

# EEE575

## ASSIGNMENT 4

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### Q1

The paper by Marcel Carvalho and Minhoto Teixeira introduces a comprehensive method to simplify the traditional Ogata lead-lag compensatory design by developing direct analytical expressions. This approach addresses a fundamental limitation of the classical root locus technique, which typically relies on an iterative trial-and-error process to determine suitable compensatory parameters

The key contribution is the derivation of closed-form mathematical expressions that allow for the direct calculation of the compensatory parameters based on desired closed-loop specifications, such as damping ratio or settling time. Unlike the iterative graphical method, these expressions enable designers to compute the parameters algebraically, eliminating the need for repetitive pole-zero placement and root locus sketching.

Building on Ogata's foundational work, this method bypasses the graphical construction steps. The authors demonstrate that by correctly formulating design constraints such as mathematical relationships, the parameters can be solved directly.

A significant advantage of this direct approach is its suitability for computer-aided design environments. The expressions can be easily implemented in software tools, enabling rapid design iteration and optimization. This is particularly valuable in education, allowing students to focus on core control concepts rather than tedious graphical procedures.

The method is validated through design examples, showing comparable performance to the traditional approach but with a substantial reduction in design time. This work effectively bridges classical control theory and modern computational tools, making advanced compensation design more accessible while preserving the physical intuition of the root locus method.

Q2

$$G(s) = \frac{5}{s(s+2)}$$

$$H(s) = 1 + \alpha s$$

$$\begin{aligned} T(s) &= \frac{G(s)}{1 + G(s)H(s)} \\ &= \frac{\frac{5}{s(s+2)}}{1 + \frac{5(1+\alpha s)}{s(s+2)}} \\ &= \frac{5}{s(s+2) + 5(1+\alpha s)} \\ &= \frac{5}{s^2 + (2+5\alpha)s + 5} \\ T(s) &= \frac{5}{s^2 + (2+5\alpha)s + 5} \end{aligned}$$

Characteristic equation

$$s^2 + (2 + 5\alpha) s + 5$$

$$1 + KL(s) = 0$$

$$(s^2 + 2s + 5) + \alpha(5s) = 0$$

$$1 + \alpha \frac{5s}{s^2 + 2s + 5} = 0$$

$$L(s) = \frac{b(s)}{a(s)} = \frac{5s}{s^2 + 2s + 5}$$

$$\begin{aligned} a(s) &= s^2 + 2s + 5 \\ b(s) &= 5s \end{aligned}$$

$$a(s) = s^2 + 2s + 5 = 0$$

$$s = -1 \pm 2i$$

$$b(s) = 5s = 0$$

$$s = 0$$

$$\begin{aligned} L(s) &= \frac{5s}{s^2 + 2s + 5} \\ a(s) &= s^2 + 2s + 5 \\ b(s) &= 5s \\ \text{pole: } s &= -1 \pm 2j \\ \text{zero: } s &= 0 \end{aligned}$$

n = 2 (number of poles)

m = 1 (number of zeros)

Root locus plotting

Starting point (K=0): Two branches start at the pole  $-1 \pm 2j$

End point (K=∞): One branch ends at zero 0, and one branch ends at ∞

The root locus on the real axis is in a region with an odd number of poles and zeros to its right:

The zero is at  $s = 0$ , and the pole is at  $s = -1 \pm 2j$  (not on the real axis).

Interval  $(-\infty, 0)$ : 1 zero to the right (odd number)  $\rightarrow$  root locus exists.

Interval  $(0, +\infty)$ : 0 poles and zeros to the right (even number)  $\rightarrow$  no root locus exists.

