

Information theory uses, definition, Uncertainty & property.

Information - It is the intelligence / ideas / message in Information theory.

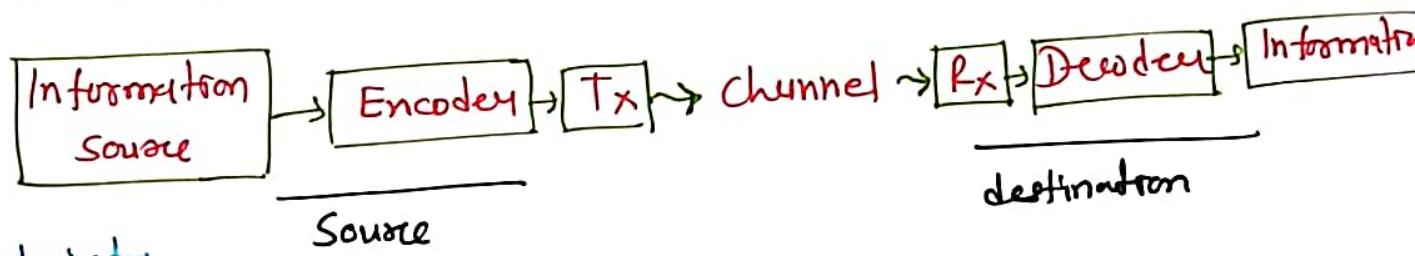
Message - Electrical signal

- Speech / voice
- Picture / Image
- Video
- text

- In communication system, Information is transmitted from source to destination

E-1

from source to destination



### Uncertainty

- $X = \{x_0, x_1, \dots, x_n\}$
  - $P = \{P_0, P_1, \dots, P_n\}$
  - total Probability  $P = \sum_{i=1}^n P_i$
  - As probability of message  $\downarrow^u$  then uncertainty  $\uparrow^u$
-

- $\rightarrow P_0 = 0.1$
- As probability of message  $\downarrow^a$  then uncertainty  $\uparrow^a$

### Measure of Information

- It is a Information Content of a message.
- Consider an Information source emitting independent messages  $m = \{m_1, m_2, \dots, m_n\}$  with probabilities of occurrence is  $P = \{P_1, P_2, \dots, P_n\}$ .
- Here,  $P_1 + P_2 + \dots + P_n = 1$ .
- Amount of Information is given by,

$$I_k = \log_2 \left( \frac{1}{P_k} \right) = \frac{\log (1/P_k)}{\log 2} \quad \boxed{\text{bits}}$$

E9

## Properties of Information

- More Uncertainty about Message then Information is more.

$$\rightarrow P_1 = \frac{1}{4} \quad , \quad P_2 = \frac{1}{2}$$
$$U_1 \qquad \qquad U_2$$

$$\rightarrow U_1 > U_2$$

$$\begin{array}{c|c|c} \rightarrow Q. I_1 = \log_2 \left( \frac{1}{P_1} \right) & I_2 = \log_2 \left( \frac{1}{P_2} \right) & I_1 > I_2 \\ = \log_2 \left( \frac{1}{(1/4)} \right) & = \log_2 \left( \frac{1}{(1/2)} \right) \\ = \log_2 4 & = \log_2 (2) \\ = 2 \log_2 2 & = 1 \text{ bit} \\ = 2 \text{ bits} & \end{array}$$

■

- Receiver knows message being transmitted then information is zero.
- $P = 1$
- $I = \log_2\left(\frac{1}{P}\right)$   
 $= \log_2 1$   
 $= 0 \text{ bit}$
- $I_1$  is the information of message  $m_1$ , and  $I_2$  of  $m_2$ , then combined information of  $m_1$  &  $m_2$  =  $I_1 + I_2$
- $I_1 = \log_2\left(\frac{1}{P_1}\right) \rightarrow$  Since messages  $m_1$  &  $m_2$  are independent. So combined probability =  $P_1 P_2$   
 $\rightarrow I_2 = \log_2\left(\frac{1}{P_2}\right) \rightarrow I = \log_2\left(\frac{1}{P}\right) = \log_2\left(\frac{1}{P_1 P_2}\right) = \log_2\left(\frac{1}{P_1}\right) + \log_2\left(\frac{1}{P_2}\right)$   
 $\Rightarrow \boxed{I = I_1 + I_2}$

- If there are  $m = 2^n$  equally likely messages, then amount of information carried by each message will be =  $n$  bits.
  - probability of each message =  $\frac{1}{m}$
  - $I = \log_2 \left(\frac{1}{p}\right)$ 
    - $\Rightarrow \log_2(m)$
- $\rightarrow m = 2^n$   
 $\Rightarrow I = \log_2 2^n$   
 $= n \log_2 2$   
 $I = N$  bits

## Examples on Information in Information Theory

① calculate amount of information for

$$\text{a) } P_1 = 1/4, \quad \text{b) } P_2 = 3/4$$

$$\rightarrow P_1 = 1/4$$

$$I_1 = \log_2 \left( \frac{1}{P_1} \right)$$

$$= \log_2 (4)$$

$$= 2 \log_2 2$$

$$= 2 \text{ bits}$$

$$\rightarrow P_2 = 3/4$$

$$I_2 = \log_2 \left( \frac{1}{P_2} \right)$$

$$= \log_2 \left( \frac{4}{3} \right)$$

$$= \frac{\log (4/3)}{\log 2}$$

$$= 0.415 \text{ bits.}$$



2) A Card is selected at random from a deck of playing cards. If you have been told that it is red in colour.

1) How much information you have received.

2) How much more information you needed to completely specify the card.

- Total Cards = 52

Total Red Cards = 26

$$P = \frac{26}{52} = \frac{1}{2}$$

$$- I = \log_2\left(\frac{1}{P}\right)$$

$$= \log_2(2)$$

$$= 1 \text{ bit}$$

→ " " we are given a card . . . .  
Cards. If you have been told that it is red in colour.

- 1) How much information you have received.
- 2) How much more information you needed to completely specify the card.

- Total Cards = 52

Total Red Cards = 26

$$P = \frac{26}{52} = \frac{1}{2}$$

-  $I = \log_2\left(\frac{1}{P}\right)$

$$= \log_2(2)$$

$$= 1 \text{ bit}$$

$$\rightarrow \text{Probability } P = \frac{1}{26}$$

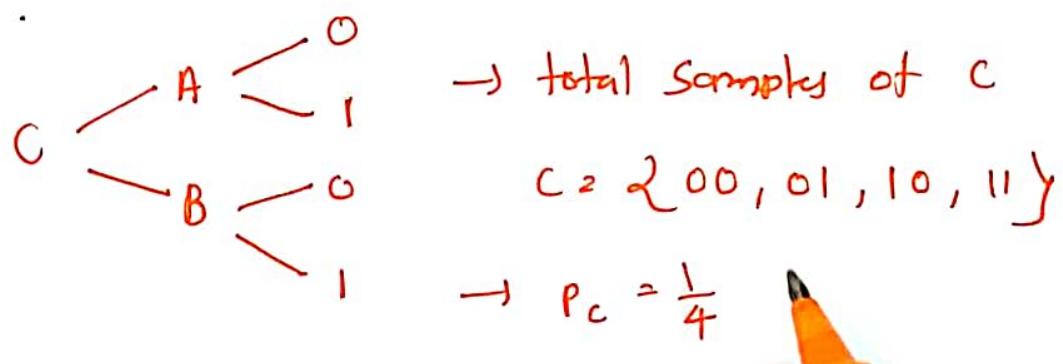
$$\rightarrow I = \log_2\left(\frac{1}{P}\right)$$

$$= \log_2(26)$$

$$= \frac{\lg 26}{\lg 2}$$

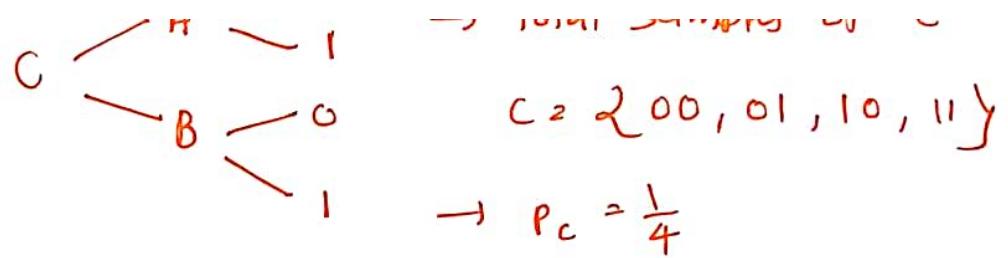
$$= 4.7 \text{ bits.}$$

③ Consider discrete memoryless source 'C' that o/p two bits at a time. This source comprises two binary sources 'A' and 'B', whose o/p are equally likely to occur and each source contributing one bit. Suppose that the sources within the source 'C' are independent. What is the information content of each o/p from source 'C'.



□

$$\begin{aligned} \rightarrow P_A &= 1/2 \\ P_B &= 1/2 \end{aligned}$$



→ Combined Probability of A & B

$$\begin{aligned} P_C &= P_A P_B \\ &= \frac{1}{2} \times \frac{1}{2} \\ &= \frac{1}{4} \end{aligned}$$

→ Information of C

$$I_C = \log_2 \left( \frac{1}{P_C} \right)$$



→ Information of C

$$I_c = \log_2 \left( \frac{1}{P_c} \right)$$

$$= \log_2 (4)$$

$$= 2 \log_2 2$$

$$\boxed{I_c = 2 \text{ bits}}$$

## Entropy basis, definition & property

Definition - It is Average Information of Symbols.

- If we have  $M = \{x_1, x_2, \dots, x_n\}$  messages with probability  $P = \{P_1, P_2, \dots, P_n\}$ .
- then Information of each messages

$$I_1 = \log_2 \left( \frac{1}{P_1} \right), I_2 = \log_2 \left( \frac{1}{P_2} \right), \dots, I_n = \log_2 \left( \frac{1}{P_n} \right)$$

- So Entropy  $H = \frac{\text{Total Information}}{\text{No of messages}}$

$$= \frac{I_1 + I_2 + \dots + I_n}{n}$$

$$= \frac{I_1 + I_2 + \dots + I_n}{n}$$

$$H = \sum_{i=1}^n p_i \lg_2 \left( \frac{1}{p_i} \right)$$

bits/  
Symbol

### Properties

i) Entropy is zero, if the event is sure.

