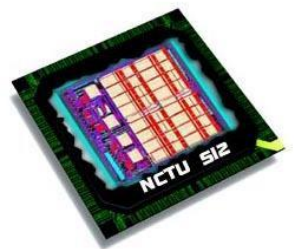


# SEQUENTIAL CIRCUITS

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**NCTU-EE IC LAB FALL-2023**



Lecturer: Yen-Teng Chuang

# Outline

- ✓ **Section 1 Sequential Circuits**
- ✓ **Section 2 Finite State Machine**
- ✓ **Section 3 Timing**
- ✓ **Section 4 Synthesis and Design Compiler**
- ✓ **Section 5 Generate and for loop**



# Outline

## ✓ Section 1 Sequential Circuits

✓ Introduction

✓ Syntax

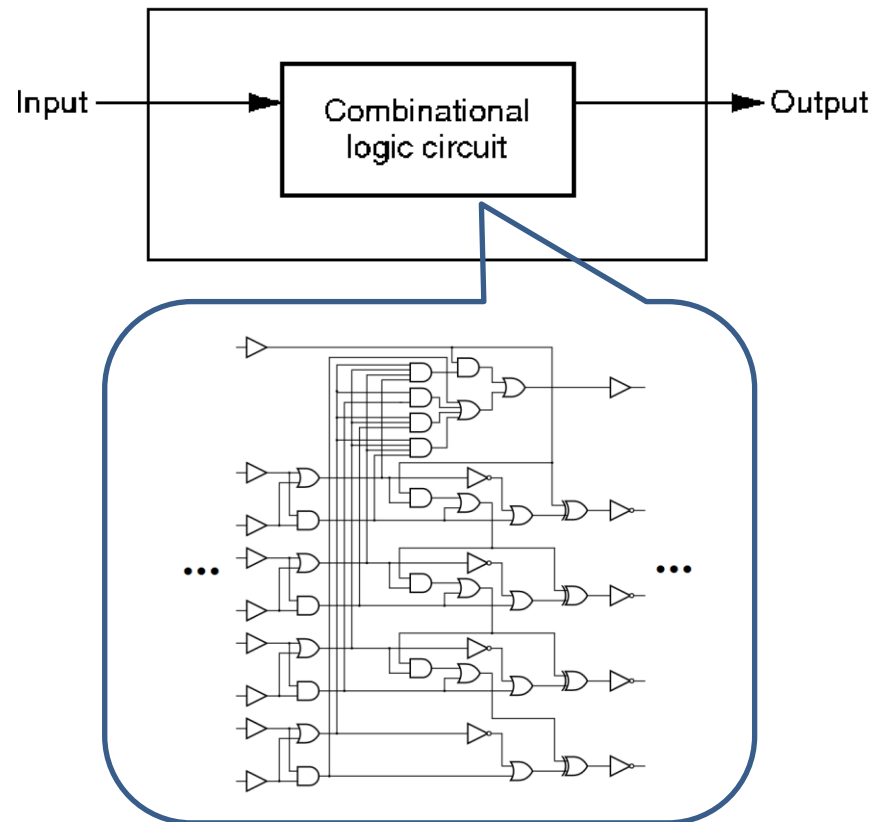
✓ Reset

✓ Coding Style



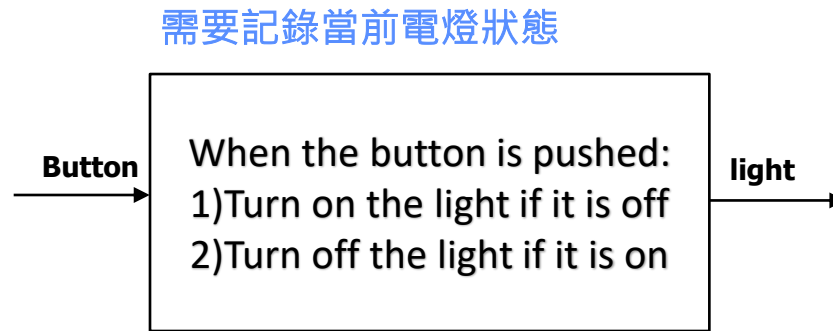
# Motivation

- ✓ **Progress so far : Combinational circuit**
  - Output is only a function of the **current** input values



# Motivation

- ✓ What if you were given the following design specification:



- ✓ What makes this circuit so different from we've discussed before?

“State”

為了記錄當前狀態，  
因此會需要用到state  
這個概念



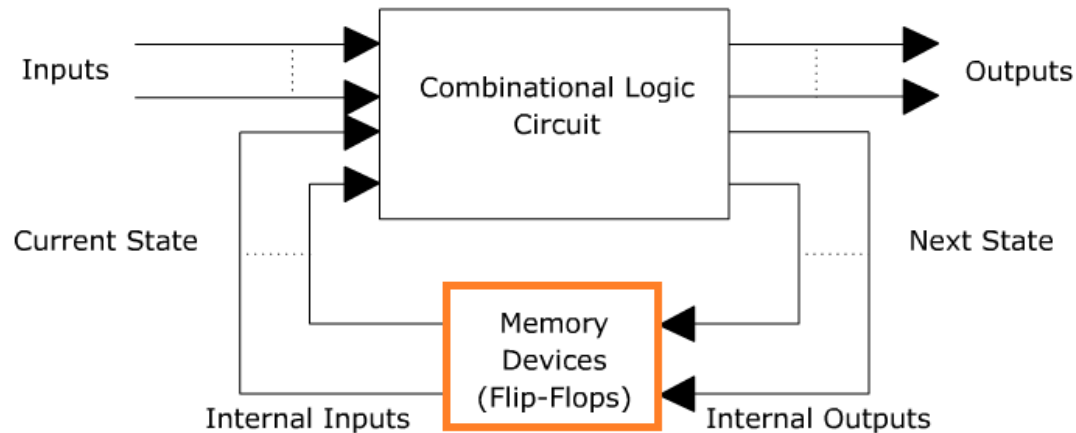
# What is Sequential Circuit ?

## ✓ Sequential circuit

- Output depends not only on the current input values, but also on **preceding** input values
- It remembers sort of the past history of the system

## ✓ How?

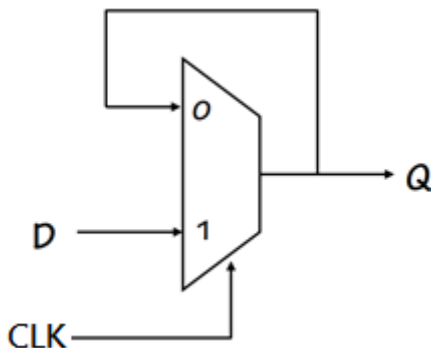
- Registers(Flip-Flops) 因此需要使用memory來儲存相關狀態



# Latch Operation

## ✓ Latch: level sensitive

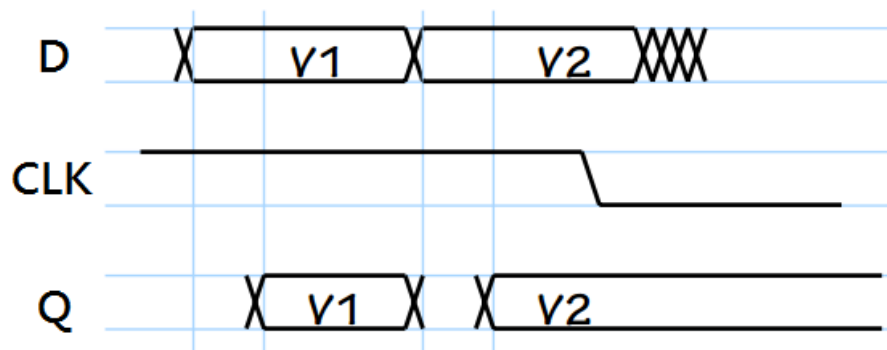
若電路中存在latch會對電路的timing產生影響  
latch本身是一個非同步電路，若存在於設計  
當中會對同步電路產生不好的影響(同步與非  
同步電路要分開)



CLK	D	Q	Q'	
0	--	0	0	} Q stable
0	--	1	1	
1	0	--	0	} Q follows D
1	1	--	1	

CLK=1 : Q follows D

CLK=0 : Q holds

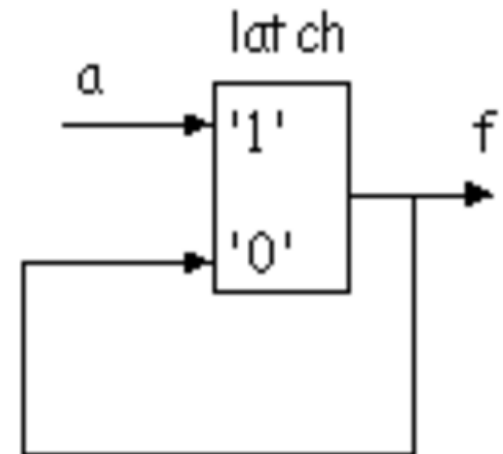
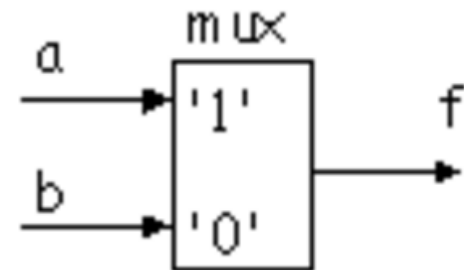


# Avoid Unintentional Latch (1/2)

✓ **Example** 因此我們要把所有case寫滿，以避免產生 combinational feedback(即latch)

```
always @(*)
begin
    if(sel == 1) f = a;
    else f = b;
end
```

```
always @(*)
begin
    if(sel == 1) f = a;
end
```





# Avoid Unintentional Latch (2/2)

## ✓ Avoid latches in combinational circuit

- Avoid incomplete if-then-else
- Avoid incomplete case statements

X

```
if(!rst_n) out = 0;  
else if(m==3'd0) out = m0_out;  
else if(m==3'd1) out = m1_out;
```

X

```
case(mode)  
  3'd0: out = m0_out;  
  3'd1: out = m1_out;  
endcase
```

O

```
if(!rst_n) out = 0;  
else if(m==3'd0) out = m0_out;  
else if(m==3'd1) out = m1_out;  
else out = default_out;
```

