

Gate Assignment 4

Yashas Tadikamalla - AI20BTECH11027

Download all python codes from

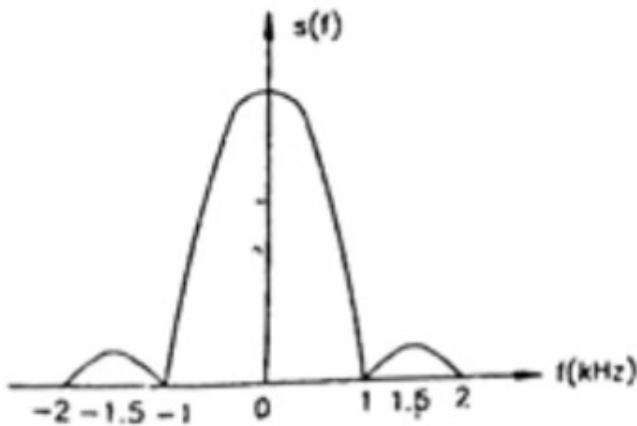
<https://github.com/YashasTadikamalla/EE3900/blob/main/GateAssignment4/codes>

and latex-tikz codes from

<https://github.com/YashasTadikamalla/EE3900/blob/main/GateAssignment4/GateAssignment4.tex>

1 PROBLEM (EC-1997 Q1.10)

A deterministic signal has the power spectrum given in the figure. The minimum sampling rate needed to completely represent this signal is



- 1) 1KHz
- 2) 2KHz
- 3) 3KHz
- 4) None

2 SOLUTION

Definition 2.1 (Normalised sinc function). A normalised sinc function is defined as

$$\text{sinc}(x) = \begin{cases} 1, & x = 0 \\ \frac{\sin(\pi x)}{\pi x}, & x \neq 0 \end{cases} \quad (2.0.1)$$

Definition 2.2 (Power spectrum). Power Spectral density, or simply, Power spectrum, denoted by $s(f)$ is defined as

$$s(f) = |X(f)|^2 \quad (2.0.2)$$

Theorem 2.1 (Sampling Theorem). If a signal contains no frequency components above W Hz, then the sampling rate at which the continuous time signal needs to be sampled uniformly, so as to completely recover the original signal is given by

$$f_s \geq 2W \quad (2.0.3)$$

Definition 2.3 (Nyquist rate). Minimum sampling rate is also called as Nyquist rate. It is given by

$$f_s = 2W \quad (2.0.4)$$

Given, power spectrum of a deterministic signal.

$$s(f) = \begin{cases} \text{sinc}(f)^2, & |f| \leq 2\text{KHz} \\ 0, & \text{else} \end{cases} \quad (2.0.5)$$

From (2.2), Fourier transform of the given band limited signal is **truncated normalised sinc pulse**.

$$X(f) = \begin{cases} \text{sinc}(f), & |f| \leq 2\text{KHz} \\ 0, & \text{else} \end{cases} \quad (2.0.6)$$

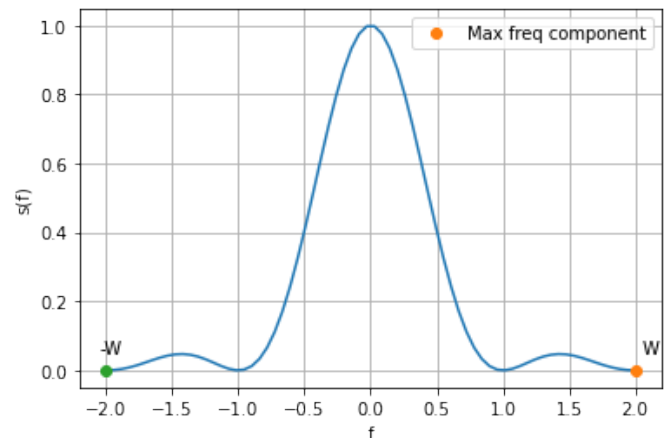
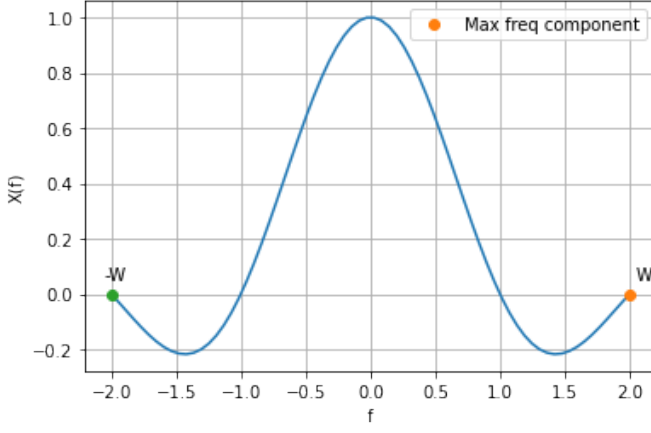