For  $n-m = 2-1 = 1$ , 1 asymptote

$$\theta_k = \frac{180^\circ + 360^\circ k}{n-m} = \frac{180^\circ + 360^\circ k}{1}$$

$$k = 0, \theta_0 = 180^\circ$$

$$\sigma_a = \frac{\sum pole - \sum zero}{n-m} = \frac{(-1+2j) + (-1-2j) - 0}{2-1} = \frac{-2}{1} = -2$$

Asymptote: A straight line passing through the point  $(-2, 0)$  with an angle of  $180^\circ$  (i.e., the negative direction of the real axis)

$$z_1 - s_2 = 0 - (-1 - 2j) = 1 + 2j$$

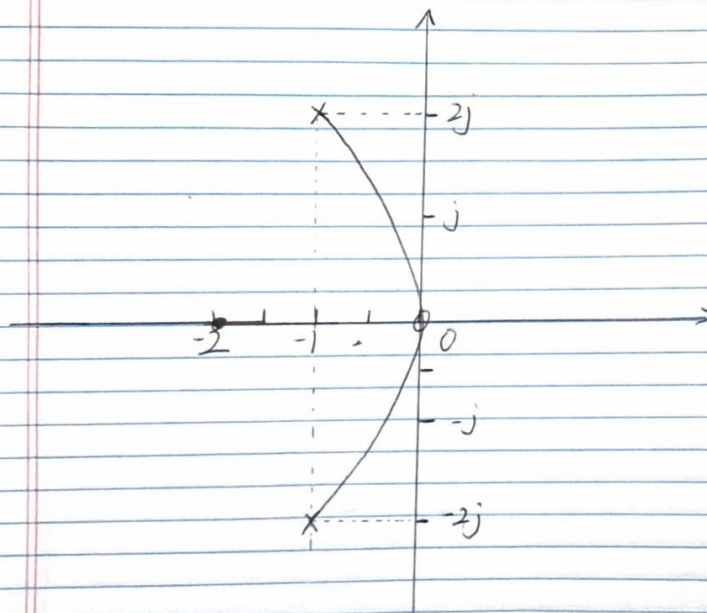
$$\angle(z_1 - s_2) = \angle(1 + 2j) = \arctan\left(\frac{2}{1}\right) = 63.43^\circ$$

$$s^1 - s^2 = (-1 + 2j) - (-1 - 2j) = 4j$$

$$\angle(s_1 - s_2) = \angle(4j) = 90^\circ$$

$$\theta_{departure} = 63.43^\circ - 90^\circ + 180^\circ$$

$$\theta_{departure} = 153.43^\circ$$



Asymptote: A straight line passing through the R point  $(-2, 0)$  with an angle of  $180^\circ$

```
>> ASS4Q2m
L
Poles:  -1.0000 - 2.0000i  -1.0000 + 2.0000i
Zero:    0
```

#### ROOT LOCUS PARAMETERS

```
Number of poles n = 2
Number of Zero m = 1
Number of asymptotes n-m = 1
```

#### ASYMPTOTE CALCULATION

```
Asymptote center:  $\sigma = -2.00$ 
Asymptote angles:  $180.0^\circ$ 
```

### DEPARTURE ANGLE CALCULATION

Departure angle from pole  $s_1 = -1+2j$ :  $26.57^\circ$

Departure angle from pole  $s_2 = -1-2j$ :  $-26.57^\circ$

### Zero REAL AXIS ROOT LOCUS ANALYSIS Zero

According to Rule 2: Real axis root locus exists to the left of odd number of poles and Zero

Zero location:  $0$

Pole locations:  $-1\pm 2j$  (not on real axis)

