



## 2.2 控制系统的传递函数

2.2.1 传递函数定义

2.2.2 典型环节传递函数

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## 2.2.1 传递函数定义

零初始条件下，系统输出量的拉氏变换与输入量的拉氏变换之比。

$$G(s) = \frac{Y(s)}{U(s)}$$

$$\begin{aligned} & \frac{d^n y(t)}{dt^n} + a_{n-1} \frac{d^{n-1} y(t)}{dt^{n-1}} + \dots + a_1 \frac{dy(t)}{dt} + a_0 y(t) \\ &= b_m \frac{d^m u(t)}{dt^m} + b_{m-1} \frac{d^{m-1} u(t)}{dt^{m-1}} + \dots + b_1 \frac{du(t)}{dt} + b_0 u(t) \end{aligned} \quad (1)$$



First step:  $y(0) = 0 \quad y'(0) = 0$

Second step:  $L\left[\frac{df(t)}{dt}\right] = sF(s) - f(0)$   $L\left[\frac{df(t)}{dt}\right] = sF(s)$

$$L\left[\frac{d^n f(t)}{dt^n}\right] = s^n \cdot F(s) - s^{n-1}f(0) - s^{n-2}f^{(1)}(0) - \dots - f^{(n-1)}(0)$$



$$L\left[\frac{d^n f(t)}{dt^n}\right] = s^n \cdot F(s)$$

$$\frac{d^n y(t)}{dt^n} + a_{n-1} \frac{d^{n-1}y(t)}{dt^{n-1}} + \dots + a_1 \frac{dy(t)}{dt} + a_0 y(t) = b_m \frac{d^m u(t)}{dt^m} + b_{m-1} \frac{d^{m-1}u(t)}{dt^{m-1}} + \dots + b_1 \frac{du(t)}{dt} + b_0 u(t)$$



$$(s^n + a_{n-1}s^{n-1} + \dots + a_1s + a_0)Y(s) = (b_m s^m + b_{m-1}s^{m-1} + \dots + b_1s + b_0)U(s)$$

Third step:

$$(s^n + a_{n-1}s^{n-1} + \dots + a_1s + a_0)Y(s) = (b_ms^m + b_{m-1}s^{m-1} + \dots + b_1s + b_0)U(s)$$

$$G(s) = \frac{Y(s)}{U(s)} = \frac{b_ms^m + b_{m-1}s^{m-1} + \dots + b_1s + b_0}{s^n + a_{n-1}s^{n-1} + \dots + a_1s + a_0} \quad (3)$$

传递函数性质：

$$G(s) = \frac{Y(s)}{U(s)} = \frac{b_ms^m + b_{m-1}s^{m-1} + \dots + b_1s + b_0}{s^n + a_{n-1}s^{n-1} + \dots + a_1s + a_0}$$

**The rational function:** the transfer function is called to be **strictly proper** if the order of the denominator polynomial is greater than that of the numerator polynomial (i.e.,  $n > m$ ). if  $n = m$ , the transfer function is called **proper**.

$$n \geq m$$

$$G(s) = L[g(t)]$$

$$R(s) = L[\delta(t)] = 1 \quad Y(s) = G(s) \cdot R(s) \quad R(s) = 1 \quad Y(s) = G(s)$$

$$g(t) = y(t) = L^{-1}[Y(s)] = L^{-1}[G(s)]$$

$$G(s) = \frac{b_m s^m + b_{m-1} s^{m-1} + \dots + b_0}{s^n + a_{n-1} s^{n-1} + \dots + a_0} = \frac{N(s)}{D(s)}$$

$$G(s) = K^* \frac{\prod_{i=1}^m (s + z_i)}{\prod_{i=1}^n (s + p_i)}$$



$$G(s) = \frac{K(\tau_1 s + 1)(\tau_2^2 s^2 + 2\zeta \tau_2 s + 1) \cdots (\tau_i s + 1)}{(T_1 s + 1)(T_2^2 s^2 + 2\zeta T_2 s + 1) \cdots (T_j s + 1)}$$

The roots of the numerator  $-z_1, -z_2, \dots, -z_m$  are called the finite **zeros** of the system.