O

```
case(mode)  
  3'd0: out = m0_out;  
  3'd1: out = m1_out;  
  default:  
    out = default_out;  
endcase
```

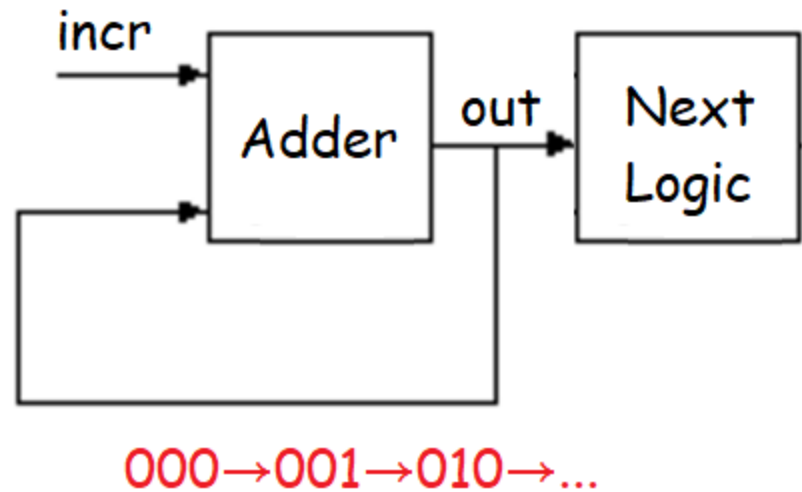


# Avoid Combinational Feedback (1/2)

## ✓ Example

```
assign out=out+1;
```

這樣寫也會變成combinational feedback  
所以要新增一條wire來儲存+1這個結果



# Avoid Combinational Feedback (2/2)

## ✓ Avoid combinational feedbacks

- Lead to unpredictable oscillated output
- NOT allowed

X

```
assign a=a+1;
```

X

```
always @(*) begin  
    a = a+1;  
end
```

X

```
always @(*) begin  
    if(in_a) a = c;  
    else a = a;  
end
```

X

```
assign out_value=out;  
always @(*) begin  
    case(mode)  
        3'd0: out = m0_out;  
        3'd1: out = m1_out;  
        default:  
            out = out_value;  
    endcase  
end
```

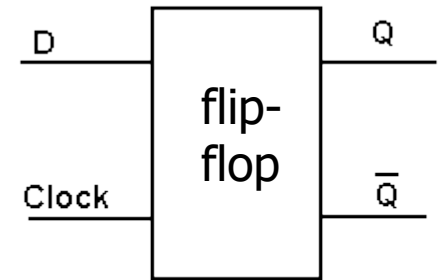


# Avoid Latch Summary

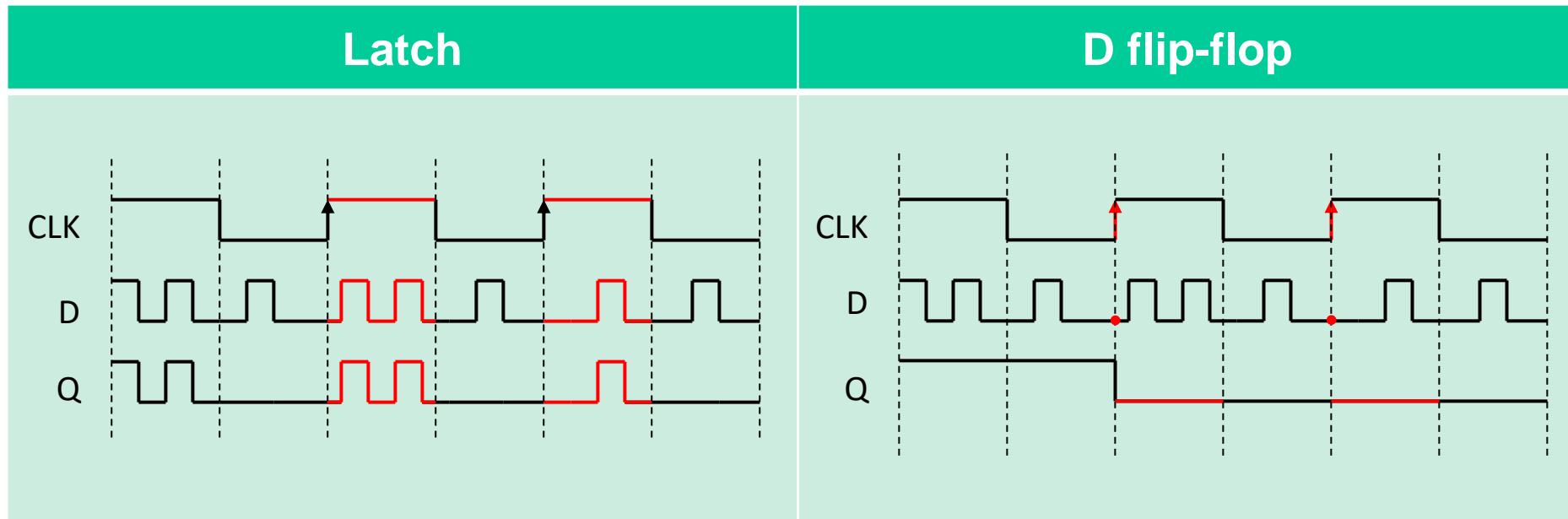
- ✓ **In a sequential circuit -- with clk control**
  - It is a flip-flop so there is not a latch problem.
- ✓ **In a combinational circuit -- without clk control**
  - If some net needs to keep its data, DC will synthesize a latch.
- ✓ **How to avoid?**
  - Conditional statement : must be full cases
  - Otherwise it will produce latches.
    - if – else work together or add default value
  - Ex: `if ( a == b ) c = 1 ;`
  - Case statement : remember default value
  - Ex: `case (a) 1'b0: c = b; endcase`
  - Avoid combinational feedback
- ✓ **Notice**
  - In a combinational circuit, no information will be stored, so latches are not allowed.
- ✓ **Latch is a memory storage device**
  - It will cause the problems of timing analysis .
    - That's why we recommend to avoid latches here!!

# Flip-Flop Operation

✓ D flip-flop: edge triggered



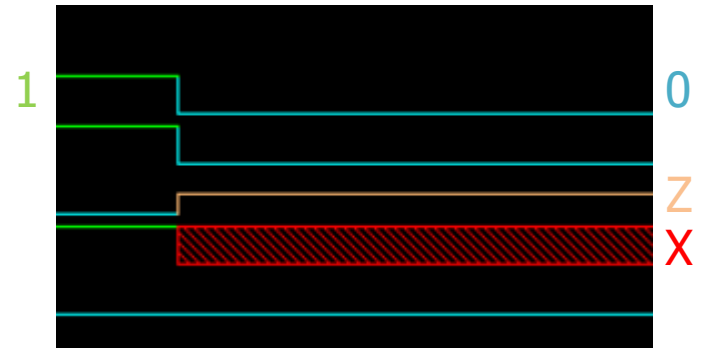
✓ Positive latch v.s. positive D flip-flop



# Flip-Flop Data Type

## ✓ Flip-flop: data storage element with 4 states (0,1, X, Z)

- **0**: logic low
- **1**: logic high
- **X**: unknown, may be a 0,1, Z, or in transition
- **Z**: high impedance, floating state



## ✓ Operations on the 4 states

- Example: AND, OR, NOT gate

AND	0	1	X	Z
0	0	0	0	0
1	0	1	X	X
X	0	X	X	X
Z	0	X	X	X

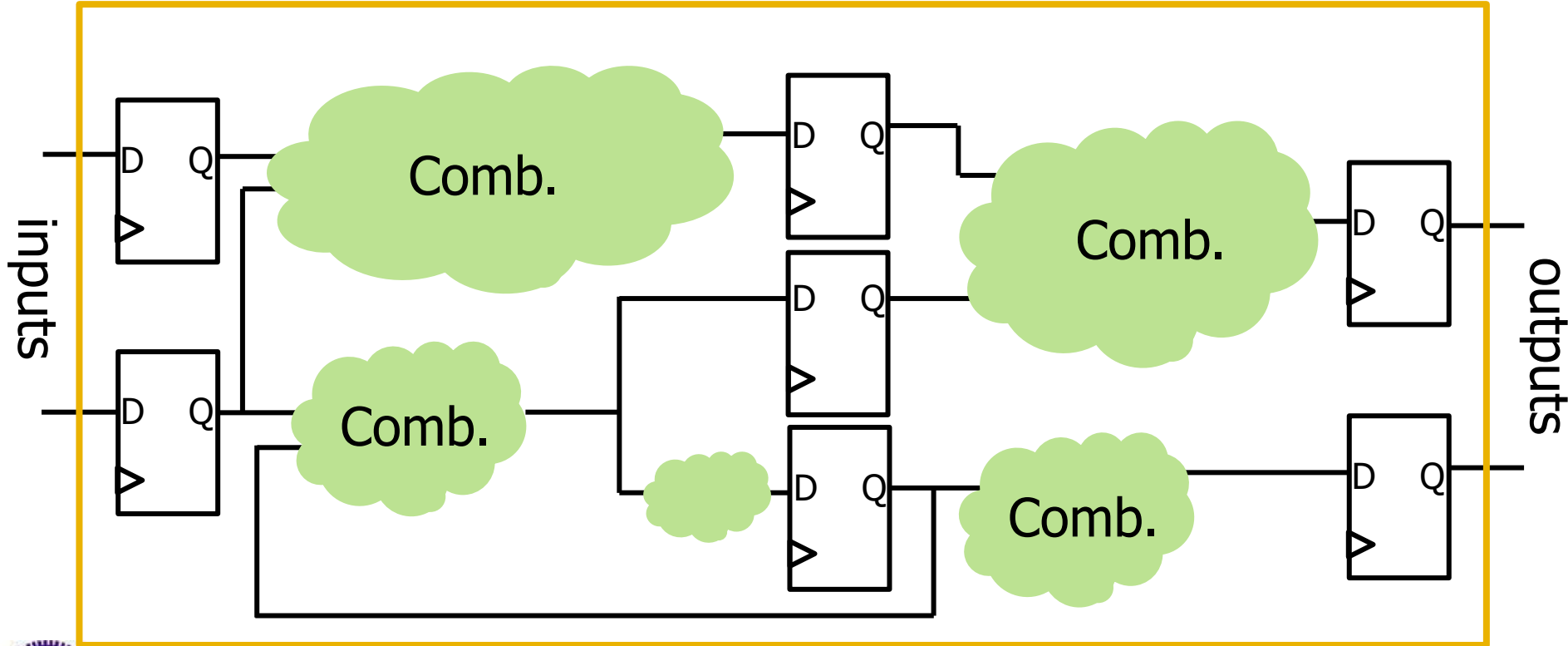
OR	0	1	X	Z
0	0	1	X	X
1	1	1	1	1
X	X	1	X	X
Z	X	1	X	X

