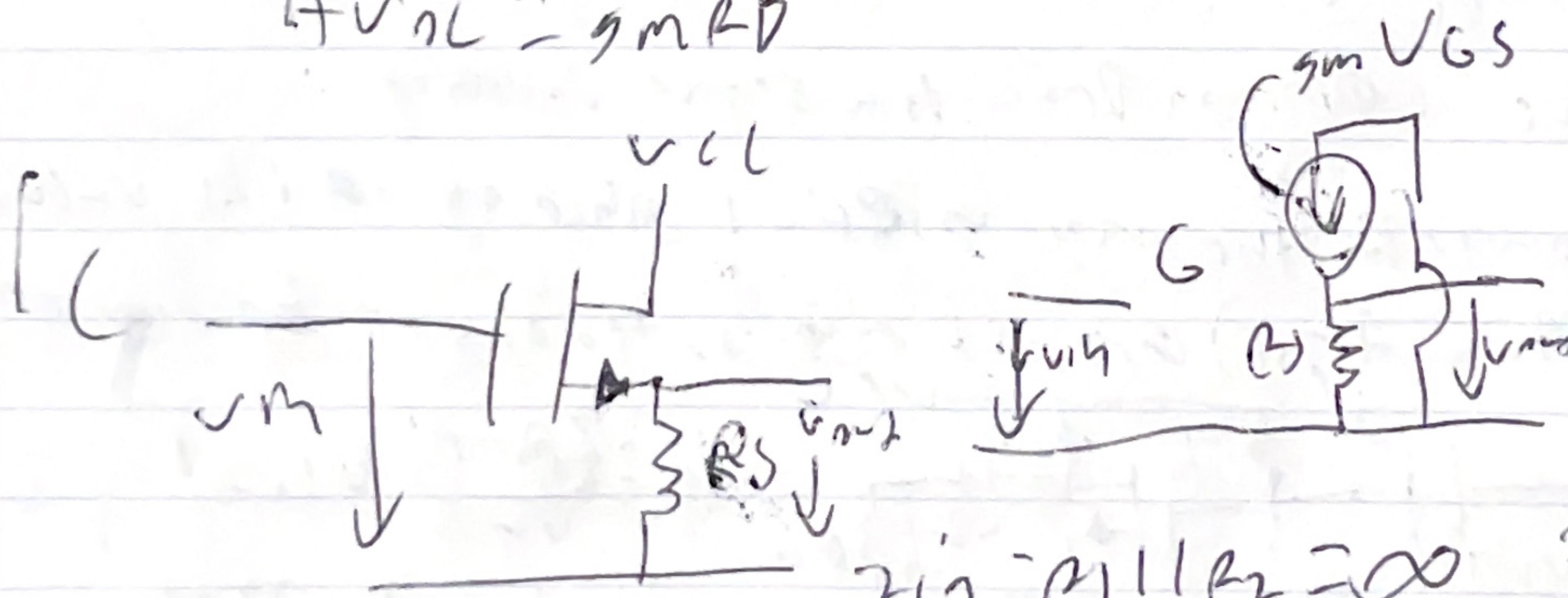


$$Z_{in} = R_1 \parallel R_2 \quad R_1, R_2 \gg \infty \quad \left(\frac{1}{R_1} + \frac{1}{R_2} \right)^{-1} = \frac{1}{2} \gg \infty$$

