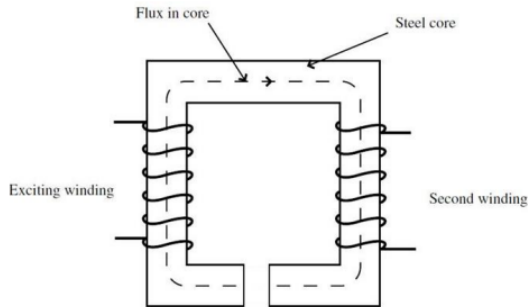
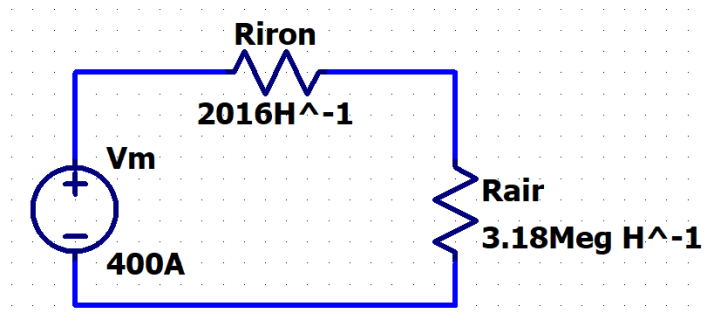


1. Transformer on an Iron Core with an Air Gap [12 Points]

The transformer below consists of two coils on an iron core with an air gap. The iron core portion has a perimeter of 475cm and the air gap is 25cm wide, giving a total perimeter through the iron core region and air gap of 500cm. Additionally, the iron core has a square cross-sectional profile with side lengths of 25cm. The permeability of the iron core region is  $3000\mu_0$  and the permeability of the air gap region is  $\mu_0$ . The exciting winding has 1000 turns and the second winding has 3000 turns.



- a. Draw a magnetic circuit to represent the structure when the current through the coil is 400mA DC and calculate and label the following:



- i. the MMF

$$\text{mmf} = N \cdot I = 1000 \cdot 400\text{mA} = 400\text{A}$$

- ii. the reluctance of the iron core and air gap region

$$R = l / (\mu \cdot A)$$

$$R_{\text{iron}} = 0.475 / (3000 \mu_0 \cdot .25^2) = 2016 \text{ H}^{-1}$$

$$R_{\text{air}} = 0.25 / (\mu_0 \cdot .25^2) = 3.18 \times 10^6 \text{ H}^{-1}$$

- iii. the total magnetic flux through the core. (For this part of the problem, ignore the second winding.)

$$\phi = V_m / (R_{\text{iron}} + R_{\text{air}}) = 400\text{A} / (2016 + 3.18 \times 10^6 \text{ H}^{-1}) = 1.26 \times 10^{-4} \text{ Wb}$$

- b. Calculate the magnitude of the H field in the air gap

$$H = B / \mu_0 = \phi_{\text{im}} / \mu_0 A = 1.26 \times 10^{-4} / (4\pi \times 10^{-9} \text{ H/m} \cdot .25\text{m}^2) = 1604 \text{ A/m}$$

- c. Calculate the magnetic force felt by the two pieces of the core on either side of the air gap. Is this an attractive or repulsive force?

$$F_m = B_{air}^2 / 2\mu_0 = 25.9 \text{ N}$$

Attractive,  $F = -\nabla(B^2 / 2\mu_0)$

- d. Suppose that the 400mA DC current is replaced with a 400mA 100Hz AC current. What is the resulting emf across the secondary coil?

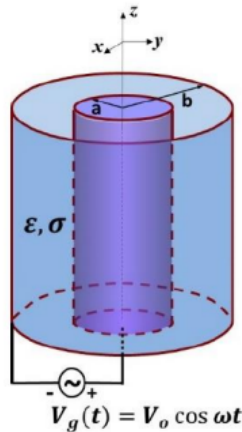
$$\text{emf} = -d/dt (N_2 \phi(t)) = -3000 * 1.26 \times 10^{-4} d/dt \sin(200\pi t) = -237.5 \cos(200\pi t) \text{ V}$$

- e. Is it preferable to use a hard or a soft ferromagnetic material for the core of this transformer? Why?

Soft, less losses to hysteresis

2. Displacement Current and the Quasi-static Approximation [16 Points]

For a coax-cable capacitor, the volume between the cylindrical copper conductors is filled with a lossy dielectric with permittivity  $\epsilon_r$  and conductivity  $\sigma$ . The radius of the inner conductor is  $a$  and the effective inner radius for the outer conductor is  $b$ . The cable length  $l$  is much shorter than the wavelength, but  $l \gg b$ . The voltage applied across the coax-cable capacitor is  $V(t) = V_0 \cos(\omega t)$  V.



- a. Determine the displacement field  $\vec{D}$  between the conductors, the displacement current  $I_d$  and the conduction current  $I_c$  passing through the capacitor. What is the phase angle between  $I_c$  and  $I_d$ ? Which current leads (or which one lags)?  
Hint: you can use either time domain or phasor domain; you need to integrate the current passing through a cylindrical surface.

