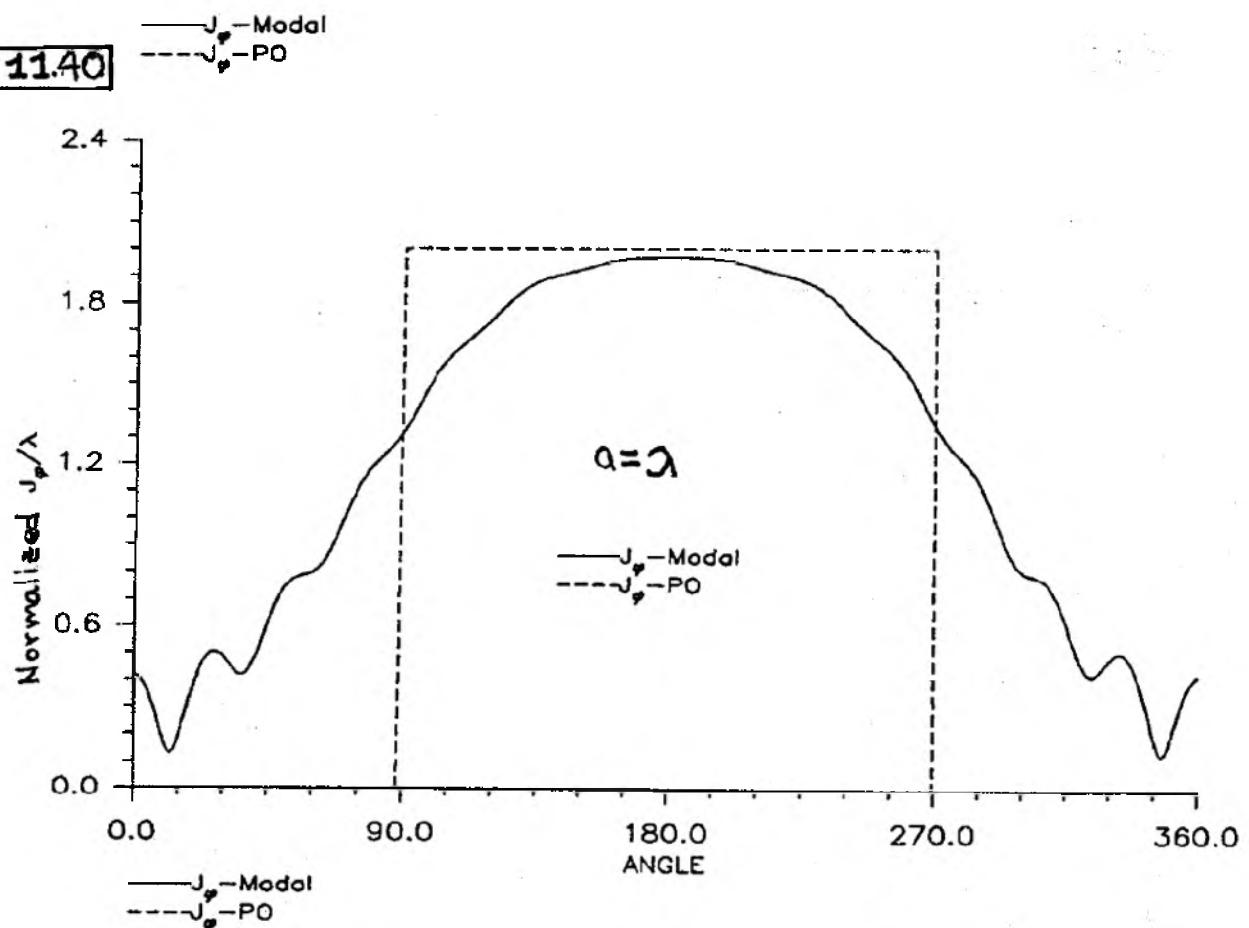
$$\sigma_{2-D} = \lim_{\rho \rightarrow \infty} \left[2\pi\rho \frac{|\underline{H}^S|^2}{|\underline{H}|^2} \right] = \frac{4}{\beta} \left| \sum_{n=-\infty}^{+\infty} \frac{J_n(\beta a)}{H_n^{(2)}(\beta a)} e^{jn\phi} \right|^2 \text{ where } \underline{H}^S = \hat{a}_\phi \frac{E_0}{\eta}$$

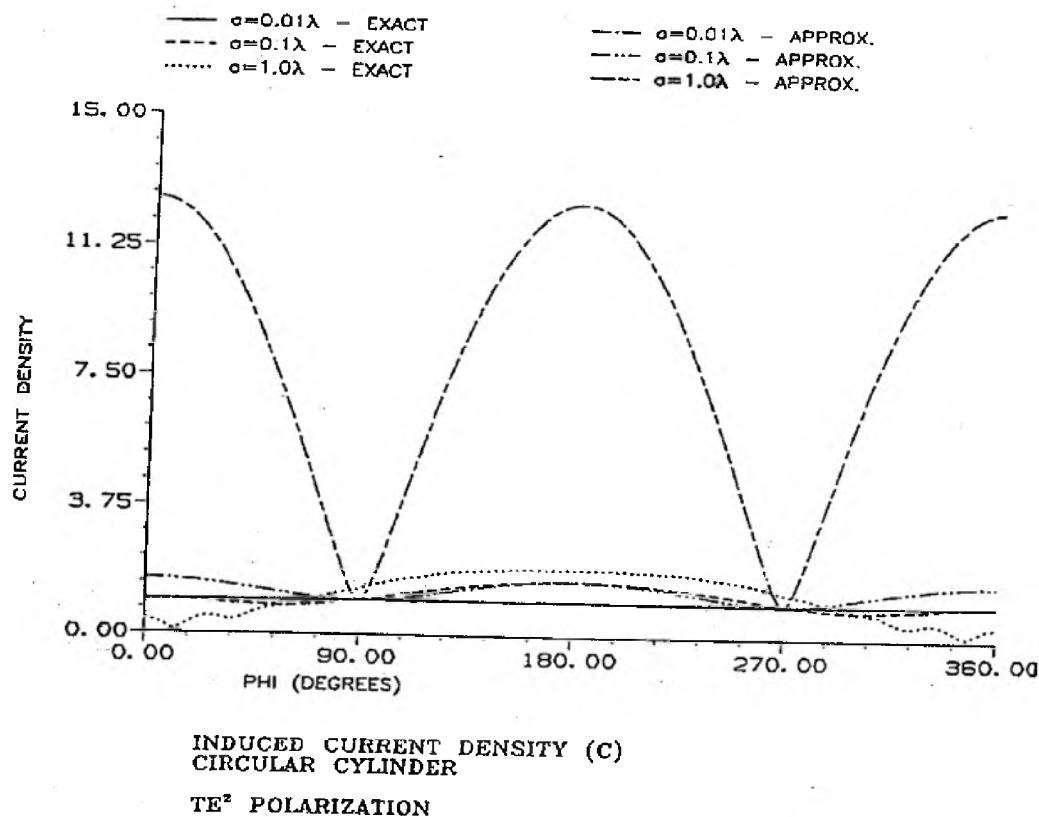
$$\begin{aligned} \underline{J}_S &= \hat{a}_\phi j \frac{2H_0}{\pi \beta a} \sum_{n=-\infty}^{+\infty} \frac{-n e^{jn\phi}}{H_n^{(2)'}(\beta a)} = \hat{a}_\phi j \frac{2H_0}{\pi \beta a} \left\{ \underbrace{\frac{1}{H_0^{(2)'}(\beta a)}}_{n=0} + \right. \\ &\quad + \underbrace{j^{-1} \frac{e^{j\phi}}{H_1^{(2)'}(\beta a)}}_{n=1} + \underbrace{j^{-2} \frac{e^{j2\phi}}{H_2^{(2)'}(\beta a)}}_{n=2} + \underbrace{j^{-3} \frac{e^{j3\phi}}{H_3^{(2)'}(\beta a)}}_{n=3} + \underbrace{j^{-4} \frac{e^{j4\phi}}{H_4^{(2)'}(\beta a)}}_{n=4} + \dots \\ &\quad + \underbrace{j^{-1} \frac{e^{-j\phi}}{H_{-1}^{(2)'}(\beta a)}}_{n=-1} + \underbrace{j^{-2} \frac{e^{-j2\phi}}{H_{-2}^{(2)'}(\beta a)}}_{n=-2} + \underbrace{j^{-3} \frac{e^{-j3\phi}}{H_{-3}^{(2)'}(\beta a)}}_{n=-3} + \underbrace{j^{-4} \frac{e^{-j4\phi}}{H_{-4}^{(2)'}(\beta a)}}_{n=-4} + \dots \\ \underline{J}_S &= \hat{a}_\phi j \frac{2H_0}{\pi \beta a} \left\{ \frac{1}{H_0^{(2)'}(\beta a)} + j^{-1} \frac{2}{H_1^{(2)'}(\beta a)} \left[\frac{e^{j\phi} + e^{-j\phi}}{2} \right] + j^{-2} \frac{2}{H_2^{(2)'}(\beta a)} \left[\frac{e^{j2\phi} + e^{-j2\phi}}{2} \right] + \dots \right\} \\ &= \hat{a}_\phi j \frac{2H_0}{\pi \beta a} \left\{ \frac{1}{H_0^{(2)'}(\beta a)} + j^{-1} \frac{2 \cos \phi}{H_1^{(2)'}(\beta a)} - \frac{2 \cos (2\phi)}{H_2^{(2)'}(\beta a)} + j^{-3} \frac{2 \cos (3\phi)}{H_3^{(2)'}(\beta a)} \dots \right\} \end{aligned}$$

$$\underline{J}_S = \hat{a}_\phi j \frac{2H_0}{\pi \beta a} \left\{ \sum_{n=0}^{\infty} \epsilon_n (-j)^n \frac{\cos(n\phi)}{H_n^{(2)'}(\beta a)} \right\}$$

$$\text{where } \epsilon_n = \begin{cases} 1, & n=0 \\ 2, & n \neq 0 \end{cases}$$

11.40





$$[11.42] \quad \sigma_{2-D} = \frac{4}{\beta} \left| \sum_{n=-\infty}^{+\infty} \frac{J_n'(\beta a)}{H_n^{(2)}(\beta a)} e^{jn\phi} \right|^2 = \frac{4}{\beta} \left| \underbrace{\frac{J_0'(\beta a)}{H_0^{(2)}(\beta a)}}_{n=0} + \underbrace{\frac{J_1'(\beta a)}{H_1^{(2)}(\beta a)} e^{j\phi}}_{n=1} + \underbrace{\frac{J_2'(\beta a)}{H_2^{(2)}(\beta a)} e^{j2\phi}}_{n=2} + \dots + \underbrace{\frac{J_{-1}'(\beta a)}{H_{-1}^{(2)}(\beta a)} e^{-j\phi}}_{n=-1} + \underbrace{\frac{J_{-2}'(\beta a)}{H_{-2}^{(2)}(\beta a)} e^{-j2\phi}}_{n=-2} + \dots \right|^2$$

$$\sigma_{2-D} = \frac{4}{\beta} \left| \frac{J_0'(\beta a)}{H_0^{(2)}(\beta a)} + 2 \frac{J_1'(\beta a)}{H_1^{(2)}(\beta a)} \left(e^{j\phi} + e^{-j\phi} \right) + 2 \frac{J_2'(\beta a)}{H_2^{(2)}(\beta a)} \left(e^{j2\phi} + e^{-j2\phi} \right) + \dots \right|^2$$

$$\sigma_{2-D} = \frac{4}{\beta} \left| \sum_{n=0}^{\infty} c_n \frac{J_n'(\beta a)}{H_n^{(2)}(\beta a)} \cos(n\phi) \right|^2 = \frac{2\pi}{\beta} \left| \sum_{n=0}^{\infty} c_n \frac{J_n'(\beta a)}{H_n^{(2)}(\beta a)} \cos(n\phi) \right|^2$$

where $c_n = \begin{cases} 1 & n=0 \\ 2 & n \neq 0 \end{cases}$

[11.43] According to (11-110b) and (11-110c) we see that the ρ component of the scattered electric field has a $1/\rho$ variation outside the summation while the ϕ component has no ρ variations outside the summation. Within the summation both have the same ρ variations. Therefore we can approximate the far zone scattered electric field by

$$\underline{E}^s = \hat{a}_\rho E_\rho^s + \hat{a}_\phi E_\phi^s \xrightarrow{\rho \gg \lambda} \hat{a}_\phi E_\phi^s = \hat{a}_\phi \frac{\beta H_0}{j\omega \epsilon} \sum_{n=-\infty}^{+\infty} \frac{-n J_n'(\beta a)}{H_n^{(2)}(\beta a)} H_n^{(2)}(\beta \rho) e^{jn\phi}$$

Since

$$H_n^{(2)}(\beta \rho) \xrightarrow{\rho \gg \lambda} \sqrt{\frac{2j}{\pi \beta \rho}} j^n e^{-j\beta \rho}$$

$$\underline{E}^s \xrightarrow{\rho \gg \lambda} \hat{a}_\phi \frac{\beta H_0}{j\omega \epsilon} \sqrt{\frac{2j}{\pi \beta \rho}} e^{-j\beta \rho} \sum_{n=-\infty}^{+\infty} \frac{J_n'(\beta a)}{H_n^{(2)}(\beta a)} e^{jn\phi} = -\hat{a}_\phi j H_0 \sqrt{\frac{2j}{\pi \beta \rho}} e^{-j\beta \rho} \sum_{n=-\infty}^{+\infty} \frac{J_n'(\beta a)}{H_n^{(2)}(\beta a)} e^{jn\phi}$$

$$\underline{E}^s \xrightarrow{\rho \gg \lambda} -\hat{a}_\phi j H_0 \sqrt{\frac{2j}{\pi \beta \rho}} e^{-j\beta \rho} \sum_{n=-\infty}^{+\infty} \frac{J_n'(\beta a)}{H_n^{(2)}(\beta a)} e^{jn\phi}, \quad \underline{E}^i = \hat{a}_y \eta H_0$$

$$\sigma_{2-D} = \lim_{\rho \rightarrow \infty} \left\{ 2\pi \rho \frac{|\underline{E}^s|^2}{|\underline{E}^s|^2} \right\} = \lim_{\rho \rightarrow \infty} \left\{ 2\pi \rho \frac{2}{\pi \beta \rho} \left| \sum_{n=-\infty}^{+\infty} \frac{J_n'(\beta a)}{H_n^{(2)}(\beta a)} e^{jn\phi} \right|^2 \right\}$$

$$\sigma_{2-D} = \frac{4}{\beta} \left| \sum_{n=-\infty}^{+\infty} \frac{J_n'(\beta a)}{H_n^{(2)}(\beta a)} e^{jn\phi} \right|^2 = \frac{2\pi}{\beta} \left| \sum_{n=0}^{\infty} \frac{J_n'(\beta a)}{H_n^{(2)}(\beta a)} e^{jn\phi} \right|^2$$

$$11.44 \quad \underline{E}^i = (\hat{a}_y + j\hat{a}_z) e^{-j\beta x} : \text{Circular Polarization, CCW}$$

(a) $\underline{\text{TE}}^z: E_y = e^{-j\beta x}$ Because plane wave $\Rightarrow H_z^i = \frac{1}{\eta} e^{-j\beta x}$

According to (11-115), the scattered far-zone H-field is

$$H_z^s = -\frac{1}{\eta} \sqrt{\frac{\pi}{\beta}} \frac{e^{-j\beta p}}{\sqrt{p}} \sum_{n=-\infty}^{+\infty} J_n^1(\beta a) e^{jn\phi} - \frac{1}{\eta} \sqrt{\frac{\pi}{\beta}} \frac{e^{-j\beta p}}{\sqrt{p}} \sum_{n=0}^{\infty} \epsilon_n H_n^{(2)}(\beta a) \cos(n\phi)$$

and the scattered far-zone electric field is

$$E_\phi^s \approx \eta H_z^s = -\sqrt{\frac{\pi}{\beta}} \frac{e^{-j\beta p}}{\sqrt{p}} \sum_{n=-\infty}^{+\infty} J_n^1(\beta a) e^{jn\phi} = -\sqrt{\frac{\pi}{\beta}} \frac{e^{-j\beta p}}{\sqrt{p}} \sum_{n=-\infty}^{+\infty} \epsilon_n H_n^{(2)}(\beta a) \cos(n\phi)$$

$\underline{\text{TM}}^z: E_z^i = j e^{-j\beta x}$ According to (11-100), the far-zone E-field is

$$E_z^s = -j \sqrt{\frac{\pi}{\beta}} \frac{e^{-j\beta p}}{\sqrt{p}} \sum_{n=-\infty}^{+\infty} J_n^1(\beta a) e^{jn\phi} = -j \sqrt{\frac{\pi}{\beta}} \frac{e^{-j\beta p}}{\sqrt{p}} \sum_{n=0}^{\infty} \epsilon_n H_n^{(2)}(\beta a) \cos(n\phi)$$

TOTAL ($\text{TE}^z + \text{TM}^z$):

$$E^s_{\text{total}} \xrightarrow{\beta p \rightarrow \text{large}} -\sqrt{\frac{\pi}{\beta}} \frac{e^{-j\beta p}}{\sqrt{p}} \sum_{n=0}^{\infty} \epsilon_n \left[\frac{\hat{a}_\phi J_n^1(\beta a)}{H_n^{(2)}(\beta a)} + j \frac{\hat{a}_z J_n(\beta a)}{H_n^{(2)}(\beta a)} \right] \cos(n\phi)$$

$$\epsilon_n = \begin{cases} 1 & n=0 \\ 2 & n \neq 0 \end{cases}$$

- (b) • $\phi = 0^\circ$: According to (11-6): $\hat{a}_\phi = [\hat{a}_x \sin\phi + \hat{a}_y \cos\phi] = \hat{a}_y$

$$\underline{E}^s_{\text{total}} \underset{\phi=0}{\approx} \underline{E}^s_{\text{total}} = \hat{a}_y E_y^s + j \hat{a}_z E_z^s \underset{\phi=0}{\approx} \hat{a}_y E_y^s$$

$$(\text{forward direction}) \underset{\text{direction}}{\approx} -j \sqrt{\frac{\pi}{\beta}} \frac{e^{-j\beta p}}{\sqrt{p}} \sum_{n=0}^{\infty} \epsilon_n \left[\frac{\hat{a}_y J_n^1(\beta a)}{H_n^{(2)}(\beta a)} + j \frac{\hat{a}_z J_n(\beta a)}{H_n^{(2)}(\beta a)} \right]$$

∴ Elliptical Polarization, CCW which agrees with what is expected (see Figure 5.16, Section 5.6)

- $\phi = 180^\circ$: According to (11-6): $\hat{a}_\phi = [\hat{a}_x \sin\phi + \hat{a}_y \cos\phi] = -\hat{a}_y$

$$\underline{E}^s_{\text{total}} \underset{\phi=180^\circ}{\approx} \underline{E}^s_{\text{total}} = \hat{a}_y E_y^s + j \hat{a}_z E_z^s$$

Backscattered direction

$$\approx -j \sqrt{\frac{\pi}{\beta}} \frac{e^{-j\beta p}}{\sqrt{p}} \sum_{n=0}^{\infty} (-1)^n \epsilon_n \left[\frac{\hat{a}_y J_n^1(\beta a)}{H_n^{(2)}(\beta a)} + j \frac{\hat{a}_z J_n(\beta a)}{H_n^{(2)}(\beta a)} \right]$$

∴ Elliptical Polarization, CW which agrees with what is expected (See Figure 5.16, Section 5.6 discussion).

$$\underline{E}^i = (\hat{a}_y + j\hat{a}_z) e^{-j\beta x} : \text{Circular Polarization, CCW}$$

11.45 Based on duality, the scattering characteristics

(a) of a PMC cylinder can be obtained from those of a PEC cylinder. Accordingly, the:

TM^z scattering from a PMC cylinder is the same as the TE^z scattering from a PEC cylinder.

See solution of Problem 11.56.

TE^z scattering from a PMC cylinder is the same as the TM^z scattering from a PEC cylinder.

See solution of Problem 11.57.

Based upon that, and without rederiving the expressions of the scattered field but rather using duality and referring to the solution of Problem 11.44, we can write the far zone scattered field from a PMC cylinder of circular cross section and radius a , as:

TOTAL ($\text{TE}^z + \text{TM}^z$):

$$\underline{E}^s(\text{total}) \xrightarrow[\phi=0]{\text{forward}} + \frac{j\sqrt{\beta}}{\pi\beta} \frac{e^{-j\beta p}}{\sqrt{p}} \sum_{n=0}^{\infty} \underline{\epsilon}_n \left[\hat{a}_y \frac{J_n(\beta a)}{H_n^{(2)}(\beta a)} + j\hat{a}_z \frac{J'_n(\beta a)}{H_n^{(2)}(\beta a)} \right] \cos(n\phi)$$

(b) $\phi = 0^\circ$: According to (11-6): $\hat{a}_\phi = [-\hat{a}_x \sin\phi + \hat{a}_y \cos\phi] \Big|_{\phi=0} = \hat{a}_y$

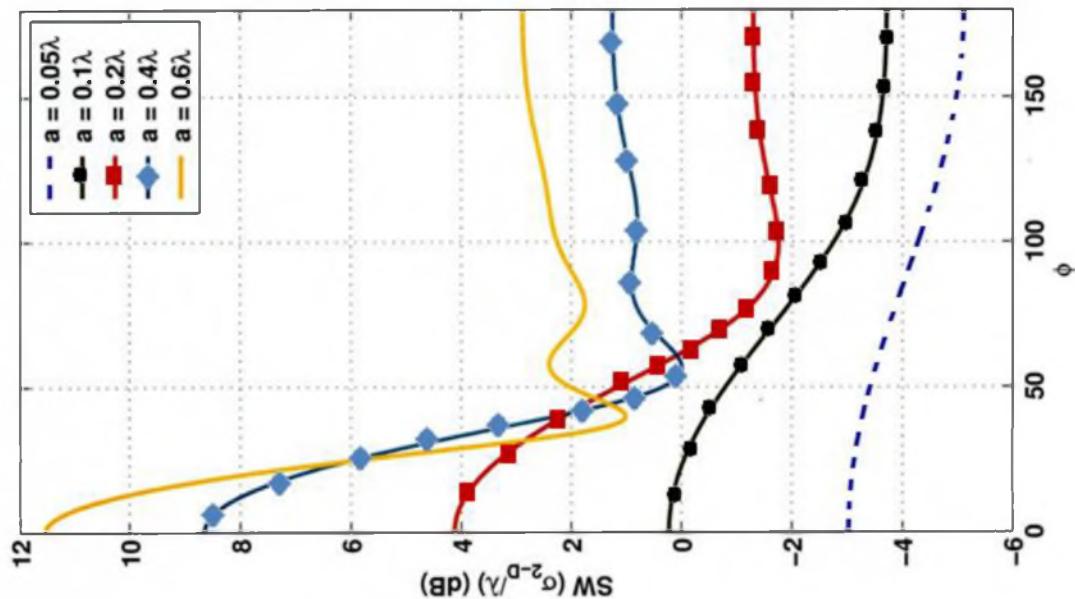
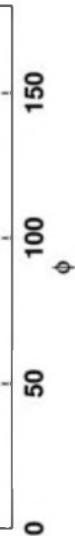
$$\underline{E}^s(\text{total}) \Big|_{\phi=0} \xrightarrow[\text{(forward direction)}]{\phi=0} \sim j \frac{j\sqrt{\beta}}{\pi\beta} \frac{e^{-j\beta p}}{\sqrt{p}} \sum_{n=0}^{\infty} \underline{\epsilon}_n \left[\hat{a}_y \frac{J_n(\beta a)}{H_n^{(2)}(\beta a)} + j\hat{a}_z \frac{J'_n(\beta a)}{H_n^{(2)}(\beta a)} \right]$$

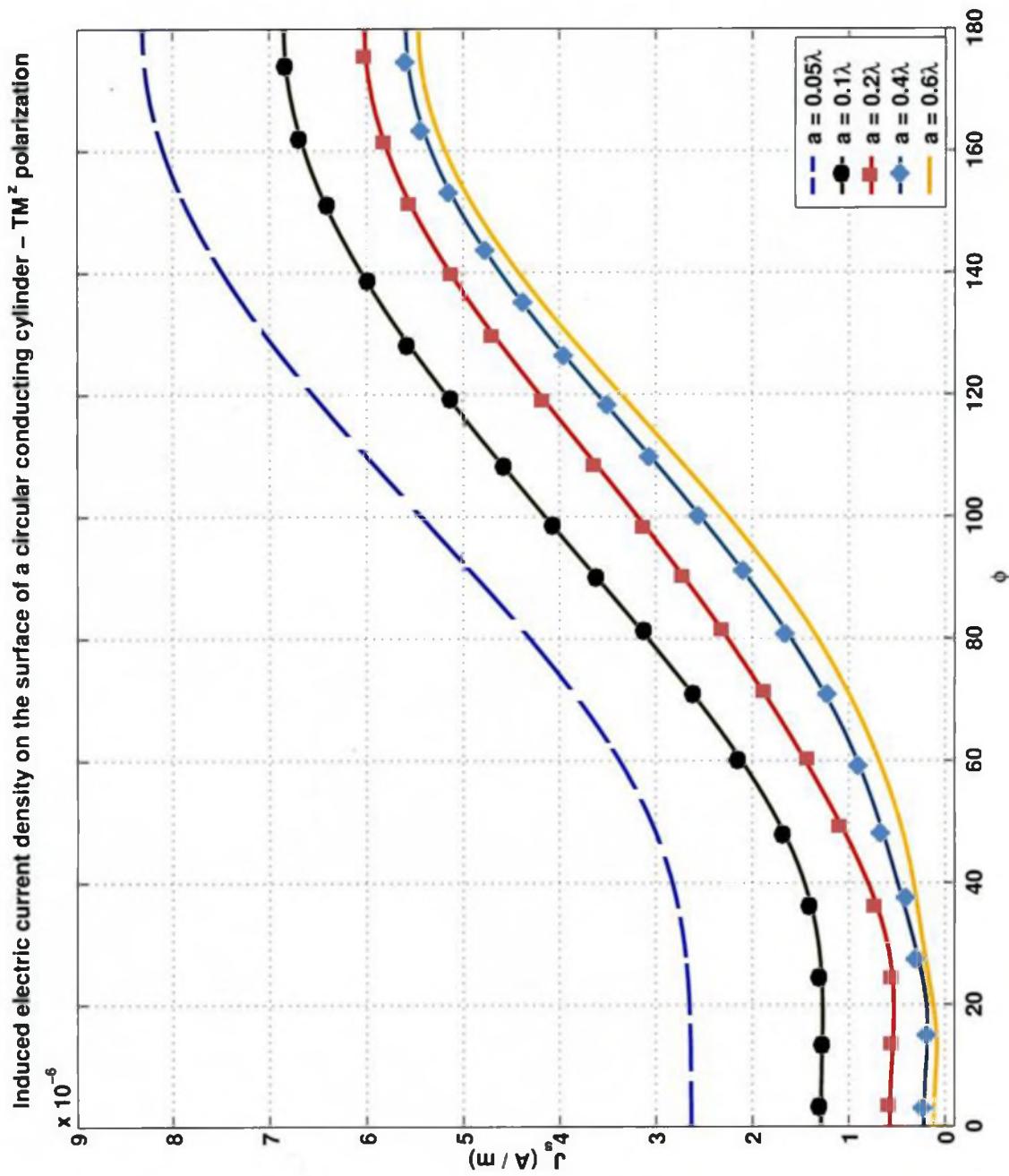
\therefore Elliptical Polarization, CCW which with what is expected (see Figure 5.16, Section 5.6 discussion).

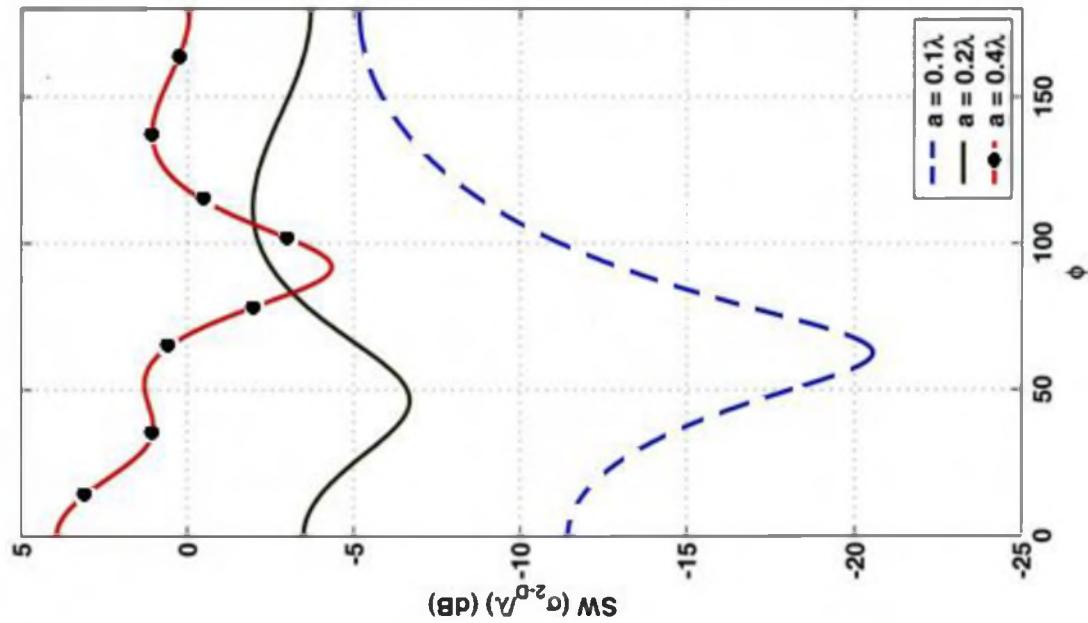
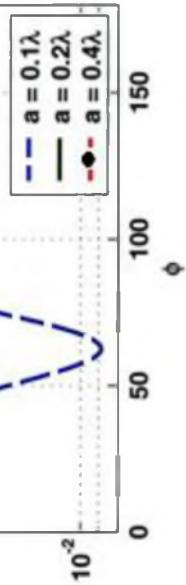
$\phi = 180^\circ$: According to (11-6): $\hat{a}_\phi = [-\hat{a}_x \sin\phi + \hat{a}_y \cos\phi] \Big|_{\phi=180^\circ} = -\hat{a}_y$

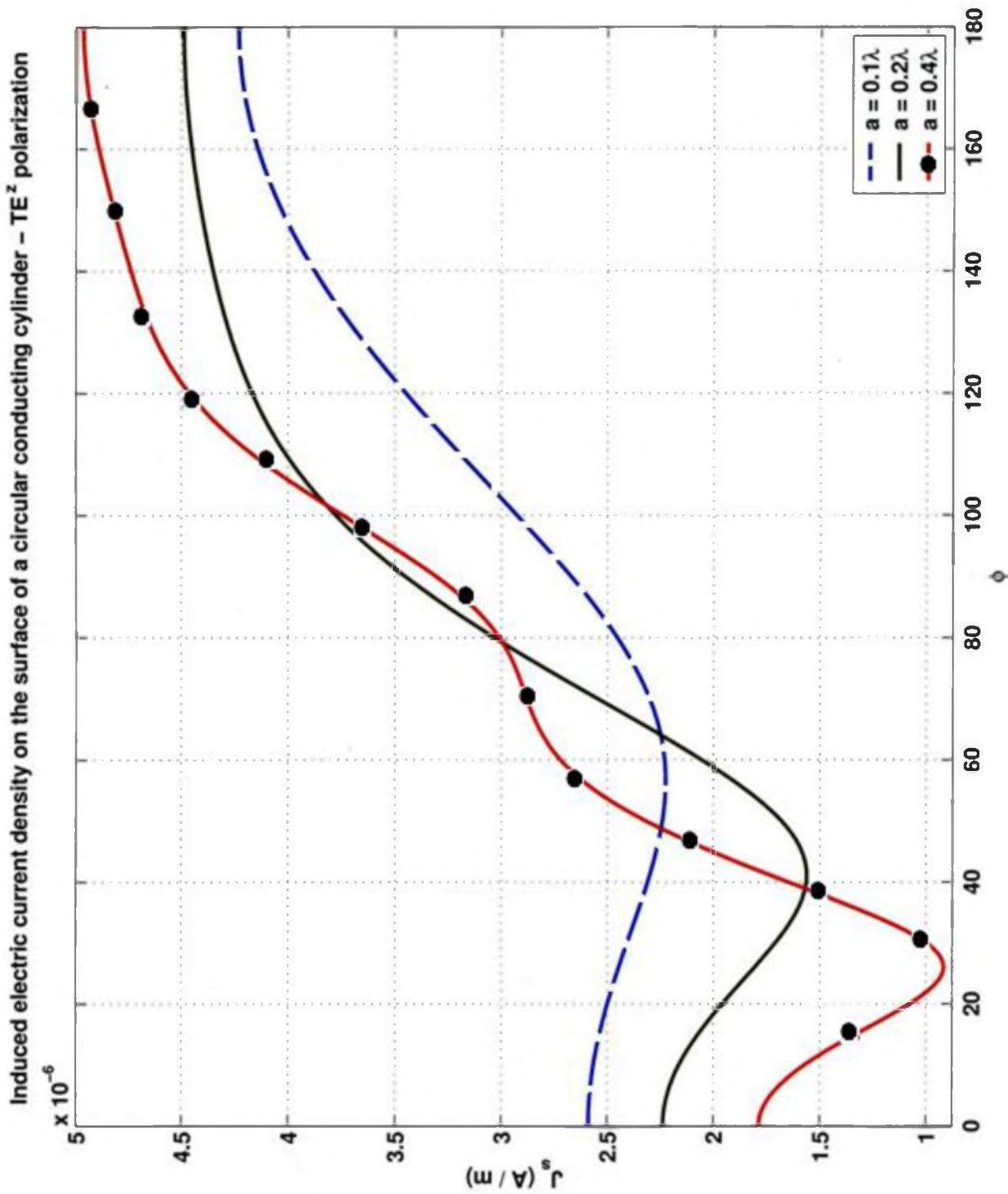
$$\underline{E}^s(\text{total}) \Big|_{\phi=180^\circ} \xrightarrow[\text{Backscattered direction}]{\phi=180^\circ} \sim +j \frac{j\sqrt{\beta}}{\pi\beta} \frac{e^{-j\beta p}}{\sqrt{p}} \sum_{n=0}^{\infty} \underline{\epsilon}_n \left[-\hat{a}_y \frac{J_n(\beta a)}{H_n^{(2)}(\beta a)} + j\hat{a}_z \frac{J'_n(\beta a)}{H_n^{(2)}(\beta a)} \right]$$

\therefore Elliptical Polarization, CW which agrees with what is expected (see Figure 5.16, Section 5.6 discussion).

Two-dimensional TM^z bistatic scattering width (SW) of a circular conducting cylinderSW ($\text{g}^{2-\text{D}}/\lambda$) (dimensionless)



Two-dimensional TE^z bistatic scattering width (SW) of a circular conducting cylinderSW (σ_{2D}/λ) (dimensionless)



[11.50] For a small ($a \ll \lambda$) PEC cylinder the bistatic scattering TM^2 width (SW) is given by (11-103a), or

$$\sigma_{2-D} \stackrel{a \ll \lambda}{\approx} \frac{\pi^2}{2} \left| \frac{1}{\ln(0.89\beta a)} \right|^2$$

which is independent of the observation angle ϕ . Therefore there is no maximum or minimum on the pattern based on this approximate expression.

However based on the results of Figure 11.46, there seems to be a

- (a) Maximum at $\phi = 0^\circ$.
- (b) Minimum at $\phi = 180^\circ$.

[11.51] For a small ($a \ll \lambda$) PEC cylinder the bistatic TE^2 scattering width (SW) is given by (11-118d), or

$$\sigma_{2-D} \stackrel{a \ll \lambda}{\approx} \frac{\pi^2}{8} (\beta a)^2 \left[1 - 2 \cos(\phi) \right]^2$$

- (a) Its maximum value occurs when

$$|1 - 2 \cos(\phi)| = 3 \Rightarrow \cos\phi = -1 \Rightarrow \phi = 180^\circ$$

This seems to be confirmed by the curves of Figure 11.48.

- (b) Its minimum value occurs when

$$|1 - 2 \cos(\phi)| = 0 \Rightarrow 2 \cos\phi = 1 \Rightarrow \cos\phi = 1/2$$

$$\phi = \cos^{-1}(1/2) = 60^\circ$$

This seems to be confirmed by the curves of Fig. 11.48.

11.52 TM^z :

For a small ($a \ll \lambda$) PEC cylinder the TM^z bistatic SW is given by

$$\sigma_{2-D} \sim \frac{\pi}{2} \left| \frac{1}{\ln(0.89\beta a)} \right|^2$$

which is independent of the observation angle ϕ . To maintain the normalized SW ($\sigma/2$) to less than -20 dB (0.01 dimensionless), the radius must be such that

$$\frac{\sigma}{2} = \frac{\pi}{2} \left| \frac{1}{\ln(0.89\beta a)} \right|^2 = \frac{1}{100} \Rightarrow \left| \ln(0.89\beta a) \right| = \pm 12.533$$

Choosing

$$\ln(0.89\beta a) = -12.533 \Rightarrow 0.89\beta a = 3.6053 \times 10^{-6} \Rightarrow a = 6.4472 \times 10^{-7}$$

11.53 TE^z :

For a small ($a \ll \lambda$) PEC cylinder the TE^z bistatic SW is given by

$$\sigma_{2-D} \left| \max \frac{\pi^2 (\beta a)^2 [1 - 2\cos(\phi)]}{8} \right| = \frac{\pi^2 (\beta a)^2 (3)}{8} @ \phi = 180^\circ$$

To maintain the normalized SW ($\sigma/2$) to less than -20 dB (0.01 dimensionless), the radius must be such that

$$\frac{\sigma}{2} = \frac{\pi^2 (\beta a)^2 (3)}{8} = 0.01 \Rightarrow \beta a = \sqrt{\frac{8}{300\pi}} \Rightarrow a = \frac{1}{2\pi} \sqrt{\frac{8}{300\pi}}$$

$$a = 0.0147\lambda$$

11.54 TM^z :

Based on the solution of Problem 11.52, to main the normalized SW ($\sigma/2$) to less than -10 dB (0.1 dimensionless), the radius must be

$$\frac{\sigma}{2} = \frac{\pi}{2} \left| \frac{1}{\ln(0.89\beta a)} \right|^2 = \frac{1}{10} \Rightarrow \left| \ln(0.89\beta a) \right| = \pm 3.9633$$

Choosing $\ln(0.89\beta a) = -3.9633 \Rightarrow 0.89\beta a = 0.019 \Rightarrow \beta a = \frac{0.019}{0.89}$

$$a = \frac{1}{2\pi} \left(\frac{0.019}{0.89} \right) = 3.398 \times 10^{-3}$$

11.55 TE^z :

Based on the solution of Problem 11.53, to maintain the normalized SW ($\sigma/2$) to less than -10 dB (0.1 dimensionless), the radius must be

$$\frac{\sigma}{2} = \frac{\pi}{2} \left| \max \frac{\pi^2 (\beta a)^2 [1 - 2\cos\phi]}{8} \right| = \frac{\pi^2 (\beta a)^2 (3)}{8} = 0.1 @ \phi = 180^\circ$$

$$\beta a = \sqrt{\frac{8}{300\pi}} \Rightarrow a = \frac{1}{2\pi} \sqrt{\frac{8}{300\pi}} = 4.637 \times 10^{-2}$$

$$11.56 \quad TM^z: \quad E^t = \hat{a}_z e^{-j\beta x} E_0 = \hat{a}_z E_0 e^{-j\beta p \cos \phi}$$

Using duality, the magnetic fields are obtained from (11-111a) – (11-111d). We can write that (from the TE^z PEC and using duality)

$$H_z^t = 0$$

$$H_p^t = \frac{H_0}{j\omega\mu} \sum_{n=-\infty}^{+\infty} n j^{-n+1} \left[J_n(\beta p) - \frac{J_n'(Ba)}{H_n^{(2)'}(Ba)} H_n^{(2)}(\beta p) \right] e^{jn\phi}$$

$$H_\phi^t = -\frac{\beta H_0}{j\omega\mu} \sum_{n=-\infty}^{+\infty} j^{-n} \left[J_n'(\beta p) - \frac{J_n'(Ba)}{H_n^{(2)'}(Ba)} H_n^{(2)}(\beta p) \right] e^{-jn\phi}$$

(a) The 2-D RCS(SW) for the TM^z PMC surface is the same, due to duality, as that of the TE^z PEC surface, or from (11-117)

$$\sigma_{2-D} = \frac{4}{\beta} \left| \sum_{n=-\infty}^{+\infty} \frac{J_n'(Ba)}{H_n^{(2)'}(Ba)} e^{jn\phi} \right|^2 = \frac{2\pi}{\pi} \left| \sum_{n=0}^{\infty} \epsilon_n \frac{J_n'(Ba)}{H_n^{(2)'}(Ba)} \cos(n\phi) \right|^2$$

$$\epsilon_n = \begin{cases} 1 & n=0 \\ 2 & n \neq 0 \end{cases}$$

(b) The current density on the surface of the cylinder ($p=a$), it is given by

$$J_s = \hat{n} \times \underline{H}^t \Big|_{p=a} = \hat{a}_p \times (\hat{a}_p H_p^t + \hat{a}_\phi H_\phi^t) \Big|_{p=a} = \hat{a}_z H_\phi^t \Big|_{p=a}$$

$$J_s = \hat{a}_z H_\phi^t(p=a) = \hat{a}_z \left[-\frac{\beta H_0}{j\omega\mu} \sum_{n=-\infty}^{+\infty} j^{-n} \left[J_n'(\beta a) - \frac{J_n'(Ba)}{H_n^{(2)'}(Ba)} H_n^{(2)}(\beta a) \right] e^{jn\phi} \right]$$

$$= \hat{a}_z \left\{ -\frac{\beta H_0}{j\omega\mu} \sum_{n=-\infty}^{+\infty} j^{-n} \left[J_n'(\beta a) H_n^{(2)'}(Ba) - J_n'(Ba) H_n^{(2)'}(\beta a) \right] e^{jn\phi} \right\} = 0$$

as it should be because the tangential \underline{H} field vanishes next to a PMC surface; i.e., PMC surface cannot support an electric current density.

$$11.57 \quad TE^z: \underline{H}^t = \hat{a}_z H_0 \sum_{n=-\infty}^{+\infty} \int^{-n} J_n(\beta p) e^{j n \phi}$$

Using duality, the magnetic fields are obtained from (11-95a) - (11-95d). We can write that (from the TM^z PEC and using duality)

$$\underline{H}_p^t = 0 = \underline{H}_\phi^t$$

$$\underline{H}_z^t = H_0 \sum_{n=-\infty}^{+\infty} \int^{-n} \left[J_n(\beta p) - \frac{J_n(\beta a)}{H_n^{(2)}(\beta a)} H_n^{(2)}(\beta p) \right] e^{j n \phi}$$

(a) The 2-D RCS (sw) for the TE^z PMC surface is the same, due to duality, as that of the TM^z PEC surface, or from (11-102)

$$R_{2-D} = \frac{4}{\beta} \left| \sum_{n=-\infty}^{+\infty} \frac{J_n(\beta a)}{H_n^{(2)}(\beta a)} e^{j n \phi} \right|^2 = \frac{29}{\pi} \left| \sum_{n=0}^{\infty} \epsilon_n \frac{J_n(\beta a)}{H_n^{(2)}(\beta a)} \cos(n\phi) \right|^2$$

$$\epsilon_n = \begin{cases} 1 & n=0 \\ 2 & n \neq 0 \end{cases}$$

(b) The current density on the surface of the cylinder ($p=a$), it is given by

$$\begin{aligned} \underline{J}_S &= \hat{n} \times \underline{H}^t \Big|_{p=a} = \hat{a}_p \times (\hat{a}_z \underline{H}_z^t) \Big|_{p=a} = -\hat{a}_\phi \underline{H}_z^t \Big|_{p=a} \\ &= -\hat{a}_\phi \underline{H}_z^t (p=a) = -\hat{a}_\phi \left[H_0 \sum_{n=-\infty}^{+\infty} \int^{-n} \left[J_n(\beta a) - \frac{J_n(\beta a)}{H_n^{(2)}(\beta a)} H_n^{(2)}(\beta a) \right] e^{j n \phi} \right] \\ &= -\hat{a}_\phi \left[H_0 \sum_{n=-\infty}^{+\infty} \frac{\left[J_n(\beta a) H_n^{(2)}(\beta a) - J_n(\beta a) H_n^{(2)}(\beta a) \right]}{H_n^{(2)}(\beta a)} e^{j n \phi} \right] = 0 \end{aligned}$$

as it should be because the tangential \underline{H} field vanishes next to a PMC surface; i.e., PMC surface cannot support an electric current density.

$$[11.58] \quad \underline{E}^L = \hat{a}_z E_0 \sum_{n=-\infty}^{+\infty} \int^{-n} J_n(\beta_0 p) e^{jn\phi} \quad p \geq a$$

$$\underline{E}^S = \hat{a}_z E_0 \sum_{n=-\infty}^{+\infty} a_n H_n^{(2)}(\beta_0 p) e^{jn\phi} \quad p \geq a$$

$$\underline{E}^d = \hat{a}_z E_0 \sum_{n=-\infty}^{+\infty} [b_n J_n(\beta_1 p) + c_n Y_n(\beta_1 p)] e^{jn\phi} \quad a \geq p \geq 0$$

a. $\nabla \times \underline{E} = -j\omega \mu \underline{H} \Rightarrow \underline{H} = -\frac{1}{j\omega \mu} \nabla \times \underline{E} = -\frac{1}{j\omega \mu} \left\{ \hat{a}_p \frac{1}{p} \frac{\partial E_z}{\partial \phi} - \hat{a}_\phi \frac{\partial E_z}{\partial p} \right\}$

$$H_p = -\frac{1}{j\omega \mu} \frac{1}{p} \frac{\partial E_z}{\partial \phi}, \quad H_\phi = \frac{1}{j\omega \mu} \frac{\partial E_z}{\partial p}$$

Therefore

$$H_p^L = -\frac{E_0}{j\omega \mu_0} \frac{1}{p} \sum_{n=-\infty}^{+\infty} n j^{-n+1} J_n(\beta_0 p) e^{jn\phi}$$

$$H_\phi^L = \frac{E_0 \beta_0}{j\omega \mu_0} \sum_{n=-\infty}^{+\infty} j^{-n} J_n'(\beta_0 p) e^{jn\phi}$$

$$H_p^S = -\frac{E_0}{j\omega \mu_0} \frac{1}{p} \sum_{n=-\infty}^{+\infty} j n a_n H_n^{(2)}(\beta_0 p) e^{jn\phi}$$

$$H_\phi^S = \frac{E_0 \beta_0}{j\omega \mu_0} \sum_{n=-\infty}^{+\infty} a_n H_n^{(2)}(\beta_0 p) e^{jn\phi}$$

$$H_p^d = -\frac{E_0}{j\omega \mu_1} \frac{1}{p} \sum_{n=-\infty}^{+\infty} j n [b_n J_n(\beta_1 p) + c_n Y_n(\beta_1 p)] e^{jn\phi}$$

$$H_\phi^d = \frac{E_0 \beta_1}{j\omega \mu_1} \sum_{n=-\infty}^{+\infty} [b_n J_n'(\beta_1 p) + c_n Y_n'(\beta_1 p)] e^{jn\phi}$$

- b. Since the fields within the dielectric cylinder must be finite everywhere (including $p=a$), then $c_n=0$. Matching tangential components of electric and magnetic fields at $p=a$, leads to
 $(E_z^L + E_z^S)|_{p=a} = E_z^d|_{p=a}; \quad (H_\phi^L + H_\phi^S)|_{p=a} = H_\phi^d|_{p=a}$

$$\sum_{n=-\infty}^{+\infty} [j^{-n} J_n(\beta_0 a) + a_n H_n^{(2)}(\beta_0 a)] e^{jn\phi} = \sum_{n=-\infty}^{+\infty} b_n J_n(\beta_1 a) e^{jn\phi}$$

$$\frac{\beta_0}{\beta_1} \sum_{n=-\infty}^{+\infty} [j^{-n} J_n'(\beta_0 a) + a_n H_n^{(2)\prime}(\beta_0 a)] = \frac{\beta_1}{\beta_0} \sum_{n=-\infty}^{+\infty} b_n J_n'(\beta_1 a) e^{jn\phi}$$

$$a_n H_n^{(2)}(\beta_0 a) - b_n J_n(\beta_1 a) = -j^{-n} J_n(\beta_0 a)$$

$$a_n H_n^{(2)\prime}(\beta_0 a) - b_n \frac{\beta_1}{\beta_0} j^{-n} J_n'(\beta_1 a) = -j^{-n} J_n'(\beta_0 a)$$

or

$$a_n H_n^{(2)}(\beta_0 a) - b_n J_n(\beta_1 a) = -j^{-n} J_n(\beta_0 a) \quad (1)$$

$$a_n H_n^{(2)\prime}(\beta_0 a) - b_n \sqrt{\frac{\epsilon_r}{\mu_r}} J_n'(\beta_1 a) = -j^{-n} J_n'(\beta_0 a) \quad (2)$$

cont'd.

11.58 cont'd. Multiplying (1) by $\sqrt{\frac{\epsilon_r}{\mu_r}} J_n'(\beta_0 a)$ and (2) by $J_n(\beta_0 a)$, we get

$$a_n \sqrt{\frac{\epsilon_r}{\mu_r}} H_n^{(2)}(\beta_0 a) J_n'(\beta_0 a) - b_n \sqrt{\frac{\epsilon_r}{\mu_r}} J_n(\beta_0 a) J_n'(\beta_0 a) = -j^{-n} J_n(\beta_0 a) J_n'(\beta_0 a) \sqrt{\frac{\epsilon_r}{\mu_r}} \quad (3)$$

$$a_n H_n^{(2)\prime}(\beta_0 a) J_n(\beta_0 a) - b_n \sqrt{\frac{\epsilon_r}{\mu_r}} J_n(\beta_0 a) J_n'(\beta_0 a) = -j^{-n} J_n'(\beta_0 a) J_n(\beta_0 a) \quad (4)$$

Subtracting (4) from (3) leads to

$$a_n \left[\sqrt{\frac{\epsilon_r}{\mu_r}} H_n^{(2)}(\beta_0 a) J_n'(\beta_0 a) - H_n^{(2)\prime}(\beta_0 a) J_n(\beta_0 a) \right] = -j^{-n} \left[\sqrt{\frac{\epsilon_r}{\mu_r}} J_n(\beta_0 a) J_n'(\beta_0 a) - J_n'(\beta_0 a) J_n(\beta_0 a) \right]$$

$$a_n = j^{-n} \frac{J_n'(\beta_0 a) J_n(\beta_0 a) - \sqrt{\frac{\epsilon_r}{\mu_r}} J_n(\beta_0 a) J_n'(\beta_0 a)}{\sqrt{\frac{\epsilon_r}{\mu_r}} J_n'(\beta_0 a) H_n^{(2)\prime}(\beta_0 a) - J_n(\beta_0 a) H_n^{(2)}(\beta_0 a)} \quad (5)$$

Multiplying (1) by $H_n^{(2)\prime}(\beta_0 a)$ and (2) by $H_n^{(2)}(\beta_0 a)$, we get

$$a_n H_n^{(2)}(\beta_0 a) H_n^{(2)\prime}(\beta_0 a) - b_n J_n(\beta_0 a) H_n^{(2)\prime}(\beta_0 a) = -j^{-n} J_n(\beta_0 a) H_n^{(2)\prime}(\beta_0 a) \quad (6)$$

$$a_n H_n^{(2)}(\beta_0 a) H_n^{(2)\prime}(\beta_0 a) - b_n \sqrt{\frac{\epsilon_r}{\mu_r}} J_n'(\beta_0 a) H_n^{(2)}(\beta_0 a) = -j^{-n} J_n'(\beta_0 a) H_n^{(2)}(\beta_0 a) \quad (7)$$

Subtracting (7) from (6) leads to

$$b_n \left[\sqrt{\frac{\epsilon_r}{\mu_r}} J_n'(\beta_0 a) H_n^{(2)\prime}(\beta_0 a) - J_n(\beta_0 a) H_n^{(2)\prime}(\beta_0 a) \right] = j^{-n} \left[J_n'(\beta_0 a) H_n^{(2)}(\beta_0 a) - J_n(\beta_0 a) H_n^{(2)\prime}(\beta_0 a) \right]$$

$$b_n = j^{-n} \frac{J_n(\beta_0 a) H_n^{(2)\prime}(\beta_0 a) - J_n'(\beta_0 a) H_n^{(2)}(\beta_0 a)}{J_n(\beta_0 a) H_n^{(2)\prime}(\beta_0 a) - \sqrt{\frac{\epsilon_r}{\mu_r}} J_n'(\beta_0 a) H_n^{(2)}(\beta_0 a)} \quad (8)$$

11.59 c. $\sigma_{2D} = \lim_{p \rightarrow \infty} \left[z \eta p \frac{|E_z^s|^2}{|E_z^u|^2} \right]$

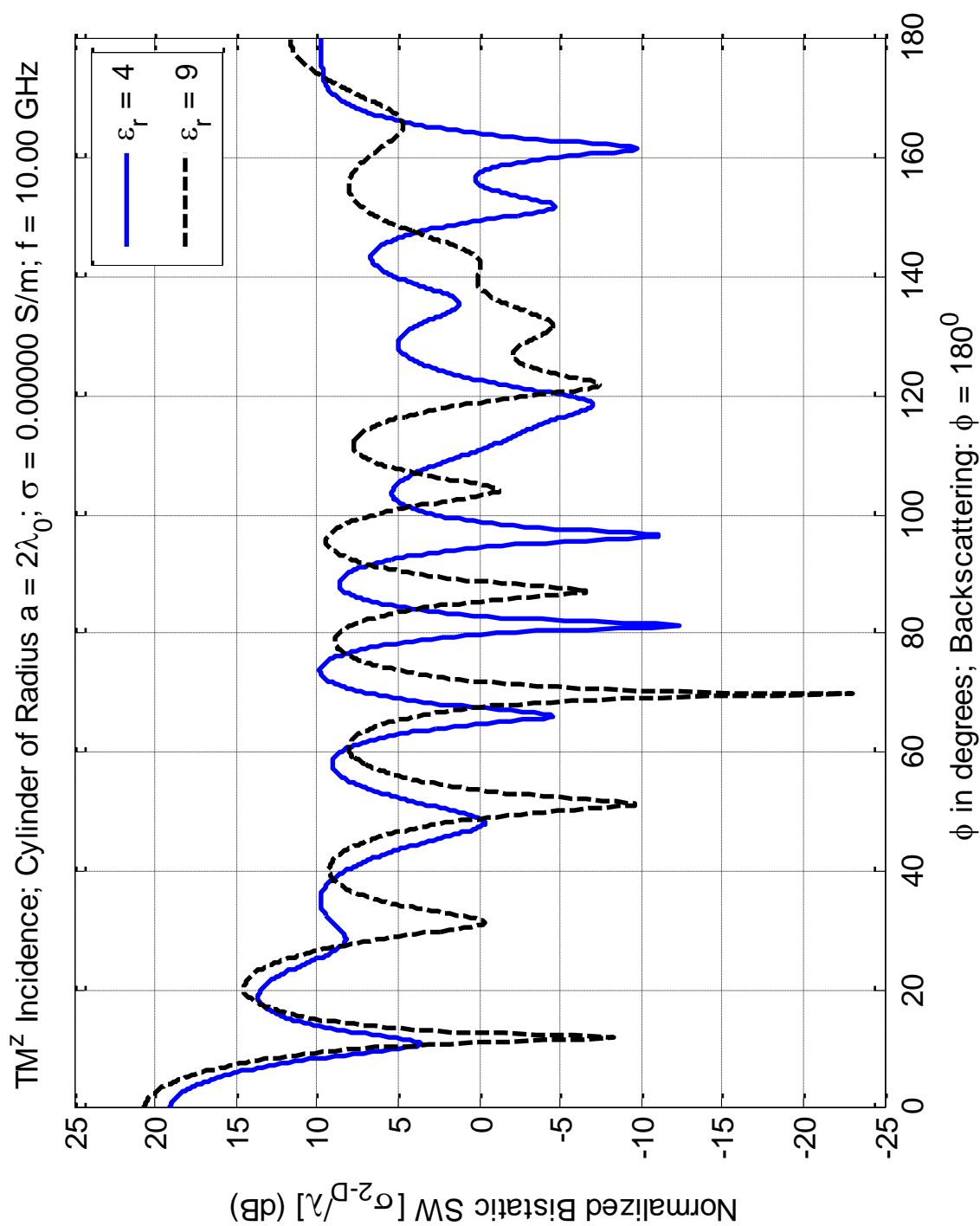
$$E_z^s = E_0 \sum_{n=-\infty}^{+\infty} a_n H_n^{(2)}(\beta_0 p) e^{jn\phi} \stackrel{p \rightarrow \infty}{\sim} E_0 \sum_{n=-\infty}^{+\infty} a_n \sqrt{\frac{2j}{\pi \beta_0 p}} j e^{jn\phi} e^{j\phi}$$

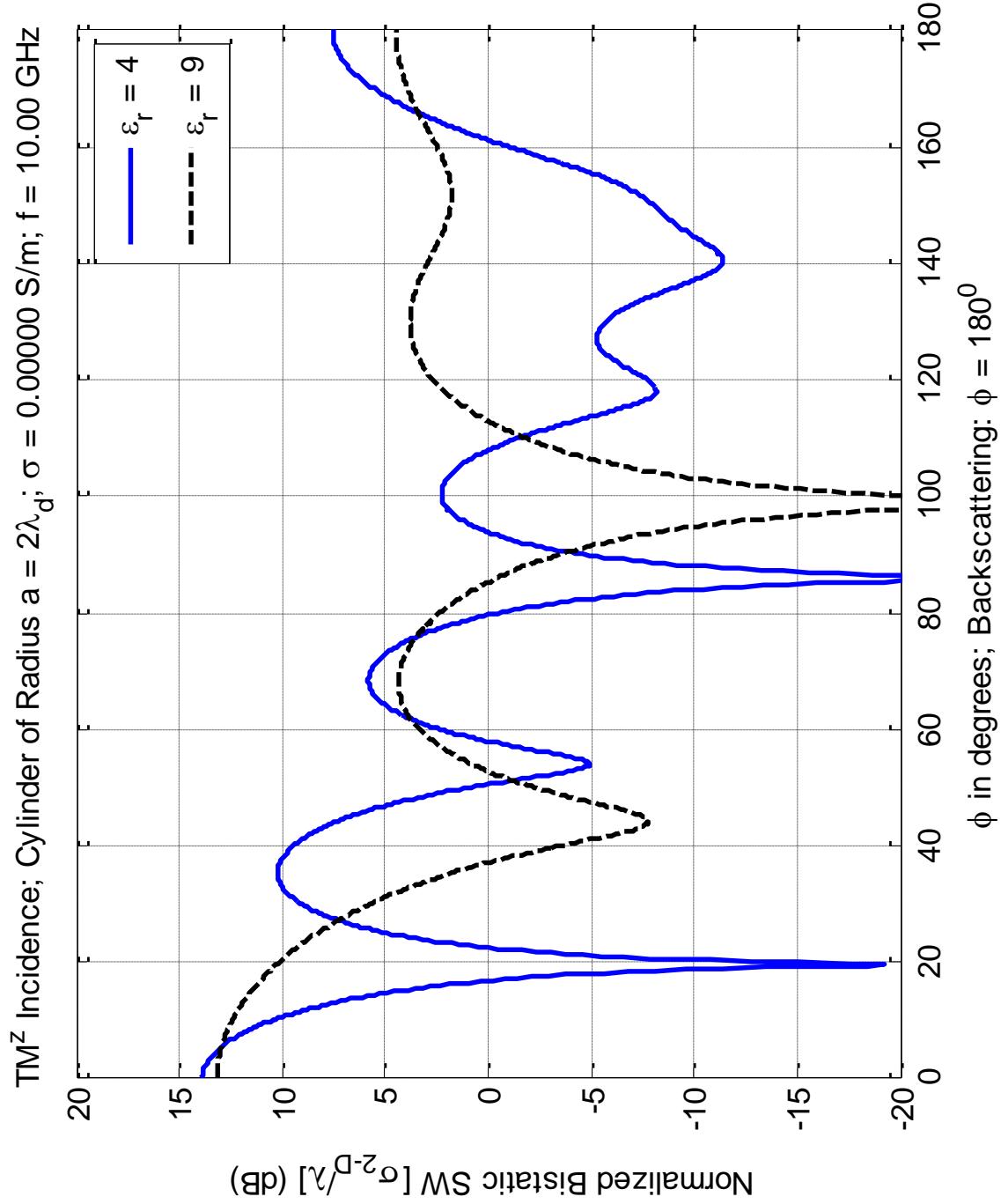
$$= E_0 \sqrt{\frac{2j}{\pi \beta_0 p}} e^{-j\beta_0 p} \sum_{n=-\infty}^{+\infty} a_n j^n e^{jn\phi}$$

$$E_z^u = E_0 \sqrt{\frac{2j}{\pi \beta_0 p}} e^{-j\beta_0 p} \sum_{n=0}^{+\infty} \varepsilon_n a_n j^n \cos(n\phi), \quad \varepsilon_n = \begin{cases} 1 & n=0 \\ 2 & n \neq 0 \end{cases}$$

$$\sigma_{2D} = \frac{4}{\beta_0} \left| \sum_{n=0}^{+\infty} \varepsilon_n a_n j^n \cos(n\phi) \right|^2 = \frac{z \lambda_0}{\pi} \left| \sum_{n=0}^{+\infty} \varepsilon_n a_n j^n \cos(n\phi) \right|^2$$

d. See attached figures.





11.60 The reflection coefficient of a plane wave from a planar interface formed by two media is given by (5-4a) (for normal incidence)

$$\Gamma = \frac{\eta_2 - \eta_1}{\eta_2 + \eta_1} = \frac{\sqrt{\mu_2/\epsilon_2} - \sqrt{\mu_1/\epsilon_1}}{\sqrt{\mu_2/\epsilon_2} + \sqrt{\mu_1/\epsilon_1}}$$

where η_2 is the intrinsic impedance of the medium the wave is impinging upon while η_1 is the intrinsic impedance of the wave is traveling in.

Assuming medium 2 has a dielectric constant $\epsilon_{r2}/\epsilon_{r1} = \epsilon_r \gg 1$, while their permeabilities are the same, the reflection coefficient can be written as

$$\Gamma = \frac{1 - \sqrt{\epsilon_2/\epsilon_1}}{1 + \sqrt{\epsilon_2/\epsilon_1}} = \frac{1 - \sqrt{\epsilon_r/\epsilon_0}}{1 + \sqrt{\epsilon_r/\epsilon_0}} = \frac{1 - \sqrt{\epsilon_r}}{1 + \sqrt{\epsilon_r}} \underset{\epsilon_r \gg 1}{\approx} -1$$

Therefore the reflecting/scattering surface has reflection characteristics similar to those of a PEC. Therefore a cylinder with $\epsilon_r \gg 1, \mu_r = 1$ has scattering characteristics similar to those of a PEC. Thus we can approximate the SW and J_s by those of a PEC.

(a) TM^z:

From (11-102), (11-102a)

$$R_{2-D} \approx \frac{4}{\beta} \left| \sum_{n=-\infty}^{+\infty} \frac{J_n(\beta a)}{H_n^{(2)}(\beta a)} e^{jnd} \right|^2 = \frac{2\pi}{\beta} \left| \sum_{n=0}^{\infty} \epsilon_n \frac{J_n(\beta a)}{H_n^{(2)}(\beta a)} \cos(n\phi) \right|^2$$

$$\epsilon_n = \begin{cases} 1 & n=0 \\ 2 & n \neq 0 \end{cases}$$

(b) TM^z:

From (11-97)

$$J_s \approx \hat{E}_z \frac{2E_0}{\pi a \omega \mu} \sum_{n=-\infty}^{+\infty} \int^n \frac{e^{jnd}}{H_n^{(2)}(\beta a)}$$

This can also be confirmed by examining the solution of Problem 11.58, especially the coefficient a_n when $\epsilon_r \gg 1, \mu_r = 1$.

11.61 When $\mu_r \gg 1$, $\epsilon_r = 1$, the reflection coefficient of the solution of Problem 11.60 reduces to ($\epsilon_1 = \epsilon_2$)

$$\Gamma = \frac{\sqrt{\mu_2/\epsilon_2} - \sqrt{\mu_1/\epsilon_1}}{\sqrt{\mu_2/\epsilon_2} + \sqrt{\mu_1/\epsilon_1}} = \frac{\sqrt{\mu_2/\mu_1} - 1}{\sqrt{\mu_2/\mu_1} + 1} = \frac{\sqrt{\mu_2/\mu_0} - 1}{\sqrt{\mu_2/\mu_0} + 1} = \frac{(\mu_r - 1)}{\sqrt{\mu_r + 1}} \underset{\mu_r \gg 1}{\sim} 1$$

Therefore the reflecting/scattering surface has reflection characteristics similar to those of a PMC. Therefore a cylinder with $\mu_r \gg 1$, $\epsilon_r = 1$ has scattering characteristics similar to those of a PMC. Thus we can approximate the SW and \underline{J}_s by those of a PMC.

(a) TM^Z:

From the solution of Problem 11.56 (scattering from a PMC cylinder) we can write the normalized scattering width (SW) as

$$T_{2-D} \approx \frac{4}{\beta} \left| \sum_{n=-\infty}^{+\infty} \frac{J_n'(ka)}{H_n^{(2)'}(ka)} e^{jnd} \right|^2 = \frac{2}{\pi} \left| \sum_{n=0}^{\infty} \epsilon_n \frac{J_n'(ka)}{H_n^{(2)'}(ka)} \cos(n\phi) \right|^2$$

$$\epsilon_n = \begin{cases} 1 & n=0 \\ 2 & n \neq 0 \end{cases}$$

(b) Also from the solution of Problem 11.56 (scattering from a PMC cylinder) we can write the current density \underline{J}_s on the surface ($\rho=a$) of the PMC cylinder as

$$\underline{J}_s = \hat{a}_z \left\{ \frac{\beta H_0}{j\omega \mu} \sum_{n=-\infty}^{+\infty} \int_0^\pi \left[\frac{J_n'(ka) H_n^{(2)'}(ka) - J_n'(ka) H_n^{(2)''}(ka)}{H_n^{(2)'}(ka)} \right] e^{jnd} \right\} = 0$$

as it should be because the tangential \underline{H} field vanishes on the surface of a PMC surface; PMC surface cannot support an electric current density.

This can also be confirmed by examining the solution of Problem 11.58, especially the coefficient a_n when $\mu_r \gg 1$, $\epsilon_r = 1$.

11.62

$$\underline{H}^i = \hat{a}_2 H_0 \sum_{n=-\infty}^{+\infty} j^{-n} J_n(\beta_0 \rho) e^{jn\phi} \quad \rho \geq a$$

$$\underline{H}^s = \hat{a}_2 H_0 \sum_{n=-\infty}^{+\infty} a_n H_n^{(2)}(\beta_0 \rho) e^{jn\phi} \quad \rho \geq a$$

$$\underline{H}^d = \hat{a}_2 H_0 \sum_{n=-\infty}^{+\infty} [b_n J_n(\beta_0 \rho) + c_n Y_n(\beta_0 \rho)] e^{jn\phi} \quad a \geq \rho \geq 0$$

a. $\nabla \times \underline{H} = j \omega \epsilon \underline{E} \Rightarrow \underline{E} = \frac{1}{j \omega \epsilon} \nabla \times \underline{H} = \frac{1}{j \omega \epsilon} \left\{ \hat{a}_2 \frac{1}{\rho} \frac{\partial H_z}{\partial \phi} - \hat{a}_2 \frac{\partial H_z}{\partial \rho} \right\}$

$$E_\rho = \frac{1}{j \omega \epsilon} \frac{1}{\rho} \frac{\partial H_z}{\partial \phi}, \quad E_\phi = - \frac{1}{j \omega \epsilon} \frac{\partial H_z}{\partial \rho}$$

Therefore

$$E_\rho^i = \frac{H_0}{j \omega \epsilon_0} \frac{1}{\rho} \sum_{n=-\infty}^{+\infty} n j^{-n+1} J_n(\beta_0 \rho) e^{jn\phi}$$

$$E_\phi^i = - \frac{H_0 \beta_0}{j \omega \epsilon_0} \sum_{n=-\infty}^{+\infty} j^{-n} J_n'(\beta_0 \rho) e^{jn\phi}$$

$$E_\rho^s = \frac{H_0}{j \omega \epsilon_0} \frac{1}{\rho} \sum_{n=-\infty}^{+\infty} j n a_n H_n^{(2)}(\beta_0 \rho) e^{jn\phi}$$

$$E_\phi^s = - \frac{H_0 \beta_0}{j \omega \epsilon_0} \sum_{n=-\infty}^{+\infty} a_n H_n^{(2)\prime}(\beta_0 \rho) e^{jn\phi}$$

cont'd.

11.52 cont'd.

$$E_\phi^d = \frac{H_0}{j\omega\epsilon_1} \frac{1}{\rho} \sum_{n=-\infty}^{+\infty} j n [b_n J_n(\beta_1 \rho) + c_n Y_n(\beta_1 \rho)] e^{jn\phi}$$

$$E_\phi^d = -\frac{H_0 \beta_1}{j\omega\epsilon_1} \sum_{n=-\infty}^{+\infty} [b_n J'_n(\beta_1 \rho) + c_n Y'_n(\beta_1 \rho)] e^{jn\phi}$$

b. Since the fields within the dielectric cylinder must be finite everywhere (including $\rho=0$), then $c_n=0$. Matching tangential components of electric and magnetic fields at $\rho=a$, leads to

$$(E_\phi^i + E_\phi^s)_{\rho=a} = E_\phi^d_{\rho=a} \quad ; \quad (H_z^i + H_z^s)_{\rho=a} = H_z^d_{\rho=a}$$

$$\frac{\mu_0}{\epsilon_0} \sum_{n=-\infty}^{+\infty} [j^{-n} J'_n(\beta_0 a) + a_n H_n^{(2)'}(\beta_0 a)] e^{jn\phi} = \frac{\beta_1}{\epsilon_1} \sum_{n=-\infty}^{+\infty} b_n J'_n(\beta_1 a) e^{jn\phi}$$

$$\sum_{n=-\infty}^{+\infty} [j^{-n} J_n(\beta_0 a) + a_n H_n^{(2)}(\beta_0 a)] e^{jn\phi} = \sum_{n=-\infty}^{+\infty} b_n J_n(\beta_1 a) e^{jn\phi}$$

$$a_n H_n^{(2)'}(\beta_0 a) - \frac{\beta_1}{\beta_0} \frac{\epsilon_0}{\epsilon_1} b_n J'_n(\beta_1 a) = -j^{-n} J'_n(\beta_0 a)$$

$$a_n H_n^{(2)}(\beta_0 a) - b_n J_n(\beta_1 a) = -j^{-n} J_n(\beta_0 a)$$

or $a_n H_n^{(2)'}(\beta_0 a) - \sqrt{\frac{\mu_r}{\epsilon_r}} b_n J'_n(\beta_1 a) = -j^{-n} J'_n(\beta_0 a)$ (1)

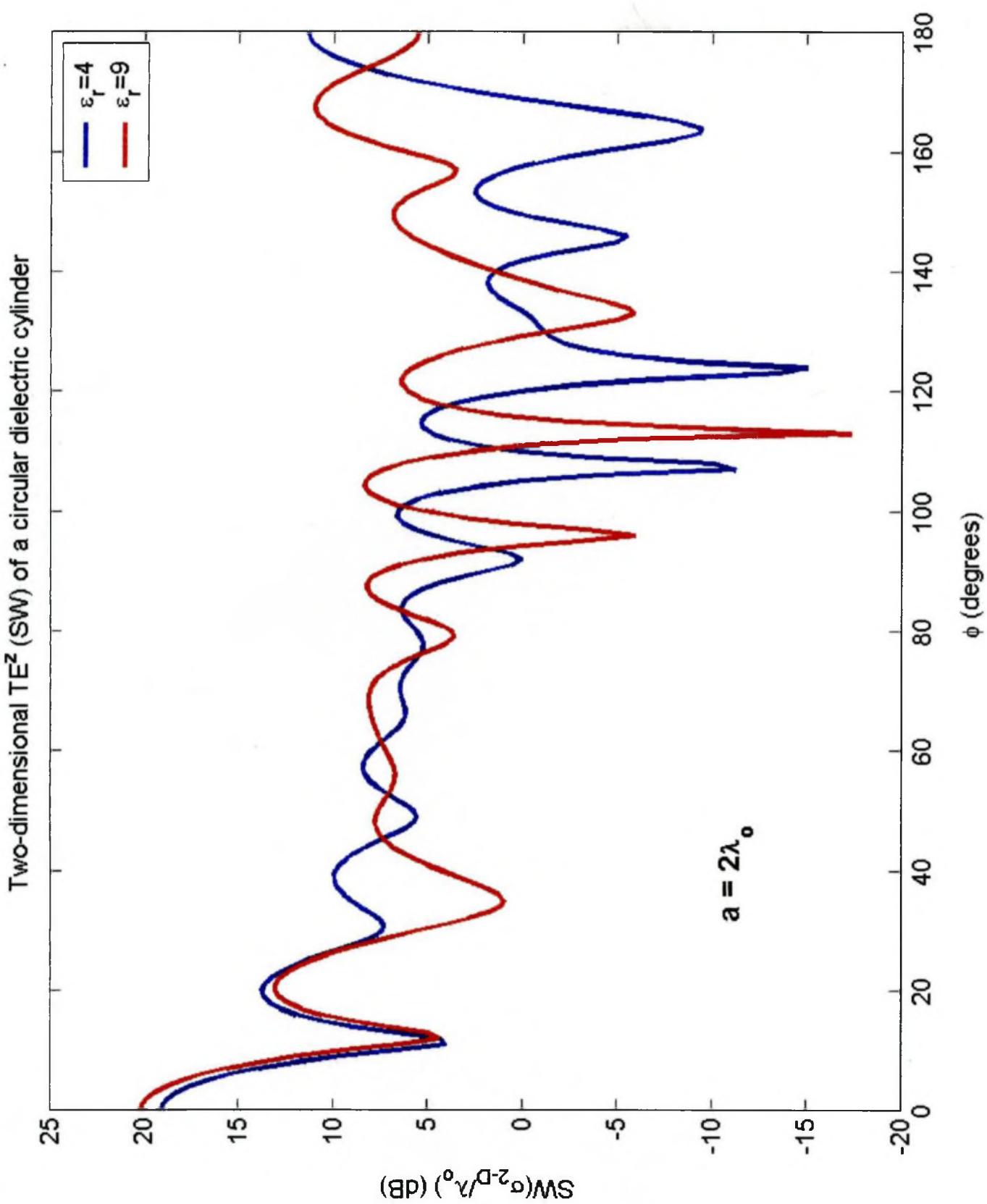
$$a_n H_n^{(2)}(\beta_0 a) - b_n J_n(\beta_1 a) = -j^{-n} J_n(\beta_0 a)$$
 (2)

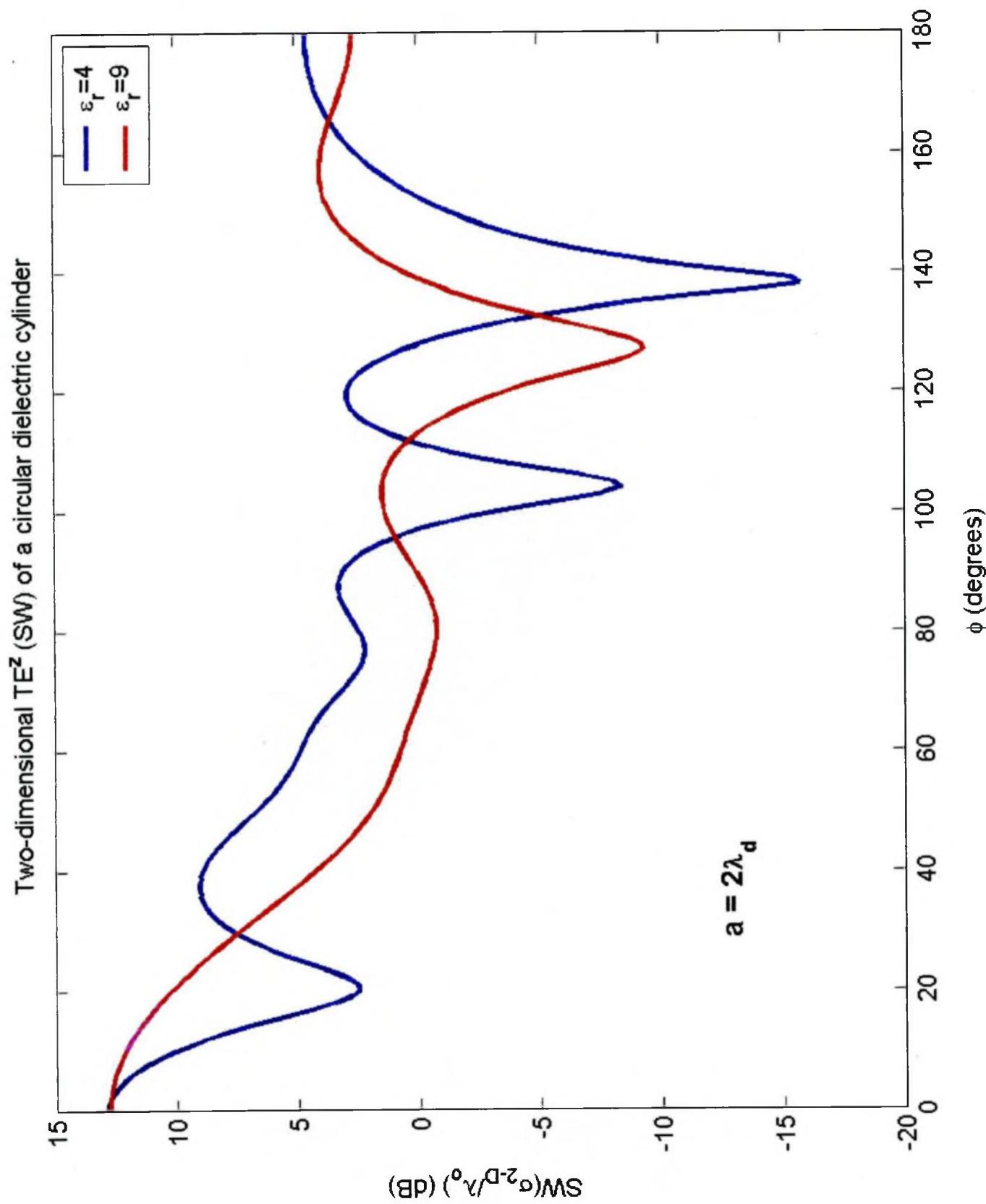
Comparing (1) and (2) of this problem solution with (1) and (2) of the solution of Problem 11.58 we see that they are identical except at μ_r should be replaced by ϵ_r and ϵ_r should be replaced with μ_r . Therefore by referring to (5) and (8) of the solution of Problem 11.58, we can write that

$$a_n = j^{-n} \frac{J'_n(\beta_0 a) J_n(\beta_1 a) - \sqrt{\mu_r/\epsilon_r} J_n(\beta_0 a) J'_n(\beta_1 a)}{\sqrt{\mu_r/\epsilon_r} J'_n(\beta_1 a) H_n^{(2)'}(\beta_0 a) - J_n(\beta_1 a) H_n^{(2)'}(\beta_0 a)} \quad (3)$$

$$b_n = j^{-n} \frac{J_n(\beta_0 a) H_n^{(2)'}(\beta_0 a) - J'_n(\beta_0 a) H_n^{(2)}(\beta_0 a)}{J_n(\beta_1 a) H_n^{(2)'}(\beta_0 a) - \sqrt{\mu_r/\epsilon_r} J'_n(\beta_1 a) H_n^{(2)}(\beta_0 a)} \quad (4)$$

11.62





11.63 The reflection coefficient of a plane wave from a planar interface from by two media, under normal wave incidence, is given by

$$\Gamma = \frac{\eta_2 - \eta_1}{\eta_2 + \eta_1} = \frac{\sqrt{\mu_2/\epsilon_2} - \sqrt{\mu_1/\epsilon_1}}{\sqrt{\mu_2/\epsilon_2} + \sqrt{\mu_1/\epsilon_1}}$$

where η_2 is the intrinsic impedance of the medium the wave is impinging upon, while η_1 is the intrinsic impedance of the wave is traveling in.

Assuming medium 2 has a dielectric constant $\epsilon_{r2}/\epsilon_1 = \epsilon_r \gg 1$, while the permeabilities are the same, the reflection coefficient can be written as

$$\Gamma = \frac{1 - \sqrt{\epsilon_r/\epsilon_0}}{1 + \sqrt{\epsilon_r/\epsilon_0}} = \frac{1 - \sqrt{\epsilon_r/\epsilon_0}}{1 + \sqrt{\epsilon_r/\epsilon_0}} = \frac{1 - \sqrt{\epsilon_r}}{\sqrt{\epsilon_r}} \frac{\epsilon_r \gg 1}{1 + \sqrt{\epsilon_r}} - 1$$

Therefore the reflecting/scattering surface has reflection characteristics similar to those of a PEC. Therefore the cylinder with $\epsilon_r \gg 1, \mu_r = 1$ has scattering characteristics similar to those of a PEC. Thus we can approximate the SW and I_s by those of a PEC.

(a) TE^z:

From (11-117), (11-117a)

$$\sigma_{2-D} \approx \frac{4}{\beta} \left| \sum_{n=-\infty}^{+\infty} \frac{J_n'(\beta a)}{H_n^{(2)'}(\beta a)} e^{jn\phi} \right|^2 = \frac{2\pi}{\beta} \left| \sum_{n=0}^{\infty} \epsilon_n \frac{J_n'(\beta a)}{H_n^{(2)'}(\beta a)} \cos(n\phi) \right|^2$$

(b) TE^x:

From (11-113)

$$I_s \approx \hat{A}_\phi j \frac{2H_0}{\pi \beta a} \sum_{n=-\infty}^{+\infty} \int_{-\pi}^{\pi} \frac{e^{jn\phi}}{H_n^{(2)'}(\beta a)} d\phi$$

This can be confirmed by examining the solution of Problem 11.62, especially the coefficient a_n when $\epsilon_r \gg 1, \mu_r = 1$.

11.64 When $\mu_r \gg 1$, $\epsilon_r = 1$, the reflection coefficient of the solution of Problem 11.63 reduces to ($\epsilon_1 = \epsilon_2$)

$$\Gamma = \frac{\sqrt{\mu_2/\epsilon_2} - \sqrt{\mu_1/\epsilon_1}}{\sqrt{\mu_2/\epsilon_2} + \sqrt{\mu_1/\epsilon_1}} = \frac{\sqrt{\mu_2/\mu_1} - 1}{\sqrt{\mu_2/\mu_1} + 1} = \frac{\sqrt{\mu_2/\mu_1} - 1}{\sqrt{\mu_2/\mu_1} + 1} = \frac{\sqrt{\mu_r - 1}}{\sqrt{\mu_r + 1}} \underset{\mu_r \gg 1}{\sim} 1$$

Therefore the reflecting/scattering surface has reflection characteristics similar to those of a PMC. Therefore a cylinder with $\mu_r \gg 1$, $\epsilon_r = 1$ has characteristics similar to those of a PMC. Thus we can approximate the SW and Js by those of a PMC.

(a) TE^z :

From the solution of Problem 11.57 (scattering from a PMC cylinder) we can write the normalized scattering width (SW) as

$$\sigma_{2-D} = \frac{4}{\beta} \left| \sum_{n=-\infty}^{+\infty} \frac{J_n(Ba)}{H_n^{(2)}(Ba)} e^{j n \phi} \right|^2 = \frac{2 \lambda}{\pi} \left| \sum_{n=0}^{\infty} \epsilon_n \frac{J_n(Ba)}{H_n^{(2)}(Ba)} \cos(n\phi) \right|^2$$

$$\epsilon_n = \begin{cases} 1 & n=0 \\ 2 & n \neq 0 \end{cases}$$

TE^z :

(b) The current density on the surface of the cylinder ($r=a$), it is given by (also from the solution of Problem 11.57; scattering of a TE^z plane wave from a PMC cylinder)

$$J_S = -\hat{a}_\phi \left\{ H_0 \sum_{n=-\infty}^{+\infty} \left[\frac{J_n(Ba) [H_n^{(2)}(Ba) - J_n(Ba) H_n^{(1)}(Ba)]}{H_n^{(2)}(Ba)} \right] e^{j n \phi} \right\} = 0$$

as it should be because the tangential H field vanishes on a PMC surface; i.e., a PMC surface cannot support an electric current density.

11.65

$$\underline{E}_z^i = \hat{a}_z E_0 \sum_{n=-\infty}^{+\infty} j^{-n} J_n(\beta_0 \rho) e^{jn\phi}$$

 $\rho \geq$

$$\underline{E}_z^s = \hat{a}_z E_0 \sum_{n=-\infty}^{+\infty} a_n H_n^{(2)}(\beta_0 \rho) e^{jn\phi}$$

 $\rho \geq b$

$$\underline{E}_z^d = \hat{a}_z E_0 \sum_{n=-\infty}^{+\infty} [b_n J_n(\beta_1 \rho) + c_n Y_n(\beta_1 \rho)] e^{jn\phi}$$

 $b \geq \rho \geq a$

a. From the solution of Problem 11.58, we can write (by using Maxwell's equation) the corresponding magnetic field components as

$$H_\rho^i = -\frac{E_0}{j\omega\mu_0} \frac{1}{\rho} \sum_{n=-\infty}^{+\infty} n j^{-n+1} J_n(\beta_0 \rho) e^{jn\phi}$$

$$H_\phi^i = \frac{E_0 \beta_0}{j\omega\mu_0} \sum_{n=-\infty}^{+\infty} j^{-n} J_n'(\beta_0 \rho) e^{jn\phi}$$

$$H_\rho^s = -\frac{E_0}{j\omega\mu_0} \frac{1}{\rho} \sum_{n=-\infty}^{+\infty} n j a_n H_n^{(2)}(\beta_0 \rho) e^{jn\phi}$$

$$H_\phi^s = \frac{E_0 \beta_0}{j\omega\mu_0} \sum_{n=-\infty}^{+\infty} a_n H_n^{(2)}(\beta_0 \rho) e^{jn\phi}$$

$$H_\rho^d = -\frac{E_0}{j\omega\mu_1} \frac{1}{\rho} \sum_{n=-\infty}^{+\infty} n j [b_n J_n(\beta_1 \rho) + c_n Y_n(\beta_1 \rho)] e^{jn\phi}$$

$$H_\phi^d = \frac{E_0 \beta_1}{j\omega\mu_1} \sum_{n=-\infty}^{+\infty} [b_n J_n'(\beta_1 \rho) + c_n Y_n'(\beta_1 \rho)] e^{jn\phi}$$

b. To find the wave amplitude coefficients a_n, b_n, c_n , the following boundary conditions must be used.

$$E_z^d(\rho=a)=0 \quad (1)$$

$$(E_z^i + E_z^s)_{\rho=b} = E_z^d_{\rho=b} \quad (2)$$

$$(H_\phi^i + H_\phi^s)_{\rho=b} = H_\phi^d)_{\rho=b} \quad (3)$$

Using (1), we can write that

$$E_0 \sum_{n=-\infty}^{+\infty} [b_n J_n(\beta_1 a) + c_n Y_n(\beta_1 a)] = 0 \Rightarrow c_n = -b_n \frac{J_n(\beta_1 a)}{Y_n(\beta_1 a)} \quad (4)$$

Using boundary condition (2), we can write that

$$E_0 \sum_{n=-\infty}^{+\infty} [j^{-n} J_n(\beta_0 b) + a_n H_n^{(2)}(\beta_0 b)] e^{jn\phi} = E_0 \sum_{n=-\infty}^{+\infty} [b_n J_n(\beta_1 b) + c_n Y_n(\beta_1 b)]$$

$$j^{-n} J_n(\beta_0 b) + a_n H_n^{(2)}(\beta_0 b) = b_n J_n(\beta_1 b) + c_n Y_n(\beta_1 b) \quad (5)$$

Using (4) reduces (5) to

$$j^{-n} J_n(\beta_0 b) + a_n H_n^{(2)}(\beta_0 b) = b_n \left[J_n(\beta_1 b) - \frac{J_n(\beta_1 a)}{Y_n(\beta_1 a)} Y_n(\beta_1 b) \right]$$

$$j^{-n} J_n(\beta_0 b) + a_n H_n^{(2)}(\beta_0 b) = b_n \left[\frac{J_n(\beta_1 b) Y_n(\beta_1 a) - J_n(\beta_1 a) Y_n(\beta_1 b)}{Y_n(\beta_1 a)} \right]$$

cont'd.

11.65 cont'd. which may be written as

$$a_n H_n^{(2)}(\beta_0 b) - b_n F_n(\beta_1 a, \beta_1 b) = -j^{-n} J_n(\beta_0 b) \quad (6)$$

$$\text{where } F_n(\beta_1 a, \beta_1 b) = \frac{J_n(\beta_1 b) Y_n(\beta_1 a) - J_n(\beta_1 a) Y_n(\beta_1 b)}{Y_n(\beta_1 a)} \quad (6a)$$

Using boundary condition (3), we can write that

$$\sum_{n=0}^{+\infty} [j^{-n} J_n'(\beta_0 b) + a_n H_n^{(2)'}(\beta_0 b)] e^{jn\phi} = \frac{\beta_1}{\mu_1} \sum_{n=0}^{+\infty} [b_n J_n'(\beta_1 b) + c_n Y_n'(\beta_1 b)]$$

$$-j^{-n} J_n'(\beta_0 b) + a_n H_n^{(2)'}(\beta_0 b) = \frac{\beta_1}{\beta_0} \frac{\mu_0}{\mu_1} [b_n J_n'(\beta_1 b) + c_n Y_n'(\beta_1 b)] \quad (7)$$

which by using (4) reduces to

$$\begin{aligned} -j^{-n} J_n'(\beta_0 b) + a_n H_n^{(2)'}(\beta_0 b) &= \sqrt{\frac{\epsilon_r}{\mu_r}} b_n \left[J_n'(\beta_1 b) - \frac{J_n(\beta_1 a)}{Y_n(\beta_1 a)} Y_n'(\beta_1 b) \right] \\ &= \sqrt{\frac{\epsilon_r}{\mu_r}} b_n \left[\frac{J_n'(\beta_1 b) Y_n(\beta_1 a) - J_n(\beta_1 a) Y_n'(\beta_1 b)}{Y_n(\beta_1 b)} \right] \end{aligned}$$

which may be written as

$$a_n H_n^{(2)'}(\beta_0 b) - \sqrt{\frac{\epsilon_r}{\mu_r}} G_n(\beta_1 a, \beta_1 b) = -j^{-n} J_n'(\beta_0 b) \quad (8)$$

$$\text{where } G_n(\beta_1 a, \beta_1 b) = \frac{J_n'(\beta_1 b) Y_n(\beta_1 a) - J_n(\beta_1 a) Y_n'(\beta_1 b)}{Y_n(\beta_1 b)} \quad (8a)$$

Now we need to solve for a_n and b_n using (6) and (8), or

$$a_n H_n^{(2)}(\beta_0 b) - b_n F_n(\beta_1 a, \beta_1 b) = -j^{-n} J_n(\beta_0 b) \quad (6)$$

$$a_n H_n^{(2)'}(\beta_0 b) - \sqrt{\frac{\epsilon_r}{\mu_r}} G_n(\beta_1 a, \beta_1 b) = -j^{-n} J_n'(\beta_0 b) \quad (8)$$

Equations (6) and (8) of this problem are identical in form to (1) and (2) of the solution of Problem 11.58. Therefore by proper interface of functions we write from the solution of Problem 11.58 that

$$a_n = j^{-n} \frac{J_n'(\beta_0 b) F_n(\beta_1 a, \beta_1 b) - \sqrt{\epsilon_r/\mu_r} J_n(\beta_0 b) G_n(\beta_1 a, \beta_1 b)}{\sqrt{\epsilon_r/\mu_r} G_n(\beta_1 a, \beta_1 b) H_n^{(2)}(\beta_0 b) - F_n(\beta_1 a, \beta_1 b) H_n^{(2)'}(\beta_0 b)} \quad (9)$$

$$b_n = j^{-n} \frac{J_n(\beta_0 b) H_n^{(2)'}(\beta_0 b) - J_n'(\beta_0 b) H_n^{(2)}(\beta_0 b)}{F_n(\beta_1 a, \beta_1 b) H_n^{(2)}(\beta_0 b) - \sqrt{\epsilon_r/\mu_r} G_n(\beta_1 a, \beta_1 b) H_n^{(2)'}(\beta_0 b)} \quad (10)$$

$$c_n = -b_n \frac{J_n(\beta_1 a)}{Y_n(\beta_1 a)} \quad (4)$$

$$F_n = \frac{J_n(\beta_1 b) Y_n(\beta_1 a) - J_n(\beta_1 a) Y_n(\beta_1 b)}{Y_n(\beta_1 a)}, \quad G_n = \frac{J_n'(\beta_1 b) Y_n(\beta_1 a) - J_n(\beta_1 a) Y_n'(\beta_1 b)}{Y_n(\beta_1 b)} \quad (7a)$$

11.66

$$\underline{H}^L = \hat{a}_2 H_0 \sum_{n=-\infty}^{+\infty} j^{-n} J_n(\beta_0 p) e^{jn\phi}$$

$$\underline{H}^S = \hat{a}_2 H_0 \sum_{n=-\infty}^{+\infty} a_n H_n^{(2)}(\beta_0 p) e^{jn\phi}$$

$$\underline{H}^d = \hat{a}_2 H_0 \sum_{n=-\infty}^{+\infty} [b_n J_n(\beta_1 p) + c_n Y_n(\beta_1 p)] e^{jn\phi}$$

 $\rho \geq a$ $\rho \geq a$ $b \geq \rho \geq a$

- a. From the solution of Problem 11.62, we can write (by using Maxwell's equations) the corresponding electric field components

$$\underline{E}_p^L = \frac{H_0}{j\omega \epsilon_0} \frac{1}{\rho} \sum_{n=-\infty}^{+\infty} n j^{-n+1} J_n(\beta_0 p) e^{jn\phi}$$

$$\underline{E}_\phi^L = - \frac{H_0 \beta_0}{j\omega \epsilon_0} \sum_{n=-\infty}^{+\infty} j^{-n} J'_n(\beta_0 p) e^{jn\phi}$$

$$\underline{E}_p^S = \frac{H_0}{j\omega \epsilon_0 \rho} \sum_{n=-\infty}^{+\infty} n j \alpha_n H_n^{(2)}(\beta_0 p) e^{jn\phi}$$

$$\underline{E}_\phi^S = - \frac{H_0 \beta_0}{j\omega \epsilon_0} \sum_{n=-\infty}^{+\infty} a_n H_n^{(2)}(\beta_0 p) e^{jn\phi}$$

$$\underline{E}_p^d = \frac{H_0}{j\omega \epsilon_1} \frac{1}{\rho} \sum_{n=-\infty}^{+\infty} j n [b_n J_n(\beta_1 p) + c_n Y_n(\beta_1 p)] e^{jn\phi}$$

$$\underline{E}_\phi^d = - \frac{H_0 \beta_1}{j\omega \epsilon_1} \sum_{n=-\infty}^{+\infty} [b_n J'_n(\beta_1 p) + c_n Y'_n(\beta_1 p)] e^{jn\phi}$$

- b. To find the wave amplitude coefficients a_n, b_n, c_n , the following boundary conditions must be used.

$$\underline{E}_\phi^d(\rho=a)=0 \quad (1)$$

$$(\underline{E}_\phi^L + \underline{E}_\phi^S)|_{\rho=b} = \underline{E}_\phi^d|_{\rho=b} \quad (2)$$

$$(\underline{H}_2^L + \underline{H}_2^S)|_{\rho=b} = \underline{H}_2^d|_{\rho=b} \quad (3)$$

Using (1), we can write

$$-\frac{H_0 \beta_1}{j\omega \epsilon_1} \sum_{n=-\infty}^{+\infty} [b_n J'_n(\beta_1 a) + c_n Y'_n(\beta_1 a)] e^{jn\phi} = 0 \Rightarrow c_n = -b_n \frac{Y'_n(\beta_1 a)}{J'_n(\beta_1 a)} \quad (4)$$

Using boundary condition (2), we can write

$$\frac{\beta_0}{\epsilon_0} \sum_{n=-\infty}^{+\infty} [j^{-n} J'_n(\beta_0 b) + a_n H_n^{(2)\prime}(\beta_0 b)] e^{jn\phi} = \frac{\beta_1}{\epsilon_1} \sum_{n=-\infty}^{+\infty} [b_n J'_n(\beta_1 b) + c_n Y'_n(\beta_1 b)]$$

$$[j^{-n} J'_n(\beta_0 b) + a_n H_n^{(2)\prime}(\beta_0 b)] = \frac{\beta_1}{\beta_0} \frac{\epsilon_0}{\epsilon_1} [b_n J'_n(\beta_1 b) + c_n Y'_n(\beta_1 b)] \quad (5)$$

which by using (4) reduces to

$$\begin{aligned} j^{-n} J'_n(\beta_0 b) + a_n H_n^{(2)\prime}(\beta_0 b) &= \frac{j n}{\epsilon_1} b_n \left[J'_n(\beta_1 b) - \frac{Y'_n(\beta_1 a)}{J'_n(\beta_1 a)} Y'_n(\beta_1 b) \right] \\ &= \sqrt{\frac{\epsilon_0}{\epsilon_1}} b_n \left[\frac{J'_n(\beta_1 b) J'_n(\beta_1 a) - Y'_n(\beta_1 a) Y'_n(\beta_1 b)}{J'_n(\beta_1 a)} \right] \end{aligned}$$

cont'd.

11.66 cont'd. which may be written as

$$a_n H_n^{(2)}(\beta_0 b) - \sqrt{\frac{\mu_r}{\epsilon_r}} b_n G_n'(\beta_0 a, \beta_1 b) = -j^{-n} J_n'(\beta_0 b) \quad (6)$$

where $G_n' = \frac{J_n'(\beta_1 b) J_n'(\beta_0 a) - Y_n'(\beta_0 a) Y_n'(\beta_1 b)}{J_n'(\beta_1 a)}$ (6a)

Using boundary condition (3), we can write

$$\sum_{n=-\infty}^{+\infty} [j^{-n} J_n(\beta_0 b) + a_n H_n^{(2)}(\beta_0 b)] e^{jn\phi} = \sum_{n=-\infty}^{+\infty} [b_n J_n(\beta_1 b) + c_n Y_n(\beta_1 b)] e^{jn\phi}$$

$$j^{-n} J_n(\beta_0 b) + a_n H_n^{(2)}(\beta_0 b) = b_n J_n(\beta_1 b) + c_n Y_n(\beta_1 b) \quad (7)$$

Using (4) reduces (7) to

$$j^{-n} J_n(\beta_0 b) + a_n H_n^{(2)}(\beta_0 b) = b_n \left[J_n(\beta_1 b) - \frac{Y_n'(\beta_1 a)}{J_n'(\beta_1 a)} Y_n(\beta_1 b) \right]$$

$$= b_n \left[\frac{J_n(\beta_1 b) J_n'(\beta_1 a) - Y_n'(\beta_1 a) Y_n(\beta_1 b)}{J_n'(\beta_1 a)} \right]$$

which may be written as

$$a_n H_n^{(2)}(\beta_0 b) - b_n F_n'(\beta_1 a, \beta_1 b) = -j^{-n} J_n(\beta_0 b) \quad (8)$$

where $F_n' = \frac{J_n(\beta_1 b) J_n'(\beta_1 a) - Y_n'(\beta_1 a) Y_n(\beta_1 b)}{J_n'(\beta_1 a)}$ (8a)

Now we need to solve for a_n and b_n using (6) and (8), or

$$a_n H_n^{(2)}(\beta_0 b) - b_n F_n'(\beta_1 a, \beta_1 b) = -j^{-n} J_n(\beta_0 b) \quad (8)$$

$$a_n H_n^{(2)\prime}(\beta_0 b) - b_n \sqrt{\frac{\mu_r}{\epsilon_r}} G_n'(\beta_0 a, \beta_1 b) = -j^{-n} J_n'(\beta_0 b) \quad (6)$$

These two equations are identical to (6) and 8, respectively, of the solution of Problem 11.65. Therefore using (9) and (10) of Problem 11.65 solution, we can write by replacing μ_r/ϵ_r by μ_r/ϵ_r that

$$a_n = j \sqrt{\frac{\mu_r}{\epsilon_r}} G_n'(\beta_0 a, \beta_1 b) H_n^{(2)}(\beta_0 b) - \sqrt{\frac{\mu_r}{\epsilon_r}} J_n(\beta_0 b) G_n'(\beta_1 a, \beta_1 b) \quad (9)$$

$$b_n = j \frac{J_n(\beta_0 b) H_n^{(2)\prime}(\beta_0 b) - J_n'(\beta_0 b) H_n^{(2)}(\beta_0 b)}{F_n'(\beta_1 a, \beta_1 b) H_n^{(2)}(\beta_0 b) - \sqrt{\frac{\mu_r}{\epsilon_r}} G_n'(\beta_1 a, \beta_1 b) H_n^{(2)}(\beta_0 b)} \quad (10)$$

$$c_n = -b_n J_n'(\beta_1 a) / Y_n'(\beta_1 a) \quad (4)$$

$$F_n' = \frac{J_n(\beta_1 b) J_n'(\beta_1 a) - Y_n'(\beta_1 a) Y_n(\beta_1 b)}{J_n'(\beta_1 a)} \quad (8a), \quad G_n' = \frac{J_n'(\beta_1 b) J_n'(\beta_1 a) - Y_n'(\beta_1 a) Y_n(\beta_1 b)}{J_n'(\beta_1 a)} \quad (6a)$$

11.67 Examining (11-134a)-(11-134c) we see that the ϕ component of the electric field has a $1/p$ variation outside the summation while the p and z components have no p variations outside the summation. Thus for large values of p ($p \gg \lambda$) the ϕ component is small compared to the other two, so small that we assume that it is negligible. Thus

$$\underline{E}_t^s = \underline{E}_p^s + \underline{E}_\phi^s + \underline{E}_z^s \xrightarrow{p \gg \lambda} \underline{E}_p^s + \underline{E}_z^s$$

Using (11-135a) and (11-135b) we can write the p and z components of the electric field, as given by (11-134a) and (11-134c), as

$$\begin{aligned} \underline{E}_p^s &\xrightarrow{p \gg \lambda} E_0 \cos\theta_i e^{j\beta(z \cos\theta_i - ps \sin\theta_i)} \sqrt{\frac{2}{\pi p s \sin\theta_i}} \sum_{n=-\infty}^{+\infty} a_n e^{jn\phi} \\ \underline{E}_z^s &\xrightarrow{p \gg \lambda} E_0 \sin\theta_i e^{j\beta(z \cos\theta_i - ps \sin\theta_i)} \sqrt{\frac{2}{\pi p s \sin\theta_i}} \sum_{n=-\infty}^{+\infty} a_n e^{jn\phi} \end{aligned}$$

$$|\underline{E}_t^s| \approx \sqrt{|\underline{E}_p^s|^2 + |\underline{E}_z^s|^2} = |E_0| \sqrt{\frac{2}{\pi p s \sin\theta_i}} \left| \sum_{n=-\infty}^{+\infty} a_n e^{jn\phi} \right|$$

Therefore using (11-21b)

$$\sigma_{2-D} = \lim_{p \rightarrow \infty} \left[2\pi p \frac{|\underline{E}_t^s|^2}{|\underline{E}_t^s|^2} \right] = \lim_{p \rightarrow \infty} \left[2\pi p \frac{2}{\pi p s \sin\theta_i} \left| \sum_{n=-\infty}^{+\infty} a_n e^{jn\phi} \right|^2 \right] = \frac{4}{s \sin\theta_i} \left| \sum_{n=-\infty}^{+\infty} a_n e^{jn\phi} \right|^2$$

11.68 Examining (11-153d)-(11-153f) we see that the ϕ component of the magnetic field has a $1/p$ variation outside the summation while the p and z components have no p variations outside the summation. Thus for large values of p ($p \gg \lambda$) the ϕ component is small compared to the other two, so small that we assume that it is negligible. Thus

$$\underline{H}_t^s = \underline{H}_p^s + \underline{H}_\phi^s + \underline{H}_z^s \xrightarrow{p \gg \lambda} \underline{H}_p^s + \underline{H}_z^s$$

Using (11-135a) and (11-135b) we can write the p and z components of the magnetic field, as given by (11-153d) and (11-153f), as

$$\begin{aligned} \underline{H}_p^s &\xrightarrow{p \gg \lambda} H_0 \cos\theta_i e^{j\beta(z \cos\theta_i - ps \sin\theta_i)} \sqrt{\frac{2}{\pi p s \sin\theta_i}} \sum_{n=-\infty}^{+\infty} b_n e^{jn\phi} \\ \underline{H}_z^s &\xrightarrow{p \gg \lambda} H_0 \sin\theta_i e^{j\beta(z \cos\theta_i - ps \sin\theta_i)} \sqrt{\frac{2}{\pi p s \sin\theta_i}} \sum_{n=-\infty}^{+\infty} b_n e^{jn\phi} \end{aligned}$$

$$|\underline{H}_t^s| \approx \sqrt{|\underline{H}_p^s|^2 + |\underline{H}_z^s|^2} = |H_0| \sqrt{\frac{2}{\pi p s \sin\theta_i}} \left| \sum_{n=-\infty}^{+\infty} b_n e^{jn\phi} \right|$$

Therefore using (11-21c)

$$\sigma_{2-D} = \lim_{p \rightarrow \infty} \left[2\pi p \frac{|\underline{H}_t^s|^2}{|\underline{H}_t^s|^2} \right] = \lim_{p \rightarrow \infty} \left[2\pi p \frac{2}{\pi p s \sin\theta_i} \left| \sum_{n=-\infty}^{+\infty} b_n e^{jn\phi} \right|^2 \right] = \frac{4}{s \sin\theta_i} \left| \sum_{n=-\infty}^{+\infty} b_n e^{jn\phi} \right|^2$$

11.69 The total field radiated by a line source next to a cylinder with large radius $a \gg \lambda$ (in this case $a = 50\lambda$) is approximately the same as an electric line source a height h above a PEC flat ground-plane, as shown in Figure 11-2.

