

# **THE AMPLIFIER PROJECT**

**ELEC 3509A – L6**

**Lab 2 Report**

**Name:** Youssef Ibrahim

**Instructor:** Qi-Jun Zhang

## 1.0 INTRODUCTION

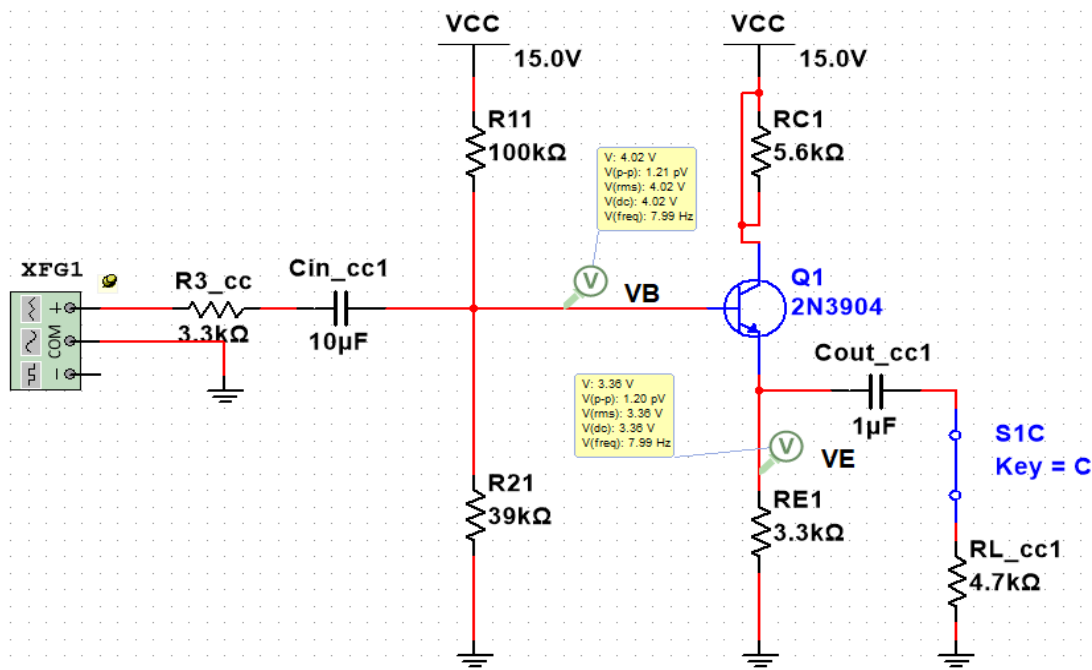
The purpose of this laboratory is to investigate the use of BJTs as amplifier circuit elements.

First, the three basic configurations (CE, CC, and CB) are observed. Then, by proper combinations and permutations, 2- transistor amplifier configurations can be studied for improved gain-bandwidth performance. Finally, a specific configuration is required to be designed to meet or exceed a prescribed set of specifications.

## 2.0 DAY 1: SINGLE-TRANSISTOR AND 2-TRANSISTOR AMPLIFIERS

### 2.1 Part 1: Circuit Construction and DC Measurements

- Common Collector:



DC Measurements:

$V_B$	4.01 V
$V_E$	3.34 V
$V_C$	9.36 V
$V_{CC}$	15.00 V

$$I_C = \frac{V_{CC} - V_C}{R_C}$$

$$I_E = \frac{V_E}{R_E}$$

$$I_B = I_{R1} - I_{R2} = \frac{V_{CC} - V_B}{R_1} - \frac{V_B}{R_2}$$

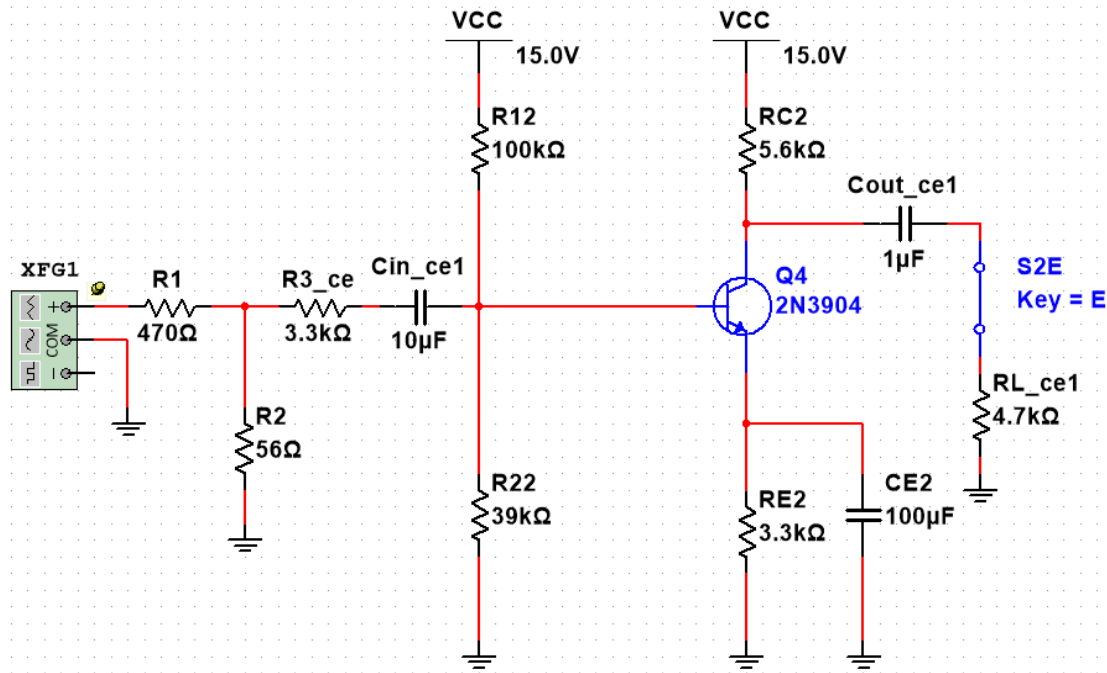
$$\beta = \frac{I_C}{I_B}$$

$$r_\pi = \frac{\beta}{gm}$$

Using the formulas shown above, I was able to find the values of  $I_C$ ,  $I_E$ ,  $I_B$ ,  $\beta$ , and  $r_\pi$  as shown below:

$I_C$	1.007 mA
$I_E$	1.012 mA
$I_B$	7.079 $\mu$ A
$\beta$	142.250 A/A
$r_\pi$	3.743 k $\Omega$

- Common Emitter:

