

EE324 CONTROL SYSTEMS LAB

Problem Sheet 5

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Question 1: Root Locus

Part A)

In this part, we are given a closed-loop transfer function connected in a unity negative feedback loop.

$$T_s = \frac{10}{s^3 + 4s^2 + 5s + 10}$$

Now, in general unity negative feedback, we get T_s as

$$T_s = \frac{G_s}{1 + G_s}$$

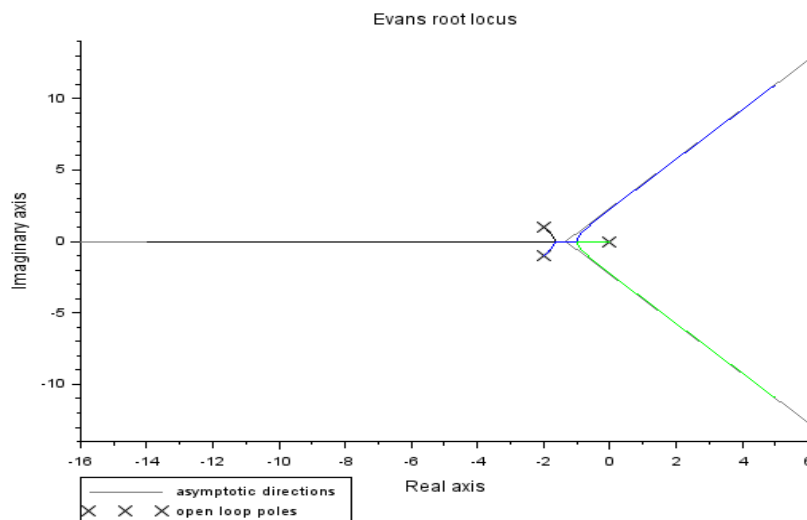
$$G_s = \frac{N}{D}$$

$$T_s = \frac{N}{N + D}$$

From comparing two equations, we get G_s as

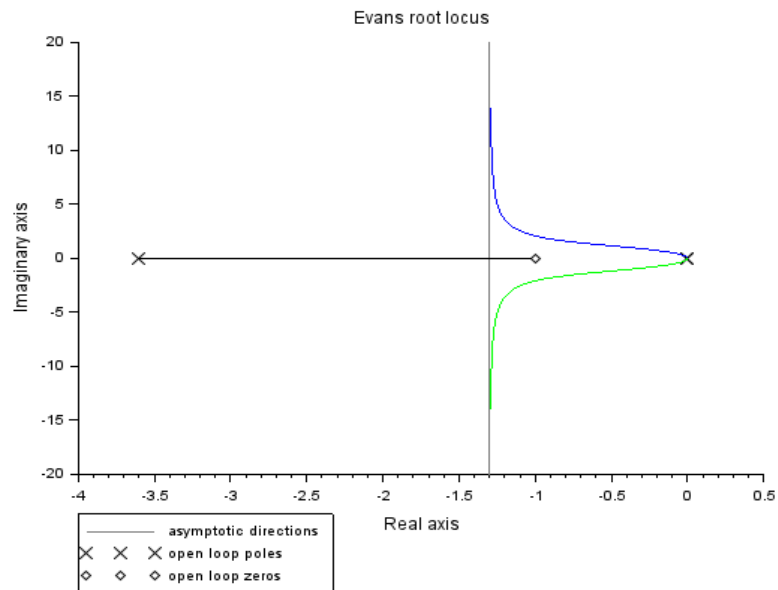
$$G_s = \frac{10}{s(s^2 + 4s + 5)}$$

Now we can get the root locus using Scilab



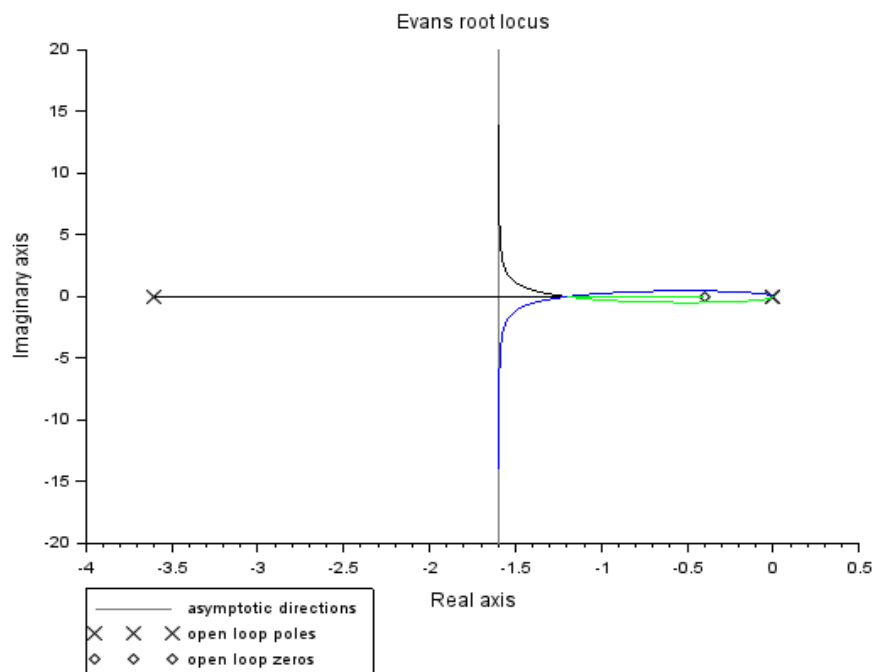
Part B)

$$G_s = \frac{s + 1}{(s^2)(s + 3.6)}$$



Part C)

$$G_s = \frac{s + 0.4}{(s^2)(s + 3.6)}$$



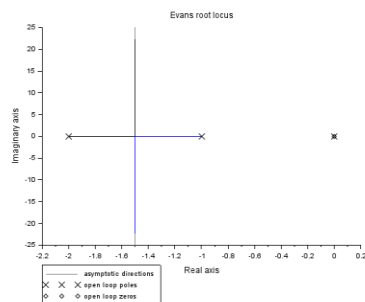
Part D)

