# Part 02 - Discrete-Time Signals and Systems

## Baboo J. Cui

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In this part, LTI(linear time invariant) will be studied due to the 2 important facts:

- a large collection of math techniques can be applied to it
- many practical complicate system can be approximated by LTI system

## 1 Discrete-Time Signal

**Discrete time signal** is a function of an independent variable that is an integer, it is not defined at instants between two samples. The following 3 methods are used to represent it:

- functional representation
- tabular representation
- sequence representation (usually with arrow to indicate t = 0)

## 1.1 Elementary DT signals

Here are some important basic signals:

• unit sample signal(also known as unit impulse), denoted as  $\delta(n)$ 

$$\delta(n) = \begin{cases} 1, & n = 0 \\ 0, & n \neq 0 \end{cases}$$

• unit step signal, denoted as u(n)

$$u(n) = \begin{cases} 1, & n \ge 0 \\ 0, & n < 0 \end{cases}$$

• unit ramp signal, denoted as  $u_r(n)$ 

$$u_r(n) = \begin{cases} n, & n \ge 0 \\ 0, & n < 0 \end{cases}$$

• **exponential** signal, denoted as x(n)

$$x(n) = a^n$$

a can be either real or complex.

#### 1.2 Complex Exponential Signal

If a is complex number and  $x(n) = a^n, \forall n \in \mathbb{Z}$ , x is said to be complex exponential signal. a generally can be expressed as

$$a=re^{j\theta}$$

therefore

$$x(n) = a^n = (re^{j\theta})^n = r^n e^{jn\theta}$$
$$= r^n (\cos(n\theta) + j\sin(n\theta))$$

Clearly, we can see that:

- real part of the signal:  $x_R(n) = r^n \cos(n\theta)$
- imaginary part of the signal:  $x_I(n) = r^n \sin(n\theta)$
- amplitude function of the signal:  $|x(n)| = A(n) = r^n$
- **phase function** of the signal:  $\angle x(n) = \phi(n) = n\theta$

## 1.3 classification of DT Signals

Discrete time signal can be classified according to its characteristic.

#### energy signal vs power signal

The energy E of a signal is defined as:

$$E = \sum_{n = -\infty}^{\infty} |x(n)|^2$$

if E is bounded, it is called **energy signal**, and its average power P is defined as:

$$P = \lim_{N \to \infty} \frac{1}{2N+1} \sum_{n=-N}^{N} |x(n)|^2$$

if P is finite and nonzero, it is called **power signal**, clearly:

- finite E leads to P = 0
- $\bullet$  infinite E may lead to either finite P or infinite P

#### periodic signal vs aperiodic signal

A signal x(n) is periodic with period N if

$$x(n+N) = x(n), \forall n \in \mathbb{Z}$$

the smallest N is called **fundamental period**, if such N doesn't exist, it is aperiodic. Periodic signals are power signals(why?).

## even signal vs odd signal

For real signal x(n), if x(-n) = x(n) it is called even signal, and if x(-n) = -x(n) it is called odd signal. Any signal x(n) can be decomposed to the sum of an even and an odd signal, namely

$$x(n) = x_e(n) + x_o(n)$$

where

$$x_e(n) = \frac{1}{2}[x(n) + x(-n)]$$

and

$$x_o(n) = \frac{1}{2}[x(n) - x(-n)]$$

## 1.4 Manipulation on DT Signals

Suppose the origin DT signal is x(n), here are some basic manipulation on it:

- $x(n) \to x(n-k)$ : delay k units if k is positive, advance k units if k is negative. It's impossible to advance a signal in real time
- $x(n) \to x(-n)$ : fold or reflect the signal about time origin n=0
- $x(n) \to x(kn)$ : down-sampling the signal by a factor of k
- $x(n) \to kx(n)$ : **amplify** the signal by factor of k

VIP: delay and reflect are not commutative!

## 2 DT Systems

A DT system is algorithm that operates on **input(excitation)** according to some well-defined rule to produce on **output(response)**. We say input is transformed by system into output, mathematically:

$$y(n) = \mathcal{T}[x(n)]$$

## 2.1 IO Description

It defines the relation between input and output explicitly, here are some examples:

- identity system:y(n) = x(n)
- delay system: y(n) = x(n-k)
- advance system: y(n) = x(n+k)
- moving average filter:  $y(n) = \frac{1}{3}[x(n+1) + x(n) + x(n-1)]$
- accumulator:  $y(n) = \sum_{k=-\infty}^{n} x(n)$

Note that some system output depends on initial condition.