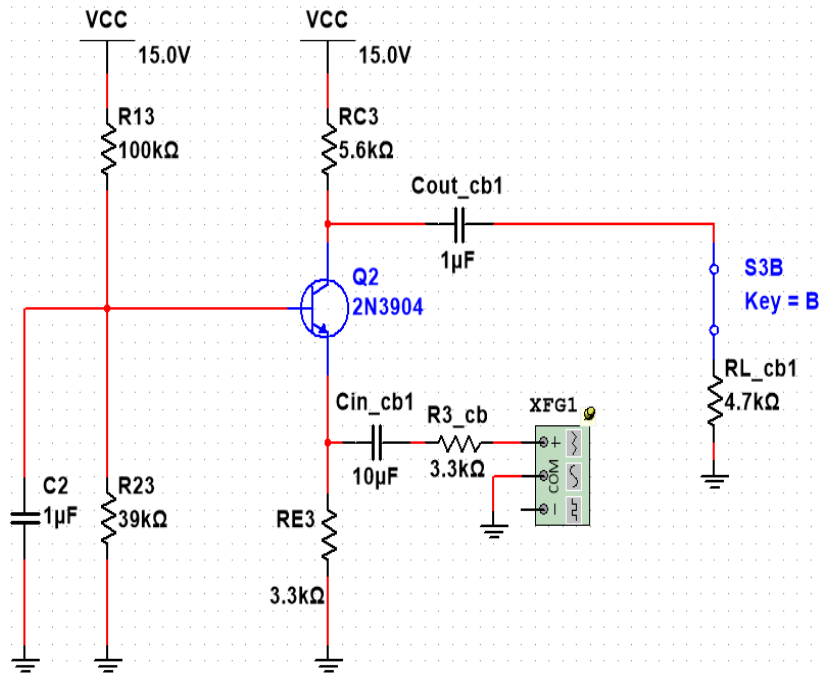


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- Common Base:



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- Why can  $R_{C1}$  be made 0 when testing the amplifier? Why do you need to have it when checking the DC operating point?

$R_{C1}$  can be made 0 when testing the amplifier since it has no effect on collector current during AC analysis. However, during the analysis, to find the DC operating point, we must find the collector current that keeps the transistor in proper biasing point, and the collector current depends on the value of  $R_{C1}$ . Hence, we need to have  $R_{C1}$  when checking the DC operating point.

- Could we instead use a capacitor like we do for the CE and CB amplifiers?

Yes, we could use a capacitor like we do for the CE and CB amplifiers to bypass the collector resistor.

- Why do we need RE2? Why can't we just short this node to ground?

We need RE2 since it stabilizes the Q-point regardless of changes in voltage supplied.

## 2.2 Part 2: AC Measurements for Single-Transistor Amplifiers

- Show a complete table summarizing the values of calculated and measured of  $R_{in}$ ,  $R_{out}$ ,  $A_{mid}$ ,  $f_H$  and  $f_L$  for all 3 amplifiers. Comment on any differences you see between measured and calculated values. Explain any differences you see. Try to be as specific as possible.

- Common Collector:  $V_{Source} = 1.15 V_{Pk} = 2.3 V_{Pk-Pk}$

$V_o$	2.00 V
$V_o'$	2.03 V
$V_s$	2.30 V
$V_x$	2.04 V
$V_{s-x}$	0.26V

- Common Emitter:  $V_{Source} = 0.195 V_{Pk} = 0.39 V_{Pk-Pk}$

$V_o$	2.0100 V
$V_o'$	4.2500 V
$V_s$	0.0412 V
$V_x$	0.0220 V
$V_{s-x}$	0.0192 V

- Common Base:  $V_{Source} = 1.325 V_{Pk} = 2.65 V_{Pk-Pk}$

$V_o$	2.0000 V
$V_o'$	4.3800 V
$V_s$	2.6500 V
$V_x$	0.0223 V
$V_{s-x}$	2.6277 V

$$R_{in} = R_3 + R_{inA} = R_3 + \frac{v_x}{i_{R_3}}$$

$$R_{out} = \frac{v_o' - v_o}{i_{R_L}}$$

$$A_{mid} = \frac{v_o}{v_s}$$

Using the formulas shown above, I was able to find the values of  $R_{in}$ ,  $R_{out}$ , and  $A_{mid}$  as shown below:

➤ Common Collector:

	Prelab	Measured
$R_{in}$	25.50 k $\Omega$	29.2 k $\Omega$
$R_{out}$	46.40 $\Omega$	70.5 $\Omega$
$A_{mid}$	0.84	0.70
f <sub>H</sub>	2.85 MHz	3.0 MHz
f <sub>L</sub>	33.5 Hz	32.1 Hz

➤ Common Emitter:

	Prelab	Measured
$R_{in}$	3.30 k $\Omega$	7.081 k $\Omega$
$R_{out}$	5.60 k $\Omega$	5.238 k $\Omega$
$A_{mid}$	48.78	48.786
f <sub>H</sub>	448 kHz	472 kHz
f <sub>L</sub>	52.18 Hz	41.9 Hz

➤ Common Base:

	Prelab	Measured
$R_{in}$	25.90 k $\Omega$	3.328 k $\Omega$
$R_{out}$	5.60 k $\Omega$	5.593 k $\Omega$
$A_{mid}$	0.755	0.755
f <sub>H</sub>	15.60 MHz	31.2 MHz
f <sub>L</sub>	26.51 Hz	17.9 Hz

Table below summarizes the values of calculated and measured of  $R_{in}$ ,  $R_{out}$ ,  $A_{mid}$ ,  $f_H$  and  $f_L$  for all three amplifiers.

	Prelab					Measured				
	$R_{in}$ ( $k\Omega$ )	$R_{out}$ ( $k\Omega$ )	$A_{mid}$	$f_H$ (MHz)	$f_L$ (Hz)	$R_{in}$ ( $k\Omega$ )	$R_{out}$ ( $k\Omega$ )	$A_{mid}$	$f_H$ (MHz)	$f_L$ (Hz)
<b>CC</b>	25.50	0.046	0.84	2.85	33.50	29.20	0.0705	0.70	3.00	32.10
<b>CE</b>	3.30	5.60	48.78	0.45	52.18	7.08	5.24	48.79	0.47	41.90
<b>CB</b>	25.90	5.60	0.76	15.60	26.51	3.33	5.59	0.76	31.20	17.90

As it can be seen from the table above, there is no major difference between the measured and calculated values of  $R_{in}$ ,  $R_{out}$ ,  $A_{mid}$ ,  $f_H$ , and  $f_L$  for all three amplifiers.

- In a normal circuit, the AC measurements could be prone to many sources of error. Try identifying some of these possible sources of error that would be encountered in a real circuit; how could we avoid these sources of error?