The roots of the denominator,  $-p_1, -p_2, \dots, -p_n$  are called the **poles** of the system. The poles are locations in the s-plane where the magnitude of the transfer function becomes infinite.



## 2.2.2 典型环节传递函数

1) 比例环节 (又称放大环节)



2) 惯性环节

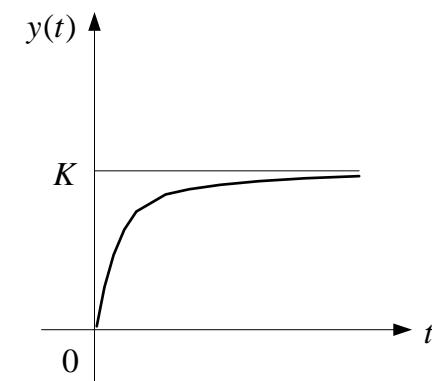
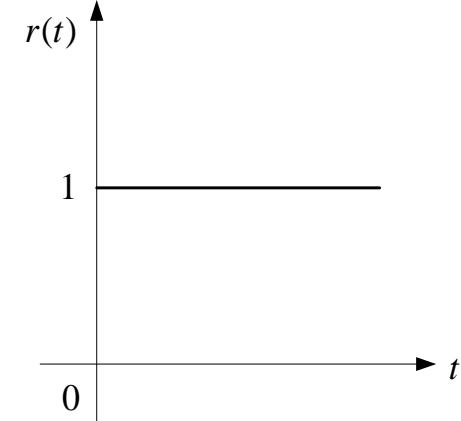
$$r(t) = 1(t)$$

$$y(t) = K \left( 1 - e^{-t/T} \right)$$

$$G(s) = \frac{Y(s)}{R(s)} = \frac{K}{Ts + 1}$$

?