Therefore the normalized radiated field is represented by the array factor $(\sin(\beta h \sin \phi))$ of $(11-20a)$ where the line source is placed directly above the PEC ground plane ($\phi = 90^\circ$).

- (a) To determine the smallest height of the line source above the cylinder, we set the sine function (array factor) of $(11-20a)$ equal to unity; that is (for a maximum)

$$\left. \sin(\beta h \sin \phi) \right|_{\phi=90^\circ} \approx \sin(\beta h) = 1 \Rightarrow \beta h = \sin^{-1}(1) = \frac{n\pi}{2}, n=1, 3, \dots$$

For $n=1$ (smallest height):

$$h \approx \frac{\pi/2}{\beta} = \frac{\pi/2}{2\pi/\lambda} = \frac{\lambda}{4}$$

Thus the line source must be placed at a height $\lambda/4$ above the cylinder surface.

- (b) To determine the smallest height for a null along $\phi = 90^\circ$, we set the sine function (array factor) of $(11-20a)$ equal to zero; that is (for a null):

$$\left. \sin(\beta h \sin \phi) \right|_{\phi=90^\circ} = \sin(\beta h) = 0 \Rightarrow \beta h = \sin^{-1}(0) = n\pi, n=0, 1, 2, \dots$$

For other than $h=0$ (trivial solution), $n=1$. Therefore

$$h \approx \frac{\pi}{\beta} = \frac{\pi}{2\pi/\lambda} = \frac{\lambda}{2}$$

- (c) To find the smallest when the pattern at $\phi = 90^\circ$ is -3 dB (0.707 field amplitude), we set the sine function [array factor of $(11-20a)$] equal to 0.707; that is

$$\left. \sin(\beta h \sin \phi) \right|_{\phi=90^\circ} = \sin(\beta h) = 0.707 \Rightarrow \beta h = \sin^{-1}(0.707) = \frac{n\pi}{4}, n=1, 3, \dots$$

For the smallest height, $n=1$. Therefore

$$h \approx \frac{\pi/4}{\beta} = \frac{\pi/4}{2\pi/\lambda} = \frac{\lambda}{8}$$

11.70 According to (11-164b) when $\rho > \rho' = s$

$$a. E_z^t = -\frac{\beta^2 I e}{4\pi\epsilon} \sum_{n=-\infty}^{+\infty} H_n^{(2)}(\beta\rho) \left[J_n(\beta s) - \frac{J_n(\beta a)}{H_n^{(2)}(\beta a)} H_n^{(2)}(\beta s) \right] e^{jn(\phi-\phi')}$$

For the two sources (one at $\phi' = 0$ and the other at $\phi' = \pi$), the total electric field can be written as

$$E_z^t(\text{total}) = -\frac{\beta^2 I e}{4\pi\epsilon} \sum_{n=-\infty}^{+\infty} H_n^{(2)}(\beta\rho) \left[J_n(\beta s) - \frac{J_n(\beta a)}{H_n^{(2)}(\beta a)} H_n^{(2)}(\beta s) \right] [e^{jn\phi} + e^{jn\phi} e^{-jn\pi}]$$

$$E_z^t(\text{total}) = -\frac{\beta^2 I e}{4\pi\epsilon} \sum_{n=-\infty}^{+\infty} H_n^{(2)}(\beta\rho) \left[J_n(\beta s) - \frac{J_n(\beta a)}{H_n^{(2)}(\beta a)} H_n^{(2)}(\beta s) \right] e^{jn\phi} (1 + e^{-jn\pi})$$

b. For large observation distances ($\rho \gg \lambda$)

$$H_n^{(2)}(\beta\rho) \xrightarrow{\rho \gg \lambda} \sqrt{\frac{2j}{\pi\beta\rho}} j^n e^{-j\beta\rho}$$

Thus

$$E_z^t(\text{total}) \xrightarrow{\rho \gg \lambda} -\frac{\beta^2 I e}{4\pi\epsilon} \sqrt{\frac{2j}{\pi\beta\rho}} e^{-j\beta\rho} \sum_{n=-\infty}^{+\infty} j^n \left[J_n(\beta s) - \frac{J_n(\beta a)}{H_n^{(2)}(\beta a)} H_n^{(2)}(\beta s) \right] e^{jn\phi} (1 + e^{-jn\pi})$$

When $s \ll \lambda$ and $a \ll \lambda$, then only the first ^{and second} terms of the infinite summation are sufficient. Therefore

$$\begin{aligned} E_z^t(\text{total}) &\xrightarrow[s \ll \lambda]{a \ll \lambda} -\frac{\beta^2 I e}{4\pi\epsilon} \sqrt{\frac{2j}{\pi\beta\rho}} e^{-j\beta\rho} \left\{ \left[J_0(\beta s) - \frac{J_0(\beta a)}{H_0^{(2)}(\beta a)} H_0^{(2)}(\beta s) \right] 2 \right. \\ &\quad + j \left[J_1(\beta s) - \frac{J_1(\beta a)}{H_1^{(2)}(\beta a)} H_1^{(2)}(\beta s) \right] e^{j\phi} (1 + e^{-jn\pi}) \\ &\quad \left. + j^{-1} \left[J_{-1}(\beta s) - \frac{J_{-1}(\beta a)}{H_{-1}^{(2)}(\beta a)} H_{-1}^{(2)}(\beta s) \right] e^{-j\phi} (1 + e^{jn\pi}) \right\} \end{aligned}$$

$$E_z^t(\text{total}) \xrightarrow[s \ll \lambda]{a \ll \lambda} -\frac{\beta^2 I e}{4\pi\epsilon} \sqrt{\frac{2j}{\pi\beta\rho}} e^{-j\beta\rho} \left\{ 2 \left[J_0(\beta s) - \frac{J_0(\beta a)}{H_0^{(2)}(\beta a)} H_0^{(2)}(\beta s) \right] \right\} = 0$$

For this to vanish

$$\left[J_0(\beta s) - \frac{J_0(\beta a)}{H_0^{(2)}(\beta a)} H_0^{(2)}(\beta s) \right] = 0 \Rightarrow s = a$$

Therefore for the field to vanish the source must be placed next to the cylinder. Of course this is obvious since the ^{electric} source does not radiate when it is placed tangentially next to a perfect electric conductor. Thus because the cylinder is small there is no other position of the source which will lead to a vanishing far zone field.

11.7.1 According to (11-176g) when $\rho > \rho' = b$

$$a. H_2^t = -\frac{\beta^2 I_m}{4\omega \mu} \sum_{n=-\infty}^{+\infty} H_n^{(2)}(\beta\rho) \left[J_n(\beta b) - \frac{J'_n(\beta a)}{H_n^{(2)'}(\beta a)} H_n^{(2)}(\beta b) \right] e^{jn\phi} e^{j\phi'}$$

For the three sources of Figure P11-33

$$H_{20}^t = -2 \frac{I_m \beta^2}{4\omega \mu} \sum_{n=-\infty}^{+\infty} H_n^{(2)}(\beta\rho) \left[J_n(\beta b) - \frac{J'_n(\beta a)}{H_n^{(2)'}(\beta a)} H_n^{(2)}(\beta b) \right] e^{jn\phi}, \quad \phi' = 0$$

$$H_{21}^t = -\frac{I_m \beta^2}{4\omega \mu} \sum_{n=-\infty}^{+\infty} H_n^{(2)}(\beta\rho) \left[J_n(\beta b) - \frac{J'_n(\beta a)}{H_n^{(2)'}(\beta a)} H_n^{(2)}(\beta b) \right] e^{jn\phi} e^{-j\phi_0}, \quad \phi' = \phi_0$$

$$H_{2-1}^t = -\frac{I_m \beta^2}{4\omega \mu} \sum_{n=-\infty}^{+\infty} H_n^{(2)}(\beta\rho) \left[J_n(\beta b) - \frac{J'_n(\beta a)}{H_n^{(2)'}(\beta a)} H_n^{(2)}(\beta b) \right] e^{jn\phi} e^{j\phi_0}, \quad \phi' = 2\pi - \phi_0$$

Thus the total field for all the three sources is equal to

$$H_2^t(\text{total}) = -\frac{I_m \beta^2}{4\omega \mu} \sum_{n=-\infty}^{+\infty} H_n^{(2)}(\beta\rho) \left[J_n(\beta b) - \frac{J'_n(\beta a)}{H_n^{(2)'}(\beta a)} H_n^{(2)}(\beta b) \right] e^{jn\phi} \left[2 + 2 \left(\frac{e^{j\phi_0} + e^{-j\phi_0}}{2} \right) \right]$$

$$= -\frac{I_m \beta^2}{4\omega \mu} \sum_{n=-\infty}^{+\infty} H_n^{(2)}(\beta\rho) \left[J_n(\beta b) - \frac{J'_n(\beta a)}{H_n^{(2)'}(\beta a)} H_n^{(2)}(\beta b) \right] 2 e^{jn\phi} \left[1 + \cos(n\phi_0) \right]$$

$$H_2^t(\text{total}) = -\frac{I_m \beta^2}{4\omega \mu} \sum_{n=-\infty}^{+\infty} H_n^{(2)}(\beta\rho) \left[J_n(\beta b) - \frac{J'_n(\beta a)}{H_n^{(2)'}(\beta a)} H_n^{(2)}(\beta b) \right] e^{jn\phi} \cos^2\left(\frac{n\phi_0}{2}\right)$$

$$b. \frac{|H_2^s|}{|H_2^t|} = \frac{\left| \sum_{n=-\infty}^{+\infty} H_n^{(2)}(\beta\rho) \frac{J'_n(\beta a)}{H_n^{(2)'}(\beta a)} H_n^{(2)}(\beta b) e^{jn\phi} \cos^2\left(\frac{n\phi_0}{2}\right) \right|}{\left| \sum_{n=-\infty}^{+\infty} H_n^{(2)}(\beta\rho) J_n(\beta b) e^{jn\phi} \cos^2\left(\frac{n\phi_0}{2}\right) \right|}$$

$$c. \text{For } \rho \gg \lambda \Rightarrow H_n^{(2)}(\beta\rho) \xrightarrow{\rho \gg \lambda} \sqrt{\frac{2j}{\pi \beta \rho}} j^n e^{-j\beta\rho}$$

$$\text{Thus } H_2^t(\text{total}) \simeq -\frac{I_m \beta^2}{4\omega \mu} \sqrt{\frac{2j}{\pi \beta \rho}} e^{-j\beta\rho} \sum_{n=-\infty}^{+\infty} j^n \left[J_n(\beta b) - \frac{J'_n(\beta a)}{H_n^{(2)'}(\beta a)} H_n^{(2)}(\beta b) \right] e^{jn\phi} \cos^2\left(\frac{n\phi_0}{2}\right)$$

$$H_2^t(\text{total}) \simeq -\frac{I_m \beta^2}{4\omega \mu} \sqrt{\frac{2j}{\pi \beta \rho}} e^{-j\beta\rho} \sum_{n=-\infty}^{+\infty} j^n \left[\frac{J_n(\beta b) H_n^{(2)'}(\beta a) - J'_n(\beta a) H_n^{(2)}(\beta b)}{H_n^{(2)'}(\beta a)} \right] e^{jn\phi} \cos^2\left(\frac{n\phi_0}{2}\right)$$

$$H_2^t(\text{total}) \underset{\substack{\text{norm} \\ \text{constant } \rho \\ \rho = C}}{\sim} \sum_{n=-\infty}^{+\infty} j^n \left[\frac{J_n(\beta b) H_n^{(2)'}(\beta a) - J'_n(\beta a) H_n^{(2)}(\beta b)}{H_n^{(2)'}(\beta a)} \right] e^{jn\phi} \cos^2\left(\frac{n\phi_0}{2}\right)$$

11.72 According to (11-191a)

$$E_z^t \xrightarrow[\beta\phi \rightarrow \infty]{\alpha=0} E_0 \sum_{m=1}^{\infty} j^{m/2} J_{m/2}(\beta\rho) \sin\left(\frac{m}{2}\phi\right) \sin\left(\frac{m}{2}\phi'\right)$$

The corresponding magnetic field is obtained using Maxwell's equation of

$$\nabla \times \vec{E} = -j\omega\mu H \Rightarrow \vec{H} = -\frac{1}{j\omega\mu} \nabla \times \vec{E} \Rightarrow \begin{cases} H_\phi = -\frac{1}{j\omega\mu} \frac{1}{\rho} \frac{\partial E_z}{\partial \phi} \\ H_\theta = \frac{1}{j\omega\mu} \frac{\partial E_z}{\partial \rho} \end{cases}$$

with the corresponding current density of

$$J_z = \hat{n} \times \vec{H} \Big|_{\phi=0} = \hat{a}_y \times (\hat{a}_\rho H_\rho + \hat{a}_\phi H_\phi) \Big|_{\phi=0} = \hat{a}_y \times (\hat{a}_x H_\rho + \hat{a}_y H_\phi) = -\hat{a}_z H_\rho$$

$$J_z = -H_\rho \Big|_{\phi=0} = \frac{1}{j\omega\mu} \frac{1}{\rho} \frac{\partial E_z}{\partial \phi} \Big|_{\phi=0} = \frac{E_0}{j2\omega\mu} \frac{1}{\rho} \sum_{m=1}^{\infty} j^{m/2} m J_{m/2}(\beta\rho) \sin\left(\frac{m}{2}\phi\right) \cos\left(\frac{m}{2}\phi'\right)$$

$$J_z = \frac{E_0}{j2\omega\mu\rho} \sum_{m=1}^{\infty} m j^{m/2} J_{m/2}(\beta\rho) \sin\left(\frac{m}{2}\phi'\right)$$

When $\beta\rho \rightarrow 0$

$$J_{m/2}(\beta\rho) \xrightarrow[\beta\rho \rightarrow 0]{m \neq 0} \frac{1}{\frac{m}{2}!} \left(\frac{\beta\rho}{2}\right)^{m/2}$$

Considering only the $m=1$ term (since the $m=0$ term does not contribute)

$$J_z \xrightarrow[m=1]{\beta\rho \rightarrow 0} \frac{E_0}{j2\omega\mu\rho} j^{1/2} \frac{1}{\frac{1}{2}!} \left(\frac{\beta\rho}{2}\right)^{1/2} \sin\left(\frac{\phi'}{2}\right)$$

Since

$$\frac{1}{2}! = \Gamma\left(\frac{3}{2}\right) = \frac{1}{2} \Gamma\left(\frac{1}{2}\right) = \frac{\sqrt{\pi}}{2}$$

then

$$J_z \xrightarrow[m=1]{\beta\rho \rightarrow 0} \frac{E_0}{j2\omega\mu\rho} j^{1/2} \frac{2}{\sqrt{\pi}} \sqrt{\frac{\beta\rho}{2}} \sin\left(\frac{\phi'}{2}\right) = \frac{E_0}{\eta\beta\rho} \sqrt{\frac{\beta\rho}{2\pi}} \sin\left(\frac{\phi'}{2}\right)$$

$$J_z \approx \frac{E_0}{\eta} \sqrt{\frac{1}{j2\pi\beta\rho}} \sin\left(\frac{\phi'}{2}\right)$$

11.73 $H_z^t = \sum_s b_s J_s(\beta\rho) H_s^{(1)}(\beta\rho') \cos[s(\phi' - \kappa)] \cos[s(\phi - \alpha)]$

When $\beta\rho' \rightarrow \infty \Rightarrow H_s^{(1)}(\beta\rho') = \sqrt{\frac{2j}{\pi\beta\rho'}} j^s e^{-j\beta\rho'}. \text{ Therefore using also (11-196)}$

$$H_z^t \approx \frac{\pi\omega\epsilon I_m}{4(\pi-\alpha)} \sqrt{\frac{2j}{\pi\beta\rho'}} e^{-j\beta\rho'} \sum_s j^s J_s(\beta\rho) \cos[s(\phi' - \kappa)] \cos[s(\phi - \alpha)]$$

$$\approx H_0 \sum_s \epsilon_s j^s J_s(\beta\rho) \cos[s(\phi' - \kappa)] \cos[s(\phi - \alpha)] \text{ where } H_0 = I_m \sqrt{\frac{\pi j}{8\beta\rho}} \frac{\omega\epsilon}{\pi-\alpha} e^{-j\beta\rho'}$$

[11.74] According to (11-199a)

$$H_z \xrightarrow[\alpha=0]{\beta p \rightarrow \infty} H_0 \sum_{m=0}^{\infty} \epsilon_{m1/2} j^{m/2} J_{m1/2}(\beta p) \cos\left(\frac{m}{2}\phi'\right) \cos\left(\frac{m}{2}\phi\right)$$

Therefore the current density is given by

$$\underline{J} = \hat{n} \times \underline{H} \Big|_{\phi=0} = \hat{a}_\phi \times \hat{a}_z H_z \Big|_{\phi=0} = \hat{a}_\phi H_z \Big|_{\phi=0} = \hat{a}_\phi H_0 \sum_{m=0}^{\infty} \epsilon_{m1/2} j^{m/2} J_{m1/2}(\beta p) \cos\left(\frac{m}{2}\phi'\right)$$

$$J_p = H_0 \sum_{m=0}^{\infty} \epsilon_{m1/2} j^{m/2} J_{m1/2}(\beta p) \cos\left(\frac{m}{2}\phi'\right)$$

When $\beta p \rightarrow 0$ only the first term of the summation is necessary. Therefore

$$J_p \xrightarrow[\beta p \rightarrow 0]{} H_0 \epsilon_0 j^0 J_0(\beta p) \quad (1)$$

$$\text{When } \beta p \rightarrow 0 \quad J_0(\beta p) \xrightarrow[\beta p \rightarrow 0]{} 1$$

$$\text{Thus } J_p \xrightarrow[\beta p \rightarrow 0]{} H_0 (1)(1)(1)(1) = H_0.$$

[11.75] The procedure follows that for the derivation of (11-221a) from (11-217a). The only difference for the derivation of (11-222a) from (11-222) is that instead of (11-219a) we need to use the integral of

$$\int_0^{\pi} e^{j\beta r \cos\theta} P_m(\cos\theta) \sin\theta d\theta = 2 j^{+m} j_m(\beta r)$$

Then the expected solution of (11-222) and (11-222a) follows.

[11.76] When $f = 8.5 \text{ GHz} \Rightarrow \lambda = \frac{3 \times 10^8}{8.5 \times 10^9} = 0.353 \times 10^{-3} = 0.0353 \text{ m}$. Therefore the

- a. diameter of 1.126 m is much greater than the wavelength. Thus the RCS of the sphere is equal to its geometric cross sectional area of $\sigma = \pi a^2 = \pi (0.564)^2 = 0.999 \approx 1 \text{ m}^2$ or $0 \text{ dB}/\text{sm}$ (dBsm)

- b. For normal incidence the RCS of any flat plate is equal to

$$\sigma = 4\pi \left(\frac{A}{\lambda}\right)^2 \Rightarrow A = \lambda \sqrt{\frac{\sigma}{4\pi}} = 0.0353 \sqrt{\frac{1}{4\pi}} = 0.0353(0.2821)$$

$$A = 0.01 \text{ m}^2$$

11.77

(a) The RCS of a sphere of radius a_s in the GO region is given by

$$RCS_{\text{sphere}} = \pi a_s^2$$

The maximum RCS of a flat circular plate of radius a_p , based on physical optics, is equal to

$$RCS_{\text{plate}} = 4\pi \left(\frac{A}{\lambda}\right)^2 = 4\pi \left(\frac{\pi a_p^2}{\lambda}\right)^2$$

Equating the 2 RCSs

$$\pi a_s^2 = 4\pi \left(\frac{\pi a_p^2}{\lambda}\right)^2 \Rightarrow a_s^2 = 4 \left(\frac{\pi a_p^2}{\lambda}\right)^2 \Rightarrow a_s = 2 \left(\frac{\pi a_p^2}{\lambda}\right)$$

$$a_s(\text{sphere}) = 2 \left[\frac{\pi a_p^2(\text{plate})}{\lambda} \right]$$

(b) For $a_p = 3\lambda$: $a_s(\text{sphere}) = 2 \left[\frac{\pi (3\lambda)^2}{\lambda} \right] = 2 \left(\frac{9\pi \lambda^2}{\lambda} \right) = 18\pi\lambda$

$$a_s(\text{sphere}) = 18\pi\lambda = 56.5487\lambda$$

11.78

(a) For a flat plate, the maximum RCS at normal incidence is

$$RCS = 4\pi \left(\frac{A}{\lambda}\right)^2 \underset{A=25\lambda^2}{=} 4\pi (25\lambda)^2 = 4(625)\pi \lambda^2 = 7,853.9816\lambda^2$$

At $f = 10 \text{ GHz}$, $\lambda = 3 \times 10^8 / 10 \times 10^9 = 0.03 \text{ m}$

$$RCS = 7,853.9816 \lambda^2 = 7,853.9816 (0.03)^2 = 7.0686 \text{ m}^2$$

which, based on 1 m^2 , is equal to

$$RCS(\text{in dBsm}) = 10 \log_{10} (7.0686) = 8.4933 \text{ dBsm}$$

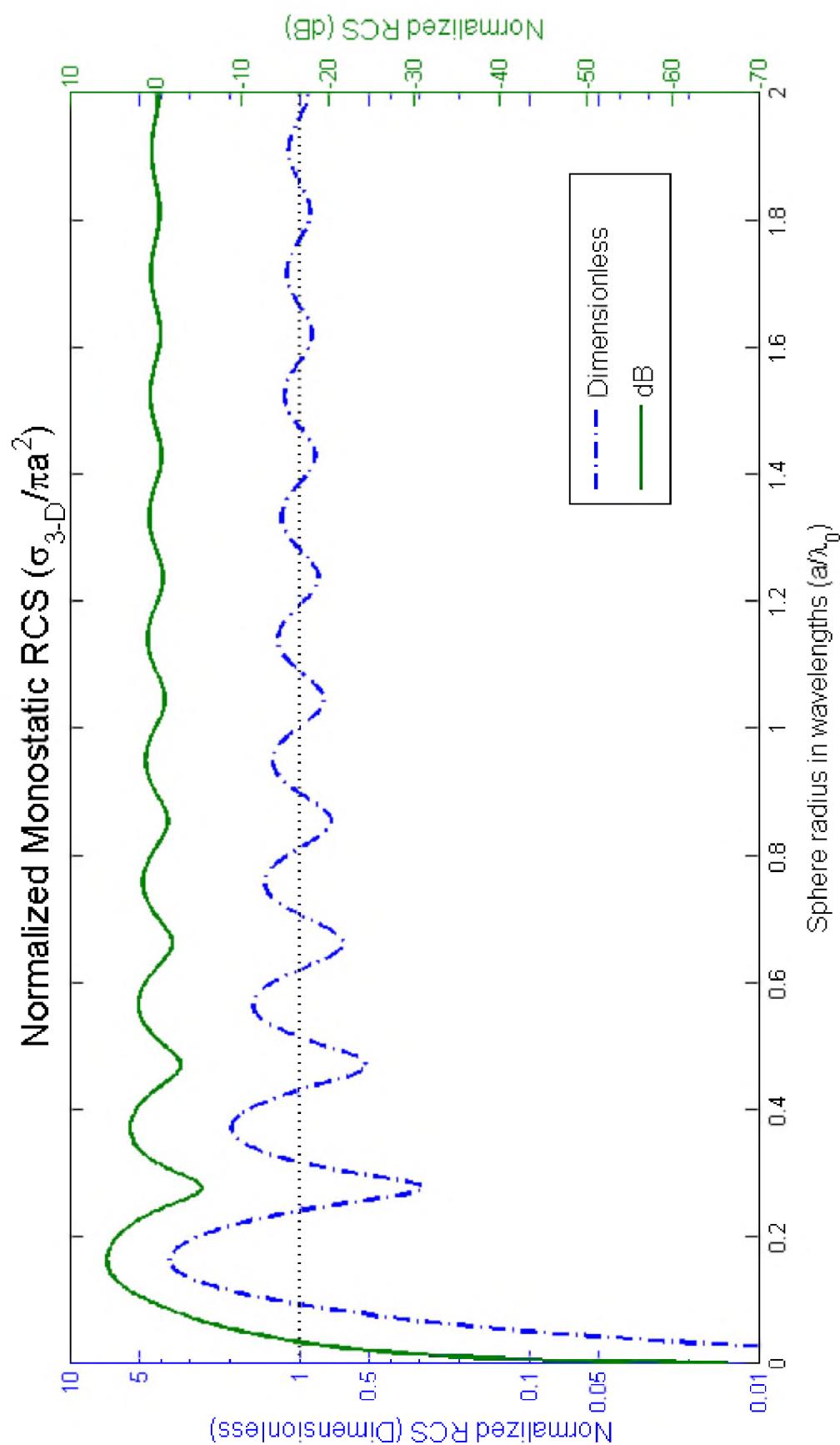
(b) The RCS of a sphere in the GO region is πa_s^2 , $a_s = \text{radius}$
Therefore

$$\pi a_s^2 = 7,853.9816 \lambda^2$$

$$a_s = \left[7,853.9816 \lambda^2 / \pi \right]^{1/2} = 50\lambda$$

Therefore the radius of the sphere must be 50λ
in order for the sphere to have the same RCS at normal incidence
as a flat plate of area $25\lambda^2$

11.79



For a plane wave, with only an x-component of the electric field incident along $\theta = \pi$ (-180°) on a PEC sphere of radius a , the far-zone scattered spherical components of the electric field are given by (11-242b) and (11-242c), or

$$E_\theta^s \approx j E_0 \frac{e^{-jkr}}{kr} \cos\phi \sum_{n=1}^{\infty} j^n \left[b_n \sin\theta P_n^{(1)}(\cos\theta) - c_n P_n^{(1)}(\cos\theta) \right] \quad (1a)$$

$$E_\phi^s \approx j E_0 \frac{e^{-jkr}}{kr} \sin\phi \sum_{n=1}^{\infty} -j^n \left[b_n \frac{P_n^{(1)}(\cos\theta)}{\sin\theta} - c_n \sin\theta P_n^{(1)}(\cos\theta) \right] \quad (1b)$$

where b_n and c_n are given, respectively, by (11-238a) and (11-238b).

The corresponding rectangular components of the scattered electric field are given by

$$E_x^s = \cos\theta \cos\phi E_\theta^s - \sin\phi E_\phi^s \quad (2a)$$

$$E_y^s = \cos\theta \sin\phi E_\theta^s + \cos\phi E_\phi^s \quad (2b)$$

Along the monostatic scattering direction ($\theta = \pi$), (2a) and (2b) reduce to

$$E_x^s|_{\theta=\pi} = -\cos\phi E_\theta^s(\theta=\pi) - \sin\phi E_\phi^s(\theta=\pi) \quad (3a)$$

$$E_y^s|_{\theta=\pi} = -\sin\phi E_\theta^s(\theta=\pi) + \cos\phi E_\phi^s(\theta=\pi) \quad (3b)$$

When $\phi = \pi$, (3a) and (3b) reduce, using (1a) and (1b), to

$$E_x^s|_{\substack{\theta=\pi \\ \phi=\pi}} = E_\phi^s(\theta=\pi, \phi=\pi)$$

$$E_y^s|_{\substack{\theta=\pi \\ \phi=\pi}} = -E_\phi^s(\theta=\pi, \phi=\pi) = 0$$

When $\phi = 3\pi/2$, (3a) and (3b) reduce, using (1a) and (1b), to

$$E_x^s|_{\substack{\theta=\pi \\ \phi=3\pi/2}} = E_\phi^s(\theta=\pi, \phi=3\pi/2)$$

$$E_y^s|_{\substack{\theta=\pi \\ \phi=3\pi/2}} = E_\phi^s(\theta=\pi, \phi=3\pi/2) = 0$$

In both cases, $E_y^s = 0$. Thus there is NO cross-polarized field along the monostatic scattering direction!!

11.81

From Equations (11-231) and (11-232), the vector potentials of the incident fields outside the dielectric sphere take the form of

$$A_r^i = E_0 \frac{\cos \phi}{\omega} \sum_{n=1}^{\infty} a_n \hat{J}_n(\beta_0 r) P_n^{(1)}(\cos \theta)$$

$$F_r^i = E_0 \frac{\sin \phi}{\omega \eta_0} \sum_{n=1}^{\infty} a_n \hat{J}_n(\beta_0 r) P_n^{(1)}(\cos \theta)$$

where, from Equation (11-231a),

$$a_n = j^{-n} \frac{(2n+1)}{n(n+1)}$$

From Equation (11-233), the vector potentials of the scattered fields outside the dielectric sphere take the form of

$$A_r^s = E_0 \frac{\cos \phi}{\omega} \sum_{n=1}^{\infty} b_n \hat{H}_n^{(2)}(\beta_0 r) P_n^{(1)}(\cos \theta)$$

$$F_r^s = E_0 \frac{\sin \phi}{\omega \eta_0} \sum_{n=1}^{\infty} c_n \hat{H}_n^{(2)}(\beta_0 r) P_n^{(1)}(\cos \theta)$$

The combined vector potentials of the fields outside the dielectric sphere equal the sum of the incident and scattered potentials, and are reproduced from Equation (11-235) as

$$A_r^{t+} = A_r^i + A_r^s = E_0 \frac{\cos \phi}{\omega} \sum_{n=1}^{\infty} [a_n \hat{J}_n(\beta_0 r) + b_n \hat{H}_n^{(2)}(\beta_0 r)] P_n^{(1)}(\cos \theta)$$

$$F_r^{t+} = F_r^i + F_r^s = E_0 \frac{\sin \phi}{\omega \eta_0} \sum_{n=1}^{\infty} [a_n \hat{J}_n(\beta_0 r) + c_n \hat{H}_n^{(2)}(\beta_0 r)] P_n^{(1)}(\cos \theta)$$

In the above equations,

$$\beta_0 = \omega \sqrt{\mu_0 \epsilon_0}$$

and

$$\eta_0 = \sqrt{\mu_0 / \epsilon_0}$$

Cont'd

11.91 cont'd

Inside the dielectric sphere, the vector potentials take the form of

$$A_r^{t-} = E_0 \frac{\cos \phi}{\omega} \sum_{n=1}^{\infty} d_n j_n(\beta_d r) P_n^{(1)}(\cos \theta)$$

$$E_r^{t-} = E_0 \frac{\sin \phi}{\omega \eta_d} \sum_{n=1}^{\infty} e_n j_n(\beta_d r) P_n^{(1)}(\cos \theta)$$

with

$$\beta_d = \omega \sqrt{\mu_d \epsilon_d}$$

and

$$\eta_d = \sqrt{\mu_d / \epsilon_d}$$

where

$$\mu_d = \mu_r \mu_0$$

and

$$\epsilon_d = \epsilon_r \epsilon_0$$

Thus

$$\beta_d = \omega \sqrt{\mu_d \epsilon_d} = \omega \sqrt{\mu_r \mu_0 \epsilon_r \epsilon_0} = \omega \sqrt{\mu_0 \epsilon_0} \sqrt{\mu_r \epsilon_r} = \beta_0 \sqrt{\mu_r \epsilon_r}$$

and

$$\eta_d = \sqrt{\mu_d / \epsilon_d} = \sqrt{\mu_r \mu_0 / \epsilon_r \epsilon_0} = \sqrt{\mu_0 / \epsilon_0} \sqrt{\mu_r / \epsilon_r} = \eta_0 \sqrt{\mu_r / \epsilon_r}$$

Cont'd

11.81 Cont'd

Drawing from Equation (11-234), the electric and magnetic fields inside or outside the sphere, using the proper vector potentials, are given by

$$E_r = \frac{1}{j\omega\mu\epsilon} \left(\frac{\partial^2}{\partial r^2} + \beta^2 \right) A_r$$

$$E_\theta = \frac{1}{j\omega\mu\epsilon r} \frac{1}{r} \frac{\partial^2 A_r}{\partial r \partial \theta} - \frac{1}{\epsilon r \sin \theta} \frac{1}{\partial \phi} \frac{\partial F_r}{\partial \phi}$$

$$E_\phi = \frac{1}{j\omega\mu\epsilon r \sin \theta} \frac{1}{\partial r \partial \phi} \frac{\partial^2 A_r}{\partial r^2} + \frac{1}{\epsilon r} \frac{1}{\partial \theta} \frac{\partial F_r}{\partial \theta}$$

$$H_r = \frac{1}{j\omega\mu\epsilon} \left(\frac{\partial^2}{\partial r^2} + \beta^2 \right) F_r$$

$$H_\theta = \frac{1}{\mu r \sin \theta} \frac{1}{\partial \phi} \frac{\partial A_r}{\partial \phi} + \frac{1}{j\omega\mu\epsilon r} \frac{1}{r} \frac{\partial^2 F_r}{\partial r \partial \theta}$$

$$H_\phi = -\frac{1}{\mu r} \frac{1}{\partial \theta} \frac{\partial A_r}{\partial \theta} + \frac{1}{j\omega\mu\epsilon r \sin \theta} \frac{1}{\partial r \partial \phi} \frac{\partial^2 F_r}{\partial r^2}$$

Now, for any Bessel function, Equation (IV-19) can be used to show

$$\frac{\partial Z_n(\beta r)}{\partial r} = -\beta Z_{n+1}(\beta r) + \frac{n}{r} Z_n(\beta r)$$

and

$$\frac{\partial Z_n(\beta r)}{\partial (\beta r)} = -Z_{n+1}(\beta r) + \frac{n}{\beta r} Z_n(\beta r)$$

then following relation will be used in the expansion of the electric and magnetic fields below:

$$\frac{\partial Z_n(\beta r)}{\partial r} = -\beta Z_{n+1}(\beta r) + \frac{n}{r} Z_n(\beta r) = \beta \left[-Z_{n+1}(\beta r) + \frac{n}{\beta r} Z_n(\beta r) \right] = \beta \frac{\partial Z_n(\beta r)}{\partial (\beta r)}$$

Cont'd

11.81 cont'd

Using Equation (11-234) with Equation (11-235), the fields transverse to the radial direction outside the sphere are

$$E_{\theta}^{t+} = \frac{1}{j\omega\mu_0\varepsilon_0} \frac{1}{r} \frac{\partial^2 A_r^{t+}}{\partial r \partial \theta} - \frac{1}{\varepsilon_0} \frac{1}{r \sin \theta} \frac{\partial F_r^{t+}}{\partial \phi}$$

$$= -j \frac{E_0}{\omega\mu_0\varepsilon_0 r} \left\{ \frac{\beta_0}{\omega} \cos \phi \sum_{n=1}^{\infty} [a_n j'_n(\beta_0 r) + b_n \hat{H}_n^{(2)}(\beta_0 r)] P_n^{(1)}(\cos \theta) \right\}$$

$$- \frac{E_0}{\varepsilon_0 r \sin \theta} \left\{ \frac{1}{\omega\eta_0} \cos \phi \sum_{n=1}^{\infty} [a_n j_n(\beta_0 r) + c_n \hat{H}_n^{(2)}(\beta_0 r)] P_n^{(1)}(\cos \theta) \right\}$$

$$E_{\phi}^{t+} = \frac{1}{j\omega\mu_0\varepsilon_0} \frac{1}{r \sin \theta} \frac{\partial^2 A_r^{t+}}{\partial r \partial \phi} + \frac{1}{\varepsilon_0} \frac{1}{r} \frac{\partial F_r^{t+}}{\partial \theta}$$

$$= j \frac{E_0}{\omega\mu_0\varepsilon_0 r \sin \theta} \left\{ \frac{\beta_0}{\omega} \sin \phi \sum_{n=1}^{\infty} [a_n j'_n(\beta_0 r) + b_n \hat{H}_n^{(2)}(\beta_0 r)] P_n^{(1)}(\cos \theta) \right\}$$

$$+ \frac{E_0}{\varepsilon_0 r} \left\{ \frac{1}{\omega\eta_0} \sin \phi \sum_{n=1}^{\infty} [a_n j_n(\beta_0 r) + c_n \hat{H}_n^{(2)}(\beta_0 r)] P_n^{(1)}(\cos \theta) \right\}$$

$$H_{\theta}^{t+} = \frac{1}{\mu_0} \frac{1}{r \sin \theta} \frac{\partial A_r^{t+}}{\partial \phi} + \frac{1}{j\omega\mu_0\varepsilon_0} \frac{1}{r} \frac{\partial^2 F_r^{t+}}{\partial r \partial \theta}$$

$$= - \frac{E_0}{\mu_0 r \sin \theta} \left\{ \frac{1}{\omega} \sin \phi \sum_{n=1}^{\infty} [a_n j_n(\beta_0 r) + b_n \hat{H}_n^{(2)}(\beta_0 r)] P_n^{(1)}(\cos \theta) \right\}$$

$$- j \frac{E_0}{\omega\mu_0\varepsilon_0 r} \left\{ \frac{\beta_0}{\omega\eta_0} \sin \phi \sum_{n=1}^{\infty} [a_n j'_n(\beta_0 r) + c_n \hat{H}_n^{(2)}(\beta_0 r)] P_n^{(1)}(\cos \theta) \right\}$$

$$H_{\phi}^{t+} = - \frac{1}{\mu_0} \frac{1}{r} \frac{\partial A_r^{t+}}{\partial \theta} + \frac{1}{j\omega\mu_0\varepsilon_0} \frac{1}{r \sin \theta} \frac{\partial^2 F_r^{t+}}{\partial r \partial \phi}$$

$$= - \frac{E_0}{\mu_0 r} \left\{ \frac{1}{\omega} \cos \phi \sum_{n=1}^{\infty} [a_n j_n(\beta_0 r) + b_n \hat{H}_n^{(2)}(\beta_0 r)] P_n^{(1)}(\cos \theta) \right\}$$

$$- j \frac{E_0}{\omega\mu_0\varepsilon_0 r \sin \theta} \left\{ \frac{\beta_0}{\omega\eta_0} \cos \phi \sum_{n=1}^{\infty} [a_n j'_n(\beta_0 r) + c_n \hat{H}_n^{(2)}(\beta_0 r)] P_n^{(1)}(\cos \theta) \right\}$$

Cont'd

11.81 contd

Similarly, Equation (11-234) is used to find the fields transverse to the radial direction inside the sphere, which are

$$E_{\theta}^{t-} = \frac{1}{j\omega\mu_d\varepsilon_d} \frac{1}{r} \frac{\partial^2 A_r^{t-}}{\partial r \partial \theta} - \frac{1}{\varepsilon_d} \frac{1}{r \sin \theta} \frac{\partial F_r^{t-}}{\partial \phi}$$

$$= -j \frac{E_0}{\omega\mu_d\varepsilon_d r} \left\{ \frac{\beta_d}{\omega} \cos \phi \sum_{n=1}^{\infty} d_n f'_n(\beta_d r) P_n^{(1)}(\cos \theta) \right\}$$

$$- \frac{E_0}{\varepsilon_d r \sin \theta} \left\{ \frac{1}{\omega\eta_d} \cos \phi \sum_{n=1}^{\infty} e_n f'_n(\beta_d r) P_n^{(1)}(\cos \theta) \right\}$$

$$E_{\phi}^{t-} = \frac{1}{j\omega\mu_d\varepsilon_d} \frac{1}{r \sin \theta} \frac{\partial^2 A_r^{t-}}{\partial r \partial \phi} + \frac{1}{\varepsilon_d} \frac{1}{r} \frac{\partial F_r^{t-}}{\partial \theta}$$

$$= j \frac{E_0}{\omega\mu_d\varepsilon_d r \sin \theta} \left\{ \frac{\beta_d}{\omega} \sin \phi \sum_{n=1}^{\infty} d_n f'_n(\beta_d r) P_n^{(1)}(\cos \theta) \right\}$$

$$+ \frac{E_0}{\varepsilon_d r} \left\{ \frac{1}{\omega\eta_d} \sin \phi \sum_{n=1}^{\infty} e_n f'_n(\beta_d r) P_n^{(1)}(\cos \theta) \right\}$$

$$H_{\theta}^{t-} = \frac{1}{\mu_d} \frac{1}{r \sin \theta} \frac{\partial A_r^{t-}}{\partial \phi} + \frac{1}{j\omega\mu_d\varepsilon_d} \frac{1}{r} \frac{\partial^2 F_r^{t-}}{\partial r \partial \theta}$$

$$= - \frac{E_0}{\mu_d r \sin \theta} \left\{ \frac{1}{\omega} \sin \phi \sum_{n=1}^{\infty} d_n f'_n(\beta_d r) P_n^{(1)}(\cos \theta) \right\}$$

$$- j \frac{E_0}{\omega\mu_d\varepsilon_d r} \left\{ \frac{\beta_d}{\omega\eta_d} \sin \phi \sum_{n=1}^{\infty} e_n f'_n(\beta_d r) P_n^{(1)}(\cos \theta) \right\}$$

$$H_{\phi}^{t-} = - \frac{1}{\mu_d} \frac{1}{r} \frac{\partial A_r^{t-}}{\partial \theta} + \frac{1}{j\omega\mu_d\varepsilon_d} \frac{1}{r \sin \theta} \frac{\partial^2 F_r^{t-}}{\partial r \partial \phi}$$

$$= - \frac{E_0}{\mu_d r} \left\{ \frac{1}{\omega} \cos \phi \sum_{n=1}^{\infty} d_n f'_n(\beta_d r) P_n^{(1)}(\cos \theta) \right\}$$

$$- j \frac{E_0}{\omega\mu_d\varepsilon_d r \sin \theta} \left\{ \frac{\beta_d}{\omega\eta_d} \cos \phi \sum_{n=1}^{\infty} e_n f'_n(\beta_d r) P_n^{(1)}(\cos \theta) \right\}$$

Contd

11.81 cont'd

where

$$' = \frac{\partial}{\partial(\beta r)}$$

for the spherical Bessel or Hankel function, and

$$' = \frac{\partial}{\partial\theta}$$

for the associated Legendre functions.

The appropriate boundary equations require the continuity of the tangential electric and magnetic fields, and are

$$E_\theta^{t-}(r = a, 0 \leq \theta \leq \pi, 0 \leq \phi \leq 2\pi) = E_\theta^{t+}(r = a, 0 \leq \theta \leq \pi, 0 \leq \phi \leq 2\pi)$$

$$E_\phi^{t-}(r = a, 0 \leq \theta \leq \pi, 0 \leq \phi \leq 2\pi) = E_\phi^{t+}(r = a, 0 \leq \theta \leq \pi, 0 \leq \phi \leq 2\pi)$$

$$H_\theta^{t-}(r = a, 0 \leq \theta \leq \pi, 0 \leq \phi \leq 2\pi) = H_\theta^{t+}(r = a, 0 \leq \theta \leq \pi, 0 \leq \phi \leq 2\pi)$$

$$H_\phi^{t-}(r = a, 0 \leq \theta \leq \pi, 0 \leq \phi \leq 2\pi) = H_\phi^{t+}(r = a, 0 \leq \theta \leq \pi, 0 \leq \phi \leq 2\pi)$$

Cont'd

11.81 cont'd

Applying the first boundary condition yields

$$\begin{aligned}
 & -j \frac{E_0}{\omega \mu_d \epsilon_d a} \left\{ \frac{\beta_d}{\omega} \cos \phi \sum_{n=1}^{\infty} d_n \hat{J}'_n(\beta_d a) P'_n^{(1)}(\cos \theta) \right\} \\
 & - \frac{E_0}{\epsilon_d a \sin \theta} \left\{ \frac{1}{\omega \eta_d} \cos \phi \sum_{n=1}^{\infty} e_n \hat{J}_n(\beta_d a) P_n^{(1)}(\cos \theta) \right\} \\
 & = -j \frac{E_0}{\omega \mu_0 \epsilon_0 a} \left\{ \frac{\beta_0}{\omega} \cos \phi \sum_{n=1}^{\infty} [a_n \hat{J}'_n(\beta_0 a) + b_n \hat{H}'_n^{(2)}(\beta_0 a)] P'_n^{(1)}(\cos \theta) \right\} \\
 & - \frac{E_0}{\epsilon_0 a \sin \theta} \left\{ \frac{1}{\omega \eta_0} \cos \phi \sum_{n=1}^{\infty} [a_n \hat{J}_n(\beta_0 a) + c_n \hat{H}_n^{(2)}(\beta_0 a)] P_n^{(1)}(\cos \theta) \right\}
 \end{aligned}$$

which leads to the following two relations

$$\begin{aligned}
 & -j \frac{E_0}{\omega \mu_d \epsilon_d a} \left\{ \frac{\beta_d}{\omega} \cos \phi \sum_{n=1}^{\infty} d_n \hat{J}'_n(\beta_d a) P'_n^{(1)}(\cos \theta) \right\} \\
 & = -j \frac{E_0}{\omega \mu_0 \epsilon_0 a} \left\{ \frac{\beta_0}{\omega} \cos \phi \sum_{n=1}^{\infty} [a_n \hat{J}'_n(\beta_0 a) + b_n \hat{H}'_n^{(2)}(\beta_0 a)] P'_n^{(1)}(\cos \theta) \right\} \\
 & - \frac{E_0}{\epsilon_d a \sin \theta} \left\{ \frac{1}{\omega \eta_d} \cos \phi \sum_{n=1}^{\infty} e_n \hat{J}_n(\beta_d a) P_n^{(1)}(\cos \theta) \right\} \\
 & = - \frac{E_0}{\epsilon_0 a \sin \theta} \left\{ \frac{1}{\omega \eta_0} \cos \phi \sum_{n=1}^{\infty} [a_n \hat{J}_n(\beta_0 a) + c_n \hat{H}_n^{(2)}(\beta_0 a)] P_n^{(1)}(\cos \theta) \right\}
 \end{aligned}$$

Using

$$\frac{\beta}{\omega^2 \mu \epsilon} = \frac{1}{\omega \eta \epsilon} = \frac{1}{\beta}$$

The above equations reduce to

$$d_n \beta_0 \hat{J}'_n(\beta_d a) = a_n \beta_d \hat{J}'_n(\beta_0 a) + b_n \beta_d \hat{H}'_n^{(2)}(\beta_0 a)$$

$$e_n \beta_0 \hat{J}_n(\beta_d a) = a_n \beta_d \hat{J}_n(\beta_0 a) + c_n \beta_d \hat{H}_n^{(2)}(\beta_0 a)$$

cont'd

11.81 cont'd]

Solving for each of the unknown coefficients yields

$$d_n = \frac{a_n \beta_d \hat{J}'_n(\beta_0 a) + b_n \beta_d \hat{H}'_n^{(2)}(\beta_0 a)}{\beta_0 \hat{J}'_n(\beta_d a)}$$

$$e_n = \frac{a_n \beta_d \hat{J}_n(\beta_0 a) + c_n \beta_d \hat{H}_n^{(2)}(\beta_0 a)}{\beta_0 \hat{J}_n(\beta_d a)}$$

$$b_n = \frac{d_n \beta_0 \hat{J}'_n(\beta_d a) - a_n \beta_d \hat{J}'_n(\beta_0 a)}{\beta_d \hat{H}'_n^{(2)}(\beta_0 a)}$$

$$c_n = \frac{e_n \beta_0 \hat{J}_n(\beta_d a) - a_n \beta_d \hat{J}_n(\beta_0 a)}{\beta_d \hat{H}_n^{(2)}(\beta_0 a)}$$

Applying the second boundary condition from above will lead to the same equations. Similarly, it is only necessary to apply one of the two boundary conditions equating the tangential magnetic fields. Applying the third boundary condition yields

$$\begin{aligned} & -\frac{E_0}{\mu_d a \sin \theta} \left\{ \frac{1}{\omega} \sin \phi \sum_{n=1}^{\infty} d_n \hat{J}_n(\beta_d a) P_n^{(1)}(\cos \theta) \right\} \\ & - j \frac{E_0}{\omega \mu_d \epsilon_d a} \left\{ \frac{\beta_d}{\omega \eta_d} \sin \phi \sum_{n=1}^{\infty} e_n \hat{J}'_n(\beta_d a) P'_n^{(1)}(\cos \theta) \right\} \\ & = -\frac{E_0}{\mu_0 a \sin \theta} \left\{ \frac{1}{\omega} \sin \phi \sum_{n=1}^{\infty} [a_n \hat{J}_n(\beta_0 a) + b_n \hat{H}_n^{(2)}(\beta_0 a)] P_n^{(1)}(\cos \theta) \right\} \\ & - j \frac{E_0}{\omega \mu_0 \epsilon_0 a} \left\{ \frac{\beta_0}{\omega \eta_0} \sin \phi \sum_{n=1}^{\infty} [a_n \hat{J}'_n(\beta_0 a) + c_n \hat{H}'_n^{(2)}(\beta_0 a)] P'_n^{(1)}(\cos \theta) \right\} \end{aligned}$$

11.81 cont'd

which leads to the following two relations

$$\begin{aligned} & -\frac{E_0}{\mu_d a \sin \theta} \left\{ \frac{1}{\omega} \sin \phi \sum_{n=1}^{\infty} d_n \hat{J}_n(\beta_d a) P_n^{(1)}(\cos \theta) \right\} \\ & = -\frac{E_0}{\mu_0 a \sin \theta} \left\{ \frac{1}{\omega} \sin \phi \sum_{n=1}^{\infty} [a_n \hat{J}_n(\beta_0 a) + b_n \hat{H}_n^{(2)}(\beta_0 a)] P_n^{(1)}(\cos \theta) \right\} \end{aligned}$$

$$\begin{aligned} & -j \frac{E_0}{\omega \mu_d \epsilon_d a} \left\{ \frac{\beta_d}{\omega \eta_d} \sin \phi \sum_{n=1}^{\infty} e_n \hat{J}'_n(\beta_d a) P'_n^{(1)}(\cos \theta) \right\} \\ & = -j \frac{E_0}{\omega \mu_0 \epsilon_0 a} \left\{ \frac{\beta_0}{\omega \eta_0} \sin \phi \sum_{n=1}^{\infty} [a_n \hat{J}'_n(\beta_0 a) + c_n \hat{H}'_n^{(2)}(\beta_0 a)] P'_n^{(1)}(\cos \theta) \right\} \end{aligned}$$

Using

$$\frac{1}{\omega \mu} = \frac{\beta}{\omega^2 \mu \epsilon \eta} = \frac{1}{\beta \eta}$$

These reduce to

$$d_n \beta_0 \eta_0 \hat{J}_n(\beta_d a) = a_n \beta_d \eta_d \hat{J}_n(\beta_0 a) + b_n \beta_d \eta_d \hat{H}_n^{(2)}(\beta_0 a)$$

$$e_n \beta_0 \eta_0 \hat{J}'_n(\beta_d a) = a_n \beta_d \eta_d \hat{J}'_n(\beta_0 a) + c_n \beta_d \eta_d \hat{H}'_n^{(2)}(\beta_0 a)$$

Solving for each of the unknown coefficients yields

$$d_n = \frac{a_n \beta_d \eta_d \hat{J}_n(\beta_0 a) + b_n \beta_d \eta_d \hat{H}_n^{(2)}(\beta_0 a)}{\beta_0 \eta_0 \hat{J}_n(\beta_d a)}$$

$$e_n = \frac{a_n \beta_d \eta_d \hat{J}'_n(\beta_0 a) + c_n \beta_d \eta_d \hat{H}'_n^{(2)}(\beta_0 a)}{\beta_0 \eta_0 \hat{J}'_n(\beta_d a)}$$

$$b_n = \frac{d_n \beta_0 \eta_0 \hat{J}_n(\beta_d a) - a_n \beta_d \eta_d \hat{J}_n(\beta_0 a)}{\beta_d \eta_d \hat{H}_n^{(2)}(\beta_0 a)}$$

$$c_n = \frac{e_n \beta_0 \eta_0 \hat{J}'_n(\beta_d a) - a_n \beta_d \eta_d \hat{J}'_n(\beta_0 a)}{\beta_d \eta_d \hat{H}'_n^{(2)}(\beta_0 a)}$$

Cont'd

11.81 cont'd

Equating the expressions of the coefficients found from applying the boundary conditions,

$$d_n = \frac{a_n \beta_d \hat{J}'_n(\beta_0 a) + b_n \beta_d \hat{H}'_n^{(2)}(\beta_0 a)}{\beta_0 \hat{J}'_n(\beta_d a)} = \frac{a_n \beta_d \eta_d \hat{J}_n(\beta_0 a) + b_n \beta_d \eta_d \hat{H}_n^{(2)}(\beta_0 a)}{\beta_0 \eta_0 \hat{J}_n(\beta_d a)}$$

$$e_n = \frac{a_n \beta_d \hat{J}_n(\beta_0 a) + c_n \beta_d \hat{H}_n^{(2)}(\beta_0 a)}{\beta_0 \hat{J}_n(\beta_d a)} = \frac{a_n \beta_d \eta_d \hat{J}'_n(\beta_0 a) + c_n \beta_d \eta_d \hat{H}'_n^{(2)}(\beta_0 a)}{\beta_0 \eta_0 \hat{J}'_n(\beta_d a)}$$

$$b_n = \frac{d_n \beta_0 \hat{J}'_n(\beta_d a) - a_n \beta_d \hat{J}'_n(\beta_0 a)}{\beta_d \hat{H}'_n^{(2)}(\beta_0 a)} = \frac{d_n \beta_0 \eta_0 \hat{J}_n(\beta_d a) - a_n \beta_d \eta_d \hat{J}_n(\beta_0 a)}{\beta_d \eta_d \hat{H}_n^{(2)}(\beta_0 a)}$$

$$c_n = \frac{e_n \beta_0 \hat{J}_n(\beta_d a) - a_n \beta_d \hat{J}_n(\beta_0 a)}{\beta_d \hat{H}_n^{(2)}(\beta_0 a)} = \frac{e_n \beta_0 \eta_0 \hat{J}'_n(\beta_d a) - a_n \beta_d \eta_d \hat{J}'_n(\beta_0 a)}{\beta_d \eta_d \hat{H}'_n^{(2)}(\beta_0 a)}$$

Cross-multiplying and factoring out the propagation constants when possible,

$$\begin{aligned} & a_n \eta_0 \hat{J}'_n(\beta_0 a) \hat{J}_n(\beta_d a) + b_n \eta_0 \hat{H}'_n^{(2)}(\beta_0 a) \hat{J}_n(\beta_d a) \\ &= a_n \eta_d \hat{J}_n(\beta_0 a) \hat{J}'_n(\beta_d a) + b_n \eta_d \hat{H}_n^{(2)}(\beta_0 a) \hat{J}'_n(\beta_d a) \end{aligned}$$

$$\begin{aligned} & a_n \eta_0 \hat{J}_n(\beta_0 a) \hat{J}'_n(\beta_d a) + c_n \eta_0 \hat{H}_n^{(2)}(\beta_0 a) \hat{J}'_n(\beta_d a) \\ &= a_n \eta_d \hat{J}'_n(\beta_0 a) \hat{J}_n(\beta_d a) + c_n \eta_d \hat{H}_n^{(2)}(\beta_0 a) \hat{J}_n(\beta_d a) \end{aligned}$$

$$\begin{aligned} & d_n \beta_0 \eta_d \hat{H}_n^{(2)}(\beta_0 a) \hat{J}'_n(\beta_d a) - a_n \beta_d \eta_d \hat{J}'_n(\beta_0 a) \hat{H}_n^{(2)}(\beta_0 a) \\ &= d_n \beta_0 \eta_0 \hat{H}'_n^{(2)}(\beta_0 a) \hat{J}_n(\beta_d a) - a_n \beta_d \eta_d \hat{J}_n(\beta_0 a) \hat{H}'_n^{(2)}(\beta_0 a) \end{aligned}$$

$$\begin{aligned} & e_n \beta_0 \eta_d \hat{H}'_n^{(2)}(\beta_0 a) \hat{J}_n(\beta_d a) - a_n \beta_d \eta_d \hat{J}_n(\beta_0 a) \hat{H}'_n^{(2)}(\beta_0 a) \\ &= e_n \beta_0 \eta_0 \hat{H}_n^{(2)}(\beta_0 a) \hat{J}'_n(\beta_d a) - a_n \beta_d \eta_d \hat{J}'_n(\beta_0 a) \hat{H}_n^{(2)}(\beta_0 a) \end{aligned}$$

Rearranging the terms,

$$\begin{aligned} & b_n \eta_0 \hat{H}'_n^{(2)}(\beta_0 a) \hat{J}_n(\beta_d a) - b_n \eta_d \hat{H}_n^{(2)}(\beta_0 a) \hat{J}'_n(\beta_d a) \\ &= -a_n \eta_0 \hat{J}'_n(\beta_0 a) \hat{J}_n(\beta_d a) + a_n \eta_d \hat{J}_n(\beta_0 a) \hat{J}'_n(\beta_d a) \end{aligned}$$

$$\begin{aligned} & c_n \eta_0 \hat{H}_n^{(2)}(\beta_0 a) \hat{J}'_n(\beta_d a) - c_n \eta_d \hat{H}'_n^{(2)}(\beta_0 a) \hat{J}_n(\beta_d a) \\ &= -a_n \eta_0 \hat{J}_n(\beta_0 a) \hat{J}'_n(\beta_d a) + a_n \eta_d \hat{J}'_n(\beta_0 a) \hat{J}_n(\beta_d a) \end{aligned}$$

Cont'd

11.81 Cont'd

$$d_n \beta_0 \eta_0 \hat{H}'_n^{(2)}(\beta_0 a) \hat{J}_n(\beta_d a) - d_n \beta_0 \eta_d \hat{H}_n^{(2)}(\beta_0 a) \hat{J}'_n(\beta_d a) \\ = a_n \beta_d \eta_d \hat{J}_n(\beta_0 a) \hat{H}'_n^{(2)}(\beta_0 a) - a_n \beta_d \eta_d \hat{J}'_n(\beta_0 a) \hat{H}_n^{(2)}(\beta_0 a)$$

$$e_n \beta_0 \eta_0 \hat{H}_n^{(2)}(\beta_0 a) \hat{J}'_n(\beta_d a) - e_n \beta_0 \eta_d \hat{H}'_n^{(2)}(\beta_0 a) \hat{J}_n(\beta_d a) \\ = a_n \beta_d \eta_d \hat{J}'_n(\beta_0 a) \hat{H}_n^{(2)}(\beta_0 a) - a_n \beta_d \eta_d \hat{J}_n(\beta_0 a) \hat{H}'_n^{(2)}(\beta_0 a)$$

Solving for the coefficients,

$$b_n = \frac{-\eta_0 \hat{J}'_n(\beta_0 a) \hat{J}_n(\beta_d a) + \eta_d \hat{J}_n(\beta_0 a) \hat{J}'_n(\beta_d a)}{\eta_0 \hat{H}'_n^{(2)}(\beta_0 a) \hat{J}_n(\beta_d a) - \eta_d \hat{H}_n^{(2)}(\beta_0 a) \hat{J}'_n(\beta_d a)} a_n$$

$$c_n = \frac{-\eta_0 \hat{J}_n(\beta_0 a) \hat{J}'_n(\beta_d a) + \eta_d \hat{J}'_n(\beta_0 a) \hat{J}_n(\beta_d a)}{\eta_0 \hat{H}_n^{(2)}(\beta_0 a) \hat{J}'_n(\beta_d a) - \eta_d \hat{H}'_n^{(2)}(\beta_0 a) \hat{J}_n(\beta_d a)} a_n$$

$$d_n = \frac{\beta_d \eta_d \hat{J}_n(\beta_0 a) \hat{H}'_n^{(2)}(\beta_0 a) - \beta_d \eta_d \hat{J}'_n(\beta_0 a) \hat{H}_n^{(2)}(\beta_0 a)}{\beta_0 \eta_0 \hat{H}'_n^{(2)}(\beta_0 a) \hat{J}_n(\beta_d a) - \beta_0 \eta_d \hat{H}_n^{(2)}(\beta_0 a) \hat{J}'_n(\beta_d a)} a_n$$

$$e_n = \frac{\beta_d \eta_d \hat{J}'_n(\beta_0 a) \hat{H}_n^{(2)}(\beta_0 a) - \beta_d \eta_d \hat{J}_n(\beta_0 a) \hat{H}'_n^{(2)}(\beta_0 a)}{\beta_0 \eta_0 \hat{H}_n^{(2)}(\beta_0 a) \hat{J}'_n(\beta_d a) - \beta_0 \eta_d \hat{H}'_n^{(2)}(\beta_0 a) \hat{J}_n(\beta_d a)} a_n$$

Factoring out the propagation constant and impedance of free space leaves

$$b_n = \frac{-\hat{J}'_n(\beta_0 a) \hat{J}_n(\beta_d a) + \sqrt{\frac{\mu_r}{\epsilon_r}} \hat{J}_n(\beta_0 a) \hat{J}'_n(\beta_d a)}{\hat{H}'_n^{(2)}(\beta_0 a) \hat{J}_n(\beta_d a) - \sqrt{\frac{\mu_r}{\epsilon_r}} \hat{H}_n^{(2)}(\beta_0 a) \hat{J}'_n(\beta_d a)} a_n$$

$$c_n = \frac{-\hat{J}_n(\beta_0 a) \hat{J}'_n(\beta_d a) + \sqrt{\frac{\mu_r}{\epsilon_r}} \hat{J}'_n(\beta_0 a) \hat{J}_n(\beta_d a)}{\hat{H}_n^{(2)}(\beta_0 a) \hat{J}'_n(\beta_d a) - \sqrt{\frac{\mu_r}{\epsilon_r}} \hat{H}'_n^{(2)}(\beta_0 a) \hat{J}_n(\beta_d a)} a_n$$

$$d_n = \frac{\mu_r [\hat{J}_n(\beta_0 a) \hat{H}'_n^{(2)}(\beta_0 a) - \hat{J}'_n(\beta_0 a) \hat{H}_n^{(2)}(\beta_0 a)]}{\hat{H}'_n^{(2)}(\beta_0 a) \hat{J}_n(\beta_d a) - \sqrt{\frac{\mu_r}{\epsilon_r}} \hat{H}_n^{(2)}(\beta_0 a) \hat{J}'_n(\beta_d a)} a_n$$

$$e_n = \frac{\mu_r [\hat{J}'_n(\beta_0 a) \hat{H}_n^{(2)}(\beta_0 a) - \hat{J}_n(\beta_0 a) \hat{H}'_n^{(2)}(\beta_0 a)]}{\hat{H}_n^{(2)}(\beta_0 a) \hat{J}'_n(\beta_d a) - \sqrt{\frac{\mu_r}{\epsilon_r}} \hat{H}'_n^{(2)}(\beta_0 a) \hat{J}_n(\beta_d a)} a_n$$

Cont'd

11.81 cont'd

Now, from Equation (IV-49), a relation for spherical Bessel and Hankel functions is

$$\hat{B}_n(\beta r) = \sqrt{\frac{\pi \beta r}{2}} B_{n+1/2}(\beta r)$$

which leads to

$$\begin{aligned} & J_n(\beta_0 a) \hat{H}'_n^{(2)}(\beta_0 a) - \hat{J}'_n(\beta_0 a) \hat{H}_n^{(2)}(\beta_0 a) \\ &= \sqrt{\frac{\pi \beta_0 a}{2}} J_{n+1/2}(\beta_0 a) \sqrt{\frac{\pi \beta_0 a}{2}} H'_{n+1/2}^{(2)}(\beta_0 a) \\ & - \sqrt{\frac{\pi \beta_0 a}{2}} J'_{n+1/2}(\beta_0 a) \sqrt{\frac{\pi \beta_0 a}{2}} H_{n+1/2}^{(2)}(\beta_0 a) \\ &= \frac{\pi \beta_0 a}{2} [J_{n+1/2}(\beta_0 a) H'_{n+1/2}^{(2)}(\beta_0 a) - J'_{n+1/2}(\beta_0 a) H_{n+1/2}^{(2)}(\beta_0 a)] \end{aligned}$$

From Equation (IV-15), the Hankel function of the second kind is defined as

$$H_n^{(2)}(\beta r) = J_n(\beta r) - j Y_n(\beta r)$$

This leads to

$$\begin{aligned} & J_{n+1/2}(\beta_0 a) H'_{n+1/2}^{(2)}(\beta_0 a) - J'_{n+1/2}(\beta_0 a) H_{n+1/2}^{(2)}(\beta_0 a) \\ &= J_{n+1/2}(\beta_0 a) [J'_{n+1/2}(\beta_0 a) - j Y'_{n+1/2}(\beta_0 a)] \\ & - J'_{n+1/2}(\beta_0 a) [J_{n+1/2}(\beta_0 a) - j Y_{n+1/2}(\beta_0 a)] \\ &= J_{n+1/2}(\beta_0 a) J'_{n+1/2}(\beta_0 a) - j J_{n+1/2}(\beta_0 a) Y'_{n+1/2}(\beta_0 a) \\ & - J'_{n+1/2}(\beta_0 a) J_{n+1/2}(\beta_0 a) + j J'_{n+1/2}(\beta_0 a) Y_{n+1/2}(\beta_0 a) \\ &= -j [J_{n+1/2}(\beta_0 a) Y'_{n+1/2}(\beta_0 a) - J'_{n+1/2}(\beta_0 a) Y_{n+1/2}(\beta_0 a)] \end{aligned}$$

Finally, using the Wronskian from Equation (IV-21), which is,

$$J_n(\beta r) Y'_n(\beta r) - J'_n(\beta r) Y_n(\beta r) = \frac{2}{\pi \beta r}$$

leads to

$$-j [J_{n+1/2}(\beta_0 a) Y'_{n+1/2}(\beta_0 a) - J'_{n+1/2}(\beta_0 a) Y_{n+1/2}(\beta_0 a)] = -j \frac{2}{\pi \beta_0 a}$$

Cont'd

11.81 cont'd

It follows that

$$\begin{aligned}
 & \hat{J}_n(\beta_0 a) \hat{H}'_n^{(2)}(\beta_0 a) - \hat{J}'_n(\beta_0 a) \hat{H}_n^{(2)}(\beta_0 a) \\
 &= \frac{\pi \beta_0 a}{2} [J_{n+1/2}(\beta_0 a) H'_{n+1/2}^{(2)}(\beta_0 a) - J'_{n+1/2}(\beta_0 a) H_{n+1/2}^{(2)}(\beta_0 a)] \\
 &= \frac{\pi \beta_0 a}{2} \left\{ -j [J_{n+1/2}(\beta_0 a) Y'_{n+1/2}(\beta_0 a) - J'_{n+1/2}(\beta_0 a) Y_{n+1/2}(\beta_0 a)] \right\} \\
 &= \frac{\pi \beta_0 a}{2} \left\{ -j \frac{2}{\pi \beta_0 a} \right\} = -j
 \end{aligned}$$

Inserting this back into the equations for the coefficients, and multiplying by the complex relative permittivity leaves

$$b_n = \frac{-\sqrt{\epsilon_r} \hat{J}'_n(\beta_0 a) \hat{J}_n(\beta_d a) + \sqrt{\mu_r} \hat{J}_n(\beta_0 a) \hat{J}'_n(\beta_d a)}{\sqrt{\epsilon_r} \hat{H}'_n^{(2)}(\beta_0 a) \hat{J}_n(\beta_d a) - \sqrt{\mu_r} \hat{H}_n^{(2)}(\beta_0 a) \hat{J}'_n(\beta_d a)} a_n$$

$$c_n = \frac{-\sqrt{\epsilon_r} \hat{J}_n(\beta_0 a) \hat{J}'_n(\beta_d a) + \sqrt{\mu_r} \hat{J}'_n(\beta_0 a) \hat{J}_n(\beta_d a)}{\sqrt{\epsilon_r} \hat{H}_n^{(2)}(\beta_0 a) \hat{J}'_n(\beta_d a) - \sqrt{\mu_r} \hat{H}'_n^{(2)}(\beta_0 a) \hat{J}_n(\beta_d a)} a_n$$

$$d_n = -j \frac{\mu_r \sqrt{\epsilon_r}}{\sqrt{\epsilon_r} \hat{H}'_n^{(2)}(\beta_0 a) \hat{J}_n(\beta_d a) - \sqrt{\mu_r} \hat{H}_n^{(2)}(\beta_0 a) \hat{J}'_n(\beta_d a)} a_n$$

$$e_n = +j \frac{\mu_r \sqrt{\epsilon_r}}{\sqrt{\epsilon_r} \hat{H}_n^{(2)}(\beta_0 a) \hat{J}'_n(\beta_d a) - \sqrt{\mu_r} \hat{H}'_n^{(2)}(\beta_0 a) \hat{J}_n(\beta_d a)} a_n$$

11.82

For a dielectric sphere of small radius, the first term may be sufficient to represent the fields. The small-argument approximations of the spherical Bessel and Hankel functions and their derivatives can be found using multiple methods. One is to use Equation (IV-49), which gives us

$$\hat{J}_n(\beta r) = (\beta r) j_n(\beta r)$$

$$\hat{H}_n^{(2)}(\beta r) = (\beta r) h_n^{(2)}(\beta r) = (\beta r) j_n(\beta r) - j(\beta r) y_n(\beta r)$$

where, from Equation (IV-48), for small arguments,

$$j_n(\beta r) \approx \frac{(\beta r)^n}{1 \cdot 3 \cdot 5 \cdots (2n+1)}$$

$$y_n(\beta r) \approx -1 \cdot 3 \cdot 5 \cdots (2n-1)(\beta r)^{-(n+1)}$$

Then

$$j_1(\beta r) \approx \frac{\beta r}{3}$$

$$y_1(\beta r) \approx -\frac{1}{(\beta r)^2}$$

and

$$\hat{J}_1(\beta r) = (\beta r) j_1(\beta r) \approx \frac{(\beta r)^2}{3}$$

$$\hat{H}_1^{(2)}(\beta r) = (\beta r) h_1^{(2)}(\beta r) = (\beta r) j_1(\beta r) - j(\beta r) y_1(\beta r) \approx \frac{(\beta r)^2}{3} + j \frac{1}{\beta r} \approx j \frac{1}{\beta r}$$

It follows that

$$\hat{J}'_1(\beta r) = \frac{\partial \hat{J}_1(\beta r)}{\partial(\beta r)} \approx \frac{\partial}{\partial(\beta r)} \left[\frac{(\beta r)^2}{3} \right] = \frac{2(\beta r)}{3}$$

and

$$\hat{H}'_1^{(2)}(\beta r) = \frac{\partial \hat{H}_1^{(2)}(\beta r)}{\partial(\beta r)} \approx \frac{\partial}{\partial(\beta r)} \left[j \frac{1}{\beta r} \right] = -j \frac{1}{(\beta r)^2}$$

Cont'd

11.82 cont'd

Therefore, the small-argument approximations for the spherical Bessel and Hankel functions and their derivatives to be used are

$$\hat{J}_1(\beta r) \approx \frac{(\beta r)^2}{3}$$

$$\hat{H}_1^{(2)}(\beta r) \approx j \frac{1}{\beta r}$$

$$\hat{J}'_1(\beta r) \approx \frac{2(\beta r)}{3}$$

$$\hat{H}'_1^{(2)}(\beta r) \approx -j \frac{1}{(\beta r)^2}$$

Also,

$$a_1 = -j1.5$$

Inserting the above into the coefficient equations,

$$\begin{aligned} b_1 &= \frac{-\sqrt{\dot{\epsilon}_r} \hat{J}'_1(\beta_0 a) \hat{J}_1(\beta_d a) + \sqrt{\dot{\mu}_r} \hat{J}_1(\beta_0 a) \hat{J}'_1(\beta_d a)}{\sqrt{\dot{\epsilon}_r} \hat{H}'_1^{(2)}(\beta_0 a) \hat{J}_1(\beta_d a) - \sqrt{\dot{\mu}_r} \hat{H}_1^{(2)}(\beta_0 a) \hat{J}'_1(\beta_d a)} a_1 \\ &\approx \frac{-\sqrt{\dot{\epsilon}_r} \left[\frac{2}{3} (\beta_0 a) \right] \left[\frac{1}{3} (\beta_d a)^2 \right] + \sqrt{\dot{\mu}_r} \left[\frac{1}{3} (\beta_0 a)^2 \right] \left[\frac{2}{3} (\beta_d a) \right]}{\sqrt{\dot{\epsilon}_r} \left[-j \frac{1}{(\beta_0 a)^2} \right] \left[\frac{1}{3} (\beta_d a)^2 \right] - \sqrt{\dot{\mu}_r} \left[j \frac{1}{\beta_0 a} \right] \left[\frac{2}{3} (\beta_d a) \right]} (-j1.5) \\ &= (-j1.5) \frac{\frac{2\dot{\mu}_r \dot{\epsilon}_r}{9} (\beta_0 a)^3 \left[-\sqrt{\dot{\epsilon}_r} + \frac{1}{\sqrt{\dot{\epsilon}_r}} \right]}{\left(-j \frac{\dot{\mu}_r \dot{\epsilon}_r}{3} \right) \left[\sqrt{\dot{\epsilon}_r} + \frac{2}{\sqrt{\dot{\epsilon}_r}} \right]} = -(\beta_0 a)^3 \frac{\dot{\epsilon}_r - 1}{\dot{\epsilon}_r + 2} \end{aligned}$$

$$\begin{aligned} c_1 &= \frac{-\sqrt{\dot{\epsilon}_r} \hat{J}_1(\beta_0 a) \hat{J}'_1(\beta_d a) + \sqrt{\dot{\mu}_r} \hat{J}'_1(\beta_0 a) \hat{J}_1(\beta_d a)}{\sqrt{\dot{\epsilon}_r} \hat{H}_1^{(2)}(\beta_0 a) \hat{J}'_1(\beta_d a) - \sqrt{\dot{\mu}_r} \hat{H}'_1^{(2)}(\beta_0 a) \hat{J}_1(\beta_d a)} a_1 \\ &\approx \frac{-\sqrt{\dot{\epsilon}_r} \left[\frac{2}{3} (\beta_0 a)^2 \right] \left[\frac{1}{3} (\beta_d a) \right] + \sqrt{\dot{\mu}_r} \left[\frac{1}{3} (\beta_0 a) \right] \left[\frac{2}{3} (\beta_d a)^2 \right]}{\sqrt{\dot{\epsilon}_r} \left[j \frac{1}{\beta_0 a} \right] \left[\frac{2}{3} (\beta_d a) \right] - \sqrt{\dot{\mu}_r} \left[-j \frac{1}{(\beta_0 a)^2} \right] \left[\frac{1}{3} (\beta_d a)^2 \right]} (-j1.5) \\ &= (-j1.5) \frac{\frac{2\dot{\mu}_r \dot{\epsilon}_r}{9} (\beta_0 a)^3 \left[-\frac{1}{\sqrt{\dot{\mu}_r}} + \sqrt{\dot{\mu}_r} \right]}{\left(j \frac{\dot{\mu}_r \dot{\epsilon}_r}{3} \right) \left[\frac{2}{\sqrt{\dot{\mu}_r}} + \sqrt{\dot{\mu}_r} \right]} = -(\beta_0 a)^3 \frac{\dot{\mu}_r - 1}{\dot{\mu}_r + 2} \end{aligned}$$

Cont'd

11.82 cont'd

$$\begin{aligned}
 d_1 &= -j \frac{\mu_r \sqrt{\epsilon_r}}{\sqrt{\epsilon_r} \hat{H}_1^{(2)}(\beta_0 a) \hat{J}_1(\beta_d a) - \sqrt{\mu_r} \hat{H}_1^{(2)}(\beta_0 a) \hat{J}'_1(\beta_d a)} a_1 \\
 &\simeq -j \frac{\mu_r \sqrt{\epsilon_r}}{\sqrt{\epsilon_r} \left[-j \frac{1}{(\beta_0 a)^2} \right] \left[\frac{1}{3} (\beta_d a)^2 \right] - \sqrt{\mu_r} \left[j \frac{1}{\beta_0 a} \right] \left[\frac{2}{3} (\beta_d a) \right]} (-j1.5) \\
 &= (-1.5) \frac{\mu_r \sqrt{\epsilon_r}}{\left(-j \frac{\mu_r \epsilon_r}{3} \right) \left[\sqrt{\epsilon_r} + \frac{2}{\sqrt{\epsilon_r}} \right]} = \frac{9}{2 j \mu_r \epsilon_r (\epsilon_r + 2)} = \frac{9}{j 2 (\epsilon_r + 2)}
 \end{aligned}$$

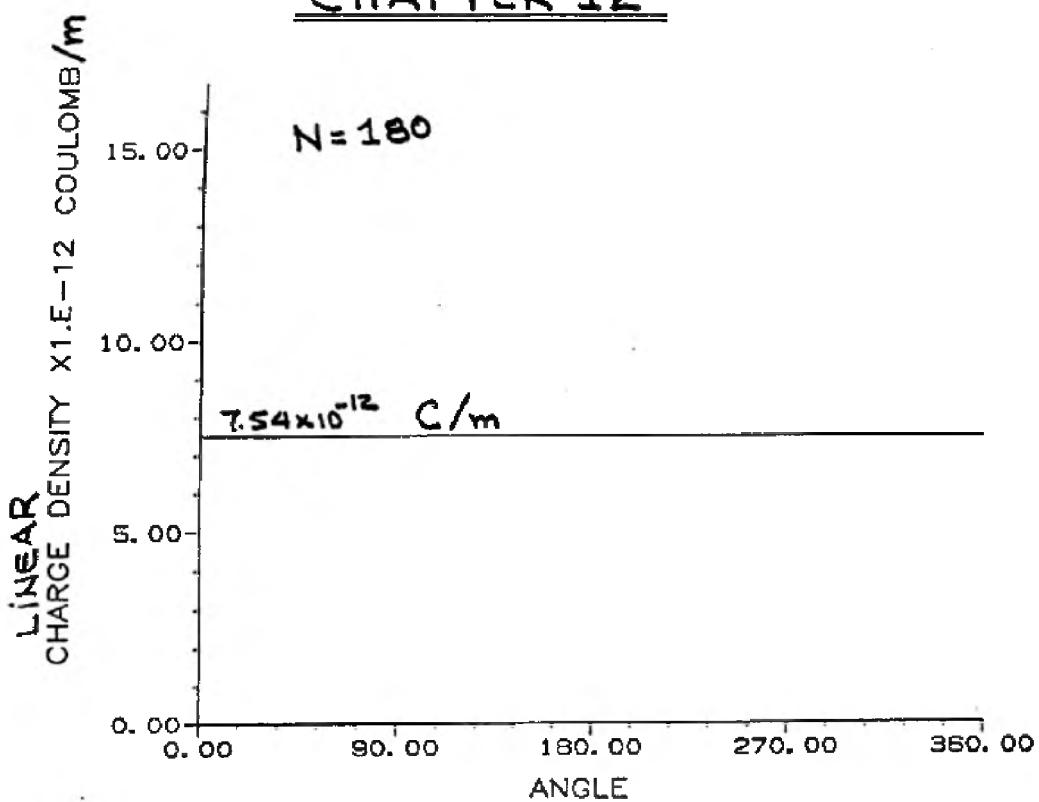
$$\begin{aligned}
 e_1 &= +j \frac{\mu_r \sqrt{\epsilon_r}}{\sqrt{\epsilon_r} \hat{H}_1^{(2)}(\beta_0 a) \hat{J}'_1(\beta_d a) - \sqrt{\mu_r} \hat{H}_1^{(2)}(\beta_0 a) \hat{J}_1(\beta_d a)} a_1 \\
 &\simeq +j \frac{\mu_r \sqrt{\epsilon_r}}{\sqrt{\epsilon_r} \left[j \frac{1}{\beta_0 a} \right] \left[\frac{2}{3} (\beta_d a) \right] - \sqrt{\mu_r} \left[-j \frac{1}{(\beta_0 a)^2} \right] \left[\frac{1}{3} (\beta_d a)^2 \right]} (-j1.5) \\
 &= (1.5) \frac{\mu_r \sqrt{\epsilon_r}}{\left(j \frac{\mu_r \epsilon_r}{3} \right) \left[\frac{2}{\sqrt{\mu_r}} + \sqrt{\mu_r} \right]} = \frac{9 \sqrt{\mu_r}}{j 2 (\sqrt{\epsilon_r}) (\mu_r + 2)}
 \end{aligned}$$

Thus for a dielectric sphere of small radius,

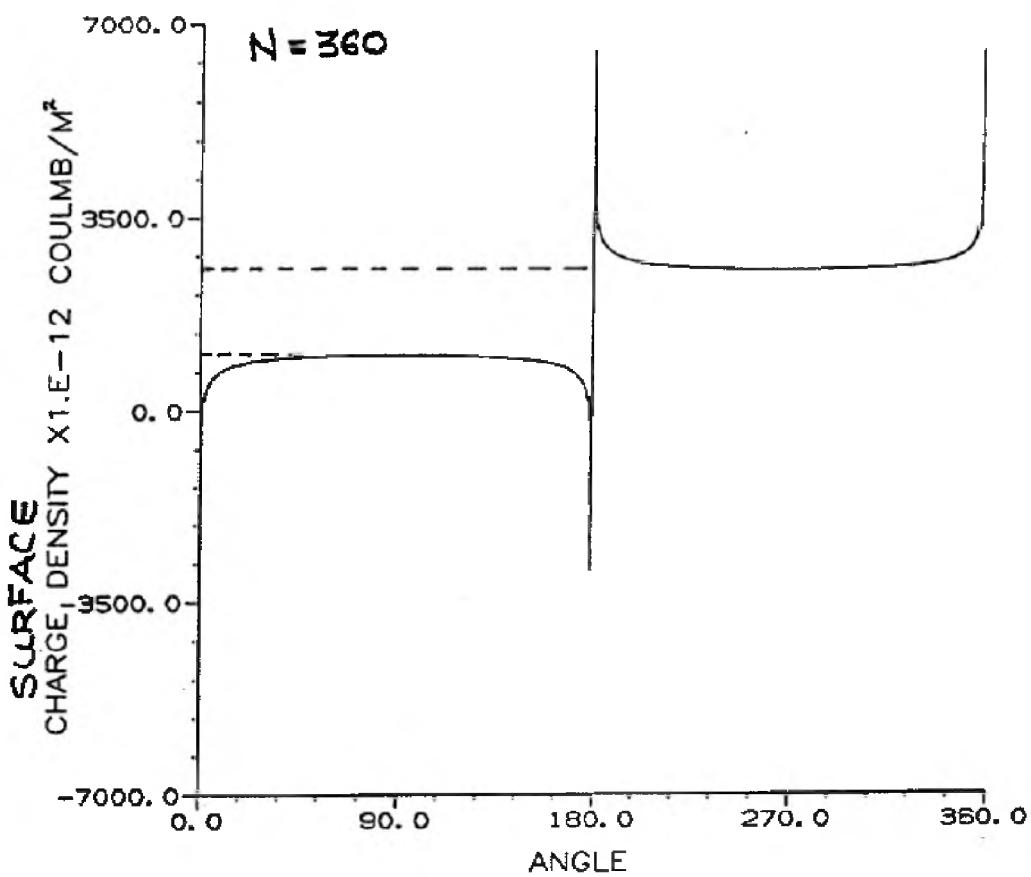
$b_1 \approx -(\beta_0 a)^3 \frac{\epsilon_r - 1}{\epsilon_r + 2}$
$c_1 \approx -(\beta_0 a)^3 \frac{\mu_r - 1}{\mu_r + 2}$
$d_1 \approx \frac{9}{j 2 (\epsilon_r + 2)}$
$e_1 \approx \frac{9 \sqrt{\mu_r}}{j 2 (\sqrt{\epsilon_r}) (\mu_r + 2)}$

CHAPTER 12

12.1



12.2



12.3

Potential

$$V_z = \frac{10a}{\sqrt{a^2 + z^2}} \quad 0 \leq z \leq 7 \quad (1)$$

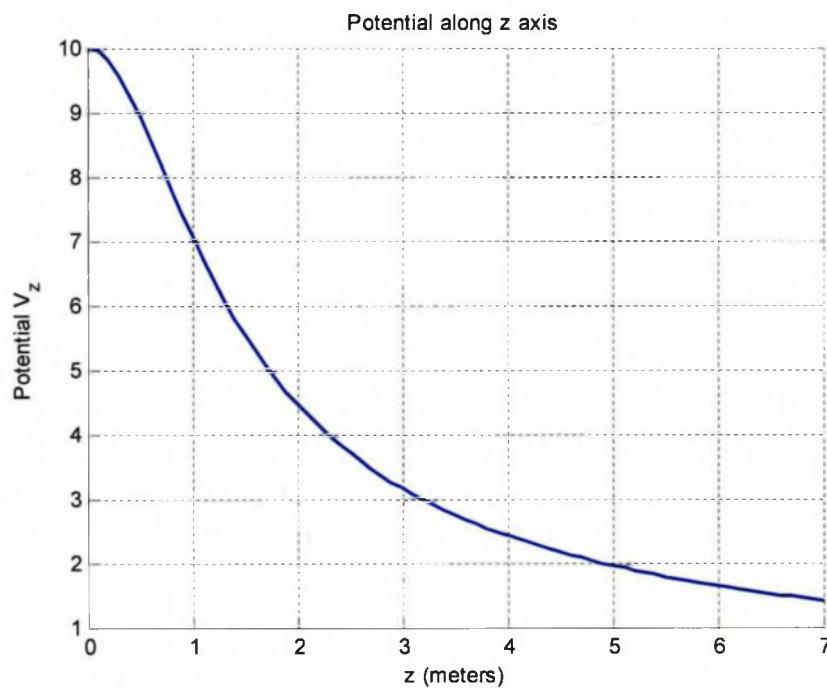


Figure 1

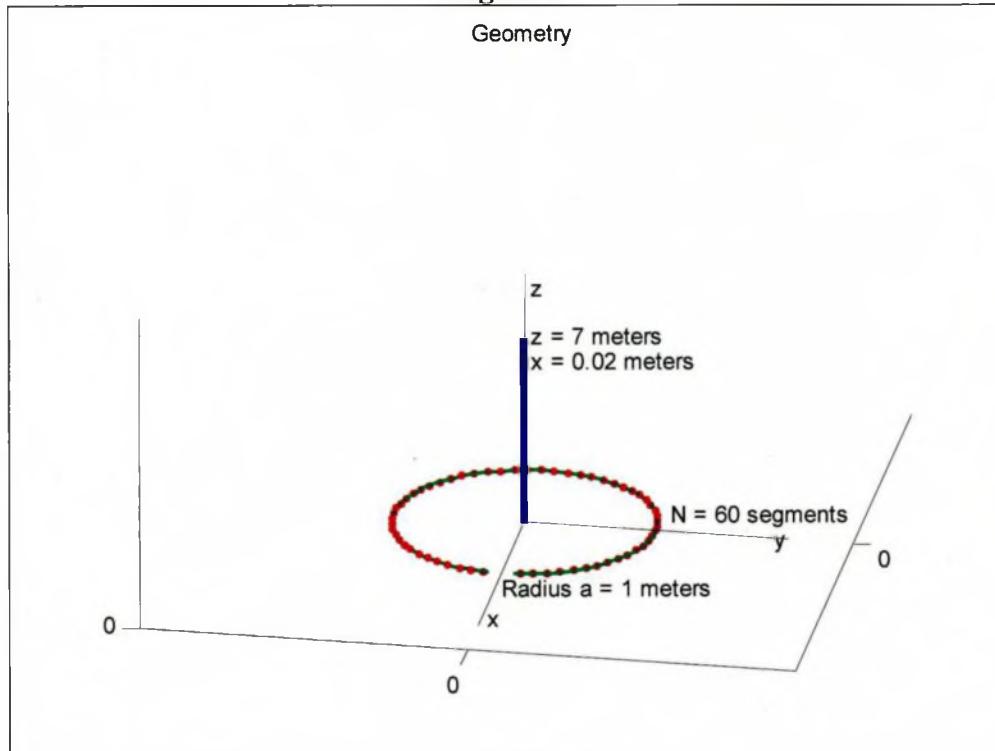


Figure 2

Cont'd

12.3 cont'd

Formulation

Integral equation based on Poisson's differential equation.

$$V = \frac{a}{4\pi\epsilon_0} \int_{\phi_i}^{\phi_f} \frac{\rho(\phi')}{R} d\phi' \quad (2)$$

$$\text{Assume } \rho(\phi') = \sum_{n=1}^N a_n g_n \quad (3)$$

$$\text{where: } g_n = \begin{cases} 1 & \phi'_{n-1} \leq \phi' \leq \phi'_n \\ 0 & \text{elsewhere} \end{cases} \quad (4)$$

Then,

$$\frac{10a}{\sqrt{a^2 + z^2}} = \frac{a}{4\pi\epsilon_0} \sum_{n=1}^N a_n \left(\int_{\phi'_{n-1}}^{\phi'_n} \frac{1}{\sqrt{a^2 + x^2 + z_m^2 - 2ax\cos(\phi')}} d\phi' \right) \quad (5)$$

Therefore, a system of equations can be set up using

$$V_m = 4\pi\epsilon_0 \frac{10}{\sqrt{a^2 + z_m^2}} \quad (6a)$$

$$I_n = a_n \quad (6b)$$

$$Z_{mn} = \sum_{n=1}^N \left(\int_{\phi'_{n-1}}^{\phi'_n} \frac{1}{\sqrt{a^2 + x^2 + z_m^2 - 2ax\cos(\phi')}} d\phi' \right) \quad (6c)$$

The system of equations is

$$\{V_m\} = [Z_{mn}] \{I_{nm}\} \quad (7)$$

This system can be solved in MatLab for the unknown coefficients $a_n = I_n$.

Contd

12.3 Cont'd

Program Output

```

Radius of the disconnected ring (meters)
=> 1
Ring starts at angle (degrees)
=>5
Ring ends at angle (degrees)
=>355
Displacement of the potential from the center along x axis (meters)
=> 0.02
Number of segments
=>60
Warning: Matrix is close to singular or badly scaled.
          Results may be inaccurate. RCOND = 6.444138e-020.
> In TakeHome1b_Kononov at 46
>>

```

Charge Density on the Ring

This problem is the inverse of a problem where it is required to find potential due to a given charge distribution. In fact, V_z in (1) is the potential along z axis due to a uniform charge distribution along a closed ring of radius a . Since our geometry is perturbed only slightly from the geometry of a closed ring, i.e. the ring is discontinuous (see Figure 2), it is reasonable to expect that the charge distribution found using the Integral Equation method will also be nearly uniform. Based on this assumption, we can predict the charge distribution as follows. Using (2)

$$V = \frac{a}{4\pi\epsilon_0} \int_{5^\circ}^{355^\circ} \frac{\rho(\phi')}{R} d\phi' = \frac{al}{4\pi\epsilon_0 R} \int_{5^\circ}^{355^\circ} \rho(\phi') d\phi' = \frac{\rho(\phi')}{4\pi\epsilon_0 R} (355 - 5) \frac{\pi}{180} \quad (8)$$

Equating to (8) to (1),

$$V_z = \frac{10a}{\sqrt{a^2 + z^2}} = \frac{a\rho(\phi')}{4\epsilon_0 R} \frac{35}{18} \quad (9)$$

where

$$\begin{aligned} a &= 1 \\ R &= \sqrt{a^2 + z^2} \end{aligned} \quad (10)$$

Then,

$$\rho(\phi') = 10 \frac{18}{35} 4\epsilon_0 = 182.14 \times 10^{-12} \quad (11)$$

Cont'd

12.3 cont'd

By solving this problem using Integral equation method, we get the following charge density:

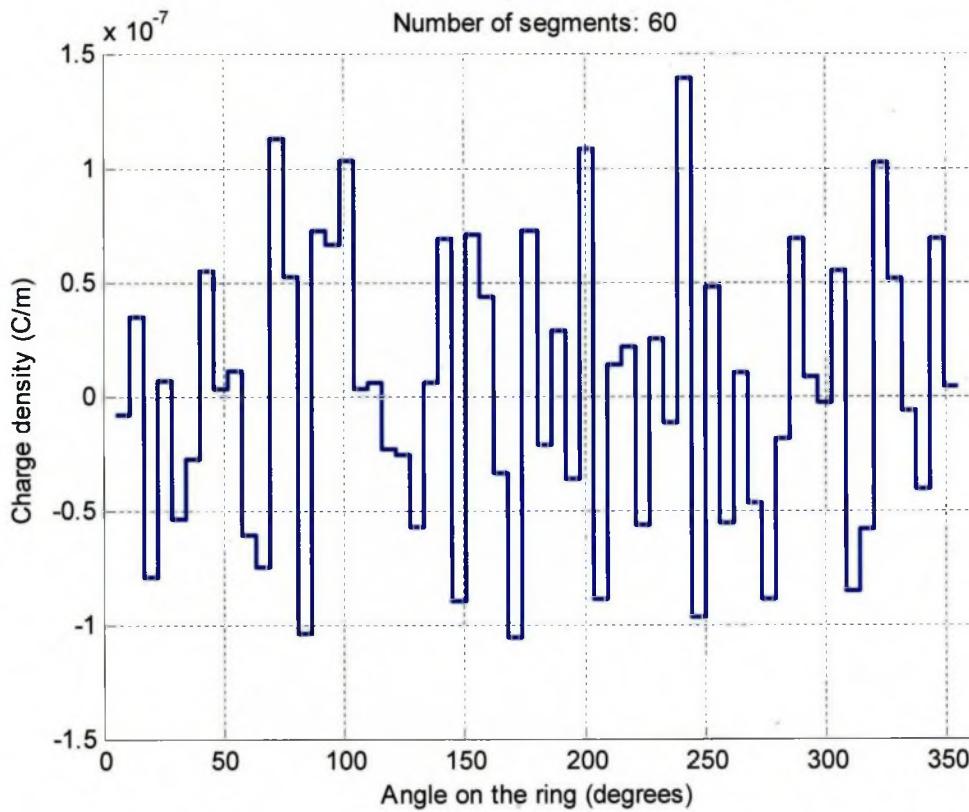


Figure 3

which is obviously non-uniform. This charge density, however, does reproduce the potential specified by (1) as can be seen in Figure 4. This observation suggests that the problem of finding the charge distribution given a potential does not always have a unique solution. In other words, it is possible to find different charge distributions that produce the same potential. In such a case it is said that the problem is 'ill posed'. In order to rectify this problem, more (or different) boundary conditions are required.

Another indication of the problem not having a unique solution is the fact that the matrix $[Z_{mn}]$ is close to singular as is indicated by the following warning message produced by MatLab:

```
Warning: Matrix is close to singular or badly scaled.  
Results may be inaccurate. RCOND = 6.444138e-020.
```

Cont'd

12.3 cont'd

Potential Integral Equation and Equation (1)

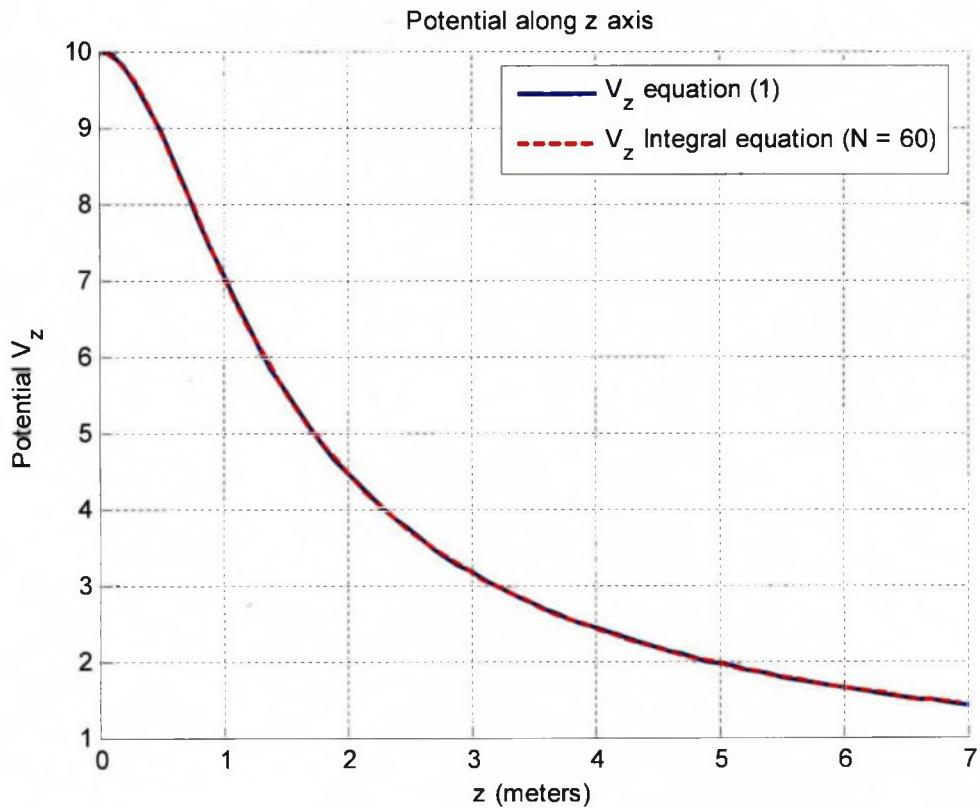
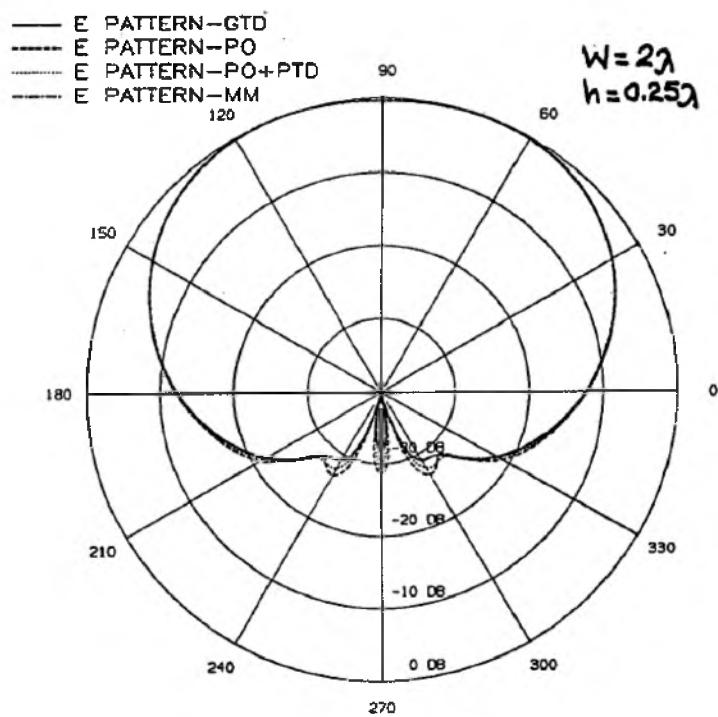
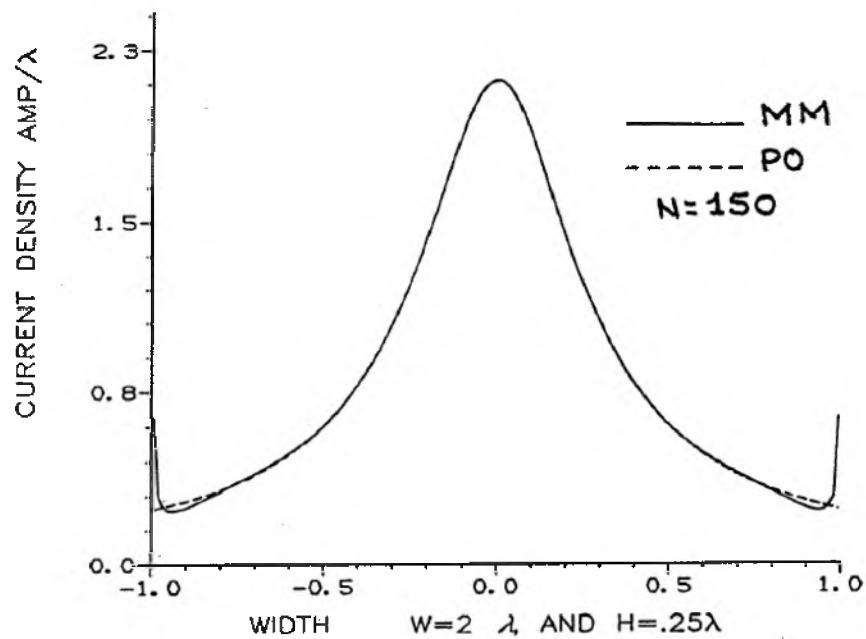


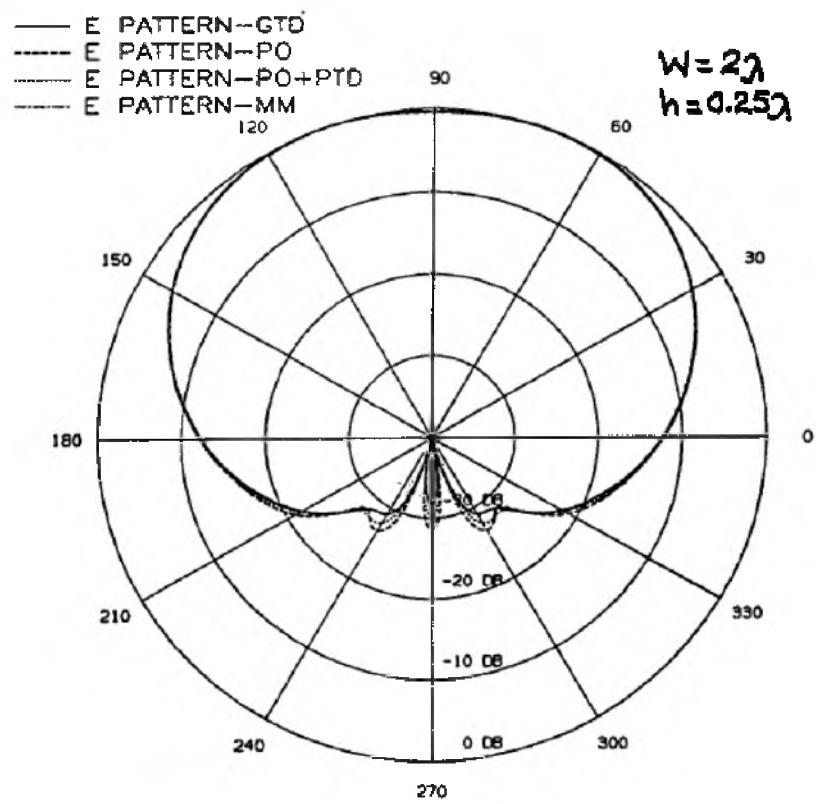
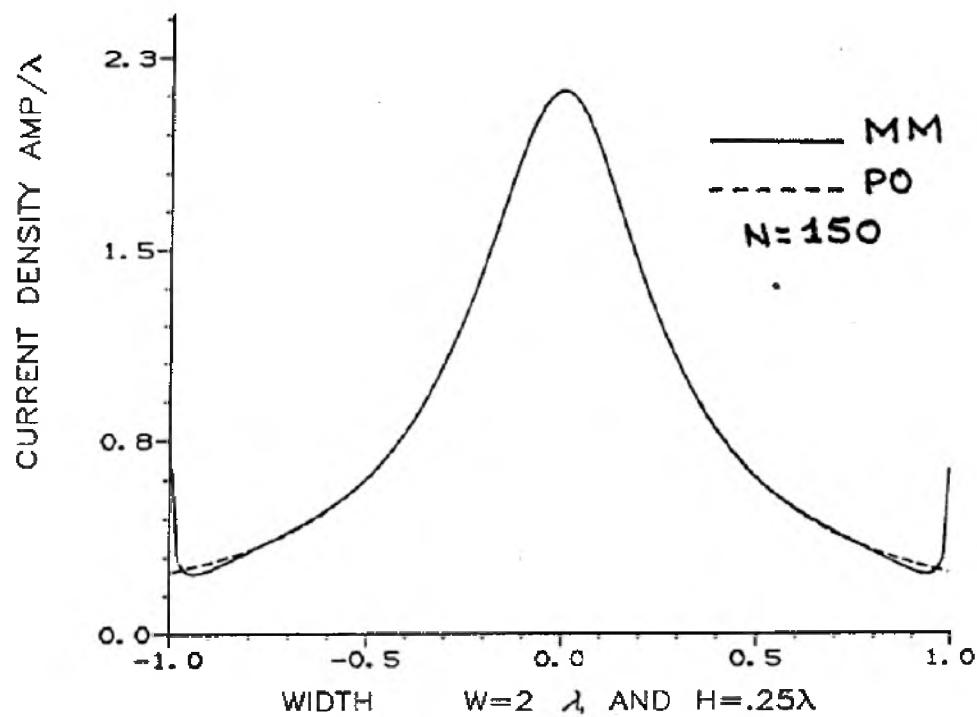
Figure 4

Cont'd

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124



12.5 The H_z component of a magnetic line source of constant current I_m is given by (11-15b)

$$H_z = -I_m \frac{B^2}{4\omega\mu} H_o^{(2)}(\beta p) \quad (11-15b)$$

which is identical in form with the E_z component of an electric line source of constant current I_e as given by (12-12), or

$$E_z = -I_e \frac{\beta^2}{4\omega\epsilon} H_o^{(2)}(\beta p) \quad (12-12)$$

The boundary conditions for a magnetic line source on a PMC, vanishing H^{\tan} on the strip, are identical to those of an electric line source on a PEC, vanishing E^{\tan} on the strip as given by (12-16).