NOT	output
0	1
1	0
X	X
Z	X



# Concept of Sequential Circuit

- ✓ Most computations are done by combinational circuit
  - ✓ Sequential elements are used for storage
- top design



# Outline

## ✓ **Section 1 Sequential Circuits**

✓ Introduction

✓ **Syntax**

✓ Reset

✓ Coding Style





# Assignment in Sequential Circuit

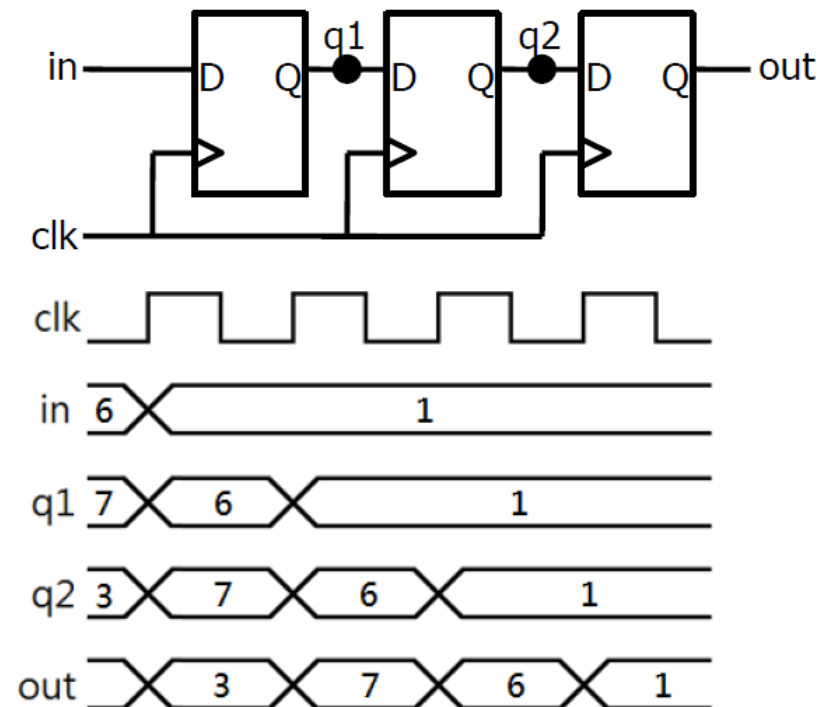
## ✓ Non-blocking assignment

- Evaluations and assignments are executed **at the same time without regard to orders or dependence upon each other**
- Syntax : **<variable> <= <expression>;**

## ✓ Example

```
always @(posedge clk)
begin
    q1 <= in;
    q2 <= q1;
    out <= q2;
end
```

寫non-blocking的時候  
要有右邊flip-flop的概念



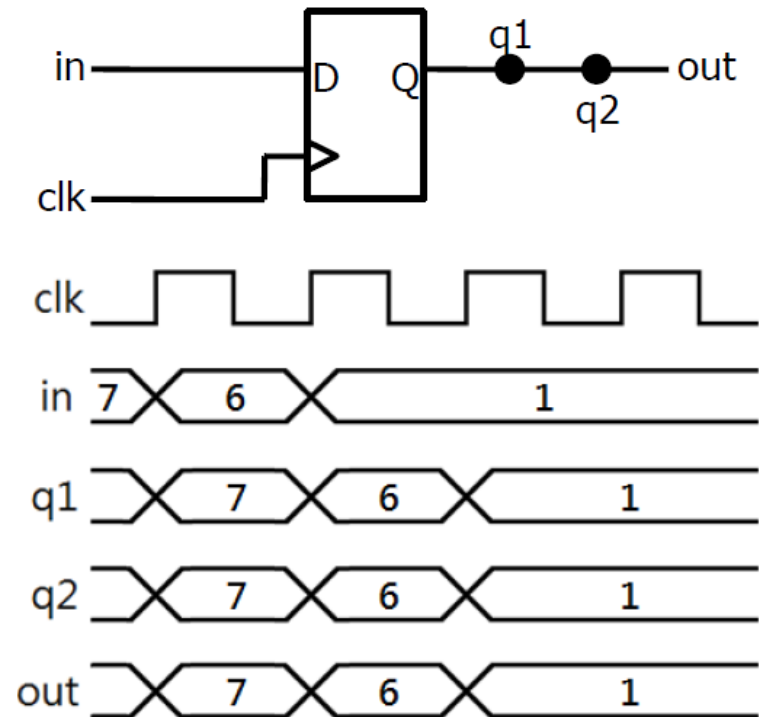
# Assignment in Sequential Circuit

## ✓ Blocking assignment

- Evaluations and assignments are **immediate** and **in order**
- Syntax : **<variable> = <expression>;**

## ✓ Example

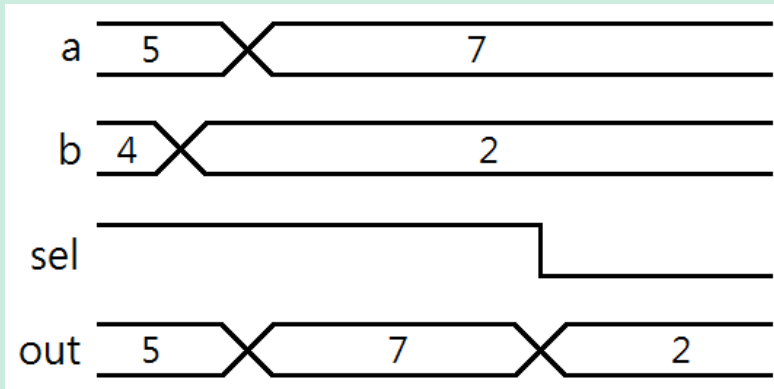
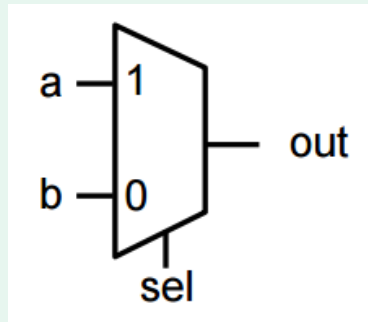
```
always @(posedge clk)
begin
    q1 = in;
    q2 = q1;
    out = q2;
end
```



# Combinational v.s. Sequential

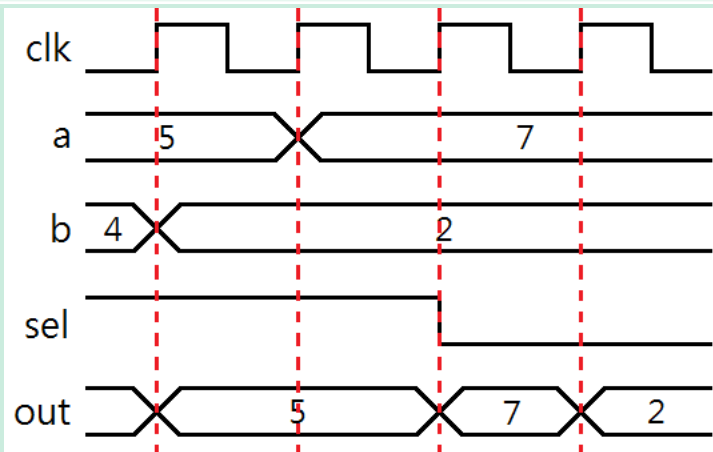
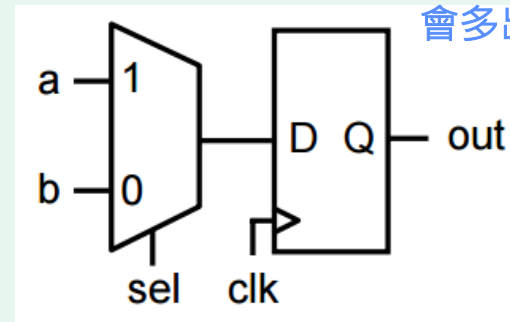
## Combinational

```
always@ (*)
begin
    if(sel) out = a;
    else    out = b;
end
```



## Sequential

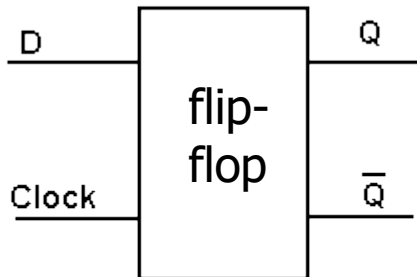
```
always@(posedge clk)
begin
    if(sel) out <= a;
    else    out <= b;
end
```



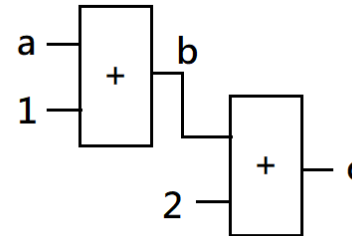
# Sequential Circuit

- ✓ **Sequential block**
  - use **non-blocking** assignments
- ✓ **Combinational block**
  - use **blocking** assignments
- ✓ **Comb./Seq. logic should be separated**

```
always@(posedge clk)
begin
    Q <= D;
end
```



```
always@*
begin
    b = a + 1;
    c = b + 2;
end
```



