Computer-Aided VLSI System Design, Fall 2018

Final Project - LED Display Controller

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A. Requirements

1. Layout Area

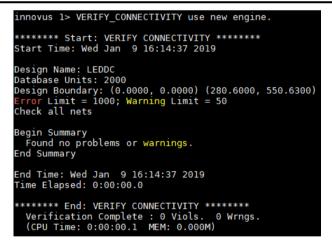
Area = 154506.78 um^2

2. Verification Results

DRC Verification Result (Verify Geometry)

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LVS Verification Result (Verify Connectivity)
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DRC Violation Count = 0



LVS Violation Count = 0

Simulation Results

RTL Level Simulation Result

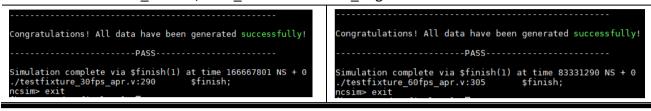
ncverilog testfixture 30|60fps.v LEDDC.v +notimingchecks +access+r

Gate Level Simulation Result

ncverilog +ncmaxdelays testfixture_30|60fps.v LEDDC_syn.v sram_512x16/sram_512x16.v sram 256x16/sram 256x16.v tsmc13 neg.v +define+SDF +access+r

Post-layout Simulation Result

ncverilog +ncmaxdelays testfixture_30|60fps_apr.v CHIP_apr.v sram_512x16/sram_512x16.v sram 256x16/sram 256x16.v tsmc13 neg.v +define+SDF +access+r



Applied Technique Explanation
See Section B. and Section C. for more details.

B. Design Description

The block diagram of our design is shown in the Figure 1.

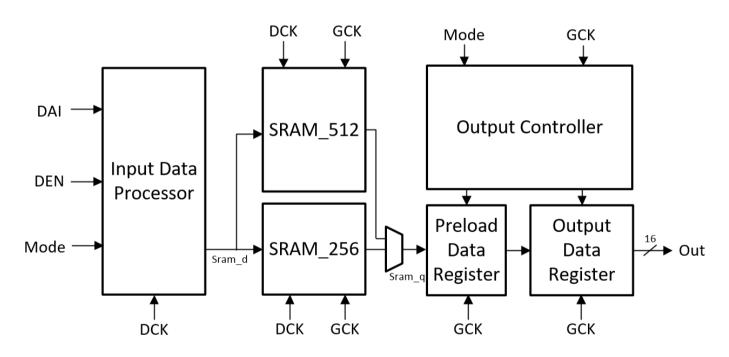


Figure 1. Block diagram of our design

1. Input Processor

Since the number of rising edges of DCK when the DEN is high is exactly 16. We use a shift register to store the data bits sequentially. When the DEN is low, we store the data into the sram.

2. Output Controller and Register

The finite state of our design is shown in the Figure 2.

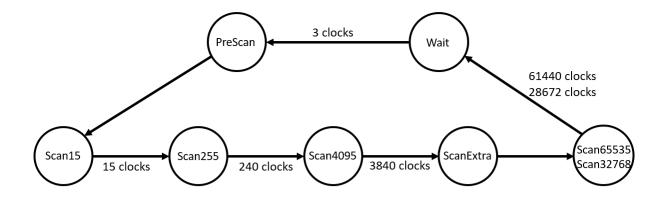


Figure 2. The Finite state machine of the output for 60FPS/30FPS

According to the specification, the number of clocks is fixed for any input sequence. Therefore the transition of states entirely depends on the number of clocks.

a. 30 FPS mode

Each output in the scanlines is determined by a 16 bits pixel value. To reduce the number of the output registers, we split the original 65536 clocks into 5 state. The number of clocks in each state is listed in the table below. The number of 1's in the outputs of each state is therefore determined by the corresponding 4 bits of the pixel value as shown in the Figure 3. Besides, before every state, the values of all pixels must be prepared. As a result we need additional preload registers for each pixel. These additional register is inevitable for either 4-bit or 16-bit register. Therefore, instead of 16*2 bits, only 4*2 bits is required for each output.