Therefore the Integral Equation for the two problems, for a normalized current $I_m = I_e = 1$ is identical, and it is given by (12-17a) which in this case can be written, to solve for the linear magnetic current density M_z , as

$$H_o^{(2)}(\beta p_m) = - \int_{-W/2}^{W/2} M_z(x') H_o^{(2)}(\beta |p_m - p'|) dx' \quad \boxed{\text{12-17a}}$$

The above equation can also be obtained directly from (12-17a), using duality by referring to Table 7-2; replacing $J_z(x')$ by $M_z(x')$ in (12-17a), or

$$H_o^{(2)}(\beta p_m) = - \int_{-W/2}^{W/2} J_z(x') H_o^{(2)}(\beta |p_m - p'|) dx' \quad (12-17a)$$

12-6 The H_z component of a TE^z polarization plane wave is identical in form to the electric field E_z for the TM^z polarization, which is given by (12-57), or

$$\underline{E}^S = \hat{a}_z E_0 e^{j\beta(x \cos \phi + y \sin \phi)} \quad (12-57)$$

Therefore \underline{H}_z^S can be written for the TE^z polarization as

$$\underline{H}^S = \hat{a}_z H_0 e^{j\beta(x \cos \phi + y \sin \phi)}$$

The vector potential F_z for the TE^z polarization of a 2-D PML strip can be written, using (12-59) and (12-60) as a guide

$$A_z = -j \frac{\hbar}{4} \int_0^W J_z(x) H_0^{(2)}(\beta |p - p'|) dx \quad (12-59)$$

$$\underline{E}^S = -\hat{a}_z \frac{\rho n}{4} \int_0^W J_z(x) H_0^{(2)}(\beta |p - p'|) dx \quad (12-60)$$

and (6-96a) with (6-96b)

$$\underline{A} = \frac{\hbar}{4\pi} \iint_S J_z(x, y, z) \frac{e^{-j\beta R}}{R} ds' \quad (6-96a)$$

$$\underline{E} = \frac{\epsilon}{4\pi} \iint_S M_z(x, y, z) \frac{e^{-j\beta R}}{R} ds' \quad (6-96b)$$

$$F_z = \frac{\epsilon}{4\pi} \iint_S M_z(x) \frac{e^{-j\beta R}}{R} ds' = -j \frac{\epsilon}{4} \int_0^W M_z(x) H_0^{(2)}(\beta |p - x|) dx$$

while by referring to (12-60), and comparing (6-32b) with (6-33b)

$$\underline{H}^S = -\hat{a}_z j \omega F_z = -\hat{a}_z \frac{\omega \epsilon}{4} \int_0^W M_z(x) H_0^{(2)}(\beta |p - x|) dx$$

$$= -\hat{a}_z j \frac{\omega \sqrt{\mu \epsilon}}{4 \sqrt{\mu \epsilon}} \int_0^W M_z(x) H_0^{(2)}(\beta |p - x|) dx$$

$$\underline{H}^S = -\hat{a}_z \frac{\beta}{4\eta} \int_0^W M_z(x) H_0^{(2)}(\beta |p - x|) dx$$

cont'd

12.6 cont'd

The boundary conditions for the PMC strip TE^z polarization (vanishing tangential magnetic field) are identical in form for the PEC strip TM^z polarization (vanishing tangential electric field) as given by (12-63), or

$$\frac{\beta n_k}{4} \int_0^W J_z(x') H_0^{(2)}(\beta|x_m - x'|) dx' = e^{j\beta x_m \cos \phi_i} \quad (12-63)$$

Using duality

$$\beta \rightarrow \beta, \eta \rightarrow \frac{1}{\eta}, J_z \rightarrow M_z$$

and (12-63), we can write the integral equation as

$$\boxed{\frac{\beta}{4\eta} \int_0^W M_z(x') H_0^{(2)}(\beta|x_m - x'|) dx' = e^{j\beta x_m \cos \phi_i}}$$

Easy Way:

The above integral equation can be obtained directly by using (12-63) and duality (Table 7-2), without having to go through all the steps. However having gone through all the steps in deriving it, it gives the reader a better appreciation and understanding.

12.7Show that :

$$\left(\frac{\partial^2}{\partial x^2} + \beta^2 \right) H_0^{(2)}(\beta R) = \frac{\beta^2}{2} [H_0^{(2)}(\beta R) + H_2^{(2)}(\beta R) \cos(2\phi'')]$$

$$\frac{\partial^2}{\partial x \partial y} H_0^{(2)}(\beta R) = \frac{\beta^2}{2} H_2^{(2)}(\beta R) \sin(2\phi'')$$

.....

$$\frac{d}{dx} H_n^{(2)}(dx) = \alpha H_{n-1}^{(2)}(\alpha x) - \frac{n}{x} H_n^{(2)}(\alpha x) \quad (1)$$

$$\frac{d}{dx} H_n^{(2)}(\alpha x) = -\alpha H_{n+1}^{(2)}(\alpha x) + \frac{n}{x} H_n^{(2)}(\alpha x) \quad (2)$$

Using (2) with $n=0$

$$\frac{d}{dx} H_0^{(2)}(\alpha x) = -\alpha H_1^{(2)}(\alpha x) \quad (3)$$

Using (2) with $n=1$

$$\frac{d}{dx} H_1^{(2)}(\alpha x) = -\alpha H_2^{(2)} + \frac{1}{x} H_1^{(2)}(\alpha x)$$

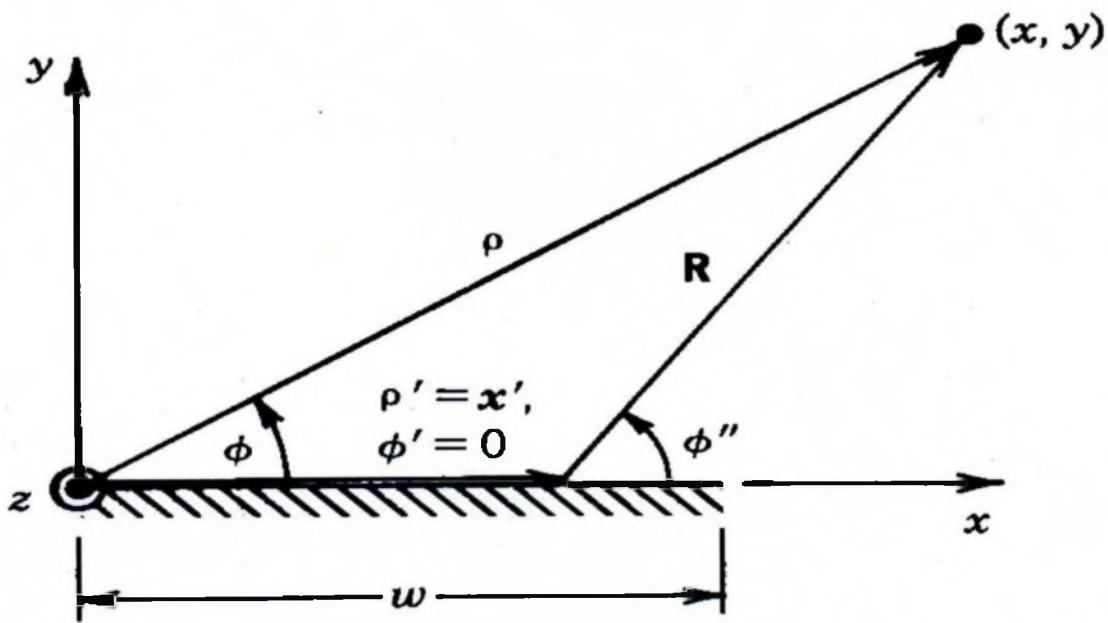
Dividing by x

$$\frac{1}{x} \frac{dH_1^{(2)}(\alpha x)}{dx} = -\frac{\alpha}{x} H_2^{(2)}(\alpha x) + \frac{1}{x^2} H_1^{(2)}(\alpha x)$$

or

$$\frac{1}{x} \frac{dH_1^{(2)}(\alpha x)}{dx} = -\frac{1}{x^2} H_1^{(2)}(\alpha x) = -\frac{\alpha}{x} H_2^{(2)}(\alpha x)$$

Conf&



$$R = \left| \underline{\rho} - \underline{\rho}' \right| = \sqrt{\rho^2 + (\rho')^2 - 2\rho\rho' \cos(\phi_s - \phi')} \Big|_{\phi'=0}$$

$$\simeq \rho - \rho' \cos(\phi_s) = \rho - x' \cos \phi_s$$

Figure 12.4 Geometry for scattering from a strip.

cont'd

12.7 cont'd

or

$$\underbrace{\frac{1}{\alpha x} \frac{d}{dx} H_1^{(2)}(\alpha x) - \frac{1}{\alpha x^2} H_1^{(2)}(\alpha x)}_{\frac{d}{dx} \left[\frac{1}{\alpha x} H_1^{(2)}(\alpha x) \right]} = -\frac{1}{x} H_2^{(2)}(\alpha x)$$

$$\frac{d}{dx} \left[\frac{1}{\alpha x} H_1^{(2)}(\alpha x) \right] = -\frac{1}{x} H_2^{(2)}(\alpha x) \quad (4)$$

$$\frac{\partial^2}{\partial x^2} H_0^{(2)} \left(\beta \left| \underline{x} - \underline{x}' \right| \right) = \frac{\partial^2}{\partial x^2} H_0^{(2)}(\beta R) = \frac{\partial}{\partial x} \left[\frac{\partial}{\partial x} H_0^{(2)}(\beta R) \right]$$

$$\frac{\partial}{\partial x} H_0^{(2)}(\beta R) = \frac{\partial R}{\partial x} \left[\frac{\partial}{\partial R} H_0^{(2)}(\beta R) \right] = \frac{\partial R}{\partial x} \left[\beta H_1^{(2)}(\beta R) \right]$$

$$\frac{\partial H_0^{(2)}}{\partial x}(\beta R) = \frac{\partial R}{\partial x} \left[-\beta H_1^{(2)}(\beta R) \right]$$

$$R = \left[(x - x')^2 + (y - y')^2 \right]^{1/2}$$

$$= \left[(x - x')^2 + y^2 \right]^{1/2}$$

$$\frac{\partial R}{\partial x} = \frac{1}{2} \left[(x - x')^2 + y^2 \right]^{-1/2} 2(x - x')$$

$$= \frac{x - x'}{\sqrt{(x - x')^2 + y^2}} = \frac{x - x'}{R}$$

Therefore: $\frac{\partial H_0^{(2)}(\beta R)}{\partial x} = \left(\frac{x - x'}{R} \right) \left[-\beta H_1^{(2)}(\beta R) \right]$

$$= -\beta \frac{x - x'}{R} H_1^{(2)}(\beta R)$$

Cont'd

12.7 cont'd

$$\frac{\partial^2}{\partial x^2} H_0^{(2)}(\beta R) = \frac{\partial}{\partial x} \left[\frac{\partial H_0^{(2)}(\beta R)}{\partial x} \right]$$

$$= -\beta \left[\frac{\partial}{\partial x} \left\{ \frac{x - x'}{R} H_1^{(2)}(\beta R) \right\} \right]$$

$$= -\beta \frac{\partial}{\partial x} \left\{ \frac{x - x'}{R} H_1^{(2)}(\beta R) \right\}$$

$$\frac{\partial^2}{\partial x^2} H_0^{(2)}(\beta R) = -\beta \frac{\partial}{\partial x} \left\{ (x - x') \frac{H_1^{(2)}(\beta R)}{R} \right\}$$

$$\frac{\partial^2}{\partial x^2} H_0^{(2)}(\beta R) = -\beta \frac{\partial}{\partial x} \left\{ (x - x') \left[\frac{H_1^{(2)}(\beta R)}{R} \right] \right\}$$

$$\frac{\partial^2}{\partial x^2} H_0^{(2)}(\beta R) = -\beta \left\{ (x - x') \frac{\partial}{\partial x} \left[\frac{H_1^{(2)}(\beta R)}{R} \right] + \frac{H_1^{(2)}(\beta R)}{R} \right\}$$

$$\frac{\partial^2}{\partial x^2} H_0^{(2)}(\beta R) = -\beta \left\{ (x - x') \frac{\partial R}{\partial x} \frac{\partial}{\partial R} \left[\frac{H_1^{(2)}(\beta R)}{R} \right] + \frac{H_1^{(2)}(\beta R)}{R} \right\}$$

$$\frac{\partial^2}{\partial x^2} H_0^{(2)}(\beta R) = -\beta \left\{ (x - x') \underbrace{\frac{\partial R}{\partial x}}_{\frac{(x-x')}{R}} \underbrace{\frac{\partial}{\partial R} \left[\frac{H_1^{(2)}(\beta R)}{R} \right]}_{-\frac{\beta}{R} H_2^{(2)}(\beta R)} + \frac{H_1^{(2)}(\beta R)}{R} \right\}$$

$$\frac{\partial^2}{\partial x^2} H_0^{(2)}(\beta R) = -\beta \left\{ \frac{(x - x')^2}{R} \left[-\frac{\beta}{R} H_2^{(2)}(\beta R) \right] + \frac{H_1^{(2)}(\beta R)}{R} \right\}$$

$$= +\beta^2 \frac{H_2^{(2)}(\beta R)}{R^2} (x - x')^2 - \beta \frac{H_1^{(2)}(\beta R)}{R}$$

cont'd

12.7 Cont'd

Subtracting (1) from (2)

$$0 = -\alpha \left[H_{n+1}^{(2)}(\alpha x) + H_{n-1}^{(2)}(\alpha x) \right] + \frac{2n}{x} H_n^{(2)}(\alpha x)$$

or

$$\frac{1}{x} H_n^{(2)}(\alpha x) = \frac{\alpha}{2} \left[H_{n+1}^{(2)}(\alpha x) + H_{n-1}^{(2)}(\alpha x) \right]$$

For n=1: $\frac{1}{R} H_1^{(2)}(\beta R) = \frac{\beta}{2} \left[H_2^{(2)}(\beta R) + H_0^{(2)}(\beta R) \right]$

$$\frac{\partial^2}{\partial x^2} H_0^{(2)}(\beta R) = \beta^2 \frac{H_2^{(2)}(\beta R)}{R^2} (x - x')^2$$

$$- \beta \frac{\beta}{2} \left[H_2^{(2)}(\beta R) + H_0^{(2)}(\beta R) \right]$$

$$\frac{\partial^2}{\partial x^2} H_0^{(2)}(\beta R) = \frac{\beta^2}{2} \left[\frac{2(x - x')^2}{R^2} - 1 \right] H_2^{(2)}(\beta R) - \frac{\beta^2}{2} H_0^{(2)}(\beta R)$$

$$\frac{\partial^2}{\partial x^2} H_0^{(2)}(\beta R) = \frac{\beta^2}{2} \left[2 \cos^2 \phi'' - 1 \right] H_2^{(2)}(\beta R) - \frac{\beta^2}{2} H_0^{(2)}(\beta R)$$

$$\left(\frac{\partial^2}{\partial x^2} + \beta^2 \right) H_0^{(2)}(\beta R) = \begin{cases} \frac{\beta^2}{2} \underbrace{\left[2 \cos^2 \phi'' - 1 \right]}_{\cos(2\phi'')} H_2^{(2)}(\beta R) \\ - \frac{\beta^2}{2} H_0^{(2)}(\beta R) + \beta^2 H_0^{(2)}(\beta R) \end{cases}$$

$$\boxed{\left(\frac{\partial^2}{\partial x^2} + \beta^2 \right) H_0^{(2)}(\beta R) = \left\{ \frac{\beta^2}{2} \left[H_0^{(2)}(\beta R) + H_2^{(2)}(\beta R) \cos(2\phi'') \right] \right\}} \quad (5)$$

Cont'd

12.7 cont'd

Using (5):

$$E_x^s = -\frac{\eta}{4\beta} \int_0^w J_x(x') \left\{ \left[\frac{\partial^2}{\partial x^2} + \beta^2 \right] H_0^{(2)}(\beta R) \right\} dx'$$

$$E_x^s = -\frac{\eta\beta}{8} \int_0^w J_x(x') \left\{ H_2^{(2)}(\beta R) \cos(2\phi'') + H_0^{(2)}(\beta R) \right\} dx'$$

$$\frac{\partial^2 H_0^{(2)}(\beta R)}{\partial y \partial x} = \frac{\partial}{\partial y} \left[\frac{\partial H_0^{(2)}}{\partial x}(\beta R) \right] = \frac{\partial}{\partial y} \left[-\beta \frac{x - x'}{R} H_1^{(2)}(\beta R) \right]$$

$$= -\beta \frac{\partial}{\partial y} \left[\frac{x - x'}{R} H_1^{(2)}(\beta R) \right] = -\beta \frac{\partial}{\partial y} \left[(x - x') \frac{H_1^{(2)}(\beta R)}{R} \right]$$

$$\frac{\partial^2 H_0^{(2)}(\beta R)}{\partial y \partial x} = -\beta \left[(x - x') \frac{\partial}{\partial y} \left\{ \frac{H_1^{(2)}(\beta R)}{R} \right\} \right.$$

$$\left. + \underbrace{\frac{\partial(x - x')}{\partial y}}_0 \frac{H_1^{(2)}(\beta R)}{R} \right]$$

$$\frac{\partial^2 H_0^{(2)}(\beta R)}{\partial y \partial x} = -\beta(x - x') \frac{\partial R}{\partial y} \frac{\partial}{\partial R} \left[\frac{H_1^{(2)}(\beta R)}{R} \right]$$

$$= -\beta(x - x') \frac{\partial R}{\partial y} \frac{\partial}{\partial R} \left[\frac{H_1^{(2)}(\beta R)}{R} \right]$$

$$\frac{\partial R}{\partial y} = \frac{\partial}{\partial R} \left[(x - x')^2 + y^2 \right]^{1/2} = \frac{1}{2} \left[(x - x')^2 + y^2 \right]^{-1/2} (2y)$$

$$\frac{\partial R}{\partial y} = \frac{y}{\sqrt{(x - x')^2 + y^2}} = \frac{y}{R}$$

Cont'd

12.7 cont'd

$$\frac{\partial^2 H_0^{(2)}(\beta R)}{\partial y \partial x} = -\beta(x - x') \frac{y}{R} \underbrace{\frac{\partial}{\partial R} \left[\frac{H_1^{(2)}(\beta R)}{R} \right]}_{-\frac{\beta}{R} H_2^{(2)}(\beta R)}$$

$$= + \frac{\beta^2 (x - x') y}{R^2} H_2^{(2)}(\beta R)$$

$$\frac{\partial^2 H_0^{(2)}(\beta R)}{\partial y \partial x} = \beta^2 \left(\frac{x - x'}{R} \right) \left(\frac{y}{R} \right) H_2^{(2)}(\beta R)$$

$$\frac{\partial^2 H_0^{(2)}(\beta R)}{\partial y \partial x} = \beta^2 \underbrace{\cos \phi'' \sin \phi''}_{\frac{1}{2} \sin(2\phi'')} H_2^{(2)}(\beta R)$$

$$\boxed{\frac{\partial^2 H_0^{(2)}(\beta R)}{\partial y \partial x} = \frac{1}{2} \beta^2 H_2^{(2)}(\beta R) \sin(2\phi'')} \quad (6)$$

Using (6):

$$\begin{aligned} E_y^s &= -\frac{\eta}{4\beta} \int_0^w J_x(x') \left\{ \frac{\partial^2}{\partial y \partial x} H_0^{(2)}(\beta R) \right\} dx' \\ &= -\frac{\eta}{4\beta} \int_0^w J_x(x') \left\{ \frac{\beta^2}{2} H_2^{(2)}(\beta R) \sin(2\phi'') \right\} dx' \end{aligned}$$

$$\boxed{E_y^s = -\frac{\eta\beta}{8} \int_0^w J_x(x') \left\{ H_2^{(2)}(\beta R) \sin(2\phi'') \right\} dx'}$$

12.8] Using Figure 12.13(b), we can write the incident magnetic field as

$$\underline{H}^L = \hat{a}_z H_0 e^{j\beta(x \cos \phi_i + y \sin \phi_i)}$$

$$\underline{J}_s^{\text{total}} = \hat{n} \times (\underline{H}^L + \underline{H}^R) \Big|_{\substack{\text{strip} \\ y=0}} = \hat{a}_y \times (\hat{a}_z H_z^L + \hat{a}_z H_z^R) \Big|_{\substack{\text{strip} \\ y=0}}$$

Assuming the current density on the strip is the same as that of a strip with infinite width, then the total current density on the strip is equal to twice that due to the incident field. Thus

$$\begin{aligned} \underline{J}_s^t &= \hat{a}_y \times (2 \hat{a}_z H_z^L) = 2 \hat{a}_x H_0 e^{j\beta x \cos \phi_i} = \hat{a}_x J_x^t \\ J_x^t &= 2 H_0 e^{j\beta x \cos \phi_i} \end{aligned}$$

Using (12-59) but for J_x , we can write the corresponding potential

$$\begin{aligned} A_x &= \frac{\mu}{4\pi} \iint_S J_x(x') \frac{e^{-j\beta R}}{R} ds' = \frac{\mu}{4\pi} \int_0^W J_x(x') \left[\int_{-\infty}^{+\infty} \frac{e^{-j\beta \sqrt{(p-p')^2 + (z-z')^2}}}{\sqrt{(p-p')^2 + (z-z')^2}} dx' \right] dz' \\ &= -j \frac{\mu}{4} \int_0^W J_x(x') H_0^{(2)}(\beta |p-p'|) dz' \end{aligned}$$

$$\underline{H}^S = \frac{1}{\mu} \nabla \times A = \hat{a}_z \left(-\frac{1}{\mu} \frac{\partial A_x}{\partial y} \right) = -\hat{a}_z \frac{1}{\mu} \left(-j \frac{\mu}{4} \right) \int_0^W J_x(x') \frac{\partial}{\partial y} \left[H_0^{(2)}(|p-p'|) \right] dz'$$

According to Figure 12-14

$$\frac{\partial}{\partial y} H_0^{(2)}(p|p-p'|) = \frac{\partial}{\partial y} H_0^{(2)}(\beta R) = -\beta \frac{y-y'}{R} H_1^{(2)}(\beta R) = -\beta \sin \phi'' H_1^{(2)}(\beta R)$$

Therefore

$$\underline{H}^S = +\hat{a}_z j \frac{1}{4} \int_0^W J_x(x') \left[-\beta \sin \phi'' H_1^{(2)}(\beta R) \right] dx' = -j \frac{\beta}{4} \int_0^W J_x(x') H_1^{(2)}(\beta R) \sin \phi'' dx'$$

$$\underline{H}^S = -j \frac{\beta}{4} \int_0^W J_x(x') H_1^{(2)}(\beta R) \sin \phi'' dx'$$

$$\text{Since } J_x = 2 H_0 e^{j\beta x \cos \phi_i} \Rightarrow \underline{H}^S = -j H_0 \frac{\beta}{2} \int_0^W H_1^{(2)}(\beta R) e^{j\beta x \cos \phi_i} \sin \phi'' dx'$$

Cont'd

12.8 cont'd

For far-field observations, according to Figure 12-14,

$$\phi' = \phi, p' = x', \phi'' = \phi$$

and the Hankel function is approximated by its asymptotic expansion of

$$R = \sqrt{p^2 + (p')^2 - 2pp' \cos(\phi - \phi')} \underset{\substack{\phi' \approx 0 \\ p' = x'}}{\approx} p - x' \cos \phi$$

$$H_1^{(2)}(\beta R) = \sqrt{\frac{2}{\pi \beta p}} e^{-j(\beta p - 3\pi/4)} e^{j\beta x' \cos \phi}$$

$$= \sqrt{\frac{2}{\pi \beta p}} e^{-j\beta p + j\pi/4} e^{j\beta x' \cos \phi}$$

$$H_1^{(2)}(\beta R) = \sqrt{\frac{2j}{\pi \beta p}} e^{-j\beta p} e^{j\beta x' \cos \phi}$$

Therefore the far zone scattered field is .

$$\underline{H}^S = -jH_0 \frac{\beta}{2} \int_0^W H_1^{(2)}(\beta R) e^{j\beta x' \cos \phi} \sin \phi'' dx'$$

$$\approx -jH_0 \sin \phi \left(\frac{\beta}{2}\right) \sqrt{\frac{2j}{\pi \beta p}} e^{-j\beta p} \int_0^W e^{j\beta x' (\cos \phi + \cos \phi'')} dx'$$

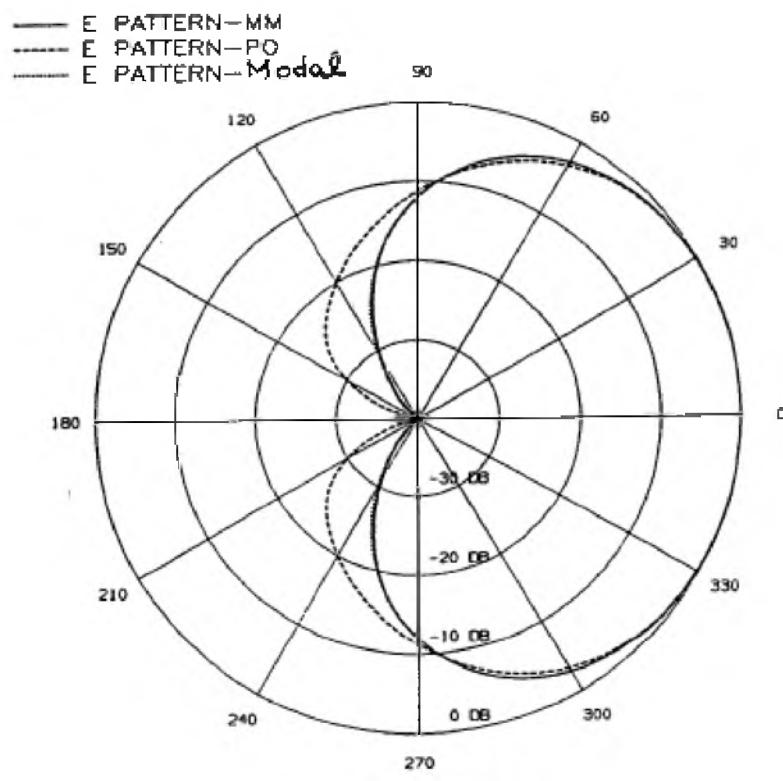
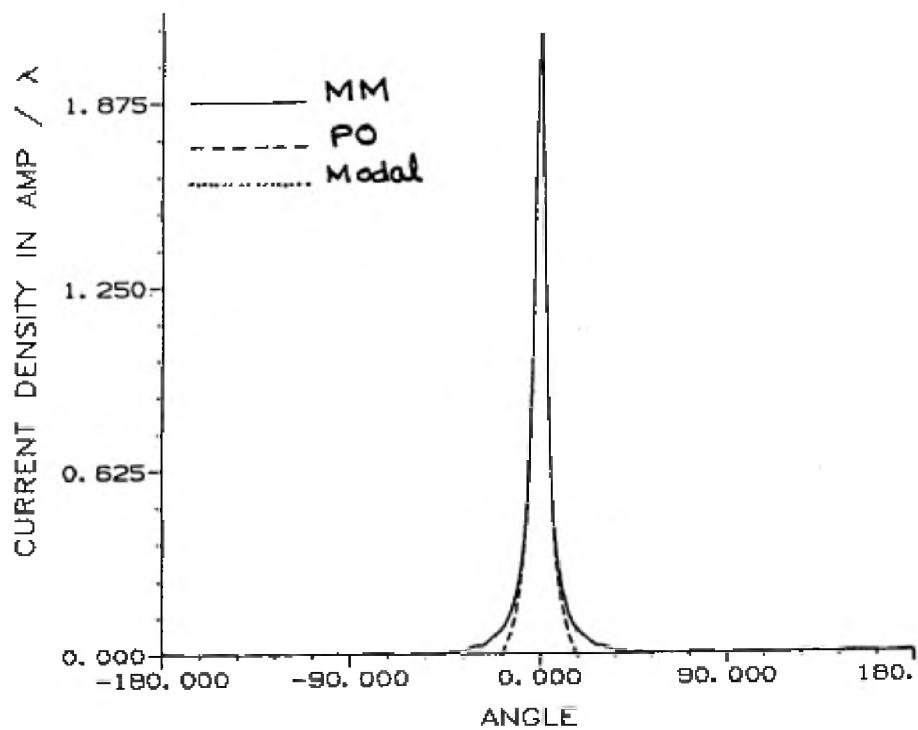
$$\underline{H}^S \approx -H_0 \sqrt{\frac{\beta \sin \phi}{j2\pi p}} e^{-j\beta p} \int_0^W e^{j\beta x' (\cos \phi + \cos \phi'')} dx'$$

According to (11-21c)

$$G_{2D} = \lim_{p \rightarrow \infty} \left[2\pi p \frac{|H^S|^2}{|H^S|^2} \right] = \left[\beta \sin^2 \phi \left| \int_0^W e^{j\beta x' (\cos \phi + \cos \phi'')} dx' \right|^2 \right]$$

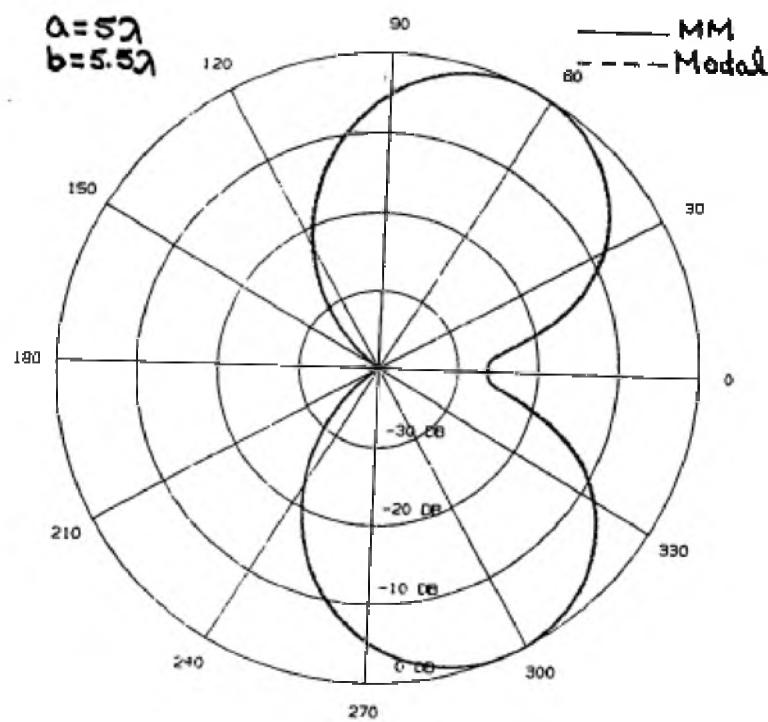
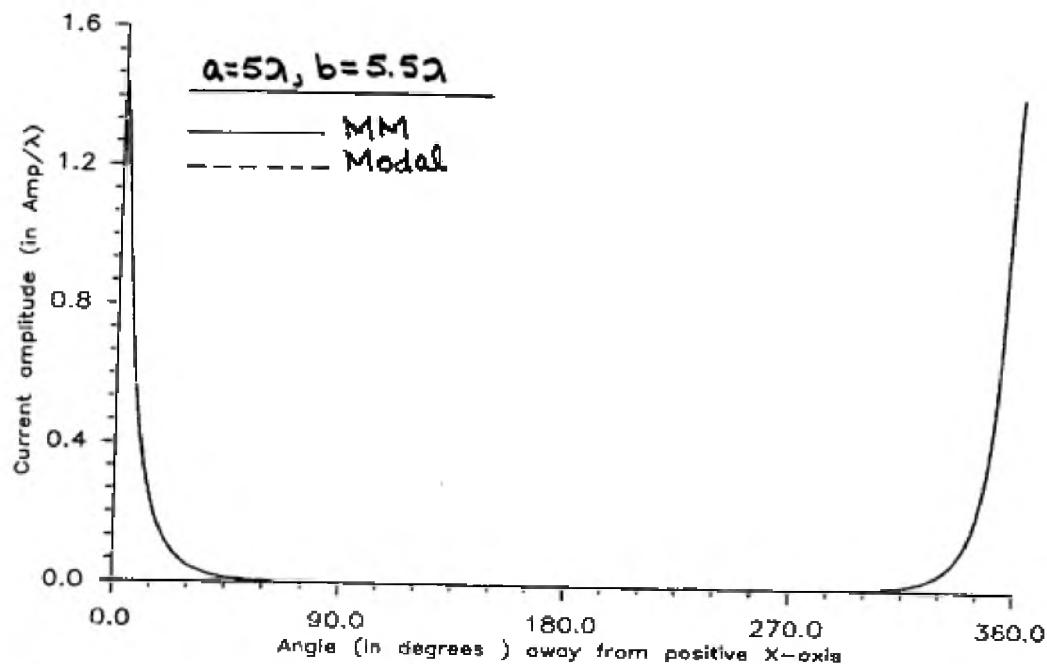
$$G_{2D} = \frac{2\pi}{\lambda} \sin^2 \phi \left| \int_0^W e^{j\beta x' (\cos \phi + \cos \phi'')} dx' \right|^2$$

12.4



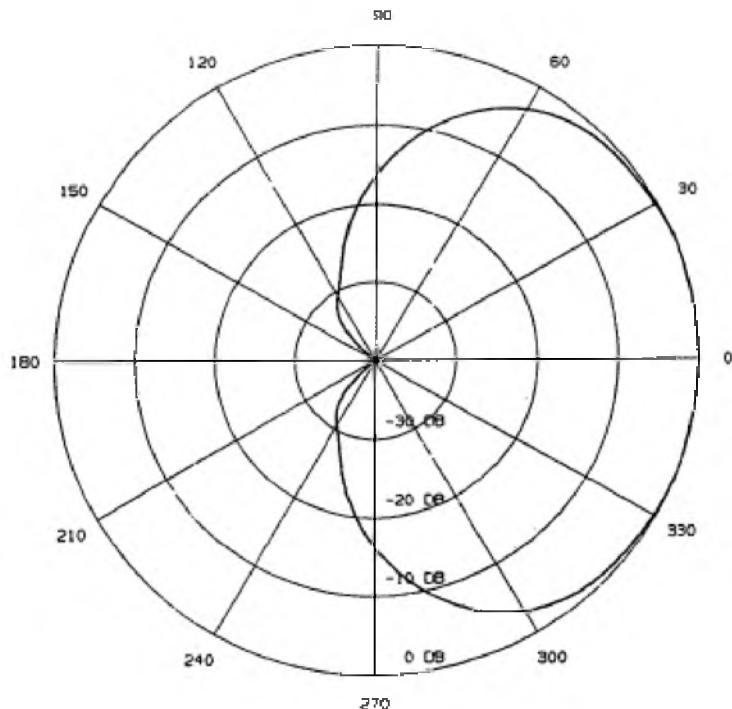
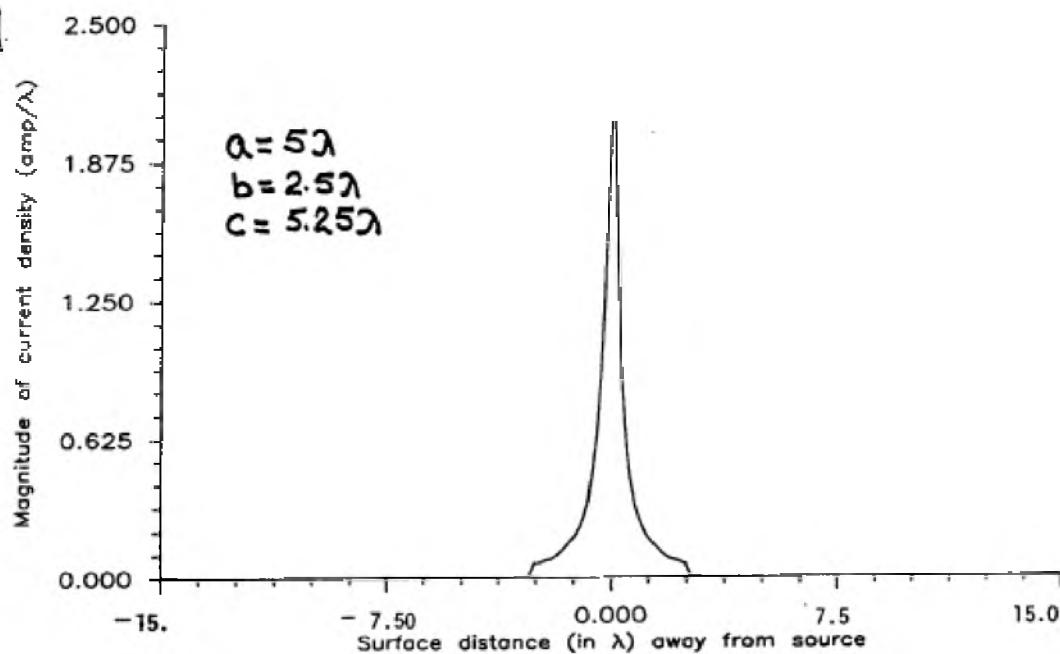
This problem can be solved using the TDRS computer program found at the end of this chapter.

12.10



This problem can be solved using the TDRS computer program at the end of this chapter.

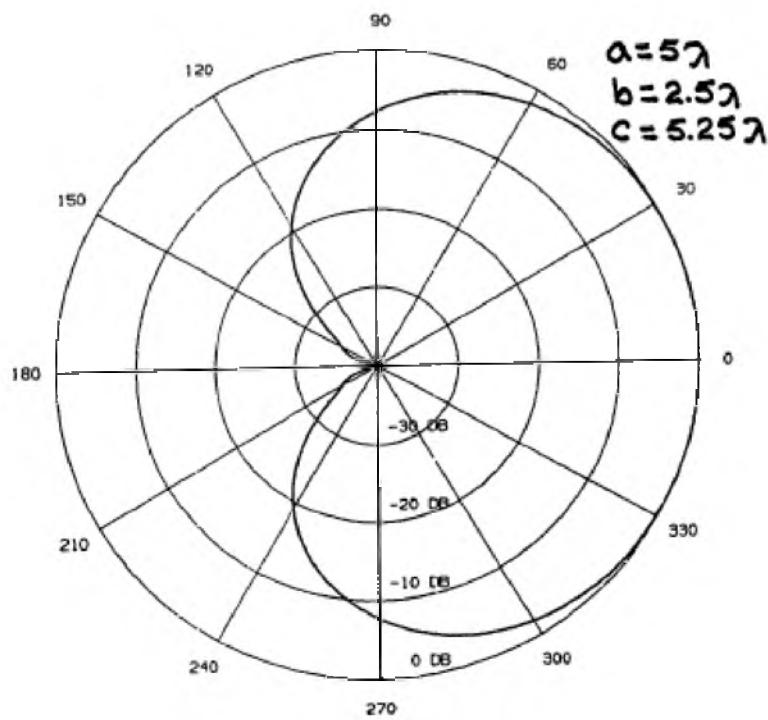
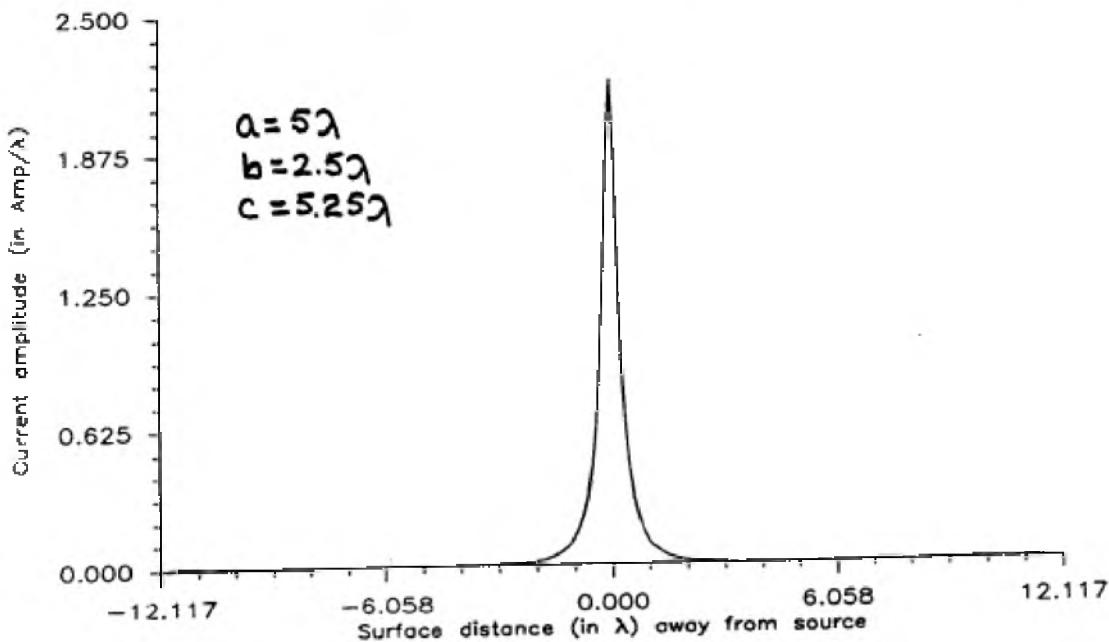
12.11



This problem can be solved using the TDRS program at the end of this chapter.

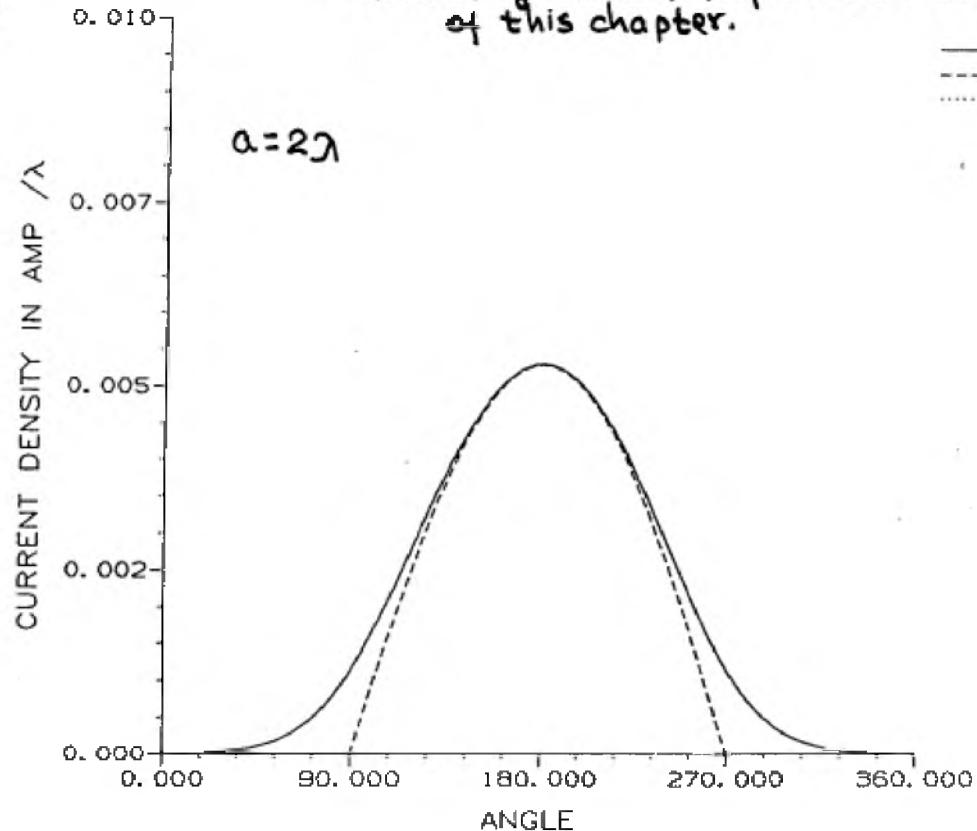
12.12

This problem can be solved using the TDRS program at the end of this chapter.

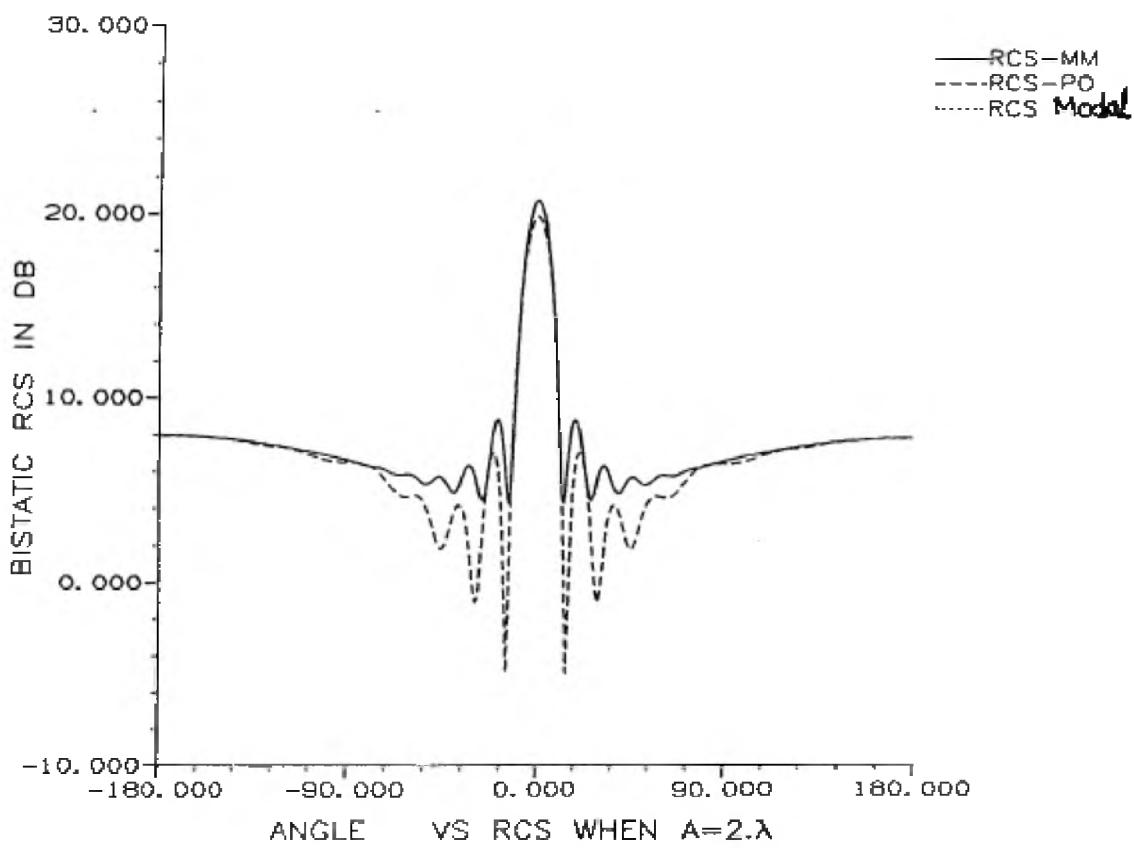


12.13

This problem can be solved using the TDRS program at the end
of this chapter.

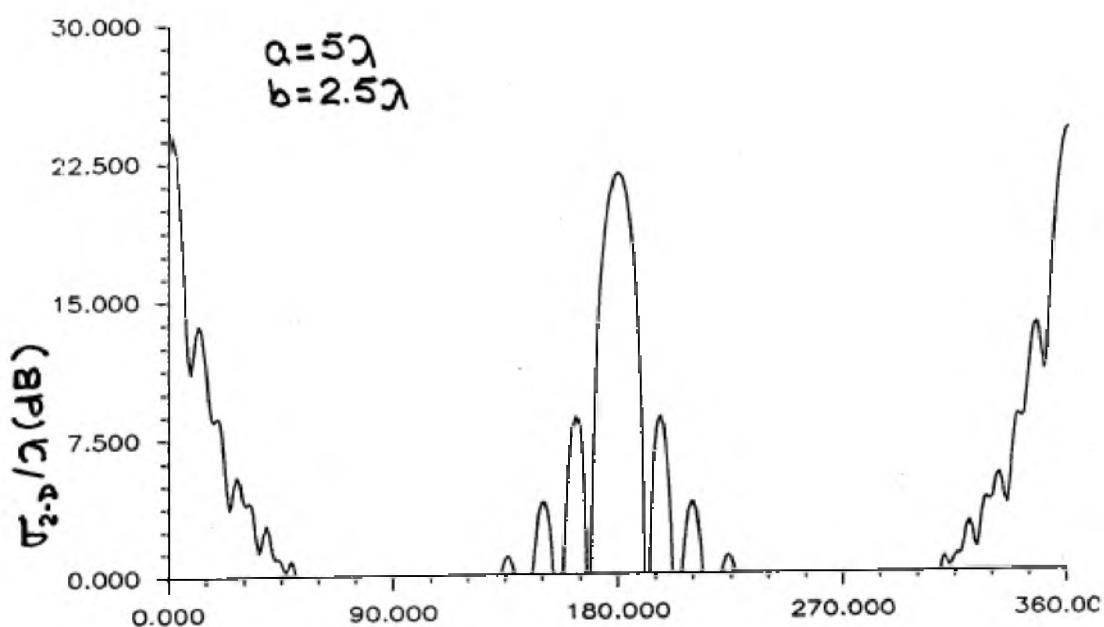
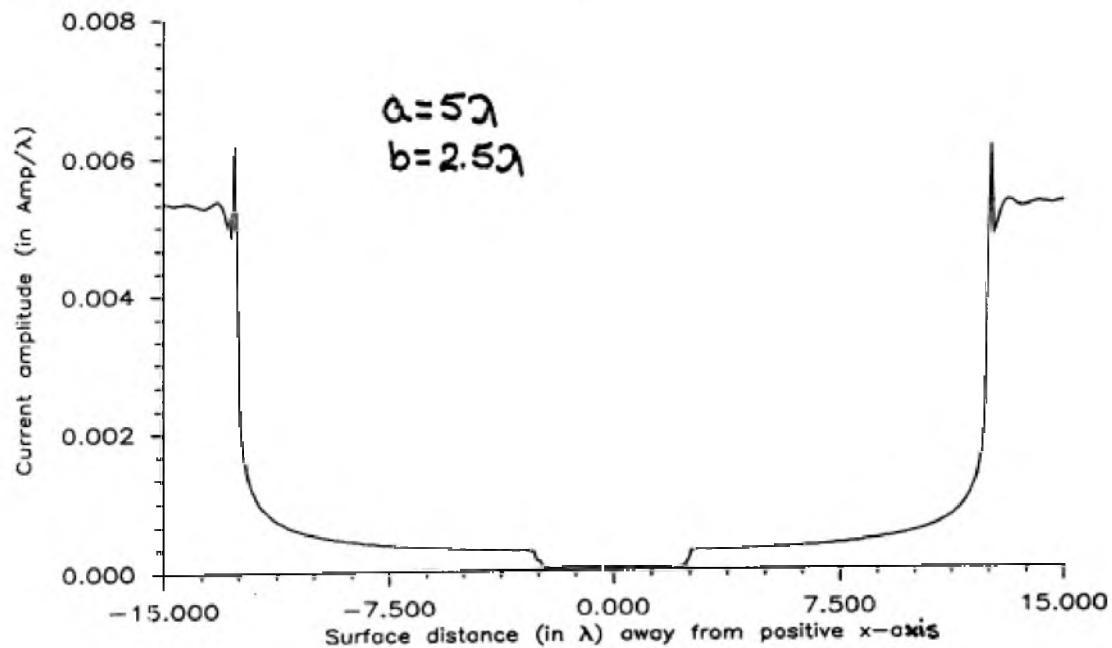


$\alpha = 2\lambda$



ANGLE VS RCS WHEN $\alpha = 2\lambda$

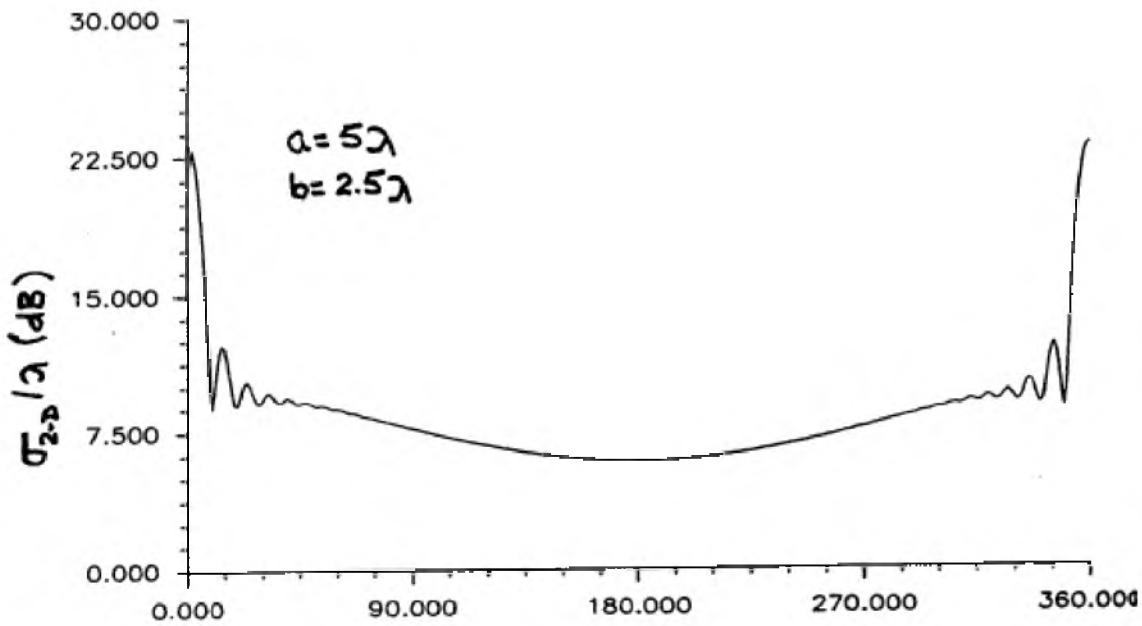
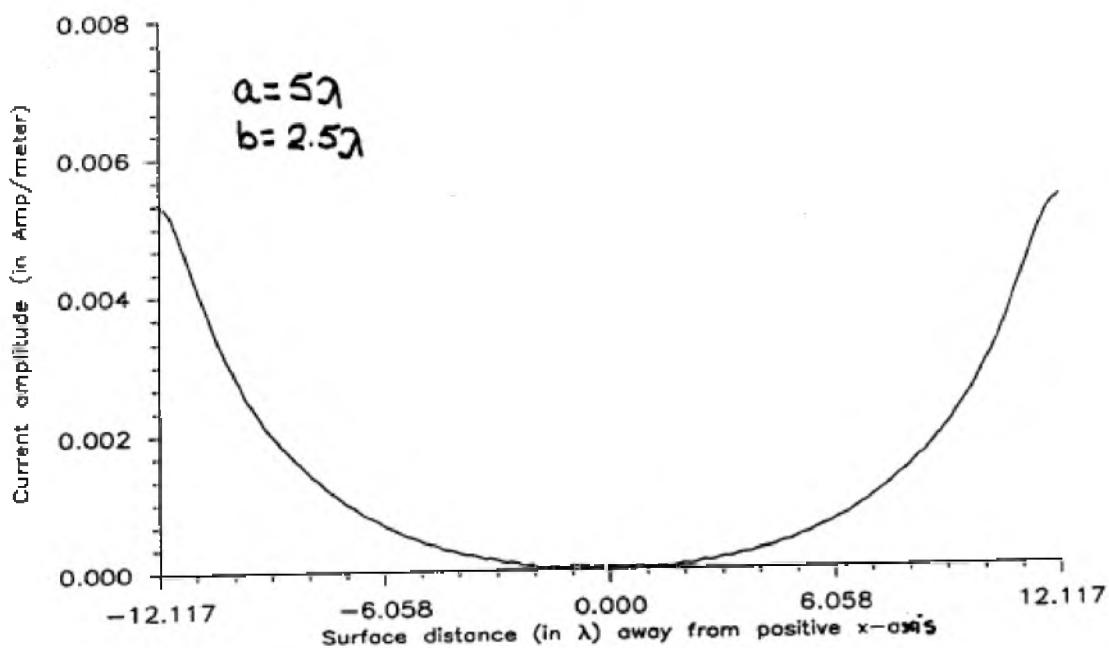
12.14



This problem can be solved using the TDRS computer program at the end of this chapter.

12.15

This problem can be solved using the TDRE program at the end of the chapter.



$$12.16 \quad \frac{d}{dx} H_0^{(2)}(\alpha x) = -\alpha H_1^{(2)}(\alpha x); \quad (1)$$

$$\frac{d}{dx} H_1^{(2)}(\alpha x) = -\alpha H_2^{(2)}(\alpha x) + \frac{1}{x} H_1^{(2)}(\alpha x)$$

Dividing both sides by x

$$\frac{1}{x} \frac{d}{dx} H_1^{(2)}(\alpha x) = -\frac{\alpha}{x} H_2^{(2)}(\alpha x) + \frac{1}{x^2} H_1^{(2)}(\alpha x)$$

or $\frac{1}{x} \frac{d}{dx} H_2^{(2)}(\alpha x) - \frac{1}{x^2} H_1^{(2)}(\alpha x) = -\frac{\alpha}{x} H_2^{(2)}(\alpha x)$

or

$$\underbrace{\frac{1}{\alpha x} \frac{d}{dx} H_1^{(2)}(\alpha x) - \frac{1}{\alpha x^2} H_1^{(2)}(\alpha x)}_{\frac{d}{dx} \left[\frac{1}{\alpha x} H_1^{(2)}(\alpha x) \right]} = -\frac{1}{x} H_2^{(2)}(\alpha x)$$

Thus

$$\frac{d}{dx} \left[\frac{1}{\alpha x} H_1^{(2)}(\alpha x) \right] = -\frac{1}{x} H_2^{(2)}(\alpha x) \quad (2)$$

$$\frac{\partial^2}{\partial x^2} H_0^{(2)}(\beta |x-x'|) = \frac{\partial^2}{\partial x^2} H_0^{(2)}(\beta R) = \frac{\partial}{\partial x} \left[\frac{\partial}{\partial x} H_0^{(2)}(\beta R) \right] \quad (3)$$

$$\frac{\partial}{\partial x} H_0^{(2)}(\beta R) = \frac{\partial R}{\partial x} \left[\frac{\partial}{\partial R} H_0^{(2)}(\beta R) \right] = \frac{\partial R}{\partial x} \left[-\beta H_1^{(2)}(\beta R) \right] \quad (4)$$

Since $R = [(x-x')^2 + (y-y')^2]^{1/2} = [(x-x')^2 + y'^2]^{1/2}$

then $\frac{\partial R}{\partial x} = \frac{1}{2} [(x-x')^2 + y'^2]^{-1/2} (x-x') = \frac{x-x'}{\sqrt{(x-x')^2 + y'^2}} = \frac{x-x'}{R} \quad (5)$

Thus we can write (4) as

$$\frac{\partial}{\partial x} H_0^{(2)}(\beta R) = \frac{\partial R}{\partial x} \left[-\beta H_1^{(2)}(\beta R) \right] = -\beta \frac{x-x'}{R} H_1^{(2)}(\beta R) \quad (6)$$

Using (6) we can write (3) as

$$\begin{aligned} \frac{\partial^2}{\partial x^2} H_0^{(2)}(\beta R) &= \frac{\partial}{\partial x} \left[-\beta \frac{x-x'}{R} H_1^{(2)}(\beta R) \right] = -\beta \frac{\partial}{\partial x} \left[\frac{x-x'}{R} H_1^{(2)}(\beta R) \right] = -\beta \frac{\partial}{\partial x} \left[(x-x') \frac{H_1^{(2)}(\beta R)}{R} \right] \\ &= -\beta \left\{ (x-x') \frac{\partial}{\partial x} \left[\frac{H_1^{(2)}(\beta R)}{R} \right] + \frac{H_1^{(2)}(\beta R)}{R} \right\} = -\beta \left\{ (x-x') \frac{\partial R}{\partial x} \frac{\partial}{\partial R} \left[\frac{H_1^{(2)}(\beta R)}{R} \right] + \frac{H_1^{(2)}(\beta R)}{R} \right\} \end{aligned}$$

Using (2) and (4) we can write it as

$$\frac{\partial^2}{\partial x^2} H_0^{(2)}(\beta R) = -\beta \left\{ (x-x') \left[-\frac{\beta}{R} H_2^{(2)}(\beta R) + \frac{H_1^{(2)}(\beta R)}{R} \right] \right\} = \beta^2 \frac{H_2^{(2)}(\beta R)}{R^2} (x-x')^2 - \beta \frac{H_1^{(2)}(\beta R)}{R} \quad (7)$$

Cont'd.

12.16 Cont'd. For the Hankel functions we can write (12-18) and (12-19) as

$$\frac{d}{dx} H_n^{(2)}(\alpha x) = \alpha H_{n+1}^{(2)}(\alpha x) - \frac{n}{x} H_n^{(2)}(\alpha x) \quad (8a)$$

$$\frac{d}{dx} H_n^{(2)}(\alpha x) = -\alpha H_{n-1}^{(2)}(\alpha x) + \frac{n}{x} H_n^{(2)}(\alpha x) \quad (8b)$$

Subtracting (8a) from (8b) reduces to

$$0 = -\alpha [H_{n+1}^{(2)}(\alpha x) + H_{n-1}^{(2)}(\alpha x)] + \frac{2n}{x} H_n^{(2)}(\alpha x)$$

or

$$\frac{1}{x} H_n^{(2)}(\alpha x) = \frac{\alpha}{2} [H_{n+1}^{(2)}(\alpha x) + H_{n-1}^{(2)}(\alpha x)] \quad (9)$$

which for $n=1$ reduces to

$$\frac{1}{x} H_1^{(2)}(\beta R) = \frac{\beta}{2} [H_2^{(2)}(\beta R) + H_0^{(2)}(\beta R)] \quad (9a)$$

Using (9a) we can write (7) as

$$\begin{aligned} \frac{\partial^2}{\partial x^2} H_0^{(2)}(\beta R) &= \beta^2 \frac{H_2^{(2)}(\beta R)}{R^2} (x-x')^2 - \beta \left\{ \frac{\beta}{2} [H_2^{(2)}(\beta R) + H_0^{(2)}(\beta R)] \right\} \\ \frac{\partial^2}{\partial x^2} H_0^{(2)}(\beta R) &= \frac{\beta^2}{2} \left\{ 2 \frac{(x-x')^2}{R^2} - 1 \right\} H_2^{(2)}(\beta R) - \frac{\beta^2}{2} H_0^{(2)}(\beta R) = \frac{\beta^2}{2} \underbrace{\left[2 \frac{(x-x')^2}{R^2} - 1 \right]}_{\cos(2\phi'')} H_2^{(2)}(\beta R) - H_0^{(2)}(\beta R) \\ \frac{\partial^2}{\partial x^2} H_0^{(2)}(\beta R) &= \frac{\beta^2}{2} \left\{ H_2^{(2)}(\beta R) \cos(2\phi'') - H_0^{(2)}(\beta R) \right\} \end{aligned} \quad (10)$$

Using (10) we can write (12-70a) as

$$\begin{aligned} E_x^s &= -\frac{\eta}{4\beta_0} \int_0^w J_x(x') \left\{ \frac{\beta^2}{2} [H_2^{(2)}(\beta R) \cos(2\phi'') - H_0^{(2)}(\beta R)] + \beta^2 H_0^{(2)}(\beta R) \right\} dx' \\ E_x^s &= -\frac{\eta\beta}{8} \int_0^w J_x(x') \left\{ H_2^{(2)}(\beta R) \cos(2\phi'') + H_0^{(2)}(\beta R) \right\} dx' \end{aligned} \quad (11)$$

Now let us look at (12-71b).

$$\begin{aligned} \frac{\partial^2 H_0^{(2)}}{\partial y^2 x} &= \frac{\partial}{\partial y} \left[\frac{\partial}{\partial x} H_0^{(2)}(\beta R) \right] = \frac{\partial}{\partial y} \left[-\beta \frac{(x-x')}{R} H_1^{(2)}(\beta R) \right] = -\beta \frac{\partial}{\partial y} \left[(x-x') \frac{H_1^{(2)}(\beta R)}{R} \right] \\ &= -\beta \left\{ (x-x') \frac{\partial}{\partial y} \left[\frac{H_1^{(2)}(\beta R)}{R} \right] + \frac{\partial}{\partial y} (x-x') \frac{H_1^{(2)}(\beta R)}{R} \right\} = -\beta (x-x') \frac{\partial}{\partial y} \frac{\partial}{\partial R} \left[\frac{H_1^{(2)}(\beta R)}{R} \right] \\ &= -\beta (x-x') \frac{\partial}{\partial R} \left[-\frac{\partial}{\partial R} H_2^{(2)}(\beta R) \right] = \beta^2 \frac{(x-x')}{R} \left(\frac{\partial}{\partial R} H_2^{(2)}(\beta R) \right) = \beta^2 \underbrace{\cos(2\phi'') \sin(\phi'')}_{\frac{1}{2} \sin(2\phi'')} H_2^{(2)}(\beta R) \\ \frac{\partial^2 H_0^{(2)}}{\partial y^2 x} &= \frac{\beta^2}{2} H_2^{(2)}(\beta R) \sin(2\phi'') \end{aligned} \quad (12)$$

Therefore we can write (12-70b) as

$$E_y^s = -\frac{\eta}{4\beta_0} \int_0^w J_x(x) \left\{ \frac{\beta^2}{2} H_2^{(2)}(\beta R) \sin(2\phi'') \right\} dx' = -\frac{\eta\beta}{8} \int_0^w J_x(x) H_2^{(2)}(\beta R) \sin(2\phi'') dx'$$

$$12.17 \quad Z_{mn} = \frac{\beta\eta}{\delta} \int_{x_n}^{x_{n+1}} [H_0^{(2)}(\beta R_{mn}) + H_2^{(2)}(\beta R_{mn}) \cos(2\phi_{mn}')] dx'$$

c. For $m=n$

$$\cos(2\phi_{nn}') = \frac{(x')^2 + t^2 - \frac{t^2}{4}}{(x')^2 + t^2} \approx 1$$

$$Z_{nn} \approx \frac{\beta\eta}{\delta} \int_{-\Delta x_n/2}^{\Delta x_n/2} [H_0^{(2)}(\beta R) + H_2^{(2)}(\beta R)] dx'$$

$$H_0^{(2)}(\beta R) \approx 1 - \frac{2}{\pi} \ln\left(\frac{\beta Y|x|}{2}\right) + O(x)^2 \Rightarrow H_0^{(2)}(\beta R) dx' \approx \int_{-\Delta x_n/2}^{\Delta x_n/2} \left[1 - \frac{2}{\pi} \ln\left(\frac{\beta Y|x|}{2}\right)\right] dx' = 4\left[\frac{1}{\pi} \ln\left(\frac{\beta Y|x|}{2}\right)\right]$$

$$Y=2.791 \quad -\Delta x_n/2 \quad -\Delta x_n/2 \quad \Delta x_n/2 \quad e=2.718$$

$$H_2^{(2)}(\beta R) \approx j \frac{1}{\pi} \left[\frac{4}{\beta^2(x'^2+t^2)} + 1 \right] + O(t)^2 \Rightarrow H_2^{(2)}(\beta R) dx' \approx \lim_{t \rightarrow 0} \int_{-\Delta x_n/2}^{\Delta x_n/2} \left[j \frac{1}{\pi} \left[\frac{4(x'^2+t^2)}{\beta^2(x'^2+t^2)} + 1 \right] \right] dx'$$

$$= j \frac{\Delta x_n}{\pi} \left[-\frac{26}{(\beta \Delta x_n)^2} + 1 \right]$$

$$Z_{nn} = \frac{\beta\eta \Delta x_n}{\delta} \left\{ 1 - j \frac{1}{\pi} \left[-1 + 2 \ln\left(\frac{\beta Y \Delta x_n}{4e}\right) + \frac{16}{(\beta \Delta x_n)^2} \right] \right\}, \quad m=n$$

b. $|m-n| \leq 2, \quad m \neq n \Rightarrow \cos(2\phi_{mn}') = 1$

$$Z_{mn} = \frac{\beta\eta}{\delta} \int_{x_n}^{x_{n+1}} [H_0^{(2)}(\beta R_{mn}) + H_2^{(2)}(\beta R_{mn})] dx' = \frac{\beta\eta}{\delta} \int_{x_n - \frac{\Delta x_n}{2}}^{x_{n+1} + \frac{\Delta x_n}{2}} [H_0^{(2)}(\beta|x'-x_m|) + H_2^{(2)}(\beta|x'-x_m|)] dx'$$

Using (III-18) and (III-19)

$$\frac{dH_1(\beta z)}{dz} = \beta H_0^{(2)}(\beta z) - \frac{2}{\pi} H_2^{(2)}(\beta z) \quad (1)$$

$$\frac{dH_2(\beta z)}{dz} = -\beta H_1(\beta z) + \frac{2}{\pi} H_0^{(2)}(\beta z) \quad (2)$$

Subtracting (1) and (2), we get

$$\beta [H_0^{(2)}(\beta z) + H_2^{(2)}(\beta z)] - \frac{2}{\pi} H_1^{(2)}(\beta z) \Rightarrow H_0^{(2)}(\beta z) + H_2^{(2)}(\beta z) = \frac{2}{\beta z} H_1^{(2)}(\beta z) \quad (3)$$

Thus

$$Z_{mn} = \frac{\beta\eta}{\delta} \int_{x_n - \frac{\Delta x_n}{2}}^{x_{n+1} + \frac{\Delta x_n}{2}} \frac{H_1(\beta|x'-x_m|)}{(\beta|x'-x_m|)} dx' \approx \frac{\beta\eta}{\delta} \int_{x_n - \frac{\Delta x_n}{2}}^{x_{n+1} + \frac{\Delta x_n}{2}} \left(\frac{\beta|x'-x_m|}{\pi} + j \frac{2}{\pi} \frac{1}{\beta|x'-x_m|} \right) dx'$$

$$Z_{mn} = \frac{\beta\eta}{\delta} \int_{x_n - \frac{\Delta x_n}{2}}^{x_{n+1} + \frac{\Delta x_n}{2}} \frac{1}{2} \left[1 + j \frac{4}{\pi} \frac{1}{\beta^2|x'-x_m|^2} \right] dx' \quad \text{Let } x' = x'' + x_n. \text{ Then}$$

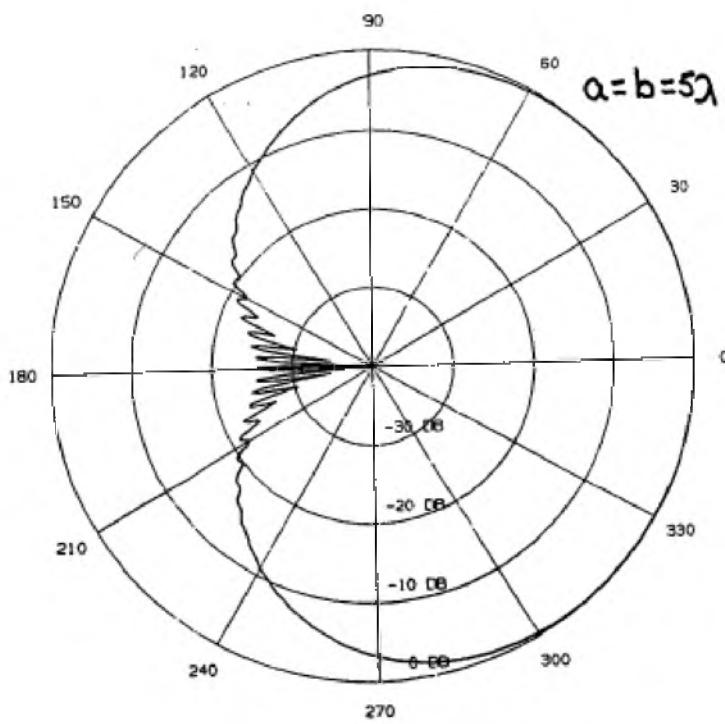
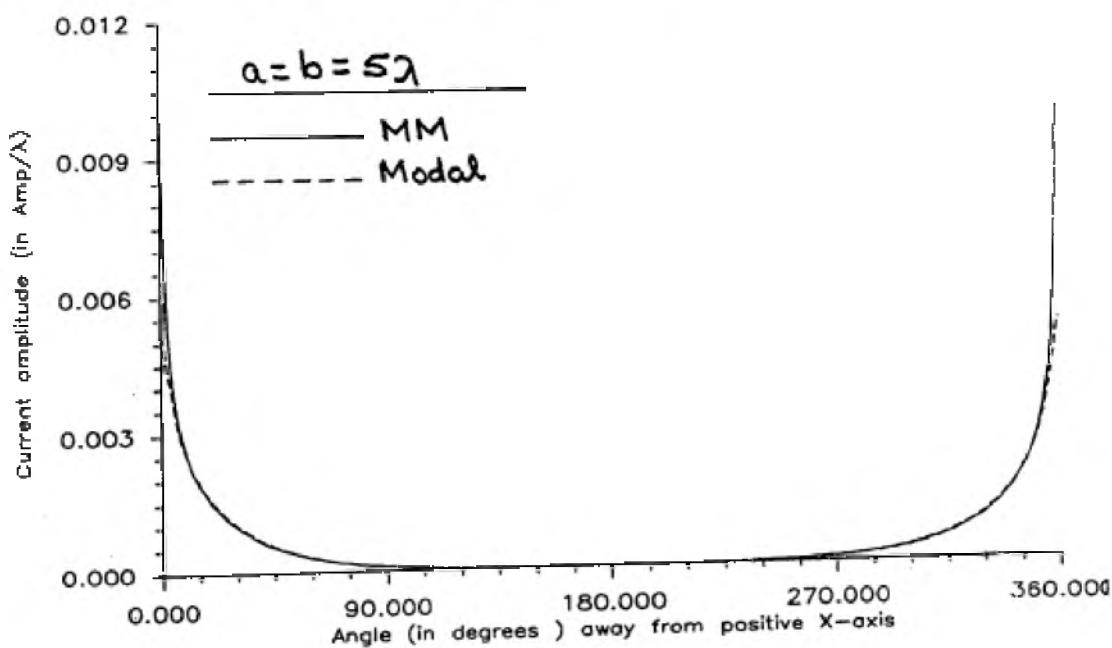
$$Z_{mn} = \frac{\beta\eta}{\delta} \int_{-\Delta x_n/2}^{\Delta x_n/2} \left[1 + j \frac{4}{\pi} \frac{1}{\beta^2(|x_m-x_n|+|x''|)^2} \right] dx' = \frac{\beta\eta \Delta x_n}{\delta} \left\{ 1 - j \frac{4}{\pi \beta^2 \Delta x_n} \left[\frac{1}{|x_m-x_n| + \frac{\Delta x_n}{2}} - \frac{1}{|x_m-x_n| - \frac{\Delta x_n}{2}} \right] \right\}$$

$$Z_{mn} = \frac{\beta\eta \Delta x_n}{\delta} \left\{ 1 + j \frac{4}{\pi \beta^2} \frac{1}{|x_m-x_n|^2 + (\Delta x_n)^2/4} \right\}, \quad |m-n| \leq 2, \quad m \neq n$$

c. $|m-n| > 2: \quad Z_{mn} = \frac{\beta\eta}{\delta} \int_{x_n - \Delta x_n/2}^{x_{n+1} + \Delta x_n/2} \frac{H_1(\beta|x'-x_m|)}{(\beta|x'-x_m|)} dx' = \frac{\beta\eta}{\delta} \int_{-\Delta x_n/2}^{\Delta x_n/2} \frac{H_1(\beta(|x_m-x_n|+x'|))}{\beta(|x_m-x_n|+x'|)} dx'$

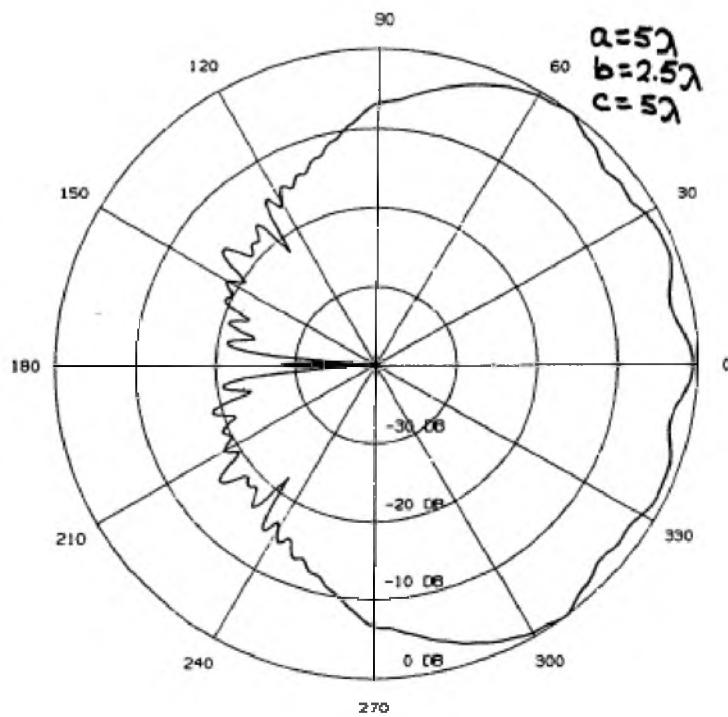
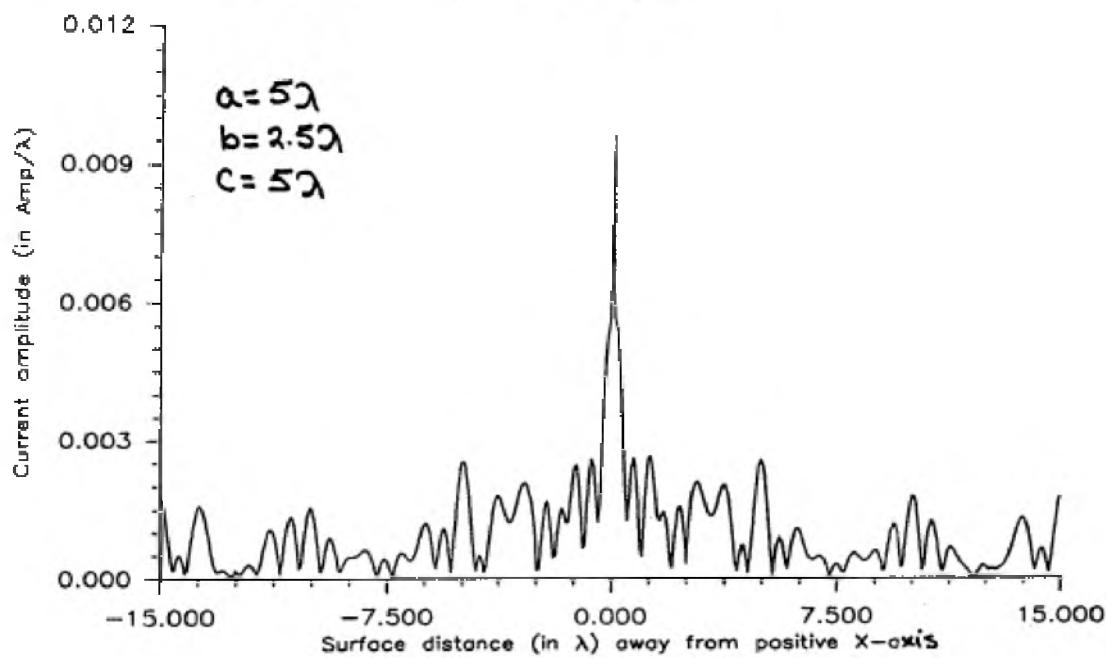
12.18

This problem can be solved using the TDRS computer program
at the end of this chapter.



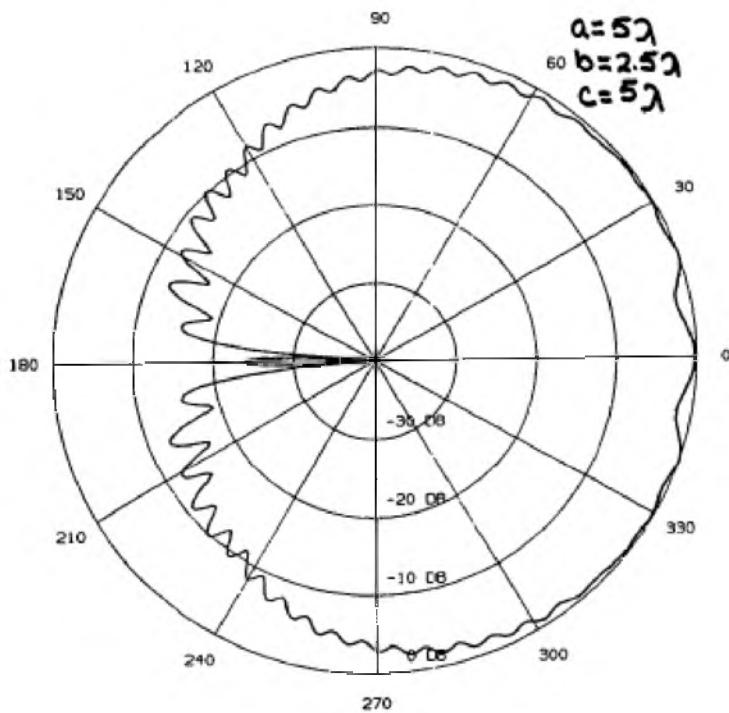
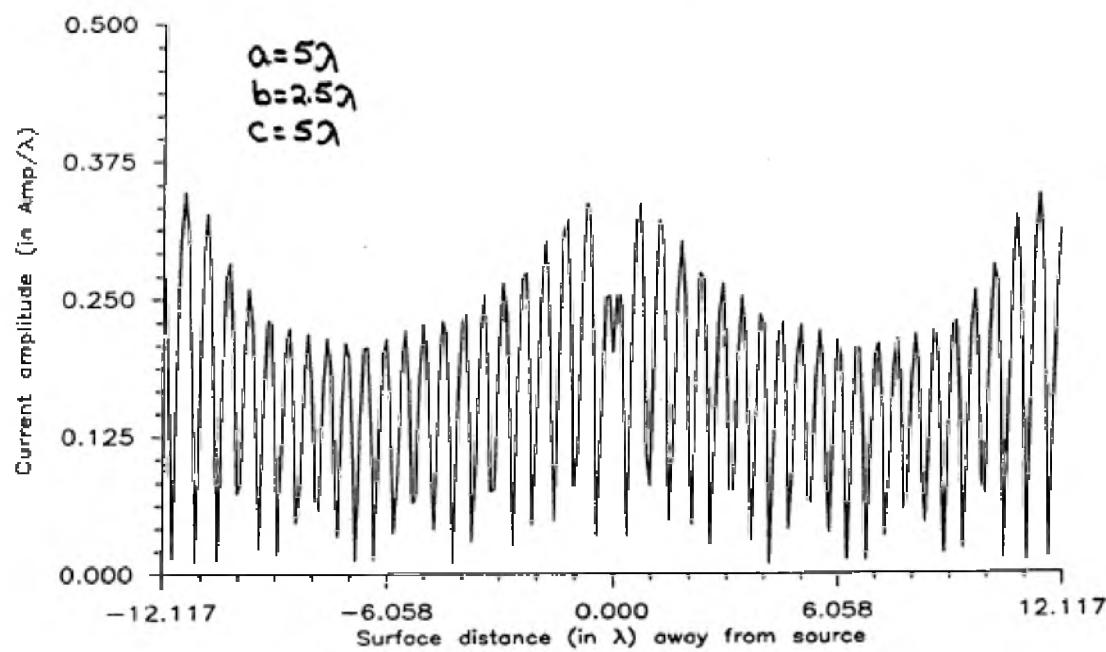
12.19

This problem can be solved using the TDRS computer program at the end of this chapter.



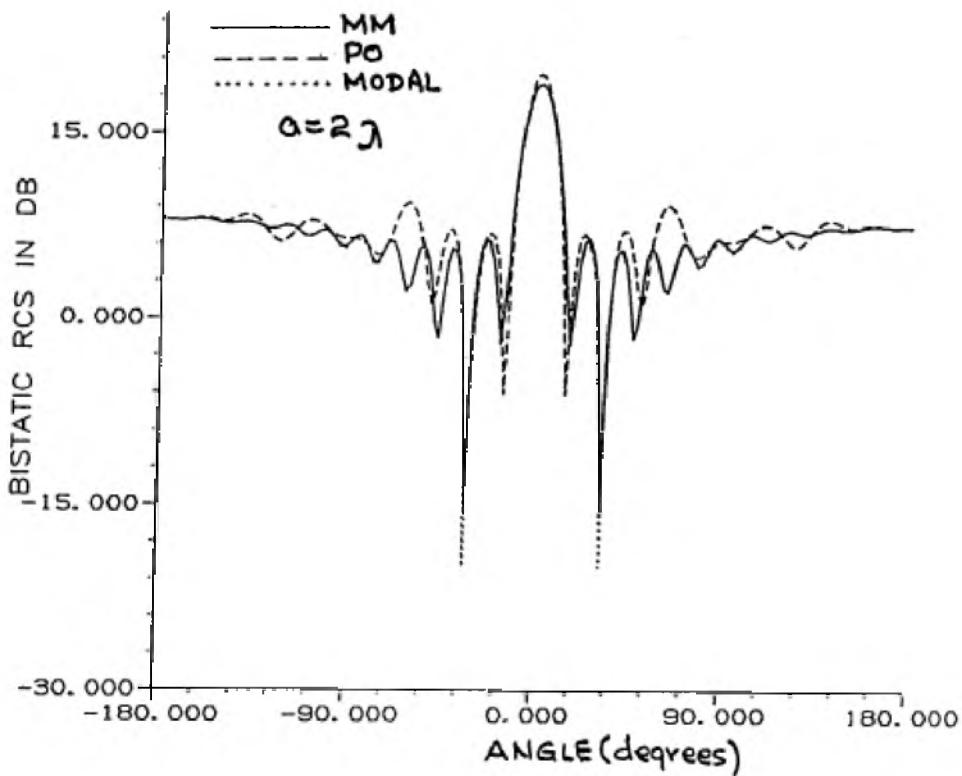
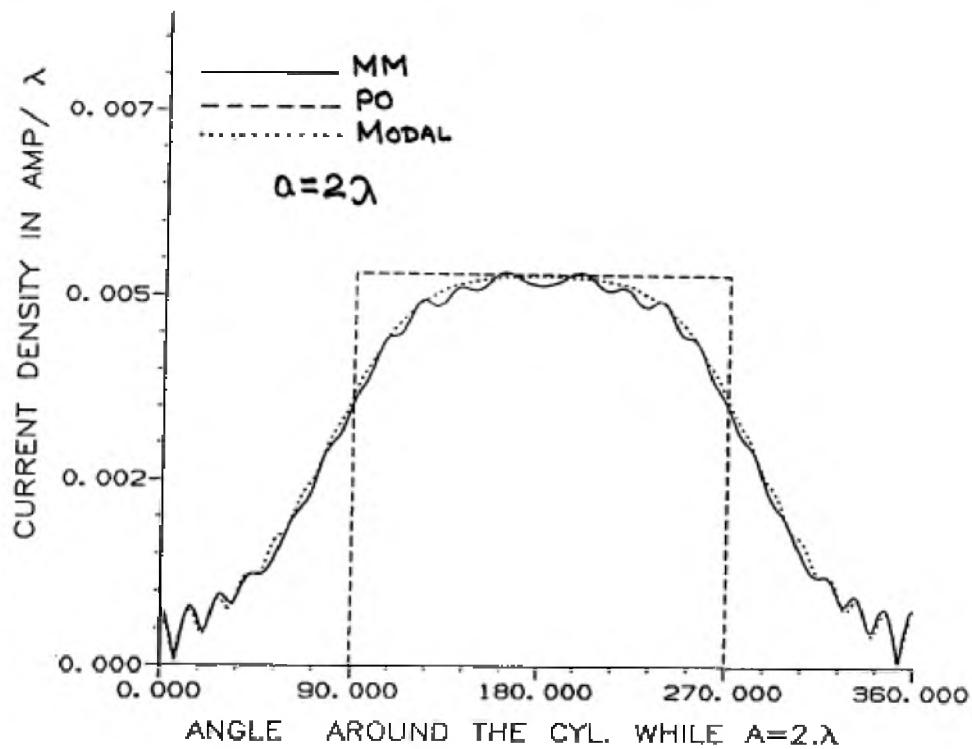
12.20

This problem can be solved using the TDRS computer program at the end of this chapter.



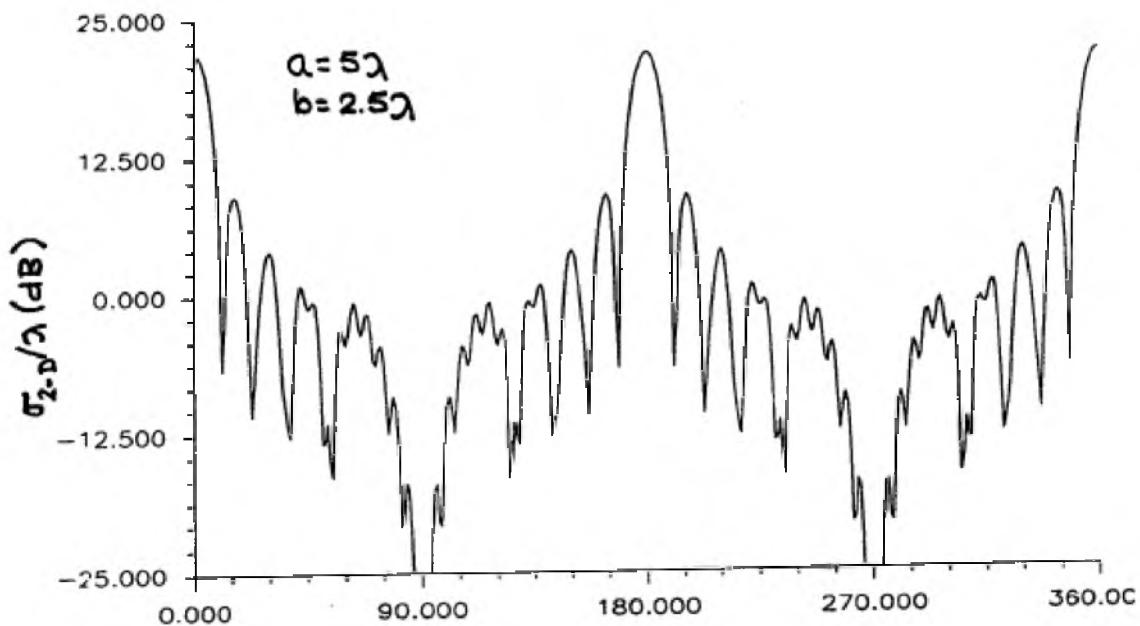
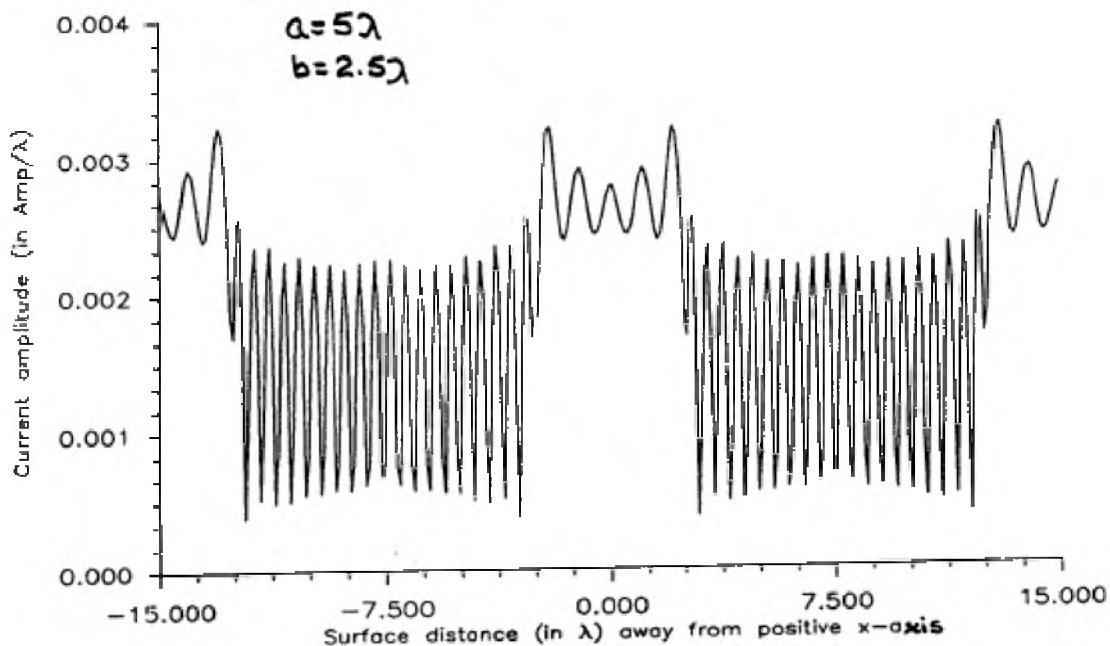
12.21

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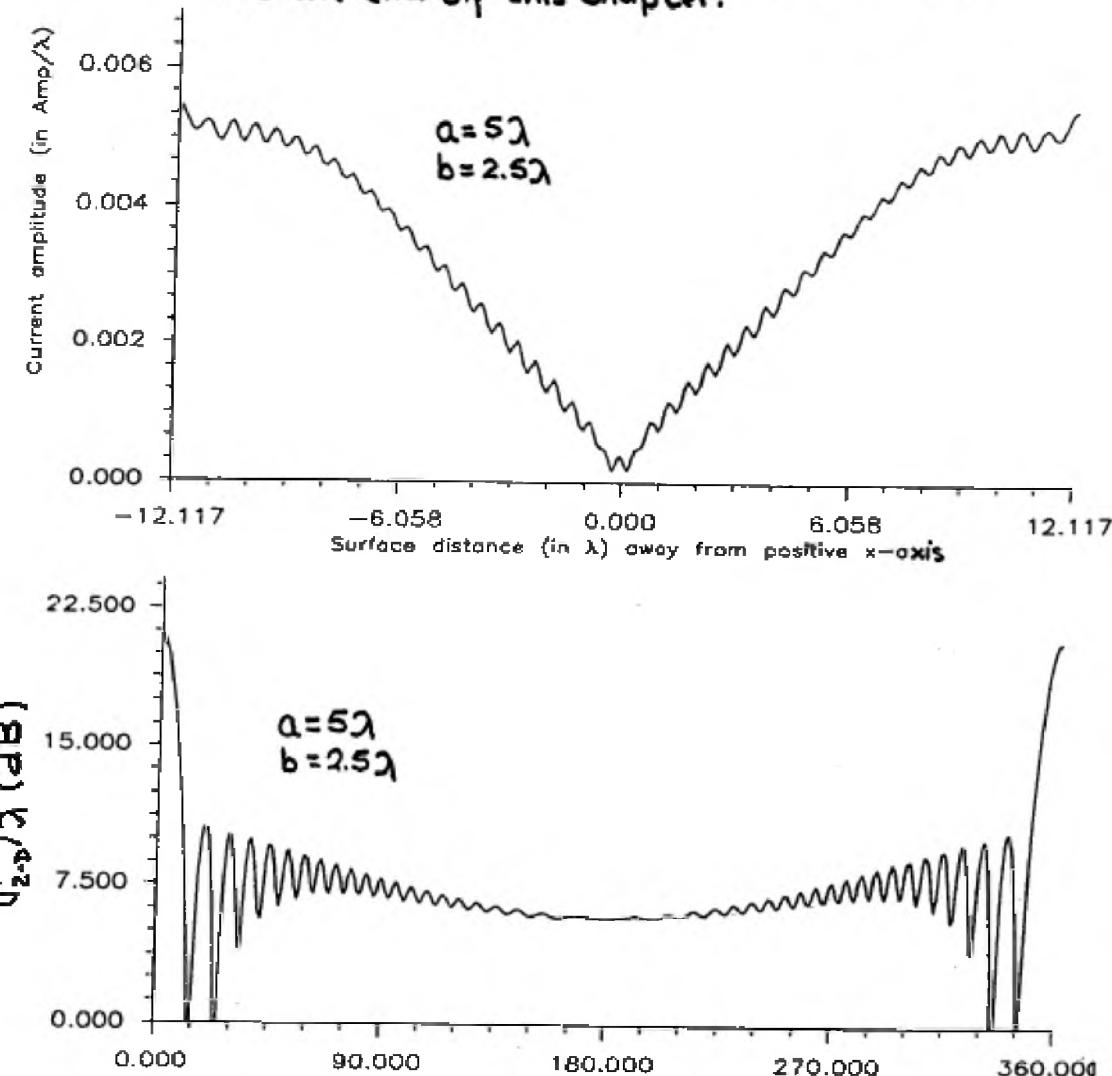
12.22

This problem can be solved using the TDRS computer program at the end of this chapter.



12.23

This problem can be solved using the TDRS computer program at the end of this chapter.

