

25 Spring ECEN 607: Advanced Analog Circuit Tech

Design Post-lab Report

Lab3: Op Amp Design - I

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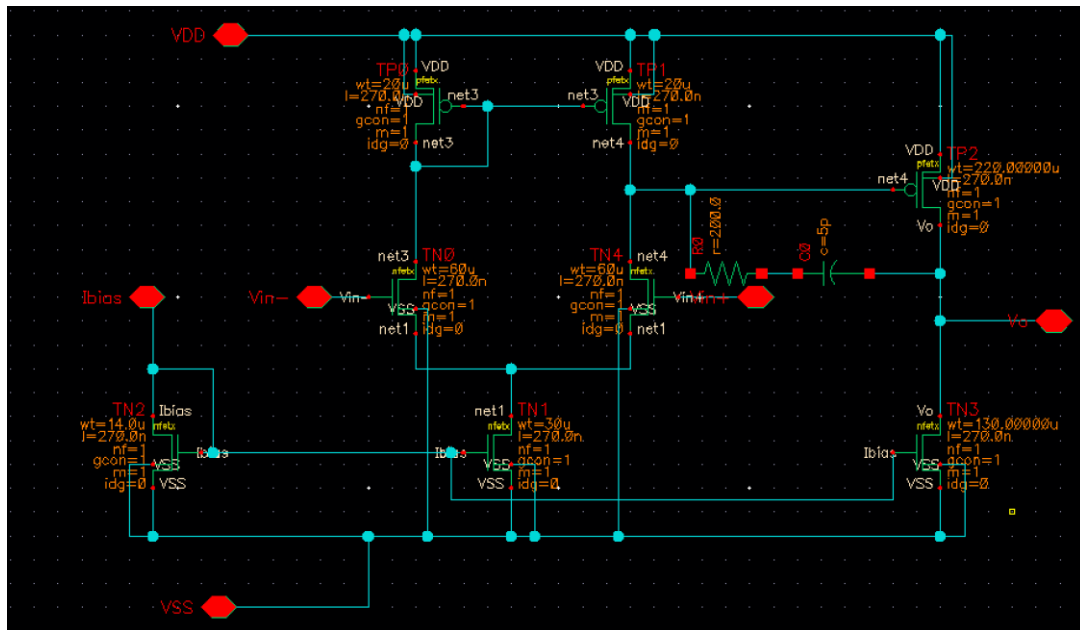
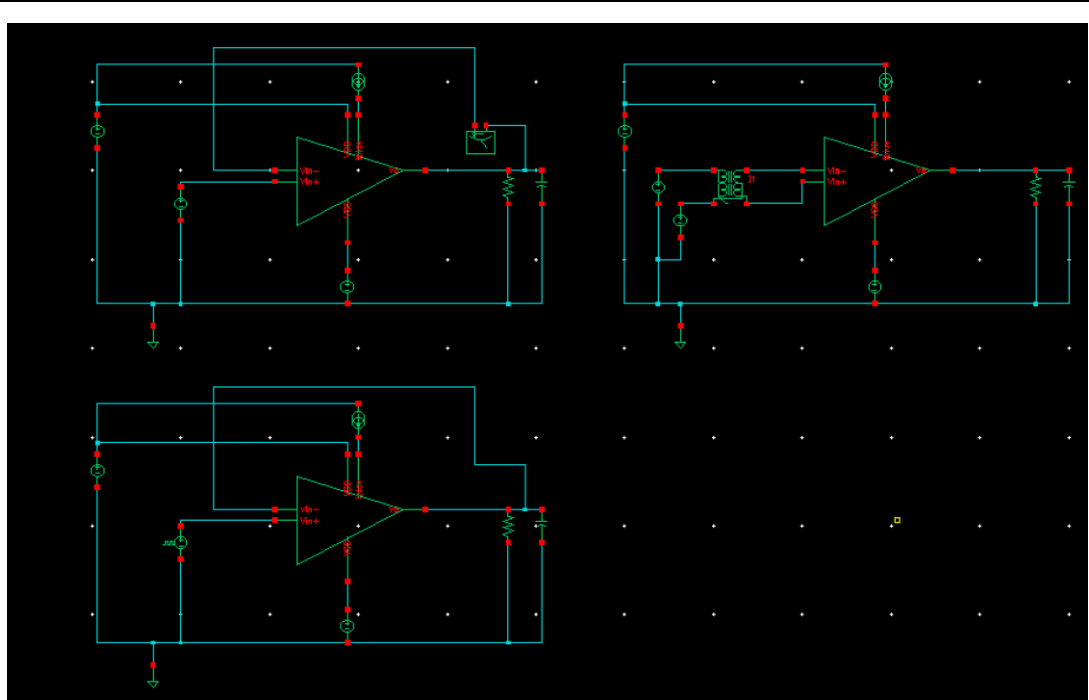
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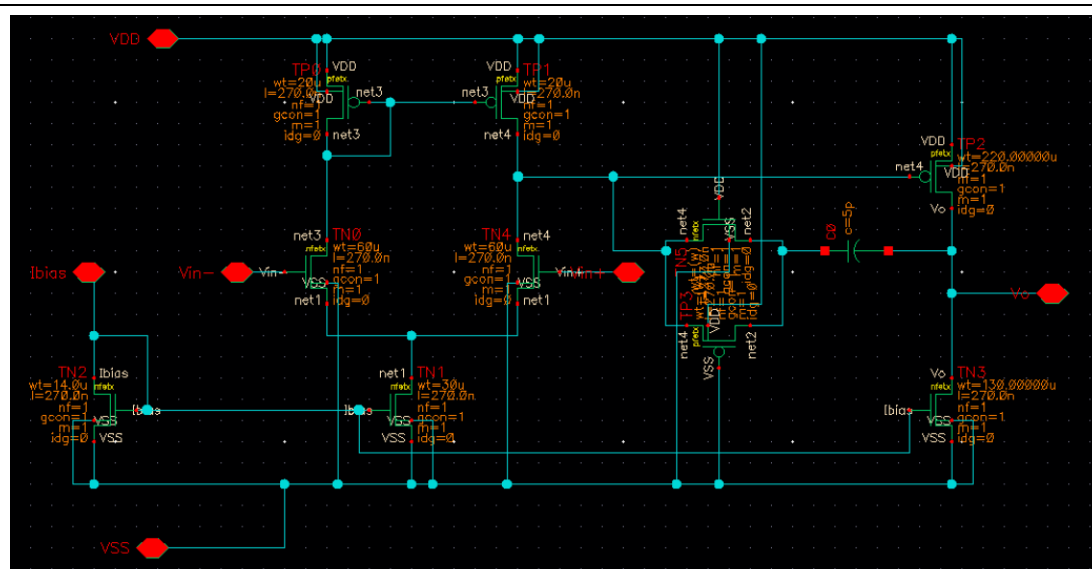
Professor: Jose Silva-Martinez