Possible sources of error that would be encountered in a real circuit is ignoring intrinsic resistance  $r_o$  and feedback resistance  $r_\mu$ . Also the  $C_\mu$  and  $C_\pi$  used in the high cut-off frequency was not exact since the values of  $C_\mu$  and  $C_\pi$  depends on the characteristics of BJT. We could avoid this source of error by including ignoring intrinsic resistance  $r_o$  and feedback resistance  $r_\mu$  in our calculations and getting the exact the  $C_\mu$  and  $C_\pi$  values from the information sheet of 2N3904.



- Comment on the differences between all 3 amplifiers. Mathematically explain any differences you see: for instance if one amplifier has a higher gain, explain why that is so. Which ones are better for which tasks and why?

Common Collector: has voltage gain of about 1 (unity gain), moderate low cut-off frequency, moderate cut-off frequency, high input impedance, and low output impedance. It is best for where a high impedance input source needs to be connected to a low impedance output load requiring a high current gain, it is usually used as a voltage buffer since the voltage gain is unity.

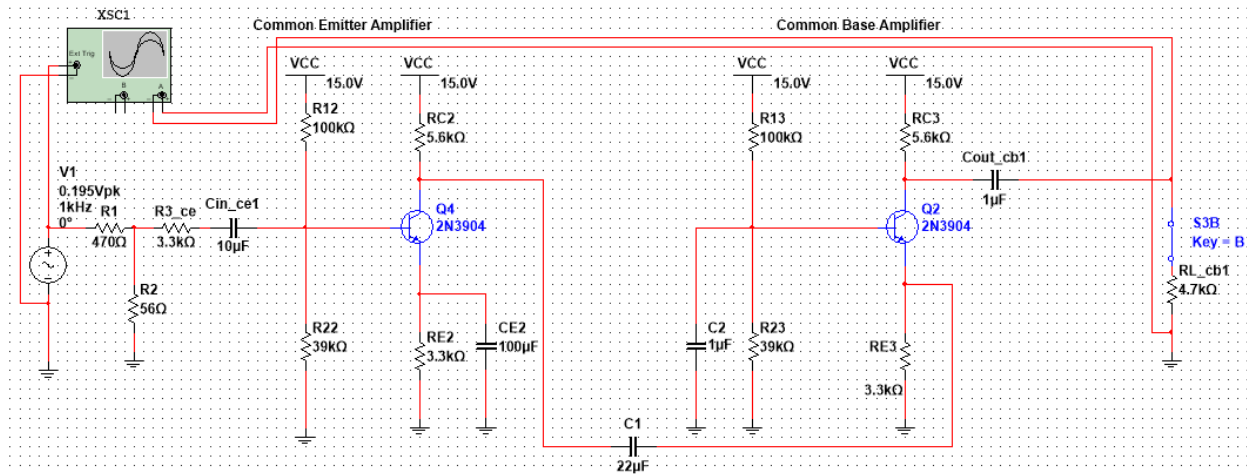
Common Emitter: has the highest voltage gain and is dependent upon  $\frac{R_C}{R_E}$ , highest low cut-off frequency due to miller's effect, lowest high cut-off frequency, moderate input impedance, and its output impedance equal collector resistance  $R_C$ . It is best for radio frequency circuits to amplify weak signals received by an antenna.

Common Base: Has a very low voltage gain since it depends on the source impedance  $R_S$ , very little low cut-off frequency, very large high cut-off frequency, low input impedance, and its output impedance equal collector resistance  $R_C$ . It is best for audio and radio frequency applications as a current buffer in order for it to match a low impedance source to a high impedance load or as a single stage amplifier in part of a cascaded or multi-stage configuration.

## 2.3 Part 3: 2-Transistor Amplifiers

- Include a table summarizing the measured values of  $R_{in}$ ,  $A_{mid}$ ,  $f_H$  and  $f_L$ . Calculate expected values for these and compare them. Like with the 1 transistor amplifiers, explain any differences between the two set of values.

➤ CE-CB Amplifier:  $V_{source} = 0.195 V_{pk} = 0.39 V_{pk-pk}$

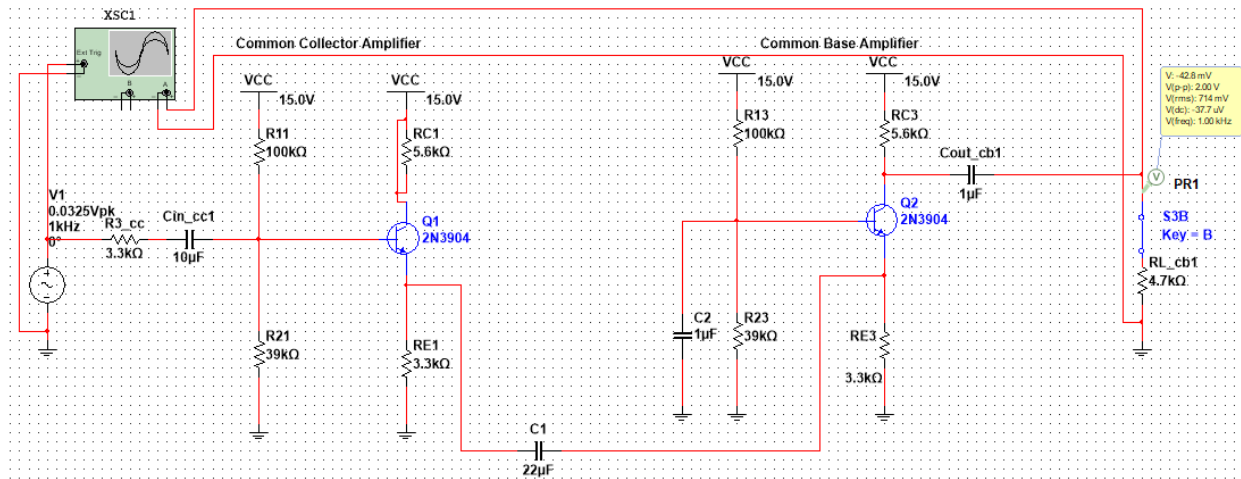


	Prelab	Measured
$V_o$	-	2.04 V
$V_o'$	-	4.48 V
$V_s$	-	0.0412 V
$V_x$	-	0.022 V
$V_{s-x}$	-	0.0192 V
$R_{in}$	3.302 kΩ	7.081 kΩ
$R_{out}$	5.6 kΩ	5.238 kΩ
$A_{mid}$	48.78	49.51
$f_H$	4.276 MHz	4.36 MHz
$f_L$	50.895 Hz	41.5 Hz

- What do you notice about the cascading amplifiers parameters?

