

CNN Sensor Analytics with Hybrid-Float6 on Low-Power Resource-Constrained Embedded FPGAs.

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ABSTRACT The use of artificial intelligence (AI) in sensor analytics is entering a new era based on the use of ubiquitous embedded connected devices. This transformation requires the adoption of design techniques that reconcile accurate results with sustainable system architectures. As such, improving efficiency of AI hardware engines must be considered. In this paper, we present the Hybrid-Float6 (HF6) quantization on shallow CNNs for sensor data analytics and its dedicated hardware design. This approach reduces over-fit on feature extraction and improves generalization. As dedicated hardware design, we propose a fully customizable tensor processor (TP) implementing a pipelined vector dot-product with HF6. This approach reduces energy consumption and resource utilization. The proposed embedded hardware/software architecture is unified with TensorFlow Lite. We evaluate the applicability of our approach with a data analytics application for structural health monitoring (SHM) for anomaly localization. The embedded hardware/software framework is demonstrated on XC7Z007S as the smallest and most inexpensive Zynq SoC device.

INDEX TERMS Convolutional neural networks, structural health monitoring, hardware accelerator, TensorFlow Lite, embedded systems, FPGA, custom floating-point

I. INTRODUCTION

THERE is a growing demand for ubiquitous AI sensor data analytics. Industry 4.0 and smart city infrastructure leverage AI solutions to increase productivity and adaptability [1]. These solutions are powered by advances in machine learning (ML), hardware engines, and big data. Hence, enhancement of these should be considered for research, as they are the machinery of the future.

CNN-based models represent the essential building blocks in 2D pattern recognition tasks. Sensor-based applications such as mechanical fault diagnosis [2], [3], structural health monitoring (SHM) [4], human activity recognition (HAR) [5], hazardous gas detection [6] have been powered by CNN models in industry and academia.

CNN-based models, as one of the main types of artificial neural networks (ANNs), have been widely used in sensor analytics with automatic learning from sensor data [7]–[10]. In this context, CNN models are applied for automatic feature learning, usually, from 1D time series as well as for 2D time-frequency spectrograms. CNN models provide advantages such as local dependency, scale invariance, and noise resilience in data analytics. However, these models represent

compute-intensive and power-hungry tasks, particularly, for low-power and resource-limited Internet-of-Things (IoT) devices.

Dedicated hardware architectures are typically used to enhance compute performance and power efficiency. In terms of computational throughput, graphics processing units (GPUs) offer the highest performance. In terms of power efficiency, ASIC and FPGA solutions are well known to be more energy efficient (than GPUs) [11]. As a result, numerous commercial ASIC and FPGA accelerators have been proposed, targeting both high performance computing (HPC) for data-centers and embedded systems applications.

However, most FPGA accelerators have been implemented to target mid- to high-range FPGAs for compute costly CNN models such as AlexNet, VGG-16, and ResNet-18. The power supply demands, physical dimensions, air cooling and heat sink requirements, and in some cases their elevated price make these implementations unsustainable and not always feasible for ubiquitous resource-constrained applications.

Furthermore, to reduce the computational cost for CNN inference there are two types of research [12]: the first one is deep compression including weight pruning, weight

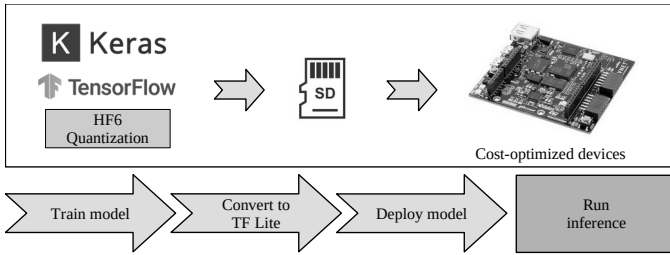


FIGURE 1. The workflow of our approach on embedded FPGAs.

quantization, and compression storage [13], [14]; the second type of research corresponds to a more efficient data representation, also known as quantization for dedicated circuit implementation. In this group, hardware implementations with customized 8-bit floating-point computation have been proposed [12], [15], [16]. However, these implementations are inadequate for embedded applications, since the target devices are high-end FPGA and PCIe architectures.

The aforementioned works have good accuracy with re-training, more aggressive data representations such as binary [17], ternary [18], and mixed precision (2-bit activations and ternary weights) [19] may suffer from great accuracy loss even with time-consuming retraining. The afforded mentioned limitations make these implementations inadequate for accurate data analytics in low-power embedded applications.

In this paper, we present the Hybrid-Float6 quantization on CNNs for sensor data analytics and its dedicated hardware design on low-power resource-constrained embedded FPGAs. The HF6 implements 6-bit floating-point quantization on the trainable parameters of convolution layers and standard floating-point on feature maps. This approach reduces over-fit on feature extraction and improves generalization. As dedicated hardware design, we propose a parameterized tensor processor implementing a pipelined vector dot-product with HF6. This approach reduces energy consumption and resource utilization facilitating on-chip stationary weights on limited footprint devices. The embedded hardware/software architecture is unified with TensorFlow Lite using delegate interface to accelerate *Conv2D* tensor operations. We evaluate the applicability of our approach with a CNN model and hardware design exploration for sensor analytics of SHM for anomaly localization based on regression. The embedded hardware/software framework is demonstrated on XC7Z007S as the smallest and most inexpensive Zynq SoC device, see Fig. 1.

To the best of our knowledge, this is the first research addressing 6-bit floating-point quantization on CNN-based models and its dedicated hardware implementation. The purpose of this work is to demonstrate the feasibility of HF6 as an alternative for ubiquitous and sustainable AI deployment.

Our main contributions are as follows:

- 1) We present the Hybrid-Float6 quantization. This approach improves generalization by reducing over-fit on

feature extraction. The HF6 is wrapped into the standard floating-point representation (IEEE 754) allowing compatibility with standard hardware. Therefore, it can be beneficial for inference in other devices.

- 2) We develop a hardware/software co-design framework for sensor analytics applications on low-power and resource-limited FPGAs. This is a scalable architecture integrating TensorFlow Lite core in the embedded system.
- 3) We present a customizable tensor processor as a dedicated hardware for HF6. This design computes *Conv2D* tensor operations employing a pipelined vector dot-product with parametrized on-chip memory utilization. The compute engine of the tensor processor is implemented with standard floating-point, fixed-point, and HF6.
- 4) We demonstrate the potential of our approach by addressing CNN model and hardware design exploration for sensor analytics of anomaly localization based on regression for SHM. We evaluate inference accuracy, compute performance, hardware resource utilization, and energy consumption.