$$D = \epsilon E$$

$$\oint \vec{D} \cdot d\vec{s} = Q_{\text{inner}}$$

$$qE \oint d\vec{s} = Q$$

$$E = Q / (\oint d\vec{s} \epsilon)$$

$$V(r) - V(b) = V(r) = - \int_b^r \frac{Q}{(2\pi r \epsilon)} dr = - \frac{Q}{(2\pi \epsilon)} \ln(r/b)$$

$$V(a) = V_0 = \frac{Q}{(2\pi \epsilon)} \ln(a/b)$$

$$V_{\text{inner}} = - \frac{2\pi \epsilon V_0 \cos(\omega t)}{\ln(a/b)}$$

$$D = \epsilon E = \epsilon \frac{V_0 \cos(\omega t)}{(r \ln(a/b))}$$

$$I_d = \frac{d}{dt} \oint \vec{D} \cdot d\vec{s} = j \oint \vec{D} \cdot d\vec{s} = j \frac{2\pi \epsilon V_0 \sin(\omega t)}{\ln(a/b)}$$

$$I_c = \oint \vec{J} \cdot d\vec{s} = \int_0^l \int_0^{2\pi} \frac{V_0 \cos(\omega t)}{(r \ln(a/b))} \sigma r d\phi dz = 2\pi \sigma V_0 \cos(\omega t) / \ln(a/b)$$

- b. Evaluate  $I_d$ ,  $I_c$  and the ratio of their amplitudes when  $\epsilon_r = 27$ ,  $\sigma = 2 \times 10^{-8}$  S/m,  $a = 0.45$  mm,  $b = 1.57$  mm,  $l = 100$  m,  $V_0 = 10$  V and  $f = 1$  MHz.

$$I_d = 2\pi \epsilon \omega V_0 \sin(\omega t) / \ln(a/b) = 7.54 \cos(2\pi \text{ Meg } t + \pi/2) \text{ A}$$

$$I_c = 2\pi \sigma V_0 \cos(\omega t) / \ln(a/b) = 100 \cos(\omega t) \text{ uA}$$

$$I_d / I_c = 7.54 \text{ A} / 100 \text{ uA} = 75.4 \text{ k}, \text{ displacement dominates}$$

- c. Using your answer from part b, determine at which frequency the amplitudes of  $I_d$  and  $I_c$  are equal.

$$I_d = I_c$$

$$2\pi \epsilon \omega V_0 / \ln(a/b) = 2\pi \sigma V_0 / \ln(a/b)$$

$$\omega = s$$

$$f = s/2\pi = 13.3 \text{ Hz}$$

- d. Using your knowledge from circuit theory, calculate  $I_d$  and  $I_c$ , then compare them with your results from part a. Hint: you will need the expressions for capacitance and conductance per unit length of a coaxial cable to calculate these quantities
- $$I_d = C \frac{dV}{dt} = I_2 \pi \epsilon \omega V_0 \sin(\omega t) / \ln(b/a)$$

### 3. Plane Wave in a Lossy Medium [13 Points]

Today's computer microprocessors (and many other electronic and photonic devices) are built on a silicon (Si) substrate. For pure Si,  $\epsilon_r \approx 12$ . Assume the Si substrate resistivity is  $\rho = 10 \, \Omega \cdot \text{cm}$ .

- a. Determine the frequency range in which the Si substrate can be treated as a good insulator (dielectric), and the frequency range in which the Si substrate can be treated as a good conductor.

$$\rho = 10 \, \Omega \cdot \text{cm} \cdot \frac{1 \text{ m}}{100 \text{ cm}} = 0.1 \, \Omega \cdot \text{m}$$

$$\text{Insulator: } \frac{\epsilon''}{\epsilon'} = \frac{s}{\omega \epsilon'} \ll 0.1$$

$$f \gg \frac{s}{2\pi \epsilon'} = 1.5 \, \text{THz}$$

$$\text{Insulator: } \frac{\epsilon''}{\epsilon'} \gg 100$$

$$f \ll 150 \, \text{GHz}$$

- b. If a plane wave is traveling in the Si substrate at 100 MHz, find the attenuation constant and phase constant ( $\alpha$  and  $\beta$ ), wavelength, and intrinsic impedance ( $\eta$ ).

Conductor

$$\alpha = \sqrt{\pi f \mu s} = 20\pi$$

$$\beta = \alpha = 20\pi \, \text{rad/m}$$

$$\lambda = 2\pi/\beta = 0.1 \, \text{m}$$

$$\eta = (1+j)\alpha/s = 2\sqrt{2}\pi \, e^{j\pi/4} \, \Omega$$

- c. Using your results from b, if the wave is traveling in the y-direction with an E-field amplitude of 10 V/m measured at  $y = 0$ , find E and H for the wave in the phasor domain.  $E = E\hat{x}$

$$E = 10 \, \text{V/m} \, e^{-20\pi y} e^{-j20\pi y} \hat{x}$$

$$H = kE/\eta = 5/(\pi\sqrt{2}) \, e^{-20\pi y} e^{-j(20\pi y - \pi/4)} \hat{z}$$

4. Wave Polarization [12 Points]

Determine the polarization of the following waves (i.e., linear, circular or elliptical) and their propagation direction. If a wave is linearly polarized, also determine the inclination angle. For nonlinear polarization, determine the rotation direction (LH or RH). Draw a polarization diagram for each case with a few data points to show your work

a.  $\tilde{E}(z) = (3\hat{x} - j3\hat{y})e^{j25\pi z}$  [V/m]

-z

Ex -pi/2 of Ey

Circular polarization

Left hand circular

b.  $\tilde{E}(z,t) = 2 \cos(106\pi t - 0.5z + 45^\circ) \hat{x} + \sin(106\pi t - 0.5z - 45^\circ) \hat{y}$  [V/m]

+z

Ex pi of Ey

Linear polarization

$\arctan(-1/2) = -26.6^\circ$

-26.6deg inclination linear

c.  $\tilde{E}(z) = (3e^{j\pi/6}\hat{x} - 3e^{j\pi/3}\hat{y}) e^{-j3\pi z}$  [V/m]

+z

Ex 7/6pi of Ey

Elliptical polarization

Right hand elliptical