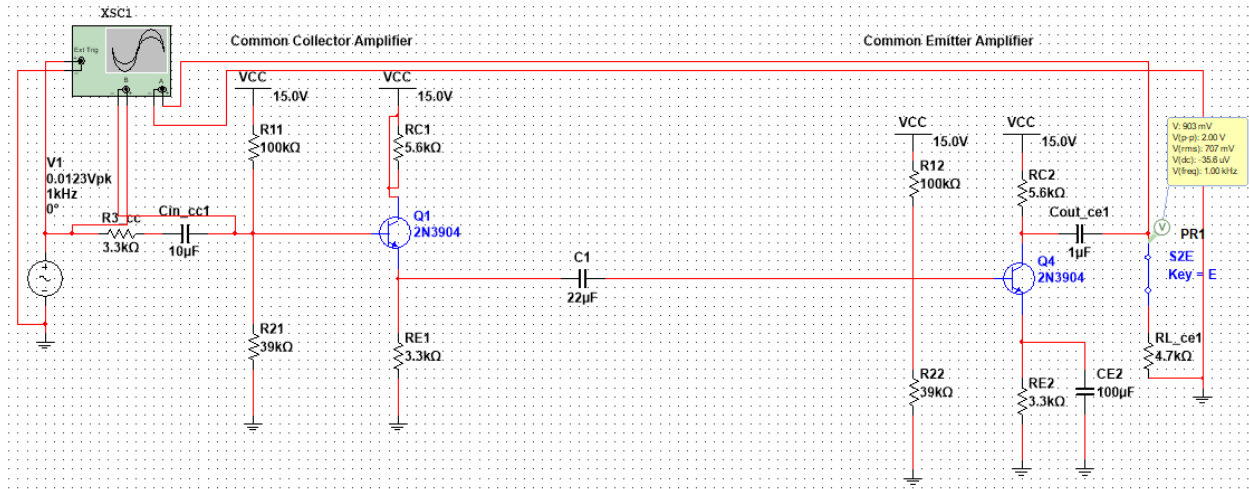
Cascading amplifiers have common parameters.

➤ CC-CB Amplifier:  $V_{source} = 0.0325 V_{Pk} = 0.065 V_{Pk-Pk}$



	Prelab	Measured
$V_o$	-	2.0 V
$V_o'$	-	4.32 V
$V_s$	-	0.0649 V
$V_x$	-	0.0444 V
$V_{s-x}$	-	0.0205 V
$R_{in}$	6 kΩ	10.45 kΩ
$R_{out}$	5.6 kΩ	5.45 kΩ
$A_{mid}$	34	30.82
$f_H$	3.68 MHz	5.4 MHz
$f_L$	136 Hz	113.7 Hz

➤ CC-CE Amplifier:  $V_{Source} = 0.0123 V_{Pk} = 0.0246 V_{Pk-Pk}$



	Prelab	Measured
$V_o$	-	2.0 V
$V_o'$	-	4.23 V
$V_s$	-	0.0246 V
$V_x$	-	0.0218 V
$V_{s-x}$	-	0.0028 V
$R_{in}$	25.5 kΩ	28.992 kΩ
$R_{out}$	5.6 kΩ	5.24 kΩ
$A_{mid}$	88.2	81.30
$f_H$	6.2 MHz	10.1 MHz
$f_L$	79 Hz	65 Hz

Table below summarizes the values of calculated and measured of  $R_{in}$ ,  $R_{out}$ ,  $A_{mid}$ ,  $f_H$  and  $f_L$  for all three 2-transistor amplifiers.

	Prelab					Measured				
	$R_{in}$ (kΩ)	$R_{out}$ (kΩ)	$A_{mid}$	$f_H$ (MHz)	$f_L$ (Hz)	$R_{in}$ (kΩ)	$R_{out}$ (kΩ)	$A_{mid}$	$f_H$ (MHz)	$f_L$ (Hz)
CE-CB	3.30	5.60	48.78	4.78	50.90	7.08	5.24	49.51	4.36	41.5
CC-CB	6.00	5.60	34.00	3.68	136.00	10.45	5.45	30.82	5.40	113.70
CC-CE	25.50	5.60	88.20	6.20	79.00	28.99	5.24	81.30	101.10	65.00

As it can be seen from the table above, there is no major difference between the measured and calculated values of  $R_{in}$ ,  $R_{out}$ ,  $A_{mid}$ ,  $f_H$ , and  $f_L$  for all three 2-transistor amplifiers.

- Comment on the differences between all three 2-transistor amplifiers. Like you did for the previous section, explain any differences in performance. Which ones are better for which tasks and why?

CE-CB: Moderate voltage gain, lowest high cut-off frequency, lowest low cut-off frequency, low input impedance, and low output impedance which is similar to CC-CE. It is best used when low input impedance is needed such as in moving coil microphones preamplifiers.

CC-CB: Lowest voltage gain, low high cut-off frequency but slightly larger than CE-CB's high cut-off frequency, highest low cut-off frequency, output impedance is slightly larger than CE-CB and CC-CE's output impedance, and moderate input impedance. It is best used in radio frequency applications with low-impedance source to high-impedance load.

CC-CE: Highest voltage gain, it has the biggest high cut-off frequency, moderate low cut-off frequency, output impedance is similar to CE-CB, and highest input impedance. It is best for voltage amplification especially at low frequencies, it is commonly used in radio frequency transceiver circuits and low noise amplifiers.

- In weeks 2 & 3 you will be designing a cascade amplifier, a variation on the CE-CB amplifier. Identify what advantages this has over any of the single transistor amplifiers you tested. What are the advantages and disadvantages of using the cascade topology over any of the 2-transistor topologies you just tested?

The benefits of using CE-CB over any single transistor amplifier is that CE-CB has an overall great voltage gain, high cut-off frequency, low cut-off frequency, input impedance, output impedance. It lowers the low cut-off frequency and lower the cost of circuit.

