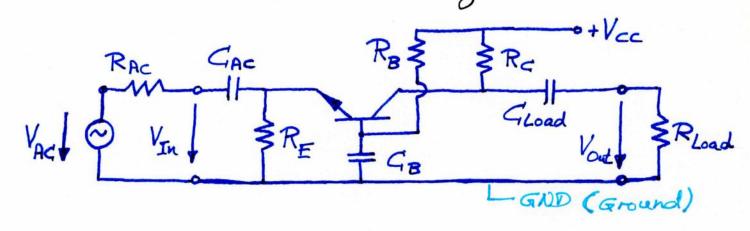
ITE - Homework 7 - Solution

Problem 1 Amplifier

$$\beta = \frac{\alpha}{1-\alpha} = \frac{0.99}{1-0.99} = \frac{0.99}{0.01} = 99 \approx 100$$

(b) Redraw circuit diagram



(c) Base is shared (common) for AG signals. Descript is common-base configuration.

(d)
$$R_{c} = R_{E} = 500\Omega$$
 $V_{Rc} = 3.0V$ $Q-point?$

$$\Rightarrow I_{c} = \frac{V_{Rc}}{R_{c}} = \frac{3.0V}{500\Omega} = 6 \text{ mA}$$

$$\Rightarrow I_{B} = \frac{I_{c}}{B} = \frac{6mA}{100} = 0.06mA = 60 \text{ pA}$$

$$\Rightarrow I_{E} = I_{c} + I_{B} = 6mA + 60 \text{ pA} = 6.06mA$$

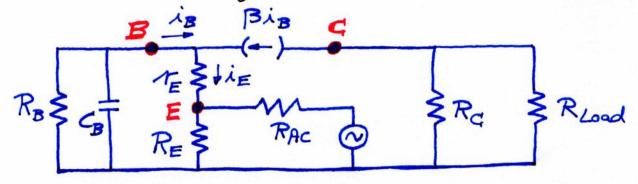
$$\Rightarrow V_{RE} = 500\Omega \times 6.06 \text{ mA} = 3.03V$$

$$\Rightarrow V_{Base} = V_{RE} + V_{BE} = 3.03V + 0.7V = 3.73V$$

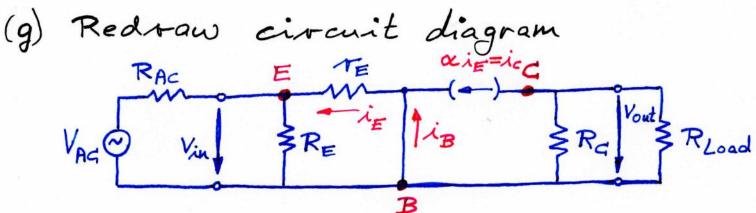
$$\Rightarrow R_{B} = \frac{V_{cc} - V_{Base}}{I_{Base}} = \frac{12V - 3.73V}{60 \mu A} = \frac{138 k \Omega}{100 \mu A}$$

(e) Differential emitter resistance $T_{\rm E} = \frac{V_{\rm E}}{T_{\rm E}} = \frac{26mV}{6.06mA} = 4.29\Omega$

(f) AG small-signal equivalent circuit



Note: Ros is shorted by GB => Base is connected to GND.



-> Linear circuit. -> We can analyze effect of one source without considering other sources.

(h)
$$Z_{in} = \frac{V_{in}}{J_{in}} = \frac{V_{in}}{J_{RE} + J_{TE}} = \frac{V_{in}}{J_{RE}} + J_{E}$$

$$= \left(\frac{V_{in}}{V_{in}} + J_{E}\right)^{-1} = \left(\frac{1}{R_{E}} + \frac{1}{J_{E}}\right)^{-1} = R_{E} / J_{E}$$

(j)
$$A_{VOC} = \frac{V_{out}}{V_{in}}$$
 when $R_{Load} = \infty$ (OG)

$$A_{Voc} = \frac{V_{out}}{V_{in}} = \frac{i_c R_c}{i_E I_E} = \frac{\alpha i_E R_c}{i_E I_E} = \alpha \frac{R_c}{I_E} \approx \frac{R_c}{I_E}$$