The rest of the paper is organized as follows. Section II covers the related work; Section III introduces the background to *Conv2D* and *DepthwiseConv2D* tensor operations; Section IV describes the system design of the hardware/software architecture and the quantized aware training method; Section V presents the experimental results thorough a design exploration flow; Section VI concludes the paper.

This design exploration framework is available to the community as an open-source project at (*hidden for double blinded review*).

II. RELATED WORK

A. HARDWARE IMPLEMENTATIONS TARGETING RESOURCE-CONSTRAINED FPGAS

In the literature we find plenty of hardware architectures dedicated to CNN accelerators implemented in FPGA and ASIC designs. However the related work on low-power and resource-limited devices is reduced. To the best of our knowledge, two research papers have been reported hardware implementations targeting XC7Z007S as the smallest device from Zynq-7000 SoC Family.

In [20], Chang Gao et al., presented EdgeDRNN, a recurrent neural network (RNN) accelerator for edge inference. This implementation adopts the spiking neural network (SNN) inspired delta network algorithm to exploit temporal sparsity in RNNs. However, this hardware architecture is dedicated to RNNs.

In [21], Paolo Meloni et al., presented a CNN inference accelerator for compact and cost-optimized devices. This implementation uses fixed-point for processing light-weight CNN architectures with a power efficiency between 2.49 to 2.98 GOPS/s/W.

B. HYBRID CUSTOM FLOATING-POINT QUANTIZATION

Reference [22] proposed a mixed data representation with floating-point for weights and fixed-point for activations (e.g., outputs of a layer). Reference [23] developed an 8-bit floating-point quantization scheme, which needs an extra inference batch to compensate for the quantization error. However, Reference [22] and Reference [23] did not present a circuit design for their approaches.

1) FPGA implementations

Reference [15] implements 16-bit floating-point in contrast to the 32-bit commonly used for computing. However, this implementation is inadequate for embedded applications, since the target device is a PCIe architecture. The 8-bit floating-point is also tried in FPGA [12]. Another 8-bit arithmetic, called block floating-point (BFP), is also applied [16], where a parameter has its own mantissa but shares a same exponent for one data block.

III. BACKGROUND

A. CONV2D TENSOR OPERATION

The *Conv2D* tensor operation is described in Eq. (1), where h is the input feature map, W is the convolution kernel (known as filter), and b is the bias for the output feature map [24]. We denote *Conv* as *Conv2D* operator.

$$\text{Conv}(W, h)_{i,j,o} = \sum_{k,l,m}^{K,L,M} h_{(i+k,j+l,m)} W_{(o,k,l,m)} + b_o \quad (1)$$

IV. SYSTEM DESIGN

In this section we describe the system design as a hardware/software co-design framework for floating-point CNN acceleration targeting resource-limited FPGAs. This is a parameterized architecture that allows design exploration with TensorFlow Lite.

A. BASE EMBEDDED SYSTEM ARCHITECTURE

As a hardware/software co-design, the system architecture is an embedded CPU+FPGA platform. Fig. 2 illustrates the top-level hardware architecture. The TPs execute the low-level tensor operations delegated from the CPU. The TPs employ AXI-Lite interface for configuration and AXI-Stream interfaces via Direct Memory Access (DMA) for data movement from off-chip DDR memory. Each TP asserts an interrupt flag once the job/transaction is complete. Interrupt events are handled by the embedded CPU to collect results and start a new transaction. The hardware architecture can resize its resource utilization by modifying the number of TP instances prior to the hardware synthesis.

B. TENSOR PROCESSOR

The TP is a dedicated hardware module to compute tensor operations. The hardware architecture is described in Fig. 3. This architecture implements high performance off-chip communication with AXI-Stream, direct CPU communication with AXI-Lite, and on-chip storage utilizing BRAM.

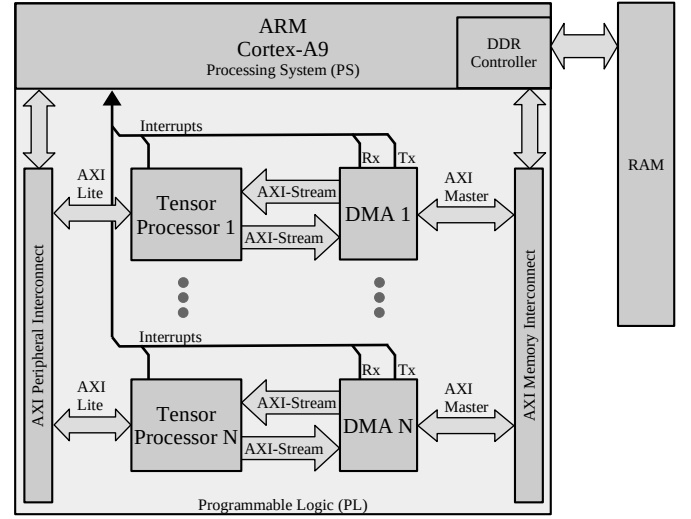


FIGURE 2. Base embedded system architecture.

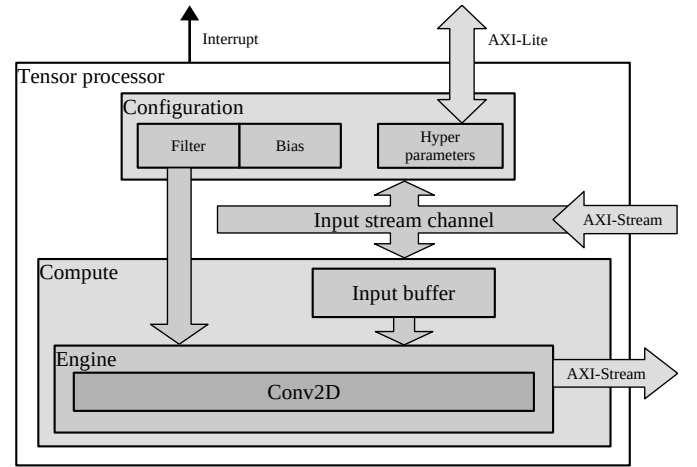


FIGURE 3. Hardware architecture of the proposed tensor processor.

This hardware architecture is implemented with high-level synthesis (HLS). The tensor operations are implemented based on the C++ TensorFlow Lite micro kernels.

The TP is an extensible hardware module that executes low-level tensor operations. This implementation creates a cooperative hardware-software architecture. In this paper, we focus on the *Conv2D* tensor operator that executes inference of convolution layers.

1) Modes of operation

This accelerator has two modes of operation: *configuration* and *execution*.

- In *configuration* mode, the TP receives the tensor operation ID for *Conv2D* and hyperparameters: stride, dilation, padding, offset, activation, depth-multiplier, input shape, filter shape, bias shape, and output shape. Afterwards, the TP receives filter and bias tensors, which are locally stored in BRAM.