### Zero ROOT LOCUS VERIFICATION Zero

Test point  $s = -0.5$ :  $\alpha = 1.700$

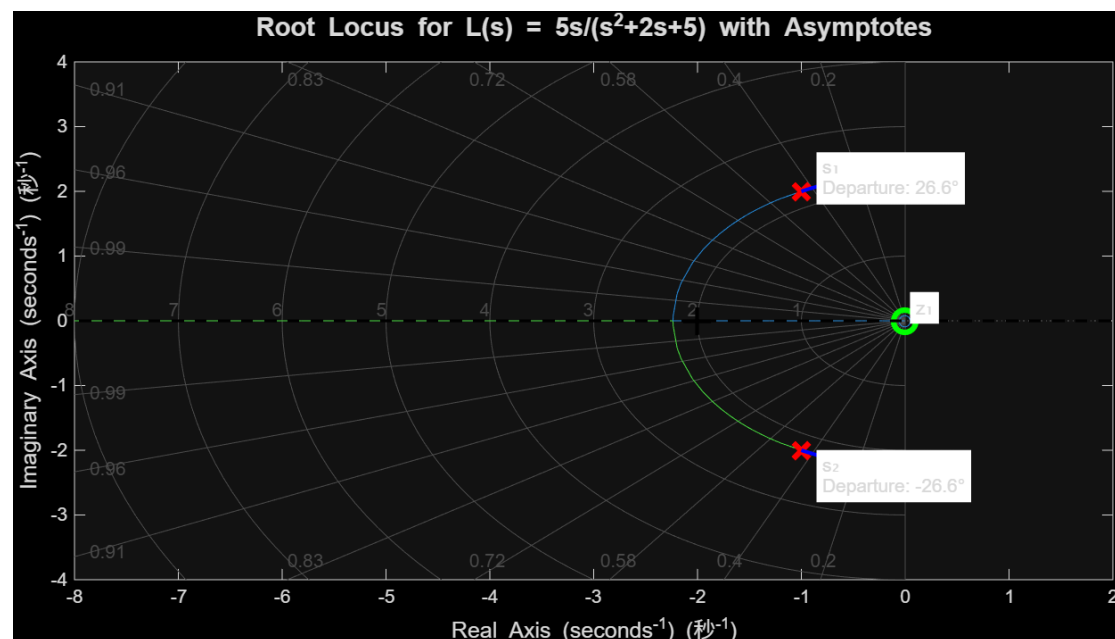
→ This point is on the root locus ( $\alpha \geq 0$ )

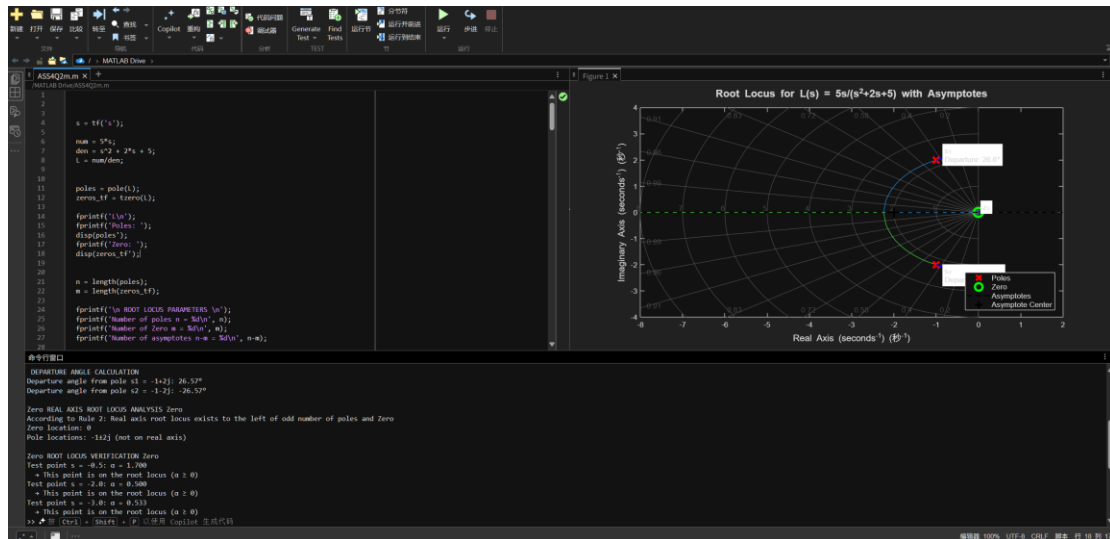
Test point  $s = -2.0$ :  $\alpha = 0.500$

→ This point is on the root locus ( $\alpha \geq 0$ )

Test point  $s = -3.0$ :  $\alpha = 0.533$

→ This point is on the root locus ( $\alpha \geq 0$ )