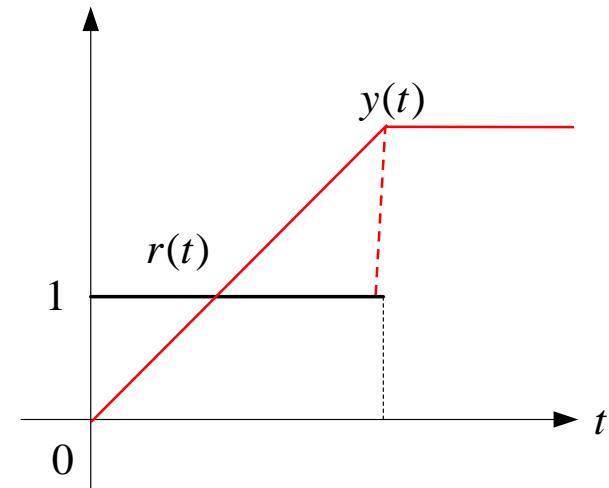
$$T \frac{dy(t)}{dt} + y(t) = Kr(t)$$





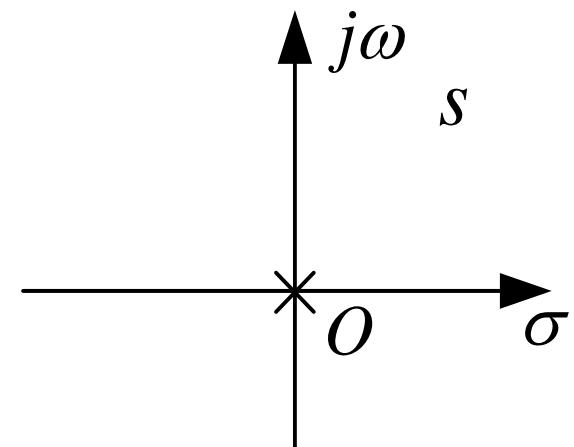
### 3) 积分环节

$$y(t) = \frac{1}{T} \int_0^t r(\tau) d\tau$$



$$r(t) = 1(t) \quad y(t) = t/T$$

$$G(s) = \frac{1}{Ts} = \frac{K}{s}$$



## 4) 微分环节

理想的纯微分环节

$$y(t) = \tau \frac{dr(t)}{dt}$$

$$r(t) = 1(t) \quad y(t) = \tau \cdot \delta(t)$$

$$G(s) = \tau s$$

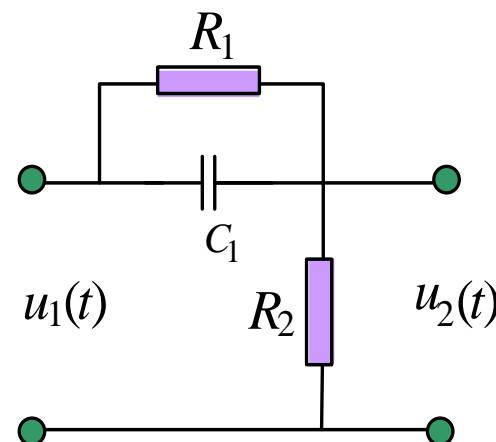
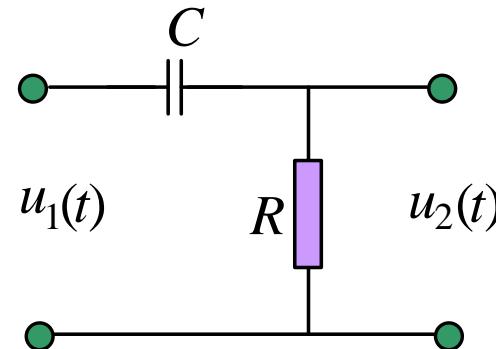
理想的一阶和二阶微分环节的传递函数分别为

$$G(s) = 1 + \tau s \quad G(s) = 1 + 2\zeta\tau s + \tau^2 s^2$$

实际物理系统

$$G(s) = \frac{RCs}{RCs + 1}$$

$$G(s) = \frac{K(\tau_1 s + 1)}{\tau_2 s + 1}$$



## 5) 振荡环节

$$T^2 \frac{d^2y(t)}{dt^2} + 2\zeta T \frac{dy(t)}{dt} + y(t) = r(t)$$

$$G(s) = \frac{1}{T^2 s^2 + 2\zeta T s + 1} = \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2}$$