$$G_s = \frac{s + p}{s(s + 1)(s + 2)}$$

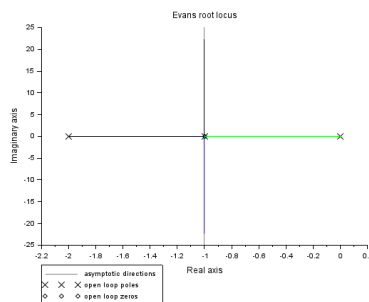
In this part, we are varying-parameter 'p.' Following are the observations for different values of parameter 'p.'

- When $p=0,1,2$:
One of the three poles of the system cancels, and G_s becomes the second-order open-loop transfer function.
This system always remains stable with any value of gain K as poles stay in LHP
- A considerable positive value of p :
This system always remains stable as all poles and zeros are in LHP, so the poles of the closed-loop function will be in LHP for any gain K
- A negative value of p ($p < 0$):
This system will become unstable for $K > K_{\text{marginal}}$ a specific gain $K_{\text{marginal}} > 0$ because one zero of G_s lies in RHP

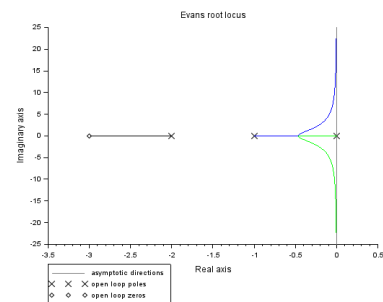
Following are Root Locus For Different values of 'p':



