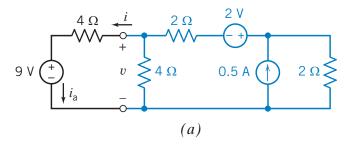
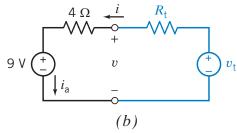
## **Section 5-2: Source Transformations**

**P 5.2-1** The circuit shown in Figure P 5.2-1*a* has been divided into two parts. The circuit shown in Figure P 5.2-1*b* was obtained by simplifying the part to the right of the terminals using source transformations. The part of the circuit to the left of the terminals was not changed.

- (a) Determine the values of  $R_t$  and  $v_t$  in Figure P 5.2-1b.
- (b) Determine the values of the current *i* and the voltage *v* in Figure P 5.2-1*b*. The circuit in Figure P 5.2-1*b* is equivalent to the circuit in Figure P 5.2-1*a*. Consequently, the current *i* and the voltage *v* in Figure P 5.2-1*a*





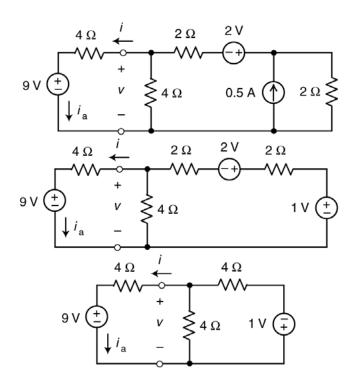
**Figure P 5.2-1** 

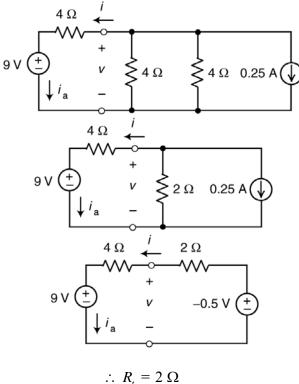
have the same values as do the current i and the voltage v in Figure P 5.2-1b.

(c) Determine the value of the current  $i_a$  in Figure P 5.2-1a.

#### **Solution:**

(a)



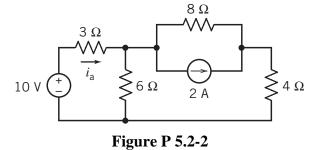


$$\therefore R_t = 2 \Omega$$
$$v_t = -0.5 \text{ V}$$

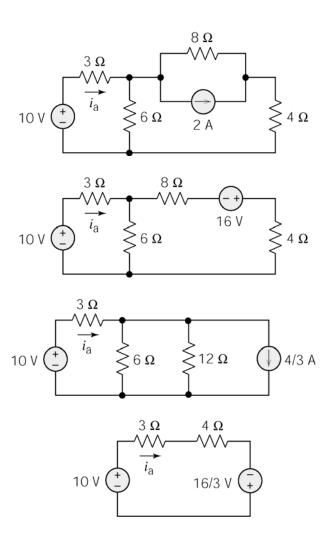
(b) 
$$-9-4i-2i+(-0.5) = 0$$
$$i = \frac{-9+(-0.5)}{4+2} = -1.58 \text{ A}$$
$$v = 9+4 i = 9+4(-1.58) = 2.67 \text{ V}$$

(c) 
$$i_a = i = -1.58 \text{ A}$$

**P 5.2-2** Consider the circuit of Figure P 5.2-2. Find  $i_a$  by simplifying the circuit (using source transformations) to a single-loop circuit so that you need to write only one KVL equation to find  $i_a$ .



# **Solution:**



Finally, apply KVL: 
$$-10 + 3 i_a + 4 i_a - \frac{16}{3} = 0$$
  $\therefore i_a = 2.19 \text{ A}$ 

**P 5.2-3** Find  $v_0$  using source transformations if i = 5/2 A in the circuit shown in Figure P 5.2-3.

*Hint:* Reduce the circuit to a single mesh that contains the voltage source labeled  $v_0$ .

**Answer:**  $v_0 = 28 \text{ V}$ 

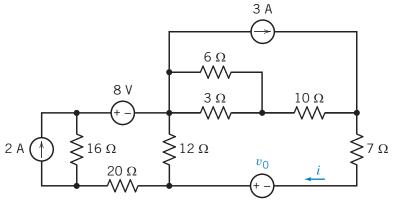
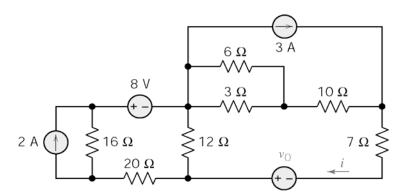
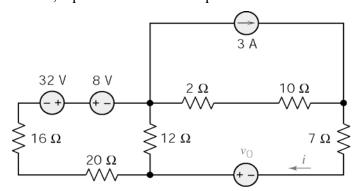


Figure P 5.2-3

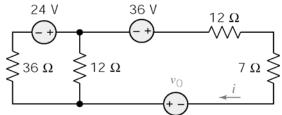
## **Solution:**



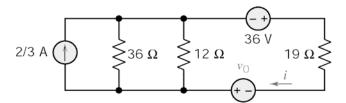
Source transformation at left; equivalent resistor for parallel 6 and 3  $\Omega$  resistors:



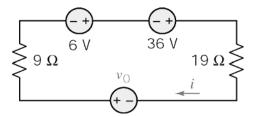
Equivalents for series resistors, series voltage source at left; series resistors, then source transformation at top:



Source transformation at left; series resistors at right:



Parallel resistors, then source transformation at left:



Finally, apply KVL to loop

$$-6 + i (9+19) - 36 - v_0 = 0$$
  
$$i = 5/2 \implies v_0 = -42 + 28 (5/2) = 28 \text{ V}$$

**P 5.2-4** Determine the value of the current  $i_a$  in the circuit shown in Figure P 5.2-4.

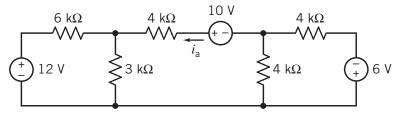
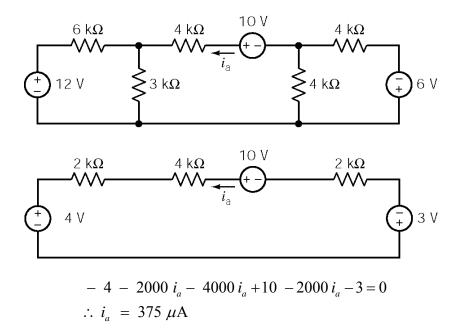


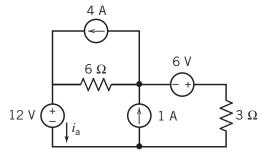
Figure P 5.2-4

# **Solution:**



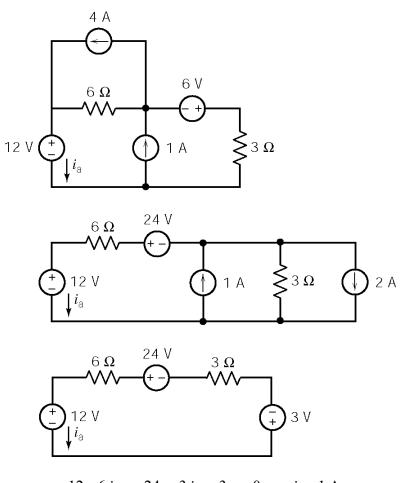
# **P 5.2-5** Use source transformations to find the current $i_a$ in the circuit shown in Figure P 5.2-5.

**Answer:**  $i_a = 1 \text{ A}$ 



**Figure P 5.2-5.** 

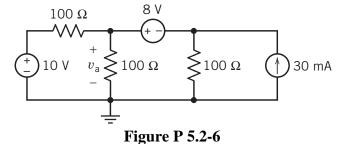
## **Solution:**



$$-12-6i_a + 24 - 3i_a - 3 = 0 \implies i_a = 1 \text{ A}$$

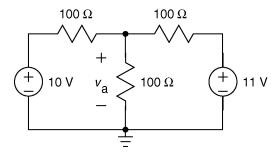
**P 5.2-6** Use source transformations to find the value of the voltage  $v_a$  in Figure P 5.2-6.

**Answer:**  $v_a = 7 \text{ V}$ 

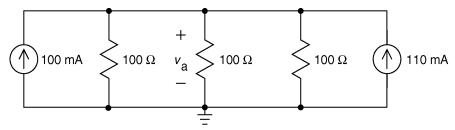


#### **Solution:**

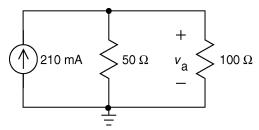
A source transformation on the right side of the circuit, followed by replacing series resistors with an equivalent resistor:



Source transformations on both the right side and the left side of the circuit:



Replacing parallel resistors with an equivalent resistor and also replacing parallel current sources with an equivalent current source:



Finally, 
$$v_a = \frac{50(100)}{50+100}(0.21) = \frac{100}{3}(0.21) = 7 \text{ V}$$

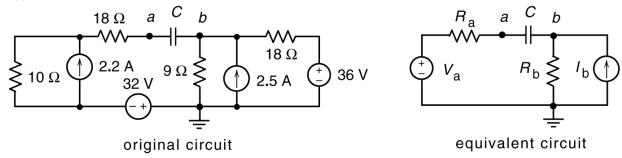
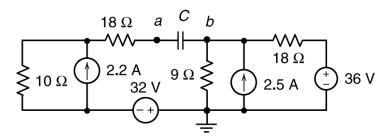


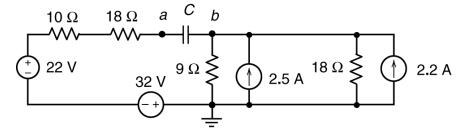
Figure P5.2-7

The equivalent circuit in Figure P5.2-7 is obtained from the original circuit using source transformations and equivalent resistances. (The lower case letters a and b identify the nodes of the capacitor in both the original and equivalent circuits.) Determine the values of  $R_a$ ,  $V_a$ ,  $R_b$  and  $I_b$  in the equivalent circuit.

## **Solution**



Performing a source transformation at each end of the circuit yields



Thenx

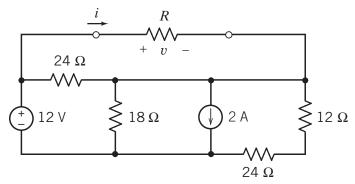
$$\begin{array}{c|cccc}
R_a & a & C & b \\
\hline
V_a & R_b & I_b
\end{array}$$

where

$$V_{\rm a} = 2.2(10) - 32 = -10 \text{ V}, \ R_{\rm a} = 18 + 10 = 28 \ \Omega, \ R_{\rm b} = 18 \parallel 9 = 6 \ \Omega \text{ and } I_{\rm b} = 2.5 + \frac{36}{18} = 4.5 \text{ A}$$

**P 5.2-8** The circuit shown in Figure P 5.2-8 contains an unspecified resistance *R*.

- (a) Determine the value of the current *i* when  $R = 4 \Omega$ .
- (b) Determine the value of the voltage v when  $R = 8 \Omega$ .
- (c) Determine the value of R that will cause i = 1 A.
- (d) Determine the value of R that will cause v = 16 V.



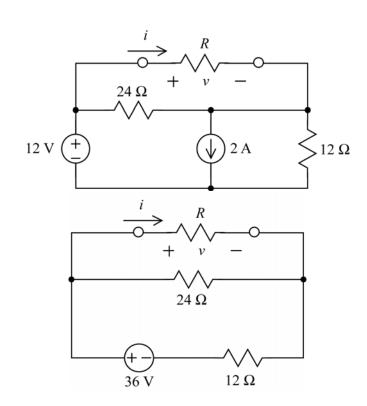
**Figure P 5.2-8** 

## **Solution:**

Replace series and parallel resistors by an equivalent resistor.

$$18 || (12 + 24) = 12 \Omega$$

Do a source transformation, then replace series voltage sources by an equivalent voltage source.



Do two more source transformations

Now current division gives

$$i = \left(\frac{8}{8+R}\right)3 = \frac{24}{8+R}$$

Then Ohm's Law gives

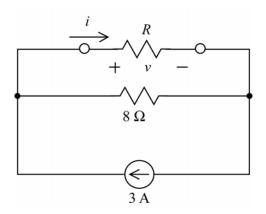
$$v = Ri = \frac{24R}{8+R}$$

(a) 
$$i = \frac{24}{8+4} = 2 \text{ A}$$

(b) 
$$v = \frac{24(8)}{8+8} = 12 \text{ V}$$

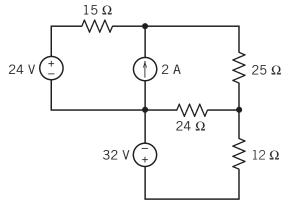
(c) 
$$1 = \frac{24}{8+R}$$
  $\Rightarrow$   $R = 16 \Omega$ 

(d) 
$$16 = \frac{24R}{8+R}$$
  $\Rightarrow$   $R = 16 \Omega$ 



(checked: LNAP 6/9/04)

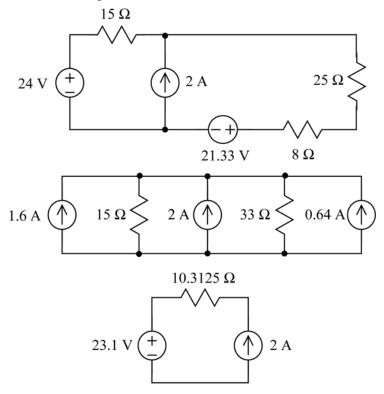
**P 5.2-9** Determine the value of the power supplied by the current source in the circuit shown in Figure P 5.2-9.



**Figure P 5.2-9** 

## **Solution:**

Use source transformations and equivalent resistances to reduce the circuit as follows



The power supplied by the current source is given by

$$p = [23.1 + 2(10.3125)]2 = 87.45 \text{ W}$$

## **Section 5-3 Superposition**

#### P5.3-1

The inputs to the circuit shown in Figure P5.3-1 are the voltage source voltages  $v_1$  and  $v_2$ . The output of the circuit is the voltage  $v_0$ . The output is related to the inputs by

$$v_0 = a v_1 + b v_2$$

where a and b are constants. Determine the values of a and b.

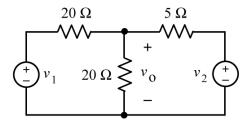


Figure P5.3-1

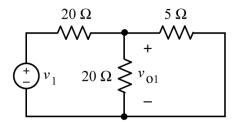
#### **Solution:**

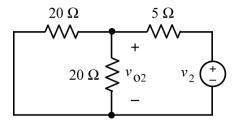
Let  $v_{01} = a v_1$  be the output when  $v_2 = 0$ . In this case, the right voltage source acts like a short circuit so we have the circuit show to the right. Then

$$v_{\text{ol}} = \frac{20 \parallel 5}{20 + (20 \parallel 5)} v_1 = \frac{4}{20 + 4} v_1 = \frac{1}{6} v_1 \implies a = \frac{1}{6}$$

Let  $v_{02} = b v_2$  be the output when  $v_1 = 0$ . In this case, the left voltage source acts like a short circuit so we have the circuit show to the right. Then

$$v_{02} = \frac{20 \parallel 20}{5 + (20 \parallel 20)} v_2 = \frac{10}{5 + 10} v_2 = \frac{2}{3} v_2 \implies b = \frac{2}{3}$$





#### P5.3-2

A particular linear circuit has two inputs,  $v_1$  and  $v_2$ , and one output,  $v_0$ . Three measurements are made. The first measurement shows that the output is  $v_0 = 4$  V when the inputs are  $v_1 = 2$  V and  $v_2 = 0$ . The second measurement shows that the output is  $v_0 = 10$  V when the inputs are  $v_1 = 0$  and  $v_2 = -2.5$  V. In the third measurement the inputs are  $v_1 = 3$  V and  $v_2 = 3$  V. What is the value of the output in the third measurement?

#### **Solution:**

The output of a linear circuit is a linear combination of the inputs:

$$v_0 = a_1 v_1 + a_2 v_2$$

From the first two measurements we have:

$$\begin{pmatrix} 4 \\ 10 \end{pmatrix} = \begin{pmatrix} 2 & 0 \\ 0 & -2.5 \end{pmatrix} \begin{pmatrix} a_1 \\ a_2 \end{pmatrix} \implies \begin{pmatrix} a_1 \\ a_2 \end{pmatrix} = \begin{pmatrix} 2 \\ -4 \end{pmatrix}$$

Now the output of the third measurement can be determine to be

$$v_0 = a_1(3) + a_2(3) = (2)(3) + (-4)(3) = -6 \text{ V}$$

#### P5.3-3

The circuit shown in Figure P5.3-3 has two inputs,  $v_s$  and  $i_s$ , and one output  $i_o$ . The output is related to the inputs by the equation

$$i_0 = ai_s + bv_s$$

Given the following two facts:

The output is  $i_0 = 0.45$  A when the inputs are  $i_s = 0.25$  A and  $v_s = 15$  V.

and

The output is  $i_0 = 0.30$  A when the inputs are  $i_s = 0.50$  A and  $v_s = 0$  V.

Determine the values of the constants a and b and the values of the resistances are  $R_1$  and  $R_2$ . Answers: a = 0.6 A/A, b = 0.02 A/V,  $R_1 = 30$   $\Omega$  and  $R_2 = 20$   $\Omega$ .

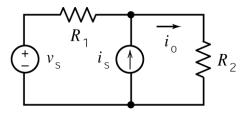


Figure P5.3-3

#### **Solution:**

From the 1st fact:

$$0.45 = a(0.25) + b(15)$$

From the 2nd fact:

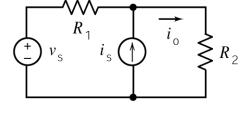
$$0.30 = a(0.50) + b(0) \implies a = \frac{0.30}{0.50} = 0.60$$

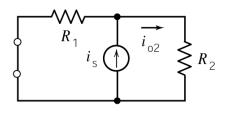
Substituting gives  $0.45 = (0.60)(0.25) + b(15) \implies b = \frac{0.45 - (0.60)(0.25)}{15} = 0.02$ 

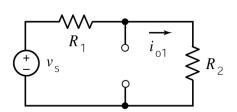
Next, consider the circuit:

$$ai_{s} = i_{o1} = i_{o}|_{v_{s}=0} = \left(\frac{R_{1}}{R_{1} + R_{2}}\right)i_{s}$$
so
$$0.60 = \frac{R_{1}}{R_{1} + R_{2}} \implies 2R_{1} = 3R_{2}$$
and
$$bv_{s} = i_{o2} = i_{o}|_{i_{s}=0} = \frac{v_{s}}{R_{1} + R_{2}}$$
so
$$0.02 = \frac{1}{R_{1} + R_{2}} \implies R_{1} + R_{2} = \frac{1}{0.02} = 50 \Omega$$

Solving these equations gives  $R_1 = 30 \Omega$  and  $R_2 = 20 \Omega$ .