# Outline

## ✓ Section 1 Sequential Circuits

✓ Introduction

✓ Syntax

✓ **Reset**

✓ Coding Style

✓ Generate & For Loop



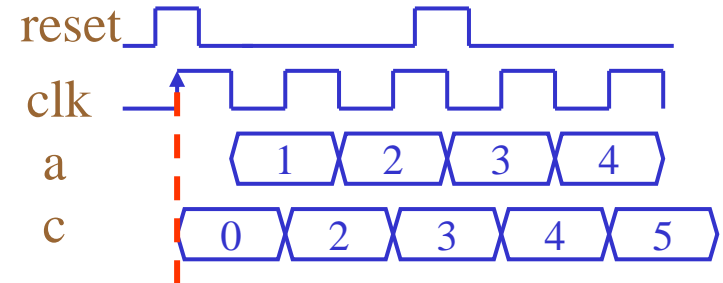
# Synchronous Reset (1/2)

## ✓ Register with **synchronous reset**

- Syntax: **always@ (posedge clk)**

代表reset只會在posedge時發生  
(與clock 同步)

```
always @(posedge clk) begin
    if (reset) c <= 0;
    else c <= a+1;
end
```

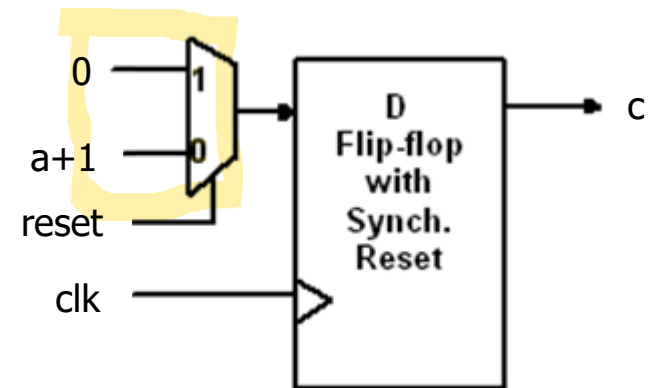


## ✓ Advantages

- **Glitch filtering** from reset combinational logic

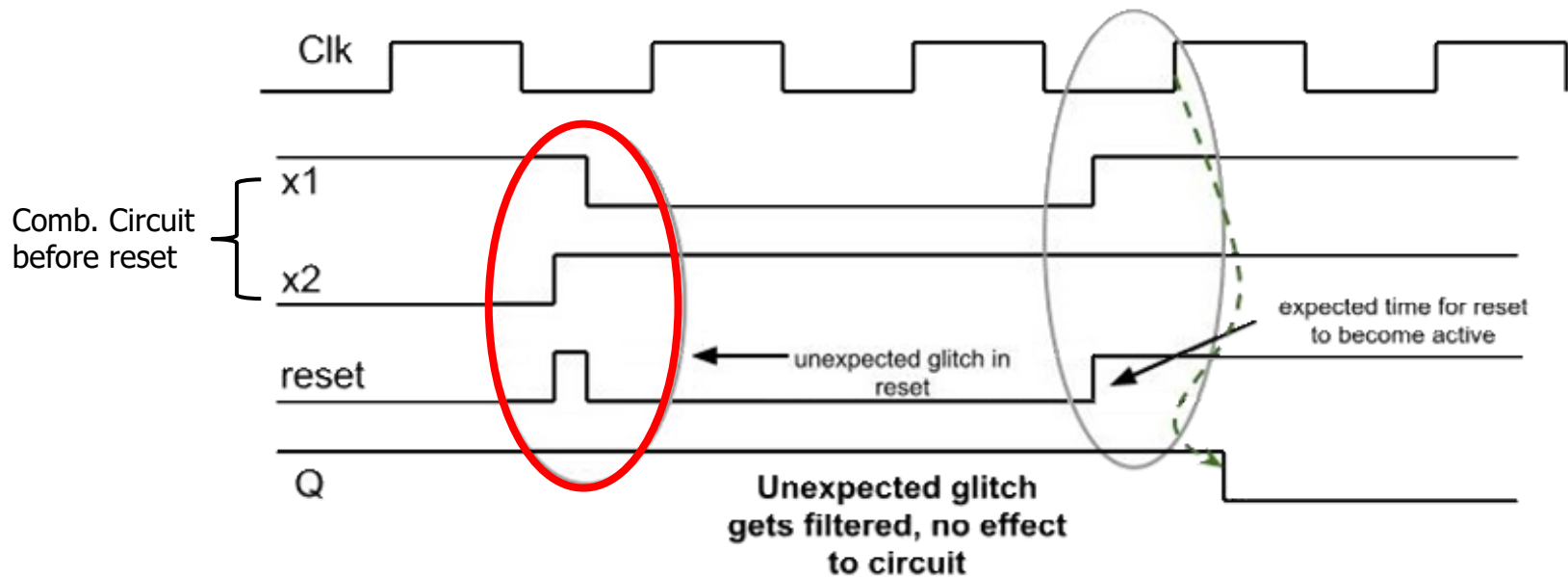
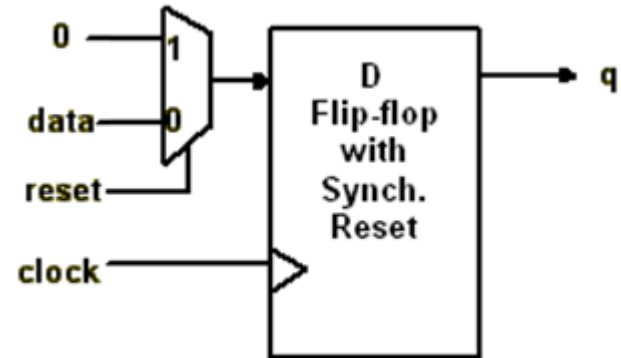
## ✓ Disadvantages

- Can't be reset without clock signal
- May need a pulse stretcher
  - Guarantee a reset pulse wide enough
- Larger area
- Increasing critical path



# Synchronous Reset (2/2)

✓ Advantage: glitch filtering

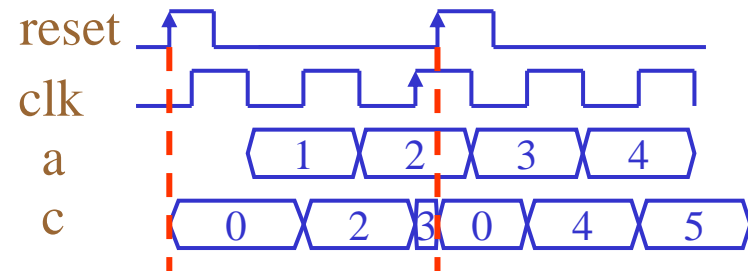


# Asynchronous Reset

## ✓ Register with asynchronous reset

- Syntax: **always @(posedge clk or posedge reset)**

```
always @(posedge clk or posedge reset)
begin
    if (reset) c <= 0;
    else c <= a+1;
end
```

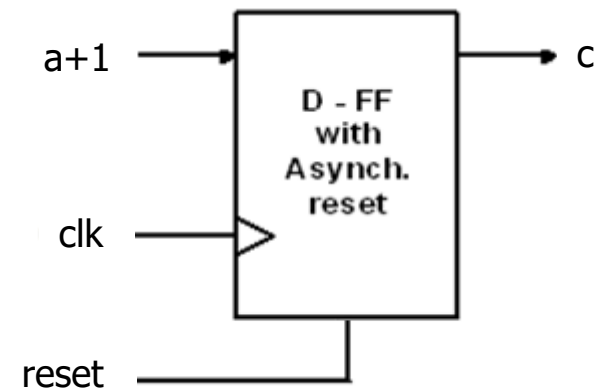


## ✓ Advantages

- Reset is independent of clock signal
- Reset is immediate
- Less area

## ✓ Disadvantages

- Noisy reset line could cause unwanted reset
- Metastability

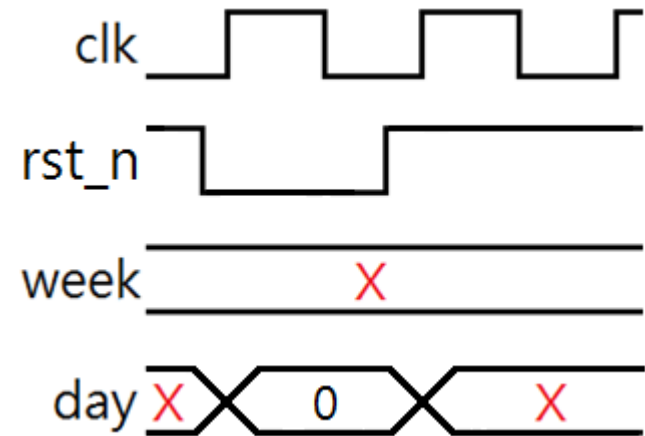




# Avoid Unknown

✓ **Reset all signals to avoid unknown propagation**

**X**always @(posedge clk) begin  
 // if(!rst\_n) week <= 0;  
 week <= week+1;  
end  
always @(posedge clk) begin  
 if(!rst\_n) day <= 0;  
 else day <= week \* 7;  
end



AND	0	1	X	Z
0	0	0	0	0
1	0	1	X	X
X	0	X	X	X
Z	0	X	X	X

OR	0	1	X	Z
0	0	1	X	X
1	1	1	1	1
X	X	1	X	X
Z	X	1	X	X

NOT	output
0	1
1	0
X	X
Z	X



# Outline

## ✓ Section 1 Sequential Circuits

✓ Introduction

✓ Syntax

✓ Reset

✓ Coding Style



# Coding Styles (1/2)

✓ Naming should be readable

✓ Synthesizable codes

- assign, always block, called sub-modules, if-then-else, cases, parameters, operators

✓ Data has to be described in one always block

- Multiple source drive is not valid

**X**

```
always @(posedge clk) begin
    out <= out+1;
end
always @(posedge clk) begin
    out <= a;
end
```

✓ Always block can't exist both blocking and non-blocking assignment

**X**

```
always @(posedge clk) begin
    if(reset) out = 0;
    else out <= out+in;
end
```



# Coding Styles (2/2)

## ✓ Do not put many variables in one always block

- Except shift registers or registers with similar properties

**bad**

```
always @(posedge CLK) begin
    q2 <= in;
    if(sel==0) out <= q2;
    else if(sel==1) out <= q3;
    else out <= out;
end
```