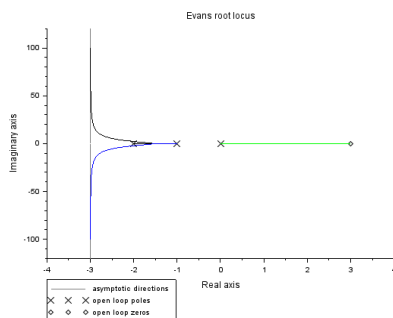
$p=0$



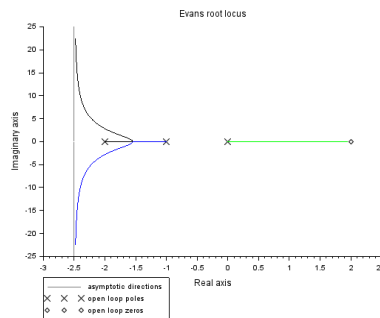
$p=1$



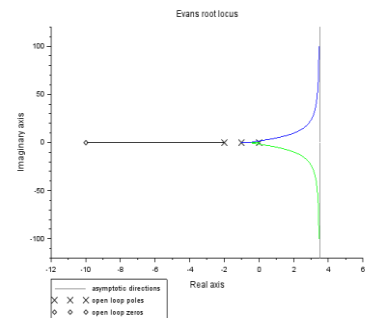
$p=3$



$p=-3$



$p=-2$



$p=10$

Scilab code for the same:

```
s=%s;  
  
// A Part  
  
Gs1=syslin('c',10,s*(s^2+4*s+5));  
evans(Gs1,200);  
show_window(1);  
  
// B Part  
  
Gs2=syslin('c',s+1,s^2*(s+3.6));  
evans(Gs2,200);  
show_window(2);  
  
// C Part  
  
Gs3=syslin('c',s+0.4,s^2*(s+3.6));  
evans(Gs3,200);  
show_window(3);  
  
// D Part  
  
Gs4=syslin('c',s+10,s*(s+1)*(s+2));  
evans(Gs4,500);  
show_window(4);
```

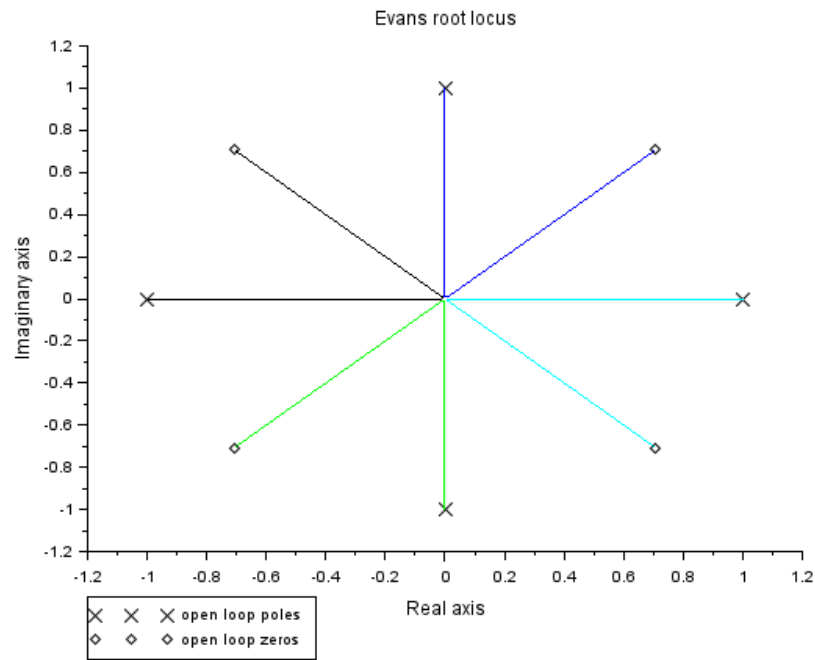
Question 2: Root Locus Analysis and Stability of System

Part A)