**P 5.3-4** Use superposition to find the value of the voltage *v* in Figure P 5.3-4.

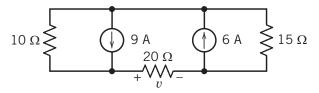
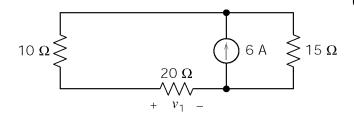


Figure P 5.3-4

#### **Solution:**

Consider 6 A source only (open 9 A source)

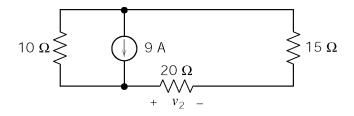


Use current division:

) 6 A 
$$\begin{cases} 15 \Omega & \frac{v_1}{20} = 6 \left[ \frac{15}{15 + 30} \right] \Rightarrow \frac{v_1}{15 + 30} \end{cases}$$

Consider 9 A source only (open 6 A source)

Use current division:



$$\frac{v_2}{20} = 9 \left[ \frac{10}{10 + 35} \right] \Rightarrow \underline{v_2} = 40 \text{ V}$$

$$v = v_1 + v_2 = 40 + 40 = 80 \text{ V}$$

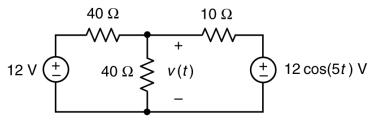
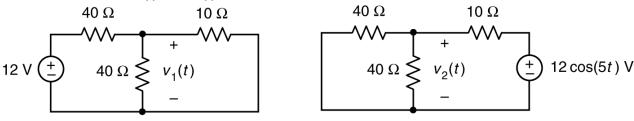


Figure P5.3-5

**P5.3-5** Determine v(t), the voltage across the vertical resistor in the circuit in Figure P5.3-5.

#### Solution;

We'll use superposition. Let  $v_1(t)$  the be the part of v(t) due to the voltage source acting alone. Similarly, let  $v_2(t)$  the be the part of v(t) due to the voltage source acting alone. We can use these circuits to calculate  $v_1(t)$  and  $v_2(t)$ .



Notice that  $v_1(t)$  is the voltage across parallel resistors. Using equivalent resistance, we calculate  $40||10 = 8 \Omega$ . Next, using voltage division we calculate

$$v_1(t) = \frac{8}{8+40}(12) = 2 \text{ V}$$

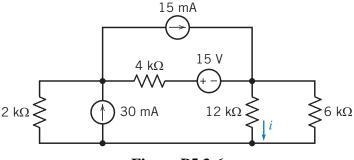
Similarly  $v_2(t)$  is the voltage across parallel resistors Using equivalent resistance we first determine  $40||40 = 20 \Omega$  and then calculate

$$v_2(t) = \frac{20}{10+20} (12\cos(5t)) = 8\cos(5t) \text{ V}$$
  
 $v(t) = v_1(t) + v_2(t) = 2 + 8\cos(5t) \text{ V}$ 

Using superposition

**P 5.3-6** Use superposition to find the value of the current *i* in Figure P 5.3-6.

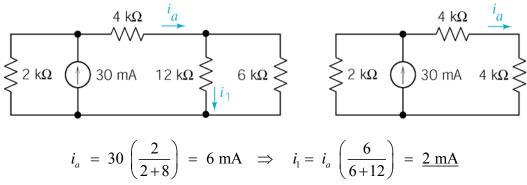
Answer: i = 3.5 mA



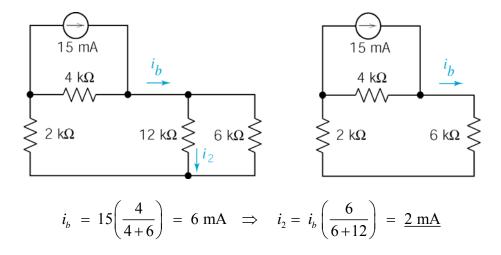
**Figure P5.3-6** 

#### **Solution:**

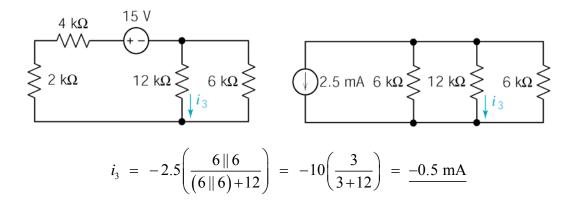
Consider 30 mA source only (open 15 mA and short 15 V sources). Let  $i_1$  be the part of i due to the 30 mA current source.



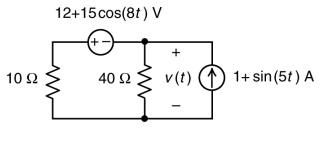
Consider 15 mA source only (open 30 mA source and short 15 V source) Let  $i_2$  be the part of i due to the 15 mA current source.



Consider 15 V source only (open both current sources). Let  $i_3$  be the part of i due to the 15 V voltage source.



Finally, 
$$i = i_1 + i_2 + i_3 = 2 + 2 - 0.5 = 3.5 \text{ mA}$$

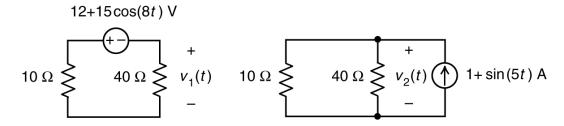


**Figure P5.3-7** 

**P5.3-7** Determine v(t), the voltage across the 40  $\Omega$  resistor in the circuit in Figure P5.3-7.

#### **Solution:**

We'll use superposition. Let  $v_1(t)$  the be the part of v(t) due to the voltage source acting alone. Similarly, let  $v_2(t)$  the be the part of v(t) due to the voltage source acting alone. We can use these circuits to calculate  $v_1(t)$  and  $v_2(t)$ .



Using voltage division we calculate

$$v_1(t) = -\frac{40}{10+40} (12+15\cos(8t)) = -9.6-12\cos(8t)$$

Using equivalent resistance we first determine  $10||40 = 8 \Omega$  and then calculate

$$v_2(t) = 8(1 + \sin(5t)) = 8 + 8\sin(5t)$$

Using superposition  $v(t) = v_1(t) + v_2(t) = -1.6 + 8\sin(5t) - 12\cos(8t)$  V

**P 5.3-8** Use superposition to find the value of the current  $i_x$  in Figure P 5.3-8.

Answer: i = 3.5 mA

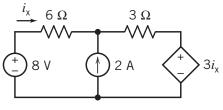
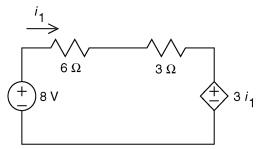


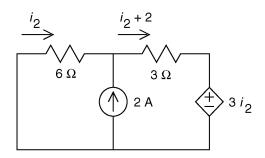
Figure P5.3-8

## **Solution:**

Consider 8 V source only (open the 2 A source)



Consider 2 A source only (short the 8 V source)



Finally,  $i_x = i_1 + i_2 = \frac{2}{3} - \frac{1}{2} = \frac{1}{6}$  A

Let  $i_1$  be the part of  $i_x$  due to the 8 V voltage source.

Apply KVL to the supermesh:

$$6(i_1) + 3(i_1) + 3(i_1) - 8 = 0$$

$$i_1 = \frac{8}{12} = \frac{2}{3}$$
 A

Let  $i_2$  be the part of  $i_x$  due to the 2 A current source.

Apply KVL to the supermesh:

$$6(i_2)+3(i_2+2)+3i_2=0$$

$$i_2 = \frac{-6}{12} = -\frac{1}{2} A$$

**P 5.3-9** The input to the circuit shown in Figure P 5.3-9 is the voltage source voltage,  $v_s$ . The output is the voltage  $v_o$ . The current source current,  $i_a$ , is used to adjust the relationship between the input and output. Design the circuit so that input and output are related by the equation  $v_o = 2v_s + 9$ .

*Hint:* Determine the required values of A and  $i_a$ .

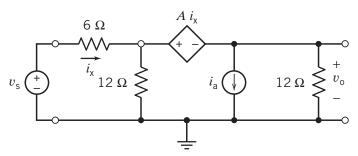


Figure P 5.3-9

**Solution:** 

$$i_{x} = \frac{v_{s} - v_{a}}{R_{1}}$$

$$v_{a} - v_{o} = A i_{x} = A \frac{v_{s} - v_{a}}{R_{1}}$$

$$v_{a} = \frac{R_{1} v_{o} + A v_{s}}{R_{1} + A}$$

$$v_{b} = \frac{R_{1} v_{o} + A v_{s}}{R_{1} + A}$$

Apply KCL to the supernode corresponding to the CCVS to get

$$\frac{v_{a} - v_{s}}{R_{1}} + \frac{v_{a}}{R_{2}} + i_{a} + \frac{v_{o}}{R_{3}} = 0$$

$$\frac{R_{1} + R_{2}}{R_{1} R_{2}} v_{a} - \frac{v_{s}}{R_{1}} + i_{a} + \frac{v_{o}}{R_{3}} = 0$$

$$\frac{R_{1} + R_{2}}{R_{1} R_{2}} \left( \frac{R_{1} v_{o} + A v_{s}}{R_{1} + A} \right) - \frac{v_{s}}{R_{1}} + i_{a} + \frac{v_{o}}{R_{3}} = 0$$

$$\left( \frac{R_{1} + R_{2}}{R_{2} (R_{1} + A)} + \frac{1}{R_{3}} \right) v_{o} + \left( \frac{(R_{1} + R_{2})A}{R_{1} R_{2} (R_{1} + A)} - \frac{1}{R_{1}} \right) v_{s} + i_{a} = 0$$

$$\frac{R_{3} (R_{1} + R_{2}) + R_{2} (R_{1} + A)}{R_{2} R_{3} (R_{1} + A)} v_{o} + \frac{A - R_{2}}{R_{2} (R_{1} + A)} v_{s} + i_{a} = 0$$

$$v_{o} = \frac{R_{3}(R_{2} - A)}{R_{3}(R_{1} + R_{2}) + R_{2}(R_{1} + A)}v_{s} - \frac{R_{2}R_{3}(R_{1} + A)}{R_{3}(R_{1} + R_{2}) + R_{2}(R_{1} + A)}i_{a}$$

When  $R_1 = 6 \Omega$ ,  $R_2 = 12 \Omega$  and  $R_3 = 12 \Omega$ 

$$v_{o} = \frac{12 - A}{24 + A} v_{s} - \frac{12(6 + A)}{24 + A} i_{a}$$

Comparing this equation to  $v_0 = 2v_s + 9$ , we requires

$$\frac{12-A}{24+A} = 2 \quad \Leftrightarrow \quad A = -12 \quad \frac{V}{A}$$

Then  $2v_s + 9 = v_o = 2v_s + 6i_a$  so we require

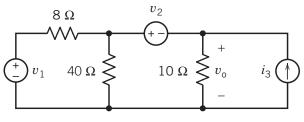
$$9 = 6i_a \implies i_a = 1.5 \text{ A}$$

(checked: LNAP 6/22/04)

**P 5.3-10** The circuit shown in Figure P 5.3-10 has three inputs:  $v_1$ ,  $v_2$ , and  $i_3$ . The output of the circuit is  $v_0$ . The output is related to the inputs by

$$v_0 = av_1 + bv_2 + ci_3$$

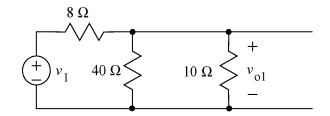
where a, b, and c are constants. Determine the values of a, b, and c.



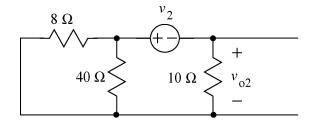
**Figure P 5.3-10** 

## **Solution:**

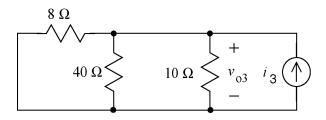
$$v_{\text{ol}} = \frac{40 \parallel 10}{8 + 40 \parallel 10} v_1 = \frac{1}{2} v_1 \implies a = \frac{1}{2}$$



$$v_{02} = -\frac{10}{8 \parallel 40 + 10} v_1 = -\frac{3}{5} v_2 \implies b = -\frac{3}{5}$$

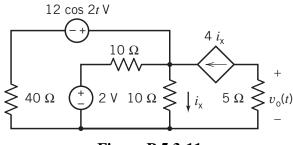


$$v_{03} = (8 \| 10 \| 40) i_3 = 4 i_3 \implies c = 4$$



(checked: LNAP 6/22/04)

**P 5.3-11** Determine the voltage  $v_0(t)$  for the circuit shown in Figure P 5.3-11.



**Figure P 5.3-11** 

**Solution:** Using superposition:

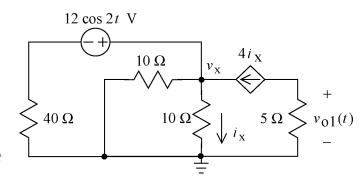
$$v_x = 10 i_x$$

and

$$\frac{v_{x} - 12\cos 2t}{40} + \frac{v_{x}}{10} + \frac{v_{x}}{10} = 4i_{x}$$

so

$$\frac{10i_{x} - 12\cos 2t}{40} = 2i_{x} \implies i_{x} = -\frac{12}{70}\cos 2t$$



Finally,

$$v_{o1} = -5(4i_x) = 3.429\cos 2t \text{ V}$$

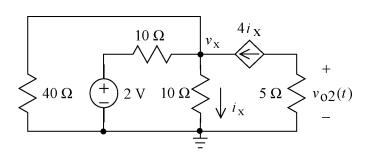
$$v_x = 10 i_x$$

and

$$\frac{v_x}{40} + \frac{v_x - 2}{10} + \frac{v_x}{10} = 4i_x$$

so

$$-0.2 = 1.75 i_x$$
  $\Rightarrow$   $i_x = -0.11429 \text{ A}$ 



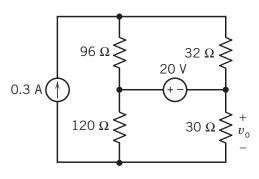
Finally,

$$v_{\rm ol} = -5(4i_{\rm x}) = 2.286 \text{ V}$$

$$v_0 = v_{01} + v_{02} = 3.429 \cos 2t + 2.286 \text{ V}$$

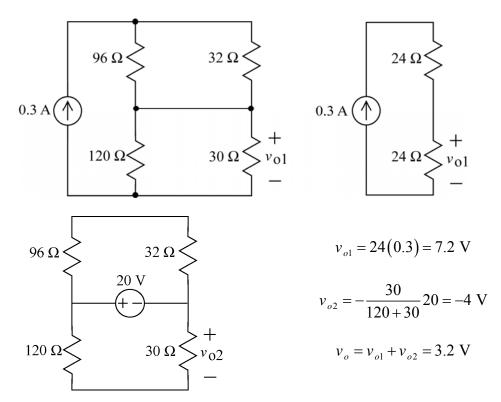
(checked: LNAP 6/22/04)

**P 5.3-12** Determine the value of the voltage  $v_0$  in the circuit shown in Figure P 5.3-12.



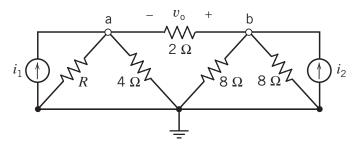
**Figure P 5.3-12** 

**Solution:** Using superposition:



(checked: LNAP 5/24/04)

**P 5.3-13** Determine the value of the voltage  $v_0$  in the circuit shown in Figure P 5.3-13.



**Figure P 5.3-13** 

#### **Solution:**

Using superposition

$$v_{o} = -2\left(\frac{R \parallel 4}{6 + (R \parallel 4)}\right)i_{1} + 2\left(\frac{4}{2 + (R \parallel 4) + 4}\right)i_{2}$$

Comparing to  $v_0 = -0.5 i_1 + 4$ , we require

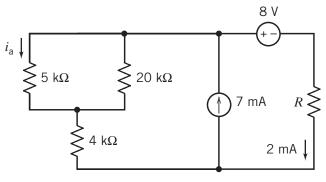
$$-2\left(\frac{R\parallel 4}{6+\left(R\parallel 4\right)}\right) = -0.5 \implies 4\left(R\parallel 4\right) = 6+\left(R\parallel 4\right) \implies R\parallel 4 = 2 \implies R = 4\Omega$$

and

$$2\left(\frac{4}{2+(R\|4)+4}\right)i_2 = 4 \implies 2\left(\frac{4}{2+(4\|4)+4}\right)i_2 = 4 \implies i_2 = 4 \text{ A}$$

(checked LNAP 6/12/04)

**P 5.3-14** Determine values of the current,  $i_a$ , and the resistance, R, for the circuit shown in Figure P 5.3-14.



**Figure P 5.3-14** 

## **Solution:**

Use units of mA,  $k\Omega$  and V.

$$4 + (5||20) = 8 k\Omega$$

(a) Using superposition

$$2 = \left(\frac{8}{R+8}\right)7 - \frac{8}{R+8} \implies 2(R+8) = 48 \implies R = 16 \text{ k}\Omega$$

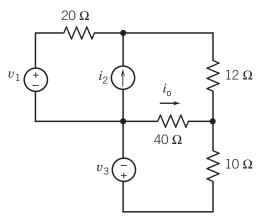
(b) Using superposition again

$$i_a = \left(\frac{5}{5+20}\right) \left[\left(\frac{16}{8+16}\right)7 + \frac{8}{8+16}\right] = \frac{4}{5}\left(\frac{2}{3}\times7 + \frac{1}{3}\right) = 4 \text{ mA}$$

**P 5.3-15** The circuit shown in Figure P 5.3-15 has three inputs:  $v_1$ ,  $i_2$ , and  $v_3$ . The output of the circuit is the current  $i_0$ . The output of the circuit is related to the inputs by

$$i_1 = av_0 + bv_2 + ci_3$$

where a, b, and c are constants. Determine the values of a, b, and c.



**Figure P 5.3-15** 

**Solution:** 

$$i_{o} = \left(-\frac{10}{10+40}\right) \left(\frac{v_{1}}{20+12+\left(40\parallel10\right)}\right) + \left(-\frac{10}{10+40}\right) \left(\frac{20}{20+\left[12+\left(40\parallel10\right)\right]}\right) i_{2} + \left(-\frac{20+12}{40+\left(20+12\right)}\right) \left(\frac{v_{3}}{10+\left[40\parallel\left(20+12\right)\right]}\right)$$

$$i_{o} = \left(-\frac{1}{200}\right)v_{1} + \left(-\frac{1}{10}\right)i_{2} + \left(-\frac{1}{62.5}\right)v_{3}$$

So

$$a = -0.05$$
,  $b = -0.1$  and  $c = -0.016$ 

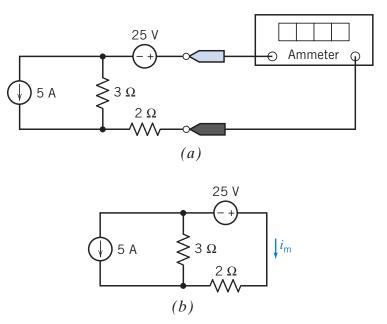
(checked: LNAP 6/19/04)

**P 5.3-16** Using the superposition principle, find the value of the current measured by the ammeter in Figure P 5.3-16*a*.

*Hint:* Figure P 5.3-16b shows the circuit after the ideal ammeter has been replaced by the equivalent short circuit and a label has been added to indicate the current measured by the ammeter,  $i_m$ .

Answer:

$$i_{\rm m} = \frac{25}{3+2} - \frac{3}{2+3} = 5 - 3 = 2 \text{ A}$$



**Figure P 5.3-16** 

**Solution:** 

$$i_m = \frac{25}{3+2} - \frac{3}{2+3} (5) = 5 - 3 = 2 \text{ A}$$

## Section 5-4: Thèvenin's Theorem

**P 5.4-1** Determine values of  $R_t$  and  $v_{oc}$  that cause the circuit shown in Figure P 5.4-1b to be the Thévenin equivalent circuit of the circuit in Figure P 5.4-1a.

*Hint:* Use source transformations and equivalent resistances to reduce the circuit in Figure P 5.4-1*a* until it is the circuit in Figure P 5.4-1*b*.

