

1) Laplace transforms/Transfer functions

Use Laplace transform tables!!!!

1.1: Find the Laplace transform of

$$f(t) = (\cos(2t) + e^{-4t}) \cdot u(t) \quad (\text{simplify into one ratio})$$

$$\frac{s}{s^2 + 4} + \frac{1}{s + 4} = \frac{(s + 4)s + (s^2 + 4)}{(s^2 + 4) \cdot (s + 4)}$$

$$\frac{2s^2 + 4s + 4}{(s^2 + 4) \cdot (s + 4)}$$

1.2: Find the poles and zeros of the following functions. Indicate any repeated poles and complex conjugate poles. Expand the transforms using partial fraction expansion.

$$1. \quad F(s) = \frac{20}{(s + 3) \cdot (s^2 + 8s + 25)}$$

$$s^2 + 8s + 25 = 0 \quad \left(\begin{array}{l} -4 + 3i \\ -4 - 3i \end{array} \right)$$

Zeros: 0
Poles: -3, -4 ± j3

$$2. \quad F(s) = \frac{2s^2 + 18s + 12}{s^4 + 9s^3 + 34s^2 + 90s + 100}$$

Zeros: double -3
Poles: -2, -5, -1+j3, -1-j3

$$F(s) = \frac{2s^2 + 18s + 12}{s^4 + 9s^3 + 34s^2 + 90s + 100} = \frac{A_1}{s + 2} + \frac{A_2}{s + 5} + \frac{A_3}{s + 1 - j3} + \frac{A_3^*}{s + 1 + j3}$$

$$s^4 + 9s^3 + 34s^2 + 90s + 100 = 0$$

$$\left(\begin{array}{l} -2 \\ -5 \\ -1 + 3i \\ -1 - 3i \end{array} \right) \quad \begin{array}{l} (s + 3)^2 \\ (s + 3)^2 \end{array}$$

$$\frac{2(s + 9s + 6)}{(s + 2)(s + 5)(s + 1 - j3)(s + 1 + j3)} \cdot (s + 2) \quad \text{for } s = -2$$

$$\frac{2[(-2)^2 + (9 \times -2) + 6]}{(-2 + 5)(-2 + 1 - 3i)(-2 + 1 + 3i)} = -0.533$$

$$2(-2 + 3)^2 = 2$$

$$2(4 + -18 + 6) = -16$$

$$\frac{2(s + 9s + 6)}{(s + 2)(s + 5)(s + 1 - j3)(s + 1 + j3)} \cdot (s + 5) \quad \text{for } s = -5$$

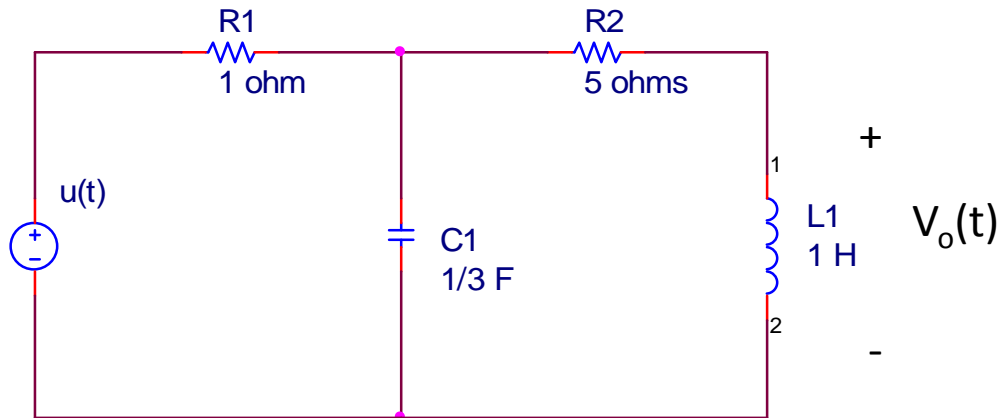
$$\frac{2[(-5)^2 + (9 \times -5) + 6]}{(-5 + 2)(-5 + 1 - 3i)(-5 + 1 + 3i)} = 0.373$$

$$\frac{2(s + 9s + 6)}{(s + 2)(s + 5)(s + 1 - j3)(s + 1 + j3)} \cdot (s + 1 - j3) \quad \text{for } s = -1 + j3$$

