LogiCORE™ IP Aurora 8B/10B v5.2

User Guide

UG353 (v5.2) July 23, 2010

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Revision History

The following table shows the revision history for this document.

Date	Version	Revision
03/24/08	2.9	Initial Xilinx release.
03/24/08	2.9.1	Post-release updates and corrections.
06/27/08	3.0	Virtex-5 FPGA Aurora v3.0 release. Added “Using the Build Script” section. Added “Using the Build Script” section. Corrected UFC_TX_MS[0:2] data in Figure 5-5. Added PMA_INIT and INIT_CLK to Table 6-1.
04/24/09	4.1	LogiCORE IP Aurora 8B/10B v4.1 release. Updated tools to v11.1. Renamed document title to include the 8B/10B protocol. Renamed Clock Source to “GT REFCLK Source1 and GT REFCLK Source2.” Replaced the perl script topic with the shell script topic in “Using the Build Script.” Deleted GREFCLK. Renamed RX/TX simplex Both to RX/TX Simplex, TX simplex only to TX-only Simplex, RX simplex only to RX-only Simplex, duplex to Duplex, and ERROR_COUNT to ERR_COUNT. Updated many figures and tables throughout. Added the following: • Framing and Streaming figures in Chapter 3, “Customizing the Aurora 8B/10B Core” • “FRAME_GEN” section • “FRAME_CHECK” section • “Designing with the Core” section • “Top Level Architecture” section • Appendix A, “Two Aurora 8B/10B Cores in the Virtex-5 Family Sharing a High-Speed Serial GTX Transceiver Tile” • Appendix B, “Handling Timing Errors Due To Far Apart Transceiver Selection” • Appendix C, “Performance and Core Latency” • Appendix D, “Generating a Wrapper File from Its Respective GTP/GTX Transceiver Wizard”

Date	Version	Revision
06/24/09	4.2	LogiCORE IP Aurora 8B/10B v4.2 release. Updated tools to v11.2. Removed callouts Figure 3-1, page 28. Added "Back Channel," page 31 and "Column Used," page 32, Corrected "can" to "cannot" in Table 6-1, page 62. Replaced all occurrences of DCM with PLL. Replaced all occurrences of DCM_NOT_LOCKED with PLL_NOT_LOCKED. Major updates to Table 7-1, page 77.
12/02/09	5.1	LogiCORE IP Aurora 8B/10B v5.1 release. Updated tools to v11.4. Added support for Spartan-6 FPGA GTP transceivers. Updated "Using the IP Customizer." Added "Framing NFC RX Interface" and "Reference Clocks for Spartan-6 FPGA GTP Transceiver Designs." Replaced bus range element 16*wn* to *wn*; REF_CLK to GT_REFCLK; and BUFDS to IBUFDS. Updated "Steps to Modify Transceiver Specific Ports and Attributes" in Appendix A and "Solutions" in Appendix B.
07/23/10	5.2	LogiCORE IP Aurora v5.2 release. Added support for the Virtex-6 HXT family. Integrated the LogiCORE Aurora 8B/10B Getting Started Guide (UG352) into this user guide (UG353). Replaced references to the getting started guide with cross-references to the corresponding sections in this user guide. Added Chapter 2, "Installing and Licensing the Core," Chapter 9, "Quick Start Example Design," and Chapter 11, "Project Directory Structure." Created Chapter 10, "Example Design Overview" from content previously in Chapter 3, "Customizing the Aurora 8B/10B Core." Replaced the content of Appendix B, "Handling Timing Errors Due To Far Apart Transceiver Selection" and Appendix D, "Generating a Wrapper File from Its Respective GTP/GTX Transceiver Wizard."

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About This Guide

The *LogiCORE IP Aurora 8B/10B v5.2 User Guide* provides information for generating a LogiCORE™ IP Aurora 8B/10B core using Virtex®-5 FPGA GTP/GTX transceivers, Virtex-6 FPGA GTX transceivers, and Spartan®-6 FPGA GTP transceivers.

The core implements the Aurora 8B/10 protocol using the high-speed serial transceivers on the Virtex-5 LXT, SXT, FXT, and TXT family, the Virtex-6 LXT, SXT, CXT, HXT, and lower-power family, and the Spartan-6 LXT family.

This user guide describes the function and operation of the LogiCORE™ IP Aurora 8B/10B v5.2 core and provides information about designing, customizing, and implementing the core.

Guide Contents

This guide contains the following chapters:

- Preface, "About this Guide" introduces the organization and purpose of the design guide, a list of additional resources, and the conventions used in this document.
- Chapter 1, "Introduction" describes the core and related information, including recommended design experience, additional resources, technical support, and submitting feedback to Xilinx.
- Chapter 2, "Installing and Licensing the Core" provides information about installing and licensing the core.
- Chapter 3, "Customizing the Aurora 8B/10B Core" provides port descriptions for the user interface.
- Chapter 4, "User Interface" provides port descriptions for the user interface.
- Chapter 5, "Flow Control" describes the user flow control and native flow control options for sending and receiving data.
- Chapter 6, "Status, Control, and the GTP/GTX Block Interface" provides details on using the Aurora 8B/10B core's status and control ports to initialize and monitor the Aurora 8B/10B channel.
- Chapter 7, "Clock Interface and Clocking" describes how to connect FPGA clocking resources.
- Chapter 8, "Clock Compensation" covers Aurora 8B/10B clock compensation, and explains how to customize it for a given system.
- Chapter 9, "Quick Start Example Design" provides an overview of the Aurora 8B/10B protocol and core, and gives a step-by-step tutorial on how to generate Aurora 8B/10B designs with the CORE Generator™ software.

- Chapter 10, "Example Design Overview" defines the main components of the example design.
- Chapter 11, "Project Directory Structure" provides detailed information about the example design, including a description of files and the directory structure generated by the Xilinx CORE Generator tool, the purpose and contents of the provided scripts, the contents of the example HDL wrappers, and the operation of the demonstration test bench.
- Appendix A, "Two Aurora 8B/10B Cores in the Virtex-5 Family Sharing a High-Speed Serial GTX Transceiver Tile"
- Appendix B, "Handling Timing Errors Due To Far Apart Transceiver Selection"
- Appendix C, "Performance and Core Latency"
- Appendix D, "Generating a Wrapper File from Its Respective GTP/GTX Transceiver Wizard"

Additional Resources

To find additional documentation, see the Xilinx website at:

http://www.xilinx.com/support/documentation/index.htm.

To search the Answer Database of silicon, software, and IP questions and answers, or to create a technical support WebCase, see the Xilinx website at:

http://www.xilinx.com/support/mysupport.htm.

Conventions

This document uses the following conventions. An example illustrates each convention.

Typographical

The following typographical conventions are used in this document:

Convention	Meaning or Use	Example
`Courier font`	Messages, prompts, and program files that the system displays	`speed grade: - 100`
`Courier bold`	Literal commands that you enter in a syntactical statement	**`ngdbuild`** *design_name*
Helvetica bold	Commands that you select from a menu	**File → Open**
	Keyboard shortcuts	**Ctrl+C**

Convention	Meaning or Use	Example
Italic font	Variables in a syntax statement for which you must supply values	**`ngdbuild`** *design_name*
	References to other manuals	See the *Development System Reference Guide* for more information.
	Emphasis in text	If a wire is drawn so that it overlaps the pin of a symbol, the two nets are *not* connected.
Dark Shading	Items that are not supported or reserved	This feature is not supported
Square brackets [ ]	An optional entry or parameter. However, in bus specifications, such as **`bus[7:0]`**, they are required.	**`ngdbuild`** [*`option_name`*] *`design_name`*
Braces { }	A list of items from which you must choose one or more	**`lowpwr ={on|off}`**
Vertical bar |	Separates items in a list of choices	**`lowpwr ={on|off}`**
Angle brackets < >	User-defined variable or in code samples	<directory name>

Vertical ellipsis
.
.
.
	Repetitive material that has been omitted	`IOB #1: Name = QOUT'`
		`IOB #2: Name = CLKIN'`
		.
		.
		.

Horizontal ellipsis ...	Repetitive material that has been omitted	**`allow block`** *block_name loc1 loc2 ... locn;*
Notations	The prefix '0x' or the suffix 'h' indicate hexadecimal notation	A read of address 0x00112975 returned 45524943h.
	An '_n' means the signal is active low	**`usr_teof_n`** is active low.

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Chapter 1

Introduction

This chapter introduces the LogiCORE™ IP Aurora 8B/10B core and provides related information, including recommended design experience, additional resources, technical support, and how to submit feedback to Xilinx. The Aurora core is a high-speed serial solution based on the Aurora protocol, Virtex®-5 FPGA GTP/GTX transceivers, Virtex-6 FPGA GTX transceivers, and Spartan®-6 FPGA GTP transceivers. The core is delivered as open-source code and supports both Verilog and VHDL design environments. Each core comes with an example design and supporting modules.

About the Core

The Aurora 8B/10B core is a CORE Generator™ IP core, included in the latest IP Update on the Xilinx IP Center. For detailed information about the core, see www.xilinx.com/aurora. For information about system requirements, installation, and licensing options, see Chapter 2, "Installing and Licensing the Core."

Recommended Design Experience

Although the Aurora core is a fully verified solution, the challenge associated with implementing a complete design varies depending on the configuration and functionality of the application. For best results, previous experience building high-performance, pipelined FPGA designs using Xilinx implementation software and user constraints files (`.ucf`) is recommended.

Read Chapter 6, "Status, Control, and the GTP/GTX Block Interface" carefully.

Consult the PCB design requirements information in:

- *Virtex-5 FPGA RocketIO GTP Transceiver User Guide*,
- *Virtex-5 FPGA RocketIO GTX Transceiver User Guide*
- *Virtex-6 FPGA GTX Transceivers User Guide*,
- *Spartan-6 FPGA GTP Transceivers User Guide*.

Contact your local Xilinx representative for a closer review and estimation for your specific requirements.

Related Xilinx Documents

Prior to generating an Aurora core, users should be familiar with the following:

- SP002, *Aurora 8B/10B Protocol Specification*
- SP006, *LocalLink Interface Specification* is located on the LocalLink product page
- UG196, *Virtex-5 FPGA RocketIO GTP Transceiver User Guide*
- UG198, *Virtex-5 FPGA RocketIO GTX Transceiver User Guide*
- UG366, *Virtex-6 FPGA GTX Transceivers User Guide*
- UG386, *Spartan-6 FPGA GTP Transceivers User Guide*
- ISE® software documentation: xilinx.com/ise

Additional Core Resources

For detailed information and updates about the Aurora core, see the following documents, located on the Aurora product page at www.xilinx.com/aurora.

- DS637: *LogiCORE IP Aurora 8B/10B v5.2 Data Sheet*
- UG058: *Aurora 8B/10B Bus Functional Model User Guide* (Contact: auroramkt@xilinx.com)
- Aurora 8B/10B Release Notes
- Aurora Solution list – AR#21263

Technical Support

For technical support, go to www.xilinx.com/support. Questions are routed to a team of engineers with expertise using the Aurora core.

Xilinx will provide technical support for use of this product as described in the *LogiCORE IP Aurora 8B/10B v5.2 User Guide*. Xilinx cannot guarantee timing, functionality, or support of this product for designs that do not follow the guidelines.

Feedback

Xilinx welcomes comments and suggestions about the Aurora core and the accompanying documentation.

Core

For comments or suggestions about the Aurora 8B/10B core, please submit a WebCase from http://www.xilinx.com/support/clearexpress/websupport.htm. Be sure to include the following information:

- Product name
- Core version number
- List of parameter settings
- Explanation of your comments

Documentation

For comments or suggestions about the Aurora 8B/10B documents, please submit a WebCase from http://www.xilinx.com/support/clearexpress/websupport.htm. Be sure to include the following information:

- Document title
- Document number
- Page number(s) to which your comments refer
- Explanation of your comments

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Chapter 2

Installing and Licensing the Core

This chapter provides instructions for installing the LogiCORE™ IP Aurora 8B/10B core in the Xilinx® CORE Generator™ tool. It is not necessary to obtain a license to use the core.

Supported Tools and System Requirements

Operating Systems

Windows

- Windows XP Professional 32-bit/64-bit
- Windows Vista Business 32-bit/64-bit

Linux

- Red Hat Enterprise Linux WS v4.0 32-bit/64-bit
- Red Hat Enterprise Desktop v5.0 32-bit/64-bit (with Workstation Option)
- SUSE Linux Enterprise (SLE) desktop and server v10.1 32-bit/64-bit

Tools

- ISE® software 12.2
- Mentor Graphics ModelSim 6.5c and above

Check the release notes for the required Service Pack; ISE Service Packs can be downloaded from www.xilinx.com/xlnx/xil_sw_updates_home.jsp?update=sp.

Before You Begin

This chapter assumes you have installed the core using either the CORE Generator™ IP Software Update installer or by performing a manual installation after downloading the core from the web. For information about installing the core, see:

www.xilinx.com/aurora

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Chapter 3

Customizing the Aurora 8B/10B Core

Introduction

The Aurora 8B/10B core can be customized to suit a wide variety of requirements using the CORE Generator™ software. This chapter details the customization parameters available to the user and how these parameters are specified within the IP Customizer interface.

Using the IP Customizer

The Aurora 8B/10B IP Customizer is presented when the user selects the Aurora 8B/10B core in the CORE Generator software. For help starting and using the CORE Generator software, see the *CORE Generator Help* in the ISE® software documentation. Each numbered item in Figure 3-1, page 28 corresponds to its respective section that describes the purpose of the feature.

IP Customizer

Figure 3-1, page 28 shows the customizer. The left side displays a representative block diagram of the Aurora 8B/10B core as currently configured. The right side consists of user-configurable parameters.

The second pages of the GUI are shown in:

- Figure 3-2, page 28 for Virtex®-5 FPGA RocketIO™ GTP transceivers
- Figure 3-3, page 29 for Virtex-5 FPGA RocketIO GTX transceivers
- Figure 3-4, page 29 for Virtex-6 FPGA GTX transceivers
- Figure 3-5, page 30 for Spartan®-6 FPGA GTP transceivers