**Answer:**  $R_t = 5 \Omega$  and  $v_{oc} = 2 V$ 

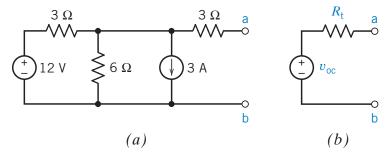
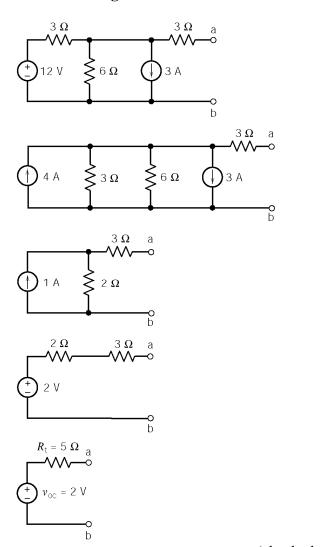


Figure P 5.4-1

**Solution:** 



**P 5.4-2** The circuit shown in Figure P 5.4-2*b* is the Thévenin equivalent circuit of the circuit shown in Figure P 5.4-2*a*. Find the value of the open-circuit voltage,  $v_{oc}$ , and Thévenin resistance,  $R_t$ .

**Answer:**  $v_{\text{oc}} = -12 \text{ V} \text{ and } R_{\text{t}} = 16 \Omega$ 

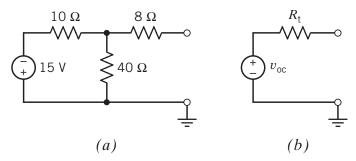
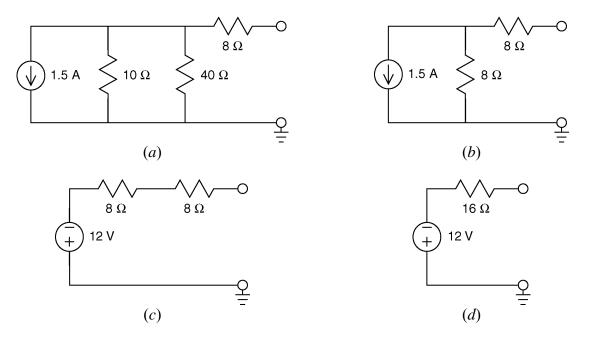


Figure P 5.4-2

#### **Solution:**

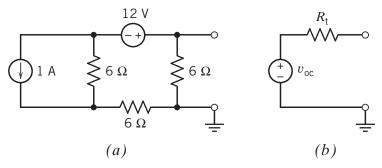
The circuit from Figure P5.4-2a can be reduced to its Thevenin equivalent circuit in four steps:



Comparing (d) to Figure P5.4-2b shows that the Thevenin resistance is  $R_t = 16 \Omega$  and the open circuit voltage,  $v_{oc} = -12 \text{ V}$ .

**P 5.4-3** The circuit shown in Figure P 5.4-3b is the Thévenin equivalent circuit of the circuit shown in Figure P 5.4-3a. Find the value of the open-circuit voltage,  $v_{oc}$ , and Thévenin resistance,  $R_{t}$ .

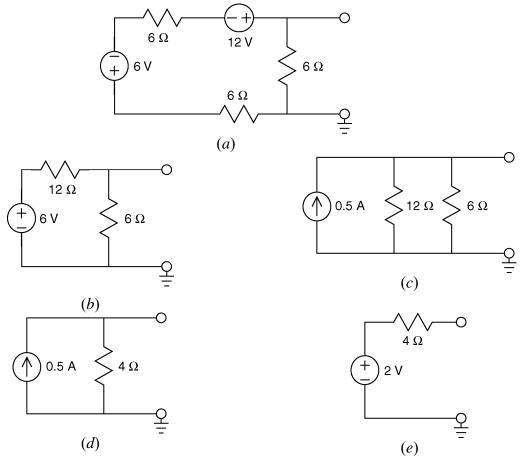
**Answer:**  $v_{oc} = 2 \text{ V} \text{ and } R_t = 4 \Omega$ 



**Figure P 5.4-3** 

#### **Solution:**

The circuit from Figure P5.4-3a can be reduced to its Thevenin equivalent circuit in five steps:



Comparing (e) to Figure P5.4-3b shows that the Thevenin resistance is  $R_t = 4 \Omega$  and the open circuit voltage,  $v_{oc} = 2 \text{ V}$ .

# **P 5.4-4** Find the Thévenin equivalent circuit for the circuit shown in Figure P 5.4-4.

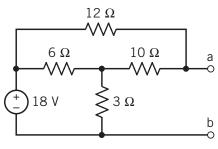
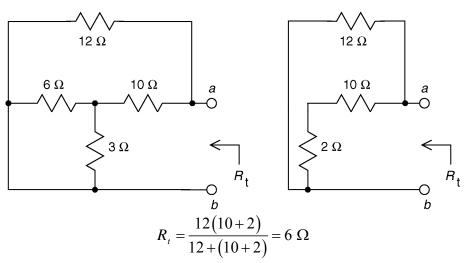
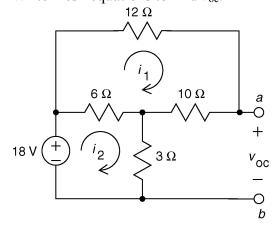


Figure P 5.4-4

Find  $R_t$ :



Write mesh equations to find  $v_{oc}$ :



Finally, 
$$v_{oc} = 3 i_2 + 10 i_1 = 3 \left(\frac{7}{3}\right) + 10 \left(\frac{1}{2}\right) = 12 \text{ V}$$

Mesh equations:

$$6(i_2 - i_1) + 3i_2 - 18 = 0$$

$$28i_1 = 6i_2$$

$$9i_2 - 6i_1 = 18$$

$$36i_1 = 18 \implies i_1 = \frac{1}{2} A$$

$$i_2 = \frac{14}{3} \left(\frac{1}{2}\right) = \frac{7}{3} A$$

 $12 i_1 + 10 i_1 - 6 (i_2 - i_1) = 0$ 

**P 5.4-5** Find the Thévenin equivalent circuit for the circuit shown in Figure P 5.4-5.

**Answer:**  $v_{oc} = -2 \text{ V} \text{ and } R_t = -8/3 \Omega$ 

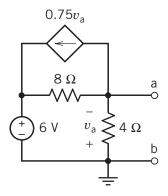


Figure P 5.4-4

#### **Solution:**

Find  $v_{oc}$ :

Notice that  $v_{oc}$  is the node voltage at node a. Express the controlling voltage of the dependent source as a function of the node voltage:

$$v_a = -v_{oc}$$

Apply KCL at node a:

$$-\left(\frac{6 - v_{oc}}{8}\right) + \frac{v_{oc}}{4} + \left(-\frac{3}{4}v_{oc}\right) = 0$$

$$-6 + v_{oc} + 2v_{oc} - 6v_{oc} = 0 \implies v_{oc} = -2 \text{ V}$$

Find  $R_t$ :

We'll find  $i_{sc}$  and use it to calculate  $R_t$ . Notice that the short circuit forces

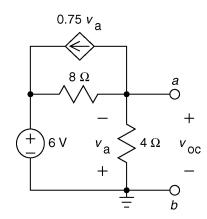
$$v_a = 0$$

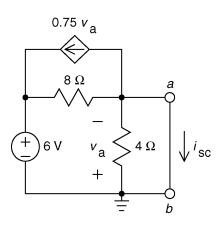
Apply KCL at node a:

$$-\left(\frac{6-0}{8}\right) + \frac{0}{4} + \left(-\frac{3}{4}0\right) + i_{sc} = 0$$

$$i_{sc} = \frac{6}{8} = \frac{3}{4} \text{ A}$$

$$R_{t} = \frac{v_{oc}}{i_{sc}} = \frac{-2}{3/4} = -\frac{8}{3}\Omega$$





**P 5.4-6** Find the Thévenin equivalent circuit for the circuit shown in Figure P 5.4-6.

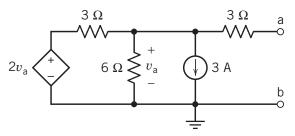
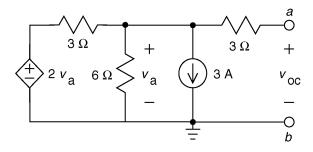


Figure P 5.4-6

#### **Solution:**

Find  $v_{oc}$ :

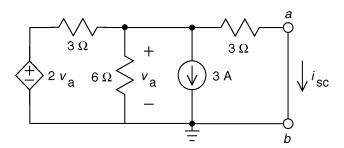


Apply KCL at the top, middle node:

$$\frac{2v_a - v_a}{3} = \frac{v_a}{6} + 3 + 0 \quad \Rightarrow \quad v_a = 18 \text{ V}$$

The voltage across the right-hand 3  $\Omega$  resistor is zero so:  $v_a = v_{oc} = 18 \text{ V}$ 

Find  $i_{sc}$ :



Apply KCL at the top, middle node:

$$\frac{2v_a - v_a}{3} = \frac{v_a}{6} + 3 + \frac{v_a}{3} \implies v_a = -18 \text{ V}$$

Apply Ohm's law to the right-hand 3  $\Omega$  resistor :

$$i_{sc} = \frac{v_a}{3} = \frac{-18}{3} = -6 \text{ V}$$

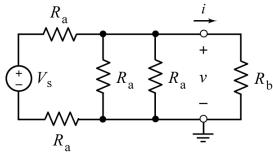
 $R_t = \frac{v_{oc}}{i_{co}} = \frac{18}{-6} = -3 \Omega$ Finally:

**P5.4-7** The equivalent circuit in Figure P5.4-7 is obtained by replacing part of the original circuit by its Thevenin equivalent circuit. The values of the parameters of the Thevenin equivalent circuit are

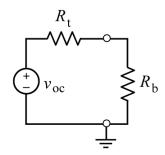
$$v_{\rm oc} = 15 \text{ V} \text{ and } R_{\rm t} = 60 \Omega$$

Determine the following:

- a.) The values of  $V_{\rm s}$  and  $R_{\rm a}$ . (Three resistors in the original circuit have equal resistance,  $R_{\rm a}$ .)
- b.) The value of  $R_b$  required to cause i = 0.2 A.
- c.) The value of  $R_b$  required to cause v = 12 V.



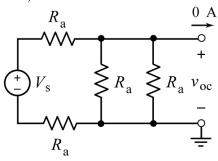
original circuit

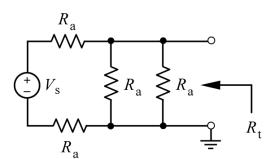


equivalent circuit

**Figure P5.4-7** 

Solution: a.) From



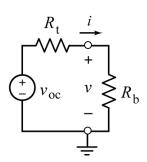


We see that  $v_{\text{oc}} = \frac{V_s}{5}$  and  $R_t = \frac{2}{5}R_a$ . With the given values of  $v_{\text{oc}}$  and  $R_t$  we calculate

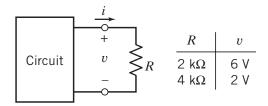
$$15 = \frac{V_s}{5}$$
  $\Rightarrow$   $V_s = 75 \text{ V}$  and  $60 = \frac{2}{5}R_a$   $\Rightarrow$   $R_a = 150 \Omega$ .

b.) 
$$i = \frac{v_{\text{oc}}}{R_{\text{t}} + R_{\text{b}}} \implies 0.2 = \frac{15}{60 + R_{\text{b}}} \implies R_{\text{b}} = 15 \Omega$$

c.) 
$$v = \frac{R_b}{R_t + R_b} v_{oc} \implies 12 = \frac{15R_b}{60 + R_b} \implies R_b = 240 \Omega$$

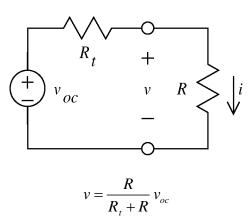


**P 5.4-8** A resistor, R, was connected to a circuit box as shown in Figure P 5.4-8. The voltage, v, was measured. The resistance was changed, and the voltage was measured again. The results are shown in the table. Determine the Thévenin equivalent of the circuit within the box and predict the voltage, v, when  $R = 8 \text{ k}\Omega$ .



**Figure P 5.4-8** 

#### **Solution:**



From the given data:

$$6 = \frac{2000}{R_t + 2000} v_{oc}$$

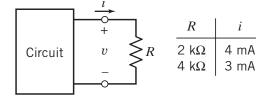
$$2 = \frac{4000}{R_t + 4000} v_{oc}$$

$$\Rightarrow \begin{cases} v_{oc} = 1.2 \text{ V} \\ R_t = -1600 \Omega \end{cases}$$

When  $R = 8000 \Omega$ ,

$$v = \frac{8000}{-1600 + 8000} (1.2) = 1.5 \text{ V}$$

**P 5.4-9** A resistor, *R*, was connected to a circuit box as shown in Figure P 5.4-9. The current, *i*, was measured. The resistance was changed, and the current was measured again. The results are shown in the table.

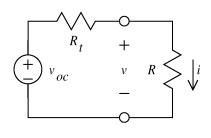


**Figure P 5.4-9** 

- (a) Specify the value of R required to cause i = 2 mA.
- (b) Given that R > 0, determine the maximum possible value of the current i.

*Hint:* Use the data in the table to represent the circuit by a Thévenin equivalent.

#### **Solution:**



$$i = \frac{v_{oc}}{R_t + R}$$

From the given data:

$$0.004 = \frac{v_{oc}}{R_t + 2000}$$

$$0.003 = \frac{v_{oc}}{R_t + 4000}$$

$$\Rightarrow \begin{cases} v_{oc} = 24 \text{ V} \\ R_t = 4000 \Omega \end{cases}$$

(a) When i = 0.002 A:

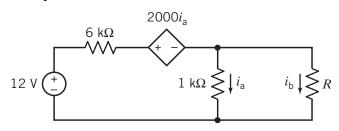
$$0.002 = \frac{24}{4000 + R} \quad \Rightarrow \quad R = 8000 \ \Omega$$

(b) Maximum i occurs when R = 0:

$$\frac{24}{4000} = 0.006 = 6 \text{ mA} \implies i \le 6 \text{ mA}$$

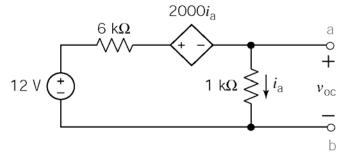
**P 5.4-10** For the circuit of Figure P 5.4-10, specify the resistance R that will cause current  $i_b$  to be 2 mA. The current  $i_a$  has units of amps.

*Hint:* Find the Thévenin equivalent circuit of the circuit connected to *R*.

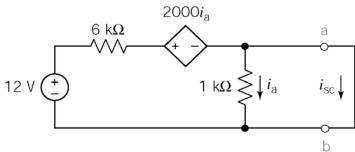


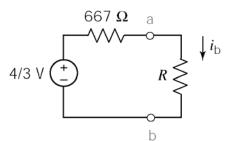
**Figure P 5.4-10** 

**Solution:** 



$$-12 + 6000 i_a + 2000 i_a + 1000 i_a = 0$$
$$i_a = 4/3000 \text{ A}$$
$$v_{oc} = 1000 i_a = \frac{4}{3} \text{ V}$$





 $i_a = 0$  due to the short circuit

$$-12 + 6000 i_{sc} = 0 \implies i_{sc} = 2 \text{ mA}$$

$$R_t = \frac{v_{oc}}{i_{sc}} = \frac{\frac{4}{3}}{.002} = 667 \ \Omega$$

$$i_b = \frac{\frac{4}{3}}{667 + R}$$

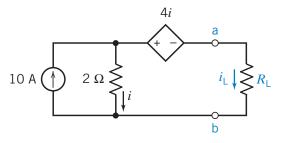
 $i_b = 0.002$  A requires

$$R = \frac{\frac{4}{3}}{0.002} - 667 = 0$$

(checked using LNAP 8/15/02)

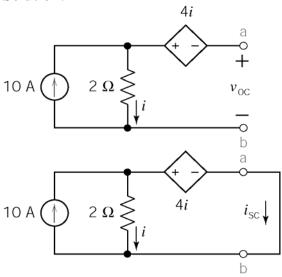
**P 5.4-11** For the circuit of Figure P 5.4-11, specify the value of the resistance  $R_L$  that will cause current  $i_L$  to be -2 A.

**Answer:**  $R_{\rm L} = 12 \Omega$ 



**Figure P 5.4-11** 

## **Solution:**



$$10 = i + 0 \implies i = 10 \text{ A}$$

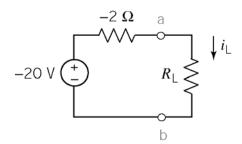
$$v_{oc} + 4i - 2i = 0$$

$$\implies v_{oc} = -2i = \underline{-20 \text{ V}}$$

$$i + i_{sc} = 10 \implies i = 10 - i_{sc}$$

$$4 i + 0 - 2 i = 0 \implies i = 0 \implies i_{sc} = 10 \text{ A}$$

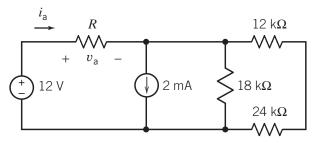
$$R_{t} = \frac{v_{oc}}{i_{sc}} = \frac{-20}{10} = -2 \Omega$$



$$-2 = i_L = \frac{-20}{R_L - 2} \implies R_L = 12 \,\Omega$$

(checked using LNAP 8/15/02)

- **P 5.4-12** The circuit shown in Figure P 5.4-12 contains an adjustable resistor. The resistance *R* can be set to any value in the range  $0 \le R \le 100 \text{ k}\Omega$ .
- (a) Determine the maximum value of the current  $i_a$  that can be obtained by adjusting R. Determine the corresponding value of R.
- (b) Determine the maximum value of the voltage  $v_a$  that can be obtained by adjusting R. Determine the corresponding value of R.
- (c) Determine the maximum value of the power supplied to the adjustable resistor that can be obtained by adjusting R. Determine the corresponding value of R.



**Figure P 5.4-12** 

**Solution:** Replace the part of the circuit that is connected to the variable resistor by its Thevenin equivalent circuit:

$$\begin{array}{c}
\stackrel{i_{a}}{\Longrightarrow} & \stackrel{R}{\nearrow} \\
+ v_{a} - \\
\downarrow^{i_{12} \text{ V}} & \downarrow^{2} \text{ mA}
\end{array}$$

$$\begin{array}{c}
12 \text{ k}\Omega \\
+ v_{a} - \\
\downarrow^{i_{12} \text{ V}} & \downarrow^{2} \text{ mA}
\end{array}$$

$$\begin{array}{c}
12 \text{ k}\Omega \\
+ v_{a} - \\
\downarrow^{i_{12} \text{ V}} & \downarrow^{2} \text{ mA}
\end{array}$$

$$\begin{array}{c}
i_{a} = \frac{36}{R + 12000} \text{ and } v_{a} = \frac{R}{R + 12000} 36 \\
\downarrow^{i_{12} \text{ k}\Omega} & \downarrow^{2} \text{ mA}
\end{array}$$

$$\begin{array}{c}
i_{a} = \frac{36}{R + 12000} \text{ and } v_{a} = \frac{R}{R + 12000} 36 \\
\downarrow^{i_{12} \text{ k}\Omega} & \downarrow^{2} \text{ mA}
\end{array}$$

$$\begin{array}{c}
i_{a} = \frac{36}{R + 12000} \text{ and } v_{a} = \frac{R}{R + 12000} 36 \\
\downarrow^{i_{12} \text{ k}\Omega} & \downarrow^{2} \text{ mA}
\end{array}$$

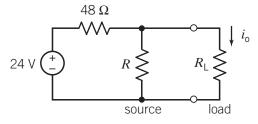
(a) 
$$i_a = \frac{36}{0 + 12000} = 3$$
 mA when  $R = 0$   $\Omega$  (a short circuit).

