

Optimal Wideband Precoder For MIMO

October 26, 2021

For a MIMO wireless communication, we have sampled transmit signal $x[n] \in \mathbb{C}^M$, receive signal $y[n] \in \mathbb{C}^N$, and a channel $h[n] \in \text{Hom}(\mathbb{C}^M, \mathbb{C}^N)$ all with lengths P

$$y[n] = \sum_{n=0}^{P-1} h[n]x[n]$$

And in the frequency domain we have

$$Y[k] = H[k]X[k]$$

where $k = 0, \dots, P-1$.

We want to find an approximation of $h[n]$ denoted by \hat{h} such that y and $\hat{h}x$ are as close as possible, or, Y and $\hat{h}X$ as close as possible. Later we will show that these two requirements are equivalent.

Note that y and Y are elements of a $\text{Hom}(\mathbb{C}^N)$ -module, h and H are elements of another $\text{Hom}(\mathbb{C}^N)$ -module, and x as well as X are elements of a $\mathbb{C}^{M \times M}$ -module. To measure how close to elements of a module are, we can define an "inner product" analogy to the usual inner product defined on Hilbert spaces.

Definition 1 (Inner product on $\text{Hom}(\mathbb{C}^N)$ -modules). An auto-correlation on a P -dimensional free $\text{Hom}(\mathbb{C}^N)$ -module \mathcal{M} is a operation $\langle \cdot, \cdot \rangle: \mathcal{M} \times \mathcal{M} \rightarrow \text{Hom}(\mathbb{C}^N)$

$$\langle x, y \rangle = \sum_{n=0}^{P-1} x[n]y[n]^*$$

We also refer C_{xy} to $\langle x, y \rangle$ for convenience under some context.

Definition 2 (Orthogonality on modules). $x, y \in \mathcal{M}$ are orthogonal if

$$\langle x, y \rangle = 0$$

It is easy to verify that $\langle x, x \rangle$ is positive semi-definite. Therefore we can define the norm induced by the inner product

Definition 3 (Euclidean norm on modules). The Euclidean norm of $x \in \mathcal{M}$ is

$$||x|| = \langle x, x \rangle^{1/2}$$

It is obvious that the norm of any element is positive semi-definite. Hence, there is a partial order \preceq induced by the positive semi-definite cone. And we are able to measure the closeness of two elements in a module by this partial order.

The theorem follows tells us that we can study the closeness in either time domain or frequency domain.

Theorem 1 (Extended Parseval's Theorem).

$$||X|| = \sqrt{P}||x||$$

w.r.t. the DFT defined by

$$X[k] = \sum_{n=0}^{P-1} x[n] \omega_P^{nk}$$

where

$$\omega_P^{nk} = e^{-j2\pi nk/P}$$

Proof.

$$\begin{aligned} ||X||^2 &= \sum_{k=0}^{P-1} X[k] X[k]^* \\ &= \sum_{k=0}^{P-1} \left(\sum_{m=0}^{P-1} x[m] \omega_P^{mk} \right) \left(\sum_{n=0}^{P-1} x[n] \omega_P^{nk} \right)^* \\ &= \sum_{k=0}^{P-1} \sum_{m=0}^{P-1} \sum_{n=0}^{P-1} x[m] x[n]^* \omega_P^{(m-n)k} \end{aligned}$$

Note that the terms being summed are non-zero only if $m = n$, thus

$$||X||^2 = \sum_{k=0}^{P-1} Px[k]x[k]^* = P||x||^2$$

completes the proof. \square

To find the optimal \hat{h} in time domain is equivalent to find it in the frequency domain. This amounts to say that we are trying to approximate the frequency-selective $H[k]$ with a constant \hat{h} over all frequencies.

We need to extend the orthogonality principle to modules. Before that we need another kind of product of elements from different modules.

Definition 4 (Cross-correlation). The cross-correlation is a binary operator defined on the product of two modules $\langle \cdot, \cdot \rangle: \mathcal{N} \times \mathcal{M} \rightarrow \text{Hom}(\mathbb{C}^M, \mathbb{C}^N)$

$$\langle x, y \rangle = \sum_{n=0}^{P-1} x[n]y[n]^*$$

And we sometimes refer C_{xy} to $\langle x, y \rangle$ for convenience under certain contexts.

When $N = M$, the cross-correlation coincides with auto-correlation.

Lemma 1. *Let $z \in \mathcal{N}$, $x \in \mathcal{M}$ s.t. $C_{zx} = 0$. Then $\forall \mu \in \text{Hom}(\mathbb{C}^M, \mathbb{C}^N)$ we have*

$$||z + \mu x|| \succeq ||z||$$

Proof. It is easy to verify that

$$||z + \mu x||^2 = ||z||^2 + ||\mu x||^2$$

and we are done. \square

It follows from lemma 1 that every $y \in \mathcal{N}$ can be decomposed into two parts

$$y = z + \mu x$$

where $C_{zx} = 0$ and $\mu \in \text{Hom}(\mathbb{C}^M, \mathbb{C}^N)$