Q3

$$G(s) = \frac{1}{(s+2)(s+3)}$$

$$D_c(s) = K \frac{s+a}{s+b}$$

$$L(s) = D_c(s)G(s) = K \frac{(s+a)}{(s+b)(s+2)(s+3)}$$

$$1 + L(s) = 0$$

$$1 + K \frac{s+a}{(s+b)(s+2)(s+3)} = 0$$

$$(s+b)(s+2)(s+3) + K(s+a) = 0$$

$$s = -1 \pm j$$

So, let's use -1-j

$$(s+b)(s+2)(s+3) + K(s+a) = 0$$

$$(b-1-j)(-1-j+2)(-1-j+3) + K(-1-j+a) = 0$$

$$(1-3i)(b-(1+i)) + K(a-(1+i)) = 0$$

$$(1 - 3j)(b - (1 + j)) + K(a - (1 + j)) = 0$$

$$(b - 4) + (2 - 3b)j + (a - 1)K - jk = 0$$

$$\begin{aligned} b - 4 + K(a - 1) &= 0 \\ (2 - 3b) - K &= 0 \end{aligned}$$

$$k = 2 - 3b$$

$$a = \frac{6 - 4b}{2 - 3b}$$

Let  $b = 0$ ,

$$K = 2, a = 3$$

$$D_c(s) = K \frac{s + a}{s + b} = \frac{2(s + 3)}{s} = \frac{2s + 6}{s}$$

$$D_c(s) = \frac{2s + 6}{s}$$

$$G(s) = \frac{1}{(s + 2)(s + 3)}$$

$$L(s) = D_c(s)G(s) = \frac{2s + 6}{s(s + 2)(s + 3)}$$

$$L(s) = D_c(s)G(s) = 2 \frac{s + 3}{s(s + 2)(s + 3)} = \frac{2}{s(s + 2)}$$

$$D_c(s) = \frac{2s + 6}{s}$$

$$1 + L(s) = 0$$

$$1 + 2L(s) = 0$$



$$1 + \frac{2}{s(s+2)} = 0$$

$$\frac{s(s+2)+2}{s(s+2)} = 0$$

$$\frac{s^2 + 2s + 2}{s(s+2)} = 0$$

$$s^2 + 2s + 2 = 0$$

$$s^2 + 2s + 2 = 0$$

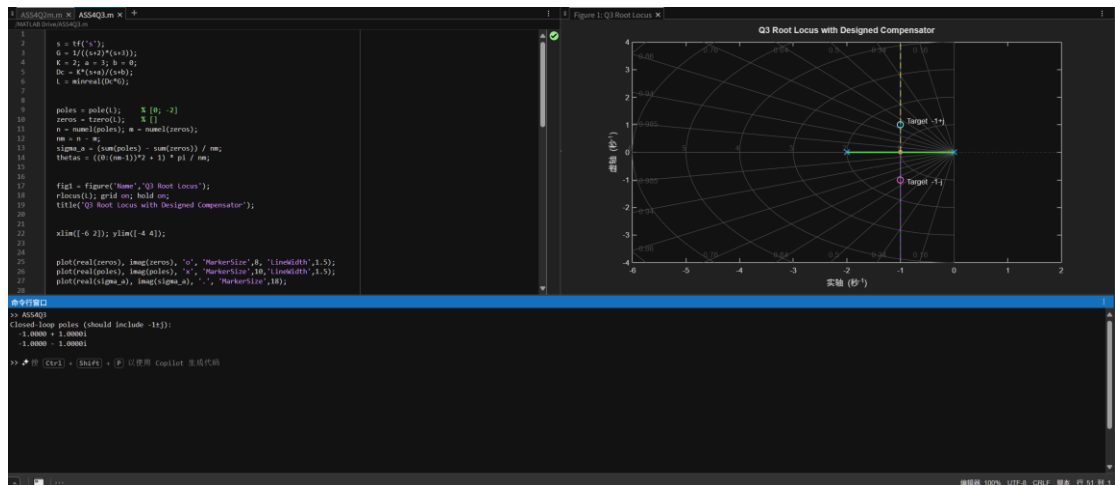
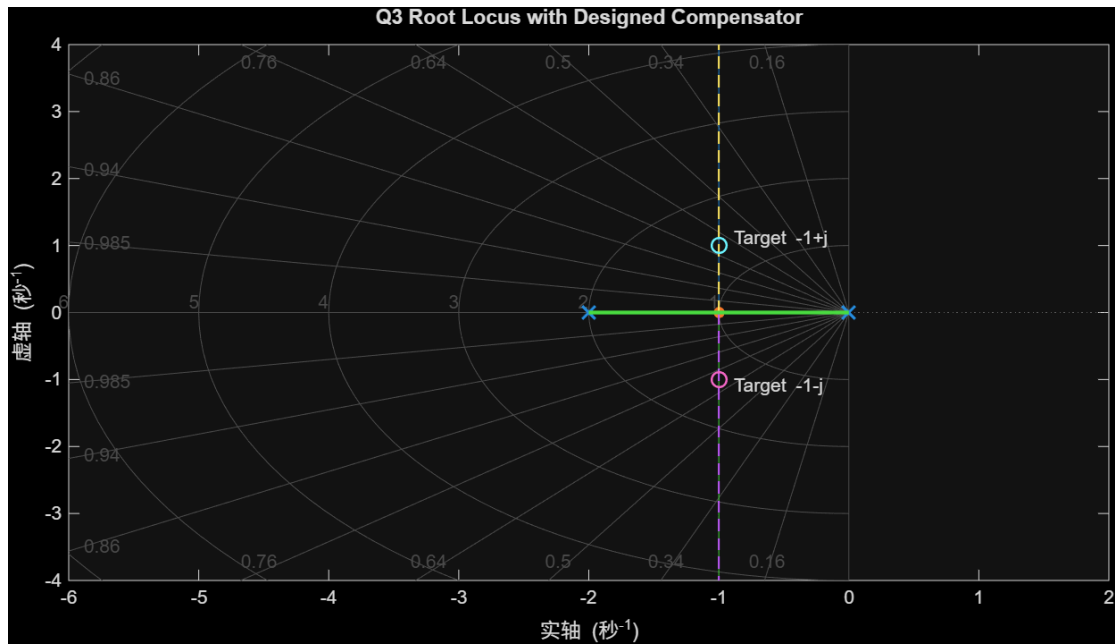