- (b)  $v_a = \frac{10^5}{10^5 + 12000} 36 = 32.14 \text{ V when } R \text{ is as large as possible, i.e. } R = 100 \text{ k}\Omega.$
- (c) Maximum power is delivered to the adjustable resistor when  $R = R_t = 12 \text{ k}\Omega$ . Then

$$p = i_a v_a = \left(\frac{36}{12000 + 12000}\right)^2 12000 = 0.027 = 27 \text{ mW}$$

(checked: LNAP 6/22/04)

**P 5.4-13** The circuit shown in Figure P 5.4-13 consists of two parts, the source (to the left of the terminals) and the load. The load consists of a single adjustable resistor having resistance  $0 \le R_{\rm L} \le 20~\Omega$ . The resistance R is fixed, but unspecified. When  $R_{\rm L} = 4~\Omega$ , the load current is measured to be  $i_0 = 0.375$  A. When  $R_{\rm L} = 8~\Omega$ , the value of the load current is  $i_0 = 0.300~{\rm A}$ .

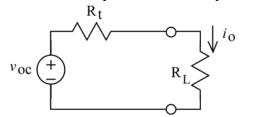


**Figure P 5.4-13** 

- (a) Determine the value of the load current when  $R_L = 10 \Omega$ .
- (b) Determine the value of R.

#### **Solution:**

Replace the source by it's Thevenin equivalent circuit to get



$$i_{\rm o} = \frac{v_{\rm oc}}{R_{\rm t} + R_{\rm L}}$$

Using the given formation

$$0.375 = \frac{v_{\text{oc}}}{R_{\text{t}} + 4}$$

$$0.300 = \frac{v_{\text{oc}}}{R_{\text{t}} + 8}$$

$$\Rightarrow 0.375(R_{\text{t}} + 4) = 0.300(R_{\text{t}} + 8)$$

So

$$R_t = \frac{(0.300)8 - (0.375)4}{0.075} = 12 \Omega \text{ and } v_{oc} = 0.3(12 + 8) = 6 \text{ V}$$

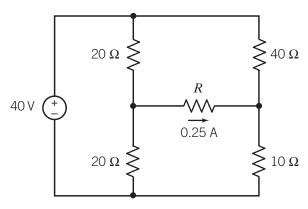
(a) When 
$$R_L = 10 \Omega$$
,  $i_o = \frac{6}{12 + 10} = 0.27\overline{27} A$ .

(b) 
$$12 \Omega = R_t = 48 11R \implies R = 16 \Omega$$
.

(checked: LNAP 5/24/04)

# **P 5.4-14** The circuit shown in Figure P 5.4-14 contains an unspecified resistance, *R*. Determine the value of *R* in each of the following two ways.

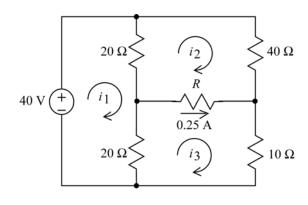
- (a) Write and solve mesh equations.
- (b) Replace the part of the circuit connected to the resistor R by a Thévenin equivalent circuit. Analyze the resulting circuit.



**Figure P 5.4-14** 

#### **Solution:**

(a)



$$i_3 - i_2 = 0.25 \text{ A}$$

Apply KVL to mesh 1 to get

$$20(i_1 - i_2) + 20(i_1 - i_3) - 40 = 0$$

Apply KVL to the supermesh corresponding to the unspecified resistance to get

$$40i_2 + 10i_3 - 20(i_1 - i_3) - 20(i_1 - i_2) = 0$$

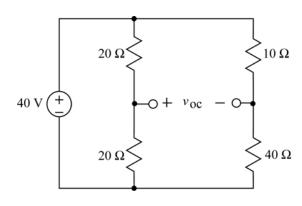
Solving, for example using MATLAB, gives

$$\begin{bmatrix} 0 & -1 & 1 \\ 40 & -20 & -20 \\ -40 & 60 & 30 \end{bmatrix} \begin{bmatrix} i_1 \\ i_2 \\ i_3 \end{bmatrix} = \begin{bmatrix} 0.25 \\ 40 \\ 0 \end{bmatrix} \implies \begin{bmatrix} i_1 \\ i_2 \\ i_3 \end{bmatrix} = \begin{bmatrix} 1.875 \\ 0.750 \\ 1.000 \end{bmatrix}$$

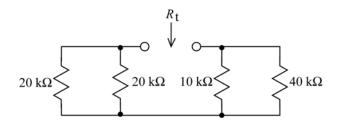
Apply KVL to mesh 2 to get

$$40i_2 + R(i_2 - i_3) - 20(i_1 - i_2) = 0 \qquad \Rightarrow \qquad R = \frac{20(i_1 - i_2) - 40i_2}{i_2 - i_3} = 30 \ \Omega$$





$$v_{\text{oc}} = \left(\frac{20}{20+20}\right) 40 - \left(\frac{40}{10+40}\right) 40 = -12 \text{ V}$$



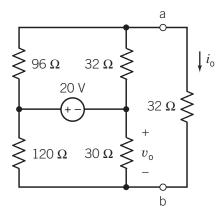
$$R_{\rm t} = 18 \ \Omega$$

$$\begin{array}{c|c}
 & R \\
\hline
0.25 \text{ A} \\
\hline
18 \Omega & 12 \text{ V}
\end{array}$$

$$0.25 = \frac{12}{18 + R} \qquad \Rightarrow \qquad R = 30 \ \Omega$$

(checked: LNAP 5/25/04)

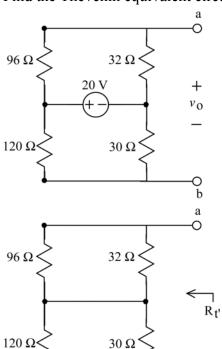
**P 5.4-15** Consider the circuit shown in Figure P 5.4-15. Replace the part of the circuit to the left of terminals a—b by its Thévenin equivalent circuit. Determine the value of the current  $i_0$ .



**Figure P 5.4-15** 

#### **Solution:**

Find the Thevenin equivalent circuit for the part of the circuit to the left of the terminals a-b.



Using voltage division twice

$$v_{\text{oc}} = \frac{32}{32 + 96} 20 - \frac{30}{120 + 30} 20 = 5 - 4 = 1 \text{ V}$$

$$R_{t} = (96 \parallel 32) + (120 \parallel 30) = 24 + 24 = 48 \Omega$$

Replacing the part of the circuit to the left of terminals a-b by its Thevenin equivalent circuit gives

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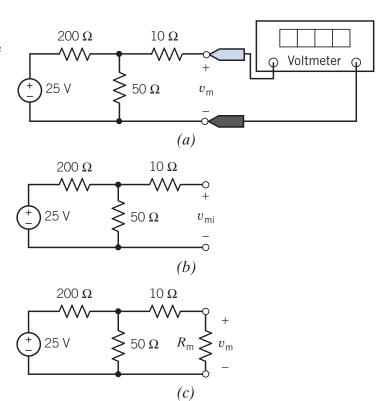
$$i_o = \frac{1}{48 + 32} = 0.0125 \text{ A} = 12.5 \text{ mA}$$

(checked: LNAP 5/24/04)

**P 5.4-16** An ideal voltmeter is modeled as an open circuit. A more realistic model of a voltmeter is a large resistance. Figure P 5.4-16a shows a circuit with a voltmeter that measures the voltage  $v_{\rm m}$ . In Figure P 5.4-16b the voltmeter is replaced by the model of an ideal voltmeter, an open circuit. The voltmeter measures  $v_{\rm mi}$ , the ideal value of  $v_{\rm m}$ .

As  $R_{\rm m} \to \infty$ , the voltmeter becomes an ideal voltmeter and  $v_{\rm m} \to v_{\rm mi}$ . When  $R_{\rm m} < \infty$ , the voltmeter is not ideal and  $v_{\rm m} > v_{\rm mi}$ . The difference between  $v_{\rm m}$  and  $v_{\rm mi}$  is a measurement error caused by the fact that the voltmeter is not ideal.

- (a) Determine the value of  $v_{\rm mi}$ .
- (b) Express the measurement error that occurs when  $R_{\rm m} = 1000~\Omega$  as a percentage of  $v_{\rm mi}$ .
- (c) Determine the minimum value of  $R_{\rm m}$  required to ensure that the measurement error is smaller than 2 percent of  $v_{\rm mi}$ .



**Figure P 5.4-16** 

**Solution**: Replace the circuit by its Thevenin equivalent circuit:

$$v_{m} = \left(\frac{R_{m}}{R_{m} + 50}\right) 5$$

$$v_{mi} = \lim_{R_{m} \to \infty} v_{m} = 5$$

$$v_{mi} = 5$$

$$v_{mi} = 0$$

$$v_{mi} = 0$$

(b) When 
$$R_{\rm m} = 1000 \,\Omega$$
,  $v_{\rm m} = 4.763 \,\text{V}$  so 
$$\% \,\text{error} = \frac{5 - 4.762}{5} \times 100 = 4.76\%$$

#### P5.4-17

Given that  $0 \le R \le \infty$  in the circuit shown in Figure P5.4-17, consider these two observations:

When  $R = 2 \Omega$  then  $v_R = 4 \text{ V}$  and  $i_R = 2 \text{ A}$ . Observation 1:

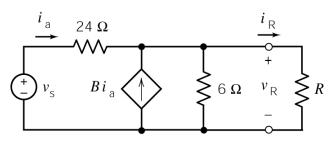
Observation 1: When  $R = 6 \Omega$  then  $v_R = 6 V$  and  $i_R = 1 A$ .

Determine the following

a) The maximum value of  $i_R$  and the value of R that causes  $i_R$  to be maximal.

b) The maximum value of  $v_R$  and the value of R that causes  $v_R$  to be maximal.

c) The maximum value of  $p_R = i_R v_R$  and the value of R that causes  $p_R$  to be maximal.

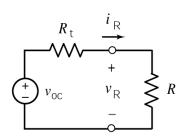


**Figure P5.4-17** 

#### **Solution:**

We can replace the part of the circuit to the left of the terminals by its Thevenin equivalent circuit:

Using voltage division  $v_R = \frac{R}{R+R} v_{oc}$  and using Ohm's law



$$i_{\rm R} = \frac{v_{\rm oc}}{R + R_{\rm t}}$$

$$i_{R} = \frac{v_{oc}}{R + R_{t}}.$$

$$k_{R} = \frac{v_{oc}}{R + R_{t}}.$$
By inspection,  $v_{R} = \frac{R}{R + R_{t}}v_{oc} = \frac{v_{oc}}{1 + \frac{R_{t}}{R}}$  will be maximum when

 $R = \infty$ . The maximum value of  $v_R$  will be  $v_{oc}$ . Similarly,

 $i_{\rm R} = \frac{v_{\rm oc}}{R + R}$  will be maximum when R = 0. The maximum value

of 
$$i_{\rm R}$$
 will be  $\frac{v_{\rm oc}}{R_{\star}} = i_{\rm sc}$ .

The maximum power transfer theorem tells use that  $p_R = i_R v_R$  will be maximum when  $R = R_t$ .

Then 
$$p_{R} = i_{R} v_{R} = \left(\frac{v_{oc}}{R + R_{t}}\right) \left(\frac{R}{R + R_{t}} v_{oc}\right) = R \left(\frac{v_{oc}}{R + R_{t}}\right)^{2}$$
.

Let's substitute the given data into the equation  $i_R = \frac{v_{oc}}{R + R_t}$ .

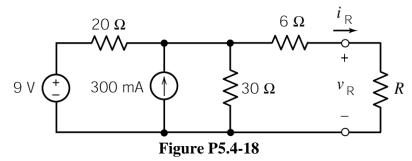
When 
$$R = 2 \Omega$$
 we get  $2 = \frac{v_{\text{oc}}}{2 + R_{\text{t}}} \implies 4 + 2R_{\text{t}} = v_{\text{oc}}$ . When  $R = 6 \Omega$  we get

$$1 = \frac{v_{\text{oc}}}{6 + R_{\text{t}}} \quad \Rightarrow \quad 6 + R_{\text{t}} = v_{\text{oc}} .$$

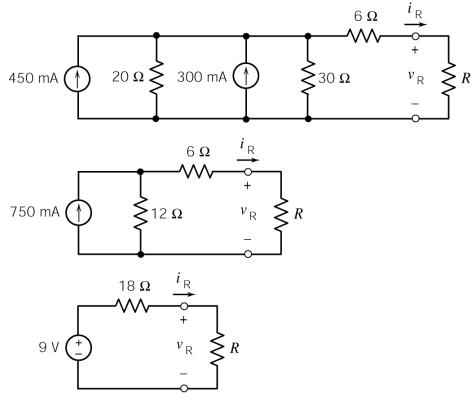
So 
$$6 + R_t = 4 + 2R_t \implies R_t = 2 \Omega$$
 and  $v_{oc} = 4 + 2R_t = 8 \text{ V}$ . Also  $i_{sc} = \frac{v_{oc}}{R_t} = \frac{8}{2} = 4 \text{ A}$ .

# P5.4-18 Consider the circuit shown in Figure P5.4-18. Determine

- a) The value of  $v_R$  that occurs when  $R = 9 \Omega$ .
- b) The value of R that causes  $v_R = 5.4 \text{ V}$ .
- c) The value of R that causes  $i_R = 300 \text{ mA}$ .



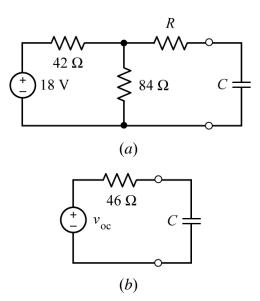
**Solution:** Reduce this circuit using source transformations and equivalent resistance:



Now  $v_R = \left(\frac{R}{R+18}\right)9$  and  $i_R = \frac{9}{R+18}$  so the questions can be easily answered:

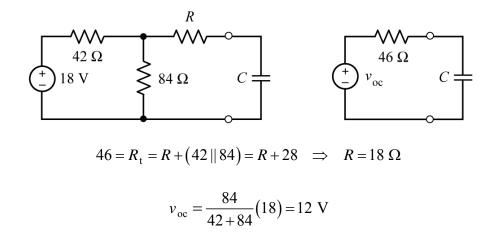
- a) When  $R = 9 \Omega$  then  $v_R = 3 \text{ V}$ .
- b) When  $R = 27 \Omega$  then  $v_R = 5.4 \text{ V}$ .
- c) When  $R = 12 \Omega$  then  $i_R = 300 \text{ mA}$ .

**P5.4-19** The circuit shown in Figure P5.4-19a can be reduced to the circuit shown in Figure P5.4-19b using source transformations and equivalent resistances. Determine the values of the source voltage  $v_{\rm oc}$  and the resistance R.



**Figure P5.4-19** 

## **Solution**

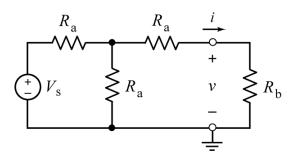


**P5.4-20** The equivalent circuit in Figure P5.4-20 is obtained by replacing part of the original circuit by its Thevenin equivalent circuit. The values of the parameters of the Thevenin equivalent circuit are

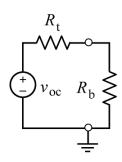
$$v_{\rm oc} = 15 \text{ V} \text{ and } R_{\rm t} = 60 \Omega$$

Determine the following:

- a.) The values of  $V_s$  and  $R_a$ . (Three resistors in the original circuit have equal resistance,  $R_a$ .)
- b.) The value of  $R_b$  required to cause i = 0.2 A.
- c.) The value of  $R_b$  required to cause v = 5 V.



original circuit

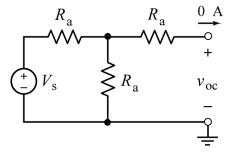


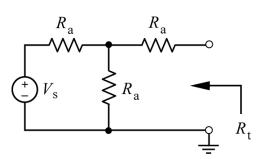
equivalent circuit

**Figure P5.4-20** 

#### **Solution**

a.) From



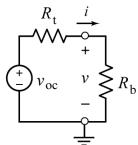


We see that  $v_{\text{oc}} = \frac{V_{\text{s}}}{2}$  and  $R_{\text{t}} = \frac{3}{2}R_{\text{a}}$ . With the given values of  $v_{\text{oc}}$  and  $R_{\text{t}}$  we calculate

$$15 = \frac{V_s}{2}$$
  $\Rightarrow$   $V_s = 30 \text{ V} \text{ and } 60 = \frac{3}{2}R_a$   $\Rightarrow$   $R_a = 40 \Omega$ .

b.) 
$$i = \frac{v_{\text{oc}}}{R_{\text{t}} + R_{\text{b}}} \implies 0.2 = \frac{15}{60 + R_{\text{b}}} \implies R_{\text{b}} = 15 \Omega$$

c.) 
$$v = \frac{R_b}{R_t + R_b} v_{oc} \implies 5 = \frac{15 R_b}{60 + R_b} \implies R_b = 30 \Omega$$



## **Section 5-5: Norton's Theorem**

**P5.5-1** The part of the circuit shown in Figure P5.3-1a to the left of the terminals can be reduced to its Norton equivalent circuit using source transformations and equivalent resistance. The resulting Norton equivalent circuit, shown in Figure P5.3-1b, will be characterized by the parameters:

$$i_{\rm sc} = 0.5 \, \text{A}$$
 and  $R_{\rm t} = 20 \, \Omega$ 

- a) Determine the values of  $v_s$  and  $R_1$ .
- b) Given that  $0 \le R_2 \le \infty$ , determine the maximum values of the voltage, v, and of the power, p = vi.

**Answers:**  $v_s = 37.5 \text{ V}, R_1 = 25 \Omega, \text{ max } v = 10 \text{ V} \text{ and } \text{max } p = 1.25 \text{ W}$ 

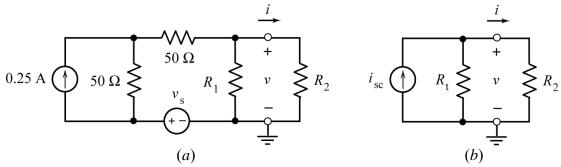
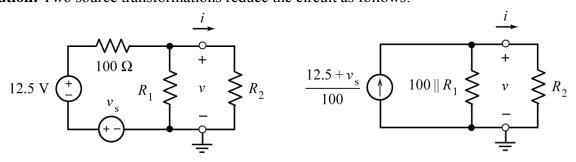


Figure P5.5-1

**Solution:** Two source transformations reduce the circuit as follows:



Recognizing the parameters of the Norton equivalent circuit gives:

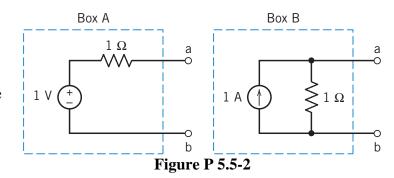
$$0.5 = i_{sc} = \frac{12.5 + v_s}{100} \implies v_s = 37.5 \text{ V} \text{ and } 20 = R_t = 100 \parallel R_1 = \frac{100 R_1}{100 + R_1} \implies R_1 = 25 \Omega$$

Next, the voltage across resistor  $R_2$  is given by  $v = i_{sc} \left( R_t \parallel R_2 \right) = \frac{R_t R_2 i_{sc}}{R_t + R_2} = \frac{R_t i_{sc}}{R_t}$  so this

voltage is maximum when  $R_2 = \infty$  and max  $v = R_t i_{sc} = 10$  V. The power p = vi will be

maximum when 
$$R_2 = R_t = 20 \ \Omega$$
. Then  $v = \frac{R_t i_{sc}}{2} = \frac{20(0.5)}{2} = 5 \ V$ ,  $i = \frac{v}{R_2} = \frac{5}{20} = 0.25 \ A$  and  $p = v i = 5(0.25) = 1.25 \ W$ .