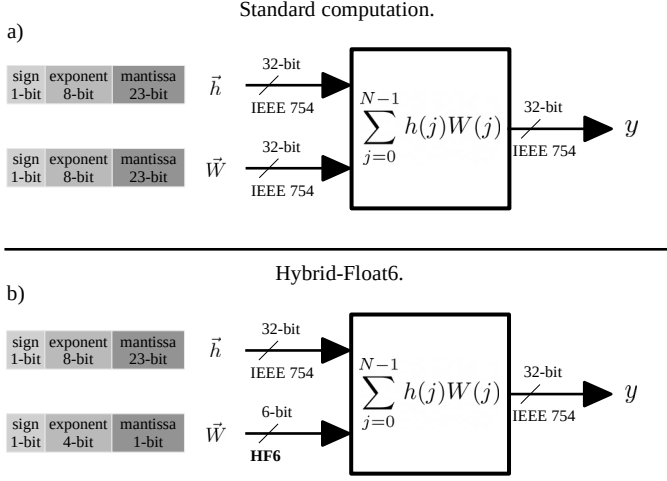


FIGURE 4. Dot-product hardware module with (a) standard floating-point and (b) Hybrid-Float6.

- In *execution* mode, the TP executes the tensor operation according to the hyperparameters given in the configuration mode. During execution, the input and output tensor-buffers are moved from/to the DDR memory via DMA.

2) Dot-product with hybrid floating-point

We implement the floating-point computation adopting the dot-product with hybrid custom floating-point and logarithmic approximation [25]. The hardware dot-product is illustrated in **Fig. 4**. This approach: (1) denormalizes input values, (2) executes computation with integer format for exponent and mantissa, and finally, (3) it normalizes the result into IEEE 754 format, see **Fig. 5**. Rather than a parallelized structure, this is a pipelined hardware design suitable for resource-limited devices. The latency in clock cycles of this hardware module is defined by **Eq. (2)** and **Eq. (3)**, where N is the vector length. The latency equations are obtained from the general pipelined hardware latency formula: $L = (N - 1)II + IL$, where II is the initiation interval (**Fig. 5(a)**), and IL is the iteration latency (**Fig. 5(b)**). Both II and IL are obtained from the high-level synthesis analysis. The hybrid logarithmic computation removes the mantissa bit-field, which removes the mantissa multiplication and correction in clock cycle 3 and 4, respectively, see **Fig. 5**.

$$L_{hf} = N + 7 \quad (2)$$

$$L_{log} = N + 6 \quad (3)$$

As a design parameter, both the exponent and mantissa bit-width of the weight/filter vector provides a tunable knob to trade-off between resource utilization and QoR [26]. These parameters must be defined before hardware synthesis.

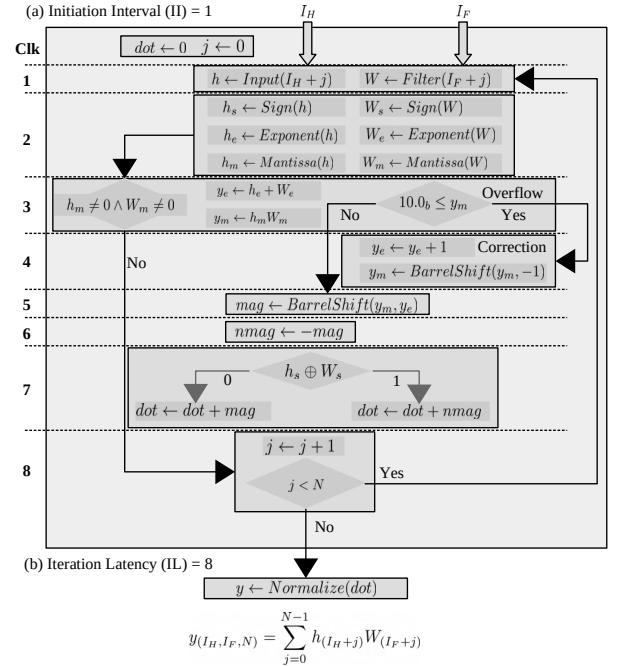


FIGURE 5. Pipelined hardware module for vector dot-product with hybrid custom floating-point, (a) exhibits the initiation interval of 1 clock cycle, and (b) presents the iteration latency of 8 clock cycles. I_H and I_F represent the input and filter buffer indexes, respectively.

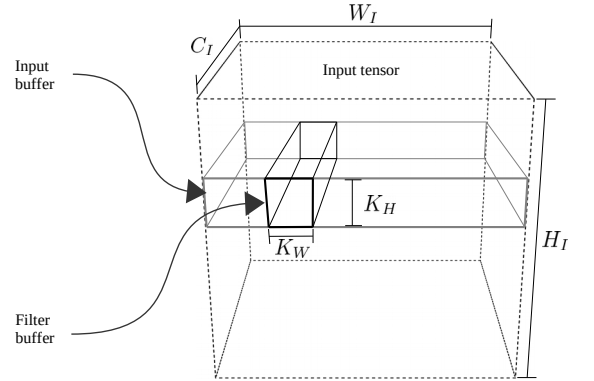


FIGURE 6. Design parameters for on-chip memory buffers on the TP.

3) On-chip memory utilization

The total on-chip memory utilization on the TP is defined by **Eq. (4)**, where $Input_M$ is the *input buffer*, $Filter_M$ is the *filter buffer*, $Bias_M$ is the *bias buffer*, and V_M represents the local variables required for the design. The on-chip memory buffers are defined in bits. **Fig. 6** illustrates the convolution operation utilizing the on-chip memory buffers.

$$TP_M = Input_M + Filter_M + Bias_M + V_M \quad (4)$$

The memory utilization of *input buffer* is defined by **Eq. (5)**, where K_H is the height of the convolution kernel, W_I is the width of the input tensor, C_I is the number of input

channels, and $BitSize_I$ is the bit size of each input tensor element.

$$Input_M = K_H W_I C_I BitSize_I \quad (5)$$

The memory utilization of *filter buffer* is defined by **Eq. (6)**, where K_W and K_H are the width and height of the convolution kernel, respectively; C_I and C_O are the number of input and output channels, respectively; and $BitSize_F$ is the bit size of each filter element.

$$Filter_M = C_I K_W K_H C_O BitSize_F \quad (6)$$

The memory utilization of *bias buffer* is defined by **Eq. (7)**, where C_O is the number of output channels, and $BitSize_B$ is the bit size of each bias element.

$$Bias_M = C_O BitSize_B \quad (7)$$

As a design trade-off, **Eq. (8)** defines the capacity of output channels based on the given design parameters. The total on-chip memory TP_M determines the TP capacity.

$$C_O = \frac{TP_M - V_M - K_H W_I C_I BitSize_I}{C_I K_W K_H BitSize_F + BitSize_B} \quad (8)$$

The floating-point formats implemented in the TP are defined by $BitSize_F$, $BitSize_B$ and $BitSize_I$. The HF6 is defined by 1-bit sign, 4-bit exponent and 1-bit mantissa. These are design parameters defined before hardware synthesis. This allows fine control of BRAM utilization, which is suitable for resource-limited devices.