TA: Yoon, Sung J

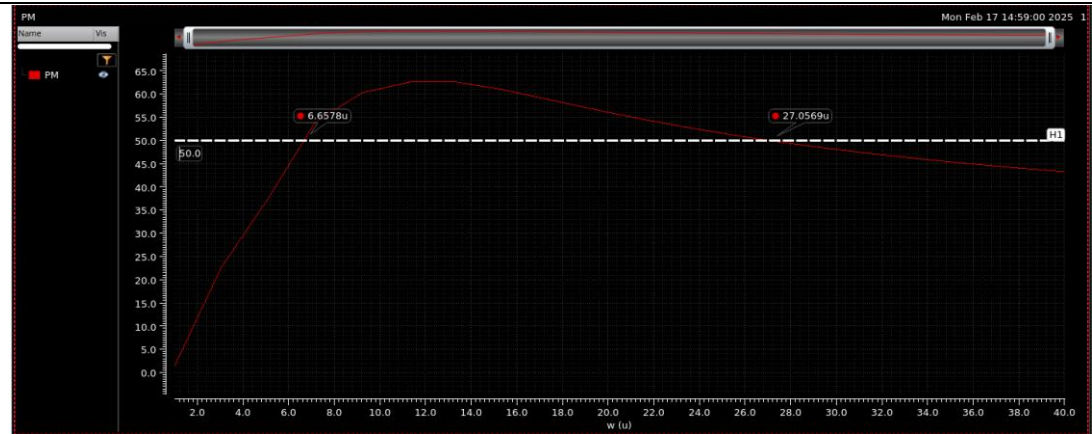
Table 2-1: Design specifications

| | |
|--|-----------------|
| $V_{DD}-V_{SS}$ | 1.8V |
| A_{v0} | > 40 dB |
| GBW | > 2 MHz |
| PM | > 45° |
| Output Swing | > 1 V |
| C_L | 20 pF |
| R_L | 20k Ω |
| Integrated input referred noise (10 Hz – 2 MHz) | < 30 μ Vrms |



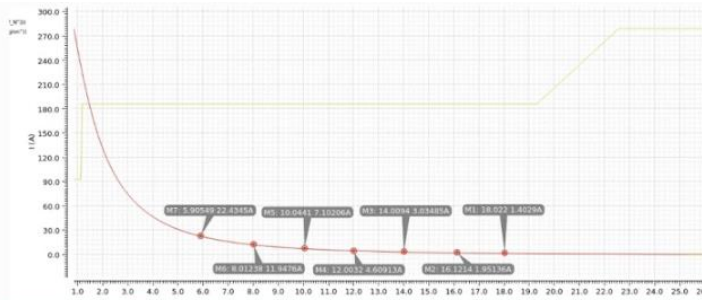


W= 27u for two triode mode transistor

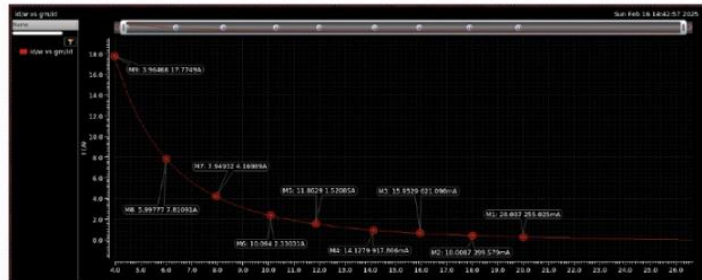


1. Simulate the circuit from the prelab and adjust the transistor sizes accordingly.

Nmos



pmos



make g_{m2} bigger for lower I_{D2}

$$\frac{I_{D2}}{W_2 (b_{00\mu})} = g (16) = 1.95 \quad g_{m2} = 1.87m \quad W_{tail} \left(\frac{g_m}{\frac{g_m}{20}} = 10 \right) = 32.95$$