P 5.5-2 Two black boxes are shown in Figure P 5.5-2. Box A contains the Thévenin equivalent of some linear circuit, and box B contains the Norton equivalent of the same circuit. With access to just the outsides of the boxes and their terminals, how can you determine which is which, using only one shorting wire?



#### **Solution:**

When the terminals of the boxes are open-circuited, no current flows in Box A, but the resistor in Box B dissipates 1 watt. Box B is therefore warmer than Box A. If you short the terminals of each box, the resistor in Box A will draw 1 amp and dissipate 1 watt. The resistor in Box B will be shorted, draw no current, and dissipate no power. Then Box A will warm up and Box B will cool off.

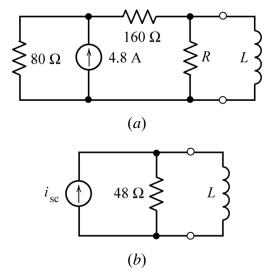
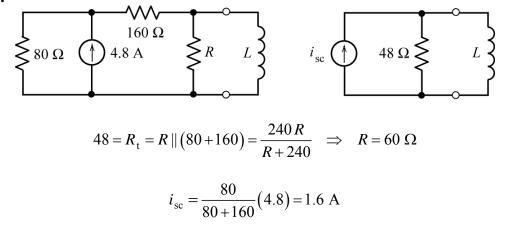


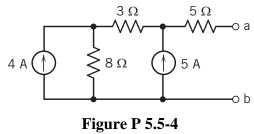
Figure P5.5-3

**P5.5-3** The circuit shown in Figure P5.5-3a can be reduced to the circuit shown in Figure P5.5-3b using source transformations and equivalent resistances. Determine the values of the source current  $i_{sc}$  and the resistance R.

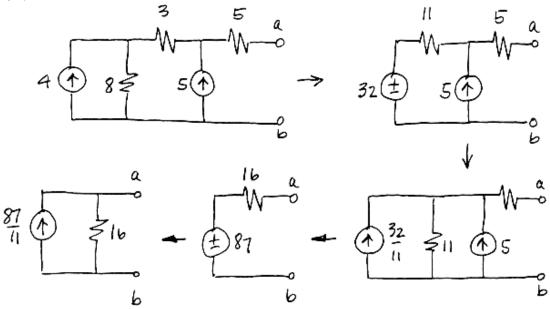
# **Solution:**



# **P 5.5-4** Find the Norton equivalent circuit for the circuit shown in Figure P 5.5-4.



# **Solution:**



**P 5.5-5** The circuit shown in Figure P 5.5-5b is the Norton equivalent circuit of the circuit shown in Figure P 5.5-5a. Find the value of the short-circuit current,  $i_{sc}$ , and Thévenin resistance,  $R_{\rm t}$ .

**Answer:**  $i_{sc} = 1.13 \text{ A} \text{ and } R_t = 7.57 \Omega$ 

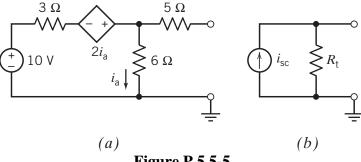


Figure P 5.5-5

#### **Solution:**

To determine the value of the short circuit current,  $i_{sc}$ , we connect a short circuit across the terminals of the circuit and then calculate the value of the current in that short circuit. Figure (a) shows the circuit from Figure 5.6-4a after adding the short circuit and labeling the short circuit current. Also, the meshes have been identified and labeled in anticipation of writing mesh equations. Let  $i_1$  and  $i_2$  denote the mesh currents in meshes 1 and 2, respectively.

In Figure (a), mesh current  $i_2$  is equal to the current in the short circuit. Consequently,  $i_2 = i_{sc}$ . The controlling current of the CCVS is expressed in terms of the mesh currents as

$$i_a = i_1 - i_2 = i_1 - i_{sc}$$

Apply KVL to mesh 1 to get

$$3i_1 - 2(i_1 - i_2) + 6(i_1 - i_2) - 10 = 0 \implies 7i_1 - 4i_2 = 10$$
 (1)

Apply KVL to mesh 2 to get

$$5i_2 - 6(i_1 - i_2) = 0 \implies -6i_1 + 11i_2 = 0 \implies i_1 = \frac{11}{6}i_2$$

Substituting into equation 1 gives

**Figure** (a) Calculating the short circuit current,  $i_{sc}$ , using mesh equations.

To determine the value of the Thevenin resistance,  $R_t$ , first replace the 10 V voltage source by a 0 V voltage source, i.e. a short circuit. Next, connect a current source across the terminals of the circuit and then label the voltage across that current source as shown in Figure (b). The Thevenin resistance will be calculated from the current and voltage of the current source as

$$R_{t} = \frac{v_{T}}{i_{T}}$$

In Figure (b), the meshes have been identified and labeled in anticipation of writing mesh equations. Let  $i_1$  and  $i_2$  denote the mesh currents in meshes 1 and 2, respectively.

In Figure (b), mesh current  $i_2$  is equal to the negative of the current source current. Consequently,  $i_2 = i_T$ . The controlling current of the CCVS is expressed in terms of the mesh currents as

$$i_a = i_1 - i_2 = i_1 + i_T$$

Apply KVL to mesh 1 to get

$$3i_1 - 2(i_1 - i_2) + 6(i_1 - i_2) = 0 \implies 7i_1 - 4i_2 = 0 \implies i_1 = \frac{4}{7}i_2$$
 (2)

Apply KVL to mesh 2 to get

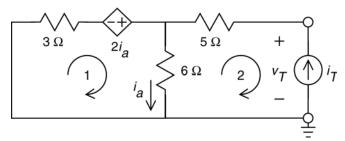
$$5i_2 + v_T - 6(i_1 - i_2) = 0 \implies -6i_1 + 11i_2 = -v_T$$

Substituting for  $i_1$  using equation 2 gives

$$-6\left(\frac{4}{7}i_2\right) + 11i_2 = -v_T \implies 7.57i_2 = -v_T$$

Finally,

$$R_{t} = \frac{v_{T}}{i_{T}} = \frac{-v_{T}}{-i_{T}} = \frac{-v_{T}}{i_{2}} = 7.57 \Omega$$



**Figure (b)** Calculating the Thevenin resistance,  $R_t = \frac{v_T}{i_T}$ , using mesh equations.

To determine the value of the open circuit voltage,  $v_{oc}$ , we connect an open circuit across the terminals of the circuit and then calculate the value of the voltage across that open circuit. Figure (c) shows the circuit from Figure 4.6-4a after adding the open circuit and labeling the

open circuit voltage. Also, the meshes have been identified and labeled in anticipation of writing mesh equations. Let  $i_1$  and  $i_2$  denote the mesh currents in meshes 1 and 2, respectively.

In Figure (c), mesh current  $i_2$  is equal to the current in the open circuit. Consequently,  $i_2 = 0$  A. The controlling current of the CCVS is expressed in terms of the mesh currents as

$$i_a = i_1 - i_2 = i_1 - 0 = i_1$$

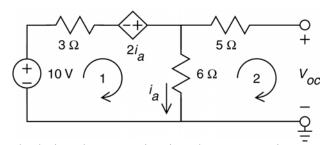
Apply KVL to mesh 1 to get

$$3 i_{1} - 2(i_{1} - i_{2}) + 6(i_{1} - i_{2}) - 10 = 0 \implies 3 i_{1} - 2(i_{1} - 0) + 6(i_{1} - 0) - 10 = 0$$

$$\Rightarrow i_{1} = \frac{10}{7} = 1.43 \text{ A}$$

Apply KVL to mesh 2 to get

$$5i_2 + v_{oc} - 6(i_1 - i_2) = 0 \implies v_{oc} = 6(i_1) = 6(1.43) = 8.58 \text{ V}$$



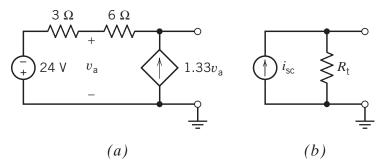
**Figure** (c) Calculating the open circuit voltage,  $v_{oc}$ , using mesh equations.

As a check, notice that  $R_t i_{sc} = (7.57)(1.13) = 8.55 \approx v_{oc}$ 

(checked using LNAP 8/16/02)

**P 5.5-6** The circuit shown in Figure P 5.5-6b is the Norton equivalent circuit of the circuit shown in Figure P 5.5-6a. Find the value of the short-circuit current,  $i_{sc}$ , and Thévenin resistance,  $R_t$ .

**Answer:**  $i_{sc} = -24 \text{ A} \text{ and } R_t = -3 \Omega$ 



**Figure P 5.5-6** 

#### **Solution:**

To determine the value of the short circuit current,  $I_{sc}$ , we connect a short circuit across the terminals of the circuit and then calculate the value of the current in that short circuit. Figure (a) shows the circuit from Figure 4.6-5a after adding the short circuit and labeling the short circuit current. Also, the nodes have been identified and labeled in anticipation of writing node equations. Let  $v_1$ ,  $v_2$  and  $v_3$  denote the node voltages at nodes 1, 2 and 3, respectively.

In Figure (a), node voltage  $v_1$  is equal to the negative of the voltage source voltage. Consequently,  $v_1 = -24$  V. The voltage at node 3 is equal to the voltage across a short,  $v_3 = 0$ . The controlling voltage of the VCCS,  $v_a$ , is equal to the node voltage at node 2, i.e.  $v_a = v_2$ . The voltage at node 3 is equal to the voltage across a short, i.e.  $v_3 = 0$ .

Apply KCL at node 2 to get

$$\frac{v_1 - v_2}{3} = \frac{v_2 - v_3}{6} \implies 2v_1 + v_3 = 3v_2 \implies -48 = 3v_a \implies v_a = -16 \text{ V}$$

Apply KCL at node 3 to get

$$\frac{v_2 - v_3}{6} + \frac{4}{3}v_2 = i_{sc} \implies \frac{9}{6}v_a = i_{sc} \implies i_{sc} = \frac{9}{6}(-16) = -24 \text{ A}$$

$$0$$

$$3\Omega + 6\Omega$$

$$+ 24 \text{ V} \quad v_a$$

$$- 1.33v_a$$

$$- 1/sc$$

**Figure** (a) Calculating the short circuit current,  $I_{sc}$ , using mesh equations.

To determine the value of the Thevenin resistance,  $R_{th}$ , first replace the 24 V voltage source by a 0 V voltage source, i.e. a short circuit. Next, connect a current source circuit across

the terminals of the circuit and then label the voltage across that current source as shown in Figure (b). The Thevenin resistance will be calculated from the current and voltage of the current source as

$$R_{th} = \frac{v_T}{i_T}$$

Also, the nodes have been identified and labeled in anticipation of writing node equations. Let  $v_1$ ,  $v_2$  and  $v_3$  denote the node voltages at nodes 1, 2 and 3, respectively.

In Figure (b), node voltage  $v_1$  is equal to the across a short circuit, i.e.  $v_1 = 0$ . The controlling voltage of the VCCS,  $v_a$ , is equal to the node voltage at node 2, i.e.  $v_a = v_2$ . The voltage at node 3 is equal to the voltage across the current source, i.e.  $v_3 = v_T$ .

Apply KCL at node 2 to get

$$\frac{v_1 - v_2}{3} = \frac{v_2 - v_3}{6} \implies 2v_1 + v_3 = 3v_2 \implies v_T = 3v_a$$

Apply KCL at node 3 to get

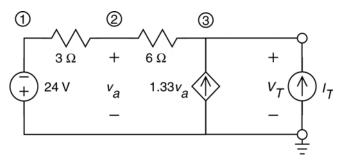
$$\frac{v_2 - v_3}{6} + \frac{4}{3}v_2 + i_T = 0 \implies 9v_2 - v_3 + 6i_T = 0$$

$$\implies 9v_a - v_T + 6i_T = 0$$

$$\implies 3v_T - v_T + 6i_T = 0 \implies 2v_T = -6i_T$$

Finally,

$$R_{t} = \frac{v_{T}}{i_{T}} = -3 \Omega$$



**Figure (b)** Calculating the Thevenin resistance,  $R_{th} = \frac{v_T}{i_T}$ , using mesh equations.

To determine the value of the open circuit voltage,  $v_{oc}$ , we connect an open circuit across the terminals of the circuit and then calculate the value of the voltage across that open circuit. Figure (c) shows the circuit from Figure P 4.6-5a after adding the open circuit and labeling the open circuit voltage. Also, the nodes have been identified and labeled in anticipation of writing node equations. Let  $v_1$ ,  $v_2$  and  $v_3$  denote the node voltages at nodes 1, 2 and 3, respectively.

In Figure (c), node voltage  $v_1$  is equal to the negative of the voltage source voltage. Consequently,  $v_1 = -24$  V. The controlling voltage of the VCCS,  $v_a$ , is equal to the node voltage at node 2, i.e.  $v_a = v_2$ . The voltage at node 3 is equal to the open circuit voltage, i.e.  $v_3 = v_{oc}$ .

Apply KCL at node 2 to get

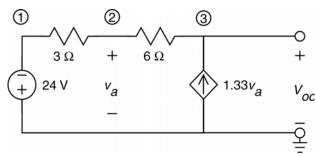
$$\frac{v_1 - v_2}{3} = \frac{v_2 - v_3}{6} \implies 2v_1 + v_3 = 3v_2 \implies -48 + v_{oc} = 3v_a$$

Apply KCL at node 3 to get

$$\frac{v_2 - v_3}{6} + \frac{4}{3}v_2 = 0 \implies 9v_2 - v_3 = 0 \implies 9v_a = v_{oc}$$

Combining these equations gives

$$3(-48 + v_{oc}) = 9 v_a = v_{oc} \implies v_{oc} = 72 \text{ V}$$



**Figure** (c) Calculating the open circuit voltage,  $v_{oc}$ , using node equations.

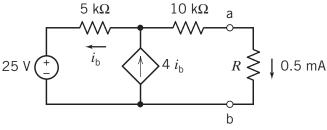
As a check, notice that

$$R_{th} I_{sc} = (-3)(-24) = 72 = V_{oc}$$

(checked using LNAP 8/16/02)

**P 5.5-7** Determine the value of the resistance *R* in the circuit shown in Figure P 5.5-7 by each of the following methods:

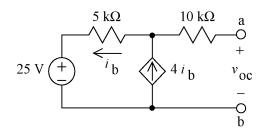
(a) Replace the part of the circuit to the left of terminals a—b by its the Norton equivalent circuit. Use current division to determine the value of *R*.



**Figure P 5.5-7** 

(b) Analyze the circuit shown Figure P 5.5-6 using mesh equations. Solve the mesh equations to determine the value of R.

**Solution:** (a) Replace the part of the circuit that is connected to the left of terminals a-b by its Norton equivalent circuit:



Apply KCL at the top node of the dependent source to see that  $i_b = 0$  A. Then

$$v_{\rm oc} = 25 + 5000 (i_{\rm b}) = 25 \text{ V}$$

8.33 kΩ

3 mA

a

Apply KVL to the supermesh corresponding to the dependent source to get

$$-5000 i_b + 10000 (3 i_b) - 25 = 0 \implies i_b = 1 \text{ mA}$$

Apply KCL to get

$$i_{\rm sc} = 3 i_{\rm b} = 3 \,\mathrm{mA}$$

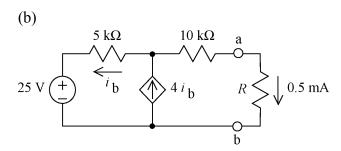
Then

0.5 mA

$$R_{\rm t} = \frac{v_{\rm oc}}{i_{\rm sc}} = 8.3\overline{3} \text{ k}\Omega$$

Current division gives

$$0.5 = \frac{8333}{R + 8333}$$
  $\Rightarrow R = 41.67 \text{ k}\Omega$ 



Notice that  $i_b$  and 0.5 mA are the mesh currents. Apply KCL at the top node of the dependent source to get

$$i_b + 0.5 \times 10^{-3} = 4i_b \implies i_b = \frac{1}{6} \text{ mA}$$

Apply KVL to the supermesh corresponding to

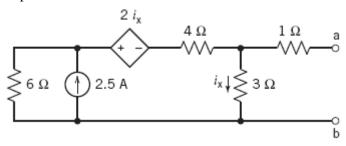
the dependent source to get

$$-5000 i_b + (10000 + R)(0.5 \times 10^{-3}) - 25 = 0$$

$$-5000 \left(\frac{1}{6} \times 10^{-3}\right) + (10000 + R)(0.5 \times 10^{-3}) = 25$$

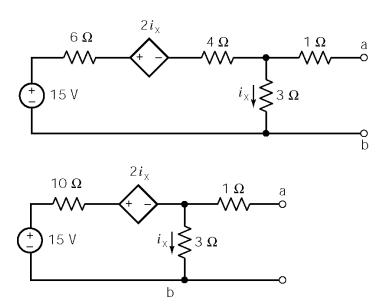
$$R = \frac{\frac{125}{6}}{0.5 \times 10^{-3}} = 41.67 \text{ k}\Omega$$

**P5.5-8** Find the Norton equivalent circuit of this circuit:

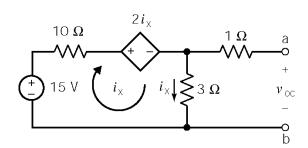


# **Solution**

Simplify the circuit using a source transformation:



Identify the open circuit voltage and short circuit current.

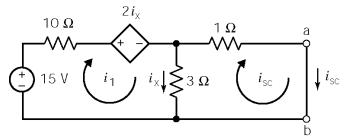


Apply KVL to the mesh to get:

$$(10+2+3)i_x-15=0 \implies i_x=1 \text{ A}$$

Then

$$v_{\rm oc} = 3 i_{\rm x} = 3 \text{ V}$$



Express the controlling current of the CCVS in terms of the mesh currents:

$$i_{x} = i_{1} - i_{sc}$$

The mesh equations are

$$10 i_1 + 2(i_1 - i_{sc}) + 3(i_1 - i_{sc}) - 15 = 0 \implies 15 i_1 - 5 i_{sc} = 15$$

and

$$i_{\rm sc} - 3(i_1 - i_{\rm sc}) = 0 \implies i_1 = \frac{4}{3}i_{\rm sc}$$

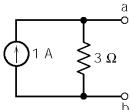
so

$$15\left(\frac{4}{3}i_{\rm sc}\right) - 5i_{\rm sc} = 15 \implies i_{\rm sc} = 1 \text{ A}$$

The Thevenin resistance is

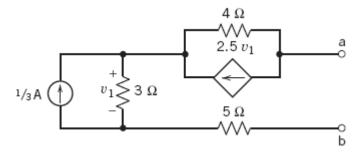
$$R_{\rm t} = \frac{3}{1} = 3 \ \Omega$$

Finally, the Norton equivalent circuit is



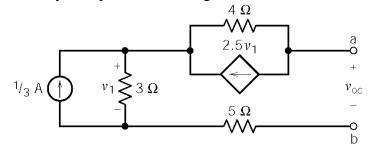
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# **P5.5-9** Find the Norton equivalent circuit of this circuit:



## **Solution**

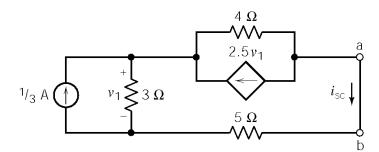
Identify the open circuit voltage and short circuit current.



$$v_1 = \left(\frac{1}{3}\right) 3 = 1 \text{ V}$$

Then

$$v_{\text{oc}} = v_1 - 4(2.5 v_1) = -9 \text{ V}$$



$$v_1 = 3\left(\frac{1}{3} - i_{\rm sc}\right) = 1 - 3 i_{\rm sc}$$

$$4(2.5 v_1 + i_{sc}) + 5i_{sc} - v_1 = 0$$

$$\Rightarrow 9 v_1 + 9 i_{sc} = 0$$

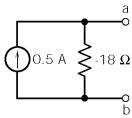
$$9(1 - 3i_{sc}) + 9i_{sc} = 0 \Rightarrow i_{sc} = \frac{1}{2} A$$

$$9(1-3i_{sc})+9i_{sc}=0 \implies i_{sc}=\frac{1}{2} A$$

The Thevenin resistance is

$$R_{\rm t} = \frac{-9}{0.5} = -18 \ \Omega$$

Finally, the Norton equivalent circuit is

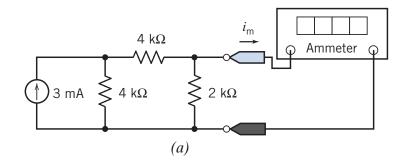


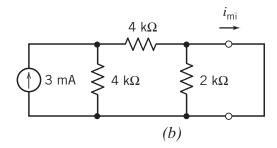
(checked: LNAP 6/21/04)

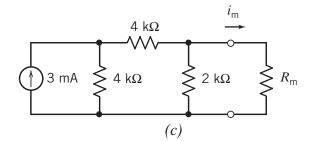
**P 5.5-10** An ideal ammeter is modeled as a short circuit. A more realistic model of an ammeter is a small resistance. Figure P 5.5-10a shows a circuit with an ammeter that measures the current  $i_{\rm m}$ . In Figure P 5.5-10*b* the ammeter is replaced by the model of an ideal ammeter, a short circuit. The ammeter measures  $i_{mi}$ , the ideal value of  $i_{\rm m}$ .