**suggested**

```
always @(posedge CLK) begin
    q2 <= in;
end
always @(posedge CLK) begin
    if(sel==0) out <= q2;
    else if(sel==1) out <= q3;
    else out <= out;
end
```

## ✓ Use FSM (Finite State Machine)



# Outline

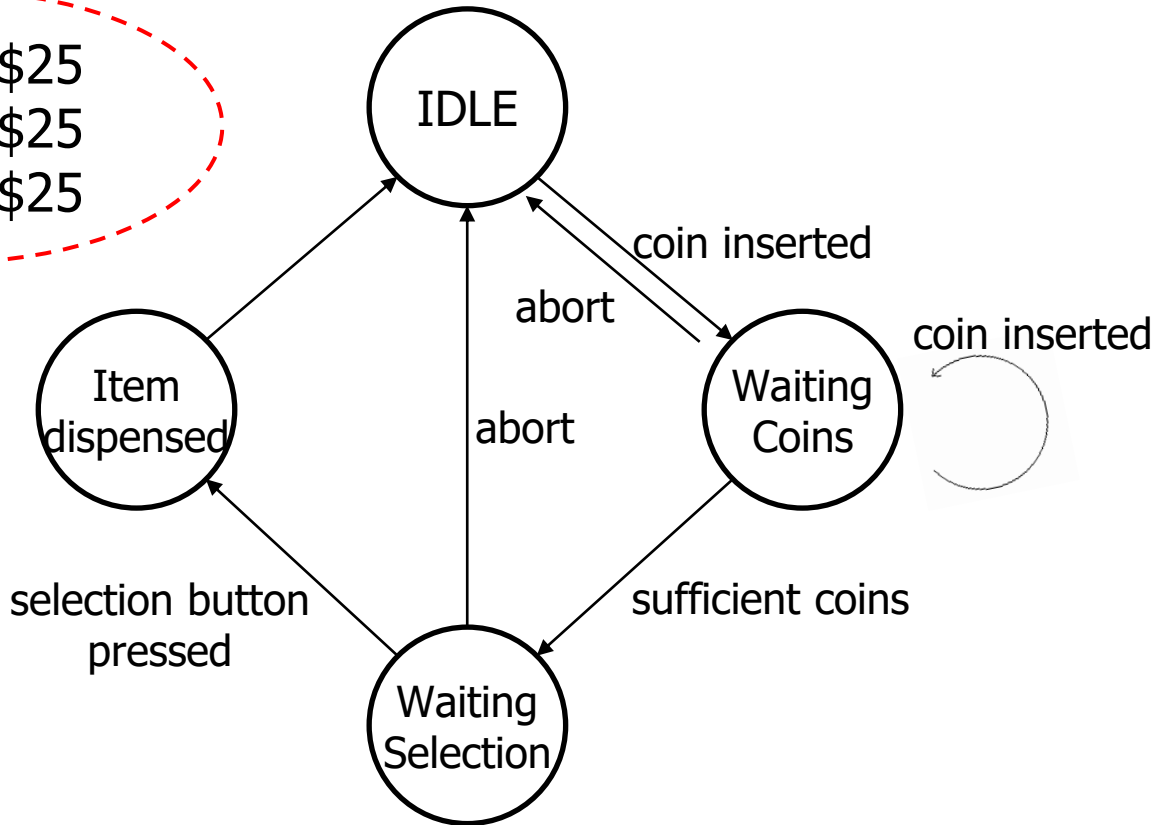
- ✓ Section 1 Sequential Circuits
- ✓ **Section 2 Finite State Machine**
- ✓ Section 3 Timing
- ✓ Section 4 Synthesis and Design Compiler
- ✓ Section 5 Generate & for loop



# Finite State Machine

## ✓ Example: Vending machine

Coke \$25  
Pepsi \$25  
Sprite \$25



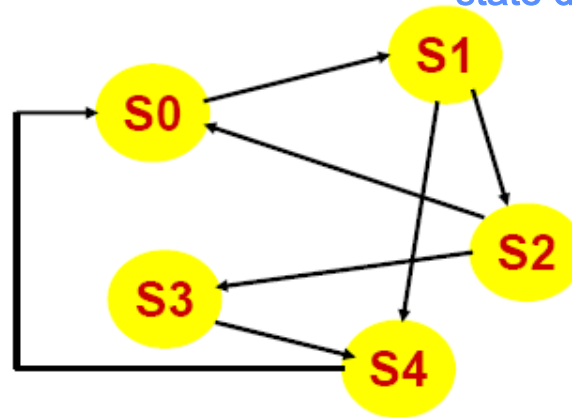
# Finite State Machine

## ✓ Finite state machine

- Powerful model for describing a sequential circuit
- Divide a sequential circuit operation into finite number of states.
- A state machine controller can output results depending on the input signal, control signal and states.
- As different input or control signal changes, the state machine will take a proper state transition.

## ✓ State diagram

看到題目的時候都可以想想要怎麼轉化成 state diagram



# Mealy and Moore Machines(1/3)

## ✓ Mealy machine

- The outputs depend on the current state and inputs
- If input changes, output also changes

## ✓ Advantages

- Less number of states are required

## ✓ Disadvantages 比較難設計

- More hardware requirements for circuit implementation

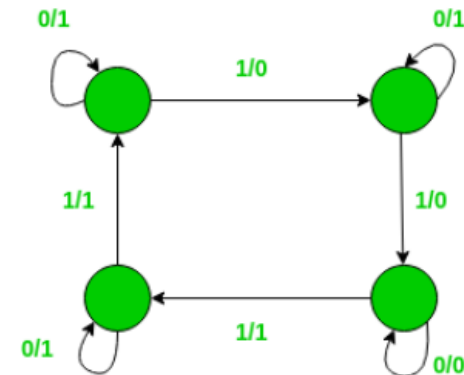
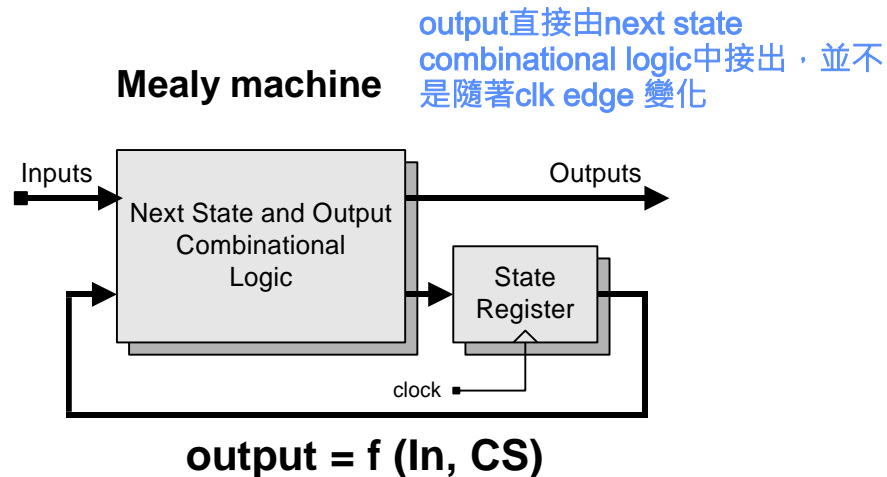


Figure - Mealy machine





# Mealy and Moore Machines(2/3)

## ✓ Moore machine

<https://www.geeksforgeeks.org/difference-between-mealy-machine-and-moore-machine/>

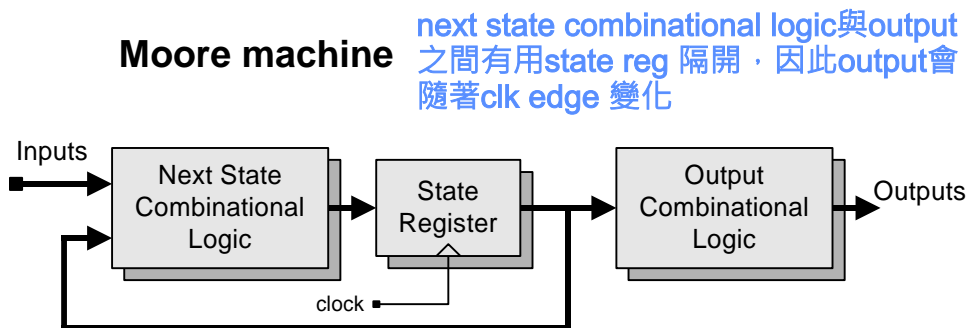
- The outputs depend on the current state only
- Inputs affect outputs but not immediately

## ✓ Advantage

- Safer. Outputs change at clock edge

## ✓ Disadvantage

- More states are required



$$\text{output} = f(\text{CS})$$

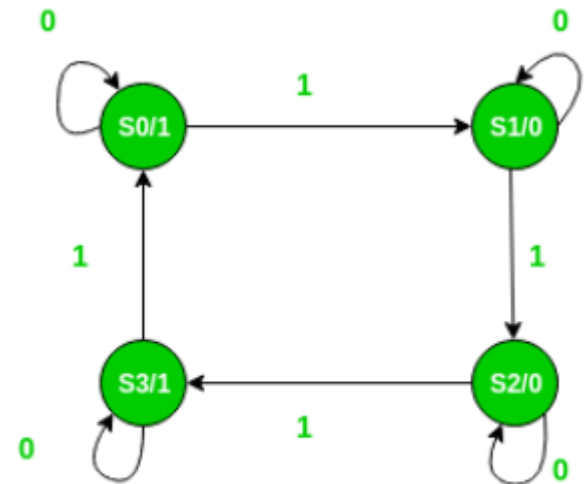


Figure - Moore machine



# Mealy and Moore Machines(3/3)

- FSM coding style
  - Separate CS, NS and OL

non-blocking

**Current  
State**

```
always @(posedge clk)
    current_state<=next_state;
```

blocking

**Next  
State**

```
always @(current_state or In)
case (current_state)
state_0: case(In)
            In0: next_state = state_value1;
            In1: next_state = state_value2;
            .....
        endcase
        .....
    default : .....
endcase
```

If it is not full case and without **default case** ,latch will be incurred!