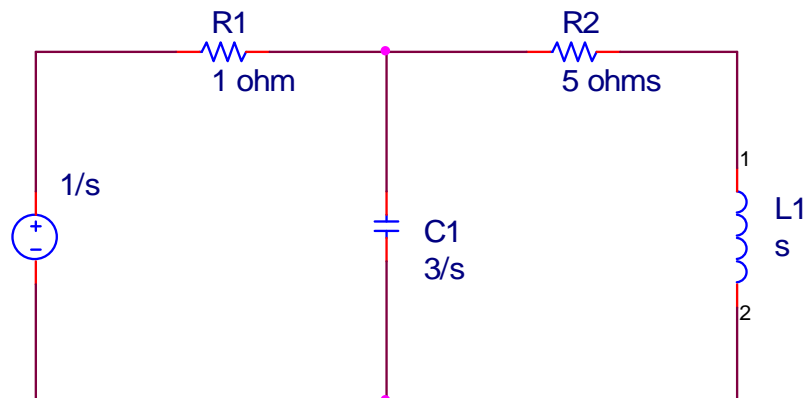
$$\frac{2[(-1 + 3i)^2 + [9(-1 + 3i)] + 6]}{(-1 + 3i + 2)(-1 + 3i + 5)(-1 + 3i + 1 + 3i)} = 0.08 - 0.493i$$

$$F(s) = \frac{2s^2 + 18s + 12}{s^4 + 9s^3 + 34s^2 + 90s + 100} = \frac{-0.533}{s + 2} + \frac{0.373}{s + 5} + \frac{0.08 - 0.493i}{s + 1 - j3} + \frac{0.08 - 0.493i}{s + 1 + j3}$$

2) Circuits and Differential Equations



2.1: Draw the s-domain equivalent circuit. Assume all initial conditions are zero and the source is an arbitrary source.



2.2 Using impedances, determine the expression for $V_o(t)$. Consider using mesh analysis then make one ratio.

Applying mesh analysis

mesh 1

$$\frac{1}{s} = \left(1 + \frac{3}{s}\right)i_1 - \frac{3}{s}i_2$$

mesh 2

$$0 = \frac{-3}{s}i_1 + \left(s + 5 + \frac{3}{s}\right)i_2$$

$$i_1 = \frac{1}{3} \cdot (s^2 + 5s + 3) \cdot i_2$$

substituting equation mesh 2 into mesh 1

$$\frac{1}{2} = \left(1 + \frac{3}{s}\right) \cdot \frac{1}{3} \cdot (s^2 + 5s + 3) \cdot i_2 - \frac{3}{s} \cdot i_2$$

multiply through by 3s

$$3 = (s^3 + 8s^2 + 18s) \cdot i_2 \quad i_2 = \frac{3}{s^3 + 8s^2 + 18s}$$

$$v_o(s) = s \cdot i_2 = \frac{3}{s^2 + 8s + 18}$$

2.3 Find $V_o(t)$ which is the $V_L(t)$ for $t > 0$ using $V_s = 1 \text{ u}(t)$.

Using Mesh analysis:

$$\frac{1}{s} = i_1 \left(1 + \frac{3}{s}\right) - i_2 \frac{3}{s}$$

$$0 = -i_1 \cdot \left(\frac{3}{s}\right) + i_2 \cdot \left(s + 5 + \frac{3}{s}\right) \quad \text{solve for } i_1 \quad i_1 = \frac{1}{3} \cdot (s^2 + 5s + 3) \cdot i_2$$

substitute in to find i_2 which can be multiplied by s to find the voltage across $L1$

$$\frac{1}{s} = \frac{1}{3} \cdot (s^2 + 5s + 3) \cdot i_2 \cdot \left(1 + \frac{3}{s}\right) - \frac{3}{s} \cdot i_2$$

multiply by 3s and solve for i_2

$$i_2 = \frac{3}{s^3 + 8s^2 + 18s}$$

$$V_L = i_2 \cdot s = \frac{3}{s^2 + 8s + 18}$$

Students will start with the end of 2.2 here...