$$s = 1 \pm j$$

$$k = 2, a = 3, b = 0$$

$$D_c(s) = 2 \frac{s+3}{s}$$

This is a **lag compensator** because it approximates PI control (pole at origin, zero at -3), improving steady-state accuracy.

```
>> ASS4Q3
Closed-loop poles (should include -1±j):
-1.0000 + 1.0000i
-1.0000 - 1.0000i
```



```

poles = pole(L);
zeros_tf = tzero(L);

fprintf('L\n');
fprintf('Poles: ');
disp(poles);
fprintf('Zero: ');
disp(zeros_tf);

n = length(poles);
m = length(zeros_tf);

fprintf('\n ROOT LOCUS PARAMETERS \n');
fprintf('Number of poles n = %d\n', n);
fprintf('Number of Zero m = %d\n', m);
fprintf('Number of asymptotes n-m = %d\n', n-m);

if n > m

    sigma_a = (sum(real(poles)) - sum(real(zeros_tf))) / (n - m);

    asymptote_angles = zeros(1, n-m);
    for k = 0:(n-m-1)
        asymptote_angles(k+1) = (180 + 360*k) / (n-m);
    end

    fprintf('\n ASYMPTOTE CALCULATION \n');
    fprintf('Asymptote center:  $\sigma = %.2f$ \n', sigma_a);
    fprintf('Asymptote angles: ');
    for i = 1:length(asymptote_angles)
        fprintf('%.1f° ', asymptote_angles(i));
    end
    fprintf('\n');
end

s1 = -1 + 2j;
s2 = -1 - 2j;
zero = 0;

angle_to_zero_from_s1 = angle(zero - s1) * 180/pi;

