

### Homework 3

Released: September 20th

Due: 11:59pm September 28th

#### 1. Lossy Transmission Lines

You are given a length of transmission line with the following properties:  $u_p = 0.66c$  and  $Z_0 = 50\Omega$ .

- a) You measure the amplitude of the voltage at two positions on the transmission line and obtain the following measurements:

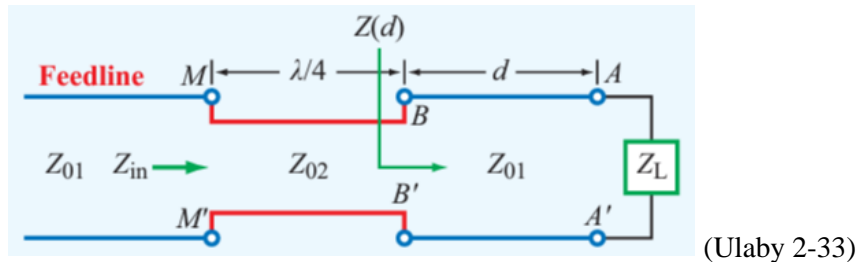
Distance (m)	Voltage (V)
0 (source)	5.00
900	4.97

What is the value of  $\alpha$ , the attenuation factor for this line?

- b) Assuming this is a low-loss transmission line, what is the value of  $r$ , the resistance per unit length of the line?
- c) Knowing both  $u_p$  and  $Z_0$  (assumed to be purely real in this case), you can calculate  $c$  (capacitance per unit length) and  $l$  (inductance per unit length) for the transmission line. What value of  $g$  (conductance per unit length) would you need to make this low-loss transmission line dispersionless?
- d) What is the new value for  $\alpha$  on this dispersionless transmission line?
- e) If you were to measure the voltage at  $z = 900\text{m}$  on this new, dispersionless transmission line, what would it be? Has the addition of conductance noticeably affected the loss of this transmission line?

## 2. Impedance Matching with Smith Charts Part I

Given the circuit below, you are to use a transmission line of length  $d$  with characteristic impedance  $Z_{01} = 50\Omega$  and a quarter-wave transformer of characteristic impedance  $Z_{02}$  to match the feedline impedance (which is also  $Z_{01} = 50\Omega$ ) to the load impedance  $Z_L = (40 + j35)\Omega$ .

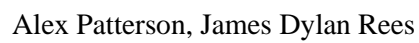


*Finding the Length  $L$  of the Transmission Line Needed to Make  $Z_{in,0}$  Real*

- Calculate the normalized impedance  $z$ .
- Using the Smith chart on the next page, plot  $z$  and the VSWR circle.
- Using the Smith chart, determine the magnitude and phase of  $\Gamma$  and the SWR. Explain how you found each of these values.
- What length of transmission line (in terms of  $\lambda$ ) is needed to transform  $z$  into a purely real impedance and what is the resulting impedance?
- The VSWR circle you drew in part b) is a curve of constant  $|\Gamma_L|$ . Why must we remain on a curve of constant  $|\Gamma_L|$  on the Smith chart when we are transforming impedances using transmission lines? (Hint: what parameters determine the magnitude of  $\Gamma_L$ ?)
- When you determined the length of the transmission line needed in part d), you traveled along the VSWR circle, on which the magnitude of  $\Gamma$  is constant, but the phase of  $\Gamma$  is changing. What does a change in the phase of  $\Gamma$  on a Smith chart physically correspond to? (Hint:  $\Gamma$  on the Smith chart can be thought of as the phase-shifted reflection coefficient  $\Gamma_d$ ).
- What do the impedances on the VSWR circle on the Smith chart represent? (Hint: recall the equation for wave impedance on the transmission line  $Z(d) = \frac{1 + \Gamma_L e^{-j2\beta d}}{1 - \Gamma_L e^{-j2\beta d}}$ , where  $\theta_r$  is the phase of  $\Gamma_L$  at the load and  $d$  is the distance measured from the load in the direction of the generator).