$$s^2 + 8s + 18 = 0$$

$$\begin{pmatrix} -4 + \sqrt{2} \cdot i \\ -4 - \sqrt{2} \cdot i \end{pmatrix}$$

$$V_L = \frac{A_1}{s - (-4 - \sqrt{2}i)} + \frac{A_1^*}{s - (-4 + \sqrt{2}i)}$$

$$\sqrt{2} = 1.414$$

$$\frac{3}{(s + 4 + \sqrt{2}i) \cdot (s + 4 - \sqrt{2}i)} \cdot [s - (-4 - \sqrt{2}i)] \quad s = -4 - \sqrt{2}i$$

$$\frac{3}{(-4 - 1.414i) + 4 - 1.414i} = 1.061i$$

$$V_L = \frac{1.061i}{(s + 4 + \sqrt{2}i)} + \frac{-1.061i}{(s + 4 - \sqrt{2}i)}$$

$$V_L = 1.061i \cdot e^{(-4 - \sqrt{2}i)t} - 1.061i \cdot e^{(-4 + \sqrt{2}i)t}$$

not necessary but to check another way manipulate with laplace transform for $e^{at} \sin(bt)$

$$\frac{3}{\sqrt{2}} \cdot \frac{\sqrt{2}}{(s + 4)^2 + (\sqrt{2})^2} = \frac{3}{\sqrt{2}} \cdot e^{-4t} \cdot \sin(\sqrt{2}t)$$

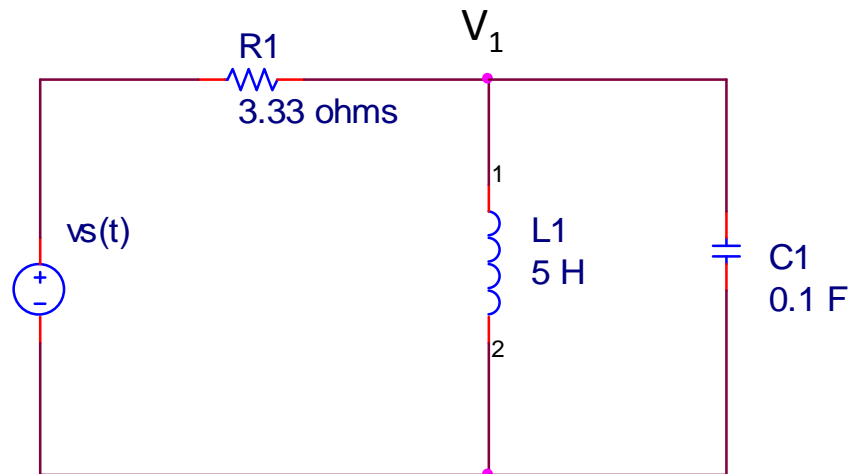
$$-2i \cdot 1.061i \cdot e^{-4t} \cdot \left[\frac{(-e)^{-\sqrt{2}it} + e^{\sqrt{2}it}}{2i} \right] \quad \text{Euler's formula} \quad \sin x = \frac{e^{ix} - e^{-ix}}{2i}$$

this should match with highlighted answer as well with some algebraic manipulation

$$2.122e^{-4t} \cdot \sin(\sqrt{2}t)$$

$$\frac{3}{\sqrt{2}} = 2.121$$

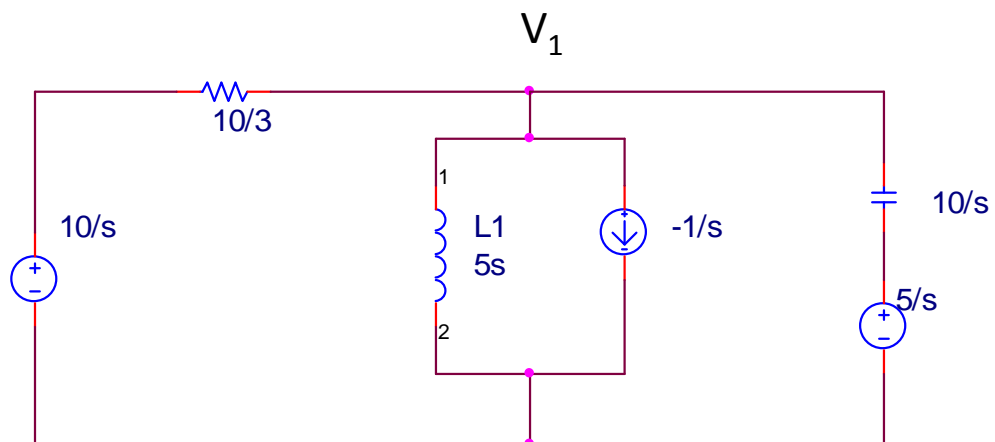
3) RLC and initial conditions



$$v_s(t) = 10u(t)$$

AND assume that -1A flows through the inductor and +5V is across the capacitor at $t=0$i.e. $v_c(0)=5$ and $i_L(0)=-1$

3.1: Draw the s-domain equivalent with initial conditions.



