

Lecture 9

Wavelet and Other Image

Transforms

Dr. Xiran Cai

Email: caixr@shanghaitech.edu.cn

Office: 3-438 SIST

Tel: 20684431

ShanghaiTech University



Outline

- ❑ **2D Unitary transform**
- ❑ **Frequency Domain Extension**
 - Discrete Cosine Transform（余弦变换）
 - Hadamard Transform（哈德马变换）
 - Discrete Wavelet Transform（小波变换）
- ❑ **Discrete Wavelet Transform (DWT)**
 - An example for 1D-DWT
 - Generalization of 1D-DWT
 - 2D-DWT



Unitary Transform

□ Forward Transform:

$$t = Af$$

$$t[k] = \sum_{n=0}^{N-1} A[k, n]f[n]$$

□ Inverse Transform:

$$f = A^H t \quad \text{if } A^H = (A^T)^* \quad \text{and} \quad AA^H = I$$



Example for 1D Unitary Transform

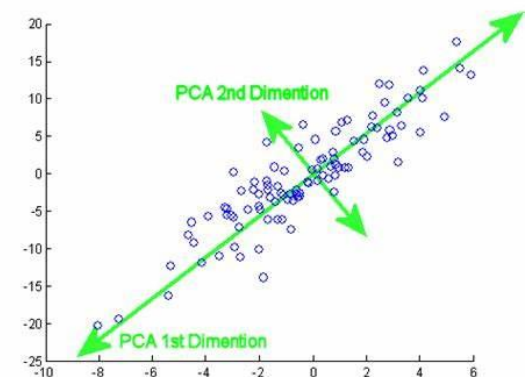
□ Image rotation:

$$\begin{bmatrix} x' \\ y' \end{bmatrix} = \begin{bmatrix} \cos\theta & \sin\theta \\ -\sin\theta & \cos\theta \end{bmatrix} * \begin{bmatrix} x \\ y \end{bmatrix}$$

□ Principle Component Analysis (PCA):

$Y = PX$ that satisfies $C = XX^T$ $D = PCP^T$

and $PP^T = I$



Discrete Fourier Transform

➤ Forward Transform:

$$t = Af; \quad t[k] = \sum_{n=0}^{N-1} A[k, n] f[n]$$

➤ Inverse Transform:

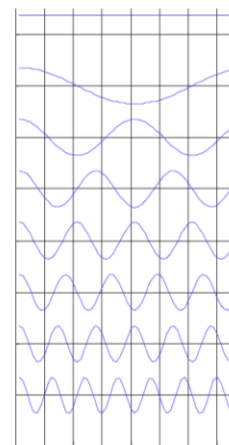
$$f = A^H t; \quad f[n] = \sum_{k=0}^{N-1} A^H[k, n] t[k]$$

➤ 1-D DFT

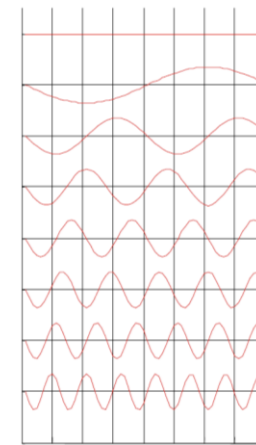
$$X[k] = \sum_{n=0}^{N-1} x[n] e^{-j\frac{2\pi kn}{N}}, \quad (k = 1, 2, \dots, N)$$

$$A[k, n] = e^{-j\frac{2\pi kn}{N}} = \cos(2\pi \frac{kn}{N}) - j\sin(2\pi \frac{kn}{N})$$

Real(A)



Imag(A)



2D Unitary Transform

□ Forward Transform:

$$F(u, v) = \sum_{x=0}^{M-1} \sum_{y=0}^{N-1} f(x, y) e^{-j2\pi(\frac{ux}{M} + \frac{vy}{N})}$$
$$= A_M f A_N$$

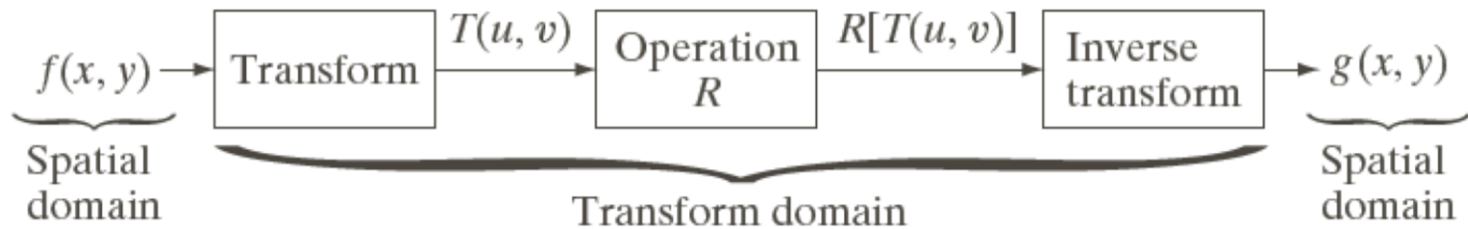
□ Inverse Transform

$$f = A_M^H F A_N^H \qquad A A^H = I$$



Image Transform

- ❑ The general approach for operating in linear transform domain



- ❑ The unitary transform satisfies

$$\sum_{x=0}^M \sum_{y=0}^N (f[x, y])^2 = \sum_{u=0}^M \sum_{v=0}^N (F[u, v])^2$$

- i.e., the energy is preserved.



Good and Bad things about DFT

❑ Positive:

- Energy is usually packed into low-frequency coefficients
- Convolution property
- Fast implementation

❑ Negative:

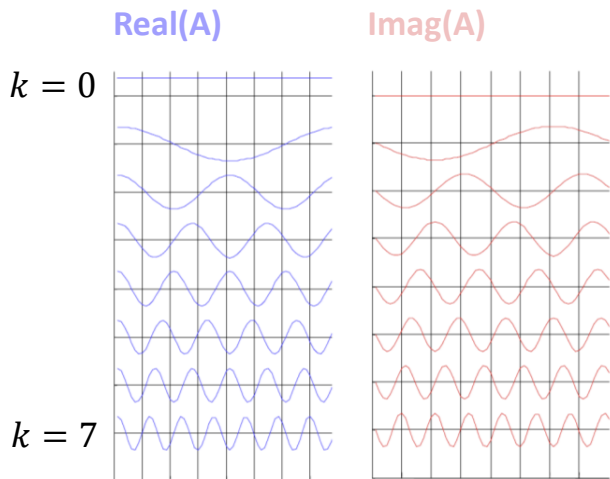
- Transform is complex, even if image is real
- The basis function span image height/width



DFT vs. DCT (Discrete Cosine Transform)

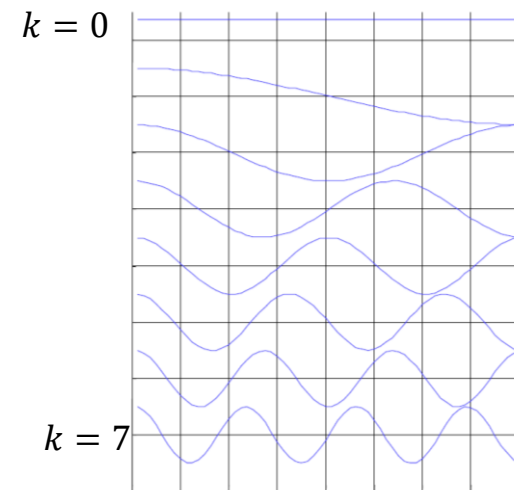
➤ 1D-DFT

$$A[k, n] = e^{-j\frac{2\pi kn}{N}}$$
$$= \cos\left(2\pi \frac{kn}{N}\right) + j\sin\left(2\pi \frac{kn}{N}\right)$$



➤ 1D-DCT

$$A[k, n] = \sqrt{\frac{2}{N}} \cos \frac{\pi(2n+1)k}{2N}$$



What's the difference???