Mealy  
machine

Moore  
machine

**Output  
Logic**

```
always @(current_state or In)
    Z = values;
```

```
always @(current_state)
    Z = values;
```

# FSM Coding Style

## ✓ Separate current state, next state and output logic

### Current State

```
always @(posedge clk or negedge rst_n) begin
    if (!rst_n) current_state <= IDLE;
    else current_state <= next_state;
end
```

Use parameters for readability

### Next State

```
always @(*) begin
    if(!rst_n) next_state=IDLE;
    else begin
        case(current_state)
            STATE_1: begin
                if (in==in_1) next_state=STATE_2;
                else next_state=current_state;
            end
            STATE_2: .....
            ....
            default: next_state=current_state;
        endcase
    end
end
```

If it's not full case and without default case, latch would be incurred!

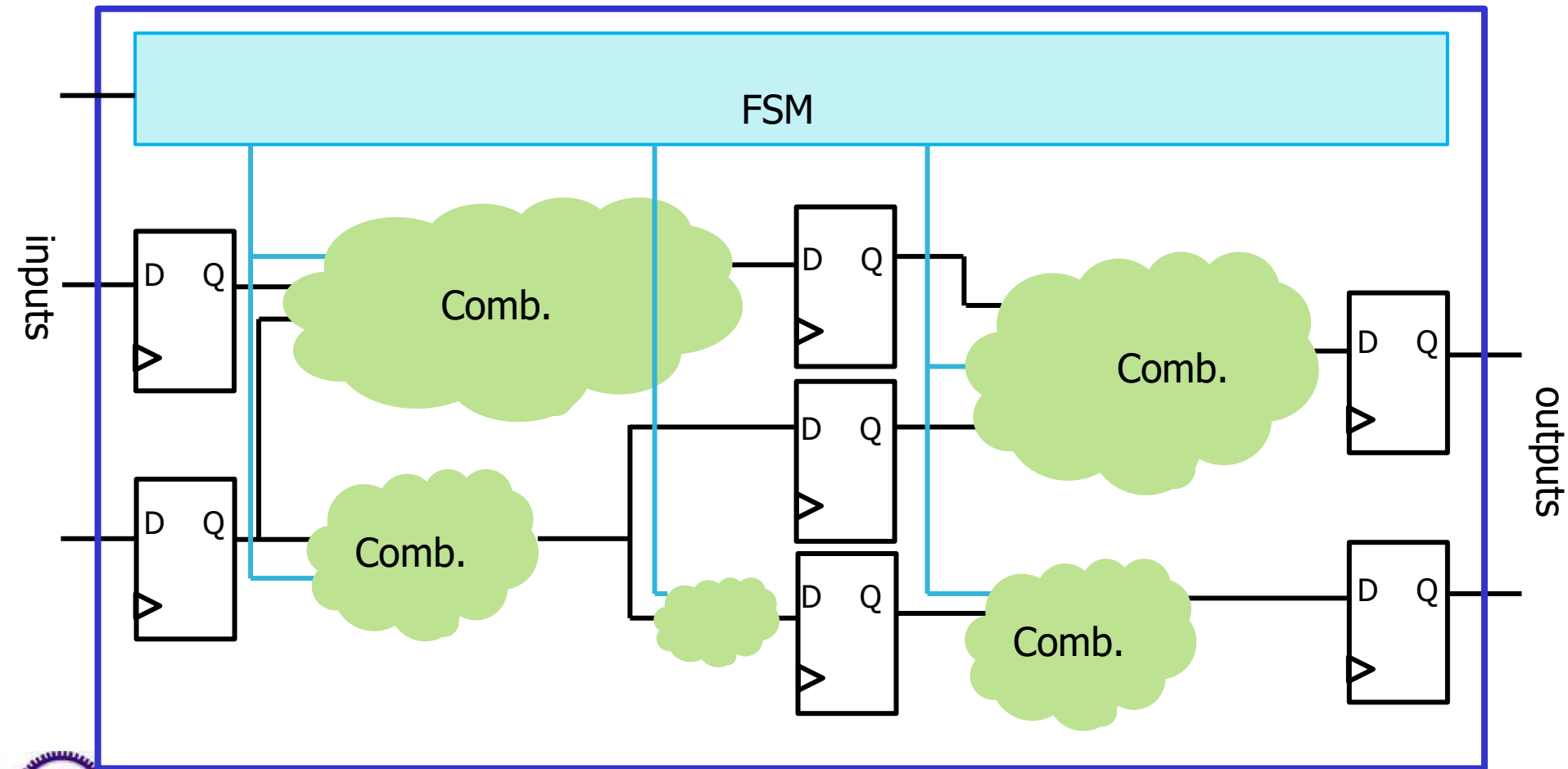
```
parameter IDLE      = 2'd0;
parameter STATE_1   = 2'd1;
parameter STATE_2   = 2'd2;
parameter STATE_3   = 2'd3;
```

### Output Logic

```
always@(posedge clk or negedge rst_n) begin
    if (!rst_n) out <= 0;
    else if (current_state==STATE_3) out <= output_value;
    else out <= out;
end
```

# Why FSM?

- ✓ FSM can be referred to as the controller and status of the whole module



# Outline

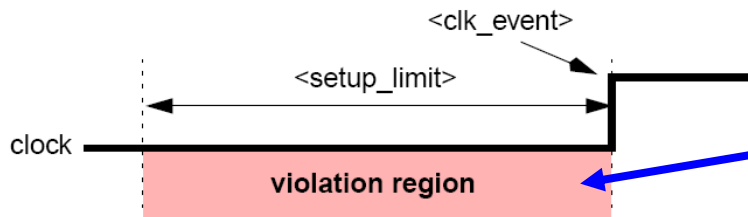
- ✓ Section 1 Sequential Circuits
- ✓ Section 2 Finite State Machine
- ✓ **Section 3 Timing**
- ✓ Section 4 Synthesis and Design Compiler
- ✓ Section 5 Generate & for loop



# Timing Check (1/3)

## ✓ Setup time check

- The `$setup` system task determines whether a data signal remains stable for a minimum specified time before a transition in an enabling, such as a clock event.

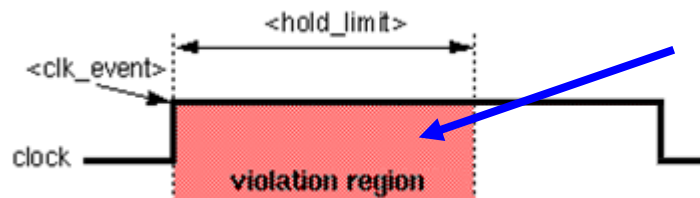


在posedge clk前的一小段時間，我們必須保證要傳送的数据是stable，否則傳輸data時有可能產生錯誤

data can't be changed in this red region

## ✓ Hold time check

- The `$hold` system task determines whether a data signal remains stable for a minimum specified time after a transition in an enabling signal, such as a clock event.



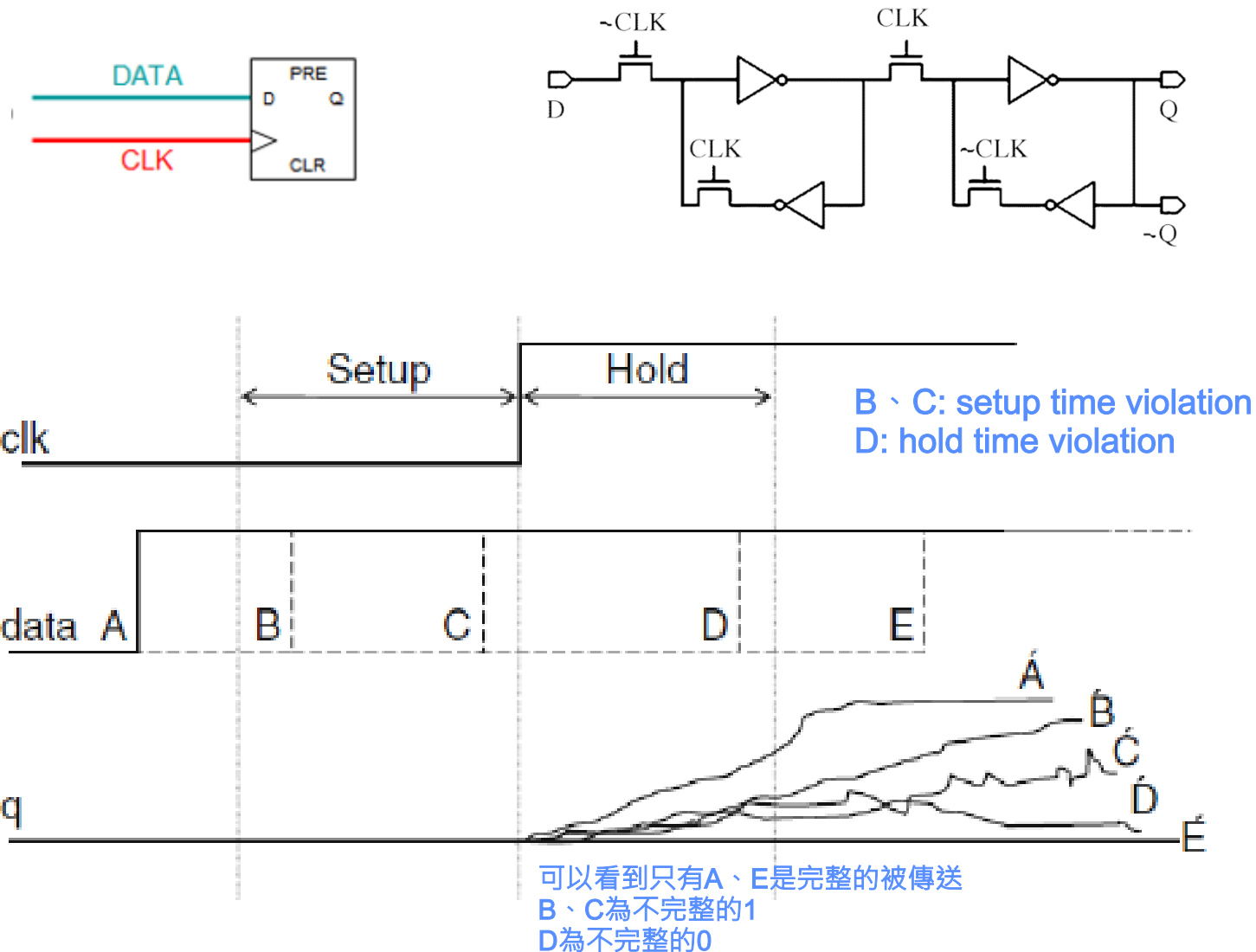
在data傳送後的一小段時間內(flip flop都是在posedge 傳送data，因此就是正緣後的一小段時間)，我們也要保證傳送的数据是stable的，不然也有可能會傳送到不完整的data

data can't be changed in this red region



# Timing Check (2/3)

## ✓ Metastability



# Timing Check (3/3)

## ✓ Timing report: setup time

clock CLK_1 (rise edge)	2.00	2.00
clock network delay (ideal)	2.00	4.00
clock uncertainty	-0.50	3.50
IN_A_reg[0]/CK (EDFFXL)	0.00	3.50 r
library setup time	-0.42	3.08
data required time		3.08
-----		
data required time		3.08
data arrival time		-3.08
-----		
slack (MET)		0.00