Fig. 4: Plot of $s(f)$

Fig. 4: Plot of $X(f)$

As no frequency component exceeds $2KHz$,

$$W = 2KHz \quad (2.0.7)$$

From (2.0.4),

$$f_s = 2W = 4KHz \quad (2.0.8)$$

Hence, option 4 is the correct answer.

To verify the validity of (2.1), let's see what happens if we sample at a rate lower than Nyquist rate.

Let our original continuous time signal be $x(t)$.

$$x(t) = \int_{-\infty}^{\infty} X(f) e^{2\pi f t} df \quad (2.0.9)$$

Consider impulse train $x_i(t)$ given by

$$x_i(t) = \sum_{n=-\infty}^{\infty} \delta(t - nT_s) \quad (2.0.10)$$

where T_s is the sampling period. ($f_s = \frac{1}{T_s}$ is the sampling frequency). The sampled signal would be

$$x_s(t) = x(t)x_i(t) \quad (2.0.11)$$

$$= x(t) \sum_{n=-\infty}^{\infty} \delta(t - nT_s) \quad (2.0.12)$$

$$= \sum_{n=-\infty}^{\infty} x(nT_s) \delta(t - nT_s) \quad (2.0.13)$$

Also, from (2.0.11)

$$X_s(f) = X(f) * X_i(f) \quad (2.0.14)$$

$$= \frac{1}{T_s} \sum_{k=-\infty}^{\infty} X(f - kf_s) \quad (2.0.15)$$

$X_s(f)$ consists periodically repeated copies of $X(f)$, shifted by integer multiples of f_s . Let W be the maximum frequency component.

- Case 1: $W > 2f_s$:

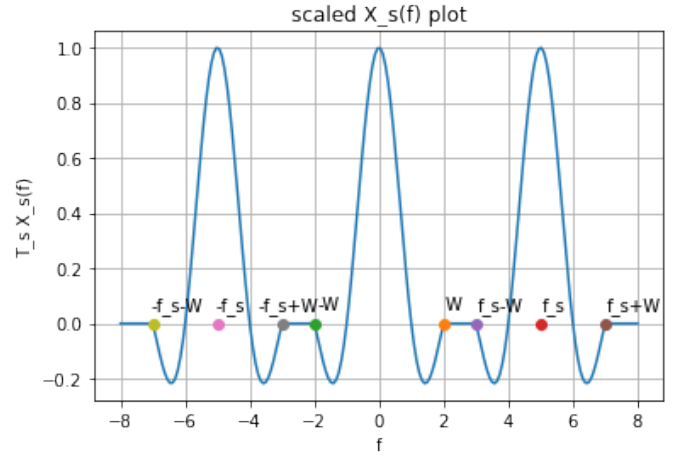
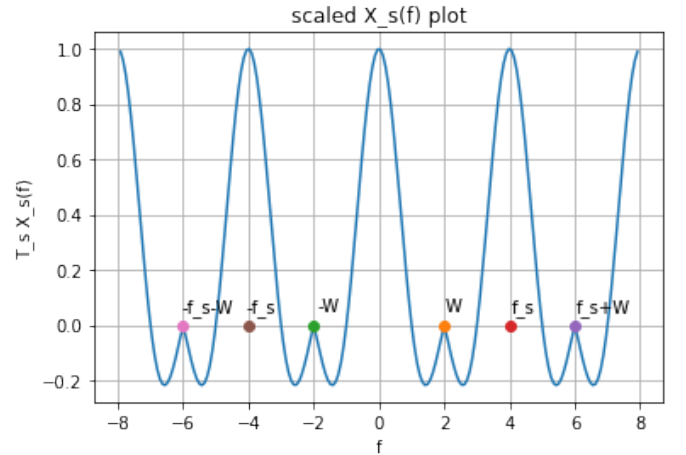
The copies don't overlap. Hence, $x(t)$ can be recovered from $x_s(t)$ using an ideal low pass filter.