$$- P = 0$$

$$, P = 1$$

$$H = \sum_{k=1}^m p_k \lg_2 \left( \frac{1}{p_k} \right)$$

$$H = \sum_{k=1}^m p_k \lg_2 \left( \frac{1}{p_k} \right)$$

$$\boxed{H = 0}$$

$$= 1 \lg_2 (1)$$

$$\boxed{H = 0}$$

□

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$$H = \sum_{k=1}^m p_k \log_2 \left( \frac{1}{p_k} \right)$$

$$\boxed{H = 0}$$

$$H = \sum_{k=1}^m p_k \log_2 \left( \frac{1}{p_k} \right)$$

$$= 1 \log_2 (1)$$

$$\boxed{H = 0}$$

2) When  $p_k = 1/m$  for all 'm' symbols, then symbols are equally. So  $H = \log_2 m$

$$\rightarrow p_k = 1/m$$

$$\rightarrow H = \sum_{k=1}^m p_k \log_2 \left( \frac{1}{p_k} \right)$$

$$= \boxed{\sum_{k=1}^m \left( \frac{1}{m} \right)} \log_2 m$$

$$\rightarrow \boxed{H = \log_2 m} = \underline{H_{\max}}$$

upper bound of entropy

## Entropy Source Efficiency , Redundancy & Information rate.

### - Source Efficiency

$$\eta = \frac{H}{H_{\max}} \quad | \quad H_{\max} = \log_2 M$$

where  $H$  = Calculated Entropy of source

$H_{\max}$  = Max. Entropy

### - Redundancy of source

$$R_e = 1 - \eta$$

### - Information Rate

$n = 11$



where  $H$  = Calculated Entropy of Source  
 $H_{\max}$  = Max. Entropy

- Redundancy of Source

$$R_c = 1 - \eta$$

- Information Rate

$$R = \gamma H = \left[ \frac{\text{messages}}{\text{sec}} \right] \left[ \frac{\text{bits}}{\text{message}} \right] = \frac{\text{bits}}{\text{sec}}.$$

where,  $\gamma$  = rate at which messages are generated  
[messages/sec]

$$H = \text{Entropy} \left[ \frac{\text{bits}}{\text{messages}} \right]$$

Q) For a discrete memoryless source there are three symbols with  $P_1 = \alpha$  and  $P_2 = P_3$ . Find the entropy of the source

$$\rightarrow P_1 = \alpha$$

$$P_2 = P_3$$

$$\rightarrow P_1 + P_2 + P_3 = 1$$

$$\Rightarrow \alpha + P_2 + P_2 = 1$$

$$\Rightarrow \alpha + 2P_2 = 1$$

$$\Rightarrow P_2 = P_3 = \frac{1-\alpha}{2}$$

$$\rightarrow H = \sum_{k=1}^3 P_k \log_2 \left( \frac{1}{P_k} \right)$$

$$H = \alpha \log_2 \left( \frac{1}{\alpha} \right) + \frac{1-\alpha}{2} \log_2 \left( \frac{2}{1-\alpha} \right)$$

$$= \alpha \log_2 \left( \frac{1}{\alpha} \right) + (1-\alpha) \log_2 \left( \frac{2}{1-\alpha} \right)$$

2] Show that the Entropy of the source with following probability distribution is  $\left[ 2 - \frac{2+n}{2^n} \right] = \frac{1}{2} + \frac{2}{2^2} + \frac{3}{2^3} + \dots + \frac{n}{2^n}$

✓

S	$s_1$	$s_2$	$s_3$	....	$s_n$
P	$\frac{1}{2}$	$\frac{1}{4}$	$\frac{1}{8}$	....	$\frac{1}{2^n}$

$$\begin{aligned} \rightarrow H &= \sum_{k=1}^n p_k \lg_2 \left( \frac{1}{p_k} \right) \\ &= p_1 \lg_2 \left( \frac{1}{p_1} \right) + p_2 \lg_2 \left( \frac{1}{p_2} \right) + p_3 \lg_2 \left( \frac{1}{p_3} \right) + \dots + p_n \lg_2 \left( \frac{1}{p_n} \right) \end{aligned}$$

$$r \quad |'2|'4|'8|\cdots|'2^n|$$

$$\begin{aligned}
\rightarrow H &= \sum_{k=1}^n p_k \lg_2 \left( \frac{1}{p_k} \right) \\
&= p_1 \lg_2 \left( \frac{1}{p_1} \right) + p_2 \lg_2 \left( \frac{1}{p_2} \right) + p_3 \lg_2 \left( \frac{1}{p_3} \right) + \dots + p_n \lg_2 \left( \frac{1}{p_n} \right) \\
&= \left( \frac{1}{2} \right) \lg_2(2) + \left( \frac{1}{4} \right) \lg_2(4) + \left( \frac{1}{8} \right) \lg_2(8) + \dots + \left( \frac{1}{2^n} \right) \lg_2(2^n) \\
&= \frac{1}{2} + \frac{2}{4} + \frac{3}{8} + \dots + \frac{n}{2^n} \\
&= 2 - \frac{2+n}{2^n}
\end{aligned}$$

3) The source emits three messages with probabilities

$$P_1 = 0.7, P_2 = 0.2 \text{ and } P_3 = 0.1$$

Calculate,

$$H = \sum_{k=1}^3 P_k \lg_2 \left( \frac{1}{P_k} \right)$$

1) Source Entropy

$$= 0.7 \lg_2 \left( \frac{1}{0.7} \right) + 0.2 \lg_2 \left( \frac{1}{0.2} \right) + 0.1 \lg_2 \left( \frac{1}{0.1} \right)$$

2) Maximum Entropy

$$= 1.1568 \text{ bits/messages}$$

3) Source Efficiency

$$H_{\max} = \log_2 m$$

$$= \log_2 3 = \frac{\log 3}{\log 2} = 1.585 \frac{\text{bits}}{\text{message}}$$



Calculator,

- 1) Source Entropy
- $H = \sum_{k=1}^3 p_k \lg_2 \left( \frac{1}{p_k} \right)$
- =  $0.7 \lg_2 \left( \frac{1}{0.7} \right) + 0.2 \lg_2 \left( \frac{1}{0.2} \right) + 0.1 \lg_2 \left( \frac{1}{0.1} \right)$
- =  $1.1568 \text{ bits/messages}$
- 2) Maximum Entropy
- $H_{\max} = \log_2 m$
- =  $\log_2 3 = \frac{\log 3}{\log 2} = 1.585 \frac{\text{bits}}{\text{message}}$
- 3) Source Efficiency
- $\eta = \frac{H}{H_{\max}} = \frac{1.1568}{1.585} = 0.73$
- 4) Redundancy
- $R_e = 1 - \eta = 1 - 0.73 = 0.27$

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4) A discrete source emits one of six symbols once every msec. The symbol probability are  $\frac{1}{2}$ ,  $\frac{1}{4}$ ,  $\frac{1}{8}$ ,  $\frac{1}{16}$ ,  $\frac{1}{32}$  &  $\frac{1}{32}$ . Find the source Entropy and Information rate.

$$- \gamma = \frac{1}{T} = \frac{1}{10^{-3}} = 10^3 \text{ messages/sec}$$

$$R = \gamma H$$

$$- H = \sum_{k=1}^6 P_k \lg_2 \left( \frac{1}{P_k} \right)$$

$$\begin{aligned} &= \frac{1}{2} \lg_2(2) + \frac{1}{4} \lg_2(4) + \frac{1}{8} \lg_2(8) + \frac{1}{16} \lg_2(16) + 2 \times \frac{1}{32} \times \lg_2^{32} \\ &= \frac{1}{2} + \frac{2}{4} + \frac{3}{8} + \frac{4}{16} + 2 \times \frac{1}{32} \times 5 \\ &= 1.9375 \text{ bits/sec} \end{aligned}$$

- $\gamma = \frac{1}{T} = \frac{1}{10^{-3}} = 10^3$  messages/sec
- $H = \sum_{k=1}^6 p_k \log_2 \left( \frac{1}{p_k} \right)$ 

$$= \frac{1}{2} \log_2(2) + \frac{1}{4} \log_2(4) + \frac{1}{8} \log_2(8) + \frac{1}{16} \log_2(16) + 2 \times \frac{1}{32} \times \log_2 32$$

$$= \frac{1}{2} + \frac{2}{4} + \frac{3}{8} + \frac{4}{16} + 2 \times \frac{1}{32} \times 5$$

$$= 1.9375 \text{ bits/sec}$$
- $R = \gamma H$ 

$$= 10^3 \times 1.9375$$

$$= 1937.5 \text{ bits/sec}$$

END

## Shannon Fano Encoding Algorithm

1. The messages are first written in the order of decreasing probability.
2. Then divide the messages set into two most equiprobable subset X and Y.
3. The message of 1st set X is given bit 0 and message in the 2nd subset is given bit 1.
4. The procedure is now applied for each set separately till end.
5. Finally we get the code word for respective symbol.
6. Calculation

probability.

2. Then divide the messages set into two most equiprobable subset X and Y.
3. The message of 1st set X is given bit 0 and message in the 2nd subset is given bit 1.
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5. Finally we get the code word for respective symbol.
6. Calculation

$$\rightarrow \text{efficiency } (\eta) = \frac{H}{\hat{H}}$$

where,  $H = \text{Entropy} = \sum_{i=1}^n p_i \log_2 \left( \frac{1}{p_i} \right)$

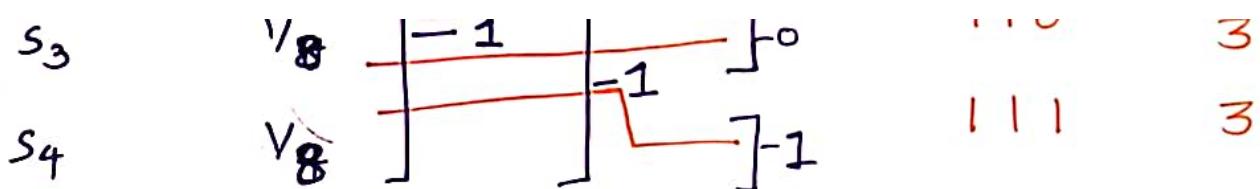
$$\hat{H} = \sum_{i=1}^n p_i n_i \quad | \rightarrow \text{Redundancy}$$

$$R_e = 1 - \eta$$

Example - Find the code words occurring in the probability  $\left\{ \frac{1}{2}, \frac{1}{4}, \frac{1}{8}, \frac{1}{8} \right\}$  for symbols  $s_1, s_2, s_3$  &  $s_4$ . Find efficiency and redundancy of code.

Symbol	Prob		Codeword	Length
$s_1$	$\frac{1}{2}$	$\begin{bmatrix} * \\ -0 \end{bmatrix}$	0	1
$s_2$	$\frac{1}{4}$	$\begin{bmatrix} 1 \\ -1 \\ -0 \end{bmatrix}$	10	2
$s_3$	$\frac{1}{8}$	$\begin{bmatrix} 1 \\ -1 \\ -1 \\ -0 \end{bmatrix}$	110	3
$s_4$	$\frac{1}{8}$	$\begin{bmatrix} 1 \\ -1 \\ -1 \\ -1 \end{bmatrix}$	111	3

■



- Efficiency  $\eta = \frac{H}{\hat{H}}$

$$\begin{aligned}
 \text{where } H &= \text{entropy} = \sum_{i=1}^4 p_i \log_2 \left( \frac{1}{p_i} \right) \\
 &= \frac{1}{2} \log_2 2 + \frac{1}{4} \log_2 4 + \\
 &\quad \left( \frac{1}{8} \log_2 8 \right) \times 2 \\
 &= \frac{1}{2} + \frac{1}{2} + \frac{3}{4} \\
 &= 1.75 \text{ bits / symbol}
 \end{aligned}$$

$$\left(\frac{1}{8} \log_2 8\right) \times 2$$

$$= \frac{1}{2} + \frac{1}{2} + \frac{3}{4}$$

$$= 1.75 \text{ bits / symbol}$$

$$\hat{H} = \sum_{i=1}^4 p_i n_i$$

$$= \frac{1}{2} \times 1 + \frac{1}{4} \times 2 + \frac{1}{8} \times 3 \times 2$$

$$= 1.75 \text{ bits / symbol}$$

- Redundancy  $R_e = 1 - \eta$

$$= 1 - 1 = 0$$

□

Example Find the code word for the probability  
 $\frac{1}{4}, \frac{1}{4}, \frac{1}{8}, \frac{1}{8}, \frac{1}{16}, \frac{1}{16}, \frac{1}{16}, \frac{1}{16}$  for symbols  $s_1, s_2, \dots, s_8$ .  
Find the code efficiency and Redundancy.