2D DCT

□ Forward Transform

$$F(0,0) = \frac{1}{N} \sum_{x=0}^{N-1} \sum_{y=0}^{N-1} f(x,y)$$

$$F(u,0) = \frac{\sqrt{2}}{N} \sum_{x=0}^{N-1} \sum_{y=0}^{N-1} f(x,y) \cos \frac{(2x+1)u\pi}{2N}$$

$$F(0,v) = \frac{\sqrt{2}}{N} \sum_{x=0}^{N-1} \sum_{y=0}^{N-1} f(x,y) \cos \frac{(2y+1)v\pi}{2N}$$

$$F(u,v) = \frac{2}{N} \sum_{x=0}^{N-1} \sum_{y=0}^{N-1} f(x,y) \cos \frac{(2x+1)u\pi}{2N} \cos \frac{(2y+1)v\pi}{2N}$$



2D IDCT

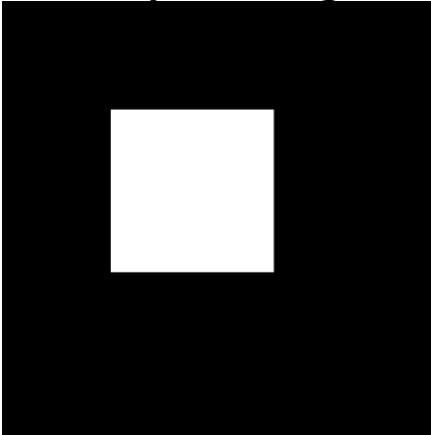
□ Inverse Transform

$$\begin{aligned} f(x, y) = & \frac{1}{N} F(0, 0) \\ & + \frac{\sqrt{2}}{N} \sum_{u=1}^{N-1} F(u, 0) \cos \frac{(2x+1)u\pi}{2N} \\ & + \frac{\sqrt{2}}{N} \sum_{v=1}^{N-1} F(0, v) \cos \frac{(2y+1)v\pi}{2N} \\ & + \frac{2}{N} \sum_{x=1}^{N-1} \sum_{y=1}^{N-1} F(u, v) \cos \frac{(2x+1)u\pi}{2N} \cos \frac{(2y+1)v\pi}{2N} \end{aligned}$$

