

Digital Design & Computer Arch.

Lecture 5a: Sequential Logic Design II Finite State Machines

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First, We Will Complete Sequential Logic Design

We Covered A Lot of Sequential Logic

■ **Circuits that can store information**

- ❑ Cross-coupled inverter
- ❑ R-S Latch
- ❑ Gated D Latch
- ❑ D Flip-Flop
- ❑ Register
- ❑ Memory

■ **Sequential logic circuits**

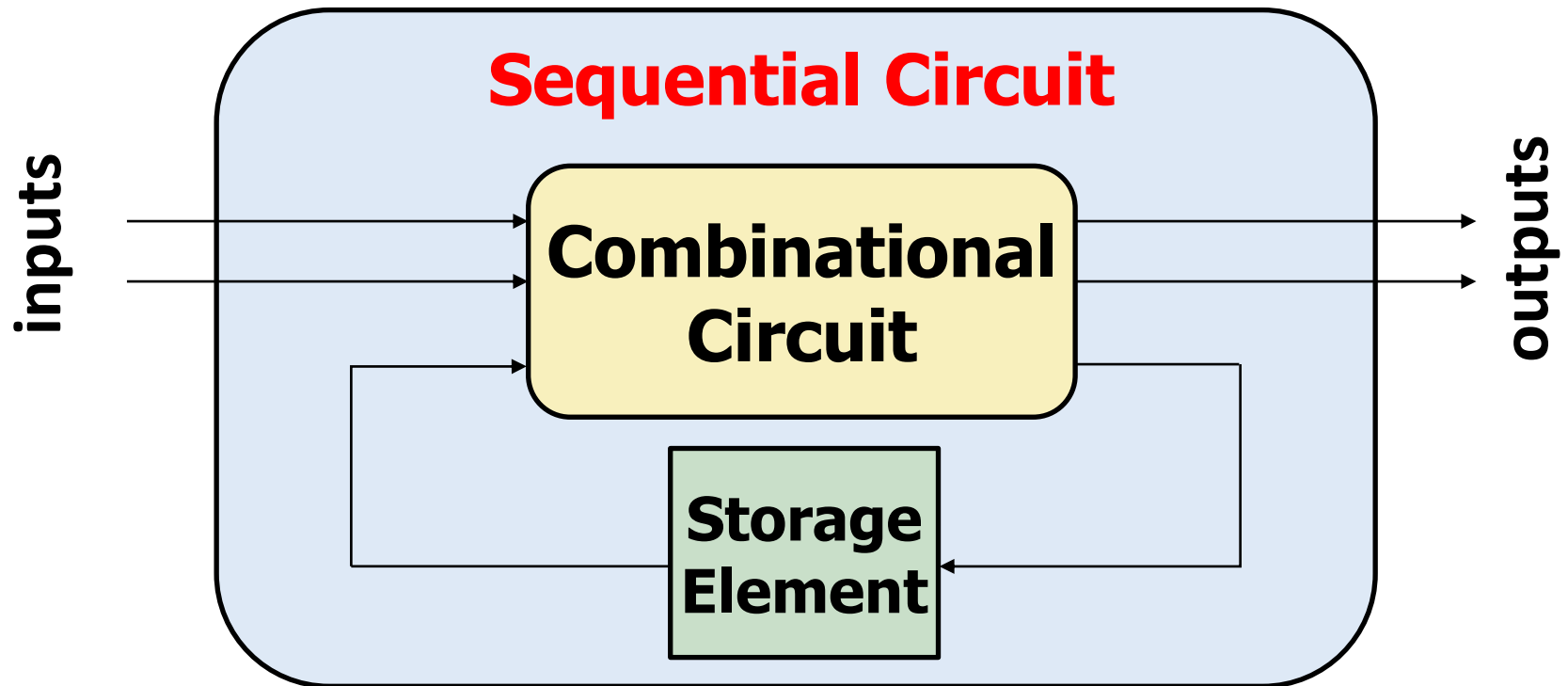
- ❑ State & Clock
- ❑ Asynchronous vs. Synchronous

■ **Finite State Machines (FSM)**

- ❑ How to design FSMs

Recall: Sequential Circuits

- Circuits that produce output depending on **current** and **past** input values – circuits with **memory**



Recall: Sequential Logic Circuits



Combinational

Only depends on current inputs

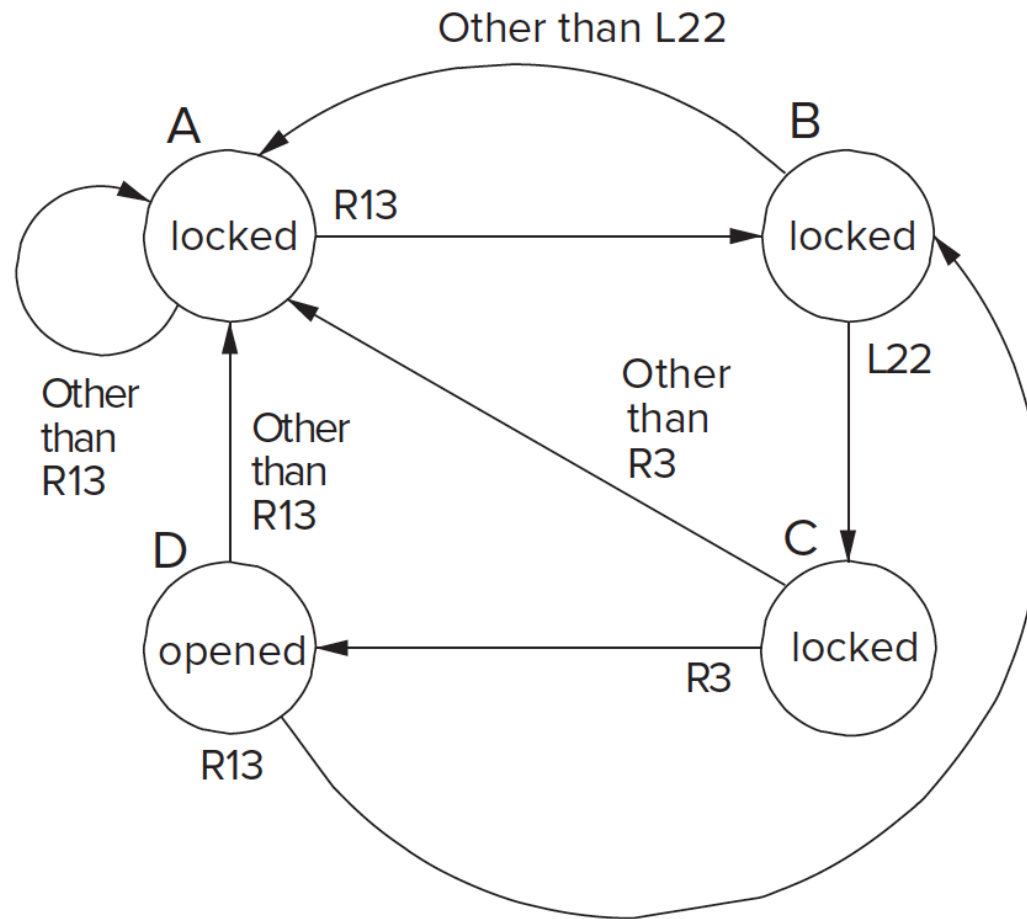


Sequential

Opens depending on past inputs

Recall: State Diagram of Our Sequential Lock

- Completely describes the operation of the sequential lock



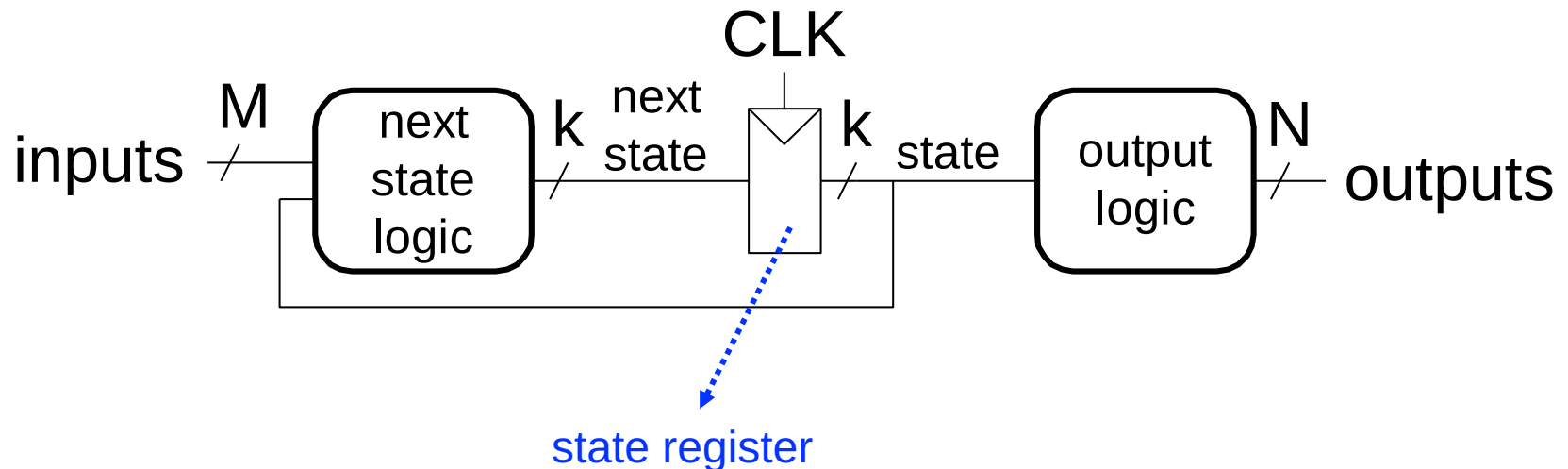
Recall: Finite State Machines (FSMs) Consist of:

■ Five elements:

1. A **finite** number of **states**
 - **State**: snapshot of all relevant elements of the system at the time of the snapshot
2. A **finite** number of external **inputs**
3. A **finite** number of external **outputs**
4. An explicit **specification of all state transitions**
 - How to get from one state to another
5. An explicit **specification of what determines each external output value**

Recall: Finite State Machines (FSMs)

- Each FSM consists of three separate parts:
 - next state logic
 - state register
 - output logic

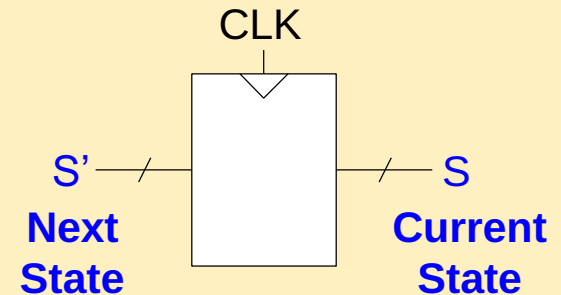


At the beginning of the clock cycle, next state is latched into the state register

Recall: Finite State Machines (FSMs) Consist of:

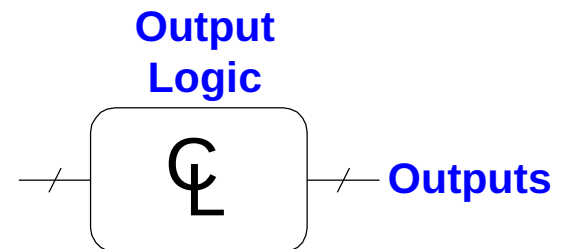
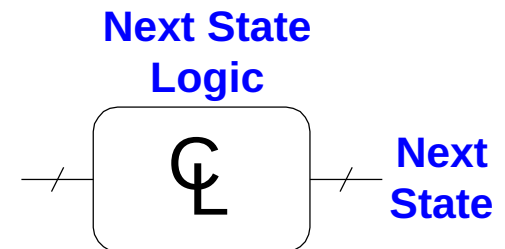
■ Sequential Circuits

- State register(s)
 - Store the current state and
 - Provide the next state at the clock edge



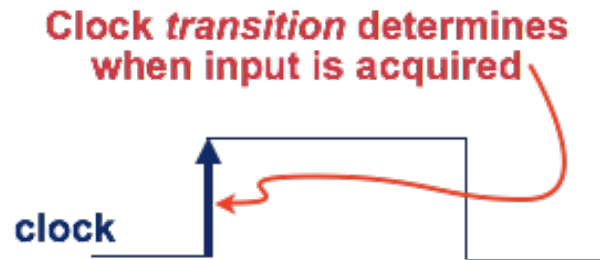
■ Combinational Circuits

- Next state logic
 - Determines what the next state will be
- Output logic
 - Generates the outputs

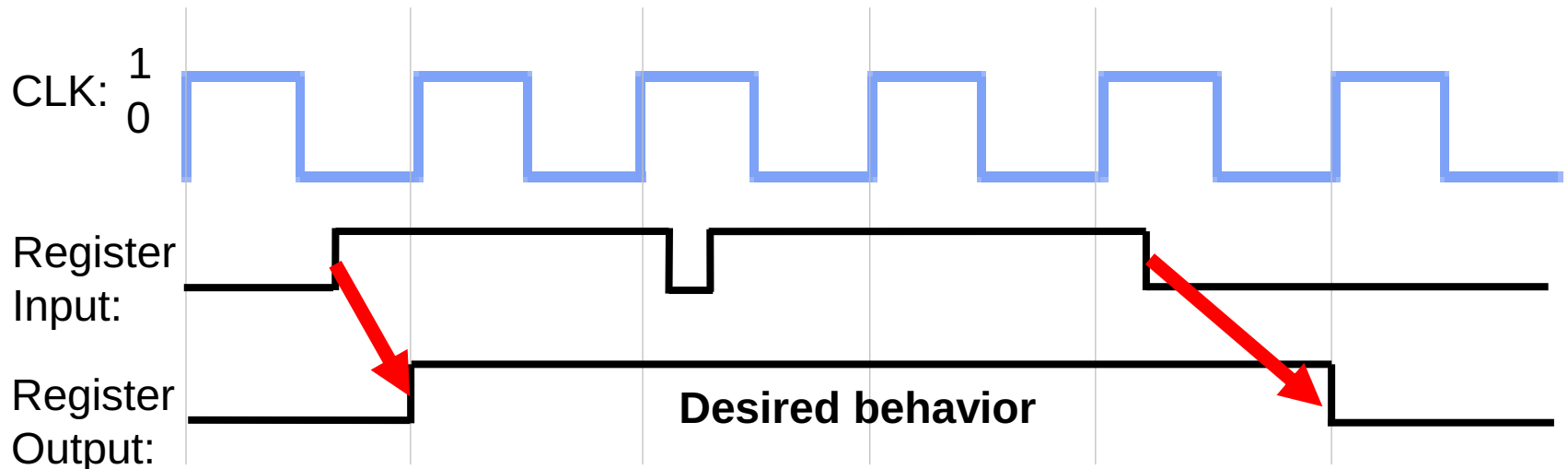


Recall: State Register Implementation

- How can we implement a **state register**? Two properties:
 - We need to store data at the **beginning** of every clock cycle

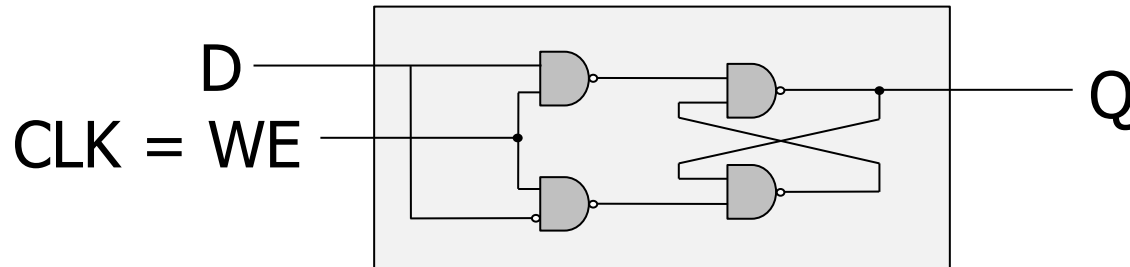


- The data must be **available** during the **entire clock cycle**



Recall: The Problem with Latches: Transparency

Recall the
Gated D Latch



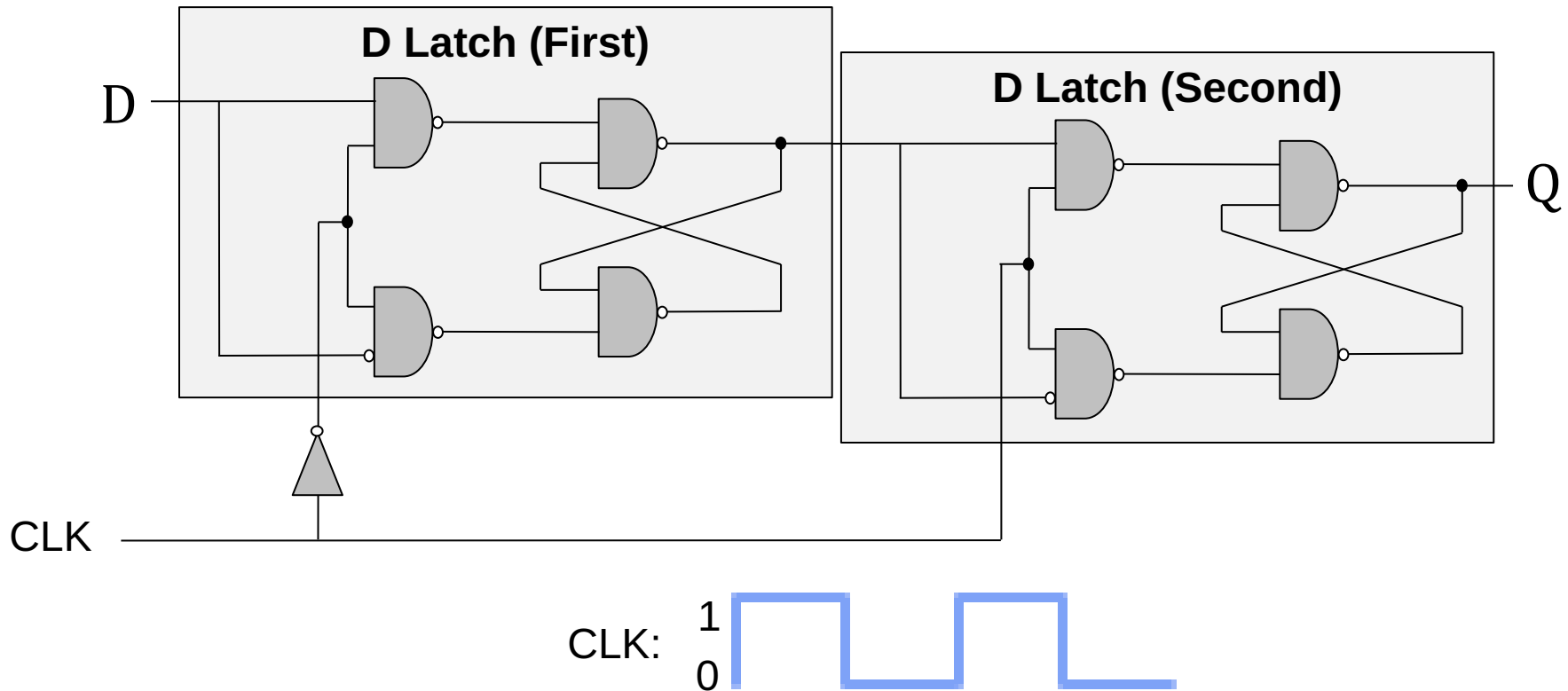
How can we change the latch, so that

1) D (input) is **observable** at **Q** (output)
only at the **beginning of next** clock cycle?