## 6) 滞后环节

$$y(t) = r(t - \tau)$$

$$G(s) = e^{-\tau s}$$



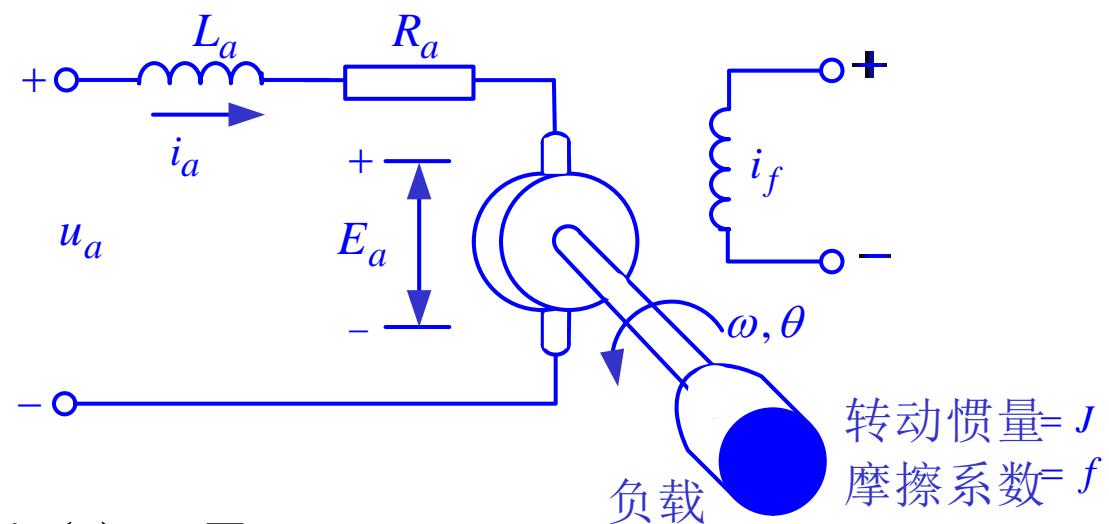
## 2.2.3 举例说明建立传递函数的方法

- 重新考虑电枢控制式直流电动机，试求以电枢控制电压 $u_a(t)$  为输入量，电动机转角 $\theta$  为输出量的传递函数。

$$E_a = C_e \omega(t)$$

$$M(t) = C_m i_a(t)$$

$$u_a(t) = L_a \frac{di_a(t)}{dt} + R_a i_a(t) + E_a$$



$$J \frac{d\omega(t)}{dt} + f\omega(t) = M(t) - M_c(t)$$

$$\omega(t) = \frac{d\theta(t)}{dt}$$



$$u_a(t) = L_a \frac{di_a(t)}{dt} + R_a i_a(t) + E_a$$

$$E_a = C_e \omega(t)$$

$$M(t) = C_m i_a(t)$$

$$J \frac{d\omega(t)}{dt} + f\omega(t) = M(t) - M_c(t)$$

$$\omega(t) = \frac{d\theta(t)}{dt}$$



$$U_a(s) = L_a s I_a(s) + R_a I_a(s) + E_a$$

$$E_a = C_e \omega(s)$$

$$M(s) = C_m I_a(s)$$

$$J s \omega(s) + f \omega(s) = M(s) - M_c(s)$$

$$\omega(s) = s \theta(s)$$



$$L_a J \frac{d^2 \omega(t)}{dt^2} + (L_a f + R_a J) \frac{d\omega(t)}{dt} + (R_a f + C_m C_e) \omega(t) =$$

$$I_a(s), E_a, M(s), \omega(s)$$



$$C_m u_a(t) - L_a \frac{dM_c(t)}{dt} - R_a M_c(t)$$



$$G(s) = \frac{\theta(s)}{U_a(s)} = \frac{C_m}{s[(L_a s + R_a)(J s + f) + C_e C_m]}$$

$$G(s) = \frac{\theta(s)}{U_a(s)} = \frac{C_m}{s[(L_a s + R_a)(J_s + f) + C_e C_m]}$$

电枢时间常数  $\tau_a = L_a / R_a$  可以忽略不计

$$G(s) = \frac{\theta(s)}{U_a(s)} = \frac{C_m}{s[R_a(J_s + f) + C_e C_m]} = \frac{K_1}{s(T_m s + 1)}$$

如果电枢电阻和电动机的转动惯量都很小，可忽略不计时

$$G(s) = \frac{\theta(s)}{U_a(s)} = \frac{K_1}{s}$$

可将直流电动机作为积分环节

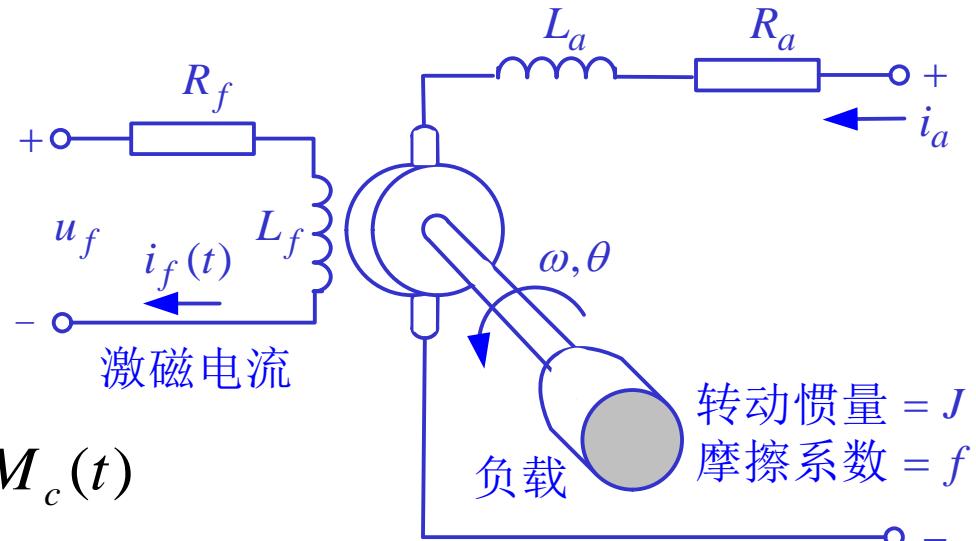
2. 重新考虑磁场控制式直流电动机，试求以激磁电压为输入量，电动机转角为输出量的传递函数。