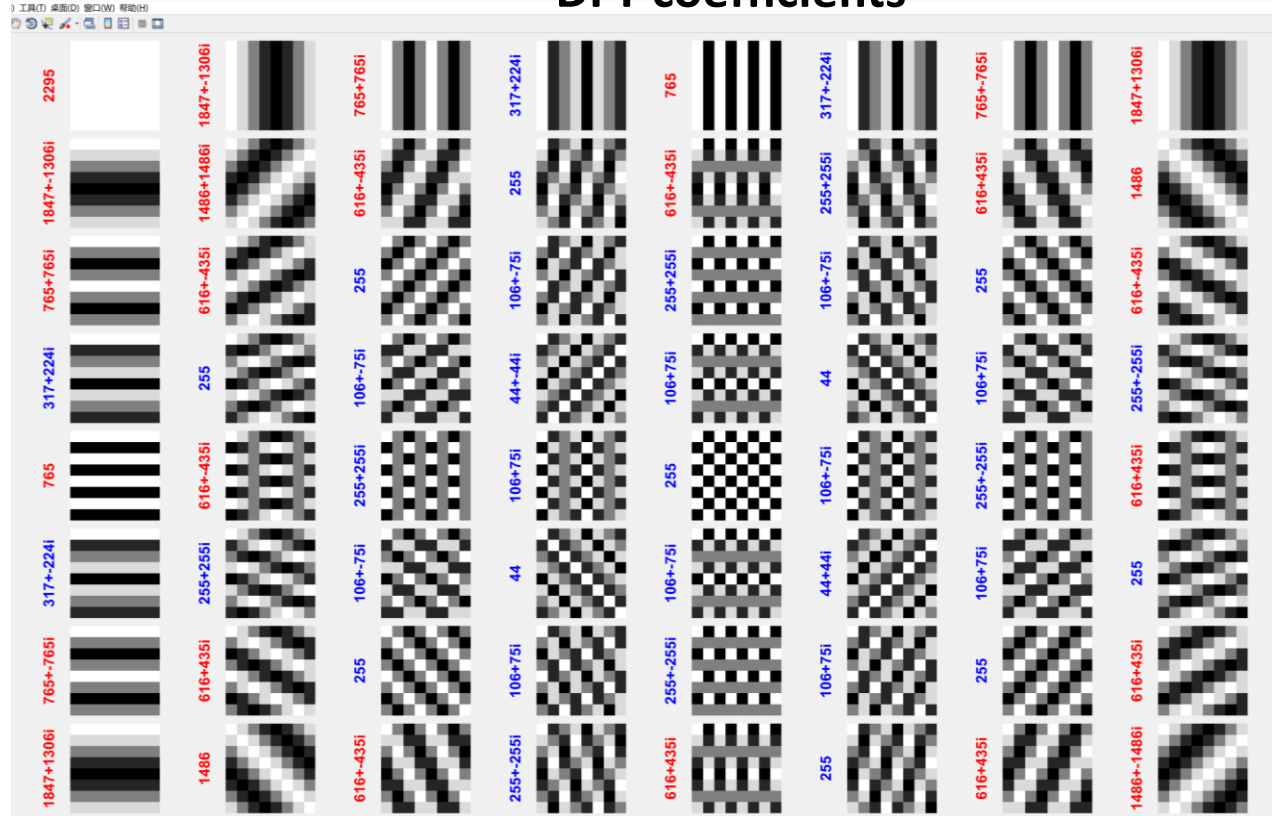


DFT example

Input image

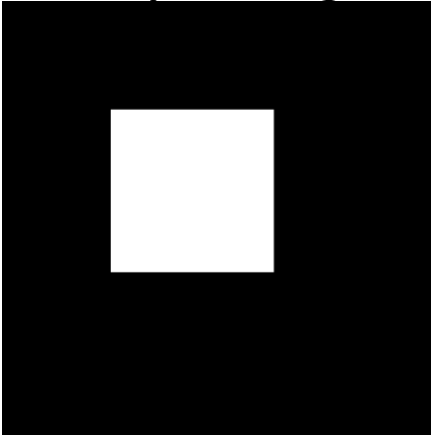


DFT coefficients

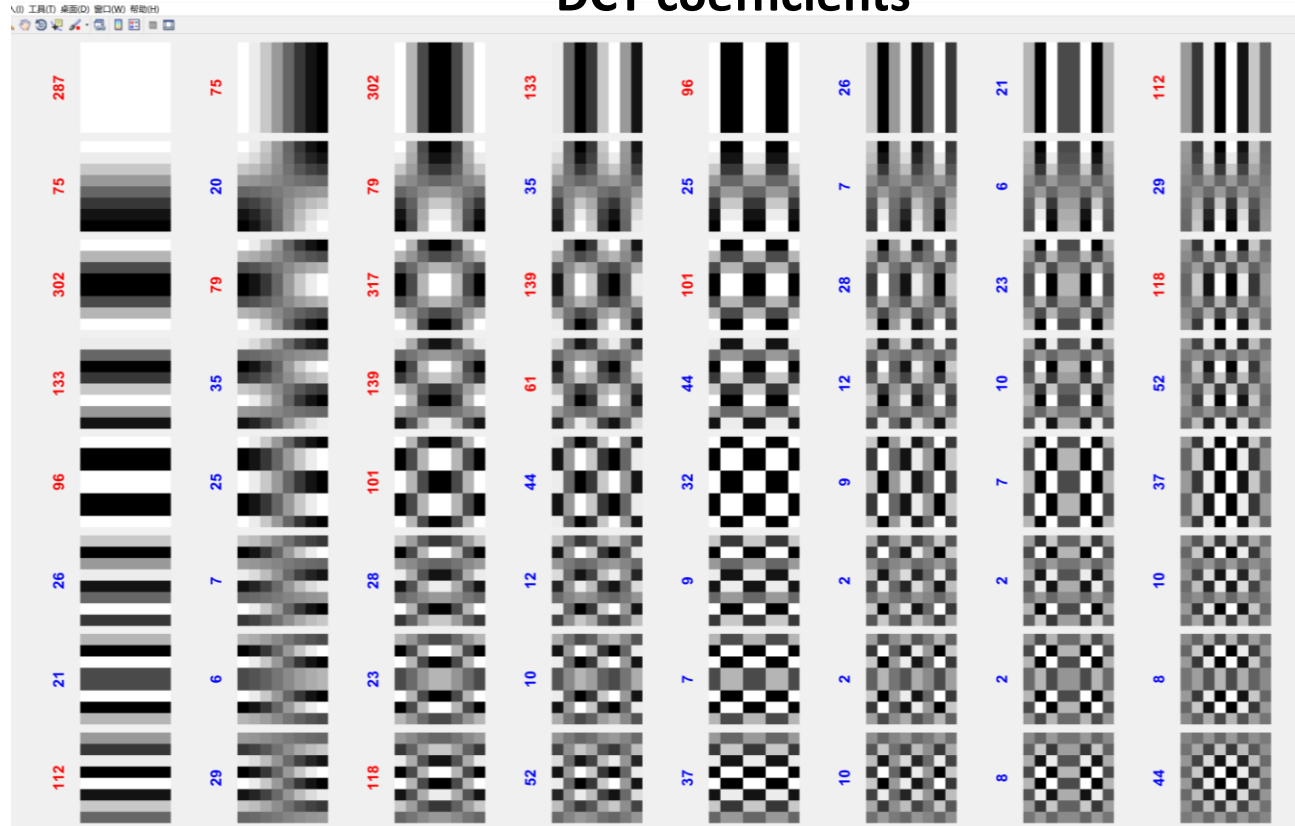


DCT example

Input image



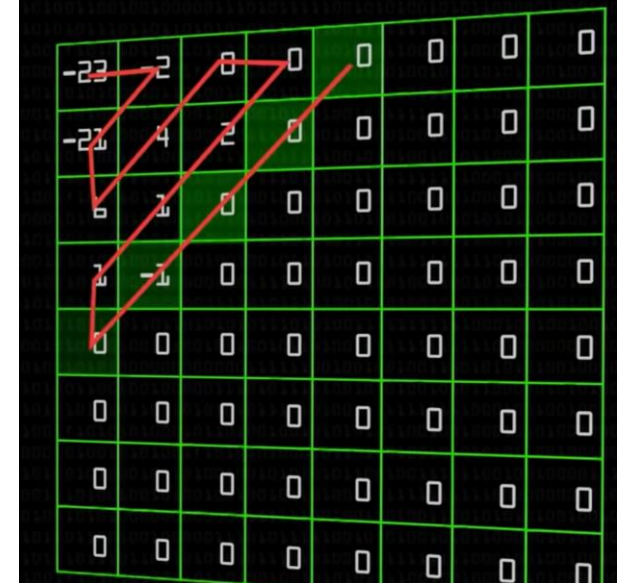
DCT coefficients



Good things about DCT

□ Positive:

- Transform is real, $C^{-1} = C^T$ (unitary transform).
- Excellent energy compaction for nature images.
- Fast transform.
- JPEG algorithms.



Walsh Transform

➤ Consist of ± 1 arranged in a checkerboard pattern.

➤ Transform:

$$W(i) = \frac{1}{N} \sum_{t=0}^{N-1} f(t) Wal(i, t)$$

$$f(t) = \sum_{i=0}^{N-1} W(i) Wal(i, t)$$

➤ Types of $Wal(i, t)$.

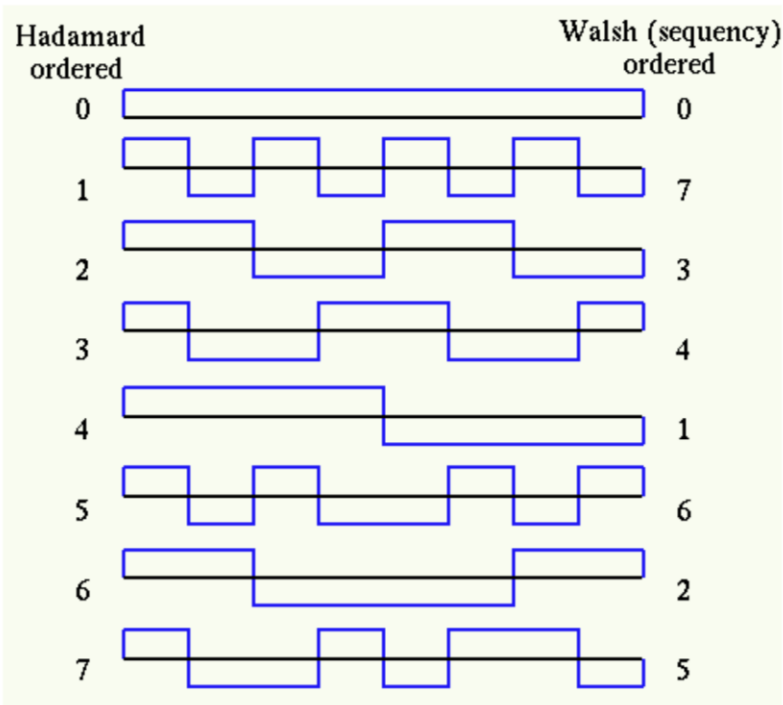
➤ Walsh Ordering (沃尔什定序)