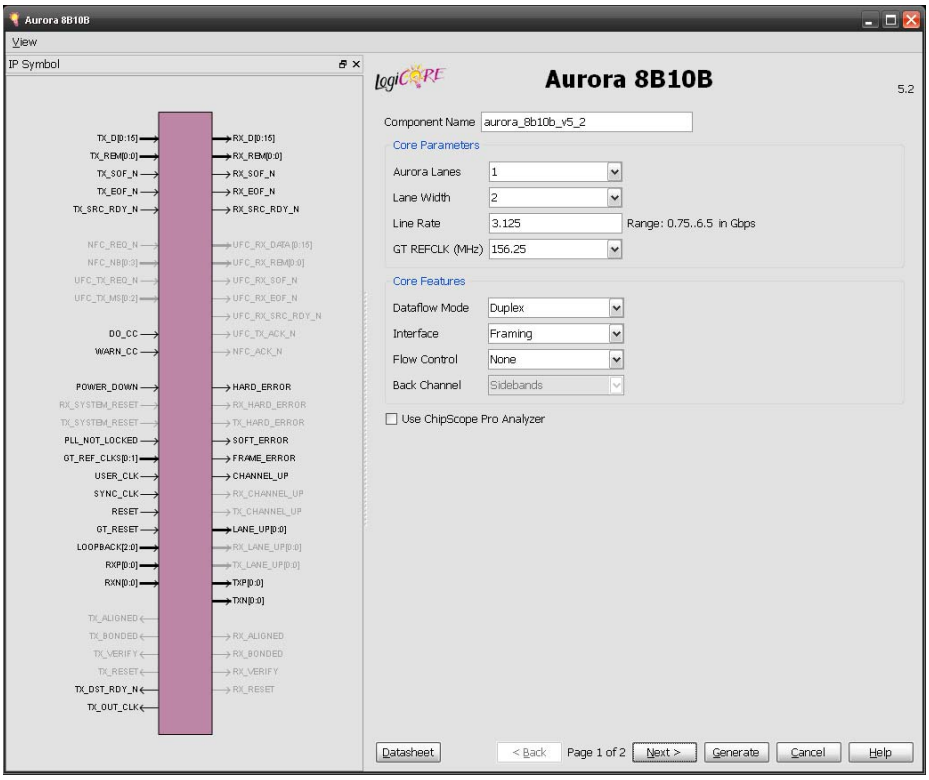

Figure 3-1: **Aurora 8B/10B IP Customizer**

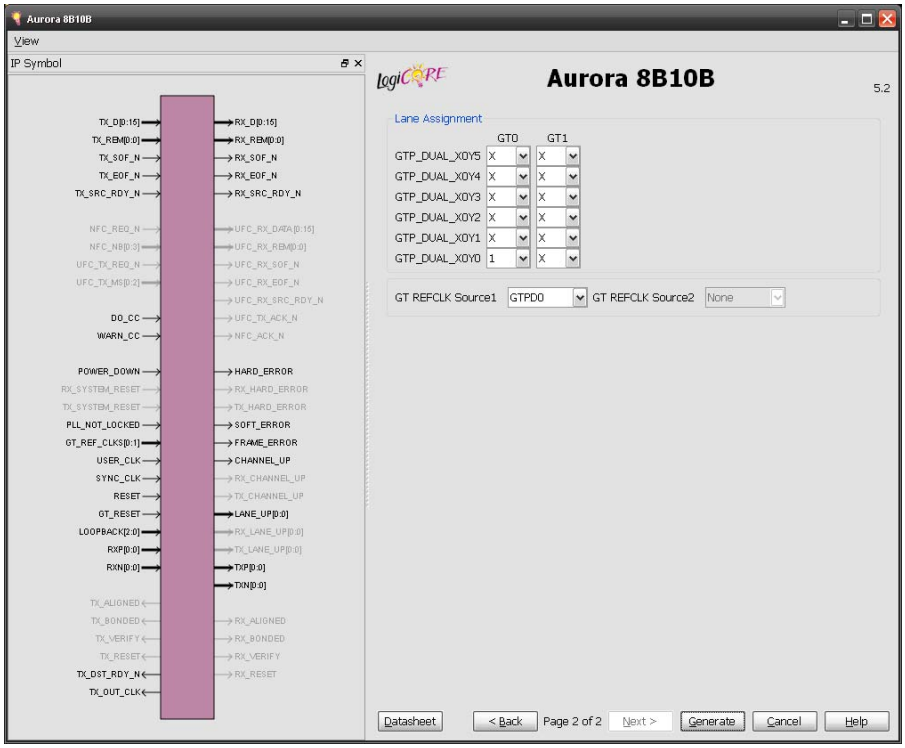

Figure 3-2: **Second GUI Page for Virtex-5 FPGA RocketIO GTP Transceivers**

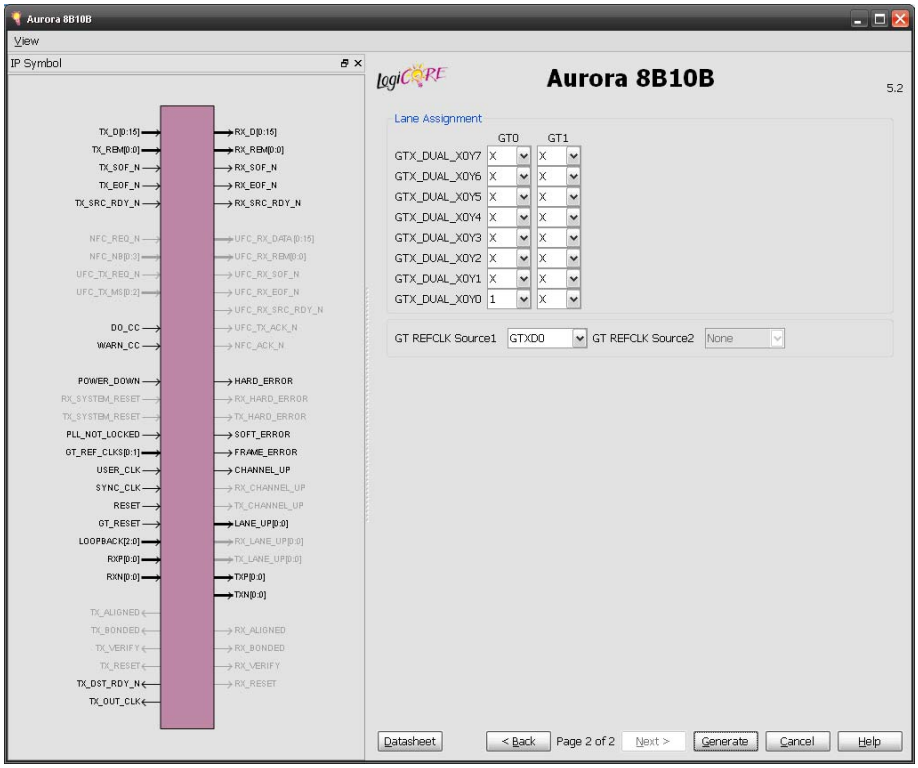

Figure 3-3: **Second GUI Page for Virtex-5 FPGA RocketIO GTX Transceivers**

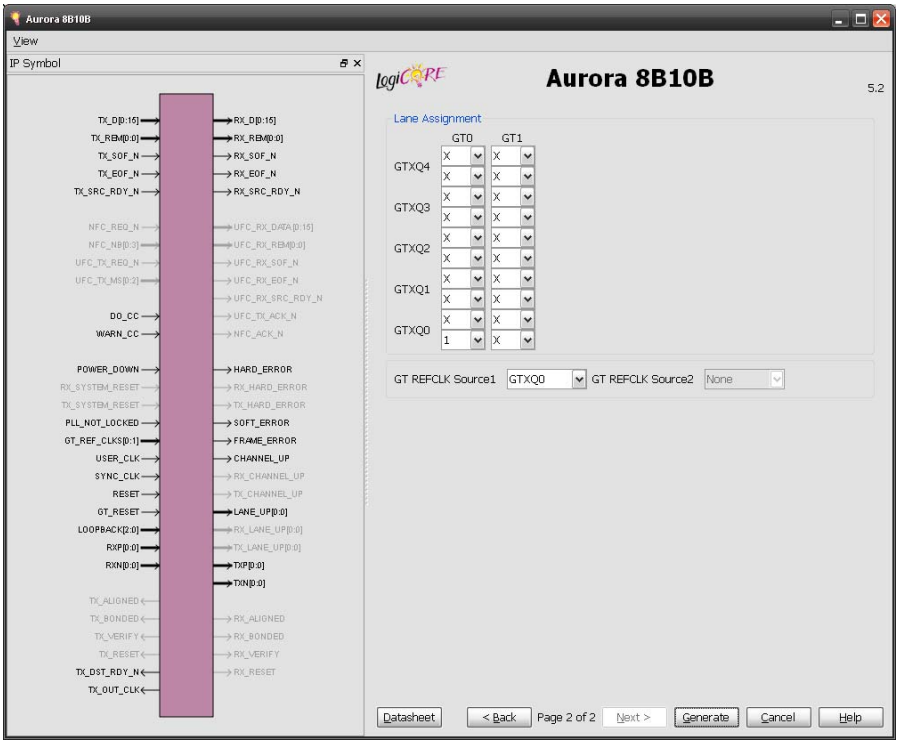

Figure 3-4: **Second GUI Page for Virtex-6 FPGA GTX Transceivers**

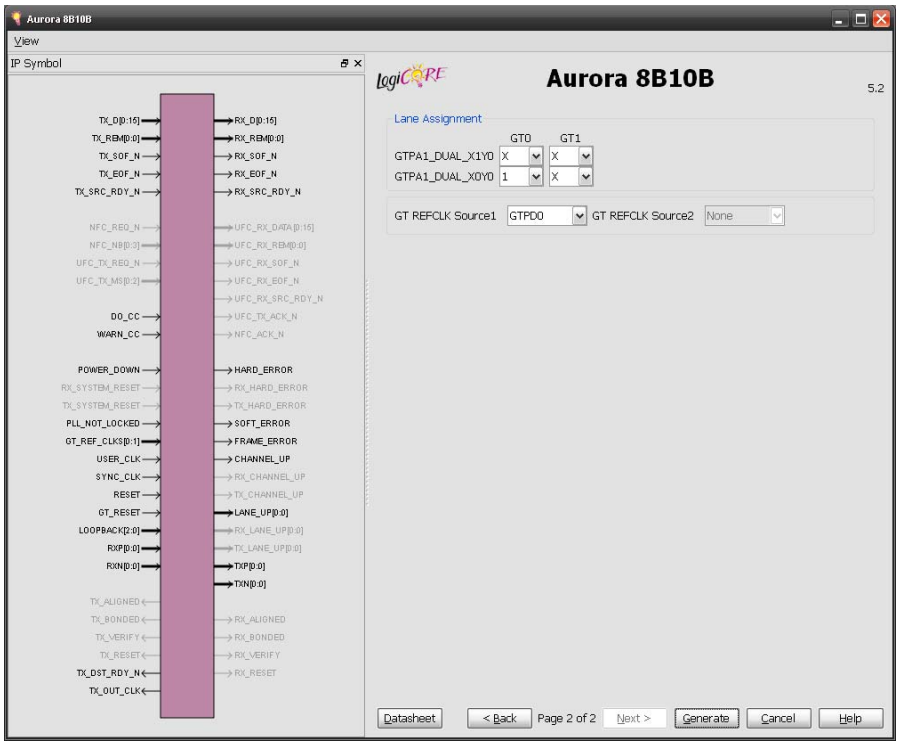

Figure 3-5: **Second GUI Page for Spartan-6 FPGA GTP Transceivers**

Component Name

Enter the top-level name for the core in this text box. Illegal names are highlighted in red until they are corrected. All files for the generated core are placed in a subdirectory using this name. The top-level module for the core also uses this name.

Default: aurora_8b10b_v5_2

Lane Assignment

Refer to the diagram in the information area in Figure 3-1, page 28. Each numbered column represents a GTP_DUAL/GTX_DUAL or two columns represents GTX Quad and each active box represents an available GTP/GTX transceiver.

Aurora Lanes

Select the number of lanes (GTP/GTX transceivers) to be used in the core. The valid range depends on the target device selected.

Default: 1

Lane Width

Select the byte width of the GTP/GTX transceivers used in the core.

Default: 2

Interface

Select the type of data path interface used for the core. Select *Framing* to use a LocalLink interface that allows encapsulation of data frames of any length. Select *Streaming* to use a simple word-based interface with a data valid signal to stream data through the Aurora 8B/10B channel. See Chapter 4, "User Interface" for more information.

Default: Framing

Dataflow Mode

Select the options for direction of the channel the Aurora 8B/10B core will support. Simplex Aurora 8B/10B cores have a single, unidirectional serial port that connects to a complementary simplex Aurora 8B/10B core. Available options are *RX-only Simplex*, *TX-only Simplex*, *RX/TX Simplex*, *and Duplex*. RX/TX Simplex creates two cores, one RX and one TX, that share a single GTP/GTX transceiver. See Chapter 6, "Status, Control, and the GTP/GTX Block Interface" for more information.

Default: Duplex

Back Channel

Select the options for Back Channel only for Simplex Aurora cores; Duplex Aurora cores does not require this option. The available options are:

- Sidebands
- Timer

Default: Sidebands

Note: There is no functionality difference between RX-only Simplex design with Sidebands option and RX-only Simplex design with Timer option.

Flow Control

Select the required option to add flow control to the core. User flow control (UFC) allows applications to send a brief, high-priority message through the Aurora 8B/10B channel. Native flow control (NFC) allows full duplex receivers to regulate the rate of the data send to them. Immediate mode allows idle codes to be inserted within data frames while completion mode only inserts idle codes between complete data frames.

Available options are listed below (See Chapter 5, "Flow Control" for more information):

- None
- UFC
- Immediate Mode - NFC
- Completion Mode - NFC
- UFC + Immediate Mode - NFC
- UFC + Completion Mode - NFC

Default: None

Line Rate

Enter a floating-point value in gigabits per second within the valid range. This determines the un-encoded bit rate at which data will be transferred over the serial link. The aggregate data rate of the core is (0.8 × line rate) × Aurora 8B/10B lanes.
Default: 3.125 Gbps

GT REFCLK (MHz)

Select a reference clock frequency for the transceiver from the drop-down list. Reference clock frequencies are given in megahertz (MHz), and depend on the line rate selected. For best results, select the highest rate that can be practically applied to the reference clock input of the target device.

Default: 156.25 MHz

GT REFCLK Source1 and GT REFCLK Source2

Select reference clock sources for the GTP/GTX_DUALs or GTX Quad from the drop-down list in this section.

- Default: GT REFCLK Source1 - GTXD0; GT REFCLK Source2 - None for Virtex-5 FPGA GTX transceivers
- Default: GT REFCLK Source1 - GTPD0; GT REFCLK Source2 - None for Virtex-5 FPGA GTP and Spartan-6 FPGA GTP transceivers
- Default: GT REFCLK Source1 - GTXQ0; GT REFCLK Source2 - None for Virtex-6 FPGA GTX transceivers
- GTXD0, GTPD0 and GTXQ0 will change based on the selected device and package.

Column Used

Select the appropriate column of transceivers used from the drop down list. The column used is enabled only for Virtex-5 TXT or Virtex-6 HXT devices and is disabled for all other devices.

Default: left

Row Used

Select the appropriate row of transceivers used from the drop down list. The row used is enabled only for Spartan-6 LXT devices and is disabled for all other devices.

Default: top

Use ChipScope Pro Analyzer

Select to add ChipScope™ Pro cores to the Aurora 8B/10B core (see "Using ChipScope Pro Cores with the Aurora 8B/10B Core"). This option provides users a debugging interface that shows the core status signals in the ChipScope Pro analyzer tool.

Default: Unchecked

Generate

Click Generate to generate the core. The modules for the Aurora 8B/10B core are written to the CORE Generator software project directory using the same name as the top level of the core. See Chapter 11, "Project Directory Structure" for details about the example_design directory and files.

Using the Build Script

A shell script called implement.sh is delivered with the Aurora 8B/10B core in the implement subdirectory. The script can be run to synthesize using XST, generate project files, or implement the core. Make sure the XILINX environment variable is set properly then run the script by entering the following command in the implement directory:

```
./implement.sh
```

Designing with the Core

This section provides a general description of how to use the Aurora 8B/10B core in your designs

General Design Guidelines

This section describes the steps required to turn an Aurora 8B/10B core into a fully functioning design with user-application logic. It is important to note that not all implementations require all of the design steps listed here. Follow the logic design guidelines in this manual carefully.

Use the Example Design as a Starting Point

Each instance of an Aurora 8B/10B core created by CORE Generator software is delivered with an example design that can be simulated and implemented in FPGA. This design can be used as a starting point for your own design or can be used to troubleshoot the user application, if necessary.

Know the Degree of Difficulty

Aurora 8B/10B design is challenging to implement in any technology, and the degree of difficulty is further influenced by

- Maximum system clock frequency
- Targeted device architecture
- Nature of the user application

All Aurora 8B/10B implementations require careful attention to system performance requirements. Pipelining, logic mappings, placement constraints and logic duplications are all methods that help boost system performance.

Keep It Registered

To simplify timing and increase system performance in an FPGA design, keep all inputs and outputs registered between the user application and the core. This means that all inputs and outputs from user application should come from, or connect to a flip-flop. Registering signals may not be possible for all paths, but doing so simplifies timing analysis and makes it easier for the Xilinx tools to place-and-route the design.

Recognize Timing Critical Signals

The UCF provided with the example design for the core identifies the critical signals and the timing constraints that should be applied.

Use Supported Design Flows

The core is delivered as Verilog or VHDL source code. The example implementation scripts provided currently use XST as synthesis tool for the example design that is delivered with the core. Other synthesis tools may also be used.

Make Only Allowed Modifications

The Aurora 8B/10B core is not user modifiable. Any modifications may have adverse effects on the system timings and protocol compliance. Supported user configurations of the Aurora 8B/10B core can only be made by selecting options from CORE Generator tool.

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Chapter 4

User Interface

Introduction

An Aurora 8B/10B core can be generated with either a *framing* or *streaming* user data interface. In addition, flow control options are available for designs with framing interfaces. See Chapter 5, "Flow Control."

The framing user interface complies with the *LocalLink Interface Specification*. It comprises the signals necessary for transmitting and receiving framed user data. The streaming interface allows users to send data without special frame delimiters. It is simple to operate and uses fewer resources than framing.

Top Level Architecture

Aurora 8B/10B top level (block level) file instantiates Aurora 8B/10B lane module, TX and RX LocalLink modules, global logic module, and wrapper for GTP/GTX transceiver. This top level wrapper file is instantiated in the example design file together with clock, reset circuit and frame generator and checker modules.

Figure 4-1, page 36 shows Aurora 8B/10B top level for a duplex configuration. The top level file is the starting point for a user design.

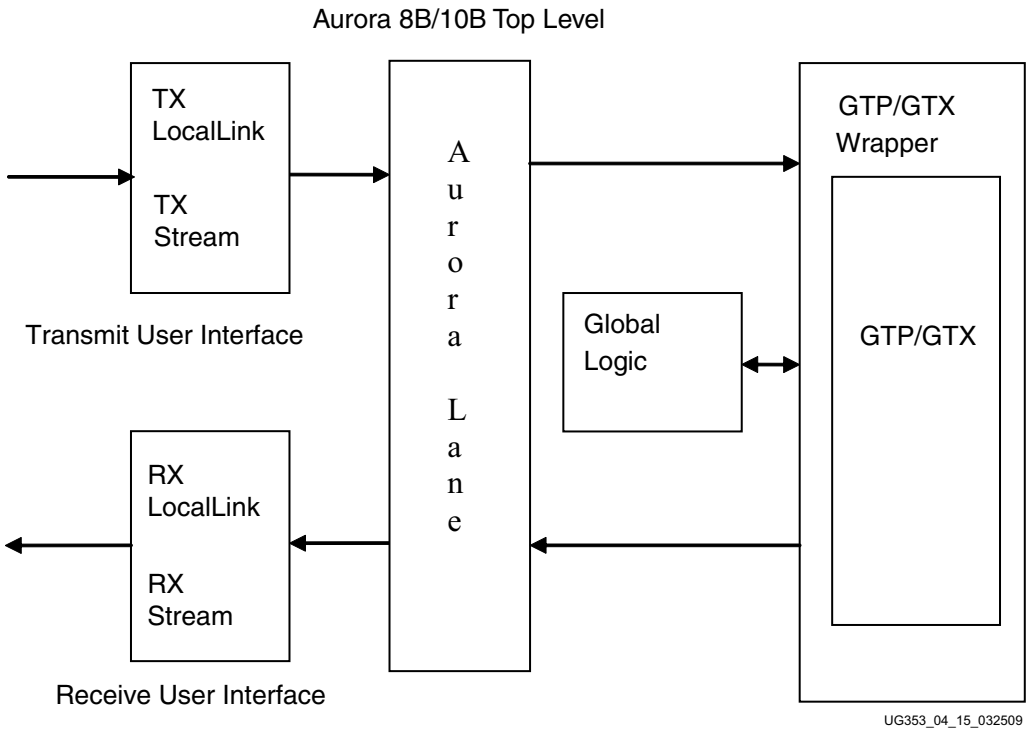

Figure 4-1: **Top-Level Architecture**

The following sections describe the streaming and framing interface in details. User interface logic should be designed to comply with the timing requirement of the respective interface as explained in the subsequent sections.

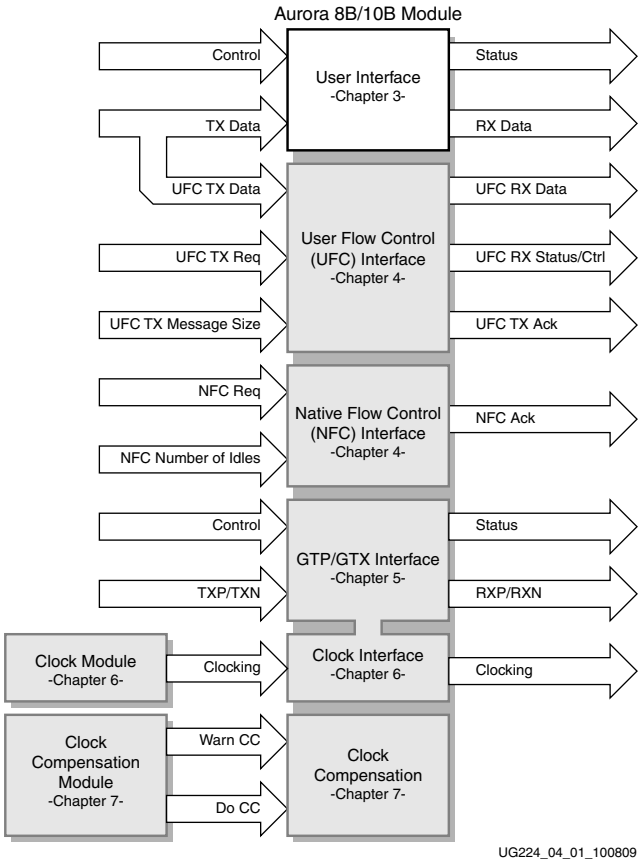

Figure 4-2: **Top-Level User Interface**

Note: The user interface signals vary depending upon the selections made when generating an Aurora 8B/10B core in the CORE Generator™ software.

Framing Interface

Figure 4-3 shows the framing user interface of the Aurora 8B/10B core, with LocalLink-compliant ports for TX and RX data.

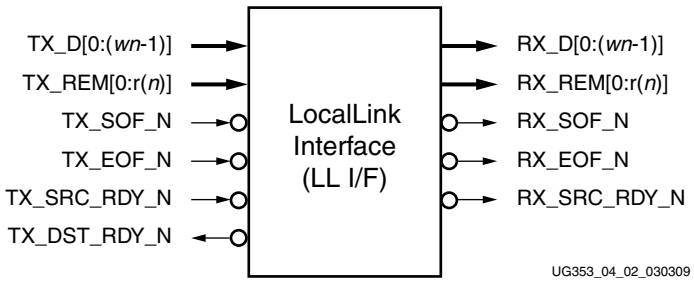

Figure 4-3: **Aurora 8B/10B Core Framing Interface (LocalLink)**

LocalLink TX Ports

Table 4-1 lists port descriptions for LocalLink TX data ports. These ports are included on full-duplex, simplex TX, and simplex Both framing cores.

Table 4-1: **LocalLink User I/O Ports (TX)**

Name	Direction	Description
TX_D[0:(*wn*-1)]	Input	Outgoing data (Ascending bit order). • *w* takes values 16 or 32 depending on whether width is 2-byte or 4-byte • *n* is the number of lanes
TX_DST_RDY_N	Output	Asserted (Low) during clock edges when signals from the source will be accepted (if TX_SRC_RDY_N is also asserted). Deasserted (High) on clock edges when signals from the source will be ignored.
TX_EOF_N	Input	Signals the end of the frame (active-Low).
TX_REM[0:r(*n*)]	Input	Specifies the number of valid bytes in the last data beat; valid only while TX_EOF_N is asserted. REM bus widths are given by [0:r(*n*)], where r(*n*) = ceiling [{$\log_2(n)$}-1].
TX_SOF_N	Input	Signals the start of the outgoing channel frame (active-Low).
TX_SRC_RDY_N	Input	Asserted (Low) when LocalLink signals from the source are valid. Deasserted (High) when LocalLink control signals and/or data from the source should be ignored (active-Low).

LocalLink RX Ports

Table 4-2 lists port descriptions for LocalLink RX data ports. These ports are included on full-duplex, simplex RX, and simplex Both framing cores.

Table 4-2: **LocalLink User I/O Ports (RX)**

Name	Direction	Description
RX_D[0:(*wn*-1)]	Output	Incoming data from channel partner (Ascending bit order).
RX_EOF_N	Output	Signals the end of the incoming frame (active-Low, asserted for a single user clock cycle). Ignored when RX_SRC_RDY_N is de-asserted (High)
RX_REM[0:r(*n*)]	Output	Specifies the number of valid bytes in the last data beat; valid only when RX_EOF_N is asserted. REM bus widths are given by [0:r(*n*)], where r(*n*) = ceiling [{$\log_2(n)$}-1].
RX_SOF_N	Output	Signals the start of the incoming frame (active-Low, asserted for a single user clock cycle). Ignored when RX_SRC_RDY_N is de-asserted (High)
RX_SRC_RDY_N	Output	Asserted (Low) when data and control signals from an Aurora 8B/10B core are valid. Deasserted (High) when data and/or control signals from an Aurora 8B/10B core should be ignored (active-Low).