C. QUANTIZED AWARE TRAINING

The quantize-aware training method is an iterative optimization. The custom CNN model is initially trained with early stop monitoring for minimal validation loss, then the CNN model is retrained including the quantization method implemented as a callback function, see **Algorithm 1**. The quantization method maps the full precision filter and bias values with rounding to the closest representable quantized values, see **Algorithm 2**. The quantize-aware training method starts with a wide exponent size target (e.g. 3-bits) and gradually reduces the target size until the model drops to a given accuracy degradation threshold (e.g. 1%). We have observed that the exponent bit size plays a more predominant influence on the model accuracy than the mantissa bit size. The mantissa bit size can be set to the minimum (e.g. 1-bit). This method quantizes the filter and bias tensors of the *Conv2D* and *SeparableConv2D* layers. This method is integrated in TensorFlow/Keras framework. The resulting quantized parameters are truncated and buffered in the on-chip memory of the TP during *configuration* mode.

Algorithm 1: Training method.

input: $MODEL$ as the CNN.
input: E_{size} as the target exponent bit size.
input: M_{size} as the target mantissa bits size.
input: D_{train} as the training data set.
input: D_{val} as the validation data set.
input: Acc_d as the accuracy degradation threshold.
input: $Loop_{max}$ as the max quantization loop iterations.
output: $MODEL$ as the quantized CNN.

```

Train( $MODEL$ ,  $D_{train}$ ,  $D_{val}$ ) // Regular training
 $acc_i \leftarrow Evaluate(MODEL, D_{val})$  // Benchmark
 $acc_q \leftarrow 0$ ,  $loop_c \leftarrow 0$  // Initialize quantize training
while ( $acc_q < acc_i - Acc_d$ )  $\wedge$  ( $loop_c < Loop_{max}$ ) do
  // Iterative optimization
   $callback \leftarrow Quantize(E_{size}, M_{size})$ 
  Train( $MODEL$ ,  $D_{train}$ ,  $D_{val}$ ,  $callback$ )
   $acc_q \leftarrow Evaluate(MODEL, D_{val})$ 
   $loop_c \leftarrow loop_c + 1$ 
end while

```

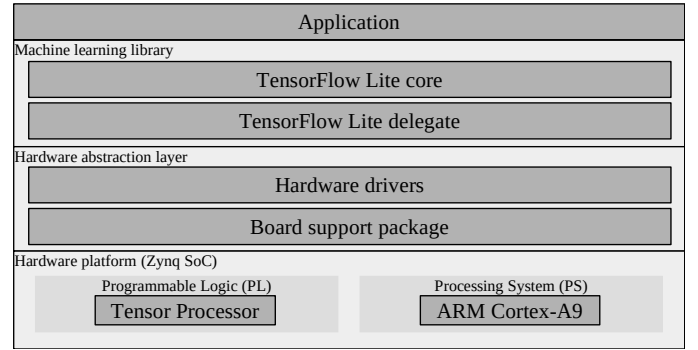


FIGURE 7. Base embedded software architecture.

D. EMBEDDED SOFTWARE ARCHITECTURE

The software architecture is a layered object-oriented application framework written in C++, see **Fig. 7**. The main characteristics of the software layers are as follows:

- **Application:** As the highest level of abstraction, this layer implements the embedded application logic with the ML library.
- **Machine learning library:** This layer consists of TensorFlow Lite micro. This offers a comprehensive high level API that allows ML inference. This provides delegate interfaces for custom hardware accelerators.
- **Hardware abstraction layer:** This layer consists of the hardware drivers to handle initialization and runtime operation of the TP and DMA.

V. EXPERIMENTAL RESULTS

In this section, we present experimental results of a sensor analytics application for SHM and its hardware deployment.

To demonstrate the HF6 quantization, we implement an analytics application to predict x- y- coordinates of struc-

Algorithm 2: Custom floating-point quantization method.

input: $MODEL$ as the CNN.
input: E_{size} as the target exponent bit size.
input: M_{size} as the target mantissa bits size.
input: $STDM_{size}$ as the IEEE 754 mantissa bit size.
output: $MODEL$ as the quantized CNN.

for $layer$ in $MODEL$ **do**
 if $layer$ is *Conv2D* or *SeparableConv2D* **then**
 $filter, bias \leftarrow GetWeights(layer)$
 for x in $filter$ and $bias$ **do**
 $sign \leftarrow GetSign(x)$
 $exp \leftarrow GetExponent(x)$
 $fullexp \leftarrow 2^{E_{size}-1} - 1$ // Get full range value
 $cman \leftarrow GetCustomMantissa(x, M_{size})$
 $leftman \leftarrow GetLeftoverMantissa(x, M_{size})$
 if $exp < -fullexp$ **then**
 $x \leftarrow 0$
 else if $exp > fullexp$ **then**
 $x \leftarrow (-1)^{sign} \cdot 2^{fullexp} \cdot (1 + (1 - 2^{-M_{size}}))$
 else
 if $2^{STDM_{size}-M_{size}-1} - 1 < leftman$ **then**
 $cman \leftarrow cman + 1$ // Above halfway
 if $2^{M_{size}} - 1 < cman$ **then**
 $cman \leftarrow 0$ // Correct mantissa overflow
 $exp \leftarrow exp + 1$
 end if
 end if
 // Build custom quantized floating-point value
 $x \leftarrow (-1)^{sign} \cdot 2^{exp} \cdot (1 + cman \cdot 2^{-M_{size}})$
 end if
 end for
 $SetWeights(layer, filter, bias)$
 end if
end for

tural anomalies based on acoustic sensor data with CNN-regression models. We address the model design exploration using TensorFlow and Keras library.

To demonstrate the HF6 hardware concept, we deploy the CNN model for low-power inference in the smallest Zynq SoC. For comparison, the hardware engine is implemented with 32-bit floating-point, 8-bit fixed-point, and HF6. We address the hardware design exploration using HLS.

We present statistical evaluations of the CNN models with standard floating-point, fixed-point, and HF6 quantization. We evaluate the computational latency, resource utilization, and power dissipation. Finally, we present a discussion of the presented results.