Same Breakaway and Breakin points

We can implement symmetric poles around the origin

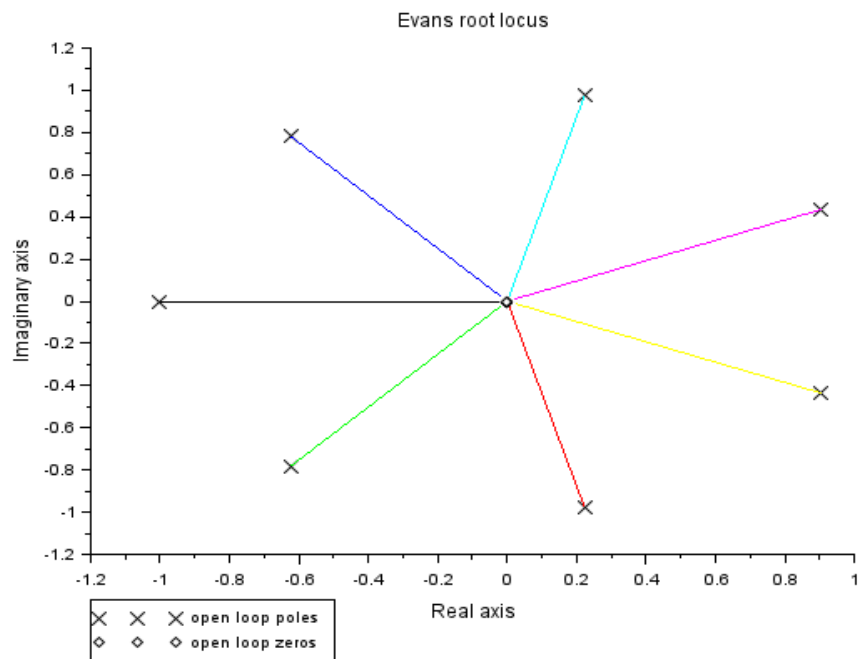
$$G_s = \frac{s^4 + 1}{s^4 - 1}$$



Part B)

The number of branches at breakaway and break-in points are more than 4

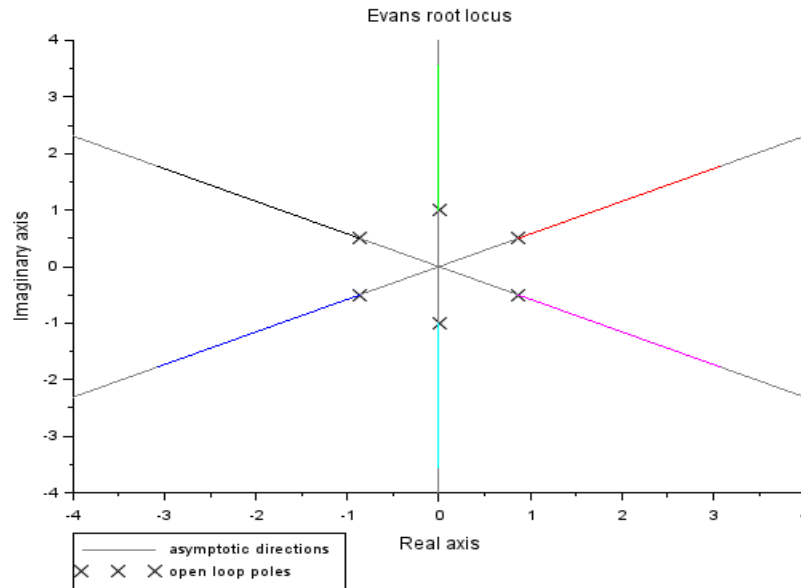
$$G_s = \frac{s^7}{s^7 + 1}$$



Part C)

Branches of Root locus coincide with their asymptotes

$$G_s = \frac{10}{s^6 + 1}$$



Part D)