$$I_{D2} = 117\mu A \quad I_{tail} = 234\mu A$$

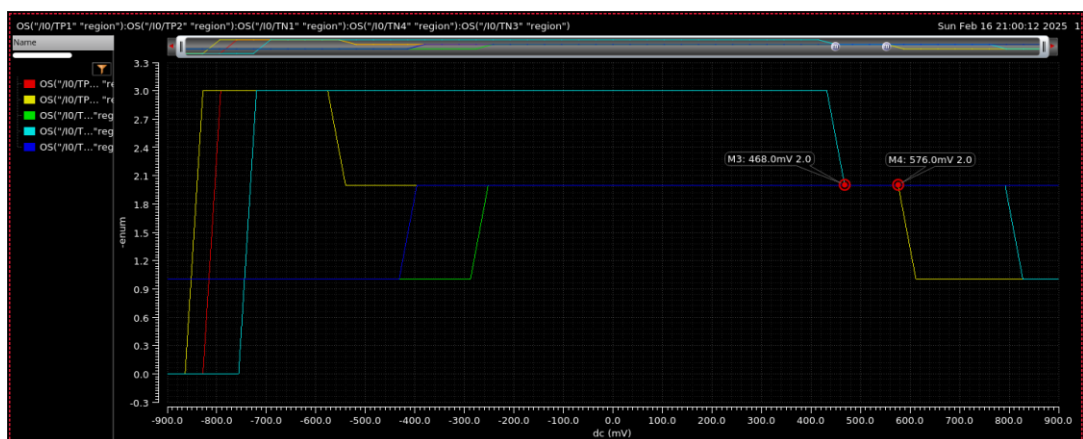
$$\frac{I_D (100\mu)}{W} = g(10) = 7.1 \quad W_{bias} = 14.08 \quad \left(\frac{32.95}{14.08} = \frac{234\mu A}{100\mu A} \right)$$

$$GBW = \frac{1.87}{2\pi \times C_c} = 59M Hz$$

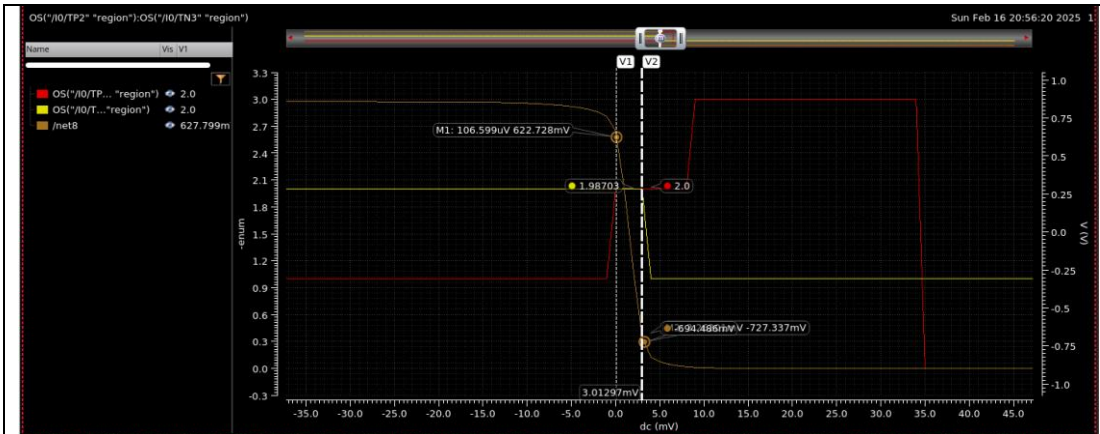
$$If \quad f_{p2} = GBW \quad (PM 45) \quad \frac{g_{m8}}{2\pi C_L} = GBW \quad g_{m8} = 7.4m$$

$$set \quad \frac{g_{m8}}{I_{D8}} = 8 \quad I_{D8} = 926\mu A \quad W_8 = \frac{I_{D8}}{g(8)} = 222\mu$$

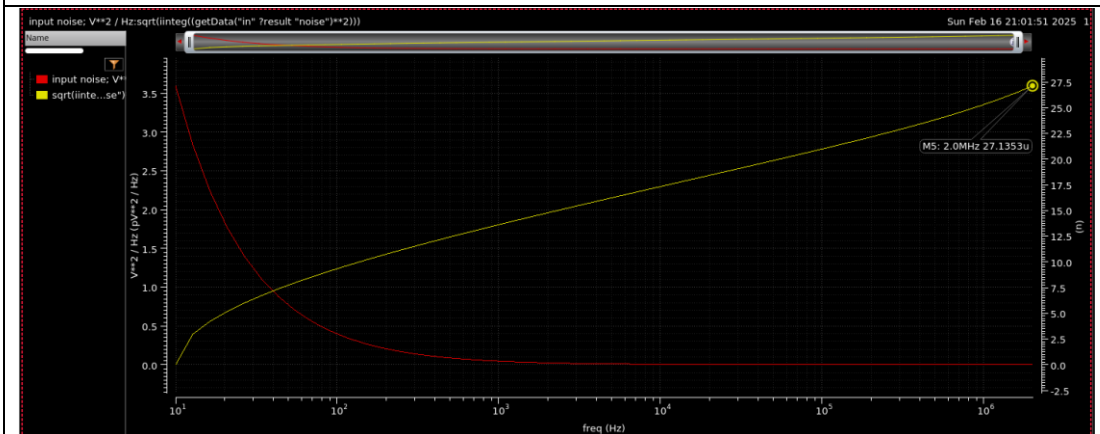
$$I_{D7} = 926\mu A \quad \frac{I_{D7}}{W_7} = 10 \quad W_7 = 130\mu \quad \left(\frac{130\mu}{14\mu} = \frac{926\mu}{100\mu} \right)$$