A. CNN MODEL DESIGN EXPLORATION

B. HARDWARE DESIGN EXPLORATION

The proposed hardware/software co-design framework is demonstrated on the MiniZed development board with a

TABLE 1. Inference on embedded CPU.

Layer	Latency (ms)
HX_IN	1.184
H1_CONV	4.865
H2_POOL	3.656
H3_CONV	20.643
H4_POOL	0.828
H5_FC	3.099
HY_OUT	0.004
TOTAL	34.279

Zynq-7007S system-on-chip (SoC). This device integrates a single ARM Cortex-A9 processing system (PS) and programmable logic (PL) equivalent to Xilinx Artix-7 (FPGA) in a single chip [27]. The Zynq-7007S SoC architecture maps the custom logic and software in the PL and PS respectively as an embedded system.

In this platform, we implement the proposed hardware architecture to deploy the CNN model for SHM shown in Fig. 8. The CNN model is created, trained, and quantized using Keras/TensorFlow with Python on a desktop computer. The resulting model is converted to TensorFlow Lite, which is deployed on the MiniZed. The Zynq-7007S SoC performs the model inference with TensorFlow Lite core API running on the PS. The computational workload of the convolution layers is delegated to the TP on the PL.

C. PERFORMANCE BENCHMARK

1) Benchmark on embedded CPU

We examine the performance of the embedded CPU for inference with no hardware acceleration. In this case, the embedded software builds the CNN as a sequential model mapping the entire computation to the CPU (ARM Cortex-A9) at 666 MHz and a power dissipation of 1.658W.

The inference on the CPU achieves a latency of 40ms. The model is computed with standard floating-point arithmetic with no accuracy degradation. The latency and schedule of the CNN inference are displayed in Tab. 1 and Fig. 9 respectively.

2) Benchmark on tensor processor with standard floating-point computation

To benchmark the computation on hardware TP with standard floating-point, we implement the system architecture with one TP. In this case, the embedded software builds the CNN as a sequential model mapping *Conv2D* tensor operations to the TP at 200 MHz as clock frequency. The hardware mapping and the computation schedule of this deployment are displayed in Tab. 2 and Fig. 10.

The post-implementation resource utilization and power dissipation are shown in Tab. 3.

The TP instantiates an on-chip weight matrix of 52,000 entries, which is sufficient to store $W \in \mathbb{R}^{5 \times 5 \times 2 \times 32}$ and $B \in \mathbb{R}^{5 \times 5 \times 32 \times 64}$ for weight and bias, respectively. In order to reduce BRAM utilization, we use a custom floating-point representation composed of 4-bit exponent and 4-bit man-

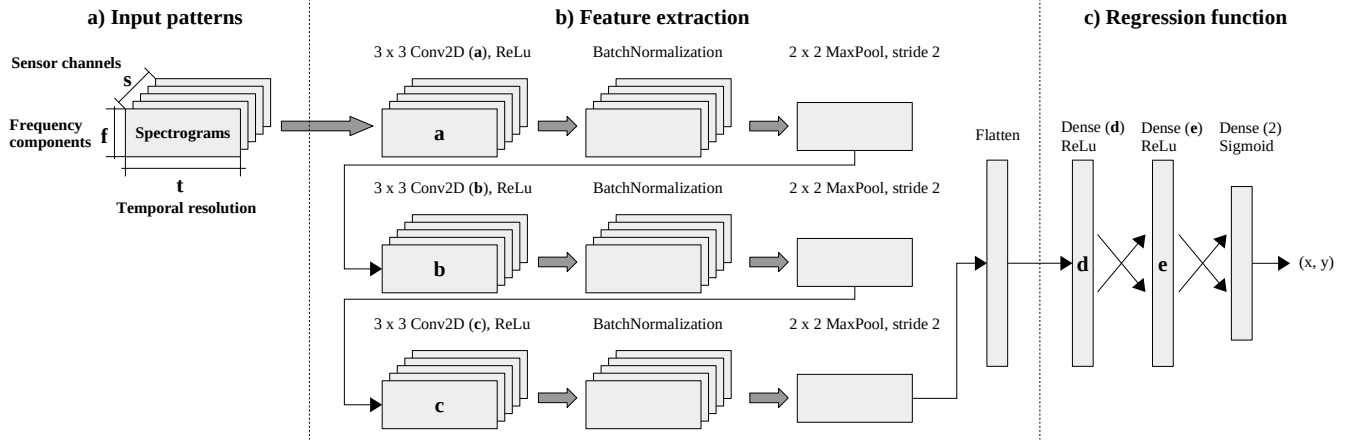


FIGURE 8. CNN model for case study.

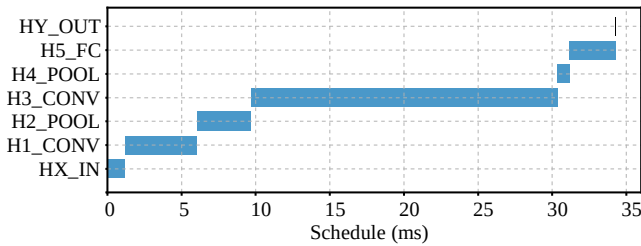


FIGURE 9. Computation on embedded CPU.

TABLE 2. Performance of TP with standard floating-point (IEEE 754) computation.

Hardware mapping		Computation schedule (ms)			
Layer	PU	t_s	t_{CPU}	t_{PU}	t_f
HX_IN	Spike	0	0.056	0.370	0.426
H1_CONV	Conv1	0.058	0.598	2.002	2.658
H2_POOL	Pool1	0.658	0.126	1.091	1.875
	Pool2	0.785	0.125	1.075	1.985
H3_CONV	Conv2	0.911	0.280	3.183	4.374
	Conv3	1.193	0.279	3.176	4.648
H4_POOL	Pool3	1.473	0.037	0.481	1.991
H5_FC	FC	1.512	0.101	1.118	2.731
HY_OUT	CPU	1.615	0.004	0	1.619

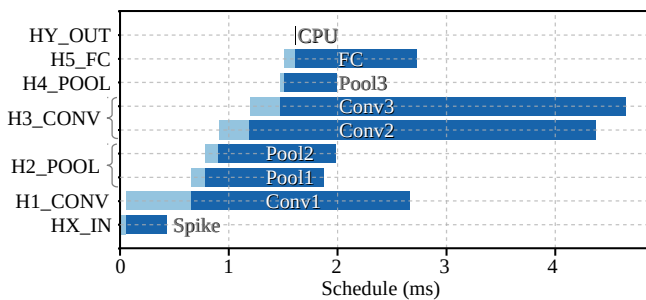


FIGURE 10. Performance of TP with standard floating-point (IEEE 754) computation.

point representation for computation.