Breakaway and Break-in points are complex numbers

Step 1] Transfer function with no zero and real, symmetric poles around jw axis

$$G_s = \frac{144}{(s-3)(s+3)(s+4)(s-4)}$$

$$G_s = \frac{144}{(s^2-9)(s^2-16)}$$

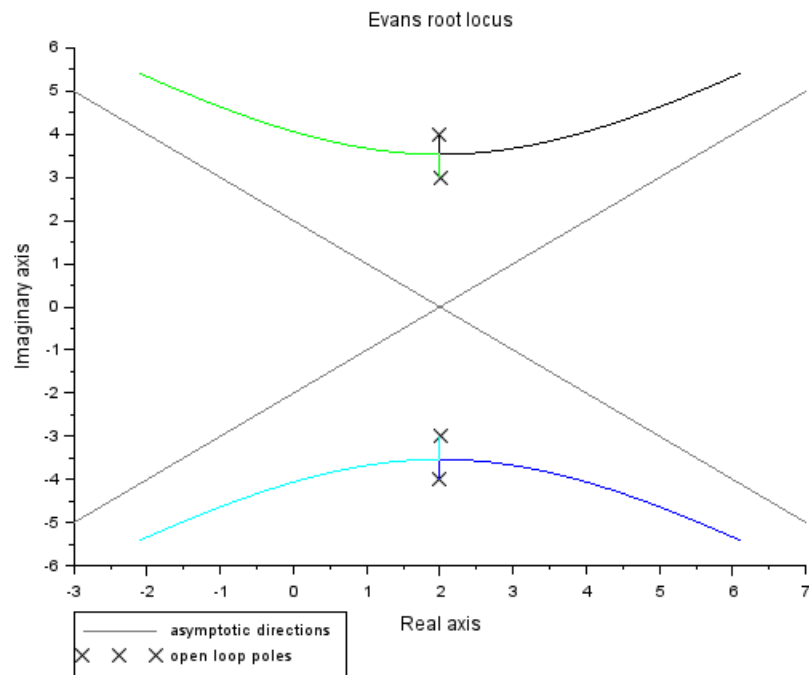
Step 2] Replace s^2 with $-s^2$

$$G_s = \frac{144}{(-s^2-9)(-s^2-4)}$$

$$G_s = \frac{144}{(s^2+9)(s^2+16)}$$

Step 3] Replaces 's' with 's-k'. Here k=2

$$G_s = \frac{144}{((s-2)^2+9)((s-2)^2+16)}$$



Scilab code for the same:

```
s=%s;
```

```
// Part A
```

```
Gs1=syslin('c',s^4+1,s^4-1);
```

```
scf(0);
```

```
evans(Gs1,200);
```

```
xs2png(0,'Q2_A.png');
```

```
// Part B
```

```
Gs2=syslin('c',s^7/(s^7+1));
```

```
scf(1);
```

```
evans(Gs2,200000000000);
```

```
xs2png(1,'Q2_B.png');
```

```
// Part C
```

```
Gs3=syslin('c',10,s^6+1);
```

```
scf(2);
```

```
evans(Gs3,200);
```

```
xs2png(2,'Q2_C.png');
```

```
// Part D
```

```
Gs4=syslin('c',1,(((s-2)^2+9)*((s-2)^2+16)));
```

```
scf(3);
```

```
evans(Gs4,2000);
```

```
xs2png(3,'Q2_D.png');
```

Question 3: Designing Proportional Gain Controller:

$$G_s = \frac{1}{s(s^2 + 3s + 5)}$$

```
Minimum Rise time is: 0.555000  
Minimum Rise time occurs for k=15.000000  
Rise Time of System is 1.5s for k=3.600000
```

Scilab code for the same:

```
s=%s;  
  
k=0.1:0.1:15;  
time=0.0:0.015:60;  
rise_times=zeros(1,length(k));  
  
for n=1:length(k)  
    k_n=k(n);  
    Gs=syslin('c',k_n/(s^3+3*s^2+5*s+k_n));  
    Ps=csim('step',time,Gs);  
    t_90=time(find(Ps>0.9))(1);  
    t_10=time(find(Ps>0.1))(1);  
    rise_times(1,n)= t_90-t_10 ;  
end  
  
Tr_min=min(rise_times(1,2:length(k)));  
k_min=k(find(rise_times == Tr_min));  
  
printf('Minimum Rise time is: %f',Tr_min);  
printf('\nMinimum Rise time occurs for k=%f',k_min);  
  
temp=find(rise_times>1.5);  
k_req=0.1*length(temp);  
  
printf('\nRise Time of System is 1.5s for k=%f',k_req);
```