$$u_f(t) = L_f \frac{di_f(t)}{dt} + R_f i_f(t)$$

$$M(t) = k_m i_f(t)$$

$$J \frac{d\omega(t)}{dt} + f\omega(t) = M(t) - M_c(t)$$

$$\omega(t) = \frac{d\theta(t)}{dt}$$

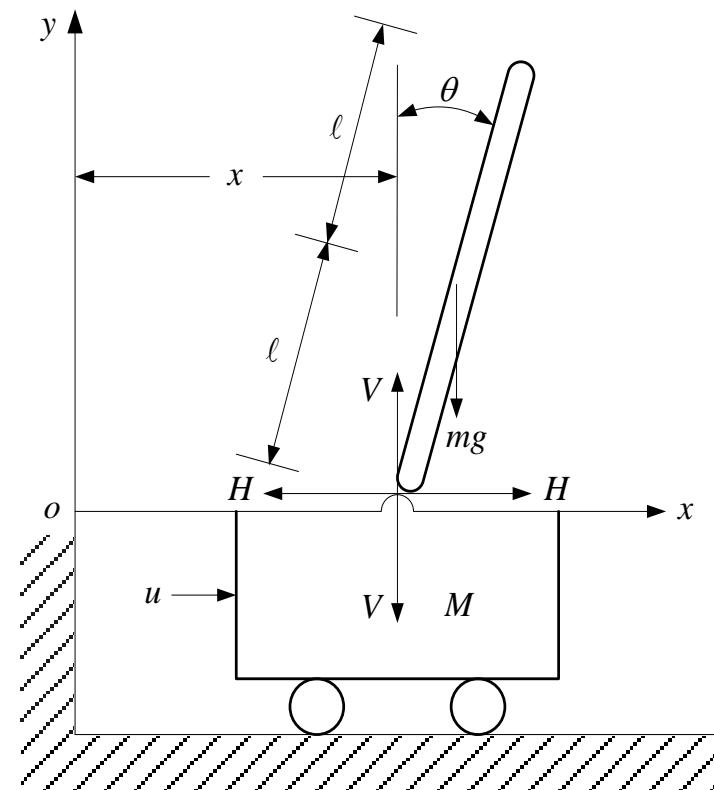
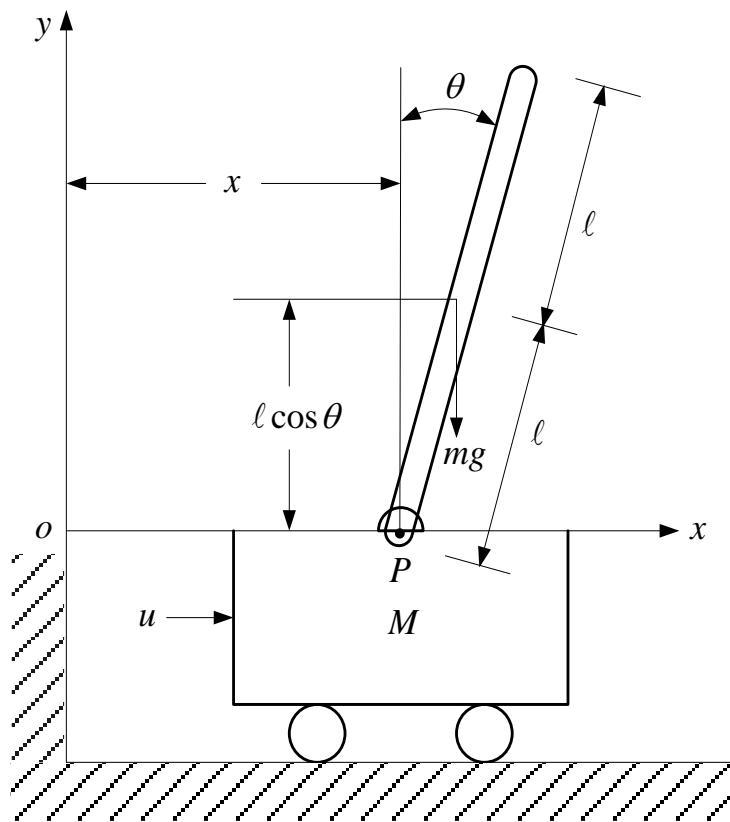


$$T_m = \frac{J}{f} \quad T_f = \frac{L_f}{R_f}$$

$$G(s) = \frac{\theta(s)}{U_f(s)} = \frac{k_m}{s(L_f s + R_f)(Js + f)} = \frac{k'_m}{s(T_m s + 1)(T_f s + 1)}$$