## 2.2 Block Diagram Representation of DT System

This representation is intuitive and useful, here are basic blocks:

- ullet adder, it is memory-less
- constant multiplier: also known as amplifier or attenuator
- signal multiplier: multiply two signals
- unit delay: denoted as  $z^{-1}$
- $\bullet$  advance system: denoted as z

Some basic DT system can be realized by these blocks easily!

#### 2.3 Classification of DT System

Systems can be classified based on some general properties that they satisfy.

- static vs dynamic system: the output of static system only depends on the current input, meaning the system is memory-less, otherwise the system is said to be dynamic or has memory
- time-invariant vs time-variant: a relaxed system is said to be time-invariant if  $x(n) \to y(n)$  implies that  $x(n-k) \to y(n-k)$ , otherwise time-variant
- linear vs non-linear system: a system is linear if and only if

$$\mathcal{T}[a_1x_1(n) + a_2x_2(n)] = a_1\mathcal{T}[x_1(n)] + a_2\mathcal{T}[x_2(n)]$$

this is also known as **superposition principle**(**multiplicative** and **additivity** properties), otherwise non-linear. Note that a relaxed linear system always give zero output if input is zero, which is known as **ZIZO** 

- **causal** vs **non-causal** system: a system is said to be causal if output only depends on current and past input, otherwise non-causal
- stable vs unstable system: for relaxed system, if any bounded input produces an bounded output(BIBO), the system is said to be stable, otherwise unstable

## 2.4 Interconnection of DT System

There are two ways to connect 2 DT systems:

- cascade(series): output is the composition of systems, usually 2 systems are note commutative
- parallel: sum of each output

These two methods are used to construct or decompose a system.

## 3 Analysis of DT LTI System

The reason why we focus on DT LTI systems is that they have lots of nice properties

#### 3.1 Techniques for the Analysis of Linear System

There are 2 methods for analyzing the response of the system:

• direct IO equation: solve y(n) explicitly by the following difference equation

$$y(n) = -\sum_{k=1}^{N} a_k y(n-k) + \sum_{k=0}^{M} b_k x(n-k)$$

• decompose method: decompose a signal to elementary ones and get the corresponding response, then use linear property to construct the output, a signal can be decompose as:

$$x(n) = \sum_{k} c_k x_k(n)$$

and corresponding response is:

$$y_k(n) = \mathcal{T}[x_k(n)]$$

therefore

$$y(n) = \mathcal{T}[x(n)] = \mathcal{T}\left[\sum_{k} c_k x_k(n)\right] = \sum_{k} c_k \mathcal{T}[x_k(n)]$$

usually, **unit impulse** and **complex exponential** signals are chosen to be the elementary signal.

## 3.2 Decompose Signal into Sum of Impulses

Any signal can be decomposed into weighted sum of impulses, choose the elementary signal  $x_k(n)$  to be:

$$x_k(n) = \delta(n-k)$$

which is the unit impulse delayed by k units, note that:

$$x(n)\delta(n-k) = x(k)\delta(n-k)$$

so that the signal can be written as:

$$x(n) = \sum_{k=-\infty}^{\infty} x(k)\delta(n-k)$$

Comments:

- n is a **sequence** and k is an integer
- must be clear about the graphic interpretation of the formula

#### 3.3 Convolution Sum

Define the system response to unit impulse at n = k by h(n, k):

$$y(n,k) = h(n,k) = \mathcal{T}[\delta(n-k)]$$

so that if the system is LTI:

$$y(n) = \mathcal{T}[x(n)] = \mathcal{T}\left[\sum_{k=-\infty}^{\infty} x(k)\delta(n-k)\right]$$

$$= \sum_{k=-\infty}^{\infty} x(k)\mathcal{T}[\delta(n-k)] \quad \text{linearity}$$

$$= \sum_{k=-\infty}^{\infty} x(k)h(n,k)$$

$$= \sum_{k=-\infty}^{\infty} x(k)h(n-k) \quad \text{time invariant}$$

#### Comments:

- LTI system is completely characterized by function h
- $\bullet$  the output y is called **convolution sum**
- x(k) is a real number, h(n-k) is a sequence

There are several ways to calculate convolution:

- software: Matlab(recommend)
- folding method(just know this)
- sequence sum method(highly recommended, intuitive)

## 3.4 Convolution Properties

Convolution is denoted by \*, and has some important properties:

- identity:  $x(n) * \delta(n) = x(n)$
- time shift:  $x(n) * \delta(n-k) = x(n-k)$
- commutative: x(n) \* h(n) = h(n) \* x(n)
- associative:  $[x(n) * h_1(n)] * h_2(n) = x(n) * [h_1(n) * h_2(n)]$ , related to cascading decomposition
- distributive:  $x(n) * [h_1(n) + h_2(n)] = x(n) * h_1(n) + x(n) * h_2(n)]$ , related to parallel decomposition

### 3.5 Causal LTI System

Causality can be translated to a condition on IR, an LTI system is causal if and only if its IR is zero for negative values of n. It can be implemented for real-time signal processing.