2) Q is **available for the full clock cycle**

Recall: The D Flip-Flop

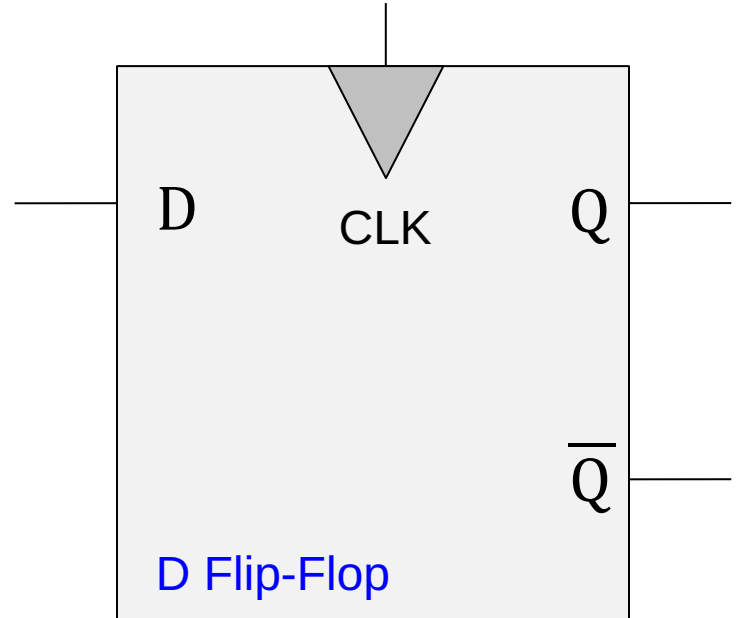
- 1) state change on clock edge, 2) data available for full cycle



- When the clock is low, 1st latch propagates **D** to the input of the 2nd (Q unchanged)
- Only when the clock is high, 2nd latch latches **D** (**Q stores D**)
 - At the rising edge of clock (clock going from 0→1), Q gets assigned D

Recall: The D Flip-Flop

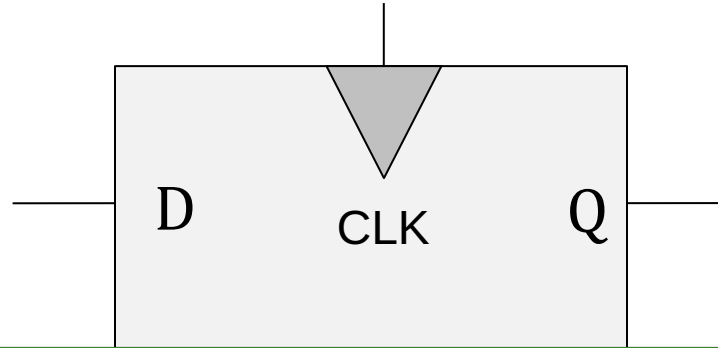
- 1) state change on clock edge, 2) data available for full cycle



- At the rising edge of clock (clock going from 0->1), **Q** gets assigned **D**
- At all other times, Q is unchanged

Recall: The D Flip-Flop

- 1) state change on clock edge, 2) data available for full cycle



We can use **D Flip-Flops**
to implement the state register

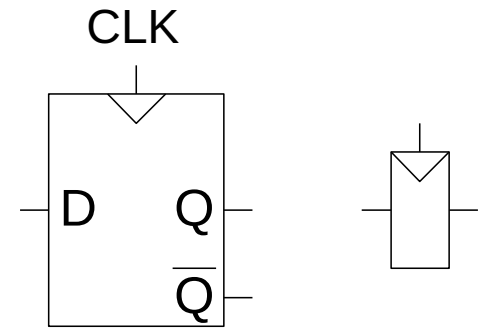
- At the rising edge of clock (clock going from 0->1), **Q** gets assigned **D**
- At all other times, **Q** is unchanged

Recall: Rising-Clock-Edge Triggered Flip-Flop

- **Two inputs:** CLK, D

- **Function**

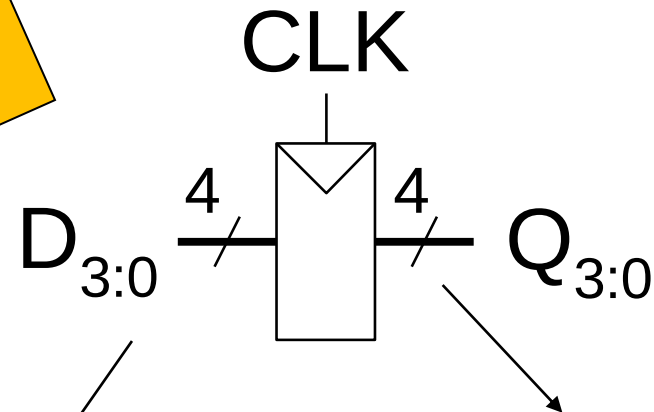
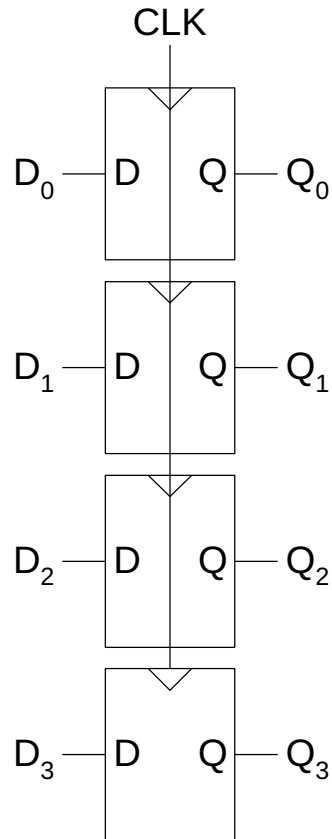
- The flip-flop “samples” **D** on the rising edge of CLK (**positive edge**)
- When CLK rises from 0 to 1, **D** passes through to **Q**
- Otherwise, **Q** holds its previous value
- **Q** changes **only** on the rising edge of CLK



- A flip-flop is called an **edge-triggered state element** because it captures data on the clock edge
 - A latch is a **level-triggered** state element

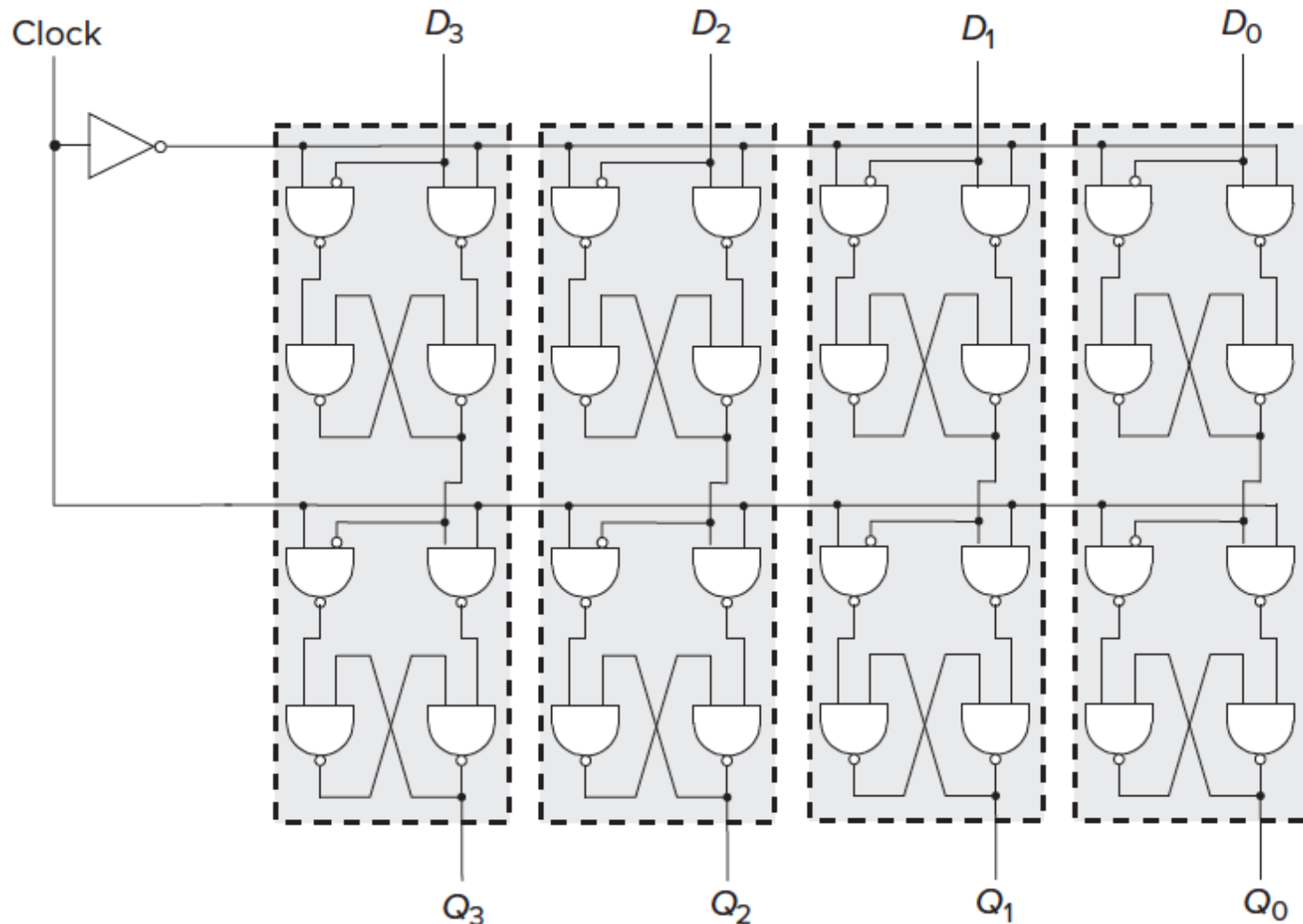
D Flip-Flop Based Register

- Multiple parallel D flip-flops, each of which storing 1 bit



This register stores 4 bits

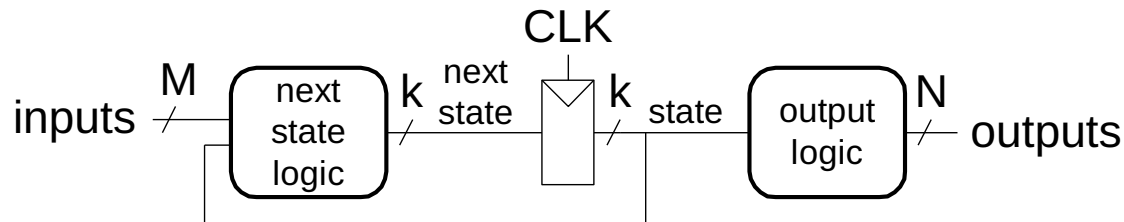
A 4-Bit D-Flip-Flop-Based Register (Internally)



Finite State Machines (FSMs)

- Next state is determined by the current state and the inputs
- Two types of finite state machines differ in the **output logic**:
 - **Moore FSM**: outputs depend only on the current state

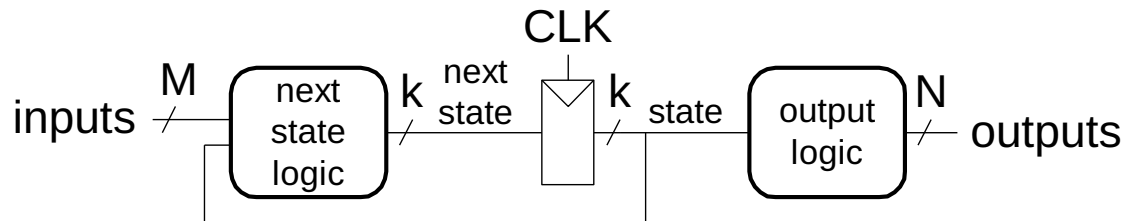
Moore FSM



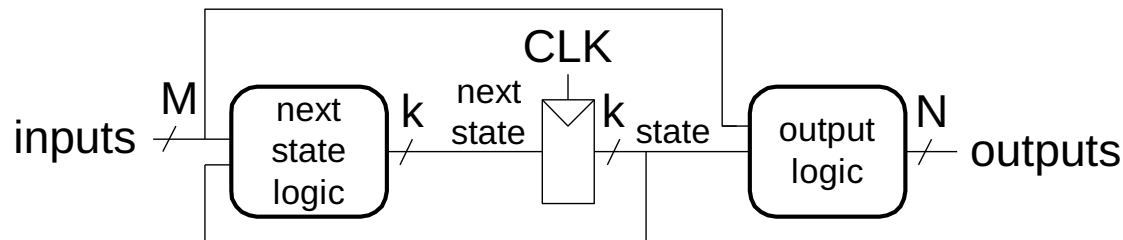
Finite State Machines (FSMs)

- Next state is determined by the current state and the inputs
- Two types of finite state machines differ in the **output logic**:
 - **Moore FSM**: outputs depend only on the current state
 - **Mealy FSM**: outputs depend on the current state and the inputs

Moore FSM

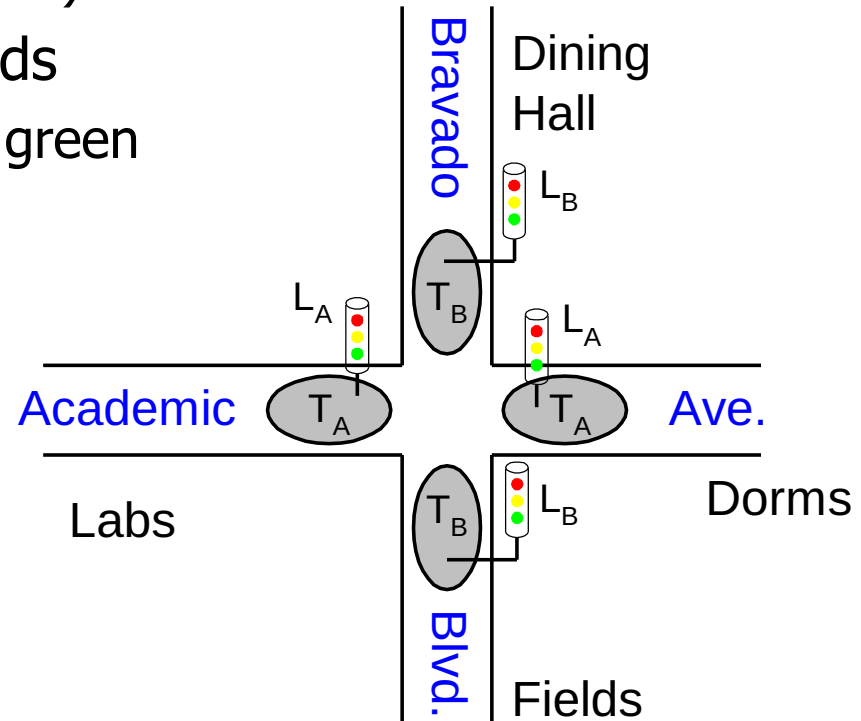


Mealy FSM



Finite State Machine Example

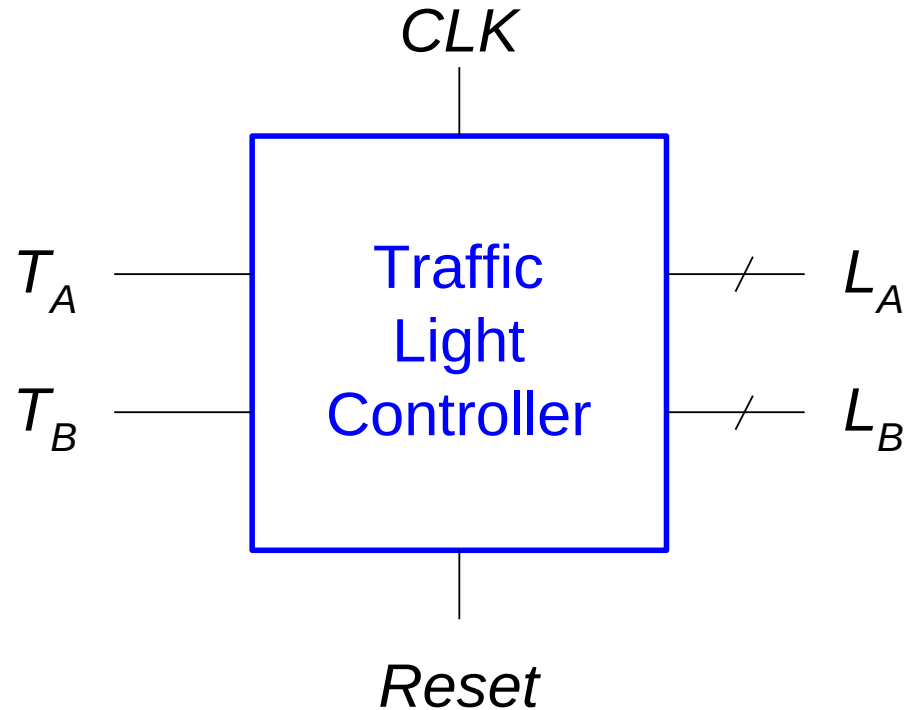
- “Smart” traffic light controller
 - **2 inputs:**
 - Traffic sensors: T_A , T_B (TRUE when there's traffic)
 - **2 outputs:**
 - Lights: L_A , L_B (Red, Yellow, Green)
 - State can change every 5 seconds
 - Except if green and traffic, stay green



From H&H Section 3.4.1

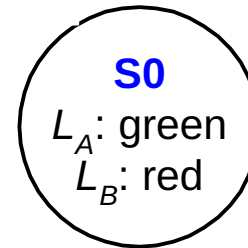
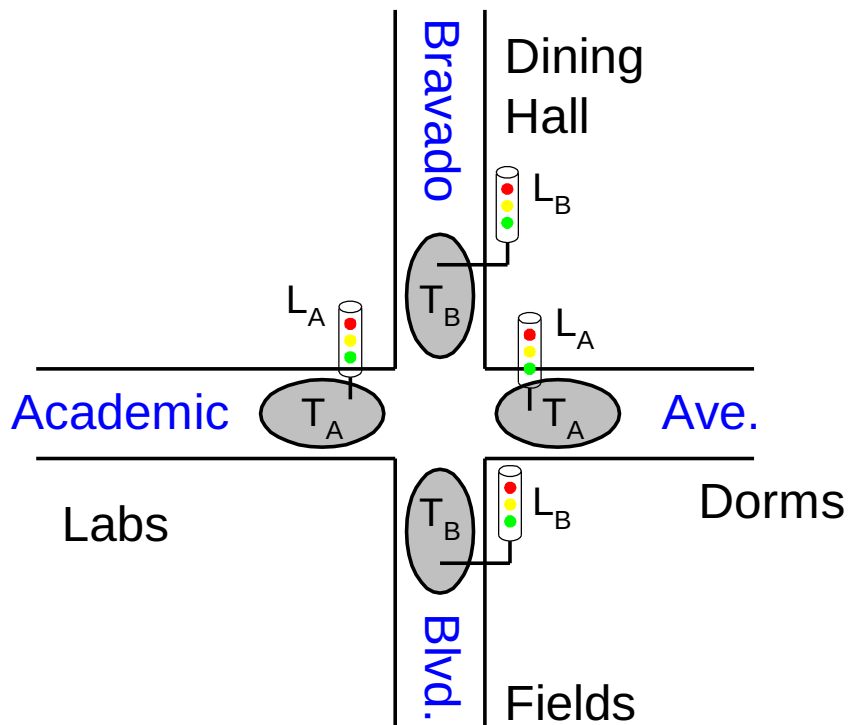
Finite State Machine Black Box