TABLE 3. Resource utilization and power dissipation with standard floating-point (IEEE 754) computation.

PU	LUT	FF	DSP	BRAM 18K	Power (mW)
Spike	2,640	4,903	2	2	38
Conv	2,765	4,366	19	37	89
Pool	2,273	3,762	5	3	59
FC	2,649	4,189	8	9	66

The implementation of dot-product with standard floating-point arithmetic (IEEE 754) utilizes proprietary multiplier and adder floating-point operator cores. Vivado HLS accomplishes floating-point arithmetic operations by mapping them onto Xilinx LogiCORE IP cores, these floating-point operator cores are instantiated in the resultant RTL [28]. In this case, the implementation of the dot-product with the standard floating-point computation reuses the multiplier and adder cores already instantiated in other compute sections of the TP. The post-implementation resource utilization and power dissipation of the floating-point operator cores are shown in Tab. 4.

TABLE 4. Resource utilization and power dissipation of multiplier and adder floating-point (IEEE 754) operator cores.

Core operation	DSP	FF	LUT	Latency (clk)	Power (mW)
Multiplier	3	151	325	4	7
Adder	2	324	424	8	6

D. DESIGN EXPLORATION WITH HYBRID CUSTOM FLOATING-POINT AND LOGARITHMIC APPROXIMATION

In this section, we address a design exploration to evaluate our approach for inference using hybrid custom floating-point and logarithmic approximation. First, we examine the weight matrix of each convolution layer in order to determine the minimum requirements for numeric representation and memory storage. Second, we implement the TP using

tissa. Each 8-bit entry is promoted to its standard floating-

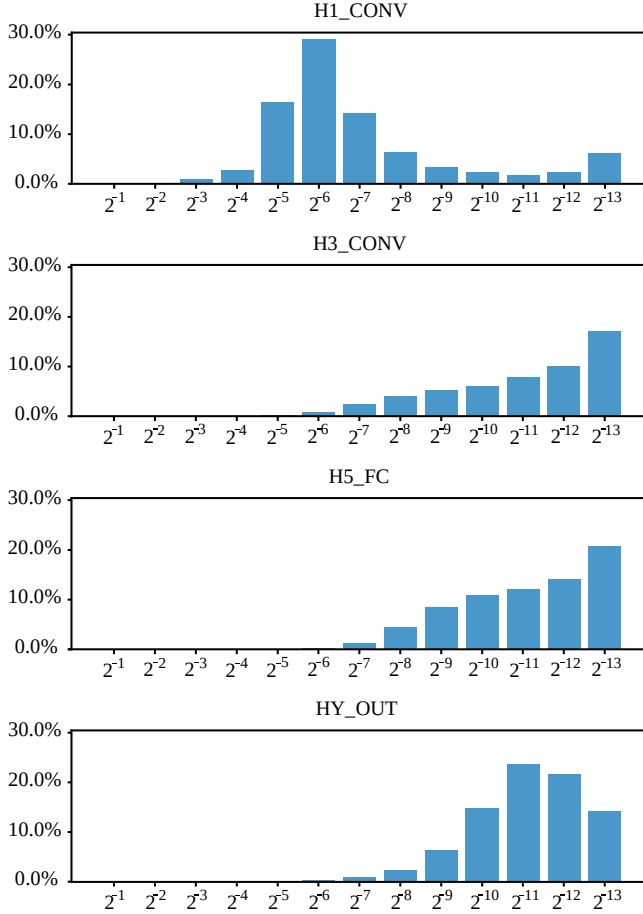


FIGURE 11. \log_2 -histogram of each synaptic weight matrix showing the percentage of matrix elements with given integer exponent.

the minimal floating-point and logarithmic representation as design parameters. Finally, we evaluate the overall performance, inference accuracy, resource utilization, and power dissipation.

1) Parameters for numeric representation of weight matrix

We obtain information for the numerical representation of the synaptic weight matrices from their \log_2 -histograms presented in **Fig. 11**. These histograms show the distribution of weight values in each matrix. We observe that the minimum integer exponent value is -13 . Hence, applying **Eq. (??)** and **Eq. (??)** to the given CNN, we obtain $E_{\min} = -13$ and $N_E = 4$, respectively. Therefore, 4-bits are required for the absolute binary representation of the exponents.

For quality configurability, the mantissa bit-width is a knob parameter that is tuned by the designer. This procedure leverages the builtin error-tolerance of neural networks and performs a trade-off between resource utilization and QoR. In the following subsection, we present a case study with 1-bit mantissa corresponding to the custom floating-point approximation.

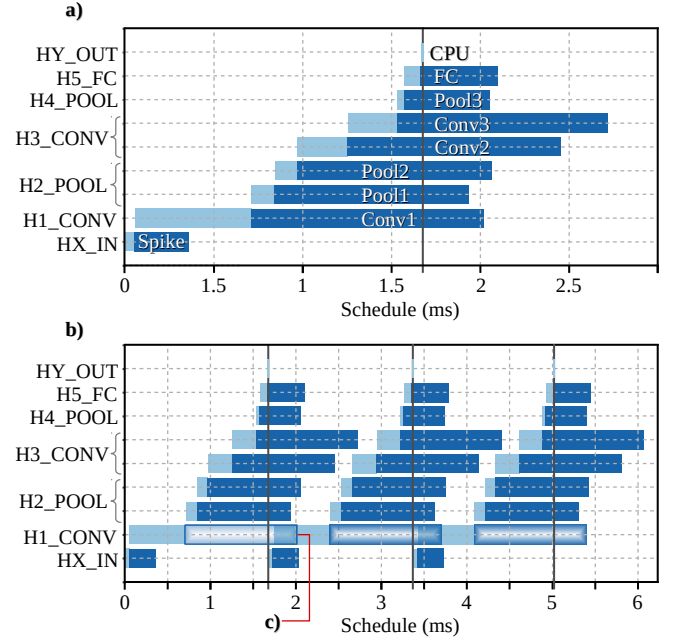


FIGURE 12. Performance on processing units with hybrid custom floating-point approximation, (a) exhibits computation schedule, (b) presents cyclic computation schedule, and (c) shows the performance of *Conv2* from a previous computation cycle during the preprocessing of *H1_CONV* on the current computation cycle without bottleneck.

2) Design exploration for dot-product with hybrid custom floating-point approximation

For this design exploration, we use a custom floating-point representation composed of 4-bit exponent and 1-bit mantissa. This format is used for both the filter matrix and bias vectors of each convolution layer. The TP instantiates on-chip stationary both the filter matrix and bias vectors for X and Y entries of 6-bit (S1E4M1). The available memory size is large enough to store $W \in \mathbb{R}^{5 \times 5 \times 2 \times 32}$ and $W \in \mathbb{R}^{5 \times 5 \times 32 \times 64}$ for \vec{F} and \vec{b} , respectively. The hardware mapping and the computation schedule of this implementation are displayed in **Tab. 6** and **Fig. 12**.