```

```

angle_from_s2_to_s1 = angle(s1 - s2) * 180/pi;
departure_s1 = angle_to_zero_from_s1 - angle_from_s2_to_s1 + 180;

angle_to_zero_from_s2 = angle(zero - s2) * 180/pi;
angle_from_s1_to_s2 = angle(s2 - s1) * 180/pi;
departure_s2 = angle_to_zero_from_s2 - angle_from_s1_to_s2 + 180;

departure_s1 = mod(departure_s1 + 180, 360) - 180;
departure_s2 = mod(departure_s2 + 180, 360) - 180;

fprintf('\n DEPARTURE ANGLE CALCULATION \n');
fprintf('Departure angle from pole s1 = -1+2j: %.2f°\n', departure_s1);
fprintf('Departure angle from pole s2 = -1-2j: %.2f°\n', departure_s2);

figure('Position', [100, 100, 1200, 800]);

rlocus(L);
hold on;
grid on;

plot(real(poles), imag(poles), 'rx', 'MarkerSize', 12, 'LineWidth', 3);
plot(real(zeros_tf), imag(zeros_tf), 'go', 'MarkerSize', 12,
'LineWidth', 3);

if n > m

    asymptote_length = 10;

    for i = 1:length(asymptote_angles)
        angle_rad = asymptote_angles(i) * pi / 180;

        x_end = sigma_a + asymptote_length * cos(angle_rad);
        y_end = 0 + asymptote_length * sin(angle_rad);

        x_start = sigma_a - asymptote_length * cos(angle_rad);
        y_start = 0 - asymptote_length * sin(angle_rad);

        plot([x_start, x_end], [y_start, y_end], 'k--', ...

```

```

        'LineWidth', 1.5, 'DisplayName', sprintf('Asymptote %.1f°',
asymptote_angles(i)));
    end

    plot(sigma_a, 0, 'k+', 'MarkerSize', 15, 'LineWidth', 2, ...
        'DisplayName', 'Asymptote Center');
end

arrow_length = 1;

departure_rad_s1 = departure_s1 * pi / 180;
x_arrow_s1 = real(s1) + arrow_length * cos(departure_rad_s1);
y_arrow_s1 = imag(s1) + arrow_length * sin(departure_rad_s1);
quiver(real(s1), imag(s1), x_arrow_s1 - real(s1), y_arrow_s1 -
imag(s1), ...
    0, 'b', 'LineWidth', 2, 'MaxHeadSize', 0.3, 'DisplayName',
'Departure Angle');

departure_rad_s2 = departure_s2 * pi / 180;
x_arrow_s2 = real(s2) + arrow_length * cos(departure_rad_s2);
y_arrow_s2 = imag(s2) + arrow_length * sin(departure_rad_s2);
quiver(real(s2), imag(s2), x_arrow_s2 - real(s2), y_arrow_s2 -
imag(s2), ...
    0, 'b', 'LineWidth', 2, 'MaxHeadSize', 0.3);

text(real(s1) + 0.2, imag(s1) + 0.2, sprintf('s1\nDeparture: %.1f°',
departure_s1), ...
    'FontSize', 10, 'BackgroundColor', 'white');
text(real(s2) + 0.2, imag(s2) - 0.4, sprintf('s2\nDeparture: %.1f°',
departure_s2), ...
    'FontSize', 10, 'BackgroundColor', 'white');
text(real(zeros_tf) + 0.1, imag(zeros_tf) + 0.2, 'z1', ...
    'FontSize', 10, 'BackgroundColor', 'white');

title('Root Locus for L(s) = 5s/(s^2+2s+5) with Asymptotes',
'FontSize', 14);
xlabel('Real Axis (seconds^{-1})', 'FontSize', 12);
ylabel('Imaginary Axis (seconds^{-1})', 'FontSize', 12);