### 3.0 DAY 2: THE AMPLIFIER PROJECT

#### Cascode Amplifier Requirements:

- $Z = 0 + 8 + 0 = 8$
- $|A_V| = 12\sqrt{8 + 35} = 78.69 \pm 10\% = 78.69 \pm 7.87 \text{ } v/v$
- $R_L = 6(8 + 40)^2 = 13824\Omega = 13.82k\Omega \approx 14k\Omega$
- $f_H > 1 \text{ MHz}$
- $V_{out} = 2 V_{pk-pk}$
- No DC current flow in  $R_L$  and signal generator
- $f_L < 200 \text{ Hz}$
- Total circuit power  $\leq 50\text{mW}$
- All transistors used are 2N3904
- $I_C = 1.0 \text{ mA} \pm 10\% = (1.0 \pm 0.1) \text{ mA}$
- Power-supply is +15V
- $\beta = 200$ ;  $\alpha = 0.995$ ;  $C_\mu = 4 \text{ pf}$ ;  $C_\pi = 8 \text{ pf}$

#### DC Analysis:

$$V_E = 0.2 * V_{CC} = 0.2 * 15 = 3 \text{ V}$$

$$\therefore V_{CE} = 2 \text{ V}$$

$$V_{C1} = V_{E1} + V_{CE1} = 3 + 2 = 5 \text{ V}$$

$$V_{C2} = \frac{V_{CC} + V_{C1}}{2} = \frac{15 + 5}{2} = 10 \text{ V}$$

$$R_{E2} = \frac{V_E}{I_E} = \frac{3}{1 \text{ mA}} = 3 \text{ k}\Omega$$

$$R_C = \frac{V_{CC} - V_{C2}}{I_{C2}} = \frac{15 - 10}{1 \text{ mA}} = 5 \text{ k}\Omega$$

$$V_{B1} = V_{E1} + 0.7 = 3 + 0.7 = 3.7 \text{ V}$$

$$V_{B2} = V_{C1} + 0.7 = 5 + 0.7 = 5.7 \text{ V}$$

$$I_{B1} = I_{B2} = \frac{I_C}{\beta} = \frac{1 \text{ mA}}{200} = 5 \mu A$$

$$I_{BB} = 10(I_{B1} + I_{B2}) = 10(5 \mu A + 5 \mu A) = 100 \mu A$$

$$R_1 = \frac{V_{CC} - V_{B2}}{I_{BB}} = \frac{15 - 5.7}{100 \mu A} = 93 \text{ k}\Omega$$

$$R_2 = \frac{V_{B2} - V_{B1}}{I_{BB} - I_{B2}} = \frac{5.7 - 3.7}{100 \mu A - 5 \mu A} = 21.05 \text{ k}\Omega$$

$$R_3 = \frac{V_{B1}}{I_{BB} - I_{B1} - I_{B2}} = \frac{3.7}{100 \mu A - 5 \mu A - 5 \mu A} = 41.11 \text{ k}\Omega$$

### **AC Analysis:**

$$g_m = \frac{I_C}{V_T} = \frac{1 \text{ mA}}{25 \text{ mV}} = 40 \frac{\text{mA}}{\text{V}}$$

$$r_\pi = \frac{\beta}{g_m} = \frac{200}{40 * 10^{-3}} = 5 \text{ k}\Omega$$

$$r_e = \frac{\alpha}{g_m} = \frac{0.995}{40 * 10^{-3}} = 24.88 \Omega$$

$$V_S = \frac{V_O}{A_{Vmid}} = \frac{2}{78.69} = 24.42 \text{ mV}$$

$$\text{Function generator amplitude} = \frac{V_S}{2} = \frac{25.42}{2} = 12.71 \text{ mV}$$

$$R_{S1} = 470 \Omega; R_{S2} = 56 \Omega$$

$$A_V = \propto g_m(R_C // R_L) \frac{(R_2 // R_S // r_\pi)}{R_{S1} + R_S + (R_2 // R_3 // r_\pi)} = 78.69 \pm 7.87 \text{ V/V}$$

$$R_S = \frac{\propto g_m(R_C // R_L)(R_2 // R_3 // r_\pi)}{A_V} - R_{S1} - (R_2 // R_3 // r_\pi) = 2.83 \text{ k}\Omega$$

$$R_{IN} = R_S + (R_2 // R_3 // r_\pi) = 6.51 \text{ k}\Omega$$

$$R_{OUT} = R_C = 5 \text{ k}\Omega$$

$$f_L = \frac{1}{2\pi} \left( \frac{1}{C_{OUT}R_{COUT}} + \frac{1}{C_{IN}R_{CIN}} + \frac{1}{C_B R_{CB}} + \frac{1}{C_E R_{CE}} \right) < 200 \text{ Hz}$$

$$\frac{1}{2\pi C_E R_{CE}} = 70\% f_L = 140 \text{ Hz}$$

$$\frac{1}{2\pi C_B R_{CB}} = \frac{1}{2\pi C_{IN} R_{CIN}} = \frac{1}{2\pi C_{OUT} R_{COUT}} = 10\% f_L = 20 \text{ Hz}$$

$$R_{CE} = R_E // \left( \frac{r_\pi + (R_2 // R_S // R_3)}{1 + \beta} \right) = 35.09 \text{ }\Omega$$

$$C_E = \frac{1}{2\pi * 140 * R_{CE}} = 32.4 \text{ }\mu\text{F}$$

$$\therefore \text{Use } C_E = 33 \text{ }\mu\text{F}$$

$$R_{CIN} = R_S + (R_2 // R_S // r_\pi) = 5.879 \text{ k}\Omega$$

$$C_{IN} = \frac{1}{2\pi * 20 * R_{CIN}} = 1.35 \text{ }\mu\text{F}$$

$$\therefore \text{Use } C_{IN} = 1 \text{ }\mu\text{F}$$

$$R_{COUT} = R_C + R_L = 19 \text{ k}\Omega$$

$$C_{OUT} = \frac{1}{2\pi * 20 * R_{COUT}} = 0.419 \text{ }\mu\text{F}$$

$$\therefore \text{Use } C_{OUT} = 1 \text{ }\mu\text{F}$$



$$R_{CB} = R_1 // R_2 // (r_e * (1 + \beta)) = 3.872 \text{ k}\Omega$$

$$C_B = \frac{1}{2\pi * 20 * R_{CB}} = 2.055 \text{ }\mu\text{F}$$

$$\therefore \text{Use } C_B = 10 \text{ }\mu\text{F}$$

$$f_L = \frac{1}{2\pi} \left( \frac{1}{(1 \text{ }\mu\text{F})(19 \text{ k}\Omega)} + \frac{1}{(1 \text{ }\mu\text{F})(5.879 \text{ k}\Omega)} + \frac{1}{(10 \text{ }\mu\text{F})(3.872 \text{ k}\Omega)} + \frac{1}{(33 \text{ }\mu\text{F})(35.09 \text{ k}\Omega)} \right)$$