Symbol Prob.

$s_1$

$\frac{1}{4}, \frac{1}{4}, \frac{1}{8}, \frac{1}{8}, \frac{1}{16}, \frac{1}{16}, \frac{1}{16}, \frac{1}{16}$  for symbols  $s_1, s_2, \dots, s_8$ .

Find the code efficiency and Redundancy.

Symbol	Prob.	Code word	Length
$s_1$	$\frac{1}{4}$	00	2
$s_2$	$\frac{1}{4}$	01	2
$s_3$	$\frac{1}{8}$	100	3
$s_4$	$\frac{1}{8}$	101	3
$s_5$	$\frac{1}{16}$	1100	4
$s_6$	$\frac{1}{16}$	1101	4
$s_7$	$\frac{1}{16}$	1110	4
$s_8$	$\frac{1}{16}$	1111	4

$$\begin{aligned}
 H &= \sum p_i \log_2 \left( \frac{1}{p_i} \right) \\
 &= 2 \left( \frac{1}{4} \log_2 (4) \right) + 2 \left( \frac{1}{8} \log_2 (8) \right) + 4 \left( \frac{1}{16} \log_2 16 \right) \\
 &= 1 + \frac{3}{4} + 1 \\
 &= 2.75 \text{ bits / symbol}
 \end{aligned}$$

$$\begin{aligned}
 \hat{H} &\approx \sum p_i n_i \\
 &= 2 \left( \frac{1}{4} \times 2 \right) + 2 \left( \frac{1}{8} \times 3 \right) + 4 \left( \frac{1}{16} \times 4 \right) \\
 &= 2.75 \text{ bits / symbol}
 \end{aligned}$$

$$\begin{aligned}
 &= 2\left(\frac{1}{4}\log_2(4)\right) + 2\left(\frac{1}{8}\log_2(8)\right) + 4\left(\frac{1}{16}\log_2 16\right) \\
 &= 1 + \frac{3}{4} + 1 \\
 &\approx 2.75 \text{ bits / symbol}
 \end{aligned}$$

- $\hat{H} = \sum p_i n_i$

$$\begin{aligned}
 &= 2\left(\frac{1}{4} \times 2\right) + 2\left(\frac{1}{8} \times 3\right) + 4\left(\frac{1}{16} \times 4\right) \\
 &\approx 2.75 \text{ bits / symbol}
 \end{aligned}$$

- efficiency  $\eta = \frac{H}{A} = \frac{2.75}{2.75} = 1$

- Redundancy  $R_c = 1 - \eta = 1 - 1 = 0$

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Example - Apply the Shannon-Fano Coding procedure for the message  $X = [x_1, x_2, x_3, x_4, x_5, x_6, x_7]$

Probability  $P = [0.4, 0.2, 0.12, 0.08, 0.08, 0.08, 0.04]$

message Prob

$x_1$  0.4

$x_2$  0.2

$x_3$  0.12

$x_4$  0.08

$x_5$  0.08

$x_6$  0.08

$x_7$  0.04



message	prob		codeword	length
$x_1$	0.4	0	0	1
$x_2$	0.2	0	100	3
$x_3$	0.12	1	101	3
$x_4$	0.08	0	1100	4
$x_5$	0.08	1	1101	4
$x_6$	0.08	0	1110	4
$x_7$	0.04	1	1111	4

- $\{x_1\} \subset \{x_2, x_3, \dots, x_7\}$
- 0.4      0.6
- $\{x_1, x_2\} \subset \{x_3, x_4, \dots, x_7\}$
- 0.6      0.4

message      Prob

$x_1$	0.4	0
$x_2$	0.2	0
$x_3$	0.12	1
$x_4$	0.08	0
$x_5$	0.08	1
$x_6$	0.08	0
$x_7$	0.04	1

- $\{x_1\} \{x_2, x_3, \dots, x_7\}$
- 0.4      0.6
- $\{x_1, x_2\} \{x_3, x_4, \dots, x_7\}$
- 0.6      0.4
- ...

codeword	length
0	1
100	3
101	3
1100	4
1101	4
1110	4
1111	4

$$H = \frac{1}{A}$$

$$\begin{aligned}
 \hat{H} &= \sum p_i n_i \\
 &= 0.4 \times 1 + 0.2 \times 3 + 0.12 \times 3 + \\
 &\quad 3 \times 0.08 \times 4 + 0.04 \times 4 \\
 &= 2.48 \text{ bits / symbol}
 \end{aligned}$$

□

message	Prob	codeword	length
$x_1$	0.4	0	2
$x_2$	0.2	1	2
$x_3$	0.12	00	3
$x_4$	0.08	01	3
$x_5$	0.08	100	3
$x_6$	0.08	101	3
$x_7$	0.04	110	3
		1110	4
		1111	4

$$\begin{aligned}
 H &= \sum p_i \log_2 \left( \frac{1}{p_i} \right) \\
 &= 0.4 \log_2 \left( \frac{1}{0.4} \right) + 0.2 \log_2 \left( \frac{1}{0.2} \right) + \\
 &\quad 0.12 \log_2 \left( \frac{1}{0.12} \right) + 3 \times 0.08 \log_2 \left( \frac{1}{0.08} \right) + \\
 &\quad 0.04 \log_2 \left( \frac{1}{0.04} \right)
 \end{aligned}$$

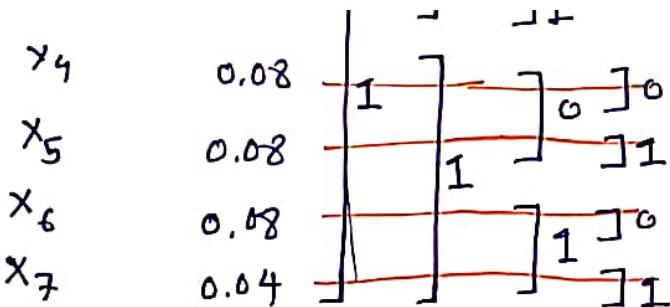
$$\begin{aligned}
 H &= \sum p_i n_i \\
 &= 0.4 \times 2 + 0.2 \times 2 + 0.12 \times 3 + \\
 &\quad 0.08 \times 3 \times 2 + 0.08 \times 4 + 0.04 \times \\
 &\quad 4 \times 3 \\
 &= 2.52 \text{ bits/symbol} \quad \times
 \end{aligned}$$

□

message	Prob	Codeword	length
$x_1$	0.4	00	2
$x_2$	0.2	01	2
$x_3$	0.12	100	3
$x_4$	0.08	101	3
$x_5$	0.08	110	3
$x_6$	0.08	1110	4
$x_7$	0.04	1111	4

$$\begin{aligned}
 H &= \sum p_i \log_2(p_i) \\
 &= 0.4 \log_2(0.4) + 0.2 \log_2(0.2) + \\
 &\quad 0.12 \log_2(0.12) + 3 \times 0.08 \log_2(0.08) + \\
 &\quad 0.04 \log_2(0.04) = 2.42 \text{ bits/symbol}
 \end{aligned}$$

$$\begin{aligned}
 H &= \sum p_i n_i \\
 &= 0.4 \times 2 + 0.2 \times 2 + 0.12 \times 3 + \\
 &\quad 0.08 \times 3 \times 2 + 0.08 \times 4 + 0.04 \times \\
 &\quad 4 = 2.52 \text{ bits/symbol} \quad \square
 \end{aligned}$$



codeword length  
1 0 0 1 4  
1 1 0 1 4  
1 1 1 0 4  
1 1 1 1 4

-  $\{x_1\} \subset \{x_2, x_3, \dots, x_7\}$

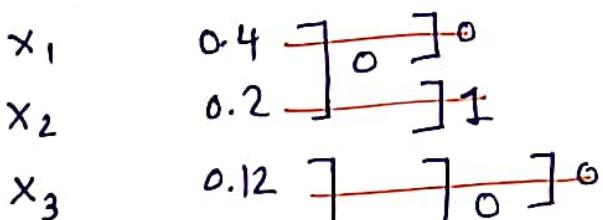
$$0.4 \quad 0.6$$

-  $\{x_1, x_2\} \subset \{x_3, x_4, \dots, x_7\}$

$$0.6 \quad 0.4$$

$$\begin{aligned} H &= \frac{H}{A} \\ &= \frac{2.42}{2.48} \\ &= [97.67] \end{aligned}$$

message Prob



codeword length

0 0	2
0 1	2
1 0 0	3
... ,	2

$$\begin{aligned} \hat{H} &= \sum P_i n_i \\ &= 0.4 \times 1 + 0.2 \times 3 + 0.12 \times 3 + \\ &\quad 3 \times 0.08 \times 4 + 0.04 \times 4 \\ &= 2.48 \text{ bits/symbol} \quad \checkmark \end{aligned}$$

## Huffman Coding

1. The source symbols are arranged in order of decreasing probability.  
Then the two of lowest probability are assigned bit 0 and 1.
2. Then combine last two symbols and move the combined symbol as high as possible.
3. Repeat the above step until end.
4. Code for each symbol is found by moving backward.
5. Calculation

$$\text{efficiency } \eta = \frac{H}{L \log_2 8}$$

$$L = \text{Avg Codeword} = \sum_{i=1}^n p_i n_i$$

(n = 1, 2, 3, 4, 5)

Then the two or lowest probability are assigned bit 0 and 1.