- **Inputs:** CLK, Reset, T_A , T_B
- **Outputs:** L_A , L_B



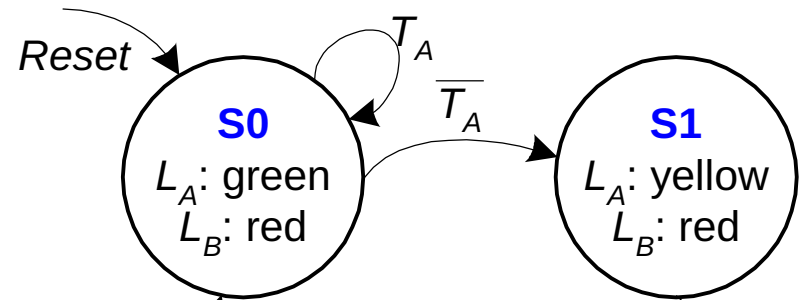
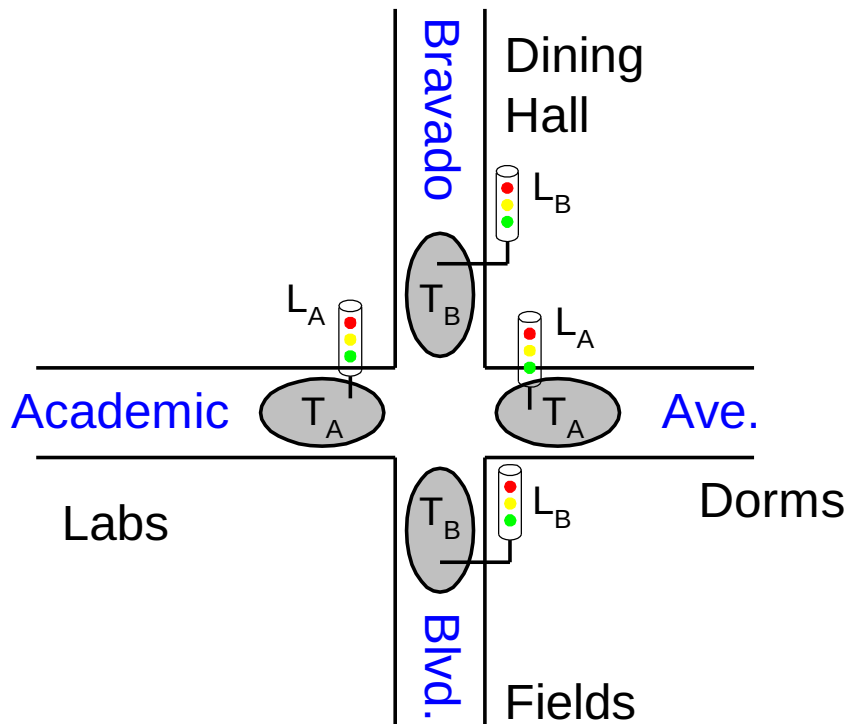
Finite State Machine Transition Diagram

- **Moore FSM:** outputs labeled in each state
 - **States:** Circles
 - **Transitions:** Arcs



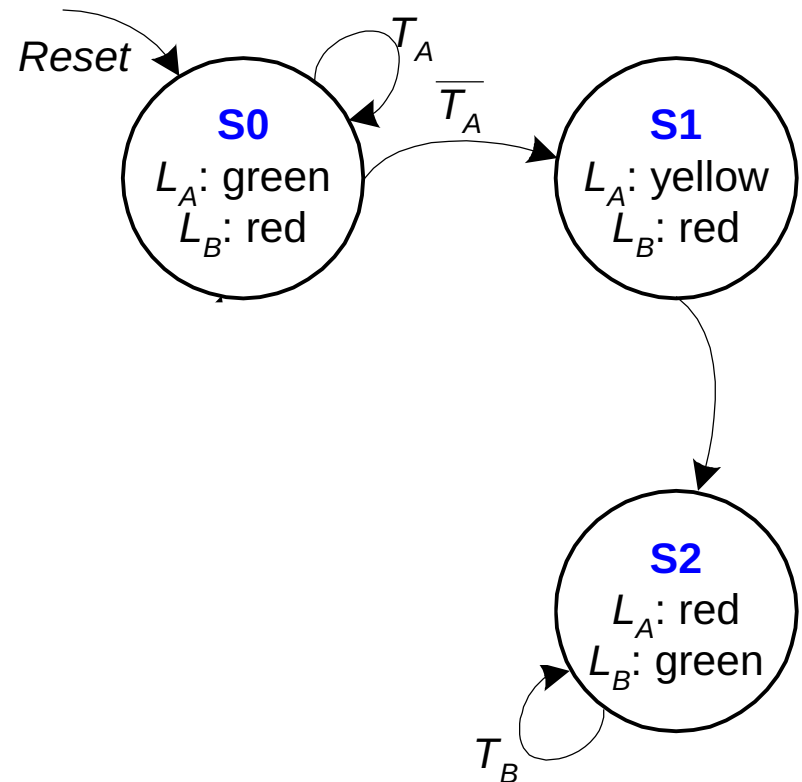
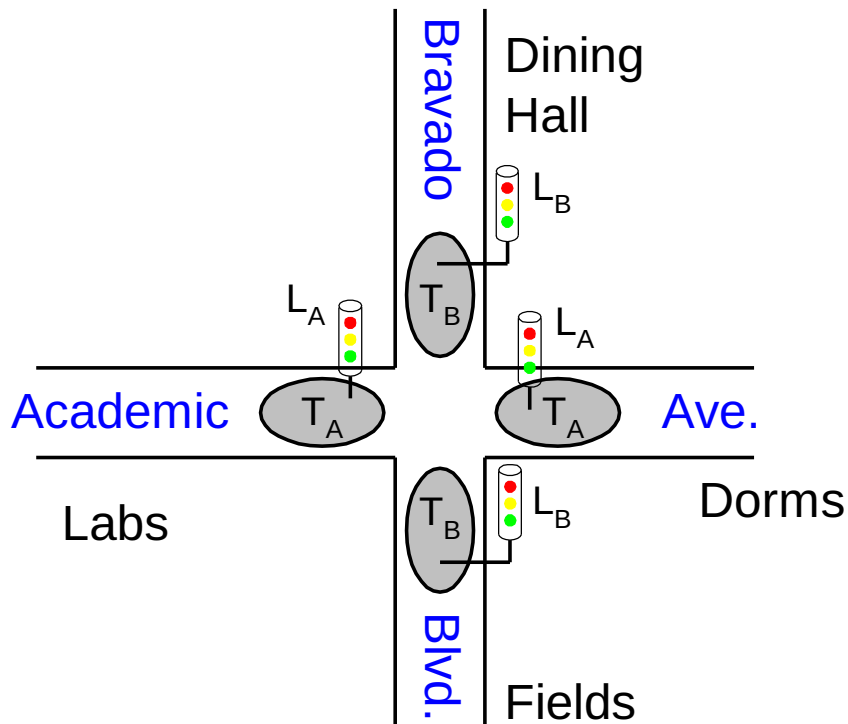
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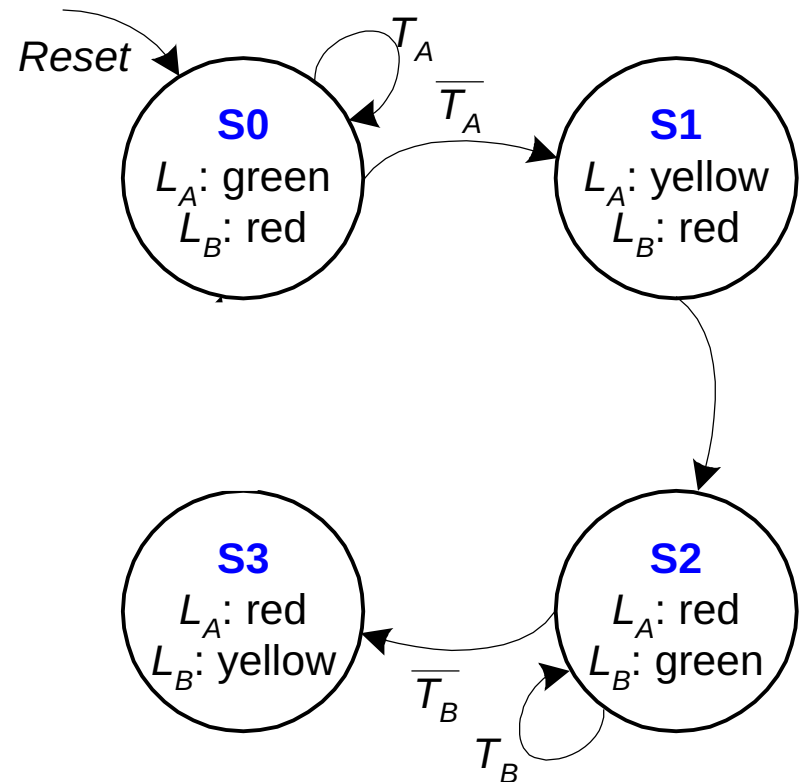
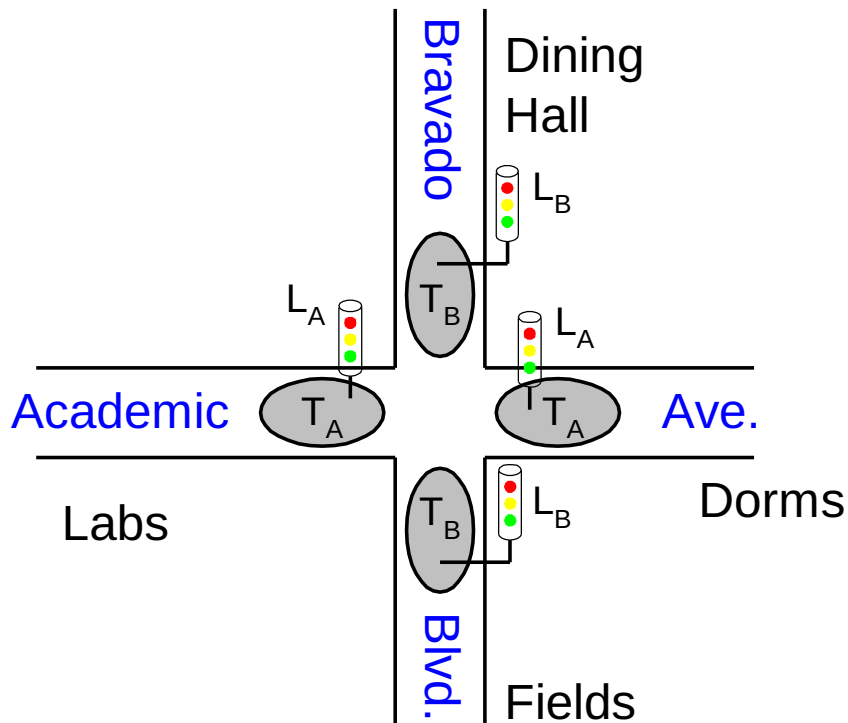
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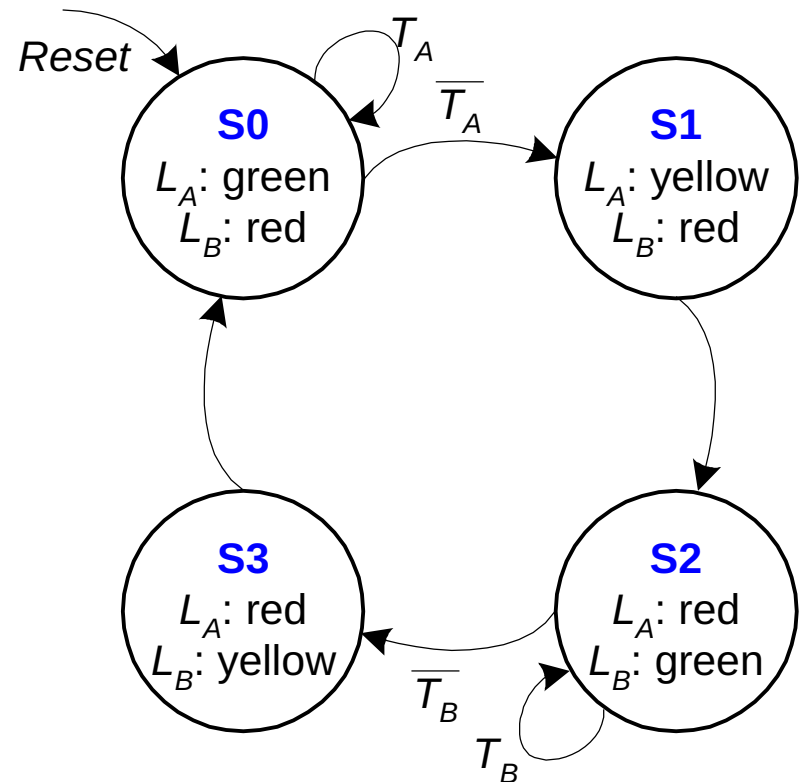
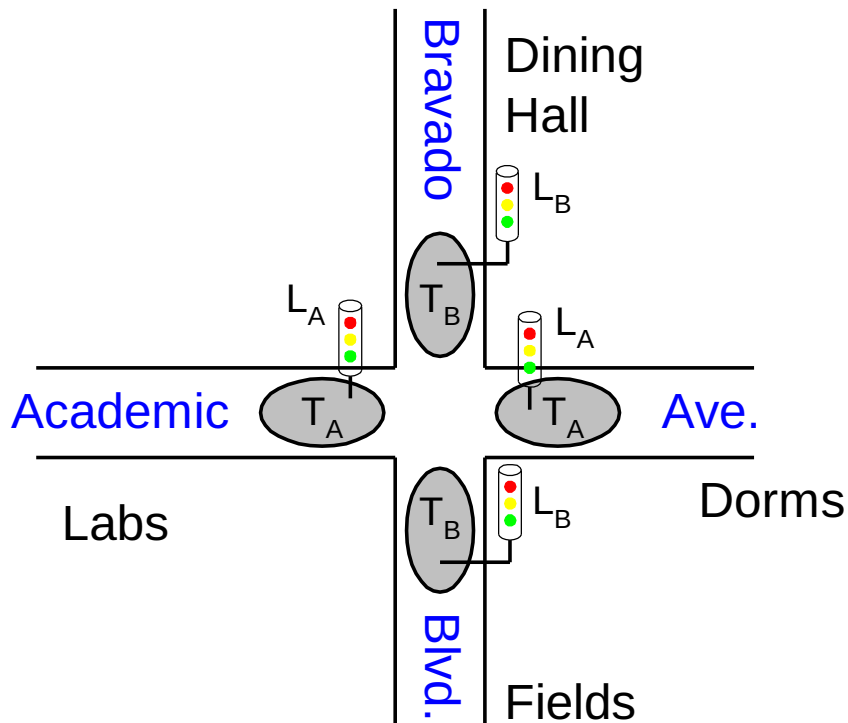
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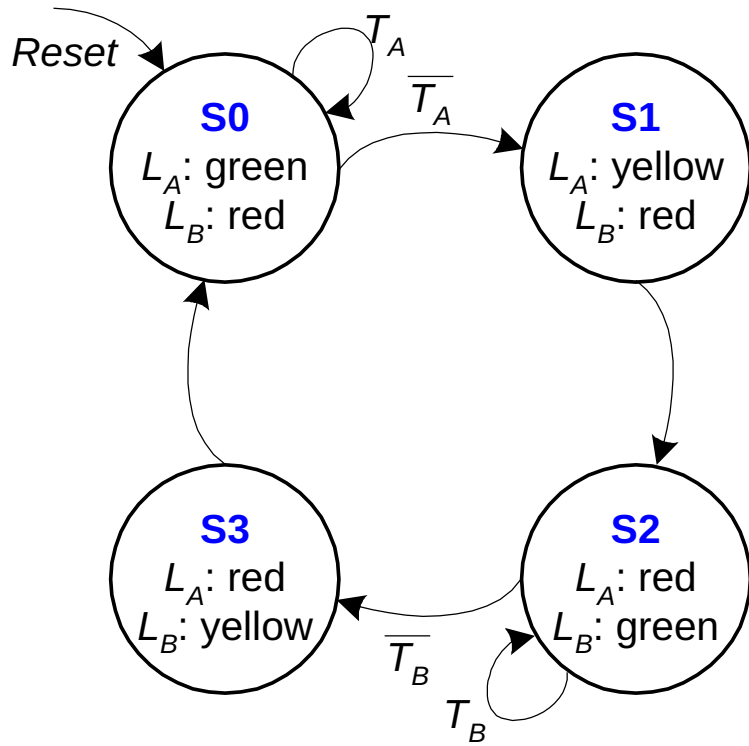
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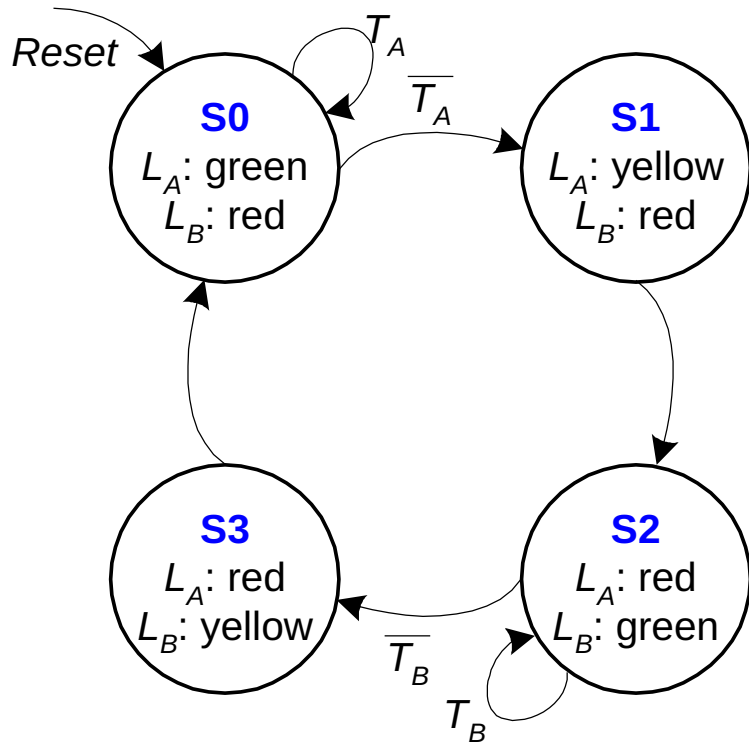
Finite State Machine: State Transition Table