## ✓ Timing report: hold time

Slacks should be **MET**!  
(non-negative)

clock CLK_2 (rise edge)	0.00	0.00
clock network delay (ideal)	4.00	4.00
clock uncertainty	1.00	5.00
IN_B_reg[20]/CK (EDFFXL)	0.00	5.00 r
library hold time	-0.19	4.81
data required time		4.81
-----		
data required time		4.81
data arrival time		-4.82
-----		
slack (MET)		0.01



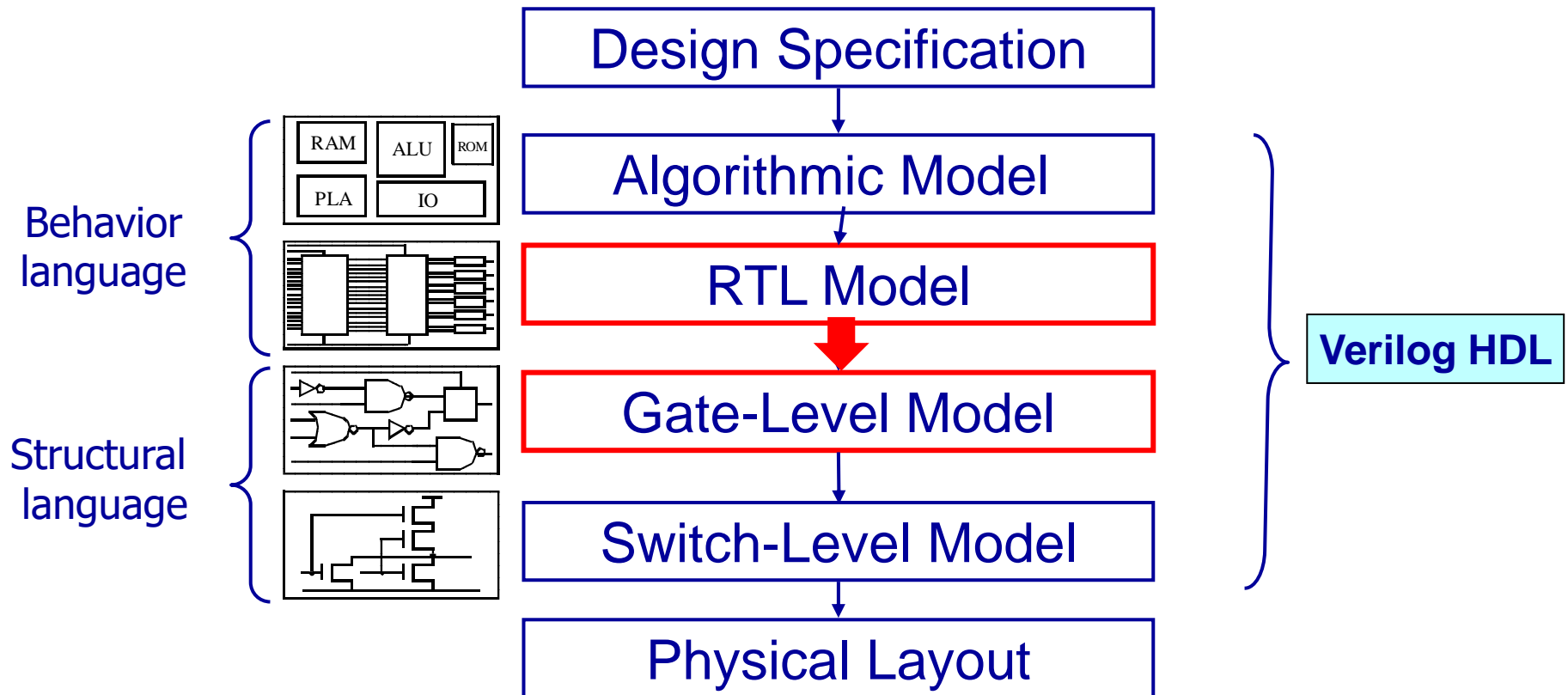


# Outline

- ✓ Section 1 Sequential Circuits
- ✓ Section 2 Finite State Machine
- ✓ Section 3 Timing
- ✓ **Section 4 Synthesis and Design Compiler**
- ✓ Section 5 Generate & for loop



# Recall: Design Flow



# Logic Synthesis

## ✓ Logic synthesis

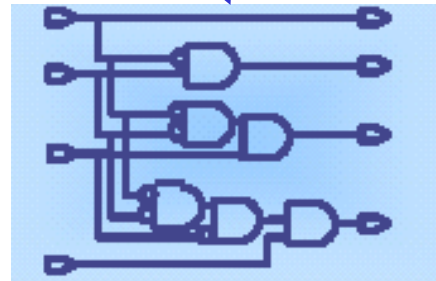
- A process by which behavioral model of a circuit is turned into an implementation in terms of logic gates
- Synthesis = **Translation+Mapping+Optimization**

第六章會詳細講

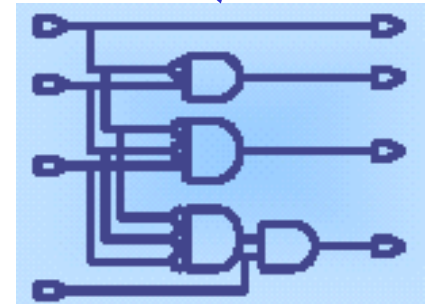
```
assign avg=sum/total;  
always_ff @(posedge clk)  
begin  
    sum=sum+score*weight;  
end
```