As shown in the computation schedule in **Tab. 6** and **Fig. 12**, this implementation achieves a peak acceleration of $55\times$, and a power efficiency of 5.5 GFLOP/s/W . This configuration achieves an accuracy of 90.97% correct regressions on the 500 validation samples. This indicates an accuracy gain of 0.33% .

The post-implementation resource utilization and power dissipation are shown in **Tab. 5**.

TABLE 5. Resource utilization and power dissipation of processing units with hybrid custom floating-point approximation.

PU	LUT	FF	DSP	BRAM 18K	Power (mW)
Conv	3,139	4,850	19	25	82
FC	3,265	5,188	8	9	66

TABLE 6. Performance of hardware processing units with hybrid custom floating-point approximation.

Hardware mapping		Computation schedule (ms)			
Layer	PU	t_s	t_{CPU}	t_{PU}	t_f
HX_IN	Spike	0	0.055	0.307	0.362
H1_CONV	Conv1	0.057	0.654	1.309	2.020
H2_POOL	Pool1	0.713	0.131	1.098	1.942
	Pool2	0.845	0.125	1.098	2.068
H3_CONV	Conv2	0.972	0.285	1.199	2.456
	Conv3	1.258	0.279	1.184	2.721
H4_POOL	Pool3	1.538	0.037	0.484	2.059
H5_FC	FC	1.577	0.091	0.438	2.106
HY_OUT	CPU	1.669	0.004	0	1.673

TABLE 7. Performance of hardware processing units with hybrid logarithmic approximation.

Hardware mapping		Computation schedule (ms)			
Layer	PU	t_s	t_{CPU}	t_{PU}	t_f
HX_IN	Spike	0	0.055	0.264	0.319
H1_CONV	Conv1	0.057	0.655	1.271	1.983
H2_POOL	Pool1	0.714	0.130	1.074	1.918
	Pool2	0.845	0.126	1.106	2.077
H3_CONV	Conv2	0.973	0.285	1.179	2.437
	Conv3	1.258	0.278	1.176	2.712
H4_POOL	Pool3	1.538	0.037	0.488	2.063
H5_FC	FC	1.577	0.091	0.388	2.056
HY_OUT	CPU	1.669	0.004	0	1.673

3) Design exploration for dot-product with hybrid logarithmic approximation

As the most efficient setup and yet the worst-case quality configuration, we use a 4-bit integer exponent for logarithmic representation of \vec{F} and \vec{b} . The hardware mapping and the computation schedule of this implementation are displayed in **Tab. 7** and **Fig. 13**. As shown in the computation schedule in **Tab. 7** and **Fig. 13**, this implementation achieves a peak acceleration of **55X** and a power efficiency of **5.5 GFLOPS/s/W**. This quality configuration achieves an accuracy degradation **0.84%** on correct regressions on the **500** validation samples.

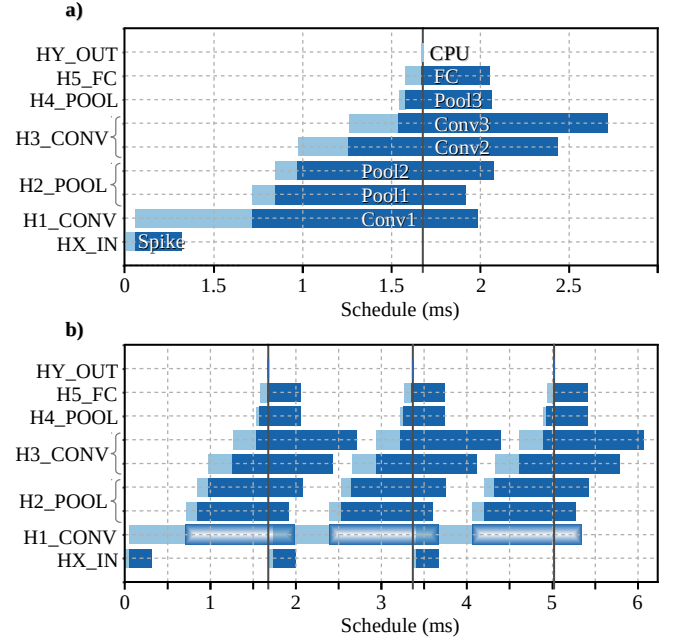
The post-implementation resource utilization and power dissipation are shown in **Tab. 8**.

TABLE 8. Resource utilization and power dissipation of processing units with hybrid logarithmic approximation.

PU	LUT	FF	DSP	BRAM 18K	Power (mW)
Conv	3,086	4,804	19	21	78
FC	3,046	4,873	8	8	66

E. RESULTS AND DISCUSSION

As a reference, the inference on embedded CPU using standard 32-bit floating-point achieves an accuracy gain of **0.3%** with a latency of **3,450.28ms**. As a second reference point, the inference on TP with standard floating-point presents a latency of **34.5ms**, as result we get a **10.7×** latency enhancement.

**FIGURE 13.** Performance of processing units with hybrid logarithmic approximation, (a) exhibits computation schedule, and (b) illustrates cyclic computation schedule.

As a demonstration of the proposed hardware/software architecture, the inference with TP using 5-bit custom floating-point (4-bit exponent, 1-bit mantissa) and 4-bit logarithmic (4-bit exponent) achieves **55.5×** latency enhancement. This results in an accuracy gain of **0.33%** and degradation of **0.46%**, respectively.

Regarding resource utilization and power dissipation, the TP with 5-bit custom floating-point has a **43.24%** reduction of BRAM, and a **12.35%** of improvement in energy efficiency over the standard floating-point implementation. However, the hybrid dot-product with custom floating-point does not reuse the available floating-point operator cores instantiated from other computational sections (see **Tab. 4**). Therefore, the logic required for the dot-product must be implemented, which is reflected as additional utilization of LUT and FF resources. The experimental results of the design exploration are summarized in **Tab. 9**. The platform implementations are summarized in **Tab. 10**, and their power dissipation breakdowns are presented in **Fig. 14**.

F. HARDWARE DESIGN EXPLORATION

To evaluate the methodology, we employ **Eq. (8)**, giving the maximum hyper parameters from models *A* and *B*: $W_I = 32$, $C_I = 60$, $C_O = 120$, $K_W = K_H = 3$. For the number formats, $BitSize_I = 32$ -b, and $BitSize_F = BitSize_B = 6$ -bits. To determine V_M , we use HLS tool, which gives an estimate of 6 RAM blocks. The performance evaluation and the hardware resource utilization are displayed in **Tab. ??** and **Tab. 11**, respectively.