As  $R_{\rm m} \rightarrow 0$ , the ammeter becomes an ideal ammeter and  $i_{\rm m}$  $\rightarrow i_{\text{mi}}$ . When  $R_{\text{m}} > 0$ , the ammeter is not ideal and  $i_{\rm m} < i_{\rm mi}$ . The difference between  $i_{\rm m}$  and  $i_{\rm mi}$  is a measurement error caused by the fact that the ammeter is not ideal.

- (a) Determine the value of  $i_{mi}$ .
- (b) Express the measurement error that occurs when  $R_{\rm m} = 20~\Omega$  as a percentage of  $i_{\rm mi}$ .
- (c) Determine the maximum value of  $R_{\rm m}$  required to ensure that the measurement error is smaller than 2 percent of  $i_{mi}$ .



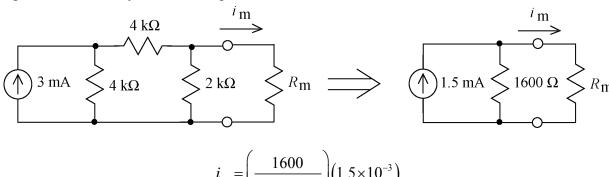




**Figure P 5.5-10** 

#### **Solution:**

Replace the circuit by its Norton equivalent circuit:



$$i_{\rm m} = \left(\frac{1600}{1600 + R_{\rm m}}\right) \left(1.5 \times 10^{-3}\right)$$

(a)

$$i_{\rm mi} = \lim_{R_{\rm m} \to 0} \quad i_{\rm m} = 1.5 \text{ mA}$$

(b) When  $R_{\rm m} = 20~\Omega$  then  $i_{\rm m} = 1.48~{\rm mA~so}$ 

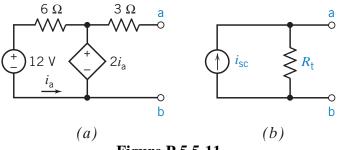
% error = 
$$\frac{1.5 - 1.48}{1.5} \times 100 = 1.23\%$$

$$0.02 \ge \frac{0.015 - \left(\frac{1600}{1600 + R_{\rm m}}\right) (0.015)}{0.015} \quad \Rightarrow \quad \frac{1600}{1600 + R_{\rm m}} \ge 0.98 \quad \Rightarrow \quad R_{\rm m} \le 32.65 \ \Omega$$

(checked: LNAP 6/18/04)

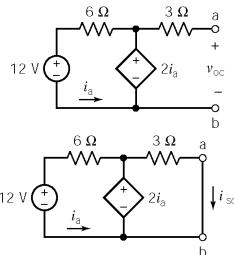
**P 5.5-11** Determine values of  $R_t$  and  $i_{sc}$ that cause the circuit shown in Figure P 5.5-11*b* to be the Norton equivalent circuit of the circuit in Figure P 5.5-11a.

**Answer:**  $R_t = 3 \Omega$  and  $i_{sc} = -2 A$ 



**Figure P 5.5-11** 

# **Solution:**



$$i_a = \frac{2i_a - 12}{6} \implies i_a = -3 \text{ A}$$

$$v_{oc} = 2i_a = -6 \text{ V}$$

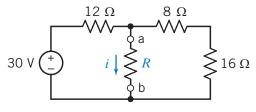
$$12 + 6i_a = 2i_a \implies i_a = -3 \text{ A}$$

$$3i_{sc} = 2i_a \implies i_{sc} = \frac{2}{3}(-3) = -2 \text{ A}$$

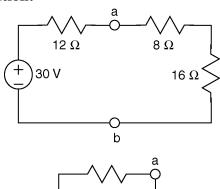
$$R_t = \frac{-6}{-2} = 3 \,\Omega$$

**P 5.5-12** Use Norton's theorem to formulate a general expression for the current *i* in terms of the variable resistance *R* shown in Figure P 5.5-12.

**Answer:** i = 20/(8 + R) A



**Figure P 5.5-12** 



$$R_t = \frac{12 \times 24}{12 + 24} = \frac{12 \times 24}{36} = 8 \Omega$$
  
 $v_{oc} = \frac{24}{12 + 24} (30) = 20 \text{ V}$ 

$$i = \frac{20}{8 + R}$$

#### **Section 5-6: Maximum Power Transfer**

**P 5.6-1** The circuit model for a photovoltaic cell is given in Figure P 5.6-1 (Edelson, 1992). The current  $i_s$  is proportional to the solar insolation (kW/m<sup>2</sup>).

- Find the load resistance,  $R_{\rm L}$ , for maximum power (a) transfer.
- Find the maximum power transferred when  $i_s = 1$  A. (b)

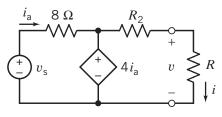
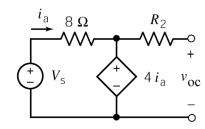


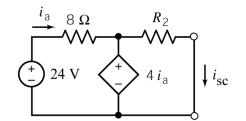
Figure P 5.6-1

#### **Solution:**

(a) The value of the current in  $R_2$  is 0 A so  $v_{oc} = 4i_a$ . Then KVL gives

$$8i_a + 4i_a - V_s = 0 \implies V_s = 12i_a = 3(4i_a) = 3(v_{oc}) = 24 \text{ V}$$





Next, KVL gives

$$8i_a + 4i_a - 24 = 0 \implies i_a = 2 \text{ A}$$

$$8i_a + 4i_a - 24 = 0 \implies i_a = 2 \text{ A}$$
 and 
$$4i_a = R_2 i_{\text{sc}} \implies 4(2) = R_2(2) \implies R_2 = 4 \Omega$$

(b) The power delivered to the resistor to the right of the terminals is maximized by setting R equal to the Thevenin resistance of the part of the circuit to the left of the terminals:

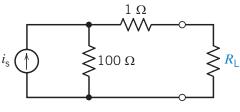
$$R = R_{\rm t} = \frac{v_{\rm oc}}{i_{\rm sc}} = \frac{8}{2} = 4 \ \Omega$$

Then

$$p_{\text{max}} = \frac{v_{\text{oc}}^2}{4R_t} = \frac{8^2}{4(4)} = 4 \text{ W}$$

**P 5.6-2** For the circuit in Figure P 5.6-2, (a) find R such that maximum power is dissipated in R and (b) calculate the value of maximum power.

**Answer:**  $R = 60 \Omega$  and  $P_{\text{max}} = 54 \text{ mW}$ 

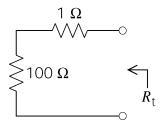


**Figure P 5.6-2** 

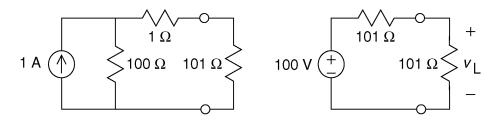
#### **Solution:**

a) For maximum power transfer, set  $R_L$  equal to the Thevenin resistance:

$$R_L = R_t = 100 + 1 = 101 \Omega$$



b) To calculate the maximum power, first replace the circuit connected to  $R_L$  be its Thevenin equivalent circuit:



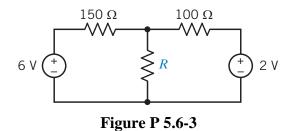
The voltage across  $R_L$  is

$$v_L = \frac{101}{101 + 101} (100) = 50 \text{ V}$$

Then

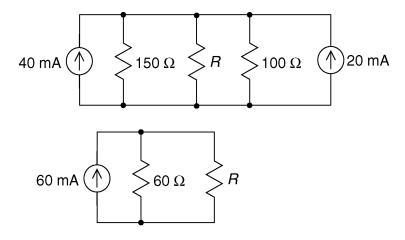
$$p_{\text{max}} = \frac{{v_L}^2}{R_L} = \frac{50^2}{101} = 24.75 \text{ W}$$

**P 5.6-3** For the circuit in Figure P 5.6-3, prove that for  $R_s$  variable and  $R_L$  fixed, the power dissipated in  $R_L$  is maximum when  $R_s = 0$ .



#### **Solution:**

Reduce the circuit using source transformations:

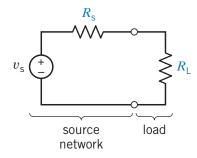


Then (a) maximum power will be dissipated in resistor R when:  $R = R_t = 60 \Omega$  and (b) the value of that maximum power is

$$P_{\text{max}} = i_R^2(R) = (0.03)^2(60) = \underline{54 \text{ mW}}$$

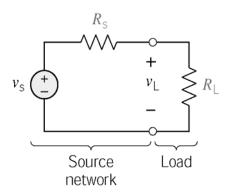
**P 5.6-4** Find the maximum power to the load  $R_L$  if the maximum power transfer condition is met for the circuit of Figure P 5.6-4.

**Answer:**  $\max p_L = 0.75 \text{ W}$ 



**Figure P 5.6-4** 

#### **Solution:**

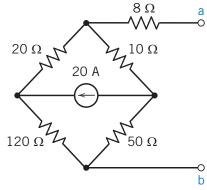


$$v_{L} = v_{S} \left[ \frac{R_{L}}{R_{S} + R_{L}} \right]$$
$$\therefore p_{L} = \frac{v_{L}^{2}}{R_{L}} = \frac{v_{S}^{2} R_{L}}{(R_{S} + R_{L})^{2}}$$

By inspection,  $p_L$  is max when you reduce  $R_S$  to get the smallest denominator.

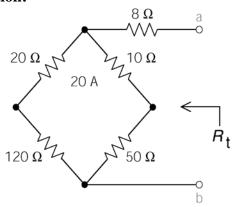
$$\therefore$$
 set  $R_S = 0$ 

**P 5.6-5** Determine the maximum power that can be absorbed by a resistor, *R*, connected to terminals a–b of the circuit shown in Figure P 5.6-5. Specify the required value of *R*.



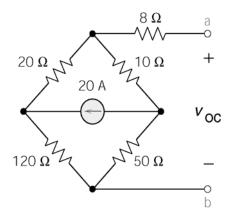
**Figure P 5.6-5** 

#### **Solution:**



The required value of R is

$$R = R_t = 8 + \frac{(20+120)(10+50)}{(20+120)+(10+50)} = 50 \Omega$$

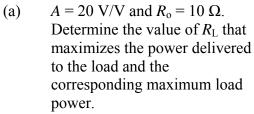


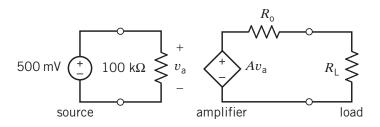
$$v_{oc} = \left[\frac{170}{170 + 30}(20)\right] 10 - \left[\frac{30}{170 + 30}(20)\right] 50$$
$$= \frac{170(20)(10) - 30(20)(50)}{200} = \frac{4000}{200} = 20 \text{ V}$$

The maximum power is given by

$$p_{\text{max}} = \frac{v_{oc}^2}{4 R_t} = \frac{20^2}{4 (50)} = 2 \text{ W}$$

**P 5.6-6** Figure P 5.6-6 shows a source connected to a load through an amplifier. The load can safely receive up to 15 W of power. Consider three cases:





**Figure P 5.6-6** 

- (b) A = 20 V/V and  $R_L = 8 \Omega$ . Determine the value of  $R_0$  that maximizes the power delivered to the load and the corresponding maximum load power.
- (c)  $R_0 = 10 \Omega$  and  $R_L = 8 \Omega$ . Determine the value of A that maximizes the power delivered to the load and the corresponding maximum load power.

#### **Solution:**

(a)  $R_t = R_o$  so  $R_L = R_o = 10 \Omega$  maximizes the power delivered to the load. The corresponding load power is

$$P_{L} = \frac{20^{2} \left(\frac{1}{2}\right)^{2} 10}{\left(10 + 10\right)^{2}} = 2.5 \text{ W}.$$

(b)  $R_0 = 0$  maximizes  $P_L$  (The numerator of  $P_L$  does not depend on  $R_0$  so  $P_L$  can be maximized by making the denominator as small as possible.) The corresponding load power is

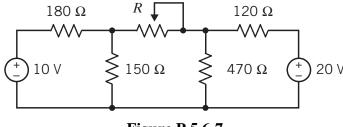
$$P_{L} = \frac{A^{2}v_{s}^{2}R_{L}}{R_{L}^{2}} = \frac{A^{2}v_{s}^{2}}{R_{L}} = \frac{20^{2}\left(\frac{1}{2}\right)^{2}}{8} = 12.5 \text{ W}.$$

(c)  $P_L$  is proportional to  $A^2$  so the load power continues to increase as A increases. The load can safely receive 15 W. This limit corresponds to

$$15 = \frac{A^2 \left(\frac{1}{2}\right)^2 8}{\left(18\right)^2} \implies A = 36\sqrt{\frac{15}{8}} = 49.3 \text{ V}.$$

(checked: LNAP 6/9/04)

**P 5.6-7** The circuit in Figure P 5.6-7 contains a variable resistance, R, implemented using a potentiometer. The resistance of the variable resistor varies over the range  $0 \le R \le 1000 \Omega$ . The variable resistor can safely receive 1/4 W power. Determine the maximum power received by the variable resistor. Is the circuit safe?



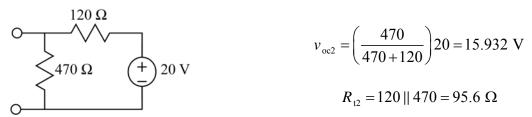
**Figure P 5.6-7** 

#### **Solution:**

Replace the part of the circuit connected to the variable resistor by its Thevenin equivalent circuit. First, replace the left part of the circuit by its Thevenin equivalent:



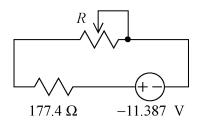
Next, replace the right part of the circuit by its Thevenin equivalent:



Now, combine the two partial Thevenin equivalents:

$$v_{\text{oc}} = v_{\text{oc1}} - v_{\text{oc2}} = -10.387 \text{ V} \text{ and } R_{\text{t}} = R_{\text{t1}} + R_{\text{t2}} = 177.4 \Omega$$

So



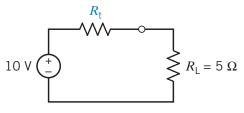
The power received by the adjustable resistor will be maximum when  $R = R_t = 177.4 \Omega$ . The maximum power received by the adjustable

resistor will be 
$$p = \frac{(-11.387)^2}{4(177.4 \Omega)} = 0.183 \text{ W}.$$

(checked LNAPDC 7/24/04)

**P 5.6-8** For the circuit of Figure P 5.6-8, find the power delivered to the load when  $R_L$  is fixed and  $R_{\rm t}$  may be varied between 1  $\Omega$  and 5  $\Omega$ . Select  $R_{\rm t}$ 

so that maximum power is delivered to  $R_{\rm L}$ . Answer: 13.9 W



**Figure P 5.6-8** 

**Solution:** 

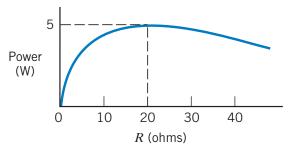
$$p = i v = \left(\frac{10}{R_t + R_L}\right) \left[\frac{R_L}{R_t + R_L}(10)\right] = \frac{100 R_L}{\left(R_t + R_L\right)^2}$$

The power increases as  $R_t$  decreases so choose  $R_t = 1 \Omega$ . Then

$$p_{\text{max}} = i v = \frac{100(5)}{(1+5)^2} = 13.9 \text{ W}$$

**P 5.6-9** A resistive circuit was connected to a variable resistor, and the power delivered to the resistor was measured as shown in Figure P 5.6-9. Determine the Thévenin equivalent circuit.

**Answer:**  $R_t = 20 \Omega$  and  $v_{oc} = 20 V$ 



**Figure P 5.6-9** 

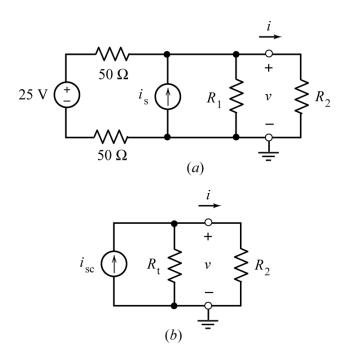
#### **Solution:**

From the plot, the maximum power is 5 W when  $R = 20 \Omega$ . Therefore:

$$R_{\rm t} = 20 \ \Omega$$

and

$$p_{\text{max}} = \frac{v_{oc}^2}{4 R_t} \implies v_{oc} = \sqrt{p_{\text{max}} 4 R_t} = \sqrt{5(4)20} = 20 \text{ V}$$



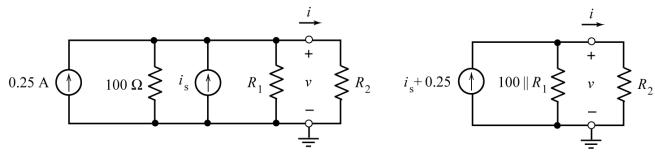
**Figure P5.6-10** 

**P5.6-10** The part circuit shown in Figure P5.6-10*a* to left of the terminals can be reduced to its Norton equivalent circuit using source transformations and equivalent resistance. The resulting Norton equivalent circuit, shown in Figure P5.6-10*b*, will be characterized by the parameters:

$$i_{\rm sc} = 1.5 \text{ A}$$
 and  $R_{\rm t} = 80 \Omega$ 

- (a) Determine the values of  $i_s$  and  $R_1$ .
- (b) Given that  $0 \le R_2 \le \infty$ , determine the maximum value of p = vi, the power delivered to  $R_2$ .