3.2: Find the value of the voltage across the capacitor, $v_c(t)$, using nodal analysis (at node V1) and laplace.

$$\frac{V_1 - \frac{10}{s}}{\frac{10}{3}} + \frac{V_1 - 0}{5s} - \frac{1}{s} + \frac{V_1 - \frac{5}{s}}{\frac{10}{s}} = 0$$

$$\frac{V_1}{\frac{10}{3}} - \frac{\frac{10}{s}}{\frac{10}{3}} + \frac{V_1}{5s} - \frac{1}{s} + \frac{V_1}{\frac{10}{s}} - \frac{\frac{5}{s}}{\frac{10}{s}} = 0$$

$$\frac{3V_1}{10} - \frac{3}{s} + \frac{V_1}{5s} - \frac{1}{s} + \frac{V_1 \cdot s}{10} - \frac{5}{10} = 0$$

group together V1 on left and other factors on right

$$V_1 \cdot \left(\frac{3}{10} + \frac{1}{5s} + \frac{s}{10} \right) = \frac{3}{s} + \frac{1}{s} + \frac{1}{2}$$

$$\frac{1}{10} \cdot \left(s + 3 + \frac{2}{s} \right) \cdot V_1 = \frac{3}{s} + \frac{1}{s} + 0.5$$

Take out 1/10 from right

$$\left(s + 3 + \frac{2}{s} \right) \cdot V_1 = \frac{30}{s} + \frac{10}{s} + 5$$

Multiply thorough by 10

$$(s^2 + 3s + 2) \cdot V_1 = 40 + 5s$$

multiply by s on both sides

$$V_1 = \frac{40 + 5s}{(s + 1)(s + 2)}$$

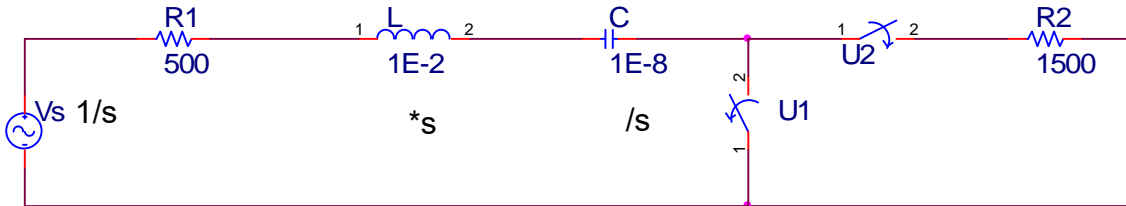
using partial fraction expansion

$$\frac{35}{s + 1} - \frac{30}{s + 2}$$

$$v_1(t) = v_c(t) = \left(35e^{-t} - 30e^{-2t} \right) \cdot u(t)$$

V

4) RLC parallel circuits



In the above circuit, the source turns on at $t=0$ with a voltage of 10V. Additionally, switch U1 is closed and switch U2 is open. At $15E-6$ s switch U1 opens and switch U2 closes. The source also turns off at $15E-6$ s.

a. Use Laplace analysis to determine the voltage across the capacitor as a function of time for $0 < t < 15E-6$ (s)

$$R_1 := 500\Omega \quad L_1 := 1 \cdot 10^{-2} \text{H} \quad C_1 := 1 \cdot 10^{-8} \text{F} \quad R_2 := 1500\Omega$$

$$V_C(s) = \frac{\frac{1}{LC}}{s^2 + \frac{R}{L} \cdot s + \frac{1}{LC}} \cdot \left(V_s(s) + L \cdot I_L(0^-) - \frac{V_C(0^-)}{s} \right) + \frac{V_C(0^-)}{s}$$

$$\frac{1}{L_1 \cdot C_1} = 1 \times 10^{10} \frac{1}{s^2} \quad \frac{R_1}{L_1} = 5 \times 10^4 \frac{1}{s}$$

open circuit for cap and short for inductor at $t = 0^-$ and source is 0 so I_{L0^-} and V_{C0^-} are both 0

$$V_C(s) = \frac{1 \cdot 10^{10}}{s^2 + 5 \cdot 10^4 \cdot s + 1 \cdot 10^{10}} \cdot \left(\frac{10}{s} \right)$$