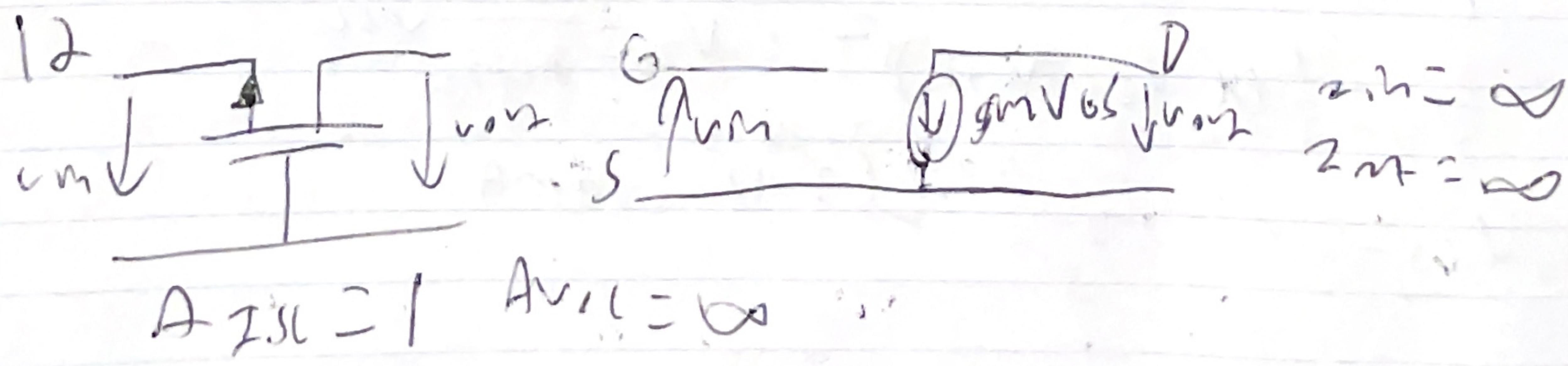
$$Z_{out} = R_D \quad A_{VSL} = g_m (R_1 \parallel R_2) \approx \infty$$

$$A_{VNL} = g_m R_D$$



$$Z_{in} = R_1 \parallel R_2 \approx \infty \quad Z_{out} = R_S$$

$$A_{VSL} = g_m (R_1 \parallel R_2) \approx \infty \quad A_{VNL} = \frac{g_m R_S}{1 + g_m R_S}$$



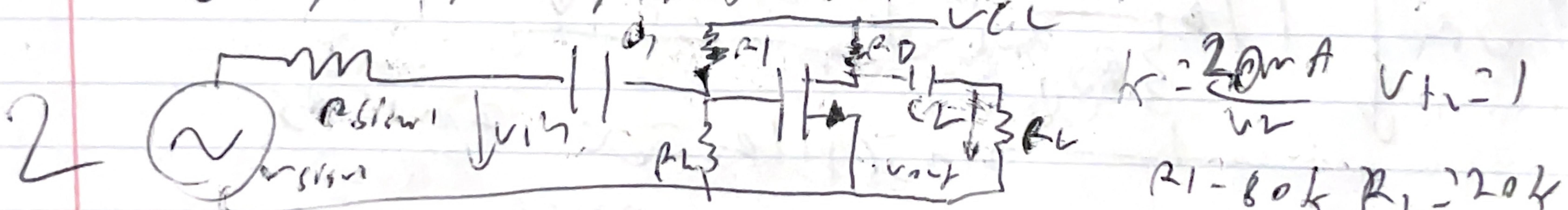
| le | S | D | G |
|------------------|-----|------------|------|
| 2m | mid | High | Low |
| 2n-2 | mid | Low | High |
| A _{VL} | mid | Low | High |
| A _{ZSL} | mid | High : Low | |

If lowest Zout is Common Drain, Shunt source follower with no A_{VL} shunts low Zout is desirable.

If lowest Zin is Common Gate, since it is a current follower, having low Zin is desirable.

If lowest A_{VL} is Common D which is roughly 1. This is good since Common Drain is a source follower.

If Common Source is the most useful since all of its values (Z_{in}, Z_{out}, A_{VL}, A_{ZSL}) are in range, letting it be very versatile.



2a this is common source:

$$I_D = \frac{1}{2} k(V_{GS} - V_{TH})^2 \quad V_{GS} = \frac{R_2}{R_1 + R_2} V_{CC}$$