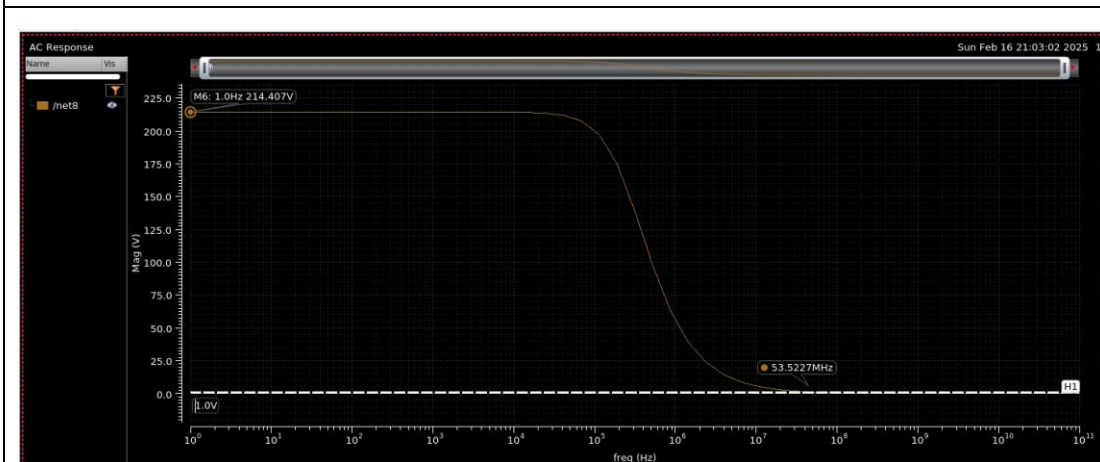
Vcm range



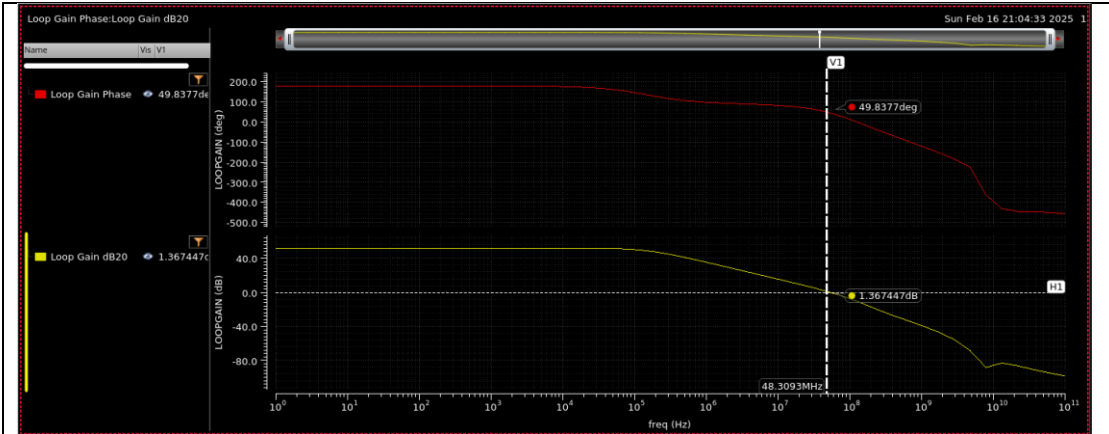
Output swing: $0.6 \sim -0.7 = 1.3\text{v}$



Input ref noise



GBW & Av0



PM

ADE XL Test Editor - gm8@n04-hades.olympus.ece.tamu.edu

Session Setup Analyses Variables Outputs Simulation Results Tools Help

cadence

Design Variables

| Name | Value |
|-----------|-------|
| 1 VDD | 900m |
| 2 VSS | -900m |
| 3 p5vonly | 0 |

Analyses

| Type | Enable | Arguments |
|------|-------------------------------------|-----------|
| 1 dc | <input checked="" type="checkbox"/> | t |

Outputs

| Name/Signal/Expr | Value | Plot | Save | Save Options |
|------------------|-------|-------------------------------------|-------------------------------------|--------------|
| 1 gm8 | 8.6m | <input checked="" type="checkbox"/> | <input checked="" type="checkbox"/> | |