3. 重新考虑倒立摆系统，试求这个系统的传递函数。

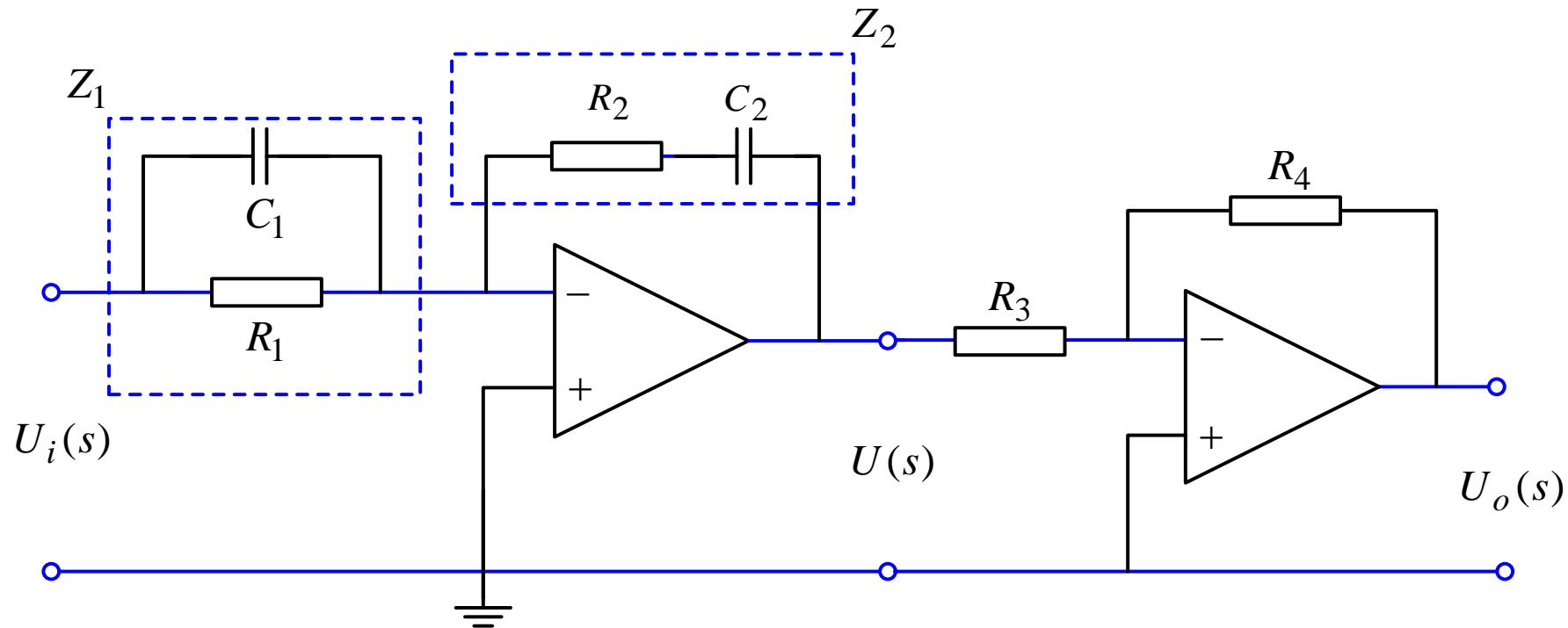


$$(M + m)\ddot{x} + m\ell\ddot{\theta} = u$$

$$(I + m\ell^2)\ddot{\theta} + m\ell\ddot{x} = mg\ell\theta$$

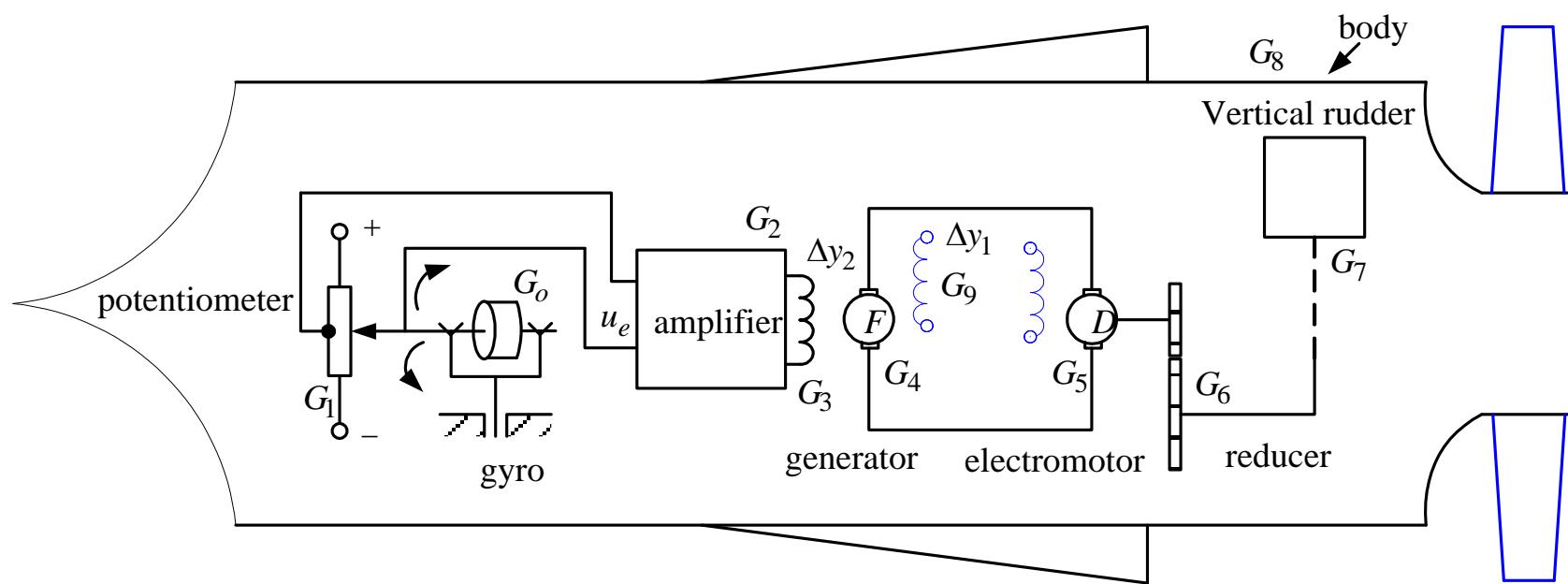