$$\therefore f_L = 177 \text{ Hz}$$

$$f_H = \frac{1}{2\pi(\tau_1 + \tau_2 + \tau_3)} > 1 \text{ MHz}$$

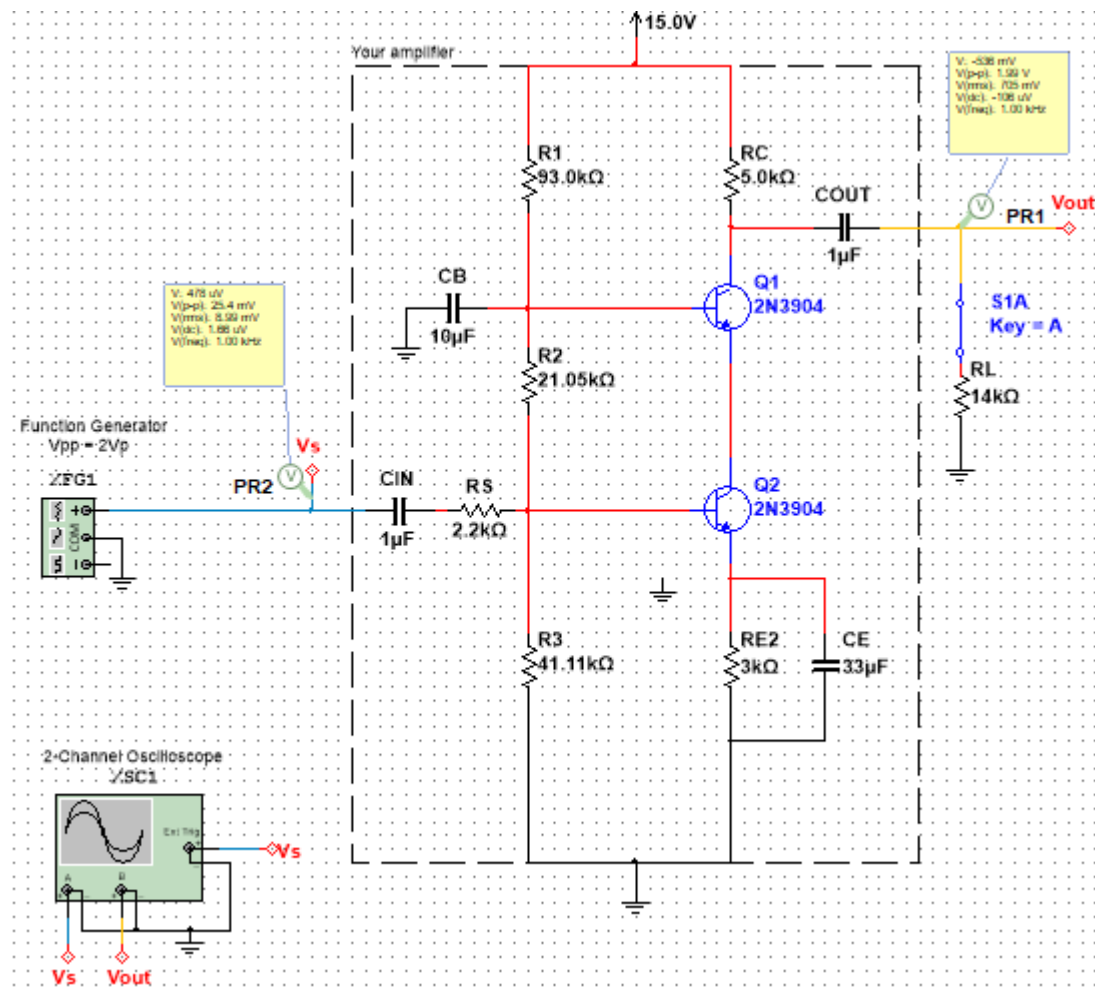
$$\tau_1 = (R_S // R_2 // R_3 // r_\pi)(2C_{\mu 1} + C_\pi) = 2.203 * 10^{-8} \text{ s}$$

$$\tau_2 = r_e(2C_{\mu 1} + C_\pi) = \frac{r_\pi}{\beta}(2C_{\mu 1} + C_\pi) = 4 * 10^{-10} \text{ s}$$

$$\tau_3 = (R_C // R_L)(C_{\mu 2} + C_L) = 1.474 * 10^{-8} \text{ s}$$

$$f_H = \frac{1}{2\pi((2.203 * 10^{-8}) + (4 * 10^{-10}) + (1.474 * 10^{-8}))} = 4.282 \text{ MHz}$$

Component	Design Value
$R_S$	2.2 k $\Omega$
$R_1$	93 k $\Omega$
$R_2$	21.05 k $\Omega$
$R_3$	41.11 k $\Omega$
$R_C$	5 k $\Omega$
$R_L$	14 k $\Omega$
$R_E$	3 k $\Omega$
$C_{IN}$	1 $\mu\text{F}$
$C_E$	33 $\mu\text{F}$
$C_B$	10 $\mu\text{f}$
$C_{OUT}$	1 $\mu\text{F}$