**Solution:** Two source transformations reduce the circuit as follows:



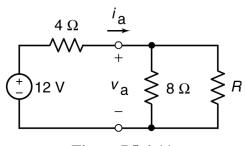
(a) Recognizing the parameters of the Norton equivalent circuit gives:

$$1.5 = i_{sc} = i_{s} + 0.25 \implies i_{s} = 1.25 \text{ A} \text{ and } 80 = R_{t} = 100 \parallel R_{1} = \frac{100 R_{1}}{100 + R_{1}} \implies R_{1} = 400 \Omega$$

(b) The maximum value of the power delivered to  $R_2$  occurs when  $R_2 = R_{\rm t} = 80~\Omega$ . Then

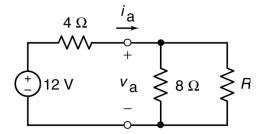
$$i = \frac{1}{2}i_{sc} = 0.75 \text{ A} \text{ and } p = \left(\frac{1}{2}i_{sc}\right)^2 R_t = \left(0.625^2\right)80 = 45 \text{ W}$$

**P5.6-11.** Given that  $0 \le R \le \infty$  in the circuit shown in Figure P5.6-12, determine (a) maximum value of  $i_a$ , (b) the maximum value of  $v_a$ , and (c) the maximum value of  $p_a = i_a v_a$ .



**Figure P5.6-11** 

#### **Solution:**



Replace the parallel combination of resistor R and the 8  $\Omega$  resistor by an equivalent resistance.

$$\begin{array}{c|c}
4 \Omega & \stackrel{i_{a}}{\longrightarrow} \\
 & \downarrow \\
 & \uparrow \\
 & \downarrow \\$$

Using voltage division

$$v_{\rm a} = \frac{R_{\rm eq}}{4 + R_{\rm eq}} (12) = \frac{1}{\frac{4}{R_{\rm eq}} + 1} (12)$$

Consequently, the maximum value of  $v_a$  corresponds to the is obtained by maximizing  $R_{eq}$ . The maximum of  $R_{eq}$  is obtained by maximizing R. Given that  $0 \le R \le \infty$ , the maximum value of  $R_{eq}$  is 8  $\Omega$  and the maximum value of  $v_a$  is

$$v_{\text{a max}} = \frac{1}{\frac{4}{8} + 1} (12) = 8 \text{ V}$$

Using Ohm's law

$$i_{\rm a} = \frac{12}{4 + R_{\rm eq}}$$

Consequently, the maximum value of  $i_a$  corresponds to the is obtained by minimizing  $R_{eq}$ . The minimum of  $R_{eq}$  is obtained by maximizing R. Given that  $0 \le R \le \infty$ , the minimum value of  $R_{eq}$  is  $0 \Omega$  and the maximum value of  $i_a$  is

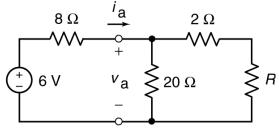
$$i_{\text{a max}} = \frac{12}{4+0} = 3 \text{ A}$$

The maximum power theorem indicates that the maximum value of  $p_a = i_a v_a$  occurs when  $R_{eq} = R_t$ . In this case,  $R_t = 4 \Omega$ . We require  $R_{eq} = 4 \Omega$  which is accomplished by making  $R = 8 \Omega$ , an acceptable value since

$$0 \le 8 \le \infty$$
. Then

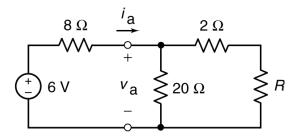
$$p_{\rm a} = \frac{\left(\frac{12}{2}\right)^2}{R_{\rm eq}} = \frac{\left(\frac{12}{2}\right)^2}{4} = 9 \text{ W}$$

**P5.6-12.** Given that  $0 \le R \le \infty$  in the circuit shown in Figure P5.6-12, determine value of R that maximizes the power  $p_a = i_a v_a$  and the corresponding maximum value of  $p_a$ .

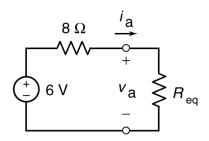


**Figure P5.6-12** 

#### **Solution:**



Replace the combination of resistor R and the 20  $\Omega$  and 2  $\Omega$  resistors by an equivalent resistance.



The maximum power theorem indicates that the maximum value of  $p_a = i_a v_a$  occurs when  $R_{eq} = R_t$ . In this case,  $R_t = 8 \Omega$ . We require

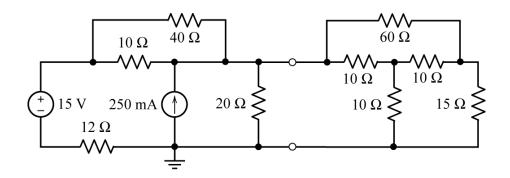
$$8 = R_{eq} = \frac{20(R+2)}{20 + (R+2)} = \frac{20R + 40}{R + 22}$$
$$8(R+22) = 20R + 40 \implies R = \frac{8(22) - 40}{20 - 8} = 11.333 \Omega$$

This isn't a standard resistance value but it is an acceptable value for this problem since  $0 \le 11.333 \le \infty$ . Then

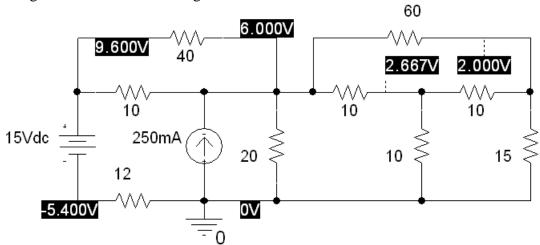
$$p_{\rm a} = \frac{\left(\frac{6}{2}\right)^2}{R_{\rm eq}} = \frac{\left(\frac{6}{2}\right)^2}{8} = 1.125 \text{ W}$$

# Section 5.8 Using PSpice to Determine the Thevenin Equivalent Circuit

P5.8-1



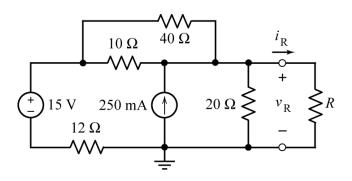
a) Here are the results of simulating the circuit in PSpice. The numbers shown in white on a black background are the node voltages.



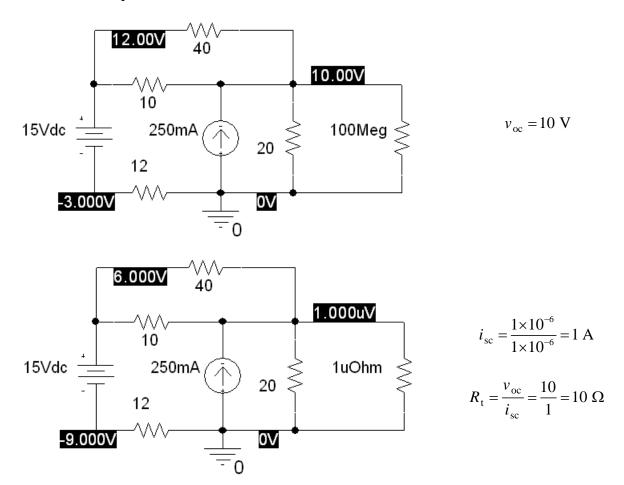
b) Add a resistor across the terminals of Circuit A. Then

$$v_{\rm oc} = v_{\rm R}$$
 when  $R \approx \infty$ 

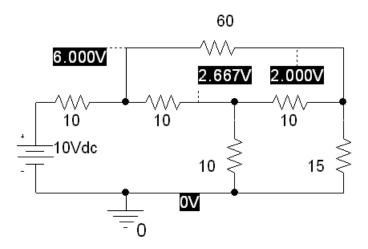
$$i_{\rm sc} = \frac{v_{\rm R}}{R}$$
 when  $R \approx 0$ 



Here are the PSpice simulation results:



c) Here is the result of simulation the circuit after replacing Circuit A by its Thevenin equivalent:



d) The node voltages of Circuit B are the same before and after replacing Circuit A by its Thevenin equivalent circuit.

#### Section 5-9 How Can We Check...?

**P 5.9-1** For the circuit of Figure P 5.9-1, the current *i* has been measured for three different values of *R* and is listed in the table. Are the data consistent?

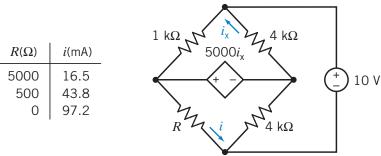
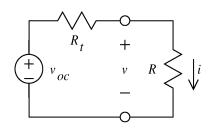


Figure P 5.9-1

#### **Solution:**



Use the data in the first two lines of the table to determine  $v_{oc}$  and  $R_t$ :

$$0.0972 = \frac{v_{oc}}{R_t + 0}$$

$$0.0438 = \frac{v_{oc}}{R_t + 500}$$

$$\Rightarrow \begin{cases} v_{oc} = 39.9 \text{ V} \\ R_t = 410 \Omega \end{cases}$$

Now check the third line of the table. When  $R=5000 \Omega$ :

$$i = \frac{v_{oc}}{R_t + R} = \frac{39.9}{410 + 5000} = 7.37 \text{ mA}$$

which disagree with the data in the table.

#### The data is not consistent.

**P 5.9-2** Your lab partner built the circuit shown in Figure P 5.9-2 and measured the current i and voltage v corresponding to several values of the resistance R. The results are shown in the table in Figure P 5.9-2. Your lab partner says that  $R_L = 8000 \Omega$  is required to cause i = 1 mA. Do you agree? Justify your answer.

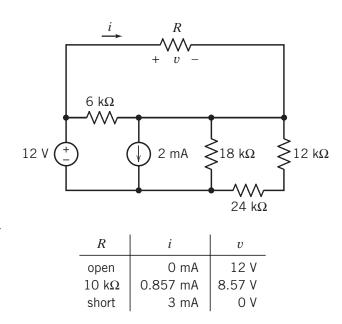
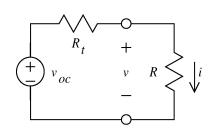


Figure P 5.9-2

#### **Solution:**



Use the data in the table to determine  $v_{oc}$  and  $i_{sc}$ :

$$v_{oc} = 12 \text{ V}$$
 (line 1 of the table)

$$i_{sc} = 3 \text{ mA}$$
 (line 3 of the table)

so 
$$R_t = \frac{v_{oc}}{i_{sc}} = 4 \text{ k}\Omega$$

Next, check line 2 of the table. When  $R = 10 \text{ k}\Omega$ :

$$i = \frac{v_{oc}}{R_t + R} = \frac{12}{10(10^3) + 5(10^3)} = 0.857 \text{ mA}$$

which agrees with the data in the table.

To cause 
$$i = 1$$
 mA requires  $0.001 = i = \frac{v_{oc}}{R_t + R} = \frac{12}{10(10^3) + R} \implies R = 8000 \Omega$ 

I agree with my lab partner's claim that R = 8000 causes i = 1 mA.

**P 5.9-3** In preparation for lab, your lab partner determined the Thévenin equivalent of the circuit connected to  $R_{\rm L}$  in Figure P 5.9-3. She says that the Thévenin resistance is  $R_{\rm t} = \frac{6}{11}R$  and the open-circuit voltage is  $v_{\rm oc} = \frac{60}{11}$  V. In lab, you built the circuit using  $R = 110~\Omega$  and  $R_{\rm L} = 40~\Omega$  and measured that i = 54.5 mA. Is this measurement consistent with the prelab calculations? Justify your answers.

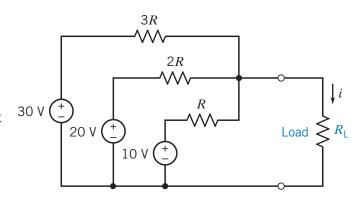
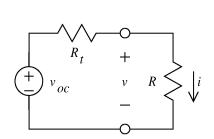


Figure P 5.9-3

#### **Solution:**



$$\frac{1}{R_t} = \frac{1}{R} + \frac{1}{2R} + \frac{1}{3R} = \frac{11}{6R} \implies R_t = \frac{6R}{11}$$

$$v_{oc} = \left(\frac{2/3}{3+2/3}\right)30 + \left(\frac{3/4}{2+3/4}\right)20 + \left(\frac{6/5}{1+6/5}\right)10 = \frac{180}{11}$$

so the prelab calculation isn't correct.

But then

$$i = \frac{v_{oc}}{R_t + R} = \frac{\frac{180}{11}}{\frac{6}{11}(110) + 40} = \frac{\frac{180}{11}}{60 + 40} = 163 \text{ mA} \neq 54.5 \text{ mA}$$

so the measurement does not agree with the corrected prelab calculation.

**P 5.9-4** Your lab partner claims that the current i in Figure P 5.9-4 will be no greater than 12.0 mA, regardless of the value of the resistance R. Do you agree?

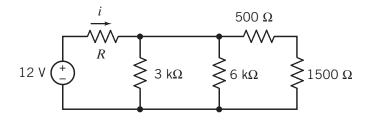


Figure P 5.9-4

**Solution:** 

6000 || 3000 || 
$$(500 + 1500) = 2000$$
 || 2000 = 1000  $\Omega$   
$$i = \frac{12}{R + 1000} \le \frac{12}{1000} = 12 \text{ mA}$$

How about that?! Your lab partner is right.

(checked using LNAP 6/21/05)

- **P 5.9-5** Figure P 5.9-5 shows a circuit and some corresponding data. Two resistances,  $R_1$  and R, and the current source current are unspecified. The tabulated data provide values of the current, i, and voltage, v, corresponding to several values of the resistance R.
- (a) Consider replacing the part of the circuit connected to the resistor R by a Thévenin equivalent circuit. Use the data in rows 2 and 3 of the table to find the values of  $R_t$  and  $v_{\rm oc}$ , the Thévenin resistance and the open-circuit voltage.
- (b) Use the results of part (a) to verify that the tabulated data are consistent.
- (c) Fill in the blanks in the table.
- Determine the values of  $R_1$  and  $i_s$ . (d)

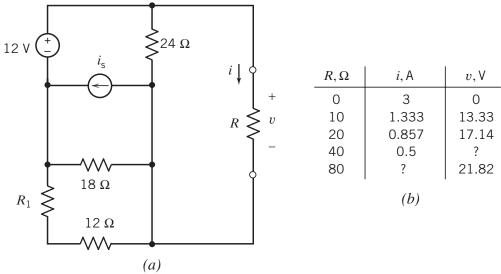
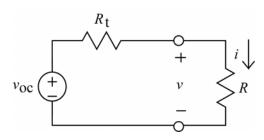


Figure P5.9-5

#### **Solution:**

(a)



KVL gives 
$$v_{oc} = (R_{t} + R)i$$
 from row 2 
$$v_{oc} = (R_{t} + 10)(1.333)$$
 from row 3

$$v_{\rm oc} = (R_{\rm t} + 20)(0.857)$$

So

$$(R_t + 10)(1.333) = (R_t + 20)(0.857)$$

$$28(R_t+10)=18(R_t+20)$$

Solving gives

$$10R_{\rm t} = 360 - 280 = 80 \quad \Rightarrow \quad R_{\rm t} = 8 \Omega$$

and

$$v_{\rm oc} = (8+10)(1.333) = 24 \text{ V}$$

$$i = \frac{v_{\text{oc}}}{R_1 + R} = \frac{24}{8 + R}$$
 and  $v = \frac{R}{R + R_1} v_{\text{oc}} = \frac{24R}{R + 8}$ 

When R = 0, i = 3 A, and v = 0 V.

When 
$$R = 40 \Omega$$
,  $i = \frac{1}{2} A$ .

When 
$$R = 80 \Omega$$
,  $v = \frac{24(80)}{88} = \frac{240}{11} = 21.82$ .

These are the values given in the tabulated data so the data is consistent.

(c) When 
$$R = 40 \Omega$$
,  $v = \frac{24(40)}{48} = 20 \text{ V}$ .

When 
$$R = 80 \Omega$$
,  $i = \frac{24}{88} = 0.2727 \text{ A}$ .

(d) First

$$8 = R_{t} = 24 || 18 || (R_{1} + 12)$$
  $\Rightarrow$   $R_{1} = 24 \Omega$ 

the, using superposition,

$$24 = v_{oc} = \frac{24}{24 + (18 || (R_1 + 12))} 12 + (24 || 18 (R_1 + 12)) i_s = 8 + 8i_s \implies i_s = 2 \text{ A}$$

(checked using LNAP 6/21/05)

### **Design Problems**

**DP 5-1** The circuit shown in Figure DP 5-1a has four unspecified circuit parameters:  $v_s$ ,  $R_1$ ,  $R_2$ , and  $R_3$ . To design this circuit, we must specify the values of these four parameters. The graph shown in Figure DP 5-1b describes a relationship between the current i and the voltage v.

Specify values of  $v_s$ ,  $R_1$ ,  $R_2$ , and  $R_3$  that cause the current i and the voltage v in Figure DP 5-1a to satisfy the relationship described by the graph in Figure DP 5-1b.

*First Hint:* The equation representing the straight line in Figure DP 5-1*b* is

$$v = -R_{\rm t}i + v_{\rm oc}$$

That is, the slope of the line is equal to -1 times the Thévenin resistance and the "v-intercept" is equal to the open-circuit voltage.

**Second Hint:** There is more than one correct answer to this problem. Try setting  $R_1 = R_2$ .

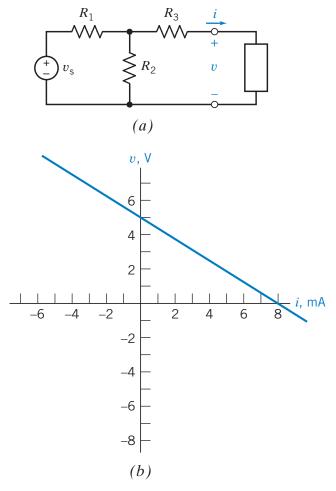


Figure DP 5-1

#### **Solution:**

The equation of representing the straight line in Figure DP 5-1b is  $v = -R_t i + v_{oc}$ . That is, the slope of the line is equal to -1 times the Thevenin resistance and the "v - intercept" is equal to the open circuit voltage. Therefore:  $R_t = -\frac{0-5}{0.008-0} = 625 \Omega$  and  $v_{oc} = 5 \text{ V}$ .

Try  $R_1 = R_2 = 1 \text{ k}\Omega$ .  $(R_1 \parallel R_2 \text{ must be smaller than } R_t = 625 \Omega$ .) Then

$$5 = \frac{R_2}{R_1 + R_2} v_s = \frac{1}{2} v_s \implies v_s = 10 \text{ V}$$

and

$$625 = R_3 + \frac{R_1 R_2}{R_1 + R_2} = R_3 + 500 \implies R_3 = 125 \Omega$$

Now  $v_s$ ,  $R_1$ ,  $R_2$  and  $R_3$  have all been specified so the design is complete.

**DP 5-2** The circuit shown in Figure DP 5.2a has four unspecified circuit parameters:  $i_s$ ,  $R_1$ ,  $R_2$ , and  $R_3$ . To design this circuit, we must specify the values of these four parameters. The graph shown in Figure DP 5-2b describes a relationship between the current i and the voltage v.

Specify values of  $i_s$ ,  $R_1$ ,  $R_2$ , and  $R_3$  that cause the current i and the voltage v in Figure DP 5-2a to satisfy the relationship described by the graph in Figure DP 5-2b.

*First Hint:* Calculate the open-circuit voltage,  $v_{oc}$ , and the Thévenin resistance,  $R_t$ , of the part of the circuit to the left of the terminals in Figure DP 5-2a.

**Second Hint:** The equation representing the straight line in Figure DP 5-2b is

$$v = -R_t i + v_{oc}$$

That is, the slope of the line is equal to -1 times the Thévenin resistance and the "v-intercept" is equal to the open-circuit voltage.

**Third Hint:** There is more than one correct answer to this problem. Try setting both  $R_3$  and  $R_1 + R_2$  equal to twice the slope of the graph in Figure DP 5-2b.