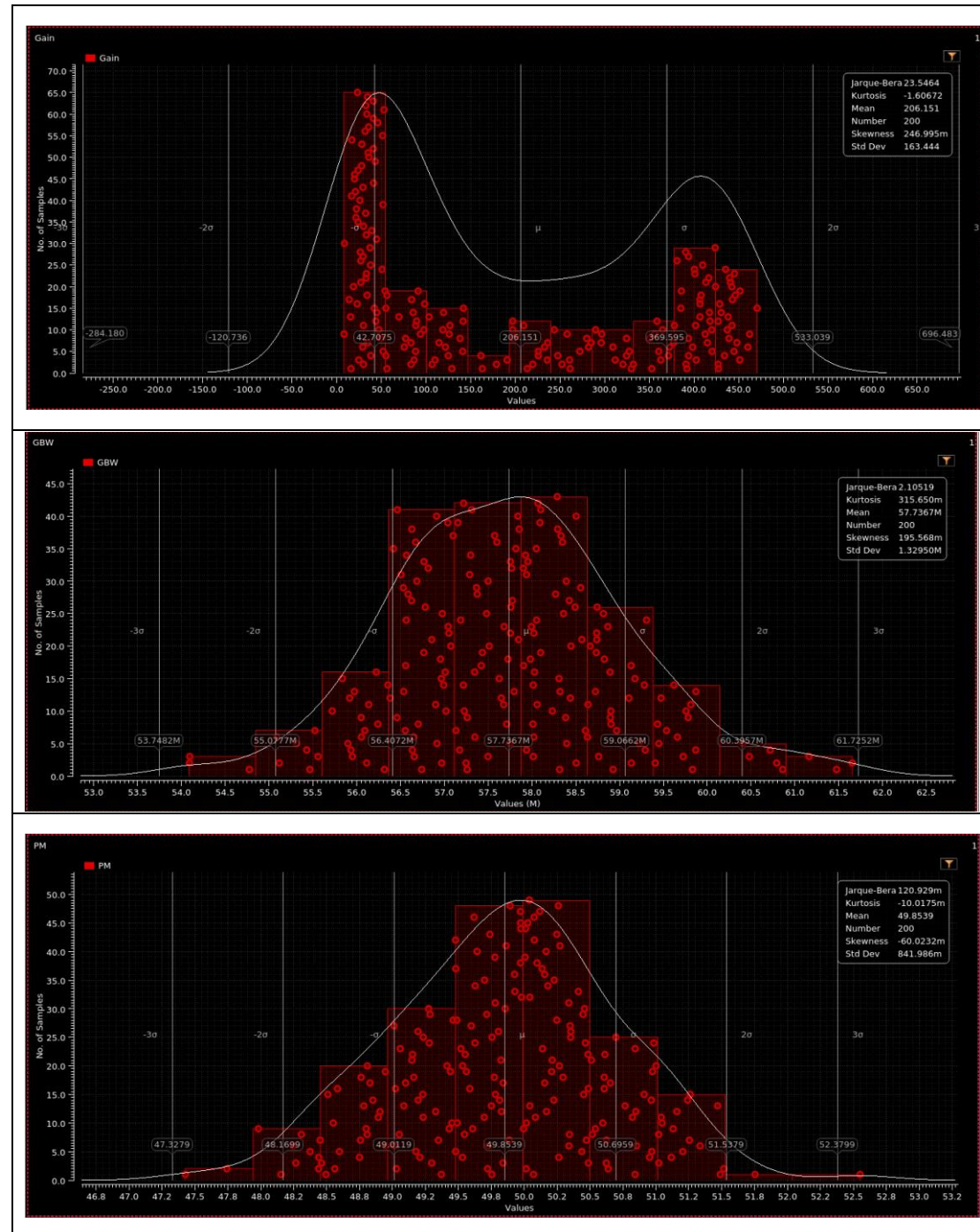
> Results in ...r_harrison900531/gm8/simulation

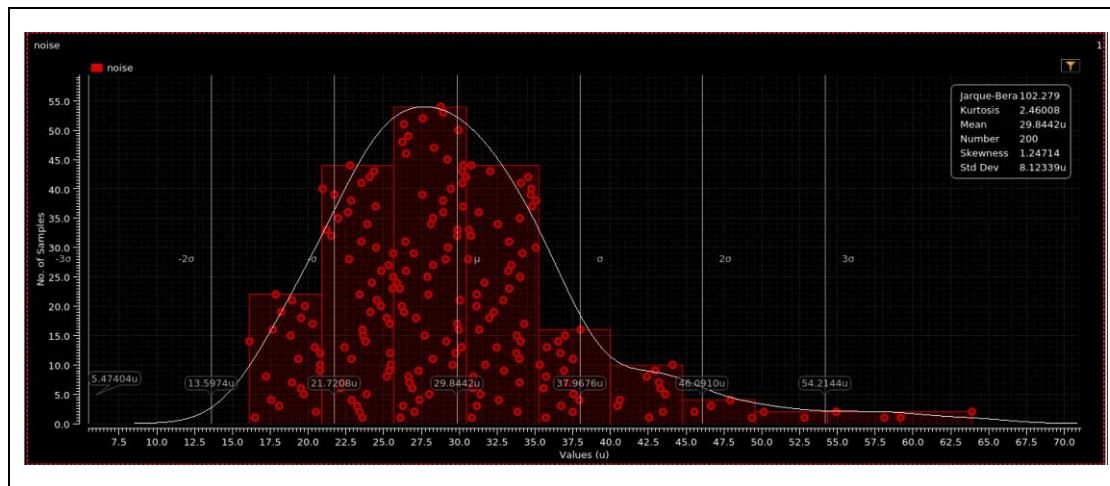
9(38) | Simulator ... | Status: Ready | T=27 C | Simulator: spectre aps | State: gm8_active