$$G(s) = \frac{\theta(s)}{U(s)} = \frac{-\frac{m\ell}{M+m}}{\left[(I + m\ell^2) - \frac{m^2\ell^2}{M+m}\right]s^2 - mg\ell}$$

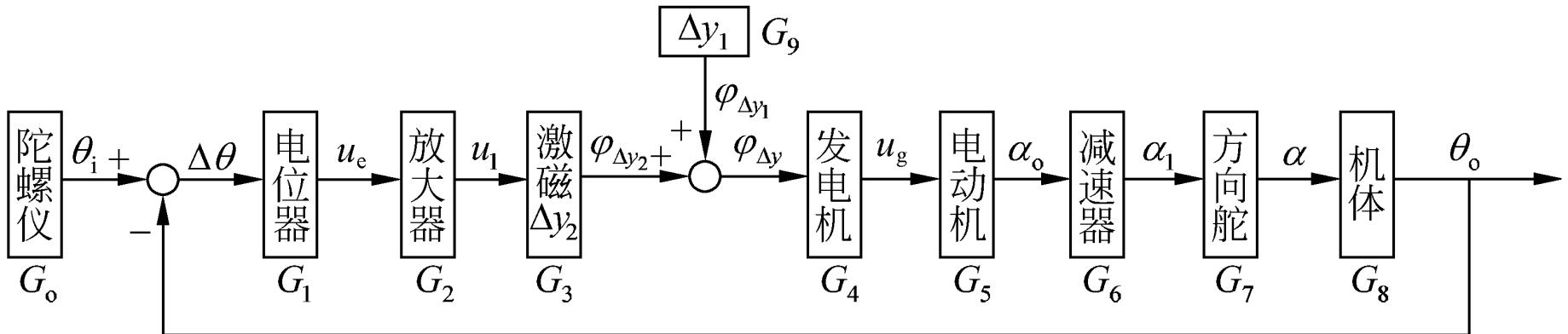
4. 比例、积分、微分（亦即PID proportion integral differential）控制器如图所示，试求其传递函数。