- Case 2: $W = 2f_s$:

The copies just touch each other, but don't overlap. So, $x(t)$ can be recovered from $x_s(t)$ using an ideal low pass filter.

- Case 3: $W < 2f_s$:

As the copies overlap, they get added. So, we cannot reconstruct the original signal $x(t)$. This gives rise to situation called aliasing.

Fig. 4: $W > 2f_s$ Fig. 4: $W = 2f_s$

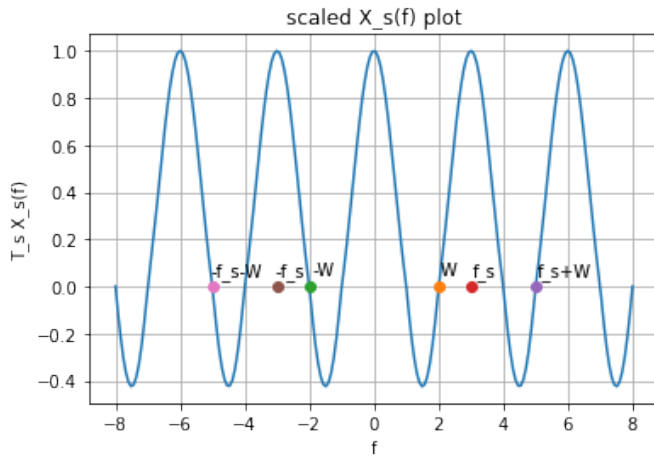


Fig. 4: $W < 2f_s$

Hence, we need to sample at a rate greater than Nyquist rate to be able to recover the original signal.

$$x(t) = \int_{-2}^2 \text{sinc}(f) e^{2\pi f t j} df \quad (2.0.16)$$

We cannot find a closed form expression for $x(t)$, but we can calculate it numerically using python.

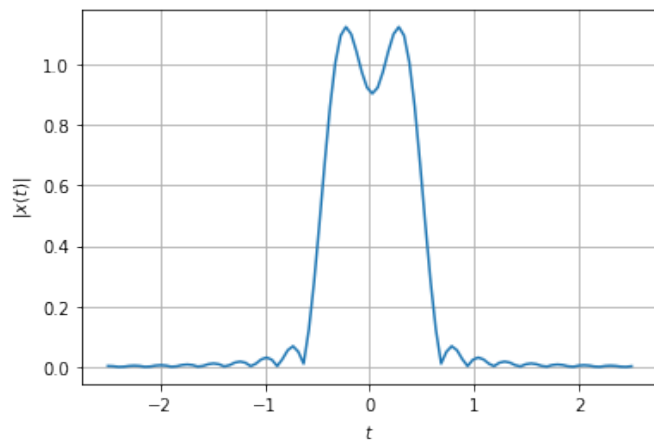


Fig. 4: Plot of $|x(t)|$

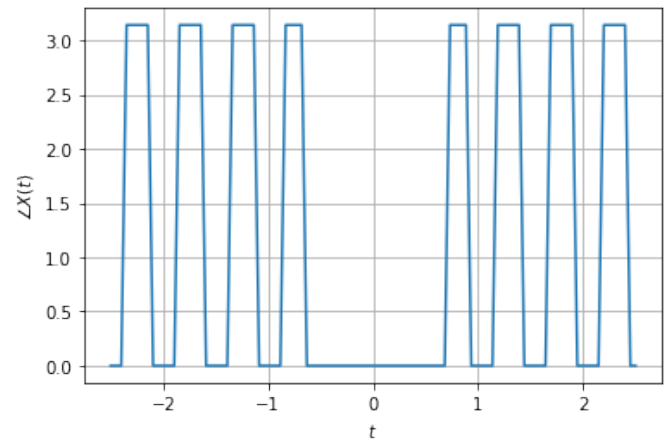


Fig. 4: Plot of $\angle x(t)$