To transmit data, the user manipulates control signals to cause the core to do the following:

- Take data from the user on the TX_D bus
- Encapsulate and stripe the data across lanes in the Aurora 8B/10B channel (TX_SOF_N, TX_EOF_N)
- Pause data (that is, insert idles) (TX_SRC_RDY_N)

When the core receives data, it does the following:

- Detects and discards control bytes (idles, clock compensation, SCP, ECP)
- Asserts framing signals (RX_SOF_N, RX_EOF_N)
- Recovers data from the lanes
- Assembles data for presentation to the user on the `RX_D` bus

LocalLink Bit Ordering

The *LocalLink Interface Specification* allows both ascending and descending bit ordering. Aurora 8B/10B cores use ascending ordering. They transmit and receive the most significant bit of the most significant byte first. Figure 4-4 shows the organization of an *n*-byte example of the LocalLink data interfaces of an Aurora 8B/10B core.

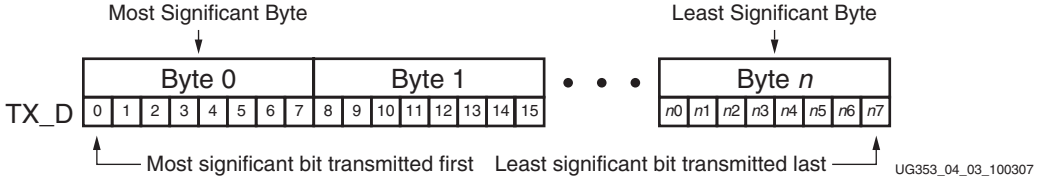

Figure 4-4: **LocalLink Interface Bit Ordering**

Transmitting Data

LocalLink is a synchronous interface. The Aurora 8B/10B core samples the data on the interface only on the positive edge of USER_CLK, and only on the cycles when both TX_DST_RDY_N and TX_SRC_RDY_N are asserted (Low).

When LocalLink signals are sampled, they are only considered valid if TX_SRC_RDY_N is asserted. The user application can deassert TX_SRC_RDY_N on any clock cycle; this will cause Aurora 8B/10B to ignore the LocalLink input for that cycle. If this occurs in the middle of a frame, idle symbols are sent through the Aurora 8B/10B channel, which eventually result in a idle cycles during the frame when it is received at the RX user interface.

LocalLink data is only valid when it is framed. Data outside of a frame is ignored. To start a frame, assert TX_SOF_N while the first word of data is on the TX_D port. To end a frame, assert TX_EOF_N while the last word (or partial word) of data is on the TX_D port.

Note: In the case of frames that are a single word long or less, `TX_SOF_N` and `TX_EOF_N` are asserted simultaneously.

Data Remainder

LocalLink allows the last word of a frame to be a partial word. This lets a frame contain any number of bytes, regardless of the word size. The TX_REM bus is used to indicate the number of valid bytes in the final word of the frame. The bus is only used when TX_EOF_N is asserted. Aurora 8B/10B uses encoded REM values. REM is the binary encoding of the number of valid bytes minus 1. A zero REM value indicates the left-most byte in the TX_D port (the MSB) is the only valid byte in the word. Table 4-3 shows the mapping between TX_REM values and valid byte counts for the TX_D port.

Table 4-3: **TX Data Remainder Values**

TX_REM Value	Number of Valid Bytes
0	1
1	2
2	3
3	4
. . .	. . .
n	*n*+1

Aurora 8B/10B Frames

The TX_LL submodule translates each user frame that it receives through the TX interface to an Aurora 8B/10B frame. The 2-byte SCP code group is added to the beginning of the frame data to indicate the start of frame, and a 2-byte ECP set is sent after the frame ends to indicate the end of frame. Idle code groups are inserted whenever data is not available. Code groups are 8B/10B encoded byte pairs. All data in Aurora 8B/10B is sent as code groups, so user frames with an odd number of bytes have a control character called PAD appended to the end of the frame to fill out the final code group. Table 4-4 shows a typical Aurora 8B/10B frame with an even number of data bytes.

Length

The user controls the channel frame length by manipulation of the TX_SOF_N and TX_EOF_N signals. The Aurora 8B/10B core responds with start-of-frame and end-of-frame ordered sets, /SCP/ and /ECP/ respectively, as shown in Table 4-4.

Table 4-4: **Typical Channel Frame**

$/SCP/_1$	$/SCP/_2$	Data Byte 0	Data Byte 1	Data Byte 2	...	Data Byte *n* -1	Data Byte *n*	$/ECP/_1$	$/ECP/_2$

Example A: Simple Data Transfer

Figure 4-5 shows an example of a simple data transfer on a LocalLink interface that is *n*-bytes wide. In this case, the amount of data being sent is 3*n* bytes and so requires three data beats. TX_DST_RDY_N is asserted, indicating that the LocalLink interface is already ready to transmit data. When the Aurora 8B/10B core is not sending data, it sends idle sequences.

To begin the data transfer, the user asserts the TX_SOF_N concurrently with TX_SRC_RDY_N and the first *n* bytes of the user frame. Since TX_DST_RDY_N is already asserted, data transfer begins on the next clock edge. An /SCP/ ordered set is placed on the first two bytes of the channel to indicate the start of the frame. Then the first *n*-2 data bytes are placed on the channel. Because of the offset required for the /SCP/, the last two bytes in each data beat are always delayed one cycle and transmitted on the first two bytes of the next beat of the channel.

To end the data transfer, the user asserts TX_EOF_N, TX_SRC_RDY_N, the last data bytes, and the appropriate value on the TX_REM bus. In this example, TX_REM is set to *n*-1 to indicate that all bytes are valid in the last data beat. One clock cycle after TX_EOF_N is asserted, the LocalLink interface deasserts TX_DST_RDY_N and uses the gap in the data flow to send the final offset data bytes and the /ECP/ ordered set, indicating the end of the frame. TX_DST_RDY_N is reasserted on the next cycle so that more data transfers can continue. As long as there is no new data, the Aurora 8B/10B core sends idles.

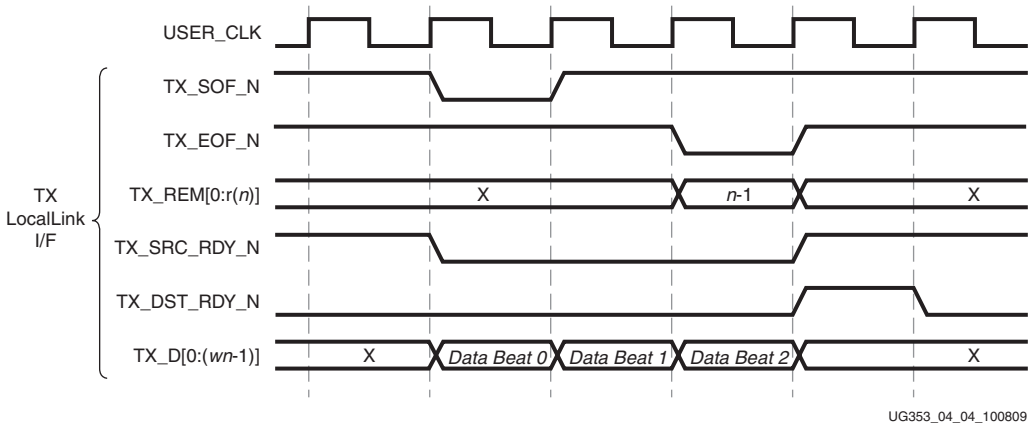

Figure 4-5: **Simple Data Transfer**

Example B: Data Transfer with Pad

Figure 4-6 shows an example of a (3*n*-1)-byte data transfer that requires the use of a pad. Since there is an odd number of data bytes, the Aurora 8B/10B core appends a pad character at the end of the Aurora 8B/10B frame, as required by the protocol. A transfer of 3*n*-1 data bytes requires two full *n*-byte data words and one partial data word. In this example, TX_REM is set to *n*-2 to indicate *n*-1 valid bytes in the last data word.

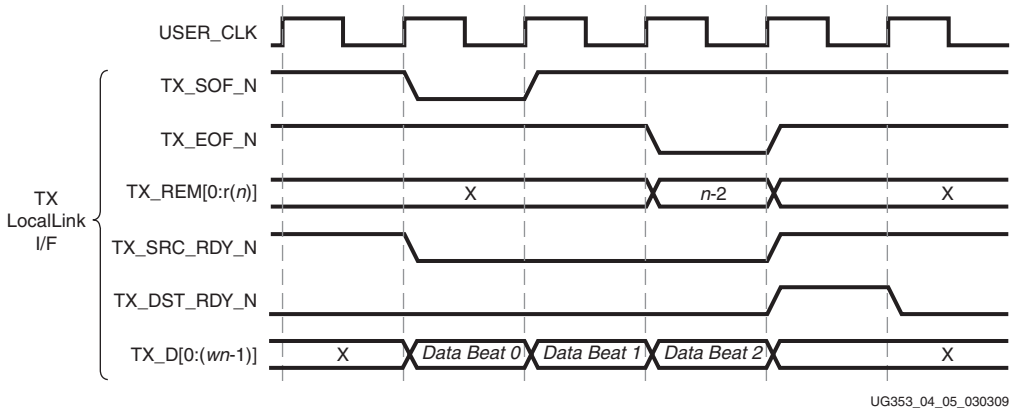

Figure 4-6: **Data Transfer with Pad**

Example C: Data Transfer with Pause

Figure 4-7 shows how a user can pause data transmission during a frame transfer. In this example, the user is sending 3*n* bytes of data, and pauses the data flow after the first *n* bytes. After the first data word, the user deasserts TX_SRC_RDY_N, causing the TX Aurora 8B/10B core to ignore all data on the bus and transmit idles instead. The offset data from the first data word in the previous cycle still is transmitted on lane 0, but the next data word is replaced by idle characters. The pause continues until TX_SRC_RDY_N is deasserted.

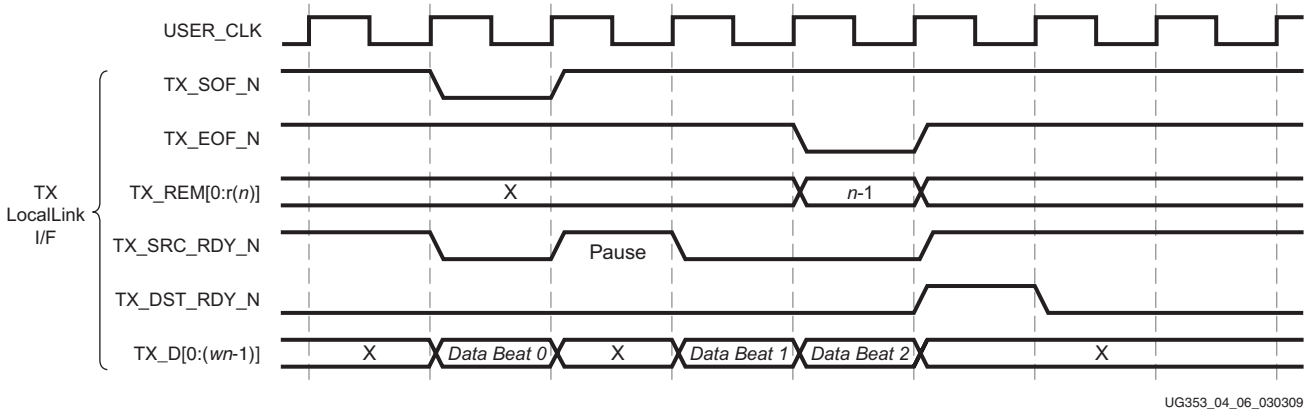

Figure 4-7: **Data Transfer with Pause**

Example D: Data Transfer with Clock Compensation

The Aurora 8B/10B core automatically interrupts data transmission when it sends clock compensation sequences. The clock compensation sequence imposes 12 bytes of overhead per lane every 10,000 bytes.

Figure 4-8 shows how the Aurora 8B/10B core pauses data transmission during the clock compensation(1) sequence.

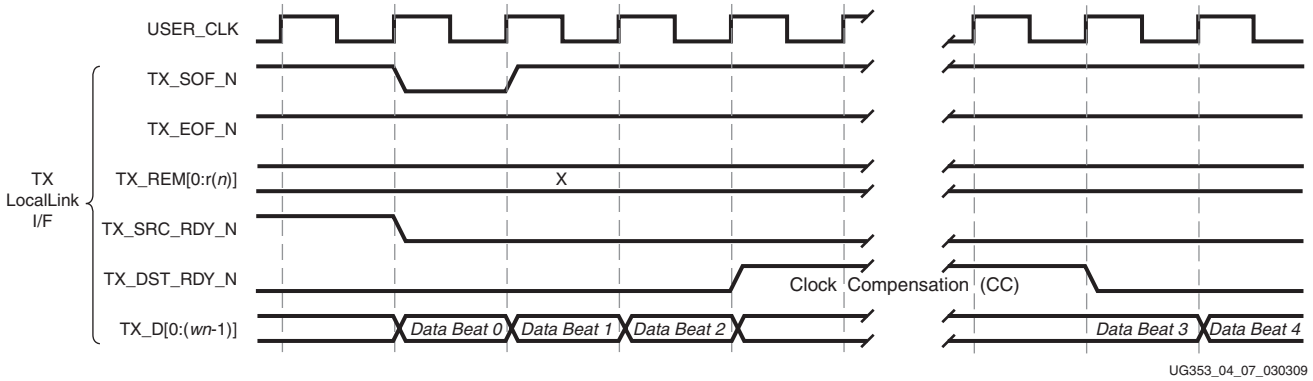

Figure 4-8: **Data Transfer Paused by Clock Compensation**

TX Interface Example

This section illustrates a simple example of how a user might design an interface between the user's transmit FIFO and the LocalLink interface of an Aurora 8B/10B core.

To review, in order to transmit data, the user asserts TX_SOF_N and TX_SRC_RDY_N. TX_DST_RDY_N indicates that the data on the TX_D bus will be transmitted on the next rising edge of the clock, assuming TX_SRC_RDY_N remains asserted.

Figure 4-9 is a diagram of a typical connection between an Aurora 8B/10B core and the user's data source (in this example, a FIFO), including the simple logic needed to generate TX_SOF_N, TX_SRC_RDY_N, and TX_EOF_N from typical FIFO buffer status signals. While RESET is false, the example application waits for a FIFO to fill and then it generates the TX_SOF_N and TX_SRC_RDY_N signals. These two signals cause the Aurora 8B/10B core to start reading the FIFO by asserting the TX_DST_RDY_N signal.

The Aurora 8B/10B core encapsulates the FIFO data and transmits it until the FIFO is empty. At this point, the example application tells the Aurora 8B/10B core to end the transmission using the TX_EOF_N signal.

1. Because of the need for clock compensation every 10,000 bytes per lane (5,000 clocks for 2-byte per lane designs; 2,500 clocks for 4-byte per lane designs), a user cannot continuously transmit data nor can data be continuously received. During clock compensation, data transfer is suspended for six clock periods.

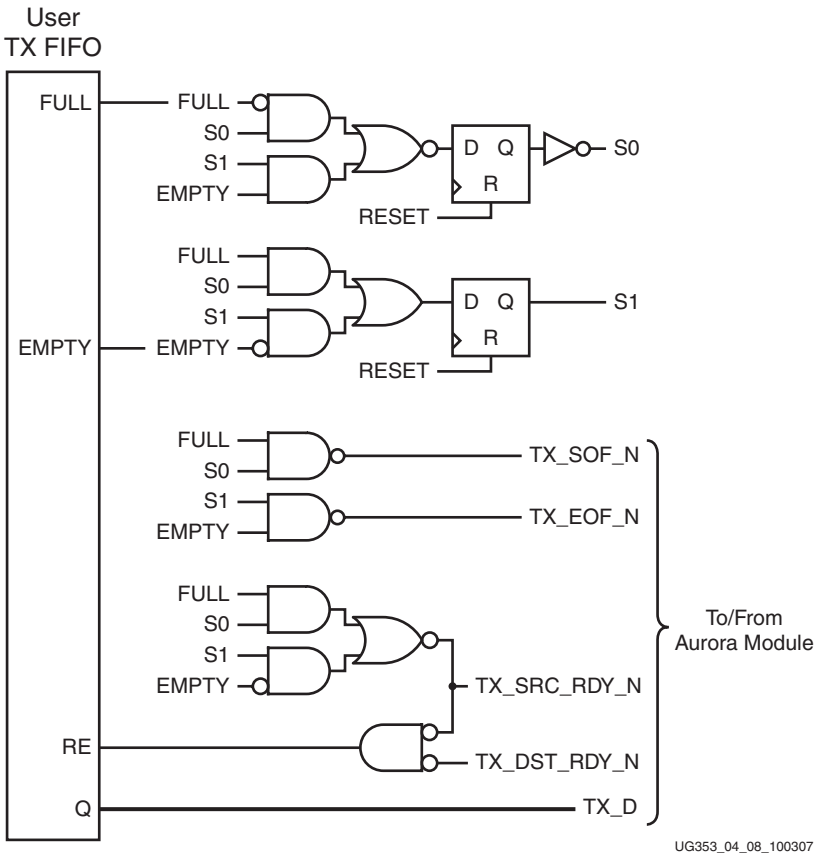

Figure 4-9: **Transmitting Data**

Receiving Data

When the Aurora 8B/10B core receives an Aurora 8B/10B frame, it presents it to the user through the RX LocalLink interface after discarding the framing characters, idles, and clock compensation sequences.

The RX_LL submodule has no built in elastic buffer for user data. As a result, there is no RX_DST_RDY_N signal on the RX LocalLink interface. The only way for the user application to control the flow of data from an Aurora 8B/10B channel is to use one of the core's optional flow control features. In most cases, a FIFO should be added to the RX data path to ensure no data is lost while flow control messages are in transit.

The Aurora 8B/10B core asserts the RX_SRC_RDY_N signal when the signals on its RX LocalLink interface are valid. Applications should ignore any values on the RX LocalLink ports sampled while RX_SRC_RDY_N is deasserted (High).

RX_SOF_N is asserted concurrently with the first word of each frame from the Aurora 8B/10B core. RX_EOF_N is asserted concurrently with the last word or partial word of each frame. The RX_REM port indicates the number of valid bytes in the final word of each frame. It uses the same encoding as TX_REM and is only valid when RX_EOF_N is asserted.

The Aurora 8B/10B core can deassert RX_SRC_RDY_N anytime, even during a frame. The timing of the RX_SRC_RDY_N deassertions is independent of the way the data was transmitted. The core can occasionally deassert RX_SRC_RDY_N even if the frame was

originally transmitted without pauses. These pauses are a result of the framing character stripping and left alignment process, as the core attempts to process each frame with as little latency as possible.

"Example A: Data Reception with Pause" shows the reception of a typical Aurora 8B/10B frame.

Example A: Data Reception with Pause

Figure 4-10 shows an example of 3*n* bytes of received data interrupted by a pause. Data is presented on the RX_D bus. When the first *n* bytes are placed on the bus, the RX_SOF_N and RX_SRC_RDY_N outputs are asserted to indicate that data is ready for the user. On the clock cycle following the first data beat, the core deasserts RX_SRC_RDY_N, indicating to the user that there is a pause in the data flow.

After the pause, the core asserts RX_SRC_RDY_N and continues to assemble the remaining data on the RX_D bus. At the end of the frame, the core asserts RX_EOF_N. The core also computes the value of RX_REM bus and presents it to the user based on the total number of valid bytes in the final word of the frame.

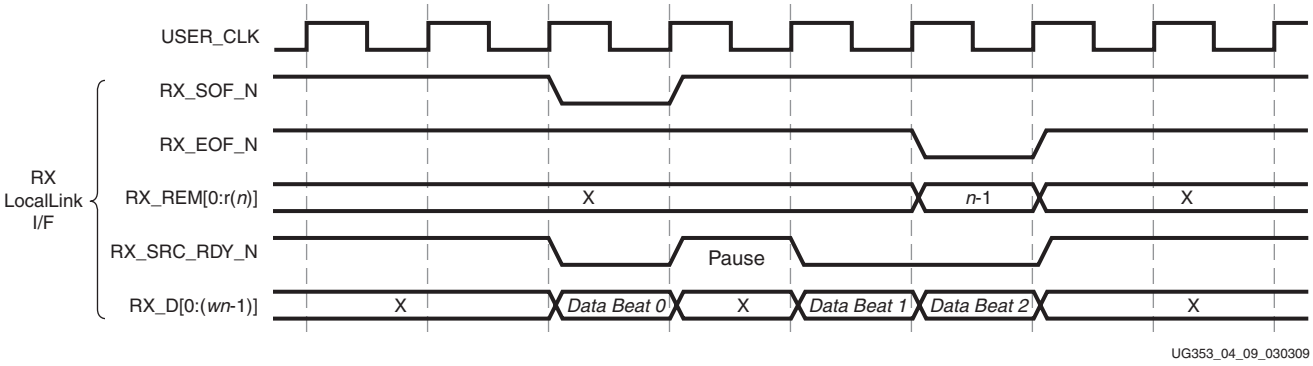

Figure 4-10: **Data Reception with Pause**

RX Interface Example

Figure 4-11 is a simple example of how a user might design an interface between the LocalLink interface of an Aurora 8B/10B core and a FIFO. To receive data, the user monitors the RX_SRC_RDY_N signal. When valid data is present on the RX_D port, RX_SRC_RDY_N is asserted. Because the inverse of RX_SRC_RDY_N is connected to the FIFO's WE port, the data and framing signals and REM value are written to the FIFO.

