

Fall Semester 2015

**Line following robot**

Group 2

2. Semester IT-Technology

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Title:

Line following robot

Project Period:

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Projectgroup:

Group 2

Medvirkende:

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Pages: TBD

Appendices: TBD

Completed TBD

# Introduction

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Project is written by:

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# Glossary

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ADC Analog-digital conversion

PID Proportional-integral-derivative controller

THT Through-hole-technology

3D print Printing something in 3 dimensions

# Requirements specification

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# 1

The following section will describe the specific requirements that have been decided to fulfil to the general requirements as shown in the project description.

- Project must include light sensors
- Implement motor control
- Should make use of the Pic32 MCU
- Software will be written in MPLAPX
- Autonomous operation
- The product must make use of feedback concept e.g. a PID algorithm
- A function & performance test is to be conducted

# Hardware section 2

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## 2.0.1 Hardware diagram

## 2.1 Description of the hardware structure and functionality

In the following section, the hardware parts and functions will be introduced and described.

## 2.2 Selection of sensor

Name	QRE1113 board	OPB706A	OPB704
Max sensor distance	3mm	1.27mm	3.8mm
Forward current	50mA	20mA	40mA
Mounting	On print	THT	In casing
Price	19.43DKK	26.90DKK	42.55DKK
Notes			

Table 2.1: Table of a selection of sensors.

TBD Beskriv sensorer og hvorfor vi har valgt denne

### 2.2.1 OPB704 Sensor

The selected sensor for the line following robot will be the OPB704. This sensor has been chosen from a range of criteria, most important being the optimal sensor distance and the easier mounting method. This method makes it easier to add several sensors in an array, which will be done with a 3D-printed mount. TBD billedede af sensor array?

TBD Beskriv hardwarestruktur og funktion samt alle underdele

## 2.3 Analog-to-digital converter (ADC)

The purpose of the ADC is to convert the analog data from the sensors to digital data that can be managed by a computer - this allows data received from the blue-tooth transmitter on the product to be processed into readable data more easily, which is great for showing how the sensors are reacting. The sensors themselves cannot discern what they actually need to read, the sensors just read anything they can see and send that signal.



Analog signals can have a significant amount of noise - since any received noise is interpreted as part of the signal, a digital signal is not only more easy to work with, it will also provide more precise data. This will make for more accurate readings on the tachometer on the robot, which allows even more finely tuned monitoring of the robot and its working processes.

### 2.3.1 ADC diagram

TBD

**This products usage of ADC**

TBD

## 2.4 The Arduino Uno32 board

Our main computing core is at the Arduino Uno32 board. The board was selected both because of previous experiences, and because it met the set requirements perfectly - most importantly it has twelve analog inputs to handle our array of seven sensors. This board utilizes the PIC32MX320F128 microcontroller, which features a 32-bit processing core running at 80Mhz with 128K of flash program memory and 16K SRAM data memory. This board allowed the most important aspects of the project to become reality; namely a GUI showing both the ADC readings as well as visualizing the sensor inputs, plenty of analog inputs to allow for the sensor array, and BlueTooth transmission of data from the robot unto a computer.

The board is compatible for use with MPLAB IDE and the PICKit3 debugger.

## 2.5 The motor shield - PKA03

The motor shield was chosen because it is compatible with the Uno32 board, and fits perfectly within the scope of the project. It controls both motors and receives power from the 7.2 V lithium-ion battery pack mounted on the chassis itself. The motor shield is instrumental in providing motor controls to the product. To do this, it utilizes a specific electric circuit, called an h bridge.

### 2.5.1 The H bridge

An h bridge is a circuit that allows a voltage to be applied across a load in either direction. The purpose of this in the case of this project is to enable controls of the two motors in a way so that they can function individually, and both drive forwards and backwards.

At first, there were experiments with custom prints and a purpose-built h bridge on the first iteration of the motor shield. However after testing, it proved faulty and was changed to accommodate time constraints.

# Software section 3

## 3.0.1 Software diagram

TBD Software diagram

## 3.1 Description of the software structure and functionality

The following section will introduce the software segment, based on our required specifications. The design will be shown in the form of a flowchart. TBD Softwarebeskrivelse og underafsnit

## 3.2 Description of the PID controller

A PID controller continuously calculates an error value as the difference to a reference point and a measured process variable.

PID is an abbreviation for a proportional-integral-derivative controller, it is a control loop feedback mechanism. The controller's job is to minimize the error value for the given devices running time. In the case of this project the reference point is the line to follow and the PID will allow the MCU to adjust the power to the engines, to steer accordingly to said reference point.

$$F(t) = K_p e(t) + K_i \int_0^t e(\tau) d\tau + K_d \frac{de(t)}{dt}$$

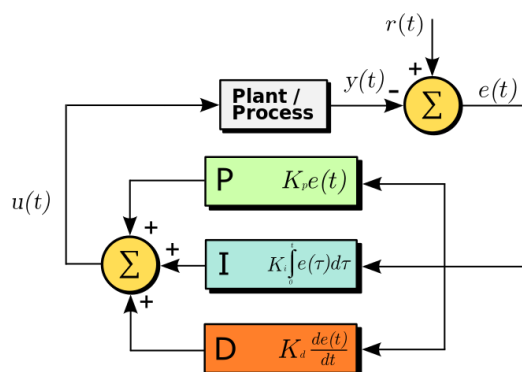


Figure 3.1: Block diagram, showing the idea of PID controller. Credits:

### 3.2.1 Proportional control(P)

The proportional part creates an output value that is proportionally related to the current error value, this value can be tuned by multiplying the error by a constant

$K_p$ . A high proportional gain results in a large change in the output for a given change in the error.

$$P_{\text{out}} = K_p e(t)$$

If the proportional gain is too high, the system can become unstable. Contrarily, a small gain will result in the device adjusting too slowly, which decreases overall efficiency and in the case of this project, it will end up being detrimental to the steering accuracy.

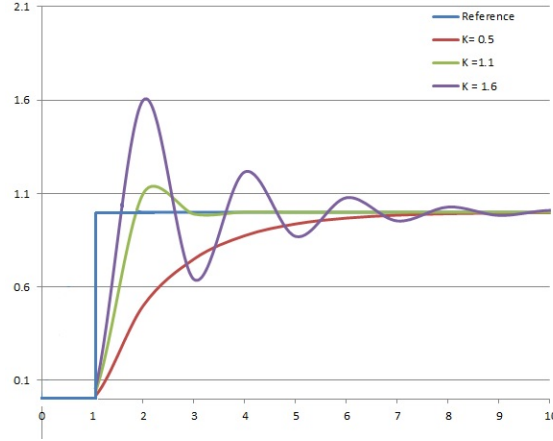


Figure 3.2:  $K_p$  with 3 values. ( $K_i$ ,  $K_d$  held constant) Credits:

### 3.2.2 Steady-state error

TBD (omskrives)

### 3.2.3 Integral control(I)

The integral controller is contributing proportionally to both the magnitude of the error and the duration of the error.

The integral in a PID controller is the sum of the instantaneous error over time and gives the accumulated offset that should have been corrected previously.

The controller output equals the accumulated error multiplied by the integral gain( $K_i$ )

$$I_{\text{out}} = K_i \int_0^t e(\tau) d\tau$$

The integral part accelerates the movement of the process towards the reference point. Since the integral term correlates to accumulated errors from the past, it can cause the present value to overshoot the reference value.

### 3.2.4 Derivative control(D)

The derivative of the process error is calculated by determining the slope of the error over time and multiplying this rate of change by the derivative gain  $K_d$ . The

magnitude of the contribution of the derivative term to the overall control action is termed the derivative gain,  $K_d$ .

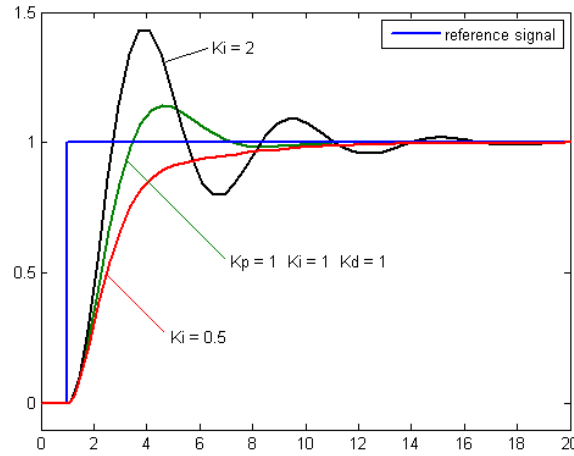


Figure 3.3:  $K_i$  shown with 3 values. Credits:

TBD (billede skal rettes ind til teksten) The derivative term is given by:

$$D_{\text{out}} = K_d \frac{de(t)}{dt}$$

The derivative action predicts system behaviour and utilizes this to improve the settling time and stability of the system. An ideal derivative is not causal, so that implementations of PID controllers include an additional low pass filtering for the derivative term, to limit the high frequency gain and noise.

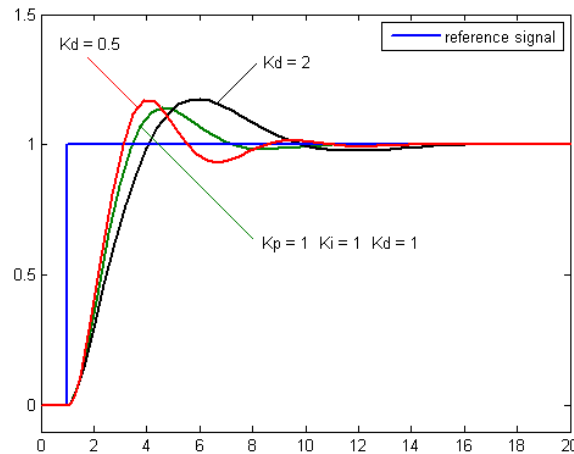


Figure 3.4:  $K_d$  shown with 3 values. Credits:

### 3.2.5 Loop tuning

Tuning the loop is the term used to describe the adjustments of the PID's control parameters (proportional band/gain, integral gain/reset, derivative gain/rate) to the optimal values for the given control scheme.