2. Then combine last two symbols and move the combined symbol as high as possible.
3. Repeat the above step until end.
4. Code for each symbol is found by moving backward.
5. Calculation

$$\text{efficiency } \eta = \frac{H}{L \log_2 \delta} \rightarrow \begin{array}{l} \text{binary } \delta=2 (0,1) \\ \text{ternary } \delta=3 (0,1,2) \\ \text{Quaternary } \delta=4 (0,1,2,3) \end{array}$$

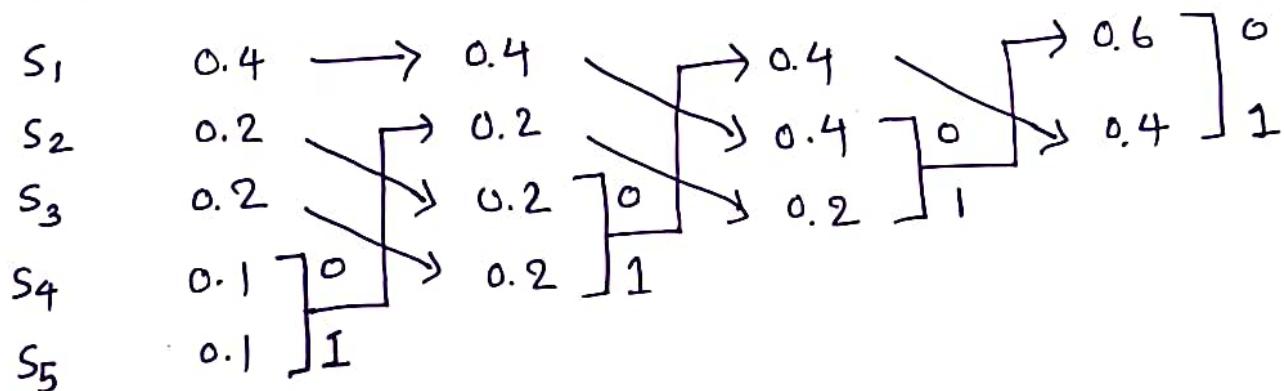
$$L = \text{Avg Codeword} = \sum_{i=1}^n p_i n_i$$

$$H = \sum_{i=1}^n p_i \log_2 \left( \frac{1}{p_i} \right)$$

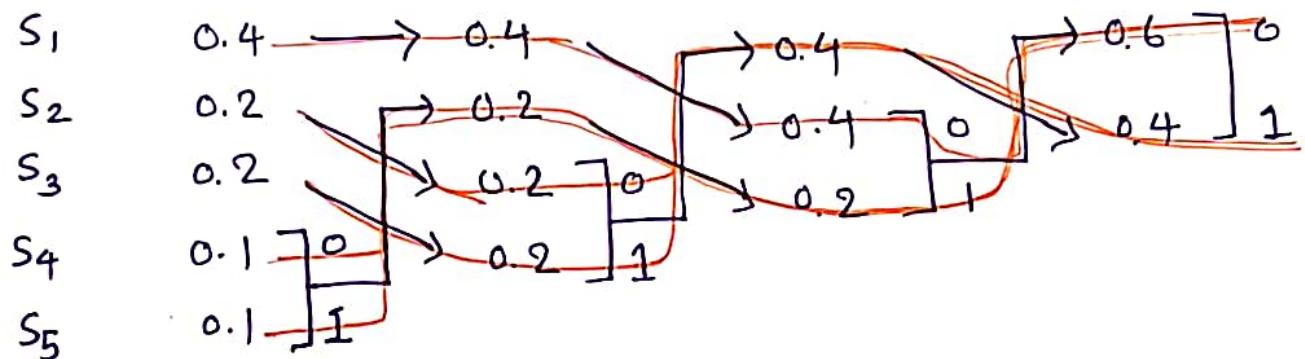
$$\text{Variance } \sigma^2 = \sum_{i=1}^n p_i (n_i - L)^2$$

Example : Alphabet with prob = { 0.4, 0.2, 0.2, 0.1, 0.1 }  
 For symbols  $\{ s_1, s_2, \dots, s_5 \}$ . Find Huffman codes  
 and also find efficiency & variance

Symbol      Prob



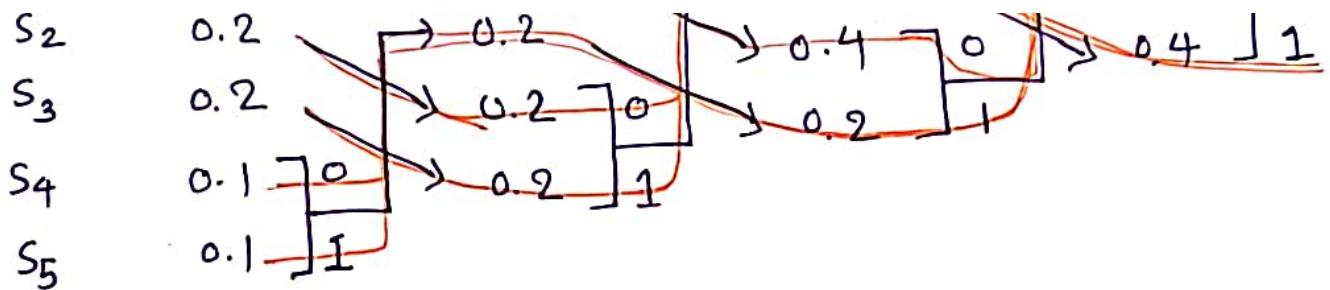
E



Symbol      Codeword      length

Symbol	Codeword	length
$s_1$	00	2
$s_2$	10	2
$s_3$	11	2
$s_4$	010	3
$s_5$	011	3





Symbol	Codeword	length
$S_1$	00	2
$S_2$	10	2
$S_3$	11	2
$S_4$	010	3
$S_5$	011	3

- Entropy

$$\begin{aligned}
 H &= \sum p_i \log_2 \left( \frac{1}{p_i} \right) \\
 &= 0.4 \log_2 \left( \frac{1}{0.4} \right) + \\
 &\quad 2 \times 0.2 \log_2 \left( \frac{1}{0.2} \right) + \\
 &\quad 2 \times 0.1 \log_2 \left( \frac{1}{0.1} \right) \\
 &= 2.1216 \text{ bits / symbol}
 \end{aligned}$$

■

$$S_5 \quad 011 \quad 3 \quad 2 \times 0.2 \log_2\left(\frac{1}{0.2}\right) + \\ 2 \times 0.1 \log_2\left(\frac{1}{0.1}\right)$$

$$- L = \sum p_i n_i \\ = 2 \times 0.4 + 2(2 \times 0.2) + 3 \times 0.1 \times 2 \\ = 2.2 \text{ bits / symbol}$$

$$- \eta^2 = \frac{H}{L \log_2 8} = \frac{2.1216}{2.2 \times \log_2 2} = 96.4 \%$$

$$- \sigma^2 = \sum p_i (n_i - L)^2 \\ = 0.4 (2 - 2.2)^2 + 2 \times 0.2 (2 - 2.2)^2 + 2 \times 0.1 (3 - 2.2)^2 \\ = 0.16$$

↑ As low as possible.

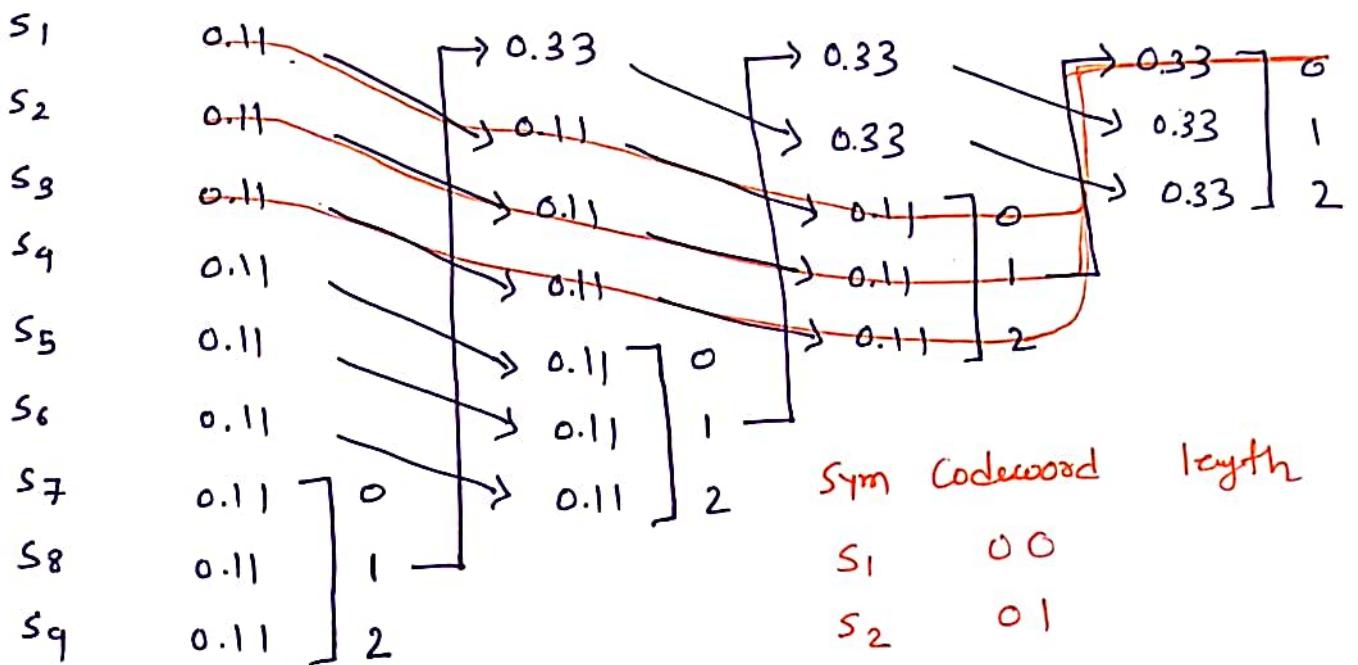
■

Huffman Coding procedure and example [ternary code]

- A source produces  $q$  symbols  $(s_1, s_2, \dots, s_q)$ . construct bi ternary & Quaternary Huffman Coding by moving symbols as high as possible. Also find efficiency & variance of the coding. Probability of symbols is given by

$$P_i = \{0.11, 0.11, 0.11, 0.11, 0.11, 0.11, 0.11, 0.11, 0.11\}$$

Symbols      Prob

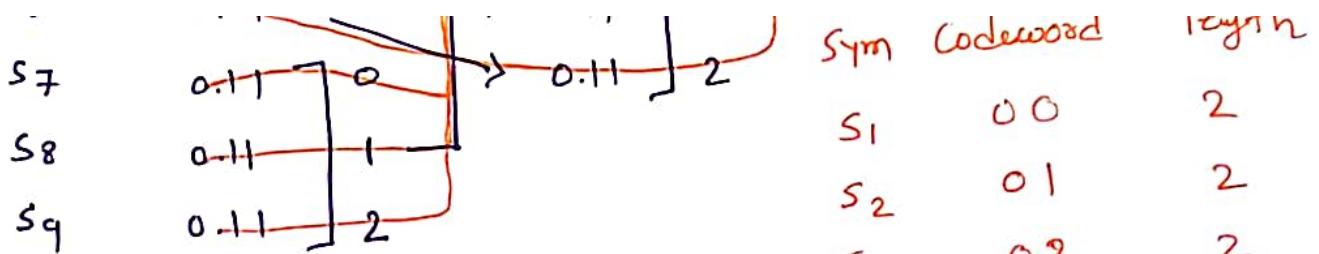


Sym Codeword Length

$s_1$	00
$s_2$	01
$s_3$	0
$s_4$	-



E



$$\begin{aligned}
 H &= \sum p_i \log_2 \left( \frac{1}{p_i} \right) \\
 &= 9 \left( 0.11 \log_2 \left( \frac{1}{0.11} \right) \right) \\
 &= 3.1525 \text{ bits / symbol}
 \end{aligned}$$

$$\begin{aligned}
 L &= \sum p_i n_i \\
 &\approx 9 (0.11 \times 2) \\
 &= 1.98 \text{ bits / symbol}
 \end{aligned}$$

Sym	Codeword	$\log_2 n$
$s_1$	00	2
$s_2$	01	2
$s_3$	02	2
$s_4$	10	2
$s_5$	11	2
$s_6$	12	2
$s_7$	20	2
$s_8$	21	2
$s_9$	22	2