➤ Paley Ordering (佩利定序)

➤ Hadamard Matrix Ordering (哈达玛矩阵定序)



Hadamard Matrix Ordering

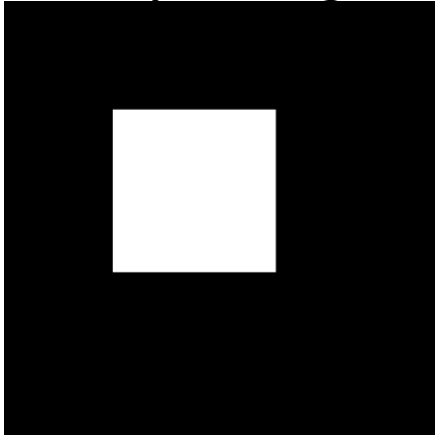


$$V_8 = \begin{pmatrix} W_4 & W_4 \\ W_4 & -W_4 \end{pmatrix} = \begin{pmatrix} 1 & 1 & 1 & 1 & 1 & 1 & 1 & 1 \\ 1 & -1 & 1 & -1 & 1 & -1 & 1 & -1 \\ 1 & 1 & -1 & -1 & 1 & 1 & -1 & -1 \\ 1 & -1 & -1 & 1 & 1 & -1 & -1 & 1 \\ 1 & 1 & 1 & 1 & -1 & -1 & -1 & -1 \\ 1 & -1 & 1 & -1 & -1 & 1 & -1 & 1 \\ 1 & 1 & -1 & -1 & -1 & -1 & 1 & 1 \\ 1 & -1 & -1 & 1 & -1 & 1 & 1 & -1 \end{pmatrix}$$