```

```
h_legend = [];  
labels_legend = {};  
  
h_rlocus = findobj(gca, 'Type', 'line', 'Color', 'b');  
if ~isempty(h_rlocus)  
    h_legend(end+1) = h_rlocus(1);  
    labels_legend{end+1} = 'Root Locus';  
end  
  
h_pole = findobj(gca, 'Marker', 'x', 'Color', 'r');  
if ~isempty(h_pole)  
    h_legend(end+1) = h_pole(1);  
    labels_legend{end+1} = 'Poles';  
end  
  
h_zero = findobj(gca, 'Marker', 'o', 'Color', 'g');  
if ~isempty(h_zero)  
    h_legend(end+1) = h_zero(1);  
    labels_legend{end+1} = 'Zero';  
end  
  
h_asym = findobj(gca, 'LineStyle', '--', 'Color', 'k');  
if ~isempty(h_asym)  
    h_legend(end+1) = h_asym(1);  
    labels_legend{end+1} = 'Asymptotes';  
end  
  
h_center = findobj(gca, 'Marker', '+', 'Color', 'k');  
if ~isempty(h_center)  
    h_legend(end+1) = h_center(1);  
    labels_legend{end+1} = 'Asymptote Center';  
end  
  
legend(h_legend, labels_legend, 'Location', 'best', 'FontSize', 10);  
  
xlim([-8, 2]);  
ylim([-4, 4]);  
  
hold off;
```

```

fprintf('\nZero REAL AXIS ROOT LOCUS ANALYSIS Zero\n');
fprintf('According to Rule 2: Real axis root locus exists to the left
of odd number of poles and Zero\n');
fprintf('Zero location: 0\n');
fprintf('Pole locations: -1±2j (not on real axis)\n');

fprintf('\nZero ROOT LOCUS VERIFICATION Zero\n');
test_points = [-0.5, -2, -3];
for i = 1:length(test_points)
    s_test = test_points(i);
    L_value = 5*s_test / (s_test^2 + 2*s_test + 5);
    alpha_required = -1 / L_value;

    fprintf('Test point s = %.1f: α = %.3f\n', s_test, alpha_required);

    if alpha_required >= 0
        fprintf(' → This point is on the root locus (α ≥ 0)\n');
    else
        fprintf(' → This point is not on the root locus (α < 0)\n');
    end
end

```

Q3

```

s = tf('s');
G = 1/((s+2)*(s+3));
K = 2; a = 3; b = 0;
Dc = K*(s+a)/(s+b);
L = minreal(Dc*G);

poles = pole(L);    % [0; -2]
zeros = tzero(L);   % []
n = numel(poles); m = numel(zeros);
nm = n - m;
sigma_a = (sum(poles) - sum(zeros)) / nm;
thetas = ((0:(nm-1))*2 + 1) * pi / nm;

```

```

fig1 = figure('Name','Q3 Root Locus');
rlocus(L); grid on; hold on;
title('Q3 Root Locus with Designed Compensator');

xlim([-6 2]); ylim([-4 4]);

plot(real(zeros), imag(zeros), 'o', 'MarkerSize',8, 'LineWidth',1.5);
plot(real(poles), imag(poles), 'x', 'MarkerSize',10,'LineWidth',1.5);
plot(real(sigma_a), imag(sigma_a), '.', 'MarkerSize',18);

XL = xlim; YL = ylim; R = max([diff(XL), diff(YL)])*1.1;
for th = thetas
    x2 = real(sigma_a) + R*cos(th);
    y2 = imag(sigma_a) + R*sin(th);
    plot([real(sigma_a) x2], [imag(sigma_a) y2], '--', 'LineWidth',1.0);
end

plot([-2 0], [0 0], '-', 'LineWidth',2.0);

plot(-1, 1, 'o', 'MarkerSize',8, 'LineWidth',1.2);
plot(-1,-1, 'o', 'MarkerSize',8, 'LineWidth',1.2);
text(-0.9, 1.1, 'Target -1{+}j', 'FontSize',10);
text(-0.9,-1.1, 'Target -1{-}j', 'FontSize',10);

T = feedback(L, 1);
disp('Closed-loop poles (should include -1±j):');
disp(pole(T));

```