FSM State Transition Table



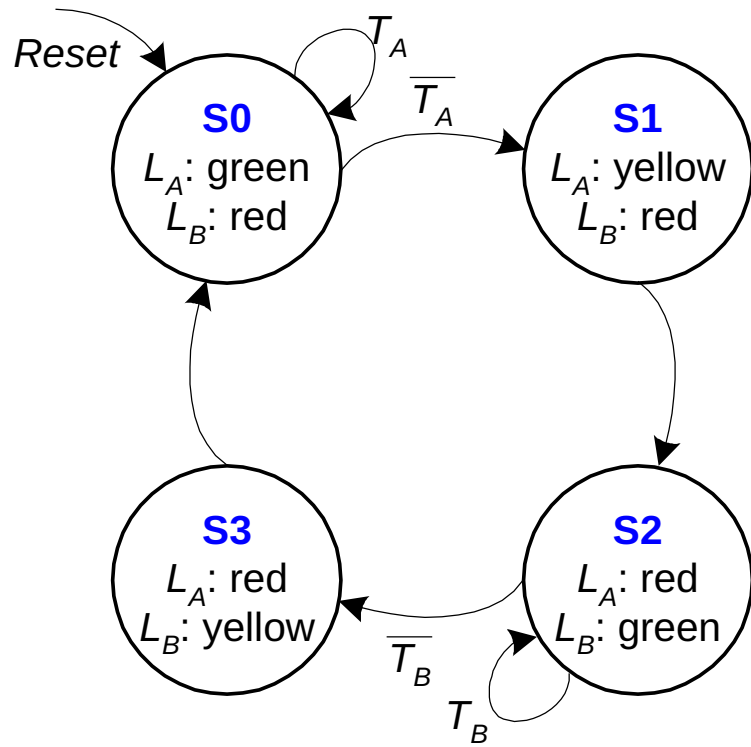
Current State	Inputs		Next State
S	T_A	T_B	S'
S0	0	X	
S0	1	X	
S1	X	X	
S2	X	0	
S2	X	1	
S3	X	X	

FSM State Transition Table



Current State	Inputs		Next State
S	T_A	T_B	S'
S0	0	X	S1
S0	1	X	S0
S1	X	X	S2
S2	X	0	S3
S2	X	1	S2
S3	X	X	S0

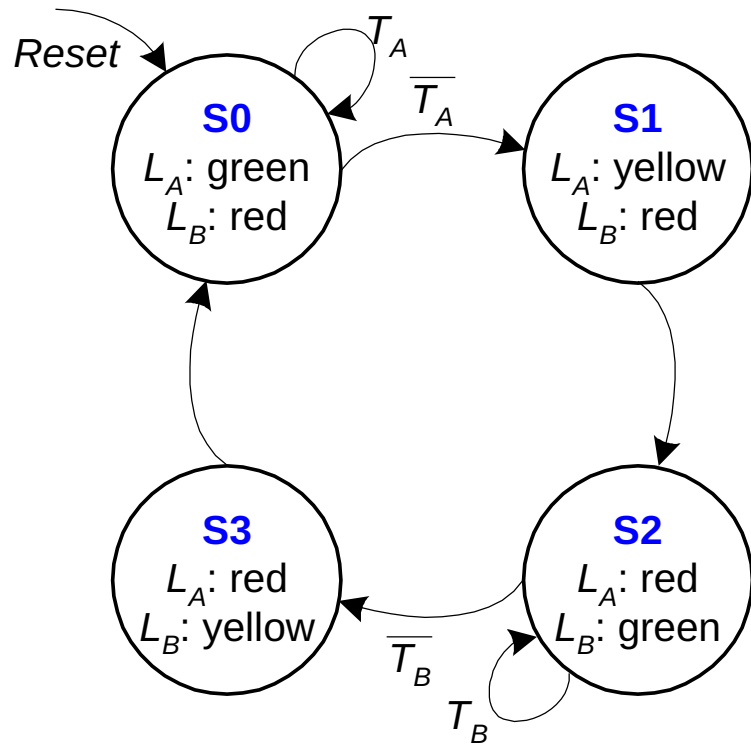
FSM State Transition Table



Current State	Inputs		Next State
S	T_A	T_B	S'
S0	0	X	S1
S0	1	X	S0
S1	X	X	S2
S2	X	0	S3
S2	X	1	S2
S3	X	X	S0

State	Encoding
S0	00
S1	01
S2	10
S3	11

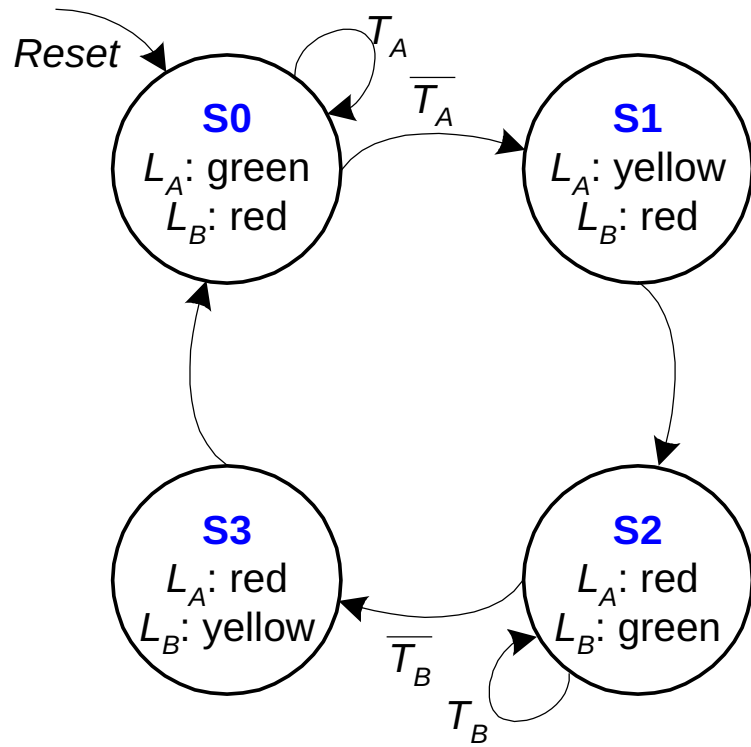
FSM State Transition Table



Current State		Inputs		Next State	
S_1	S_0	T_A	T_B	S'_1	S'_0
0	0	0	X	0	1
0	0	1	X	0	0
0	1	X	X	1	0
1	0	X	0	1	1
1	0	X	1	1	0
1	1	X	X	0	0

State	Encoding
S0	00
S1	01
S2	10
S3	11

FSM State Transition Table

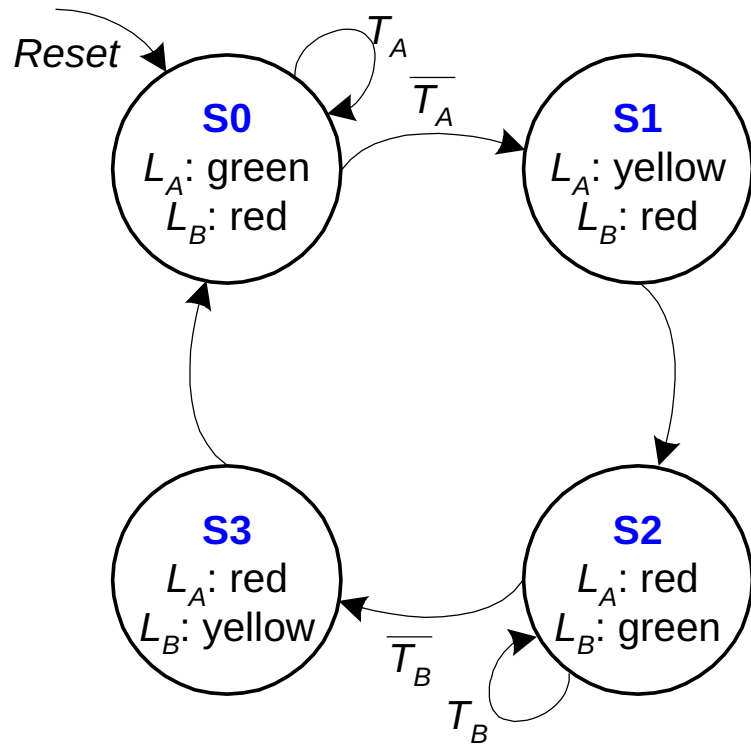


$S'_1 = ?$

Current State		Inputs		Next State	
S_1	S_0	T_A	T_B	S'_1	S'_0
0	0	0	X	0	1
0	0	1	X	0	0
0	1	X	X	1	0
1	0	X	0	1	1
1	0	X	1	1	0
1	1	X	X	0	0

State	Encoding
S0	00
S1	01
S2	10
S3	11

FSM State Transition Table

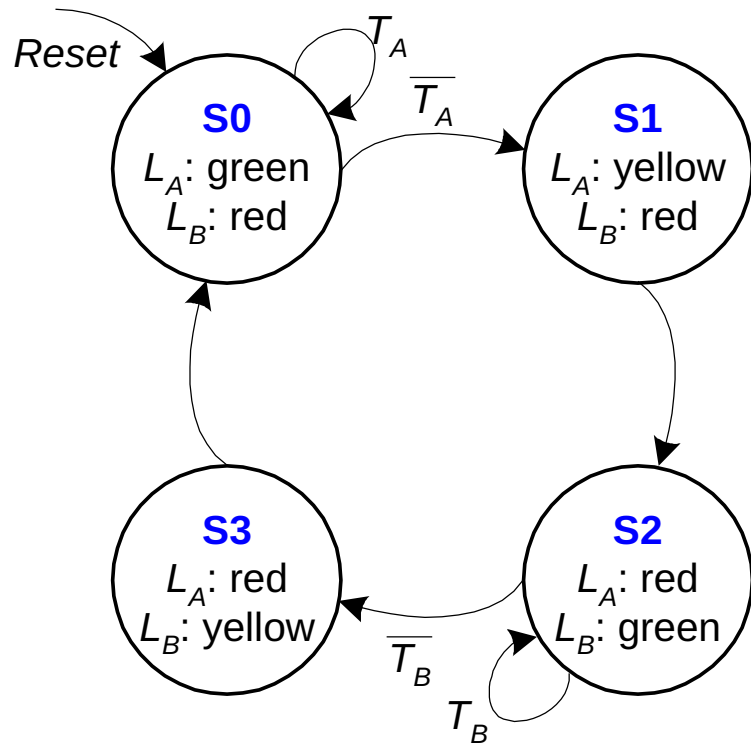


Current State		Inputs		Next State	
S_1	S_0	T_A	T_B	S'_1	S'_0
0	0	0	X	0	1
0	0	1	X	0	0
0	1	X	X	1	0
1	0	X	0	1	1
1	0	X	1	1	0
1	1	X	X	0	0

State	Encoding
S0	00
S1	01
S2	10
S3	11

$$S'_1 = (\overline{S_1} \cdot S_0) + (S_1 \cdot \overline{S_0} \cdot \overline{T_B}) + (S_1 \cdot \overline{S_0} \cdot T_B)$$

FSM State Transition Table



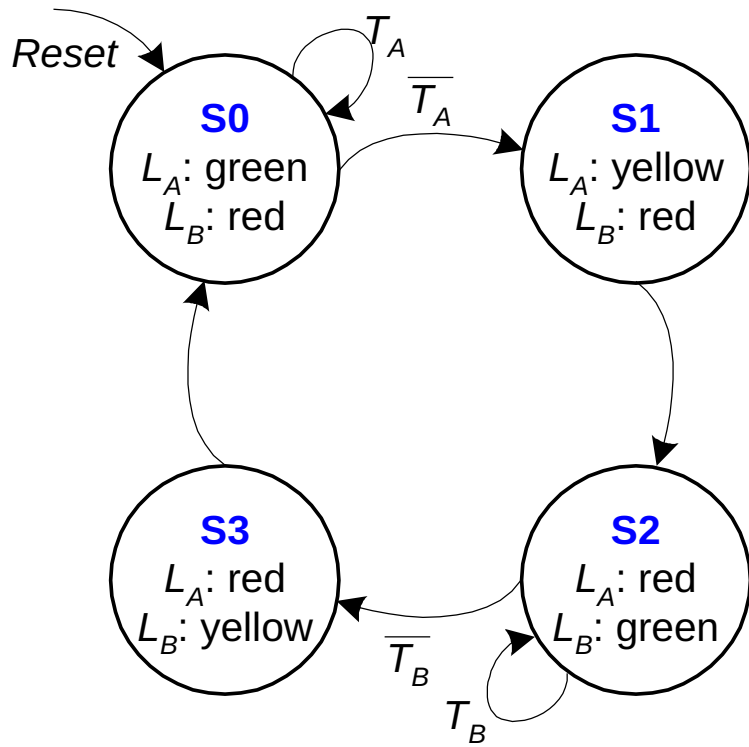
Current State		Inputs		Next State	
S_1	S_0	T_A	T_B	S'_1	S'_0
0	0	0	X	0	1
0	0	1	X	0	0
0	1	X	X	1	0
1	0	X	0	1	1
1	0	X	1	1	0
1	1	X	X	0	0

State	Encoding
S0	00
S1	01
S2	10
S3	11

$$S'_1 = (\overline{S_1} \cdot S_0) + (S_1 \cdot \overline{S_0} \cdot \overline{T_B}) + (S_1 \cdot \overline{S_0} \cdot T_B)$$

$$S'_0 = ?$$

FSM State Transition Table



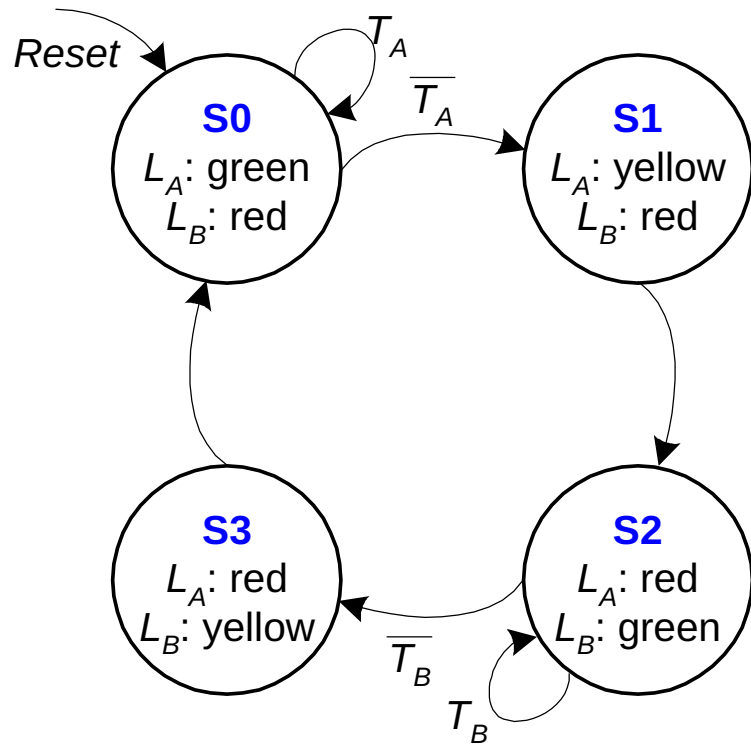
Current State		Inputs		Next State	
S_1	S_0	T_A	T_B	S'_1	S'_0
0	0	0	X	0	1
0	0	1	X	0	0
0	1	X	X	1	0
1	0	X	0	1	1
1	0	X	1	1	0
1	1	X	X	0	0

State	Encoding
S0	00
S1	01
S2	10
S3	11

$$S'_1 = (\overline{S_1} \cdot S_0) + (S_1 \cdot \overline{S_0} \cdot \overline{T_B}) + (S_1 \cdot \overline{S_0} \cdot T_B)$$

$$S'_0 = (\overline{S_1} \cdot \overline{S_0} \cdot \overline{T_A}) + (S_1 \cdot \overline{S_0} \cdot \overline{T_B})$$

FSM State Transition Table



Current State		Inputs		Next State	
S_1	S_0	T_A	T_B	S'_1	S'_0
0	0	0	X	0	1
0	0	1	X	0	0
0	1	X	X	1	0
1	0	X	0	1	1
1	0	X	1	1	0
1	1	X	X	0	0

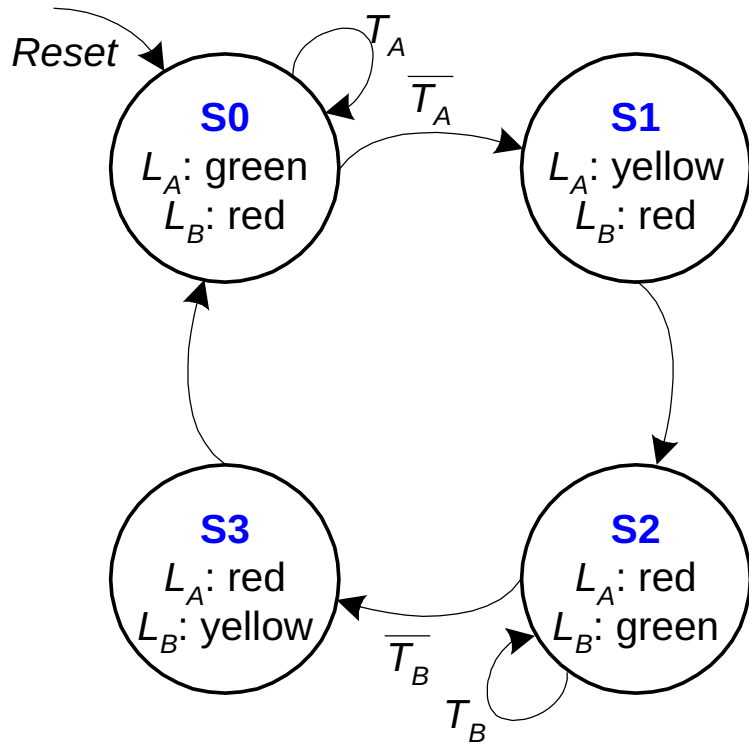
State	Encoding
S0	00
S1	01
S2	10
S3	11

$$S'_1 = S_1 \text{ xor } S_0 \quad \textbf{(Simplified)}$$

$$S'_0 = (\overline{S_1} \cdot \overline{S_0} \cdot \overline{T_A}) + (S_1 \cdot \overline{S_0} \cdot \overline{T_B})$$