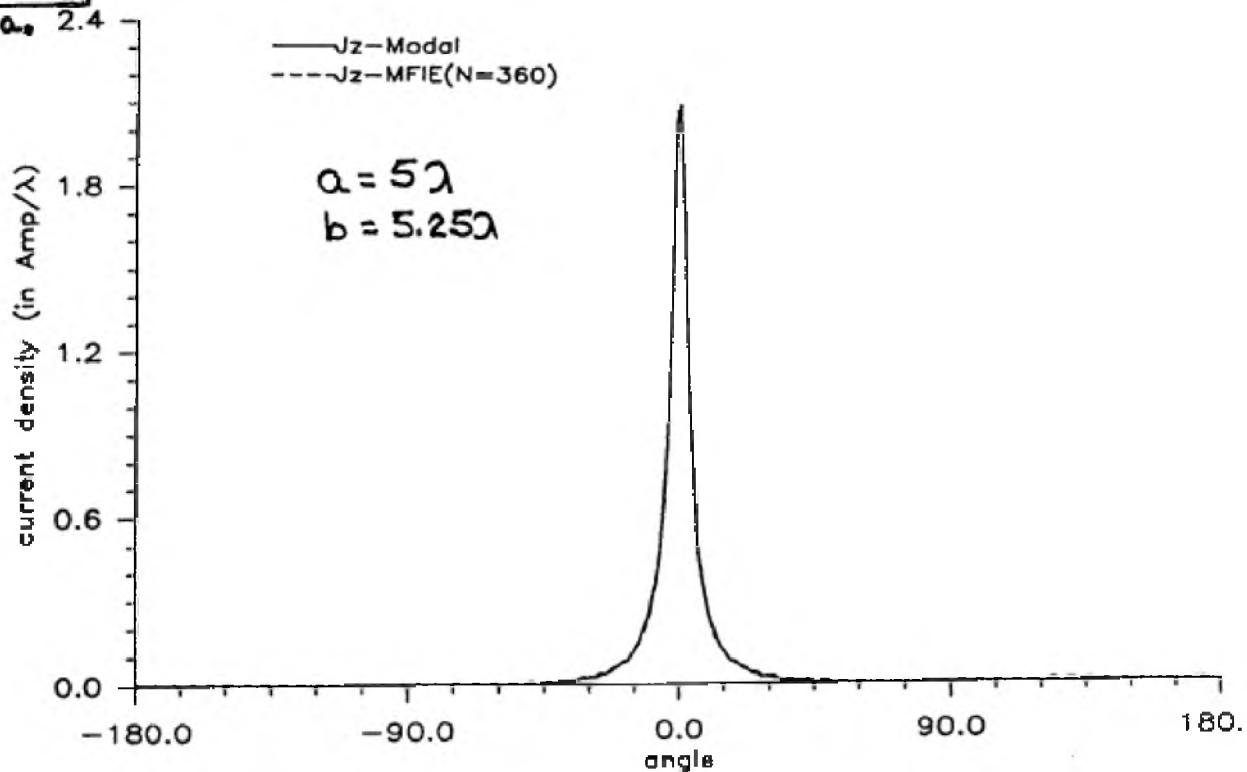


12.24

$$\begin{aligned}\hat{n} \cdot \nabla H_0^{(2)}(\beta R) &= \hat{n} \cdot \left\{ \hat{a}_R \frac{\partial}{\partial R} + \hat{a}_\phi \frac{1}{R} \frac{\partial}{\partial \phi} + \hat{a}_z \frac{\partial^2}{\partial z^2} \right\} H_0^{(2)}(\beta R) = (\hat{n} \cdot \hat{a}_R) \frac{\partial}{\partial R} H_0^{(2)}(\beta R) \\ &= (\hat{n} \cdot \hat{a}_R) \left\{ -\beta H_1^{(2)}(\beta R) \right\} = -\beta (\hat{n} \cdot \hat{a}_R) H_1^{(2)}(\beta R)\end{aligned}$$

$$\hat{n} \cdot \nabla H_0^{(2)}(\beta R) = -\beta \cos \psi H_1^{(2)}(\beta R)$$

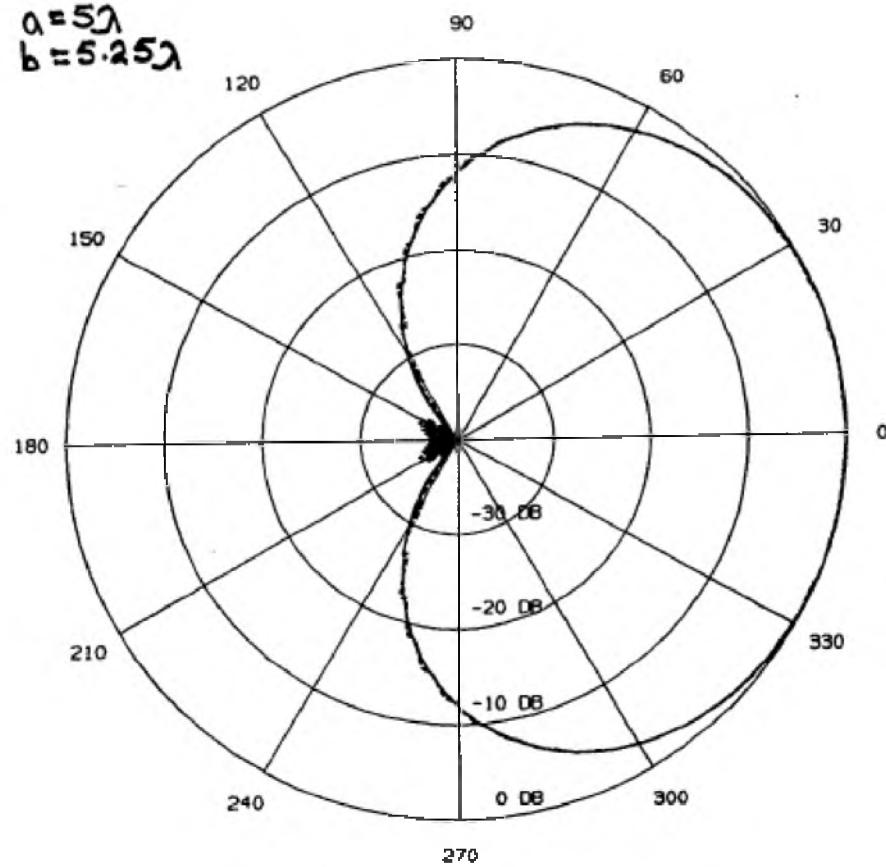
12.25

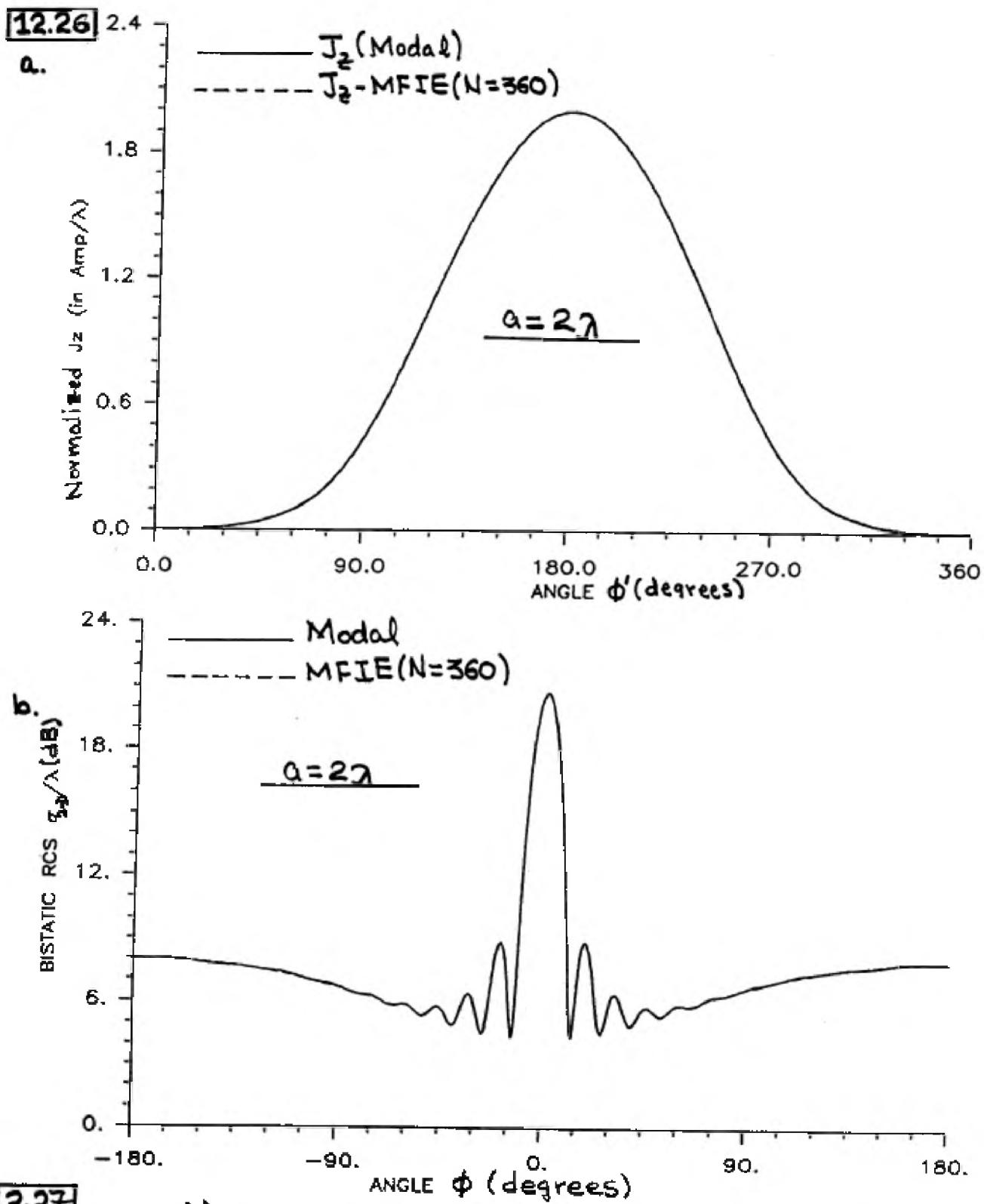


b.

----- E_z -MFIE(N=360)
 ——— E_z -Modal

$a = 5\lambda$
 $b = 5.25\lambda$





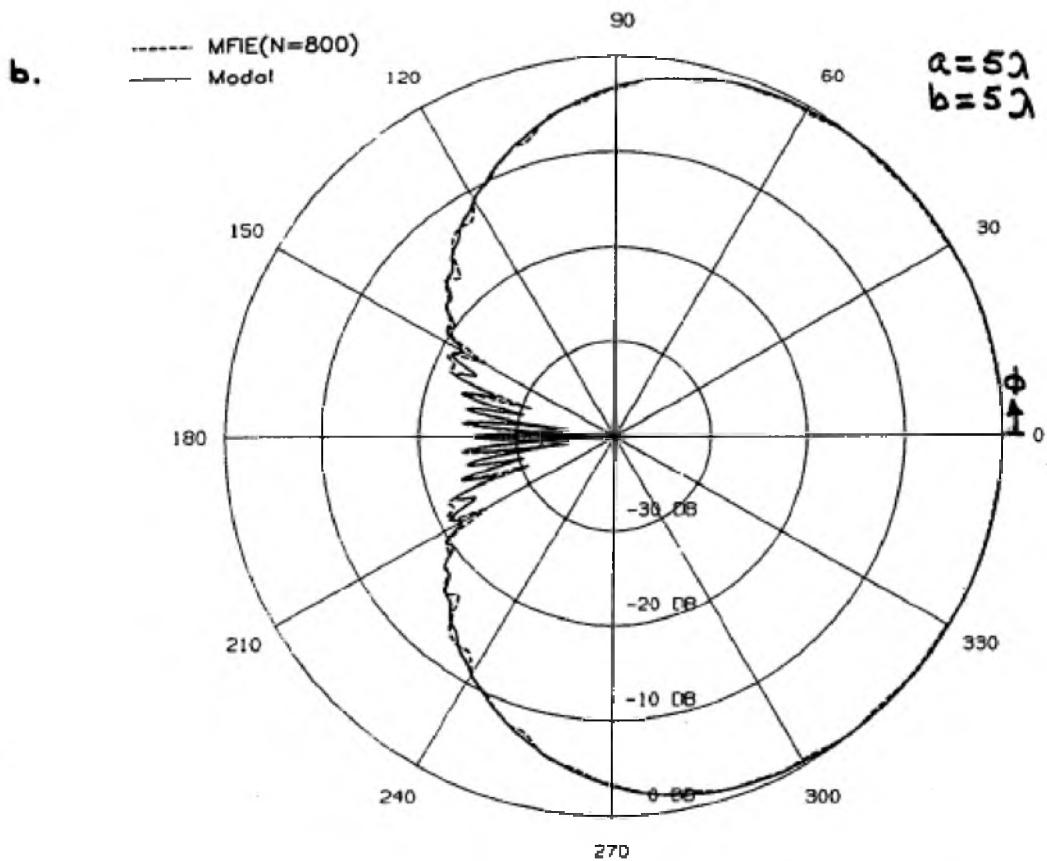
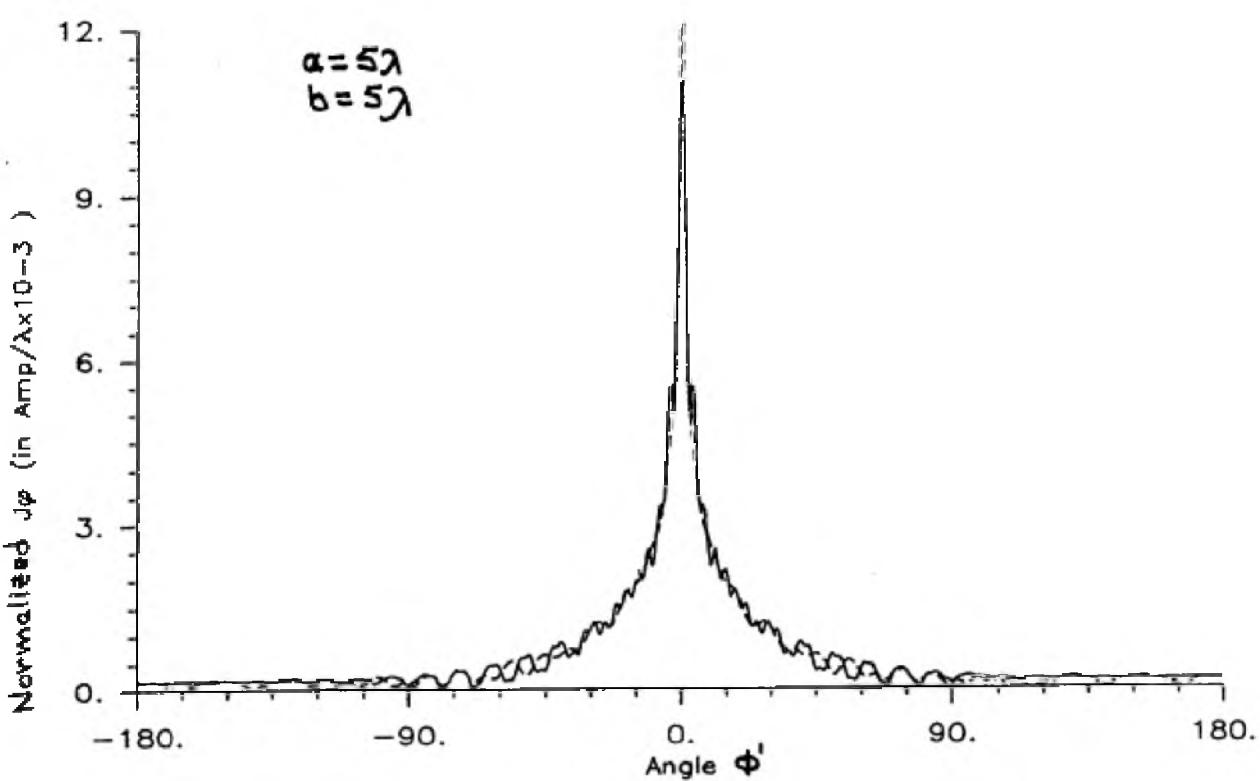
[12.27]

$$\hat{n}' \cdot \nabla H_0^{(1)}(\beta R) = \hat{n}' \cdot [\hat{a}_x \frac{\partial}{\partial R} + \hat{a}_\phi \frac{\partial}{\partial \phi} + \hat{a}_z \frac{\partial}{\partial z}] H_0^{(1)}(\beta R) = (\hat{n}' \cdot \hat{a}_x) \frac{\partial}{\partial R} H_0^{(1)}(\beta R)$$

$$\hat{n}' \cdot \nabla H_0^{(1)}(\beta R) = (\hat{n}' \cdot \hat{R}) [-\beta H_1^{(1)}(\beta R)] = -\beta \cos \varphi' H_1^{(1)}(\beta R)$$

12.28

— J_φ-Modal
 - - - J_φ-MFIE(N=800)



$$12.29 \quad \frac{1}{4\pi} \int_{-\ell/2}^{\ell/2} I(z') \left[\left(\frac{\partial^2}{\partial z'^2} + \beta^2 \right) \frac{e^{-j\beta R}}{R} \right] dz' = -j\omega \epsilon E_z^L (\rho=a), \quad R^2 = (z-z')^2 + a^2$$

$$\frac{\partial^2}{\partial z'^2} \left(\frac{e^{-j\beta R}}{R} \right) = -(1+j\beta R) \left(\frac{z-z'}{R^3} \right) e^{-j\beta R}$$

$$\begin{aligned} \frac{\partial^2}{\partial z'^2} \left(\frac{e^{-j\beta R}}{R} \right) &= -(1+j\beta R) \frac{e^{-j\beta R}}{R^3} - \left[j\beta \frac{(z-z')^2}{R^4} - (1+j\beta R)(z+j\beta R) \frac{(z-z')^2}{R^6} \right] e^{-j\beta R} \\ &= -(1+j\beta R) \frac{e^{-j\beta R}}{R^3} - \left[j\beta (R^2-a^2)R - (3+j4\beta R-(\beta R)^2)(R^2-a^2) \right] \frac{e^{-j\beta R}}{R^5} \\ &= -(1+j\beta R) \frac{e^{-j\beta R}}{R^3} - (R^2-a^2) \left[(\beta R)^2 - 3(1+j\beta R) \right] \frac{e^{-j\beta R}}{R^5} \\ &= \frac{e^{-j\beta R}}{R^5} \left[-R^2(1+j\beta R) - (R^2-a^2)(\beta R)^2 + 3(R^2-a^2)(1+j\beta R) \right] \end{aligned}$$

$$\frac{\partial^2}{\partial z'^2} \left(\frac{e^{-j\beta R}}{R} \right) = \frac{e^{-j\beta R}}{R^5} \left\{ (1+j\beta R)(2R^2-3a^2) - (R^2-a^2)(\beta R)^2 \right\}$$

Therefore

$$\begin{aligned} \frac{1}{4\pi} \int_{-\ell/2}^{\ell/2} I(z') \left[\left(\frac{\partial^2}{\partial z'^2} + \beta^2 \right) \frac{e^{-j\beta R}}{R} \right] dz' &= \frac{1}{4\pi} \int_{-\ell/2}^{\ell/2} I(z') \frac{e^{-j\beta R}}{R^5} \left\{ (1+j\beta R)(2R^2-3a^2) - k^2(kR)^2 + a^2(\beta R)^2 + \beta^2 R^4 \right\} dz' \\ &= \frac{1}{4\pi} \int_{-\ell/2}^{\ell/2} I(z') \frac{e^{-j\beta R}}{R^5} \left[(1+j\beta R)(2R^2-3a^2) + (\beta a R)^2 \right] dz' \end{aligned}$$

12.30 According to (12-121) we have that

$$\left(\frac{\partial^2}{\partial z'^2} + \beta^2 \right) \int_{-\ell/2}^{\ell/2} I_z(z') \frac{e^{-j\beta R}}{4\pi R} dz' = -j\omega \epsilon E_z^L (\rho=a) = -j\frac{\beta}{\eta} E_z^L \quad (1)$$

The above equation is an inhomogeneous ordinary harmonic differential equation where solution is obtained as the sum of complementary and particular solutions. Thus we can write the solution as

$$\int_{-\ell/2}^{\ell/2} I_z(z') \frac{e^{-j\beta R}}{4\pi R} dz' = -j \underbrace{\frac{\beta}{\eta} [B \cos(\beta z) + C \sin(\beta z)]}_{\text{Complementary}} - j \underbrace{\frac{1}{\eta} \int_0^z E_z^L \sin[\beta(z-z')] dz'}_{\text{Particular Solution}} \quad (2)$$

At $z=0$ the particular solution vanishes. Therefore (2) term reduces to

$$\begin{aligned} \int_{-\ell/2}^{\ell/2} I_z(z') \frac{e^{-j\beta R}}{4\pi R} dz' &= -j \frac{\beta}{\eta} [B \cos(\beta z) + C \sin(\beta z)] = -j \frac{1}{\eta} [\beta B \cos(\beta z) + \beta C \sin(\beta z)] \\ &= -j \sqrt{\frac{\epsilon}{\mu}} [B_1 \cos(\beta z) + C_1 \sin(\beta z)] \end{aligned}$$

1231

$$\underline{E}_f = \hat{a}_\phi \frac{\underline{V}_i}{2\pi^2 n(b/a)}, \underline{M}_f = -2\pi \times \underline{E}_f = -2\hat{a}_z \times \hat{a}_\phi, \underline{E}_p = -\hat{a}_\phi \frac{\underline{V}_i}{\pi^2 n(b/a)}$$

$$M_\phi = -\frac{\underline{V}_i}{\pi^2 n(b/a)} \quad (2)$$

The solution presented here is that reported in the following journal paper:
L. L. Tsai, "A numerical solution for the near and far fields of an annular ring of magnetic current", IEEE Transactions Antennas and Propagation, Vol. AP-20, No. 5, Sept. 1972, pp. 569-576.

The radiated electric and magnetic fields can be found using

$$\underline{E} = -j\omega \underline{A} - j \frac{1}{wpe} \nabla (\nabla \cdot \underline{A}) - \frac{1}{\epsilon} \nabla \times \underline{F} \quad (2a)$$

$$\underline{H} = -j\omega \underline{F} - j \frac{1}{wpe} \nabla (\nabla \cdot \underline{F}) + \frac{1}{\mu} \nabla \times \underline{A} \quad (2b)$$

where

$$\underline{E} = \frac{\underline{E}}{4\pi} \iint_S M_f \frac{e^{-j\beta R}}{R} dS' \quad (2c)$$

$$\underline{A} = \frac{\underline{A}}{4\pi} \iint_S J \frac{e^{-j\beta R}}{R} dS' \quad (2d)$$

Since $\underline{J} = 0$, then $\underline{A} = 0$ and (2a) and (2b) reduce to

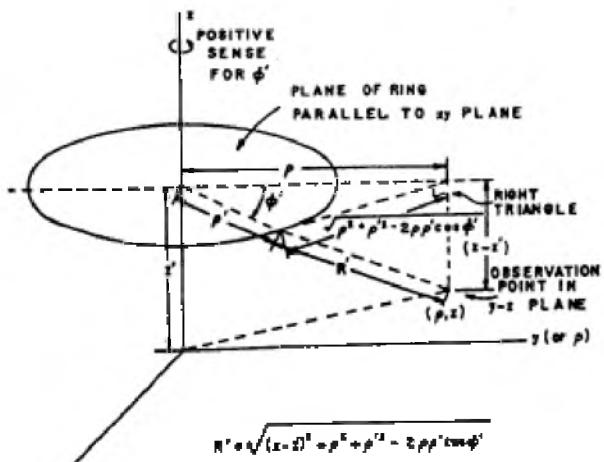
$$\underline{E} = -\frac{1}{\epsilon} \nabla \times \underline{F} \quad (3a)$$

$$\underline{H} = -j\omega \underline{F} - j \frac{1}{wpe} \nabla (\nabla \cdot \underline{F}) \quad (3b)$$

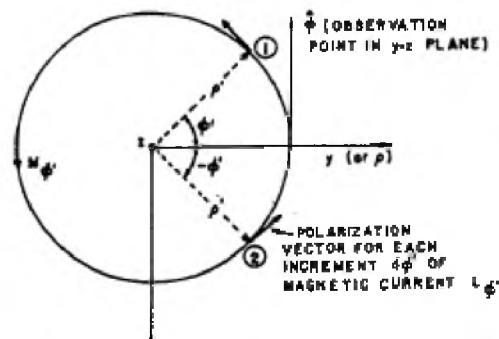
To evaluate (2c) the differential area dS' on the frill is

$$dS' = p' dp' d\phi' \quad (4)$$

Dividing the frill into dp' wide magnetic ring current, the contribution of each ring will be evaluated first. This is accomplished by integrating with respect to ϕ' before summing over all the rings.



Geometry for ring.



Polarization for source and observation.
cont'd.

Cont'd

12.31 cont'd. The problem possesses symmetry since there are no ϕ variations. Therefore the observation can be located at any plane. Without loss of generality we choose the yz plane for the observations. For this plane $\rho = y$ and the E potential is ϕ -polarized (F_ϕ). Using (1) and (2c) we can then write that

$$F_\phi = -\frac{\epsilon}{2\pi} \frac{1}{4n(b/a)} \int_0^b \int_0^\pi \cos\phi' \frac{e^{-j\beta R'}}{R'} d\phi' dp' \quad (5)$$

where $R' = [(z-z')^2 + p'^2 - (p')^2 - 2pp'\cos\phi']^{1/2}$ (5a)

Because the ring is on the xy -plane, each pair of incremental current element $d\phi'$ (symmetrically located at $\pm\phi'$) has only a net \hat{a}_ϕ contribution. This leads to the $2\cos\phi'$ factor with an integration in ϕ' from 0 to π . Now using (2a) and (2b) in cylindrical coordinates, we can write that

$$E_z = -\frac{1}{2} \frac{1}{p} \frac{\partial}{\partial p} (p F_\phi) \quad (6a)$$

$$E_p = \frac{1}{2} \frac{\partial}{\partial z} (F_\phi) \quad (6b)$$

$$H_\phi = -j\omega F_\phi \quad (6c)$$

$$E_\phi = H_p = H_z = 0 \quad (6d)$$

Closed form expressions for the electric and magnetic fields can be derived when $p \gg b$ without making far field approximations. From (5a)

$$R_0 = [(z-z')^2 + p^2]^{1/2}$$

After suitable approximations on (5) we can write that

$$F_\phi(p, z) \approx -\frac{\epsilon}{4\ln(b/a)} pp \frac{e^{-j\beta R}}{R_0^2} (b^2 - a^2) \left\{ -j \frac{(b^2 + a^2)}{4R_0^2} + \frac{1}{2\beta R_0} + j \frac{1}{2} \right\} \quad (7)$$

Using (7), we can write (6a) as

$$E_z \approx +V_i \frac{\epsilon (b^2 - a^2)}{8\ln(b/a)} \frac{e^{-j\beta R_0}}{R_0^2} \left\{ 2 \left[\frac{1}{\beta R_0} + j - j \frac{(b^2 + a^2)}{2R_0^2} \right] + \frac{\beta^2}{R_0} \left[\left(\frac{1}{\beta R_0} + j - j \frac{(b^2 + a^2)}{2R_0^2} \right) \left(-j \frac{2}{\beta R_0} \right) \right. \right. \\ \left. \left. + \left(-\frac{1}{\beta R_0^2} + j \frac{(b^2 + a^2)}{R_0^3} \right) \right] \right\}$$

$$E_p \approx -V_i \frac{(b^2 - a^2)p}{8\ln(b/a)} \frac{(z-z')}{R_0} \frac{e^{-j\beta R_0}}{R_0^2} \left\{ \beta - \left[\frac{2}{\beta} + \frac{\beta(b^2 + a^2)}{2} \right] \frac{1}{R_0^2} + j \left[\frac{2(b^2 + a^2)}{R_0^3} - \frac{2}{R_0} \right] \right\} \quad (8)$$

12.32 On the axis of the ring ($\rho=0$), the symmetry of the problem along the ϕ direction requires that $E_\phi(\rho=0, z) = 0$. However equations (8) and (9) of the solution of Problem 12.31 are no longer accurate when $\rho \rightarrow 0$. Therefore we will derive a simpler form for the field which will be accurate only for $\rho=0$. When $\rho=0$, we have that

$$R' = \left[(z - z')^2 + (\rho')^2 \right]^{1/2} \quad (1)$$

Therefore

$$F_\phi(0, z) = -\frac{\epsilon}{4\pi} \frac{1}{2n(b/a)} \int_a^b \frac{e^{-j\beta R'}}{R'} \left[\int_0^{2\pi} \cos \phi' d\phi' \right] d\rho' \equiv 0 \quad (2)$$

Then

$$H_\phi(0, z) = -j\omega F_\phi = 0 \quad (3)$$

From (6a) of the solution of Problem 12.32

$$E_z(0, z) = -\frac{1}{\epsilon} \frac{F_\phi}{\rho} - \frac{1}{\epsilon} \frac{\partial F_\phi}{\partial \rho} = -\frac{2}{\epsilon} \frac{\partial F_\phi}{\partial \rho} \Big|_{\rho=0} \quad (4)$$

In deriving (4), l'Hopital's rule was used to take care of the $0/0$ term.

Using (2) and (4) and interchanging the order of integration and differentiation, it can be shown that (4) can be written as

$$E_z(0, z) = \frac{1}{\pi} \frac{1}{2n(b/a)} \int_0^b \int_0^{\pi} \cos \phi' \left[\frac{\partial}{\partial \rho} \frac{e^{-j\beta R'}}{R'} \right]_{\rho=0} d\phi' d\rho'$$

$$E_z(0, z) = \frac{1}{2 \cdot 2n(b/a)} \left\{ \frac{e^{-j\beta [(z-z')^2 + a^2]^{1/2}}}{[(z-z')^2 + a^2]^{1/2}} - \frac{e^{-j\beta [(z-z')^2 + b^2]^{1/2}}}{[(z-z')^2 + b^2]^{1/2}} \right\} \quad (5)$$

The above solution is that found in :

1. L. L. Tsai, "A numerical solution for the near and far fields of an annular ring of magnetic current," IEEE Transactions Antennas and Propagation, Vol. AP-20, No. 5, Sept. 1972, pp. 569-576.

$Z(1, 1) = -0.96043E+03 + J - 0.29443E+00$	$Z(16, 3) = 0.21923E+00 + J - 0.20708E+00$
$Z(2, 1) = 0.41849E+03 + J - 0.29383E+00$	$Z(17, 3) = 0.16554E+00 + J - 0.19502E+00$
$Z(3, 1) = 0.44756E+02 + J - 0.29209E+00$	$Z(18, 3) = 0.12239E+00 + J - 0.18262E+00$
$Z(4, 1) = 0.13232E+02 + J - 0.28921E+00$	$Z(19, 3) = 0.87235E-01 + J - 0.16995E+00$
$Z(5, 1) = 0.57993E+01 + J - 0.28520E+00$	$Z(20, 3) = 0.58352E-01 + J - 0.15711E+00$
$Z(6, 1) = 0.31310E+01 + J - 0.28012E+00$	$Z(21, 3) = 0.34510E-01 + J - 0.14419E+00$
$Z(7, 1) = 0.19197E+01 + J - 0.27398E+00$	$Z(1, 4) = 0.13232E+02 + J - 0.28921E+00$
$Z(8, 1) = 0.12790E+01 + J - 0.26686E+00$	$Z(2, 4) = 0.44756E+02 + J - 0.29209E+00$
$Z(9, 1) = 0.98103E+00 + J - 0.25880E+00$	$Z(3, 4) = 0.41849E+03 + J - 0.29383E+00$
$Z(10, 1) = 0.65908E+00 + J - 0.24988E+00$	$Z(4, 4) = -0.96043E+03 + J - 0.29443E+00$
$Z(11, 1) = 0.49402E+00 + J - 0.24016E+00$	$Z(5, 4) = 0.41849E+03 + J - 0.29383E+00$
$Z(12, 1) = 0.37563E+00 + J - 0.22974E+00$	$Z(6, 4) = 0.44756E+02 + J - 0.29209E+00$
$Z(13, 1) = 0.28728E+00 + J - 0.21868E+00$	$Z(7, 4) = 0.13232E+02 + J - 0.28921E+00$
$Z(14, 1) = 0.21923E+00 + J - 0.20708E+00$	$Z(8, 4) = 0.57993E+01 + J - 0.28520E+00$
$Z(15, 1) = 0.16554E+00 + J - 0.19502E+00$	$Z(9, 4) = 0.31310E+01 + J - 0.28012E+00$
$Z(16, 1) = 0.12239E+00 + J - 0.18262E+00$	$Z(10, 4) = 0.19197E+01 + J - 0.27398E+00$
$Z(17, 1) = 0.87235E-01 + J - 0.16995E+00$	$Z(11, 4) = 0.12790E+01 + J - 0.26686E+00$
$Z(18, 1) = 0.58352E-01 + J - 0.15711E+00$	$Z(12, 4) = 0.98103E+00 + J - 0.25880E+00$
$Z(19, 1) = 0.34510E-01 + J - 0.14419E+00$	$Z(13, 4) = 0.65908E+00 + J - 0.24988E+00$
$Z(20, 1) = 0.14817E-01 + J - 0.13130E+00$	$Z(14, 4) = 0.49402E+00 + J - 0.24016E+00$
$Z(21, 1) = -0.13950E-02 + J - 0.11852E+00$	$Z(15, 4) = 0.37563E+00 + J - 0.22974E+00$
$Z(1, 2) = 0.41849E+03 + J - 0.29383E+00$	$Z(16, 4) = 0.28728E+00 + J - 0.21868E+00$
$Z(2, 2) = -0.96043E+03 + J - 0.29443E+00$	$Z(17, 4) = 0.21923E+00 + J - 0.20708E+00$
$Z(3, 2) = 0.41849E+03 + J - 0.29383E+00$	$Z(18, 4) = 0.16554E+00 + J - 0.19502E+00$
$Z(4, 2) = 0.44756E+02 + J - 0.29209E+00$	$Z(19, 4) = 0.12239E+00 + J - 0.18262E+00$
$Z(5, 2) = 0.13232E+02 + J - 0.28921E+00$	$Z(20, 4) = 0.87235E-01 + J - 0.16995E+00$
$Z(6, 2) = 0.57993E+01 + J - 0.28520E+00$	$Z(21, 4) = 0.58352E-01 + J - 0.15711E+00$
$Z(7, 2) = 0.31310E+01 + J - 0.28012E+00$	$Z(1, 5) = 0.57993E+01 + J - 0.28520E+00$
$Z(8, 2) = 0.19197E+01 + J - 0.27398E+00$	$Z(2, 5) = 0.13232E+02 + J - 0.28921E+00$
$Z(9, 2) = 0.12790E+01 + J - 0.26686E+00$	$Z(3, 5) = 0.44756E+02 + J - 0.29209E+00$
$Z(10, 2) = 0.98103E+00 + J - 0.25880E+00$	$Z(4, 5) = 0.41849E+03 + J - 0.29383E+00$
$Z(11, 2) = 0.65908E+00 + J - 0.24988E+00$	$Z(5, 5) = -0.96043E+03 + J - 0.29443E+00$
$Z(12, 2) = 0.49402E+00 + J - 0.24016E+00$	$Z(6, 5) = 0.41849E+03 + J - 0.29383E+00$
$Z(13, 2) = 0.37563E+00 + J - 0.22974E+00$	$Z(7, 5) = 0.44756E+02 + J - 0.29209E+00$
$Z(14, 2) = 0.28728E+00 + J - 0.21868E+00$	$Z(8, 5) = 0.13232E+02 + J - 0.28921E+00$
$Z(15, 2) = 0.21923E+00 + J - 0.20708E+00$	$Z(9, 5) = 0.57993E+01 + J - 0.28520E+00$
$Z(16, 2) = 0.16554E+00 + J - 0.19502E+00$	$Z(10, 5) = 0.31310E+01 + J - 0.28012E+00$
$Z(17, 2) = 0.12239E+00 + J - 0.18262E+00$	$Z(11, 5) = 0.19197E+01 + J - 0.27398E+00$
$Z(18, 2) = 0.87235E-01 + J - 0.16995E+00$	$Z(12, 5) = 0.12790E+01 + J - 0.26686E+00$
$Z(19, 2) = 0.58352E-01 + J - 0.15711E+00$	$Z(13, 5) = 0.98103E+00 + J - 0.25880E+00$
$Z(20, 2) = 0.34510E-01 + J - 0.14419E+00$	$Z(14, 5) = 0.65908E+00 + J - 0.24988E+00$
$Z(21, 2) = 0.14817E-01 + J - 0.13130E+00$	$Z(15, 5) = 0.49402E+00 + J - 0.24016E+00$
$Z(1, 3) = 0.44756E+02 + J - 0.29209E+00$	$Z(16, 5) = 0.37563E+00 + J - 0.22974E+00$
$Z(2, 3) = 0.41849E+03 + J - 0.29383E+00$	$Z(17, 5) = 0.28728E+00 + J - 0.21868E+00$
$Z(3, 3) = -0.96043E+03 + J - 0.29443E+00$	$Z(18, 5) = 0.21923E+00 + J - 0.20708E+00$
$Z(4, 3) = 0.41849E+03 + J - 0.29383E+00$	$Z(19, 5) = 0.16554E+00 + J - 0.19502E+00$
$Z(5, 3) = 0.44756E+02 + J - 0.29209E+00$	$Z(20, 5) = 0.12239E+00 + J - 0.18262E+00$
$Z(6, 3) = 0.13232E+02 + J - 0.28921E+00$	$Z(21, 5) = 0.87235E-01 + J - 0.16995E+00$
$Z(7, 3) = 0.57993E+01 + J - 0.28520E+00$	$Z(1, 6) = 0.31310E+01 + J - 0.28012E+00$
$Z(8, 3) = 0.31310E+01 + J - 0.28012E+00$	$Z(2, 6) = 0.57993E+01 + J - 0.28520E+00$
$Z(9, 3) = 0.19197E+01 + J - 0.27398E+00$	$Z(3, 6) = 0.13232E+02 + J - 0.28921E+00$
$Z(10, 3) = 0.12790E+01 + J - 0.26686E+00$	$Z(4, 6) = 0.44756E+02 + J - 0.29209E+00$
$Z(11, 3) = 0.98103E+00 + J - 0.25880E+00$	$Z(5, 6) = 0.41849E+03 + J - 0.29383E+00$
$Z(12, 3) = 0.65908E+00 + J - 0.24988E+00$	$Z(6, 6) = -0.96043E+03 + J - 0.29443E+00$
$Z(13, 3) = 0.49402E+00 + J - 0.24016E+00$	$Z(7, 6) = 0.41849E+03 + J - 0.29383E+00$
$Z(14, 3) = 0.37563E+00 + J - 0.22974E+00$	$Z(8, 6) = 0.44756E+02 + J - 0.29209E+00$
$Z(15, 3) = 0.28728E+00 + J - 0.21868E+00$	$Z(9, 6) = 0.13232E+02 + J - 0.28921E+00$

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$Z(14, 6) = 0.98103E+00+j-0.25880E+00$
 $Z(15, 6) = 0.65908E+00+j-0.24988E+00$
 $Z(16, 6) = 0.49402E+00+j-0.24016E+00$
 $Z(17, 6) = 0.37563E+00+j-0.22974E+00$
 $Z(18, 6) = 0.28728E+00+j-0.21868E+00$
 $Z(19, 6) = 0.21923E+00+j-0.20708E+00$
 $Z(20, 6) = 0.16554E+00+j-0.19502E+00$
 $Z(21, 6) = 0.12238E+00+j-0.18262E+00$
 $Z(1, 7) = 0.19197E+01+j-0.27398E+00$
 $Z(2, 7) = 0.31310E+01+j-0.28012E+00$
 $Z(3, 7) = 0.57993E+01+j-0.28520E+00$
 $Z(4, 7) = 0.13232E+02+j-0.28921E+00$
 $Z(5, 7) = 0.44756E+02+j-0.29209E+00$
 $Z(6, 7) = 0.41849E+03+j-0.29383E+00$
 $Z(7, 7) = -0.96043E+03+j-0.29443E+00$
 $Z(8, 7) = 0.41849E+03+j-0.29383E+00$
 $Z(9, 7) = 0.44756E+02+j-0.29209E+00$
 $Z(10, 7) = 0.13232E+02+j-0.28921E+00$
 $Z(11, 7) = 0.57993E+01+j-0.28520E+00$
 $Z(12, 7) = 0.31310E+01+j-0.28012E+00$
 $Z(13, 7) = 0.19197E+01+j-0.27398E+00$
 $Z(14, 7) = 0.12790E+01+j-0.26686E+00$
 $Z(15, 7) = 0.98103E+00+j-0.25880E+00$
 $Z(16, 7) = 0.65908E+00+j-0.24988E+00$
 $Z(17, 7) = 0.49402E+00+j-0.24016E+00$
 $Z(18, 7) = 0.37563E+00+j-0.22974E+00$
 $Z(19, 7) = 0.28728E+00+j-0.21868E+00$
 $Z(20, 7) = 0.21923E+00+j-0.20708E+00$
 $Z(21, 7) = 0.16554E+00+j-0.19502E+00$
 $Z(1, 8) = 0.12790E+01+j-0.26686E+00$
 $Z(2, 8) = 0.19197E+01+j-0.27398E+00$
 $Z(3, 8) = 0.31310E+01+j-0.28012E+00$
 $Z(4, 8) = 0.57993E+01+j-0.28520E+00$
 $Z(5, 8) = 0.13232E+02+j-0.28921E+00$
 $Z(6, 8) = 0.44756E+02+j-0.29209E+00$
 $Z(7, 8) = 0.41849E+03+j-0.29383E+00$
 $Z(8, 8) = -0.96043E+03+j-0.29443E+00$
 $Z(9, 8) = 0.41849E+03+j-0.29383E+00$
 $Z(10, 8) = 0.44756E+02+j-0.29209E+00$
 $Z(11, 8) = 0.13232E+02+j-0.28921E+00$
 $Z(12, 8) = 0.57993E+01+j-0.28520E+00$
 $Z(13, 8) = 0.31310E+01+j-0.28012E+00$
 $Z(14, 8) = 0.19197E+01+j-0.27398E+00$
 $Z(15, 8) = 0.12790E+01+j-0.26686E+00$
 $Z(16, 8) = 0.98103E+00+j-0.25880E+00$
 $Z(17, 8) = 0.65908E+00+j-0.24988E+00$
 $Z(18, 8) = 0.49402E+00+j-0.24016E+00$
 $Z(19, 8) = 0.37563E+00+j-0.22974E+00$
 $Z(20, 8) = 0.28728E+00+j-0.21868E+00$
 $Z(21, 8) = 0.21923E+00+j-0.20708E+00$
 $Z(1, 9) = 0.98103E+00+j-0.25880E+00$
 $Z(2, 9) = 0.12790E+01+j-0.26686E+00$
 $Z(3, 9) = 0.19197E+01+j-0.27398E+00$
 $Z(4, 9) = 0.31310E+01+j-0.28012E+00$
 $Z(5, 9) = 0.57993E+01+j-0.28520E+00$
 $Z(6, 9) = 0.13232E+02+j-0.28921E+00$
 $Z(7, 9) = 0.44756E+02+j-0.29209E+00$
 $Z(8, 9) = 0.41849E+03+j-0.29383E+00$
 $Z(9, 9) = -0.96043E+03+j-0.29443E+00$
 $Z(10, 9) = 0.41849E+03+j-0.29383E+00$
 $Z(11, 9) = 0.44756E+02+j-0.29209E+00$

$Z(12, 9) = 0.13232E+02+j-0.28921E+00$
 $Z(13, 9) = 0.57993E+01+j-0.28520E+00$
 $Z(14, 9) = 0.31310E+01+j-0.28012E+00$
 $Z(15, 9) = 0.19197E+01+j-0.27398E+00$
 $Z(16, 9) = 0.12790E+01+j-0.26686E+00$
 $Z(17, 9) = 0.98103E+00+j-0.25880E+00$
 $Z(18, 9) = 0.65908E+00+j-0.24988E+00$
 $Z(19, 9) = 0.49402E+00+j-0.24016E+00$
 $Z(20, 9) = 0.37563E+00+j-0.22974E+00$
 $Z(21, 9) = 0.28728E+00+j-0.21868E+00$
 $Z(1, 10) = 0.65908E+00+j-0.24988E+00$
 $Z(2, 10) = 0.98103E+00+j-0.25880E+00$
 $Z(3, 10) = 0.12790E+01+j-0.26686E+00$
 $Z(4, 10) = 0.19197E+01+j-0.27398E+00$
 $Z(5, 10) = 0.31310E+01+j-0.28012E+00$
 $Z(6, 10) = 0.57993E+01+j-0.28520E+00$
 $Z(7, 10) = 0.13232E+02+j-0.28921E+00$
 $Z(8, 10) = 0.44756E+02+j-0.29209E+00$
 $Z(9, 10) = 0.41849E+03+j-0.29383E+00$
 $Z(10, 10) = -0.96043E+03+j-0.29443E+00$
 $Z(11, 10) = 0.41849E+03+j-0.29383E+00$
 $Z(12, 10) = 0.44756E+02+j-0.29209E+00$
 $Z(13, 10) = 0.13232E+02+j-0.28921E+00$
 $Z(14, 10) = 0.57993E+01+j-0.28520E+00$
 $Z(15, 10) = 0.31310E+01+j-0.28012E+00$
 $Z(16, 10) = 0.19197E+01+j-0.27398E+00$
 $Z(17, 10) = 0.12790E+01+j-0.26686E+00$
 $Z(18, 10) = 0.98103E+00+j-0.25880E+00$
 $Z(19, 10) = 0.65908E+00+j-0.24988E+00$
 $Z(20, 10) = 0.49402E+00+j-0.24016E+00$
 $Z(21, 10) = 0.37563E+00+j-0.22974E+00$
 $Z(1, 11) = 0.49402E+00+j-0.24016E+00$
 $Z(2, 11) = 0.65908E+00+j-0.24988E+00$
 $Z(3, 11) = 0.98103E+00+j-0.25880E+00$
 $Z(4, 11) = 0.12790E+01+j-0.26686E+00$
 $Z(5, 11) = 0.19197E+01+j-0.27398E+00$
 $Z(6, 11) = 0.31310E+01+j-0.28012E+00$
 $Z(7, 11) = 0.57993E+01+j-0.28520E+00$
 $Z(8, 11) = 0.13232E+02+j-0.28921E+00$
 $Z(9, 11) = 0.44756E+02+j-0.29209E+00$
 $Z(10, 11) = 0.41849E+03+j-0.29383E+00$
 $Z(11, 11) = -0.96043E+03+j-0.29443E+00$
 $Z(12, 11) = 0.41849E+03+j-0.29383E+00$
 $Z(13, 11) = 0.44756E+02+j-0.29209E+00$
 $Z(14, 11) = 0.13232E+02+j-0.28921E+00$
 $Z(15, 11) = 0.57993E+01+j-0.28520E+00$
 $Z(16, 11) = 0.31310E+01+j-0.28012E+00$
 $Z(17, 11) = 0.19197E+01+j-0.27398E+00$
 $Z(18, 11) = 0.12790E+01+j-0.26686E+00$
 $Z(19, 11) = 0.98103E+00+j-0.25880E+00$
 $Z(20, 11) = 0.65908E+00+j-0.24988E+00$
 $Z(21, 11) = 0.49402E+00+j-0.24016E+00$
 $Z(1, 12) = 0.37563E+00+j-0.22974E+00$
 $Z(2, 12) = 0.49402E+00+j-0.24016E+00$
 $Z(3, 12) = 0.65908E+00+j-0.24988E+00$
 $Z(4, 12) = 0.98103E+00+j-0.25880E+00$
 $Z(5, 12) = 0.12790E+01+j-0.26686E+00$
 $Z(6, 12) = 0.19197E+01+j-0.27398E+00$
 $Z(7, 12) = 0.31310E+01+j-0.28012E+00$
 $Z(8, 12) = 0.57993E+01+j-0.28520E+00$
 $Z(9, 12) = 0.13232E+02+j-0.28921E+00$

Cont'd.

12.33 cont'd.

$Z(10,12) = 0.44756E+02+J-0.29209E+00$	$Z(8,15) = 0.12790E+01+J-0.26686E+00$
$Z(11,12) = 0.41849E+03+J-0.29383E+00$	$Z(9,15) = 0.19197E+01+J-0.27398E+00$
$Z(12,12) = -0.96043E+03+J-0.29443E+00$	$Z(10,15) = 0.31310E+01+J-0.28012E+00$
$Z(13,12) = 0.41849E+03+J-0.29383E+00$	$Z(11,15) = 0.57993E+01+J-0.28520E+00$
$Z(14,12) = 0.44756E+02+J-0.29209E+00$	$Z(12,15) = 0.13232E+02+J-0.28921E+00$
$Z(15,12) = 0.13232E+02+J-0.28921E+00$	$Z(13,15) = 0.44756E+02+J-0.29209E+00$
$Z(16,12) = 0.57993E+01+J-0.28520E+00$	$Z(14,15) = 0.41849E+03+J-0.29383E+00$
$Z(17,12) = 0.31310E+01+J-0.28012E+00$	$Z(15,15) = -0.96043E+03+J-0.29443E+00$
$Z(18,12) = 0.19197E+01+J-0.27398E+00$	$Z(16,15) = 0.41849E+03+J-0.29383E+00$
$Z(19,12) = 0.12790E+01+J-0.26686E+00$	$Z(17,15) = 0.44756E+02+J-0.29209E+00$
$Z(20,12) = 0.90103E+00+J-0.25880E+00$	$Z(18,15) = 0.13232E+02+J-0.28921E+00$
$Z(21,12) = 0.65908E+00+J-0.24988E+00$	$Z(19,15) = 0.57993E+01+J-0.28520E+00$
$Z(1,13) = 0.28728E+00+J-0.21868E+00$	$Z(20,15) = 0.31310E+01+J-0.28012E+00$
$Z(2,13) = 0.37563E+00+J-0.22974E+00$	$Z(21,15) = 0.19197E+01+J-0.27398E+00$
$Z(3,13) = 0.49402E+00+J-0.24016E+00$	$Z(1,16) = 0.12239E+00+J-0.18262E+00$
$Z(4,13) = 0.65908E+00+J-0.24988E+00$	$Z(2,16) = 0.16554E+00+J-0.18502E+00$
$Z(5,13) = 0.90103E+00+J-0.25880E+00$	$Z(3,16) = 0.21923E+00+J-0.20708E+00$
$Z(6,13) = 0.12790E+01+J-0.26686E+00$	$Z(4,16) = 0.28728E+00+J-0.21868E+00$
$Z(7,13) = 0.19197E+01+J-0.27398E+00$	$Z(5,16) = 0.37563E+00+J-0.22974E+00$
$Z(8,13) = 0.31310E+01+J-0.28012E+00$	$Z(6,16) = 0.49402E+00+J-0.24016E+00$
$Z(9,13) = 0.57993E+01+J-0.28520E+00$	$Z(7,16) = 0.65908E+00+J-0.24988E+00$
$Z(10,13) = 0.13232E+02+J-0.28921E+00$	$Z(8,16) = 0.90103E+00+J-0.25880E+00$
$Z(11,13) = 0.44756E+02+J-0.29209E+00$	$Z(9,16) = 0.12790E+01+J-0.26686E+00$
$Z(12,13) = 0.41849E+03+J-0.29383E+00$	$Z(10,16) = 0.19197E+01+J-0.27398E+00$
$Z(13,13) = -0.96043E+03+J-0.29443E+00$	$Z(11,16) = 0.31310E+01+J-0.28012E+00$
$Z(14,13) = 0.41849E+03+J-0.29383E+00$	$Z(12,16) = 0.57993E+01+J-0.28520E+00$
$Z(15,13) = 0.44756E+02+J-0.29209E+00$	$Z(13,16) = 0.13232E+02+J-0.28921E+00$
$Z(16,13) = 0.13232E+02+J-0.28921E+00$	$Z(14,16) = 0.44756E+02+J-0.29209E+00$
$Z(17,13) = 0.57993E+01+J-0.28520E+00$	$Z(15,16) = 0.41849E+03+J-0.29383E+00$
$Z(18,13) = 0.31310E+01+J-0.28012E+00$	$Z(16,16) = -0.96043E+03+J-0.29443E+00$
$Z(19,13) = 0.19197E+01+J-0.27398E+00$	$Z(17,16) = 0.41849E+03+J-0.29383E+00$
$Z(20,13) = 0.12790E+01+J-0.26686E+00$	$Z(18,16) = 0.44756E+02+J-0.29209E+00$
$Z(21,13) = 0.90103E+00+J-0.25880E+00$	$Z(19,16) = 0.13232E+02+J-0.28921E+00$
$Z(1,14) = 0.21923E+00+J-0.20708E+00$	$Z(20,16) = 0.57993E+01+J-0.28520E+00$
$Z(2,14) = 0.28728E+00+J-0.21868E+00$	$Z(21,16) = 0.31310E+01+J-0.28012E+00$
$Z(3,14) = 0.37563E+00+J-0.22974E+00$	$Z(1,17) = 0.87235E-01+J-0.16995E+00$
$Z(4,14) = 0.49402E+00+J-0.24016E+00$	$Z(2,17) = 0.12239E+00+J-0.18262E+00$
$Z(5,14) = 0.65908E+00+J-0.24988E+00$	$Z(3,17) = 0.16554E+00+J-0.19502E+00$
$Z(6,14) = 0.90103E+00+J-0.25880E+00$	$Z(4,17) = 0.21923E+00+J-0.20708E+00$
$Z(7,14) = 0.12790E+01+J-0.28686E+00$	$Z(5,17) = 0.28728E+00+J-0.21868E+00$
$Z(8,14) = 0.19197E+01+J-0.27398E+00$	$Z(6,17) = 0.37563E+00+J-0.22974E+00$
$Z(9,14) = 0.31310E+01+J-0.28012E+00$	$Z(7,17) = 0.49402E+00+J-0.24016E+00$
$Z(10,14) = 0.57993E+01+J-0.28520E+00$	$Z(8,17) = 0.65908E+00+J-0.24988E+00$
$Z(11,14) = 0.13232E+02+J-0.28921E+00$	$Z(9,17) = 0.90103E+00+J-0.25880E+00$
$Z(12,14) = 0.44756E+02+J-0.29209E+00$	$Z(10,17) = 0.12790E+01+J-0.26686E+00$
$Z(13,14) = 0.41849E+03+J-0.29383E+00$	$Z(11,17) = 0.19197E+01+J-0.27398E+00$
$Z(14,14) = -0.96043E+03+J-0.29443E+00$	$Z(12,17) = 0.31310E+01+J-0.28012E+00$
$Z(15,14) = 0.41849E+03+J-0.29383E+00$	$Z(13,17) = 0.57993E+01+J-0.28520E+00$
$Z(16,14) = 0.44756E+02+J-0.29209E+00$	$Z(14,17) = 0.13232E+02+J-0.28921E+00$
$Z(17,14) = 0.13232E+02+J-0.28921E+00$	$Z(15,17) = 0.44756E+02+J-0.29209E+00$
$Z(18,14) = 0.57993E+01+J-0.28520E+00$	$Z(16,17) = 0.41849E+03+J-0.29383E+00$
$Z(19,14) = 0.31310E+01+J-0.28012E+00$	$Z(17,17) = -0.96043E+03+J-0.29443E+00$
$Z(20,14) = 0.19197E+01+J-0.27398E+00$	$Z(18,17) = 0.41849E+03+J-0.29383E+00$
$Z(21,14) = 0.12790E+01+J-0.26686E+00$	$Z(19,17) = 0.44756E+02+J-0.29209E+00$
$Z(1,15) = 0.16554E+00+J-0.19502E+00$	$Z(20,17) = 0.13232E+02+J-0.28921E+00$
$Z(2,15) = 0.21923E+00+J-0.20708E+00$	$Z(21,17) = 0.57993E+01+J-0.28520E+00$
$Z(3,15) = 0.28728E+00+J-0.21868E+00$	$Z(1,18) = 0.58352E-01+J-0.15711E+00$
$Z(4,15) = 0.37563E+00+J-0.22974E+00$	$Z(2,18) = 0.87235E-01+J-0.16995E+00$
$Z(5,15) = 0.49402E+00+J-0.24016E+00$	$Z(3,18) = 0.12239E+00+J-0.18262E+00$
$Z(6,15) = 0.65908E+00+J-0.24988E+00$	$Z(4,18) = 0.16554E+00+J-0.19502E+00$
$Z(7,15) = 0.90103E+00+J-0.25880E+00$	$Z(5,18) = 0.21923E+00+J-0.20708E+00$

Cont'd.

12.33 cont'd.

Z(6,18) =	0.28728E+00+j-0.21868E+00	Z(4,21) =	0.58352E-01+j-0.15711E+00
Z(7,18) =	0.37563E+00+j-0.22974E+00	Z(5,21) =	0.87235E-01+j-0.16995E+00
Z(8,18) =	0.49402E+00+j-0.24016E+00	Z(6,21) =	0.12239E+00+j-0.18262E+00
Z(9,18) =	0.65908E+00+j-0.24988E+00	Z(7,21) =	0.16554E+00+j-0.19502E+00
Z(10,18) =	0.90103E+00+j-0.25880E+00	Z(8,21) =	0.21923E+00+j-0.20708E+00
Z(11,18) =	0.12790E+01+j-0.26686E+00	Z(9,21) =	0.28728E+00+j-0.21868E+00
Z(12,18) =	0.19197E+01+j-0.27398E+00	Z(10,21) =	0.37563E+00+j-0.22974E+00
Z(13,18) =	0.31310E+01+j-0.28012E+00	Z(11,21) =	0.49402E+00+j-0.24016E+00
Z(14,18) =	0.57993E+01+j-0.28520E+00	Z(12,21) =	0.65908E+00+j-0.24988E+00
Z(15,18) =	0.13232E+02+j-0.28921E+00	Z(13,21) =	0.90103E+00+j-0.25880E+00
Z(16,18) =	0.44756E+02+j-0.29209E+00	Z(14,21) =	0.12790E+01+j-0.26686E+00
Z(17,18) =	0.41849E+03+j-0.29383E+00	Z(15,21) =	0.19197E+01+j-0.27398E+00
Z(18,18) =	-0.96043E+03+j-0.29443E+00	Z(16,21) =	0.31310E+01+j-0.28012E+00
Z(19,18) =	0.41849E+03+j-0.29383E+00	Z(17,21) =	0.57993E+01+j-0.28520E+00
Z(20,18) =	0.44756E+02+j-0.29209E+00	Z(18,21) =	0.13232E+02+j-0.28921E+00
Z(21,18) =	0.13232E+02+j-0.28921E+00	Z(19,21) =	0.44756E+02+j-0.29209E+00
Z(1,19) =	0.34510E-01+j-0.14419E+00	Z(20,21) =	0.41849E+03+j-0.29383E+00
Z(2,19) =	0.58352E-01+j-0.15711E+00	Z(21,21) =	-0.96043E+03+j-0.29443E+00
Z(3,19) =	0.87235E-01+j-0.16995E+00		
Z(4,19) =	0.12239E+00+j-0.18262E+00		
Z(5,19) =	0.16554E+00+j-0.19502E+00		
Z(6,19) =	0.21923E+00+j-0.20708E+00		
Z(7,19) =	0.28728E+00+j-0.21868E+00		
Z(8,19) =	0.37563E+00+j-0.22974E+00		
Z(9,19) =	0.49402E+00+j-0.24016E+00		
Z(10,19) =	0.65908E+00+j-0.24988E+00		
Z(11,19) =	0.90103E+00+j-0.25880E+00		
Z(12,19) =	0.12790E+01+j-0.26686E+00		
Z(13,19) =	0.19197E+01+j-0.27398E+00		
Z(14,19) =	0.31310E+01+j-0.28012E+00		
Z(15,19) =	0.57993E+01+j-0.28520E+00		
Z(16,19) =	0.13232E+02+j-0.28921E+00		
Z(17,19) =	0.44756E+02+j-0.29209E+00		
Z(18,19) =	0.41849E+03+j-0.29383E+00		
Z(19,19) =	-0.96043E+03+j-0.29443E+00		
Z(20,19) =	0.41849E+03+j-0.29383E+00		
Z(21,19) =	0.44756E+02+j-0.29209E+00		
Z(1,20) =	0.14817E-01+j-0.13130E+00		
Z(2,20) =	0.34510E-01+j-0.14419E+00		
Z(3,20) =	0.58352E-01+j-0.15711E+00		
Z(4,20) =	0.87235E-01+j-0.16995E+00		
Z(5,20) =	0.12239E+00+j-0.18262E+00		
Z(6,20) =	0.16554E+00+j-0.19502E+00		
Z(7,20) =	0.21923E+00+j-0.20708E+00		
Z(8,20) =	0.28728E+00+j-0.21868E+00		
Z(9,20) =	0.37563E+00+j-0.22974E+00		
Z(10,20) =	0.49402E+00+j-0.24016E+00		
Z(11,20) =	0.65908E+00+j-0.24988E+00		
Z(12,20) =	0.90103E+00+j-0.25880E+00		
Z(13,20) =	0.12790E+01+j-0.26686E+00		
Z(14,20) =	0.19197E+01+j-0.27398E+00		
Z(15,20) =	0.31310E+01+j-0.28012E+00		
Z(16,20) =	0.57993E+01+j-0.28520E+00		
Z(17,20) =	0.13232E+02+j-0.28921E+00		
Z(18,20) =	0.44756E+02+j-0.29209E+00		
Z(19,20) =	0.41849E+03+j-0.29383E+00		
Z(20,20) =	-0.96043E+03+j-0.29443E+00		
Z(21,20) =	0.41849E+03+j-0.29383E+00		
Z(1,21) =	-0.13950E-02+j-0.11552E+00		
Z(2,21) =	0.14817E-01+j-0.13130E+00		
Z(3,21) =	0.34510E-01+j-0.14419E+00		

12.34

$$l = \pi/50, a = 0.005\pi, N = 21$$

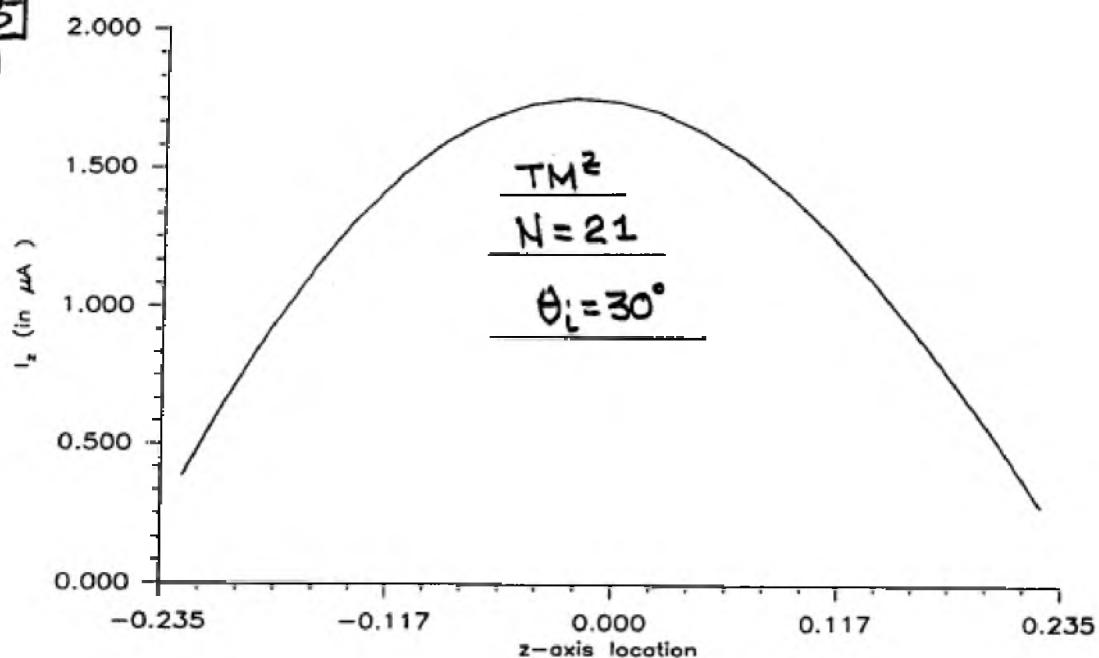
Delta Gap

Magnetic Frill

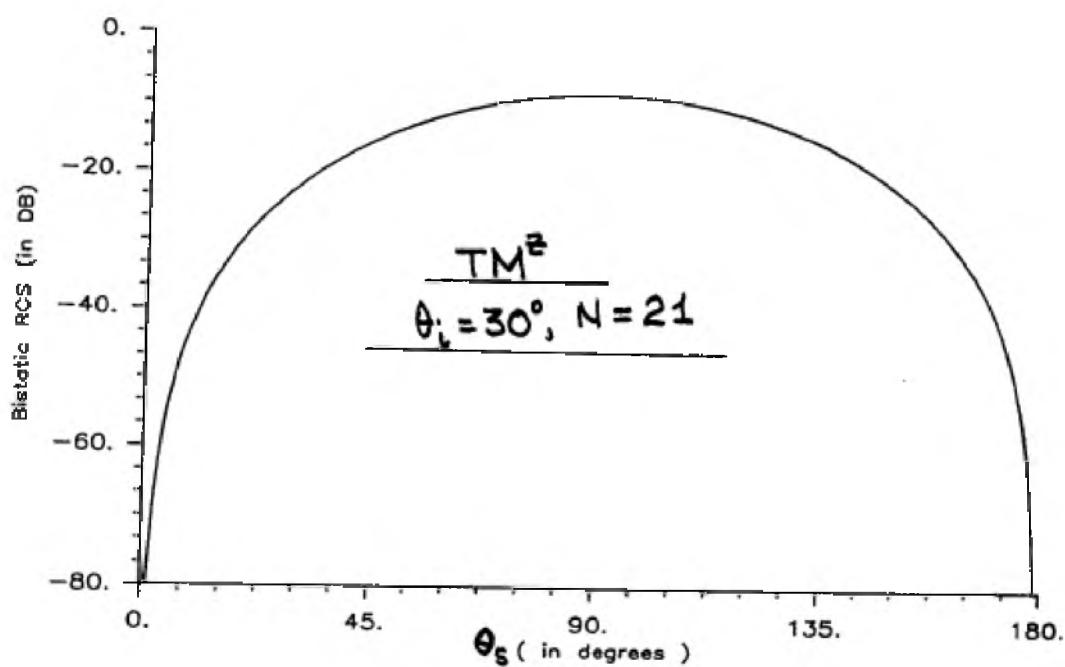
$$Z_{in} = 8.17 \times 10^6 - j 2.15 \times 10^{-2} \text{ Ohms}; \quad Z_{in} = 0.2199 - j 421.4 \text{ Ohms}$$

12.35

(a)



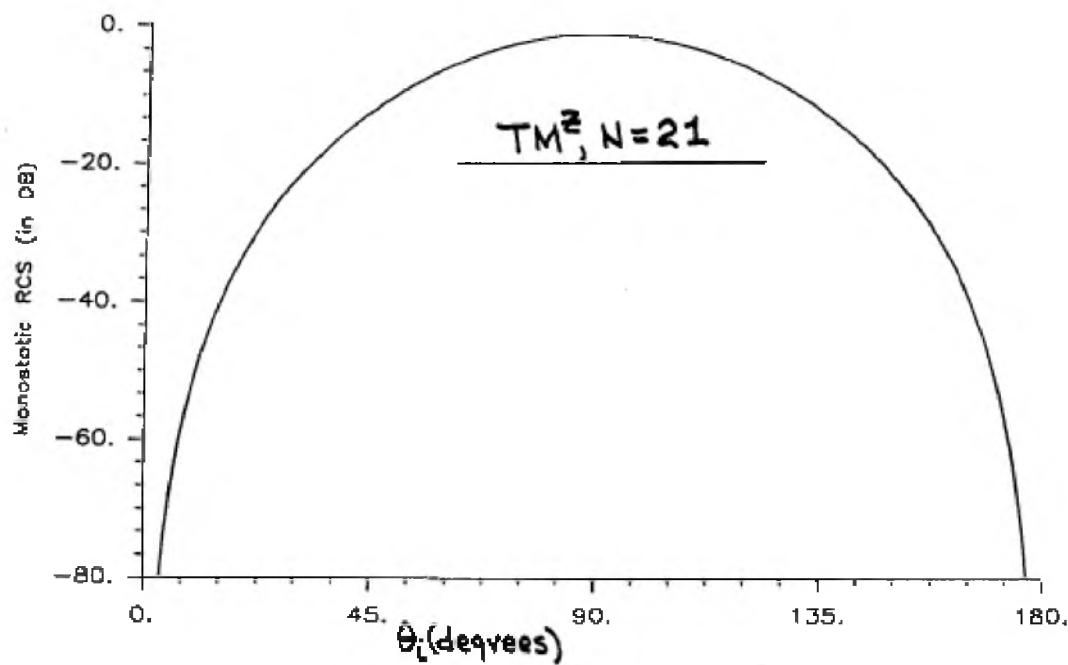
(b)



Cont'd.

12.35 cont'd.

(c)



Stability is achieved and maintained when the number of segments does not exceed about 100 ($N \leq 100$). However $N=21$ gives very good results.

12.36

For TE^2 uniform plane wave incidence, the solutions of parts a,b,c of Problem 12.35 are all zero because the incident electric field is perpendicular to the wire.

CHAPTER 13

- 13.1** Let α = angular space from Q-Q' along z axis
 β = angular space from P-P' along z axis

Therefore each area, dA and dA_0 , is defined as the product of the arc lengths in the xz and yz planes. Thus

$$dA_0 = (p_1 \beta)(p_2 \alpha) = \alpha \beta p_1 p_2$$

$$dA = [(p_1 + s)\beta][(p_2 + s)\alpha] = \alpha \beta (p_1 + s)(p_2 + s)$$

Therefore we can write that

$$\frac{|E|}{|E_0|} = \sqrt{\frac{dA_0}{dA}} = \sqrt{\frac{\alpha \beta p_1 p_2}{\alpha \beta (p_1 + s)(p_2 + s)}} = \sqrt{\frac{p_1 p_2}{(p_1 + s)(p_2 + s)}}$$

- 13.2** $z = f - \frac{x^2 + y^2}{4f}$. Since $z = g(u)$ where $u = \frac{x^2 + y^2}{2}$

$$\text{then } z = f - \frac{1}{2f} u$$

$$\frac{dg(u)}{du} = \frac{d}{du} \left[f - \frac{1}{2f} u \right] = -\frac{1}{2f}$$

$$\frac{d^2g(u)}{du^2} = \frac{d}{du} \left[-\frac{1}{2f} \right] = 0$$

Thus

$$K^2 = 1 + 2u \left[\frac{dg(u)}{du} \right]^2 = 1 + 2u \left[-\frac{1}{2f} \right]^2 = 1 + \frac{2u}{4f^2} = 1 + \frac{u}{2f^2} = 1 + \frac{x^2 + y^2}{4f^2}$$

$$K = \sqrt{1 + \frac{x^2 + y^2}{4f^2}}$$

$$\frac{1}{R_1} = \frac{1}{K} \frac{dg(u)}{du} = -\frac{1}{2f} \frac{1}{\sqrt{1 + \frac{x^2 + y^2}{4f^2}}}$$

$$\frac{1}{R_2} = \frac{1}{K^3} \left[\frac{dg(u)}{du} + 2u \frac{d^2g(u)}{du^2} \right]^0 = \frac{1}{K^3} \frac{dg(u)}{du}$$

$$= -\frac{1}{2f} \frac{1}{\left[1 + \frac{x^2 + y^2}{4f^2} \right]^{3/2}}$$

$$[13.3] S_0(0) = 10 \text{ mWatts/cm}^2$$

$$r_0 = 5 \text{ m} = 500 \text{ cm}$$

$$r_1 = 50 \text{ m} = 5,000 \text{ cm}$$

$$\frac{|E|^2}{|E_0|^2} = \frac{dA_0}{dA_1} = \frac{S(s)}{S_0(0)}$$

$$(a) dA_0 = \int_0^{\pi/6} \int_0^{r_0} r_0^2 \sin\theta d\theta d\phi = 2\pi r_0^2 \int_0^{\pi/6} \sin\theta d\theta$$

$$= 2\pi r_0^2 (-\cos\theta)$$

$$= 2\pi r_0^2 (-\cos 30^\circ + \cos 0)$$

$$= 2\pi r_0^2 \left(-\frac{\sqrt{3}}{2} + 1\right) = 2\pi r_0^2 (1 - 0.866)$$

$$dA_0 = 2\pi r_0^2 (0.13397) = 0.84179 r_0^2$$

$$dA_0 = 0.841797 (500)^2 = 21,044.68 \times 10^4 = 21,044.68 \times 10^3 \text{ cm}^2$$

$$W = S_0(r_0) dA_0 = 10 \times 10^{-3} (21,044.68 \times 10^3) = 210.4468 \text{ Watts}$$

$$W = 210.4468 \text{ Watts}$$

$$(b) \frac{S(r)}{S(r_0)} = \frac{dA_0}{dA_1} = \frac{0.841797 r_0^2}{0.841797 r_1^2} = \left(\frac{r_0}{r_1}\right)^2 = \left(\frac{5}{50}\right)^2 = \left(\frac{1}{10}\right)^2 = \frac{1}{100}$$

$$S(r_1) = \frac{1}{100} S(r_0) = \frac{10 \times 10^{-3}}{100} = 0.1 \times 10^{-3}$$

$$S(r_1) = 0.1 \times 10^{-3} = 0.1 \text{ mWatts/cm}^2$$

Alternate:

$$\frac{S(r)}{S(r_0)} = \frac{P_1 P_2}{(P_1 + S)(P_2 + S)} = \frac{S(s)}{(S+45)(S+45)} = \frac{S(s)}{(50)(50)} = \frac{1}{100}$$

$$S(r_1) = \frac{1}{100} S(r_0) = 0.1 \times 10^{-3} = 0.1 \text{ mWatts/cm}^2$$

13.4 To derive (13-30) we look at the geometry of Figure 13-8.

The length of the arc between points $Q_R - Q_R$ is equal to

$$\rho_a \Delta\phi = \frac{\Delta\Psi_0}{\cos\theta_i} = \frac{\rho_0 \Delta\Psi_1}{\cos\theta_i}$$

Since

$$\Psi_1 = \pi - [\phi + \frac{1}{2}(2\pi - 2\theta_i)] = \pi - [\phi + (\pi - \theta_i)] = \theta_i - \phi$$

then

$$\Delta\Psi_1 = \Delta\theta_i - \Delta\phi$$

and we can write that

$$(1) \quad \rho_a \Delta\phi = \frac{\rho_0 (\Delta\theta_i - \Delta\phi)}{\cos\theta_i}$$

Also

$$\rho_a \Delta\phi = \frac{\rho^r \Delta\Psi_2}{\cos\theta_i}$$

Since

$$\Delta\Psi_2 = \Delta\theta_i + \Delta\phi$$

then

$$(2) \quad \rho_a \Delta\phi = \frac{\rho^r \Delta\Psi_2}{\cos\theta_i} = \frac{\rho^r (\Delta\theta_i + \Delta\phi)}{\cos\theta_i}$$

Equating (1) and (2) we have that

$$\rho_a \Delta\phi \cos\theta_i = \rho_0 (\Delta\theta_i - \Delta\phi) = \rho^r (\Delta\theta_i + \Delta\phi)$$

or

$$(3) \quad \Delta\theta_i - \Delta\phi = \frac{\rho_a}{\rho_0} \Delta\phi \cos\theta_i \Rightarrow \Delta\phi \left(1 + \frac{\rho_a}{\rho_0} \cos\theta_i \right) = \Delta\theta_i \Rightarrow \Delta\phi = \frac{\rho_0 \Delta\theta_i}{\rho_0 + \rho_a \cos\theta_i}$$

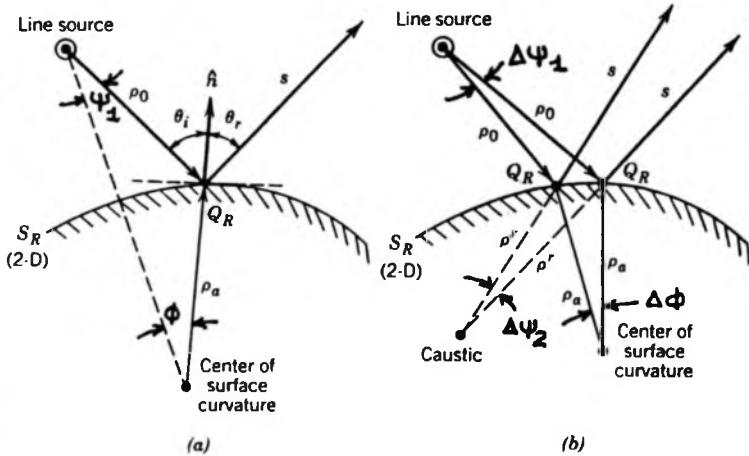
$$(4) \quad \Delta\theta_i + \Delta\phi = \frac{\rho_a}{\rho^r} \Delta\phi \cos\theta_i \Rightarrow \Delta\phi \left(-1 + \frac{\rho_a}{\rho^r} \cos\theta_i \right) = \Delta\theta_i \Rightarrow \Delta\phi = \frac{\rho^r \Delta\theta_i}{-\rho^r + \rho_a \cos\theta_i}$$

Equating (3) and (4) leads to

$$(5) \quad \frac{\rho_0 \Delta\theta_i}{\rho_0 + \rho_a \cos\theta_i} = \frac{\rho^r \Delta\theta_i}{-\rho^r + \rho_a \cos\theta_i} \Rightarrow \frac{\rho_0}{\rho_0 + \rho_a \cos\theta_i} = \frac{\rho^r}{-\rho^r + \rho_a \cos\theta_i}$$

Solving (5) leads to

$$\frac{1}{\rho^r} = \frac{1}{\rho_0} + \frac{2}{\rho_a \cos\theta_i}$$



13.5(a) $\frac{1}{p_r} = \frac{1}{p_0} + \frac{2}{p_a \cos \theta_i}$, $p_0 = a$, $p_a = +a$, $\theta_i = 0$

Therefore

$$\frac{1}{p_r} = \frac{1}{a} + \frac{2}{a} = +\frac{3}{a} \Rightarrow p_r = a/3$$

Thus the caustic is at the origin of the circle.

(b) $E^r = E^i(\theta_R) \cdot R \sqrt{\frac{p^r}{p^r + s}} e^{-js}$

$$E^i(\theta_R) = \hat{a}_z 10^{-3} \text{ V/m}$$

$$R = -\hat{a}_z \hat{a}_t$$

$$p^r = +a/3 = 5\lambda/3$$

$$s = p + a = 50\lambda + 5\lambda = 55\lambda$$

$$E^r(p=50\lambda, \phi=180^\circ) = (\hat{a}_z 10^{-3}) \cdot (-\hat{a}_z \hat{a}_t) \sqrt{\frac{+5\lambda/3}{5\lambda/3 + 55\lambda}} e^{-j\frac{2\pi}{\lambda}(55\lambda)}$$

$$= -\hat{a}_z 10^{-3} \sqrt{\frac{5}{570}} e^{-j110\pi} = -\hat{a}_z 0.1715 \times 10^{-3}$$

13.6(a) $\frac{1}{p_r} = \frac{1}{p_0} - \frac{2}{p_a \cos \theta_i}$, $p_0 = a$, $p_a = -a$, $\theta_i = 0$

Therefore

$$\frac{1}{p_r} = \frac{1}{a} - \frac{2}{a} = -\frac{1}{a} \Rightarrow p_r = -a$$

Thus the caustic is at the origin of the circle.

(b) $E^r = E^i(\theta_R) \cdot R \sqrt{\frac{p^r}{p^r + s}} e^{-js}$

$$E^i(\theta_R) = \hat{a}_z 10^{-3} \text{ V/m}$$

$$R = -\hat{a}_z \hat{a}_t$$

$$p^r = -a = -5\lambda$$

$$s = p + a = 50\lambda + 5\lambda = 55\lambda$$

$$E^r(p=50\lambda, \phi=180^\circ) = (\hat{a}_z 10^{-3}) \cdot (-\hat{a}_z \hat{a}_t) \sqrt{\frac{-5\lambda}{-5\lambda + 55\lambda}} e^{-j\frac{2\pi}{\lambda}(55\lambda)}$$

$$= -\hat{a}_z 10^{-3} \sqrt{\frac{5}{50}} e^{-j110\pi} = -\hat{a}_z 0.3162 \times 10^{-3} e^{\pm j\pi/2}$$

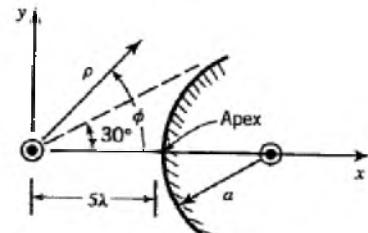


Figure P13-5

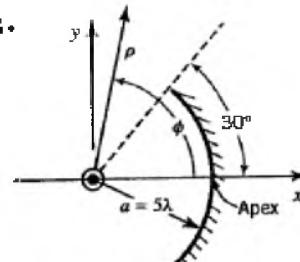


Figure P13-6

[13.7]

According to (13-20)

$$E^r(s) = E^i(z=0) \cdot R \sqrt{\frac{P_1 P_2}{(P_1 + s)(P_2 + s)}} e^{-js}$$

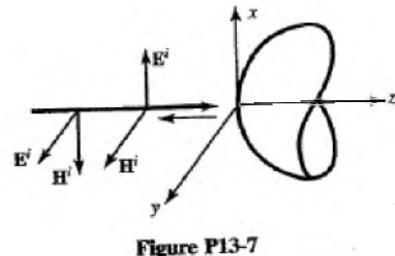


Figure P13-7

$$\stackrel{s \gg P_1 P_2}{\sim} E^i(z=0) \cdot R \sqrt{\frac{P_1 P_2}{s}} e^{-js} = E^i(z=0) \cdot R \sqrt{\frac{P_1 P_2}{s}} \frac{e^{-js}}{s}$$

According to (13-23a)

$$(a) E^i_{||} = \hat{a}_x e^{-j\beta z} \Big|_{z=0} = \hat{a}_x, \bar{R}_{||} = -\hat{a}_x \hat{a}_x$$

$$|E^s| = \hat{a}_x \cdot (-\hat{a}_x \hat{a}_x) \sqrt{P_1 P_2} \frac{e^{-js}}{s} = -\hat{a}_x \sqrt{\frac{R_1 R_2}{4}} \frac{e^{-js}}{s} = -\hat{a}_x \sqrt{\frac{196(19.6)}{4}} \frac{e^{-js}}{s}$$

$$\approx -\hat{a}_x \sqrt{\frac{196\pi^2}{4}} \frac{e^{-js}}{s} = -\hat{a}_x \mp \lambda \frac{e^{-js}}{s}$$

$$\sigma = \lim_{s \rightarrow \infty} \left[4\pi s^2 \frac{|E^s|^2}{|E^i|^2} \right] = 4\pi s^2 \left[\frac{(\mp \lambda)^2}{s^2} \right] = 4\pi (49) \lambda^2 = 196\pi \lambda^2$$

$$\frac{\sigma}{\lambda^2} = 196\pi = 615.752 \approx 10 \log_{10}(615.752) = 27.894 \text{ dB}$$

$$\frac{\sigma}{\lambda^2} = 615.752 = 27.894 \text{ dB}$$

$$(b) E^i_{\perp} = \hat{a}_y e^{-j\beta z} \Big|_{z=0} = \hat{a}_y, \bar{R}_{\perp} = -\hat{a}_y \hat{a}_y$$

$$E^s \approx \hat{a}_y \cdot (-\hat{a}_y \hat{a}_y) \sqrt{P_1 P_2} \frac{e^{-js}}{s} = -\hat{a}_y \sqrt{P_1 P_2} \frac{e^{-js}}{s} = -\hat{a}_y \sqrt{\frac{R_1 R_2}{4}} \frac{e^{-js}}{s}$$

$$|E^s| \approx -\hat{a}_y \sqrt{\frac{196\pi^2}{4}} \frac{e^{-js}}{s} = -\hat{a}_y \mp \lambda \frac{e^{-js}}{s}$$

$$\sigma = \lim_{s \rightarrow \infty} \left[4\pi s^2 \frac{|E^s|^2}{|E^i|^2} \right] = 4\pi s^2 \left[\frac{(\mp \lambda)^2}{s^2} \right] = 196\pi \lambda^2$$

$$\frac{\sigma}{\lambda^2} = 196\pi = 615.752 = 27.894 \text{ dB}$$

13.8

$$\underline{E}_{G0}^r(\text{strip}) = \underline{E}^L(z=0) \bar{R} A e^{-j\beta s}$$

$$= \underline{E}^L(z=0) (-1) \sqrt{\frac{p^r}{p^r+s}} e^{-j\beta s}$$

$$\underline{E}_{G0}^r(\text{strip}) = -\underline{E}^L(z=0) \sqrt{\frac{s'}{s'+s}} e^{-j\beta s}$$

$$= -\underline{E}^L(z=0) \sqrt{\frac{s'/s}{s'/s+1}} e^{-j\beta s} - \underline{E}^L(z=0) \sqrt{\frac{s'}{s}} e^{-j\beta s}$$

$$\underline{E}_{G0}^r(\text{cylinder}) = \underline{E}^L(z=0) \cdot \bar{R} A e^{-j\beta s}$$

$$= \underline{E}^L(z=0) (-1) \sqrt{\frac{p^r}{p^r+s}} e^{-j\beta s} = -\underline{E}^L(z=0) \sqrt{\frac{p^r}{p^r+s}} e^{-j\beta s}$$

$$\frac{1}{p^r} = \frac{1}{p_0} + \frac{2}{p_0 \cos \theta_L} = \frac{1}{s'} + \frac{2}{a} = \frac{a+s'(2)}{as'} \xrightarrow{s' \gg a} \frac{2}{a} \Rightarrow p^r = \frac{a}{2}$$

$$\sqrt{\frac{p^r}{p^r+s}} = \sqrt{\frac{a/2}{a/2+s}} \xrightarrow{s' \gg a} \sqrt{\frac{a}{2s}}$$

$$\underline{E}_{G0}^r(\text{cylinder}) = -\underline{E}^L(z=0) \sqrt{\frac{p^r}{p^r+s}} e^{-j\beta s} = -\underline{E}^L(z=0) \sqrt{\frac{a}{2s}} e^{-j\beta s} = -\underline{E}^L(z=0) \sqrt{\frac{a/2}{s}} e^{-j\beta s}$$

$$\left| \frac{\underline{E}_{G0}^r(\text{cylinder})}{\underline{E}_{G0}^r(\text{strip})} \right| = \left| \frac{\underline{E}^L(z=0) \sqrt{\frac{a/2}{s}} e^{-j\beta s}}{-\underline{E}^L(z=0) \sqrt{\frac{a}{s'}} e^{-j\beta s}} \right| = \sqrt{\frac{a/2}{s}} \left(\frac{s}{s'} \right) = \sqrt{\frac{a/2}{s'}}$$

$$\left| \frac{\underline{E}_{G0}^r(\text{cylinder})}{\underline{E}_{G0}^r(\text{strip})} \right| = -20 \text{ dB} = +20 \log_{10} \left| \frac{\sqrt{a/2}}{s'} \right| \Rightarrow \frac{\sqrt{a/2}}{s'} = \frac{1}{10} \Rightarrow \frac{a/2}{s'} = \frac{1}{100}$$

$$\frac{a}{2} = \frac{s'}{100} \Rightarrow a = \frac{25}{100} = \frac{2(10\lambda)}{100} = \frac{20\lambda}{100} = \frac{2}{5}$$

$$\frac{s'}{100} = 10\lambda$$

$$\therefore a = \frac{2}{5}, \quad s' = 10\lambda$$

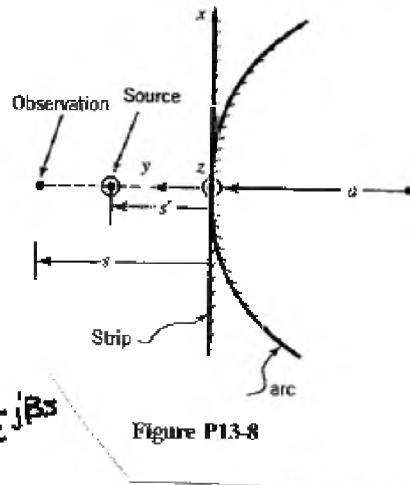
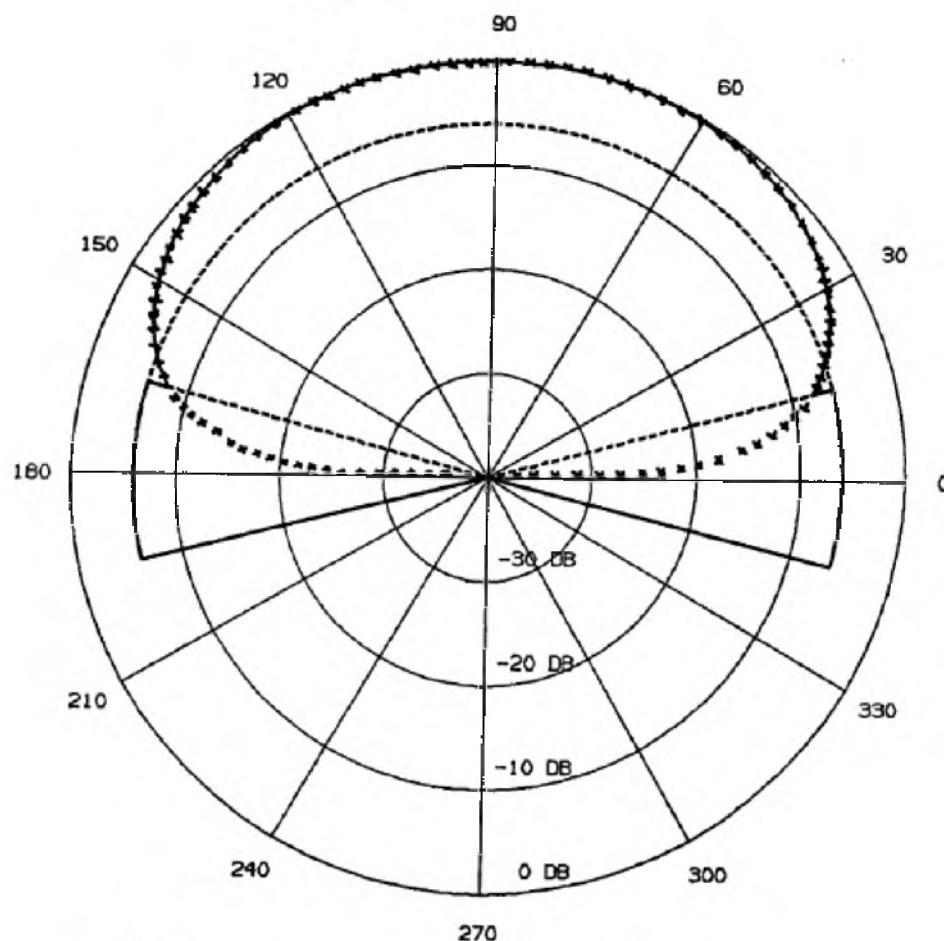


Figure P13.8

13.9

..... INCIDENT FIELD
 ----- REFLECTED FIELD
 —— TOTAL FIELD
 * * * * TOTAL FIELD ($\omega = \infty$)

(W = 2.0 λ , H = 0.25 λ)

$$13.10 \quad F_1(\beta\theta) \Big|_{SDP_{\pm n}} = \frac{e^{-j\pi/4}}{2n\sqrt{2\pi}\beta} \left\{ \cot\left[\frac{\pi + (\phi - \theta)}{2n}\right] + \cot\left[\frac{\pi - (\phi - \theta)}{2n}\right] \right\} e^{-j\beta\theta}$$

Since

$$\cot(x+y) = \frac{\cos(x+y)}{\sin(x+y)}, \quad \cot(x-y) = \frac{\cos(x-y)}{\sin(x-y)}$$

then $\cot(x+y) + \cot(x-y) = \frac{\cos(x+y)}{\sin(x+y)} + \frac{\cos(x-y)}{\sin(x-y)} = \frac{\cos(2x)\sin(x-y) + \sin(2x)\cos(x-y)}{\sin(x+y)\sin(x-y)}$

Also $\cos(x+y)\sin(x-y) = \frac{1}{2} [\sin(2x) - \sin(2y)]$

$$\sin(x+y)\cos(x-y) = \frac{1}{2} [\sin(2x) + \sin(2y)]$$

$$\sin(x+y)\sin(x-y) = -\frac{1}{2} [\cos(2x) - \cos(2y)]$$

Thus $\cot(x+y) + \cot(x-y) = \frac{\frac{1}{2} [\sin(2x) - \sin(2y)] + \frac{1}{2} [\sin(2x) + \sin(2y)]}{-\frac{1}{2} [\cos(2x) - \cos(2y)]} = -\frac{2 \sin(2x)}{\cos(2x) - \cos(2y)}$

cont'd.

13.10 cont'd. Therefore

$$\cot\left(\frac{\pi}{2n} + \frac{\phi-\phi'}{2n}\right) + \cot\left(\frac{\pi}{2n} - \frac{\phi-\phi'}{2n}\right) = -\frac{2 \sin\left(\frac{\pi}{n}\right)}{\cos\left(\frac{\pi}{n}\right) - \cos\left(\frac{\phi-\phi'}{n}\right)}$$

$$\text{Thus } F_1(p_p) = -\frac{e^{-j\pi/4}}{n\sqrt{2\pi\beta}} \frac{\sin\left(\frac{\pi}{n}\right)}{\cos\left(\frac{\pi}{n}\right) - \cos\left(\frac{\phi-\phi'}{n}\right)} \frac{e^{-jp_p}}{\sqrt{\beta}}$$

13.11 Following the procedure used to evaluate (13-48a) along $SDP_{\pm\pi}$ and along C_T in order to derive, respectively, (13-61) and (13-58) we can show that by simply replacing $\phi-\phi'$ by $\phi+\phi'$ everywhere in that derivation we arrive at (13-62a) when we evaluate (13-48b) along C_T and at (13-62b) when we evaluate (13-48b) along $SDP_{\pm\pi}$.

13.12(a) For $p \geq p'$ the Green's function corresponding to (13-38) is given by (11-208b) which for the geometry of Figure 13-13(b) can be written as

$$G(p, p'; \phi, \phi') = \frac{1}{n} \sum_{m=0}^{\infty} \epsilon_m J_m(p_p) H_m^{(2)}(ap) \{ \cos\left[\frac{m}{n}(\phi-\phi')\right] \pm \cos\left[\frac{m}{n}(\phi+\phi')\right] \}$$

$$\text{where } \epsilon_m = \begin{cases} 1 & m=0 \\ 2 & m \neq 0 \end{cases}$$

It is obvious that the above Green's function is identical to that of (13-38) except that p and p' have been interchanged. Therefore for $p \gg p'$ the Green's function is identical to (13-40) and (13-40a) but with p and p' interchanged.

- (b) The same reasoning as in Part a is used to justify that the geometrical optics for $p \gg p'$ are those given by (13-65) but with p and p' interchanged.
- (c) Similarly the diffracted fields for $p \gg p'$ are identical to of (13-67) but with p and p' interchanged.

The answers to all three parts above also justified by the application of the reciprocity principle as demonstrated graphically in Figure 13-17.

[13.13] Because the ISB and RSB are far away from the observation angle, we can use Keller's forms to compute the diffracted fields.

$$(a) E_{G0}^i = e^{j\beta p \cos(\phi - \theta)} = e^{j\frac{2\pi}{\lambda}(5\lambda) \cos(180^\circ - 45^\circ)} = e^{j10\pi \cos(135^\circ)} = -0.9752 + j0.2214$$

(b) $E_{G0}^r = 0$ because the reflected GO field does not exist at $\phi = 180^\circ$.

$$(c) E_{G0}^t = E_{G0}^i - E_{G0}^r = -0.9752 + j0.2214$$

$$(d) E_D^i = \frac{e^{-j(\beta p + \pi/4)}}{\sqrt{2\pi\beta p}} \frac{1}{n} \sin\left(\frac{\pi}{n}\right) \left\{ \frac{1}{\cos\left(\frac{\pi}{n}\right) - \cos\left(\frac{\phi-\theta}{n}\right)} \right\}_{\substack{n=2 \\ \phi=180^\circ \\ \theta=45^\circ}} = \frac{e^{-j\left(\frac{\pi}{3}\lambda + \pi/4\right)}}{2\sqrt{2\pi\left(\frac{\pi}{3}\lambda\right)}} \left\{ -\frac{1}{\cos\left(\frac{135}{2}\right)} \right\} = -\frac{e^{-j(10\pi + \pi/4)}}{4\pi\sqrt{5}(0.3827)} = -\frac{e^{-j10.25\pi}}{10.753}$$

$$E_D^i = -0.093(0.707 - j0.707) = -0.0657 + j0.0657$$

$$(e) E_D^r = \frac{e^{-j(\beta p + \pi/4)}}{\sqrt{2\pi\beta p}} \frac{1}{n} \sin\left(\frac{\pi}{n}\right) \left\{ \frac{1}{\cos\left(\frac{\pi}{n}\right) - \cos\left(\frac{\phi+\theta}{n}\right)} \right\}_{\substack{n=2 \\ \phi=180^\circ \\ \theta=45^\circ}} =$$

$$E_D^r = \frac{e^{-j(10\pi + \pi/4)}}{4\pi\sqrt{5}} \left\{ -\frac{1}{\cos\left(\frac{225}{2}\right)} \right\} = \frac{e^{-j(10\pi + \pi/4)}}{4\pi\sqrt{5}(0.3827)} = 0.0657 - j0.0657$$

$$(f) E_D^t = E_D^i - E_D^r = 2E_D^i = -0.1314 + j0.1314$$

$$(g) E^t = E_{G0}^t + E_D^t = (-0.9752 + j0.2214) + (-0.1314 + j0.1314) = -1.1066 + j0.3568$$

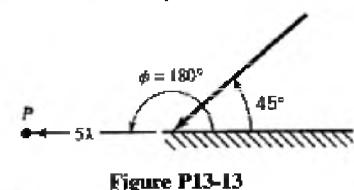


Figure P13-13

[13.14] For the hard polarization only the total diffracted field and the total field are different from those of the soft polarization. From Problem 13.13

$$(a) E_{G0}^i = -0.9752 + j0.2214$$

$$(b) E_{G0}^r = 0$$

$$(c) E_{G0}^t = E_{G0}^i + E_{G0}^r = -0.9752 + j0.2214$$

$$(d) E_D^i = -0.0657 + j0.0657$$

$$(e) E_D^r = 0.0657 - j0.0657$$

$$(f) E_D^t = E_D^i + E_D^r = 0$$

$$(g) E^t = E_{G0}^t + E_D^t = -0.9752 + j0.2214$$

13.15 The solution of this problem can be obtained by referring to the reciprocity principle of Figure 13-17 and to the solutions of Problems 13.12 and 13.13. Doing this, we can write the following:

(a) $E_{G0}^L = -0.9752 + j0.2214$

(b) $E_{G0}^R = 0$

(c) $E_{G0}^t = E_{G0}^L - E_{G0}^R = -0.9752 + j0.2214$

(d) $E_D^L = -0.0657 + j0.0657$

(e) $E_D^R = 0.0657 - j0.0657$

(f) $E_D^t = E_D^L - E_D^R = 2E_D^L = -0.1314 + j0.1314$

(g) $E^t = E_{G0}^t + E_D^t = -1.1066 + j0.3568$

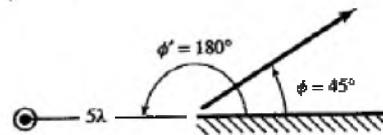


Figure P13-15

13.16 The solution of this problem can be obtained by referring to the reciprocity principle of Figure 13-17 and to the solutions of Problems 13.12 and 13.14. Doing this, we can write the following:

(a) $E_{G0}^L = -0.9752 + j0.2214$

(b) $E_{G0}^R = 0$

(c) $E_{G0}^t = E_{G0}^L + E_{G0}^R = -0.9752 + j0.2214$

(d) $E_D^L = -0.0657 + j0.0657$

(e) $E_D^R = 0.0657 - j0.0657$

(f) $E_D^t = E_D^L + E_D^R = 0$

(g) $E^t = E_{G0}^t + E_D^t = -0.9752 + j0.2214$

13.17 $\underline{E} = \hat{a}_z E^L$ (soft polarization)

Reflection Shadow Boundary (RSB):

$$\text{RSB: } \pi - \phi' = 180^\circ - 60^\circ = 120^\circ$$

$$(a) \underline{E}_{\text{go}}^L(\phi=120^\circ^-) = \hat{a}_z E^L e^{j\beta p \cos(\phi-\phi')} \Big|_{\begin{array}{l} \phi=120^\circ \\ \phi'=55^\circ, \phi=60^\circ \end{array}}$$

$$= \hat{a}_z E^L e^{j\frac{\pi}{3}(55)} \cos(120-60) = \hat{a}_z E^L e^{j\frac{11\pi}{6}} \cos(60^\circ) = \hat{a}_z E^L e^{j\frac{11\pi}{6}(0.5)}$$

$$\underline{E}_{\text{go}}^L(\phi=120^\circ) = \hat{a}_z E^L e^{j5.5\pi} = \hat{a}_z E^L e^{j3\pi/2} = -j \hat{a}_z E^L$$

$$\underline{E}_{\text{go}}^L(\phi=120^\circ^+) = \underline{E}_{\text{go}}^L(\phi=120^\circ^-) = -j \hat{a}_z E^L$$

$$(b) \underline{E}_{\text{go}}^r(\phi=120^\circ^-) = \hat{a}_z E^L e^{j\beta p \cos(\phi+\phi')} = \hat{a}_z E^L e^{j\frac{11\pi}{6} \cos(120+60^\circ)} = \hat{a}_z E^L e^{j\frac{11\pi}{6} \cos(\pi)}$$

$$= \hat{a}_z E^L e^{-j\frac{11\pi}{6}} = \hat{a}_z E^L e^{-j\pi} = -\hat{a}_z E^L$$

$$\underline{E}_{\text{go}}^r(\phi=120^\circ^+) = 0$$

$$(c) \underline{E}_D^L(\phi=120^\circ^-) = \text{small} \approx 0 \text{ because far away from ISB} \quad \left. \begin{array}{l} \text{continuous} \\ \text{across} \\ \text{RSB} \end{array} \right\}$$

$$\underline{E}_D^L(\phi=120^\circ^+) = \text{small} \approx 0 \text{ because far away from ISB}$$

$$(d) \underline{E}_D^r(\phi=120^\circ^-) = \hat{a}_z E^L \left[-0.5 e^{-j\beta p} \text{sgn}(2) \right] \Big|_{\epsilon > 0} = \hat{a}_z E^L \left[-0.5 e^{-j\frac{3\pi}{8}(55)} \right]$$

$$= \hat{a}_z E^L \left(-0.5 e^{-j\frac{11\pi}{8}} \right) = \hat{a}_z E^L \left(-0.5 e^{-j\pi} \right) = +\hat{a}_z E^L (0.5)$$

$$\underline{E}_D^r(\phi=120^\circ^+) = 0.5 E^L \hat{a}_z$$

$$\underline{E}_D^r(\phi=120^\circ^+) = \hat{a}_z E^L \left[-0.5 e^{-j\beta p} \text{sgn}(-2) \right] \Big|_{\epsilon < 0} = -0.5 E^L \hat{a}_z$$

$$(e) \underline{E}^{\text{total}}(\phi=120^\circ^-) = \underline{E}_{\text{go}}^L(\phi=120^\circ^-) + \underline{E}_{\text{go}}^r(\phi=120^\circ^-) + \underline{E}_D^L(\phi=120^\circ^-) + \underline{E}_D^r(\phi=120^\circ^-)$$

$$= -j \hat{a}_z E^L - \hat{a}_z E^L + 0.5 E^L \hat{a}_z = -\hat{a}_z (0.5 + j) E^L$$

$$\underline{E}^{\text{total}}(\phi=120^\circ^+) = \underline{E}_{\text{go}}^L(\phi=120^\circ^+) + \underline{E}_{\text{go}}^r(\phi=120^\circ^+) + \underline{E}_D^L(\phi=120^\circ^+) + \underline{E}_D^r(\phi=120^\circ^+)$$

$$= j \hat{a}_z E^L + 0 + 0 + (-0.5 E^L \hat{a}_z) = -\hat{a}_z (0.5 + j) E^L$$

which shows continuity of the field across the RSB ($\phi=120^\circ$).

A similar procedure can be used to compute the fields and illustrate the continuity of them at the ISB ($\phi=240^\circ$).

- * A complete set of linear and polar amplitude patterns follow, assuming $E^L = 1$.