TABLE 9. Experimental results.

Dot-product implementation	PU	Post-implementation resource utilization				Power (mW)	Latency		Accuracy (%) ^e	
		LUT	FF	DSP	BRAM 18K		T_{SC} (ms)	Gain ^d	Noise 0%	50%
Standard floating-point computation ^a	Conv	2,765	4,366	19	37	89	3.18	10.7x	98.98	98.63
	FC	2,649	4,189	8	9	66				
Hybrid custom floating-point approx ^b	Conv	3,139	4,850	19	25	82	1.67	20.5x	98.97	98.47
	FC	3,265	5,188	8	9	66				
Hybrid logarithmic approximation ^c	Conv	3,086	4,804	19	21	78	1.67	20.5x	98.84	95.22
	FC	3,046	4,873	8	8	66				

^a Reference with standard floating-point arithmetic (IEEE 754).^b Synaptic weight with number representation composed of 4-bit exponent and 1-bit mantissa.^c Synaptic weight with number representation composed of 4-bit exponent.^d Acceleration with respect to the computation on embedded CPU (ARM Cortex-A9 at 666 MHz) with latency $T_{SC} = 34.28ms$.^e Accuracy on 10,000 image test set with 1000 spikes.

TABLE 10. Platform implementations.

Platform implementation	Post-implementation resource utilization				Power (W)	Clock (MHz)	Latency		Accuracy (%) ^f
	LUT	FF	DSP	BRAM 18K			T_{SC} (ms)	Gain ^e	
Ref. [?] ^a	42,740	57,118	49	92	2.519	250	4.65	7.4x	99.02
This work (standard floating-point computation) ^b	39,514	56,036	82	180	2.420	200	3.18	10.7x	98.98
This work (hybrid custom floating-point approx) ^c	42,021	58,759	82	156	2.369	200	1.67	20.5x	98.97
This work (hybrid logarithmic approximation) ^d	41,060	57,862	82	148	2.324	200	1.67	20.5x	98.84

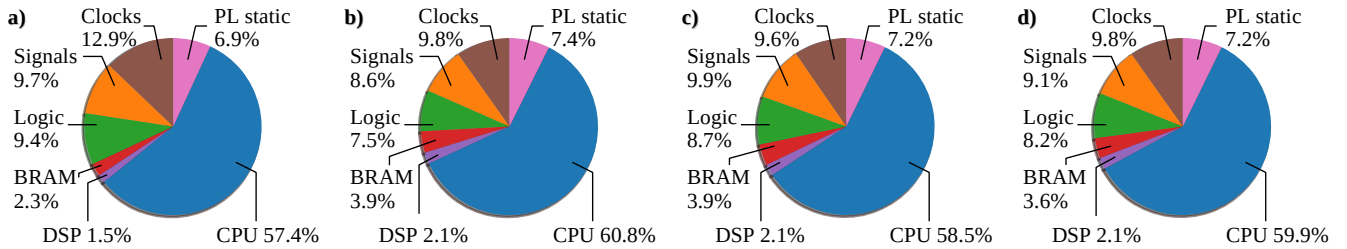
^a Reference architecture with homogeneous AUs using standard floating-point arithmetic (IEEE 754).^b Reference architecture with specialized heterogeneous PUs using standard floating-point arithmetic (IEEE 754).^c Proposed architecture with specialized heterogeneous PUs using synaptic weight with number representation composed of 4-bit exponent and 1-bit mantissa.^d Proposed architecture with specialized heterogeneous PUs using synaptic weight with number representation composed of 4-bit exponent.^e Acceleration with respect to the computation on embedded CPU (ARM Cortex-A9 at 666 MHz) with latency $T_{SC} = 34.28ms$.^f Accuracy on 10,000 image test set with 1000 spikes.

FIGURE 14. Power dissipation breakdown of platform implementations. (a) Ref. [?] architecture with homogeneous AUs using standard floating-point arithmetic (IEEE 754), (b) reference architecture with specialized heterogeneous PUs using standard floating-point arithmetic (IEEE 754), (c) proposed architecture with hybrid custom floating-point approximation, and (d) proposed architecture with hybrid logarithmic approximation.

- 1) **XC7Z007S**: As a resource-limited FPGA, this device has a capacity of 14,400 LUTs and 1.8Mb of BRAM. This limitation allows to instantiate one TP with *Conv* due to its LUT capacity. With Eq. (4), we obtain a BRAM utilization of 789.84Kb. This implementation presents a peak runtime acceleration of $55\times$ in model *A* at the tensor operation (3A) *Conv* with a power reduction of $808\times$.
- 2) **XC7Z010**: This device has a capacity of 17,600 LUTs and 2.1Mb of BRAM. These resources allow to instantiate two TPs with *Conv*, and one TP with *Conv* and *DConv* engines. With Eq. (4), we obtain a BRAM utilization of 1,580Kb. This implementation presents a peak runtime acceleration of $105\times$ in model *A* at the tensor operation (3A) *Conv* with a power reduction of $1121\times$. On model *B*, (6B) *Conv* presents a peak acceleration of $43.8\times$. The *DConv* tensor operator yields

an acceleration of $6.75\times$, which is limited since the pipelined vector dot-product performs on channel wise.

TABLE 11. Hardware resource utilization and estimated power dissipation.

Device	TP	Post-implementation resource utilization				Power (W)
		LUT	FF	DSP	BRAM 36Kb	
XC7Z007S	1	7,939	8,955	20	25	1.44
		55%	31%	30%	50%	
XC7Z010	2	13,542	15,279	36	46	1.880
		77%	43%	45%	76%	

VI. CONCLUSIONS

In this paper, we present a design exploration framework for floating-point CNNs acceleration on low-power, resource-limited embedded FPGAs. This design targets inexpensive IoT and near-sensor data analytic applications. We propose

a scalable hardware architecture with customizable tensor processors integrated with TensorFlow Lite. The implemented hardware optimization realizes a pipelined vector dot-product using hybrid custom floating-point and logarithmic approximation with fully parametrized on-chip memory utilization. This approach accelerates computation, reduces energy consumption and resource utilization. We proposed a quantized-aware training method to maintain and increase inference accuracy with custom reduced floating-point formats. This approach is fundamentally more efficient compared to equivalent fixed-point number representations. Experimental results on XC7Z007S (MiniZed) and XC7Z010 (Zybo) demonstrate peak acceleration and power efficiency of 105X and 5.5 GFLOP/s/W, respectively.

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