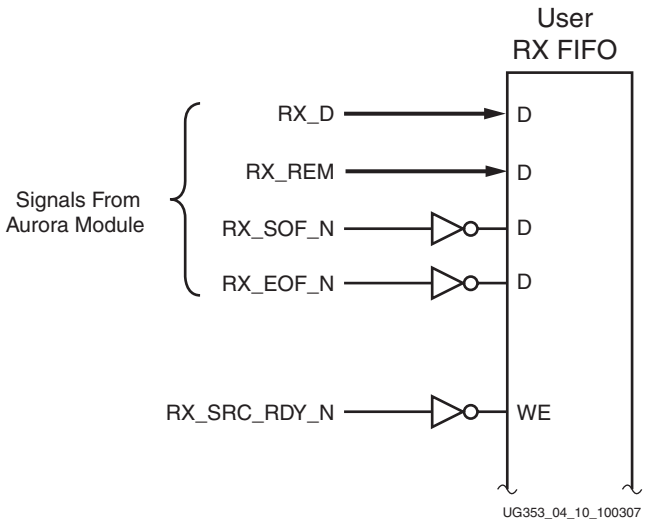

Figure 4-11: **Receiving Data**

Framing Efficiency

There are two factors that affect framing efficiency in the Aurora 8B/10B core:

- Size of the frame
- Width of the data path

The CC sequence, which uses 12 bytes on every lane every 10,000 bytes, consumes about 0.12% of the total channel bandwidth.

All bytes in Aurora 8B/10B are sent in 2-byte code groups. Aurora 8B/10B frames with an even number of bytes have four bytes of overhead, two bytes for SCP (start of frame) and two bytes for ECP (end of frame). Aurora 8B/10B frames with an odd number of bytes have five bytes of overhead, four bytes of framing overhead plus an additional byte for the pad byte that is sent to fill the second byte of the code group carrying the last byte of data in the frame.

Like many parallel interfaces, LocalLink processes data from only one frame at a time. The core must drop data when it arrives on the GTP/GTX transceiver interface at the same time as data from a previous cycle. The *LocalLink Interface Specification* includes advanced options for handling multiple frames on a single cycle, but these options are not implemented in this core.

The core transmits frame delimiters only in specific lanes of the channel. SCP is only transmitted in the left-most (most-significant) lane, and ECP is only transmitted in the right-most (least-significant) lane. Any space in the channel between the last code group with data and the ECP code group is padded with idles. The result is reduced resource cost for the design, at the expense of a minimal additional throughput cost. Though SCP and ECP could be optimized for additional throughput, the single frame per cycle limitation imposed by the user interface would make this improvement unusable in most cases.

Use the formula shown in Figure 4-12 to calculate the efficiency for a design of any number of lanes, any width of interface, and frames of any number of bytes.

Note: This formula includes the overhead for clock compensation.

$$E = \frac{100n}{n + 4 + 0.5 + IDLEs + \frac{12n}{9,988}}$$

Where:

E = The average efficiency of a specified PDU
n = Number of user data bytes
12*n*/9,988 = Clock correction overhead
4 = The overhead of SCP + ECP
0.5 = Average PAD overhead
IDLEs = The overhead for IDLEs = (*w*/2)-1
(*w* = The interface width)

UG353_04_11_100307

Figure 4-12: **Formula for Calculating Overhead**

Example

Table 4-5 is an example calculated from the formula given in Figure 4-12. It shows the efficiency for an 8-byte, 4-lane channel and illustrates that the efficiency increases as the length of channel frames increases.

Table 4-5: **Efficiency Example**

User Data Bytes	Efficiency
100	92.92%
1,000	99.14%
10,000	99.81%

Table 4-6 shows the overhead in an 8-byte, 4-lane channel when transmitting 256 bytes of frame data across the four lanes. The resulting data unit is 264 bytes long due to start and end characters, and due to the idles necessary to fill out the lanes. This amounts to 3.03% of overhead in the transmitter. In addition, a 12-byte clock compensation sequence occurs on each lane every 10,000 bytes, which adds a small amount more to the overhead. The receiver can handle a slightly more efficient data stream because it does not require any idle pattern.

Table 4-6: **Typical Overhead for Transmitting 256 Data Bytes**

Lane	Clock	Function	Character or Data Byte	
			Byte 1	Byte 2
0	1	Start of channel frame	/SCP/$_1$	/SCP/$_2$
1	1	Channel frame data	D0	D1
2	1	Channel frame data	D2	D3
3	1	Channel frame data	D4	D5
. . .				
0	33	Channel frame data	D254	D255
1	33	Transmit idles	/I/	/I/
2	33	Transmit idles	/I/	/I/
3	33	End of channel frame	/ECP/$_1$	/ECP/$_2$

Table 4-7 shows the overhead that occurs with each value of TX_REM.

Table 4-7: **TX_REM Value and Corresponding Bytes of Overhead**

TX_REM Bus Value	SCP	Pad	ECP	Idles	Total
0	2	1	2	6	11
1		0			10
2		1		4	9
3		0			8
4		1		2	7
5		0			6
6		1		0	5
7		0			4

Streaming Interface

Figure 4-13 shows an example of an Aurora 8B/10B core configured with a streaming user interface.

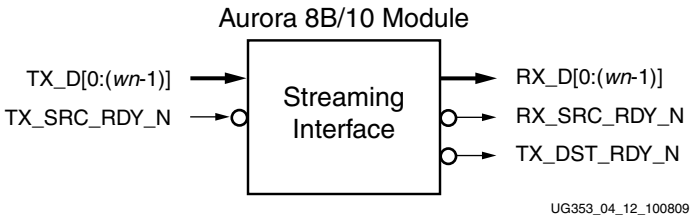

Figure 4-13: **Aurora 8B/10B Core Streaming User Interface**

Streaming TX Ports

Table 4-8 lists the streaming TX data ports. These ports are included on full-duplex, simplex TX, and simplex Both framing cores.

Table 4-8: **Streaming User I/O Ports (TX)**

Name	Direction	Description
TX_D[0:(*wn*-1)]	Input	Outgoing data (Ascending bit order).
TX_DST_RDY_N	Output	Asserted (Low) during clock edges when signals from the source will be accepted (if TX_SRC_RDY_N is also asserted). Deasserted (High) on clock edges when signals from the source will be ignored.
TX_SRC_RDY_N	Input	Asserted (Low) when LocalLink signals from the source are valid. Deasserted (High) when LocalLink control signals and/or data from the source should be ignored (active-Low).

Streaming RX Ports

Table 4-9 lists the streaming RX data ports. These ports are included on full-duplex, simplex RX, and simplex Both framing cores.

Table 4-9: **Streaming User I/O Ports (RX)**

Name	Direction	Description
RX_D[0:(*wn*-1)]	Output	Incoming data from channel partner (Ascending bit order).
RX_SRC_RDY_N	Output	Asserted (Low) when data and control signals from an Aurora 8B/10B core are valid. Deasserted (High) when data and/or control signals from an Aurora 8B/10B core should be ignored (active-Low).

Transmitting and Receiving Data

The streaming interface allows the Aurora 8B/10B channel to be used as a pipe. Words written into the TX side of the channel are delivered, in order after some latency, to the RX side. After initialization, the channel is always available for writing, except when the DO_CC signal is asserted to send clock compensation sequences. Applications transmit data through the TX_D port, and use the TX_SRC_RDY_N port to indicate when the data is valid (asserted Low). The Aurora 8B/10B core will deassert TX_DST_RDY_N (High) when the channel is not ready to receive data. Otherwise, TX_DST_RDY_N will remain asserted.

When TX_SRC_RDY_N is deasserted, gaps are created between words. These gaps are preserved, except when clock compensation sequences are being transmitted. Clock compensation sequences are replicated or deleted by the GTP/GTX transceiver to make up for frequency differences between the two sides of the Aurora 8B/10B channel. As a result, gaps created when DO_CC is asserted can shrink and grow. For details on the DO_CC signal, see Chapter 8, "Clock Compensation."

When data arrives at the RX side of the Aurora 8B/10B channel it is presented on the RX_D bus and RX_SRC_RDY is asserted. The data must be read immediately or it is lost. If this is unacceptable, a buffer must be connected to the RX interface to hold the data until it can be used.

Figure 4-14 shows a typical example of streaming data. The example begins with neither of the ready signals asserted, indicating that both the user logic and the Aurora 8B/10B core are not ready to transfer data. During the next clock cycle, the Aurora 8B/10B core indicates that it is ready to transfer data by asserting TX_DST_RDY_N. One cycle later, the user logic indicates that it is ready to transfer data by asserting the TX_D bus and the TX_SRC_RDY_N signal. Because both ready signals are now asserted, data D0 is transferred from the user logic to the Aurora 8B/10B core. Data D1 is transferred on the following clock cycle. In this example, the Aurora 8B/10B core deasserts its ready signal, TX_DST_RDY_N, and no data is transferred until the next clock cycle when, once again, the TX_DST_RDY_N signal is asserted. Then the user deasserts TX_SRC_RDY_N on the next clock cycle, and no data is transferred until both ready signals are asserted.

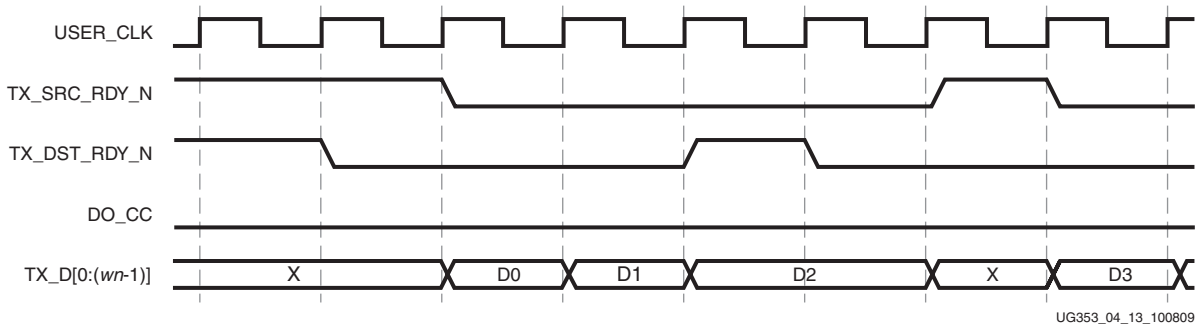

Figure 4-14: **Typical Streaming Data Transfer**

Figure 4-15 shows the receiving end of the data transfer that is shown in Figure 4-14.

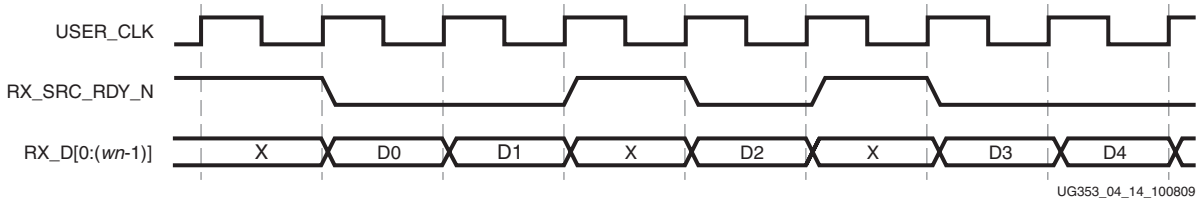

Figure 4-15: **Typical Data Reception**

Chapter 5

Flow Control

Introduction

This chapter explains how to use Aurora 8B/10B flow control. Two flow control interfaces are available as options on cores that use a framing interface. *Native flow control* (NFC) is used for regulating the data transmission rate at the receiving end a full-duplex channel. *User flow control* (UFC) is used to accommodate high priority messages for control operations.

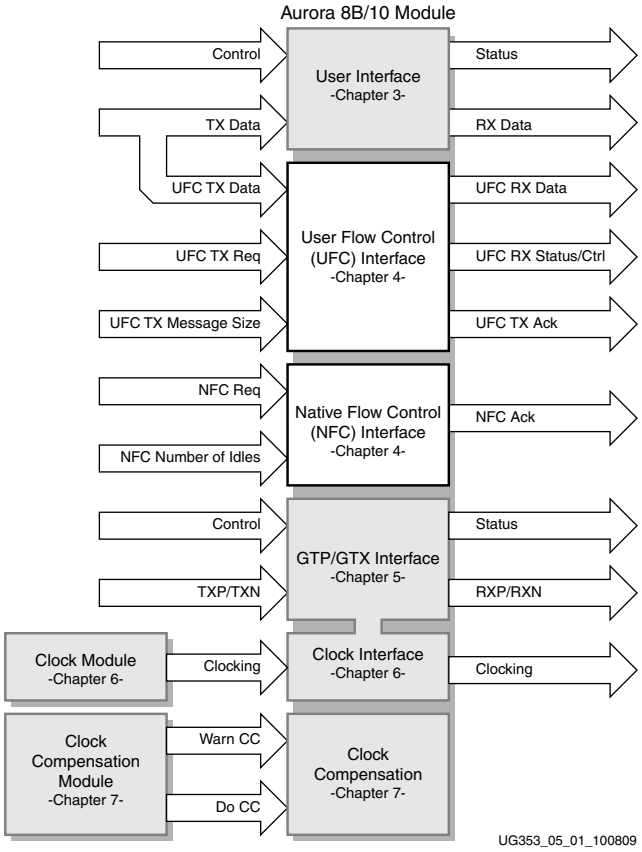

Figure 5-1: **Top-Level Flow Control**

Native Flow Control

Table 5-1 shows the codes for native flow control (NFC).

Table 5-1: **NFC Codes**

NFC_NB	Idle Cycles Requested
0000	0 (XON)
0001	2
0010	4
0011	8
0100	16
0101	32
0110	64
0111	128
1000	256
1001 to 1110	Reserved
1111	Infinite (XOFF)

Table 5-2 lists the ports for the NFC interface available only in full-duplex Aurora 8B/10B cores.

Table 5-2: **NFC I/O Ports**

Name	Direction	Description
NFC_ACK_N	Output	Asserted when an Aurora 8B/10B core accepts an NFC request (active-Low).
NFC_NB[0:3]	Input	Indicates the number of PAUSE idles the channel partner must send when it receives the NFC message. Must be held until NFC_ACK_N is asserted.
NFC_REQ_N	Input	Asserted to request an NFC message be sent to the channel partner (active-Low). Must be held until NFC_ACK_N is asserted.

The Aurora 8B/10B protocol includes native flow control (NFC) to allow receivers to control the rate at which data is sent to them by specifying a number of idle data beats that must be placed into the data stream. The data flow can even be turned off completely by requesting that the transmitter temporarily send only idles (XOFF). NFC is typically used to prevent FIFO overflow conditions. For detailed explanation of NFC operation and NFC codes, see the *Aurora 8B/10B Protocol Specification*.

To send an NFC message to a channel partner, the user application asserts NFC_REQ_N and writes an NFC code to NFC_NB. The NFC code indicates the minimum number of idle cycles the channel partner should insert in its TX data stream. The user application must hold NFC_REQ_N and NFC_NB until NFC_ACK_N is asserted on a positive USER_CLK edge, indicating the Aurora 8B/10B core will transmit the NFC message. Aurora 8B/10B cores cannot transmit data while sending NFC messages. TX_DST_RDY_N is always deasserted on the cycle following an NFC_ACK_N assertion.

Example A: Transmitting an NFC Message

Figure 5-2 shows an example of the transmit timing when the user sends an NFC message to a channel partner.

Note: TX_DST_RDY_N is deasserted for one cycle (assumes that *n* is at least 2) to create the gap in the data flow in which the NFC message is placed.

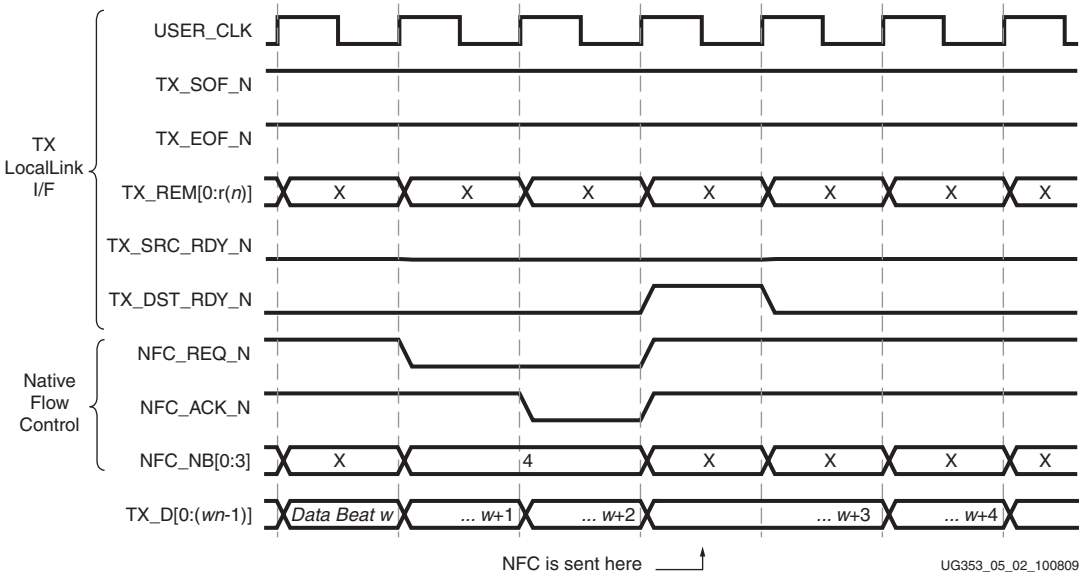

Figure 5-2: **Transmitting an NFC Message**

Example B: Receiving a Message with NFC Idles Inserted

Figure 5-3 shows an example of the signals on the TX user interface when an NFC message is received. In this case, the NFC message has a code of `0001`, requesting two data beats of idles. The core deasserts TX_DST_RDY_N on the user interface until enough idles have been sent to satisfy the request. In this example, the core is operating in immediate NFC mode. Aurora 8B/10B cores can also operate in completion mode, where NFC idles are only inserted between frames. If a completion mode core receives an NFC message while it is transmitting a frame, it finishes transmitting the frame before deasserting TX_DST_RDY_N to insert idles.

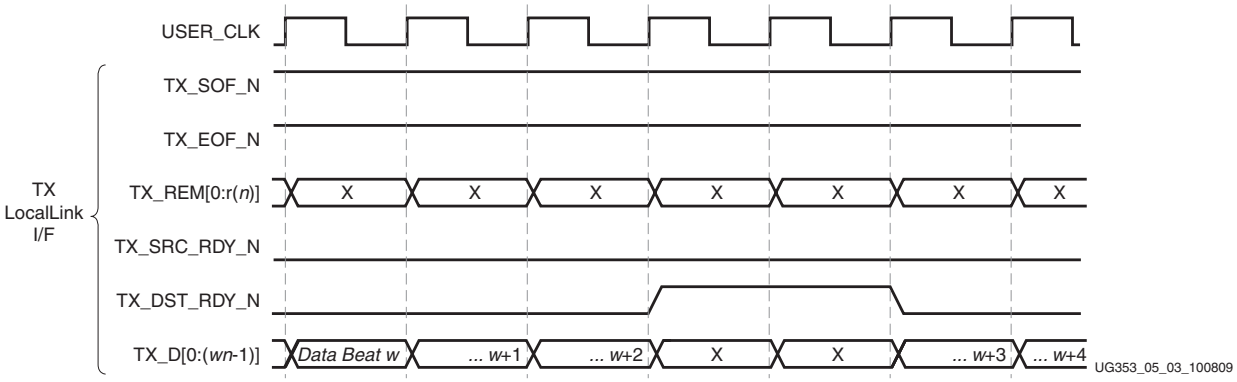

Figure 5-3: **Transmitting a Message with NFC Idles Inserted**

User Flow Control

The Aurora 8B/10B protocol includes user flow control (UFC) to allow channel partners to send control information using a separate in-band channel. The user can send short UFC messages to the core's channel partner without waiting for the end of a frame in progress. The UFC message shares the channel with regular frame data, but has a higher priority.