E-1

$$\begin{aligned}
 H &= \sum p_i \log_2 \left( \frac{1}{p_i} \right) \\
 &= 9 (0.11 \log_2 \left( \frac{1}{0.11} \right)) \\
 &= 3.1525 \text{ bits / symbol}
 \end{aligned}$$

$$\begin{aligned}
 L &= \sum p_i n_i \\
 &\Rightarrow 9 (0.11 \times 2) \\
 &= 1.98 \text{ bits / symbol}
 \end{aligned}$$

$$\eta = \frac{H}{L \lg_2 3}$$

$$= \frac{3.1525}{1.98 \times \lg_2 3} = \frac{3.1525 \times \lg 2}{1.98 \times \lg 3} = 100\%$$

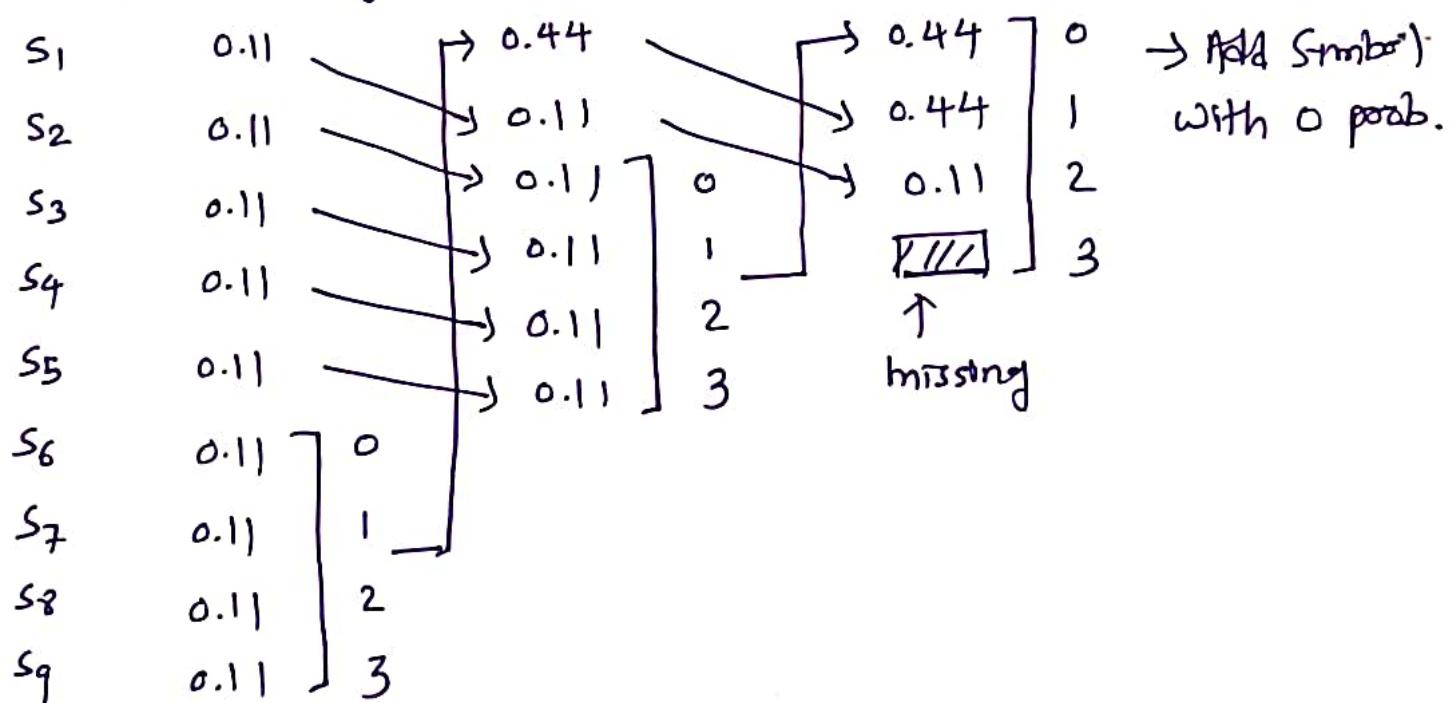
$s_3$	02	2
$s_4$	10	2
$s_5$	11	2
$s_6$	12	2
$s_7$	20	2
$s_8$	21	2
$s_9$	22	2

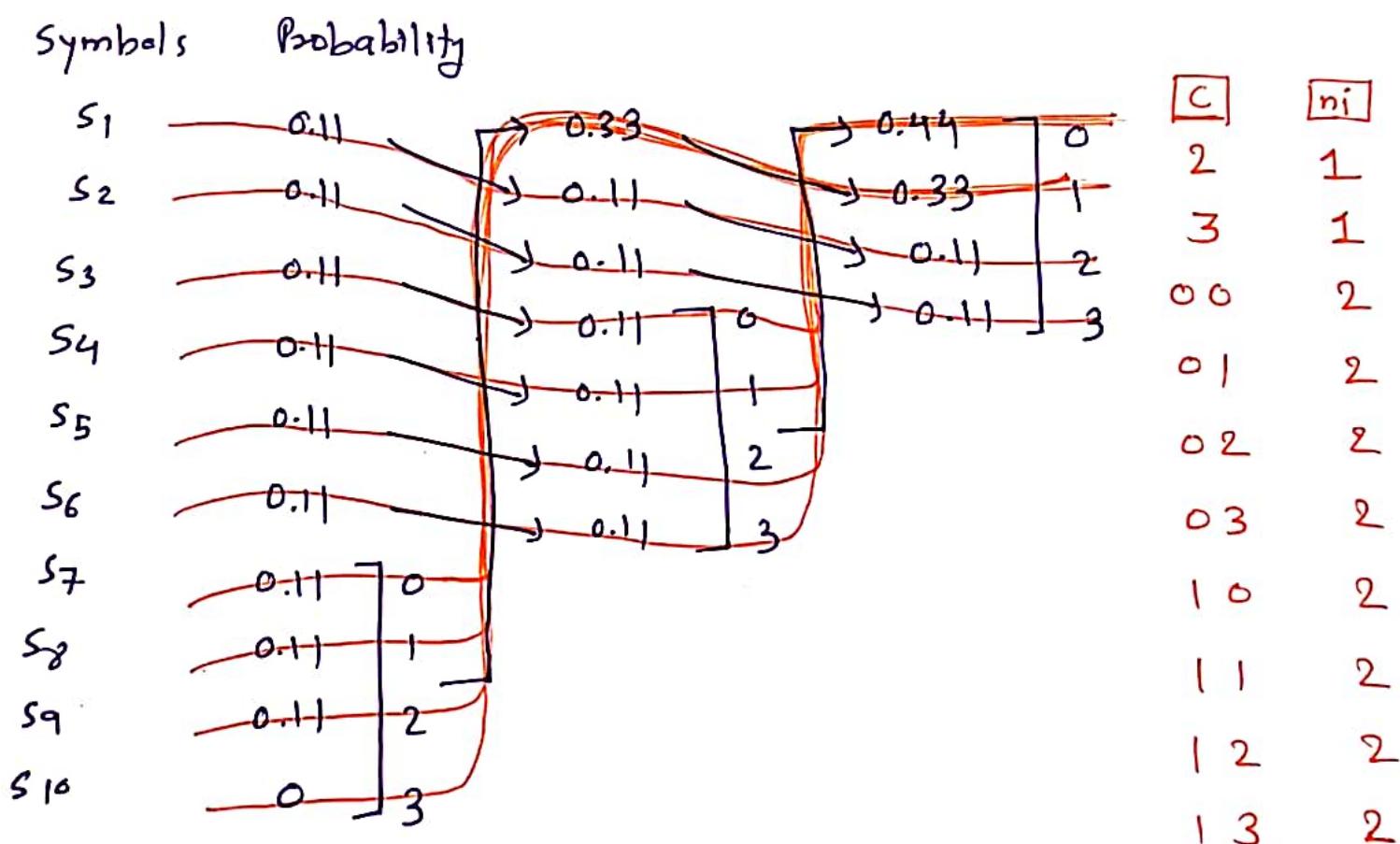
$$\begin{aligned}
 \sigma^2 &= \sum p_i (n_i - L)^2 \\
 &= 9 \times 0.11 \times (1.98 - 2)^2
 \end{aligned}$$

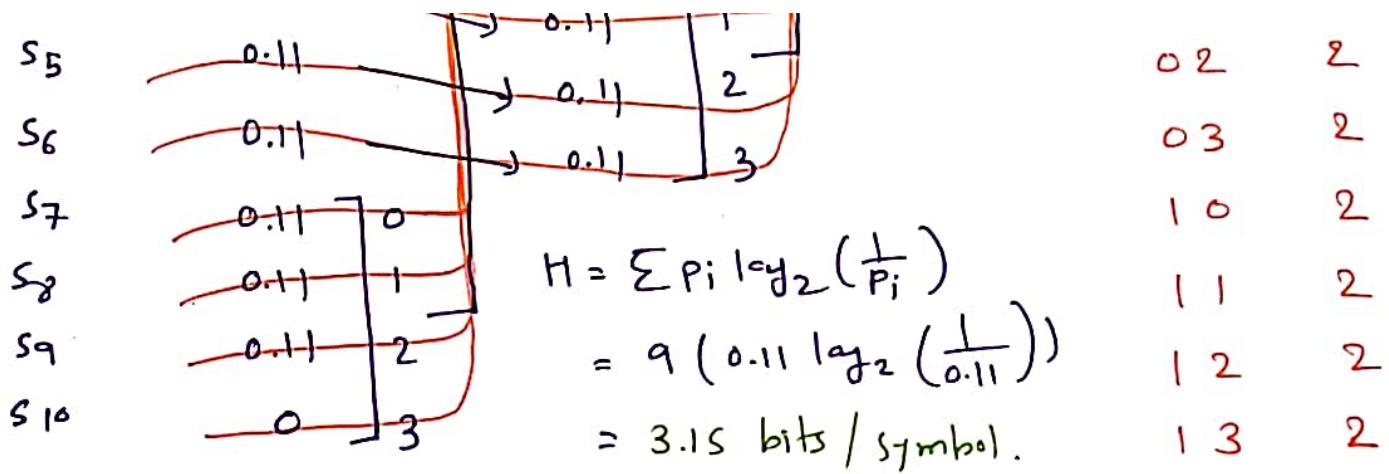
■

## Huffman Coding Procedure and example [Quaternary Code]

Symbols Probability







$$\begin{aligned}
 \rightarrow L &= \sum p_i n_i \\
 &= 2(0.11 \times 1) + 0.11 \times 2 \times 2 \\
 &= 1.76 \text{ bits / symbol}
 \end{aligned}$$

$$\rightarrow \eta^2 = \frac{H}{L \log_2 8} = \frac{3.15}{1.76 \log_2 4} = 0.8948$$

$$\begin{aligned}
 \eta^2 &= \sum p_i (n_i - L)^2 \\
 &= \boxed{\quad}
 \end{aligned}$$

■

# Sources of noise

1. Natural noise sources
2. Man made noise sources
3. Fundamental sources (internal sources)

## Natural source of noise

- # the natural phenomena that give rise to noise are Electronic storms, solar flares and radiation in space
- # the noise originating from the Sun and the outer space ,  
Is known as extraterrestrial noise.
- # the extra Terrestrial noise is subdivided into two groups .
  - 1. Solar noise
  - 2. Cosmic noise

# our sun radiates lots of noise.

- # the temperature changes follow cycle 11 years
- # cosmic noise comes from stars
- # it is identical to noise by sun
- # it is uniformly distributed over entire Sky
- # we also receives noise from centre of our Galaxy and  
From other galaxies such as Quasars and Pulsars

## Man made noise

# it is generated because of make and break process in  
a current carrying circuit.

# Examples

- × electrical Motors,
- × welding machines,
- × ignition system of automobiles,
- × fluorescent lights,
- × switching gears

## Fundamental sources

- # it is within the electronic equipment
- # they are called fundamental sources because they are integral part of material and electrical component.
- # this type of noise follows certain rules.
- # hence it can be eliminated by designing properties

## Classifications of noise

- ✗ thermal noise
- ✗ partition noise
- ✗ shot noise
- ✗ flicker noise or low frequency noise
- ✗ transient noise or high frequency noise

## Shot Noise

- # it is produced due to shot effect
- # it is produced in amplifying devices rather than in all.
- # it is produced because of random variation of electrons or holes (charge carrier)
- # it has uniform power spectral density like thermal noise
- # the exact formula for shot noise only can be calculated for diodes.

$$\overline{I_n^2} = 2 \left( \frac{1}{g_w} + 2 I_0 \right) q \beta \frac{1}{T}$$

$\overline{I_n^2}$        $\frac{1}{g_w}$        $I_0$        $q \beta \frac{1}{T}$

Shot noise      Direct current      Reverse sat. current      Charge

- # For amplifying devices, shot noise is
  - × directly proportional to output current
  - × inversely proportional to transconductance of device

## Partition noise

# partition noise is generated when current gets divided into two or more parts.

# show partition noise is higher in transistor than the diode.

# the device is like gallium arsenide FET draw almost zero gate bias current and therefore, keeping the partition noise to its minimum value.