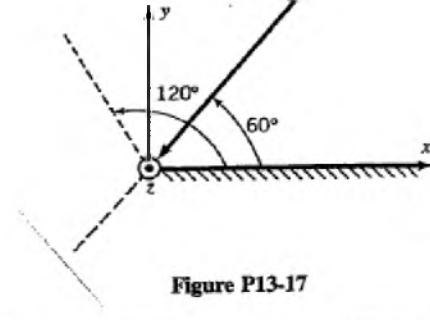
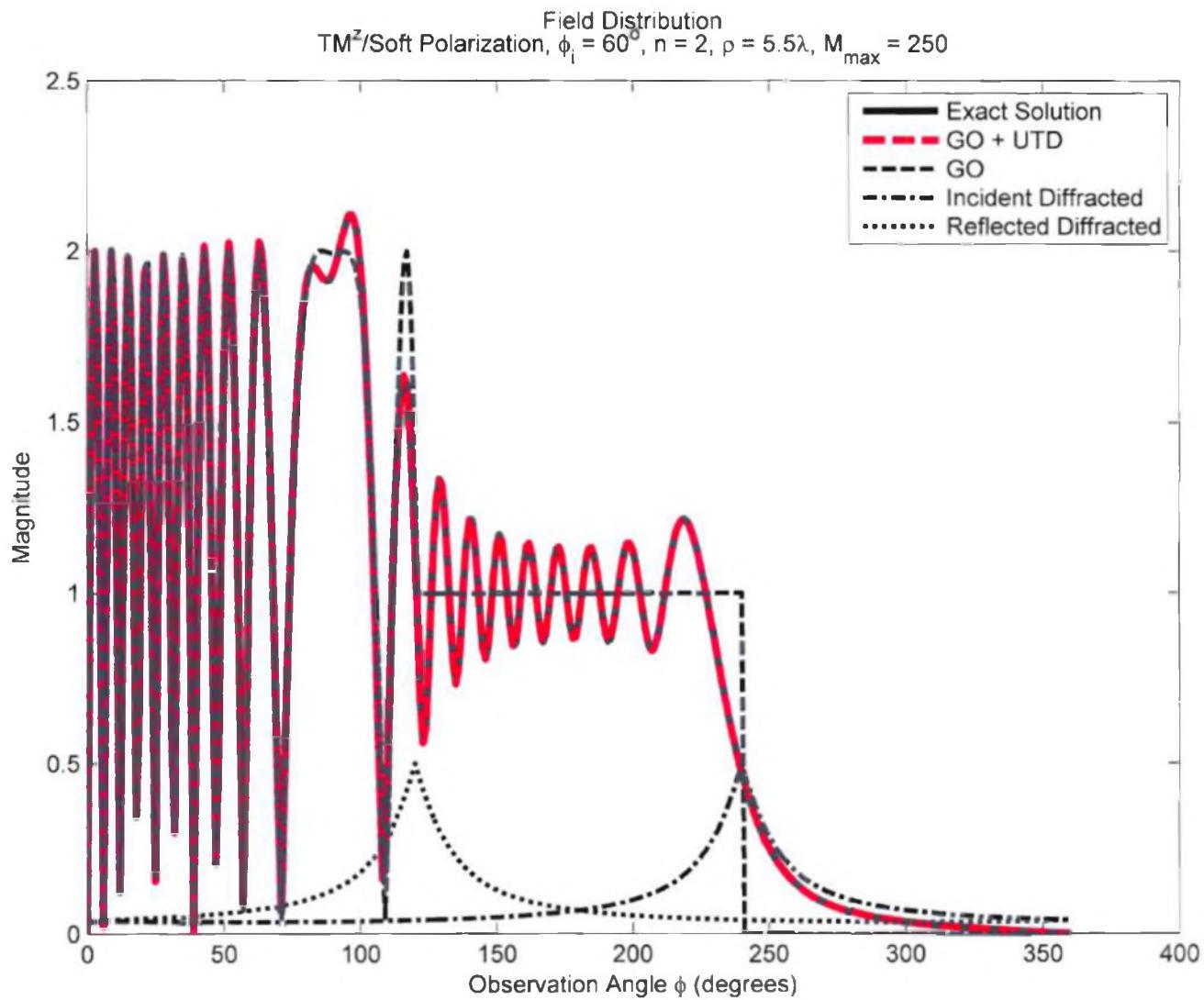


Figure P13-17

13.17 cont'd

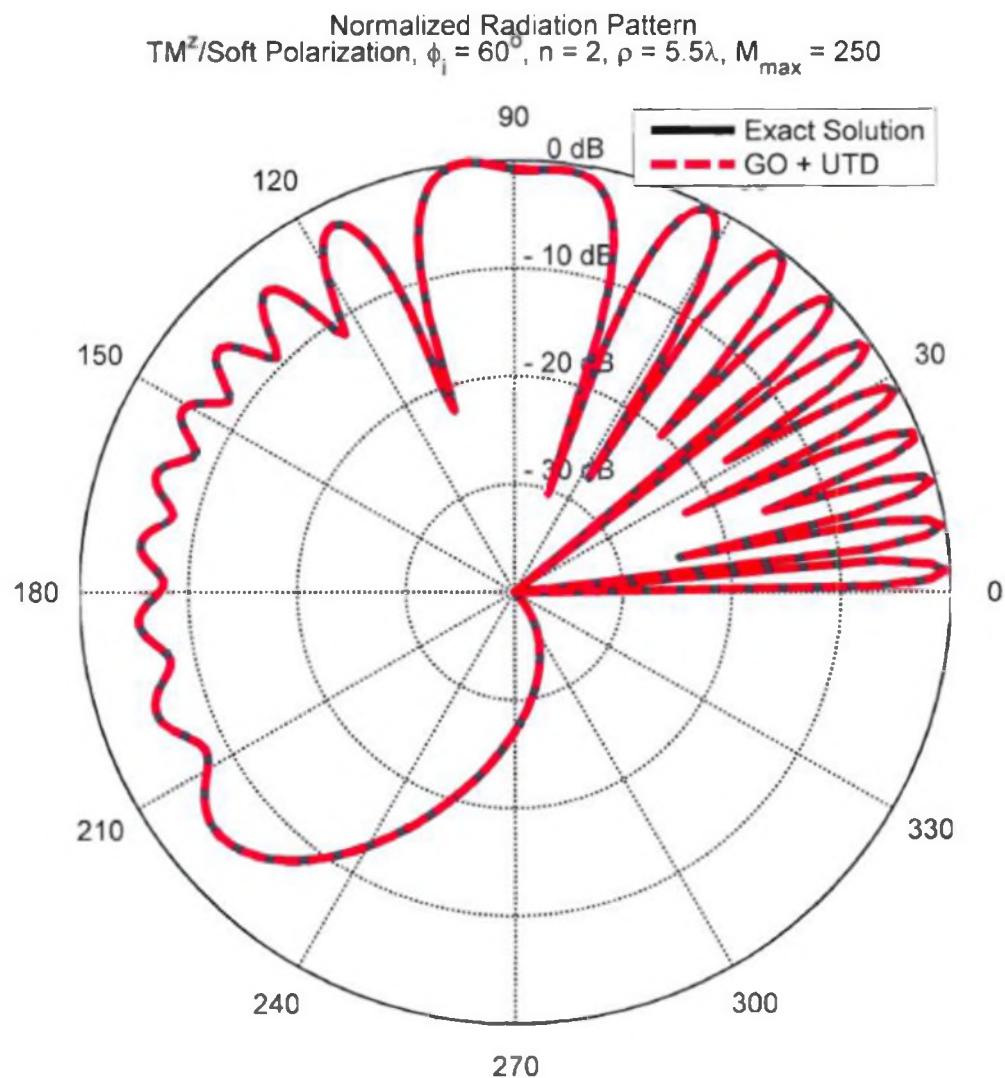
TM^z - Soft Polarization



Cont'd

13.17 cont'd

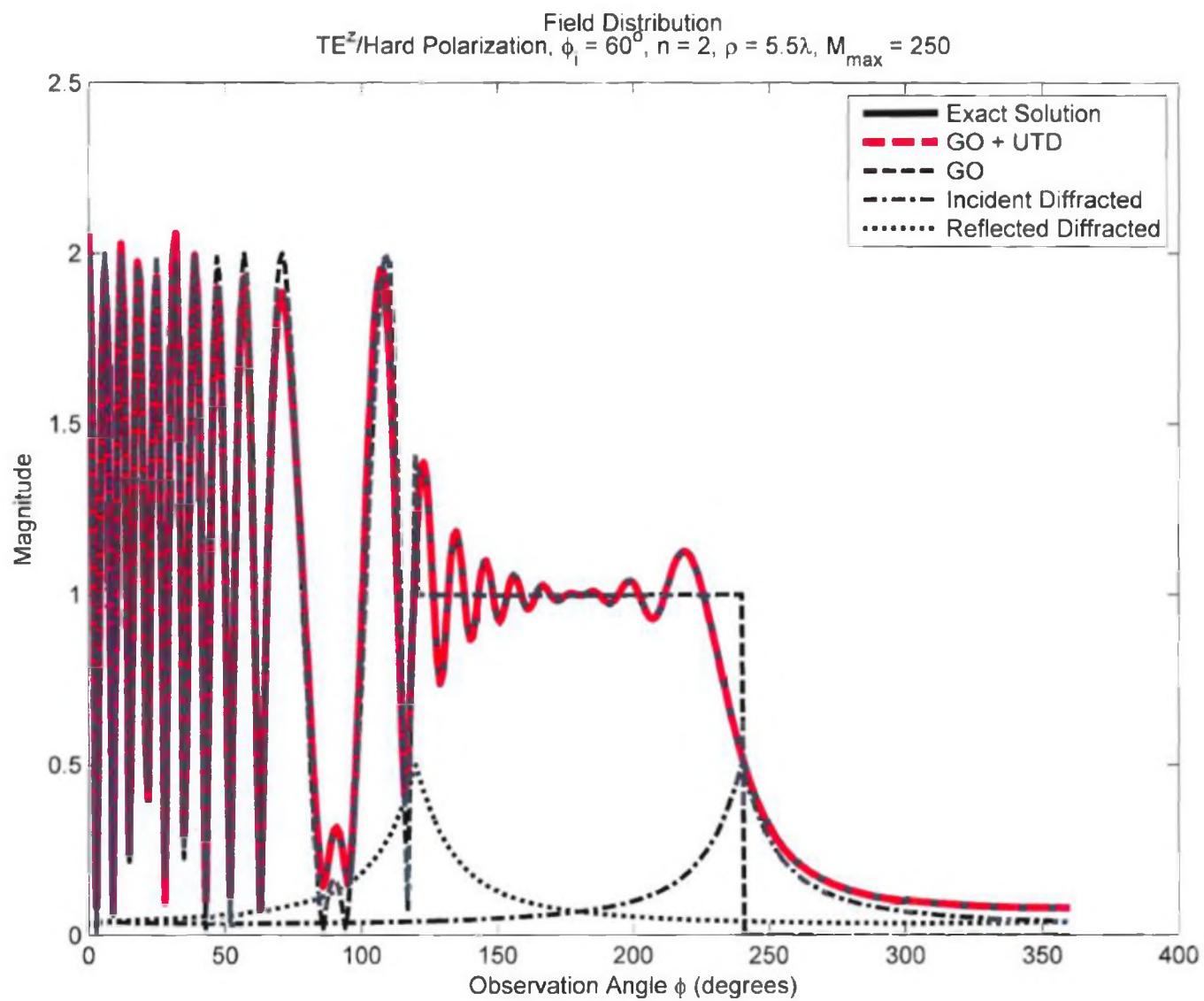
TM^z- Soft Polarization



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13.17 cont'd

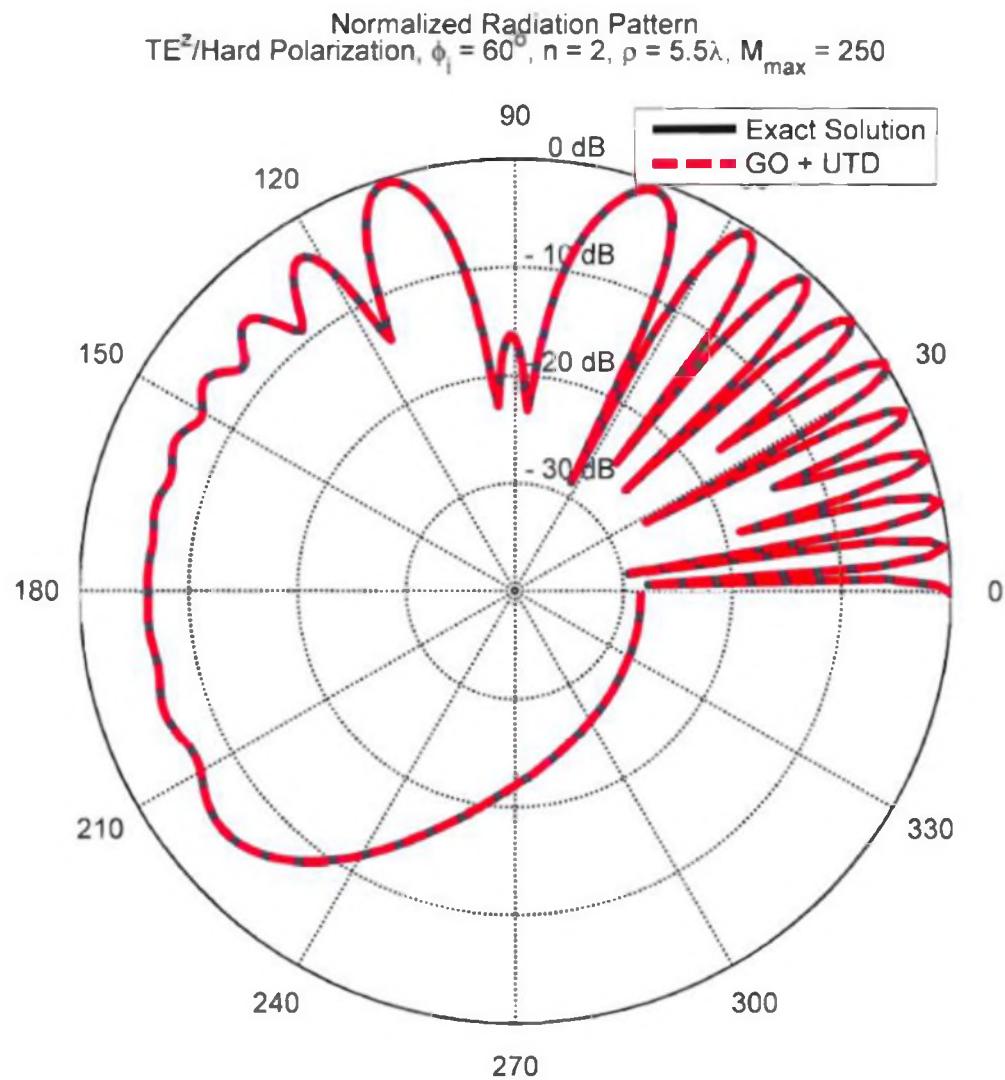
TE^z - Hard Polarization



Cont'd

13.17 cont'd

TE^z-Hard Polarization



$$13.18 \quad V_B^L = \frac{e^{-j(\pi/4 + \beta p)}}{\sqrt{2\pi\beta p}} \frac{\frac{1}{n} \sin(\frac{\pi}{n})}{\cos(\frac{\pi}{n}) - \cos(\frac{\phi - \phi'}{n})}$$

$$\eta = 3/2, \phi' = 60^\circ, \phi = 180^\circ, p = 81\lambda$$

$$(a) V_B^L = \frac{e^{-j(\pi/4 + \frac{3\pi}{2} \cdot 81\lambda)}}{\sqrt{2\pi(\frac{3\pi}{2})81\lambda}} \frac{\frac{2}{3} \sin(\frac{2}{3}\pi)}{\cos(\frac{2}{3}\pi) - \cos[\frac{2}{3}(180 - 60)]}$$

$$= \frac{2e^{-j\pi/4}}{3(2\pi)(9)} \frac{\sin(120^\circ)}{\cos(120^\circ) - \cos(120^\circ)} = \frac{e^{-j\pi/4}}{27\pi} \frac{0.866}{-0.5 - 0.17365} = \frac{e^{-j\pi/4}}{27\pi} (-1.28554)$$

$$\boxed{V_B^L = -0.01516(0.707 - j0.707) = -0.01071(1-j)} \quad GTD$$

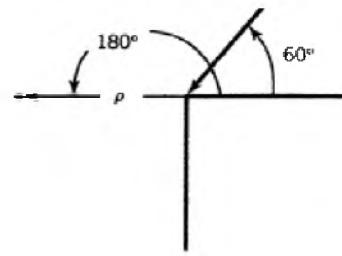


Figure P13-18

$$(b) V_B^L = \frac{e^{-j\beta p}}{\sqrt{p}} \left\{ -\frac{e^{j\pi/4}}{2n\sqrt{2\pi\beta}} \left[\cot\left[\frac{\pi + (\phi - \phi')}{2n}\right] F[\beta pg^-(\phi - \phi')] + \cot\left[\frac{\pi - (\phi - \phi')}{2n}\right] F[\beta pg^+(\phi - \phi')] \right] \right\}$$

$$\approx -\frac{e^{-j(\beta p + \pi/4)}}{2n\sqrt{2\pi\beta p}} \left\{ \cot\left[\frac{\pi + (\phi - \phi')}{2n}\right] + \cot\left[\frac{\pi - (\phi - \phi')}{2n}\right] \right\}$$

$$\cot\left[\frac{\pi + (\phi - \phi')}{2n}\right] = \cot\left[\frac{180^\circ + (180^\circ - 60^\circ)}{2(3/2)}\right] = \cot\left(\frac{300}{3}\right) = \cot(100^\circ) = -0.17633$$

$$\cot\left[\frac{\pi - (\phi - \phi')}{2n}\right] = \cot\left[\frac{180^\circ - (180^\circ - 60^\circ)}{2(3/2)}\right] = \cot\left(\frac{60}{3}\right) = \cot(20^\circ) = 2.74748$$

$$V_B^L = -\frac{e^{-j\left[\frac{\pi}{3}(81\lambda) + \frac{\pi}{4}\right]}}{2(3/2)\sqrt{2\pi(\frac{3\pi}{2})81\lambda}} \left[-0.17633 + 2.74748 \right] = -\frac{e^{-j\pi/4}}{3(2\pi)(9)} [2.5716] = -0.01516 e^{-j\pi/4}$$

$$\boxed{V_B^L = -0.01516(0.707 - j0.707) = -0.01072(1-j)} \quad UTG$$

$\therefore V_B^L(GTD) = V_B^L(UTG)$ because observations are made away from the ISB and RSB.

$$\frac{\pi}{3} = \phi - \phi' = 180^\circ - 60^\circ = 120^\circ = \frac{2\pi}{3}, n = 1.5 \Rightarrow \begin{cases} \text{Using Fig. 13-19(a)} \Rightarrow N^- = 0 \\ g^- = 1 + \cos\left(\frac{\pi}{3} - 2\pi n N^-\right) = 1 + \cos(120^\circ) = 0.5 \end{cases}$$

$$\begin{cases} \text{Using Fig. 13-19(b)} \Rightarrow N^+ = 1 \\ g^+ = 1 + \cos\left(\frac{\pi}{3} - 2\pi n N^+\right) = 1 + \cos\left(\frac{2\pi}{3} - 2\pi \frac{3}{2}\right) \\ g^+ = 1 + \cos\left(\frac{8\pi}{3} - 3\pi\right) = 1 + \cos\left(-\frac{\pi}{3}\right) \approx 1 + 0.5 \\ g^+ = 1.5 \end{cases}$$

$$|\beta pg^-| = \frac{2\pi}{3}(81\lambda)(0.5) = 81\pi \gg 1$$

$$g^- = 0.5 \quad \text{Thus } F(\beta pg^-) = F(81\pi) \approx 1$$

$$|\beta pg^+| = \frac{2\pi}{3}(81\lambda)(1.5) = 243\pi \gg 1$$

$$g^+ = 1.5 \quad \text{Thus } F(\beta pg^+) = F(243\pi) \approx 1$$

Because observations are made far away from ISB

13.19

(a) $\underline{H}^i = \hat{a}_z e^{-j\beta x}$ Half plane diffraction ($n=2$)

(b) $\underline{E}^i = \hat{a}_y e^{-j\beta x}$

(c) $\underline{H}^d = \hat{a}_z (1) D^h (s, \phi, \phi', n=2) \frac{1}{\sqrt{s}} e^{-j\beta s}$

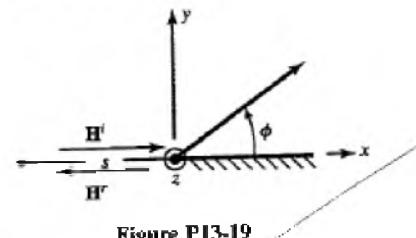


Figure P13-19

$$D_1^h = \frac{e^{-j\pi/4}}{\sqrt{2\pi\beta}} \frac{1}{n} \sin\left(\frac{\pi}{n}\right) \left[\frac{1}{\cos\left(\frac{\pi}{n}\right) - \cos\left(\frac{\phi-\phi'}{n}\right)} + \frac{1}{\cos\left(\frac{\pi}{n}\right) - \cos\left(\frac{\phi+\phi'}{n}\right)} \right]$$

$$= \frac{e^{-j\pi/4} \sin\left(\frac{\pi}{2}\right)}{2\sqrt{2\pi\beta}} \left[\frac{1}{\cos\left(\frac{\pi}{2}\right) - \cos\left(\frac{0}{2}\right)} + \frac{1}{\cos\left(\frac{\pi}{2}\right) - \cos\left(\frac{2\pi}{2}\right)} \right] \stackrel{n=2}{\phi=\phi'=180^\circ}$$

$$D_1^h = \frac{e^{-j\pi/4}}{2\sqrt{2\pi\beta}} [-1+1] = 0$$

$$\underline{H}^d = 0$$

(d) $\underline{E}^d = 0$

(e) $\left| \frac{\underline{H}^d}{\underline{H}^i} \right| = \left| \frac{0}{1} \right| = 0 = 10 \log_{10}(0) = -\infty \text{ dB}$

13.20

(a) $\underline{E}_{40}^i = e^{j\beta p' \omega s (\phi - \phi')} e^{j\beta p \cos(45^\circ - 180^\circ)}$

$$= e^{j\frac{2\pi}{\lambda}(5\lambda) \cos(-135^\circ)} e^{j10\pi \cos(-135^\circ)}$$

$$\underline{E}_{40}^i = e^{j10\pi (-0.707)} = e^{-j\frac{7}{2}\pi} = -0.975 + j0.2214$$

(b) $\underline{E}_{40}^r = 0$ Because of reciprocity:

Also because of RSBs:

$$\phi' = \pi - \frac{(k-n)\pi}{2} \Big|_{n=2} = \pi - 0 = \pi$$

$$\phi = \pi - \phi' = \pi - \pi = 0$$

See Fig. 13-24(b)

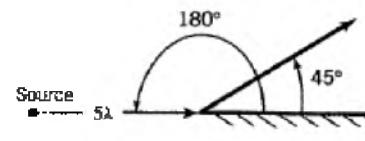
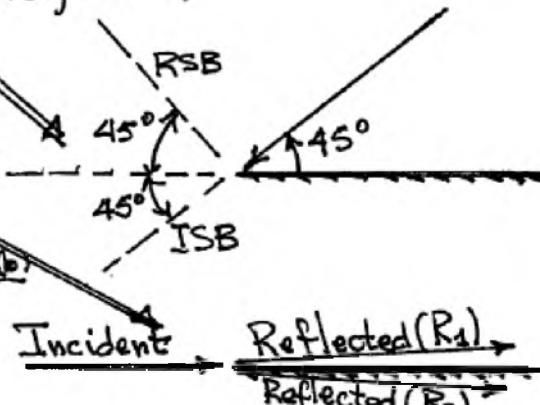


Figure P13-20

(c)

$$\underline{E}_{40}^t = \underline{E}_{40}^i + \underline{E}_{40}^r = \underline{E}_{40}^i + 0 = \underline{E}_{40}^i$$

$$= -0.975 + j0.2214$$



13.20 cont'd

$$(d) E_D^i = \frac{e^{-j(\beta p + \pi/4)}}{\sqrt{2\pi\beta p}} \left| \frac{\frac{1}{n} \sin\left(\frac{\pi}{n}\right)}{\cos\left(\frac{\pi}{n}\right) - \cos\left(\frac{\phi-\psi}{n}\right)} \right| = \frac{e^{-j\left(\frac{2\pi s}{\lambda} + \frac{\pi}{4}\right)}}{\sqrt{2\pi\left(\frac{2\pi}{\lambda}\right)(s)}} \frac{\frac{1}{2} \sin\left(\frac{\pi}{2}\right)}{\cos\left(\frac{\pi}{2}\right) - \cos\left(45^\circ - 180^\circ\right)}$$

$$= \frac{e^{-j(10\pi + \pi/4)}}{2\pi\sqrt{s}(2)} \frac{1}{-\cos(135^\circ/2)} = -\frac{e^{-j\pi/4}}{4\pi\sqrt{s}} \frac{1}{\cos(-67.5^\circ)} = -\frac{e^{-j\pi/4}}{4\pi\sqrt{s}(0.38268)}$$

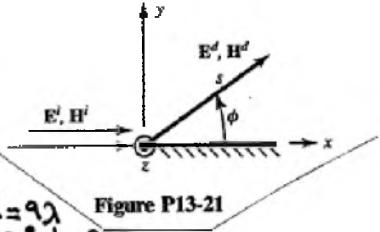
$$E_D^i = 0.06575(-1+j)$$

$$(e) E_D^r = \frac{e^{j(\beta p + \pi/4)}}{\sqrt{2\pi\beta p}} \left| \frac{\frac{1}{n} \sin\left(\frac{\pi}{n}\right)}{\cos\left(\frac{\pi}{n}\right) - \cos\left(\frac{\phi+\psi}{n}\right)} \right| = \frac{e^{j\pi/4}}{28.0993 \cos(112.5^\circ)} = 0.06575(1-j)$$

$$(f) E_D^t = E_D^i - E_D^r = 0.06575(-1+j) - 0.06575(1-j) = 2[0.06575(-1+j)]$$

$$E_D^t = 0.1315(-1+j) = -0.1315(1-j)$$

$$(g) E^t = E_{go}^t + E_D^t = (-0.975 + j0.2214) + (-0.1315 + j0.1315) \\ = -1.1065 + j0.3529$$



13.21

$$(a) E_D = \hat{a}_z \frac{e^{-j(\beta p + \pi/4)} \left\{ \frac{1}{n} \sin\left(\frac{\pi}{n}\right) \right.}{\sqrt{2\pi\beta p}} \left. - \frac{1}{n} \sin\left(\frac{\pi}{n}\right) \right\}_{n=2} \left. \begin{array}{l} \rho=s=9.7 \\ \phi=180^\circ, \psi=90^\circ \end{array} \right.$$

$$= \hat{a}_z \frac{e^{-j(\beta s + \pi/4)} \frac{1}{2} \sin\left(\frac{\pi}{2}\right)}{\sqrt{2\pi\beta s}} \left\{ -\frac{1}{\cos\left(\frac{90-180}{2}\right)} + \frac{1}{\cos\left(\frac{90+180}{2}\right)} \right\}$$

$$= \hat{a}_z \frac{e^{-j\left(\frac{2\pi s}{\lambda} + \frac{\pi}{4}\right)}}{2\sqrt{2\pi\left(\frac{2\pi}{\lambda}\right)(s)}} \left\{ -\frac{1}{\cos(-45^\circ)} + \frac{1}{\cos(135^\circ)} \right\} = \hat{a}_z \frac{e^{-j\pi/4}}{2(2\pi)(3)} \left\{ -\frac{1}{0.707} - \frac{1}{0.707} \right\}$$

$$E_D = -\hat{a}_z \frac{e^{-j\pi/4}(2)}{2(2\pi)(3)(0.707)} = -\hat{a}_z \frac{0.707(-1-j)}{2\pi(3)(0.707)} = -\hat{a}_z \frac{1}{6\pi}(1-j) = -\hat{a}_z 0.053(1-j)$$

$$(b) H_D = \hat{a}_z \frac{e^{-j(\beta p + \pi/4)} \left\{ \frac{1}{n} \sin\left(\frac{\pi}{n}\right) \right.}{\sqrt{2\pi\beta p}} \left. + \frac{1}{n} \sin\left(\frac{\pi}{n}\right) \right\}_{n=2} \left. \begin{array}{l} \rho=s=9.7 \\ \phi=180^\circ, \psi=90^\circ \end{array} \right.$$

Based on the calculation of Part (a), by changing the minus(-) sign within the brackets to a plus(+), we can write the diffracted magnetic field as

$$\underline{H}_D = \hat{a}_z \frac{e^{-j\pi/4}}{2(2\pi)(s)} \left\{ -\frac{1}{0.707} + \frac{1}{0.707} \right\} = 0$$

[13.22] (a) $\underline{E}^L|_{\phi=225^-} = \hat{a}_z e^{j\beta\rho \cos(\phi-\theta)} = \hat{a}_z e^{j\frac{\pi}{3}\sin(60^\circ)(225^\circ-45^\circ)} = \hat{a}_z e^{-j\pi} = -\hat{a}_z$

(b) $\underline{E}^L|_{\phi=225^+} = 0$

(c) $\underline{E}^L|_{\phi=225^-} \approx \hat{a}_z [-0.5 e^{-j\beta\rho} \operatorname{sgn}(\epsilon)] = \hat{a}_z (-0.5 e^{-j\pi} \operatorname{sgn}\epsilon) = \hat{a}_z (0.5), \epsilon > 0$

(d) $\underline{E}^L|_{\phi=225^+} \approx \hat{a}_z [-0.5 e^{j\beta\rho} \operatorname{sgn}(\epsilon)] = \hat{a}_z (-0.5 e^{j\pi} \operatorname{sgn}\epsilon) = -\hat{a}_z (0.5), \epsilon < 0$

(e) $\underline{E}^t = \underline{E}_{G0}^t + \underline{E}_D^t = \underline{E}_{G0}^t - \underline{E}_D^t \approx \underline{E}_{G0}^t + \underline{E}_D^t = \hat{a}_z (-1+0.5) = -\hat{a}_z (0.5)$

(f) $\underline{E}^t|_{\phi=225^+} = \underline{E}_{G0}^t + \underline{E}_D^t = \underline{E}_{G0}^t + \underline{E}_D^t - \underline{E}_D^t \stackrel{\sigma=\infty}{=} \underline{E}_{G0}^t = \hat{a}_z (0-0.5) = -\hat{a}_z (0.5)$

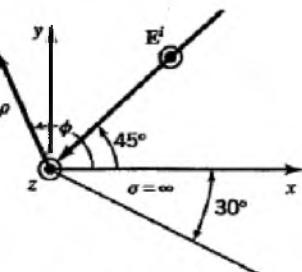


Figure P13-22

[13.23]

(a) $\underline{H}_{G0}^L|_{\phi=225^-} = \hat{a}_z e^{j\beta\rho \cos(\phi-\theta)} = \hat{a}_z e^{j\frac{\pi}{3}\sin(60^\circ)(225^\circ-45^\circ)} = \hat{a}_z e^{-j\pi} = -\hat{a}_z$

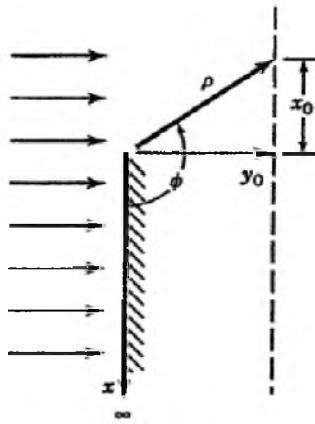
(b) $\underline{H}_{G0}^t|_{\phi=225^+} = 0$

(c) $\underline{H}_D^L|_{\phi=225^-} \approx \hat{a}_z [-0.5 e^{-j\beta\rho} \operatorname{sgn}\epsilon] = \hat{a}_z (-0.5 e^{-j\pi} \operatorname{sgn}\epsilon) = \hat{a}_z 0.5, \epsilon > 0$

(d) $\underline{H}_D^L|_{\phi=225^+} \approx \hat{a}_z [-0.5 e^{j\beta\rho} \operatorname{sgn}\epsilon] = \hat{a}_z (-0.5 e^{j\pi} \operatorname{sgn}\epsilon) = -\hat{a}_z 0.5, \epsilon < 0$

(e) $\underline{H}^t|_{\phi=225^-} = \underline{H}_{G0}^t + \underline{H}_D^t = \underline{H}_{G0}^t + \underline{H}_D^t - \underline{H}_D^t \stackrel{\sigma=\infty}{=} \underline{H}_{G0}^t + \underline{H}_D^t = \hat{a}_z (0.5-1) = -\hat{a}_z 0.5$

(f) $\underline{H}^t|_{\phi=225^+} = \underline{H}_{G0}^t + \underline{H}_D^t = \underline{H}_{G0}^t + \underline{H}_D^t - \underline{H}_D^t \stackrel{\sigma=\infty}{=} \underline{H}_{G0}^t + \underline{H}_D^t = \hat{a}_z (0-0.5) = -\hat{a}_z 0.5$



$$13.24 \quad (a) \quad E_{G0}^L = \begin{cases} e^{j\beta\rho \cos(\phi - 90^\circ)} & x_0 \geq 0 \\ 0 & x_0 < 0 \end{cases}$$

$$E_{G0}^r = 0 \quad -\infty < x < +\infty$$

Figure P13-24

$$E_D^L = V_B^L = D^L \frac{e^{-j\beta\rho}}{\sqrt{\rho}} = -\frac{e^{-j(\beta\rho + \pi/4)}}{2n\sqrt{2\pi}\beta\rho} \left\{ C^+(\xi, n=2) F[\beta\rho g^+(\xi)] + C^-(\xi, n=2) F[\beta\rho g^-(\xi)] \right\}_{n=2}$$

$$E_D^r = V_B^r = D^r \frac{e^{-j\beta\rho}}{\sqrt{\rho}} = -\frac{e^{-j(\beta\rho + \pi/4)}}{2n\sqrt{2\pi}\beta\rho} \left\{ C^+(\xi, n=2) F[\beta\rho g^+(\xi)] + C^-(\xi, n=2) F[\beta\rho g^-(\xi)] \right\}_{n=2}$$

$$\underline{E}^t = E_{G0}^L + E_D^L - E_D^r = e^{j\beta\rho \cos(\phi - 90^\circ)} + E_D^L - E_D^r, \quad x_0 \geq 0$$

$$\underline{E}^t = E_{G0}^L + E_D^L - E_D^r = 0 + E_D^L - E_D^r, \quad x_0 < 0$$

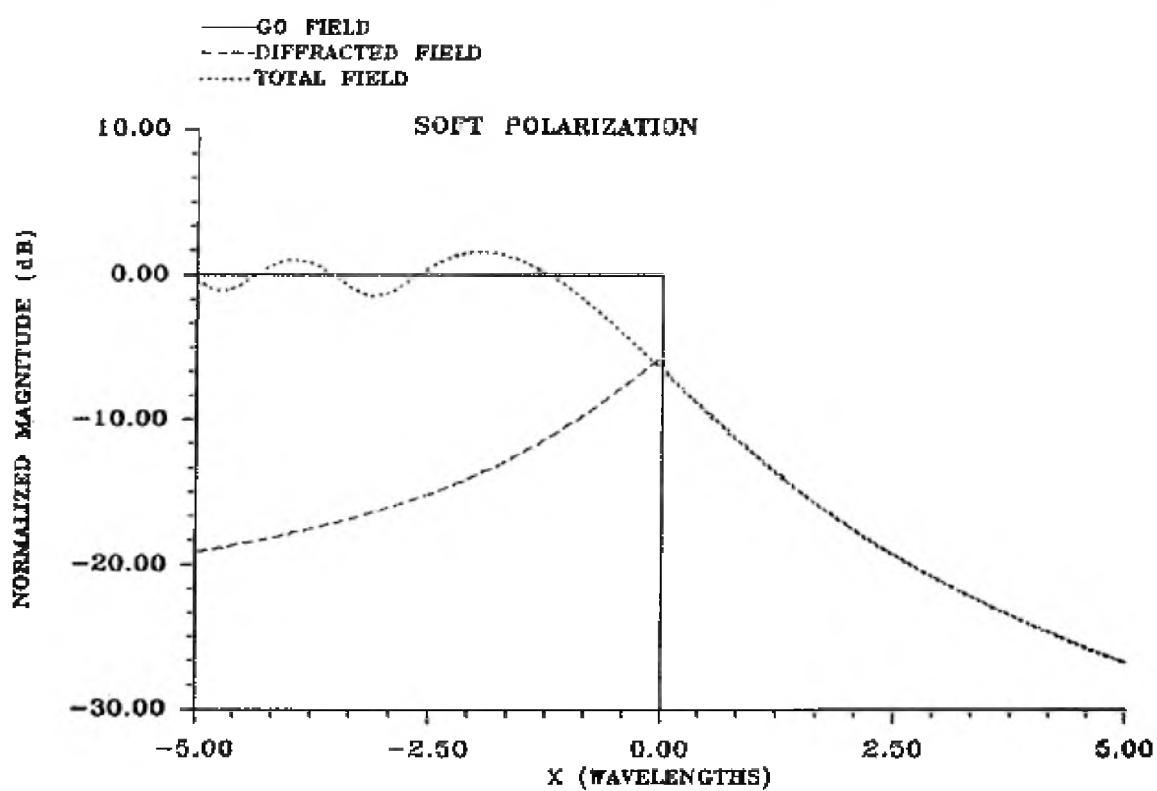
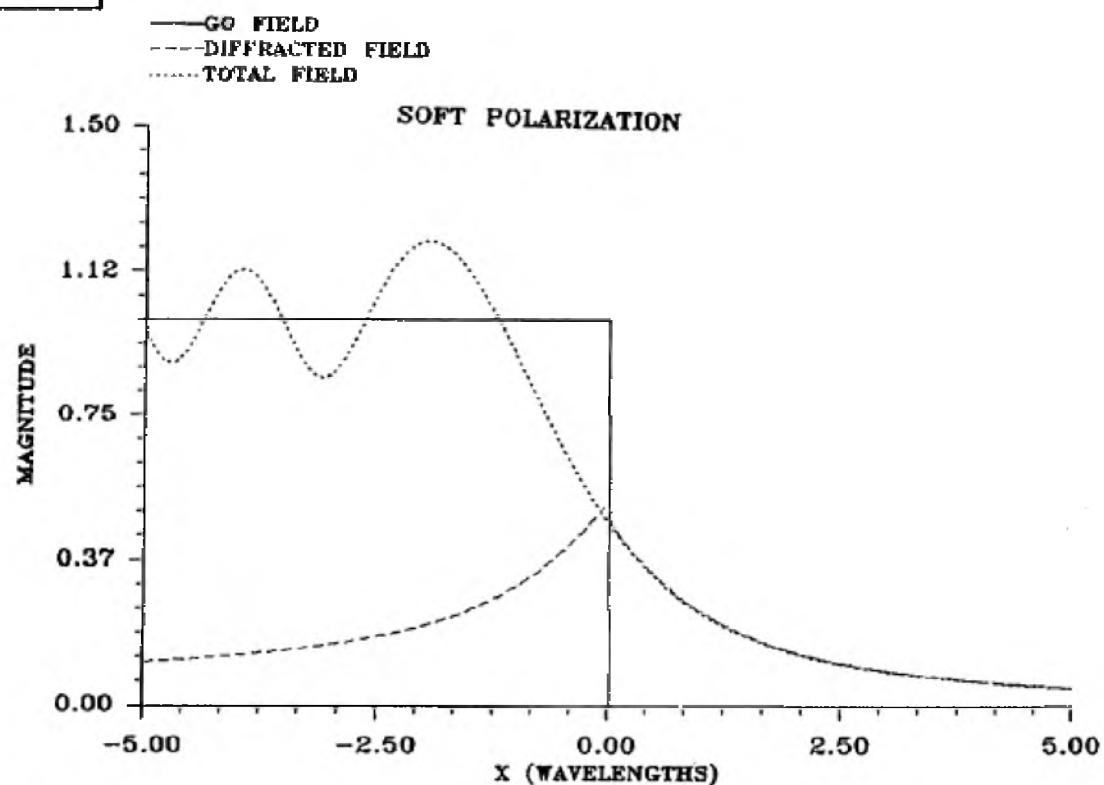
$$\text{where } \rho = \sqrt{x_0^2 + y_0^2}, \quad \phi = 90^\circ + \tan^{-1}\left(\frac{x_0}{y_0}\right)$$

(b) See Graphs on the next page.

(c) At $y_0 = 5\lambda$ and $x_0 \geq 0$ the total field $|E^t| = 0.475 \text{ V/m}$ while $|E_D^L| = 0.526 \text{ V/m}$. When $y_0 = 500\lambda$ and $x_0 = 0$, the field is $|E^t| = 0.497 \text{ V/m}$ while $|E_D^L| = 0.502 \text{ V/m}$. We see that as y_0 becomes very large, $|E^t|$ approaches the expected value of 0.5 or -6 dB.

cont'd.

13.24 cont'd.



13.25

$$(a) E_{g0}^i = \begin{cases} e^{j\beta_0 \rho \cos(\phi - 90^\circ)} & x_0 \geq 0 \\ 0 & x_0 < 0 \end{cases}$$

$$E_{g0}^r = 0 \quad -\infty \leq x_0 \leq +\infty$$

$$E_D^i = V_B^i = D^i \frac{e^{-j\beta_0 \rho}}{\sqrt{\rho}} = -\frac{e^{-j(\beta_0 \rho + \pi/4)}}{2\pi\sqrt{2\pi}\beta_0 \rho} \left\{ C^+(\xi, n=2) F[\beta_0 g^+(\xi)] + C^-(\xi, n=2) F[\beta_0 g^-(\xi)] \right\}_{n=2}$$

$$E_D^r = V_B^r = D^r \frac{e^{-j\beta_0 \rho}}{\sqrt{\rho}} = -\frac{e^{-j(\beta_0 \rho + \pi/4)}}{2\pi\sqrt{2\pi}\beta_0 \rho} \left\{ C^+(\xi^+, n=2) F[\beta_0 g^+(\xi^+)] + C^-(\xi^+, n=2) F[\beta_0 g^-(\xi^+)] \right\}_{n=2}$$

$$E^t = E_{g0}^i + E_D^i + E_D^r = e^{j\beta_0 \rho \cos(\phi - 90^\circ)} + E_D^i + E_D^r \quad , x_0 \geq 0$$

$$E^t = E_{g0}^i + E_D^i + E_D^r = 0 + E_D^i + E_D^r \quad , x_0 < 0$$

where $\rho = \sqrt{x_0^2 + y_0^2}$, $\phi = 90^\circ + \tan^{-1}\left(\frac{x_0}{y_0}\right)$

(b) See Graphs on the next page

(c) At $y_0 = 5\lambda$ and $x_0 = 0$ the total field $|E^t| = 0.526$ while

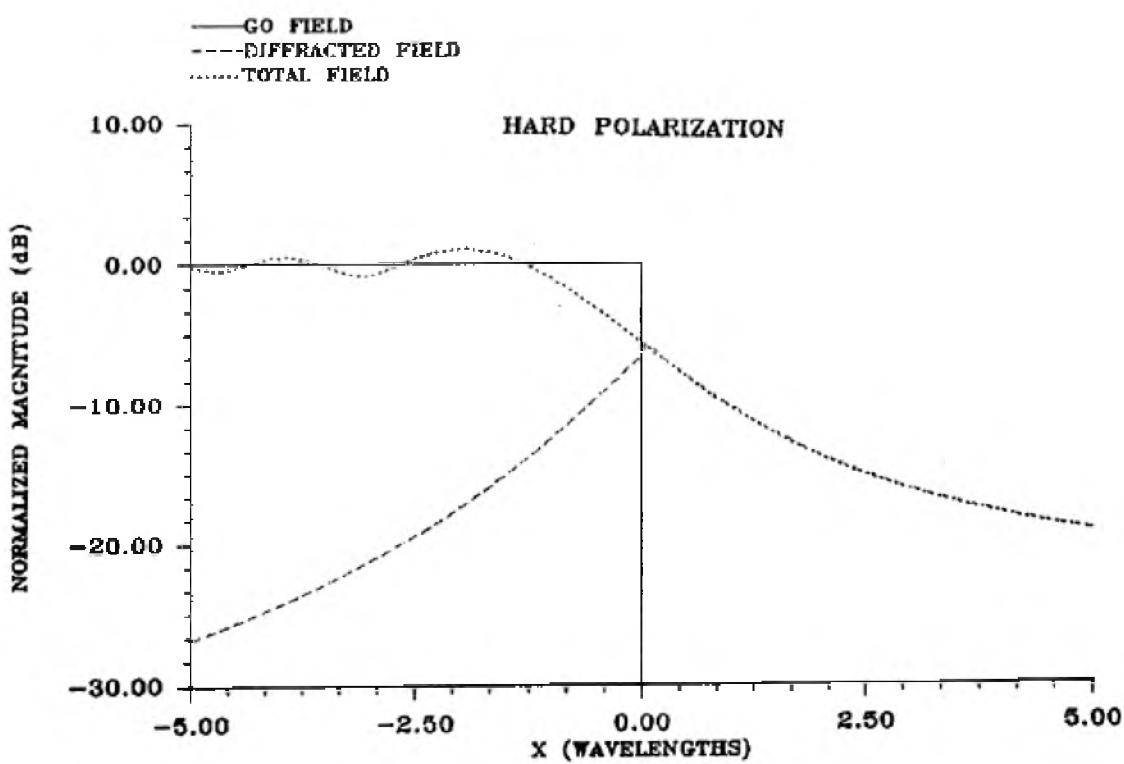
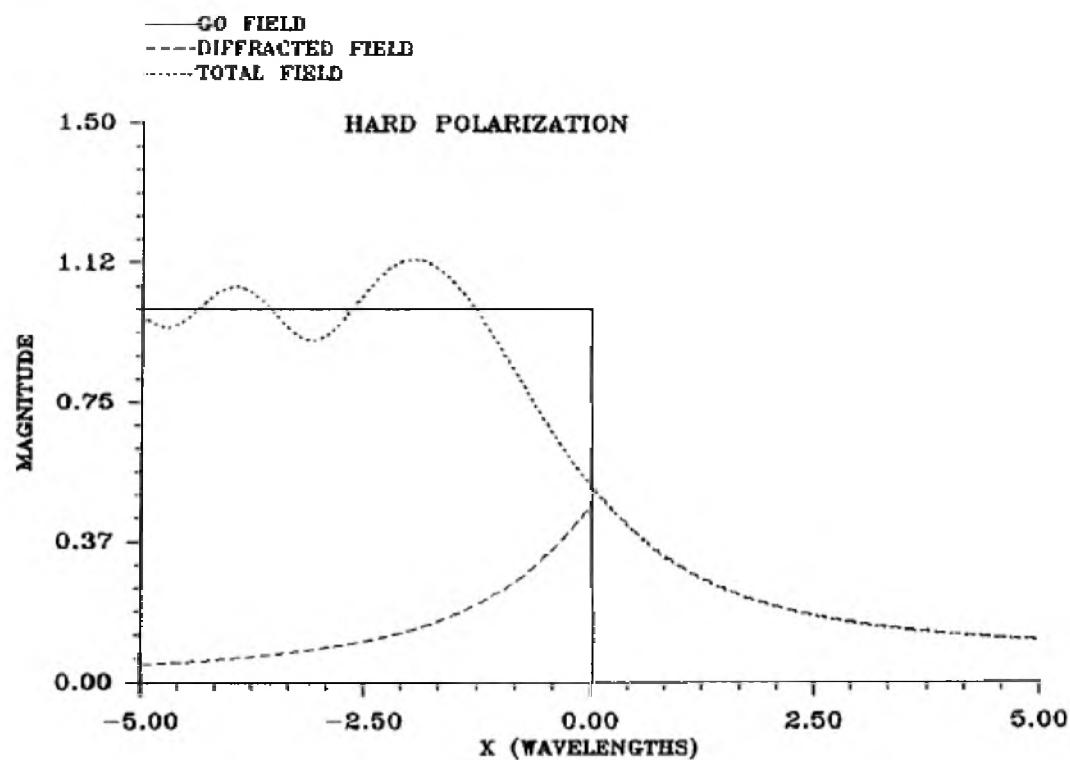
$|E_D^t| = 0.475$. When $y_0 = 500\lambda$ and $x_0 = 0$, the total field is

$|E^t| = 0.502$ while $|E_D^t| = 0.497$ V/m. We see that as y_0

becomes very large $|E^t|$ approaches the expected value of 0.5 or -6 dB.

Cont'd.

15.25 Cont'd.



13.26

(a) Away from the incident and reflected shadow boundaries we can use Keller's diffraction coefficients and functions.

$$\phi' = \frac{n\pi}{2}$$

$$0 \leq \phi \leq n\pi$$

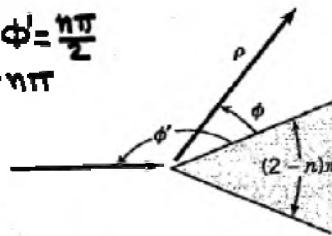


Figure P13-26

$$\phi' = \frac{n\pi}{2}$$

$$0 \leq \phi \leq n\pi$$

$$V_B^L = \frac{e^{-j(\beta p + \pi/4)}}{\sqrt{2\pi\beta p}} \frac{\frac{1}{n} \sin\left(\frac{\pi}{n}\right)}{\cos\left(\frac{\pi}{n}\right) - \cos\left(\frac{\phi-\phi'}{n}\right)} = D^L \frac{e^{-j\beta p}}{\sqrt{\beta}}$$

$$V_B^R = \frac{e^{-j(\beta p + \pi/4)}}{\sqrt{2\pi\beta p}} \frac{\frac{1}{n} \sin\left(\frac{\pi}{n}\right)}{\cos\left(\frac{\pi}{n}\right) - \cos\left(\frac{\phi+\phi'}{n}\right)} = D^R \frac{e^{-j\beta p}}{\sqrt{\beta}}$$

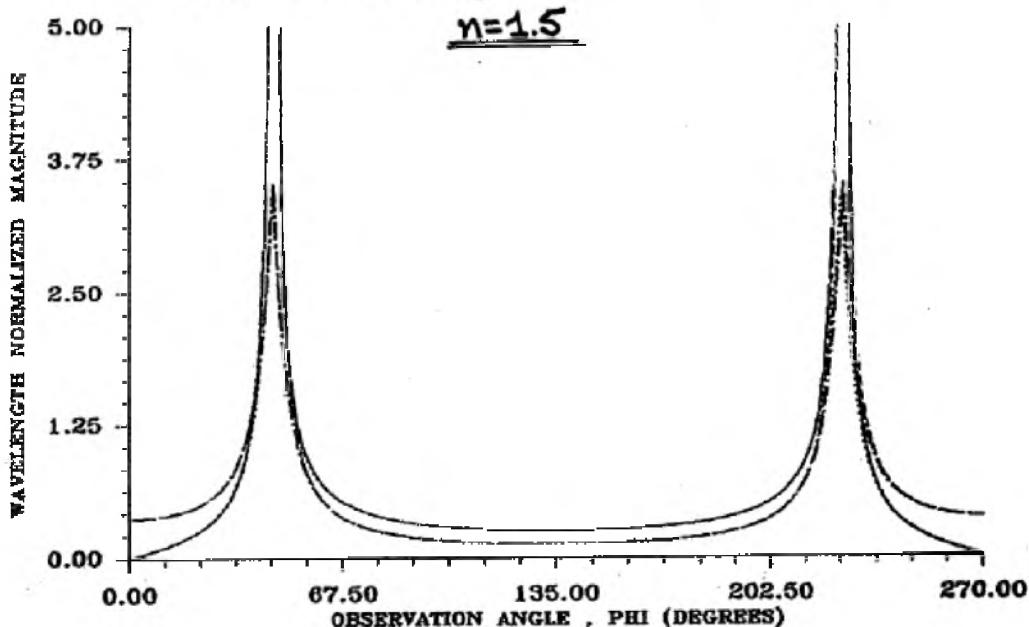
$$V_B^{S,h} = \frac{e^{-j(\beta p + \pi/4)}}{n\sqrt{2\pi\beta p}} \sin\left(\frac{\pi}{n}\right) \left\{ \frac{1}{\cos\left(\frac{\pi}{n}\right) - \cos\left(\frac{\phi-\phi'}{n}\right)} \mp \frac{1}{\cos\left(\frac{\pi}{n}\right) - \cos\left(\frac{\phi+\phi'}{n}\right)} \right\} = D^{S,h} \frac{e^{-j\beta p}}{\sqrt{\beta}}$$

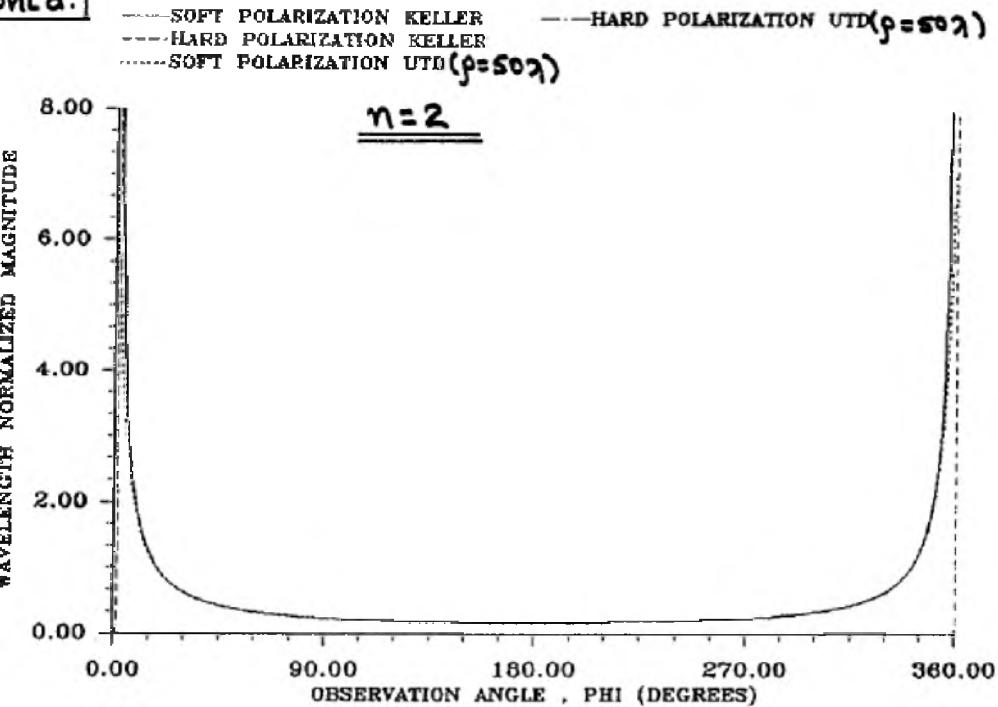
$$D^{S,h} = \frac{e^{-j\pi/4}}{n\sqrt{2\pi\beta}} \sin\left(\frac{\pi}{n}\right) \left\{ \frac{1}{\cos\left(\frac{\pi}{n}\right) - \cos\left(\frac{\phi-\phi'}{n}\right)} \mp \frac{1}{\cos\left(\frac{\pi}{n}\right) - \cos\left(\frac{\phi+\phi'}{n}\right)} \right\}$$

$$D^{S,h} \sqrt{\beta} = \frac{e^{-j\pi/4}}{n(2\pi)^2} \sin\left(\frac{\pi}{n}\right) \left\{ \frac{1}{\cos\left(\frac{\pi}{n}\right) - \cos\left(\frac{\phi-\phi'}{n}\right)} \mp \frac{1}{\cos\left(\frac{\pi}{n}\right) - \cos\left(\frac{\phi+\phi'}{n}\right)} \right\}$$

(b)

— SOFT POLARIZATION KELLER
 - - HARD POLARIZATION UTD
 HARD POLARIZATION KELLER
 SOFT POLARIZATION UTD ($\rho = 50\lambda$)

Cont'd.

(c) For $n=2$

$$\sqrt{s,h} = -\frac{e^{-j(\beta\rho + \pi/4)}}{2\sqrt{2}\pi\beta\rho} \left\{ \sec\left(\frac{\phi-\phi'}{2}\right) \mp \sec\left(\frac{\phi+\phi'}{2}\right) \right\}$$

(d)

$$\sigma_w(2-D) = \lim_{\rho \rightarrow \infty} \left\{ 2\pi\rho \frac{|\sqrt{s,h}|^2}{|1|^2} \right\} = \frac{1}{4\beta} \left\{ \sec\left(\frac{\phi-\phi'}{2}\right) \mp \sec\left(\frac{\phi+\phi'}{2}\right) \right\}^2$$

$$\sigma_w(2-D) = \frac{1}{8\pi} \left| \sec\left(\frac{\phi-\phi'}{2}\right) \mp \sec\left(\frac{\phi+\phi'}{2}\right) \right|^2$$

(e) For $\phi = \phi'$

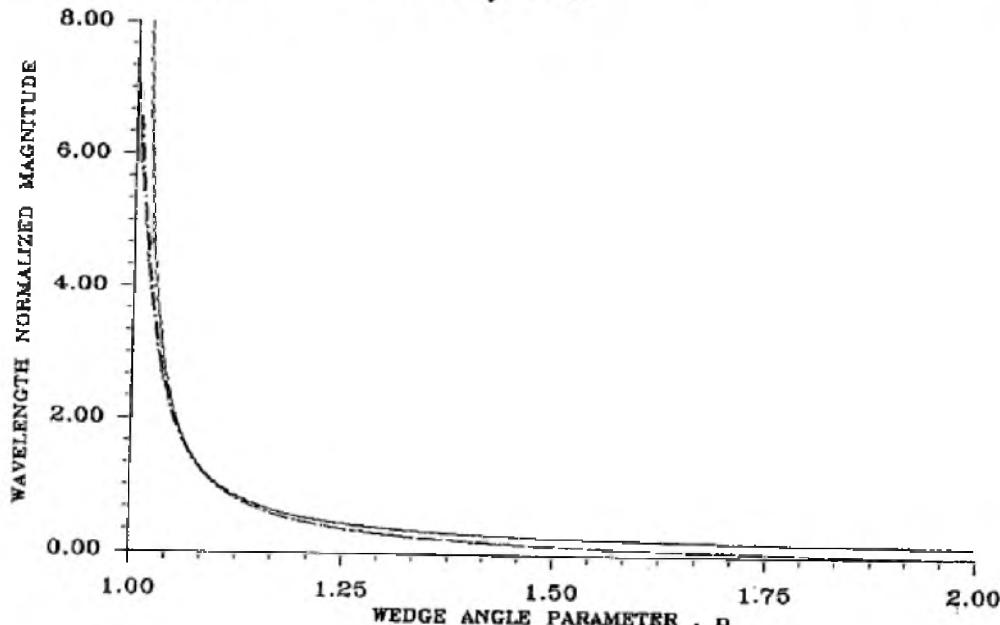
$$\sigma_w(2-D) = \frac{1}{8\pi} \left| 1 \mp \sec(\phi') \right|^2$$

It is seen that when $\phi' = 180^\circ$ the hard polarized scattering width vanishes. This implies that the leading edge of a half plane at grazing angle ($\phi = 180^\circ$) is invisible to the radar for hard polarized incident waves. This can be used to design low observable targets.

13.27 From the formulations of Problem 13.26, we can write for $\phi = \phi'$

$$D^{s,h} \sqrt{\lambda} = \frac{e^{-j\pi/4}}{n(2\pi)^2} \sin\left(\frac{\pi}{n}\right) \left\{ \frac{1}{\cos\left(\frac{\pi}{n}\right) - 1} \mp \frac{1}{\cos\left(\frac{\pi}{n}\right) - \cos\left(\frac{2\phi'}{n}\right)} \right\}$$

— SOFT POLARIZATION KELLER
 --- HARD POLARIZATION KELLER
 SOFT POLARIZATION UTD ($\rho = 502$)



13.28 From the formulations of Problem 13.26, we can write for $\phi = 0$ and $\phi' = n\pi/2$

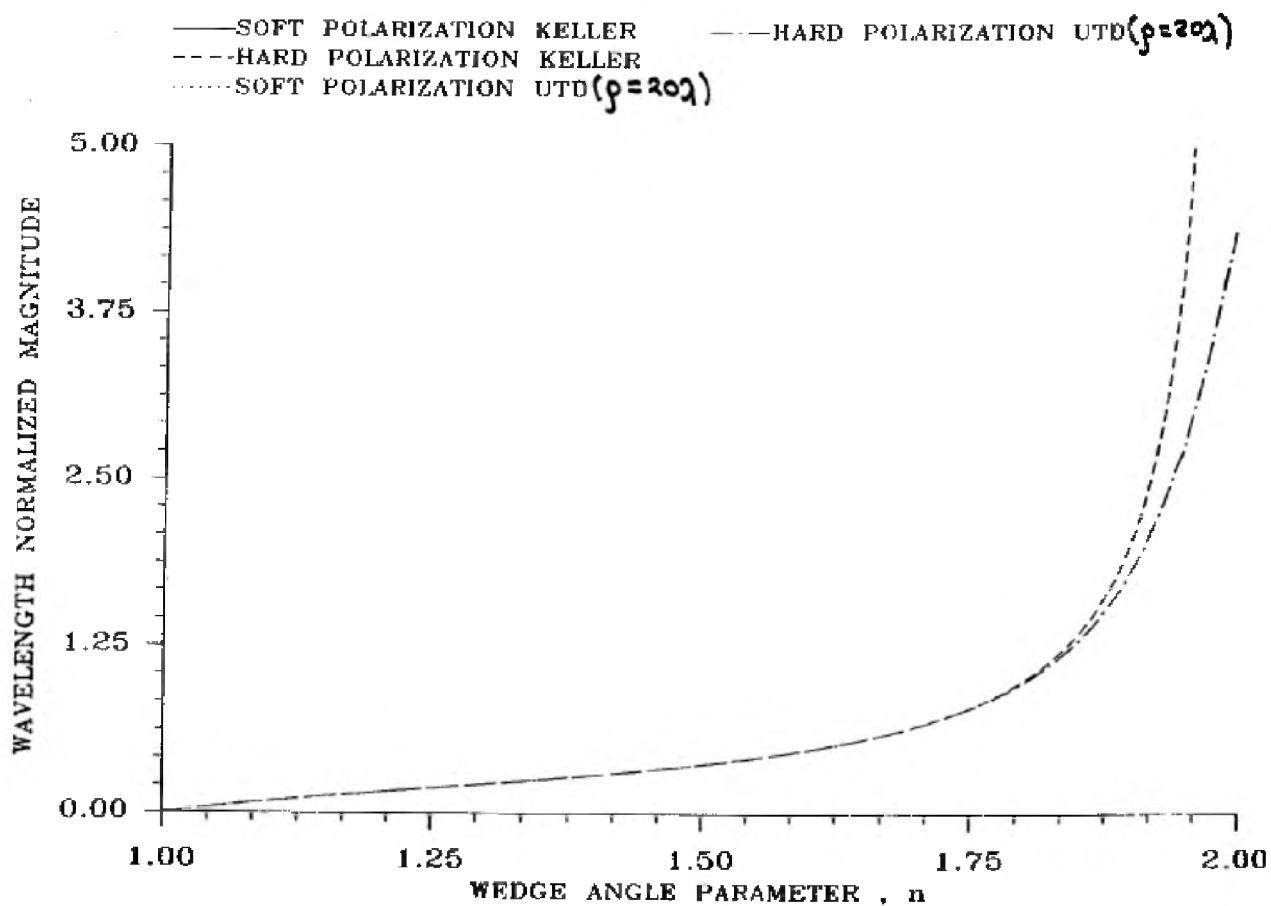
$$D^{s,h} \sqrt{\lambda} = \frac{e^{-j\pi/4}}{n(2\pi)^2} \sin\left(\frac{\pi}{n}\right) \left\{ \frac{1}{\cos\left(\frac{\pi}{n}\right) - \cos\left(\frac{\phi'}{n}\right)} \mp \frac{1}{\cos\left(\frac{\pi}{n}\right) - \cos\left(\frac{\phi'}{n}\right)} \right\}_{\phi = \frac{n\pi}{2}}$$

$$= \frac{e^{-j\pi/4}}{n(2\pi)^2} \sin\left(\frac{\pi}{n}\right) \left\{ \frac{1}{\cos\left(\frac{\pi}{n}\right) - \cos\left(\frac{\pi}{2}\right)} \mp \frac{1}{\cos\left(\frac{\pi}{n}\right) - \cos\left(\frac{\pi}{n}\right)} \right\}$$

The plots of these are shown on the next page.

Cont'd.

13.2B cont'd.



$$13.29 \quad F(x) = 2j\sqrt{x} e^{jx} \int_{-\infty}^{\infty} e^{-j\tau^2} d\tau$$

Let us write that

$$2j\sqrt{x} e^{jx} \int_0^{\infty} e^{-j\tau^2} d\tau = 2j\sqrt{x} e^{jx} \int_0^{\sqrt{x}} e^{-j\tau^2} d\tau + 2j\sqrt{x} e^{jx} \int_{\sqrt{x}}^{\infty} e^{-j\tau^2} d\tau$$

$$\text{or } \underbrace{2j\sqrt{x} e^{jx} \int_0^{\infty} e^{-j\tau^2} d\tau}_{F(x)} = \underbrace{2j\sqrt{x} e^{jx} \int_0^{\sqrt{x}} e^{-j\tau^2} d\tau}_{F_\infty(x)} + \underbrace{2j\sqrt{x} e^{jx} \int_{\sqrt{x}}^{\infty} e^{-j\tau^2} d\tau}_{F_0(x)}$$

$$F_\infty = 2j\sqrt{x} e^{jx} \int_0^{\sqrt{x}} e^{-j\tau^2} d\tau = 2j\sqrt{x} e^{jx} \left[\frac{\sqrt{\pi}}{2} e^{-j\pi/4} \right] = j\sqrt{\pi x} e^{-j\pi/4} e^{jx} = \sqrt{\pi x} e^{jx}$$

$$F_0 = 2j\sqrt{x} e^{jx} \int_{\sqrt{x}}^{\infty} e^{-j\tau^2} d\tau \approx 2j\sqrt{x} e^{jx} \int_{\sqrt{x}}^{\infty} \sum_{n=0}^M \frac{(-j\tau^2)^n}{n!} d\tau \approx 2j\sqrt{x} e^{jx} \sum_{n=0}^M \int_{\sqrt{x}}^{\infty} \frac{(-j\tau^2)^n}{n!} d\tau$$

$$F_0 \approx 2j\sqrt{x} e^{jx} \sum_{n=0}^M \frac{(-j\tau^2)^n \tau}{n! (2n+1)} \Big|_{\sqrt{x}}^{\infty} \approx 2jx e^{jx} \sum_{n=0}^M \frac{(-jx)^n}{n! (2n+1)}$$

Therefore

$$F(x) \approx \sqrt{\pi x} e^{j\pi/4} e^{jx} - 2jx e^{jx} \sum_{n=0}^M \frac{(-jx)^n}{n! (2n+1)} = e^{jx} \left\{ \sqrt{\pi x} e^{j\pi/4} - 2jx \sum_{n=0}^M \frac{(-jx)^n}{n! (2n+1)} \right\}$$

$$\approx e^{j(x+\pi/4)} \left\{ \sqrt{\pi x} - 2x e^{j\pi/4} \sum_{n=0}^M \frac{(-jx)^n}{n! (2n+1)} \right\}$$

$$13.30 \quad F(x) = 2j\sqrt{x} e^{jx} \int_{-\infty}^{\infty} e^{-j\tau^2} d\tau$$

Let us first look at the integral which can be written by letting $w = \sqrt{x}$, and repeatedly integrating by parts as

$$\int_w^{\infty} e^{-j\tau^2} d\tau = - \frac{e^{-j\tau^2}}{j2\tau} \Big|_w^{\infty} - \frac{1}{2j} \int_w^{\infty} \frac{e^{-j\tau^2}}{\tau^2} d\tau$$

$$= - \frac{e^{-j\tau^2}}{j2\tau} \Big|_w^{\infty} - \frac{1}{2j} \left\{ - \frac{e^{-j\tau^2}}{j2\tau^3} \Big|_w^{\infty} - \frac{3}{j2} \int_w^{\infty} \frac{e^{-j\tau^2}}{\tau^4} d\tau \right\}$$

$$\int_w^{\infty} e^{-j\tau^2} d\tau = - \frac{e^{-j\tau^2}}{j2\tau} \Big|_w^{\infty} - \frac{1}{2j} \left\{ \frac{e^{-j\tau^2}}{j2\tau^3} \Big|_w^{\infty} - \frac{3}{j2} \left[- \frac{e^{-j\tau^2}}{j2\tau^5} \Big|_w^{\infty} - \frac{5}{j2} \int_w^{\infty} \frac{e^{-j\tau^2}}{\tau^6} d\tau \right] \right\}$$

Continuing this process we can write that

$$\int_w^{\infty} e^{-j\tau^2} d\tau = \frac{e^{-jw^2}}{j2w} \left[1 + \sum_{m=1}^N \frac{1 \cdot 3 \cdot 5 \cdots (2m-1)}{(-j2w)^m} \right] + R_N \text{ where } R_N = \frac{1 \cdot 3 \cdot 5 \cdots (2N+1)}{(-j2)^{N+1}} \int_w^{\infty} \frac{e^{-j\tau^2}}{\tau^{2N+2}} d\tau$$

Since R_N decreases as N increases, then

$$F(x) = 2j\sqrt{x} e^{jx} \int_{-\infty}^{\infty} e^{-j\tau^2} d\tau \approx 2j\sqrt{x} e^{jx} \left\{ \frac{e^{-jx}}{j2\sqrt{x}} \left[1 + \sum_{m=1}^N \frac{1 \cdot 3 \cdot 5 \cdots (2m-1)}{(-j2x)^m} \right] \right\} = 1 + \sum_{m=1}^N \frac{1 \cdot 3 \cdot 5 \cdots (2m-1)}{(-j2x)^m}$$

a. The procedure to show that there is a finite discontinuity of the same amplitude and phase along the RSB of Figure 13-24a, you follow the same procedure as was done on pages 789 and 790 for the ISB of Figure 13-24a. Along the RSB of Figure 13-24a

$$\xi^+ = \phi + \phi' = \pi \quad (1)$$

and in the neighborhood of it

$$\xi^+ = \phi + \phi' = \pi - \epsilon \quad (2)$$

where ϵ is positive in the illuminated side of the reflection shadow boundary.

Using (2) we can write (13-66f)

$$2n\pi N^- - (\phi + \phi') = 2n\pi N^- - (\pi - \epsilon) = (2nN^- - 1)\pi + \epsilon = -\pi \quad (3)$$

For this situation the cotangent function that becomes singular is that shown in second row of Table 13-1 whose $N^- = 0$. Thus (3) reduces for $N^- = 0$ to

$$\phi + \phi' = \pi - \epsilon \quad (4)$$

which we can use to write the cotangent of the second row of Table 13-1 as

$$C^-(\phi + \phi', n) = \cot \left[\frac{\pi - (\phi + \phi')}{2n} \right] = \cot \left[\frac{\pi - (\pi - \epsilon)}{2n} \right] = \cot \left(\frac{\epsilon}{2n} \right) \approx \frac{2n}{\epsilon} = \frac{2n}{|\epsilon| \operatorname{sgn}(\epsilon)} \quad (5)$$

According to (13-66d)

$$g^-(\xi^+) = g^-(\phi + \phi') = 1 + \cos [(\phi + \phi') - 2n\pi N^-] = 1 + \cos(\phi + \phi') = 1 + \cos(\pi - \epsilon) = 1 - \cos(\epsilon) \\ \approx 1 - \left[1 - \frac{\epsilon^2}{2} \right] = \frac{\epsilon^2}{2} \quad (6)$$

The transition function of (13-69f) can also be written using (6) as

$$F[\beta pg^-(\xi^+)] = F[\beta pg^-(\phi + \phi')] = F[\beta p \left(\frac{\epsilon^2}{2} \right)] \quad (7)$$

which for small values of its argument can be approximated by the first term of (13-74a). Thus

$$F[\beta p \left(\frac{\epsilon^2}{2} \right)] \approx \sqrt{\frac{\pi \beta p \epsilon^2}{2}} e^{j\pi/4} = |\epsilon| \sqrt{\frac{\pi \beta p}{2}} e^{j\pi/4} \quad (8)$$

Thus the product of $C^-(\phi + \phi', n) F[\beta pg^-(\phi + \phi')]$ can be approximated by

$$C^-(\phi + \phi', n) F[\beta pg^-(\phi + \phi')] = \frac{2n}{|\epsilon| \operatorname{sgn}(\epsilon)} |\epsilon| \sqrt{\frac{\pi \beta p}{2}} e^{j\pi/4} = n \sqrt{2\pi \beta p} \operatorname{sgn}(\epsilon) e^{j\pi/4} \quad (9)$$

Therefore the corresponding reflected diffracted field of (13-69) can be approximated along the reflection shadow boundary by the second term within the brackets of (13-69b). Thus we can write that

$$V_B^r(g, \phi + \phi' = \pi - \epsilon, n) = \frac{e^{-j\beta p}}{\sqrt{p}} \left\{ -\frac{e^{-j\pi/4}}{2n\sqrt{2\pi \beta p}} n \sqrt{2\pi \beta p} \operatorname{sgn}(\epsilon) e^{j\pi/4} \right\} = -\frac{e^{-j\beta p}}{2} \operatorname{sgn}(\epsilon)$$

Cont'd.

13.31 Cont'd.

b. The same procedure is used along the ISB of Figure 13-24 b. The only difference is that you use (13-66c) with $\phi-\phi'$; the third row of Table 13-1, (13-66c) with $(\phi-\phi')^0$, (13-68e), and (13-69) can be approximated by the first term within the brackets of (13-69b). The result is the same as (13-83).

c. The same procedure is used along the RSB of Figure 13-24b. The only difference is that you use (13-66c) with $\phi+\phi'$, the fourth row of Table 13-1, (13-66c) with $\phi+\phi'$, (13-69a), and (13-69) can be approximated by the first term within the brackets of (13-69b). The result is the same as (13-83)

13.32

(a) ISB $(\phi-\phi')=\pi$ See Figure 13-24(a)

$$\cot \left[\frac{\pi - (\phi - \phi')}{2n} \right] = \cot \left(\frac{\pi - \pi}{2n} \right) = \cot \left(\frac{0}{2n} \right) = \infty, F[\beta \rho q^-(\phi - \phi')] = 0$$

Product: $\cot \left[\frac{\pi - (\phi - \phi')}{2n} \right] F[\beta \rho q^-(\phi - \phi')] =$ Discontinuous but finite bounded; removes GO discontinuity
 $\phi - \phi' = \pi$

(b)

RSB $(\phi + \phi') = \pi$ See Figure 13-24(a)

$$\cot \left[\frac{\pi - (\phi + \phi')}{2n} \right] = \cot \left(\frac{\pi - \pi}{2n} \right) = \cot \left(\frac{0}{2n} \right) = \infty, F[\beta \rho q^+(\phi + \phi')] = 0$$

Product: $\cot \left[\frac{\pi - (\phi + \phi')}{2n} \right] F[\beta \rho q^+(\phi + \phi')] =$ Discontinuous but finite bounded; removes GO discontinuity
 $\phi + \phi' = \pi$

13.33

(a) ISB $(\phi - \phi' = -\pi)$ See Figure 13-24(b)

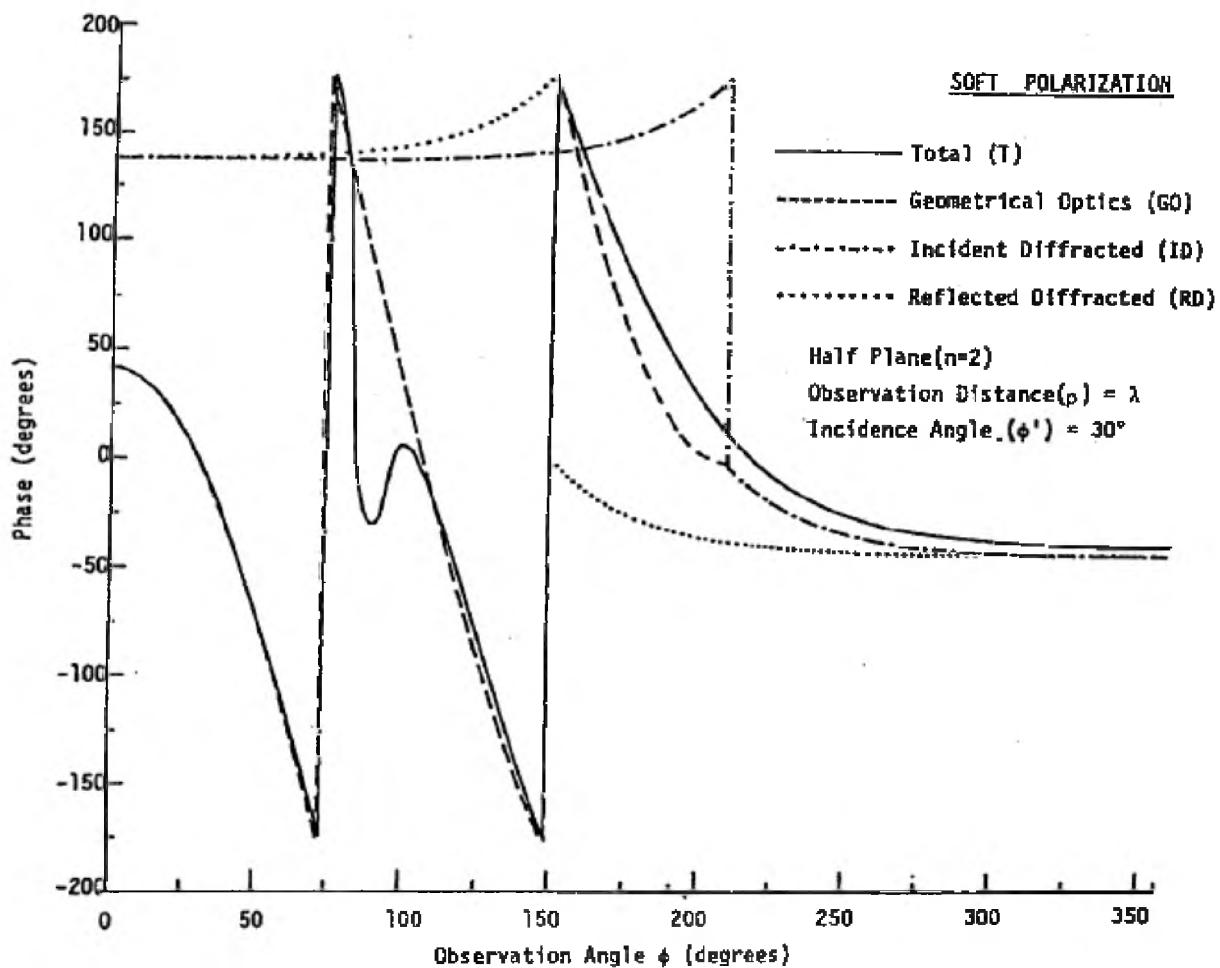
$$\cot \left[\frac{\pi + (\phi - \phi')}{2n} \right] = \cot \left(\frac{\pi - \pi}{2n} \right) = \cot \left(\frac{0}{2n} \right) = \infty; F[\beta \rho q^+(\phi - \phi')] = 0$$

Product: $\cot \left[\frac{\pi + (\phi - \phi')}{2n} \right] F[\beta \rho q^+(\phi - \phi')] =$ Discontinuous but finite bounded; removes GO discontinuity
 $\phi - \phi' = -\pi$

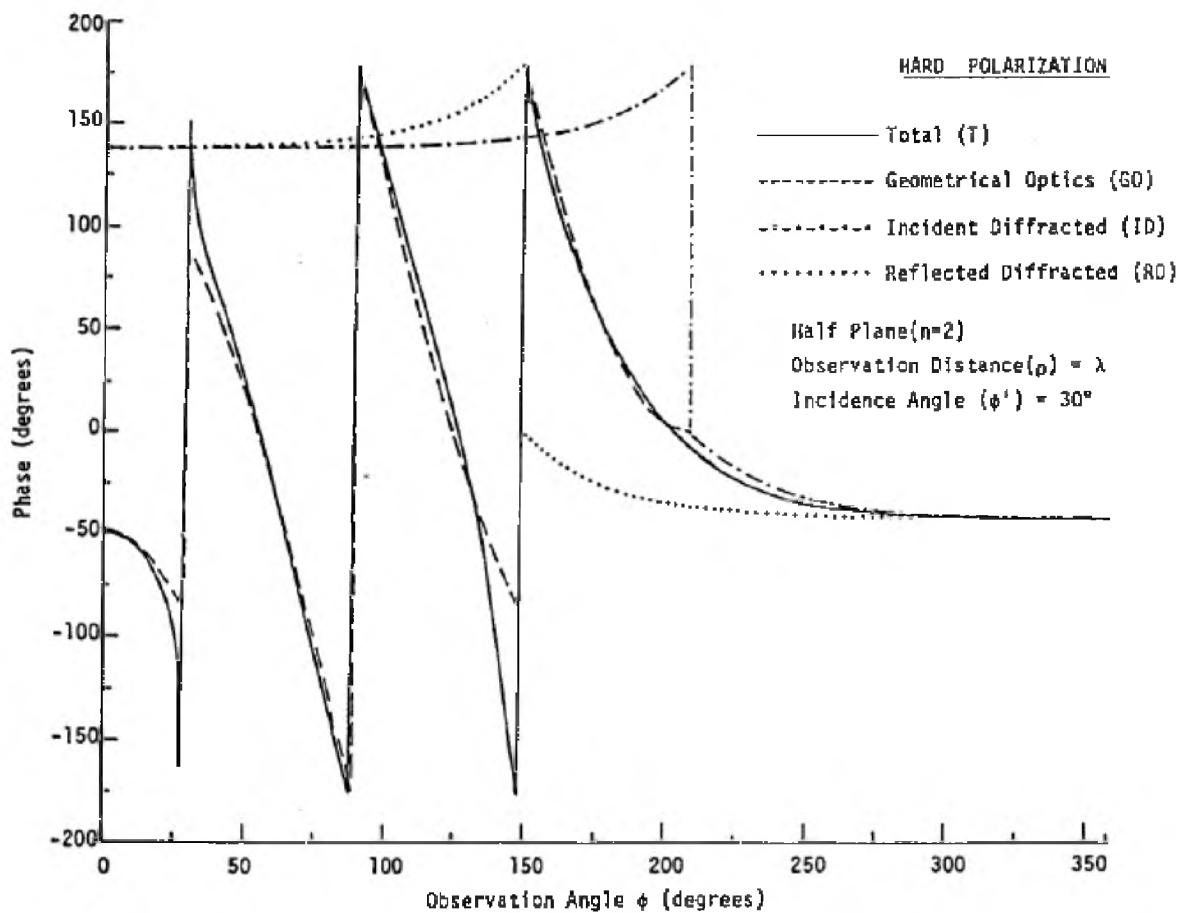
(b) RSB $[\phi + \phi' = (2n-1)\pi]$ See Figure 13-24(b)

$$\cot \left[\frac{\pi + (\phi + \phi')}{2n} \right] = \cot \left[\frac{\pi + (2n-1)\pi}{2n} \right] = \cot(\pi) = \infty; F[\beta \rho q^+(\phi + \phi')] = 0$$

Product: $\cot \left[\frac{\pi + (\phi + \phi')}{2n} \right] F[\beta \rho q^+(\phi + \phi')] =$ Discontinuous but finite bounded; removes GO discontinuity.
 $\phi + \phi' = (2n-1)\pi$

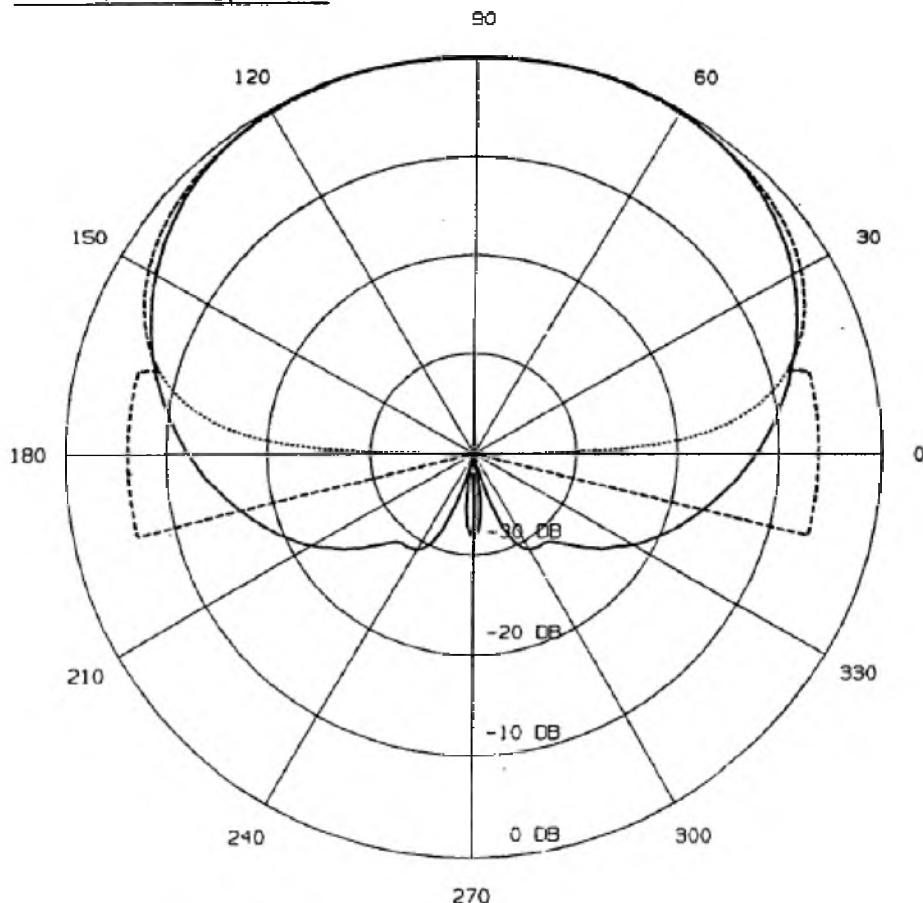
(a) Soft Polarization PhaseCont'd.

(b) Hard Polarization Phase



13.55

----- Total GO Field ($W=\infty$)
 ----- Total GO Field ($W=2\lambda$)
 ——— Total Field (GO+DF) ($W=2\lambda$)
 $h = 0.25\lambda$



13.56

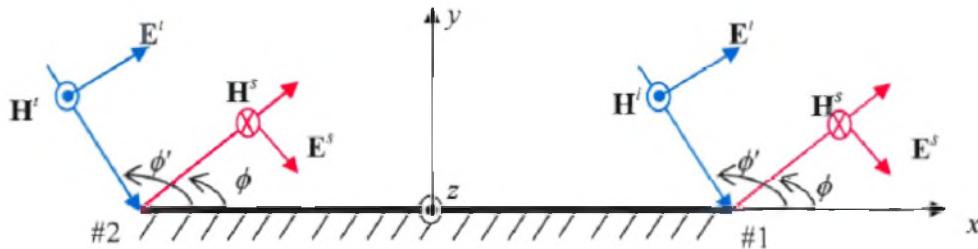
The only difference between the soft and hard polarized fields is sign in front of the second term within the brackets in the diffraction coefficient on pages 796-798. Changing that sign, we can eventually write the total magnetic field based on the electric field of page 798 as

$$\underline{H}^d = -\hat{\alpha}_2 \underline{H}_0 \frac{e^{-j\pi/4}}{\sqrt{2\pi\rho}} \left[\cos(\beta w \cos\phi) - j\beta w \frac{\sin(\beta w \cos\phi)}{\beta w \cos\phi} \right] \frac{e^{-j\beta\rho}}{\sqrt{\rho}}$$

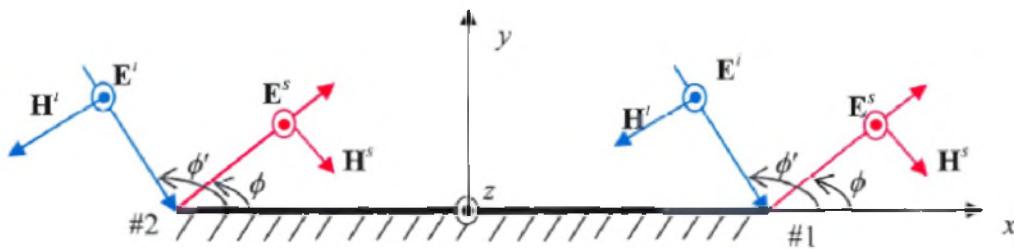
They are identical in form except for the sign (minus) in front of the second term within the brackets. Therefore the scattering width for the hard polarization is identical polarization since it involves the magnitude of the diffracted field.

Geometrical/Uniform Theory of Diffraction (GTD/UTD):Geometry

Hard Polarization:



Soft Polarization:

**(a) (Hard Polarization)**

- Backscattered /Monostatic SW for a hard polarized uniform plane wave incident upon a PEC strip of width w .

Using the geometry of the hard polarization, we can state that the incident field is:

$$\mathbf{H}^i = \hat{a}_z H_0 e^{j\beta(x\cos\phi' + y\sin\phi')}$$

Wedge #1:

In the wedge #1, the incident field is given by,

$$\mathbf{H}^i \Big|_{\substack{x=\frac{w}{2} \\ y=0}} = \hat{a}_z H_0 e^{j\beta(w/2\cos\phi')}$$

The backscattered field is:

$$\mathbf{H}_1^d = \mathbf{H}^i \cdot \bar{\mathbf{D}}_1^h \cdot A_1(\rho_1) e^{-j\beta\rho_1}$$

cont'd

13.37 cont'd

where

$$D_1^h = \frac{e^{-\frac{j\pi}{4}} \left(\frac{1}{n}\right) \sin\left(\frac{\pi}{n}\right)}{\sqrt{2\pi\beta}} \left[\frac{1}{\cos\left(\frac{\pi}{n}\right) - \cos\left(\frac{\psi_1 - \psi_1'}{n}\right)} + \frac{1}{\cos\left(\frac{\pi}{n}\right) - \cos\left(\frac{\psi_1 + \psi_1'}{n}\right)} \right]$$

For this geometry and wedge #1, $n = 2$ and $\psi_1 = \pi - \phi$ and $\psi_1' = \pi - \phi'$

$$\begin{aligned} D_1^h &= \frac{e^{-\frac{j\pi}{4}} \left(\frac{1}{2}\right) \sin\left(\frac{\pi}{2}\right)}{\sqrt{2\pi\beta}} \left[\frac{1}{\cos\left(\frac{\pi}{2}\right) - \cos\left(\frac{\pi - \phi - \pi + \phi'}{2}\right)} \right. \\ &\quad \left. + \frac{1}{\cos\left(\frac{\pi}{2}\right) - \cos\left(\frac{\pi - \phi + \pi - \phi'}{2}\right)} \right] \\ &= \frac{e^{-\frac{j\pi}{4}}}{2\sqrt{2\pi\beta}} \left[\frac{1}{-\cos\left(\frac{-\phi + \phi'}{2}\right)} + \frac{1}{-\cos\left(\frac{2\pi - \phi - \phi'}{2}\right)} \right] \\ &= \frac{e^{-\frac{j\pi}{4}}}{2\sqrt{2\pi\beta}} \left[\frac{1}{\cos\left(\frac{-\phi + \phi'}{2}\right)} + \frac{1}{-\cos\left(\frac{2\pi - \phi - \phi'}{2}\right)} \right] \\ D_1^h &= \frac{-e^{-\frac{j\pi}{4}}}{2\sqrt{2\pi\beta}} \left[\frac{1}{\cos\left(\frac{-\phi + \phi'}{2}\right)} + \frac{1}{\cos\left(\frac{2\pi - \phi - \phi'}{2}\right)} \right] \end{aligned}$$

Then, the diffracted field from wedge #1 is given by,

$$\mathbf{H}_1^d = -\hat{a}_z H_0 e^{j\beta(w/2\cos\phi')} \frac{e^{-\frac{j\pi}{4}}}{2\sqrt{2\pi\beta}} \left[\frac{1}{\cos\left(\frac{-\phi + \phi'}{2}\right)} + \frac{1}{\cos\left(\frac{2\pi - \phi - \phi'}{2}\right)} \right] \frac{e^{-j\beta(p-w/2\cos\phi)}}{\sqrt{\rho}}$$

or,

$$\boxed{\mathbf{H}_1^d = -\hat{a}_z H_0 e^{j\beta w/2(\cos\phi' + \cos\phi)} \frac{e^{-\frac{j\pi}{4}}}{2\sqrt{2\pi\beta}} \left[\frac{1}{\cos\left(\frac{-\phi + \phi'}{2}\right)} + \frac{1}{\cos\left(\frac{2\pi - \phi - \phi'}{2}\right)} \right] \frac{e^{-j\beta\rho}}{\sqrt{\rho}}}$$

cont'd

13.37 cont'd

Wedge #2:

In the wedge #2, the incident field is given by,

$$\mathbf{H}^i \Big|_{\substack{x=-\frac{w}{2} \\ y=0}} = \hat{a}_z H_0 e^{-j\beta(w/2\cos\phi')}$$

The backscattered field is:

$$\mathbf{H}_2^d = \mathbf{H}^i \cdot \bar{\mathbf{D}}_2^h \cdot A_2(\rho_2) e^{-j\beta\rho_2}$$

where

$$D_2^h = \frac{e^{-j\pi/4} \left(\frac{1}{n}\right) \sin\left(\frac{\pi}{n}\right)}{\sqrt{2\pi\beta}} \left[\frac{1}{\cos\left(\frac{\pi}{n}\right) - \cos\left(\frac{\psi_2 - \psi'_2}{n}\right)} + \frac{1}{\cos\left(\frac{\pi}{n}\right) - \cos\left(\frac{\psi_2 + \psi'_2}{n}\right)} \right]$$

For this geometry and wedge #2, $n = 2$ and $\psi_2 = \phi, \psi'_2 = \phi'$

$$D_2^h = \frac{e^{-j\pi/4} \left(\frac{1}{2}\right) \sin\left(\frac{\pi}{2}\right)}{\sqrt{2\pi\beta}} \left[\frac{1}{\cos\left(\frac{\pi}{2}\right) - \cos\left(\frac{\phi - \phi'}{2}\right)} + \frac{1}{\cos\left(\frac{\pi}{2}\right) - \cos\left(\frac{\phi + \phi'}{2}\right)} \right]$$

$$D_2^h = \frac{e^{-j\pi/4}}{2\sqrt{2\pi\beta}} \left[\frac{1}{-\cos\left(\frac{\phi - \phi'}{2}\right)} + \frac{1}{-\cos\left(\frac{\phi + \phi'}{2}\right)} \right]$$

$$D_2^h = -\frac{e^{-j\pi/4}}{2\sqrt{2\pi\beta}} \left[\frac{1}{\cos\left(\frac{\phi - \phi'}{2}\right)} + \frac{1}{\cos\left(\frac{\phi + \phi'}{2}\right)} \right]$$

Then, the diffracted field from wedge #2 is given by,

$$\mathbf{H}_2^d = -\hat{a}_z H_0 e^{-j\beta(w/2\cos\phi')} \frac{e^{-j\pi/4}}{2\sqrt{2\pi\beta}} \left[\frac{1}{\cos\left(\frac{\phi - \phi'}{2}\right)} + \frac{1}{\cos\left(\frac{\phi + \phi'}{2}\right)} \right] \frac{1}{\sqrt{\rho}} e^{-j\beta(\rho+w/2\cos\phi)}$$

or,

$$\boxed{\mathbf{H}_2^d = -\hat{a}_z H_0 e^{-j\beta w/2(\cos\phi' + \cos\phi)} \frac{e^{-j\pi/4}}{2\sqrt{2\pi\beta}} \left[\frac{1}{\cos\left(\frac{\phi - \phi'}{2}\right)} + \frac{1}{\cos\left(\frac{\phi + \phi'}{2}\right)} \right] \frac{1}{\sqrt{\rho}} e^{-j\beta\rho}}$$

cont'd

13.37 cont'd]

Now, the total diffracted field is given by,

$$\begin{aligned}
 \mathbf{H}^d &= \mathbf{H}_1^d + \mathbf{H}_2^d \\
 \mathbf{H}^d &= -\hat{a}_z H_0 e^{j\beta w/2(\cos\phi' + \cos\phi)} \frac{e^{-j\pi/4}}{2\sqrt{2\pi\beta}} \left[\frac{1}{\cos\left(\frac{-\phi + \phi'}{2}\right)} + \frac{1}{\cos\left(\frac{2\pi - \phi - \phi'}{2}\right)} \right] \frac{1}{\sqrt{\rho}} e^{-j\beta\rho} \\
 &\quad - \hat{a}_z H_0 e^{-j\beta w/2(\cos\phi' + \cos\phi)} \frac{e^{-j\pi/4}}{2\sqrt{2\pi\beta}} \left[\frac{1}{\cos\left(\frac{\phi - \phi'}{2}\right)} + \frac{1}{\cos\left(\frac{\phi + \phi'}{2}\right)} \right] \frac{1}{\sqrt{\rho}} e^{-j\beta\rho} \\
 \mathbf{H}^d &= -\hat{a}_z H_0 \frac{e^{-j\pi/4}}{2\sqrt{2\pi\beta}} \left\{ e^{j\beta w/2(\cos\phi' + \cos\phi)} \left[\frac{1}{\cos\left(\frac{-\phi + \phi'}{2}\right)} + \frac{1}{\cos\left(\frac{2\pi - \phi - \phi'}{2}\right)} \right] \right. \\
 &\quad \left. + e^{-j\beta w/2(\cos\phi' + \cos\phi)} \left[\frac{1}{\cos\left(\frac{\phi - \phi'}{2}\right)} + \frac{1}{\cos\left(\frac{\phi + \phi'}{2}\right)} \right] \right\} \frac{1}{\sqrt{\rho}} e^{-j\beta\rho} \\
 \mathbf{H}^d &= -\hat{a}_z H_0 \frac{e^{-j\pi/4}}{2\sqrt{2\pi\beta}} \left\{ e^{j\beta w/2(\cos\phi' + \cos\phi)} \left[\frac{1}{\cos\left(\frac{\phi - \phi'}{2}\right)} - \frac{1}{\cos\left(\frac{\phi + \phi'}{2}\right)} \right] \right. \\
 &\quad \left. + e^{-j\beta w/2(\cos\phi' + \cos\phi)} \left[\frac{1}{\cos\left(\frac{\phi - \phi'}{2}\right)} + \frac{1}{\cos\left(\frac{\phi + \phi'}{2}\right)} \right] \right\} \frac{1}{\sqrt{\rho}} e^{-j\beta\rho}
 \end{aligned}$$

Regrouping similar terms,

$$\begin{aligned}
 \mathbf{H}^d &= -\hat{a}_z H_0 \frac{e^{-j\pi/4}}{\sqrt{2\pi\beta}} \left\{ \frac{1}{\cos\left(\frac{\phi - \phi'}{2}\right)} \left[\frac{e^{j\beta w/2(\cos\phi' + \cos\phi)} + e^{-j\beta w/2(\cos\phi' + \cos\phi)}}{2} \right] \right. \\
 &\quad \left. - \frac{1}{\cos\left(\frac{\phi + \phi'}{2}\right)} \left[\frac{e^{j\beta w/2(\cos\phi' + \cos\phi)} - e^{-j\beta w/2(\cos\phi' + \cos\phi)}}{2} \right] \right\} \frac{1}{\sqrt{\rho}} e^{-j\beta\rho} \\
 \mathbf{H}^d &= -\hat{a}_z H_0 \frac{e^{-j\pi/4}}{\sqrt{2\pi\beta}} \left\{ \frac{1}{\cos\left(\frac{\phi - \phi'}{2}\right)} [\cos(\beta w/2(\cos\phi' + \cos\phi))] \right. \\
 &\quad \left. - \frac{j}{\cos\left(\frac{\phi + \phi'}{2}\right)} [\sin(\beta w/2(\cos\phi' + \cos\phi))] \right\} \frac{1}{\sqrt{\rho}} e^{-j\beta\rho}
 \end{aligned}$$

cont'd

13.37 cont'd

For backscattered, when $\phi = \phi'$, the diffracted field, that is equal to the backscattered field, reduces to:

$$\mathbf{H}^s = \mathbf{H}^d = -\hat{a}_z H_0 \frac{e^{-\frac{j\pi}{4}}}{\sqrt{2\pi\beta}} \left\{ \cos(\beta w \cos\phi) - j\beta w \frac{\sin(\beta w \cos\phi)}{(\beta w \cos\phi)} \right\} \frac{e^{-j\beta\rho}}{\sqrt{\rho}}$$

Therefore, the **monostatic scattering width** can be obtained as

$$\begin{aligned} \sigma_{2-D} &= \lim_{\rho \rightarrow \infty} \left[2\pi\rho \left| \frac{\mathbf{H}^s}{\mathbf{H}^i} \right|^2 \right] \\ \sigma_{2-D} &= \lim_{\rho \rightarrow \infty} \left[2\pi\rho \left| \frac{-\hat{a}_z H_0 \frac{e^{-\frac{j\pi}{4}}}{\sqrt{2\pi\beta}} \left\{ \cos(\beta w \cos\phi) - j\beta w \frac{\sin(\beta w \cos\phi)}{(\beta w \cos\phi)} \right\} \frac{e^{-j\beta\rho}}{\sqrt{\rho}}}{H_0 e^{j\beta(x \cos\phi' + y \sin\phi')}} \right|^2 \right] \\ \sigma_{2-D} &= \lim_{\rho \rightarrow \infty} \left[\frac{1}{\beta} \left| \cos(\beta w \cos\phi') - j\beta w \frac{\sin(\beta w \cos\phi')}{(\beta w \cos\phi')} \right|^2 \right] \\ \boxed{\sigma_{2-D} = \frac{\lambda}{2\pi} \left| \cos(\beta w \cos\phi') - j\beta w \frac{\sin(\beta w \cos\phi')}{(\beta w \cos\phi')} \right|^2} \end{aligned}$$

- **Bistatic scattering field and its bistatic scattered width SW for a hard polarized uniform plane wave incident upon a PEC strip of width w.**

From the previous formulation, we have the total diffracted field is given by,

$$\mathbf{H}^s = -\hat{a}_z H_0 \frac{e^{-\frac{j\pi}{4}}}{\sqrt{2\pi\beta}} \left\{ \frac{\cos(\beta w/2(\cos\phi' + \cos\phi))}{\cos(\frac{\phi - \phi'}{2})} - \frac{j\sin(\beta w/2(\cos\phi' + \cos\phi))}{\cos(\frac{\phi + \phi'}{2})} \right\} \frac{1}{\sqrt{\rho}} e^{-j\beta\rho}$$

cont'd

13.37 Cont'd

Therefore, the **bistatic scattering width** is given by,

$$\begin{aligned}\sigma_{2-D} &= \lim_{\rho \rightarrow \infty} \left[2\pi\rho \left| \frac{\mathbf{H}^s}{\mathbf{H}^i} \right|^2 \right] \\ &= \lim_{\rho \rightarrow \infty} \left[2\pi\rho \left| \frac{-\hat{a}_z H_0 \frac{e^{-j\frac{\pi}{4}}}{\sqrt{2\pi\beta}} \left\{ \frac{\cos(\beta w/2(\cos\phi' + \cos\phi))}{\cos(\frac{\phi - \phi'}{2})} - \frac{j\sin(\beta w/2(\cos\phi' + \cos\phi))}{\cos(\frac{\phi + \phi'}{2})} \right\} \frac{e^{-j\beta\rho}}{\sqrt{\rho}}}{H_0 e^{j\beta(x\cos\phi' + y\sin\phi')}} \right|^2 \right] \\ \sigma_{2-D} &= \lim_{\rho \rightarrow \infty} \left[\frac{1}{\beta} \left| \frac{\cos(\beta w/2(\cos\phi' + \cos\phi))}{\cos(\frac{\phi - \phi'}{2})} - \frac{j\sin(\beta w/2(\cos\phi' + \cos\phi))}{\cos(\frac{\phi + \phi'}{2})} \right|^2 \right] \\ \sigma_{2-D} &= \boxed{\frac{\lambda}{2\pi} \left| \frac{\cos(\beta w/2(\cos\phi' + \cos\phi))}{\cos(\frac{\phi - \phi'}{2})} - \frac{j\sin(\beta w/2(\cos\phi' + \cos\phi))}{\cos(\frac{\phi + \phi'}{2})} \right|^2}\end{aligned}$$

(b) **Bistatic scattering field and its bistatic scattered width for a soft polarized uniform plane wave incident upon a PEC strip of width w .**

Using the geometry of the soft polarization, we can state that the incident field is:

$$\mathbf{E}^i = \hat{a}_z \eta H_0 e^{j\beta(x\cos\phi' + y\sin\phi')}$$

Wedge #1:

In the wedge #1, the incident field is given by,

$$\mathbf{E}^i \Big|_{\substack{x=\frac{w}{2} \\ y=0}} = \hat{a}_z \eta H_0 e^{j\beta(w/2\cos\phi_i)}$$