Table 5-3 describes the ports for the UFC interface.

Table 5-3: **UFC I/O Ports**

Name	Direction	Description
UFC_TX_REQ_N	Input	Asserted to request a UFC message be sent to the channel partner (active-Low). Must be held until UFC_TX_ACK_N is asserted. Do not assert this signal unless the entire UFC message is ready to be sent; a UFC message cannot be interrupted once it has started.
UFC_TX_MS[0:2]	Input	Specifies the size of the UFC message that will be sent. The SIZE encoding is a value between 0 and 7. See Table 5-4, page 57.
UFC_TX_ACK_N	Output	Asserted when an Aurora 8B/10B core is ready to read the contents of the UFC message (active-Low). On the cycle after the ACK signal is asserted, data on the TX_D port will be treated as UFC data. TX_D data continues to be used to fill the UFC message until enough cycles have passed to send the complete message. Unused bytes from a UFC cycle are discarded.
UFC_RX_DATA[0:(8*n*-1)]	Output	Incoming UFC message data from the channel partner (n = 16 bytes max).
UFC_RX_SRC_RDY_N	Output	Asserted when the values on the UFC_RX ports are valid. When this signal is not asserted, all values on the UFC_RX ports should be ignored (active-Low).
UFC_RX_SOF_N	Output	Signals the start of the incoming UFC message (active-Low).
UFC_RX_EOF_N	Output	Signals the end of the incoming UFC message (active-Low).
UFC_RX_REM[0:r(n)]	Output	Specifies the number of valid bytes of data presented on the UFC_RX_DATA port on the last word of a UFC message. Valid only when UFC_RX_EOF_N is asserted. n = 16 bytes max. REM bus widths are given by [0:r(n)], where r(n) = ceiling [{$\log_2(n)$}-1].

Transmitting UFC Messages

UFC messages can carry an even number of data bytes from 2 to 16. The user application specifies the length of the message by driving a SIZE code on the UFC_TX_MS port. Table 5-4 shows the legal SIZE code values for UFC.

Table 5-4: **SIZE Encoding**

SIZE Field Contents	UFC Message Size
000	2 bytes
001	4 bytes
010	6 bytes
011	8 bytes
100	10 bytes
101	12 bytes
110	14 bytes
111	16 bytes

To send a UFC message, the user application asserts UFC_TX_REQ_N while driving the UFC_TX_MS port with the desired SIZE code. UFC_TX_REQ_N must be held until the Aurora 8B/10B core asserts the UFC_TX_ACK_N signal, indicating that the core is ready to send the UFC message. The data for the UFC message must be placed on the TX_D port of the data interface, starting on the first cycle after UFC_TX_ACK_N is asserted. The core deasserts TX_DST_RDY_N while the TX_D port is being used for UFC data.

Figure 5-4 shows a useful circuit for switching TX_D from sending regular data to UFC data.

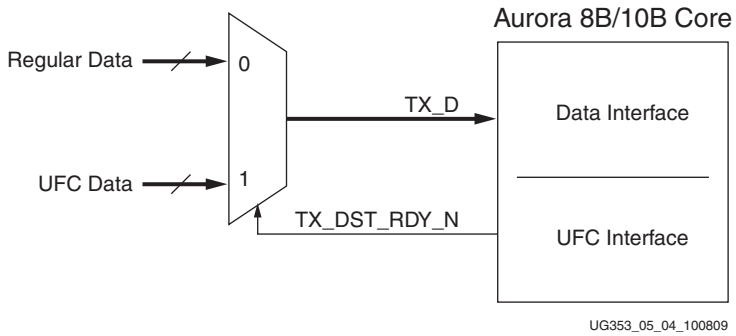

Figure 5-4: **Data Switching Circuit**

Table 5-5, page 58 shows the number of cycles required to transmit UFC messages of different sizes based on the width of the LocalLink data interface. UFC messages should never be started until all message data is available. Unlike regular data, UFC messages cannot be interrupted after UFC_TX_ACK_N has been asserted.

Table 5-5: **Number of Data Beats Required to Transmit UFC Messages**

UFC Message	UFC_TX_MS Value	LL I/F Width	Number of Data Beats	LL I/F Width	Number of Data Beats
2 Bytes	0	2 Bytes	1	10 Bytes	1
4 Bytes	1		2		
6 Bytes	2		3		
8 Bytes	3		4		
10 Bytes	4		5		
12 Bytes	5		6		2
14 Bytes	6		7		
16 Bytes	7		8		
2 Bytes	0	4 Bytes	1	12 Bytes	1
4 Bytes	1				
6 Bytes	2		2		
8 Bytes	3				
10 Bytes	4		3		
12 Bytes	5				
14 Bytes	6		4		2
16 Bytes	7				
2 Bytes	0	6 Bytes	1	14 Bytes	1
4 Bytes	1				
6 Bytes	2				
8 Bytes	3		2		
10 Bytes	4				
12 Bytes	5				
14 Bytes	6		3		
16 Bytes	7				2
2 Bytes	0	8 Bytes	1	16 Bytes or more	1
4 Bytes	1				
6 Bytes	2				
8 Bytes	3				
10 Bytes	4		2		
12 Bytes	5				
14 Bytes	6				
16 Bytes	7				

Example A: Transmitting a Single-Cycle UFC Message

The procedure for transmitting a single cycle UFC message is shown in Figure 5-5. In this case a 4-byte message is being sent on a 2-byte interface.

Note: TX_DST_RDY_N is deasserted for two cycles. Aurora 8B/10B cores use this gap in the data flow to transmit the UFC header and message data.

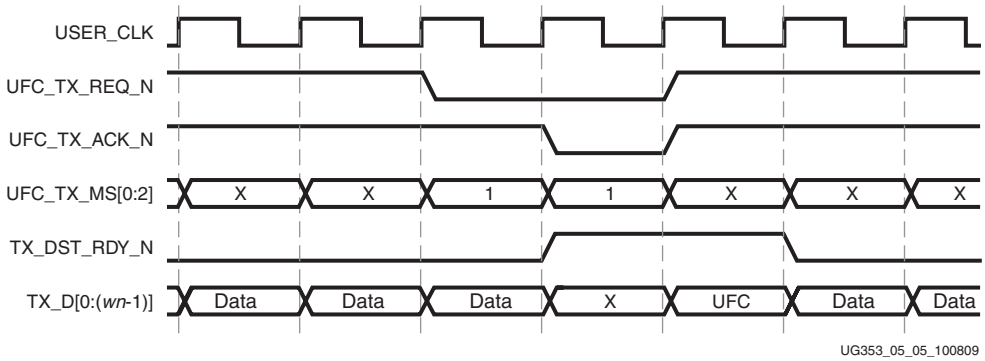

Figure 5-5: **Transmitting a Single-Cycle UFC Message**

Example B: Transmitting a Multi-Cycle UFC Message

The procedure for transmitting a two-cycle UCF message is shown in Figure 5-6. In this case the user application is sending a 16-byte message using a 2-byte interface. TX_DST_RDY_N is asserted for three cycles; one cycle for the UFC header which is sent during the UFC_TX_ACK_N cycle, and two cycles for data.

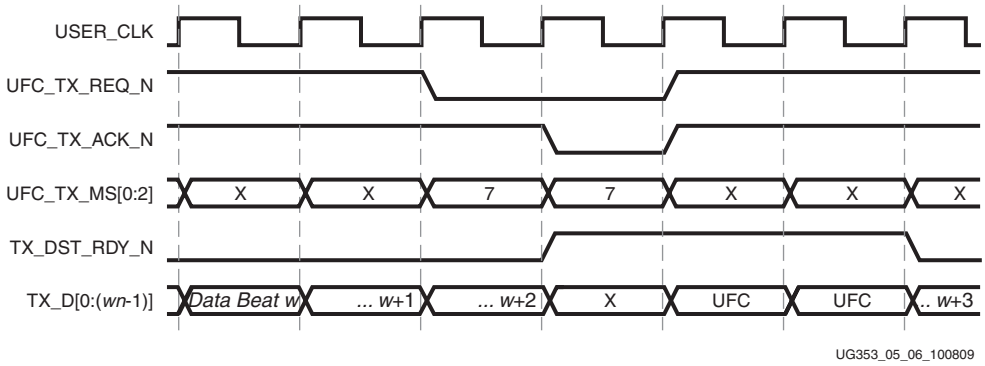

Figure 5-6: **Transmitting a Multi-Cycle UFC Message**

Receiving User Flow Control Messages

When the Aurora 8B/10B core receives a UFC message, it passes the data from the message to the user application through a dedicated UFC LocalLink interface. The data is presented on the UFC_RX_DATA port; UFC_RX_SOF_N indicates the start of the message data and UFC_RX_EOF_N indicates the end. UFC_RX_REM is used to show the number of valid bytes on UFC_RX_DATA during the last cycle of the message (for example, while UFC_RX_EOF_N is asserted). Signals on the UFC_RX LocalLink interface are only valid when UFC_RX_SRC_RDY_N is asserted.

Example C: Receiving a Single-Cycle UFC Message

Figure 5-7 shows an Aurora 8B/10B core with a 4-byte data interface receiving a 4-byte UFC message. The core presents this data to the user application by asserting UFC_RX_SRC_RDY_N, UFC_RX_SOF_N and UFC_RX_EOF_N to indicate a single cycle frame. The UFC_RX_REM bus is set to 3, indicating only the four most significant bytes of the interface are valid.

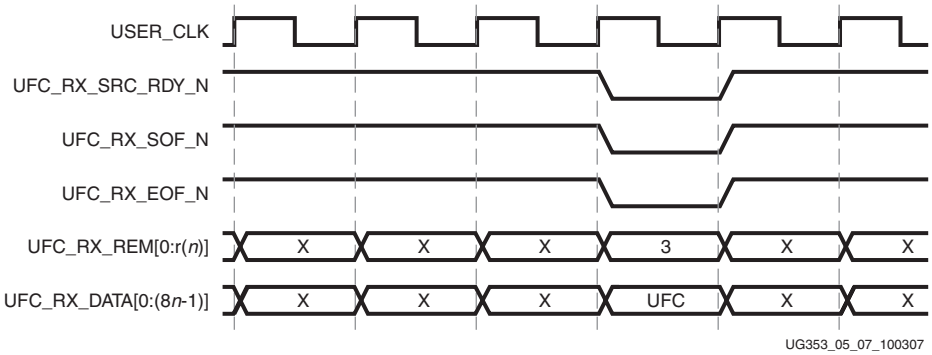

Figure 5-7: **Receiving a Single-Cycle UFC Message**

Example D: Receiving a Multi-Cycle UFC Message

Figure 5-8 shows an Aurora 8B/10B core with a 4-byte interface receiving an 8-byte message.

Note: The resulting frame is two cycles long, with UFC_RX_REM set to 3 on the second cycle indicating that all eight bytes of the data are valid.

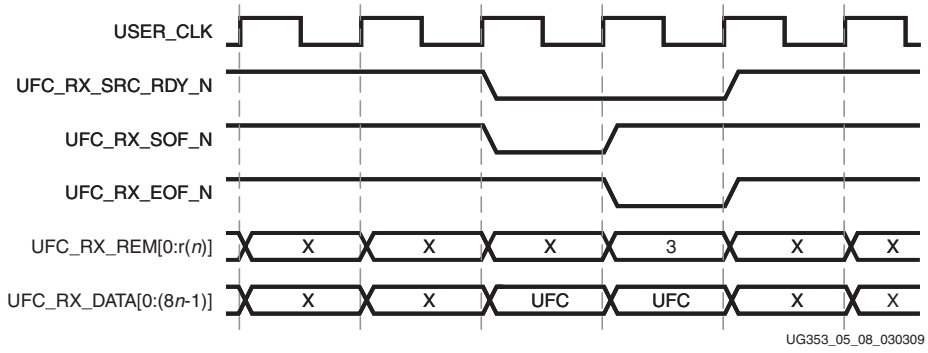

Figure 5-8: **Receiving a Multi-Cycle UFC Message**

Chapter 6

Status, Control, and the GTP/GTX Block Interface

Introduction

The status and control ports of the Aurora 8B/10B core allow user applications to monitor the Aurora 8B/10B channel and use built-in features of the GTP/GTX transceivers. Aurora 8B/10B cores can be configured as full-duplex or simplex modules. Full-duplex modules provide high-speed TX and RX links. Simplex modules provide a link in only one direction and are initialized using sideband ports. This chapter provides diagrams and port descriptions for the Aurora 8B/10B core's status and control interface, along with the GTP/GTX transceiver serial I/O interface and the sideband initialization ports that are used exclusively for simplex modules.

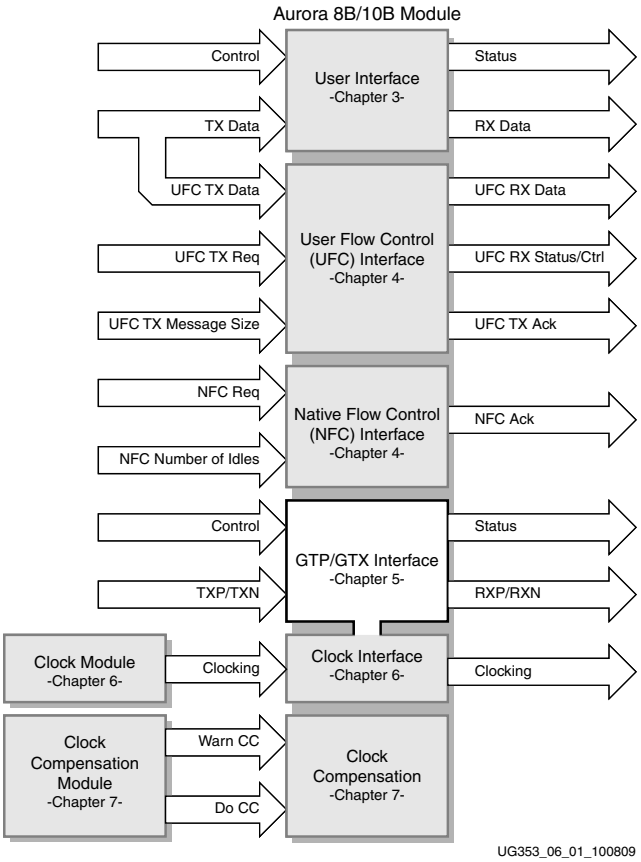

Figure 6-1: **Top-Level GTP/GTX Block Interface**

Full-Duplex Cores

Full-Duplex Status and Control Ports

Full-duplex cores provide a TX and an RX Aurora 8B/10B channel connection. Figure 6-2 shows the status and control interface for a full-duplex Aurora 8B/10B core. Table 6-1 describes the function of each of the ports in the interface.

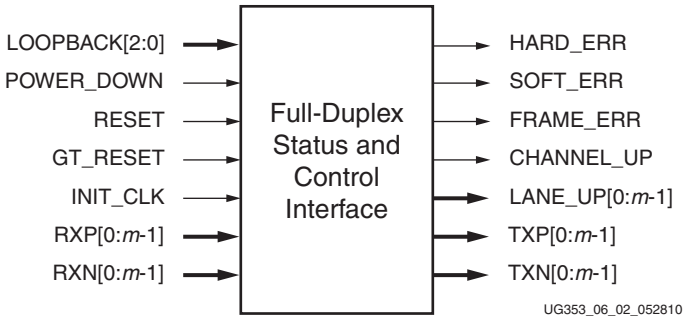

Figure 6-2: **Status and Control Interface for Full-Duplex Cores**

Table 6-1: **Status and Control Ports for Full-Duplex Cores**

Name	Direction	Description
CHANNEL_UP	Output	Asserted when Aurora 8B/10B channel initialization is complete and channel is ready to send data. The Aurora 8B/10B core cannot receive data before CHANNEL_UP.
LANE_UP[0:*m*-1]	Output	Asserted for each lane upon successful lane initialization, with each bit representing one lane (active-High). The Aurora 8B/10B core can only receive data after all LANE_UP signals are High.
FRAME_ERR	Output	Channel frame/protocol error detected. This port is active-High and is asserted for a single clock.
HARD_ERR	Output	Hard error detected. (Active-High, asserted until Aurora 8B/10B core resets). See "Error Signals in Full-Duplex Cores," page 63 for more details.
LOOPBACK[2:0]	Input	The LOOPBACK[2:0] port selects between the normal operation mode and the different loopback modes. See the *Virtex-5 FPGA RocketIO GTP Transceiver User Guide*, *Virtex-5 FPGA RocketIO GTX Transceiver User Guide*, *Virtex-6 FPGA GTX Transceivers User Guide*, and *Spartan-6 FPGA GTP Transceivers User Guide* for details about loopback. See "Related Xilinx Documents" in Chapter 1.
POWER_DOWN	Input	Drives the power-down input of the GTP/GTX transceiver (active-High).
RESET	Input	Resets the Aurora 8B/10B core (active-High). This signal must be synchronous to USER_CLK and must be asserted for at least one USER_CLK cycle.

Table 6-1: **Status and Control Ports for Full-Duplex Cores *(Cont'd)***

Name	Direction	Description
SOFT_ERR	Output	Soft error detected in the incoming serial stream. See "Error Signals in Full-Duplex Cores," page 63 for more details. (Active-High, asserted for a single clock).
RXP[0:*m*-1]	Input	Positive differential serial data input pin.
RXN[0:*m*-1]	Input	Negative differential serial data input pin.
TXP[0:*m*-1]	Output	Positive differential serial data output pin.
TXN[0:*m*-1]	Output	Negative differential serial data output pin.
GT_RESET	Input	The reset signal for the PMA modules in the transceivers is connected to the top level through a debouncer. The GT_RESET should be asserted (active-High) when the module is first powered up in hardware. This systematically resets all PCS and PMA subcomponents of the transceiver. The signal is debounced using the INIT_CLK. See the Reset section in the respective transceiver user guide for further details.
INIT_CLK	Input	INIT_CLK is used to register and debounce the GT_RESET signal in cores targeted for the Virtex- 5 device. INIT_CLK is required since USER_CLK stops when GT_RESET is asserted. INIT_CLK should be set to a slow rate, preferably slower than the reference clock. INIT_CLK is a board clock. For example, the ML523 board has a 50 MHz crystal oscillator and it is constrained for this frequency by default in the <component_name>_example_design.ucf file. Users needs to update this clock constraint with respect to their board clock frequency.

Notes:

1. *m* is the number of GTP/GTX transceivers

Error Signals in Full-Duplex Cores

Equipment problems and channel noise can cause errors during Aurora 8B/10B channel operation. 8B/10B encoding allows the Aurora 8B/10B core to detect all single bit errors and most multi-bit errors that occur in the channel. The core reports these errors by asserting the SOFT_ERR signal on every cycle they are detected.

The core also monitors each GTP/GTX transceiver for hardware errors such as buffer overflow and loss of lock. The core reports hardware errors by asserting the HARD_ERR signal. Catastrophic hardware errors can also manifest themselves as burst of soft errors. The core uses the leaky bucket algorithm described in the *Aurora 8B/10B Protocol Specification* to detect large numbers of soft errors occurring in a short period of time, and will assert the HARD_ERR signal when it detects them.

Whenever a hard error is detected, the Aurora 8B/10B core automatically resets itself and attempts to reinitialize. In most cases, this will allow the Aurora 8B/10B channel to be

reestablished as soon as the hardware issue that caused the hard error is resolved. Soft errors do not lead to a reset unless enough of them occur in a short period of time to trigger the Aurora 8B/10B leaky bucket algorithm.

Aurora 8B/10B cores with a LocalLink data interface can also detect errors in Aurora 8B/10B frames. Errors of this type include frames with no data, consecutive Start of Frame symbols, and consecutive End of Frame symbols. When the core detects a frame problem, it asserts the FRAME_ERR signal. This signal is usually asserted close to a SOFT_ERR assertion, with soft errors being the main cause of frame errors.

Table 6-2 summarizes the error conditions the Aurora 8B/10B core can detect and the error signals used to alert the user application.