$$G(s) = \frac{U_o(s)}{U_i(s)} = K_p + \frac{K_i}{s} + K_d s$$

5. 导弹航向控制系统。试建立以陀螺仪角度为输入量，导弹机体航向角度为输出量的系统传递函数。



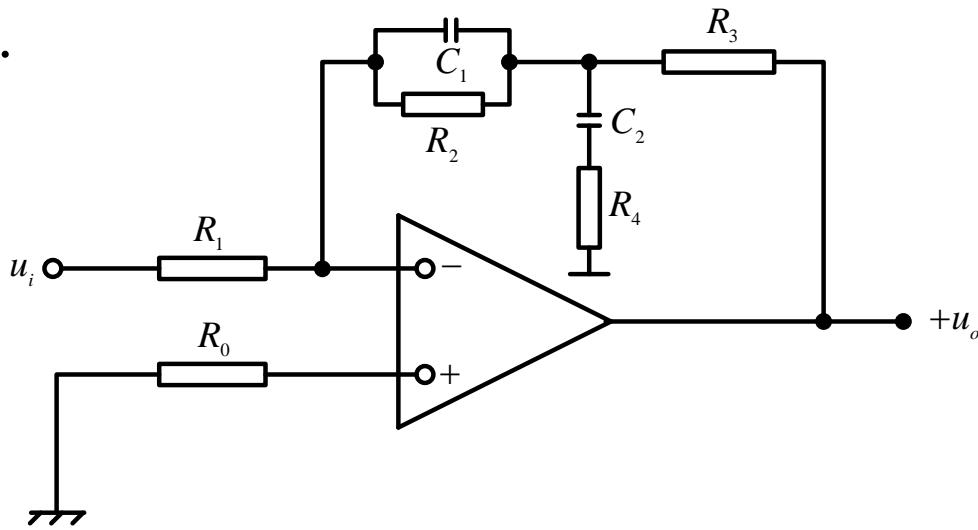


导弹航向控制系统的传递函数为：

$$G(s) = \frac{\theta_o(s)}{\theta_i(s)} = \frac{G'(s)}{1 + G'(\textcolor{red}{?})} \cdot \frac{k}{(T_m s + 1)(T_f s + 1) + k}$$



6.



$$Z_1 = \frac{R_2 \cdot \frac{1}{C_1 s}}{R_2 + \frac{1}{C_1 s}} = \frac{R_2}{1 + R_2 C_1 s}$$

$$Z_2 = R_4 + \frac{1}{C_2 s}$$

$$Z = \frac{Z_1 \cdot Z_2}{Z_1 + Z_2} = \frac{R_2 (1 + R_4 C_2 s)}{R_2 C_2 s + (1 + R_2 C_1 s)(1 + R_4 C_2 s)} \quad \frac{u_i}{R_1} = -\frac{u_o}{R_3 + Z}$$

$$G(s) = \frac{u_o}{u_i} = -\frac{R_3 + Z}{R_1}$$

$$G(s) = -k_1 - \frac{k_2 + k_3 s}{T_1 s^2 + T_2 s + 1}$$



### 三、传递函数

#### 1. 初始条件非零时系统的传递函数如何求取？

例1：当  $r(t) = 1(t)$ ， $y(t) = 1 - 2e^{-t} + e^{-2t}$  时，求系统的传递函数  $G(s)$ 。

例2：已知系统初始条件为  $y(0) = -1$ ,  $\dot{y}(0) = 0$ ，  
在  $t = 0$  给系统加入  $r(t) = 1(t)$ ，响应为  
 $y(t) = 1 - 4e^{-t} + 2e^{-2t}$ ，求传递函数。

#### 2. 传递函数分解为典型环节，有什么意义？