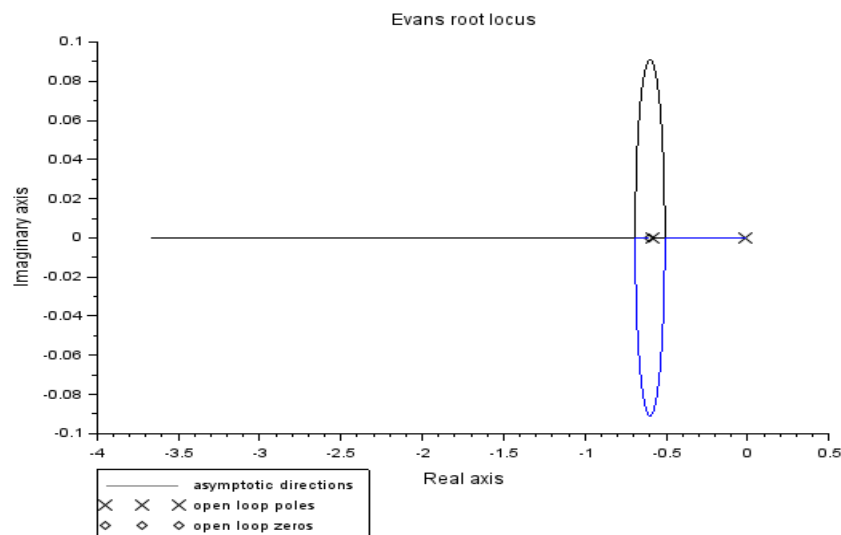
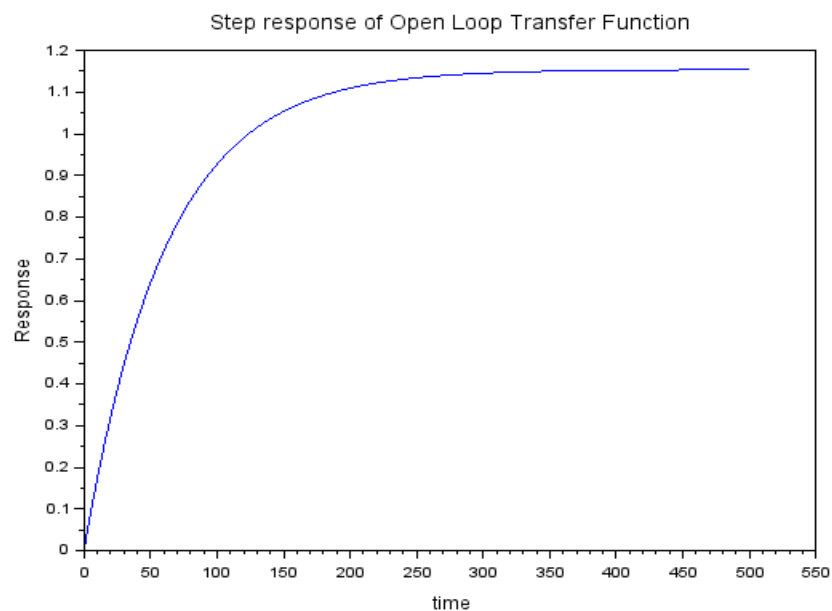

Question 4: Proportional Controller for desired steady-state error:

Part A)

$$G = \frac{0.11(s + 0.6)}{6s^2 + 3.6127s + 0.0572}$$

At $s=0$

$$\text{Steady State Error} = \frac{1}{1+kG} = 0.01$$



Value of proportional gain K_p for 1 percent steady state error: $K_p=85.800000$

Scilab code for the same:

```
s=%s;
Gs=0.11*(s+0.6)/(6*s^2+3.6127*s+0.0572);
Gs=syslin('c',Gs);
evans(Gs,200);

time=0.1:0.1:500;
Ps=csim('step',time,Gs);
scf(2);
plot(time,Ps);
xlabel("time");
ylabel("Response");
xtitle("Step response of Open Loop Transfer Function");

gainK=0.1:0.1:100;
for i=1:length(gainK)
    Es=1/(1+gainK(i)*0.066/0.0572);
    if Es==0.01
        Kp=gainK(i);
    end
end

printf("Value of roportional gain Kp for 1 percent steady state error:
Kp=%f",Kp);
```

Part B)

We can use the RH table for determining the marginal stability of the closed-loop system