Table 6-2: **Error Signals in Full-Duplex Cores**

Signal	Description
HARD_ERR	TX Overflow/Underflow: The elastic buffer for TX data overflows or underflows. This can occur when the user clock and the reference clock sources are not running at the same frequency. RX Overflow/Underflow: The elastic buffer for RX data overflows or underflows. This can occur when the clock source frequencies for the two channel partners are not within ± 100 ppm. Bad Control Character: The protocol engine attempts to send a bad control character. This is an indication of design corruption or catastrophic failure. Soft Errors: There are too many soft errors within a short period of time. The Aurora 8B/10B protocol defines a leaky bucket algorithm for determining the acceptable number of soft errors within a given time period. When this number is exceeded, the physical connection may be too poor for communication using the current voltage swing and pre-emphasis settings.
SOFT_ERR	Invalid Code: The 10-bit code received from the channel partner was not a valid code in the 8B/10B table. This usually means a bit was corrupted in transit, causing a good code to become unrecognizable. Typically, this will also result in a frame error or corruption of the current channel frame. Disparity Error: The 10-bit code received from the channel partner did not have the correct disparity. This error is also usually caused by corruption of a good code in transit, and can result in a frame error or bad data if it occurs while a frame is being sent.
FRAME_ERR	Truncated Frame: A channel frame is started without ending the previous channel frame, or a channel frame is ended without being started. Invalid Control Character: The protocol engine receives a control character that it does not recognize. No Data in Frame: A channel frame is received with no data.

Full-Duplex Initialization

Full-duplex cores initialize automatically after power up, reset, or hard error. Full-duplex modules on each side of the channel perform the Aurora 8B/10B initialization procedure until the channel is ready for use. The LANE_UP bus indicates which lanes in the channel have finished the lane initialization portion of the initialization procedure. This signal can be used to help debug equipment problems in a multi-lane channel. CHANNEL_UP is asserted only after the core completes the entire initialization procedure.

Aurora 8B/10B cores cannot receive data before CHANNEL_UP is asserted. Only the RX_SRC_RDY_N signal on the user interface should be used to qualify incoming data. CHANNEL_UP can be inverted and used to reset modules that drive the TX side of a full-duplex channel, since no transmission can occur until after CHANNEL_UP. If user application modules need to be reset before data reception, one of the LANE_UP signals can be inverted and used. Data cannot be received until after all the LANE_UP signals are asserted.

Simplex Cores

Simplex TX Status and Control Ports

Simplex TX cores allow user applications to transmit data to a simplex RX core. They have no RX connection. Figure 6-3 shows the status and control interface for a simplex TX core. Table 6-3 describes the function of each of the ports in the interface.

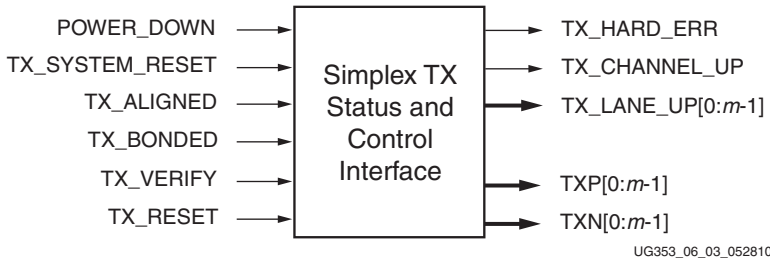

Figure 6-3: **Status and Control Interface for Simplex TX Core**

Table 6-3: **Status and Control Ports for Simplex TX Cores**

Name	Direction	Description
TX_ALIGNED	Input	Asserted when RX channel partner has completed lane initialization for all lanes. Typically connected to RX_ALIGNED.
TX_BONDED	Input	Asserted when RX channel partner has completed channel bonding. Not needed for single-lane channels. Typically connected to RX_BONDED.
TX_VERIFY	Input	Asserted when RX channel partner has completed verification. Typically connected to RX_VERIFY.

Table 6-3: **Status and Control Ports for Simplex TX Cores *(Cont'd)***

Name	Direction	Description
TX_RESET	Input	Asserted when reset is required because of initialization status of RX channel partner. This signal must be synchronous to USER_CLK and must be asserted for at least one USER_CLK cycle. Typically connected to RX_RESET.
TX_CHANNEL_UP	Output	Asserted when Aurora 8B/10B channel initialization is complete and channel is ready to send data. The Aurora 8B/10B core cannot receive data before TX_CHANNEL_UP.
TX_LANE_UP[0:*m*-1]	Output	Asserted for each lane upon successful lane initialization, with each bit representing one lane (active-High).
TX_HARD_ERR	Output	Hard error detected. (Active-High, asserted until Aurora 8B/10B core resets). See "Error Signals in Simplex Cores," page 71 for more details.
POWER_DOWN	Input	Drives the powerdown input of the GTP/GTX transceiver (active-High).
TX_SYSTEM_RESET	Input	Resets the Aurora 8B/10B core (active-High).
TXP[0:*m*-1]	Output	Positive differential serial data output pin.
TXN[0:*m*-1]	Output	Negative differential serial data output pin.

Notes:

1. *m* is the number of GTP/GTX transceivers.

Simplex RX Status and Control Ports

Simplex RX cores allow user applications to receive data from a simplex TX core. Figure 6-4 shows the status and control interface for a simplex RX core. Table 6-4, page 67 describes the function of each of the ports in the interface.

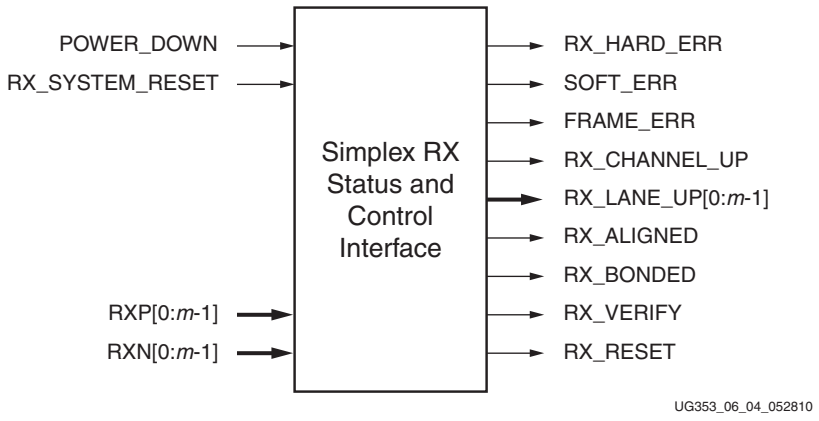

Figure 6-4: **Status and Control Interface for Simplex RX Core**

Table 6-4: **Status and Control Ports for Simplex RX Cores**

Name	Direction	Description
RX_ALIGNED	Output	Asserted when RX module has completed lane initialization. Typically connected to TX_ALIGNED.
RX_BONDED	Output	Asserted when RX module has completed channel bonding. Not used for single-lane channels. Typically connected to TX_BONDED.
RX_VERIFY	Output	Asserted when RX module has completed verification. Typically connected to TX_VERIFY.
RX_RESET	Output	Asserted when the RX module needs the TX module to restart initialization. Typically connected to TX_RESET.
RX_CHANNEL_UP	Output	Asserted when Aurora 8B/10B channel initialization is complete and channel is ready to send data. The Aurora 8B/10B core cannot receive data before RX_CHANNEL_UP.
RX_LANE_UP[0:*m*-1]	Output	Asserted for each lane upon successful lane initialization, with each bit representing one lane (active-High). The Aurora 8B/10B core can only receive data after all RX_LANE_UP signals are High.
FRAME_ERR	Output	Channel frame/protocol error detected. This port is active-High and is asserted for a single clock.
RX_HARD_ERR	Output	Hard error detected. (Active-High, asserted until Aurora 8B/10B core resets). See "Error Signals in Simplex Cores," page 71 for more details.
POWER_DOWN	Input	Drives the power-down input of the GTP/GTX transceiver (active-High).
RX_SYSTEM_RESET	Input	Resets the Aurora 8B/10B core (active-High).
SOFT_ERR	Output	Soft error detected in the incoming serial stream. See "Error Signals in Simplex Cores," page 71 for more details. (Active-High, asserted for a single clock).
RXP[0:*m*-1]	Input	Positive differential serial data input pin.
RXN[0:*m*-1]	Input	Negative differential serial data input pin.

Notes:

1. *m* is the number of GTP/GTX transceivers.
2. RX_ALIGNED, RX_BONDED, RX_VERIFY, and RX_RESET are available as output signals even when the simplex partner is timer based, but functionally these signals are not required.

Simplex Both (RX and TX) Status and Control Ports

Simplex Both cores consist of a simplex TX core and a simplex RX core sharing the same set of GTP/GTX transceivers. Like a full-duplex core, the simplex Both core transmits and receives data. Two key differences are as follows:

- The TX and RX sides of the simplex Both core initialize and run independently of each other, unlike the duplex core, where both TX and RX must be operational for either direction to work.
- Simplex Both cores only connect to other simplex cores, while full-duplex cores only connect to other full-duplex cores. The TX side of a simplex Both core connects to a simplex RX core, or to the RX side of a simplex Both core; the RX side of a simplex Both core connects to a simplex TX core, or to the TX side of a simplex Both core.

Figure 6-5 shows the status and control interface for a simplex Both core. Table 6-5, page 69 describes the function of each of the ports in the interface.

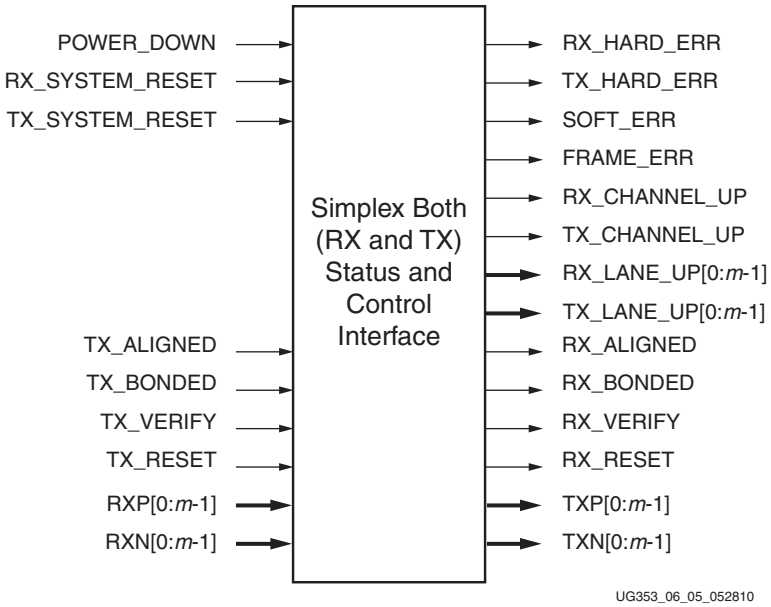

Figure 6-5: **Status and Control Interface for Simplex Both Cores**

Table 6-5: **Status and Control Ports for Simplex Both Cores**

Name	Direction	Description
TX_ALIGNED	Input	Asserted when RX channel partner has completed lane initialization for all lanes. Typically connected to RX_ALIGNED.
RX_ALIGNED	Output	Asserted when RX module has completed lane initialization. Typically connected to TX_ALIGNED.
TX_BONDED	Input	Asserted when RX channel partner has completed channel bonding. Not needed for single-lane channels. Typically connected to RX_BONDED.
RX_BONDED	Output	Asserted when RX module has completed channel bonding. Not used for single-lane channels. Typically connected to TX_BONDED.
TX_VERIFY	Input	Asserted when RX channel partner has completed verification. Typically connected to RX_VERIFY.
RX_VERIFY	Output	Asserted when RX module has completed verification. Typically connected to TX_VERIFY.
TX_RESET	Input	Asserted when reset is required because of initialization status of RX channel partner. Typically connected to RX_RESET.
RX_RESET	Output	Asserted when the RX module needs the TX module to restart initialization. Typically connected to TX_RESET.
RX_CHANNEL_UP	Output	Asserted when Aurora 8B/10B channel initialization is complete and channel is ready to receive data. The Aurora 8B/10B core cannot receive data before CHANNEL_UP.
TX_CHANNEL_UP	Input	Asserted when Aurora 8B/10B channel initialization is complete and channel is ready to send data. The Aurora 8B/10B core cannot receive data before CHANNEL_UP.
RX_LANE_UP[0:*m*-1]	Output	Asserted for each lane upon successful lane initialization, with each bit representing one lane (active-High). The Aurora 8B/10B core can only receive data after all LANE_UP signals are High.
TX_LANE_UP[0:*m*-1]	Input	Asserted for each lane upon successful lane initialization, with each bit representing one lane (active-High). The Aurora 8B/10B core can only transmit data after all LANE_UP signals are High.
FRAME_ERR	Output	Channel frame/protocol error detected. This port is active-High and is asserted for a single clock.
RX_HARD_ERR TX_HARD_ERR	Output	Hard error detected. (Active-High, asserted until Aurora 8B/10B core resets). See "Error Signals in Simplex Cores," page 71 for more details.

Table 6-5: **Status and Control Ports for Simplex Both Cores *(Cont'd)***

Name	Direction	Description
POWER_DOWN	Input	Drives the power-down input of the GTP/GTX transceiver (active-High).
RX_SYSTEM_RESET TX_SYSTEM_RESET	Input	Resets the Aurora 8B/10B core (active-High).
SOFT_ERR	Output	Soft error detected in the incoming serial stream. See "Error Signals in Simplex Cores," page 71 for more details. (Active-High, asserted for a single clock).
RXP[0:*m*-1]	Input	Positive differential serial data input pin.
RXN[0:*m*-1]	Input	Negative differential serial data input pin.
TXP[0:*m*-1]	Output	Positive differential serial data output pin.
TXN[0:*m*-1]	Output	Negative differential serial data output pin.

Notes:

1. *m* is the number of GTP/GTX transceivers.

Error Signals in Simplex Cores

The 8B/10B encoding allows RX simplex cores and the RX sides of simplex Both cores to detect all single bit errors and most multi-bit errors in a simplex channel. The cores report these errors by asserting the SOFT_ERR signal on every cycle an error is detected. The TX simplex cores do not include a SOFT_ERR port. All transmit data is assumed correct at transmission unless there is an equipment problem.

All simplex cores monitor their GTP/GTX transceivers for hardware errors such as buffer overflow and loss of lock. Hardware errors on the TX side of the channel are reported by asserting the TX_HARD_ERR signal; RX side hard errors are reported using the RX_HARD_ERR signal. Simplex RX and simplex Both cores use the Aurora 8B/10B protocol's leaky bucket algorithm to evaluate bursts of soft errors. If too many soft errors occur in a short span of time, RX_HARD_ERR is asserted.

Whenever a hard error is detected, the Aurora 8B/10B core automatically resets itself and attempts to reinitialize. In simplex Both cores, TX hard errors reset only the TX side, and RX errors reset only the RX side. Resetting allows the Aurora 8B/10B channel to be re-established as soon as the hardware issue that caused the hard error is resolved in most cases. Soft errors do not lead to a reset unless enough of them occur in a short period of time to trigger the Aurora 8B/10B leaky bucket algorithm.

Simplex RX and simplex Both cores with a LocalLink data interface can also detect errors in Aurora 8B/10B frames when they are received. Errors of this type include frames with no data, consecutive Start of Frame symbols, and consecutive End of Frame symbols. When the core detects a frame problem, it asserts the FRAME_ERR signal. This signal will usually be asserted close to a SOFT_ERR assertion, as soft errors are the main cause of frame errors. Simplex TX modules do not use the FRAME_ERR port.

Table 6-6 summarizes the error conditions simplex Aurora 8B/10B cores can detect and the error signals uses to alert the user application.

Table 6-6: **Error Signals in Simplex Cores**

Signal	Description	TX	RX	Both
HARD_ERR	TX Overflow/Underflow: The elastic buffer for TX data overflows or underflows. This can occur when the user clock and the reference clock sources are not running at the same frequency.	x		x
	RX Overflow/Underflow: The elastic buffer for RX data overflows or underflows. This can occur when the clock source frequencies for the two channel partners are not within ± 100 ppm.		x	x
	Bad Control Character: The protocol engine attempts to send a bad control character. This is an indication of design corruption or catastrophic failure.	x		x
	Soft Errors: There are too many soft errors within a short period of time. The Aurora 8B/10B protocol defines a leaky bucket algorithm for determining the acceptable number of soft errors within a given time period. When this number is exceeded, the physical connection may be too poor for communication using the current voltage swing and pre-emphasis settings.		x	x

Table 6-6: **Error Signals in Simplex Cores *(Cont'd)***

Signal	Description	TX	RX	Both
SOFT_ERR	Invalid Code: The 10-bit code received from the channel partner was not a valid code in the 8B/10B table. This usually means a bit was corrupted in transit, causing a good code to become unrecognizable. Typically, this will also result in a frame error or corruption of the current channel frame.		x	x
	Disparity Error: The 10-bit code received from the channel partner did not have the correct disparity. This error is also usually caused by corruption of a good code in transit, and can result in a frame error or bad data if it occurs while a frame is being sent.		x	x
	No Data in Frame: A channel frame is received with no data.		x	x
FRAME_ERR	Truncated Frame: A channel frame is started without ending the previous channel frame, or a channel frame is ended without being started.	x		x
	Invalid Control Character: The protocol engine receives a control character that it does not recognize.		x	x
	Invalid UFC Message Length: A UFC message is received with an invalid length.		x	x

Simplex Initialization

Simplex cores do not depend on signals from an Aurora 8B/10B channel for initialization. Instead, the TX and RX sides of simplex channels communicate their initialization state through a set of sideband initialization signals. The initialization ports are called ALIGNED, BONDED, VERIFY, and RESET; one set for the TX side with a TX_ prefix, and one set for the RX side with an RX_ prefix. The bonded port is only used for multi-lane cores.

There are two ways to initialize a simplex module using the sideband initialization signals:

- Send the information from the RX sideband initialization ports to the TX sideband initialization ports
- Drive the TX sideband initialization ports independently of the RX sideband initialization ports using timed initialization intervals

Both initialization methods are described in the following sections.

Using a Back Channel

If there is no communication channel available from the RX side of the connection to the TX side, using a back channel is the safest way to initialize and maintain a simplex channel. There are very few requirements on the back channel; it need only deliver messages to the TX side to indicate which of the sideband initialization signals is asserted when the signals change.

The aurora_example design included in the example_design directory with simplex Aurora 8B/10B cores shows a simple side channel that uses 3 or 4 I/O pins on the device.

Using Timers

For some systems a back channel is not possible. In these cases, serial channels can be initialized by driving the TX simplex initialization with a set of timers. The timers must be designed carefully to meet the needs of the system since the average time for initialization depends on many channel specific conditions such as clock rate, channel latency, skew between lanes, and noise.

Some of the initialization logic in the Aurora 8B/10B module uses watchdog timers to prevent deadlock. These watchdog timers are used on the RX side of the channel, and can interfere with the proper operation of TX initialization timers. If the RX simplex module goes from ALIGNED, BONDED or VERIFY, to RESET, make sure that it is not because the TX logic spend too much time in one of those states. If a particularly long timer is required to meet the needs of the system, the watchdog timers can be adjusted by editing the lane_init_sm module and the channel_init_sm module. For most cases, this should not be necessary and is not recommended.

Aurora 8B/10B channels normally reinitialize only in the case of failure. When there is no back channel available, event-triggered re-initialization is impossible for most errors since it is usually the RX side that detects a failure and the TX side that must handle it. The solution for this problem is to make timer-driven TX simplex modules reinitialize on a regular basis. If a catastrophic error occurs, the channel will be reset and running again once the next re-initialization period arrives. System designers should balance the average time required for re-initialization against the maximum time their system can tolerate an inoperative channel to determine the optimum re-initialization period for their systems.

Reset and Power Down

Reset

The reset signals on the control and status interface are used to set the Aurora 8B/10B core to a known starting state. Resetting the core stops any channels that are currently operating; after reset, the core attempts to initialize a new channel.

On full-duplex modules, the RESET signal resets both the TX and RX sides of the channel when asserted on the positive edge of USER_CLK. On simplex modules, the resets for the TX and RX channels are separate. TX_SYSTEM_RESET resets TX channels; RX_SYSTEM_RESET resets RX channels. The TX_SYSTEM_RESET is separate from the TX_RESET and RX_RESET signals used on the simplex sideband interface.

Power Down

This is an active-High signal. When POWER_DOWN is asserted, the GTP/GTX transceivers in the Aurora 8B/10B core are turned off, putting them into a non-operating low-power mode. When POWER_DOWN is deasserted, the core automatically resets. Be careful when asserting this signal on cores that use TX_OUT_CLK (see the Chapter 7, "Clock Interface and Clocking"). TX_OUT_CLK will stop when the GTP/GTX transceivers are powered down. See the *Virtex-5 FPGA RocketIO GTP Transceiver User Guide*, the *Virtex-5 FPGA RocketIO GTX Transceiver User Guide*, the *Virtex-6 FPGA GTX Transceivers User Guide*, and the *Spartan-6 FPGA GTP Transceivers User Guide* for the device you are using for details about powering down GTP/GTX transceivers.