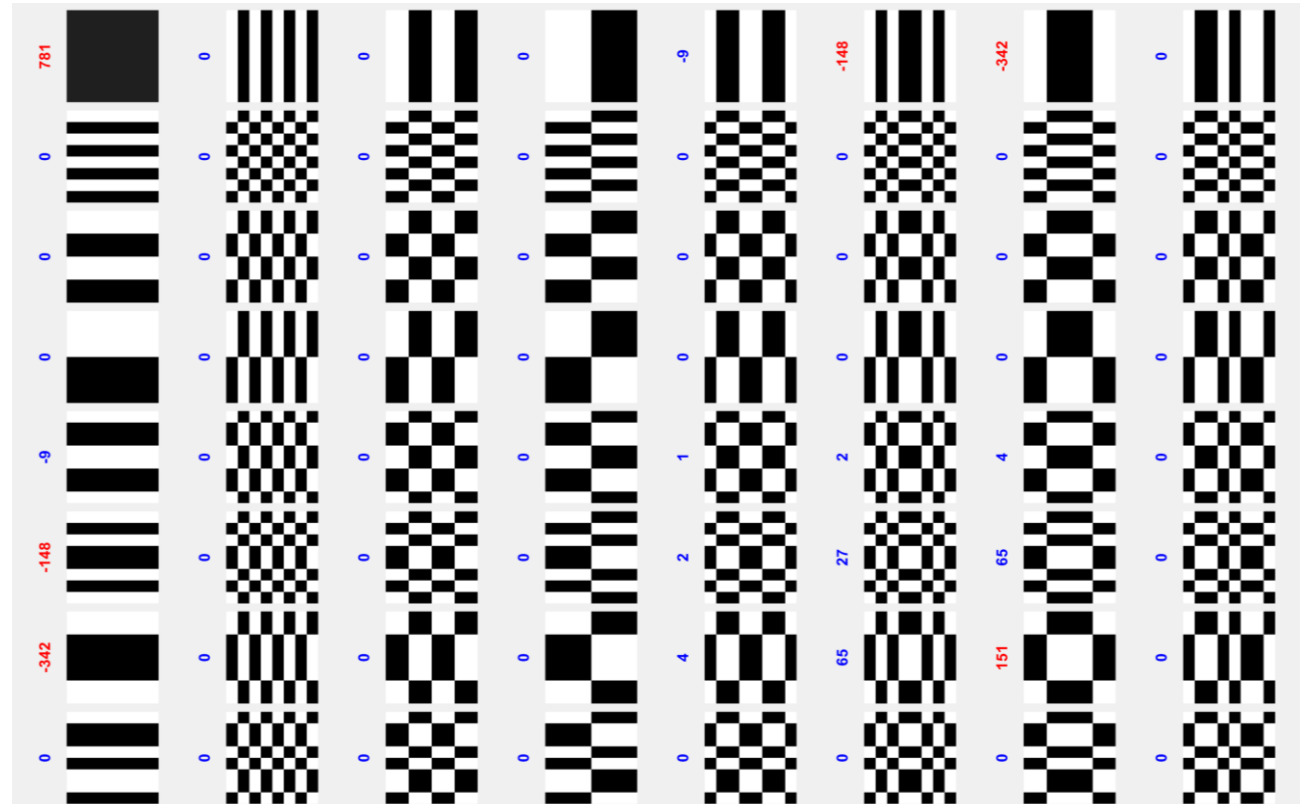


Hadarmad Transform

Input image

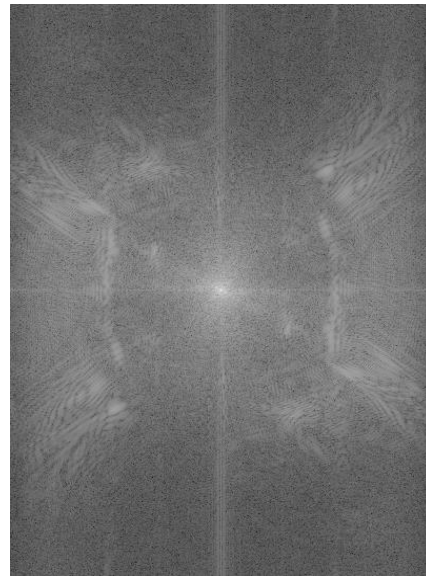


Hadarmad coefficients

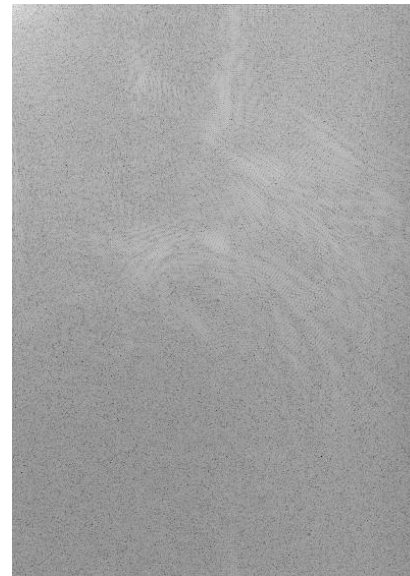


DFT, DCT, & Hadamard

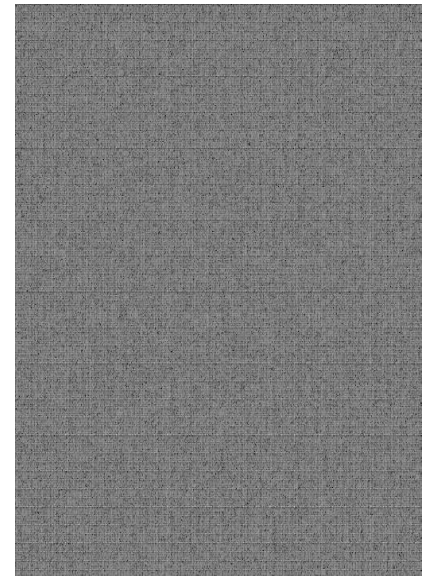
DFT



DCT



Hadamard



Any connection
between DFT and DCT?



Take home message

- ❑ The key idea for unitary transform is to find a proper basis for data decomposition.
- ❑ DCT provides better frequency consistency than DFT.
- ❑ Hadamard transform is able to present a simple image with simple coefficients. But can not keep energy compact for image full of details.



Wavelet transform Outline

□ Discrete Wavelet Transform (DWT) (小波变换)

- An example for 1D-DWT
- Generalization of 1D-DWT
- 2D-DWT



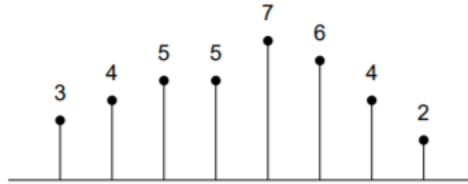
Discrete Wavelet Transform (DWT)

- ❑ Based on small waves called Wavelets-1) limited; 2) oscillation.
- ❑ Key idea: Translation & Scaling.
- ❑ Localized both time/space and frequency.
- ❑ Efficient for noise reduction and image compression.
- ❑ Two types of DWT one for image processing (easy invertible) and one for signal processing (invertible but computational expensive).

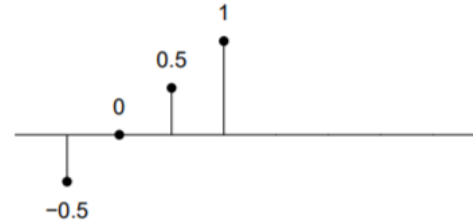
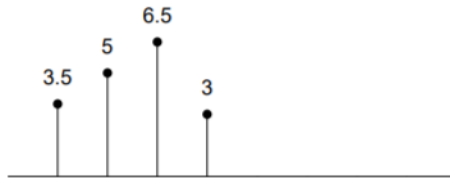


A simplest example

- We can decompose an eight-point signal $x(n]$:



into two four-point signals:



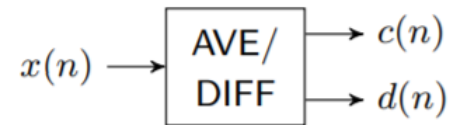
$$c(n) = 0.5 x(2n) + 0.5 x(2n + 1)$$

$$d(n) = 0.5 x(2n) - 0.5 x(2n + 1)$$



A simplest example

- The above process can be represented by a block diagram:



It is clear that this decomposition can be easily reversed:

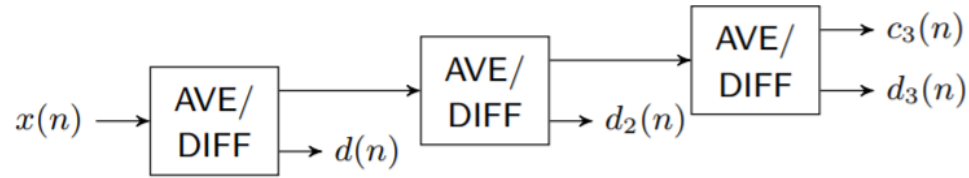
$$y(2n) = c(n) + d(n)$$
$$y(2n + 1) = c(n) - d(n)$$

Which is also represented by a block diagram:



A simplest example

□ When we repeat the simple AVE/DIFF signal decomposition:



$$x(n) = [x_1, x_2, x_3, x_4, x_5, x_6, x_7, x_8] = [3, 4, 5, 5, 7, 6, 4, 2]$$

Level 1

$$c_1 = \frac{1}{2} [x_1 + x_2, x_3 + x_4, x_5 + x_6, x_7 + x_8]$$

$$d = d_1 = \frac{1}{2} [x_1 - x_2, x_3 - x_4, x_5 - x_6, x_7 - x_8]$$

Level 2

$$c_2 = \frac{1}{4} [x_1 + x_2 + x_3 + x_4, x_5 + x_6 + x_7 + x_8]$$

$$d_2 = \frac{1}{4} [x_1 - x_3 + x_2 - x_4, x_5 - x_7 + x_6 - x_8]$$

Level 3

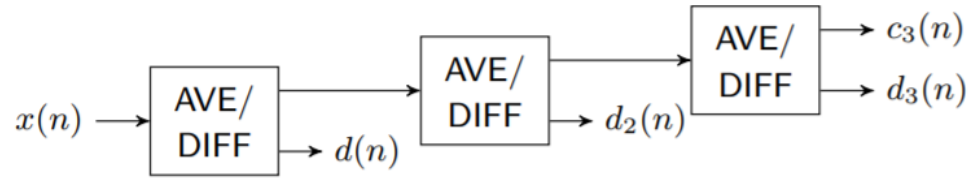
$$c_3 = \frac{1}{8} [x_1 + x_2 + x_3 + x_4 + x_5 + x_6 + x_7 + x_8]$$

$$d_3 = \frac{1}{8} [x_1 + x_2 + x_3 + x_4 - x_5 - x_6 - x_7 - x_8]$$



A simplest example

□ When we repeat the simple AVE/DIFF signal decomposition:



The Haar wavelet representation of the eight-point signal $x[n]$ is simply the set of four output signals produced by this three-level operation :

$$c_3 = [4.5]$$

$$d_3 = [-0.25]$$

$$d_2 = [-0.75, \quad 1.75]$$

$$d = [-0.5, \quad 0, \quad 0.5, \quad 1]$$

$$\mathbf{H}_8 = \frac{1}{\sqrt{8}} \begin{bmatrix} 1 & 1 & 1 & 1 & 1 & 1 & 1 & 1 \\ 1 & 1 & 1 & 1 & -1 & -1 & -1 & -1 \\ \sqrt{2} & \sqrt{2} & -\sqrt{2} & -\sqrt{2} & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & \sqrt{2} & \sqrt{2} & -\sqrt{2} & -\sqrt{2} \\ 2 & -2 & 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 2 & -2 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 2 & -2 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 & 2 & -2 \end{bmatrix} \begin{bmatrix} c_3 \\ d_3 \\ d_2(1) \\ d_2(2) \\ d(1) \\ d(2) \\ d(3) \\ d(4) \end{bmatrix}$$



Haar Transform matrix

➤ When N=2 we have:
$$\mathbf{H}_2 = \frac{1}{\sqrt{2}} \begin{bmatrix} 1 & 1 \\ 1 & -1 \end{bmatrix}$$

➤ When N=4 we have:
$$\mathbf{H}_4 = \frac{1}{2} \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & 1 & -1 & -1 \\ \sqrt{2} & -\sqrt{2} & 0 & 0 \\ 0 & 0 & \sqrt{2} & -\sqrt{2} \end{bmatrix}$$

➤ When N=8 we have:
$$\mathbf{H}_8 = \frac{1}{\sqrt{8}} \begin{bmatrix} 1 & 1 & 1 & 1 & 1 & 1 & 1 & 1 \\ 1 & 1 & 1 & 1 & -1 & -1 & -1 & -1 \\ \sqrt{2} & \sqrt{2} & -\sqrt{2} & -\sqrt{2} & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & \sqrt{2} & \sqrt{2} & -\sqrt{2} & -\sqrt{2} \\ 2 & -2 & 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 2 & -2 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 2 & -2 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 & 2 & -2 \end{bmatrix}$$



Haar Transform matrix

- The family of N Haar functions $h_u(x)$, ($u = 0, \dots, N - 1$) are defined on the interval $0 \leq x \leq 1$. The shape of the specific function $h_u(x)$ of a given index u depends on two parameters p and q :

$$u = 2^p + q$$

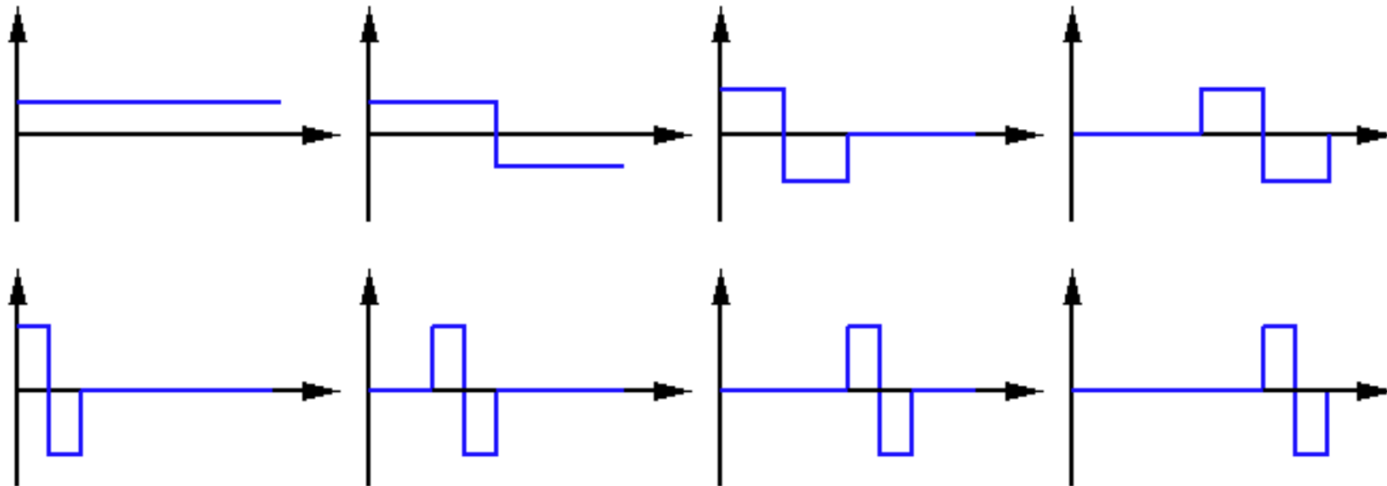
u	p	q
1	0	0
2	1	0
3	1	1

- The Haar basis functions are defined by:

$$h_u(x) = \begin{cases} 1 & u = 0 \text{ and } 0 \leq x < 1 \\ 2^{p/2} & u > 0 \text{ and } q/2^p \leq x < (q + 0.5)/2^p \\ -2^{p/2} & u > 0 \text{ and } (q + 0.5)/2^p \leq x < (q + 1)/2^p \\ 0 & \text{otherwise} \end{cases}$$



Haar Transform matrix



$$H_8 = \frac{1}{\sqrt{8}} \begin{bmatrix} 1 & 1 & 1 & 1 & 1 & 1 & 1 & 1 \\ 1 & 1 & 1 & 1 & -1 & -1 & -1 & -1 \\ \sqrt{2} & \sqrt{2} & -\sqrt{2} & -\sqrt{2} & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & \sqrt{2} & \sqrt{2} & -\sqrt{2} & -\sqrt{2} \\ 2 & -2 & 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 2 & -2 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 2 & -2 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 & 2 & -2 \end{bmatrix}$$



Generalization of 1D-DWT

➤ Discrete Wavelet Transform (DWT):

$$W_{\varphi}(j_0, k) = \frac{1}{\sqrt{M}} \sum_n f(n) \varphi_{j_0, k}(n)$$

$$W_{\psi}(j, k) = \frac{1}{\sqrt{M}} \sum_n f(n) \psi_{j, k}(n) \quad j \geq j_0$$

➤ Inverse Discrete Wavelet Transform (IDWT):

$$f(n) = \frac{1}{\sqrt{M}} \sum_k W_{\varphi}(j_0, k) \varphi_{j_0, k}(n) + \frac{1}{\sqrt{M}} \sum_{j=j_0}^{\infty} \sum_k W_{\psi}(j, k) \psi_{j, k}(n)$$

Where

$\varphi_{j_0, k}(n)$: scaling function (尺度函数)

$\psi_{j, k}(n)$: Wavelet (小波)

$W_{\varphi}(j_0, k)$: Approximation coefficients (近似系数) $W_{\psi}(j, k)$: detail coefficients (细节系数)



2D-DWT

➤ **Define 2D wavelet function: Directionally sensitive wavelet**

$$\psi^H(x, y) = \psi(x)\varphi(y) \quad \psi^V(x, y) = \varphi(x)\psi(y) \quad \psi^D(x, y) = \psi(x)\psi(y)$$