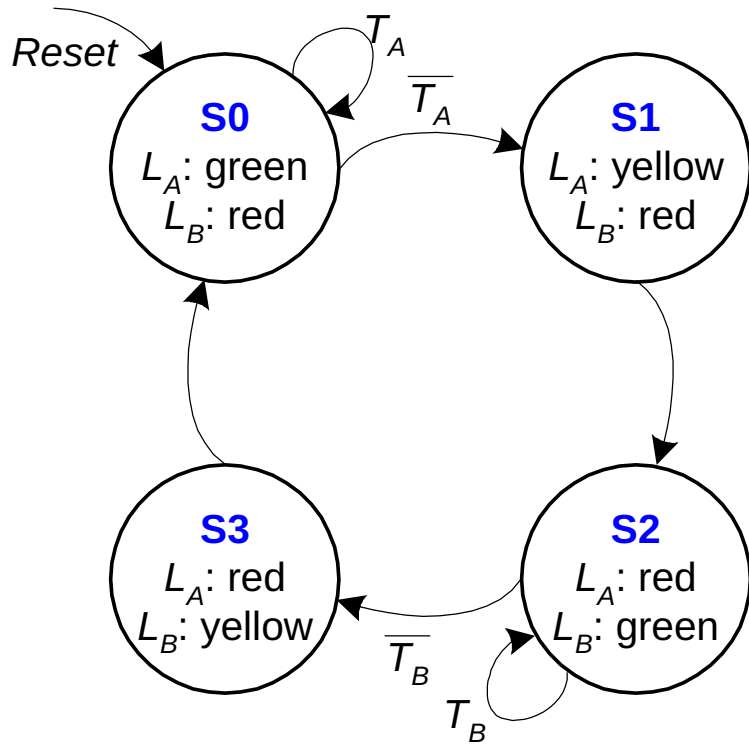
Finite State Machine: Output Table

FSM Output Table



Current State		Outputs	
S_1	S_0	L_A	L_B
0	0	green	red
0	1	yellow	red
1	0	red	green
1	1	red	yellow

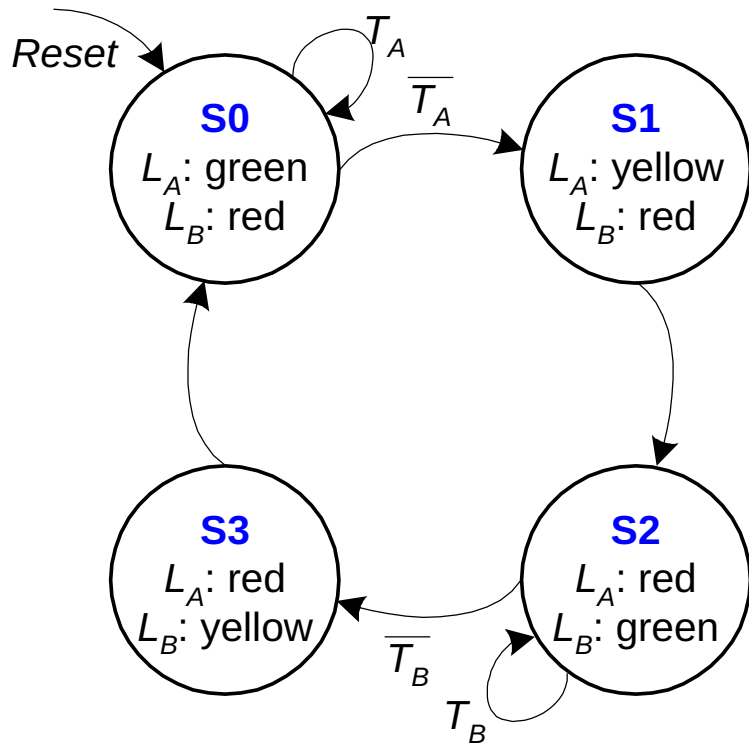
FSM Output Table



Current State		Outputs	
S_1	S_0	L_A	L_B
0	0	green	red
0	1	yellow	red
1	0	red	green
1	1	red	yellow

Output	Encoding
green	00
yellow	01
red	10

FSM Output Table

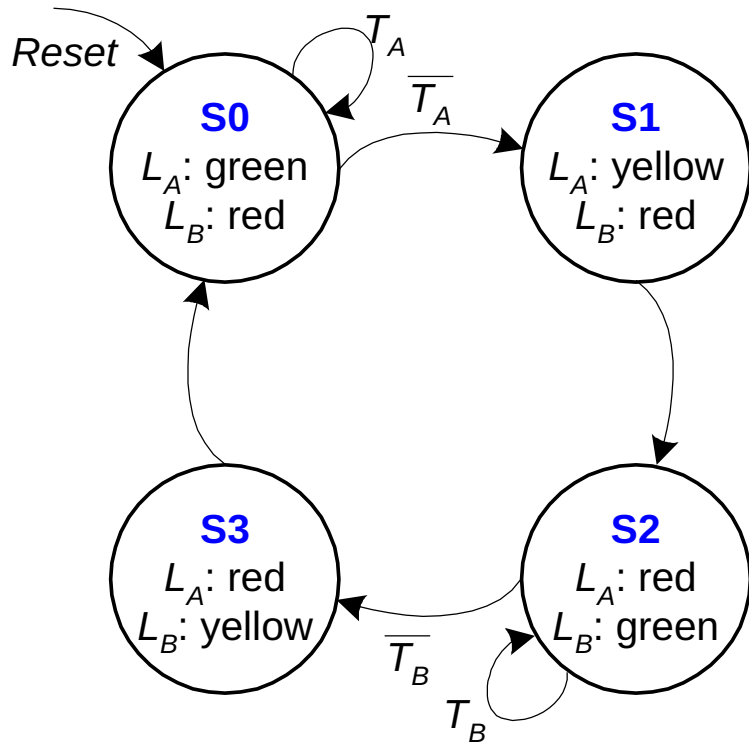


$$L_{A1} = S_1$$

Current State		Outputs			
S_1	S_0	L_{A1}	L_{A0}	L_{B1}	L_{B0}
0	0	0	0	1	0
0	1	0	1	1	0
1	0	1	0	0	0
1	1	1	0	0	1

Output	Encoding
green	00
yellow	01
red	10

FSM Output Table



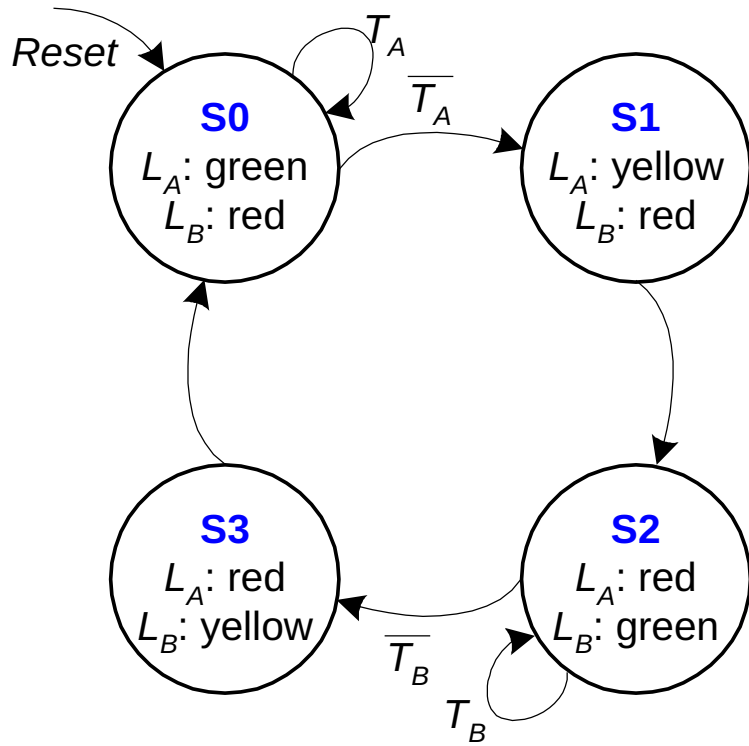
$$L_{A1} = S_1$$

$$L_{A0} = \overline{S_1} \cdot S_0$$

Current State		Outputs			
S_1	S_0	L_{A1}	L_{A0}	L_{B1}	L_{B0}
0	0	0	0	1	0
0	1	0	1	1	0
1	0	1	0	0	0
1	1	1	0	0	1

Output	Encoding
green	00
yellow	01
red	10

FSM Output Table

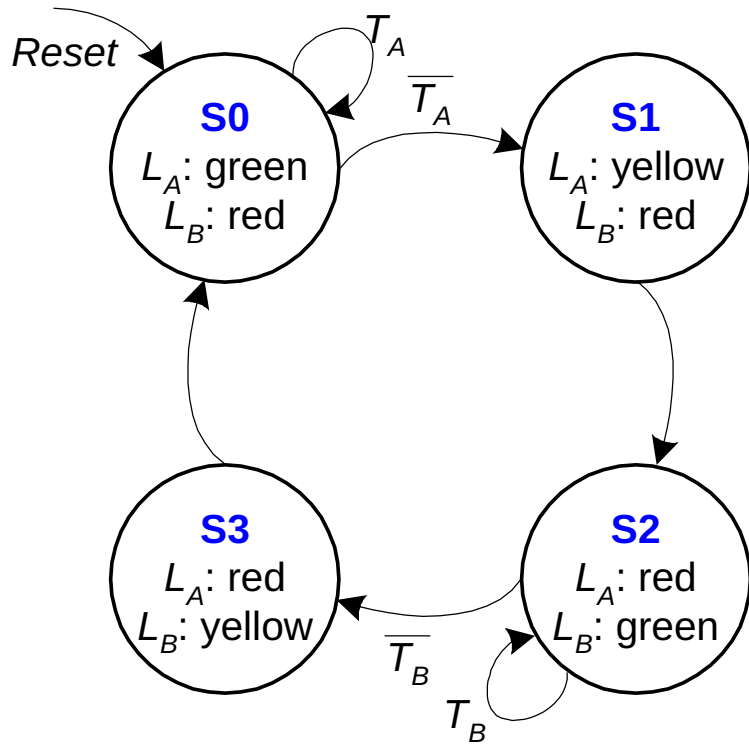


$$\begin{aligned}
 L_{A1} &= S_1 \\
 L_{A0} &= \overline{S_1} \cdot S_0 \\
 L_{B1} &= \overline{S_1}
 \end{aligned}$$

Current State		Outputs			
S_1	S_0	L_{A1}	L_{A0}	L_{B1}	L_{B0}
0	0	0	0	1	0
0	1	0	1	1	0
1	0	1	0	0	0
1	1	1	0	0	1

Output	Encoding
green	00
yellow	01
red	10

FSM Output Table



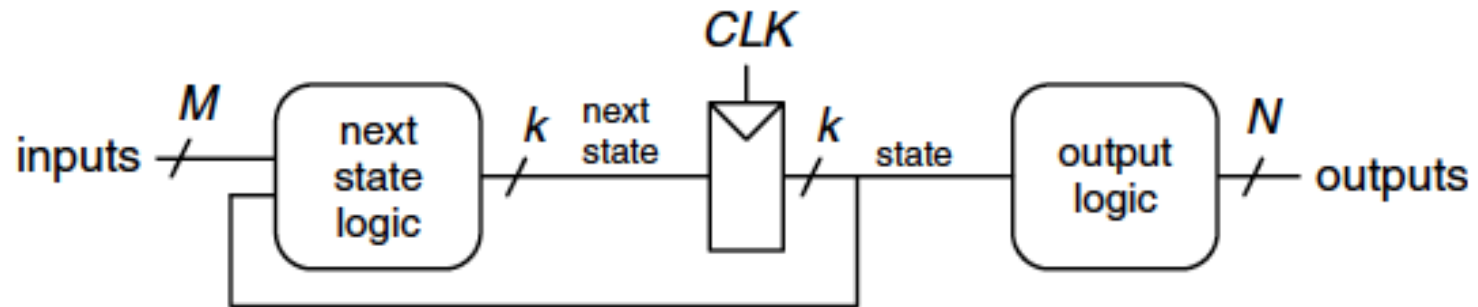
Current State		Outputs			
S_1	S_0	L_{A1}	L_{A0}	L_{B1}	L_{B0}
0	0	0	0	1	0
0	1	0	1	1	0
1	0	1	0	0	0
1	1	1	0	0	1

$$\begin{aligned}
 L_{A1} &= S_1 \\
 L_{A0} &= \overline{S_1} \cdot S_0 \\
 L_{B1} &= \overline{S_1} \\
 L_{B0} &= S_1 \cdot S_0
 \end{aligned}$$

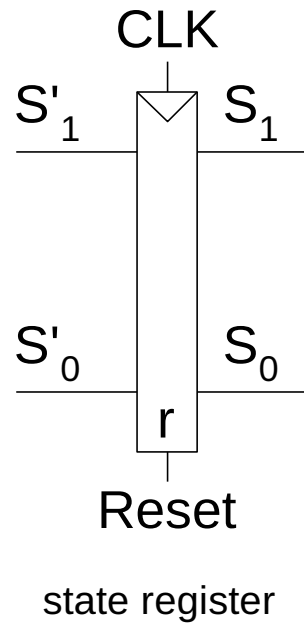
Output	Encoding
green	00
yellow	01
red	10

Finite State Machine: Schematic

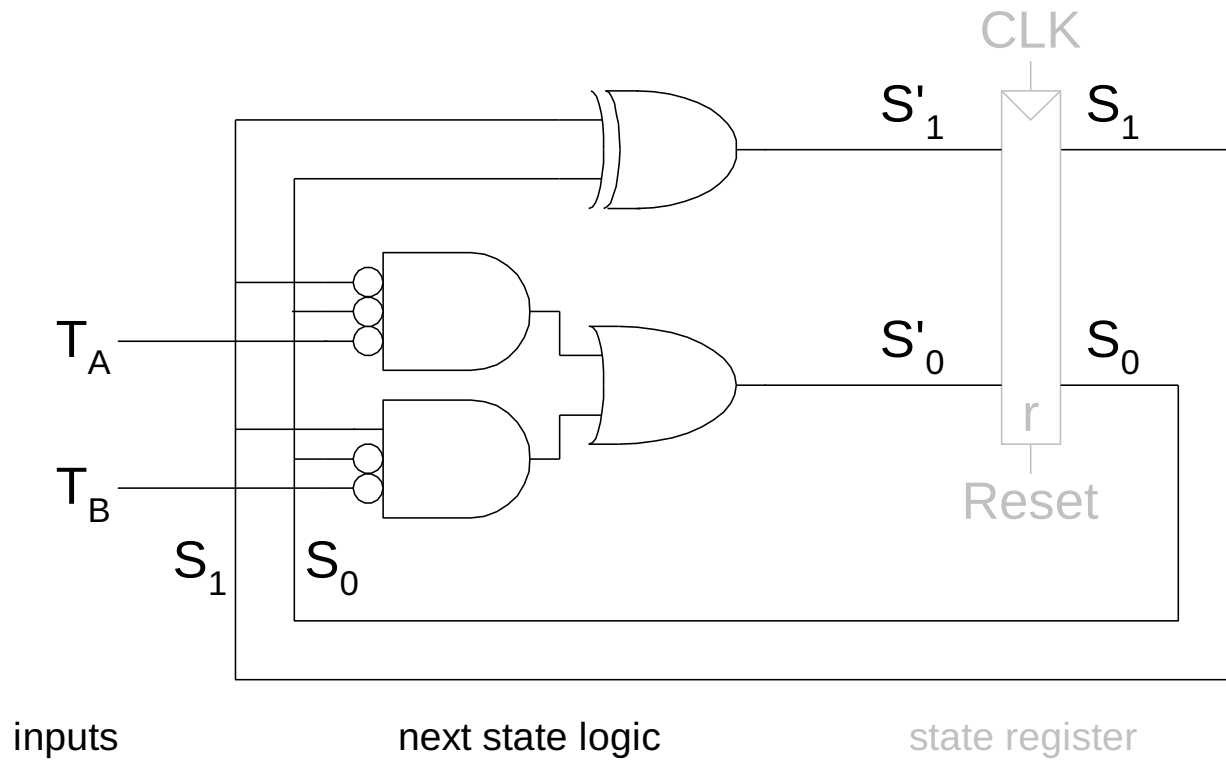
FSM Schematic: State Register



FSM Schematic: State Register



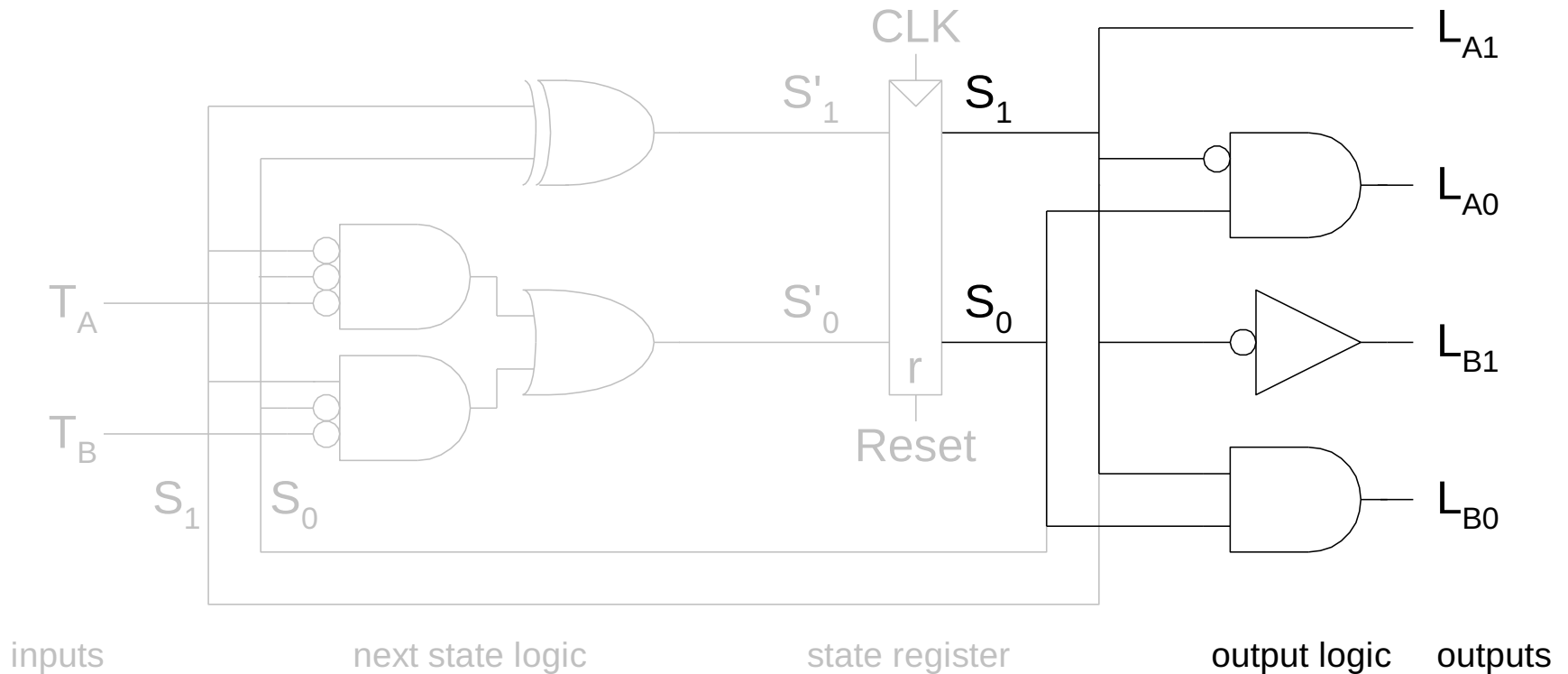
FSM Schematic: Next State Logic



$$S'_1 = S_1 \text{ xor } S_0$$

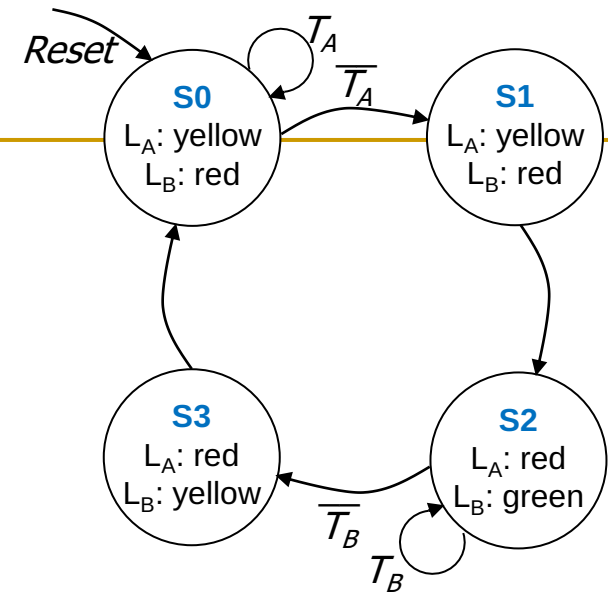
$$S'_0 = (\overline{S_1} \cdot \overline{S_0} \cdot \overline{T_A}) + (S_1 \cdot \overline{S_0} \cdot \overline{T_B})$$