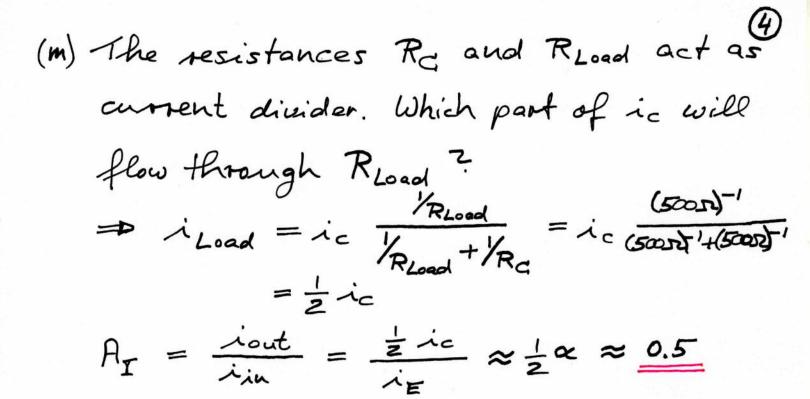
$$= \frac{500 \Omega}{4.29 \Omega} = 117$$

$$A_{ISC} = \frac{\dot{\lambda}_{out}}{\dot{\lambda}_{iu}} = \frac{\dot{\lambda}_{c}}{\dot{\lambda}_{E}} \approx \frac{\dot{\lambda}_{c}}{\dot{\lambda}_{E}} = \alpha = 0.99$$

(1)
$$A_V = ?$$
 $R_{Load} = 500 \Omega$ $R_C = 500 \Omega$

$$A_V = \frac{V_{out}}{V_{in}} = \frac{ic \left(R_c || R_{Load}\right)}{iE} \approx \frac{R_c || R_{Load}}{V_E}$$

$$= \frac{500 \Omega || 500 \Omega}{4.29 \Omega} = \frac{250 \Omega}{4.29 \Omega} = \frac{58.5}{4.29 \Omega}$$



(n)

Res

Cload

Res

RLoad

GLoad blocks DG This is a high-fre-GLoad lets pass AG J quency pass filter

Relevant R is the R that GLoad will discharge through.

Inspection of circuit diagram reveals that relevant $R = R_C + R_{Load} = 1 \text{ kSZ}$

Recall $\omega_{3dB} = \omega_{cutoff} = \frac{1}{RC}$

=> f3dB = fcutoff = = 1 RC

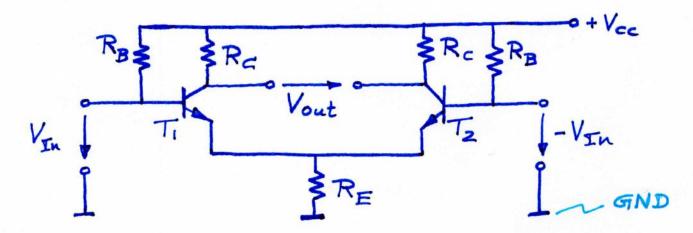
(0) fortaff =
$$\frac{1}{2\pi RC}$$
 \Rightarrow $C' = \frac{1}{2\pi R + 3dB}$

$$G = \frac{1}{2\pi I k R I k Hz} = 159 nF$$

Problem 2

Differential amplifier

(a) Circuit diagram



(b) Q-point
$$V_{RE} = IV$$
 $V_{CE} = 5V$ $V_{CC} = 12V$

Current through $R_E = I_{RE} = 20 \text{ mA}$

$$\Rightarrow R_E = \frac{V_{RE}}{I_{RE}} = \frac{IV}{20 \text{ mA}} = \frac{50 \Omega}{20 \text{ mA}}$$

$$I_{CI} \approx I_{EI} = \frac{1}{2} I_{RE} = \frac{10 \text{ mA}}{10 \text{ mA}}$$

$$KVL \quad V_{CC} = V_{RE} + V_{CE} + V_{RC}$$

$$\Rightarrow V_{RC} = V_{CC} - V_{CE} - V_{RE} = 12V - 5V - IV = 6V$$

$$\Rightarrow R_C = \frac{V_{RC}}{I_{RC}} = \frac{6V}{10 \text{ mA}} = \frac{600 \Omega}{100 \text{ mA}}$$

