

Basic:

Voltage divider:  $V_2 = \frac{R_2}{R_1 + R_2}$

Capacitor:  $Z_C = \frac{1}{j\omega C}$ ,  $I = C \frac{dV}{dt}$ ,  $E = \frac{1}{2} CV^2$

LCR: resonance:  $\omega = \frac{1}{\sqrt{LC}}$

Superposition: Short voltage, open current, analyze the thing, add.

Voltage to current:  $I_{IS} = V_{VS}$ ,  $R_{CS} = \frac{1}{RVS}$

OP AMP: Inverting Amp:  $A = \frac{V_{out}}{V_{in}} = -\frac{R_F}{R_{in}}$

Summing Inverting Amp:  $V_{out} = -R_F \left( \frac{V_1}{R_1} + \frac{V_2}{R_2} + \frac{V_3}{R_3} \right)$

Subtraction: Inverting amp with voltage divider on plus:  $V_{out} = V_{plus} \frac{R_G}{R_{in,plus} + R_G} \cdot \frac{R_{in,minus} + R_F}{R_{in,minus}} - V_{minus} \frac{R_F}{R_{in,minus}}$

H response:  $H(\omega) = \frac{V_{out}}{V_{in}}$

Integrator:  $R_F$  is a capacitor,  $V_{out} = -\frac{1}{R_{in}C} \int_0^t V_{in} dt$

Detailed amplification: op amp

diff:  $V_{out} = A_{diff}(V_+ - V_-)$

CM:  $V_{out} = A_{CM}(V_+ + V_-)$

total amplification:  $V_{out} = A_{diff}(V_+ - V_-) + A_{CM}(V_+ + V_-)$

CMRR: common mode rejection ratio:  $\frac{A_{diff}}{A_{CM}}$ ,  $20 \log(CMRR) dB$

Gain Bandwidth Product: Gain  $\cdot$  Bandwidth,  $A \cdot f$

Constant gain at low frequency, decreases linearly because slew rate at high frequency

knee frequency:  $\omega = \frac{1}{RC} = \frac{1}{T} = \frac{1}{2\pi RC}$ ,  $\Rightarrow H(\omega) = \frac{1}{\sqrt{2}}$ ,  $-3dB$

above knee frequency:  $H(\omega) = \frac{1}{\omega RC}$ ,  $-20dB$  per decade

high pass:  $H(\omega) = \omega RC$ ,  $20dB$  per decade, slope multiplied by N stages.

Slew rate:  $= dV_{out}/dt$

ex:  $SR = 1V/\mu s$ , pulse  $0 - 2V$ ,  $T = 2\mu s$ ,  $1V$  triangle wave

good sine wave:  $SR \geq 2\pi f V_0$

DIODES: forward bias, apply  $+V$  to holes, push holes(p) and electrons(n) together, get conduction

IV:  $I = I_0(e^{\frac{V}{V_t}} - 1)$

Thermal voltage:  $V_t = \frac{kT}{e} = 26mV$

Threshold voltage:  $V_{th} = 0.7V$

Reverse saturation current:  $I_0$  around  $10^{-10}$

For  $V \gg V_t$ ,  $I = I_0 e^{\frac{V}{V_t}}$

Differential resistance:  $r_D = \frac{V_t}{I}$  (I from above equation)

Basic DR circuit:  $V_B = IR + V_D$ ,  $I = I_D = I_0 e^{\frac{V}{V_t}}$

Graphically solve:  $V_B = I_0 e^{\frac{V}{V_t}} R + V_D$ ,  $V_B - V_D = I_0 e^{\frac{V}{V_t}}$

Approximate analytic solution:  $V_D = V_{th} = 0.7V$ ,  $I = \frac{V_B - 0.7}{R}$

Linearization:  $r_D = \frac{V_t}{I_D}$ , can treat as this resistor for an AC signal, just account for DC offset with superposition.

find I with DC (probably  $\frac{V_B - 0.7}{R}$ ), and  $r_D = \frac{0.026}{I}$ , then add AC signal with  $r_D$

Rectification:  $P = V_{th} I$  AM: through diode and R and past C to filter out frequency, left with audio. Also LED and solar cell.

ZENER DIODE: width of depletion region changed with doping

Breakdown: at  $V = V_{breakdown} = V_Z$

IV: turn on at  $V_{th} = 0.7$ , reverse turn on at  $-V_Z$ , typically used backwards, so flip V, expect  $+ | < -$

$V < V_Z \Rightarrow r_Z = \frac{dV}{dI} = \infty$  OR  $V \geq V_Z$ ,  $r_Z = \frac{dV}{dI} = 0$

$I = I_0 e^{\frac{V - V_Z}{V_t}}$

Basic ZDR circuit:

Graphical:  $V_{Bat} - IR = V_Z(I)$

Approx analytical assume  $V_Z = \text{constant}$ ,  $I = \frac{V_B - V_Z}{R}$

Voltage stabilization:  $V_{out} = V_Z = \text{constant}$ , wastes power, must have significant  $R_{Load}$

Voltage clipper: two parallel to load, facing away, V clipped at  $V_Z + V_{th}$

Voltage shifter: one series, cuts  $V_Z$  off of input