## Low frequency of flicker noise

- # it will appear at frequency is below few kilohertz
- # it is also referred as  $1/f$  noise
- # it is generated because of fluctuation in current density
- # this will change conductivity of material
- # this will produce fluctuation in voltage and current
- # the mean square value of flicker noise is directly proportional to the square of direct current flowing through the device

## Transient noise or high frequency noise

- # if the time taken by electron from emitter to collector becomes comparable to period of signal then the transit time effect takes place.
- # this effect can be observed at very high frequency.
- # due to this some carriers may diffuse back to emitter.
- # this gives rise to input impedance.
- # so minut change at input generates random fluctuations at output.
- # once the noise appears, it goes on increasing with frequency at a rate of 6dB per octave



## Thermal noise/ Johnson noise/ white noise/ resistor noise

- # it arises due to random motion of free charge particles.
- # mainly electrons in conducting media.
- # the intensity of random motion is proportional to thermal (heat energy) supplied. E.g. temperature.
- # the net motion of the electrons gives rise to an electric current to flow through the resistor causing the noise.
- # calculation

$$\text{Power } P = kTB$$

$$\text{Voltage } V = \sqrt{4kTB R}$$

$$\text{Current } I = \sqrt{\frac{4k}{R} kTB}$$

## Signal to Noise ratio (SNR)

- It is defined as a ratio of signal power to noise power.

$$\frac{S}{N} = \frac{P_S}{P_N}$$

where,  $P_S$  = Signal Power

$P_N$  = Noise Power.

- In dB,

$$\left( \frac{S}{N} \right)_{dB} = 10 \log \left( \frac{P_S}{P_N} \right)$$

- In dB,

$$\boxed{\left( \frac{S}{N} \right)_{dB} = 10 \log \left( \frac{P_S}{P_N} \right)}$$

- Power in terms of voltage.

$$P_S = V_S^2 / R , \quad P_N = V_N^2 / R$$

- So SNR will be.

$$\begin{aligned} \left( \frac{S}{N} \right)_{dB} &= 10 \log \left( \frac{V_S^2 / R}{V_N^2 / R} \right) \\ &= 10 \log \left( \frac{V_S}{V_N} \right)^2 \end{aligned}$$

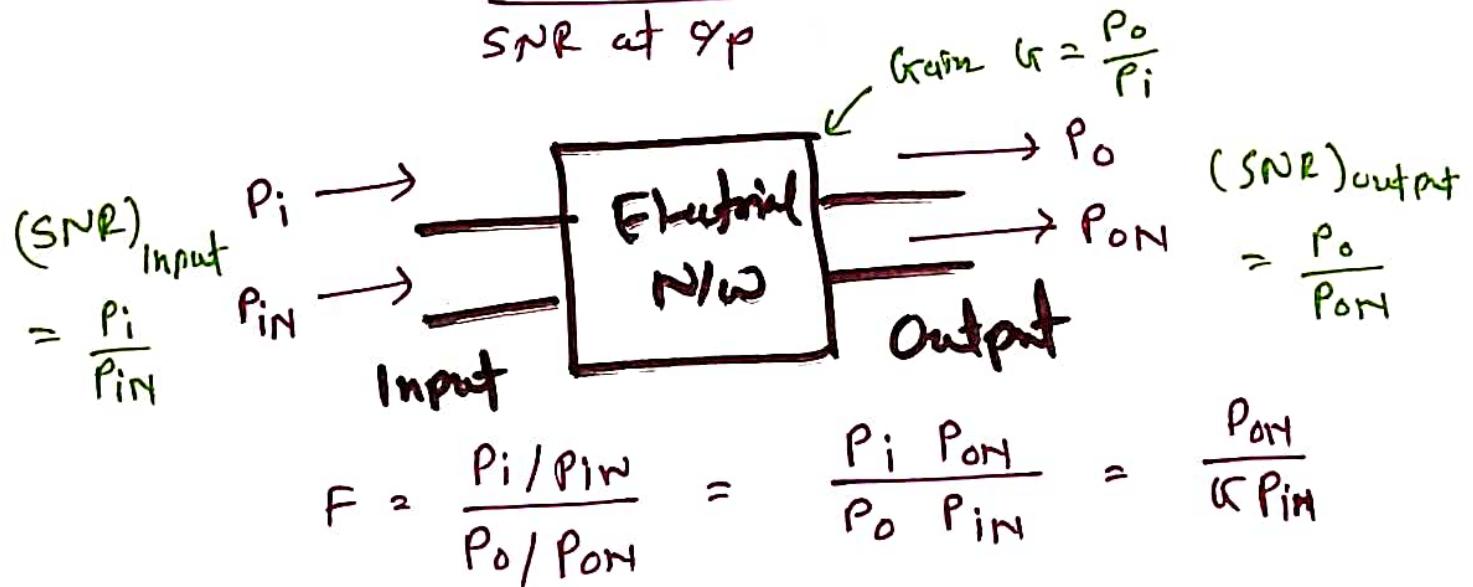
$$\boxed{\left( \frac{S}{N} \right)_{dB} = 20 \log \left( \frac{V_S}{V_N} \right)}$$

- High SNR is good for Tx and Rx

## Noise Figure

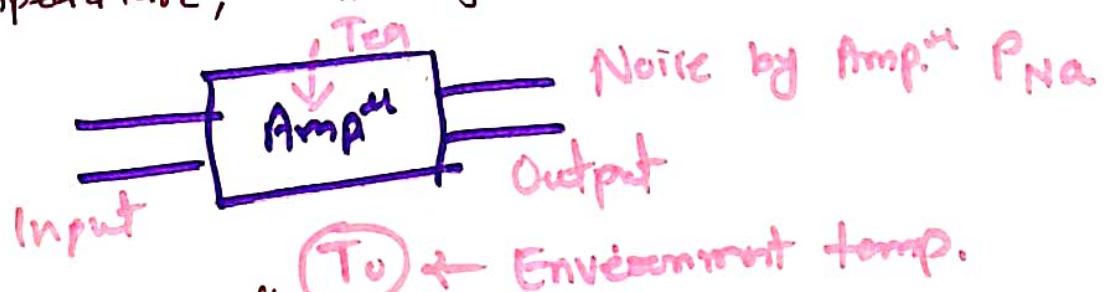
- It is a ratio of SNR at Input to SNR at Output

$$F = \frac{\text{SNR at IP}}{\text{SNR at OP}}$$



## Noise Temperature

- It is temperature, which generates noise power in system.



- Noise power by comp<sup>tr</sup>.

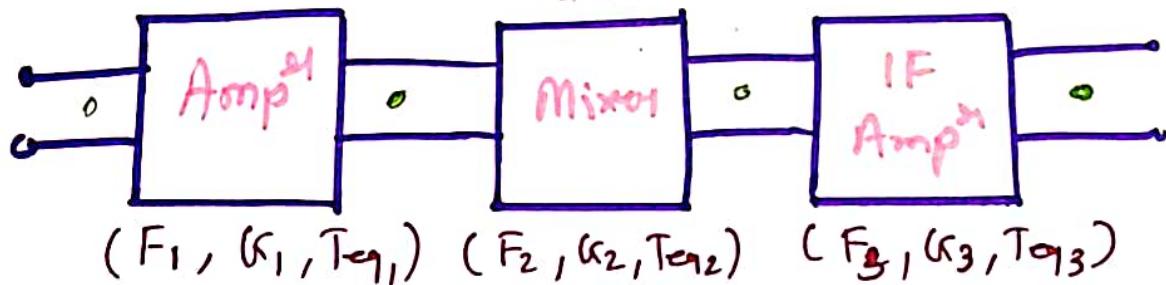
$$\Rightarrow P_{NA} = (F-1) k T_0 B$$

- Noise power power in terms of noise temp.  $T_{eq}$ .

$$\Rightarrow \underline{k T_{eq} B} = (F-1) \underline{k T_0 B}$$

$$\Rightarrow \boxed{T_{eq} = (F-1) T_0} \quad \checkmark$$

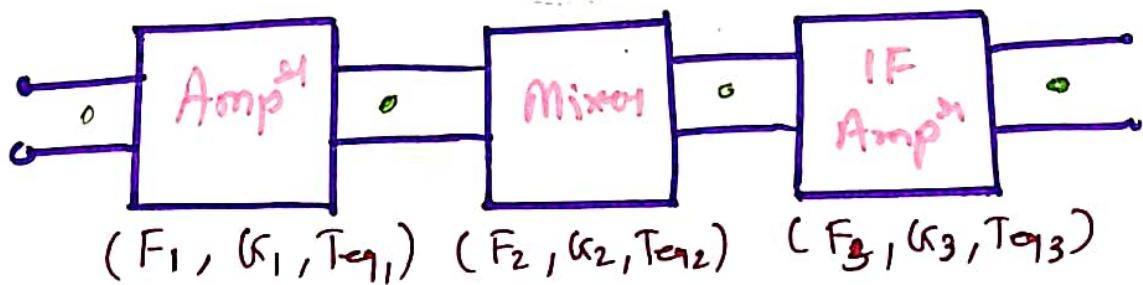
## Equivalent Noise Temp. and Noise Figure in Cascade Comm. System



→ Equivalent Noise Figure.

$$F = F_1 + \frac{(F_2 - 1)}{K_1} + \frac{(F_3 - 1)}{K_1 K_2}$$

→ Do not take  
 $F_1, F_2$  &  $F_3$   
 in terms of dB.



→ Equivalent Noise Figure.

$$F = \underline{F_1} + \frac{(F_2 - 1)}{K_1} + \frac{(F_3 - 1)}{K_1 K_2}$$

→ Do not take  
 $F_1, F_2$  &  $F_3$   
 in terms of dB.