➤ **2D-DWT**

$$W_\varphi(j_0, m, n) = \frac{1}{\sqrt{MN}} \sum_{x=0}^{M-1} \sum_{y=0}^{N-1} f(x, y) \varphi_{j_0, m, n}(x, y)$$

$$W_\psi(j, m, n) = \frac{1}{\sqrt{MN}} \sum_{x=0}^{M-1} \sum_{y=0}^{N-1} f(x, y) \psi_{j, m, n}^i(x, y) \quad i = \{H, V, D\}$$

➤ **2D-IDWT**

$$f(x, y) = \frac{1}{\sqrt{MN}} \sum_m \sum_n W_\varphi(j_0, m, n) \varphi_{j_0, m, n}(x, y) \\ + \frac{1}{\sqrt{MN}} \sum_{i=\{H, V, D\}} \sum_{j=j_0}^{\infty} \sum_{x=0}^{M-1} \sum_{y=0}^{N-1} W_\psi(j, m, n) \psi_{j, m, n}^i(x, y)$$



Haar Transform matrix

Input: image size 8X8 I_{in}

Generate a Haar matrix of 8X8 as shown right

Then clip it into 4 part:

$H_{L1}(dim = 4 * 8); H_{L2}(dim = 2 * 8);$

$H_{L3}(dim = 1 * 8); L_{L3}(dim = 1 * 8);$

$$H_8 = \frac{1}{\sqrt{8}} \begin{bmatrix} 1 & 1 & 1 & 1 & 1 & 1 & 1 & 1 \\ 1 & 1 & 1 & 1 & -1 & -1 & -1 & -1 \\ \sqrt{2} & \sqrt{2} & -\sqrt{2} & -\sqrt{2} & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & \sqrt{2} & \sqrt{2} & -\sqrt{2} & -\sqrt{2} \\ 2 & -2 & 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 2 & -2 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 2 & -2 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 & 2 & -2 \end{bmatrix} \begin{matrix} H_{L3} \\ H_{L2} \\ H_{L1} \end{matrix}$$

For computing **level 1** components:

LL_1 is down-sample of I_{in} on both X and Y direction	$HL_1 = H_{L1} * I_{in}$ + downsample on Y direction
$LH_1 = I_{in} * H_{L1}$ + downsample on X direction	$HH_1 = H_{L1} * I_{in} * H_{L1}$

For computing **level 2** components:

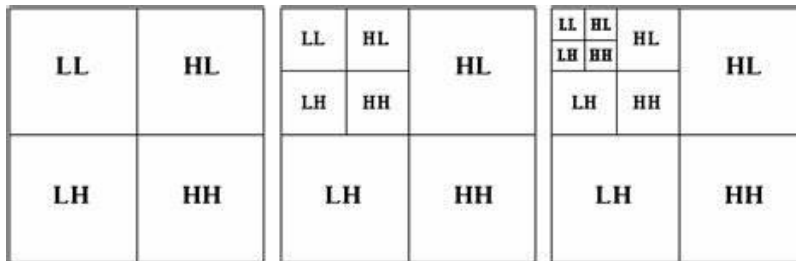
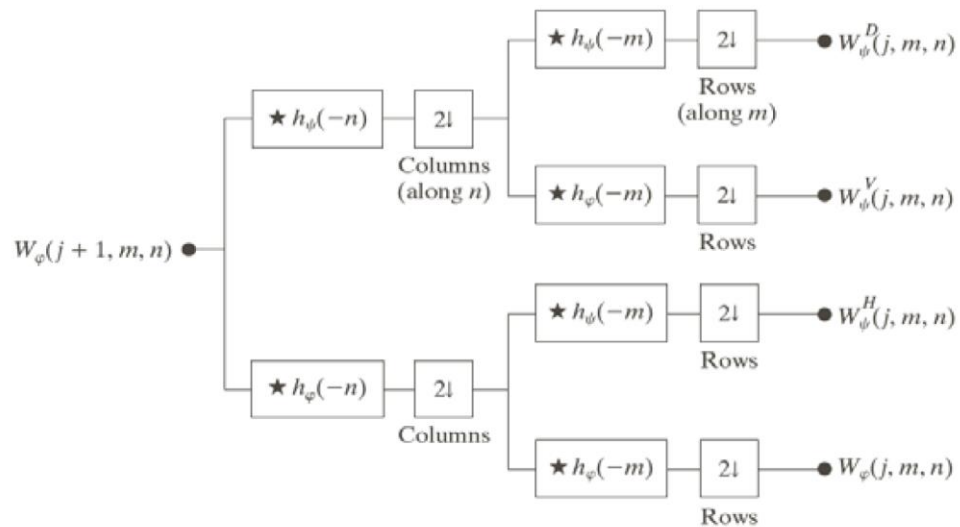
LL_2 is down-sample of LL_1 on both X and Y direction	$HL_2 = H_{L2} * I_{in}$ + downsample twice on Y direction
$LH_2 = I_{in} * H_{L2}$ + downsample twice on X direction	$HH_2 = H_{L2} * I_{in} * H_{L2}$

For computing **level 3** components:

LL_3 is down-sample of LL_2 on both X and Y direction	$HL_3 = H_{L3} * I_{in}$ + downsample 3 times on Y direction
$LH_3 = I_{in} * H_{L3}$ + downsample 3 times on X direction	$HH_3 = H_{L3} * I_{in} * H_{L3}$