FSM Schematic: Output Logic



$$\begin{aligned} L_{A1} &= S_1 \\ L_{A0} &= \overline{S_1} \cdot S_0 \\ L_{B1} &= \overline{S_1} \\ L_{B0} &= S_1 \cdot S_0 \end{aligned}$$

FSM Timing Diagram



CLK_

Reset_

T_A _

T_B _

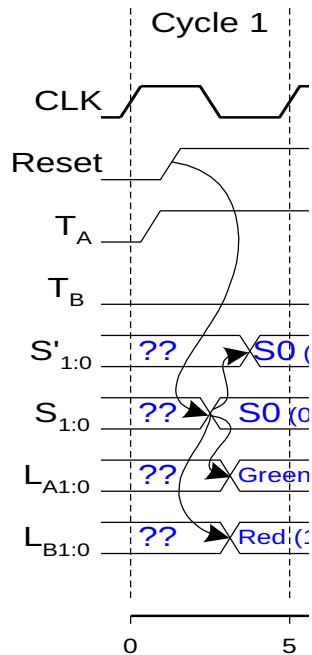
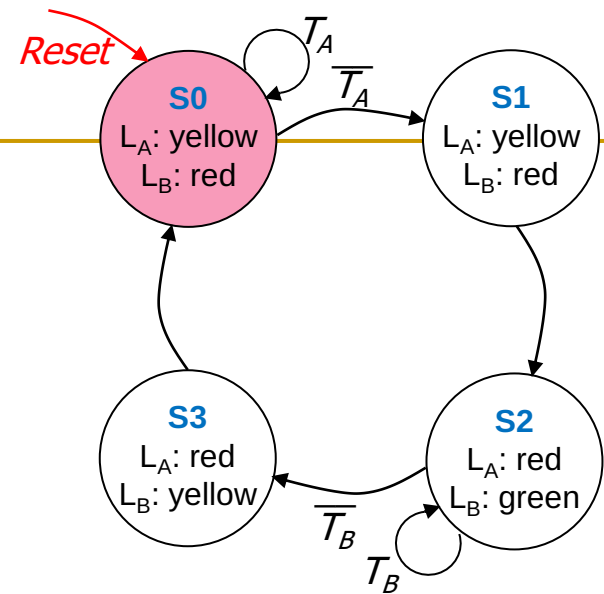
$S'_{1:0}$ _

$S_{1:0}$ _

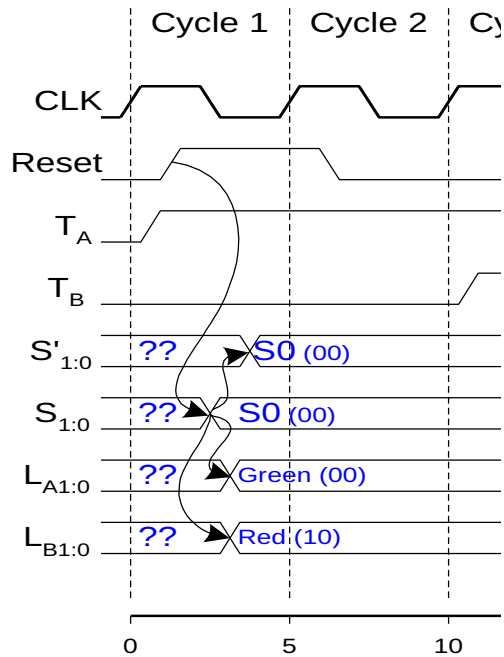
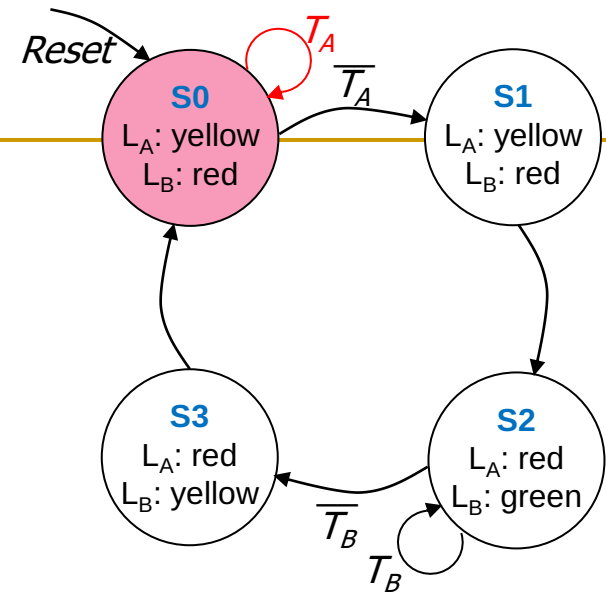
$L_{A1:0}$ _

$L_{B1:0}$ _

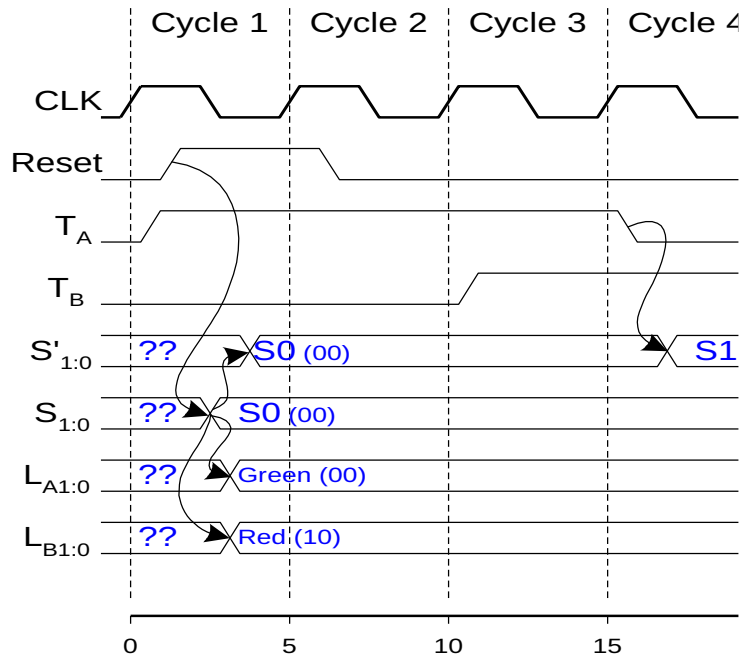
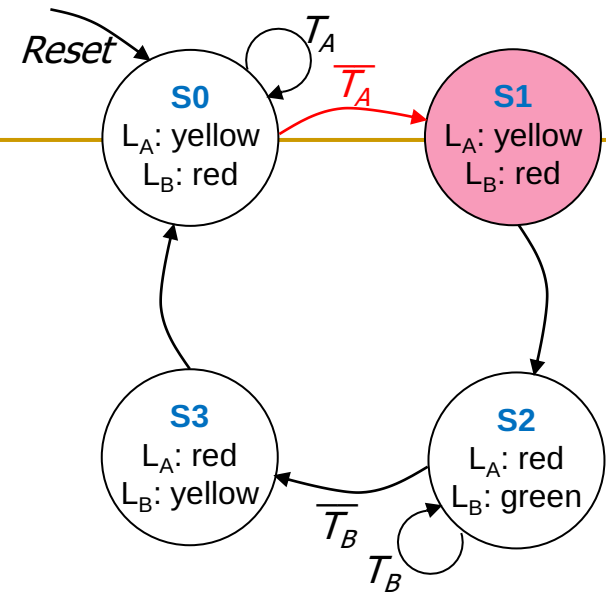
FSM Timing Diagram



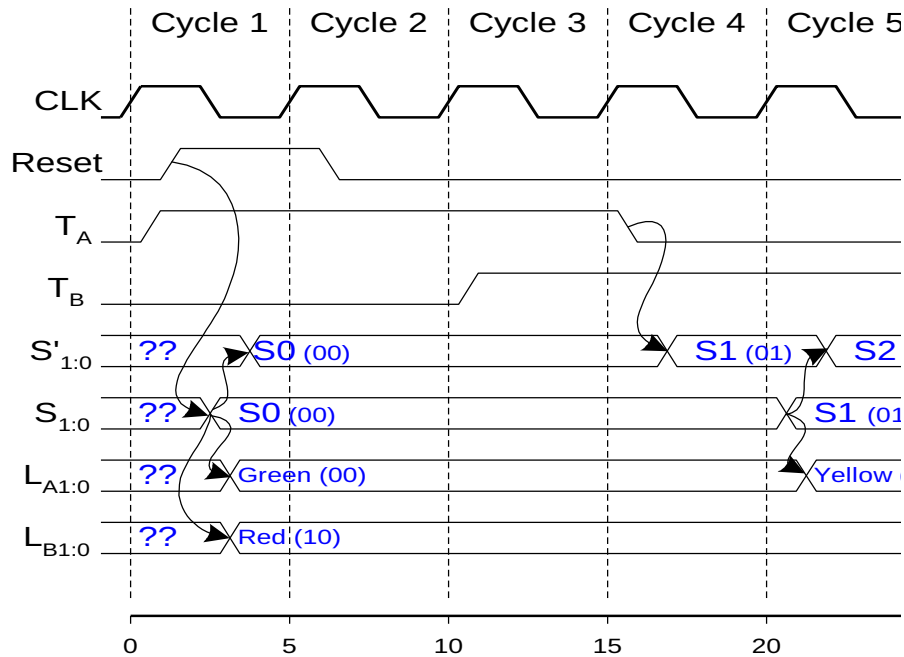
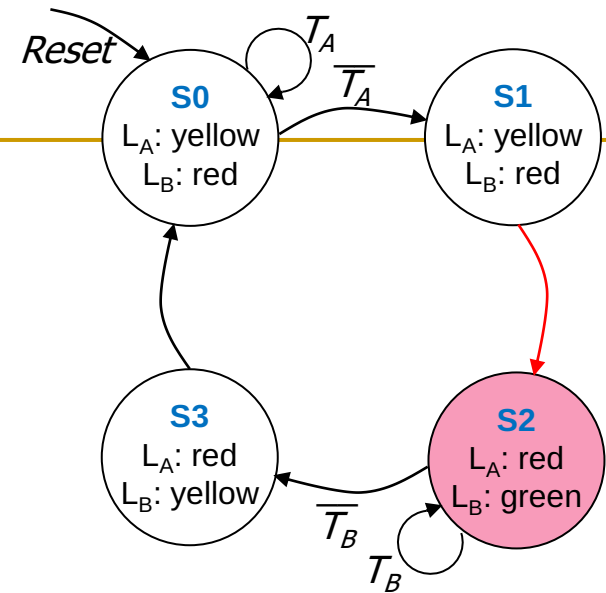
FSM Timing Diagram



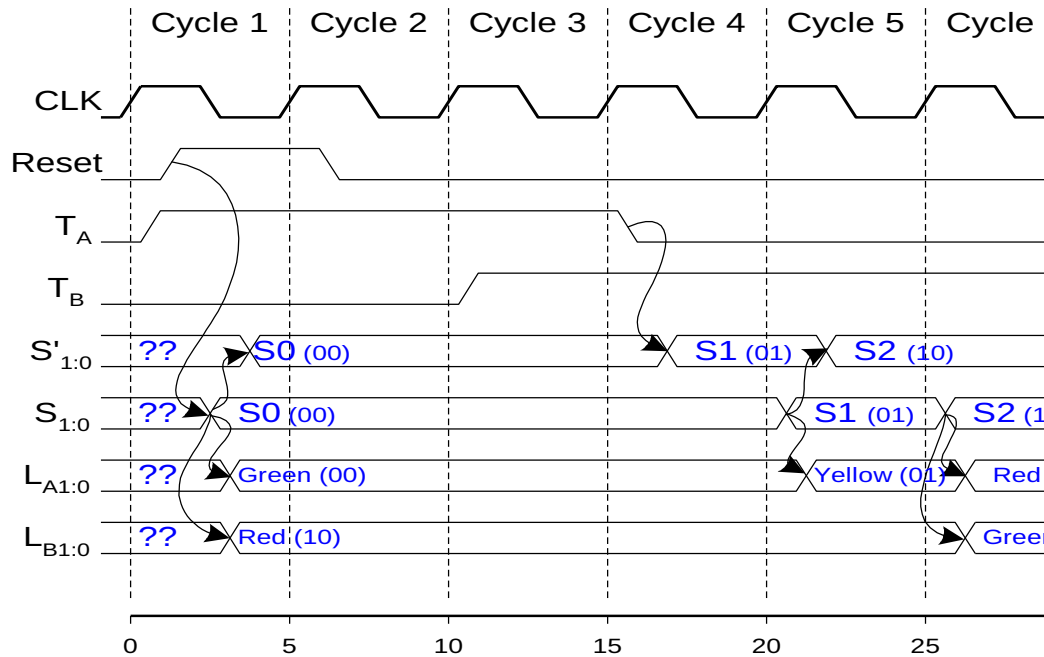
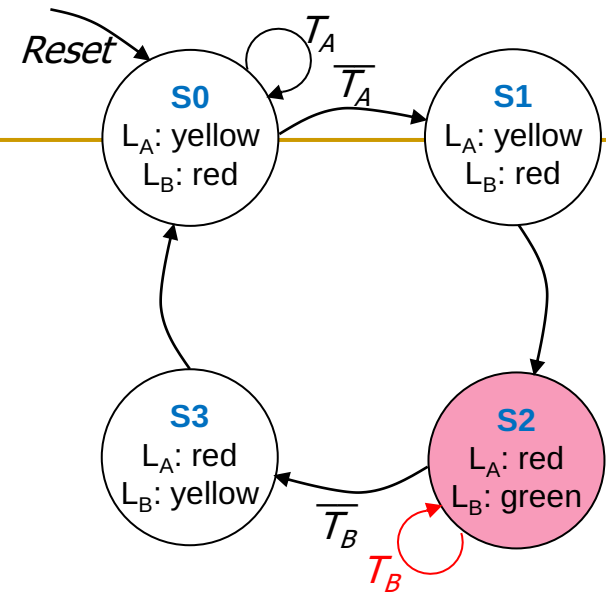
FSM Timing Diagram



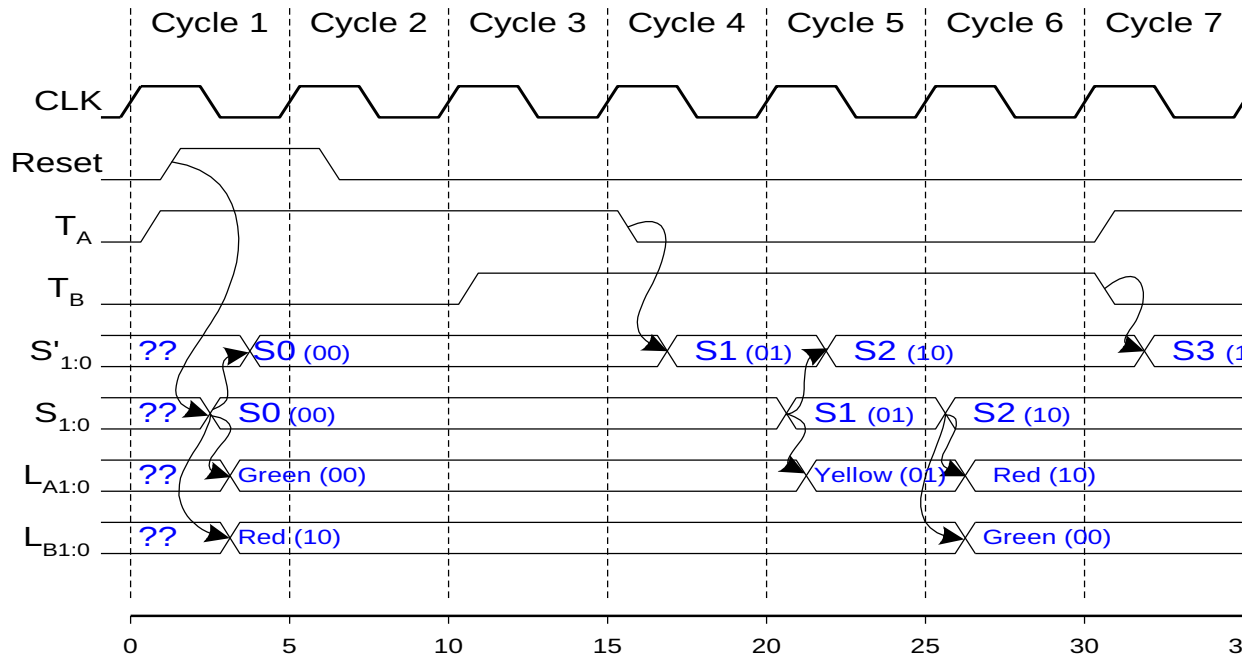
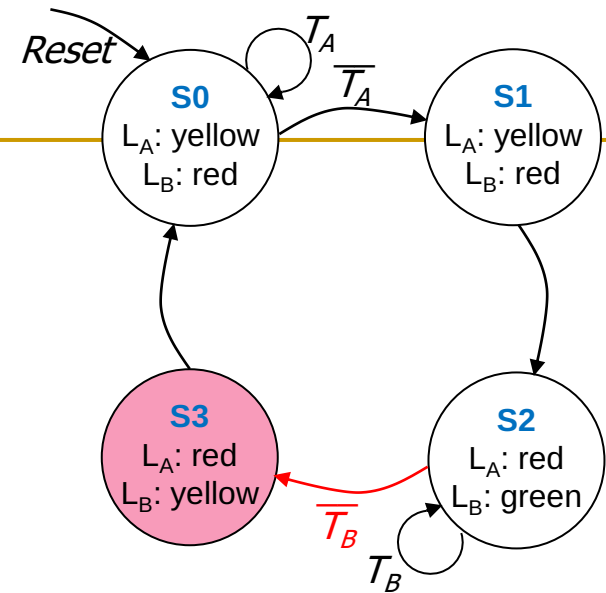
FSM Timing Diagram