Timing

Figure 6-6 shows the timing for the RESET and POWER_DOWN signals. In a quiet environment, t_{CU} is generally less than 800 clocks; in a noisy environment, t_{CU} can be much longer.

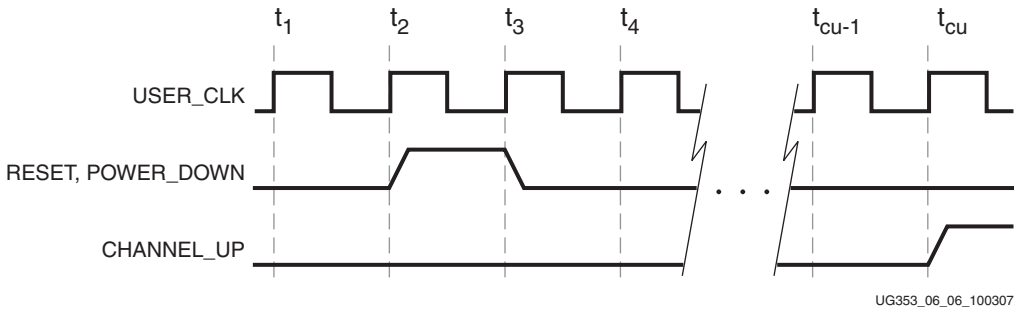

Figure 6-6: **Reset and Power Down Timing**

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Chapter 7

Clock Interface and Clocking

Introduction

Good clocking is critical for the correct operation of the Aurora 8B/10B core. The core requires a high-quality, low-jitter reference clock to drive the high-speed TX clock and clock recovery circuits in the GTP/GTX transceiver. It also requires at least one frequency locked parallel clock for synchronous operation with the user application.

The Virtex®-5 FPGA GTP/GTX architecture has a pair of transceivers in each GTP/GTX_DUAL tile and the Virtex-6 FPGA has four GTX transceivers in a Quad. The Spartan®-6 FPGA GTP architecture has a pair of transceivers in each GTPA1_DUAL tile. The Virtex-5 FPGA has a shared PMA PLL architecture inside the GTP/GTX_DUAL tile, and each reference clock sources both channels. The Virtex-6 FPGA GTX Quad has individual PLLs for both TX and RX portion of the transceivers. The Spartan-6 FPGA has individual PLLs for each transceiver in a GTPA1_DUAL tile.The reference clock is used to produce the PLL clock, which is divided to make individual TX and RX serial clocks and parallel clocks in each GTP/GTX transceiver.

Each Aurora 8B/10B core is generated in the example_design directory that includes a design called aurora_example. This design by instantiating the generated Aurora 8B/10B core, demonstrates a working clock configuration of the core. First-time users should examine the aurora example design and use it as a template when connecting the clock interface.

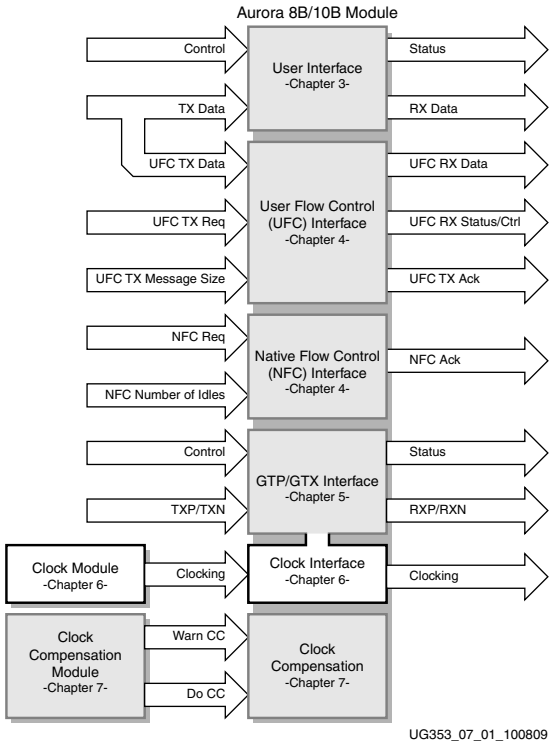

Figure 7-1: **Top-Level Clocking**

Clock Interface Ports for GTP/GTX Transceiver Cores

Table 7-1 describes the Aurora 8B/10B core clock ports.

Table 7-1: **Clock Ports for a GTP/GTX Aurora 8B/10B Core**

Clock Ports	Direction	Description
PLL_NOT_LOCKED	Input	If a PLL is used to generate clocks for the Aurora 8B/10B core, the PLL_NOT_LOCKED signal should be connected to the inverse of the PLL's LOCKED signal. The clock module provided with the Aurora 8B/10B core uses the PLL for clock division. The PLL_NOT_LOCKED signal from the clock module should be connected to the PLL_NOT_LOCKED signal on the Aurora 8B/10B core. If PLL is not used to generate clock signals for the Aurora 8B/10B core, tie PLL_NOT_LOCKED to ground.
USER_CLK	Input	Parallel clock shared by the Aurora 8B/10B core and the user application. In Aurora 8B/10B cores, USER_CLK and SYNC_CLK are the output of a PLL whose input is derived from TX_OUT_CLK. These clock generations are available in <component_name>_clock_module file. The Spartan-6 FPGA uses the GTPCLKOUT port to derive USER_CLK and SYNC_CLK outputs. USER_CLK goes as TXUSRCLK2 input to the transceiver tile. Refer to the respective transceiver user guide for more information.
SYNC_CLK	Input	Parallel clock used by the internal synchronization logic of the GTP/GTX transceivers in the Aurora 8B/10B core. SYNC_CLK goes as TXUSRCLK input to the transceiver tile. Refer to the respective transceiver user guide for more information.
GT_REFCLK	Input	GT_REFCLK (CLKP/CLKN) is a dedicated external clock generated from an oscillator. This clock is fed through IBUFDS. To minimize the number of oscillators, the GTP/GTX transceiver architecture has a NORTH/SOUTH clock routing matrix using CLKP/CLKN. A GTP/GTX_DUAL tile shares its clock with its neighbors using dedicated clock routing resources.

Reference Clocks for Virtex-5 FPGA GTP/GTX Transceiver Designs

In Virtex-5 FPGA transceiver designs, the reference clock is GTPD/GTXD, which is a differential input clock for each GTP/GTX_DUAL. In Virtex-6 FPGA transceiver designs, the reference clock is GTXQ, which is a differential input clock for each GTX QUAD. The reference clock for GTP/GTX_DUAL is provided through the CLKIN port.

The two possible use models for distributing a reference clock to drive the CLKIN port are:

- "Clocking from an External Source"
- "Clocking from a Neighboring GTP/GTX_DUAL Tile"

Clocking from an External Source

Each GTP/GTX_DUAL tile has a pair of dedicated pins that can be connected to an external clock source. To use these pins, a IBUFDS primitive is instantiated. In the user constraints file (`.ucf`), the IBUFDS input pins are set to the dedicated clock pins for the tile. In the design, the output of the IBUFDS is connected to the CLKIN port.

Each GTP/GTX_DUAL takes differential GTPD/GTXD clock inputs, which are directly bonded to the FPGA pins. GTPD/GTXD clock should be used for high data rate applications (1 Gb/s or higher). For multilane Aurora 8B/10B designs, GTPD/GTXD clock of any GTP/GTX_DUAL can be used as the reference clock for the Aurora 8B/10B design. Using a low-jitter oscillator delivers a high-quality clock suitable for top-speed operation. Figure 7-2 shows a differential GTP/GTX_DUAL clock pin pair sourced by an external oscillator on the board.

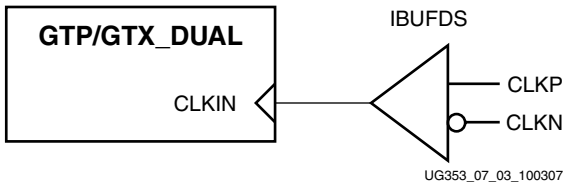

Figure 7-2: **Single GTP/GTX_DUAL Tile Clocked Externally**

Clocking from a Neighboring GTP/GTX_DUAL Tile

The external clock from one tile can be used to drive the CLKIN ports of neighboring tiles. The example in Figure 7-3, page 79 uses the clock from one GTP/GTX_DUAL tile to clock six neighboring tiles. A GTP/GTX_DUAL tile shares its clock with its neighbors using dedicated clock routing resources. GREFCLK cannot be shared using these resources.

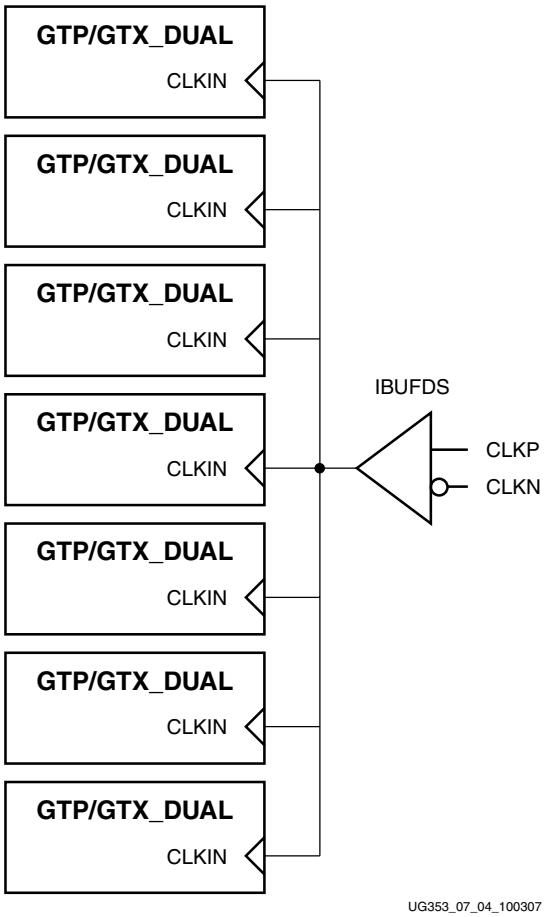

Figure 7-3: **Multiple Tiles with a Shared Reference Clock**

To keep the jitter of this configuration within the jitter margin for high-speed designs, the following rules must be observed:

1. The number of GTP/GTX_DUAL tiles above the sourcing GTP/GTX_DUAL tile must not exceed three.
2. The number of GTP/GTX_DUAL tiles below the sourcing GTP/GTX_DUAL tile must not exceed three.
3. The total number of GTP/GTX_DUAL tiles sourced by the external clock pin pair (CLKN/CLKP) must not exceed seven.

The maximum number of GTP/GTX transceivers that can be sourced by a single clock pin pair translates to a maximum of 14 pairs. Designs with more than 14 GTP/GTX transceivers require the use of multiple external clock pins to ensure that the rules for controlling jitter are followed. When multiple clock pins are used, it is recommended that a low-skew LVDS buffer be used externally to drive all the pins from the same low-jitter oscillator. This is critical when several GTP/GTX transceivers are combined to form a single channel.

Note: In the current release, the CORE Generator™ software GUI does not allow selecting an external clock from a tile in which no GTP/GTX transceivers are selected.

Clocking from a Neighboring GTX_QUAD Tile for Virtex-6 FPGA Designs

The Xilinx implementation tools will make necessary adjustments to the north-south routing shown in Figure 7-4, page 81 as well as pin swapping necessary to GTXE1 clock inputs in order to route clocks from one Quad to another when required.

The following rules must be observed when sharing a reference clock to ensure that jitter margins for high-speed designs are met:

1. The number of GTX Quads above the sourcing Quad must not exceed one.
2. The number of GTX Quads below the sourcing Quad must not exceed one.
3. The total number of GTX Quads sourced by an external clock pin pair (MGTREFCLKN/MGTREFCLKP) must not exceed three or 12 GTXE1 transceivers.

The maximum number of GTX transceivers that can be sourced by a single clock pin pair is 12. Designs with more than 12 transceivers require the use of multiple external clock pins to ensure that the rules for controlling jitter are followed. When multiple clock pins are used, an external buffer can be used to drive them from the same oscillator.

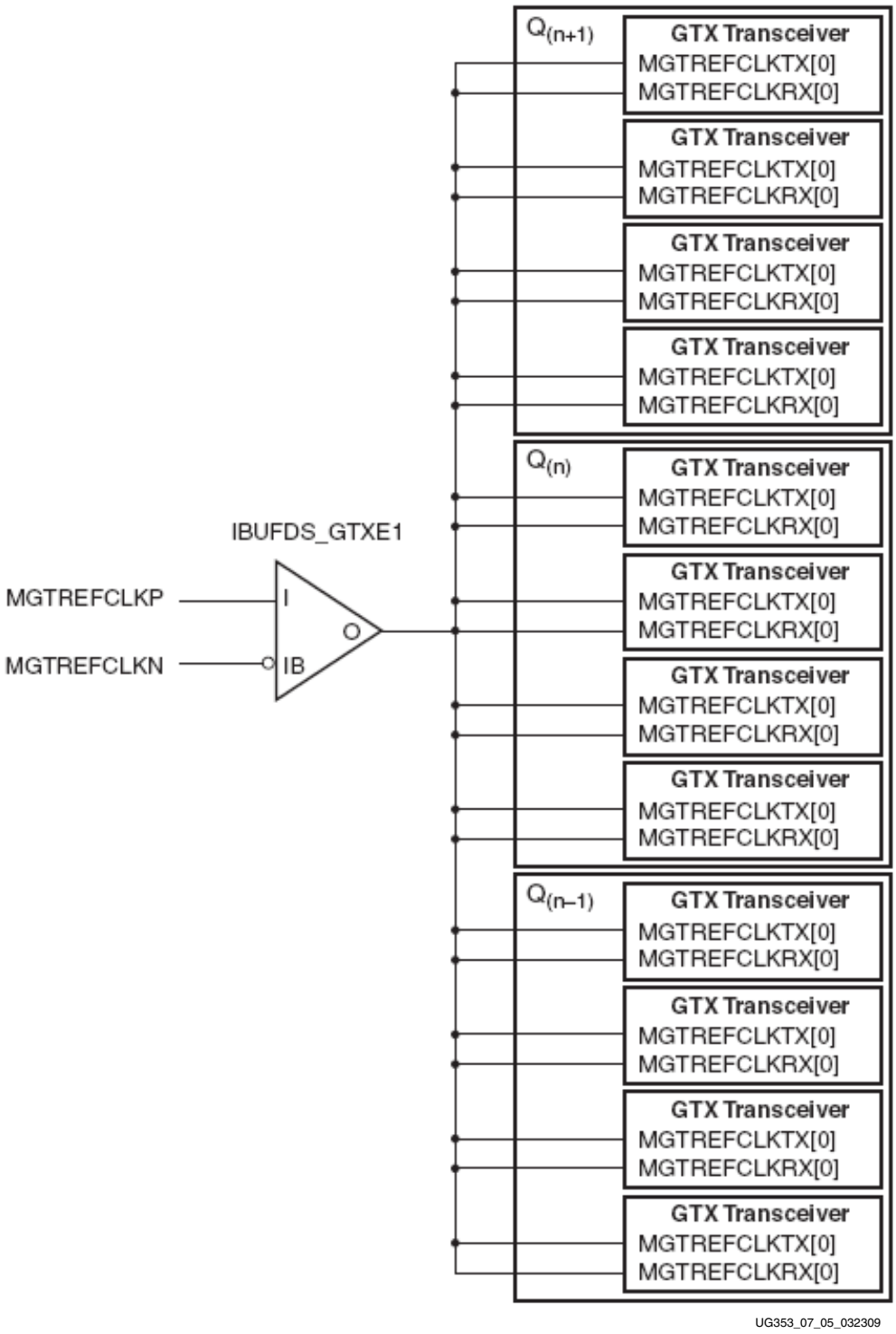

Figure 7-4: **North-South Routing Adjustments in Virtex-6 FPGAs**

Reference Clocks for Spartan-6 FPGA GTP Transceiver Designs

In Spartan-6 FPGA transceiver designs, the reference clock is GTPD, which is a differential input clock for each GTPA1_DUAL. The reference clock for GTPA1_DUAL is provided through the CLK00 and CLK01 ports. The two possible use models for distributing a reference clock to drive the CLK00 and CLK01 ports are:

- "Clocking from and External Source"
- "Clocking from a Neighboring GTPA1_DUAL Tile"

Clocking from and External Source

Each GTPA1_DUAL tile has a pair of dedicated pins that can be connected to an external clock source. To use these pins, a IBUFDS primitive is instantiated. In the user constraints file (`.ucf`), the IBUFDS input pins are set to the dedicated clock pins for the tile. In the design, the output of the IBUFDS is connected to the CLK00 and CLK01 ports. Each GTPA1_DUAL takes differential GTPD clock inputs, which are directly bonded to the FPGA pins. For multilane Aurora 8B/10B designs, GTPD clock of any GTPA1_DUAL can be used as the reference clock for the Aurora 8B/10B design. Using a low-jitter oscillator delivers a high-quality clock suitable for top-speed operation. Figure 7-5 shows a differential GTPA1_DUAL clock pin pair sourced by an external oscillator on the board. This clocking mechanism is used for Spartan-6 single lane and 2-lane designs.

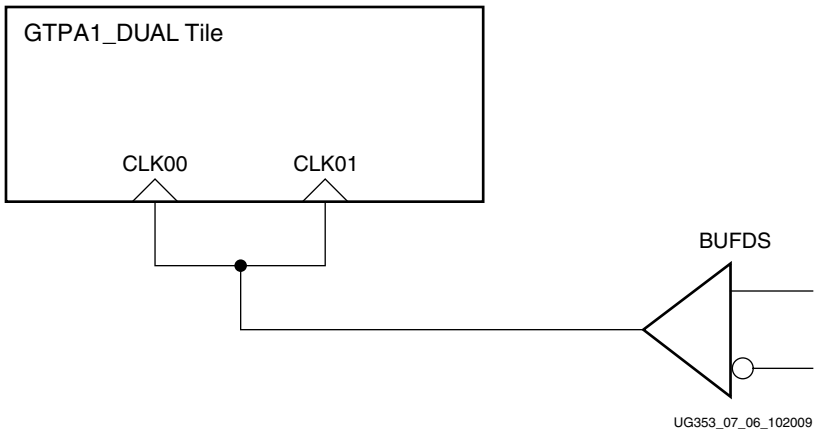

Figure 7-5: **Single GTPA1_DUAL Tile Clocked Externally**

Clocking from a Neighboring GTPA1_DUAL Tile

The external clock from one tile can be used to drive the CLK00 and CLK01 ports of neighboring tiles.

The example in Figure 7-6 uses the clock from one GTPA1_DUAL tile to clock neighboring tiles. A GTPA1_DUAL tile shares its clock with its neighbors using dedicated clock routing resources. This clocking mechanism is used for Spartan-6 FPGA 4-lane design.

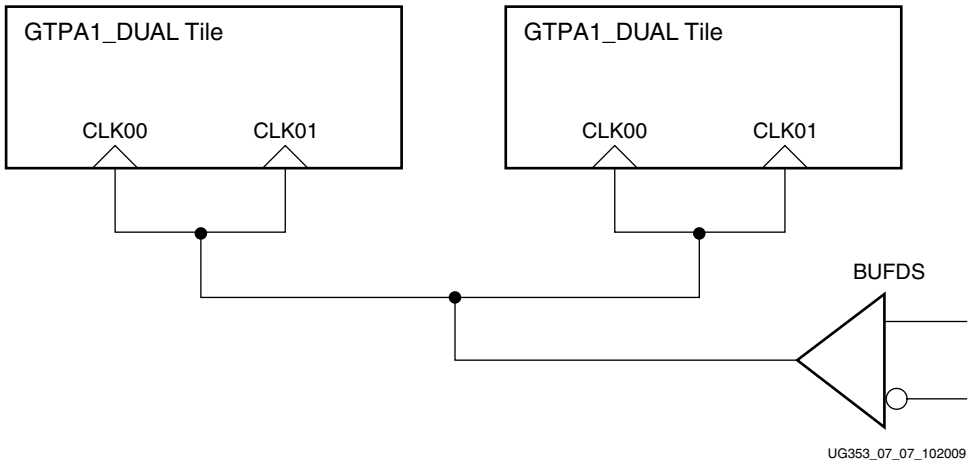

Figure 7-6: **GTPA1_DUAL Tiles with Shared Reference Clock**

Clock Rates for GTP/GTX Transceiver Designs

GTP/GTX transceivers support a wide range of serial rates. The attributes used to configure the GTP/GTX transceivers in the Aurora 8B/10B core for a specific line rate are kept in the transceiver_wrapper module for simulation. These attributes are set automatically by the CORE Generator software in response to the line rate and reference clock selections made in the Configuration GUI window for the core. Manual edits of the attributes are not recommended, but are possible using the recommendations in the *Virtex-5 FPGA RocketIO GTP Transceiver User Guide*, the *Virtex-5 FPGA RocketIO GTX Transceiver User Guide*, the *Virtex-6 FPGA GTX Transceivers User Guide*, and the *Spartan-6 FPGA GTP Transceivers User Guide*

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Chapter 8

Clock Compensation

Introduction

Clock compensation is a feature that allows up to ± 100 ppm difference in the reference clock frequencies used on each side of an Aurora 8B/10B channel. This feature is used in systems where a separate reference clock source is used for each device connected by the channel, and where the same USER_CLK is used for transmitting and receiving data.