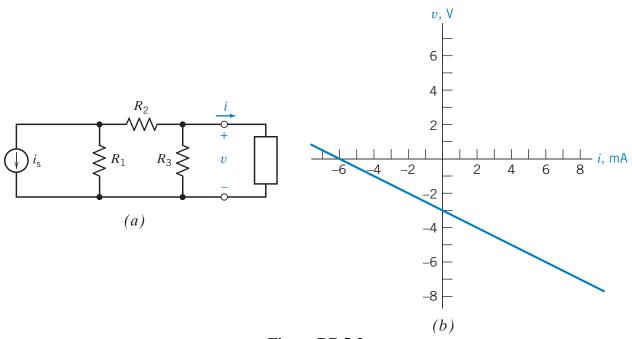


Figure DP 5-2

#### **Solution:**

The equation of representing the straight line in Figure DP 5-2b is  $v = -R_t i + v_{oc}$ . That is, the slope of the line is equal to -1 times the Thevenin resistance and the "v - intercept" is equal to the open circuit voltage. Therefore:  $R_t = -\frac{0 - \left(-3\right)}{-0.006 - 0} = 500 \,\Omega$  and  $v_{oc} = -3 \,\mathrm{V}$ .

From the circuit we calculate

$$R_t = \frac{R_3(R_1 + R_2)}{R_1 + R_2 + R_3}$$
 and  $v_{oc} = -\frac{R_1R_3}{R_1 + R_2 + R_3}i_s$ 

500 
$$\Omega = \frac{R_3 (R_1 + R_2)}{R_1 + R_2 + R_3}$$
 and  $-3 \text{ V} = -\frac{R_1 R_3}{R_1 + R_2 + R_3} i_s$ 

Try  $R_3 = 1 \text{k}\Omega$  and  $R_1 + R_2 = 1 \text{k}\Omega$ . Then  $R_t = 500 \Omega$  and

$$-3 = -\frac{1000R_1}{2000}i_s = -\frac{R_1}{2}i_s \implies 6 = R_1i_s$$

This equation can be satisfied by taking  $R_1 = 600 \Omega$  and  $i_s = 10$  mA. Finally,  $R_2 = 1 \text{ k}\Omega$  -  $400 \Omega = 400 \Omega$ . Now  $i_s$ ,  $R_1$ ,  $R_2$  and  $R_3$  have all been specified so the design is complete.

#### **DP 5-3** The circuit shown in

Figure DP 5-3a has four unspecified circuit parameters:  $v_s$ ,  $R_1$ ,  $R_2$ , and  $R_3$ . To design this circuit, we must specify the values of these four parameters. The graph shown in Figure DP 5-3b describes a relationship between the current i and the voltage v.

Is it possible to specify values of  $v_s$ ,  $R_1$ ,  $R_2$ , and  $R_3$  that cause the current i and the voltage v in Figure DP 5-1a to satisfy the relationship described by the graph in Figure DP 5-3b? Justify your answer.

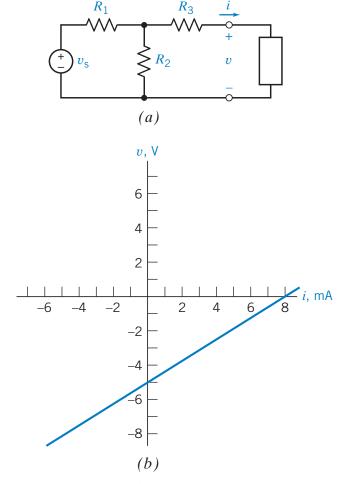


Figure DP 5-3

#### **Solution:**

The slope of the graph is positive so the Thevenin resistance is negative. This would require  $R_3 + \frac{R_1 R_2}{R_1 + R_2} < 0$ , which is not possible since  $R_1$ ,  $R_2$  and  $R_3$  will all be non-negative.

Is it not possible to specify values of  $v_s$ ,  $R_1$ ,  $R_2$  and  $R_3$  that cause the current i and the voltage v in Figure DP 5-3a to satisfy the relationship described by the graph in Figure DP 5-3b.

#### **DP 5-4** The circuit shown in

Figure DP 5-4a has four unspecified circuit parameters:  $v_s$ ,  $R_1$ ,  $R_2$ , and d, where d is the gain of the CCCS. To design this circuit, we must specify the values of these four parameters. The graph shown in Figure DP 5-4b describes a relationship between the current i and the voltage v.

Specify values of  $v_s$ ,  $R_1$ ,  $R_2$ , and d that cause the current i and the voltage v in Figure DP 5-4a to satisfy the relationship described by the graph in Figure DP 5-4b.

*First Hint:* The equation representing the straight line in Figure DP 5-4b is

$$v = -R_t i + v_{oc}$$

That is, the slope of the line is equal to -1 times the Thévenin resistance and the "v-intercept" is equal to the open-circuit voltage.

**Second Hint:** There is more than one correct answer to this problem. Try setting  $R_1 = R_2$ .

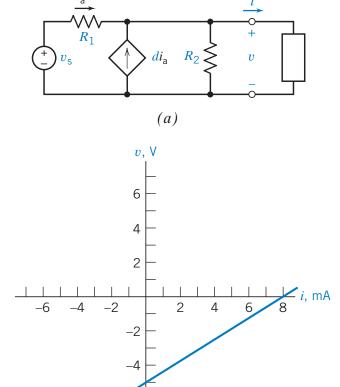


Figure DP 5-4

-6

-8

(b)

#### **Solution:**

The equation of representing the straight line in Figure DP 5-4b is  $v = -R_t i + v_{oc}$ . That is, the slope of the line is equal to the Thevenin impedance and the "v - intercept" is equal to the open circuit voltage. Therefore:  $R_t = -\frac{-5-0}{0-0.008} = -625 \Omega$  and  $v_{oc} = -5 \text{ V}$ .

The open circuit voltage,  $v_{oc}$ , the short circuit current,  $i_{sc}$ , and the Thevenin resistance,  $R_t$ , of this circuit are given by

$$v_{oc} = \frac{R_2(d+1)}{R_1 + (d+1)R_2} v_s$$

$$i_{sc} = \frac{(d+1)}{R_1} v_s$$

and

$$R_{t} = \frac{R_{1}R_{2}}{R_{1} + (d+1)R_{2}}$$

Let  $R_1 = R_2 = 1 \text{ k}\Omega$ . Then

$$-625 \Omega = R_t = \frac{1000}{d+2} \implies d = \frac{1000}{-625} - 2 = -3.6 \text{ A/A}$$

 $\quad \text{and} \quad$ 

$$-5 = \frac{(d+1)v_s}{d+2} \implies v_s = \frac{-3.6+2}{-3.6+1}(-5) = -3.077 \text{ V}$$

Now  $v_s$ ,  $R_1$ ,  $R_2$  and d have all been specified so the design is complete.

# **Chapter 5 Circuit Theorems**

#### **Exercises**

**Exercise 5.2-1** Determine values of R and  $i_s$  so that the circuits shown in Figures E 5.2-1a,b are equivalent to each other due to a source transformation.

**Answer:**  $R = 10 \Omega$  and  $i_s = 1.2 A$ 

**Exercise 5.2-2** Determine values of R and  $i_s$  so that the circuits shown in Figures E 5.2-2a,b are equivalent to each other due to a source transformation.

*Hint:* Notice that the polarity of the voltage source in Figure E 5.2-2a is not the same as in Figure E 5.2-1a.

**Answer:**  $R = 10 \Omega$  and  $i_s = -1.2 A$ 

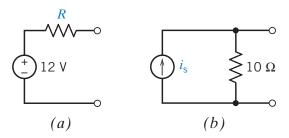
**Exercise 5.2-3** Determine values of R and  $v_s$  so that the circuits shown in Figures E 5.2-3a,b are equivalent to each other due to a source transformation.

**Answer:**  $R = 8 \Omega$  and  $v_s = 24 \text{ V}$ 

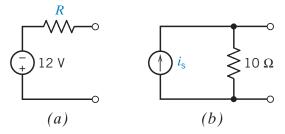
**Exercise 5.2-4** Determine values of R and  $v_s$  so that the circuits shown in Figures E 5.2-4a,b are equivalent to each other due to a source transformation.

*Hint:* Notice that the reference direction of the current source in Figure E 5.2-4*b* is not the same as in Figure E 5.2-3*b*.

**Answer:**  $R = 8 \Omega$  and  $v_s = -24 \text{ V}$ 



Figures E 5.2-1



Figures E 5.2-2

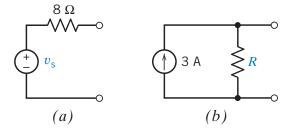


Figure E 5.2-3

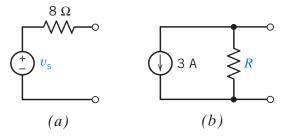


Figure E 5.2-4

**Exercise 5.4-1** Determine values of  $R_t$  and  $v_{oc}$  that cause the circuit shown in Figure E 5.4-1b to be the Thévenin equivalent circuit of the circuit in Figure E 5.4-1a.

**Answer:**  $R_t = 8 \Omega$  and  $v_{oc} = 2 V$ 

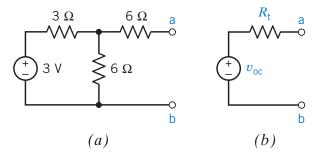
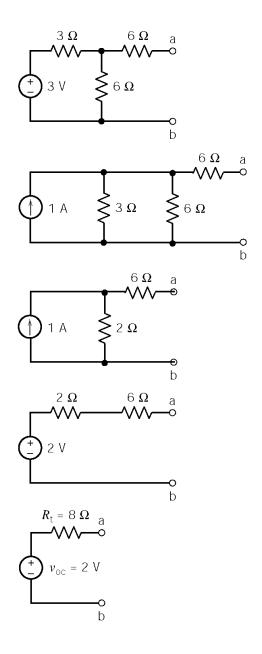
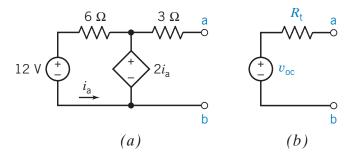


Figure E 5.2-1

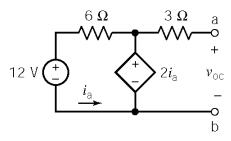


**Exercise 5.4-2** Determine values of  $R_t$  and  $v_{oc}$  that cause the circuit shown in Figure E 5.4-2b to be the Thévenin equivalent circuit of the circuit in Figure E 5.4-2a.

**Answer:**  $R_t = 3 \Omega$  and  $v_{oc} = -6 \text{ V}$ 



**Figure E 5.2-2** 



$$i_a = \frac{2i_a - 12}{6} \implies i_a = -3 \text{ A}$$
  
 $v_{oc} = 2i_a = -6 \text{ V}$ 

12 
$$V \stackrel{+}{\overset{}{\stackrel{}}} 2i_a$$

$$12 + 6i_a = 2i_a \implies i_a = -3 \text{ A}$$

$$3i_{sc} = 2i_a \implies i_{sc} = \frac{2}{3}(-3) = -2 \text{ A}$$

$$R_t = \frac{-6}{-2} = 3 \,\Omega$$

Exercise 5.5-1 Determine values of  $R_t$  and  $i_{sc}$  that cause the circuit shown in Figure E 5.5-1b to be the Norton equivalent circuit of the circuit in Figure E 5.5-1a.

**Answer:**  $R_t = 8 \Omega$  and  $i_{sc} = 0.25 A$ 

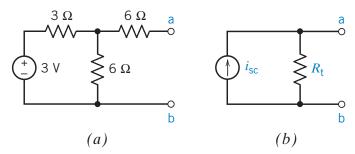
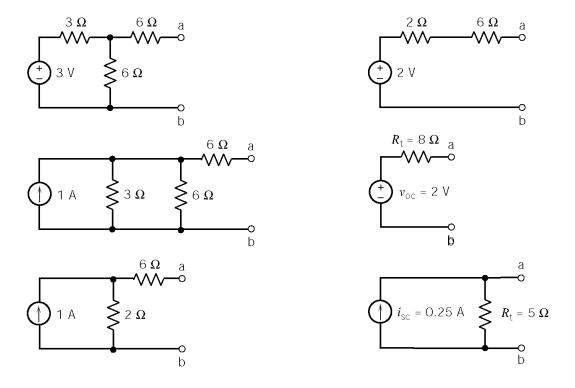
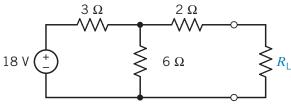


Figure E 5.5-1



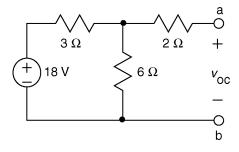
*Exercise 5.6-1* Find the maximum power that can be delivered to  $R_L$  for the circuit of Figure E 5.6-1 using a Thévenin equivalent circuit.

**Answer:** 9 W when  $R_L = 4 \Omega$ 

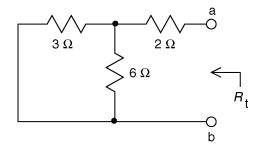


**Figure E 5.6-1** 

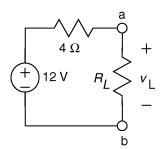
#### **Solution:**



$$v_{oc} = \frac{6}{6+3} (18) = 12 \text{ V}$$



$$R_t = 2 + \frac{(3)(6)}{3+6} = 4 \Omega$$



For maximum power, we require

$$R_L = R_t = 4 \Omega$$

Then

$$p_{\text{max}} = \frac{v_{oc}^2}{4 R_t} = \frac{12^2}{4(4)} = 9 \text{ W}$$

# **PSpice Problems**

**SP 5-1** The circuit in Figure SP 5.1 has three inputs:  $v_1$ ,  $v_2$ , and  $i_3$ . The circuit has one output,  $v_0$ . The equation

$$v_0 = av_1 + bv_2 + ci_3$$

expresses the output as a function of the inputs. The coefficients a, b, and c are real constants.

- (a) Use PSpice, and the principle of superposition, to determine the values of a, b, and c.
- (b) Suppose  $v_1 = 10 \text{ V}$ ,  $v_2 = 8 \text{ V}$ , and we want the output to be  $v_0 = 7 \text{ V}$ . What is the required value of  $i_3$ ?

*Hint:* The output is given by  $v_0 = a$  when  $v_1 = 1$  V,  $v_2 = 0$  V, and  $i_3 = 0$  A.

**Answer:** (a)  $v_0 = 0.3333v_1 + 0.3333v_2 + 33.33i_3$ , (b)  $i_3 = 30$  mA

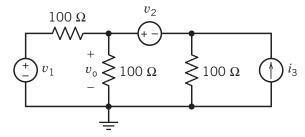
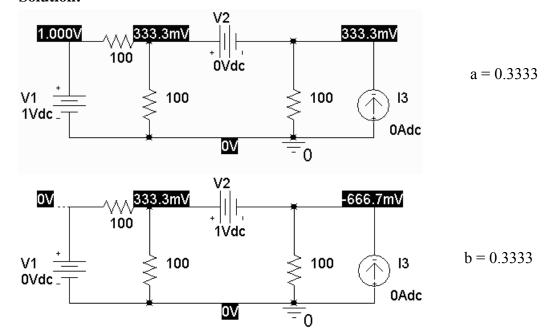


Figure SP 5.1



(a) 
$$v_o = 0.3333 v_1 + 0.3333 v_2 + 33.33 i_3$$

(b) 
$$7 = 0.3333(10) + 0.3333(8) + 33.33 i_3 \implies i_3 = \frac{7 - \frac{18}{3}}{\frac{100}{3}} = \frac{3}{100} = 30 \text{ mA}$$

**SP 5-2** The pair of terminals a—b partitions the circuit in Figure SP 5.2 into two parts. Denote the node voltages at nodes 1 and 2 as  $v_1$  and  $v_2$ . Use PSpice to demonstrate that performing a source transformation on the part of the circuit to the left of the terminal does not change anything to the right of the terminals. In particular, show that the current,  $i_0$ , and the node voltages,  $v_1$  and  $v_2$ , have the same values after the source transformation as before the source transformation.

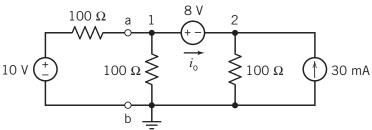
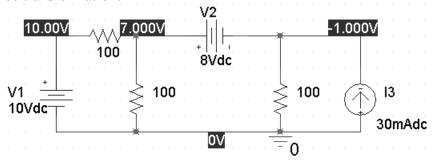


Figure SP 5.2

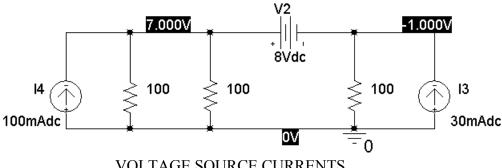
#### **Solution:**

Before the source transformation:



# VOLTAGE SOURCE CURRENTS NAME CURRENT

After the source transformation:

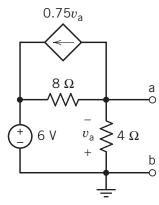


VOLTAGE SOURCE CURRENTS
NAME CURRENT

V\_V2 -4.000E-02

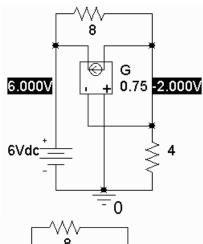
**SP 5-3** Use PSpice to find the Thévenin equivalent circuit for the circuit shown in Figure SP 5.3.

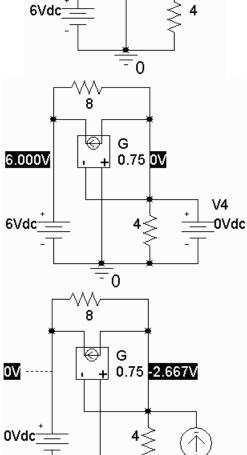
**Answer:**  $v_{\text{oc}} = -2 \text{ V} \text{ and } R_{\text{t}} = -8/3 \Omega$ 



# Figure SP 5.3

## **Solution:**





$$v_{\rm oc} = -2 \text{ V}$$

## VOLTAGE SOURCE CURRENTS NAME CURRENT

V V3	-7.500E-01
$\overline{V}V4$	7 500E-01

$$i_{\rm sc} = 0.75 \text{ A}$$

$$R_{\rm t}$$
 =  $-2.66~\Omega$ 

1Adc

**SP 5-4** The circuit shown in Figure SP 5-4b is the Norton equivalent circuit of the circuit shown in Figure SP 5-4a. Find the value of the short-circuit current,  $i_{sc}$ , and Thévenin resistance,  $R_t$ .

**Answer:**  $i_{sc} = 1.13 \text{ V} \text{ and } R_t = 7.57 \Omega$ 

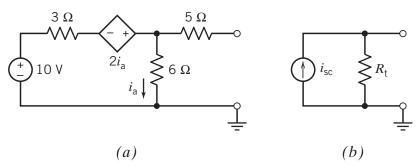
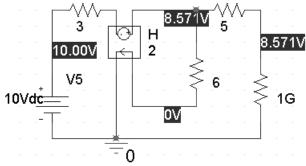
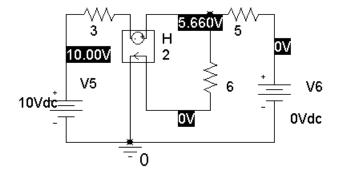


Figure SP 5-4

#### **Solution:**

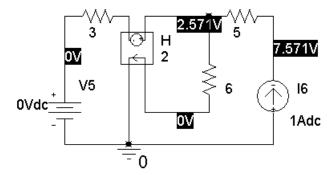


$$v_{\rm oc} = 8.571 \text{ V}$$



#### **VOLTAGE SOURCE CURRENTS**

NAME	CURRENT
V_V5 V_V6	-2.075E+00 1.132E+00
_	H_H1 9.434E-01
	$i_{sc} = 1.132 \text{ A}$



$$R_{\rm t} = 7.571 \ \Omega$$