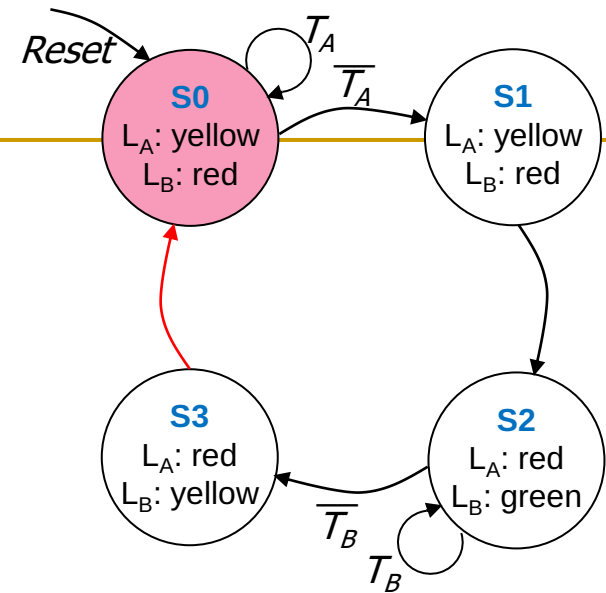
FSM Timing Diagram



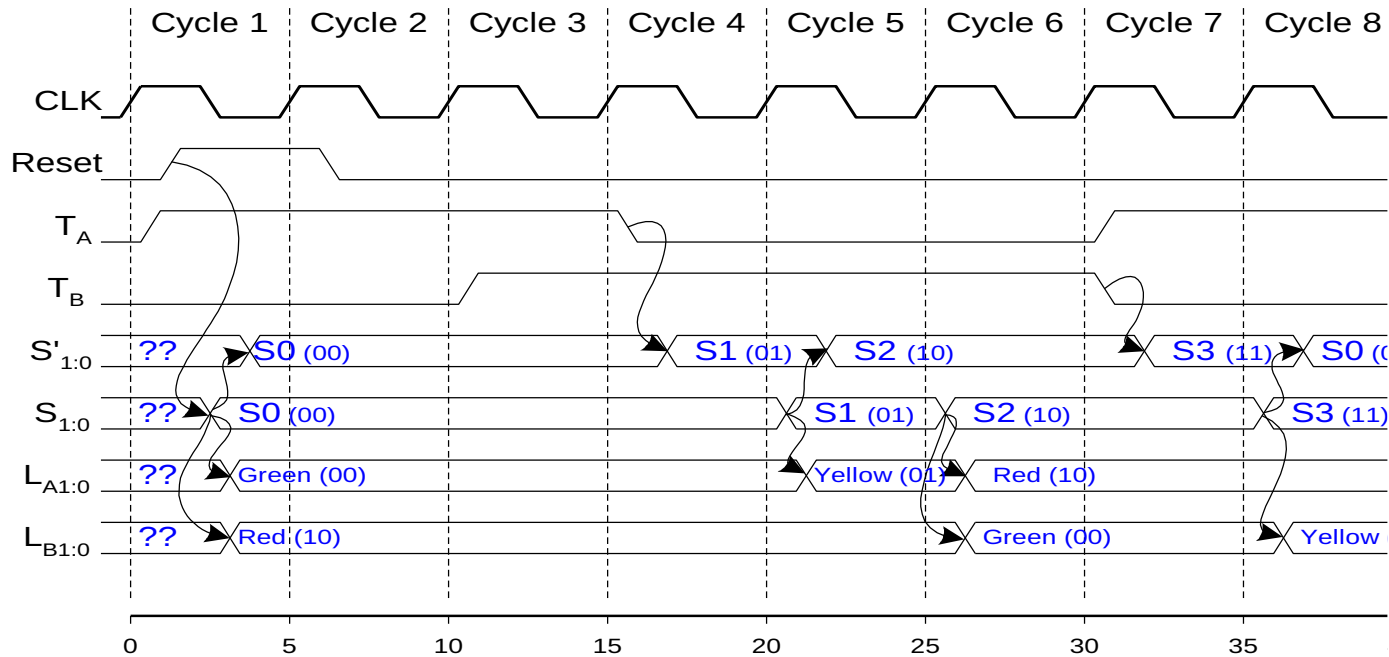
FSM Timing Diagram



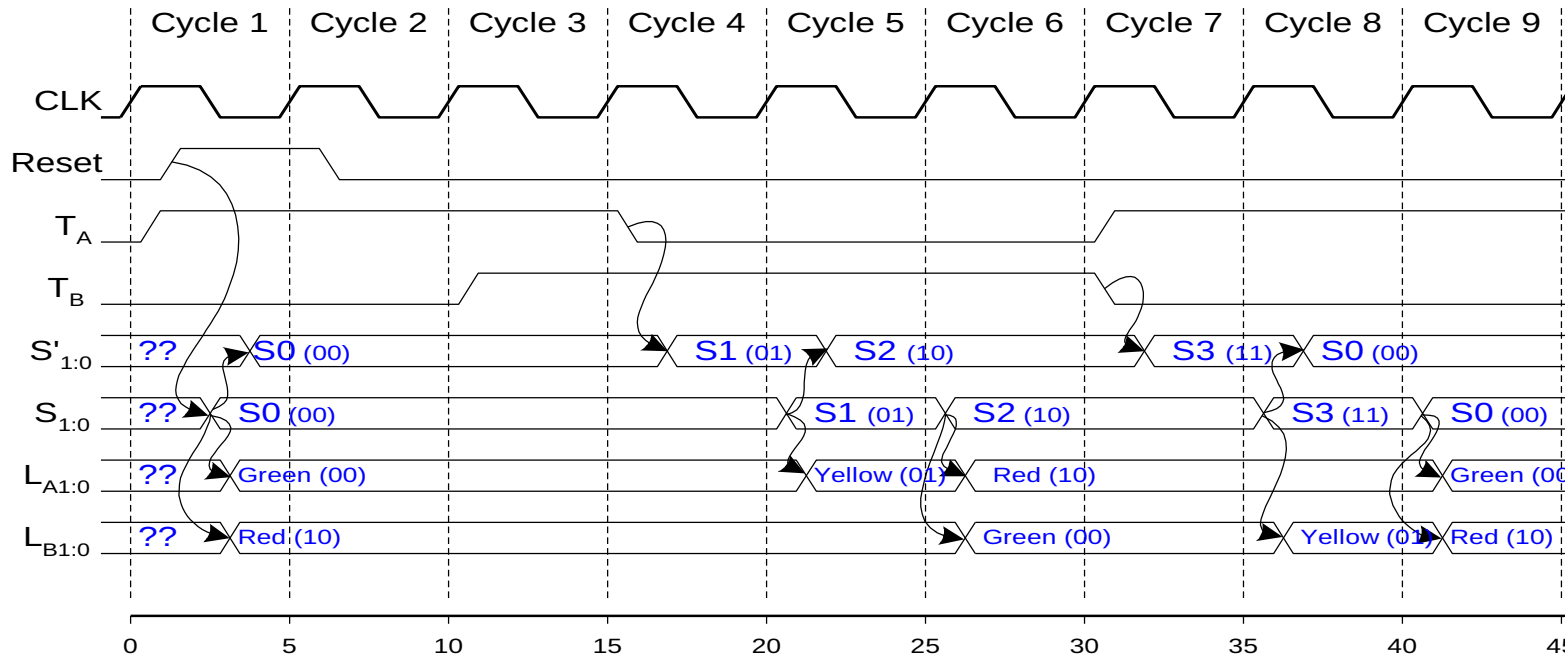
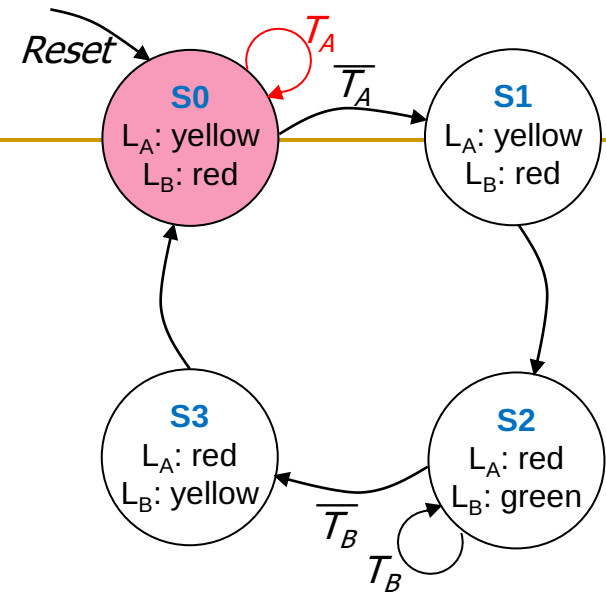
FSM Timing Diagram



This is from H&H Section 3.4.1

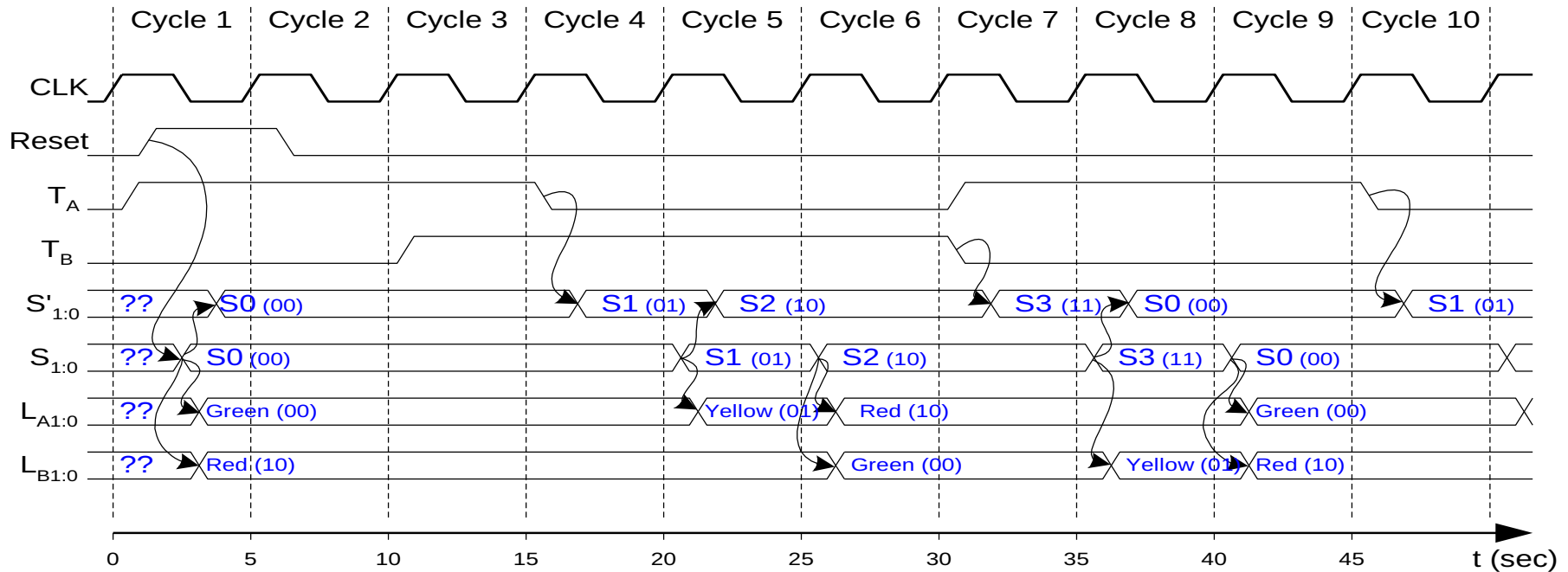
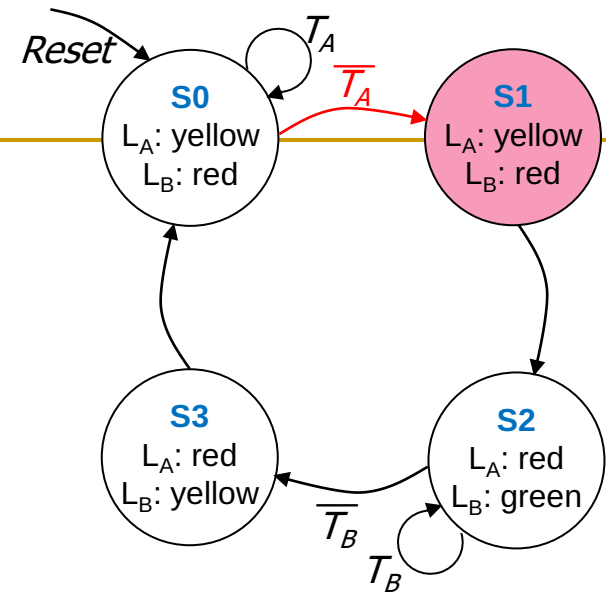


FSM Timing Diagram



FSM Timing Diagram

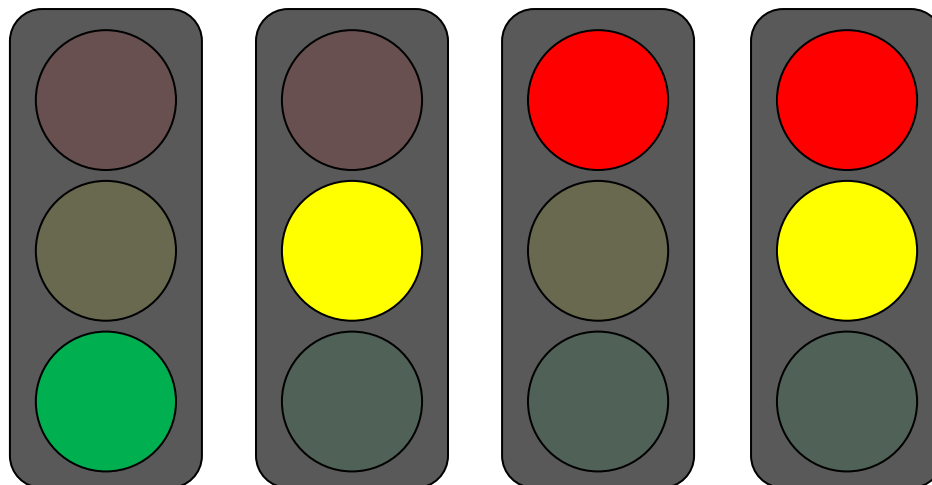
See H&H Chapter 3.4



Finite State Machine: State Encoding

FSM State Encoding

- How do we encode the state bits?
 - Three common state binary encodings with different tradeoffs
 1. **Fully Encoded**
 2. **1-Hot Encoded**
 3. **Output Encoded**
- Let's see an example **Swiss** traffic light with 4 states
 - Green, Yellow, Red, Yellow+Red



FSM State Encoding (II)

1. Binary Encoding (Full Encoding):

- Use the minimum possible number of bits
 - Use $\log_2(num_states)$ bits to represent the states
- *Example state encodings: 00, 01, 10, 11*
- **Minimizes** # flip-flops, but not necessarily output logic or next state logic

2. One-Hot Encoding:

- Each bit encodes a different state
 - Uses num_states bits to represent the states
 - Exactly 1 bit is “hot” for a given state
- *Example state encodings: 0001, 0010, 0100, 1000*
- **Simplest design process** – very automatable
- **Maximizes** # flip-flops, **minimizes** next state logic

FSM State Encoding (III)

3. Output Encoding:

- ❑ Outputs are **directly accessible** in the state encoding
- ❑ For example, since we have **3 outputs** (light color), encode state with **3 bits**, where each bit represents a color
- ❑ *Example states:* 001, 010, 100, 110
 - Bit₀ encodes **green** light output,
 - Bit₁ encodes **yellow** light output
 - Bit₂ encodes **red** light output
- ❑ **Minimizes** output logic
- ❑ Only works for Moore Machines (output function of state)

FSM State Encoding (III)

3. Output Encoding:

- ❑ Outputs are **directly accessible** in the state encoding

The **designer** must **carefully** choose an encoding scheme to **optimize** the design under given constraints

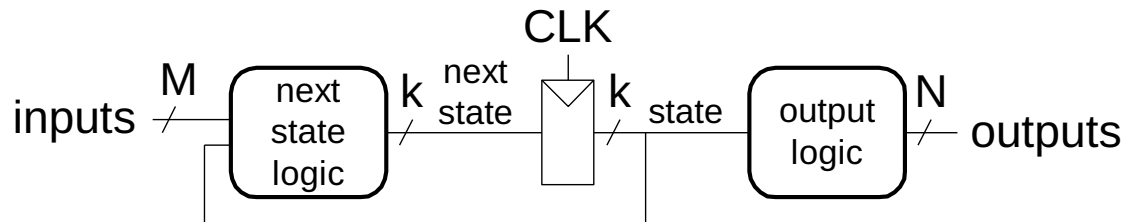
- ❑ **Minimizes** output logic
- ❑ Only works for Moore Machines (output depends only on state)

Moore vs. Mealy Machines

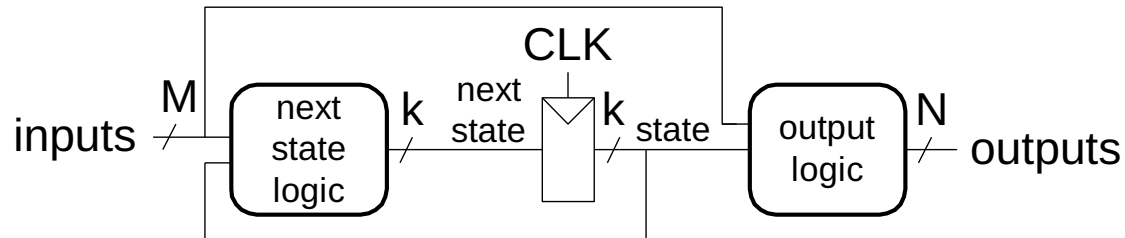
Recall: Moore vs. Mealy FSMs

- Next state is determined by the current state and the inputs
- Two types of FSMs differ in the **output logic**:
 - **Moore FSM**: outputs depend only on the current state
 - **Mealy FSM**: outputs depend on the current state and the inputs

Moore FSM



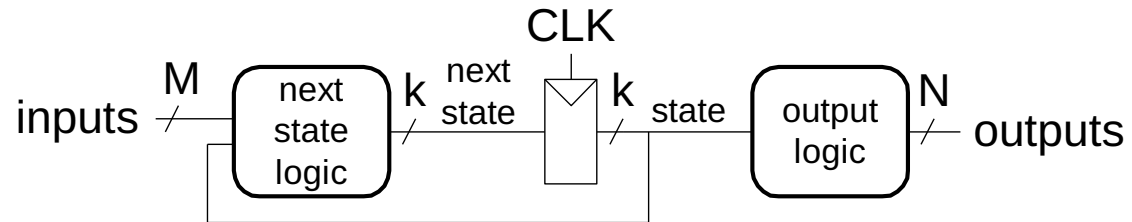
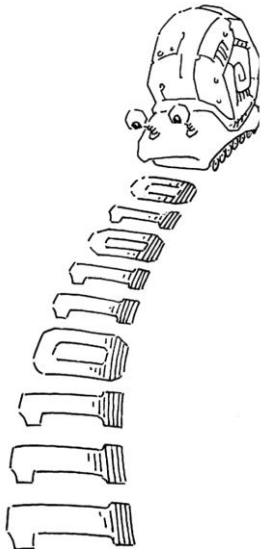
Mealy FSM



Moore vs. Mealy FSM Examples

- Alyssa P. Hacker has a snail that crawls down a paper tape with 1's and 0's on it.
- The snail smiles whenever the last four digits it has crawled over are **1101**.
- Design Moore and Mealy FSMs of the snail's brain.