Function generator-XFG1

Waveforms

Signal options

Frequency: 1 kHz

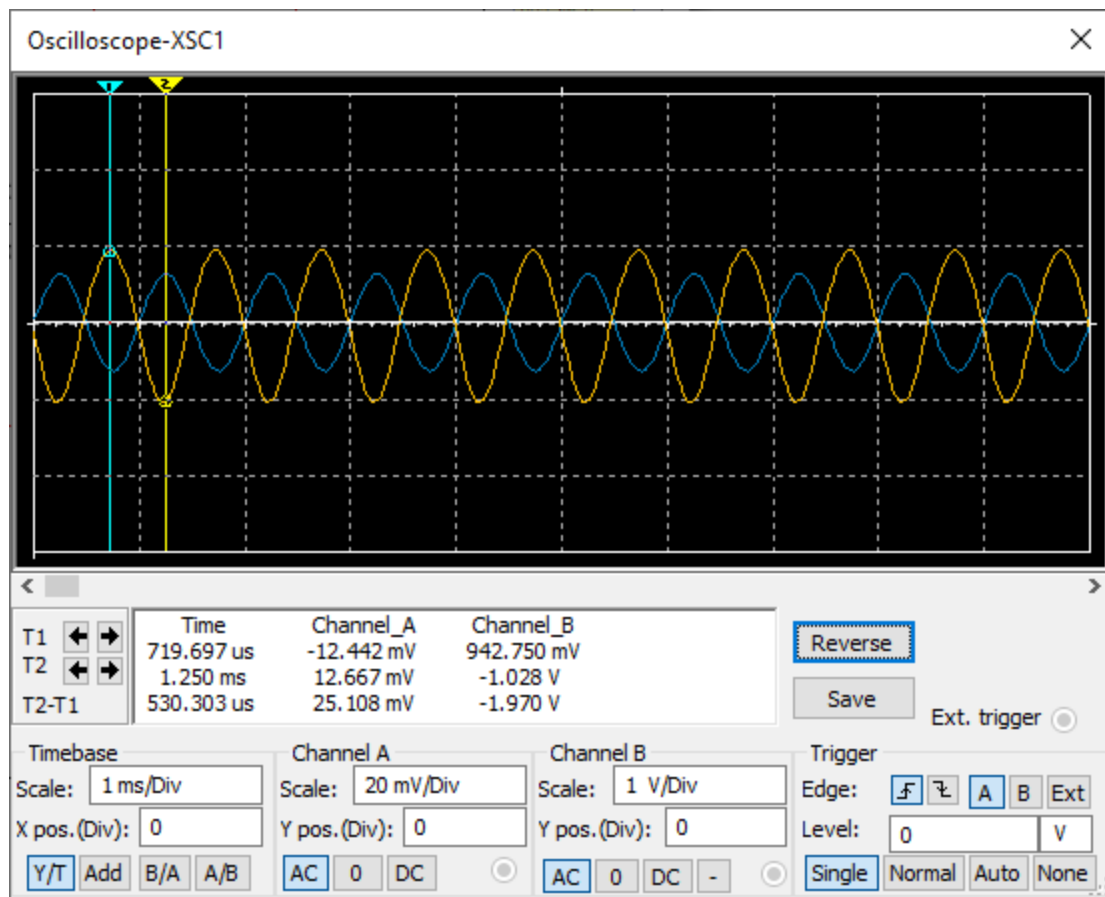
Duty cycle: 50 %

Amplitude: 12.71 mVp

Offset: 0 V

Set rise/Fall time

+ Common -



Variable	Simulated Result
$V_s$	25.4 mV
$V_s - V_x$	10.2 mV
$V_o$	1.99 V
$V_o'$	2.69 V
$I_{out}$	0.142 mA
$I_{in}$	4.61 $\mu$ A

- Measure  $R_{in}$ ,  $R_{out}$ ,  $A_{mid}$ ,  $f_H$  and  $f_L$ .

Variable	Prelab Results	Simulated Results
$A_v$	78.69 V/V	78.35 V/V
$R_{in}$	6.51 k $\Omega$	5.51 k $\Omega$
$R_{out}$	5 k $\Omega$	4.93 k $\Omega$
$f_L$	177 Hz	140 Hz
$f_H$	4.282 MHz	5 MHz

- Measure the input and output peak-peak voltage swings at a low input signal amplitude

(6 mV<sub>pk-pk</sub> to 82 mV<sub>pk-pk</sub>).

$V_{in}$ (mV)	$V_{out}$ (mV)
6	474
12	946
18	1420
22	1720
26	2030
30	2340
34	2650
38	2950
42	3240
48	3680
54	4110
62	4670
68	5070
76	5590
78	5710
82	5970

- Create a frequency response plot of gain vs frequency. Take 50 - 60 measurements starting lower than  $f_L$  and going significantly beyond than  $f_H$ .

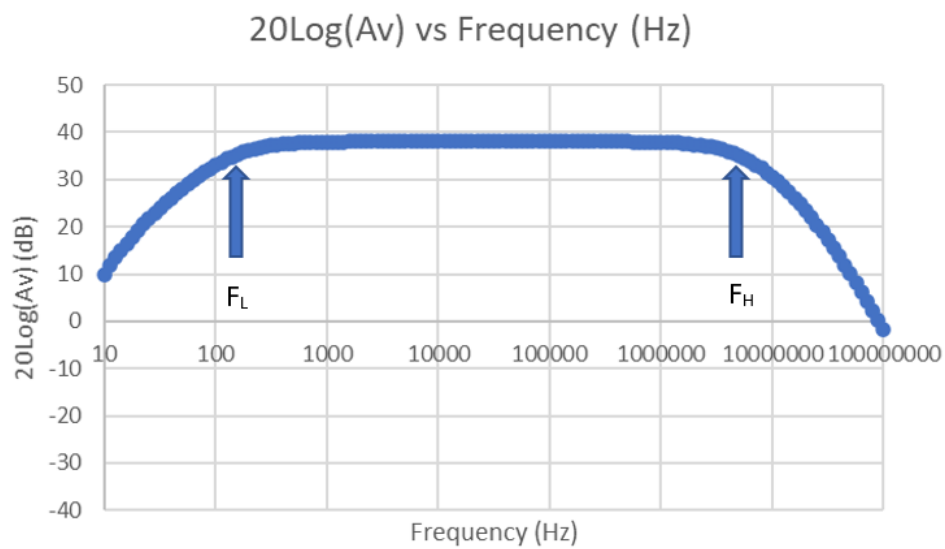
Frequency (Hz)	Vout (Vp)	Vin (Vp)	Vout (Vpk-pk)	Vin (Vpk-pk)	Av (V/V)	Gain = 20*log(Av) (dB)
1	0.0002112	0.01271	0.0004224	0.02542	0.016616833	-35.58903492
1.122018454	0.000265423	0.01271	0.000530845	0.02542	0.020882975	-33.60415261
1.258925412	0.00033636	0.01271	0.00067272	0.02542	0.026464204	-31.54682313
1.412537545	0.000429972	0.01271	0.000859944	0.02542	0.03382941	-29.41411143
1.584893192	0.000554416	0.01271	0.001108832	0.02542	0.043620465	-27.20619414
1.77827941	0.000720822	0.01271	0.001441644	0.02542	0.056713	-24.9263475
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17.7827941	0.099620619	0.01271	0.199241238	0.02542	7.837971584	17.88407369
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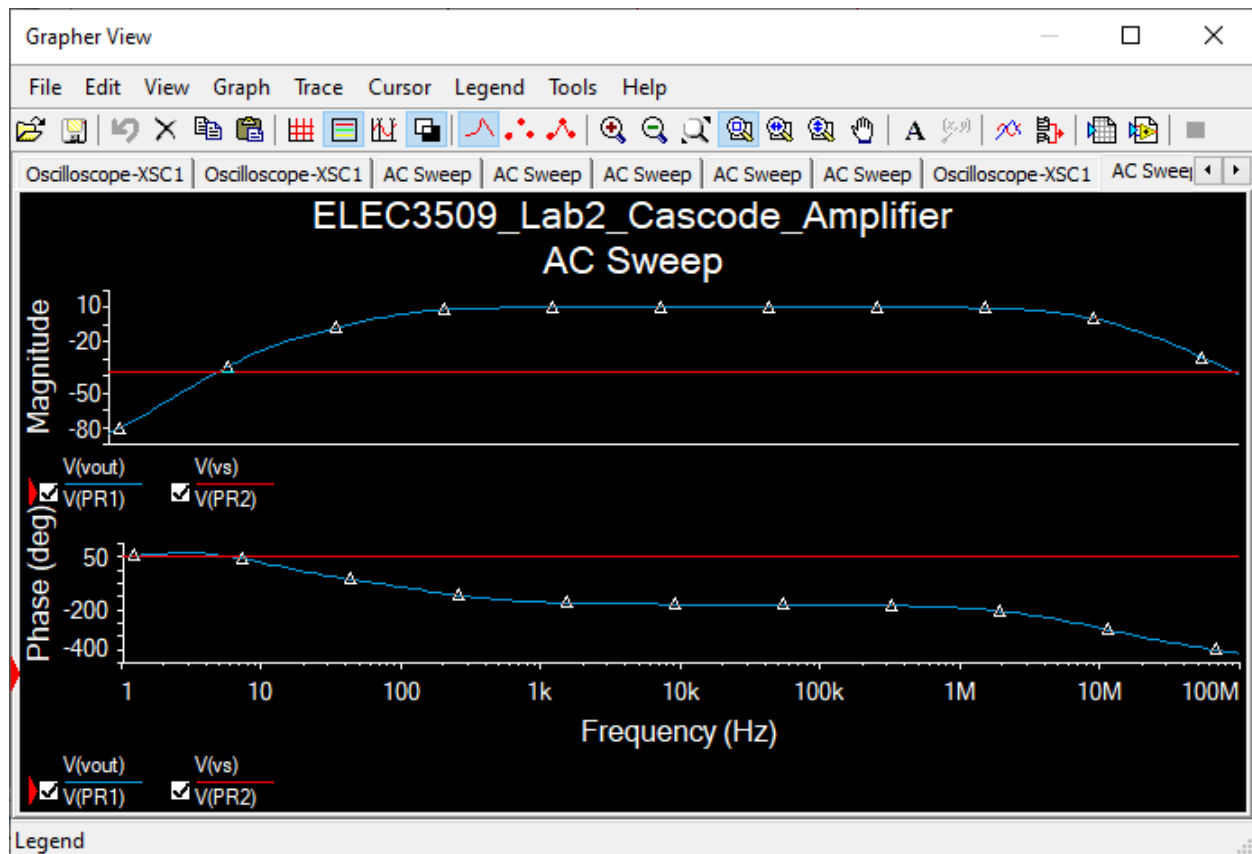
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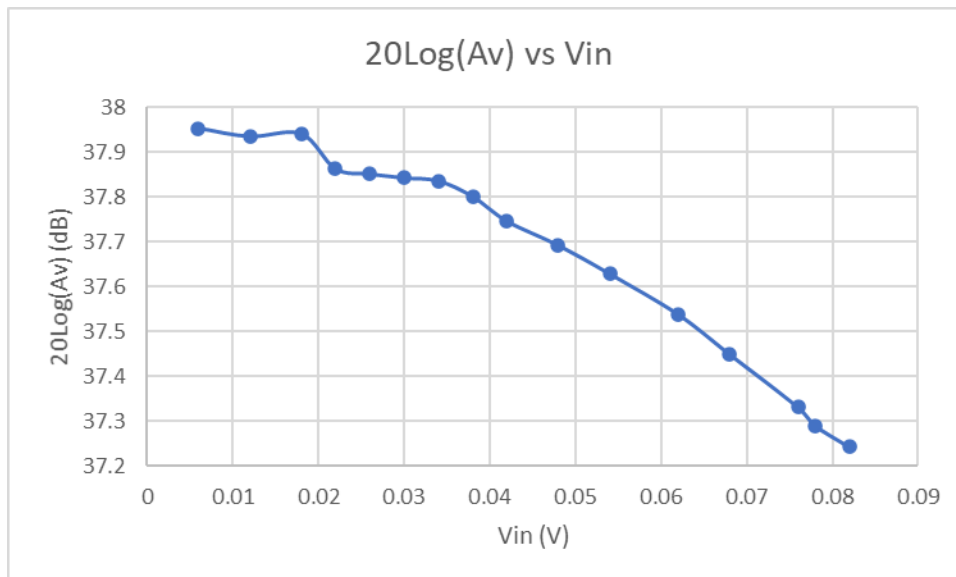