*Finding the Characteristic Impedance  $Z_{02}$  Needed to Match  $Z_{01}$  to  $Z_L$*

- Determine the characteristic impedance  $Z_{02}$  of the quarter-wave transformer required to match the generator resistance to the transmission line you determined the length of in part d).
- If the frequency of the signal is  $f = 10\text{MHz}$  and the velocity factor of both transmission lines is 0.66, how long is your transmission line from part d)? How long is your quarter wave transformer from part g)?



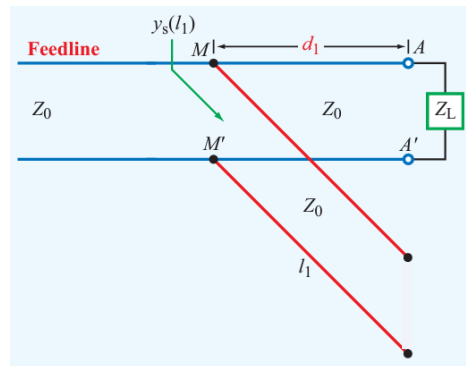
### 3. Impedance Matching with Smith Charts Part II

In this question, you will verify your results from Problem 2 using the equations for the impedance of lossless transmission lines, instead of the Smith Chart.

- Using the equation  $\Gamma_L = \frac{z-1}{z+1}$ , where  $z$  is the normalized impedance from 2a, calculate the magnitude and phase of the reflection coefficient at the load. Do they match the values you determined using the Smith chart in 2c?
- Insert the transmission line length  $L$  you determined in 2d into the equation for input impedance:  $Z_{in}(L) = Z_0 \frac{z + j \tan(\beta L)}{1 + jz \tan(\beta L)}$ , where  $Z_0 = 50\Omega$  as in problem 2. Does  $Z_{in}$  agree with the impedance you found on the Smith chart in 2d?
- Assuming that your quarter-wave transformer was appropriately designed in 2h, what impedance does the generator circuit “see” looking to the right from  $R_g$ , into the transmission line?
- Even though you’ve matched  $Z_L$  to  $R_g$  to achieve maximum power transfer to the load via a quarter-wave transformer (no reflections at the input to the transmission line in steady-state), the reflection coefficient at the load  $\Gamma_L$  (and at the generator  $\Gamma_g$ ) is non-zero, meaning that reflections still occur on the  $\lambda/4$  line. How is it possible that maximum power is still being transferred despite  $\Gamma_L \neq 0$ ?

#### 4. Impedance Matching with Stubs (also using Smith Charts)

You are to match a transmission line with a characteristic impedance of  $Z_0 = 100\Omega$  to a load impedance  $Z_L = (50 - j25)\Omega$  using an open-circuited stub, as shown below:



(Adapted from Ulaby 2-33e)

- Find the normalized impedance of the load. Plot it on the Smith chart.
- Using the Smith chart, find and plot the normalized admittance  $y$ .
- What length of transmission line  $d_1$  is required to transform the real part of  $y$  to be 1? What is  $y$  at this point? What is the significance of  $\text{Re}\{y\} = 1$ ?
- Plot the location of the open-circuit admittance and the susceptance that we will need for matching the load on the Smith chart. What length of stub  $l_1$  do we need in order to provide this susceptance for matching?
- Now we need to verify that our procedure on the Smith chart has actually matched the load impedance to the transmission line characteristic impedance of  $100\Omega$ . The input impedance seen by the transmission line is  $Z_{in} = \frac{Z_{stub}^{oc} \cdot Z_{d1}}{Z_{stub}^{oc} + Z_{d1}}$ , where  $Z_{stub}^{oc}$  is the input impedance of the open-circuited stub and  $Z_{d1}$  is the input impedance of the transmission line of length  $d_1$ .