→ Noise temp.  $T_{eq} = T_0(F - 1)$

$$\rightarrow T_{eq} = T_{eq1} + \frac{T_{eq2}}{K_1} + \frac{T_{eq3}}{K_1 K_2}$$

Two resistors 20k and 50k are at room temperature 290K. Determine for the bandwidth of 100kHz, the thermal noise for the conditions:

I. For each resistor

II. For two resistors in series

III. For two resistors in parallel

$$\rightarrow R_1 = 20 \text{ k}\Omega$$

$$R_2 = 50 \text{ k}\Omega$$

$$T = 290 \text{ K}$$

$$B = 100 \text{ kHz}$$

$$\rightarrow R_s = R_1 + R_2 = 70 \text{ k}\Omega$$

$$\rightarrow R_p = \frac{R_1 R_2}{R_1 + R_2} = 14.28 \text{ k}\Omega$$

$$\begin{aligned} &\text{For } 20 \text{ k}\Omega \\ V_n &\propto \sqrt{4KTBR} \\ &= \sqrt{4 \times 1.38 \times 10^{-23} \times} \\ &\quad 290 \times 20 \times 10^3 \\ &\quad 100 \times 10^3 \\ &= 5.66 \text{ mV} \end{aligned}$$

$$\rightarrow \text{For } 50 \text{ k}\Omega$$

$$V_n \propto \sqrt{4KTBR} = ?$$

$$\rightarrow \text{For Series } 70 \text{ k}\Omega$$

$$V_n = \sqrt{4KTBR_s} = ?$$

$$\rightarrow \text{For Parallel } 14.28 \text{ k}\Omega$$

Calculate the thermal noise power available from any resistor at room temperature 290K for a bandwidth 2 MHz. Also calculate the corresponding noise voltage given that R=100.

$$\rightarrow T = 290 \text{ K}$$

$$B = 2 \times 10^6 \text{ Hz}$$

$$R = 100 \Omega$$

$$k = 1.38 \times 10^{-23} \text{ J/K}$$

$$\rightarrow \underline{\underline{V_{rms}}} = \sqrt{4kTB}$$

$$\rightarrow \boxed{P = kTB}$$

$\rightarrow$  Thermal Noise power

$$\begin{aligned} P &= kTB \\ &= 1.38 \times 10^{-23} \times 290 \times \\ &\quad 2 \times 10^6 \\ &= 8 \times 10^{-15} \text{ W} \end{aligned}$$

$$\Rightarrow P = V^2 / R$$

$$\rightarrow V = \sqrt{PR}$$

$$\rightarrow \sqrt{8 \times 10^{-15} \times 100} = \boxed{0.896 \text{ mV}}$$

An amplifier has a bandwidth of 4 MHz with 10k as the input resistor. Calculate The RMS noise voltage at the input to this amplifier if the room temperature is 25°C.

$$- B = 4 \text{ MHz}$$

$$R = 10 \text{ k}\Omega$$

$$T = 25^\circ\text{C} = 25 + 273$$

$$= 298 \text{ K}$$

$$K = 1.38 \times 10^{-23} \text{ J/K}$$

$$\Rightarrow V_n = \sqrt{4 k T B R}$$

$$\Rightarrow V_n = \sqrt{4 \times 1.38 \times 10^{-23} \times 298 \times 4 \times 10^6 \times 10 \times 10^3}$$

$$\Rightarrow V_n = 25.64 \text{ nV}$$

The signal power and noise power measured at input of an amplifier are 150 mW and 1.5 μW, respectively. If the signal power at the output 1.5 W and noise power is 40mW, calculate amplifier noise factor and noise figure.

$$\rightarrow P_{Si} = 150 \text{ mW} \\ P_{Ni} = 1.5 \mu\text{W} \quad \left. \begin{array}{l} (S/N)_i \\ \end{array} \right\} = \frac{P_{Si}}{P_{Ni}} = \frac{150 \text{ mW}}{1.5 \mu\text{W}} = 100$$

$$P_{So} = 1.5 \text{ W} \\ P_{No} = 40 \text{ mW} \quad \left. \begin{array}{l} (S/N)_o \\ \end{array} \right\} = \frac{P_{So}}{P_{No}} = \frac{1.5 \text{ W}}{40 \times 10^{-3} \text{ W}} = 37.5$$

$$\rightarrow F = \frac{(S/N)_i}{(S/N)_o} = \frac{100}{37.5} = 2.66$$

$$\rightarrow (F)_{dB} = 10 \log F = 10 \log 2.66 = 4.26 \text{ dB}$$



The signal to noise ratio at the input of amplifier is 40dB.  
If the noise figure of an amplifier is 20dB, calculate the  
Signal to noise ratio at the amplifier output.

$$\rightarrow (S/N)_i \text{ (dB)} = 40 \text{ dB}$$

$$(F) \text{ (dB)} = 20 \text{ dB}$$

$$\Rightarrow (F) \text{ (dB)} = (S/N)_i \text{ (dB)} - (S/N)_o \text{ (dB)}$$

$$\Rightarrow (S/N)_o \text{ (dB)} = (S/N)_i \text{ (dB)} - (F) \text{ (dB)}$$
$$= 40 - 20$$

$$= \boxed{20 \text{ dB}}$$

An amplifier has noise figure of 3dB. Determine its equivalent noise temperature.

$$\rightarrow (F)_{dB} = 3 dB$$

$$\Rightarrow (F)_{dB} = 10 \log F$$

$$\Rightarrow F = \text{Antilog} \left( \frac{(F)_{dB}}{10} \right)$$

, Antilog (0.3)

$$\Rightarrow \boxed{F = 2}$$

$$\rightarrow \boxed{T_{eq} = (F-1)T_0}$$

$$\rightarrow \text{Room temp}$$

300 K.

$$\rightarrow T_{eq} = (2-1)300$$

$$\Rightarrow \boxed{T_{eq} = 300 \text{ K}}$$

Radio receiver with 10kHz bandwidth has a noise figure of 30dB. Determine the signal power required at input of receiver to achieve input SNR of 30dB.

$$\rightarrow B = 10 \text{ kHz}$$

$$F_{dB} = 30 \text{ dB} \Rightarrow F = 1000$$

$$(S/N)_i = 30 \text{ dB} \Rightarrow (S/N)_i = \underline{1000}$$

$$\rightarrow P_{Ni} = F k T B$$

$$= 1000 \times 1.38 \times 10^{-23} \times 300 \times 10 \times 10^3$$

$$= 4.14 \times 10^{-4} \text{ W}$$

$$\rightarrow (S/N)_i = \frac{P_{Si}}{P_{Ni}} \Rightarrow P_{Si} = (S/N)_i P_{Ni}$$

$$= 1000 \times 4.14 \times 10^{-14}$$

$$= 4.14 \times 10^{-11} \text{ W} = 41.4 \mu\text{W}$$

If each stage has gain of 10 dB and noise figure of 10dB. Calculate the overall noise figure of two stage amplifier.

$$\rightarrow (K_1)_{dB} = 10 dB$$

$$\Rightarrow K_1 = 10$$

$$\rightarrow (K_2)_{dB} = 10 dB$$

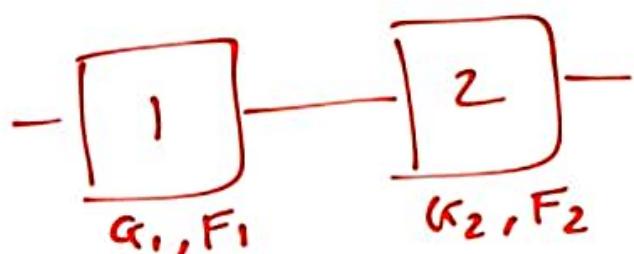
$$\Rightarrow K_2 = 10$$

$$\rightarrow (F_1)_{dB} = 10 dB$$

$$\Rightarrow F_1 = 10$$

$$\rightarrow (F_2)_{dB} = 10 dB$$

$$\Rightarrow F_2 = 10$$



$$\rightarrow F = F_1 + \frac{(F_2 - 1)}{K_1}$$

$$= 10 + \left( \frac{10 - 1}{10} \right)$$

$$= 10.9$$

$\rightarrow$  Noise Figure

$$F_{dB} = 10 \log F \Rightarrow 10 \log 10.9 \\ = 10.37 dB$$

Radio receiver with equivalent noise bandwidth of 10kHz  
 Has a noise figure of 20dB. If input SNR to receiver is  
40dB, determine the output SNR. What is the equivalent  
 Noise temperature, if ambient temperature is 300K.

$$\rightarrow B = 10 \text{ kHz}$$

$$F_{dB} = 20 \text{ dB} \Rightarrow F = \frac{100}{10}$$

$$(S/N)_i = 40 \text{ dB} \Rightarrow (S/N)_i = 10^4$$

$$T_0 = 300 \text{ K}$$

$$\rightarrow T_{eq} = (F-1)T_0$$

$$= (100-1) 300$$

$$= 29700 \text{ K}$$

$$\rightarrow F \Rightarrow \frac{(S/N)_i}{(S/N)_o} = (S/N)_o = \frac{(S/N)_i}{F}$$

$$\Rightarrow (S/N)_o = 10^4 / 100$$

$$\rightarrow (S/N)_o = 100 \Rightarrow (S/N)_o^{(dB)} = 20 \text{ dB}$$

An amplifier with 10dB noise figure and 4dB power gain is cascaded with a second amplifier which has a 10dB power gain and 10dB noise figure. What is the overall noise figure and power gain?

$$\rightarrow F_1 = 10 \text{ dB} = 10$$

$$G_1 = 4 \text{ dB} = 2.5$$

$$F_2 = 10 \text{ dB} = 10$$

$$G_2 = 10 \text{ dB} = 10$$

$$\rightarrow F = F_1 + \frac{(F_2 - 1)}{G_1}$$

$$= 10 + \frac{9}{2.5}$$

$$= 13.6 = 11.33 \text{ dB} \rightarrow T_{eq} = (F - 1) T_0$$



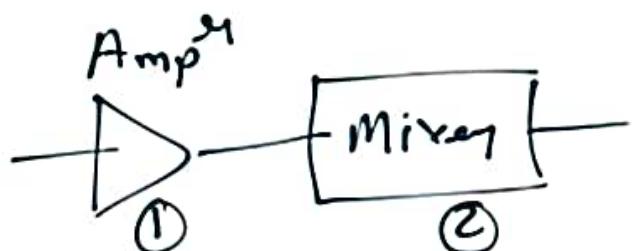
$$\begin{aligned} \rightarrow G &= G_1 G_2 \\ &= 2.5 \times 10 \\ &= 25 \\ &= 13.9 \text{ dB} \end{aligned}$$

A mixer stage has a noise figure of 20dB and it is preceded by an another amplifier with a noise figure of 9dB and an available power gain of 15dB. Calculate noise figure.