Use partial fraction expansion

$$s^2 + 5 \cdot 10^4 \cdot s + 1 \cdot 10^{10} = 0$$

$$\begin{pmatrix} -25000 + 25000i \cdot \sqrt{15} \\ -25000 - 25000i \cdot \sqrt{15} \end{pmatrix}$$

$$V_C(s) = \frac{-5 + 1.29i}{s + 2.5 \cdot 10^4 - 9.68 \cdot 10^4 i} + \frac{-5 - 1.29i}{s + 2.5 \cdot 10^4 + 9.68 \cdot 10^4 i} + \frac{10}{s}$$

$$V_C(t) = \left[(-5 + 1.29i) \cdot \exp(-2.5 \cdot 10^4 t + 9.68 \cdot 10^4 i t) + (-5 - 1.29i) \cdot \exp(-2.5 \cdot 10^4 t - 9.68 \cdot 10^4 i t) + 10 \right] u(t)$$

Note: The solution can be left in this form

b. Use Laplace analysis to determine the voltage across the capacitor as a function of time for $t > 15E-6$ s

Use the final conditions of the first region as the initial conditions of the second region (at $15E-6$).

$$V_C(s) = \frac{\frac{1}{LC}}{s^2 + \frac{R}{L}s + \frac{1}{LC}} \cdot \left(V_s(s) + L \cdot I_L(15 \cdot 10^{-6}) - \frac{V_C(15 \cdot 10^{-6})}{s} \right) + \frac{V_C(15 \cdot 10^{-6})}{s}$$

INITIAL CONDITIONS:

Must find $V_C(15 \cdot 10^{-6})$ and $L \cdot I_L(15 \times 10^{-6})$

plug $15E-6$ into equation from part a to get $V_C(t$ at $15E-6)$

Then use $I_C = C \cdot \frac{dV_C}{dt}$ to find I_L (which is the same as I_C)

$$(-5 + 1.29i) \cdot \exp[-2.5 \cdot 10^4(15 \cdot 10^{-6}) + 9.68i \cdot 10^4(15 \cdot 10^{-6})] + (-5 - 1.29i) \cdot \exp[-2.5 \cdot 10^4(15 \cdot 10^{-6}) - 9.68i \cdot 10^4(15 \cdot 10^{-6})] + 10 :$$

$$V_C(0) = 7.425$$

Taking derivative of $V_C(t)$ from previous solution.

$$(128.0 - 516250.0i) \cdot e^{-(25000.0 - 96800.0i) \cdot (15 \cdot 10^{-6})} + (128.0 + 516250.0i) \cdot e^{-(25000.0 + 96800.0i) \cdot (15 \cdot 10^{-6})} = 7.046 \times 10^5$$

$$L \cdot C \cdot \frac{dV_C}{dt} = 1 \cdot 10^{-2} \cdot 1 \cdot 10^{-8} \cdot 7.046 \times 10^5 = 7.046 \times 10^{-5}$$

$$1 \cdot 10^{-2} \cdot 1 \cdot 10^{-8} \cdot 7.046 \times 10^5 = 7.046 \times 10^{-5}$$

$$L \cdot I_L(0) = 7.046 \cdot 10^{-5}$$

$V_s(s) = 0$ source turned off...

$$R_{12} := R_1 + R_2 = 2 \times 10^3 \Omega$$

$$\frac{R_{12}}{L_1} = 2 \times 10^5 \frac{1}{s}$$

$$\frac{1 \cdot 10^{10}}{s^2 + 2 \cdot 10^5 s + 1 \cdot 10^{10}} \cdot \left(7.046 \cdot 10^{-5} - \frac{7.43}{s} \right) + \frac{7.43}{s}$$

$$\frac{s \cdot 7.046 \cdot 10^5 - 7.43 \cdot 10^{10}}{s \cdot (s^2 + 2 \cdot 10^5 s + 1 \cdot 10^{10})} + \frac{7.43}{s}$$

$$s^2 + 2 \cdot 10^5 s + 1 \cdot 10^{10} = 0$$

$$\begin{pmatrix} -100000 \\ -100000 \end{pmatrix}$$

$$\frac{7.43}{s + 1 \cdot 10^5} + \frac{1.447 \cdot 10^6}{(s + 1 \cdot 10^5)^2} + \frac{-7.43}{s} + \frac{7.43}{s}$$

PFE,
take inverse laplace and add time delay

$$\frac{7.43}{s + 1 \cdot 10^5} + \frac{1.447 \cdot 10^6}{(s + 1 \cdot 10^5)^2}$$

$$V_{C2} = 7.43 \cdot \exp[-10^5(t - 15 \cdot 10^{-6})] + 1.447 \cdot 10^6 \cdot (t - 15 \cdot 10^{-6}) \cdot \exp[-10^5(t - 15 \cdot 10^{-6})]$$