Gm8

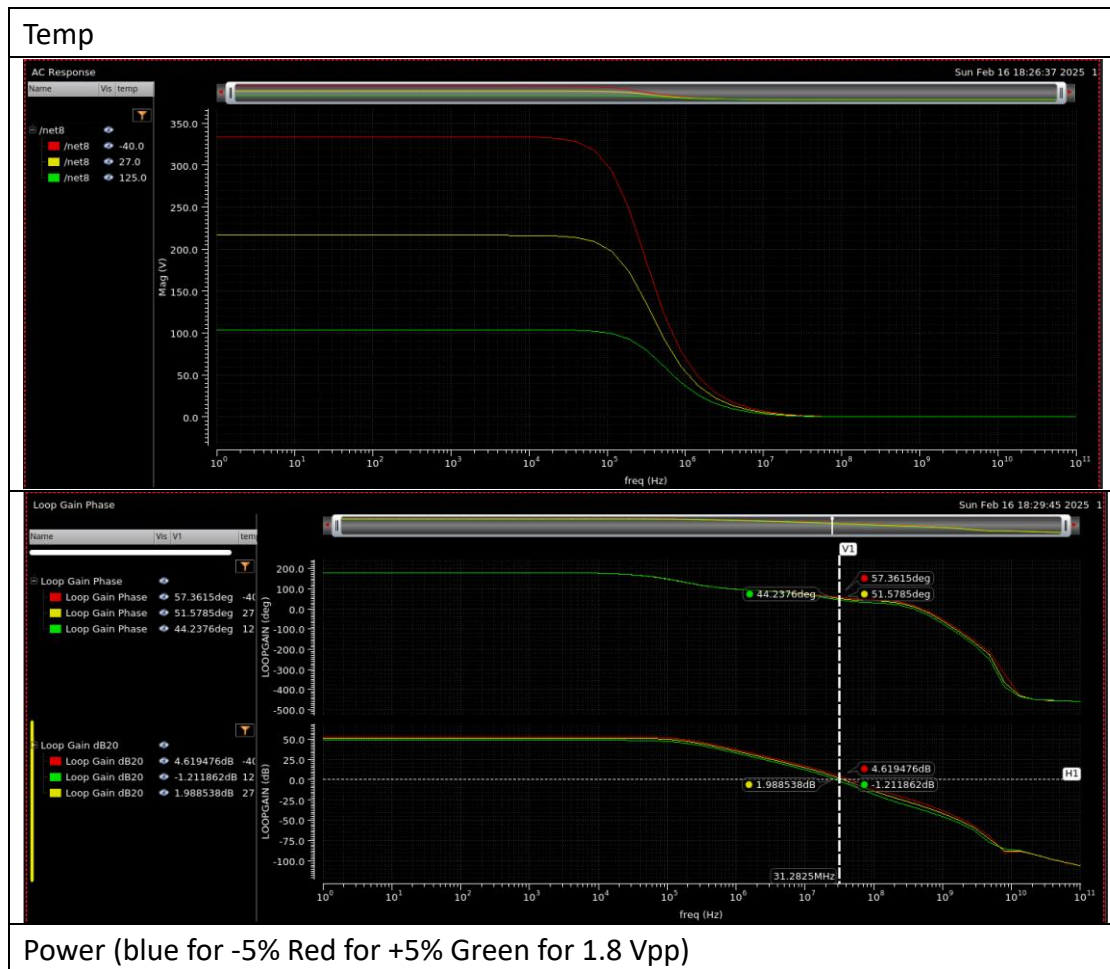
| Test | Output | Nominal | Spec | Weight | Pass/Fail |
|-----------|----------|---------|------|--------|-----------|
| AC | /net8 | | | | |
| AC | Gain | 214.4 | | | |
| AC | net8 | | | | |
| stability | GBW | 61.95M | | | |
| stability | PM | 45.49 | | | |
| stability | /I/O/TP2 | | | | |
| noise | noise | 27.14u | | | |
| gm8 | gm8 | 8.6m | | | |
| gm8 | /I/O/TP2 | | | | |