- As with all labs, you must show your pre-lab work and calculations for both parts of the lab. You need to show several plots and tables in order to clearly demonstrate that your circuit meets specifications. For  $R_{in}$  and  $R_{out}$ , show a table comparing measured and calculated results. Explain any significant differences.

		CC	CE	CB	CE-CB	CC-CB	CC-CE
Prelab	$R_{in} (k\Omega)$	25.50	3.30	25.90	3.30	6.00	25.50
	$R_{out} (k\Omega)$	0.046	5.60	5.60	5.60	5.60	5.60
Measured	$R_{in} (k\Omega)$	29.20	7.08	3.33	7.08	10.45	28.99
	$R_{out} (k\Omega)$	0.705	5.24	5.59	5.24	5.45	5.24

- Show a plot measuring gain as a function of input peak-peak voltage swing.

V <sub>in</sub> (V)	V <sub>out</sub> (V)	A <sub>v</sub> (V/V)	Gain = 20Log(A <sub>v</sub> ) (dB)
0.006	0.474	79	37.95254183
0.012	0.946	78.83333333	37.93419781
0.018	1.42	78.88888889	37.94031679
0.022	1.72	78.18181818	37.86211532
0.026	2.03	78.07692308	37.8504538
0.03	2.34	78	37.84189205
0.034	2.65	77.94117647	37.83533914
0.038	2.95	77.63157895	37.80076839
0.042	3.24	77.14285714	37.7459144
0.048	3.68	76.66666667	37.69213163
0.054	4.11	76.11111111	37.62896124
0.062	4.67	75.32258065	37.53850382
0.068	5.07	74.55882353	37.44998093
0.076	5.59	73.55263158	37.33196431
0.078	5.71	73.20512821	37.29083011
0.082	5.97	72.80487805	37.24320957



The location where the gain differs from the low signal gain by 1 dB is when  $V_{in} = 25.42$  mV and Gain = 29.56 dB

## DISCUSSION

Overall, prelab values and measured values were very close to each other, and in day 2 all the simulated results successfully achieved the cascode amplifier requirements.

Comparing the three fundamental BJT configurations, we can conclude that CE has the highest mid-band gain and lowest high cut-off frequency also it was determined that it can be used to amplify weak signals, CC has the lowest mid-band gain and highest input impedance and lowest output impedance was determined that it can as a voltage buffer, CB has the lowest input impedance and lowest low cut-off frequency and highest high cut-off frequency also it was determined that it can act as a current buffer.

While out of the three 2-transistor configurations, the CC-CE has the highest mid-band gain and highest input impedance and it acts like a good voltage amplifier circuit, CE-CB has moderate mid-band gain and lowest high cut-off frequency and it acts as moderate amplifier circuit, while CC-CB has the lowest mid-band gain and highest low cut-off frequency and it acts like a poor voltage amplifier circuit.