$$I_B = I_{CC} = \frac{10 \text{ mA}}{100} = 0.1 \text{ mA}$$

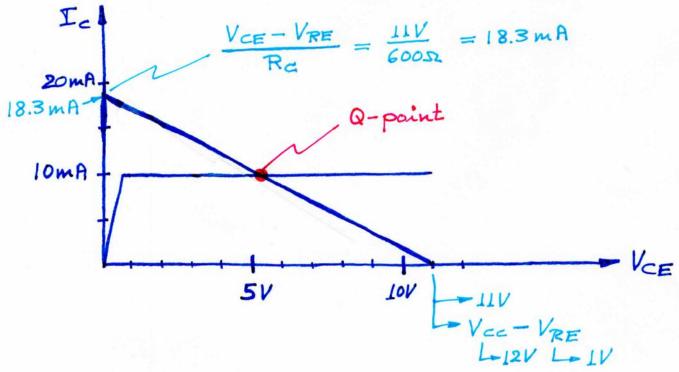
$$V_{RB} = V_{CC} - V_{RE} - V_{BE} = 12V - 1V - 0.7V = 10.3V$$

$$\Rightarrow R_B = V_{RB} = \frac{10.3V}{0.1 \text{ mA}} = \frac{103 \text{ k}\Omega}{0.1 \text{ mA}}$$

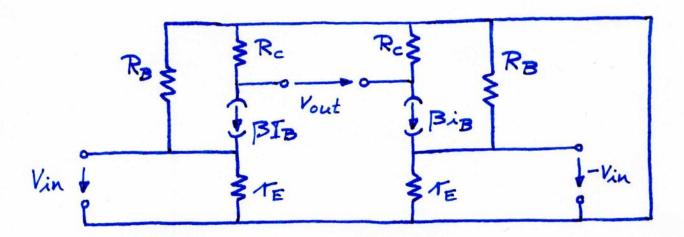
(c) At Q-point:
$$V_E = IV$$
 $V_C = 6V$ $V_{CE} = 5V$ $I_C = 10 \text{ mA}$ $I_E = 10 \text{ mA}$

 V_E can be considered "virtual ground" because V_E does not change. Currents through transistors change but sum $I_{EL} + I_{E2} = I_{RE}$ does not $\Rightarrow V_E = constant$

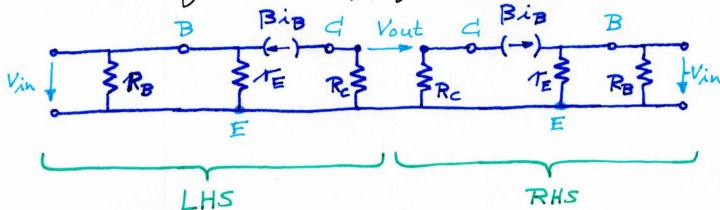
Output characteristic



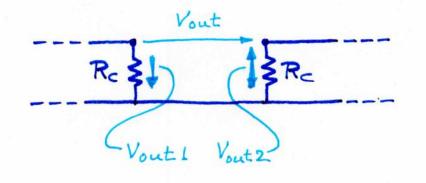
(d) Differential emitter resistance T_E (may also be called dynamic emitter resistance) $T_E = V_{\overline{L}E} = 26 \, \text{mV}$ $T_E = V_{\overline{L}E} = 10 \, \text{mA} = 2.6 \, \Omega$



Redrowing (simplifying) the circuit



LHS and RHS are identical. We therefore analyze only the LHS and double the output voltage



Vout = Vout 1 + Vout 2 = 2 Vout 1

8

(e) LHS circuit:
$$A_{VOC} = \frac{V_{out}}{V_{in}} = \frac{\beta_{iB}R_{c}}{(\beta+1)i_{B}T_{E}} \approx \frac{R_{c}}{T_{E}}$$

Same Avoc for RHS circuit due to symmetry

Total amplification
$$A_{voc} = \frac{2v_{out}}{v_{in}} = \frac{2R_E}{T_E}$$

(f) Numerical value
$$A_{VOC} = 2 \frac{RE}{E} = 2 \frac{600\pi}{2.6\pi} = 461$$

(g) Two amplifier stages in series
$$A_{Voc} = A_{Voc}^2 = 461^2 = 212000$$
2 stages

$$I_c = \alpha I_E$$

Ic = & I = 1.0 Will not work properly

a and B are reduced

Civanit will likely still work

a and B are reduced

→ IB < Ic and IB < IE

BE - Forward bias

CE => Reverse bias.