Stability is the first requirement; however systems can differ greatly, and different applications may have different requirements and these may even conflict with each

other. For example, high speed and high accuracy often cancel each other out, because high speed may cause overshooting, while high accuracy is slow. The ideal realistic behaviour is both as fast as possible, while also having minimum overshoot and oscillation.

Even though the process seems simple, with only three variables, it can be challenging to achieve, because it must satisfy the criteria despite being within the limitations of PID control. While adjusting the PID can seem conceptually intuitive, and while most PIDs may perform acceptably with default controls, they may very well also have an unsatisfactory performance.

This can generally be fixed through optimisation and tuning, either through computer simulations or manual testing. In our case, we used manual tuning of the numbers.

### 3.2.6 Stability

If the parameters of the PID controller are set incorrectly the process input can become unstable. This means the controllers output becomes divergent, which can be limited by saturation and mechanical breaking.

### 3.2.7 Manual tuning

When a system must be online at all times a method for tuning is to first set  $K_i$  and  $K_d$  values to zero. Increase  $K_p$  until the loop output oscillates, setting  $K_p$  at approximately half the value for a "quarter amplitude decay" type response.

Then increase  $K_i$  until any set off is corrected in sufficient time for the process. Adding too much  $K_i$  will however cause an instability. Finally, increase  $K_d$ , if required at all, until the loop is acceptably quick to reach its reference after a load disturbance.

A fast PID loop tuning process usually overshoots slightly to reach the reference point faster.

In the case of systems that can't accept overshoot, an over-damped closed-loop system is best suited, which requires  $K_p$  setting significantly less than half that of the  $K_p$  setting that was causing the oscillation.

TBD kilde (<http://blog.opticontrols.com/archives/1066>)

TBD (Vores fremgangsmåde med PID alt efter om "D" skal bruges)

Table 3.1: Manual tuning

Parameter	Rise time	Overshoot	Settling time	Steady-state error	Stability
$K_p$	Decrease	Increase	Small change	Decrease	Degrade
$K_i$	Decrease	Increase	Increase	Eliminate	Degrade
$K_d$	Minor change	Decrease	Decrease	No effect in theory	Improves if $K_d$ is small

**Table 3.1 explained**

Table 3.1 gives an informative overview of what the different parameters does when tuned manually.

- To minimize the rise time, decrease  $K_p$
- To eliminate the steady-state error, increase  $K_i$
- To reduce the overshoot and settling time, decrease  $K_d$

### 3.3 Description of the Pulse Width Modulator - PWM

A pulse width modulation technique used to encode a message into a pulsing signal. Its primary use is to control the power supply of electronic devices - the case of this project; this means the motors. The average voltage and amplitude output to the motors is altered by rapidly switching between an 'on' and 'off' state. This way, the average output is easily altered by changing the proportions between the on and off state. The longer the switch is set to on compared to off, the higher the average supply will be. Pulse-width modulation utilizes a rectangular pulse where the width of the pulse is modulated to get the variation in the average value of the waveform. Given the pulse waveform  $f(t)$  over the period  $T$ , low value  $y_{\min}$  and high value  $y_{\text{high}}$  and duty cycle  $D$ , the average waveform is given by:

$$\bar{y} = \frac{1}{T} \int_0^T f(t) dt$$

This expression can be simplified where  $y_{\min} = 0$  as  $\bar{y} = D \cdot y_{\max}$ . From this it can be observed that the average value of the signal ( $\bar{y}$ ) has a direct correlation with the duty cycle  $D$ .

$$\begin{aligned} \bar{y} &= \frac{1}{T} \left( \int_0^{DT} y_{\max} dt + \int_{DT}^T y_{\min} dt \right) \\ &= \frac{1}{T} (D \cdot T \cdot y_{\max} + T(1 - D)y_{\min}) \\ &= D \cdot y_{\max} + (1 - D) y_{\min} \end{aligned}$$

This expression can be simplified where  $y_{\min} = 0$  as  $\bar{y} = D \cdot y_{\max}$ . From this it can be observed that the average value of the signal ( $\bar{y}$ ) has a direct correlation with the duty cycle  $D$ .

#### 3.3.1 Duty cycles

The duty cycle describes the proportion of 'on' compared to any given period of runtime for the device. The duty cycle is described as a percentage, where 100% means that it's turned on the entire time, where 10% would be a tenth of the time.

TBD Vores duty cycle

# Test 4

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## 4.1 Test



# Conclusion 5

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TBD Konklusion

# Appendices 6

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## 6.1 Group collaboration agreement

### 6.1.1 Contact Information

Table 6.1: Contacts

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### 6.1.2 Workflow

- Every friday after 12:00 is expected work consisting of three hours.
- If you aren't able of attending for scheduled study day. - Notice must be given to the project team.

### 6.1.3 Milestones and goals

TBD Milestones

### 6.1.4 Deadline

- Hand in June 7th.

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TBD list of references PID:

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