$$G = \frac{0.11(s + 0.6)}{6s^2 + 3.6127s + 0.0572}$$

Characteristic Polynomial of the system is

$$P = 6s^2 + (3.6127 + 0.11k)s + (0.0572 + 0.066k)$$

s^2	6	$0.0572+0.066k$	0
s^1	$3.6127+0.11k$	0	0
s^0	$0.0572+0.066k$	0	0

$$3.6127+0.11k > 0 \rightarrow k > -32.84$$

$$0.0572+0.066k > 0 \rightarrow k > -0.867$$

Hence For Marginal Stabiltly, Gain K = -0.867

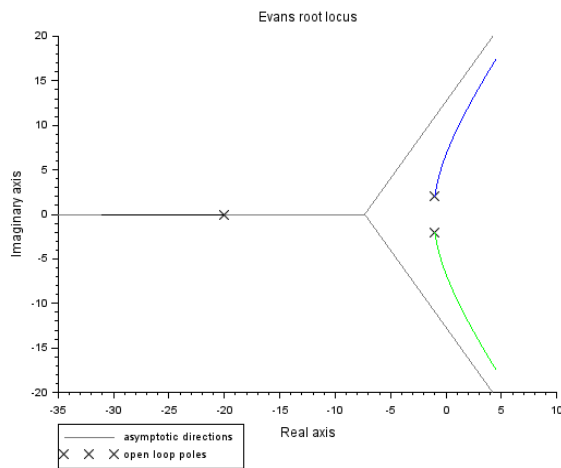
Question 5:

Part A)

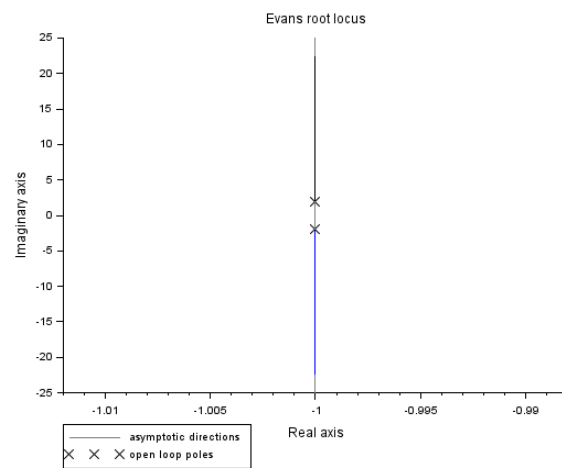
In this part, we have to consider two systems. The first one is the third-order system with two dominant and one pole on the extreme left side of the origin

$$Gs1 = \frac{1}{\left(\frac{s}{20} + 1\right)(s^2 + 2s + 5)}$$

$$Gs2 = \frac{1}{s^2 + 2s + 5}$$



Third-Order System



Second-Order System

Scilab code for the same:

```
s=%s;  
  
// 3 Pole System  
  
Gs1=syslin('c',1,(s^2+2*s+5)*(s/20+1));  
scf(1);  
evans(Gs1,500);  
  
// 2 Pole System  
  
Gs2=syslin('c',1,(s^2+2*s+5));  
scf(2);  
evans(Gs2,500);
```

Part B)

It is evident from the root locus of both the systems that the second-order system will always remain stable for any value of gain K, but the third-order system will become unstable after reaching the specific maximum value of gain K. We can find this value using the RH table as follow