HDL Source

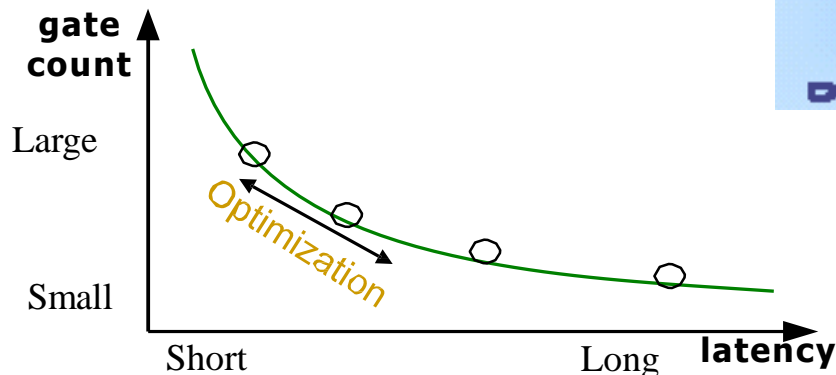
Translate



Map+Optimize



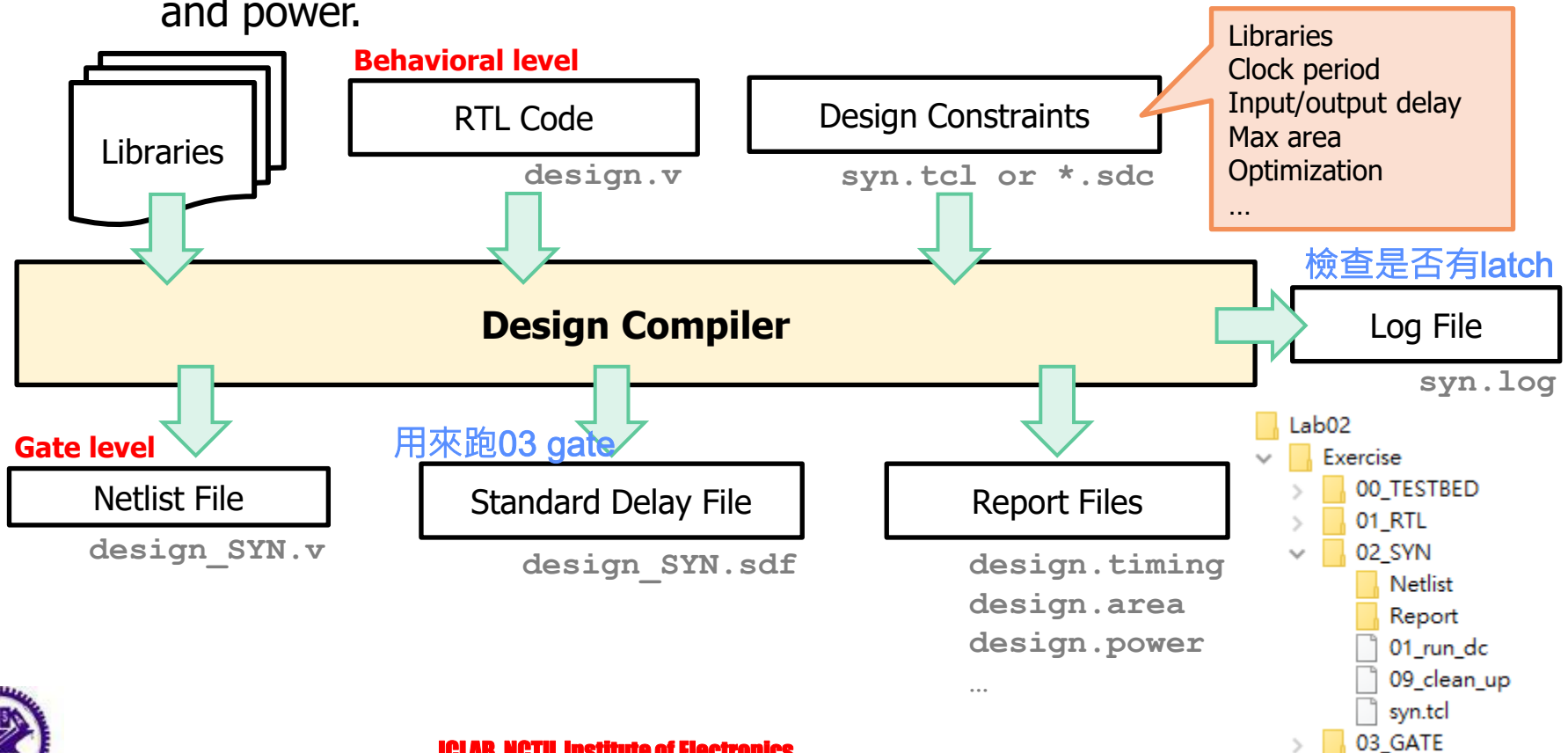
Target Technology



# Design Compiler

## ✓ Design compiler

- A tool by Synopsys, Inc. that synthesizes your HDL designs (**Verilog**) into optimized technology-dependent, **gate-level** designs.
- It can optimize both combinational and sequential designs for speed, area, and power.



# .lib

- ✓ Cell name
- ✓ Drive strength
- ✓ Area
- ✓ Pin
- ✓ Leakage power
- ✓ Timing for each pin
- ✓ Internal power



**Same information as .db file**

.db是給電腦讀的  
與.lib紀錄完全相同的資訊

```
cell (NANDX1) {
  pin(A1) {
    direction : input;
    capacitance : 0.00683597;
  }
  pin(A2) {
    direction : input;
    capacitance : 0.00798456;
  }
  pin(ZN) {
    direction : output;
    capacitance : 0.0;
    internal_power() {
      timing() {
        cell_rise(table10){
          values ("0.020844,0.02431,0.030696,0.039694,0.048205,0.072168,0.10188",\
                  "0.024677,0.027942,0.035042,0.045467,0.054973,0.082349,0.11539",\
                  "0.032068,0.035394,0.042758,0.055361,0.065991,0.090936,0.13847",\
                  "0.046811,0.049968,0.057164,0.064754,0.086481,0.11676,0.15744",\
                  "0.073919,0.078805,0.080873,0.091007,0.11655,0.1579,0.21448",\
                  "0.13162,0.13363,0.1383,0.14793,0.1685,0.22032,0.30054",\
                  "0.24661,0.24835,0.25294,0.26221,0.282,0.32417,0.42783");
          out_capacitance
          input_transition_time
        }
      }
    }
  }
}

lu_table_template(table10){
  variable_1 : total_output_net_capacitance;
  variable_2 : input_transition_time;
  index_1 ("0.001400,0.003000,0.006200,0.012500,0.025100,0.050400,0.101000");
  index_2 ("0.0208,0.0336,0.06,0.1112,0.2136,0.4192,0.8304");
}
```

# Outline

- ✓ Section 1 Sequential Circuits
- ✓ Section 2 Finite State Machine
- ✓ Section 3 Timing
- ✓ Section 4 Synthesis and Design Compiler
- ✓ **Section 5 Generate & for loop**



# Generate

## SystemVerilog

3.1a	{	assertions	mailboxes	from C / C++		
		test program blocks	semaphores	classes	dynamic arrays	
3.0	{	clocking domains	constrained random values	inheritance	associative arrays	
		process control	direct C function calls	strings	references	
3.0	{	interfaces	packages	int	globals	break
		nested hierarchy	2-state modeling	shortint	enum	continue
3.0	{	unrestricted ports	packed arrays	longint	typedef	return
		automatic port connect	array assignments	byte	structures	do-while
3.0	{	enhanced literals	queues	shortreal	unions	++ -- += -= *= /=
		time values and units	unique/priority case/if	void	casting	>>= <<= >>>= <<<=
3.0	{	specialized procedures	compilation unit space	alias	const	&=  = ^= %=

## Verilog-2001

ANSI C style ports  
**generate**  
 localparam  
 constant functions

standard file I/O  
 \$value\$plusargs  
 `ifndef `elsif `line  
 @\*

(\* attributes \*)  
 configurations  
 memory part selects  
 variable part select

multi dimensional arrays  
 signed types  
 automatic  
 \*\* (power operator)

## Verilog-1995

modules  
 parameters  
 function/tasks  
 always @  
 assign  
 \$finish \$fopen \$fclose  
 \$display \$write  
 \$monitor  
 `define `ifdef `else  
 `include `timescale

initial  
 disable  
 events  
 wait # @  
 fork-join  
 wire reg  
 integer real  
 time  
 packed arrays  
 2D memory

begin-end  
 while  
 for forever  
 if-else  
 repeat  
 + = \* /  
 %  
 >> <<



# For Loop

- For loop in Verilog
  - Duplicate same function
  - Very useful for doing reset and iterated operation
  - Unrolling

```
reg [3:0] temp[0:2];
integer i;
always @(posedge clk) begin
    for (i = 0; i < 3 ; i = i + 1) begin: for_name
        temp[i] <= 4'b0;
    end
end
```

=

```
always @(posedge clk) begin
    temp[0] <= 4'b0;
    temp[1] <= 4'b0;
    temp[2] <= 4'b0;
end
```

```
reg [3:0] temp[0:2];
reg [3:0] data;
integer i;
always @(posedge clk) begin
    for (i = 0; i < 3 ; i = i + 1) begin: for_name
        temp[i+1] <= temp[i];
    end
    temp[0] <= data;
end
```

=

```
always @(posedge clk) begin
    temp[0] <= data;
    temp[1] <= temp[0];
    temp[2] <= temp[1];
    temp[3] <= temp[2];
end
```



# Generate

- How to use for loop with generate?
  - For loop in generate : four always blocks
  - Regular for loop : one always block

```
reg [3:0] temp;  
genvar i;  
generate  
for (i = 0; i < 4 ; i = i + 1) begin: for_name  
    always @(posedge clk) begin  
        temp[i] <= 1'b0;  
    end  
end  
endgenerate
```

Generate block

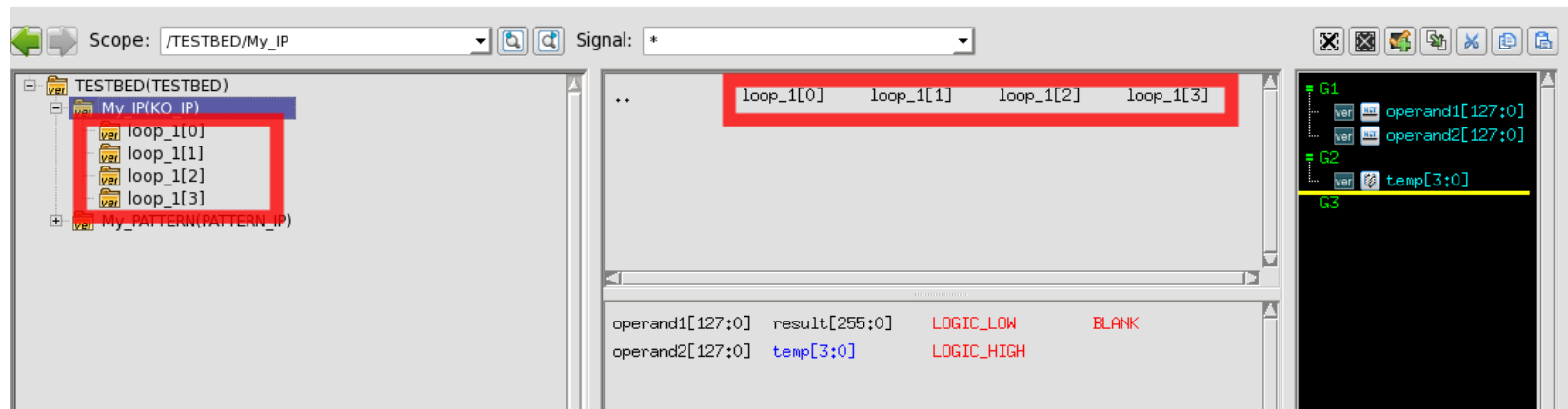
```
reg [3:0] temp;  
integer i;  
always @(posedge clk) begin  
    for (i = 0; i < 4 ; i = i + 1) begin:  
        temp[i] <= 1'b0;  
    end  
end
```

Regular for loop

# Generate

```
reg [3:0] temp;  
  
genvar i;  
generate  
for (i=0 ; i <4; i = i+1)begin loop_1  
    always@(*)begin  
        temp[i] = operand1[i] & operand2[i];  
    end  
end  
endgenerate
```

always block in for loop with  
genvar

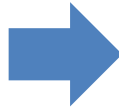


4 always block  
instance

# For Loop/Generate Example

- ✓ **Example**
  - Copy a module for 3 times
- ✓ **Generate:**

```
module A();  
endmodule  
  
module B();  
  genvar i;  
  generate  
    for(i=0; i<3; i=i+1) begin  
      A uA(...)  
    end  
  endgenerate  
endmodule
```



```
module A();  
...  
endmodule  
  
module A();  
...  
endmodule  
  
module A();  
...  
endmodule
```

<https://www.chipverify.com/verilog/verilog-generate-block>

可以參考更多的generate 用法



# For Loop/Generate Example

- ✓ **Example**
  - Copy a module for 3 times
- ✓ **Generate:**

如果要複製硬體，就勢必只能用generate寫，  
因此下面寫法是錯的

```
module A();  
endmodule  
  
module B();  
for(i=0; i<3; i=i+1) begin  
    A uA(...)  
end  
endmodule
```