The backscattered field is:

$$\mathbf{E}_1^d = \mathbf{E}^i \cdot \bar{\mathbf{D}}_1^s \cdot A_1(\rho_1) e^{-j\beta\rho_1}$$

cont'd

13.37 cont'd

where

$$D_1^s = \frac{e^{-j\pi/4} \left(\frac{1}{n}\right) \sin\left(\frac{\pi}{n}\right)}{\sqrt{2\pi\beta}} \left[\frac{1}{\cos\left(\frac{\pi}{n}\right) - \cos\left(\frac{\psi_1 - \psi_1'}{n}\right)} - \frac{1}{\cos\left(\frac{\pi}{n}\right) - \cos\left(\frac{\psi_1 + \psi_1'}{n}\right)} \right]$$

For this geometry and wedge #1, $n=2$ and $\psi_1 = \pi - \phi$ and $\psi_1' = \pi - \phi'$

$$\begin{aligned} D_1^s &= \frac{e^{-j\pi/4} \left(\frac{1}{2}\right) \sin\left(\frac{\pi}{2}\right)}{\sqrt{2\pi\beta}} \left[\frac{1}{\cos\left(\frac{\pi}{2}\right) - \cos\left(\frac{\pi - \phi - \pi + \phi'}{2}\right)} \right. \\ &\quad \left. - \frac{1}{\cos\left(\frac{\pi}{2}\right) - \cos\left(\frac{\pi - \phi + \pi - \phi'}{2}\right)} \right] \\ D_1^s &= \frac{e^{-j\pi/4}}{2\sqrt{2\pi\beta}} \left[\frac{1}{-\cos\left(\frac{-\phi + \phi'}{2}\right)} - \frac{1}{-\cos\left(\frac{2\pi - \phi - \phi'}{2}\right)} \right] \\ &= \frac{e^{-j\pi/4}}{2\sqrt{2\pi\beta}} \left[-\frac{1}{\cos\left(\frac{-\phi + \phi'}{2}\right)} - \frac{1}{-\cos\left(\frac{2\pi - \phi - \phi'}{2}\right)} \right] \\ D_1^s &= \frac{-e^{-j\pi/4}}{2\sqrt{2\pi\beta}} \left[\frac{1}{\cos\left(\frac{-\phi + \phi'}{2}\right)} - \frac{1}{\cos\left(\frac{2\pi - \phi - \phi'}{2}\right)} \right] \end{aligned}$$

Then, the diffracted field from wedge #1 is given by,

$$\mathbf{E}_1^d = -\hat{a}_z \eta H_0 e^{j\beta(w/2\cos\phi')} \frac{e^{-j\pi/4}}{2\sqrt{2\pi\beta}} \left[\frac{1}{\cos\left(\frac{-\phi + \phi'}{2}\right)} - \frac{1}{\cos\left(\frac{2\pi - \phi - \phi'}{2}\right)} \right] \frac{e^{-j\beta(\rho-w/2\cos\phi)}}{\sqrt{\rho}}$$

or,

$$\boxed{\mathbf{E}_1^d = -\hat{a}_z \eta H_0 e^{j\beta w/2(\cos\phi' + \cos\phi)} \frac{e^{-j\pi/4}}{2\sqrt{2\pi\beta}} \left[\frac{1}{\cos\left(\frac{-\phi + \phi'}{2}\right)} + \frac{1}{\cos\left(\frac{\phi + \phi'}{2}\right)} \right] \frac{e^{-j\beta\rho}}{\sqrt{\rho}}}$$

cont'd

13.37 cont'd

Wedge #2:

In the wedge #2, the incident field is given by,

$$\mathbf{E}^i \Big|_{\substack{x=-\frac{w}{2} \\ y=0}} = \hat{a}_z \eta H_0 e^{-j\beta(w/2\cos\phi)}$$

The backscattered field is:

$$\mathbf{E}_2^d = \mathbf{E}^i \cdot \bar{\mathbf{D}}_2^s \cdot A_2(\rho_2) e^{-j\beta\rho_2}$$

where

$$D_2^s = \frac{e^{-j\pi/4} \left(\frac{1}{n}\right) \sin\left(\frac{\pi}{n}\right)}{\sqrt{2\pi\beta}} \left[\frac{1}{\cos\left(\frac{\pi}{n}\right) - \cos\left(\frac{\psi_2 - \psi_2'}{n}\right)} - \frac{1}{\cos\left(\frac{\pi}{n}\right) - \cos\left(\frac{\psi_2 + \psi_2'}{n}\right)} \right]$$

For this geometry and wedge #2, $n = 2$ and $\psi_2 = \phi, \psi_2' = \phi'$

$$\begin{aligned} D_2^s &= \frac{e^{-j\pi/4} \left(\frac{1}{2}\right) \sin\left(\frac{\pi}{2}\right)}{\sqrt{2\pi\beta}} \left[\frac{1}{\cos\left(\frac{\pi}{2}\right) - \cos\left(\frac{\phi - \phi'}{2}\right)} - \frac{1}{\cos\left(\frac{\pi}{2}\right) - \cos\left(\frac{\phi + \phi'}{2}\right)} \right] \\ &= \frac{e^{-j\pi/4}}{2\sqrt{2\pi\beta}} \left[\frac{1}{-\cos\left(\frac{\phi - \phi'}{2}\right)} - \frac{1}{-\cos\left(\frac{\phi + \phi'}{2}\right)} \right] \\ D_2^s &= -\frac{e^{-j\pi/4}}{2\sqrt{2\pi\beta}} \left[\frac{1}{\cos\left(\frac{\phi - \phi'}{2}\right)} - \frac{1}{\cos\left(\frac{\phi + \phi'}{2}\right)} \right] \end{aligned}$$

Then, the diffracted field from wedge #2 is given by,

$$\mathbf{E}_2^d = -\hat{a}_z \eta H_0 e^{-j\beta(w/2\cos\phi)} \frac{e^{-j\pi/4}}{2\sqrt{2\pi\beta}} \left[\frac{1}{\cos\left(\frac{\phi - \phi'}{2}\right)} - \frac{1}{\cos\left(\frac{\phi + \phi'}{2}\right)} \right] \frac{1}{\sqrt{\rho}} e^{-j\beta(\rho+w/2\cos\phi)}$$

Cont'd

13.37 cont'd

or,

$$\mathbf{E}_2^d = -\hat{a}_z \eta H_0 e^{-j\beta w/2(\cos\phi' + \cos\phi)} \frac{e^{-j\pi/4}}{2\sqrt{2\pi\beta}} \left[\frac{1}{\cos\left(\frac{\phi - \phi'}{2}\right)} - \frac{1}{\cos\left(\frac{\phi + \phi'}{2}\right)} \right] \frac{1}{\sqrt{\rho}} e^{-j\beta\rho}$$

Now, the total diffracted field is given by,

$$\begin{aligned} \mathbf{E}^d &= \mathbf{E}_1^d + \mathbf{E}_2^d \\ \mathbf{E}^d &= -\hat{a}_z \eta H_0 e^{j\beta w/2(\cos\phi' + \cos\phi)} \frac{e^{-j\pi/4}}{2\sqrt{2\pi\beta}} \left[\frac{1}{\cos\left(\frac{\phi - \phi'}{2}\right)} + \frac{1}{\cos\left(\frac{\phi + \phi'}{2}\right)} \right] \frac{e^{-j\beta\rho}}{\sqrt{\rho}} \\ &\quad - \hat{a}_z \eta H_0 e^{-j\beta w/2(\cos\phi' + \cos\phi)} \frac{e^{-j\pi/4}}{2\sqrt{2\pi\beta}} \left[\frac{1}{\cos\left(\frac{\phi - \phi'}{2}\right)} - \frac{1}{\cos\left(\frac{\phi + \phi'}{2}\right)} \right] \frac{1}{\sqrt{\rho}} e^{-j\beta\rho} \\ \mathbf{E}^d &= -\hat{a}_z \eta H_0 \frac{e^{-j\pi/4}}{2\sqrt{2\pi\beta}} \left\{ e^{j\beta w/2(\cos\phi' + \cos\phi)} \left[\frac{1}{\cos\left(\frac{\phi - \phi'}{2}\right)} + \frac{1}{\cos\left(\frac{\phi + \phi'}{2}\right)} \right] \right. \\ &\quad \left. + e^{-j\beta w/2(\cos\phi' + \cos\phi)} \left[\frac{1}{\cos\left(\frac{\phi - \phi'}{2}\right)} - \frac{1}{\cos\left(\frac{\phi + \phi'}{2}\right)} \right] \right\} \frac{1}{\sqrt{\rho}} e^{-j\beta\rho} \end{aligned}$$

Regrouping similar terms,

$$\begin{aligned} \mathbf{E}^d &= -\hat{a}_z \eta H_0 \frac{e^{-j\pi/4}}{\sqrt{2\pi\beta}} \left\{ \frac{1}{\cos\left(\frac{\phi - \phi'}{2}\right)} \left[\frac{e^{j\beta w/2(\cos\phi' + \cos\phi)} + e^{-j\beta w/2(\cos\phi' + \cos\phi)}}{2} \right] \right. \\ &\quad \left. + \frac{1}{\cos\left(\frac{\phi + \phi'}{2}\right)} \left[\frac{e^{j\beta w/2(\cos\phi' + \cos\phi)} - e^{-j\beta w/2(\cos\phi' + \cos\phi)}}{2} \right] \right\} \frac{1}{\sqrt{\rho}} e^{-j\beta\rho} \end{aligned}$$

$$\boxed{\mathbf{E}^d = -\hat{a}_z \eta H_0 \frac{e^{-j\pi/4}}{\sqrt{2\pi\beta}} \left\{ \frac{\cos(\beta w/2(\cos\phi' + \cos\phi))}{\cos\left(\frac{\phi - \phi'}{2}\right)} + \frac{j\sin(\beta w/2(\cos\phi' + \cos\phi))}{\cos\left(\frac{\phi + \phi'}{2}\right)} \right\} \frac{1}{\sqrt{\rho}} e^{-j\beta\rho}}$$

cont'd

13.37 cont'd

Then, the bistatic scattering width is given by,

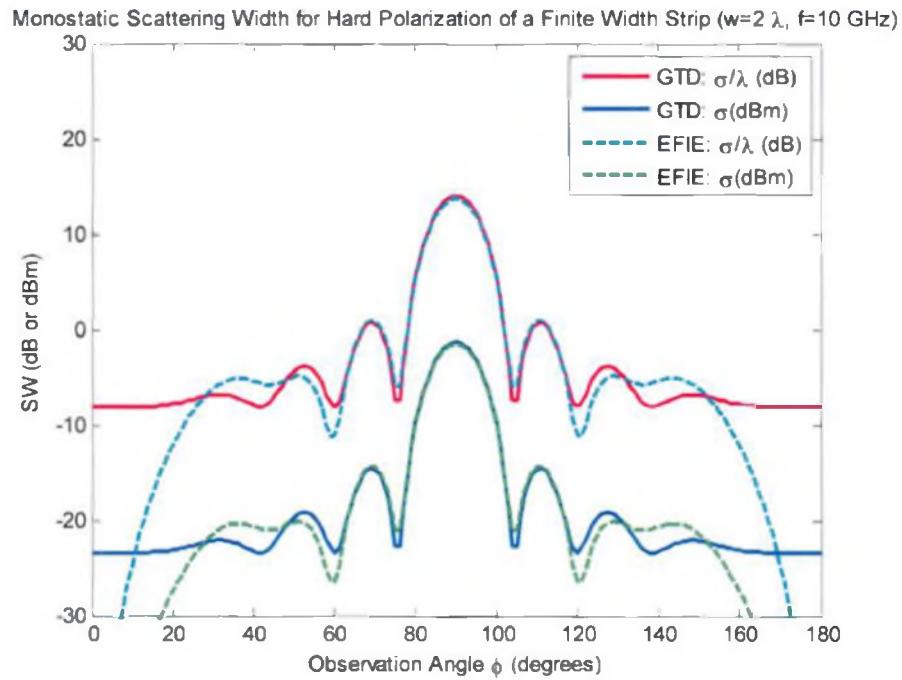
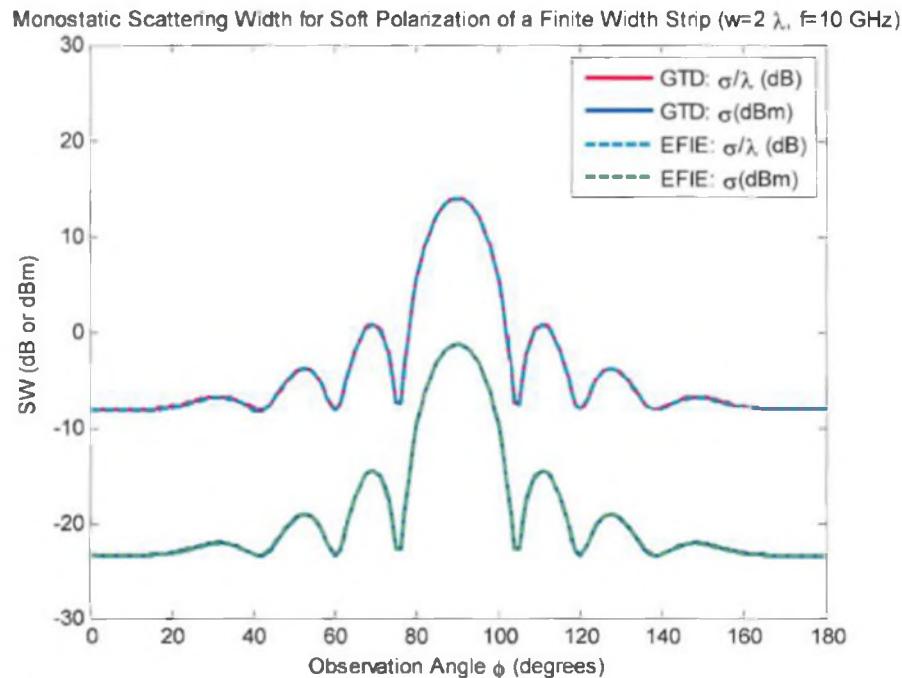
$$\sigma_{2-D} = \lim_{\rho \rightarrow \infty} \left[2\pi\rho \left| \frac{E^s}{E^i} \right|^2 \right]$$

$$\sigma_{2-D} = \lim_{\rho \rightarrow \infty} \left[2\pi\rho \left| \frac{-\hat{d}_z \eta H_0 \frac{e^{-j\frac{\pi}{4}}}{\sqrt{2\pi\beta}} \left\{ \frac{\cos(\beta w/2(\cos\phi' + \cos\phi)) + j\sin(\beta w/2(\cos\phi' + \cos\phi))}{\cos(\frac{\phi - \phi'}{2})} \right\} e^{-j\beta\rho}}{H_0 e^{j\beta(x\cos\phi' + y\sin\phi')}} \right|^2 \right]$$

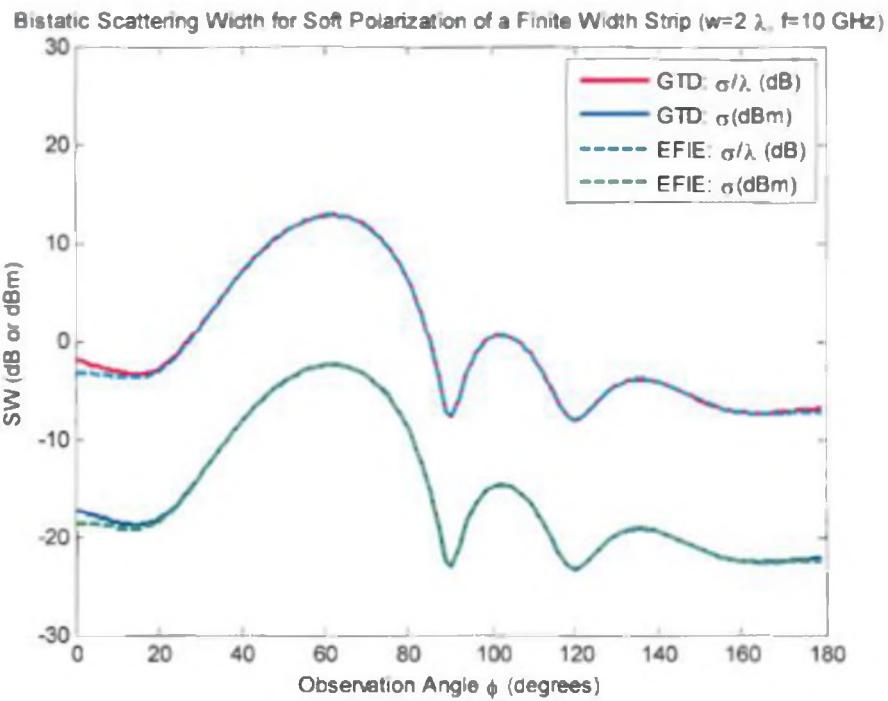
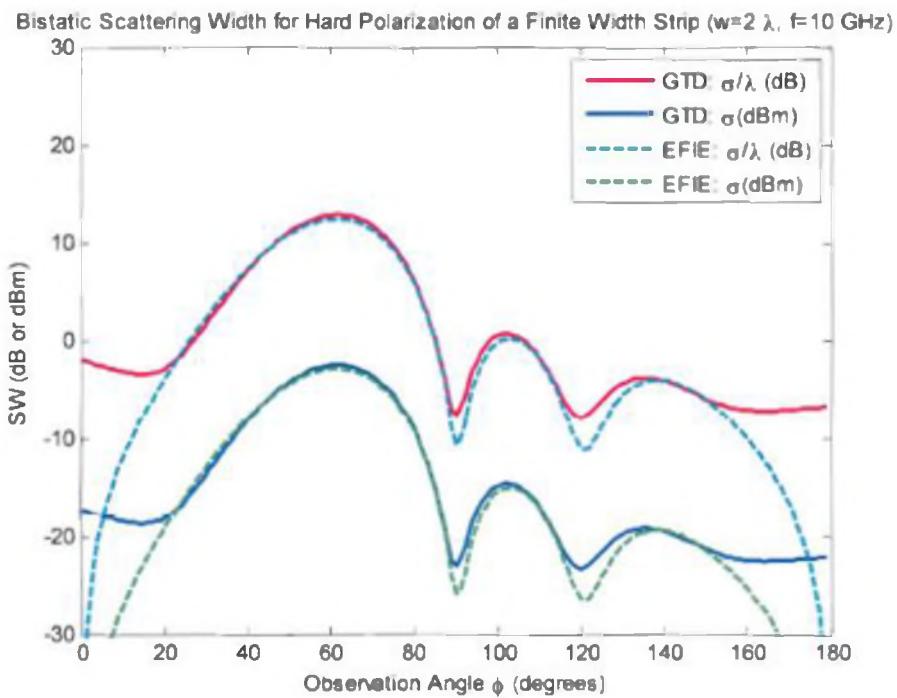
$$\sigma_{2-D} = \lim_{\rho \rightarrow \infty} \left[\frac{1}{\beta} \left| \frac{\cos(\beta w/2(\cos\phi' + \cos\phi)) + j\sin(\beta w/2(\cos\phi' + \cos\phi))}{\cos(\frac{\phi - \phi'}{2})} \right|^2 \right]$$

$$\boxed{\sigma_{2-D} = \frac{\lambda}{2\pi} \left| \frac{\cos(\beta w/2(\cos\phi' + \cos\phi)) + j\sin(\beta w/2(\cos\phi' + \cos\phi))}{\cos(\frac{\phi - \phi'}{2})} \right|^2}$$

cont'd

(c) Scattering Patterns:**Case 1: $w = 2\lambda$, $f = 10$ GHz, $\phi' = 120^\circ$, Monostatic, Hard****Case 2: $w = 2\lambda$, $f = 10$ GHz, $\phi' = 120^\circ$, Monostatic, Soft**

cont'd

Case 3: $w = 2\lambda$, $f = 10$ GHz, $\phi' = 120^\circ$, Bistatic, SoftCase 4: $w = 2\lambda$, $f = 10$ GHz, $\phi' = 120^\circ$, Bistatic, Hard

13.38

$$E^L = \hat{a}_z E_0 e^{+j\beta y} = \hat{a}_z(1) e^{+j\beta y}$$

(a)

$$\underline{E}^d = \underline{E}^L(y=0) \cdot \frac{\tilde{D}}{\sqrt{s}} e^{-j\beta s}$$

$\phi = -330^\circ$

$\phi' = 90^\circ$

$$= \hat{a}_z E^L(y=0) \cdot \hat{a}_z \hat{a}_z \frac{s}{\sqrt{s}} e^{-j\beta s}$$

$$= \hat{a}_z(1) \cdot \hat{a}_z \hat{a}_z \frac{e^{-j\pi/4}}{\sqrt{2\pi\beta}} \frac{1}{n} \sin\left(\frac{\pi}{n}\right) \begin{bmatrix} 1 & & 1 \\ \cos\left(\frac{\pi}{n}\right) - \cos\left(\frac{\phi-\phi'}{n}\right) & \cos\left(\frac{\pi}{n}\right) - \cos\left(\frac{\phi+\phi'}{n}\right) \end{bmatrix} \frac{e^{-j\beta s}}{\sqrt{s}}$$

$$\underline{E}^d = \hat{a}_z e^{-j(Bs + \pi/4)} \begin{bmatrix} 1 & & 1 \\ \cos\left(\frac{\pi}{n}\right) - \cos\left(\frac{\phi-\phi'}{n}\right) & \cos\left(\frac{\pi}{n}\right) - \cos\left(\frac{\phi+\phi'}{n}\right) \end{bmatrix} \begin{matrix} n=2 \\ s=9\lambda \\ \phi=330^\circ \\ \phi'=90^\circ \end{matrix}$$

$$= \hat{a}_z e^{-j\left[\frac{2\pi}{3}(9\lambda) + \frac{\pi}{4}\right]} \begin{bmatrix} 1 & & 1 \\ 0 - j\omega\left(\frac{240}{2}\right) & 0 - j\omega\left(\frac{420}{2}\right) \end{bmatrix}$$

$$= \hat{a}_z \frac{e^{-j\pi/4}}{2\pi(3)(2)} \begin{bmatrix} 1 & 1 \\ (-1/2) & (-\sqrt{3}/2) \end{bmatrix} = +\hat{a}_z \frac{e^{-j\pi/4}}{6\pi(2)} \begin{bmatrix} z & z \\ 1 & \sqrt{3} \end{bmatrix}$$

$$\underline{E}^d = \hat{a}_z \frac{e^{-j\pi/4}}{6\pi} \left\{ 1 - \frac{1}{\sqrt{3}} \right\} = \hat{a}_z \frac{e^{-j\pi/4}}{6\pi} (1 - 0.57735) = \hat{a}_z \frac{e^{-j\pi/4}}{6\pi} (0.42265)$$

$$\underline{E}^d = \hat{a}_z 0.02242 e^{-j\pi/4} = \hat{a}_z 0.01585(1-j)$$

$$\underline{E}^d(\phi=330^\circ, s=5\lambda) = \underbrace{\hat{a}_z 0.02242 e^{-j\pi/4}}_{-j\pi/4} = \underbrace{\hat{a}_z 0.01585(1-j)}$$

(b)

$$\frac{|\underline{E}^d|}{|\underline{E}^L|} = \frac{|0.02242 e^{-j\pi/4}|}{|e^{+j\beta y}|} = 0.02242$$

$$\frac{|\underline{E}^d|}{|\underline{E}^L|} (\text{dB}) = 20 \log_{10}(0.02242) = 20(-1.64936) = -32.987$$

$$18.39 \quad U = U_1^d + U_2^d$$

where U is used to represent the electric field for soft polarization and the magnetic field for the hard polarization.

(a)

$$U_1^d = U_0 \frac{e^{-j\beta p_1}}{\sqrt{2\pi\beta}} \left\{ \frac{e^{-j\pi/4}}{\sqrt{2\pi\beta}} \frac{1}{n} \sin\left(\frac{\pi}{n}\right) \cdot \right.$$

$$\left. \cdot \left[\frac{1}{\cos\left(\frac{\pi}{n}\right) - j\sin\left(\frac{\psi_1 - \psi_2}{n}\right)} \pm \frac{1}{\cos\left(\frac{\pi}{n}\right) + j\sin\left(\frac{\psi_1 + \psi_2}{n}\right)} \right] \right\}_{n=2} \quad \begin{matrix} \psi_2 = \phi \\ \psi_1 = 90^\circ \end{matrix}$$

$$U_1^d = U_0 \frac{e^{-j(\beta p_1 + \pi/4)}}{2\sqrt{2\pi\beta p_1}} \left[\frac{1}{\cos\left(\frac{\psi_1 - 90^\circ}{2}\right)} \pm \frac{1}{\cos\left(\frac{\psi_1 + 90^\circ}{2}\right)} \right]_{\psi_1 = \frac{\pi}{2} - \phi}$$

$$U_2^d = -U_0 \frac{e^{-j(\beta p_2 + \pi/4)}}{2\sqrt{2\pi\beta p_2}} \left[\frac{1}{\cos\left(\frac{\psi_2 - 90^\circ}{2}\right)} \pm \frac{1}{\cos\left(\frac{\psi_2 + 90^\circ}{2}\right)} \right]_{\psi_2 = \phi}$$

$$(b) \quad \omega = 3\lambda, d = 5\lambda \Rightarrow p_1 = p_2 = \sqrt{d^2 + (\omega/2)^2} = \sqrt{(5\lambda)^2 + (3\lambda/2)^2} = 5.22\lambda$$

$$\psi_1 = \psi_2 = 360^\circ - \tan^{-1}(d/\omega/2) = 360^\circ - 73.3^\circ = 286.7^\circ$$

$$U_1^d = -\frac{e^{-j(\frac{\pi}{2} \cdot 5.22\lambda + \pi/4)}}{2(2\pi)\sqrt{5.22}} \left[\frac{1}{\cos\left(\frac{286.7 - 90}{2}\right)} \pm \frac{1}{\cos\left(\frac{286.7 + 90}{2}\right)} \right] = -U_2^d$$

$$U = 2U_1^d = 2U_2^d = -\frac{e^{-j(10.69\pi)}}{2\pi\sqrt{5.22}} \left[\frac{1}{\cos(483.5)} \pm \frac{1}{\cos(188.35)} \right]$$

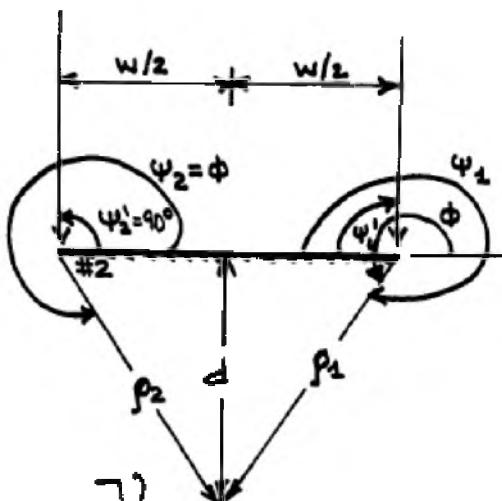
$$= -\frac{e^{-j0.69\pi}}{14.355} \left[-\frac{1}{0.1452} \pm \frac{1}{-0.9894} \right]$$

$$= -\frac{e^{-j124.2^\circ}}{14.355} [-6.886 \mp 1.0107]$$

$$U^t = U_1^d + U_2^d = -0.0697 (-6.886 \mp 1.0107) \underline{-124.2^\circ}$$

$$U^t(\text{soft}) = -0.0697 (-6.886 + 1.0107) \underline{-124.2^\circ} = 0.4045 \underline{-124.2^\circ}$$

$$U^t(\text{hard}) = -0.0697 (-6.886 - 1.0107) \underline{-124.2^\circ} = 0.5504 \underline{-124.2^\circ}$$



13.40

$$U^d = U^i(Q_D) D^{s,h} A e^{-j\beta s}$$

where U represents the electric field for soft polarization and the magnetic field for hard polarization.

$$(a) U^i(Q_D) = \frac{e^{-j\beta r_2}}{r_2}$$

$$A = \sqrt{\frac{s'}{s(s'+s)}} = \sqrt{\frac{r_1}{r_2(r_1+r_2)}}$$

$$A = \sqrt{\frac{r_2}{2r_1}} = \sqrt{\frac{r_2}{2r_2}}$$

$$D^{s,h} = \frac{e^{-j\pi/4}}{\sqrt{2\pi\beta}} \frac{1}{n} \sin\left(\frac{\pi}{n}\right) \left\{ \frac{1}{\cos\left(\frac{\pi}{n}\right) - \cos\left(\frac{\psi-\psi'}{n}\right)} + \frac{1}{\cos\left(\frac{\pi}{n}\right) - \cos\left(\frac{\psi+\psi'}{n}\right)} \right\}_{n=2}$$

$$D^{s,h} = -\frac{e^{-j\pi/4}}{2\sqrt{2\pi\beta}} \left\{ \frac{1}{\cos\left(\frac{\psi-\psi'}{2}\right)} + \frac{1}{\cos\left(\frac{\psi+\psi'}{2}\right)} \right\}$$

$$U_{s,h}^d = -\frac{e^{-j\beta r_1}}{r_1} \frac{e^{-j\pi/4}}{2\sqrt{2\pi\beta}} \left\{ \frac{1}{\cos\left(\frac{\psi-\psi'}{2}\right)} + \frac{1}{\cos\left(\frac{\psi+\psi'}{2}\right)} \right\} \frac{1}{\sqrt{2r_2}} \frac{e^{-j\beta r_2}}{1}$$

$$U_{s,h}^d = -\frac{e^{-j[\beta(r_1+r_2)+\pi/4]}}{4r_1\sqrt{\pi\beta r_2}} \left[\frac{1}{\cos\left(\frac{\psi-\psi'}{2}\right)} + \frac{1}{\cos\left(\frac{\psi+\psi'}{2}\right)} \right]$$

(b) For $r_1 = r_2 = \sqrt{2}d$, $d = 5\lambda$, $h = 5\lambda$, $\psi' = 45^\circ$, $\psi = 315^\circ \Rightarrow \psi - \psi' = 135^\circ$, $\psi + \psi' = 180^\circ$

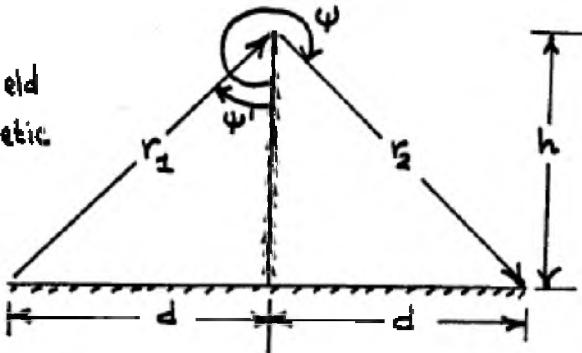
$$U_{s,h}^d = -\frac{e^{-j(2\beta r_2 + \pi/4)}}{4r_2\sqrt{\pi\beta r_2}} \left[\frac{1}{\cos(135^\circ)} + \frac{1}{\cos(180^\circ)} \right] = -\frac{e^{-j(2\beta r_2 + \pi/4)}}{4r_2\sqrt{\pi\beta r_2}} (-1.414 \pm 1)$$

$$\left| \frac{U_{s,h}^d}{U(2d)} \right| = \left| \frac{(-1.414 \pm 1)}{4r_2\sqrt{\pi\beta r_2}} \right| = \left| \frac{(-1.414 \pm 1)(d)}{2r_2\sqrt{\pi\beta r_2}} \right| \Big|_{r_2 = \sqrt{2}5\lambda} = \left| \frac{(-1.414 \pm 1)}{2\sqrt{2}\sqrt{\pi\beta}2d} \right| = \left| \frac{(-1.414 \pm 1)}{4\pi\sqrt{2}\lambda} \right|$$

$$\left| \frac{U_{s,h}^d}{U(2d)} \right| = \left| \frac{(-1.414 \pm 1)}{33.416} \right|$$

$$\left| \frac{U_s^d}{U(2d)} \right| = \left| \frac{-1.414 \pm 1}{33.416} \right| = \frac{0.414}{33.416} = 0.01239 = -38.139 \text{ dB}$$

$$\left| \frac{U_h^d}{U(2d)} \right| = \left| \frac{-1.414 - 1}{33.416} \right| = \frac{2.414}{33.416} = 0.07224 = -22.824 \text{ dB}$$



13.41 The geometry for this problem

(a) is identical to that of Figure 13-32. Therefore the diffraction formulation is identical to that of Example 13-6 except for the following:

1. The incident field on wedge #2 is the opposite of that incident on wedge #1 (in Example 13-6 they were identical). Therefore the diffracted field for this problem will have an additional negative sign.

2. The incident GO field for this problem is given by

$$E^i(\theta_1) = \frac{1}{2} E_{0A} (\theta=\pi/2, r=w/2) = \frac{E_0}{2} \frac{\sin\left(\frac{\beta b}{2}\right)}{\frac{\beta b}{2}} e^{-j\beta w/2}$$

whereas for Example 13-6 was given by

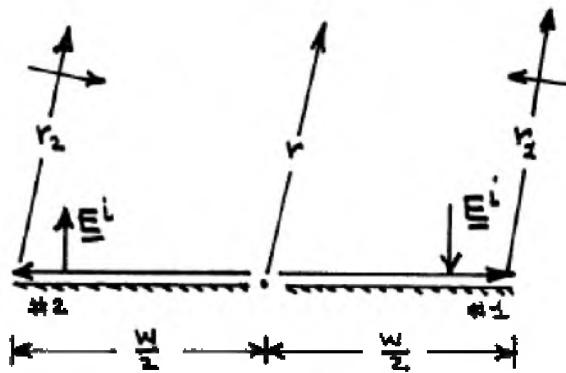
$$E^i(\theta_1) = \frac{1}{2} E_{0A} (\theta=\pi/2, r=w/2) = \frac{E_0}{2} \frac{\sin\left(\frac{\beta b}{2}\right)}{\frac{\beta b}{2}} e^{-j\beta w/2}$$

Taking into account these differences and using the results of Example 13-6 on page 806, we can write for this problem that

$$E_{01}^d = E_0 \frac{\sin\left(\frac{\beta b}{2}\right)}{\frac{\beta b}{2}} V_B^i \left(\frac{w}{2}, \psi_1, n_i=2\right) e^{+j(\beta w/2)\sin\theta} \frac{e^{-j\beta r}}{r}, \quad \psi_1 = \frac{\pi}{2} + \theta, 0 \leq \theta \leq \pi$$

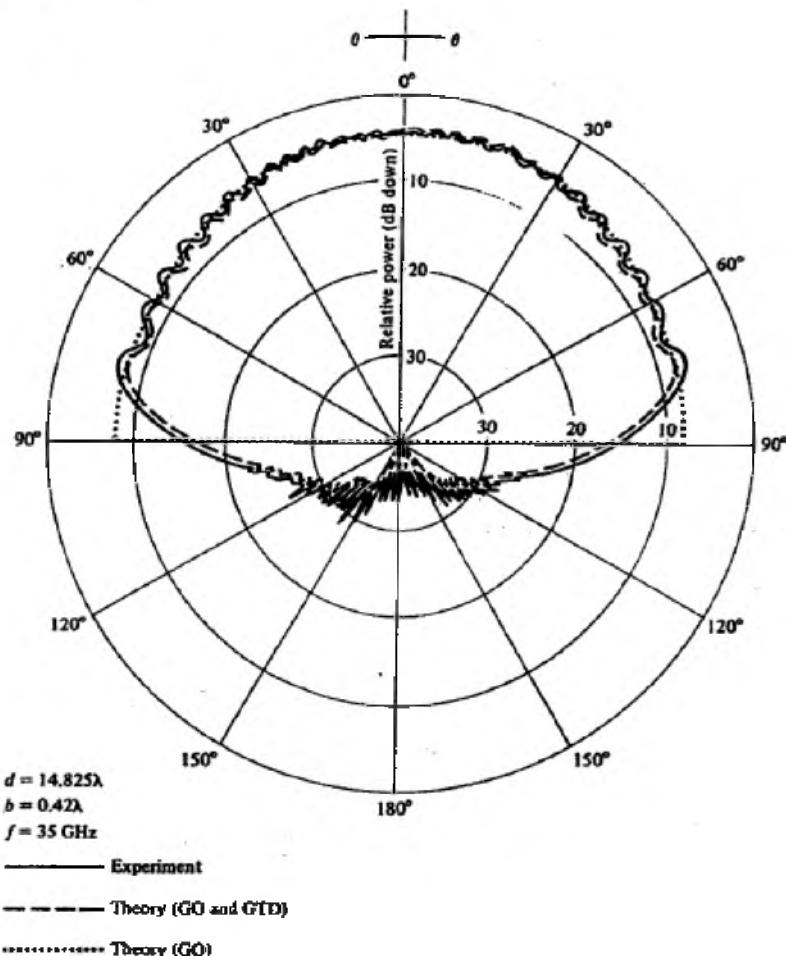
$$E_{02}^d = E_0 \frac{\sin\left(\frac{\beta b}{2}\right)}{\frac{\beta b}{2}} V_B^i \left(\frac{w}{2}, \psi_2, n_i=2\right) e^{-j(\beta w/2)\sin\theta} \frac{e^{-j\beta r}}{r}, \quad \psi_2 = \begin{cases} \frac{\pi}{2} - \theta, & 0 \leq \theta \leq \pi/2 \\ \frac{3\pi}{2} - \theta, & \pi/2 \leq \theta \leq \pi \end{cases}$$

- (b) The normalized amplitude pattern (in decibels) for $0^\circ \leq \theta \leq 180^\circ$ when $w/2 = 14.825$, and $b = 0.42\lambda$ is shown on the next page.

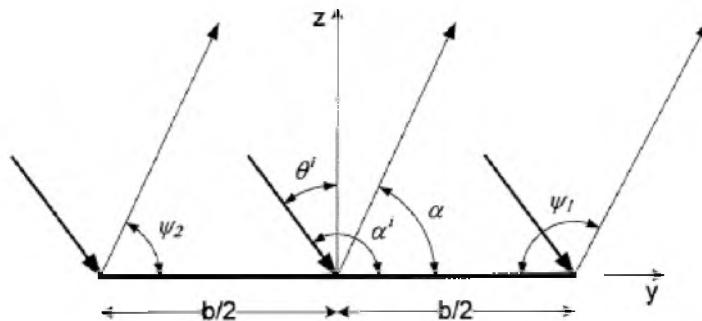


Cont'd.

13.41 cont'd.



13.42 Uniform plane wave incident upon a rectangular ground plane.



$\theta^i \in [0^\circ, 90^\circ]$; $\alpha \in [0^\circ, 180^\circ]$; Plate Dimensions ($a \times b$)

$$\theta^i = \alpha^i - 90^\circ; \quad \psi_1 = 180^\circ - \alpha; \quad \psi_2 = \alpha$$

$$(\sigma_{3-D}) \cong \sigma_{2-D} (2l^2 / \lambda) = \sigma_{2-D} (2a^2 / \lambda)$$

$$\sigma_{2-D} = \lim_{r \rightarrow \infty} \left[2\pi\rho \frac{|E^s|^2}{|E^i|^2} \right] = \lim_{r \rightarrow \infty} \left[2\pi\rho \frac{|H^s|^2}{|H^i|^2} \right]$$

$$E^i = E_0 e^{j\beta(y\cos\alpha^i + z\sin\alpha^i)}; \quad |H^i|^2 = \left(\frac{E_0}{\eta} \right)^2; \quad |E^i|^2 = E_0^2$$

For UTD

Hard polarization	Soft polarization
$ E^s ^2 = E_{hard}^{1st(1,2)} + E_{hard}^{2nd(1,2)} ^2$	$ E^s ^2 = E_{soft}^{1st(1,2)} ^2$

For Physical Optics

Monostatic RCS (soft and hard polarizations)

$$\sigma_{3-D} = 4\pi \left(\frac{ab}{\lambda} \right)^2 \cos^2 \theta_i \left| \frac{\sin(\beta b \sin \theta_i)}{\beta b \sin \theta_i} \right|^2$$

Bistatic RCS (hard/IE^x polarization)

$$\sigma_{3-D} = 4\pi \left(\frac{ab}{\lambda} \right)^2 \cos^2 \theta^s \left| \frac{\sin \left(\frac{\beta b}{2} (\sin \theta^s - \sin \theta^i) \right)}{\frac{\beta b}{2} (\sin \theta^s - \sin \theta^i)} \right|^2$$

Cont'd

13.42 cont'd

Hard polarization (TE^x)

- First-order diffractions from edges #1 and #2 for general bistatic scattering.

Use H field.

$$H_{hard}^{1st(1,2)} = H_{hard}^{d1} = H_1^d + H_2^d$$

Hard polarization

H_1^d $Q1 = (x = 0, y = b/2, z = 0)$	H_2^d $Q2 = (x = 0, y = -b/2, z = 0)$
$H_1^d = \tilde{H}^i(Q1) D_1^h A_1 e^{-j\beta r_1}$	$H_2^d = H^i(Q2) D_2^h A_2 e^{-j\beta r_2}$
$H^i(Q1) = \frac{E_0}{\eta} e^{j\beta(b/2)\cos\alpha'}$	$H^i(Q2) = \frac{E_0}{\eta} e^{-j\beta(b/2)\cos\alpha'}$
$r_1 = r - (b/2)\cos\alpha \quad \text{phase}$	$r_2 = r + (b/2)\cos\alpha \quad \text{phase}$
$r_1 = r \quad \text{amplitude}$	$r_2 = r \quad \text{amplitude}$
$A_1 = \frac{1}{\sqrt{r_1}} = \frac{1}{\sqrt{r}}$	$A_2 = \frac{1}{\sqrt{r_2}} = \frac{1}{\sqrt{r}}$
$A_1 e^{-j\beta r_1} = e^{j\beta(b/2)\cos\alpha} \frac{e^{-j\beta r}}{\sqrt{r}}$	$A_2 e^{j\beta r_2} = e^{-j\beta(b/2)\cos\alpha} \frac{e^{-j\beta r}}{\sqrt{r}}$
Keller $D_1^h(180^\circ - \alpha, 180^\circ - \alpha', 2)$	Keller $D_2^h(\alpha, \alpha', 2)$

$$H_{hard}^{1st(1,2)} = \frac{E_0}{\eta} \left(D_1^h e^{j\beta(b/2)\gamma^+} + D_2^h e^{-j\beta(b/2)\gamma^+} \right) \frac{e^{-j\beta r}}{\sqrt{r}}$$

$$\gamma^+ = \cos\alpha' + \cos\alpha$$

Cont'd

13.42 cont'd

- Second-order bistatic diffractions from edges #1 and #2.

Use H field

$$H_{hard}^{2nd(1,2)} = H_{hard}^{d2} = H_{21r}^d + H_{12r}^d$$

Hard polarization

H_{21r}^d $Q1 = (x = 0, y = b/2, z = 0)$	H_{12r}^d $Q2 = (x = 0, y = -b/2, z = 0)$
$H_{21r}^d = H_{i21}^d D_{21r}^h A_{21r} e^{-j\beta r_1}$	$H_{12r}^d = H_{i12}^d D_{12r}^h A_{12r} e^{-j\beta r_2}$
$r_1 = r - (b/2) \cos \alpha$ phase	$r_2 = r + (b/2) \cos \alpha$ phase
$r_1 = r$ amplitude	$r_2 = r$ amplitude
$A_{21r} = \frac{1}{\sqrt{r_1}} = \frac{1}{\sqrt{r}}$	$A_{12r} = \frac{1}{\sqrt{r_2}} = \frac{1}{\sqrt{r}}$
$A_{21r} e^{-j\beta r_1} = e^{j\beta(b/2) \cos \alpha} \frac{e^{-j\beta r}}{\sqrt{r}}$	$A_{12r} e^{-j\beta r_2} = e^{-j\beta(b/2) \cos \alpha} \frac{e^{-j\beta r}}{\sqrt{r}}$
UTD $D_{21r}^h(b, 180^\circ - \alpha, 0, 90, 2)$	UTD $D_{12r}^h(b, \alpha, 0, 90, 2)$

Cont'd

13.4.2 cont'd]

where

H_{i21}^d $Q2 = (x = 0, y = -b/2, z = 0)$ $H_{i21}^d = H^i(Q2)D_{i21}^h A_{i21} e^{-j\beta r_i}$ $H^i(Q2) = \frac{E_0}{\eta} e^{-j\beta(b/2)\cos\alpha^i}$ $r_2 = b \quad \text{phase, amplitude}$ $A_{i21} = \frac{1}{\sqrt{r_2}} = \frac{1}{\sqrt{b}}$ $A_{i21} e^{-j\beta r_i} = \frac{e^{-j\beta b}}{\sqrt{b}}$ UTD $D_{i21}^h(b, 0, \alpha^i, 90, 2)$	H_{i12}^d $Q1 = (x = 0, y = b/2, z = 0)$ $H_{i12}^d = H^i(Q1)D_{i12}^h A_{i12} e^{-j\beta r_i}$ $H^i(Q1) = \frac{E_0}{\eta} e^{j\beta(b/2)\cos\alpha^i}$ $r_1 = b \quad \text{phase, amplitude}$ $A_{i12} = \frac{1}{\sqrt{r_1}} = \frac{1}{\sqrt{b}}$ $A_{i12} e^{-j\beta r_i} = \frac{e^{-j\beta b}}{\sqrt{b}}$ UTD $D_{i12}^h(b, 0, 180^\circ - \alpha^i, 90, 2)$
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$$H_{21r}^d = \frac{E_0}{\eta} \frac{e^{-j\beta b}}{\sqrt{b}} e^{-j\beta(b/2)\gamma^-} D_{i21}^h D_{21r}^h \frac{e^{-j\beta r}}{\sqrt{r}}$$

$$H_{12r}^d = \frac{E_0}{\eta} \frac{e^{-j\beta b}}{\sqrt{b}} e^{j\beta(b/2)\gamma^-} D_{i12}^h D_{12r}^h \frac{e^{-j\beta r}}{\sqrt{r}}$$

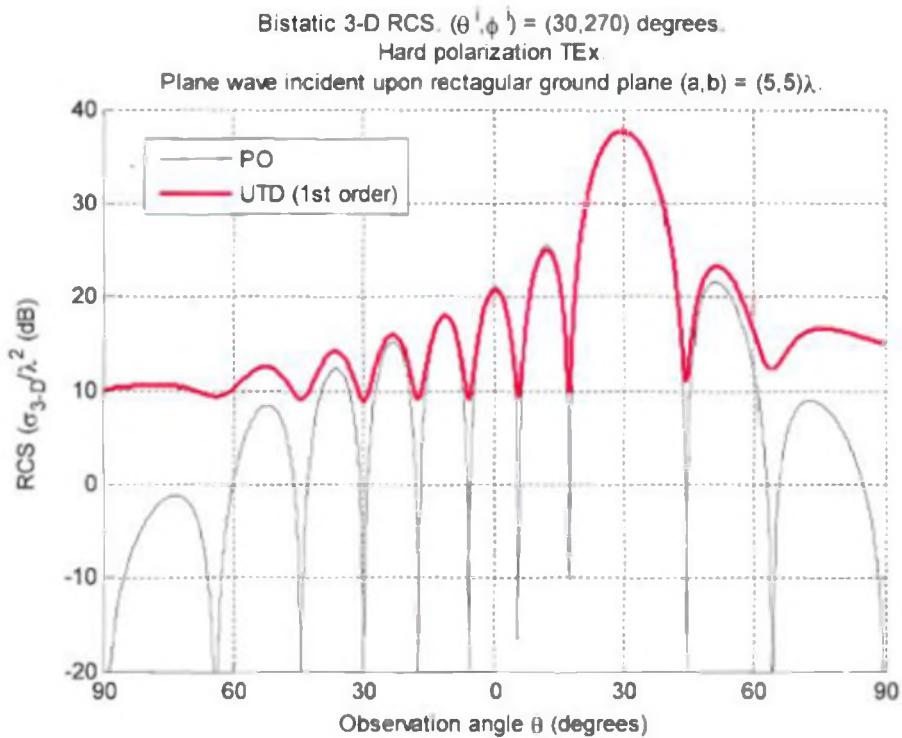
$$H_{hard}^{2nd(1,2)} = \frac{E_0}{\eta} \frac{e^{-j\beta b}}{\sqrt{b}} \left(D_{i21}^h D_{21r}^h e^{-j\beta(b/2)\gamma^-} + D_{i12}^h D_{12r}^h e^{j\beta(b/2)\gamma^-} \right) \frac{e^{-j\beta r}}{\sqrt{r}}$$

$$\gamma^- = \cos\alpha^i - \cos\alpha$$

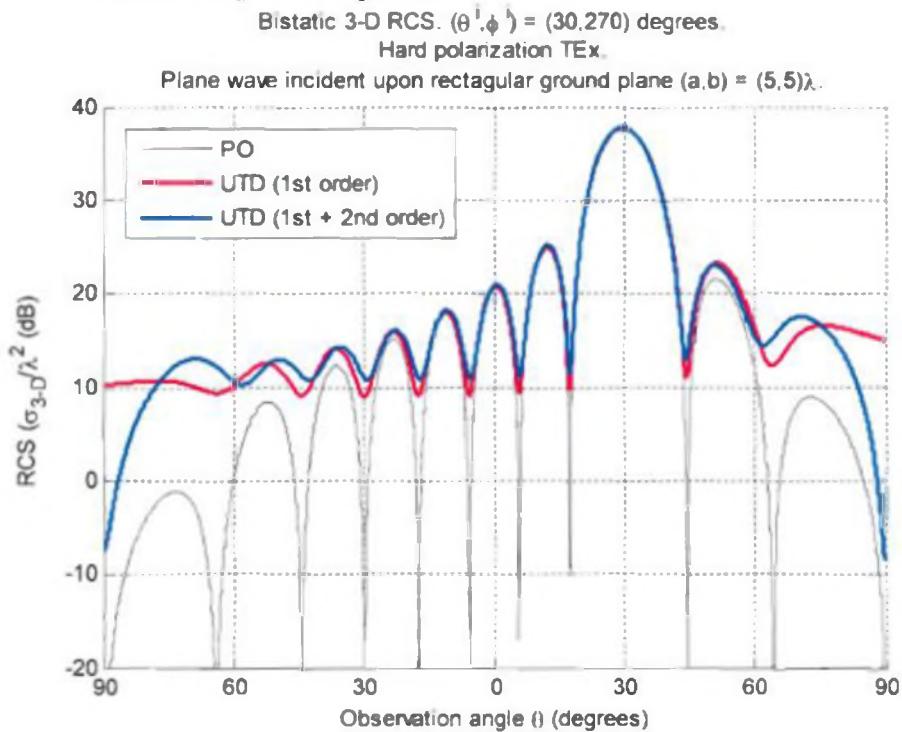
- Scattering bistatic patterns.

cont'd

13.42 cont'd



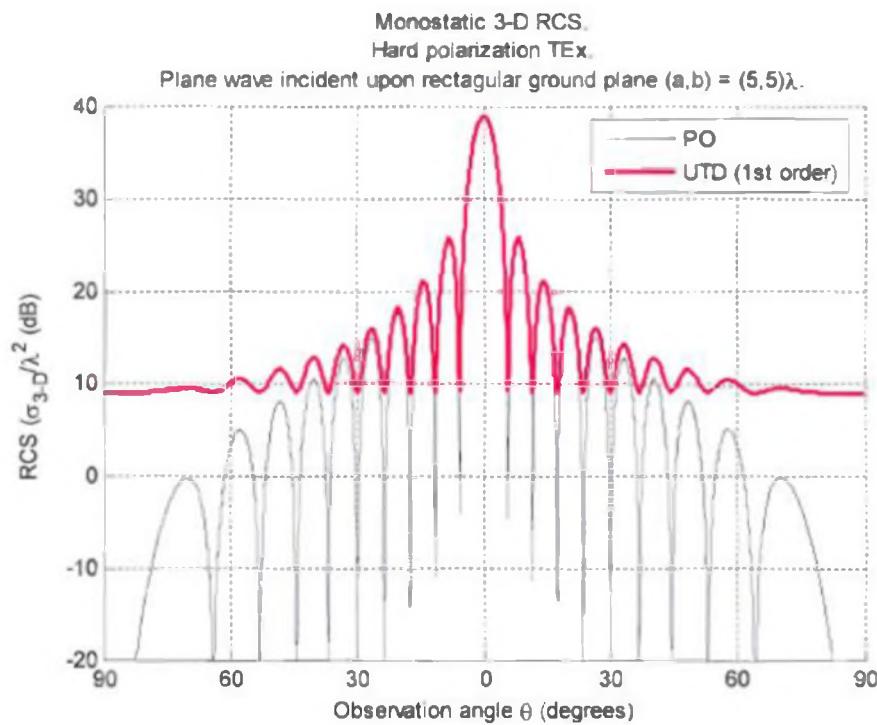
- Scattering bistatic patterns.



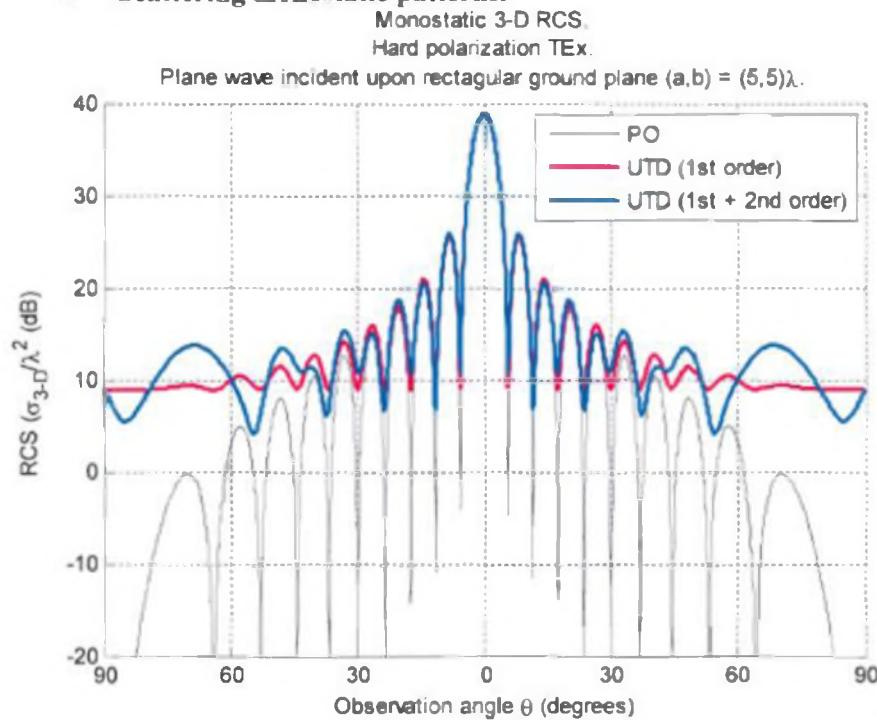
- Scattering monostatic patterns.

Cont'd

13.42 cont'd

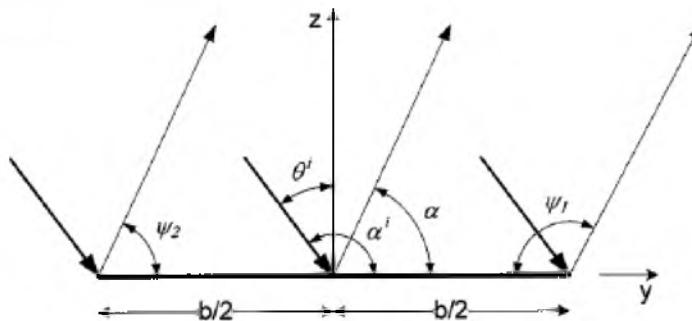


- Scattering monostatic patterns.



cont'd

13.43 Uniform plane wave incident upon a rectangular ground plane.



$\theta^i \in [0^\circ, 90^\circ]$; $\alpha \in [0^\circ, 180^\circ]$; Plate Dimensions ($a \times b$)

$$\theta^i = \alpha^i - 90^\circ; \quad \psi_1 = 180^\circ - \alpha; \quad \psi_2 = \alpha$$

$$(\sigma_{3-D} / \lambda^2) \cong \sigma_{2-D} 2l^2 = \sigma_{2-D} 2a^2$$

$$\sigma_{2-D} = \lim_{r \rightarrow \infty} \left[2\pi r \frac{|E^s|^2}{|E^i|^2} \right] = \lim_{r \rightarrow \infty} \left[2\pi r \frac{|H^s|^2}{|H^i|^2} \right]$$

$$E^i = E_0 e^{j\beta(y\cos\alpha^i + z\sin\alpha^i)}; \quad |H^i|^2 = \left(\frac{E_0}{\eta} \right)^2; \quad |E^i|^2 = E_0^2$$

For UTD

Hard polarization	Soft polarization
$ E^s ^2 = E_{hard}^{1st(1,2)} + E_{hard}^{2nd(1,2)} ^2$	$ E^s ^2 = E_{soft}^{1st(1,2)} ^2$

For Physical optics

Monostatic RCS (soft/TM^x and hard/TE^x polarizations)

$$\sigma_{3-D} = 4\pi \left(\frac{ab}{\lambda} \right)^2 \cos^2 \theta_i \left| \frac{\sin(\beta b \sin \theta_i)}{\beta b \sin \theta_i} \right|^2$$

Bistatic RCS (soft/TM^x polarization)

$$\sigma_{3-D} = 4\pi \left(\frac{ab}{\lambda} \right)^2 \cos^2 \theta^i \left| \frac{\sin\left(\frac{\beta b}{2}(\sin \theta^s - \sin \theta^i)\right)}{\frac{\beta b}{2}(\sin \theta^s - \sin \theta^i)} \right|^2$$

Contd

Soft polarization (TM^x)

- First-order diffractions from edges #1 and #2 for general bistatic scattering.

Use E field

$$E_{hard}^{1st(1,2)} = E_{hard}^{d1} = E_1^d + E_2^d$$

Soft polarization

E_1^d $Q1 = (x = 0, y = b/2, z = 0)$	E_2^d $Q2 = (x = 0, y = -b/2, z = 0)$
$E_1^d = E^i(Q1) D_1^s A_1 e^{-j\beta r_1}$	$E_2^d = E^i(Q2) D_2^s A_2 e^{-j\beta r_2}$
$E^i(Q1) = E_0 e^{j\beta(b/2)\cos\alpha'}$	$E^i(Q2) = E_0 e^{-j\beta(b/2)\cos\alpha'}$
$r_1 = r - (b/2)\cos\alpha \quad \text{phase}$	$r_2 = r + (b/2)\cos\alpha \quad \text{phase}$
$r_1 = r \quad \text{amplitude}$	$r_2 = r \quad \text{amplitude}$
$A_1 = \frac{1}{\sqrt{r_1}} = \frac{1}{\sqrt{r}}$	$A_2 = \frac{1}{\sqrt{r_2}} = \frac{1}{\sqrt{r}}$
$A_1 e^{-j\beta r_1} = e^{j\beta(b/2)\cos\alpha} \frac{e^{-j\beta r}}{\sqrt{r}}$	$A_2 e^{j\beta r_2} = e^{-j\beta(b/2)\cos\alpha} \frac{e^{-j\beta r}}{\sqrt{r}}$
Keller $D_1^s(180^\circ - \alpha, 180^\circ - \alpha^i, 2)$	Keller $D_2^s(\alpha, \alpha^i, 2)$

$$E_{hard}^{1st(1,2)} = E_0 \left(D_1^s e^{j\beta(b/2)\gamma^+} + D_2^s e^{-j\beta(b/2)\gamma^+} \right) \frac{e^{-j\beta r}}{\sqrt{r}}$$

$$\gamma^+ = \cos\alpha^i + \cos\alpha$$

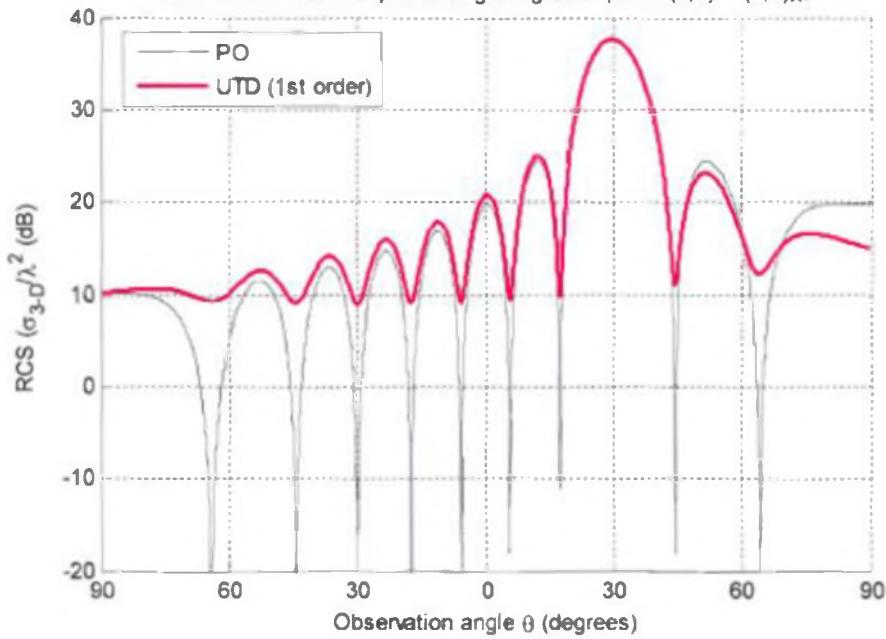
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- Scattering bistatic patterns.**

Bistatic 3-D RCS. $(\theta^i, \phi^i) = (30, 270)$ degrees.

Soft polarization TM_x.

Plane wave incident upon rectangular ground plane $(a, b) = (5, 5)\lambda$.

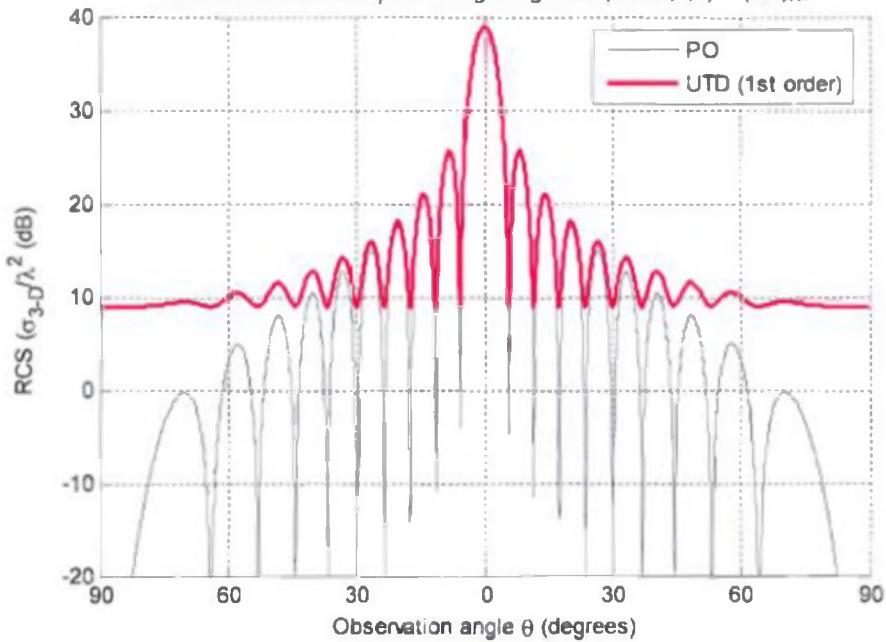


- Scattering monostatic patterns.**

Monostatic 3-D RCS.

Soft polarization TM_x.

Plane wave incident upon rectangular ground plane $(a, b) = (5, 5)\lambda$.



$$13.44 \quad \underline{E}^d(s) = \underline{E}^i(s_i) \cdot \tilde{D} A(s, s) e^{-j\beta s}$$

(a) For a straight edge: $A_s(s, s') = \frac{1}{s}$

For a curved edge: $A_c(s, s') = \frac{1}{s} \sqrt{\rho_c}$

$$\frac{1}{\rho_c} = \frac{1}{\rho_e} - \frac{\hat{n}_e \cdot (\hat{s}' - \hat{s})}{\rho_d} = \frac{1}{s'} - \frac{\hat{n}_e \cdot \hat{s}' - \hat{n}_e \cdot \hat{s}}{a}$$

$$\frac{1}{\rho_c} = \frac{1}{s'} - \frac{\cos(180^\circ) - \cos(0)}{a} = \frac{1}{s'} - \frac{-1-1}{a} = \frac{1}{s'} + \frac{2}{a} = \frac{a+2s'}{as'}$$

$$\frac{A_c(s, s')}{A_s(s, s')} = \frac{\sqrt{\rho_c}/s}{\sqrt{s'}/s} = \frac{\sqrt{\rho_c}}{\sqrt{s'}} = \frac{1}{\sqrt{s'}} \sqrt{\frac{as'}{a+2s'}} = \sqrt{\frac{a}{a+2s'}} \quad | \quad \rho_c = \frac{as'}{a+2s'}$$

$$(b) \frac{A_c(s, s')}{A_s(s, s')} = \sqrt{\left| \frac{a}{a+2s'} \right|} = \sqrt{\frac{25}{25+200}} = \sqrt{\frac{25}{225}} = \sqrt{\frac{1}{9}} = \frac{1}{3} = 20 \log_{10} \left(\frac{1}{3} \right) = 20(-0.47712)$$

$$\frac{A_c(s, s')}{A_s(s, s')} = \frac{1}{3} = 20(-0.47712) = -9.54 \text{ dB}$$

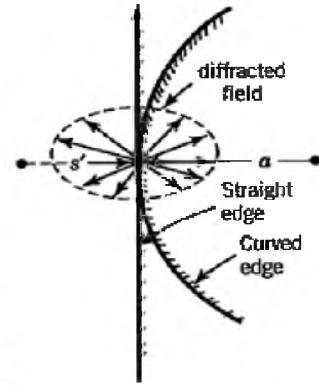


Figure P13-44

13.45 For uniform plane wave illumination,

$$A_s(s, s') \Big|_{\substack{\text{straight} \\ \text{edge}}} = \frac{1}{\sqrt{s}}, \quad A_c(s, s') \Big|_{\substack{\text{curved} \\ \text{edge}}} = \frac{1}{s} \sqrt{\rho_c}$$

$$\frac{1}{\rho_c} = \frac{1}{\rho_e} - \frac{\hat{n} \cdot (\hat{s}' - \hat{s})}{\rho_d \sin \beta_0} \Big|_{\substack{\beta_0 = 90^\circ \\ \rho_e = \infty \text{ (plane wave)}}} = -\frac{\hat{n} \cdot \hat{s}' - \hat{n} \cdot \hat{s}}{a} = -\frac{-\hat{a}_x \cdot \hat{a}_x - [\hat{a}_x \cdot (\hat{a}_x)]}{a} = -\frac{-1-1}{a}$$

$$\frac{1}{\rho_c} = -\frac{-1-1}{a} = \frac{2}{a} \Rightarrow \rho_c = \frac{a}{2}$$

$$\frac{A_c(s, s')}{A_s(s, s')} = \frac{\sqrt{\rho_c}/s}{1/\sqrt{s}} = \sqrt{\frac{\rho_c}{s}} = \sqrt{\frac{a}{2s}}$$

$$\frac{A_c}{A_s} = -20 \text{ dB} = 20 \log_{10} \left(\frac{A_c}{A_s} \right) \Rightarrow \frac{A_c}{A_s} = \frac{1}{20} = \sqrt{\frac{a}{2s}}$$

$$\frac{a}{2s} = \left(\frac{1}{10} \right)^2 = \frac{1}{100} \Rightarrow a = \frac{2s}{100} = \frac{2(50\lambda)}{100} = \lambda$$

$$\boxed{a = \lambda}$$

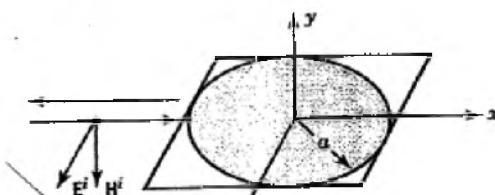


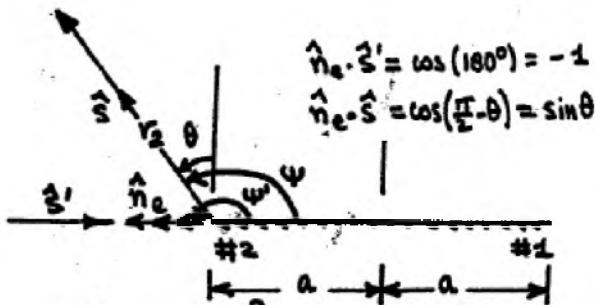
Figure P13-45

13.46

$$(a) U_{sh}^d = \frac{U^d(\theta_2)}{\sqrt{2\pi\beta}} D^{sh} A e^{-j\beta r_2}$$

$$D^{sh} = \frac{e^{-j\pi/4}}{\sqrt{2\pi\beta}} \left\{ \frac{1}{\sin(\frac{\pi}{n})} \left[\frac{1}{\cos(\frac{\pi}{n}) - i\omega(\frac{\psi-\psi'}{n})} + \frac{1}{\cos(\frac{\pi}{n}) - i\omega(\frac{\psi+\psi'}{n})} \right] \right\}_{n=2}$$

$$D^{sh} = -\frac{e^{-j\pi/4}}{2\sqrt{2\pi\beta}} \left\{ \frac{1}{\cos(\frac{\theta-\pi/2}{2})} + \frac{1}{\cos(\frac{\theta+3\pi/2}{2})} \right\}$$



$$A = \sqrt{\frac{\rho_c}{r_2(\rho_c + r_2)}} \xrightarrow{r_2 \gg \rho_c} \frac{\sqrt{\rho_c}}{r_2}$$

$$\frac{1}{\rho_c} = \frac{1}{\rho_c} - \frac{\hat{n}_e \cdot (\hat{s}' - \hat{s})}{\rho_g \sin^2 \theta'_0} \Big|_{\theta'_0 = 90^\circ} = \frac{1}{\rho_c} - \frac{\hat{n}_e \cdot \hat{s}' - \hat{n}_e \cdot \hat{s}}{\rho_g},$$

$$\frac{1}{\rho_c} = \frac{1}{\omega} - \frac{-1 - \sin \theta}{a} = \frac{1 + \sin \theta}{a} \Rightarrow \rho_c = \frac{a}{1 + \sin \theta}$$

$$A = \frac{1}{r_2} \sqrt{\frac{a}{1 + \sin \theta}}$$

$$U_{sh}^d = \frac{e^{-j\pi/4}}{2\sqrt{2\pi\beta}} \sqrt{\frac{a}{1 + \sin \theta}} \left\{ \frac{1}{\cos(\frac{\theta-\pi/2}{2})} + \frac{1}{\cos(\frac{\theta+3\pi/2}{2})} \right\} \frac{e^{-j\beta r_2}}{r_2}$$

- (b) The caustic is located at a point ρ_c to the right of point #2 (i.e., between the diffraction point and the center of the plate). From part a, the caustic is at

$$\rho_c = \frac{a}{1 + \sin \theta}$$

from the edge of wedge #2.

- (c) For $\theta = \pi/2$, $r_2 = 50\lambda$ and $a = 2\lambda$ when $U(\theta_2) = 10^{-3} \text{ V/m}$

$$U_{sh}^d = \frac{10^{-3} e^{-j\pi/4}}{2\sqrt{2\pi\beta}} \sqrt{\frac{a}{2}} \left\{ \frac{1}{\cos(0^\circ)} + \frac{1}{\cos(\pi)} \right\} \frac{e^{-j\beta r_2}}{r_2} = -\frac{10^{-3} e^{-j\pi/4}}{2\sqrt{2\pi\beta}} \sqrt{\frac{a}{2}} \left\{ 1 \pm 1 \right\} \frac{e^{-j\beta r_2}}{r_2}$$

$$U_{sh}^d = -\frac{10^{-3} e^{-j\pi/4}}{2(2\pi)} \sqrt{\frac{a}{2}} \left\{ 1 \pm 1 \right\} \frac{e^{-j\beta r_2}}{r_2}$$

$$U_s^d = -\frac{10^{-3} e^{-j\pi/4}}{2(2\pi)} (2) \frac{e^{-j\beta r_2}}{r_2} = -\frac{10^{-3} e^{-j\pi/4}}{40\pi} = -\frac{10^{-4}}{\pi} e^{-j\pi/4}$$

$$U_h^d = -\frac{10^{-3} e^{-j\pi/4}}{2(2\pi)} \left\{ 1 - 1 \right\} \frac{e^{-j\beta r_2}}{r_2} = 0$$

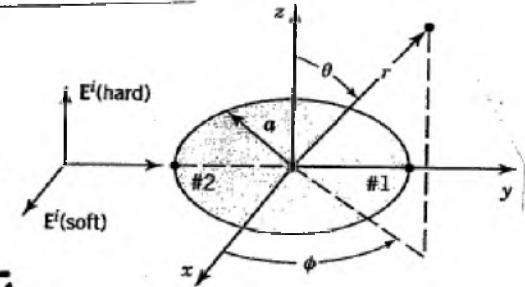


Figure P13-46

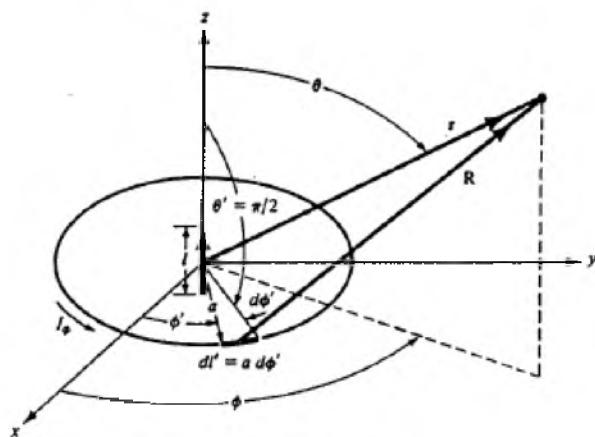
13.47 The vector potential is given by (6-97a) or

$$\underline{A} = \frac{\mu_0}{4\pi} \int_C I_e \frac{e^{-j\frac{\mu_0}{R}l'}}{R} dl' \quad , \quad dl' = ad\phi' \text{ for a circle}$$

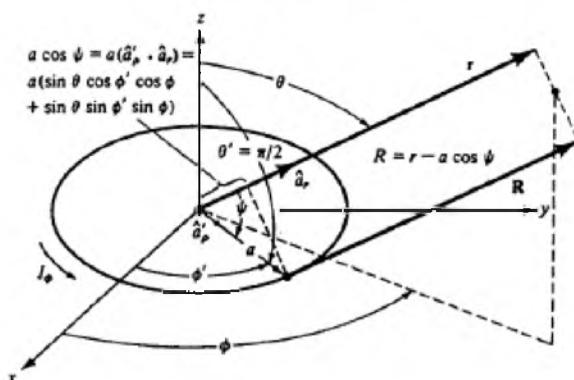
where is represented, in general, by

$$R = [r^2 + a^2 - 2ra \sin\theta \cos(\phi - \phi')]^{1/2} \xrightarrow{r \gg a} r - a \sin\theta \cos(\phi - \phi')$$

R is the distance from any point on the loop to the observation point, r is the distance from the center to the observation point and a is the radius. See figure below for geometry.



(a) Geometry for circular loop



(b) Geometry for far-field observations

Geometrical arrangement for loop antenna analysis.

In general

$$\underline{I}_e(x', y', z') = \hat{a}_x I_x^e + \hat{a}_y I_y^e + \hat{a}_z I_z^e$$

Using the transformation of (II-7a) or (II-7b)

$$I_x^e = I_p \cos\phi' - I_\phi \sin\phi'$$

$$I_y^e = I_p \sin\phi' + I_\phi \cos\phi'$$

$$I_z^e = I_\phi$$

Cont'd.

13.47 Cont'd. and that of (II-19b) for the unit vectors, or

$$\hat{a}_x = \hat{a}_r \sin\theta \cos\phi + \hat{a}_\theta \cos\theta \cos\phi - \hat{a}_\phi \sin\phi$$

$$\hat{a}_y = \hat{a}_r \sin\theta \sin\phi + \hat{a}_\theta \cos\theta \sin\phi + \hat{a}_\phi \cos\phi$$

$$\hat{a}_z = \hat{a}_r \cos\theta - \hat{a}_\theta \sin\theta$$

we can write that

$$\begin{aligned} \underline{I}_e &= \hat{a}_r \left[I_g^e \sin\theta \cos(\phi-\phi') + I_\theta^e \sin\theta \sin(\phi-\phi') + I_z^e \cos\theta \right] \\ &\quad + \hat{a}_\theta \left[I_g^e \cos\theta \cos(\phi-\phi') + I_\theta^e \cos\theta \sin(\phi-\phi') - I_z^e \sin\theta \right] \\ &\quad + \hat{a}_\phi \left[-I_g^e \sin(\phi-\phi') + I_\theta^e \cos(\phi-\phi') \right] \end{aligned}$$

For a circular loop the current is flowing in the ϕ direction (I_ϕ) so that

$$\underline{I}_e = \hat{a}_r I_\phi \sin\theta \sin(\phi-\phi') + \hat{a}_\theta I_\phi \cos\theta \sin(\phi-\phi') + \hat{a}_\phi I_\phi \cos(\phi-\phi')$$

The only component which contributes to the integral for the potential A is the ϕ component; the radial and θ components integrate to zero (see Chapter 5, pp. 232-237 of *Antenna Theory: Analysis and Design* by C.A. Balanis, John Wiley, 2005). Thus

$$A \approx \hat{a}_\phi \frac{j\omega e^{-jBr}}{4\pi r} \int_0^{2\pi} I_\phi^e e^{jB\sin\theta\cos(\phi-\phi')} \sin(\phi-\phi') d\phi$$

The far zone field is obtained by using (6-101a) or

$$E_r^e \approx 0, E_\theta^e \approx -j\omega A_\phi = 0$$

$$E_\phi^e \approx -j\omega A_\phi = -j \frac{\omega e^{-jBr}}{4\pi r} \int_0^{2\pi} I_\phi^e \sin(\phi-\phi') e^{jB\sin\theta\cos(\phi-\phi')} d\phi$$

Making use of the duality theorem as outlined in Table 7-2, we can write that for a magnetic current I_ϕ^m the magnetic field is given by

$$H_r^m \approx H_\theta^m \approx 0$$

$$H_\phi^m = -j \frac{\omega e^{-jBr}}{4\pi r} \int_0^{2\pi} I_\phi^m \cos(\phi-\phi') e^{jB\sin\theta\cos(\phi-\phi')} d\phi$$

Cont'd.

13.47 cont'd. For a uniform current I_ϕ^e and I_ϕ^m are constants. Therefore we can choose any observation point because of symmetry (let $\phi = 0$). Therefore

$$E_\phi^e \approx -j \frac{w h e^{-j\beta r}}{4\pi r} I_\phi^e \times \int_0^{2\pi} e^{j\beta s \sin \theta \cos \phi} \cos \phi' d\phi'$$

Since $\int_0^{2\pi} \cos \phi' e^{j\beta s \sin \theta \cos \phi'} d\phi' = j = \pi J_1(\text{Bessel})$ where $J_1(x)$ is the Bessel function of the first kind of order 1.

Thus $E_\phi^e = \frac{w h e^{-j\beta r}}{2r} I_\phi^e J_1(\text{Bessel})$

Similarly

$$H_\phi^m = \frac{w e^{-j\beta r}}{2r} I_\phi^m J_1(\text{Bessel})$$

13.48 (a) $H_{\phi G} = R(s) e^{j\beta s \cos(\alpha/2) \cos(\theta)}$

The field radiated by the source in the direction of wedge #1 is given by

$$H_{\phi G}^i = R(\theta = \pi - \frac{\alpha}{2}) \frac{e^{-j\beta s}}{s}$$

The field diffracted by wedge #1 is given by

$$H_\phi^d = H_{\phi G}^i D_h(s, \psi, n) A(s, \psi) e^{-j\beta r_1}$$

$$= R(\theta = \pi - \frac{\alpha}{2}) \frac{e^{-j\beta s}}{s} D_h(s, \psi, n) \sqrt{s} e^{-j\beta r_1} = R(\theta = \pi - \frac{\alpha}{2}) \left[D_h(s, \psi, n) \frac{e^{-j\beta s}}{\sqrt{s}} \right] \frac{e^{-j\beta r_1}}{r_1}$$

$$= R(\theta = \pi - \frac{\alpha}{2}) \left[2 D_h^i(s, \psi, n) \frac{e^{-j\beta s}}{\sqrt{s}} \right] \frac{e^{-j\beta r_1}}{r_1} = 2 R(\theta = \pi - \frac{\alpha}{2}) \left[D_h^i(s, \psi, n) \frac{e^{-j\beta s}}{\sqrt{s}} \right] \frac{e^{-j\beta r_1}}{r_1}$$

$$H_\phi^d = 2 R(\theta = \pi - \frac{\alpha}{2}) V_B^i(s, \psi, n) \frac{e^{-j\beta r_1}}{r_1}$$

Since

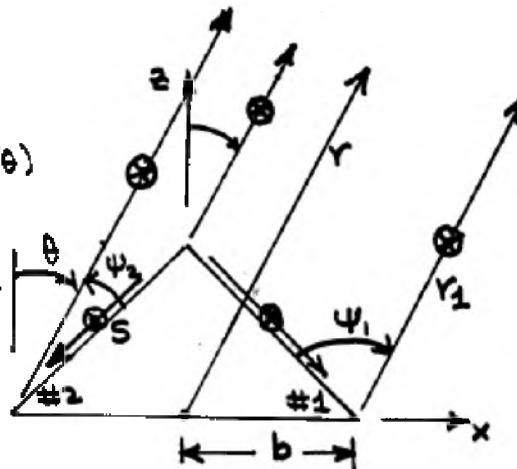
$$r_1 \approx r - b \sin \theta \cos \phi \quad \text{for phase terms}$$

$$r_1 \approx r \quad \text{for amplitude terms}$$

then

$$H_\phi^d = 2 R(\theta = \pi - \frac{\alpha}{2}) V_B^i(s, \psi, n) \frac{e^{-j\beta r}}{r} e^{j\beta b \sin \theta \cos \phi}$$

cont'd.



13.48 cont'd.

$$H_{\phi}^d = 2 R (\theta = \pi - \frac{\alpha}{2}) \frac{e^{-j\beta r}}{r} V_B^L(s, \psi, n) e^{j\beta b \sin \theta \cos \phi'}$$

Since the above is only the contribution from only one point on the rim, the total field is the sum from all points or in the limit the integral around the periphery of the rim. Thus using the form of (13-105b)

$$H_{\phi}^d = 2 R (\theta = \pi - \frac{\alpha}{2}) \frac{e^{-j\beta r}}{r} \int_0^{2\pi} V_B^L(s, \psi, n) \cos \phi' e^{j\beta b \sin \theta \cos \phi'} b d\phi'$$

which by suppressing the $e^{-j\beta r}/r$ factor can be written as

$$H_{\phi}^d = 2 b R (\theta = \pi - \frac{\alpha}{2}) \int_0^{2\pi} V_B^L(s, \psi, n) \cos \phi' e^{j\beta b \sin \theta \cos \phi'} d\phi'$$

(b) Since the integral

$$I(\beta) = \int_0^{2\pi} V_B^L(s, \psi, n) \cos \phi' e^{j\beta b \sin \theta \cos \phi'} d\phi' = \int_0^{2\pi} F(\phi') e^{j\beta f(\phi')} d\phi'$$

$$F(\phi') = V_B^L(s, \psi, n) \cos \phi', f(\phi') = j b \sin \theta \cos \phi'$$

is of the (VI-1) form, then using (VI-18), it can be written as

$$I(\beta) = \sqrt{\frac{2\pi}{\beta}} \sum_s \sqrt{-\frac{1}{f''(\phi'_s)}} F(\phi'_s) e^{j\beta f(\phi'_s)}$$

$$f'(\phi') = -j b \sin \theta \sin \phi' \Rightarrow f'(\phi'_s) = -j b \sin \theta \sin \phi'_s = 0 \Rightarrow \phi'_s = 0, \pi$$

$$\underline{\phi'_s = 0}: \quad \psi = \psi_1 = \frac{\alpha}{2} + \theta, \quad f''(\phi'_s=0) = -j b \sin(\theta) \cos(\theta) = -j b \sin \theta$$

$$F(\phi'_s=0) = V_B^L(s, \frac{\alpha}{2} + \theta, n) \cos(\theta) = V_B^L(s, \frac{\alpha}{2} + \theta, n)$$

$$\underline{\phi'_s = \pi}: \quad \psi = \psi_2 = \frac{\alpha}{2} - \theta \quad f''(\phi'_s=\pi) = -j b \sin \theta \cos(\pi) = +j b \sin \theta$$

$$F(\phi'_s=\pi) = V_B^L(s, \frac{\alpha}{2} - \theta, n) \cos(\pi) = -V_B^L(s, \frac{\alpha}{2} - \theta, n)$$

Thus

$$I(\beta) \approx \sqrt{\frac{2\pi}{\beta}} \left\{ \sqrt{\frac{1}{jb \sin \theta}} V_B^L(s, \frac{\alpha}{2} + \theta, n) e^{j\beta b \sin \theta} - \sqrt{\frac{-1}{jb \sin \theta}} V_B^L(s, \frac{\alpha}{2} - \theta, n) e^{-j\beta b \sin \theta} \right\}$$

$$\approx \sqrt{\frac{2\pi}{\beta b \sin \theta}} \left\{ e^{-j\pi/4} V_B^L(s, \frac{\alpha}{2} + \theta, n) e^{j\beta b \sin \theta} - e^{+j\pi/4} V_B^L(s, \frac{\alpha}{2} - \theta, n) e^{-j\beta b \sin \theta} \right\}$$

Therefore the diffracted field can be written as

$$H_{\phi}^d = 2 b R (\theta = \pi - \frac{\alpha}{2}) e^{j\pi/4} \sqrt{\frac{2\pi}{\beta b \sin \theta}} \left\{ V_B^L(s, \frac{\alpha}{2} + \theta, n) e^{j\beta b \sin \theta} - j V_B^L(s, \frac{\alpha}{2} - \theta, n) e^{-j\beta b \sin \theta} \right\}$$

The first term represents the diffraction contribution from edge #1 while the second is the contribution from edge #2.

13.49 For the solution of this problem refer to Figure 13-37

$$\left. \begin{aligned} E_1^d &= E^i(\theta_1) D_h^2(L, \xi_1, n=2) \sqrt{\frac{\rho_{c1}}{r_1(\rho_{c1} + r_1)}} e^{-j\beta r_1} \\ E_2^d &= -E^i(\theta_2) D_h^2(L, \xi_2, n=2) \sqrt{\frac{\rho_{c2}}{r_2(\rho_{c2} + r_2)}} e^{j\beta r_2} \end{aligned} \right\} 0 \leq \theta \leq \pi$$

where $E^i(\theta_1) = E^i(\theta_2) = \frac{1}{2} E_{G_0}(r=a, \theta=\pi/2) = \frac{1}{2} \frac{\cos\left(\frac{\pi}{2}\cot\theta\right)}{\sin\theta} \Big|_{\theta=\pi/2} = \frac{1}{2} \frac{e^{-j\beta a}}{a}$

θ_1 and θ_2 are the diffraction points (#1 and #2).

According to (13-68)-(13-71d) for $n=2$

$$D_h^2(L, \xi_1, 2) = D^L(L, \xi_1^-, 2) + D^R(L, \xi_1^+, 2)$$

$$D^L(L, \xi_1^-, 2) = -\frac{e^{-j\pi/4}}{4\sqrt{2\pi\beta}} \left\{ C_1^+(\xi_1^-, 2) F_1[\beta L g_1^+(\xi_1^-)] + C_1^-(\xi_1^-, 2) F_1[\beta L g_1^-(\xi_1^-)] \right\}$$

$$D^R(L, \xi_1^+, 2) = -\frac{e^{-j\pi/4}}{4\sqrt{2\pi\beta}} \left\{ C_1^+(\xi_1^+, 2) F_1[\beta L g_1^+(\xi_1^+)] + C_1^-(\xi_1^+, 2) F_1[\beta L g_1^-(\xi_1^+)] \right\}$$

where $C_1^+(\xi_1, 2) = \cot\left(\frac{\pi+\xi_1}{4}\right)$, $C_1^-(\xi_1, 2) = \cot\left(\frac{\pi-\xi_1}{4}\right)$

$$F_1[\beta L g_1^+(\xi)] = 2j \sqrt{\beta L g_1^+(\xi)} e^{j\beta L g_1^+(\xi)} \int_{-\infty}^{\infty} e^{-j\tau^2} d\tau$$

$$\sqrt{\beta L g_1^+(\xi)}$$

$$\xi_1^+ = \psi_1 + \psi_1' = \psi_1 = \pi/2 + \theta \quad g_1^+ = 1 + [\cos(\psi_1) - z(2)\pi N^+]$$

$$\xi_1^- = \psi_1 - \psi_1' = \psi_1 = \pi/2 + \theta \quad g_1^- = 1 + [\cos(\psi_1) - z(2)\pi N^-]$$

with N^+ and N^- being positive or negative integers or zero which most closely satisfy the equations

$$z(2)\pi N^+ - \psi_1 = +\pi \quad \text{and} \quad z(2)\pi N^- - \psi_1 = -\pi$$

For curved edge diffraction, the distance parameter L (see Reference[10]) is given for far-field observations by

$$L = \sin\theta \sin^2 \beta_0 \Big|_{\beta_0=\pi/2} = a \sin\theta$$

$$D_h^2(L, \xi_1, 2) = D^L(L, \xi_1^-, 2) + D^R(L, \xi_1^+, 2)$$

$$D^L(L, \xi_1^-, 2) = -\frac{e^{-j\pi/4}}{4\sqrt{2\pi\beta}} \left\{ C_2^+(\xi_1^-, 2) F_2[\beta L g_2^+(\xi_1^-)] + C_2^-(\xi_1^-, 2) F_2[\beta L g_2^-(\xi_1^-)] \right\}$$

$$D^R(L, \xi_1^+, 2) = -\frac{e^{-j\pi/4}}{4\sqrt{2\pi\beta}} \left\{ C_2^+(\xi_1^+, 2) F_2[\beta L g_2^+(\xi_1^+)] + C_2^-(\xi_1^+, 2) F_2[\beta L g_2^-(\xi_1^+)] \right\}$$

cont'd.

13.49 Cont'd. where $C_2^+(\xi, z) = \cot\left(\frac{\pi + \xi}{4}\right)$, $C_2^-(\xi, z) = \cot\left(\frac{\pi - \xi}{4}\right)$

$$F_2[\beta L g^+(\xi)] = 2j \sqrt{\beta L g^+(\xi)} e^{j\beta L g^+(\xi)} \int_{-\infty}^{\infty} e^{-j\eta^2} d\eta$$

$$\begin{aligned} \xi_2^+ &= \psi_2 + \psi_2' = \psi_2 \\ \xi_2^- &= \psi_2 - \psi_2' = \psi_2 \\ g_2^+ &= 1 + [2\cos\psi_2 - 2(2)\pi N^+] \\ g_2^- &= 1 + [2\cos\psi_2 - 2(2)\pi N^-] \end{aligned} \quad \left. \begin{array}{l} \psi_2 = \frac{\pi}{2} - \theta, \quad 0 \leq \theta \leq \pi/2 \\ \psi_2 = \frac{3\pi}{2} - \theta, \quad \pi/2 \leq \theta \leq \pi \end{array} \right.$$

with N^+ or N^- being a positive or negative integer or zero which most closely satisfies the equation

$$2(2)\pi N^+ - \psi_2 = +\pi \quad \text{and} \quad 2(2)\pi N^- - \psi_2 = -\pi$$

For far field observations

$$r_1 \approx r - a \cos\left(\frac{\pi}{2} - \theta\right) = r - a \sin\theta \quad \left. \begin{array}{l} \text{For phase terms} \\ \text{For amplitude terms} \end{array} \right\}$$

$$r_2 \approx r + a \cos\left(\frac{\pi}{2} - \theta\right) = r + a \sin\theta$$

$$r_1 \approx r_2 \approx r \quad \text{For amplitude terms}$$

ρ_{c1} and ρ_{c2} can be derived by using (13-100) and (13-100a). They are given on pages 821-822 as

$$\rho_{c1} = \frac{a}{\sin\theta} \Rightarrow A_1 = \frac{\sqrt{\rho_{c1}}}{r_1} = \frac{1}{r_1} \sqrt{\frac{a}{\sin\theta}}$$

$$\rho_{c2} = -\frac{a}{\sin\theta} \Rightarrow A_2 = \frac{\sqrt{\rho_{c2}}}{r_2} = \frac{1}{r_2} \sqrt{-\frac{a}{\sin\theta}}$$

To model the ground plane as a ring radiator, the equivalent current concept of Section 13.3.5 will be used. Toward $\theta = 0, \pi$

$$\rho_{c1} = \frac{a}{0} = \infty \Rightarrow A_1 = \frac{1}{r_1} \sqrt{\frac{a}{0}} = \infty$$

Therefore

$$\sqrt{\frac{\rho_{c1}}{r_1(r_1 + r)}} \underset{r \rightarrow \infty}{\approx} \sqrt{\frac{\rho_{c1}}{r(r_0 + r)}} = \frac{1}{\sqrt{r}} \sqrt{\frac{\rho_{c1}}{\rho_0 + r}} = \frac{1}{\sqrt{r}} \sqrt{\frac{1}{1 + \frac{\rho_0}{\rho_{c1}}}} \xrightarrow{\rho_0 \rightarrow \infty} \frac{1}{\sqrt{r}}$$

and

$$E_1^d = E^d(\theta_1) D_h^1(L, \xi_1, z) \sqrt{\frac{\rho_{c1}}{r_1(r_0 + r_1)}} e^{-j\beta r_1} \underset{r \rightarrow \infty}{\approx} E^d(\theta_1) D_h^1(L, \xi_1, z) \sqrt{\frac{\rho_0}{r(r_0 + r)}} e^{-j\beta r}$$

which for $\theta \approx 0, \pi$ reduces to

$$E_1^d|_{\theta \approx 0} \underset{\text{cont'd.}}{\approx} E^d(\theta_1) D_h^1(L, \xi_1, z) \Big|_{\theta=0, r} e^{-j\beta r}; \quad E_2^d|_{\theta \approx \pi} \underset{\text{cont'd.}}{\approx} E^d(\theta_1) D_h^1(L, \xi_1, z) \Big|_{\theta=\pi, r} e^{-j\beta r}$$

cont'd.

13.49 Cont'd.) and each has the same form as the left side of (13-102b).

Equating each of the above to the right side of (13-101b) or (13-102b), we can write that

$$E_{\text{axial}} = E_1^i(\theta=0, \pi) \approx E^i(\theta_1) D_h^2(L, \xi, z) \Big|_{\substack{\theta=0 \\ \theta=\pi}} \frac{e^{-j\beta r}}{\sqrt{r}} = I_e^m \frac{\beta}{2\eta} \sqrt{\frac{z}{2\pi\beta}} \frac{e^{-j\beta r}}{\sqrt{r}}$$

where I_e^m is the equivalent magnetic current, and it is equal to

$$I_e^m = 2\eta \sqrt{\frac{2\pi}{\beta}} E^i(\theta_1) D_h^2(L, \xi, z) \Big|_{\substack{\theta=0, \pi}} = 2\eta \sqrt{\frac{2\pi}{\beta}} e^{-j\pi/4} E^i(\theta_1) D_h^2(L, \xi, z) \Big|_{\theta=0, \pi}$$

which is identical in form to (13-103b).

Since the equivalent current is distributed along a circular ring at the edge of the ground plane, the field radiated by this equivalent current is given by (13-104b), or

$$E_g^m = j \frac{a w e e^{-j\beta r}}{4\pi r} \int_0^{2\pi} I_e^m(\phi') \cos(\phi - \phi') e^{j\beta a \sin\theta \cos(\phi-\phi')} d\phi'$$

The ground plane is symmetrical, and the diffractions at any point along the rim are identical to those of any other point. Thus the current can be assumed to be uniform (with respect to ϕ'). Also the observations are not a function of ϕ . Choosing $\phi = 0$, reduces to

$$E_g^m = - \frac{a w e e^{-j\beta r}}{2r} J_1(\beta a \sin\theta)$$

which is of the form of (13-106b). Using the equivalent current from above, we can write the diffracted field due to the ring source of the ground plane as

$$E_g^m = - \frac{a w e e^{-j\beta r}}{2r} 2\eta \sqrt{\frac{2\pi}{\beta}} e^{-j\pi/4} E^i(\theta_1) D_h^2(L, \xi, z) \Big|_{\theta=0, \pi} J_1(\beta a \sin\theta)$$

$$E_g^m = - a \sqrt{2\pi\beta} e^{-j\pi/4} E^i(\theta_1) D_h^2(L, \xi, z) \Big|_{\theta=0, \pi} J_1(\beta a \sin\theta) \frac{e^{-j\beta r}}{r}$$

- (b) The pattern computed using the equivalent current near the axis ($\theta=0^\circ$ and 180°) of the ground plane when $d=4.0642$ is shown plotted in Figure 13-39 where it is combined with that of the 2-point diffraction (which is used for points removed from the axis).

Formulation

2-D monostatic scattering width of the wedge for any included angle WA for both soft and hard polarization

Soft polarization

Incident field

$$\bar{E}^i = \hat{a}_z e^{-j\beta x} \quad (1)$$

Diffracted field

$$\bar{E}^d = \bar{E}^i(Q) \cdot D^s A e^{-j\beta\rho} \quad (2)$$

where

$$\bar{E}^i(Q) = \bar{E}^i(x=0) = \hat{a}_z 1 \quad (3)$$

$$A = \frac{1}{\sqrt{\rho}} \quad (4)$$

$$D^s = \frac{e^{-j\pi/4} \sin\left(\frac{\pi}{4}\right)}{n\sqrt{2\pi\beta}} \left[\frac{1}{\cos\left(\frac{\pi}{n}\right) - \cos\left(\frac{\phi-\phi'}{n}\right)} - \frac{1}{\cos\left(\frac{\pi}{n}\right) - \cos\left(\frac{\phi+\phi'}{n}\right)} \right] \quad (5)$$

For monostatic scattering

$$\phi = \phi' \quad (6)$$

then,

$$D^s = \frac{e^{-j\pi/4} \sin\left(\frac{\pi}{4}\right)}{n\sqrt{2\pi\beta}} \left[\frac{1}{\cos\left(\frac{\pi}{n}\right) - 1} - \frac{1}{\cos\left(\frac{\pi}{n}\right) - \cos\left(\frac{2\phi'}{n}\right)} \right] \quad (7)$$

Since

$$\sigma_{2-D} = \lim_{\rho \rightarrow \infty} \left[2\pi\rho \frac{|\bar{E}^s|^2}{|\bar{E}^i|^2} \right] \quad (8)$$

and

$$\vec{E}^s = \vec{E}^d \quad (9)$$

then,

$$\begin{aligned}\sigma_{2-D} &= 2\pi\rho \left| D^s \frac{1}{\sqrt{\rho}} e^{-j\beta\rho} \right|^2 = 2\pi |D^s|^2 \\ \sigma_{2-D} &= \frac{\lambda}{n^2 2\pi} \left| \sin\left(\frac{\pi}{4}\right) \left(\frac{1}{\cos\left(\frac{\pi}{n}\right) - 1} - \frac{1}{\cos\left(\frac{\pi}{n}\right) - \cos\left(\frac{2\phi'}{n}\right)} \right) \right|^2\end{aligned}\quad (10)$$

Finally, for the **soft polarization**

$$\boxed{\frac{\sigma_{2-D}}{\lambda} = \frac{1}{n^2 2\pi} \left| \sin\left(\frac{\pi}{4}\right) \left(\frac{1}{\cos\left(\frac{\pi}{n}\right) - 1} - \frac{1}{\cos\left(\frac{\pi}{n}\right) - \cos\left(\frac{2\phi'}{n}\right)} \right) \right|^2}\quad (11)$$

Hard polarization

For the hard polarization we should choose \vec{H} field instead of \vec{E} and then follow the same procedure as for the soft polarization to obtain

$$\boxed{\frac{\sigma_{2-D}}{\lambda} = \frac{1}{n^2 2\pi} \left| \sin\left(\frac{\pi}{4}\right) \left(\frac{1}{\cos\left(\frac{\pi}{n}\right) - 1} + \frac{1}{\cos\left(\frac{\pi}{n}\right) - \cos\left(\frac{2\phi'}{n}\right)} \right) \right|^2}\quad (12)$$

Formulation

2-D monostatic scattering width of the cylinder for any radius a for both soft and hard polarization

Soft polarization

Incident field

$$\bar{E}^i = \hat{a}_z E_o \sum_{n=-\infty}^{\infty} j^{-n} J_n(\beta\rho) e^{jn\phi} \quad (1)$$

Scattered field

$$\bar{E}^s = \hat{a}_z E_o \sum_{n=-\infty}^{\infty} c_n H_n^{(2)}(\beta\rho) \quad (2)$$

Use the boundary condition

$$\bar{E}^r(\rho = a, 0 \leq \phi \leq 2\pi) = \bar{E}^i + \bar{E}^s = 0 \quad (3)$$

to solve for the unknown coefficient

$$E_o \sum_{n=-\infty}^{\infty} \left(j^{-n} J_n(\beta\rho) e^{jn\phi} + c_n H_n^{(2)}(\beta\rho) \right) = 0 \quad (4)$$

then,

$$c_n = -j^{-n} \frac{J_n(\beta a)}{H_n^{(2)}(\beta a)} e^{jn\phi} \quad (5)$$

The scattered field is

$$\bar{E}^s = -\hat{a}_z E_o \sum_{n=-\infty}^{\infty} j^{-n} \frac{J_n(\beta a)}{H_n^{(2)}(\beta a)} H_n^{(2)}(\beta\rho) e^{jn\phi} \quad (6)$$

Use the far-zone approximation

$$H_n^{(2)}(\beta\rho) \xrightarrow{\beta\rho \rightarrow \infty} \sqrt{\frac{2j}{\pi\beta\rho}} j^n e^{-j\beta\rho} \quad (7)$$

then,

$$\bar{E}^s = -\hat{a}_z E_o \sqrt{\frac{2j}{\pi\beta}} \frac{e^{-j\beta\rho}}{\sqrt{\rho}} \sum_{n=-\infty}^{\infty} \frac{J_n(\beta a)}{H_n^{(2)}(\beta a)} e^{jn\phi} \quad (8)$$

Thus,

$$\sigma_{2-D} = \lim_{\rho \rightarrow \infty} \left[2\pi\rho \frac{|\bar{E}^s|^2}{|\bar{E}^i|^2} \right] = 2\pi \frac{2}{\pi\beta} \left| \sum_{n=-\infty}^{\infty} \frac{J_n(\beta a)}{H_n^{(2)}(\beta a)} e^{jn\phi} \right|^2 \quad (9)$$

For the monostatic scattering

$$\phi = \phi' \quad (10)$$

then,

$$\sigma_{2-D} = \frac{2\lambda}{\pi} \left| \sum_{n=-\infty}^{\infty} \frac{J_n(\beta a)}{H_n^{(2)}(\beta a)} e^{jn\phi'} \right|^2 \quad (11)$$

Finally, for the **soft polarization**

$$\frac{\sigma_{2-D}}{\lambda} = \frac{2}{\pi} \left| \sum_{n=0}^{\infty} \varepsilon_n \frac{J_n(\beta a)}{H_n^{(2)}(\beta a)} \cos(n\phi') \right|^2 \quad (12)$$

where

$$\varepsilon_n = \begin{cases} 1 & n = 0 \\ 2 & n \neq 0 \end{cases} \quad (13)$$

Hard polarization

Use \bar{H} field instead of \bar{E} .