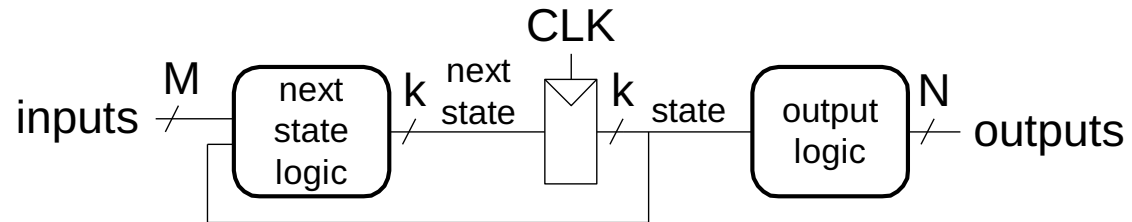
Moore FSM



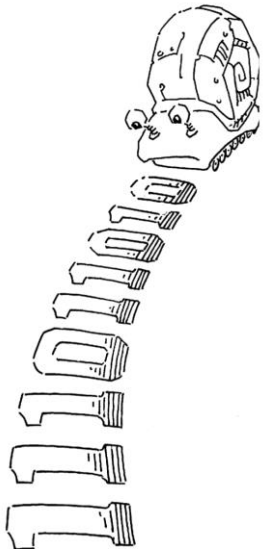
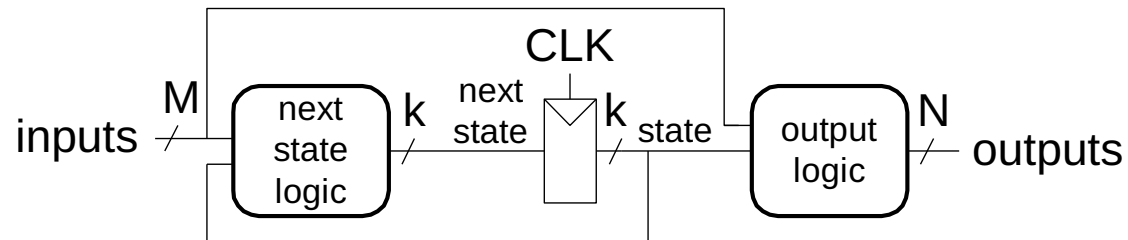
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Moore FSM

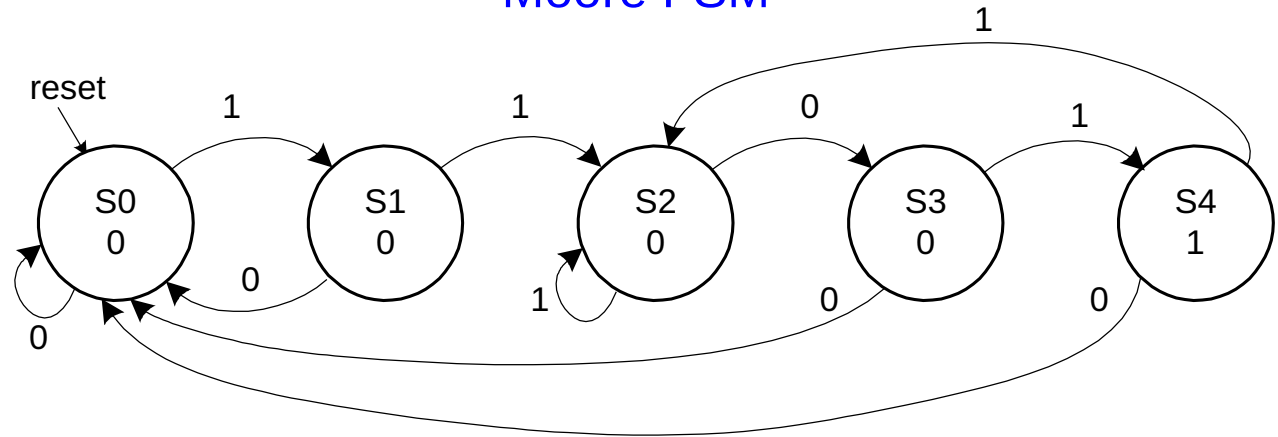


Mealy FSM



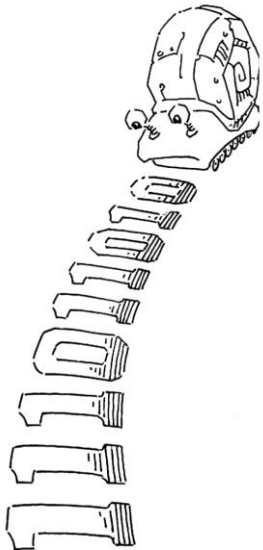
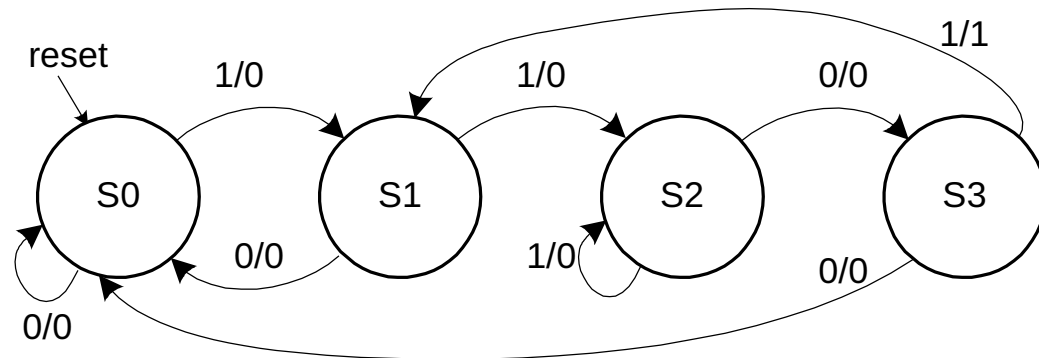
State Transition Diagrams

Moore FSM



What are the tradeoffs?

Mealy FSM



FSM Design Procedure

- **Determine** all possible states of your machine
- **Develop** a **state transition diagram**
 - Generally this is done from a textual description
 - You need to 1) determine the **inputs** and **outputs** for each **state** and 2) figure out how to get from one state to another
- **Approach**
 - Start by defining the **reset state** and what happens from it – this is typically an easy point to start from
 - Then continue to add **transitions** and **states**
 - Picking **good state names** is very important
 - Building an FSM is **like** programming (but it *is not* programming!)
 - An FSM has a sequential “control-flow” like a program with conditionals and goto’s
 - The if-then-else construct is controlled by one or more inputs
 - The outputs are controlled by the state or the inputs
 - In hardware, we typically have many concurrent FSMs

What is to Come: LC-3 Processor

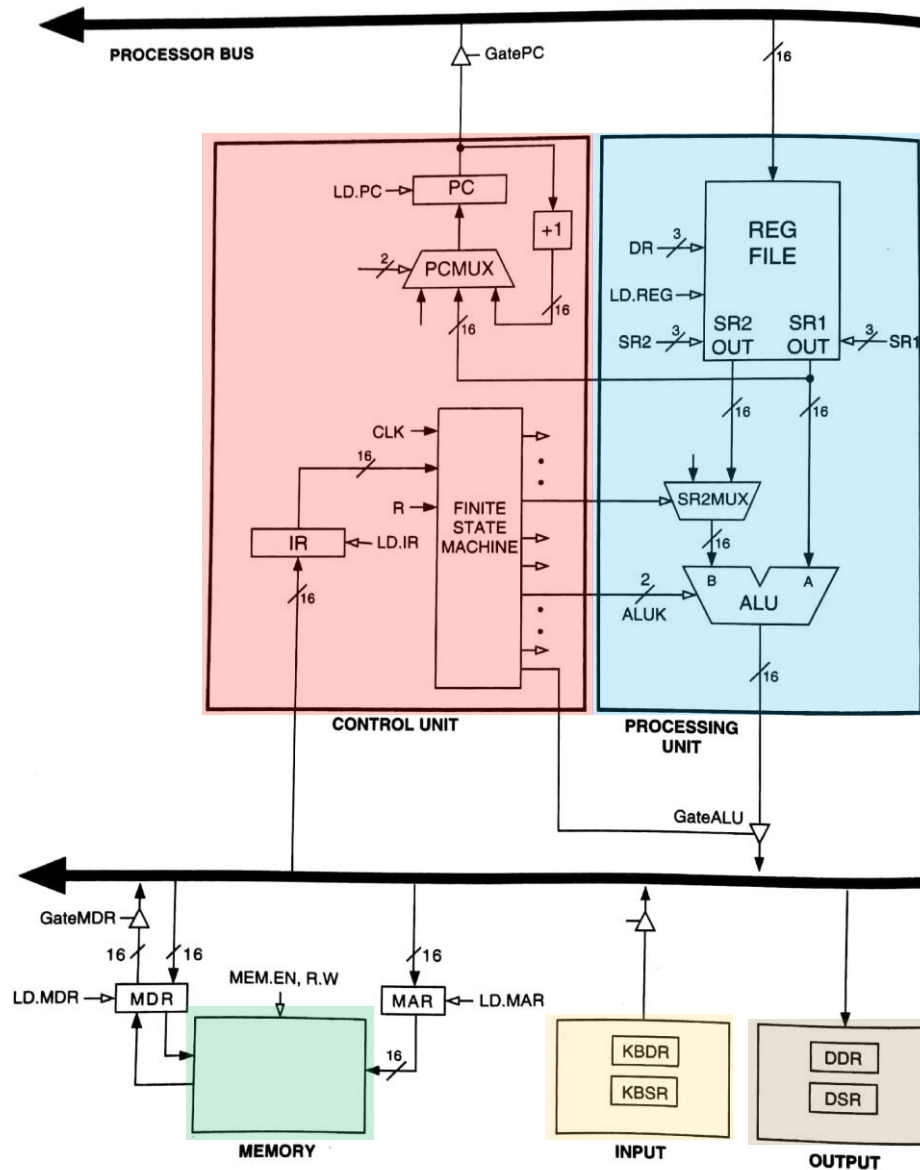
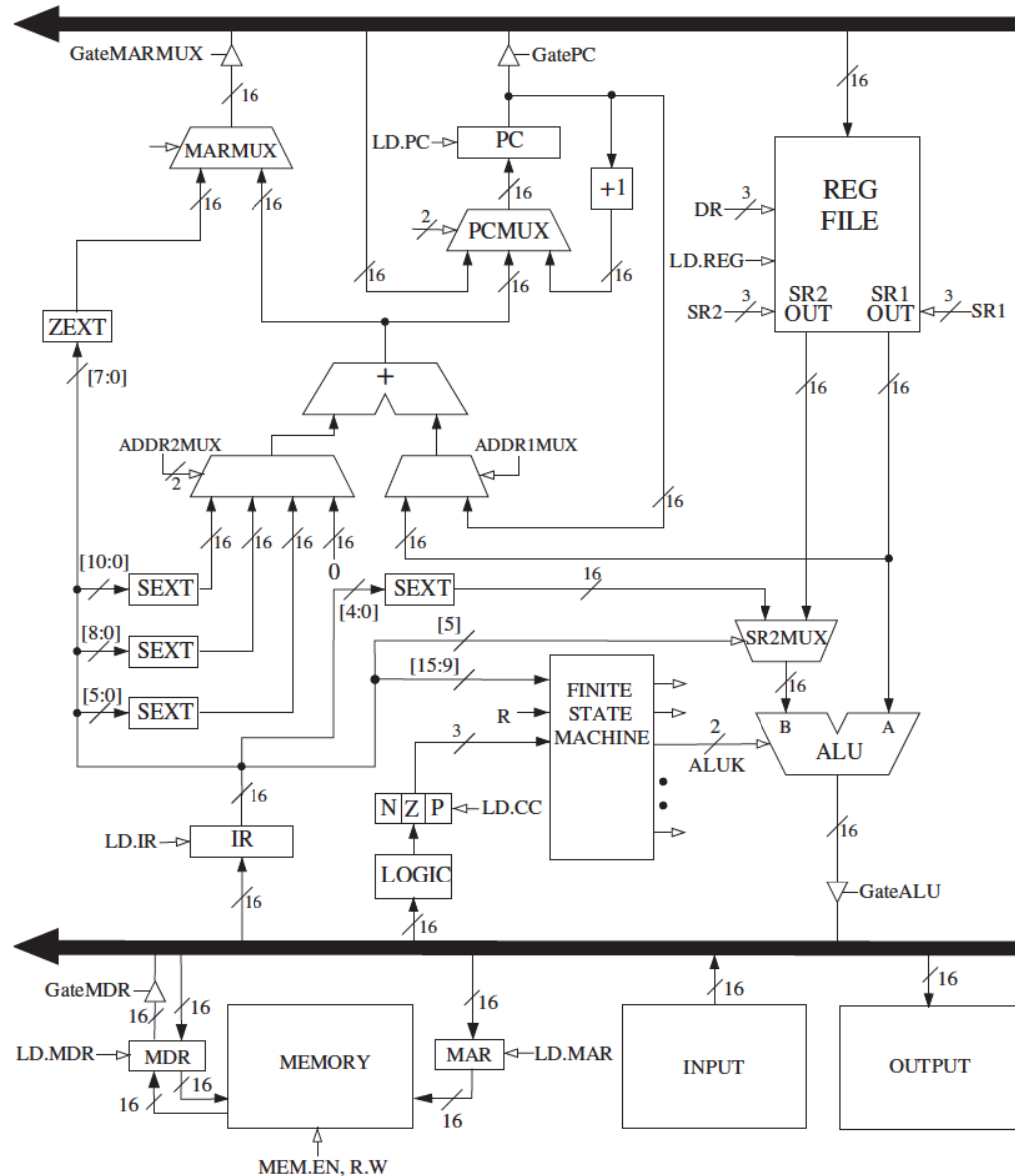


Figure 4.3 The LC-3 as an example of the von Neumann model

What is to Come: LC-3 Datapath



Digital Design & Computer Arch.

Lecture 5a: Sequential Logic Design II Finite State Machines

Prof. Onur Mutlu

ETH Zürich

Spring 2023

9 March 2023

Backup Slides:

Different Flip-Flop Types

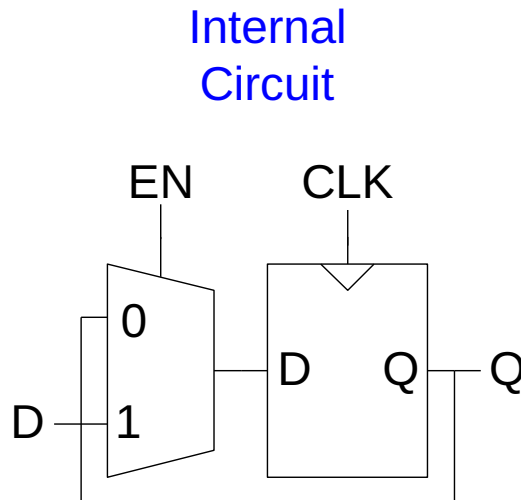
Enabled Flip-Flops

- **Inputs:** CLK, D, EN

- The enable input (EN) controls when new data (D) is stored

- **Function:**

- **EN = 1:** D passes through to Q on the clock edge
 - **EN = 0:** the flip-flop retains its previous state



Resettable Flip-Flop

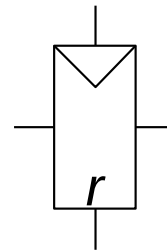
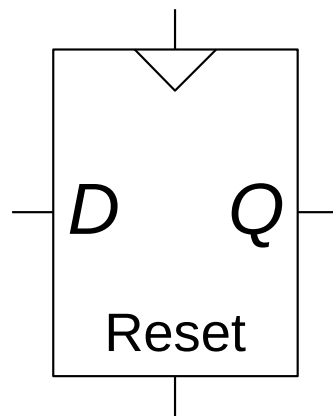
- **Inputs:** CLK, D, Reset

- The Reset is used to set the output to 0.

- **Function:**

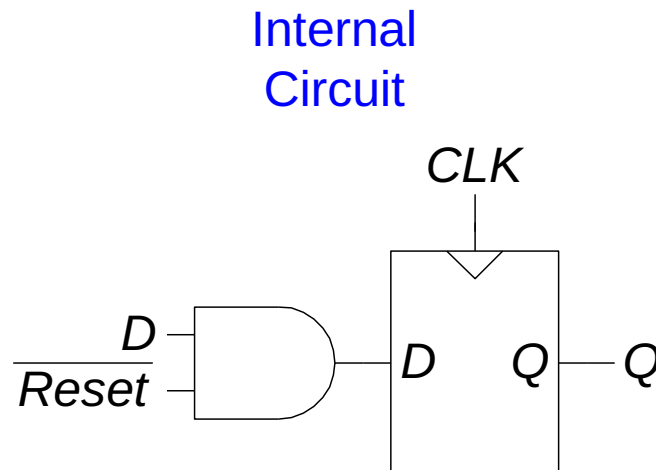
- **Reset = 1:** Q is forced to 0
 - **Reset = 0:** the flip-flop behaves like an ordinary D flip-flop

Symbols



Resettable Flip-Flops

- Two types:
 - **Synchronous:** resets at the clock edge only
 - **Asynchronous:** resets immediately when Reset = 1
- Asynchronously resettable flip-flop requires changing the internal circuitry of the flip-flop (see Exercise 3.10)
- Synchronously resettable flip-flop?



Settable Flip-Flop

- **Inputs:** CLK, D, Set
- **Function:**
 - **Set = 1:** Q is set to 1
 - **Set = 0:** the flip-flop behaves like an ordinary D flip-flop

Symbols