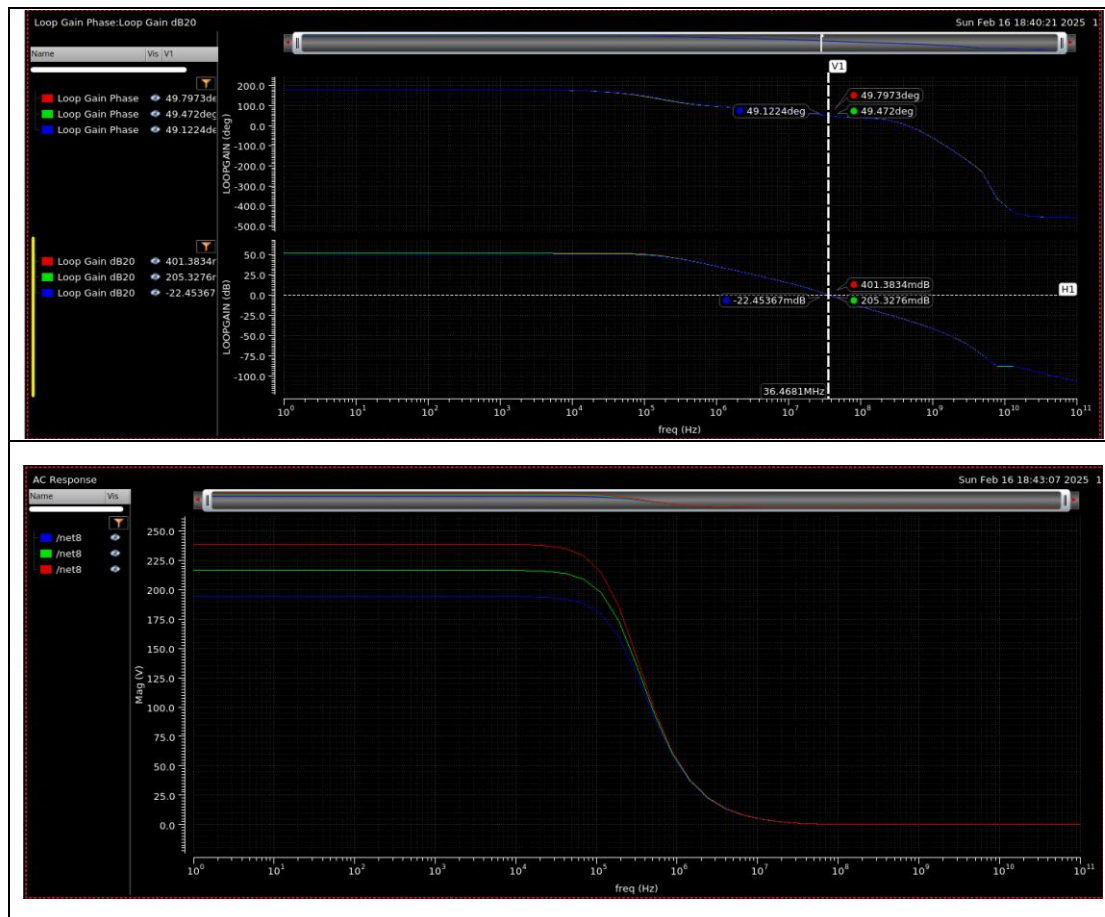
- Run MC simulation and adjust your design to target a 3-sigma yield for DC Gain, GBW, Phase Margin and input referred noise until all specifications are met. Provide necessary screen shots such that you clearly show the specs are met. (including histogram plots).