$$\rightarrow F_1 = 9 \text{ dB} = 7.94$$

$$G_1 = 15 \text{ dB} = 31.62$$

$$F_2 = 20 \text{ dB} = 100$$

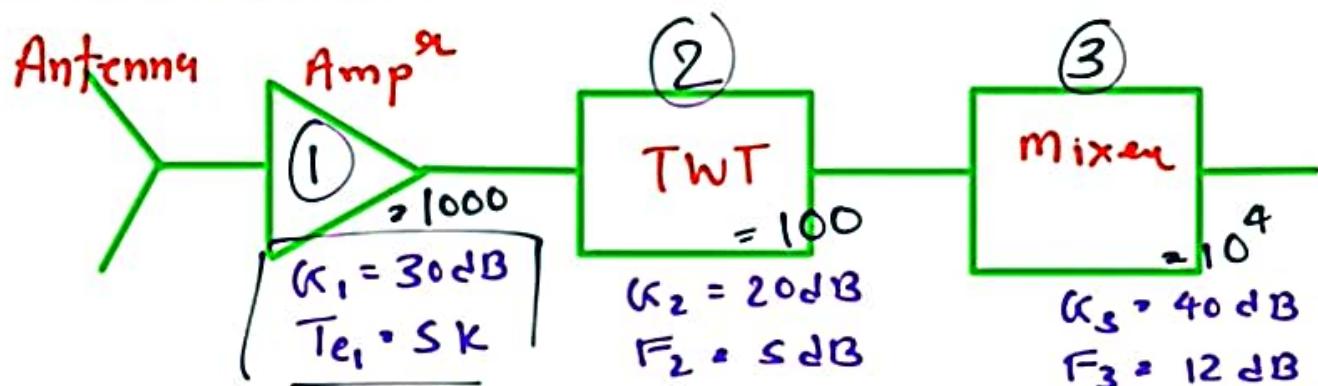


$$\rightarrow F = F_1 + \frac{(F_2 - 1)}{G_1}$$

$$= 7.94 + \frac{99}{31.62}$$

$$= 11.67 = \underline{\underline{10.44 \text{ dB}}}$$

Calculate equivalent noise figure and noise temperature for given diagram.



$$\Rightarrow T_{eq1} = (F_1 - 1) T_0$$

$$\begin{aligned}\Rightarrow F_1 &= 1 + \frac{T_{eq1}}{T_0} \\ &= 1 + \frac{5}{300} \\ &= 1.02\end{aligned}$$

$$\begin{aligned}&\rightarrow F = F_1 + \frac{(F_2 - 1)}{F_1} + \frac{(F_3 - 1)}{F_1 \cdot F_2} \\ &= 1.02 + \frac{(4 - 1)}{1.02} + \frac{16 - 1}{1.02 \times 4} \\ &= \dots \\ \rightarrow T_{eq} &= (F - 1) T_0.\end{aligned}$$

Channel Capacity by Shannon - Hartley and It's proof

$$C = B \log_2 \left[ 1 + \frac{S}{N} \right]$$

where,  $B$  = Bandwidth of channel

$S$  = signal Power

$N$  = Noise power.

E

Proof :- Received Signal = Signal Power (S) + Noise power (N)  
and It's mean Square Value. is  $\sqrt{S+N}$

- Noise power is N and It's mean Square value is  $\sqrt{N}$
- So Number of Levels can be separated without error is

$$m = \frac{\sqrt{N+S}}{\sqrt{N}} > \sqrt{1 + \frac{S}{N}}$$

- So digital Information is

$$\begin{aligned} I &= \log_2 m \\ &= \log_2 \sqrt{1 + \frac{S}{N}} \\ &= \frac{1}{2} \log_2 \left( 1 + \frac{S}{N} \right) \end{aligned}$$

E

$$m = \frac{\sqrt{N+S}}{\sqrt{N}} \Rightarrow \sqrt{1 + \frac{S}{N}}$$

- So digital information is

$$\begin{aligned} I &= \log_2 m \\ &= \log_2 \sqrt{1 + \frac{S}{N}} \\ &= \frac{1}{2} \log_2 \left( 1 + \frac{S}{N} \right) \end{aligned}$$

- If channel transmits  $K$  pulses per second then channel capacity is

$$C = IK = \frac{K}{2} \log_2 \left( 1 + \frac{S}{N} \right)$$

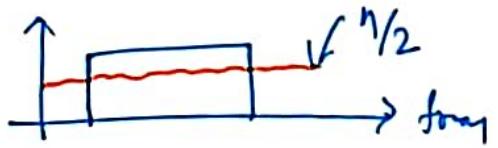
- Nyquist Bandwidth is  $K = 2B$

$$\boxed{C = B \log_2 \left( 1 + \frac{S}{N} \right)}$$

Practically if we increase  $B$  then noise  $N$  will also increase. So capacity of channel can not be infinite.

If  $\eta/2$  is power density then

$$N = \eta B$$



So channel capacity is

$$C = B \log_2 \left( 1 + \frac{S}{\eta B} \right)$$

$$= \frac{\eta B}{S} \left( \frac{S}{\eta} \right) \log_2 \left( 1 + \frac{S}{\eta B} \right)$$

- So Channel capacity is

$$\begin{aligned} C &= B \log_2 \left( 1 + \frac{S}{\eta B} \right) \\ &= \frac{\eta B}{S} \left( \frac{S}{\eta} \right) \log_2 \left( 1 + \frac{S}{\eta B} \right) \\ &= \frac{S}{\eta} \left[ \frac{\log_2 \left( 1 + \frac{S}{\eta B} \right)}{\frac{S}{\eta B}} \right] \quad \underline{B \rightarrow \infty} \end{aligned}$$

- For  $\lim_{S \rightarrow 0} \frac{\log_2 (1+S)}{(1+S)} = \log_2 e = 1.44$

$$C = \frac{S}{\eta} \log_2 e = \boxed{1.44 \frac{S}{\eta}}$$

## Examples on Channel Capacity by Shannon-Heath

① For a typical telephone line with a signal to noise ratio of 30 dB and an audio bandwidth 3 kHz, max. data rate

$$\rightarrow \text{SNR} = 30 \text{ dB} = 10^{\frac{3}{2}}$$

dB  
 of ...  
 10 → 10  
 20 → 10<sup>2</sup>  
 30 → 10<sup>3</sup>  
 40 → 10<sup>4</sup>

$B = 3 \text{ kHz}$

$C = ?$

$$\begin{aligned}\rightarrow C &= B \log_2 (1 + \text{SNR}) \\ &= 3 \times 10^3 \log_2 (1 + 10^3) \\ &= 3 \times 10^3 \frac{\log 1001}{\log 2} = 3 \times 10^6 \text{ bps} = \boxed{3 \text{ Mbps}}\end{aligned}$$

→ to noise ratio  $\square$

② For a satellite TV channel with a signal to noise ratio of 20 dB and a video bandwidth of 10 MHz, find max. data rate?

$$\rightarrow \text{SNR} = 20 \text{ dB} = 10^2$$

$$B = 10 \text{ MHz} = 10 \times 10^6 \text{ Hz}$$

$$C = ?$$

$$\rightarrow C = B \log_2 (1 + \text{SNR})$$

$$\Rightarrow C = 10 \times 10^6 \log_2 (1 + 10^2)$$

$$= 10^7 \frac{\log(101)}{\log 2}$$

$$= 6.6 \times 10^7 \text{ bps}$$

$$\approx 66 \text{ Mbps}$$

③ A Gaussian channel has 1 MHz bandwidth. calculate the channel capacity. If the signal power to noise spectral density ratio S/N is  $10^5$  Hz. Also find the max. information rate.

E

③ A Gaussian channel has 1 MHz bandwidth. calculate the channel capacity. If the signal power to noise spectral density ratio S/N is  $10^5$  Hz. Also find the max. Information rate.

$$\begin{array}{l}
 - B = 1 \text{ MHz} = 10^6 \text{ Hz} \\
 \text{SNR} = 10^5 \\
 C = ?
 \end{array}
 \quad \left| \quad \begin{aligned}
 \Rightarrow C &= B \log_2 (1 + \text{SNR}) \\
 &= 10^6 \log \frac{100001}{1 + 2} \\
 &= 16.6 \times 10^6 \text{ bps} \\
 \boxed{C = 16.6 \text{ Mbps}}
 \end{aligned}
 \right.$$

\* Calculate channel capacity with  $B \rightarrow \infty$ .

- Channel capacity

$$C = B \log_2 \left( 1 + \frac{S}{N} \right)$$

- If we have noise power  $N = \eta B$  where  $\frac{\eta}{2}$  is power density.

Power density .

$$\begin{aligned} C &= B \log_2 \left( 1 + \frac{s}{nB} \right) \\ &= \frac{nB}{s} \times \frac{s}{n} \log_2 \left( 1 + \frac{s}{nB} \right) \\ &= \frac{s}{n} \left[ \frac{nB}{s} \log_2 \left( 1 + \frac{s}{nB} \right) \right] \end{aligned}$$

- Here  $B \rightarrow \infty$ .

$$C = \frac{s}{n} \lim_{B \rightarrow \infty} \left[ \frac{\log_2 \left( 1 + \frac{s}{nB} \right)}{\left( s/nB \right)} \right]$$

E

$$- \frac{S}{\eta} \left\lfloor \frac{\eta_B}{S} \log_2 \left( 1 + \frac{S}{\eta_B} \right) \right\rfloor$$

- Here  $B \rightarrow \infty$ .

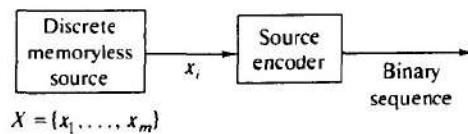
$$C = \frac{S}{\eta} \lim_{B \rightarrow \infty} \left[ \frac{\log_2 \left( 1 + \frac{S}{\eta_B} \right)}{\left( S / \eta_B \right)} \right]$$

$$\Rightarrow C = \frac{S}{\eta} \log_2 e$$

$$\Rightarrow \boxed{C = (1.44) \frac{S}{\eta}}$$

## 10.7 SOURCE CODING

A conversion of the output of a DMS into a sequence of binary symbols (binary code word) is called *source coding*. The device that performs this conversion is called the *source encoder* (Fig. 10-6).



**Fig. 10-6** Source coding

An objective of source coding is to minimize the average bit rate required for representation of the source by reducing the redundancy of the information source.

### A. Code Length and Code Efficiency:

Let  $X$  be a DMS with finite entropy  $H(X)$  and an alphabet  $\{x_1, \dots, x_m\}$  with corresponding probabilities of occurrence  $P(x_i)$  ( $i = 1, \dots, m$ ). Let the binary code word assigned to symbol  $x_i$  by the encoder have length  $n_i$ , measured in bits. The length of a code word is the number of binary digits in the code word. The average code word length  $L$ , per source symbol is given by

$$L = \sum_{i=1}^m P(x_i)n_i \quad (10.49)$$

The parameter  $L$  represents the average number of bits per source symbol used in the source coding process.

The *code efficiency*  $\eta$  is defined as

$$\eta = \frac{L_{\min}}{L} \quad (10.50)$$

where  $L_{\min}$  is the minimum possible value of  $L$ . When  $\eta$  approaches unity, the code is said to be *efficient*.

The *code redundancy*  $\gamma$  is defined as

$$\gamma = 1 - \eta \quad (10.51)$$

### B. Source Coding Theorem:

The source coding theorem states that for a DMS  $X$  with entropy  $H(X)$ , the average code word length  $L$  per symbol is bounded as

$$L \geq H(X) \quad (10.52)$$

and further,  $L$  can be made as close to  $H(X)$  as desired for some suitably chosen code.

Thus, with  $L_{\min} = H(X)$ , the code efficiency can be rewritten as

$$\eta = \frac{H(X)}{L} \quad (10.53)$$