Incident field

$$\hat{H}^i = \hat{a}_z H_o \sum_{n=-\infty}^{\infty} j^{-n} J_n(\beta\rho) e^{jn\phi} \quad (1)$$

Scattered field

$$\bar{H}^s = \hat{a}_z H_o \sum_{n=-\infty}^{\infty} d_n H_n^{(2)}(\beta\rho) \quad (2)$$

Use the boundary condition

$$\bar{E}_\phi^t(\rho=a, 0 \leq \phi \leq 2\pi) = \bar{E}_\phi^i + \bar{E}_\phi^s = 0 \quad (3)$$

where

$$\bar{E}_\phi^s = -\frac{1}{j\omega\varepsilon} \frac{\partial \bar{H}_z^s}{\partial \rho} = -\hat{a}_z \frac{\beta H_o}{j\omega\varepsilon} \sum_{n=-\infty}^{\infty} d_n H_n^{(2)}(\beta\rho) \quad (4)$$

$$\bar{E}_\phi^i = -\frac{1}{j\omega\varepsilon} \frac{\partial \bar{H}_z^i}{\partial \rho} = -\hat{a}_z \frac{\beta H_o}{j\omega\varepsilon} \sum_{n=-\infty}^{\infty} j^{-n} J_n'(\beta\rho) e^{jn\phi}$$

Solve for the unknown coefficients

$$-\frac{\beta H_o}{j\omega\varepsilon} \sum_{n=-\infty}^{\infty} \left(j^{-n} J_n'(\beta\rho) e^{jn\phi} + d_n H_n^{(2)}(\beta\rho) \right) = 0 \quad (5)$$

then,

$$d_n = -j^{-n} \frac{J_n'(\beta a)}{H_n^{(2)}(\beta a)} e^{jn\phi} \quad (6)$$

The scattered field is

$$\bar{H}^s = -\hat{a}_z H_o \sum_{n=-\infty}^{\infty} j^{-n} \frac{J_n'(\beta a)}{H_n^{(2)}(\beta a)} H_n^{(2)}(\beta\rho) e^{jn\phi} \quad (7)$$

Use the far-zone approximation

$$H_n^{(2)}(\beta\rho) \xrightarrow{\beta\rho\rightarrow\infty} \sqrt{\frac{2j}{\pi\beta\rho}} j^n e^{-j\beta\rho} \quad (8)$$

then,

$$\bar{H}^s = -\hat{a}_z H_o \sqrt{\frac{2j}{\pi\beta}} \frac{e^{-j\beta\rho}}{\sqrt{\rho}} \sum_{n=-\infty}^{\infty} \frac{J'_n(\beta a)}{H_n^{(2)}(\beta a)} e^{jn\phi} \quad (9)$$

For the monostatic scattering

$$\phi = \phi' \quad (10)$$

Thus,

$$\sigma_{2-D} = \lim_{\rho\rightarrow\infty} \left[2\pi\rho \frac{|\bar{H}^s|^2}{|\bar{H}^i|^2} \right] = 2\pi \frac{2}{\pi\beta} \left| \sum_{n=-\infty}^{\infty} \frac{J'_n(\beta a)}{H_n^{(2)}(\beta a)} e^{jn\phi} \right|^2 \quad (11)$$

Finally, for the **hard polarization**

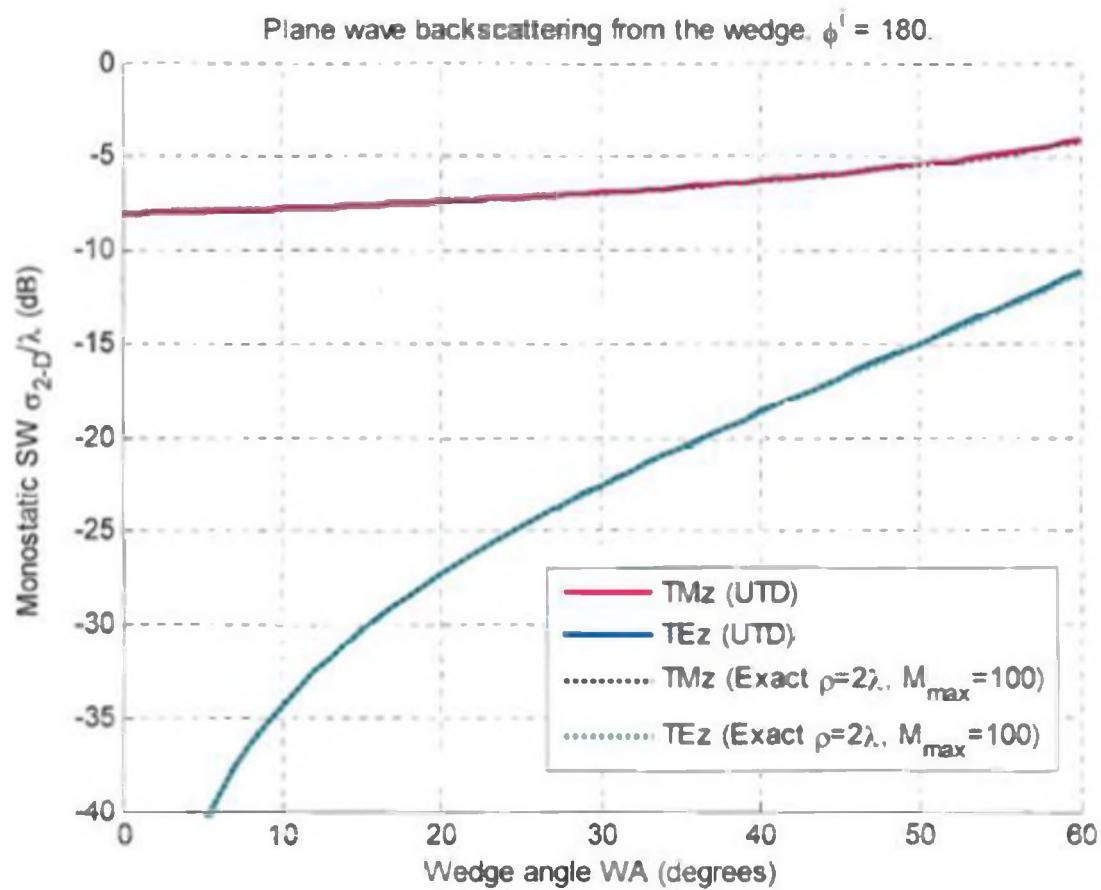
$$\boxed{\frac{\sigma_{2-D}}{\lambda} = \frac{2}{\pi} \left| \sum_{n=0}^{\infty} \varepsilon_n \frac{J'_n(\beta a)}{H_n^{(2)}(\beta a)} \cos(n\phi') \right|^2} \quad (12)$$

where

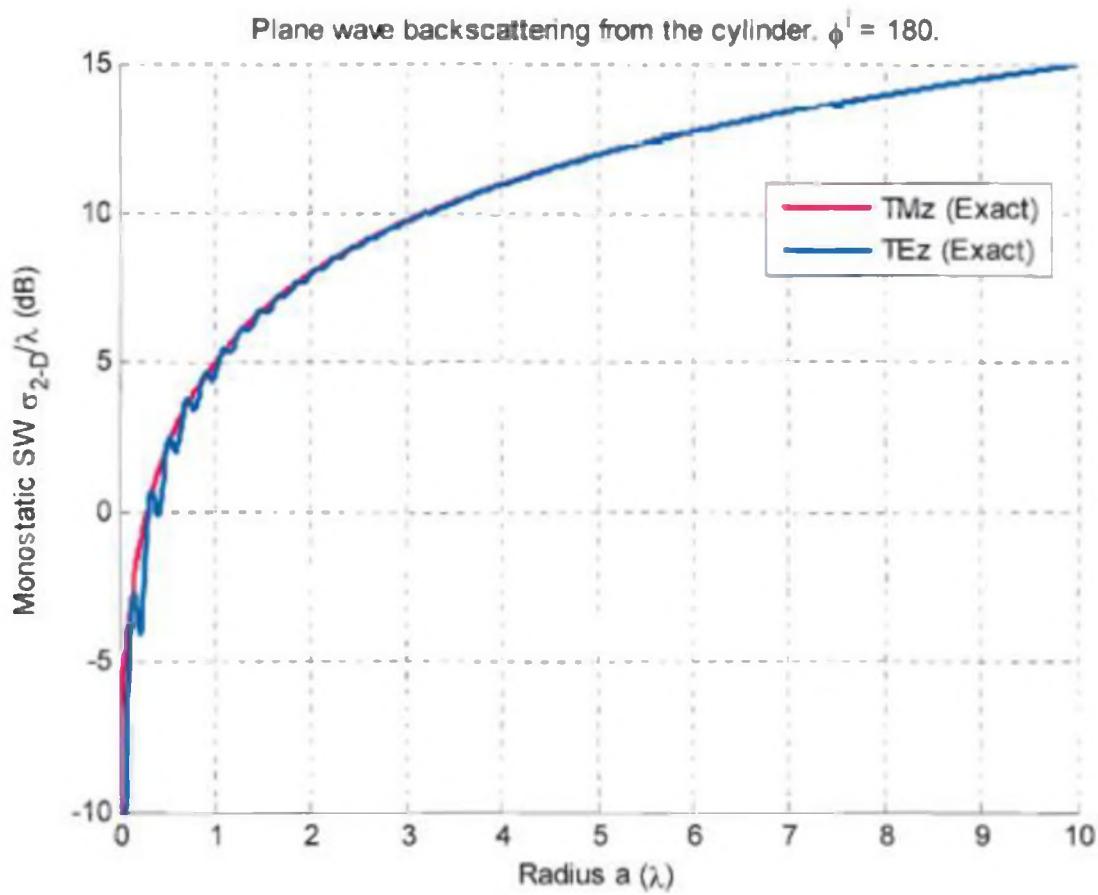
$$\varepsilon_n = \begin{cases} 1 & n=0 \\ 2 & n \neq 0 \end{cases} \quad (13)$$

(13.50a)

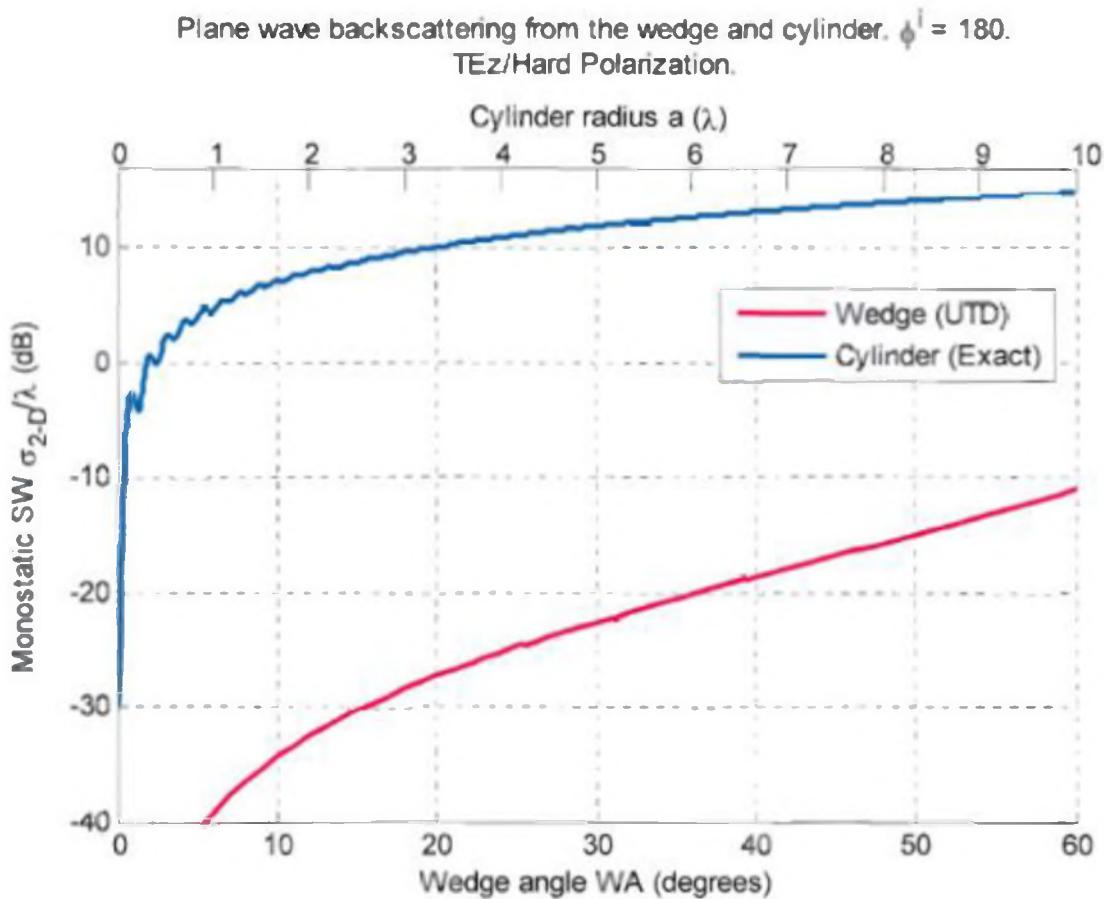
Normalized 2-D monostatic scattering width SW of the wedge vs. wedge angle WA [$0 \leq WA \leq 60^\circ$] for both hard and soft polarization.



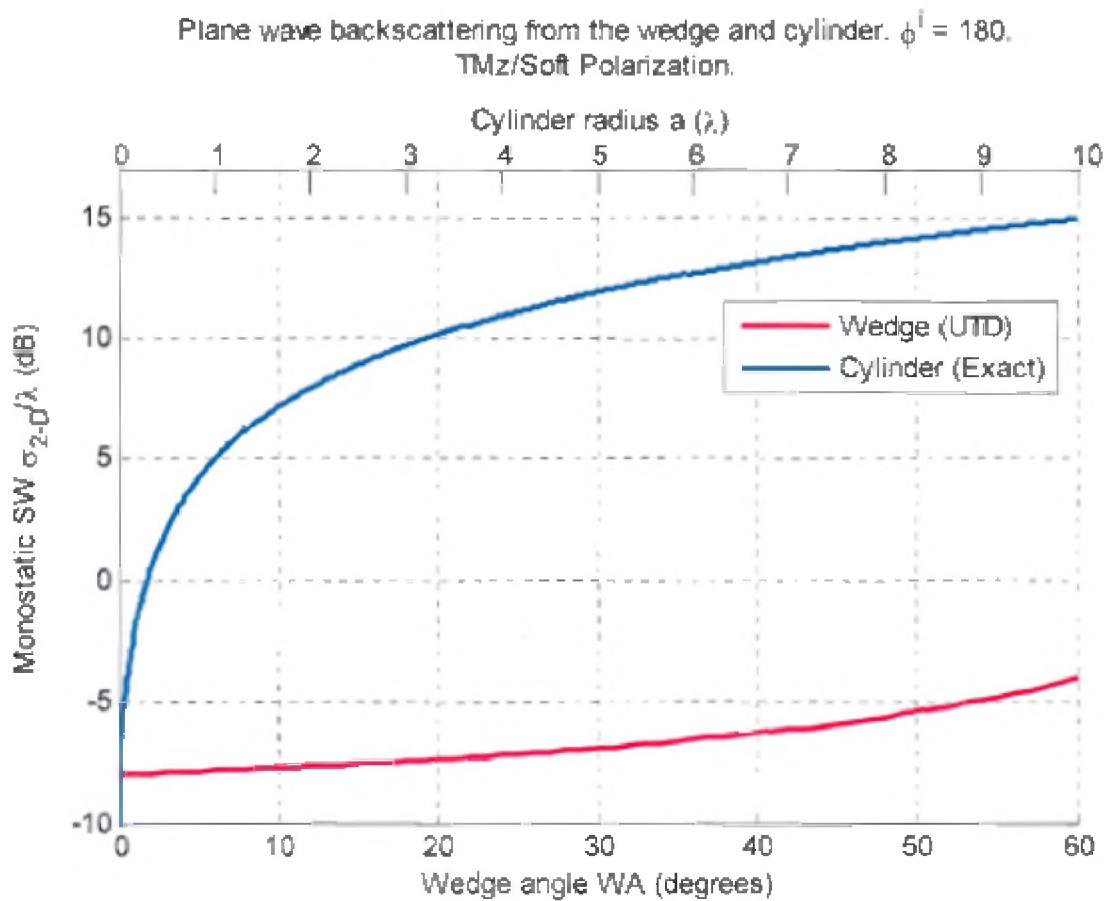
Normalized 2-D monostatic scattering width SW of the cylinder vs. radius a [$0 \leq a \leq 10\lambda$] for both hard and soft polarization



Normalized 2-D monostatic scattering width SW of the wedge [$0 \leq WA \leq 60^\circ$] and the cylinder [$0 \leq a \leq 10\lambda$] for hard polarization

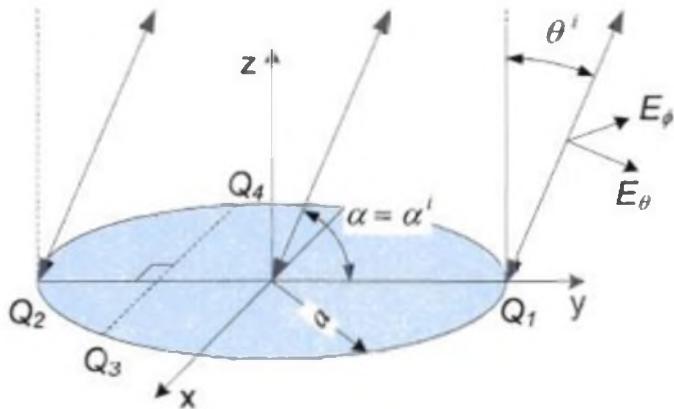


Normalized 2-D monostatic scattering width SW of the wedge [$0 \leq WA \leq 60^\circ$] and the cylinder [$0 \leq a \leq 10\lambda$] for soft polarization



13.51 Uniform plane wave incident upon a circular ground plane.

The numbered equations below are the same equations as in [36]



$$\theta' \in [0, 90^\circ]; \quad \alpha \in [0, 180^\circ]; \quad \alpha = 90^\circ - \theta; \quad \theta = \theta'$$

$$\sigma_{3-D} = \lim_{r \rightarrow \infty} \left[4\pi r^2 \frac{|E^s|^2}{|E^i|^2} \right]$$

$$\bar{E}_\theta^i = \hat{a}_\theta E_0 e^{j\beta(y \sin \theta' + z \cos \theta')} \quad (1)$$

$$\bar{E}_\phi^i = \hat{a}_\phi E_0 e^{j\beta(y \sin \theta' + z \cos \theta')} \quad (2)$$

$$|E^i|^2 = E_0^2$$

For Physical optics

Monostatic RCS (hard and soft polarizations)

$$\sigma_{3-D} = 4\pi \left(\frac{\pi a^2}{\lambda} \right)^2 \cos^2 \theta \left| 2 \frac{J_1(2\beta a \sin \theta)}{2\beta a \sin \theta} \right|^2$$

For UTD

Hard polarization	Soft polarization
$ E^s ^2 = E_{hard}^{1st(1,2)} + E_{hard}^{2nd(1,2)} + E_{hard}^{2nd(3,4)} ^2$	$ E^s ^2 = E_{soft}^{1st(1,2)} + E_{hard}^{2nd(3,4)} ^2$

cont'd

13.51 cont'd

Hard polarization (TE^x)

- First-order 3-D diffractions from edges #1 and #2 for monostatic scattering.

$$E_{hard}^{1st(1,2)} = E_{hard}^{d1} = E_1^d + E_2^d$$

Hard polarization

E_1^d $Q1 = (z = 0, y = a)$ $E_1^d = E_\theta^i(Q1) D_1^h A_1 e^{-j\beta r_1}$ $E_\theta^i(Q1) = E_0 e^{j\beta a \sin \theta}$ $r_1 = r - a \sin \theta$ phase $r_1 = r$ amplitude $A_1 = \frac{\sqrt{\rho_{c1}}}{r_1} = \frac{\sqrt{\rho_{c1}}}{r}$ $\rho_{c1} = \frac{a}{2 \sin \theta}$ (7)	E_2^d $Q2 = (z = 0, y = -a)$ $E_2^d = E_\theta^i(Q2) D_2^h A_2 e^{-j\beta r_2}$ $E_\theta^i(Q2) = E_0 e^{-j\beta a \sin \theta}$ $r_2 = r + a \sin \theta$ phase $r_2 = r$ amplitude $A_2 = \frac{\sqrt{\rho_{c2}}}{r_2} = \frac{\sqrt{\rho_{c2}}}{r}$ $\rho_{c2} = \frac{-a}{2 \sin \theta}$ (8)
$A_1 e^{-j\beta r_1} = e^{j\beta a \sin \theta} \sqrt{\frac{a}{2 \sin \theta}} \frac{e^{-j\beta r}}{r}$ Keller $D_1^h(90^\circ + \theta, 90^\circ + \theta, 2)$ $D_1^h = \frac{e^{-j\pi/4}}{2\sqrt{2\pi\beta}} \left(-1 + \frac{1}{\sin \theta} \right)$ (3)	$A_2 e^{-j\beta r_2} = e^{-j\beta a \sin \theta} \sqrt{\frac{-a}{2 \sin \theta}} \frac{e^{-j\beta r}}{r}$ Keller $D_2^h(90^\circ - \theta, 90^\circ - \theta, 2)$ $D_2^h = \frac{e^{-j\pi/4}}{2\sqrt{2\pi\beta}} \left(-1 - \frac{1}{\sin \theta} \right)$ (4)

$$E_{hard}^{1st(1,2)} = -E_0 \frac{1}{2} \sqrt{\frac{a}{\pi \beta \sin \theta}} \left(\cos \chi - j \frac{\sin \chi}{\sin \theta} \right) \frac{e^{-j\beta r}}{r} \quad (10)$$

$$\chi = 2\beta a \sin \theta - \pi/4$$

cont'd

13.51 cont'd

- Second-order 3-D diffractions from edges #1 and #2 for monostatic scattering.

$$E_{hard}^{2nd(1,2)} = E_{hard}^{d2} = E_{21r}^d + E_{12r}^d$$

Hard polarization

E_{21r}^d $Q1 = (z = 0, y = a)$ $E_{21r}^d = E_{i21}^d D_{21r}^h A_{21r} e^{-j\beta r_1}$ $r_1 = r - a \sin \theta \quad \text{phase}$ $r_1 = r \quad \text{amplitude}$ $A_{21r} = \frac{\sqrt{\rho_{21r}}}{r_1} = \frac{\sqrt{\rho_{21r}}}{r}$ $\rho_{21r} = \frac{a(1 + 2 \sin \theta)}{2 \sin^2 \theta} \quad (27)$	E_{12r}^d $Q2 = (z = 0, y = -a)$ $E_{12r}^d = E_{i12}^d D_{12r}^h A_{12r} e^{-j\beta r_2}$ $r_2 = r + a \sin \theta \quad \text{phase}$ $r_2 = r \quad \text{amplitude}$ $A_{12r} = \frac{\sqrt{\rho_{12r}}}{r_2} = \frac{\sqrt{\rho_{12r}}}{r}$ $\rho_{12r} = \frac{a(1 - 2 \sin \theta)}{2 \sin^2 \theta} \quad (26)$
$A_{21r} e^{-j\beta r_1}$ $= e^{j\beta a \sin \theta} \sqrt{\frac{a(1 + 2 \sin \theta)}{2 \sin^2 \theta}} \frac{e^{-j\beta r}}{r}$ UTD $D_{21r}^h(2a, 90^\circ + \theta, 0, 90, 2)$	$A_{12r} e^{-j\beta r_2}$ $= e^{-j\beta a \sin \theta} \sqrt{\frac{a(1 - 2 \sin \theta)}{2 \sin^2 \theta}} \frac{e^{-j\beta r}}{r}$ UTD $D_{12r}^h(2a, 90^\circ - \theta, 0, 90, 2)$

cont'd

13.51 cont'd

where

E_{i21}^d $Q2 = (z = 0, y = -a)$ $E_\theta^i(Q2) = E_0 e^{-j\beta a \sin \theta}$ $r_2 = 2a \quad \text{phase, amplitude}$ $A_{i21} = \sqrt{\frac{\rho_c}{r_2(\rho_c + r_2)}} = \sqrt{\frac{\rho_{i21}}{2a(\rho_{i21} + 2a)}}$ $\rho_{i21} = \frac{-a}{1 + \sin \theta} \quad (25)$ $A_{i21} e^{-j\beta r_2} = \sqrt{\frac{-2a}{1 + 2\sin \theta}} \frac{e^{-j\beta 2a}}{2a}$ UTD $D_{i21}^h(2a, 0, 90^\circ - \theta, 90, 2)$	E_{i12}^d $Q1 = (z = 0, y = a)$ $E_\theta^i(Q1) = E_0 e^{j\beta a \sin \theta}$ $r_1 = 2a \quad \text{phase, amplitude}$ $A_{i12} = \sqrt{\frac{\rho_c}{r_1(\rho_c + r_1)}} = \sqrt{\frac{\rho_{i12}}{2a(\rho_{i12} + 2a)}}$ $\rho_{i12} = \frac{-a}{1 - \sin \theta} \quad (24)$ $A_{i12} e^{-j\beta r_1} = \sqrt{\frac{-2a}{1 - 2\sin \theta}} \frac{e^{-j\beta 2a}}{2a}$ UTD $D_{i12}^h(2a, 0, 90^\circ + \theta, 90, 2)$
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$$E_{12r}^d = E_0 e^{j\beta a \sin \theta} \sqrt{\frac{-2a}{1 - 2\sin \theta}} \frac{e^{-j\beta 2a}}{2a} e^{-j\beta a \sin \theta} \sqrt{\frac{a(1 - 2\sin \theta)}{2\sin^2 \theta}} \frac{e^{-j\beta r}}{r} D_{i12}^h D_{12r}^h$$

$$E_{12r}^d = E_0 \frac{ja}{\sin \theta} D_{i12}^h D_{12r}^h \frac{e^{-j\beta 2a}}{2a} \frac{e^{-j\beta r}}{r}$$

$$E_{12r}^d = E_0 \frac{jae^{j\beta 2a}}{\sin \theta} D_{i12}^h D_{12r}^h \frac{e^{-j\beta 2a}}{\sqrt{2a}} \frac{e^{-j\beta 2a}}{\sqrt{2a}} \frac{e^{-j\beta r}}{r}$$

cont'd

13.51 cont'd

$$E_{21r}^d = E_0 e^{-j\beta \sin \theta} \sqrt{\frac{-2a}{1+2 \sin \theta}} \frac{e^{-j\beta 2a}}{2a} e^{j\beta \sin \theta} \sqrt{\frac{a(1+2 \sin \theta)}{2 \sin^2 \theta}} \frac{e^{-j\beta r}}{r} D_{i21}^h D_{21r}^h$$

$$E_{21r}^d = \frac{j E_0 a}{\sin \theta} D_{i21}^h D_{21r}^h \frac{e^{-j\beta 2a}}{2a} \frac{e^{-j\beta r}}{r}$$

$$E_{21r}^d = E_0 \frac{j a e^{j\beta 2a}}{\sin \theta} D_{i21}^h D_{21r}^h \frac{e^{-j\beta 2a}}{\sqrt{2a}} \frac{e^{-j\beta 2a}}{\sqrt{2a}} \frac{e^{-j\beta r}}{r}$$

Hard polarization

E_{21r}^d	E_{12r}^d
$E_{21r}^d = E_0 \frac{j a e^{j\beta 2a}}{\sin \theta} V_{i21}^h V_{21r}^h \frac{e^{-j\beta r}}{r}$	$E_{12r}^d = E_0 \frac{j a e^{j\beta 2a}}{\sin \theta} V_{i12}^h V_{12r}^h \frac{e^{-j\beta r}}{r}$

where

$$V_{i12}^h = \frac{e^{-j\beta 2a}}{\sqrt{2a}} D_{i12}^h (2a, 0, 90^\circ + \theta, 90, 2) \quad (20)$$

$$V_{i21}^h = \frac{e^{-j\beta 2a}}{\sqrt{2a}} D_{i21}^h (2a, 0, 90^\circ - \theta, 90, 2) \quad (21)$$

$$V_{12r}^h = \frac{e^{-j\beta 2a}}{\sqrt{2a}} D_{12r}^h (2a, 90^\circ - \theta, 0, 90, 2) \quad (22)$$

$$V_{21r}^h = \frac{e^{-j\beta 2a}}{\sqrt{2a}} D_{21r}^h (2a, 90^\circ + \theta, 0, 90, 2) \quad (23)$$

$$E_{hard}^{2nd(1,2)} = E_0 \frac{j a e^{j\beta 2a}}{\sin \theta} (V_{i21}^h V_{21r}^h + V_{i12}^h V_{12r}^h) \frac{e^{-j\beta r}}{r} \quad (28)$$

cont'd

13.51 cont'd

- Second-order 3-D diffractions from the migrating points #3 and #4 for monostatic scattering.

$$E_{hard}^{2nd(3,4)} = E_{hard}^{d2} = E_{34r}^d + E_{43r}^d$$

Hard polarization

E_{43r}^d	E_{34r}^d
$Q3 = (x = a \cos \phi_3, y = a \sin \phi_3, z = 0)$	$Q4 = (x = a \cos \phi_4, y = a \sin \phi_4, z = 0)$
$\phi_3 = \arccos\left(\frac{-1}{\sqrt{1 + \sin^2 \theta}}\right) \quad (12)$	$\phi_4 = \arccos\left(\frac{1}{\sqrt{1 + \sin^2 \theta}}\right) \quad (13)$
$E_{43r}^d = E_{i43}^d D_{43r}^h A_{43r} e^{-j\beta r_3}$	$E_{34r}^d = E_{i34}^d D_{34r}^h A_{34r} e^{-j\beta r_4}$
$r_3 = r - a \sin \theta \sin \phi_3 \quad \text{phase}$	$r_4 = r + a \sin \theta \sin \phi_4 \quad \text{phase}$
$r_3 = r \quad \text{amplitude}$	$r_4 = r \quad \text{amplitude}$
$A_{43r} = \frac{\sqrt{\rho_{43r}}}{r_3} = \frac{\sqrt{\rho_{43r}}}{r}$	$A_{34r} = \frac{\sqrt{\rho_{34r}}}{r_4} = \frac{\sqrt{\rho_{34r}}}{r}$
$\rho_{43r} = \frac{-a(1+2\sin^2 \theta)}{2\sin^2(\theta)(1+\sin^2 \theta)^{3/2}} \quad (15)$	$\rho_{34r} = \frac{-a(1+2\sin^2 \theta)}{2\sin^2(\theta)(1+\sin^2 \theta)^{3/2}} \quad (15)$
Keller D_{43r}^h	Keller D_{34r}^h

$$E_{hard}^{2nd(3,4)} = -\frac{j \sin^2(\theta) e^{-j2\beta a\sqrt{1+\sin^2 \theta}}}{\pi \beta \sin \theta \sqrt{1+\sin^2 \theta}} \frac{e^{-j\beta r}}{r} \quad (19)$$

cont'd

13.51 cont'd

- Ring-radiator contributions.

$$E_{hard}^{ring} = E_\theta^m + E_\theta^e$$

where

$$E_\theta^e = \frac{a}{4\pi} \frac{e^{-j\beta r}}{r} \cos^2 \theta \int_0^{2\pi} \left(1 + \frac{\sqrt{g}}{h}\right) \frac{\cos^2(\phi) e^{-j2\beta ah}}{g} d\phi = 0 \quad (49a)$$

$$E_\theta^m = -\frac{a}{4\pi} \frac{e^{-j\beta r}}{r} \int_0^{2\pi} \left(1 - \frac{\sqrt{g}}{h}\right) \frac{\sin^2(\phi) e^{j2\beta ah}}{g} d\phi \quad (49b)$$

$$g = 1 - \sin^2 \theta \cos^2 \phi$$

$$h = \sin \theta \sin \phi$$

$$E_{hard}^{ring} = -\frac{a}{4\pi} \frac{e^{-j\beta r}}{r} \int_0^{2\pi} \left(1 - \frac{\sqrt{1 - \sin^2 \theta \cos^2 \phi}}{h}\right) \frac{\sin^2(\phi) e^{j2\beta a \sin \theta \sin \phi}}{1 - \sin^2 \theta \cos^2 \phi} d\phi$$

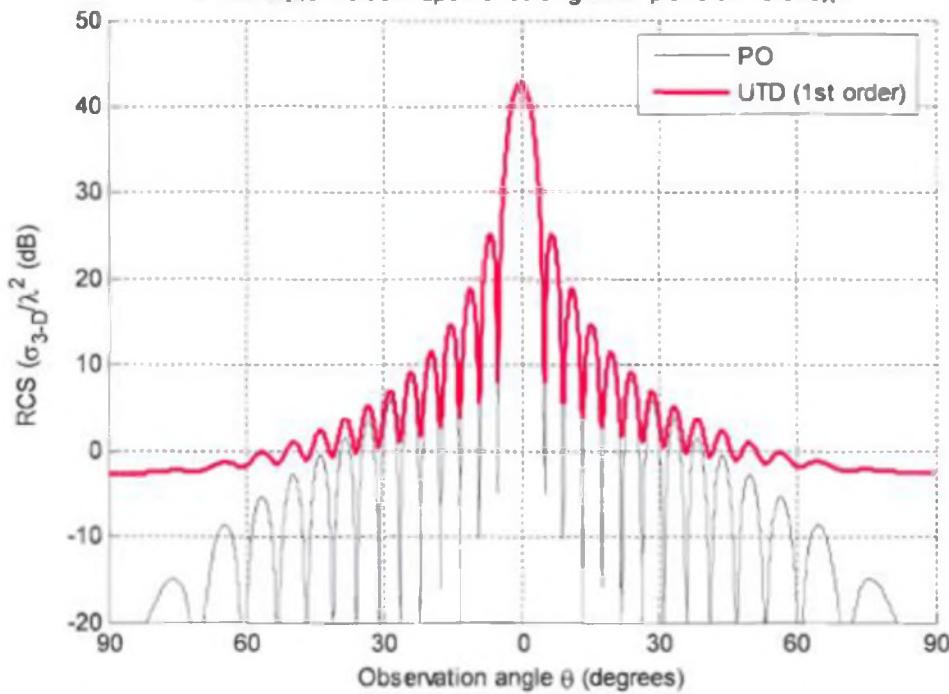
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- Scattering monostatic patterns.

Monostatic 3-D RCS.

Hard polarization TEx.

Plane wave incident upon circular ground plane $a = 3.516\lambda$.

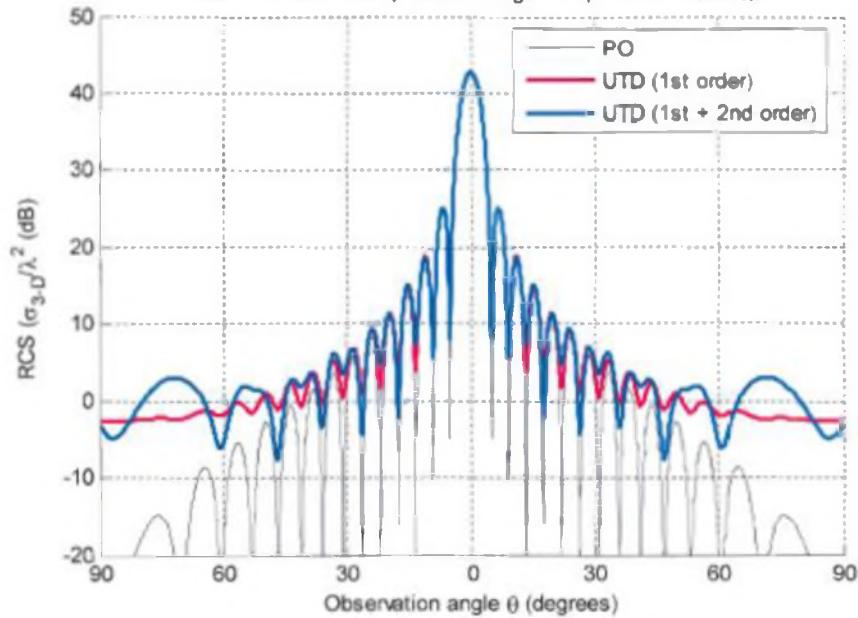


- Scattering monostatic patterns.

Monostatic 3-D RCS.

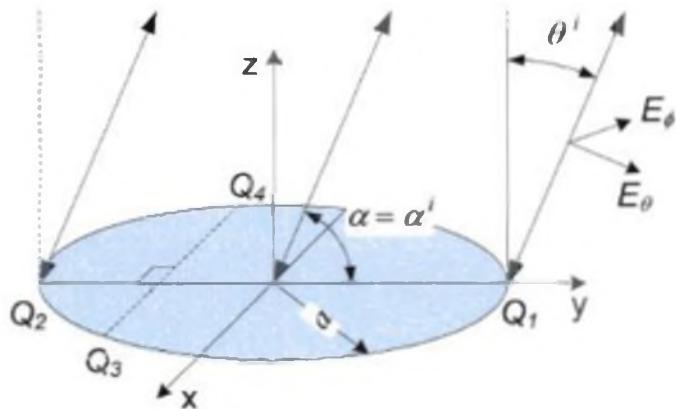
Hard polarization TEx.

Plane wave incident upon circular ground plane $a = 3.516\lambda$.



13.52 Uniform plane wave incident upon a circular ground plane.

The numbered equations below are the same equations as in [36].



$$\theta' \in [0, 90^\circ]; \quad \alpha \in [0, 180^\circ]; \quad \alpha = 90^\circ - \theta; \quad \theta = \theta'$$

$$\sigma_{3-D} = \lim_{r \rightarrow \infty} \left[4\pi r^2 \frac{|E^s|^2}{|E^i|^2} \right]$$

$$\bar{E}_\theta^i = \hat{a}_\theta E_0 e^{j\beta(y\sin\theta' + z\cos\theta')} \quad (1)$$

$$\bar{E}_\phi^i = \hat{a}_\phi E_0 e^{j\beta(y\sin\theta' + z\cos\theta')} \quad (2)$$

$$|E^i|^2 = E_0^2$$

For Physical optics

Monostatic RCS (hard and soft polarizations)

$$\sigma_{3-D} = 4\pi \left(\frac{\pi a^2}{\lambda} \right)^2 \cos^2 \theta \left| 2 \frac{J_1(2\beta a \sin \theta)}{2\beta a \sin \theta} \right|^2$$

For UTD

Hard polarization	Soft polarization
$ E^s ^2 = E_{hard}^{1st(1,2)} + E_{hard}^{2nd(1,2)} + E_{hard}^{2nd(3,4)} ^2$	$ E^s ^2 = E_{soft}^{1st(1,2)} + E_{hard}^{2nd(3,4)} ^2$

cont'd

13.52 Cont'd

Soft polarization (TM^x)

- First-order 3-D diffractions from edges #1 and #2 for monostatic scattering.

$$E_{soft}^{1st(1,2)} = E_{soft}^{d1} = E_1^d + E_2^d$$

Soft polarization

E_1^d $Q1 = (z = 0, y = a)$ $E_1^d = E_\phi^i(Q1) D_1^s A_1 e^{-j\beta r_1}$ $E_\phi^i(Q1) = E_0 e^{j\beta a \sin \theta}$ $r_1 = r - a \sin \theta$ phase $r_1 = r$ amplitude $A_1 = \frac{\sqrt{\rho_{c1}}}{r_1} = \frac{\sqrt{\rho_{c1}}}{r}$ $\rho_{c1} = \frac{a}{2 \sin \theta}$ (7)	E_2^d $Q2 = (z = 0, y = -a)$ $E_2^d = E_\phi^i(Q2) D_2^s A_2 e^{-j\beta r_2}$ $E_\phi^i(Q2) = E_0 e^{-j\beta a \sin \theta}$ $r_2 = r + a \sin \theta$ phase $r_2 = r$ amplitude $A_2 = \frac{\sqrt{\rho_{c2}}}{r_2} = \frac{\sqrt{\rho_{c2}}}{r}$ $\rho_{c2} = \frac{-a}{2 \sin \theta}$ (8) $A_2 e^{-j\beta r_2} = e^{j\beta a \sin \theta} \sqrt{\frac{-a}{2 \sin \theta}} \frac{e^{-j\beta r}}{r}$ Keller $D_1^s(90^\circ + \theta, 90^\circ + \theta, 2)$ $D_1^s = \frac{e^{-j\pi/4}}{2\sqrt{2\pi\beta}} \left(-1 - \frac{1}{\sin \theta} \right)$ (3)
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$$E_{soft}^{1st(1,2)} = E_0 \frac{1}{2} \sqrt{\frac{a}{\pi \beta \sin \theta}} \left(\cos \chi + j \frac{\sin \chi}{\sin \theta} \right) \frac{e^{-j\beta r}}{r} \quad (11)$$

$$\chi = 2\beta a \sin \theta - \pi/4$$

Cont'd.

13.52 cont'd

- Second-order 3-D diffractions from the migrating points #3 and #4 for monostatic scattering.

$$E_{soft}^{2nd(3,4)} = E_{soft}^{d2} = E_{34r}^d + E_{43r}^d$$

Soft polarization

E_{43r}^d $Q3 = (x = a \cos \phi_3, y = a \sin \phi_3, z = 0)$	E_{34r}^d $Q4 = (x = a \cos \phi_4, y = a \sin \phi_4, z = 0)$
$\phi_3 = \arccos\left(\frac{-1}{\sqrt{1 + \sin^2 \theta}}\right)$ (12)	$\phi_4 = \arccos\left(\frac{1}{\sqrt{1 + \sin^2 \theta}}\right)$ (13)
$E_{43r}^d = E_{i43}^d D_{43r}^s A_{43r} e^{-j\beta r},$ $r_3 = r - a \sin \theta \sin \phi_3 \quad \text{phase}$ $r_3 = r \quad \text{amplitude}$	$E_{34r}^d = E_{i34}^d D_{34r}^s A_{34r} e^{-j\beta r},$ $r_4 = r + a \sin \theta \sin \phi_4 \quad \text{phase}$ $r_4 = r \quad \text{amplitude}$
$A_{43r} = \frac{\sqrt{\rho_{43r}}}{r_3} = \frac{\sqrt{\rho_{43r}}}{r}$ $\rho_{43r} = \frac{-a(1+2\sin^2 \theta)}{2\sin^2(\theta)(1+\sin^2 \theta)^{3/2}}$ (15)	$A_{34r} = \frac{\sqrt{\rho_{34r}}}{r_4} = \frac{\sqrt{\rho_{34r}}}{r}$ $\rho_{34r} = \frac{-a(1+2\sin^2 \theta)}{2\sin^2(\theta)(1+\sin^2 \theta)^{3/2}}$ (15)
Keller D_{43r}^s	Keller D_{34r}^s

$$E_{soft}^{2nd(3,4)} = -\frac{j \cos^2(\theta) e^{-j2\beta a\sqrt{1+\sin^2 \theta}}}{\pi \beta \sin \theta \sqrt{1+\sin^2 \theta}} \frac{e^{-j\beta r}}{r}$$
 (18)

cont'd

13.52 cont'd

- Ring-radiator contributions.

$$E_{soft}^{ring} = E_\phi^m + E_\phi^e$$

where

$$E_\phi^e = \frac{a}{4\pi} \frac{e^{-j\beta r}}{r} \int_0^{2\pi} \left(1 + \frac{\sqrt{g}}{h} \right) \frac{\sin^2(\phi) e^{-j2\beta ah}}{g} d\phi \quad (48a)$$

$$E_\phi^m = -\frac{a}{4\pi} \frac{e^{-j\beta r}}{r} \cos^2 \theta \int_0^{2\pi} \left(1 - \frac{\sqrt{g}}{h} \right) \frac{\cos^2(\phi) e^{j2\beta ah}}{g} d\phi = 0 \quad (48b)$$

$$\begin{aligned} g &= 1 - \sin^2 \theta \cos^2 \phi \\ h &= \sin \theta \sin \phi \end{aligned}$$

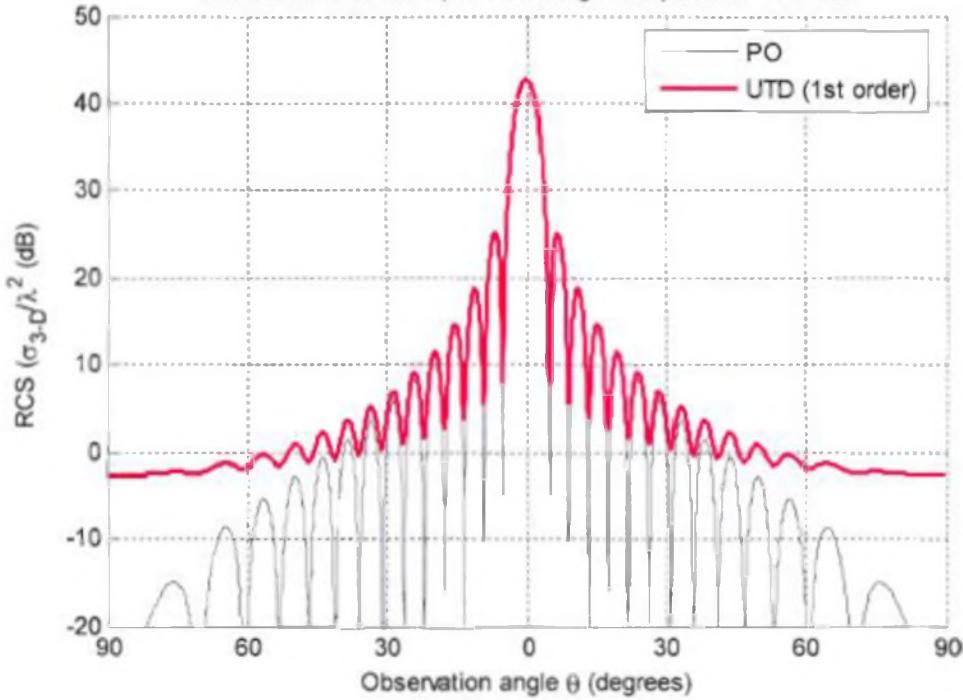
$$E_{soft}^{ring} = \frac{a}{4\pi} \frac{e^{-j\beta r}}{r} \int_0^{2\pi} \left(1 + \frac{\sqrt{1 - \sin^2 \theta \cos^2 \phi}}{h} \right) \frac{\sin^2(\phi) e^{j2\beta a \sin \theta \sin \phi}}{1 - \sin^2 \theta \cos^2 \phi} d\phi$$

- Scattering monostatic patterns.

Monostatic 3-D RCS.

Soft polarization TM_x.

Plane wave incident upon circular ground plane $a = 3.516\lambda$.

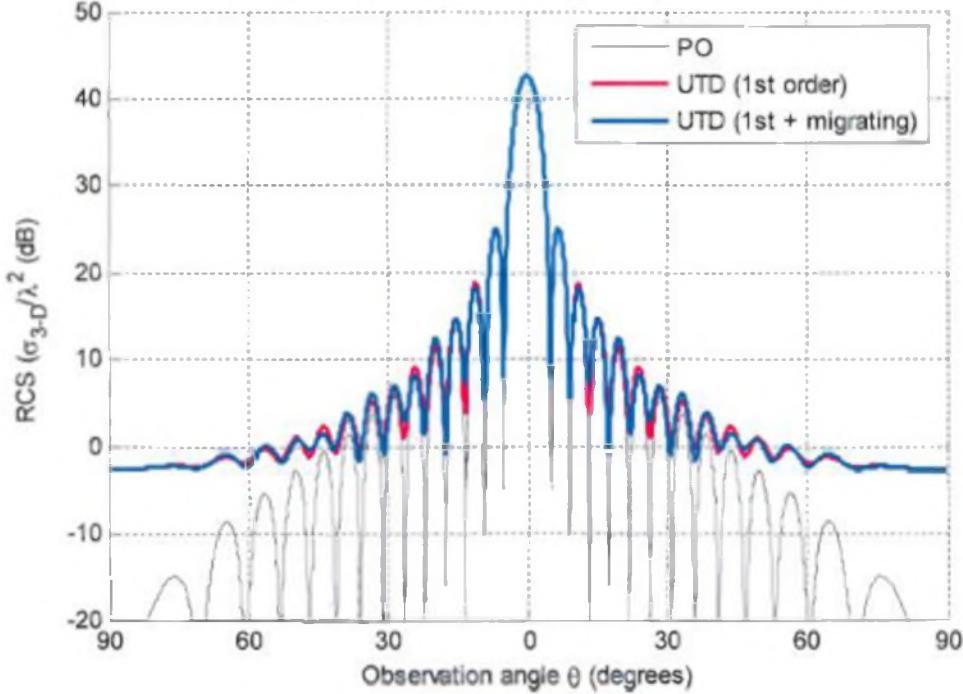


- Scattering monostatic patterns.

Monostatic 3-D RCS.

Soft polarization TM_x.

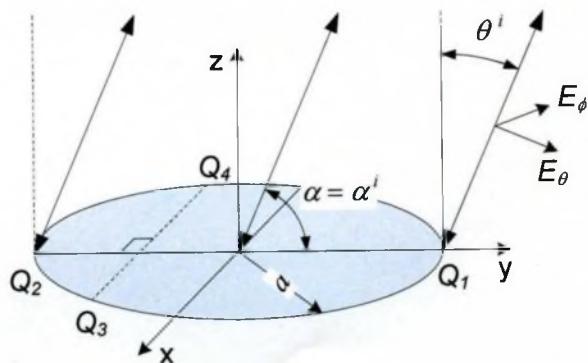
Plane wave incident upon circular ground plane $a = 3.516\lambda$.



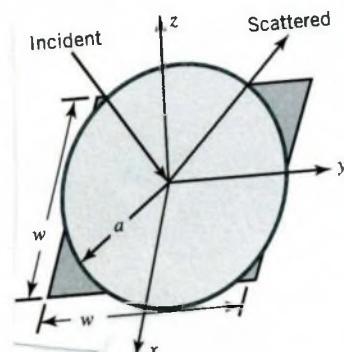
13.53 Uniform plane wave incident upon a circular and square ground plane.

The numbered equations below are the same equations as in [36].

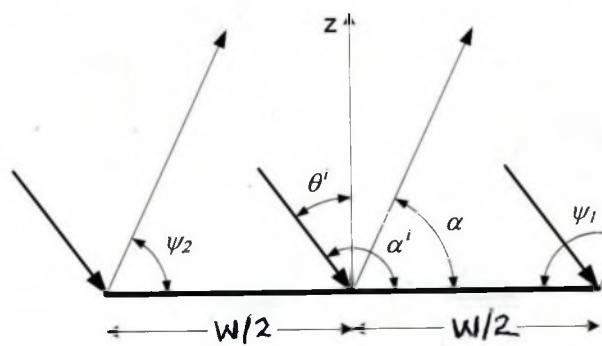
Circular Ground Plane (radius = a)



$$\theta^i \in [0, 90^\circ]; \alpha \in [0, 180^\circ]; \alpha = 90^\circ - \theta; \theta = \theta^i$$

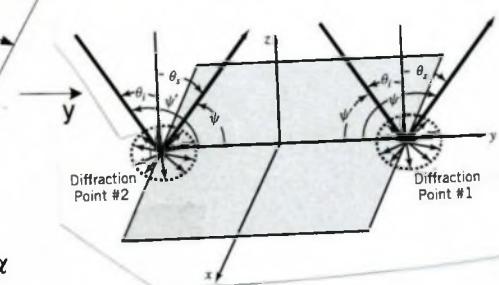


Square Ground Plane ($w \times w$)



$$\theta^i \in [0^\circ, 90^\circ]; \alpha \in [0^\circ, 180^\circ]$$

$$\theta^i = \alpha^i - 90^\circ; \psi_1 = 180^\circ - \alpha; \psi_2 = \alpha$$



Area of a square ground plane with side length w

$$A_s = w^2$$

Area of a circular ground plane of radius a

$$A_c = \pi a^2$$

$$\text{If } A_s = A_c, \text{ then } \Rightarrow a = \frac{w}{\sqrt{\pi}} = \frac{5\lambda}{\sqrt{\pi}} = 2.8209\lambda$$

cont'd

13.53 cont'd

$$\sigma_{3-D} = \lim_{r \rightarrow \infty} \left[4\pi r^2 \frac{|E^s|^2}{|E^i|^2} \right]$$

$$\begin{aligned}\bar{E}_\theta^i &= \hat{a}_\theta E_0 e^{j\beta(y \sin \theta' + z \cos \theta')} \\ \bar{E}_\phi^i &= \hat{a}_\phi E_0 e^{j\beta(y \sin \theta' + z \cos \theta')} \quad (1), (2) \\ |E^i|^2 &= E_0^2\end{aligned}$$

For Physical optics

Monostatic RCS (hard and soft polarizations)

$$\sigma_{3-D} = 4\pi \left(\frac{\pi a^2}{\lambda} \right)^2 \cos^2 \theta \left| 2 \frac{J_1(2\beta a \sin \theta)}{2\beta a \sin \theta} \right|^2$$

For UTD

Hard polarization	Soft polarization
$ E^s ^2 = \left E_{hard}^{1st(1,2)} + E_{hard}^{2nd(1,2)} + E_{hard}^{2nd(3,4)} \right ^2$	$ E^s ^2 = \left E_{soft}^{1st(1,2)} + E_{hard}^{2nd(3,4)} \right ^2$

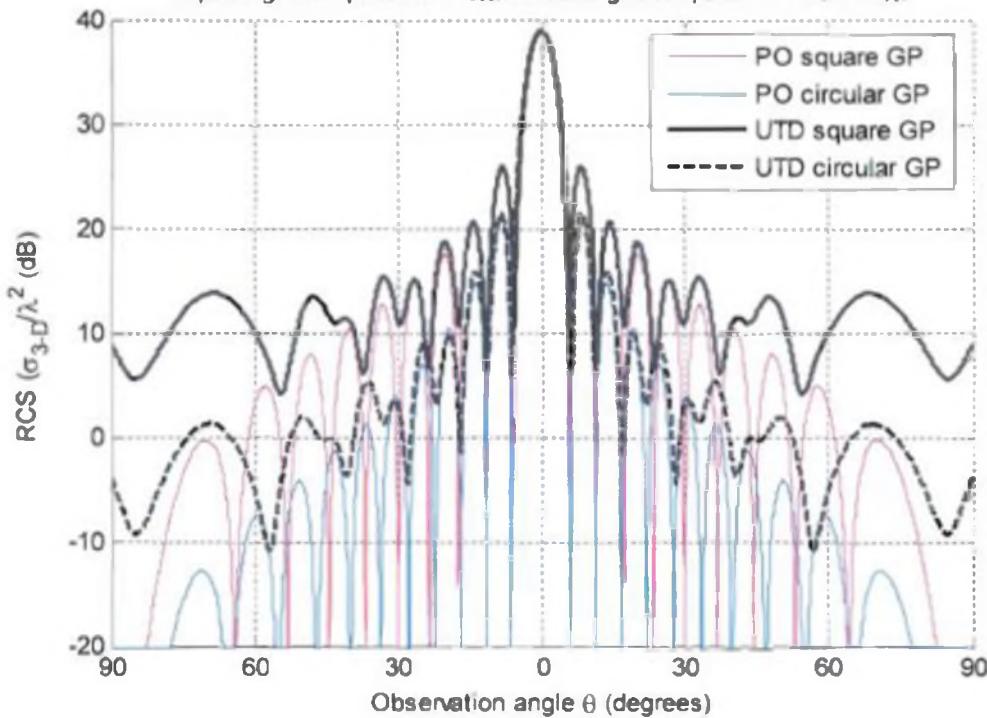
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- Scattering monostatic patterns.

Monostatic 3-D RCS

Hard polarization TEx.

Square ground plane $w = 5\lambda$, Circular ground plane $a = 2.8209\lambda$.

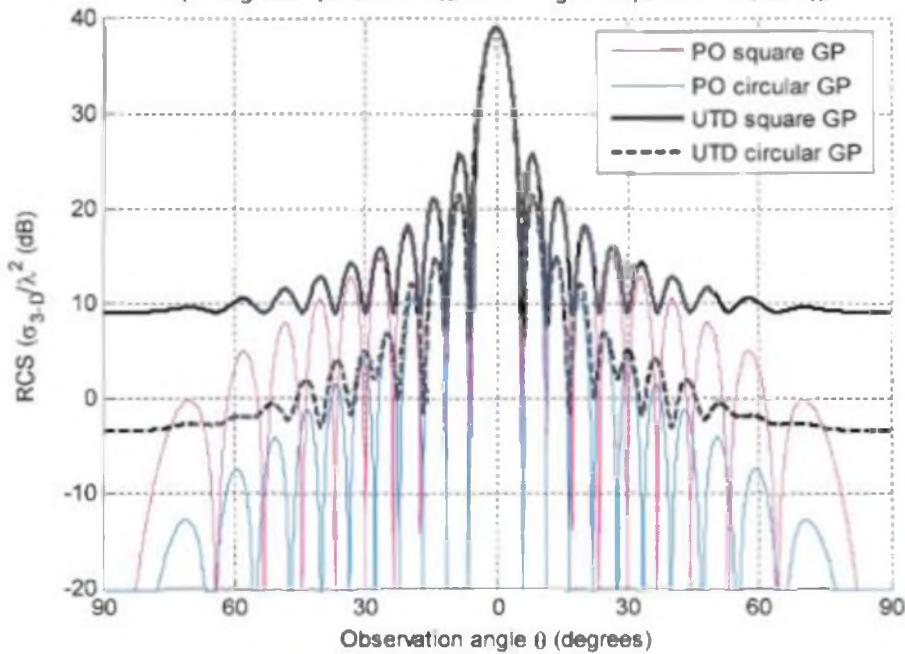


- Scattering monostatic patterns.

Monostatic 3-D RCS

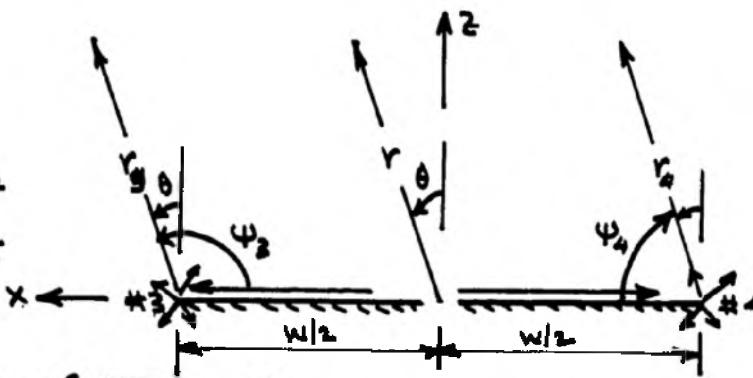
Soft polarization TMx.

Square ground plane $w = 5\lambda$, Circular ground plane $a = 2.8209\lambda$.



(a) $\Psi_3 = \pi/2 + \theta$

$$\Psi_4 = \begin{cases} \frac{\pi}{2} - \theta, & 0 \leq \theta \leq \pi/2 \\ \frac{3\pi}{2} - \theta, & \pi/2 \leq \theta \leq \pi \end{cases}$$



$$E^d = \frac{1}{j\beta} \frac{\partial E^i(\theta_3)}{\partial n} \frac{\partial D^s}{\partial \phi'} \sqrt{\frac{\rho_c}{s(\rho_c+s)}} e^{-j\beta s}$$

$$\frac{\partial E^i(\theta_3)}{\partial n} = \frac{1}{s'} \frac{\partial E^i}{\partial \phi'} \Big|_{\Psi_3}, \quad \frac{\partial D^s}{\partial \phi'} = \text{slope diffraction coefficient}$$

For Wedge #3

$$\begin{aligned} \frac{\partial E^i(\theta_3)}{\partial n} &= \frac{1}{s'} \frac{\partial E^i}{\partial \Psi_3} \Big|_{s'=w/2} \\ &\quad \Psi_3 = \theta + \pi/2 \\ &= \frac{1}{s'} \frac{\partial E^i}{\partial \theta} \Big|_{\theta=90^\circ} \end{aligned}$$

$$\frac{\partial E^i}{\partial \theta} = \frac{2}{\partial \theta} \left[\cos \theta \frac{\cos \left(\frac{\beta a}{2} \sin \theta \right)}{\left(\frac{\beta a}{2} \sin \theta \right)^2 - \left(\frac{\pi}{2} \right)^2} \right] e^{-j\beta r} = \frac{2}{\partial \theta} \left[\frac{N}{D} \right] e^{-j\beta r} = \left[\frac{DN' - ND'}{D^2} \right] e^{-j\beta r}$$

$$\text{where } N = \cos \theta \cos \left(\frac{\beta a}{2} \sin \theta \right)$$

$$D = \left(\frac{\beta a}{2} \sin \theta \right)^2 - \left(\frac{\pi}{2} \right)^2$$

$$N' = \frac{2}{\partial \theta} N = -\sin \theta \cos \left(\frac{\beta a}{2} \sin \theta \right) - \cos \theta \sin \left(\frac{\beta a}{2} \sin \theta \right) \left(\frac{\beta a}{2} \cos \theta \right)$$

$$D' = \frac{2}{\partial \theta} D = 2 \left(\frac{\beta a}{2} \sin \theta \right) \left(\frac{\beta a}{2} \cos \theta \right)$$

Therefore

$$\frac{\partial E^i}{\partial \theta} = \frac{e^{-j\beta r}}{r} \frac{\left[\left(\frac{\beta a}{2} \sin \theta \right)^2 - \left(\frac{\pi}{2} \right)^2 \right] \left\{ -\sin \theta \cos \left(\frac{\beta a}{2} \sin \theta \right) - \cos \theta \sin \left(\frac{\beta a}{2} \sin \theta \right) \left(\frac{\beta a}{2} \cos \theta \right) \right\}}{\left[\left(\frac{\beta a}{2} \sin \theta \right)^2 - \left(\frac{\pi}{2} \right)^2 \right]^2} \leftarrow$$

$$\begin{aligned} \frac{\partial E^i}{\partial \theta} \Big|_{\theta=90^\circ} &= \frac{\left[\left(\frac{\beta a}{2} \right)^2 - \left(\frac{\pi}{2} \right)^2 \right] \left[-\cos \left(\frac{\beta a}{2} \right) \right]}{\left[\left(\frac{\beta a}{2} \right)^2 - \left(\frac{\pi}{2} \right)^2 \right]} e^{-j\beta r} \\ &= -\frac{\cos \left(\frac{\beta a}{2} \right)}{\left[\left(\frac{\beta a}{2} \right)^2 - \left(\frac{\pi}{2} \right)^2 \right]} e^{-j\beta r} \end{aligned}$$

$$\cos \theta \cos \left(\frac{\beta a}{2} \sin \theta \right) \left[2 \left(\frac{\beta a}{2} \sin \theta \right) \left(\frac{\beta a}{2} \cos \theta \right) \right]$$

Cont'd.

13.54 Cont'd. Since $\left. \frac{\partial E^L}{\partial \theta} \right|_{\theta=90^\circ}$ becomes indeterminate when $\frac{\partial \theta}{\partial z} = \frac{\pi}{2}$ or $a = \frac{\pi}{2}$,

then using L'Hopital's rule leads to

$$\left. \frac{\partial E^L}{\partial \theta} \right|_{\theta=90^\circ, a=\pi/2} = \frac{\sin(\frac{\beta a}{2})}{z(\frac{\pi}{2})} \left. \frac{e^{-j\beta r}}{r} \right|_{a=\pi/2} = \frac{1}{\pi} e^{-j\beta r}$$

Thus the field diffracted from edge #3 can be written as

$$E_3^d = \frac{1}{j\beta} \left[\frac{\partial E^L(\theta_3)}{\partial \theta} \right] \frac{\partial D^s}{\partial \phi'} \frac{\sqrt{s'}}{s} e^{-j\beta s} = \frac{1}{2j\beta} \left[\frac{1}{s'} \frac{\partial E^L(\theta_3)}{\partial \theta} \right] \frac{\partial D^s}{\partial \phi'} \frac{\sqrt{s'}}{s} e^{-j\beta s} \Big|_{\substack{s'=w/2 \\ s=r_3}}$$

Since $r_3 \approx r - \frac{w}{2} \sin \theta$ for phase terms

$r_3 \approx r$ for amplitude terms

$$E_3^d = \frac{1}{2j\beta} \left[\frac{1}{w/2} - \frac{\cos(\beta a/2)}{\left(\frac{\beta a}{2}\right)^2 - (\frac{\pi}{2})^2} e^{-j\beta w/2} \right] \frac{\partial D^s(\phi, \phi', n, \beta_0')}{\partial \phi'} \frac{\sqrt{w/2}}{r} e^{-j\beta(r - \frac{w}{2} \sin \theta)}$$

$$E_3^d = j \frac{(z/w)^{3/2} \cos(\beta a/2)}{\left(\beta a/2\right)^2 - (\pi/2)^2} \frac{\partial D^s(\phi, \phi', n, \beta_0')}{\partial \phi'} e^{-j\beta \frac{w}{2}(1 - \sin \theta)} \frac{e^{-j\beta r}}{r}$$

where $\phi = \psi_3, \phi' = \psi_3' = 0^\circ, n = 2, \beta_0' = 90^\circ$

Similarly

$$E_4^d = \frac{1}{j\beta^2} \left[-\frac{1}{s'} \frac{\partial E^L(\theta_4)}{\partial \theta} \right] \frac{\partial D^s(\phi, \phi', n, \beta_0')}{\partial \phi'} \frac{\sqrt{s'}}{s} e^{-j\beta s} \Big|_{\substack{s'=w/2 \\ s=r_3}}$$

Since $r_4 \approx r + \frac{w}{2} \sin \theta$ for phase terms

$r_4 \approx r$ for amplitude terms

then

$$E_4^d = -j \frac{(z/w)^{3/2} \cos(\beta a)}{\left(\frac{\beta a}{2}\right)^2 - (\frac{\pi}{2})^2} \frac{\partial D^s(\phi, \phi', n, \beta_0')}{\partial \phi'} e^{-j\beta \frac{w}{2}(1 + \sin \theta)} \frac{e^{-j\beta r}}{r}$$

where $\phi = \psi_4, \phi' = \psi_4' = 0^\circ, n = 2, \beta_0' = 90^\circ$

The total diffracted field is then equal to

$$E_t^d = E_3^d + E_4^d$$

To compute the total field, the diffracted field must be combined with the geometrical optics field.

Cont'd.

(b) To plot the field for $a = \lambda/2$, the term

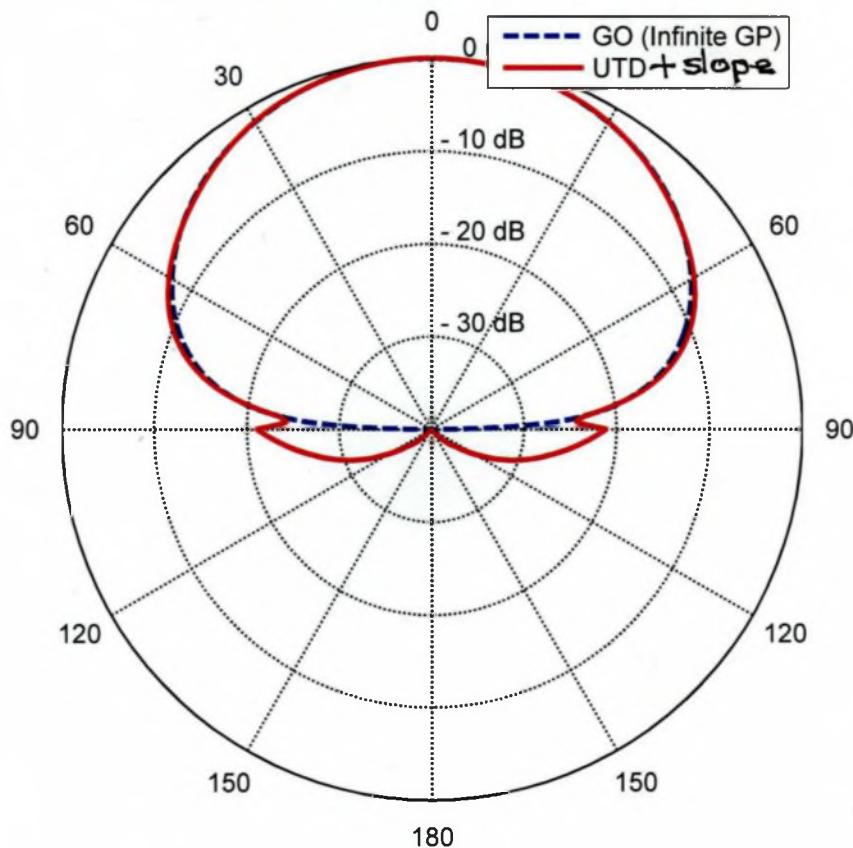
$\frac{\cos(\beta a/2)}{(\frac{\beta a}{2})^2 - (\frac{\pi}{2})^2}$ is indeterminate and must be evaluated by

L'Hopital's rule as $\lim_{a \rightarrow \lambda/2} \frac{\cos(\beta a/2)}{(\frac{\beta a}{2})^2 - (\frac{\pi}{2})^2} = -\frac{\sin(\beta a/2)}{2(\beta a/2)} \Big|_{\beta a/2 = \pi/2} = -\frac{1}{\pi}$

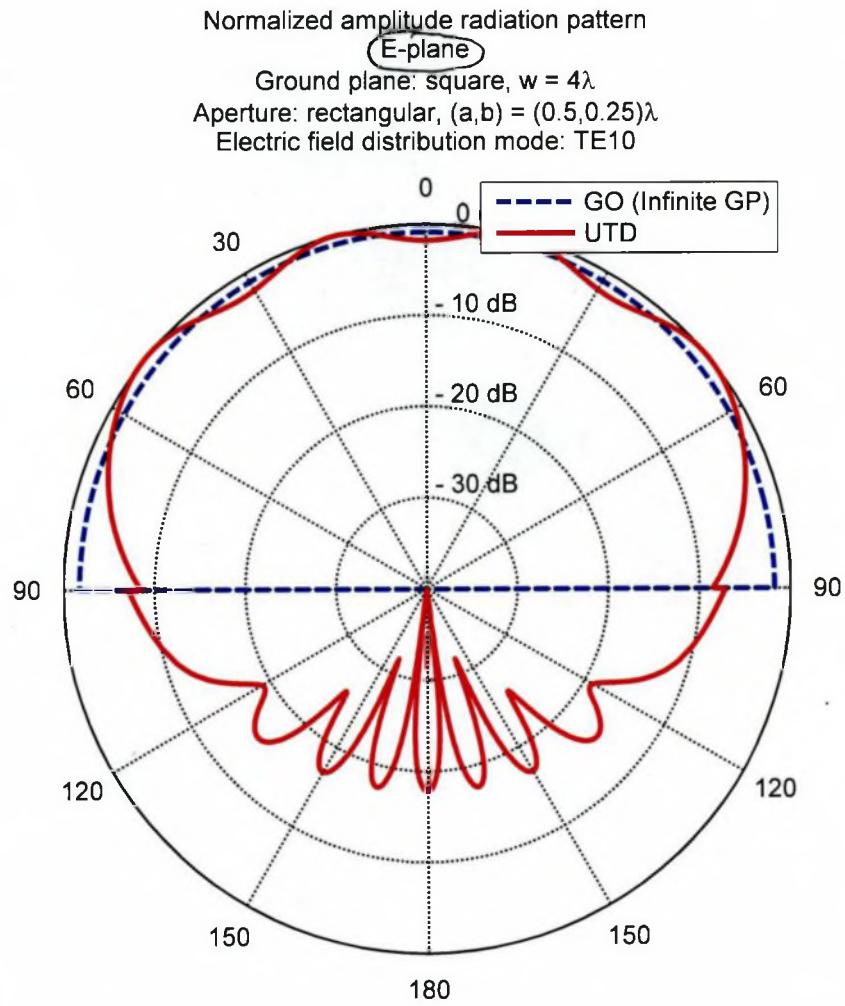
Normalized amplitude radiation pattern

(H-plane)

Ground plane: square, $w = 4\lambda$
Aperture: rectangular, $(a,b) = (0.5, 0.25)\lambda$
Electric field distribution mode: TE10

Cont'd

13.54 Cont'd



$$1355 \quad (a) \quad H_2^d = H_0 H_0^{(0)}(\beta p) \xrightarrow{p \rightarrow \infty} H_0 \sqrt{\frac{2j}{\pi \beta p}} e^{-j \beta p} = H_1 \frac{e^{-j \beta p}}{\sqrt{\beta p}} \text{ where } H_1 = H_0 \sqrt{\frac{2j}{\pi}}$$

$$H_{21}^d = H_2^d(p = \frac{w}{2}) D_h A_1 e^{-j \beta p \psi_1}, \quad H_2^d(p = \frac{w}{2}) = H_1 \frac{e^{-j \beta w/2}}{\sqrt{\beta w/2}}$$

$$D_h^2 = \frac{e^{-j \pi/4}}{\sqrt{2 \pi \beta}} \frac{\sin(\frac{\pi}{n})}{\left\{ \cos(\frac{\pi}{n}) - \cos(\frac{\psi_1 - \psi_1'}{n}) \right\}} \left\{ \frac{1}{\cos(\frac{\pi}{n}) - \cos(\frac{\psi_1 + \psi_1'}{n})} \right\}_{\substack{n=3/2 \\ \psi_1' = 0}}$$

$$= \frac{2}{3} \frac{e^{-j \pi/4}}{\sqrt{2 \pi \beta}} \frac{\sin(\frac{2\pi}{3})}{\left\{ \cos(\frac{2\pi}{3}) - \cos(\frac{2\psi_1}{3}) \right\}} + \frac{1}{\cos(\frac{2\pi}{3}) - \cos(\frac{2\psi_1}{3})}$$

For Observation at edge #2, $\psi_1 = 270^\circ$. Thus

$$D_h^2 = \frac{4}{3} \frac{e^{-j \pi/4}}{\sqrt{2 \pi \beta}} \frac{\sin(\frac{2\pi}{3})}{\left\{ \cos(120^\circ) - \cos(180^\circ) \right\}} = \frac{4(0.867)}{3 \sqrt{2 \pi \beta}} e^{-j \pi/4} \left\{ \frac{1}{-0.5 - (-1)} \right\}$$

$$D_h^2 = \frac{2.312 e^{-j \pi/4}}{\sqrt{2 \pi \beta}}$$

$$A_1 = \frac{1}{\sqrt{w}},$$

Thus at the edge #2, the field is given by

$$H_{21}^d (\text{at } \#2) = H_1 \frac{e^{-j \beta w/2}}{\sqrt{\beta w/2}} \frac{2.312 e^{-j \pi/4}}{\sqrt{2 \pi \beta}} \frac{e^{-j \beta w}}{\sqrt{w}} = H_1 \frac{2.312}{\sqrt{\pi}} \frac{e^{-j 3 \beta w/2}}{\beta w} e^{-j \pi/4}$$

$$H_{21}^d = 1.3044 H_1 \frac{e^{-j 3 \beta w/2}}{\beta w} e^{-j \pi/4}$$

The field diffracted by wedge #2 is given by

$$H_{22}^d = \frac{H_{21}^d (\text{at } \#2)}{2} D_h^2 A_2 e^{-j \beta w/2}$$

Because of the geometry

$$D_h^2 = D_h^1 = \frac{2.312 e^{-j \pi/4}}{\sqrt{2 \pi \beta}}, \quad A_2 = \frac{1}{\sqrt{w/2}}$$

Thus

$$H_{22}^d = \frac{1.3044 H_1}{2} \frac{e^{-j 3 \beta w/2}}{\beta w} e^{-j \pi/4} \left[\frac{2.312 e^{-j \pi/4}}{\sqrt{2 \pi \beta}} \right] \frac{1}{\sqrt{w/2}} e^{-j \beta w/2}$$

$$H_{22}^d = \frac{0.6522(2.312)}{\sqrt{\pi}} H_1 \frac{e^{-j 2 \beta w}}{(\beta w)^{3/2}} e^{-j \pi/2} = -j 0.85073 H_1 \frac{e^{-j 2 \beta w}}{(\beta w)^{3/2}}$$

Using the symmetry of the problem

$$H_{24}^d = H_{22}^d = -j 0.85073 H_1 \frac{e^{-j 2 \beta w}}{(\beta w)^{3/2}}$$

cont'd.

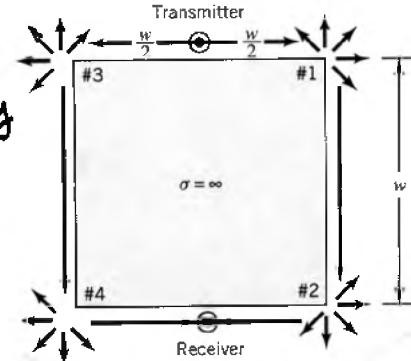


Figure P13-55

13.55 cont'd. Therefore the total field at the observation is

$$H_{2t}^d = H_{22}^d + H_{24}^d = -j 1.70147 H_1 \frac{e^{-j 2\beta w}}{(\beta w)^{3/2}}$$

$$(b) H_2^L = H_1 \frac{e^{-j \beta p}}{\sqrt{\beta p}} \Big|_{p=w} = H_1 \frac{e^{-j \beta w}}{\sqrt{\beta w}}$$

$$\left| \frac{H_{2t}^d}{H_2^L} \right| = \left| \frac{-j 1.70147 H_1 \frac{e^{-j 2\beta w}}{(\beta w)^{3/2}}}{H_1 \frac{e^{-j \beta w}}{(\beta w)^{1/2}}} \right| = \left| \frac{1.70147}{\beta w} \right| \Big|_{w=5} = \frac{1.70147}{10\pi} = 0.05416$$

$$\left| \frac{H_{2t}^d}{H_2^L} \right| = 0.05416 \text{ or } 20 \log_{10}(0.05416) = -25.326 \text{ dB}$$

CHAPTER 14

14.1 The duality in electromagnetics, as summarized in Table 7-2, going from the fields due to electric current \underline{J} to those due to a magnetic current density \underline{M} , is accomplished by basically replacing E by \underline{H} , H by $-E$, η by $1/\eta$, μ by ϵ , ϵ by μ , β by β , J by M . When this is done in (14-2), we can write its dual as

$$\underline{H} - (\hat{n} \cdot \underline{H}) \hat{n} = -\frac{\eta_0}{\eta} \hat{n} \times \underline{E}$$

where $\eta = Z/Z_0$, where Z_0 is the free space value of 377 (constant)

The basic objective of this is that if we obtain the solution of one polarization with normalized impedance η , this solution is identical to the solution of a polarization with normalized impedance $1/\eta$.

14.2 For TM^z (soft polarization), the impedance boundary conditions based on the geometry of 14-1 (both interior and exterior wedges) are given by

$$E_z = -\eta_0 Z_0 H_p \text{ for } \phi = 0$$

$$E_z = \eta_n Z_0 H_p \text{ for } \phi = n\pi$$

TM^z : Using Maxwell's equations

$$\frac{1}{\rho} \frac{\partial E_z}{\partial \phi} = -j\omega \mu H_p \Rightarrow H_p = -\frac{4}{j\omega \mu \rho} \frac{1}{\rho} \frac{\partial E_z}{\partial \phi}$$

the above two conditions can be reduced to

$$E_z = -\eta_0 Z_0 \left(-\frac{1}{j\omega \mu \rho} \frac{1}{\rho} \frac{\partial E_z}{\partial \phi} \right) \Rightarrow \frac{\partial E_z}{\partial \phi} - j \frac{\beta_p}{\eta_0} E_z = 0 \quad @ \phi = 0$$

$$E_z = +\eta_n Z_0 \left(-\frac{1}{j\omega \mu \rho} \frac{1}{\rho} \frac{\partial E_z}{\partial \phi} \right) \Rightarrow \frac{\partial E_z}{\partial \phi} + j \frac{\beta_p}{\eta_n} E_z = 0 \quad @ \phi = n\pi$$

14.3] For the TE^z polarization (hard), the impedance boundary conditions based on the geometry of (14-1) (both interior and exterior wedges) are given by

$$E_p = \eta_0 Z_0 H_z \text{ for } \phi=0$$

$$E_p = -\eta_n Z_0 H_z \text{ for } \phi=\pi$$

TE^z : Using Maxwell's equations

$$\frac{1}{j\rho} \frac{\partial H_z}{\partial \phi} = j\omega E_p \Rightarrow E_p = \frac{1}{j\omega \epsilon} \frac{1}{\rho} \frac{\partial H_z}{\partial \phi}$$

the above two conditions can be reduced to

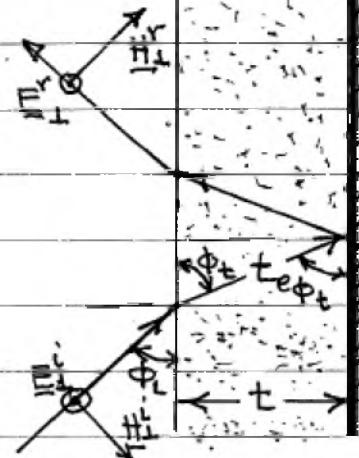
$$\frac{1}{j\omega \epsilon} \frac{1}{\rho} \frac{\partial H_z}{\partial \phi} = \eta_0 Z_0 H_z \Rightarrow \frac{\partial H_z}{\partial \phi} - j\beta \rho \eta_0 H_z = 0 @ \phi=0$$

$$\frac{1}{j\omega \epsilon} \frac{1}{\rho} \frac{\partial H_z}{\partial \phi} = \eta_n Z_0 H_z \Rightarrow \frac{\partial H_z}{\partial \phi} + j\beta \rho \eta_n H_z = 0 @ \phi=\pi$$

14.4 Soft Polarization (\perp polarization) ϵ_1, μ_1

Based on (14-4a), the reflection coefficient of a flat surface with a normalized surface impedance η is

$$\Gamma_s = \frac{E^r}{E^i} = \frac{\eta \sin \phi_i - 1}{\eta \sin \phi_i + 1} \quad (1)$$



If we allow for refraction, the above equation can be written based on (15-17a) as

$$\Gamma_s = \frac{E^r}{E^i} = \frac{\eta \sin \phi_i - \sin \phi_t}{\eta \sin \phi_i + \sin \phi_t} \quad (2)$$

Based on Snell's law of refraction

$$\beta_0 \cos^2 \phi_i = \beta_1 \cos^2 \phi_t \Rightarrow \phi_t = \cos^{-1} \left[\frac{\beta_0 \cos \phi_i}{\beta_1} \right]^{1/2} = \cos^{-1} \left[\sqrt{\frac{\mu_0}{\mu_1}} \cos \phi_i \right]^{1/2} \quad (3)$$

If ϵ_1, μ_1 , or $\epsilon_1 \mu_1$ are each large ($\epsilon_1 \gg 1, \mu_1 \gg 1, \mu_1 \epsilon_1 \gg 1$), then (3) reduces to

$$\phi_t = \cos^{-1} \left[\sqrt{\frac{\mu_0 \epsilon_0}{\mu_1 \epsilon_1}} \cos \phi_i \right]^{1/2} \xrightarrow{\epsilon_1, \mu_1 \gg 1} \cos^{-1}(0) = 90^\circ \Rightarrow \text{normal incidence.}$$

which indicates that (2) reduces to (1). This implies that the impedance boundary conditions are most accurate near normal incidence which will be most valid for materials with high permittivities and/or high permeabilities.

If we place a dielectric slab of thickness t to cover the PEC ground plane, as shown above, the reflection coefficient can be written as that of (2) but with η replaced with an equivalent normalized surface impedance η_e . For this case, we can rewrite (1) as

$$\Gamma_s = \frac{E^r}{E^i} = \frac{\eta_e \sin \phi_i - \sin \phi_t}{\eta_e \sin \phi_i + \sin \phi_t} \quad (4)$$

One way to compute η_e is to use transmission line theory.

(continued)

14.4 cont'd

The input reflection coefficient of a short transmission line, with characteristic impedance Z_c and length l , is

$$Z_{in} = jZ_c \tan(\beta l)$$

Treating the wave transmission in the dielectric covered ground plane as a shorted transmission line, we can write based on the figure of the previous page that the equivalent surface impedance η_e is

$$\eta_e = j\eta \tan(\beta_1 t \sin\phi_t) \quad (5)$$

since the wave travels in the slab through an oblique distance t_e

$$\sin\phi_t = t/t_e \Rightarrow t = t_e \sin\phi_t \quad (6)$$

and where $\beta_1 = \omega \sqrt{\mu_s \epsilon_s}$.

Therefore based on (5), the reflection coefficient of (4) can be written, using (5), as

$$\Gamma_s = \frac{j\eta \tan(\beta_1 t \sin\phi_t) \sin\phi_i - \sin\phi_t}{j\eta \tan(\beta_1 t \sin\phi_t) \sin\phi_i + \sin\phi_t} \quad (7)$$

If the permeability μ_s or permittivity ϵ_s , or both are very large, then (7) reduces to ($\phi_t \approx 90^\circ$, normal incidence)

$$\Gamma_s \approx \frac{j\eta \tan(\beta_1 t) \sin\phi_i - \sin\phi_t}{j\eta \tan(\beta_1 t) \sin\phi_i + \sin\phi_t} \quad (8)$$

14.5 Hard Polarization (II Polarization)

Using (14-4b)

$$\Gamma_h = \frac{E^r}{E^i} = \frac{\eta - \sin\phi_i}{\eta + \sin\phi_i} \quad (14-4b)$$

(5-24c) but permit refraction, we can write it as (with $\phi_i = \frac{\pi}{2} - \theta_i$ and $\phi_t = \frac{\pi}{2} - \theta_t$)

$$\Gamma_h = \frac{E^r}{E^i} = \frac{\eta \sin\phi_t - \sin\phi_i}{\eta \sin\phi_t + \sin\phi_i}$$

Following the procedure of the solution of Problem 14.4, by using transmission line theory where we can write the above equation as

$$\Gamma_h = \frac{E^r}{E^i} = \frac{\eta \sin\phi_t - \sin\phi_i}{\eta \sin\phi_t + \sin\phi_i}$$

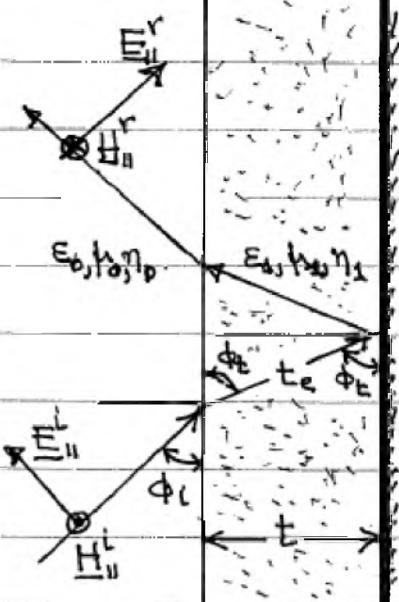
where $\eta = \text{equivalent surface impedance} = j\eta \tan(\beta_1 t \sin\phi_t)$
then it can be expressed as

$$\boxed{\Gamma_h = \frac{E^r}{E^i} = \frac{j\eta \tan(\beta_1 t \sin\phi_t) - \sin\phi_i}{j\eta \tan(\beta_1 t \sin\phi_t) + \sin\phi_i}}$$

which, if the permeability μ_2 or permittivity ϵ_2 , or both are very large, then the above equation reduces (because $\phi_t \approx 90^\circ$, near normal incidence) to

$$\boxed{\Gamma_h = \frac{E^r}{E^i} \approx \frac{j\eta \tan(\beta_1 t) - \sin\phi_i}{j\eta \tan(\beta_1 t) + \sin\phi_i}}$$

where $\beta_1 = \omega \sqrt{\mu_1 \epsilon_1}$.



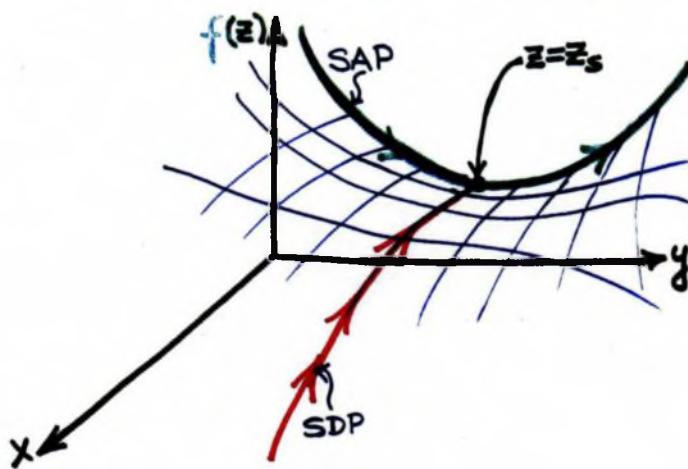
14.6

The saddle-point is so named because it lies at a point in the complex plane at which:

$$\left| e^{\beta \rho h(z)} \right|_{z=z_s} = \left| e^{\beta \rho h(z_s)} \right| = \begin{cases} \text{maximum} & (\text{along one direction}) \\ \text{minimum} & (\text{along a perpendicular direction}) \end{cases}$$

In three dimensions, a plot of $|e^{\beta \rho h(z)}|$ would take the shape of a saddle with $z = z_s$ at the center of the saddle.

If the function $H(z)$ is smoothly varying near the saddle point and if the contour C follows the steepest *descent* path through $z = z_s$, then the integrand is maximum at the saddle point and negligible elsewhere along the path. If the contour follows the steepest *ascent* path through $z = z_s$, then the integrand is minimum at the saddle point and large elsewhere along the path.



The steepest descent (ascent) path equation states that the *imaginary* part of $h(z)$ is *constant* (phase variations are nearly constant) because along these paths the *real part* of $h(z)$ (amplitude variation) is varying *most rapidly*.

Hence $|e^{\beta\rho h(z)}|$ either increases (or decreases) as rapidly as possible along the path.

$$h(z) = h_r(z) + h_i(z); \quad z = x + jy$$

The steepest descent path is obtained by equatin

$$h_i(z) = h_i(z_s); \text{ Steepest Descent Path}$$

The same equation is used for the ascent path.

Example: $e^{\beta\rho h(z)} = e^{j\beta\rho \cos(z)} = e^{\beta\rho[j \cos(z)]}$

$$h(z) = j \cos(z) = h_r(z) + h_i(z)$$

$$j \cos(z) = j \cos(x + jy) = h_r(z) + h_i(z)$$

$$\begin{aligned} j \cos(z) &= j \cos(x + jy) = j \cos(x) \cos(jy) \\ &\quad - j \sin(x) \sin(jy) \end{aligned}$$

$$j \cos(z) = j \cos(x) \cosh(y) + \sin(x) \sinh(y)$$

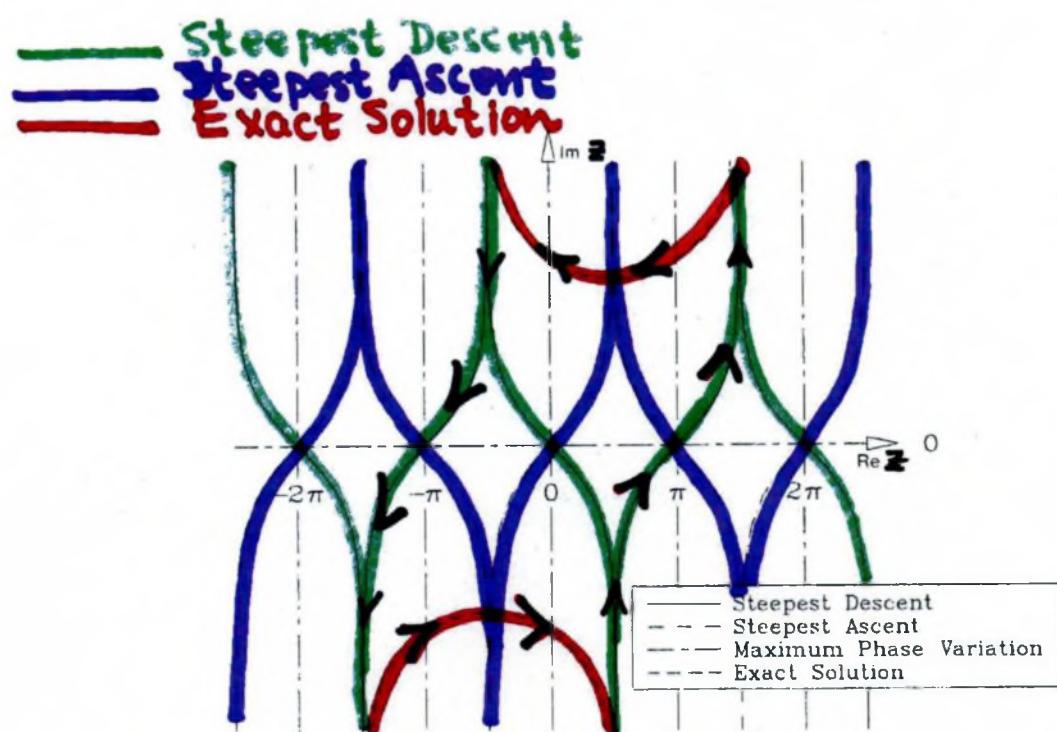
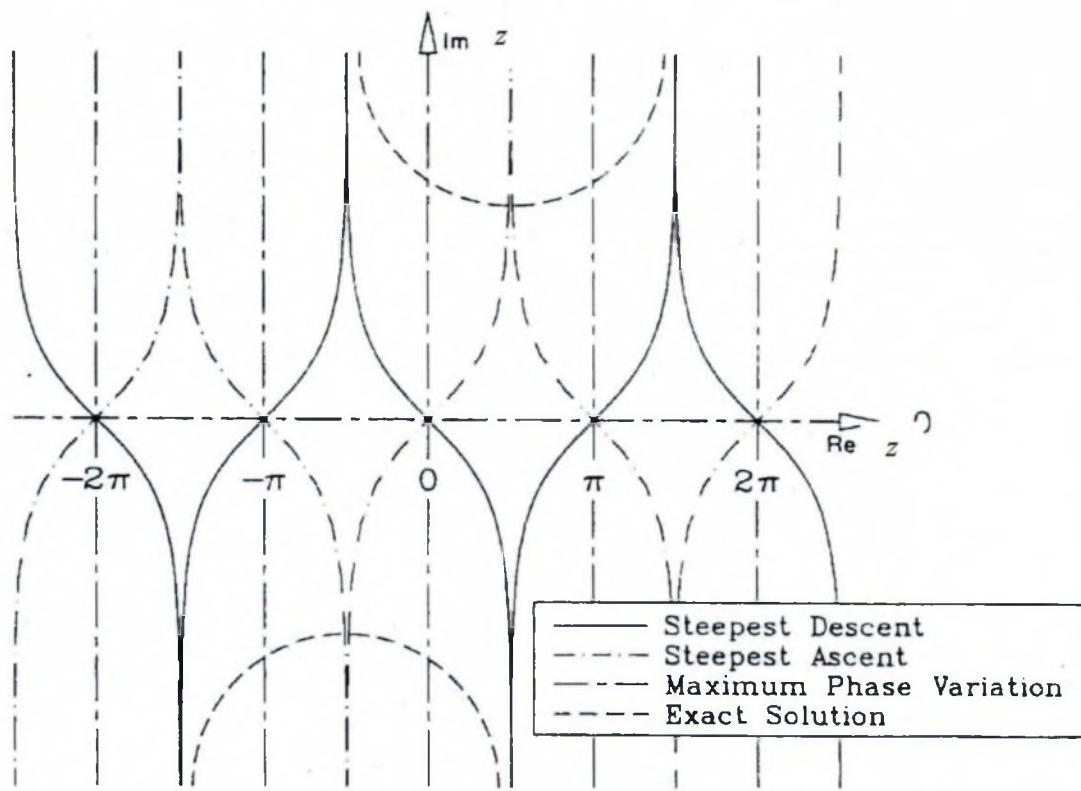
$$h_i(z) = \cos(x) \cosh(y)$$

$$h_r(z) = \sin(x) \sinh(y)$$

$$\begin{aligned} I_{\max}[j \cos(z)] &= I_{\max}[j \cos(\pm\pi)] \\ &= I_{\max}\left[j(-1)^m\right]; m = \pm 1 \end{aligned}$$

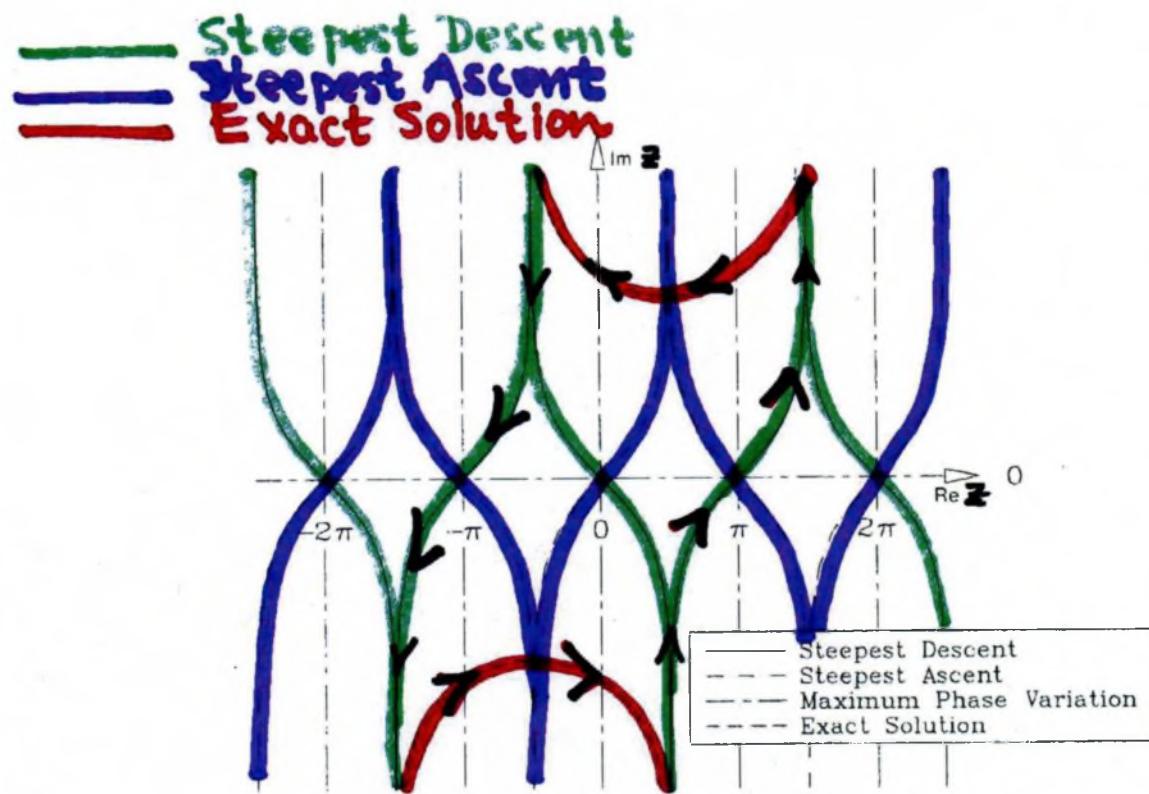
$h_i(z) = h_i(z_s) \Rightarrow$ Steepest Descent Paths

$\cos(x) \cosh(y) = -1 \quad$ Steepest Descent Paths



Continuation of the Solution of Problem 14.6

In contrast, the rapidly varying phase paths used for the Method of Stationary Phase are those for which the *real part* of $h(z)$ (amplitude variation) is *constant*, which means that the *imaginary part* of $h(z)$ (phase variations) vary as *rapidly as possible* away from the stationary saddle points.



Example: $e^{\beta\rho h(z)} = e^{j\beta\rho \cos(z)} = e^{\beta\rho[j \cos(z)]}$

$$h(z) = j \cos(z) = h_r(z) + h_i(z)$$

$$j \cos(z) = j \cos(x + jy) = h_r(z) + h_i(z)$$

$$\begin{aligned} j \cos(z) &= j \cos(x + jy) = j \cos(x) \cos(jy) \\ &\quad - j \sin(x) \sin(jy) \end{aligned}$$

$$j \cos(z) = j \cos(x) \cosh(y) + \sin(x) \sinh(y)$$

$$h_i(z) = \cos(x) \cosh(y)$$

$$h_r(z) = \sin(x) \sinh(y)$$

$$I_{\max}[j \cos(z)] = I_{\max}[j \cos(\pm\pi)]$$

$$= I_{\max}\left[j(-1)^m\right]; m = \pm 1$$

$$h_i(z) = h_i(z_s) \Rightarrow \text{steepest descent paths}$$

$$\boxed{\cos(x) \cosh(y) = -1} \quad \text{Steepest Descent Paths}$$

$$h_r(z) = h_r(z_s) \Rightarrow \text{stationary phase paths}$$

$$\boxed{\sin(x) \sinh(y) = 0} \quad \text{Stationary Phase Paths}$$

14.8 Wedge = $90^\circ \Rightarrow n = 3/2$, $\phi' = 60^\circ$, $\phi = 180^\circ$

$$\varepsilon_1 = 4\varepsilon_0, f_{\perp} = 1 \Rightarrow \eta_1 = \sqrt{\frac{f_{\perp}}{\varepsilon_1}} = \sqrt{\frac{f_{\perp}}{4\varepsilon_0}} = \frac{\eta_0}{2} \Rightarrow \eta = \frac{\eta_1}{\eta_0} = \frac{1}{2}$$

Hard Polarization

$$\theta_0 = \theta_n = \sin^{-1}(\eta_0) = \sin^{-1}(n_0) = \sin^{-1}(1/2) = 30^\circ \quad \begin{cases} \sin \theta_0 = 1/2 \\ \sin \theta_n = 1/2 \end{cases}$$

- The incident diffracted field, based on the asymptotic form, is given by the first term within the brackets of (14-59).

Using a Matlab computer program developed by the author and his students, the computed value for the incident diffracted field based on the above parameters is:

$$U^i = (-9.4724 + j9.4339) \times 10^{-2}$$

It should be noted that the observation point $\phi = 180^\circ$ is far removed from the incident ($\phi = 240^\circ$) and the reflection ($\phi = 120^\circ$) shadow boundaries, such that the

Fresnel transition functions of (14-59) are basically nearly unity.

- The reflected diffracted field, based on the asymptotic form, is given by the second term within the brackets of (14-59). Again using a Matlab program developed by the author and his students, the computed value for the reflected diffracted field based on the above parameters is:

$$U^r = (4.107 - j4.094) \times 10^{-2}$$

It should be noted again, that the observation point $\phi = 180^\circ$ is far removed from the incident ($\phi = 240^\circ$) and the reflection ($\phi = 120^\circ$) shadow boundaries such that the Fresnel transition functions of (14-59) are basically nearly unity.

14.9 Wedge = $90^\circ \Rightarrow n = 3/2, \phi' = 60^\circ, \phi = 180^\circ$

$$\epsilon_1 = 4\epsilon_0, \mu_1 = \mu_0 \Rightarrow \eta_1 = \sqrt{\frac{\mu_1}{\epsilon_1}} = \sqrt{\frac{\mu_0}{4\epsilon_0}} = \frac{\eta_0}{2} \Rightarrow \eta = \frac{\eta_1}{\eta_0} = \frac{1}{2}$$

Soft Polarization:

$$t_0 = \theta_n = \sin^{-1}\left(\frac{1}{\eta_0}\right) = \sin^{-1}\left(\frac{1}{\eta_1}\right) = \sin^{-1}\left(\frac{1}{\sqrt{2}}\right) = \sin(45^\circ) \quad \begin{cases} \sin t_0 = 2 \\ \sin \theta_n = 2 \end{cases}$$

- The incident diffracted field, based on the asymptotic form, is given by the first term within the brackets of (14-59).