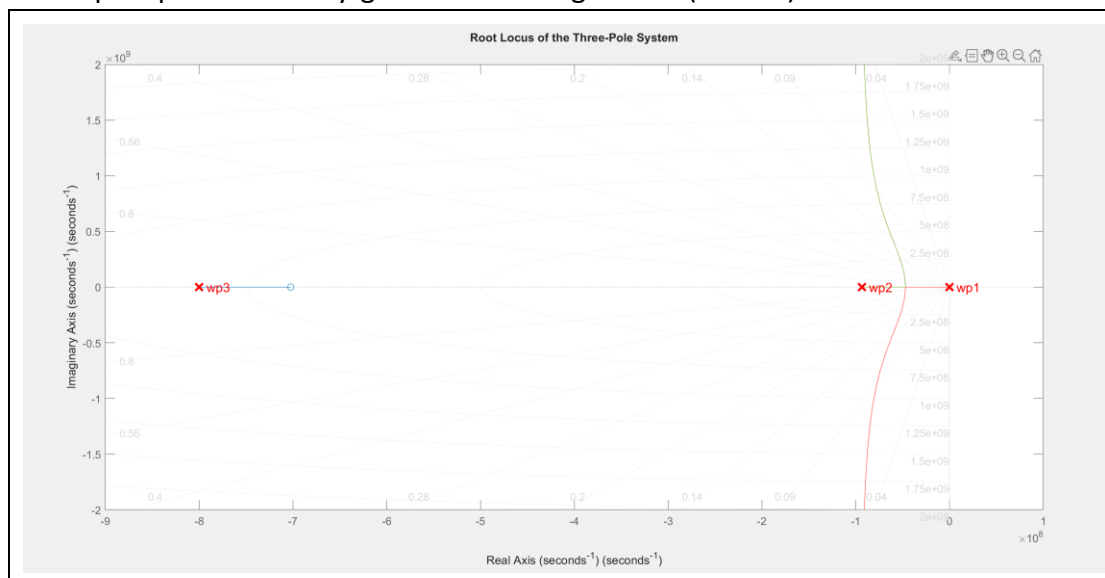


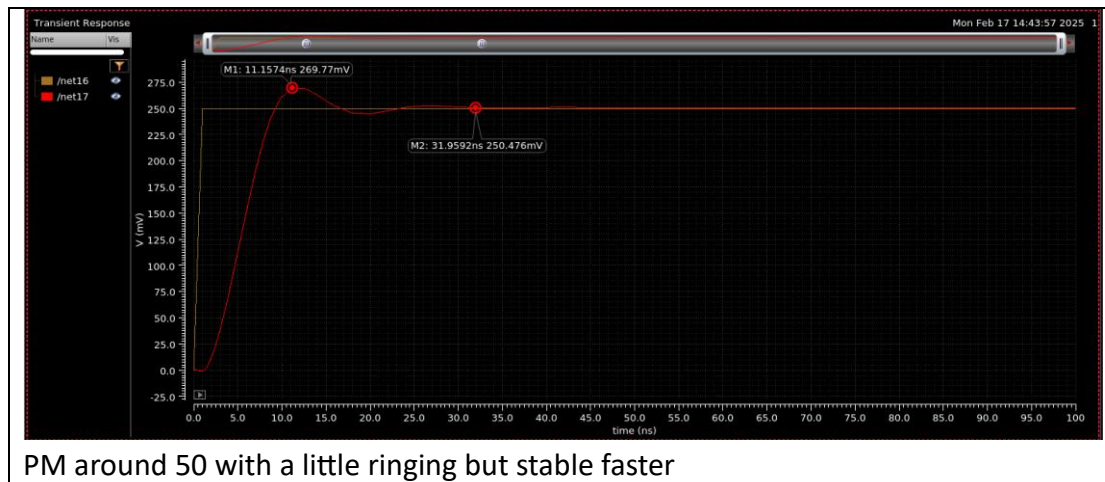
3. Simulate the supply voltage variations (+5, -5%) and temperature (-40, 125C) for DC Gain, GBW, Phase Margin. You do not have to adjust the design if specs are not met.





4. Plot the root-locus of the transfer function from pre-lab (Over design the RHZ, $2 \cdot g_{m8}/C_c$). Use MATLAB function `rlocus`. Comment on the result with 250mV step response on unity-gain buffer configuration ($\beta=1$).





● Conclusion

Miller compensation is essential for stabilizing OPAs but introduces an undesirable **Right-Half Plane (RHP) zero**, which degrades phase margin and stability. To counteract this, **zero canceling** techniques introduce a **Left-Half Plane (LHP) zero** via a series resistor, effectively mitigating the RHP zero's impact. Proper **Miller RC compensation** shapes the **root locus**, optimizing pole placement, loop stability, and transient response. By carefully designing **zero placement**, designers can enhance **bandwidth, phase margin, and gain performance**, ensuring a well-behaved amplifier with minimal overshoot and ringing.

● Appendix

```

Current Folder
Name -
~$EN607 postlab3.docx
ECEN607 postlab3.docx
ECEN607 prelab3 Yu_Hao Chen.pdf
prelab.docx
untitled.m

Details
~$EN607 postlab3.docx
ECEN607 postlab3.docx
ECEN607 prelab3 Yu_Hao Chen.pdf
prelab.docx
untitled.m

Editor - C:\Users\harm\OneDrive\桌面\TAMU\ECEN 607 Advanced Analog circuit\lab3\untitled.m
read_sparam.m channel_data.m xfr_fn_to_imp.m untitled.m +

1 klc; clear; close all;
2
3 gm2 = 1.87e-3;
4 gm8 = 8e-3;
5 R1 = 65e3;
6 R2 = 20e3;
7 Cc = 5e-12;
8 CL = 20e-12;
9 R = 2/gm8; % move to LHP zero
10
11 % === calculate poles and zero ===
12 wp1 = -1 / (gm2 * R1 * R2 * Cc); % wp1
13 wp2 = -gm2 / CL; % wp2
14 wp3 = -1 / (R * Cc); % wp3
15 wz = 1 / (Cc * ((1/gm2)-R)); % RHZ WZ
16
17 % === H(S) ===
18 s = tf('s');
19 H_corrected = (s + wz) / ((s - wp1) * (s - wp2) * (s - wp3)); % H(s)
20
21 % === Root Locus ===
22 figure;
23 rlocus(H_corrected); hold on;
24 title('Root Locus of the Three-Pole System');
25 xlabel('Real Axis (seconds^{-1})');
26 ylabel('Imaginary Axis (seconds^{-1})');
27
28 title('Root Locus of the Three-Pole System');
29 xlabel('Real Axis (seconds^{-1})');
30 ylabel('Imaginary Axis (seconds^{-1})');
31 grid on;
32
33 % === mark poles and zero ===
34 plot(real([wp1, wp2, wp3]), imag([wp1, wp2, wp3]), 'rx', 'MarkerSize', 10, 'LineWidth', 2);
35 plot(real(wz), imag(wz), 'go', 'MarkerSize', 10, 'LineWidth', 2);
36
37 % === marker ===
38 text(real(wp1), imag(wp1), 'wp1', 'FontSize', 12, 'Color', 'r');
39 text(real(wp2), imag(wp2), 'wp2', 'FontSize', 12, 'Color', 'r');
40 text(real(wp3), imag(wp3), 'wp3', 'FontSize', 12, 'Color', 'r');
41 text(real(wz), imag(wz), 'wz', 'FontSize', 12, 'Color', 'g');
42
43 hold off;

```