$$Gs1 = \frac{20}{(s + 20)(s^2 + 2s + 5)}$$

Characteristic Polynomial becomes

$$Ps = s^3 + 22s^2 + 45s + 100 + 20k$$

s^3	1	45	0
s^2	22	100+20k	0
s^1	$\frac{890 - 20k}{22}$	0	0
s^0	100+20k	0	0

$$890 - 20k > 0$$

$$k < 44.5$$

Hence, for K up to 44.5, the unit step response of both systems will be similar

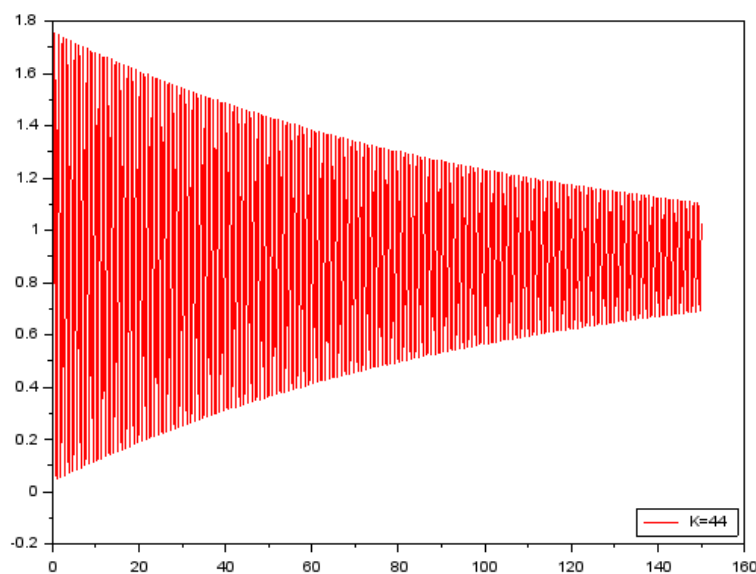


Fig. Stable System

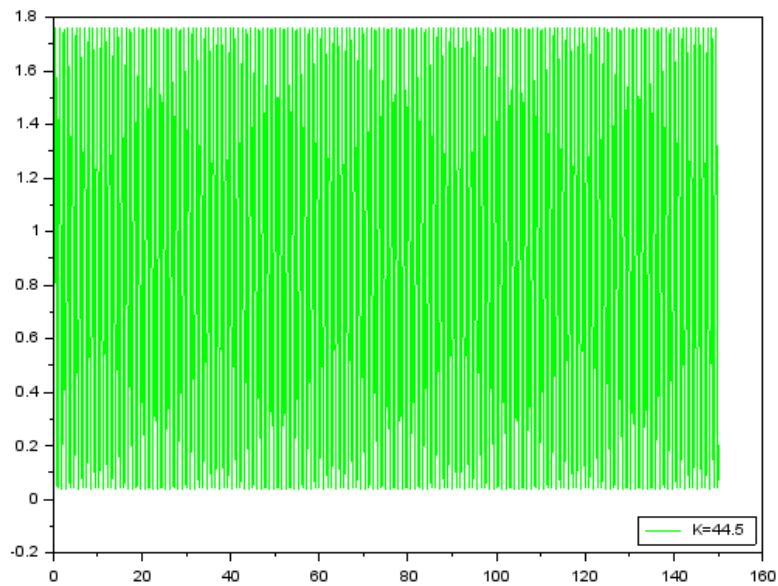


Fig. Marginally Stable System

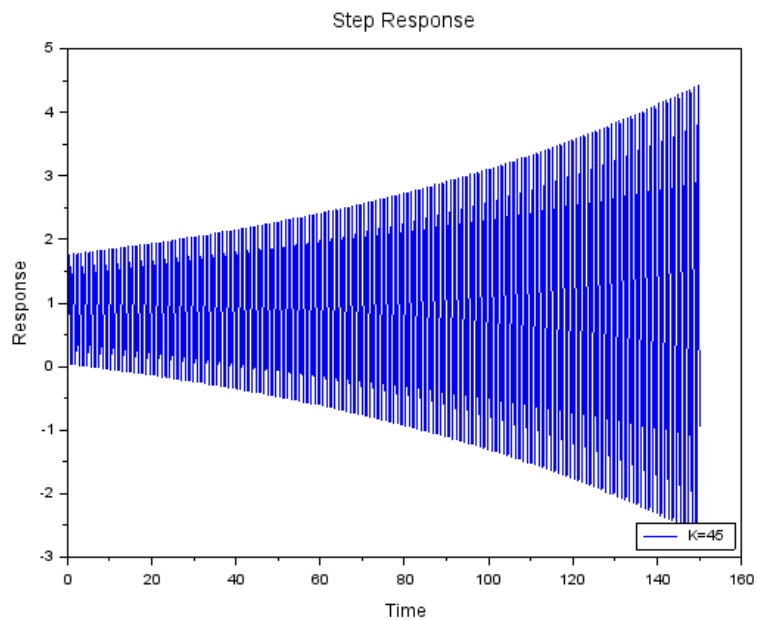


Fig. Unstable System

Scilab code for the same:

```
s=%s;

Gs1=syslin('c',1,(s^2+2*s+5)*(s/20+1));
time=0:0.001:150;
```

```
Ps1=csim('step',time,44*Gs1/(1+44*Gs1));  
Ps2=csim('step',time,44.5*Gs1/(1+44.5*Gs1));  
Ps3=csim('step',time,45*Gs1/(1+45*Gs1));  
  
plot(time,Ps1,'r');  
legends(['K=44'],[5],opt="lr");  
scf(2);  
plot(time,Ps2,'g');  
legends(['K=44.5'],[3],opt="lr");  
scf(3);  
plot(time,Ps3,'b');  
legends(['K=45'],[2],opt="lr");  
xlabel("Time");  
ylabel("Response");  
title("Step Response");
```