## 3.6 Stability of LTI System

Recall relaxed BIBO stable was introduced. For LTI system, it is **stable** if is IR is absolutely summable:

$$\sum_{k=-\infty}^{\infty} |h(k)| < \infty$$

this is the necessary and sufficient condition to ensure the stability of system, this also indicate that the unit impulse response to it decays with time and dies out eventually.

## 4 DT System Described by Difference Equation

The response of **FIR** LTI system can s implied be calculated by definition of convolution, however **IIR** LTI system cannot be. Difference equation can solve this problem.

## 4.1 Recursive and Non-recursive DT System

Two definition:

- if the output of a system only depends on the past input, it is call **non-recursive sytem**, which is related to FIR
- If the output of a system not only depends on the past input but also the past output(feedback), it is called **recursive system**, which is related to IIR

For recursive system, **initial condition** is required to determine the response, it summarize all past history of the system, it is also called **state variable** in control area.

## 4.2 LTI System Characterized by Constant-Coefficient Difference Equations

LTI system can be characterized by IR as mentioned before, it can also characterized by constant-coefficient difference equation, which is a subclass of the recursive and non-recursive systems. The general equation is:

$$y(n) = -\sum_{k=1}^{N} a_k y(n-k) + \sum_{k=0}^{M} b_k x(n-k)$$

- $\bullet$  N is called the **order** of the system
- the negative sign on the right-side first is for convenience to use

The response can be decomposed into 2 parts:

- zero-input response: also called natural response
- zero-state response: also called forced response

Here is a list of properties:

- linearity: 3 requirements response can be decomposed into the two parts, superposition applies to both natural and forced response
- causality: systems describe by constant-coefficient difference equation must be linear and time-invariant
- **stability**: BIBO stable is satisfied if every bounded input and every bounded initial condition leads to bounded output

# 4.3 Solution to Linear Constant-Coefficient Difference Equations

**Direct method** is given here(by z-transformation is called indirect method). The total solution is the sum of two parts:

$$y(n) = y_h(n) + y_p(n)$$

•  $y_h(n)$ : homogeneous or complementary solution, it's the zero-input response, it is related to sum of exponential, it has general form:

$$y_h(n) = C_1 \lambda_1^n + C_2 n \lambda_1^n + C_3 n^2 \lambda_1^n + \dots + C_m n^{m-1} \lambda_1^n + C_{m+1} \lambda_{m+1}^n + \dots + C_N \lambda_N^n$$

- n is the sequence index, it may be confusing to put it at position of power of  $\lambda$
- $-\lambda_i$  are roots to the characteristic polynomial of the system, the number of roots is N, which is also the degree of the system
- $-\lambda_1$  has multiplicity of m, which shows the general situation
- $-C_i$  are determined by initial condition
- $y_p(n)$ : **particular** solution, which take the basic form of the input x(n) and here is a list:

x(n)	$y_p(n)$
A (constant)	K(constant)
$AM^n$	$KM^n$
$An^M$	$K_0 n^M + K_1 n^{M-1} + \dots + K_M$
$A^n n^M$	$A^{n}(K_{0}n^{M}+K_{1}n^{M-1}+\cdots+K_{M})$
$A\cos\omega_0 n$ or $A\sin\omega_0 n$	$K_1\cos\omega_0 n + K_2\sin\omega_0 n$

put the  $y_p$  back to the difference equation and the constants K can be solved, and  $y_p$  is obtained.

One more thing: difference equation can be obtained from the zero-state response of the system.

#### 4.4 The IR of LTI Recursive System

Essentially, the response of a unit impulse to relaxed system is the special zero-state output (IR):

$$y_{zs}(n) = h(n)$$

consequently, the response of system to a unit impulse consists only of the solution to homogeneous equation. When the system is N-th order, the solution is:

$$y_h(n) = \sum_{k=1}^{N} C_k \lambda^n k$$

when the roots  $\lambda_k$  are distinct, the IR is:

$$h(n) = \sum_{k=1}^{N} C_k \lambda_k^n$$

So the necessary and sufficient condition for stability of causal IIR system is that all roots of characteristic polynomial are less than unity in magnitude.

## 5 Extra

## 5.1 Complicate Signal Manipulation