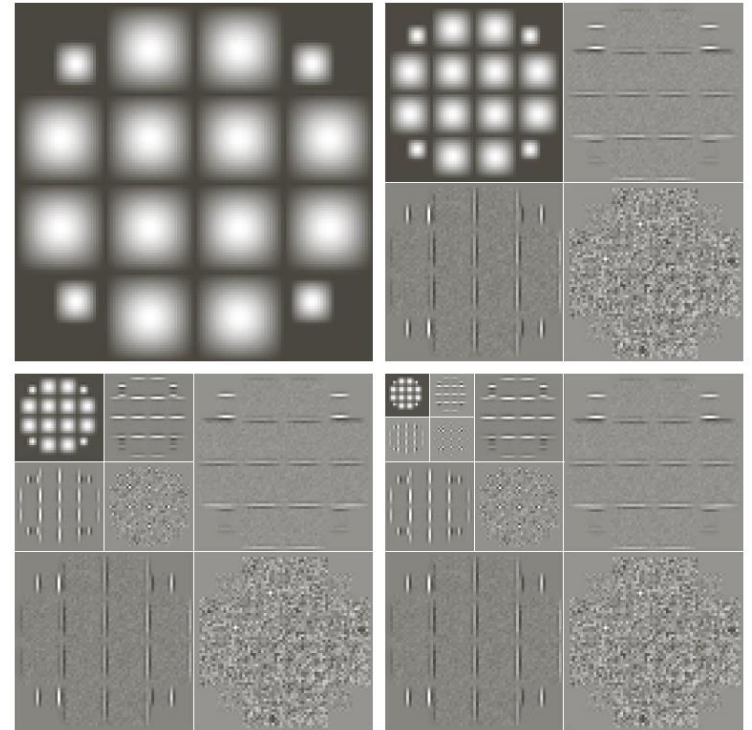
2D-DWT



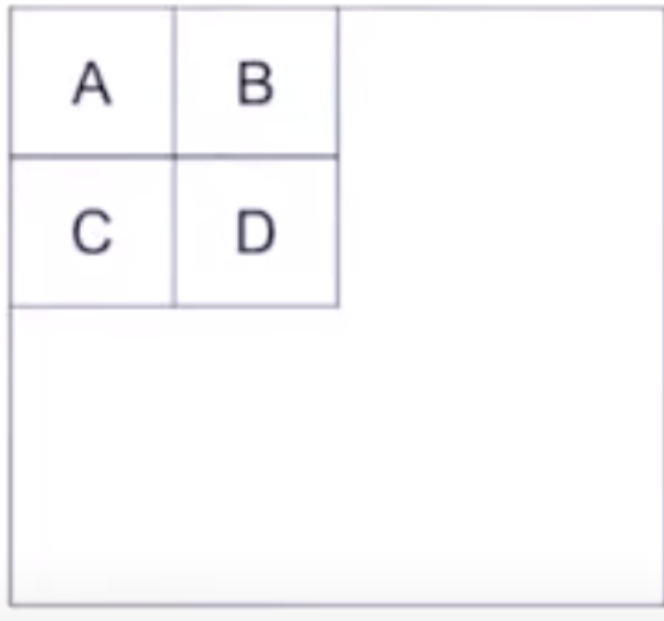
(a) Single Level Decomposition

(b) Two Level Decomposition

(c) Three Level Decomposition



2D Haar Transform



2D Haar Transform

$$A = \begin{pmatrix} 88 & 88 & 89 & 90 & 92 & 94 & 96 & 97 \\ 90 & 90 & 91 & 92 & 93 & 95 & 97 & 97 \\ 92 & 92 & 93 & 94 & 95 & 96 & 97 & 97 \\ 93 & 93 & 94 & 95 & 96 & 96 & 96 & 96 \\ 92 & 93 & 95 & 96 & 96 & 96 & 96 & 95 \\ 92 & 94 & 96 & 98 & 99 & 99 & 98 & 97 \\ 94 & 96 & 99 & 101 & 103 & 103 & 102 & 101 \\ 95 & 97 & 101 & 104 & 106 & 106 & 105 & 105 \end{pmatrix}$$

3 level Haar Transform for the first row

$$r_1 = (88 \quad 88 \quad 89 \quad 90 \quad 92 \quad 94 \quad 96 \quad 97)$$

Group r_1 in pair $[88, 88], [89, 90], [92, 94], [96, 97]$

$$r_1 h_1 = (\underline{88 \quad 89.5 \quad 93 \quad 96.5} \quad \underline{0 \quad -0.5 \quad -1 \quad -0.5})$$

Approximation coefficients **Detail coefficients**

Group the first 4 columns in pair $[88, 89.5], [93, 96.5]$

$$r_1 h_1 h_2 = (88.75 \quad 94.75 \quad -0.75 \quad -1.75 \quad 0 \quad -0.5 \quad -1 \quad -0.5)$$

Group the first 2 columns in pair $[88, 94.75]$

$$r_1 h_1 h_2 h_3 = (91.75 \quad -3 \quad -0.75 \quad -1.75 \quad 0 \quad -0.5 \quad -1 \quad -0.5)$$



2D Haar Transform

Repeat the same processing for all the columns and for the rows of the resulting matrix, we get

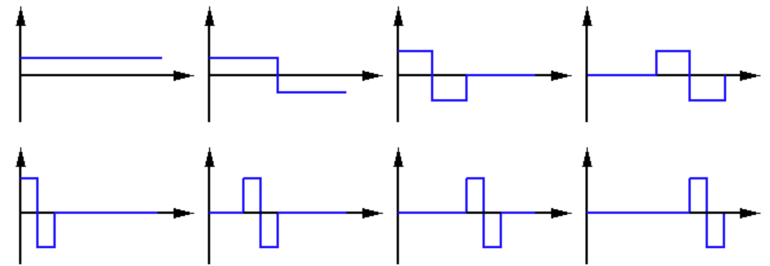
$$\begin{pmatrix} 96 & -2.0312 & -1.5312 & -0.2188 & -0.4375 & -0.75 & -0.3125 & 0.125 \\ -2.4375 & -0.0312 & 0.7812 & -0.7812 & 0.4375 & 0.25 & -0.3125 & -0.25 \\ -1.125 & -0.625 & 0 & -0.625 & 0 & 0 & -0.375 & -0.125 \\ -2.6875 & 0.75 & 0.5625 & -0.0625 & 0.125 & 0.25 & 0 & 0.125 \\ -0.6875 & -0.3125 & 0 & -0.125 & 0 & 0 & 0 & -0.25 \\ -0.1875 & -0.3125 & 0 & -0.375 & 0 & 0 & -0.25 & 0 \\ -0.875 & 0.375 & 0.25 & -0.25 & 0.25 & 0.25 & 0 & 0 \\ -1.25 & 0.375 & 0.375 & 0.125 & 0 & 0.25 & 0 & 0.25 \end{pmatrix}$$



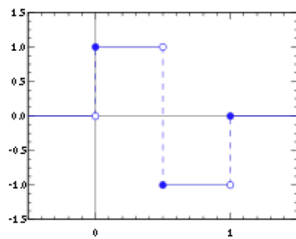
Mother Wavelet (母小波)

➤ Mother Wavelet should satisfy:

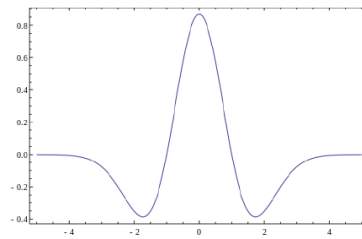
- $\int_{-\infty}^{\infty} |\psi(t)|^2 dt = 1$
- $\int_{-\infty}^{\infty} |\psi(t)| dt < \infty$
- $\int_{-\infty}^{\infty} \psi(t) dt = 0$



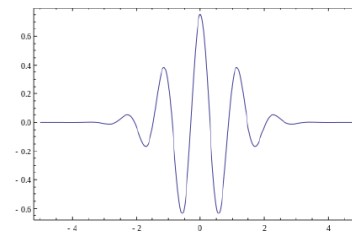
Haar



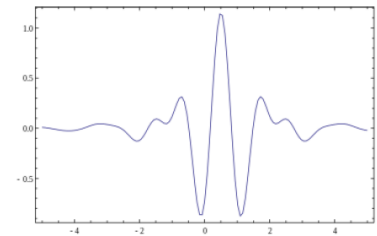
Mexican Hat



Morlet



Meyer



Take home message

- ❑ Based on small waves called Wavelets-1) limited; 2) oscillation.
- ❑ Key idea: Translation & Scaling.
- ❑ Localized both time/space and frequency.
- ❑ Efficient for noise reduction and image compression.
- ❑ JPEG2000, FBI finger printing databased.