Using a Matlab computer program developed by the author and his students, the computed value for the incident diffracted field based on the above parameters is:

$$U^i = (-10.0249 + j 8.4852) \times 10^{-2}$$

It should be noted that the observation point $\phi = 180^\circ$ is far removed from the incident ($\phi = 240^\circ$) and the reflection ($\phi = 120^\circ$) shadow boundaries such that the

Fresnel transition functions of (14-59) are basically nearly unity.

- The reflected diffracted field, based on the asymptotic form, is given by the second term within the brackets of (14-59). Again using a Matlab program developed by the author and his students, the computed value for the reflected diffracted field based on the above parameters is:

$$U^r = (-5.0221 + j 6.5065) \times 10^{-2}$$

It should be noted again, that the observation point $\phi = 180^\circ$ is far removed from the incident ($\phi = 240^\circ$) and the reflection ($\phi = 120^\circ$) shadow boundaries such that the Fresnel transition functions of (14-59) are basically nearly unity.

14.10

The surface wave residue for the z_n pole is given by

$$U_{SW}^n = \frac{U_0 R e}{n} \left[\frac{\psi(z + \frac{n\pi}{2} - \phi) \sin(\phi/n)}{\psi(\frac{n\pi}{2} - \phi')} \cdot e^{j\beta p \cos z}, z_n \right]$$

Assuming that no EO poles coincide with the surface wave pole, and since the exponential has no finite poles, the residue can be determined as

$$z_n = \phi - n\pi - \pi - \theta_n$$

$$\begin{aligned} U_{SW}^n &= \frac{U_0}{n} \lim_{z \rightarrow z_n} \left[(z - z_n) \frac{\psi(z + \frac{n\pi}{2} - \phi) \sin(\phi/n)}{\psi(\frac{n\pi}{2} - \phi')} \cdot e^{j\beta p \cos z} \right] \\ &= \frac{U_0}{n} \left[e^{j\beta p \cos(\phi - n\pi - \pi - \theta_n)} \frac{\sin(\phi/2)}{\cos(-\pi - n\pi - \theta_n) - \cos(\phi/n)} \right] \\ &\quad \cdot \lim_{z \rightarrow z_n} \left[(z - z_n) \psi(z + \frac{n\pi}{2} - \phi) \right] \end{aligned}$$

Since:

$$\cos(\phi - n\pi - \pi - \theta_n) = \cos(\phi - n\pi - \theta_n - \pi) = -\cos(\phi - n\pi - \theta_n)$$

$$\cos(-\pi - n\pi - \theta_n) = \cos(\pi + n\pi + \theta_n)$$

thus

$$\begin{aligned} U_{SW}^n &= \frac{U_0}{n} \left[e^{j\beta p \cos(\phi - n\pi - \theta_n)} \frac{\sin(\phi/2)}{\cos(\pi + n\pi + \theta_n) - \cos(\phi/n)} \right] \\ &\quad \cdot \lim_{z \rightarrow z_n} \left[(z - z_n) \psi(z + \frac{n\pi}{2} - \phi) \right] \end{aligned}$$

$$\lim_{z \rightarrow z_n} \left[(z - z_n) \psi(z + \frac{n\pi}{2} - \phi) \right] = \lim_{z \rightarrow z_n} \left[(z - z_n) \psi_n(z + n\pi - \phi + \frac{\pi}{2} - \theta_0) \right].$$

$$\psi_n(z + n\pi - \phi - \frac{\pi}{2} + \theta_0) \cdot \psi_n(z - \phi + \frac{\pi}{2} - \theta_n) \cdot \psi_n(z - \phi - \frac{\pi}{2} + \theta_n)$$

Because $\psi_n(-z) = \psi_n(z)$

Cont'd

14.10 (cont'd) Because $\psi_n(-z) = \psi_n(z)$

$$\psi_n(z + n\pi - \phi + \frac{\pi}{2} - \theta_0) \Big|_{z \rightarrow z_n} = \psi_n\left(\frac{\pi}{2} - (\theta_n + \theta_0)\right) = \psi_n\left(\frac{\pi}{2} + \theta_0 + \theta_n\right)$$

$$\psi_n(z + n\pi - \phi - \frac{\pi}{2} + \theta_0) \Big|_{z \rightarrow z_n} = \psi_n\left[-\frac{3\pi}{2} - (\theta_n - \theta_0)\right] = \psi_n\left[\frac{3\pi}{2} + \theta_n - \theta_0\right]$$

$$\psi_n(z - \phi + \frac{\pi}{2} - \theta_n) \Big|_{z \rightarrow z_n} = \psi_n\left(-\frac{\pi}{2} - n\pi - 2\theta_n\right) = \psi_n\left(\frac{\pi}{2} + n\pi + 2\theta_n\right)$$

$$\psi_n(z - \phi - \frac{\pi}{2} + \theta_n) \Big|_{z \rightarrow z_n} = \psi_n\left(-n\pi - \frac{3\pi}{2}\right) = \psi_n\left(n\pi + \frac{3\pi}{2}\right)$$

As $z \rightarrow z_n$, the fourth Maluzhinets function $\psi_n(n\pi + \frac{3\pi}{2})$ has a singularity.

Therefore

$$\lim_{z \rightarrow z_n} [(z - z_n) \psi_n(z + \frac{n\pi}{2} - \phi)] = \psi_n\left(\frac{\pi}{2} + \theta_0 + \theta_n\right) \psi_n\left(\frac{3\pi}{2} + \theta_n - \theta_0\right) \psi_n\left(\frac{\pi}{2} + n\pi + 2\theta_n\right)$$

$$\lim_{z \rightarrow z_n} [(z - z_n) \psi_n(z - \phi - \frac{\pi}{2} + \theta_n)]$$

$$\lim_{z \rightarrow z_n} [(z - z_n) \psi_n(z - \phi - \frac{\pi}{2} + \theta_n)] = \lim_{z \rightarrow z_n} [(z - z_n) \psi_n(z - z_0 + n\pi + \frac{3\pi}{2})]$$

$$= \lim_{z \rightarrow z_n} [z \psi_n(z + n\pi + \frac{3\pi}{2})] = \sin\left(\frac{\pi}{2n}\right) \psi_n\left(n\pi - \frac{\pi}{2}\right) \lim_{z \rightarrow z_n} [z \csc\left(\frac{\pi}{2n}\right)] \\ = 2n \sin\left(\frac{\pi}{2n}\right) \psi\left(n\pi - \frac{\pi}{2}\right)$$

$$\text{Substituting these } \lim_{z \rightarrow z_n} [(z - z_n) \psi_n(z + \frac{n\pi}{2} - \phi)] = \lim_{z \rightarrow z_n} [(z - z_n) \psi_n(z - \phi - \frac{\pi}{2} + \theta_n)]$$

$$= \sin\left(\frac{\pi}{2n}\right) \psi_n\left(n\pi - \frac{\pi}{2}\right) \lim_{z \rightarrow z_n} [z \csc\left(\frac{\pi}{2n}\right)]$$

$$\lim_{z \rightarrow z_n} [(z - z_n) \psi_n(z + \frac{n\pi}{2} - \phi)] = 2n \sin\left(\frac{\pi}{2n}\right) \psi\left(n\pi - \frac{\pi}{2}\right)$$

We can write the surface wave contribution from face n as

$$U_{SW}^n = \frac{U_0}{n} \begin{bmatrix} e^{-jBPC \cos(\phi - n\pi - \theta_n)} & \sin(\phi/2) \\ \psi\left(\frac{n\pi}{2} - \phi\right) & \frac{\cos\left(\frac{\pi}{2} + n\pi - \theta_n\right) - \cos(\phi/n)}{n} \end{bmatrix} \cdot 2n \sin\left(\frac{\pi}{2n}\right)$$

$$\cdot \psi_n\left(n\pi - \frac{\pi}{2}\right) \psi_n\left(\frac{\pi}{2} + \theta_0 + \theta_n\right) \cdot \psi_n\left(\frac{3\pi}{2} + \theta_n - \theta_0\right) \cdot \psi_n\left(\frac{\pi}{2} + n\pi + 2\theta_n\right)$$

cont'd

14.10 cont'd

The final form can be written, by rearranging the terms and their arguments, as :

$$U_{sw}^n = U_0 \begin{bmatrix} 2 \sin\left(\frac{\pi}{2n}\right) & \sin\left(\frac{\phi}{2}\right) e^{j\beta p \cos(\phi - n\pi - \theta_n)} \\ \psi\left(\frac{n\pi}{2} - \phi'\right) & \cos\left(\frac{n\pi + \pi + \theta_n}{n}\right) - \cos\left(\frac{\phi}{n}\right) \end{bmatrix}.$$

$$\bullet \Psi_n\left(n\pi - \frac{\pi}{2}\right) \Psi_n\left(\frac{\pi}{2} + n\pi + 2\theta_n\right) \Psi_n\left(\frac{3\pi}{2} + \theta_n - \theta_0\right) \Psi_n\left(\frac{\pi}{2} + \theta_0 + \theta_n\right)$$

14.11 Using (14-47) and the second term from (14-53)

$$H_2(z) = \frac{\Psi(z + \frac{n\pi}{2} - \phi)}{\Psi(\frac{n\pi}{2} - \phi')} \cot\left(\frac{\xi - z}{\frac{2\pi}{2n}}\right)$$

$$h_2(z) = j \cos(z), z_s = +\pi, z_p = \xi - 2\pi N^- n$$

$$h_2'(z) = -j \sin(z), h_2''(z) = -j \cos(z), h_2(z_s = \pi) = +j$$

$$\left| \sqrt{\frac{-2\pi}{\beta p h_2''(z_s)}} \right| = \left| \sqrt{\frac{-2\pi}{\beta p(j)}} \right| = \sqrt{\frac{2\pi}{\beta p}}$$

$$U_{SDP}^{(2)} = j U_0 \frac{1}{4\pi n} \frac{e^{-j\beta p}}{\sqrt{p}} \sqrt{\frac{2\pi}{\beta}} e^{-j\pi/4} \frac{\Psi(\pi + \frac{n\pi}{2} - \phi)}{\Psi(\frac{n\pi}{2} - \phi')} \cot\left(\frac{\xi - \pi}{\frac{2\pi}{2n}}\right) F\left\{ \beta p [g(\xi)] \right\}$$

$$= j U_0 \frac{e^{-j\beta p}}{\sqrt{p}} \left[-j \frac{e^{-j\pi/4}}{2n\sqrt{2\pi\beta}} \right] \frac{\Psi(\pi + \frac{n\pi}{2} - \phi)}{\Psi(\frac{n\pi}{2} - \phi')} (-1) \cot\left(\frac{\pi - \xi}{\frac{2\pi}{2n}}\right) F\left\{ \beta p [g(\xi)] \right\}$$

$$U_{SDP}^{(2)} = + U_0 \frac{e^{-j\beta p}}{\sqrt{p}} \left[- \frac{e^{-j\pi/4}}{2n\sqrt{2\pi\beta}} \right] \frac{\Psi(\pi + \frac{n\pi}{2} - \phi)}{\Psi(\frac{n\pi}{2} - \phi')} \cot\left(\frac{\pi - \xi}{\frac{2\pi}{2n}}\right)$$

$$\cdot F\left\{ \beta p \left[1 + \cos\left(\xi - 2\pi N^- n\right) \right] \right\}$$

14.12] Using (14-47) and the third term from (14-53)

$$H_3(z) = \frac{\Psi(z + \frac{n\pi}{2} - \phi)}{\Psi(\frac{n\pi}{2} - \phi)} \cot\left(\frac{\xi^+ - z}{2n}\right)$$

$$h_3(z) = j \cos z, z_s = -\pi, z_p = \xi^+ - 2\pi N_+^+ n, h'_3(z) = -j \sin z, h''_3(z) = -j \cos(z)$$

$$\left| \sqrt{\frac{-2\pi}{\pi \rho h''_3(\xi^+)}} \right| = \left| \sqrt{\frac{2\pi}{\pi \rho(j)}} \right| = \sqrt{\frac{2\pi}{\beta \rho}}$$

$$U_{SDP}^{(3)} = -j U_0 \frac{1}{4\pi n} e^{-j\beta p} \frac{\Psi(-\pi + \frac{n\pi}{2} - \phi)}{\Psi(\frac{n\pi}{2} - \phi)} \cot\left(\frac{\xi^+ + \pi}{2n}\right) e^{j\pi/4} \sqrt{\frac{2\pi}{\beta \rho}} F\{\beta \rho g(\xi^+)\}$$

$$= -j U_0 \frac{e^{-j\beta p}}{4\pi n} \frac{\Psi(-\pi + \frac{n\pi}{2} - \phi)}{\Psi(\frac{n\pi}{2} - \phi)} \cot\left(\frac{\pi + \xi^+}{2n}\right) e^{-j\frac{\pi}{2}} \frac{e^{j\pi/4}}{-j} \sqrt{\frac{2\pi}{\beta \rho}} F\{\beta \rho g(\xi^+)\}$$

$$U_{SDP}^{(3)} = -U_0 \frac{e^{-j\beta p}}{\sqrt{\rho}} \left[\frac{e^{-j\pi/4}}{2n\sqrt{2\pi\rho}} \right] \frac{\Psi(\pi + \frac{n\pi}{2} - \phi)}{\Psi(\frac{n\pi}{2} - \phi)} \cot\left(\frac{\pi + \xi^+}{2n}\right) F\{\beta \rho [1 + \cos(\xi^+ - 2\pi N_+^+ n)]\}$$

14.13] Using (14-47) and the fourth term from (14-53)

$$H_4(z) = \frac{\Psi(z + \frac{n\pi}{2} - \phi)}{\Psi(\frac{n\pi}{2} - \phi)} \cot\left(\frac{\xi^+ - z}{2n}\right)$$

$$h_4(z) = j \cos z, z_s = +\pi, z_p = \xi^+ - 2\pi N_-^- n, h'_4(z) = -j \sin(z), h''_4(z) = -j \cos z$$

$$\left| \sqrt{\frac{-2\pi}{\pi \rho h''_4(\xi^+)}} \right| = \left| \sqrt{\frac{-2\pi}{\pi \rho(j)}} \right| = \sqrt{\frac{2\pi}{\beta \rho}}$$

$$U_{SDP}^{(4)} = -j U_0 \frac{e^{-j\beta p}}{4\pi n} \frac{\Psi(\pi + \frac{n\pi}{2} - \phi)}{\Psi(\frac{n\pi}{2} - \phi)} \cot\left(\frac{\xi^+ - \pi}{2n}\right) e^{-j\frac{3\pi}{4}} \sqrt{\frac{2\pi}{\beta \rho}} F\{\beta \rho g(\xi^+)\}$$

$$= -j U_0 \frac{e^{-j\beta p}}{4\pi n} \frac{\Psi(\pi + \frac{n\pi}{2} - \phi)}{\Psi(\frac{n\pi}{2} - \phi)} (-1) \cot\left(\frac{\pi - \xi^+}{2n}\right) e^{-j\frac{\pi}{2}} e^{-j\frac{\pi}{4}} F\{\beta \rho g(\xi^+)\}$$

$$U_{SDP}^{(4)} = -U_0 \frac{e^{-j\beta p}}{\sqrt{\rho}} \left[\frac{e^{-j\pi/4}}{2n\sqrt{2\pi\rho}} \right] \frac{\Psi(\pi + \frac{n\pi}{2} - \phi)}{\Psi(\frac{n\pi}{2} - \phi)} \cot\left(\frac{\pi - \xi^+}{2n}\right) F\{\beta \rho [1 + \cos(\xi^+ - 2\pi N_-^- n)]\}$$

CHAPTER 15

15.1 $\frac{d^2u}{dx^2} = \frac{1}{T} F(x) = f(x), \quad \frac{d^2G}{dx^2} = \frac{1}{T} \delta(x-x'), \quad \rho(x') = T$

The homogeneous form is

$$\frac{d^2u}{dx^2} = 0 \Rightarrow u_1(x) = A_1 x + B_1, \\ u_2(x) = A_2 x + B_2$$

$$\text{Since } u_1(x=0) = 0 \Rightarrow u_1(x) = A_1 x, \quad u_1'(x) = A_1, \\ u_2(x=\ell) = 0 \Rightarrow u_2(x) = A_2(x-\ell), \quad u_2'(x) = A_2$$

$$W(x) = u_1(x') u_2'(x') - u_2(x') u_1'(x) = A_1 x' A_2 - A_2 (x'-\ell) A_1 = A_1 A_2 \ell$$

$$G(x, x') = \begin{cases} \frac{1}{T} \frac{A_2(x'-\ell)}{A_1 A_2 \ell} A_1 x = \frac{1}{T} \frac{x'-\ell}{\ell} x = \frac{2}{T} \left(\frac{x'-\ell}{\ell} \right) x, & 0 \leq x \leq x' \\ \frac{1}{T} \frac{A_1 x'}{A_1 A_2 \ell} A_2 (x-\ell) = \frac{1}{T} \frac{x'}{\ell} (x-\ell) = \frac{1}{T} \frac{(x-\ell)}{\ell} x', & x' \leq x \leq \ell \end{cases}$$

15.2 $\frac{d^2u}{dx^2} + \beta^2 u(x) = f(x), \quad \beta^2 = \omega/c, \quad u(x=0) = u(x=\ell) = 0, \quad \gamma = \beta^2$

$$\frac{d^2u}{dx^2} + \beta^2 u(x) = 0 \Rightarrow u_1(x) = A_1 \cos(\beta x) + B_1 \sin(\beta x) \\ u_2(x) = A_2 \cos[\beta(x-\ell)] + B_2 \sin[\beta(x-\ell)]$$

$$\text{Since } u_1(x=0) = 0 \Rightarrow u_1(x) = B_1 \sin(\beta x), \quad u_1'(x) = B_1 \beta \cos(\beta x)$$

$$u_2(x=\ell) = 0 \Rightarrow u_2(x) = B_2 \sin[\beta(x-\ell)], \quad u_2'(x) = B_2 \beta \cos[\beta(x-\ell)]$$

$$W(x') = u_1(x') u_2'(x') - u_2(x') u_1'(x) = B_1 B_2 \beta \left\{ \sin(\beta x) \cos[\beta(x'-\ell)] - \sin[\beta(x'-\ell)] \cos(\beta x) \right\} \\ = \beta B_1 B_2 \left\{ \sin \beta x' \left[\cos \beta x' \cos \beta \ell + \sin \beta x' \sin \beta \ell \right] - \cos(\beta x') \left[\sin(\beta x') \cos \beta \ell - \cos(\beta x') \sin \beta \ell \right] \right\}$$

$$W(x) = \beta B_1 B_2 \sin \beta \ell \left[\sin^2 \beta x' + \cos^2 \beta x' \right] = \beta B_1 B_2 \sin(\beta \ell)$$

$$G(x, x') = \begin{cases} \frac{B_2 \sin[\beta(x-\ell)]}{\beta B_1 B_2 \sin(\beta \ell)} B_1 \sin(\beta x) = \frac{\sin[\beta(x-\ell)]}{\beta \sin(\beta \ell)} \sin(\beta x), & 0 \leq x \leq x' \\ \frac{B_1 \sin(\beta x')}{\beta B_1 B_2 \sin(\beta \ell)} B_2 \sin[\beta(x-\ell)] = \frac{\sin(\beta x')}{\beta \sin(\beta \ell)} \sin[\beta(x-\ell)], & x' \leq x \leq \ell \end{cases}$$

15.3 Closed Form

$$\nabla^2 V(x, x') = -\frac{1}{\epsilon_0} g(x) = \frac{\partial^2 V(x)}{\partial x^2} \Rightarrow \frac{\partial^2 V}{\partial x^2} = -\frac{1}{\epsilon_0} g(x)$$

(a) $V(x=0) = V(x=w) = 0$

$$P(x) = 1$$

$$\nabla^2 G(x, x') = \delta(x-x') = \frac{\partial^2 G(x, x')}{\partial x^2}$$

$$G(x=0, x') = G(x=w, x') = 0$$

(b) $\frac{\partial^2 G(x, x')}{\partial x^2} = 0 \Rightarrow G(x, x') = \begin{cases} A_1 x + B_1 & 0 \leq x \leq x' \\ A_2 x + B_2 & x' \leq x \leq w \end{cases}$

Applying the boundary conditions:

$$G(x=0, x') = A_1(0) + B_1 = 0 \Rightarrow B_1 = 0$$

$$G(x=w, x') = A_2(w) + B_2 = 0 \Rightarrow B_2 = -w A_2$$

$$G(x, x') = \begin{cases} A_1 x & y_1 = A_1 x, y'_1 = A_1 \\ A_2(x-w) & y_2 = A_2(x-w), y'_2 = A_2 \end{cases}$$

$$W = y_1(x)y'_2(x') - y_2(x)y'_1(x') = A_1 x'(A_2) - A_2(x'-w)A_1$$

$$W = A_1 A_2 x' - A_1 A_2 (x'-w) = A_1 A_2 (x' - x' + w) = w A_1 A_2$$

$$G(x, x') = \begin{cases} \frac{y_2(x')y_1(x)}{P(x)W(x')} = \frac{A_2(x-w)A_1(x)}{(1) w A_1 A_2} = \frac{1}{w}(x-w)x, & 0 \leq x \leq x' \\ \frac{y_1(x)y'_2(x)}{P(x)W(x)} = \frac{A_1 x [A_2(x-w)]}{(2) A_1 A_2 w} = \frac{1}{w} x(x-w), & x' \leq x \leq w \end{cases}$$

(c)

$$V(x) = \int_0^w f(x') G(x, x') dx' = -\frac{1}{\epsilon_0} \int_0^w g(x) G(x, x') dx'$$

$$= -\frac{1}{\epsilon_0 w} \left[\int_0^x g(x') \frac{1}{w} (x-w)x dx' + \int_x^w g(x') \frac{1}{w} (x-w)x' dx' \right]$$

$$V(x) = -\frac{1}{\epsilon_0 w} \left[\int_0^x g(x')(x-w)x dx' + \int_x^w g(x')(x-w)x' dx' \right]$$

15.4 Series Form

$$\nabla^2 V(x, x') = -\frac{1}{\epsilon_0} g(x') = \frac{d^2 V}{dx^2} \Rightarrow \frac{d^2 V}{dx^2} = -\frac{1}{\epsilon_0} g(x')$$

(a) $V(x=0) = V(x=W) = 0$, up ($x=1$)

$$G(x=0, x') = G(x=W, x') = 0$$

(b) $\frac{d^2 \psi_n(x)}{dx^2} + \beta_n^2 \psi_n(x) = 0 \Rightarrow \psi_n(x) = A \cos(\beta_n x) + B \sin(\beta_n x)$

Boundary conditions: $\psi_n(x=0) = 0 = A_1(1) + B_1(0) \Rightarrow A_1 = 0$

$$\psi_n(x=W) = 0 = B \sin(\beta_n W) = 0 \Rightarrow \sin(\beta_n W) = 0$$

$$\beta_n W = \sin^{-1}(0) = n\pi \Rightarrow \beta_n = \frac{n\pi}{W}$$

$$\beta_n = \frac{n\pi}{W}, n=1, 2, 3, \dots$$

$$\psi_n(x) = B \sin(\beta_n x), \lambda_n = (\beta_n)^2 = \left(\frac{n\pi}{W}\right)^2$$

$$G(x, x') = \sum_n \frac{\psi_n(x) \psi_n(x')}{2 - 2\lambda_n}$$

$$\int_0^W \psi_m(x) \psi_n(x) dx = \int_0^W B \sin(\sqrt{\lambda_n} x) B \sin(\sqrt{\lambda_m} x) dx = \begin{cases} 1 & m=n \\ 0 & m \neq n \end{cases}$$

$$\int_0^W \psi_n(x) \psi_n(x) dx = B^2 \int_0^W \sin^2(\sqrt{\lambda_n} x) dx = B^2 \left(\frac{W}{2}\right) = 1 \Rightarrow B^2 = \frac{2}{W}$$

$$G(x, x') = \frac{2}{W} \sum_{n=1}^{\infty} \frac{\sin\left(\frac{n\pi}{W} x'\right) \sin\left(\frac{n\pi}{W} x\right)}{2 - 2\lambda_n} = -\frac{2}{W} \sum_{n=1}^{\infty} \frac{\sin\left(\frac{n\pi}{W} x'\right) \sin\left(\frac{n\pi}{W} x\right)}{\lambda_n} \quad \lambda_n = \left(\frac{n\pi}{W}\right)^2$$

$$G(x, x') = -\frac{2W}{\pi^2} \sum_{n=1}^{\infty} \frac{1}{n^2} \sin\left(\frac{n\pi}{W} x'\right) \sin\left(\frac{n\pi}{W} x\right)$$

c) $V(x) = -\frac{1}{\epsilon_0} \int_0^W G(x, x') g(x') dx' = -\frac{1}{\epsilon_0} \int_0^W (-1) \frac{2W}{\pi^2} \sum_{n=1}^{\infty} \frac{1}{n^2} \sin\left(\frac{n\pi}{W} x'\right) \sin\left(\frac{n\pi}{W} x\right) g(x') dx'$

$$V(x) = \frac{2Wg}{\epsilon_0 \pi^2} \sum_{n=1}^{\infty} \frac{1}{n^2} \sin^2\left(\frac{n\pi}{W} x\right)$$

$$15.5 \quad \frac{d^2V}{d\phi^2} = f(\phi) = -\frac{1}{\epsilon_0} g(\phi), \quad V(\phi=0) = V(\phi=\alpha) = 0, \quad -p(\phi) = 1$$

The homogeneous form can be written as

$$\frac{d^2V}{d\phi^2} = 0 \Rightarrow V_1(\phi) = A_1\phi + B_1$$

$$V_2(\phi) = A_2\phi + B_2$$

$$\text{Since } V_1(\phi=0) = 0 \Rightarrow V_1(\phi) = A_1\phi, \quad V'_1(\phi) = A_1$$

$$V_2(\phi=\alpha) = 0 \Rightarrow V_2(\phi) = A_2(\phi-\alpha), \quad V'_2(\phi) = A_2$$

$$W(\phi') = V_1(\phi')V'_2(\phi') - V_2(\phi')V'_1(\phi')$$

$$W(\phi') = A_1\phi' A_2 - A_2(\phi'-\alpha)A_1 = A_1 A_2 \alpha$$

Now we can write the Green's function of (15-45a) and (15-45b) as

$$G(\phi, \phi') = \begin{cases} \frac{A_2(\phi'-\alpha)}{A_1 A_2 \alpha} A_1 \phi = \left(\frac{\phi'-\alpha}{\alpha}\right) \phi & , \quad 0 \leq \phi \leq \phi' \\ \frac{A_1 \phi}{A_1 A_2 \alpha} A_2(\phi-\alpha) = \left(\frac{\phi}{\alpha}\right)(\phi-\alpha) & , \quad \phi' \leq \phi \leq \alpha \end{cases}$$

$$15.6 \quad \frac{d^2V}{d\phi^2} = f(\phi) = -\frac{1}{\epsilon_0} g(\phi), \quad V(\phi=0) = 0, \quad V(\phi=\alpha) = V_0$$

$$\frac{d^2G(\phi, \phi')}{d\phi^2} = \delta(\phi-\phi'), \quad G(\phi=0) = 0, \quad G(\phi=\alpha) = V_0$$

$$V_1(\phi) = A_1\phi + B_1, \quad V'_1(\phi) = A_1, \quad V_1(\phi=0) = A_1(0) + B_1 = 0 \Rightarrow B_1 = 0, \quad \phi < \phi'$$

$$V_2(\phi) = A_2\phi + B_2, \quad V'_2(\phi) = A_2, \quad V_2(\phi=\alpha) = A_2(\alpha) + B_2 = V_0 \Rightarrow B_2 = V_0 - \alpha A_2, \quad \phi > \phi'$$

$$V_1(\phi) = A_1\phi, \quad V'_1(\phi) = A_1$$

$$V_2(\phi) = V_0 + A_2(\phi-\alpha), \quad V'_2(\phi) = A_2$$

$$\text{Continuity at } \phi = \phi': (1) \quad A_1\phi' = V_0 + A_2(\phi'-\alpha) : G_1(\phi') = G_2(\phi')$$

$$\text{Discontinuity at } \phi = \phi': (2) \quad A_2 - A_1 = 1 : \left| \frac{dG_1}{d\phi} \right|_{\phi=\phi'}, \left| \frac{dG_2}{d\phi} \right|_{\phi=\phi'} = 1$$

Solving (1) and (2) for A_1 and A_2 :

$$A_1 = \frac{1}{\alpha} [V_0 + (\phi' - \alpha)], \quad A_2 = \frac{1}{\alpha} [V_0 + \phi']$$

Thus

$$V_1(\phi) = A_1\phi = \frac{\phi}{\alpha} [V_0 + (\phi' - \alpha)] \quad \phi < \phi'$$

$$V_2(\phi) = V_0 + \frac{(\phi-\alpha)}{\alpha} [V_0 + \phi'] \quad \phi > \phi'$$

$$15.7 \quad \frac{\partial^2 V}{\partial x^2} + \frac{\partial^2 V}{\partial y^2} = f(x, y) = -\frac{1}{\epsilon_0} g(x, y), \quad V(x=0, 0 \leq y \leq b) = V(x=a, 0 \leq y \leq b) = 0$$

$$V(0 \leq x \leq a, y=0) = 0$$

$$V(0 \leq x \leq a, y=b) = V_0$$

The Green's function will be chosen to satisfy the partial differential equation of

$$\frac{\partial^2 G}{\partial x^2} + \frac{\partial^2 G}{\partial y^2} = \delta(x-x')\delta(y-y')$$

subject to the boundary conditions of

$$G(x=0, 0 \leq y \leq b) = G(x=a, 0 \leq y \leq b) = 0$$

$$G(0 \leq x \leq a, y=0) = 0, G(0 \leq x \leq a, y=b) = V_0$$

The potential can then be found using (15-127).

$$G(x, y; x', y') = \sum_{m=1, 2, \dots}^{\infty} g_m(y; x, y') \sin\left(\frac{m\pi}{a}x\right)$$

thus

$$\sum_{m=1, 2, \dots}^{\infty} \left[-\left(\frac{m\pi}{a}\right)^2 g_m(y; x, y') \sin\left(\frac{m\pi}{a}x\right) + \sin\left(\frac{m\pi}{a}x\right) \frac{d^2 g_m}{dy^2}(y; x, y') \right] = \delta(x-x')\delta(y-y')$$

Multiplying both sides by $\sin(n\pi x/a)$ and integrating with respect to x from 0 to a , and using (15-48a) we can write that

$$\frac{d^2 g_m}{dy^2}(y; x, y') - \left(\frac{m\pi}{a}\right)^2 g_m(y; x, y') = \frac{2}{a} \sin\left(\frac{m\pi}{a}x'\right) \delta(y-y')$$

$$g_m^{(1)} = A_m(x', y') \cosh\left(\frac{m\pi}{a}y\right) + B_m(x', y') \sinh\left(\frac{m\pi}{a}y\right) \quad y < y'$$

$$g_m^{(2)} = C_m(x', y') \cosh\left(\frac{m\pi}{a}y\right) + D_m(x', y') \sinh\left(\frac{m\pi}{a}y\right) \quad y > y'$$

$$g_m^{(1)}(y=0) = 0 = A_m(1) + B_m(0) \Rightarrow A_m = 0$$

$$g_m^{(2)}(y=b) = V_0 = C_m \cosh\left(\frac{m\pi}{a}b\right) + D_m \sinh\left(\frac{m\pi}{a}b\right) \Rightarrow D_m = \frac{1}{\sinh\left(\frac{m\pi}{a}b\right)} \left\{ V_0 - C_m \cosh\left(\frac{m\pi}{a}b\right) \right\}$$

Thus

$$g_m^{(1)} = B_m(x', y') \sinh\left(\frac{m\pi}{a}y\right)$$

$$g_m^{(2)} = C_m \cosh\left(\frac{m\pi}{a}y\right) + \left[V_0 - C_m \cosh\left(\frac{m\pi}{a}b\right) \right] \frac{\sinh\left(\frac{m\pi}{a}y\right)}{\sinh\left(\frac{m\pi}{a}b\right)}$$

Continuity of $g_m^{(1)}$ and $g_m^{(2)}$ at $y=y'$:

$$B_m \sinh\left(\frac{m\pi}{a}y'\right) = C_m \cosh\left(\frac{m\pi}{a}y'\right) + \left[V_0 - C_m \cosh\left(\frac{m\pi}{a}b\right) \right] \frac{\sinh\left(\frac{m\pi}{a}y'\right)}{\sinh\left(\frac{m\pi}{a}b\right)} \quad (1)$$

Discontinuity of $d g_m^{(1)}/d\phi|_{\phi=\phi'}$ and $d g_m^{(2)}/d\phi|_{\phi=\phi'}$:

$$\frac{m\pi}{a} \left\{ C_m \sinh\left(\frac{m\pi}{a}y'\right) + \left[V_0 - C_m \cosh\left(\frac{m\pi}{a}b\right) \right] \frac{\cosh\left(\frac{m\pi}{a}y'\right)}{\sinh\left(\frac{m\pi}{a}b\right)} - B_m \cosh\left(\frac{m\pi}{a}y'\right) \right\} = 1 \quad (2)$$

cont'd.

15.7 cont'd. (1) and (2) can be written as:

$$B_m \sinh\left(\frac{m\pi}{a}y'\right) - C_m \left[\cosh\left(\frac{m\pi}{a}y'\right) - \coth\left(\frac{m\pi b}{a}\right) \sinh\left(\frac{m\pi}{a}y'\right) \right] = V_0 \frac{\sinh\left(\frac{m\pi}{a}y\right)}{\sinh\left(\frac{m\pi b}{a}\right)} \quad (3)$$

$$-B_m \cosh\left(\frac{m\pi}{a}y'\right) + C_m \left[\sinh\left(\frac{m\pi}{a}y'\right) - \coth\left(\frac{m\pi b}{a}\right) \cosh\left(\frac{m\pi}{a}y'\right) \right] = \frac{a}{m\pi} - V_0 \frac{\cosh\left(\frac{m\pi}{a}y\right)}{\sinh\left(\frac{m\pi b}{a}\right)} \quad (4)$$

These can be written as

$$\alpha_1 B_m - \beta_1 C_m = \gamma_1 \quad (5)$$

$$-\alpha_2 B_m + \beta_2 C_m = \gamma_2 \quad (6)$$

where $\alpha_1 = \sinh\left(\frac{m\pi}{a}y'\right)$, $\beta_1 = \left[\cosh\left(\frac{m\pi}{a}y'\right) - \coth\left(\frac{m\pi b}{a}\right) \sinh\left(\frac{m\pi}{a}y'\right) \right]$, $\gamma_1 = V_0 \frac{\sinh\left(\frac{m\pi}{a}y\right)}{\sinh\left(\frac{m\pi b}{a}\right)}$
 $\alpha_2 = \cosh\left(\frac{m\pi}{a}y'\right)$, $\beta_2 = \left[\sinh\left(\frac{m\pi}{a}y'\right) - \coth\left(\frac{m\pi b}{a}\right) \cosh\left(\frac{m\pi}{a}y'\right) \right]$, $\gamma_2 = \frac{a}{m\pi} - V_0 \frac{\cosh\left(\frac{m\pi}{a}y\right)}{\sinh\left(\frac{m\pi b}{a}\right)}$

Solving (5) and (6) for B_m and C_m leads to:

$$C_m = \frac{\alpha_1 \gamma_2 + \alpha_2 \gamma_1}{\beta_2 \alpha_1 - \beta_1 \alpha_2} \quad (7)$$

$$B_m = \frac{\beta_1 \gamma_2 + \beta_2 \gamma_1}{\beta_2 \alpha_1 - \alpha_2 \beta_1} \quad (8)$$

Thus

$$g_m^{(1)} = B_m(x', y') \sinh\left(\frac{m\pi}{a}y\right), \quad (9)$$

$$g_m^{(2)} = C_m(x', y') \cosh\left(\frac{m\pi}{a}y\right) + \left[V_0 - C_m \cosh\left(\frac{m\pi b}{a}\right) \right] \frac{\sinh\left(\frac{m\pi}{a}y\right)}{\sinh\left(\frac{m\pi b}{a}\right)}, \quad y > y' \quad (10)$$

Thus the Green's function

$$G(x, y; x', y') = \sum_{m=1,2,\dots}^{\infty} g_m(y; x', y') \sin\left(\frac{m\pi}{a}x\right)$$

can be written as

$$G(x, y; x', y') = \begin{cases} \sum_{m=1,2,\dots}^{\infty} B_m(x', y') \sinh\left(\frac{m\pi}{a}y\right) \sin\left(\frac{m\pi}{a}x\right) & ; y < y' \\ \sum_{m=1,2,\dots}^{\infty} \left\{ C_m(x', y') \cosh\left(\frac{m\pi}{a}y\right) + \left[V_0 - C_m \cosh\left(\frac{m\pi b}{a}\right) \right] \frac{\sinh\left(\frac{m\pi}{a}y\right)}{\sinh\left(\frac{m\pi b}{a}\right)} \right\} \sin\left(\frac{m\pi}{a}x\right), & y > y' \end{cases}$$

15.8 The Green's function will be chosen to satisfy the partial differential equation of $\frac{\partial^2 G}{\partial x^2} + \frac{\partial^2 G}{\partial y^2} = \delta(x-x')(y-y')$ (1) subject to the boundary conditions of

$$G(x=0, 0 \leq y \leq \infty) = G(x=a, 0 \leq y \leq \infty) = 0 \quad (2a)$$

$$G(0 \leq x \leq a, y=0) = 0; G(0 \leq x \leq a, y=\infty) = \text{Finite} \quad (2b)$$

Choose the Green's function to satisfy the boundary conditions of (2a) at $x=0$ and $x=a$. Thus

$$G(x, y; x', y') = \sum_{m=1,2,\dots}^{\infty} g_m(y; x', y') \sin\left(\frac{m\pi}{a}x\right) \quad (3)$$

Substituting (3) into (1) leads to

$$\sum_{m=1,2,\dots}^{\infty} \left[-\left(\frac{m\pi}{a}\right)^2 g_m(y; x', y') + \frac{d^2 g_m(y; x', y')}{dy^2} \right] \sin\left(\frac{m\pi}{a}x\right) = \delta(x-x')\delta(y-y') \quad (4)$$

Multiplying both sides of (4) by $\sin(m\pi x/a)$, integrating with respect to x from 0 to a , and using (15-48a) and (15-48b), we can write that

$$\frac{d^2 g_m}{dy^2} - \left(\frac{m\pi}{a}\right)^2 g_m = \frac{2}{a} \sin\left(\frac{m\pi}{a}x'\right) \delta(y-y') \quad (5)$$

For the homogeneous form of (5), the two solutions can be written as

$$g_m^{(1)}(y; x', y') = A_m(x', y') e^{-\frac{(m\pi/a)y}{2}} + B_m(x', y') e^{\frac{(m\pi/a)y}{2}}, \quad y < y' \quad (6)$$

$$g_m^{(2)}(y; x', y') = C_m(x', y') e^{-\frac{(m\pi/a)y}{2}} + D_m(x', y') e^{\frac{(m\pi/a)y}{2}}, \quad y > y' \quad (7)$$

Now apply on (6) and (7) the boundary conditions of (2b).

$$g_m^{(1)}(y=0) = A_m + B_m = 0 \Rightarrow B_m = -A_m \quad (8)$$

$$g_m^{(2)}(y=\infty) = \text{finite} \Rightarrow D_m = 0 \quad (9)$$

Thus

$$g_m^{(1)}(y; x, y') = 2A_m(x', y') \left[\frac{e^{-\frac{(m\pi/a)y}{2}} - e^{\frac{(m\pi/a)y}{2}}}{2} \right] = 2A_m(x', y') \sinh\left(\frac{m\pi}{a}y\right) \quad (10)$$

$$g_m^{(2)}(y; x, y') = C_m(x', y') e^{-\frac{(m\pi/a)y}{2}} \quad y > y' \quad (11)$$

Continuity of $g_m^{(1)}$ and $g_m^{(2)}$ at $y=y'$:

$$2A_m(x', y') \sinh\left(\frac{m\pi}{a}y'\right) = C_m(x', y') e^{-\frac{(m\pi/a)y'}{2}} \quad (12)$$

Discontinuity of $dg_m^{(1)}/dy|_{y=y'}$ and $dg_m^{(2)}/dy|_{y=y'}$:

$$\left(\frac{m\pi}{a} \right) \left[-C_m e^{-\frac{(m\pi/a)y'}{2}} - 2A_m \cosh\left(\frac{m\pi}{a}y'\right) \right] = 1 \quad (13)$$

cont'd.

15.8 cont'd. Equations (12) and (13) can be written as

$$C_m = 2 A_m \sinh\left(\frac{m\pi}{a} y'\right) e^{-(m\pi/a)y'} \quad (14)$$

$$C_m e^{-(m\pi/a)y'} + 2 A_m \cosh\left(\frac{m\pi}{a} y'\right) = -\frac{a}{m\pi} \quad (15)$$

Solving these two equations for A_m and C_m leads to

$$A_m = -\frac{2}{m\pi} \frac{1}{2[\cosh\left(\frac{m\pi}{a} y'\right) + \sinh\left(\frac{m\pi}{a} y'\right)]} \quad (16)$$

$$C_m = -\frac{2}{m\pi} \frac{\sinh\left(\frac{m\pi}{a} y'\right) e^{(m\pi/a)y'}}{\cosh\left(\frac{m\pi}{a} y'\right) + \sinh\left(\frac{m\pi}{a} y'\right)} \quad (17)$$

Thus (10) and (11) can be written as

$$g_m^{(1)}(y; x, y') = -\frac{2}{m\pi} \frac{\sinh\left(\frac{m\pi}{a} y\right)}{\cosh\left(\frac{m\pi}{a} y'\right) + \sinh\left(\frac{m\pi}{a} y'\right)}, \quad y < y' \quad (18)$$

$$g_m^{(2)}(y; x, y') = -\frac{2}{m\pi} \frac{\sinh\left(\frac{m\pi}{a} y'\right) e^{(m\pi/a)y'}}{\cosh\left(\frac{m\pi}{a} y'\right) + \sinh\left(\frac{m\pi}{a} y'\right)} e^{-(m\pi/a)y}, \quad y > y' \quad (19)$$

Thus the Green's function of (3) can be expressed as

$$G(x, y; x, y') = \begin{cases} \sum_{n=1,3,\dots}^{\infty} -\left(\frac{2}{m\pi}\right) \frac{1}{\cosh\left(\frac{m\pi}{a} y\right) + \sinh\left(\frac{m\pi}{a} y\right)} \sinh\left(\frac{m\pi}{a} y\right) \sin\left(\frac{m\pi}{a} x\right), & y < y' \\ \sum_{n=1,3,\dots}^{\infty} -\left(\frac{2}{m\pi}\right) \frac{\sinh\left(\frac{m\pi}{a} y'\right) e^{(m\pi/a)y'}}{\cosh\left(\frac{m\pi}{a} y'\right) + \sinh\left(\frac{m\pi}{a} y'\right)} e^{-(m\pi/a)y} \sin\left(\frac{m\pi}{a} x\right), & y > y' \end{cases}$$

$$15.9 \quad G(x, y; x', y') = \sum_{m=1,2,\dots}^{\infty} g_m(x; x', y') \sin\left(\frac{m\pi}{b}y\right) \quad (1)$$

The coefficients $g_m(x; x', y')$ of the Fourier series will be determined by first substituting (1) into (15-72). This leads to

$$\sum_{n=1,2,\dots}^{\infty} \left[-\left(\frac{n\pi}{b}\right)^2 g_m(x; x', y') \sin\left(\frac{n\pi}{b}y\right) + \sin\left(\frac{n\pi}{b}y\right) \frac{d^2 g_m(x; x', y')}{dx'^2} \right] = \delta(x-x')\delta(y-y') \quad (2)$$

Multiplying both sides of (2) by $\sin\left(\frac{m\pi}{b}y\right)$, integrating with respect to y from 0 to b , and using (15-48a) and (15-48b), we can write that

$$\frac{d^2 g_m(x; x', y')}{dx'^2} - \left(\frac{m\pi}{b}\right)^2 g_m(x; x', y') = \frac{2}{a} \sin\left(\frac{m\pi}{b}y'\right) \delta(x-x') \quad (3)$$

Equation (3) is recognized as a one-dimensional differential equation for $g_m(x; x', y')$ which can be solved using the recipe of Section 3.5.3.1 as provided by (15-44c) and (15-45a)-(15-45b).

For the homogeneous form of (3), or

$$\frac{d^2 g_m(x; x', y')}{dx'^2} - \left(\frac{m\pi}{b}\right)^2 g_m(x; x', y') \quad (4)$$

two solutions that satisfy, respectively, the boundary conditions at $x=0$ and $x=a$ are

$$g_m^{(1)} = A_n(x', y') \sinh\left(\frac{m\pi}{b}x\right) \quad \text{for } x \leq x' \quad (5a)$$

$$g_m^{(2)} = B_n(x', y') \sinh\left[\frac{m\pi}{b}(a-x)\right] \quad \text{for } x \geq x' \quad (5b)$$

Using (4-44c) where $y_1 = g_m^{(1)}$ and $y_2 = g_m^{(2)}$, we can write the Wronskian as

$$W(x; x', y') = -\left(\frac{m\pi}{b}\right) A_n B_n \left\{ \sinh\left(\frac{m\pi}{b}x\right) \cosh\left[\frac{m\pi}{b}(a-x)\right] + \cosh\left(\frac{m\pi}{b}x\right) \sinh\left[\frac{m\pi}{b}(a-x)\right] \right\} \\ = -\left(\frac{m\pi}{b}\right) A_n B_n \sinh\left(\frac{m\pi a}{b}\right) \quad (6)$$

By comparing (3) with the form of (15-39), it is apparent that

$$\begin{aligned} p(x) &= 1 \\ q(x) &= 0 \\ r(x) &= -\left(\frac{m\pi}{b}\right)^2 \end{aligned} \quad (7)$$

Using (5a)-(7) the solution for $g_m(x; x', y')$ of (3) can be written by referring to (15-45a) and (15-45b) as

$$-\frac{2}{n\pi} \sin\left(\frac{n\pi}{b}y\right) \frac{\sinh\left[\frac{m\pi}{b}(a-x')\right]}{\sinh\left(\frac{m\pi a}{b}\right)} \sinh\left(\frac{n\pi}{b}x\right), \quad 0 \leq x \leq x' \quad (8a)$$

$$g_m(x; x', y') = \begin{cases} -\frac{2}{n\pi} \sin\left(\frac{n\pi}{b}y\right) \frac{\sinh\left(\frac{m\pi}{b}x'\right)}{\sinh\left(\frac{m\pi a}{b}\right)} \sinh\left[\frac{n\pi}{b}(a-x)\right], & x' \leq x \leq a \\ \end{cases} \quad (8b)$$

Using (1), (8a) and (8b) we can then write the Green's function of (15-83a)+(15-83b).

$$15.10 \quad \nabla^2 V = f(r, \phi) = -\frac{1}{\epsilon_0} \frac{\partial^2 V}{\partial r^2} \Rightarrow \frac{1}{r} \frac{\partial}{\partial r} \left(r \frac{\partial V}{\partial r} \right) + \frac{1}{r^2} \frac{\partial^2 V}{\partial \phi^2} = f(r, \phi) = -\frac{1}{\epsilon_0} q(r, \phi) \quad (1)$$

$$\nabla^2 G = \delta(r-r') \Rightarrow \frac{1}{r} \frac{\partial}{\partial r} \left(r \frac{\partial G}{\partial r} \right) + \frac{1}{r^2} \frac{\partial^2 G}{\partial \phi^2} = \delta(r-r') = \frac{1}{r} \delta(r-r') \delta(\phi-\phi') \quad (2)$$

$G(r=a, 0 \leq \phi \leq 2\pi) = 0$, G = finite everywhere within the cylinder
Choose a solution of the form

$$G(r, \phi; r', \phi') = \sum_{m=-\infty}^{\infty} g_m(r, r') e^{im(\phi-\phi')} = \sum_{m=1, \dots}^{\infty} \epsilon_m g_m(r, r') \cos[m(\phi-\phi')], \epsilon_m = \begin{cases} 1, & m=0 \\ 2, & m \neq 0 \end{cases} \quad (3)$$

Substituting (3) into (2) leads to

$$\sum_{m=-\infty}^{+\infty} \left[\frac{1}{r} \frac{d}{dr} \left(r \frac{dg_m}{dr} \right) - \frac{1}{r^2} m^2 g_m \right] e^{im(\phi-\phi')} = \frac{1}{r} \delta(r-r') \delta(\phi-\phi') \quad (4)$$

Multiplying both sides of (4) by $e^{-im(\phi-\phi')}$, integrating from 0 to 2π , and using (15-155a) and (15-155b), we can write that

$$\text{or } 2\pi \left[\frac{1}{r} \frac{d}{dr} \left(r \frac{dg_m}{dr} \right) - \frac{1}{r^2} m^2 g_m \right] = \frac{1}{r} \delta(r-r') \quad (5)$$

$$r \frac{d^2 g_m}{dr^2} + \frac{dg_m}{dr} - \frac{m^2}{r^2} g_m = \frac{1}{2\pi} \delta(r-r') \quad (5a)$$

The two solutions for the homogeneous form of (5a) are

$$g_m^{(1)} = A_m r^m + B_m \frac{1}{r^m} \quad \text{for } r < r' \quad (6a)$$

$$g_m^{(2)} = C_m r^m + D_m \frac{1}{r^m} \quad \text{for } r > r' \quad (6b)$$

Since the Green's function must be finite everywhere, including $r=0$, the $B_m = 0$. Thus

$$g_m^{(1)} = A_m r^m \quad \text{for } r < r' \quad (7a)$$

$$g_m^{(2)} = C_m r^m + D_m \frac{1}{r^m} \quad \text{for } r > r' \quad (7b)$$

Also $g_m^{(2)}(r=a)=0=C_m a^m + D_m a^{-m}=0 \Rightarrow D_m = -C_m a^{2m}$. Thus

$$g_m^{(1)} = A_m r^m \Rightarrow g_m^{(1)} = A_m m r^{m-1} \quad (8a)$$

$$g_m^{(2)} = C_m \left[r^m - a^{2m} r^{-m} \right] \Rightarrow g_m^{(2)} = C_m \left[m r^{m-1} + m a^{2m} r^{-(m+1)} \right] \quad (8b)$$

Using (4-44c), we can write the Wronskian to

$$W(r') = A_m C_m \left\{ (r')^m [m(r')^{m-1} + m a^{2m} (r')^{-(m+1)}] - [(r')^m - a^{2m} (r')^m] m (r')^{m-1} \right\} = A_m C_m 2ma^{2m} \frac{1}{r'} \quad (9)$$

cont'd.

15.10 cont'd.

Therefore using (5a), (15-45a) and (15-45b), we can write $g_m(p, p')$ with $p(p') = p'$ as

$$g_m(p, p') = \begin{cases} \frac{1}{4\pi m a^{2m}} \left[(p')^m - a^{2m} (p')^{-m} \right] p^m & , p < p' \\ \frac{1}{4\pi m a^{2m}} (p')^m \left[p^m - a^{2m} p^{-m} \right] & , p > p' \end{cases} \quad (10a)$$

With (10a) and (10b) we can express the Green's function of (3) to

$$G(p, \phi; p', \phi') = \begin{cases} \frac{1}{4\pi} \sum_{m=-\infty}^{+\infty} \frac{1}{m a^{2m}} \left[(p')^m - \frac{a^{2m}}{(p')^m} \right] p^m e^{im(\phi-\phi')} & , p < p' \\ \frac{1}{4\pi} \sum_{m=-\infty}^{+\infty} \frac{1}{m a^{2m}} (p')^m \left[p^m - \frac{a^{2m}}{p^m} \right] e^{im(\phi-\phi')} & , p > p' \end{cases} \quad (11a)$$

$$15.11 \quad \frac{\partial^2 G}{\partial x^2} + \frac{\partial^2 G}{\partial y^2} + \beta^2 G = \delta(x-x') \delta(y-y') \quad (1)$$

$$G(x=0, 0 \leq y \leq b) = G(x=a, 0 \leq y \leq b) = 0 \quad (1a)$$

$$G(0 \leq x \leq a, y=0) = G(0 \leq x \leq a, y=b) = 0 \quad (1b)$$

$$G(x, y; x', y') = \sum_m g_m(y; x, y') [A_m \sin(\beta_x x) + B_m \cos(\beta_x x)] \quad (2)$$

Applying (1a) on (2) leads to

$$G(x=a) = \sum_m g_m(y; x, y) [A_m(0) + B_m(1)] = 0 \Rightarrow B_m = 0$$

$$G(x=a) = \sum_m g_m(y; x, y') [\bar{A}_m \sin(\beta_x a)] = 0 \Rightarrow \beta_x a = \sin^{-1}(0) = m\pi \Rightarrow \beta_x = \left(\frac{m\pi}{a}\right), m=1, 2, \dots$$

Thus

$$G(x, y; x', y') = \sum_{m=1, 2, \dots}^{\infty} A_m g_m(y; x, y') \sin\left(\frac{m\pi}{a} x\right) \quad (3)$$

Substituting (3) into (1), we can write that

$$\sum_{m=1, 2, \dots}^{\infty} A_m \left\{ -\left(\frac{m\pi}{a}\right)^2 g_m + \frac{d^2 g_m}{dy^2} + \beta^2 g_m \right\} \sin\left(\frac{m\pi}{a} x\right) = \delta(x-x') \delta(y-y')$$

$$\sum_{m=1, 2, \dots}^{\infty} A_m \left\{ \frac{d^2 g_m}{dy^2} + \left[\beta^2 - \left(\frac{m\pi}{a}\right)^2\right] g_m \right\} \sin\left(\frac{m\pi}{a} x\right) = \delta(x-x') \delta(y-y') \quad (4)$$

Multiplying both sides by $\sin(n\pi x/a)$, integrating between 0 to a in x, and using (15-48a) and (15-48b), leads to

cont'd.

$$15.11 \text{ cont'd.} \quad \frac{\partial}{\partial y^2} \left\{ \frac{d^2 g_m}{dy^2} + \left[\beta_y^2 - \left(\frac{m\pi}{a} \right)^2 \right] g_m \right\} = \sin\left(\frac{m\pi}{a}x'\right) \delta(y-y')$$

$$\text{or } \frac{d^2 g_m}{dy^2} + \left[\beta_y^2 - \left(\frac{m\pi}{a} \right)^2 \right] g_m = \frac{d^2 g_m}{dy^2} + \beta_y^2 g_m = \frac{2}{a} \sin\left(\frac{m\pi}{a}x'\right) \delta(y-y), \text{ where } \beta_y^2 = \beta_y^2 - \left(\frac{m\pi}{a} \right)^2 \quad (5)$$

The two solutions for the homogeneous form of (5) are

$$g_m^{(1)} = A_1 \cos(\beta_y y) + B_1 \sin(\beta_y y) \quad , \quad y < y' \quad (6a)$$

$$g_m^{(2)} = C_1 \cos[\beta_y(y-b)] + D_1 \sin[\beta_y(y-b)] \quad , \quad y > y' \quad (6b)$$

Applying the boundary conditions of (1b) on (6a) and (6b) leads to

$$g_m^{(1)}(y=0) = A_1(1) + B_1(0) = 0 \Rightarrow A_1 = 0$$

$$g_m^{(2)}(y=b) = C_1(1) + D_1(0) = 0 \Rightarrow C_1 = 0$$

Thus (6a) and (6b) reduce to

$$g_m^{(1)} = B_1 \sin(\beta_y y) \quad , \quad g_m^{(1)'} = \beta_y B_1 \cos(\beta_y y) \quad , \quad y < y' \quad (7a)$$

$$g_m^{(2)} = D_1 \sin[\beta_y(y-b)] \quad , \quad g_m^{(2)'} = \beta_y D_1 \cos[\beta_y(y-b)] \quad , \quad y > y' \quad (7b)$$

The Wronskian of (15-44c) can now be written as

$$W(y') = \beta_y B_1 D_1 \left[\sin(\beta_y y') \cos[\beta_y(y'-b)] - \sin[\beta_y(y'-b)] \cos(\beta_y y') \right]$$

$$W(y') = \beta_y B_1 D_1 \left[\sin(\beta_y y' - \beta_y b + \beta_y b) \right] = \beta_y B_1 D_1 \sin(\beta_y b) \quad (8)$$

Using (15-45a), (15-45b) and (5) with $\rho(y') = 1$, we can write

$$g_m(y; x', y') = \begin{cases} \frac{2}{a \beta_y} \frac{\sin[\beta_y(y'-b)]}{\sin(\beta_y b)} \sin(\beta_y y) \sin\left(\frac{m\pi}{a}x'\right) & , \quad y < y' \\ \frac{2}{a \beta_y} \frac{\sin(\beta_y y')}{\sin(\beta_y b)} \sin[\beta_y(y-b)] \sin\left(\frac{m\pi}{a}x'\right) & , \quad y > y' \end{cases} \quad (9)$$

Thus the Green's function of (3) can now be expressed as

$$\sum_{m=1,2,\dots}^{\infty} \frac{2}{a \beta_y} \frac{\sin[\beta_y(y'-b)]}{\sin(\beta_y b)} \sin(\beta_y y) \sin\left(\frac{m\pi}{a}x'\right) \sin\left(\frac{m\pi}{a}x\right), \quad y < y' \quad (10a)$$

$$G(x, y; x', y') = \begin{cases} \sum_{m=1,2,\dots}^{\infty} \frac{2}{a \beta_y} \frac{\sin(\beta_y y')}{\sin(\beta_y b)} \sin[\beta_y(y-b)] \sin\left(\frac{m\pi}{a}x'\right) \sin\left(\frac{m\pi}{a}x\right) & , \quad y > y' \end{cases} \quad (10b)$$

15.12 Since the closed form recipe of (15-45a) and (15-45b) is only one-dimensional, and the subject problem is three-dimensional, the solution for Figure 15-7 can proceed in many different ways, including the one outlined in Section 15.6.1 which starts with a two-dimensional Fourier series of sine function in x and cosine function in y to represent the Green's function $G(x,y,z; x',y',z')$. The solution basically concludes by using the closed form recipe to represent the z -variations of $g_{mn}(z; x',y',z')$, which is represented by the differential equation of (15-143) with solutions of closed form as given by (15-146a) for $z < z'$ and (15-146b) for $z > z'$. The solution concludes with the final form of the Green's function as given by (15-147).

An alternate solution would be to represent the y -variations in (15-140) by a sine function $[\sin(\frac{n\pi}{b}y)]$ since we have no boundary conditions in the y -direction as represented by (15-137a).

15.13 The Green's function when the excitation in Example 15-5 is a magnetic line source is the same as that of the electric line source, as given by, in Example 15-5,

$$G(p,\phi; p',\phi') = -\frac{1}{4\pi} H_0^{(2)}(\beta |p-p'|)$$

However, for the electric line source the Green's function is related to the electric field E_z by

$$E_z = j\omega \mu I_e G = -\frac{\omega \mu}{4} I_e H_0^{(2)}(\beta |p-p'|)$$

while it is related to the magnetic field H_z for the magnetic line source by

$$H_z = j\omega \epsilon I_m G = -\frac{\omega \epsilon}{4} I_m H_0^{(2)}(\beta |p-p'|)$$

15.14 The Green's function for this problem must satisfy the partial differential equation that $H\Phi$ must satisfy; that is

$$\nabla^2 G + \beta_0^2 G = \delta(\rho - \rho') \quad (1)$$

The boundary conditions for the Green's function are chosen to be the same as those for $H\Phi$; that is

$$G(\rho=a, 0 \leq \phi \leq 2\pi, 0 \leq z \leq h) = G(\rho=b, 0 \leq \phi \leq 2\pi, 0 \leq z \leq h) = 0 \quad (2)$$

Let

$$G(\rho, \phi; \rho', \phi') = \sum_{m=-\infty}^{+\infty} g_m(\rho; \rho', \phi) e^{im\phi} \quad (3)$$

Choosing (15-152d) to represent $\delta(\rho - \rho')$, we can write (1) in expanded form, assuming no z variations, as (15-153).

Following the procedure outlined on page 929-930 we can write the two solutions for $g_m(\rho; \rho', \phi)$ as given by (15-159a) and (15-159b), or

$$g_m^{(1)} = A_m J_m(\beta_0 \rho) + B_m Y_m(\beta_0 \rho) \quad \text{for } \rho < \rho' \quad (3a)$$

$$g_m^{(2)} = C_m J_m(\beta_0 \rho) + D_m Y_m(\beta_0 \rho) \quad \text{for } \rho > \rho' \quad (3b)$$

The Green's function of (3) must satisfy the boundary conditions of (2). Thus

$$g_m^{(1)}(\rho=a) = A_m J_m(\beta_0 a) + B_m Y_m(\beta_0 a) = 0 \Rightarrow B_m = -A_m \frac{J_m(\beta_0 a)}{Y_m(\beta_0 a)} \quad (4a)$$

$$g_m^{(2)}(\rho=b) = C_m J_m(\beta_0 b) + D_m Y_m(\beta_0 b) = 0 \Rightarrow D_m = -C_m \frac{J_m(\beta_0 b)}{Y_m(\beta_0 b)} \quad (4b)$$

Thus (3a) and (3b) reduce to

$$g_m^{(1)} = A_m \left[J_m(\beta_0 \rho) - \frac{J_m(\beta_0 a)}{Y_m(\beta_0 a)} Y_m(\beta_0 \rho) \right] = A_m \left[J_m(\beta_0 \rho) - \alpha Y_m(\beta_0 \rho) \right] \quad (5a)$$

$$g_m^{(2)} = C_m \left[J_m(\beta_0 \rho) - \frac{J_m(\beta_0 b)}{Y_m(\beta_0 b)} Y_m(\beta_0 \rho) \right] = C_m \left[J_m(\beta_0 \rho) - \gamma Y_m(\beta_0 \rho) \right] \quad (5b)$$

Using (15-44c) where $y_1 = g_m^{(1)}$ and $y_2 = g_m^{(2)}$, we can write the Wronskian as

$$W(\rho') = \beta_0 A_m \left[J_m(\beta_0 \rho') - \alpha Y_m(\beta_0 \rho') \right] C_m \left[J_m'(\beta_0 \rho') - \gamma Y_m'(\beta_0 \rho') \right] \\ - \beta_0 C_m \left[J_m(\beta_0 \rho') - \gamma Y_m(\beta_0 \rho') \right] A_m \left[J_m'(\beta_0 \rho') - \alpha Y_m'(\beta_0 \rho') \right]$$

$$W(\rho') = \beta_0 A_m C_m \left[J_m(\beta_0 \rho') J_m'(\beta_0 \rho') + \alpha \gamma Y_m(\beta_0 \rho') Y_m'(\beta_0 \rho') - \alpha Y_m(\beta_0 \rho') J_m'(\beta_0 \rho') - \gamma J_m(\beta_0 \rho') Y_m'(\beta_0 \rho') \right] \\ - \beta_0 C_m A_m \left[J_m(\beta_0 \rho') J_m'(\beta_0 \rho') + \alpha \gamma Y_m(\beta_0 \rho') Y_m'(\beta_0 \rho') - \gamma Y_m(\beta_0 \rho') J_m'(\beta_0 \rho') - \alpha J_m(\beta_0 \rho') Y_m'(\beta_0 \rho') \right]$$

cont'd.

15.14 Contd.

$$W(p') = \beta_0 A_m C_m \left\{ -\alpha [J_m^1 Y_m - J_m Y_m^1] + \gamma [J_m^1 Y_m - J_m Y_m^1] \right\}, \quad 1 = \frac{2}{\pi \beta_0 p} \quad (6)$$

By using the Wronskian for Bessel functions of (11-95), (6) reduces to

$$W(p') = + \frac{2}{\pi} A_m C_m \frac{1}{p'} \{ + \alpha - \gamma \} = \frac{2}{\pi p'} A_m C_m \left[\frac{J_m(\beta_0 a)}{Y_m(\beta_0 a)} - \frac{J_m(\beta_0 b)}{Y_m(\beta_0 b)} \right]$$

$$W(p') = \frac{2}{\pi} \frac{1}{p'} A_m C_m \frac{J_m(\beta_0 a) Y_m(\beta_0 b) - J_m(\beta_0 b) Y_m(\beta_0 a)}{Y_m(\beta_0 a) Y_m(\beta_0 b)} = \frac{2}{\pi} \frac{A_m C_m (JY)_m}{p'} \quad (7)$$

$$\text{where } (JY)_m = \frac{J_m(\beta_0 a) Y_m(\beta_0 b) - J_m(\beta_0 b) Y_m(\beta_0 a)}{Y_m(\beta_0 a) Y_m(\beta_0 b)} \quad (7a)$$

Finally $g_m(p; p', \phi)$ of (15-156) can be written using (15-158), (5a)-(7a) by referring to (15-45a) and (15-45b), as

$$g_m(p; p', \phi) = \begin{cases} \frac{1}{4} \left[\frac{J_m(\beta_0 p) - \gamma Y_m(\beta_0 p)}{(JY)_m} \right] [J_m(\beta_0 p) - \alpha Y_m(\beta_0 p)] e^{j m \phi}, & a \leq p \leq p' \\ \frac{1}{4} \left[\frac{J_m(\beta_0 p') - \alpha Y_m(\beta_0 p')}{(JY)_m} \right] [J_m(\beta_0 p') - \gamma Y_m(\beta_0 p')] e^{-j m \phi}, & p' \leq p \leq b \end{cases} \quad (8a)$$

Thus the Green's function of (3) can be written as

$$G(p, \phi; p', \phi') = \frac{1}{4} \sum_{m=-\infty}^{+\infty} \left[\frac{J_m(\beta_0 p') - \gamma Y_m(\beta_0 p')}{(JY)_m} \right] \left[\frac{J_m(\beta_0 p) - \alpha Y_m(\beta_0 p)}{J_m(\beta_0 p) - \alpha Y_m(\beta_0 p)} \right] e^{j m (\phi - \phi')} \quad (9a)$$

$$G(p, \phi; p', \phi') = \frac{1}{4} \sum_{m=-\infty}^{+\infty} \left[\frac{J_m(\beta_0 p') - \alpha Y_m(\beta_0 p')}{(JY)_m} \right] \left[\frac{J_m(\beta_0 p) - \gamma Y_m(\beta_0 p)}{J_m(\beta_0 p) - \gamma Y_m(\beta_0 p)} \right] e^{j m (\phi - \phi')} \quad (9b)$$

15.15 The solution to this problem follows that of Problem 15.12. One of differences are the boundary conditions which for this problem are.

$$G(p=a, 0 \leq \phi \leq \phi_0, 0 \leq z \leq h) = G(p=b, 0 \leq \phi \leq \phi_0, 0 \leq z \leq h) = 0 \quad (1a)$$

$$G(0 \leq p \leq b, \phi=0, 0 \leq z \leq h) = G(a \leq p \leq b, \phi=\phi_0, 0 \leq z \leq h) = 0 \quad (1b)$$

To meet the boundary conditions on ϕ , we select a solution for the Green's function of the form

$$G(p, \phi; p', \phi') = \sum_{m=1,2,\dots}^{\infty} g_m(p, p', \phi) \sin\left(\frac{m\pi}{\phi_0}\phi\right) = \sum_{m=1,2,\dots}^{\infty} g_m(p, p', \phi') \sin\left(\frac{m\pi}{\phi_0}\phi'\right) \quad (2)$$

Substituting (2) into (15-15a) leads to

$$\sum_{m=1,2,\dots}^{\infty} \left[\frac{d^2}{dp^2} + \frac{1}{p} \frac{d}{dp} - \frac{r^2}{p^2} + \beta_0^2 \right] g_m(p, p', \phi') \sin\left(\frac{m\pi}{\phi_0}\phi'\right) = \frac{1}{p} \delta(p-p') \delta(\phi-\phi') \quad (3)$$

Multiplying both sides of (3) by $\sin(s\phi)$, integrating from 0 to ϕ_0 , and using the orthogonality conditions of (15-48a) and (15-48b) leads to

$$p \frac{d^2 g_m}{dp^2} + \frac{d g_m}{dp} + \left(p \beta_0^2 - \frac{r^2}{p} \right) g_m = \frac{2}{\phi_0} \sin(r\phi) \delta(p-p') \quad (4)$$

The homogeneous form of (4) is

$$p \frac{d^2 g_m}{dp^2} + \frac{d g_m}{dp} + \left(p \beta_0^2 - \frac{r^2}{p} \right) g_m = 0 \quad (5)$$

Thus its two solutions are

$$g_m^{(1)} = A_m J_m(\beta_0 p) + B_m Y_m(\beta_0 p) \quad \text{for } p < p' \quad (6a)$$

$$g_m^{(2)} = C_m J_m(\beta_0 p) + D_m Y_m(\beta_0 p) \quad \text{for } p > p' \quad (6b)$$

Since the boundary conditions in p for this problem are the same as those of Problem 15.12, thus we can write using (6a)-(6b), (4) and (6a)-(6b) the solution for g_m as

$$g_m(p, p', \phi) = \begin{cases} \frac{\pi}{\phi_0} \left[\frac{J_r(\beta_0 p') - \gamma Y_r(\beta_0 p')}{(JY)_r} \right] \left[J_r(\beta_0 p) - \alpha Y_r(\beta_0 p) \right] \sin(r\phi) & a \leq p \leq p' \\ \frac{\pi}{\phi_0} \left[\frac{J_r(\beta_0 p) - \alpha Y_r(\beta_0 p)}{(JY)_r} \right] \left[J_r(\beta_0 p') - \gamma Y_r(\beta_0 p') \right] \sin(r\phi) & p' \leq p \leq b \end{cases} \quad (7a)$$

$$g_m(p, p', \phi) = \begin{cases} \frac{\pi}{\phi_0} \left[\frac{J_r(\beta_0 p') - \gamma Y_r(\beta_0 p')}{(JY)_r} \right] \left[J_r(\beta_0 p) - \alpha Y_r(\beta_0 p) \right] \sin(r\phi) & a \leq p \leq p' \\ \frac{\pi}{\phi_0} \left[\frac{J_r(\beta_0 p) - \alpha Y_r(\beta_0 p)}{(JY)_r} \right] \left[J_r(\beta_0 p') - \gamma Y_r(\beta_0 p') \right] \sin(r\phi) & p' \leq p \leq b \end{cases} \quad (7b)$$

cont'd.

15.15 cont'd. where $\alpha = \frac{J_r(\beta_0 a)}{Y_r(\beta_0 a)}$ (sa)

$$\gamma = \frac{J_r(\beta_0 b)}{Y_r(\beta_0 b)} \quad (sb)$$

$$(JY)_r = \frac{J_r(\beta_0 a) Y_r(\beta_0 b) - J_r(\beta_0 b) Y_r(\beta_0 a)}{Y_r(\beta_0 a) Y_r(\beta_0 b)} \quad (sc)$$

Thus the Green's function of (2) can be written as

$$G(p, \phi; p', \phi') = \left\{ \begin{array}{l} \frac{\pi}{\Phi_0} \sum_{m=1,2,\dots}^{\infty} \left\{ \frac{J_{m\pi/\beta_0}(p, \phi) - \gamma Y_{m\pi/\beta_0}(p, \phi')}{(JY)_r} \right\} \left[J_{m\pi/\beta_0}(p', \phi) - \alpha Y_{m\pi/\beta_0}(p', \phi) \right] \sin\left(\frac{m\pi}{\beta_0} \phi\right) \sin\left(\frac{m\pi}{\beta_0} \phi'\right) \\ \qquad \qquad \qquad a \leq p \leq p' \end{array} \right. \quad (qa)$$

$$G(p, \phi; p', \phi') = \left\{ \begin{array}{l} \frac{\pi}{\Phi_0} \sum_{m=1,2,\dots}^{\infty} \left\{ \frac{J_{m\pi/\beta_0}(p, \phi) - \alpha Y_{m\pi/\beta_0}(p, \phi')}{(JY)_r} \right\} \left[J_{m\pi/\beta_0}(p', \phi) - \gamma Y_{m\pi/\beta_0}(p', \phi) \right] \sin\left(\frac{m\pi}{\beta_0} \phi\right) \sin\left(\frac{m\pi}{\beta_0} \phi'\right) \\ \qquad \qquad \qquad p' \leq p \leq b \end{array} \right. \quad (qb)$$

15.16 This problem is a special case of that of Problem 15.13 except that the boundary condition in p is

$$G(p=b, 0 \leq \phi \leq \phi_0, 0 \leq z \leq h) = 0 \quad (1)$$

Thus we can copy the solution from Problem 15.13 by letting $\alpha = \frac{J_r(\beta_0 a)}{Y_r(\beta_0 a)} = 0$.

Thus the Green's function can be written from (qa) and (qb) of Problem 9.16 as

$$G(p, \phi; p', \phi') = \left\{ \begin{array}{l} \frac{\pi}{\Phi_0} \sum_{m=1,2,\dots}^{\infty} \left\{ \frac{J_{m\pi/\beta_0}(p, \phi) - \gamma Y_{m\pi/\beta_0}(p, \phi')}{(JY)_r} \right\} J_{m\pi/\beta_0}(p', \phi) \sin\left(\frac{m\pi}{\beta_0} \phi\right) \sin\left(\frac{m\pi}{\beta_0} \phi'\right) \\ \qquad \qquad \qquad a \leq p \leq p' \end{array} \right. \quad (2a)$$

$$G(p, \phi; p', \phi') = \left\{ \begin{array}{l} \frac{\pi}{\Phi_0} \sum_{m=1,2,\dots}^{\infty} \left\{ \frac{J_{m\pi/\beta_0}(p, \phi)}{(JY)_r} \right\} \left[J_{m\pi/\beta_0}(p', \phi) - \gamma Y_{m\pi/\beta_0}(p', \phi) \right] \sin\left(\frac{m\pi}{\beta_0} \phi\right) \sin\left(\frac{m\pi}{\beta_0} \phi'\right) \\ \qquad \qquad \qquad p' \leq p \leq b \end{array} \right. \quad (2b)$$

where $\gamma = \frac{J_{m\pi/\beta_0}(p, \phi)}{Y_{m\pi/\beta_0}(p, \phi)}$ (3a)

$$(JY)_r = - \frac{J_{m\pi/\beta_0}(p, \phi)}{Y_{m\pi/\beta_0}(p, \phi)} \quad (3b)$$

15.17 The solution to this problem proceeds in the same way until we reach (15-156), or

$$G(g, \phi; g', \phi') = \sum_{m=-\infty}^{+\infty} g_m(g, g', \phi) e^{jm\phi} \quad (1)$$

$$\rho \frac{d^2 g_m}{d\rho^2} + \frac{dg_m}{d\rho} + \left(\rho \beta_0^2 - \frac{m^2}{\rho} \right) g_m = \frac{e^{-jm\phi'}}{2\pi} \delta(\rho - \rho') \quad (2)$$

Now instead of using a closed form for the solution of $g_m(g, g', \phi)$ of (2), we will select a series form. We write the desired series solutions as

$$g_m(g, g', \phi) = \sum_{n=1,2,\dots}^{\infty} [A_{mn} J_m(\beta_{mn}\rho) + B_{mn} Y_m(\beta_{mn}\rho)] \quad (3)$$

Since the Green's function must be finite everywhere, including $\rho=0$, then $g_m(g, g', \phi)$ must also be finite everywhere (including $\rho=0$). Thus $B_{mn}=0$. Therefore

$$g_m(g, g', \phi) = \sum_{n=1,2,\dots}^{\infty} A_{mn} J_m(\beta_{mn}\rho) \quad \text{of the form of (15-56)} \quad (4)$$

Now we apply the boundary condition of (15-149a) which leads to

$$g_m(g=a; g', \phi) = \sum_{n=1,2,\dots}^{\infty} A_{mn}(g', \phi) J_m(\beta_{mn}a) = 0 \Rightarrow \beta_{mn} = \alpha_{mn} = \text{roots of Bessel function } J_m \quad (5)$$

thus

$$g_m(g, g', \phi) = \sum_{n=1,2,\dots}^{\infty} A_{mn}(g', \phi) J_m(\beta_{mn}\rho) , \quad \beta_{mn} = \frac{\alpha_{mn}}{a} \quad (6)$$

The solution for $g_m(g, g', \phi)$ as given by (3) or (4) was chosen to satisfy the homogeneous differential equation of

$$\rho \frac{d^2 J_m}{d\rho^2} + \frac{dJ_m}{d\rho} + \left(\rho \beta_{mn}^2 - \frac{m^2}{\rho} \right) J_m = 0 \quad \text{where } \psi_m(\rho) \text{ in (15-56) is } \psi_m(\rho) = J_m(\beta_{mn}\rho) \quad (7)$$

Substituting (6) in (2) we can write that of the form of (14-54)

$$\sum_{n=1,2,\dots}^{\infty} \left\{ \rho \frac{d^2 J_m}{d\rho^2} + \frac{dJ_m}{d\rho} + \left(\rho \beta_{mn}^2 - \frac{m^2}{\rho} \right) J_m + \left(\rho \beta_0^2 - \rho \beta_{mn}^2 \right) J_m \right\} A_{mn} = \frac{e^{-jm\phi'}}{2\pi} \delta(\rho - \rho') \quad (8)$$

Substituting (7) in (8) leads to

$$\sum_{n=1,2,\dots}^{\infty} A_{mn} (\beta_0^2 - \beta_{mn}^2) \rho J_m(\beta_{mn}\rho) = \frac{e^{-jm\phi'}}{2\pi} \delta(\rho - \rho') \quad (9)$$

Since the eigenfunction $J_m(\beta_{mn}\rho)$ are orthogonal over the interval 0 to a for the same value of m but different value of n (with the cont'd.

15.17 cont'd. weighting function ρ) as [9], [13]

$$\int_0^a \rho J_m(\beta_{mn}\rho) J_m(\beta_{mp}\rho) d\rho = \delta_{np} N \text{ where } \delta_{np} = \begin{cases} 1 & p=n \\ 0 & p \neq n \end{cases} \quad (10)$$

then we multiply both sides of (9) by $J_m(\beta_{mp}\rho)$ and integrate from 0 to a. Doing this and using that

$$\int_0^a \rho J_m(\beta_{mn}\rho) J_m(\beta_{mp}\rho) d\rho = \begin{cases} \frac{a^2}{2} J_{m+1}^2(\beta_{mn}a) & p=n \\ 0 & p \neq n \end{cases} \quad (11a)$$

we can write that

$$A_{mn}(\beta_0^2 - \beta_{mn}^2) \frac{a^2}{2} J_{m+1}^2(\beta_{mn}a) = \frac{e^{-j m \phi}}{2\pi} J_m(\beta_{mn}\rho') \quad (12)$$

or

$$A_{mn} = \frac{1}{\pi a^2 (\beta_0^2 - \beta_{mn}^2) J_{m+1}^2(\beta_{mn}a)} e^{-j m \phi} J_m(\beta_{mn}\rho') \quad (12a)$$

Therefore $g_m(g; g'; \phi)$ of (6) can be written as

$$g_m(g; g'; \phi) = \sum_{n=1}^{\infty} \frac{1}{\pi a^2 (\beta_0^2 - \beta_{mn}^2) J_{m+1}^2(\beta_{mn}a)} e^{-j m \phi} J_m(\beta_{mn}g') J_m(\beta_{mn}g) \quad (13)$$

Ultimately the Green's function $(g, \phi; g', \phi')$ of (15-151) can be written as

$$G(g, \phi; g', \phi') = \frac{1}{\pi a^2} \sum_{n=-\infty}^{+\infty} \left[\sum_{m=1}^{\infty} \frac{1}{(\beta_0^2 - \beta_{mn}^2) J_{m+1}^2(\beta_{mn}a)} J_m(\beta_{mn}g') J_m(\beta_{mn}g) \right] e^{j m(\phi - \phi')} \quad (14)$$

[13] D.C. Stinson, Intermediate Mathematics of Electromagnetics, Prentice-Hall, 1976, pp. 13, 46-47, 262.

[9] R.Collin, Field Theory of Guided Waves, McGraw-Hill Book Co., p. 196.

15.18

The boundary conditions of this problem are

$$E_z(\rho=a, 0 \leq \phi \leq 2\pi) = 0 \quad (1)$$

The Green's function will be selected to have the same boundary conditions as those of the electric field, or

$$G(\rho=a, 0 \leq \phi \leq 2\pi) = 0 \quad (2)$$

The solution of this problem follows that on pages 928-930. Thus we can write using (15-151), and (15-159a)-(15-159b) that

$$G(\rho_0, \phi; \rho_1, \phi') = \sum_{m=-\infty}^{+\infty} g_m(\rho_0, \rho_1, \phi') e^{jm\phi} \quad (3)$$

$$g_m^{(1)} = A_m J_m(\beta_0 \rho) + B_m H_m^{(2)}(\beta_0 \rho) \quad (3a)$$

$$g_m^{(2)} = C_m H_m^{(2)}(\beta_0 \rho) \quad (3b)$$

Applying the boundary condition of (2) leads to

$$g_m^{(1)}(\rho=a) = 0 = A_m J_m(\beta_0 a) + B_m H_m^{(2)}(\beta_0 a) \Rightarrow B_m = -A_m \frac{J_m(\beta_0 a)}{H_m^{(2)}(\beta_0 a)} \quad (4)$$

Thus

$$g_m^{(1)} = A_m \left\{ J_m(\beta_0 \rho) - \frac{J_m(\beta_0 a)}{H_m^{(2)}(\beta_0 a)} H_m^{(2)}(\beta_0 \rho) \right\} = A_m \left\{ J_m(\beta_0 \rho) - \alpha H_m^{(2)}(\beta_0 \rho) \right\} \quad (5a)$$

$$g_m^{(2)} = C_m H_m^{(2)}(\beta_0 \rho) \quad \alpha = \frac{J_m(\beta_0 a)}{H_m^{(2)}(\beta_0 a)} \quad (5b)$$

The Wronskian of (15-44c) can be written as

$$\begin{aligned} W(\rho') &= \beta_0 A_m C_m \left[J_m(\beta_0 \rho) - \alpha H_m^{(2)}(\beta_0 \rho) \right] H_m^{(2)}(\beta_0 \rho') - \beta_0 A_m C_m H_m^{(2)}(\beta_0 \rho) \left[J_m'(\beta_0 \rho') - \alpha H_m^{(2)}(\beta_0 \rho') \right] \\ &= \beta_0 A_m C_m \left\{ J_m(\beta_0 \rho') H_m^{(2)}(\beta_0 \rho') - J_m'(\beta_0 \rho) H_m^{(2)}(\beta_0 \rho') \right\} \end{aligned}$$

$$W(\rho') = -j \beta_0 A_m C_m \left[J_m(\beta_0 \rho') Y_m'(\beta_0 \rho') - J_m'(\beta_0 \rho) Y_m(\beta_0 \rho') \right], \quad ' = \frac{\partial}{\partial(\beta_0 \rho')} \quad (6)$$

Using the Wronskian for Bessel functions of (11-95), we can write (6) as

$$W(\rho') = -j \frac{2}{\pi} A_m C_m \frac{1}{\rho'} \quad (7)$$

Thus we can write $g_m(\rho; \rho', \phi)$, using (7), (15-156), (15-45a) and (15-45b), as

$$g_m(\rho; \rho', \phi) = \left\{ j \frac{1}{4} \left[H_m^{(2)}(\beta_0 \rho') \right] \left[J_m(\beta_0 \rho) - \alpha H_m^{(2)}(\beta_0 \rho) \right] e^{jm\phi}, \alpha \leq \rho \leq \rho' \right\} \quad (8a)$$

$$g_m(\rho; \rho', \phi) = \left\{ j \frac{1}{4} \left[J_m(\beta_0 \rho') - \alpha H_m^{(2)}(\beta_0 \rho') \right] H_m^{(2)}(\beta_0 \rho) e^{jm\phi}, \rho' \leq \rho \leq \infty \right\} \quad (8b)$$

cont'd.

15.18 cont'd. Thus the Green's function of (3) can now be written as

$$G(p, \phi; p', \phi') = \begin{cases} j \frac{1}{4} \sum_{m=-\infty}^{+\infty} H_m^{(2)}(p, p') [J_m(p, p') - \alpha H_m^{(2)}(p, p')] e^{jm(\phi-\phi')}, & p \leq p' \\ j \frac{1}{4} \sum_{m=-\infty}^{+\infty} [J_m(p, p') - \alpha H_m^{(2)}(p, p')] H_m^{(2)}(p, p') e^{jm(\phi-\phi')}, & p' \leq p \leq \infty \end{cases} \quad (9a)$$

$$G(p, \phi; p', \phi') = \begin{cases} j \frac{1}{4} \sum_{m=-\infty}^{+\infty} [J_m(p, p') - \alpha H_m^{(2)}(p, p')] H_m^{(2)}(p, p') e^{jm(\phi-\phi')}, & p' \leq p \leq \infty \end{cases} \quad (9b)$$

The forms of (9a) and (9b) which are representative of the electric field E_ϕ are identical in form of (11-64a) and (11-64b).

15.19 The boundary conditions for this problem are

$$E_\phi(p=a, 0 \leq \phi \leq 2\pi) = 0 \quad (1)$$

For this polarization the ϕ component of the electric field is related to the z component of the magnetic field by (11-106b), or

$$\bar{E}_\phi = -\frac{1}{j\omega \epsilon_0} \frac{\partial H_z}{\partial p} \quad (2)$$

The Green's function will be selected for this problem to represent the z component of the magnetic field (H_z). Therefore according to (1) and (2) the Green's function will be selected to have the boundary condition of

$$\left. \frac{\partial G}{\partial p} \right|_{p=a} = G'(p=a) = 0 \quad (3)$$

The solution of this problem follows that on pages 928-930. Thus we can write using (15-151), and (15-159a)-(15-159b) that

$$G(p, \phi; p', \phi') = \sum_{m=-\infty}^{+\infty} g_m(p, p', \phi') e^{jm\phi} \quad (4)$$

$$g_m^{(1)} = A_m J_m(p, p') + B_m H_m^{(2)}(p, p') \quad (4a)$$

$$g_m^{(2)} = C_m H_m^{(2)}(p, p') \quad (4b)$$

Applying the boundary condition of (3) leads to

$$g_m'(p=a) = A_m J_m'(p, a) + B_m H_m^{(2)'}(p, a) = 0 \Rightarrow B_m = -A_m \frac{J_m'(p, a)}{H_m^{(2)'}(p, a)} \quad (5)$$

cont'd.

15.19 cont'd. Thus $g_m^{(1)}$ and $g_m^{(2)}$ can be expressed as

$$g_m^{(1)} = A_m \left\{ J_m(\beta_0 p) - \frac{J'_m(\beta_0 a)}{H_m^{(2)}(\beta_0 a)} H_m^{(2)}(\beta_0 p) \right\} = A_m \left\{ J_m(\beta_0 p) - \gamma H_m^{(2)}(\beta_0 p) \right\} \quad (6a)$$

$$g_m^{(2)} = C_m H_m^{(2)}(\beta_0 p) \quad (6b)$$

$$\gamma = \frac{J'_m(\beta_0 a)}{H_m^{(2)}(\beta_0 a)} \quad (6c)$$

The Wronskian of (15-44c) can be written as

$$W(p) = \beta_0 A_m C_m \left\{ [J_m(\beta_0 p) - \gamma H_m^{(2)}(\beta_0 p)] H_m^{(2)}(\beta_0 p) - H_m^{(2)}(\beta_0 p) [J'_m(\beta_0 p) - \gamma H_m^{(2)}(\beta_0 p)] \right\}$$

$$W(p) = \beta_0 A_m C_m \left\{ J_m(\beta_0 p) H_m^{(2)}(\beta_0 p) - J'_m(\beta_0 p) H_m^{(2)}(\beta_0 p) \right\} = -j \beta_0 A_m C_m [J_m(\beta_0 p) Y_m(\beta_0 p) - J'_m(\beta_0 p) Y'_m(\beta_0 p)] \quad (7)$$

Using the Wronskian for Bessel functions of (11-15), we can write (7) as

$$W(p) = -j \frac{2}{\pi} A_m C_m \frac{1}{p} \quad (8)$$

Thus we can write $g_m(p; p'; \phi)$, using (8), (15-156), (15-45a) and (15-45b) as

$$g_m(p; p'; \phi) = \left\{ j \frac{1}{4} \left[H_m^{(2)}(\beta_0 p) \right] [J_m(\beta_0 p) - \gamma H_m^{(2)}(\beta_0 p)] e^{-jm\phi'}, \quad a \leq p \leq p' \right\} \quad (9a)$$

$$g_m(p; p'; \phi) = \left\{ j \frac{1}{4} \left[J_m(\beta_0 p) - \gamma H_m^{(2)}(\beta_0 p) \right] H_m^{(2)}(\beta_0 p) e^{-jm\phi'} \quad p' \leq p \leq \infty \right\} \quad (9b)$$

Thus the Green's function of (4) can now be written as

$$G(p, \phi; p', \phi') = \left\{ j \frac{1}{4} \sum_{n=-\infty}^{+\infty} H_m^{(2)}(\beta_0 p) \left[J_m(\beta_0 p) - \frac{J'_m(\beta_0 a)}{H_m^{(2)}(\beta_0 a)} H_m^{(2)}(\beta_0 p) \right] e^{jn(\phi-\phi')} \quad a \leq p \leq p' \right\} \quad (10a)$$

$$G(p, \phi; p', \phi') = \left\{ j \frac{1}{4} \sum_{n=-\infty}^{+\infty} \left[J_m(\beta_0 p) - \frac{J'_m(\beta_0 a)}{H_m^{(2)}(\beta_0 a)} H_m^{(2)}(\beta_0 p) \right] H_m^{(2)}(\beta_0 p) e^{jn(\phi-\phi')} \quad p' \leq p \leq \infty \right\} \quad (10b)$$

The forms of (10a) and (10b), which are representative of the \mathbf{z} component of the magnetic field (H_z), are identical in form of (11-116f) and (11-116g).

15.20 The boundary conditions of this problem are
 $E_z(0 \leq r \leq \infty, \phi = \alpha) = E_z(1 \leq r \leq \infty, \phi = 2\pi - \alpha) = 0$ (1)

The Green's function will be selected to have the same boundary conditions as those of the electric field, or
 $G(r, \phi; r', \phi') = G(r, \phi; r', 2\pi - \phi') = 0$ (2)

To meet the boundary conditions on ϕ , we select a solution for the Green's function of the form

$$G(r, \phi; r', \phi') = \sum_m g_m(r, \phi; r', \phi') \{ A_m \cos[m(\phi - \alpha)] + B_m \sin[m(\phi - \alpha)] \} \quad (3)$$

Applying (2) leads to

$$G(\phi = \alpha) = \sum_m g_m(r, \phi; r', \alpha) \{ A_m(1) + B_m(0) \} = 0 \Rightarrow A_m = 0 \quad (4a)$$

$$G(\phi = 2\pi - \alpha) = \sum_m g_m(r, \phi; r', 2\pi - \alpha) B_m \sin[2m(\pi - \alpha)] = 0 \Rightarrow 2m(\pi - \alpha) = r\pi \quad (4b)$$

$$m = \frac{r\pi}{2(\pi - \alpha)}, \quad r = 1, 2, 3, \dots$$

Thus (3) reduces to

$$G(r, \phi; r', \phi') = \sum_m B_m g_m(r, \phi; r', \phi') \sin[m(\phi - \alpha)] \quad (5)$$

Substituting (5) into (15-153) leads to

$$\sum_m \left[\frac{\partial^2}{\partial r^2} + \frac{1}{r} \frac{\partial}{\partial r} - \frac{m^2}{r^2} + \beta_0^2 \right] g_m(r, \phi; r', \phi') \sin[m(\phi - \alpha)] = \frac{1}{r} \delta(r - r') \delta(\phi - \phi') \quad (6)$$

Multiplying both sides of (6) by $\sin[n(\phi - \alpha)]$, integrating both sides from α to $(2\pi - \alpha)$, and using the orthogonality of (15-48a) and (15-48b)

$$(2\pi - \alpha) \left[\frac{\partial^2 g_m}{\partial r^2} + \frac{1}{r} \frac{\partial g_m}{\partial r} + \left(\beta_0^2 - \frac{m^2}{r^2} \right) g_m \right] = \frac{1}{r} \sin[n(\phi - \alpha)] \delta(r - r') \quad (7)$$

or $\frac{d^2 g_m}{dr^2} + \frac{1}{r} \frac{dg_m}{dr} + \left(\beta_0^2 - \frac{m^2}{r^2} \right) g_m = \frac{1}{r} \sin[n(\phi - \alpha)] \delta(r - r') \quad (7a)$

The homogeneous form of (7a) is

$$\frac{d^2 g_m}{dr^2} + \frac{1}{r} \frac{dg_m}{dr} + \left(\beta_0^2 - \frac{m^2}{r^2} \right) g_m = 0 \quad (8)$$

whose two solutions can be written as

$$g_m^{(1)} = A_m J_m(\beta_0 r) \quad 0 \leq r \leq r' \quad (9a)$$

$$g_m^{(2)} = C_m H_m^{(1)}(\beta_0 r) \quad r \geq r' \quad (9b)$$

cont'd.

15.20 cont'd. The Wronskian of (15-44c) can be written as

$$W(p') = A_m C_m \beta_0 J_m(\beta_0 p') H_m^{(1)}(\beta_0 p) - A_m C_m \beta_0 H_m^{(2)}(\beta_0 p) J_m'(\beta_0 p')$$

$$W(p) = -j \beta_0 A_m C_m [J_m(\beta_0 p') Y_m'(\beta_0 p) - J_m'(\beta_0 p) Y_m(\beta_0 p')] \quad (10)$$

Using the Wronskian for Bessel functions of (11-95), we can write (10) as

$$W(p') = -j \frac{\pi}{\alpha} A_m C_m \frac{1}{p'} \quad (11)$$

Thus we can write $g_m(p; p', \phi)$, using (11), (7a), (15-45a) and (15-45b) as

$$g_m(p; p', \phi) = \begin{cases} j \frac{\pi}{2(\pi-\alpha)} H_m^{(1)}(\beta_0 p') J_m(\beta_0 p) \sin[m(\phi' - \alpha)] & 0 \leq p \leq p' \\ j \frac{\pi}{2(\pi-\alpha)} J_m(\beta_0 p) H_m^{(2)}(\beta_0 p) \sin[m(\phi' - \alpha)] & p' \leq p \leq \infty \end{cases} \quad (12a)$$

Thus the Green's function of (5) can then be written as

$$\left\{ j \frac{\pi}{2(\pi-\alpha)} \sum_m B_m H_m^{(1)}(\beta_0 p) J_m(\beta_0 p) \sin[m(\phi' - \alpha)] \sin[m(\phi - \alpha)], 0 \leq p \leq p' \right\} \quad (12a)$$

$$G(p, \phi; p', \phi') = \left\{ j \frac{\pi}{2(\pi-\alpha)} \sum_m B_m J_m(\beta_0 p') H_m^{(2)}(\beta_0 p) \sin[m(\phi' - \alpha)] \sin[m(\phi - \alpha)], p' \leq p \leq \infty \right\} \quad (12b)$$

$$\text{where } m = \frac{r\pi}{2(\pi-\alpha)}, \quad r = 1, 2, 3, \dots \quad (12c)$$

The forms of (12a) and (12b), which are representative of the \mathbf{z} -component of the electric field (E_z), are identical in form to (11-182a) and (11-182b).

15.21 The boundary conditions for this problem are

$$E_\phi(0 \leq \rho \leq \infty, \phi = \alpha) = E_\phi(0 \leq \rho \leq \infty, \phi = 2\pi - \alpha) = 0 \quad (1)$$

For this polarization the ϕ component of the electric field is related to the z component of the magnetic field by (11-106a), or

$$\mathbf{E} = +\frac{1}{j\omega\epsilon_0} \frac{\partial \mathbf{H}_z}{\partial \rho} \quad (2)$$

The Green's function will be selected for this problem to represent the z component of the magnetic field (H_z). Therefore according to (1) and (2) the Green's function will be selected to have the boundary conditions of

$$\left. \frac{\partial G}{\partial \phi} \right|_{\phi=\alpha} = G'(\phi=\alpha) = \left. \frac{\partial G}{\partial \phi} \right|_{\phi=2\pi-\alpha} = G'(\phi=2\pi-\alpha) = 0 \quad (3)$$

To meet the boundary conditions on ϕ , we select a solution for the Green's function of the form

$$G(\rho, \phi; \rho', \phi') = \sum_m g_m(\rho, \rho', \phi') \left\{ A_m \cos[m(\phi - \alpha)] + B_m \sin[m(\phi - \alpha)] \right\} \quad (4)$$

Applying (3) leads to

$$\begin{aligned} G'(\phi = \alpha) &= \sum_m m g_m(\rho, \rho', \phi') \left\{ -A_m \sin[m(\phi - \alpha)] + B_m \cos[m(\phi - \alpha)] \right\} \Big|_{\phi=\alpha} \\ &= \sum_m m g_m(\rho, \rho', \phi') \left\{ -A_m(0) + B_m(1) \right\} = 0 \Rightarrow B_m = 0 \end{aligned}$$

$$G'(\phi = 2\pi - \alpha) = \sum_m m g_m(\rho, \rho', \phi') \left\{ -A_m \sin[2m(\pi - \alpha)] \right\} = 0 \Rightarrow m = \frac{s\pi}{2(\pi - \alpha)}, \quad s = 0, 1, 2, \dots \quad (5)$$

Thus (4) is reduced to

$$G(\rho, \phi; \rho', \phi') = \sum_m A_m g_m(\rho, \rho', \phi') \cos[m(\phi - \alpha)] \quad (6)$$

$$m = \frac{s\pi}{2(\pi - \alpha)}, \quad s = 0, 1, 2, \dots \quad (6a)$$

Substituting (6) into (15-153) leads to

$$\sum_m \left[\frac{2^2}{\rho^2} + \frac{1}{\rho} \frac{\partial}{\partial \rho} - \frac{m^2}{\rho^2} + \beta_0^2 \right] g_m(\rho, \rho', \phi') \cos[m(\phi - \alpha)] = \frac{1}{\rho} \delta(\rho - \rho') \delta(\phi - \phi') \quad (7)$$

Multiplying both sides of (7) by $\cos[n(\phi - \alpha)]$, integrating both sides in ϕ from α to $(2\pi - \alpha)$, and using the orthogonality conditions for cosine functions, we can write that

cont'd.

15.21 cont'd.

$$(r-\alpha) \left[\frac{d^2 g_m}{dr^2} + \frac{1}{r} \frac{dg_m}{dr} + \left(\beta_0^2 - \frac{m^2}{r^2} \right) g_m \right] = \frac{1}{r} \cos[m(\phi'-\alpha)] \delta(r-r') \quad (8)$$

or

$$\frac{d^2 g_m}{dr^2} + \frac{1}{r} \frac{dg_m}{dr} + \left(\beta_0^2 - \frac{m^2}{r^2} \right) g_m = \frac{1}{r-\alpha} \sin[m(\phi'-\alpha)] \frac{1}{r} \delta(r-r') \quad (8a)$$

The homogeneous form of (8a) is

$$\frac{d^2 g_m}{dr^2} + \frac{1}{r} \frac{dg_m}{dr} + \left(\beta_0^2 - \frac{m^2}{r^2} \right) g_m = 0 \quad (9)$$

whose two solutions can be written as

$$g_m^{(1)} = A_m J_m(\beta_0 r) \quad 0 \leq r \leq r' \quad (10a)$$

$$g_m^{(2)} = C_m H_m^{(1)}(\beta_0 r) \quad r' \leq r \leq \infty \quad (10b)$$

The Wronskian of (15-44c) can be written as

$$W(r') = \beta_0 A_m C_m [J_m(\beta_0 r') H_m^{(1)}(\beta_0 r') - H_m^{(1)}(\beta_0 r') J_m'(\beta_0 r')] \quad (11)$$

$$W(r) = -j \beta_0 A_m C_m [J_m(\beta_0 r) Y_m'(\beta_0 r) - J_m'(\beta_0 r) Y_m(\beta_0 r)] \quad (11)$$

Using the Wronskian for Bessel functions of (11-95), we can write (11) as

$$W(r') = -j \frac{2}{\pi} A_m C_m \frac{1}{r'} \quad (12)$$

Thus we can write $g_m(r; r', \phi')$, using (12), (8a), (45-45a) and (45-45b), as

$$g_m(r; r', \phi') = \begin{cases} j \frac{\pi}{2(\pi-\alpha)} H_m^{(1)}(\beta_0 r') J_m(\beta_0 r) \cos[m(\phi'-\alpha)], & 0 \leq r \leq r' \\ j \frac{\pi}{2(\pi-\alpha)} J_m(\beta_0 r') H_m^{(1)}(\beta_0 r) \cos[m(\phi'-\alpha)], & r' \leq r \leq \infty \end{cases} \quad (13a)$$

$$G(r, \phi; r', \phi') = \begin{cases} j \frac{\pi}{2(\pi-\alpha)} \sum_m A_m H_m^{(1)}(\beta_0 r') J_m(\beta_0 r) \cos[m(\phi'-\alpha)] \cos[m(\phi-\alpha)], & 0 \leq r \leq r' \\ m = \frac{s\pi}{2(\pi-\alpha)}, s = 0, 1, 2, \dots \\ j \frac{\pi}{2(\pi-\alpha)} \sum_m A_m J_m(\beta_0 r') H_m^{(1)}(\beta_0 r) \cos[m(\phi'-\alpha)] \cos[m(\phi-\alpha)], & r' \leq r \leq \infty \end{cases} \quad (14b)$$

The forms of (14a) and (14b), which are representative of the z component of the magnetic field (H_z), are identical in form to (11-192a) and (11-192b).

15.22 To solve this problem, it is easier to assume that the point source of Figure P15.20 is located at the origin. When that is done, the field radiated by the point source is only a function of r (not a function of θ or ϕ). Thus in expanded form (15-166) can be written as

$$\frac{1}{r^2} \frac{\partial}{\partial r} \left(r^2 \frac{\partial G}{\partial r} \right) + \beta^2 G = \frac{\partial^2 G}{\partial r^2} + \frac{2}{r} \frac{\partial G}{\partial r} + \beta^2 G = \frac{1}{4\pi r^2} \delta(r-r') \quad (1)$$

with $G(r=\infty) = \text{finite}$ (1a)

Since $G(r, r')$ is only a function of r , (1) can be written as

$$\frac{d^2 G}{dr^2} + \frac{2}{r} \frac{dG}{dr} + \beta^2 G = \frac{1}{4\pi r^2} \delta(r-r') \Rightarrow r \frac{d^2 G}{dr^2} + 2r \frac{dG}{dr} + \beta^2 r^2 G = \frac{\delta(r-r')}{4\pi} \quad (2)$$

A solution of the homogeneous form of (2) takes the form of

$$G(r) = A \frac{e^{-j\beta r}}{r} + B \frac{e^{+j\beta r}}{r} \quad (3)$$

For an instant time convention, the boundary condition of (1a) allows us to write that $B=0$. Thus (3) reduces to

$$G(r) = A \frac{e^{-j\beta r}}{r} \quad (4)$$

To evaluate the coefficient A in (4), we can integrate (2) over an infinitesimal volume which surrounds the origin and to then allow the radial distance r to approach zero ($r \rightarrow 0$). Doing this we find that $A = -\frac{1}{4\pi}$ which allows us to write (4) as

$$G(r) = -\frac{e^{-j\beta r}}{4\pi r} \quad (5)$$

When the source is removed from the origin, as shown in Figure P15.20, the Green's function of (5) can be written as

$$G = -\frac{1}{4\pi} \frac{e^{-j\beta R}}{R} = -\frac{1}{4\pi} \frac{e^{-j\beta |\mathbf{r}-\mathbf{r}'|}}{|\mathbf{r}-\mathbf{r}'|} \quad (6)$$

(b). The reduction of the three-dimensional Green's function of (6) to that of the two-dimensional one of Example is obtained by the use of the integral of (11-28a). This allows us to reduce the three-dimensional problem to a two-dimensional one.