Figure 3. 4-bit data required in each state in 60 FPS mode

State	Clocks	Accumulated Clocks
Scan15	15	15
Scan255	240	255
Scan4095	3840	4095
ScanExtra	1	4096
Scan65535	61440	65536

Table 1. the PWM clock distribution in each state (30FPS)

b. 60 FPS mode

In the 60 FPS mode, the concept of our design is similar to that in the 30 FPS mode. However, in 60 FPS mode, when the pixel value is odd, there will be an additional bit in the first round compared to the second round. Hence, we store the data slightly differently in the 60 FPS mode by rotating the data to right. As the result, The first bit is an indicator of whether the value is odd and this indicator will be used to determine the output in the ScanExtra state, as shown in the Figure 5.

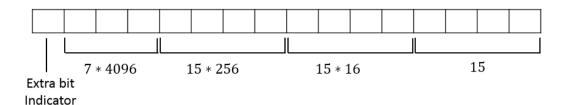


Figure 4. 4-bit data required in each state in 60 FPS mode

State	Clocks	Accumulated Clocks
Scan15	15	15
Scan255	240	255
Scan4095	3840	4095
ScanExtra	1	4096
Scan32767	28672	32768

Table 2. the PWM clock distribution in each state(60FPS)

3. Sram Control:

In the 60FPS mode, the data have to be read from the sram again in the second round, which means all data of the frame have to kept intact in the sram until the second round begins. Therefore a 512-word sram is insufficient because the data of next frame is read in simultaneously.

To address this issue, we combine a 512-word sram and a 256-word sram into a 768-word sram. The address of sram is 0-767.

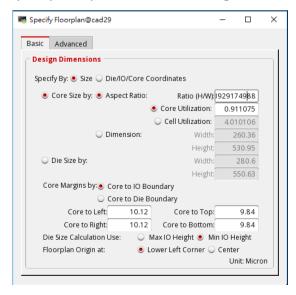
To decode the address, since when the MSB is 1, the address is larger than 511, we can easily use its MSB as the chip enable, and the real address to the sram is the remaining bits.

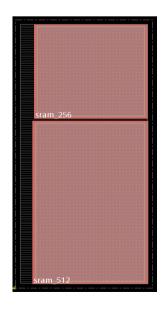
C. APR Details & Results

Most of the settings and APR flow follow the ones described in Lab7 and Lab8. We focus on planning the position of macros (e.g. SRAMs) and techniques for fixing DRC errors.

4. Floorplan Design

a. Specify floorplan & SRAM locating





b. Halo settings{Top, Bottom, Left, Right} = {6um, 6um, 5um, 5um} for both sram 256 & sram 512

5. Power Design

a. Power ring

Most of the settings follow those in Lab7, except setting "Number of bits" to 1 in order to reduce the chip margin, and also the design is small so such power planning is enough.

b. Power stripe

Set the stripe at the middle of the design, and "Number of bits" is set to 1.

6. Routing & Fixing DRC errors

- a. "Cut short error" is the most frequently encountered DRC error, which occurs when an via is directly set between SRAM pin (metal 3) and wire (metal 4). A reasonable work-around is to insert a "routing blockage" if metal 3 and metal 4 covering all SRAM pins before nano-route, and thus forces the EDA tool to avoid using metal 4 wire to gain access to SRAM.
- b. We found that the floorplan design strongly affects the result of routing. The way we locate SRAMs and the halo we set will latter cause the routing to success or fail with lot of DRC errors.
- c. As doing routing, we found that there sometimes exist hold time violation. However, such violation may not be activated referring to the functionality specified by the project. For example, some violated path is from mode pin to register, but the signal is stable during the usage of module. Another example is the path from DEN to registers, however the DEN is set before the clock triggered, and thus not influences the functionality correctness.

7. Layout Result