$$V_{GS} = 2V \quad I_D = \frac{20}{2} \cdot \frac{1}{2} (2 - 1)^2 = 10mA$$

R_{DS} = 5k R_L = 1k V_{DS} = 10V

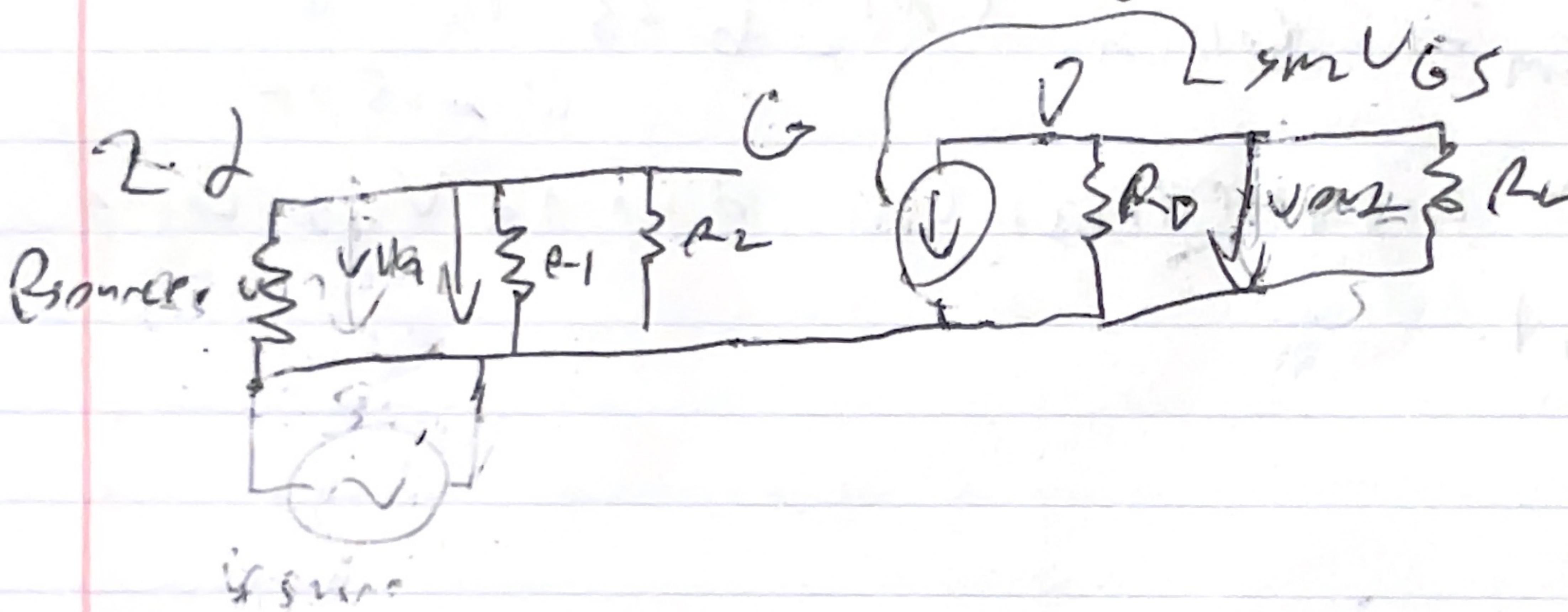
$$R_2$$

$$2L \quad V_{DS} = 5, I_D = 10mA$$

$$V_{RD} = V_{DS} + I_D R_D$$

$$I_D R_D = V_{CC} - V_{DS} \quad 10mA \cdot R_D = 5 \quad R_D = 500\Omega$$

$$2C \quad g_m = k(V_{GS} - V_{TH}) = \frac{20mA}{V_2}$$



$$2e \quad Z_M = R_D || R_L = 38.19\Omega$$

$$Z_{out} = R_D || R_L = 33.3\Omega$$

$$A_{VOL} = \frac{V_{DS}}{V_{IN}} = \frac{V_{GS} gm (R_D || R_L)}{V_{GS}} = 20 \cdot 33.3 = 666$$

$$A_V = \frac{V_{OUT}}{V_{IN}} = \frac{V_{GS} gm (R_D || R_L)}{V_{GS}} = 20 \cdot 33.3 = 666$$

2f) Drain-gate capacitance C_{DG} is a parasitic capacitance insite the FET between the drain and gate. It adds two further Miller capacitors will "exist" between my two sources of voltage that have resistors between them.

Miller capacitor C_{M} is C_{DG} but with different coefficients to fit it to the equivalent circuit better.



$$20 \text{ MHz} = \omega_m \cdot R$$

$$20 \text{ MHz} = \frac{1}{R_1 C_m}$$

$$20 \text{ MHz} = \frac{1}{C} \quad C = 5 \cdot 10^{-8} \text{ Farads} \quad 50 \text{ pF}$$

$$24 \quad 20 \text{ MHz} = \frac{1}{C_m} = \frac{1}{60 \text{ pF}} \quad C_m = 6.25 \cdot 10^{-13} \text{ F}$$

21 Common drain h- $s + v^p$ bias + C_m 2 up to V_{GD} being

very close to 1.