The Aurora 8B/10B core's clock compensation interface enables full control over the core's clock compensation features. A standard clock compensation module is generated with the Aurora 8B/10B core to provide Aurora 8B/10B-compliant clock compensation for systems using separate reference clock sources; users with special clock compensation requirements can drive the interface with custom logic. If the same reference clock source is used for both sides of the channel, the interface can be tied to ground to disable clock compensation.

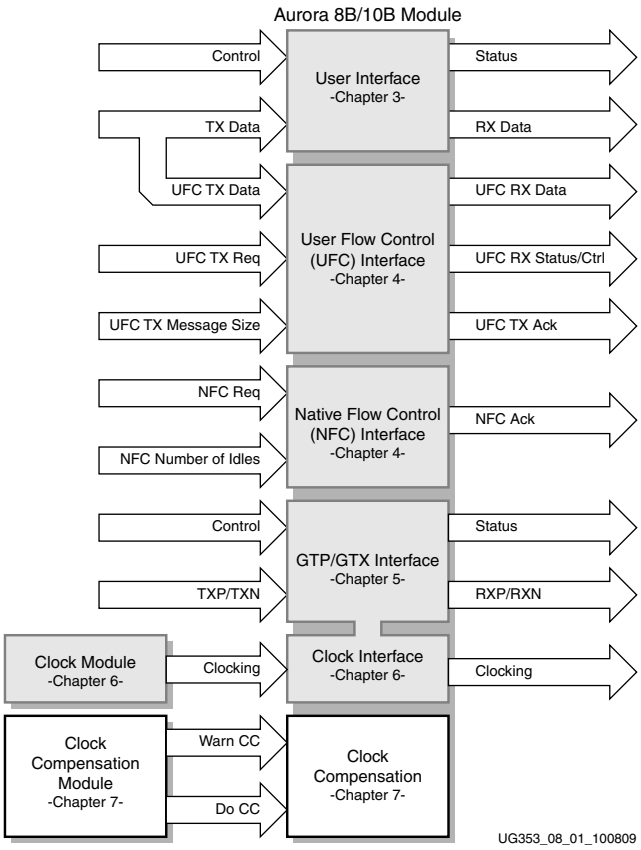

Figure 8-1: **Top-Level Clock Compensation**

Clock Compensation Interface

All Aurora 8B/10B cores include a clock compensation interface for controlling the transmission of clock compensation sequences. Table 8-1 describes the function of the clock compensation interface ports.

Table 8-1: **Clock Compensation I/O Ports**

Name	Direction	Description
DO_CC	Input	The Aurora 8B/10B core sends CC sequences on all lanes on every clock cycle when this signal is asserted. Connects to the DO_CC output on the CC module.
WARN_CC	Input	The Aurora 8B/10B core will not acknowledge UFC requests while this signal is asserted. It is used to prevent UFC messages from starting too close to CC events. Connects to the WARN_CC output on the CC module.

Figure 8-2 and Figure 8-3 are waveform diagrams showing how the DO_CC signal works.

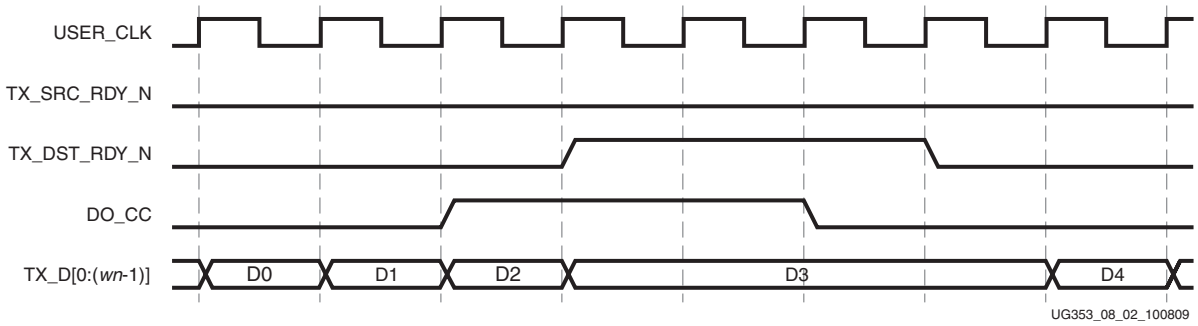

Figure 8-2: **Streaming Data with Clock Compensation Inserted**

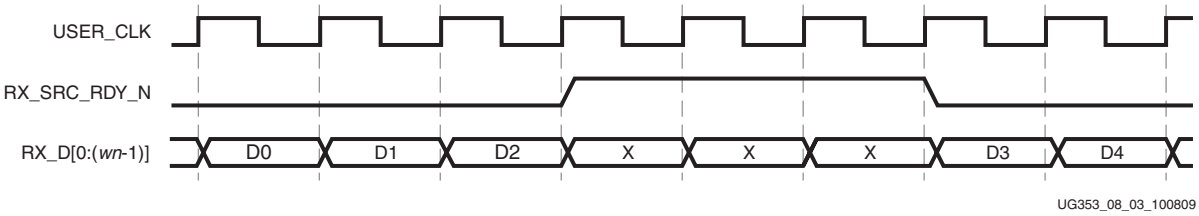

Figure 8-3: **Data Reception Interrupted by Clock Compensation**

The Aurora 8B/10B protocol specifies a clock compensation mechanism that allows up to ± 100 ppm difference between reference clocks on each side of an Aurora 8B/10B channel. To perform Aurora 8B/10B-compliant clock compensation, DO_CC must be asserted for several cycles every clock compensation period. The duration of the DO_CC assertion and the length of time between assertions is determined based on the width of the GTP/GTX transceiver data interface. While DO_CC is asserted, TX_DST_RDY_N on the user interface for modules with TX while the channel is being used to transmit clock compensation sequences. Table 8-2 shows the required durations and periods for 2-byte and 4-byte wide lanes.

Table 8-2: **Clock Compensation Cycles**

Lane Width	USER_CLK Cycles Between DO_CC	DO_CC Duration (USER_CLK cycles)
2	5000	6
4	3000	3

WARN_CC is for cores with user flow control (UFC) and/or native flow control (NFC). Driving this signal before DO_CC is asserted prevents the UFC interface from acknowledging and sending UFC messages too close to a clock correction sequence. This precaution is necessary because data corruption occurs when CC sequences and UFC messages overlap. The number of lookahead cycles required to prevent a 16-byte UFC message from colliding with a clock compensation sequence depends on the number of lanes in the channel and the width of each lane. Table 8-3 shows the number of lookahead cycles required for each combination of lane width, channel width, and maximum UFC message size.

Table 8-3: **Lookahead Cycles**

Data Interface Width	Max UFC Size	WARN_CC Lookahead
2	2	3
2	4	4
2	6	5
2	8	6
2	10	7
2	12	8
2	14	9
2	16	10
4	2-4	3
4	6-8	4
4	10-12	5
4	14-16	6
6	2-6	3
6	8-12	4
6	14-16	5
8	2-8	3
8	10-16	4
10	2-10	3
10	12-16	4
12	2-12	3
12	14-16	4
14	2-14	3
14	16	4
≥16	2-16	3

Native flow control message request will not be acknowledged during assertion of WARN_CC and DO_CC signals. This helps to prevent the collision of NFC message and clock compensation sequence.

To make Aurora 8B/10B compliance easy, a standard clock compensation module is generated along with each Aurora 8B/10B core from the CORE Generator™ software in the cc_manager subdirectory. It automatically generates pulses to create Aurora 8B/10B compliant clock compensation sequences on the DO_CC port and sufficiently early pulses on the WARN_CC port to prevent UFC collisions with maximum-sized UFC messages. This module always be connected to the clock compensation port on the Aurora 8B/10B module, except in special cases. Table 8-4 shows the port description for the standard CC module.

Table 8-4: **Standard CC I/O Port**

Name	Direction	Description
WARN_CC	Output	Connect this port to the WARN_CC input of the Aurora 8B/10B core when using UFC.
DO_CC	Output	Connect this port to the DO_CC input of the Aurora 8B/10B core.
CHANNEL_UP	Input	Connect this port to the CHANNEL_UP output of a full-duplex core, or to the TX_CHANNEL_UP output of a simplex TX or a simplex Both port.

Clock compensation is not needed when both sides of the Aurora 8B/10B channel are being driven by the same clock (see Figure 8-3, page 87) because the reference clock frequencies on both sides of the module are locked. In this case, WARN_CC and DO_CC should both be tied to ground. Additionally, the CLK_CORRECT_USE attribute can be set to false in the transceiver interface module for the core. This can result in lower latencies for single lane modules.

Other special cases when the standard clock compensation module is not appropriate are possible. The DO_CC port can be used to send clock compensation sequences at any time, for any duration to meet the needs of specific channels. The most common use of this feature is scheduling clock compensation events to occur outside of frames, or at specific times during a stream to avoid interrupting data flow. In general, customizing the clock compensation logic is not recommended, and when it is attempted, it should be performed with careful analysis, testing, and consideration of the following guidelines:

- Clock compensation sequences should last at least two cycles to ensure they are recognized by all receivers
- Be sure the duration and period selected is sufficient to correct for the maximum difference between the frequencies of the clocks that will be used
- Do not perform multiple clock correction sequences within eight cycles of one another
- Replacing long sequences of idles (>12 cycles) with CC sequences will result in increased EMI
- DO_CC will have no effect until after CHANNEL_UP; DO_CC should be asserted immediately after CHANNEL_UP since no clock compensation can occur during initialization

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Chapter 9

Quick Start Example Design

This chapter introduces the example design that is included with the LogiCORE™ IP Aurora 8B/10B core. The quick start instructions are a step-by-step procedure for generating an Aurora 8B/10B core, implementing the core in hardware using the accompanying example design, and simulating the core with the provided demonstration test bench (demo_tb). For detailed information about the example design provided with the Aurora 8B/10B core, see Chapter 11, "Project Directory Structure."

Overview

The quick start example consists of the following components:

- An instance of the Aurora 8B/10B core generated using the default parameters
  - Full-duplex with a single GTP/GTX transceiver
  - LocalLink interface
- A demonstration test bench to simulate two instances of the example design

The Aurora 8B/10B example design has been tested with XST for synthesis and Mentor Graphics ModelSim for simulation.

Generating the Core

To generate an Aurora 8B/10B core with default values using the CORE Generator™ tool:

1. Start the CORE Generator software from a required directory.

   For help starting and using the CORE Generator software, see *CORE Generator Help* in the ISE® software documentation.
2. Choose **File** → **New Project**.
3. Type a project name.
4. To set project options:
   - From the Part tab, select a silicon family, part, speed grade, and package that supports the Aurora 8B/10B core, for example, Virtex®-5 FPGAs.

     Note: If an unsupported silicon family is selected, the Aurora 8B/10B appears light grey in the taxonomy tree and cannot be customized. Only devices containing GTP/GTX transceivers are supported by the core. For a list of supported architectures, see the *LogiCORE IP Aurora 8B/10B v5.2 Data Sheet*.
   - No further project options need to be set.
   - Optionally, on the Generation tab, set the Design Entry pull-down to **Verilog**.

5. After creating the project, locate the Aurora 8B/10B core v5.2 in the taxonomy tree under:

   `/Communication_&_Networking/Serial_Interfaces`

6. Double-click the core.
7. In the Component Name field (Figure 9-1), enter a name for the core instance. This example uses the name **aurora_8b10b_v5_2**.

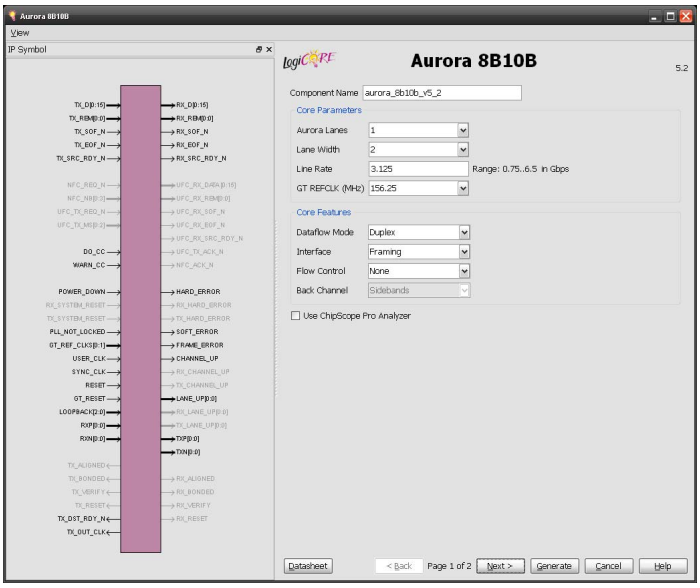

Figure 9-1: **CORE Generator Aurora 8B/10B Customization Screen**

8. Click **Generate**.

The core and its supporting files, including the example design, are generated in the project directory. For detailed information about the example design files and directories, see Chapter 11, "Project Directory Structure."

Simulating the Example Design

The Aurora 8B/10B core provides a quick way to simulate and observe the behavior of the core using the provided example design. Prior to simulating the core, the functional (gate-level) simulation models must be generated. You must compile all source files in the following directories to a single library as shown in Table 9-1. Refer to the *Synthesis and Verification Design Guide* for ISE 12.2 software for instructions on how to compile ISE software simulation libraries.

Table 9-1: **Required Simulation Libraries**

HDL	Library	Source Directories
Verilog	UNISIMS_VER	<*Xilinx dir*>/verilog/src/unisims <*Xilinx dir*>/ secureip/<*SIMULATOR*>
VHDL	UNISIM	<*Xilinx dir*>/vhdl/src/unisims <*Xilinx dir*>/ secureip/<*SIMULATOR*>

Notes:

1. SIMULATOR can be Modelsim.

The Aurora 8B/10B core provides a command line script to simulate the example design. To run a VHDL or Verilog ModelSim simulation of the Aurora 8B/10B core, use the following instructions:

1. Launch the ModelSim simulator and set the current directory to:

   ***<project directory>*/aurora_8b10b_v5_2/simulation/functional**

2. Set the MTI_LIBS variable:

   `modelsim>` **setenv MTI_LIBS *<path to compiled libraries>***

3. Launch the simulation script:

   `modelsim>` **do simulate_mti.do**

The ModelSim script compiles the example design and test bench, and adds the relevant signals to the wave window. After the design is compiled and the wave window is displayed, run the simulation to see the Aurora 8B/10B core power up, followed by Aurora 8B/10B channel initialization and data transfer. Data transfer begins after the CHANNEL_UP signal goes High.

Implementing the Example Design

After the core is generated, the design can be processed by the Xilinx implementation tools. The generated output files include several scripts to assist the user in running the Xilinx software.

From the command prompt, navigate to the project directory and type the following:

For Windows

```
ms-dos> cd aurora_8b10b_v5_2\implement
ms-dos> .\implement.bat
```

For Linux

```
% cd aurora_8b10b_v5_2/implement
%./implement.sh
```

These commands execute a script that synthesizes, translates, maps, place-and-routes the example design and produces a bitstream file. The resulting files are placed in the results directory created within the implement directory.

Using ChipScope Pro Cores with the Aurora 8B/10B Core

Description

The ChipScope™ Pro ICON and VIO cores aid in debugging and validating the design in boards and are provided with the Aurora 8B/10B core. Select the **Use ChipScope Pro Analyzer** checkbox from the core GUI to include it as a part of the example design.

Chapter 10

Example Design Overview

Introduction

Each Aurora 8B/10 B core includes an example design (aurora_example) that uses the core in a simple data transfer system. For more details about the example_design directory, see Chapter 11, "Project Directory Structure."

The example design consists of two main components:

- Frame generator ("FRAME_GEN," page 97) connected to the TX user interface
- Frame checker ("FRAME_CHECK," page 102) connected to the RX user interface

Figure 10-1 illustrates the block diagram of the example design for a full-duplex core. Table 10-1, page 96 describes the ports of the example design.

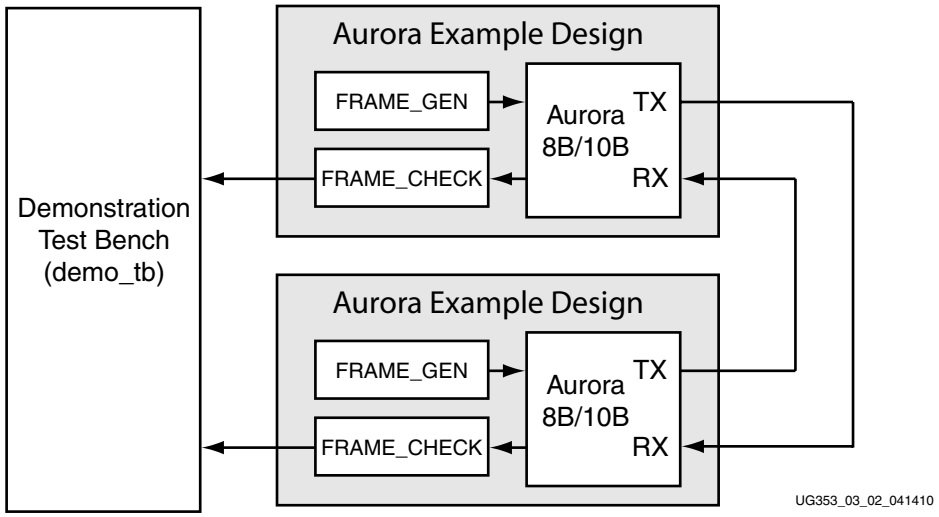

Figure 10-1: **Example Design**

The example designs uses all the interfaces of the core. Simplex cores without a TX or RX interface will have no FRAME_GEN or FRAME_CHECK block, respectively. The frame generator produces a constant stream of data for cores with a streaming interface.

Using the scripts provided in the implement subdirectory, the example design can be used to quickly get an Aurora 8B/10B design up and running on a board, or perform a quick simulation of the module. The design can also be used as a reference for the connecting the trickier interfaces on the Aurora 8B/10B core, such as the clocking interface.

When using the example design on a board, be sure to edit the <component_name>_example_design.ucf file in the example_design subdirectory to supply the correct pins and clock constraints.

Table 10-1: **Example Design I/O Ports**

Port	Direction	Description
RXN[0:*m*-1]	Input	Negative differential serial data input pin.
RXP[0:*m*-1]	Input	Positive differential serial data input pin.
TXN[0:*m*-1]	Output	Negative differential serial data output pin.
TXP[0:*m*-1]	Output	Positive differential serial data output pin.
ERR_COUNT[0:7]	Output	Count of the number of data words received by the frame checker that did not match the expected value.
RESET	Input	Reset signal for the example design. The reset is debounced using a USER_CLK signal generated from the reference clock input.
<reference clock(s)>	Input	The reference clocks for the Aurora 8B/10B core are brought to the top level of the example design. See Chapter 7, "Clock Interface and Clocking" for details about the reference clocks.
<core error signals>	Output	The error signals from the Aurora 8B/10B core's Status and Control interface are brought to the top level of the example design and registered. See Chapter 6, "Status, Control, and the GTP/GTX Block Interface" for details.
<core channel up signals>	Output	The channel up status signals for the core are brought to the top level of the example design and registered. Full-duplex cores will have a single channel up signal; simplex cores will have one for each channel direction supported. See Chapter 6, "Status, Control, and the GTP/GTX Block Interface" for details.
<core lane up signals>	Output	The lane up status signals for the core are brought to the top level of the example design and registered. Cores have a lane up signal for each GTP/GTX transceiver they use. Simplex cores have a separate lane up signal per GTP/GTX transceiver they use for each channel direction supported. See Chapter 6, "Status, Control, and the GTP/GTX Block Interface" for details.
<simplex initialization signals>	Input/ Output	If the core is a simplex core, its sideband initialization ports will be registered and brought to the top level of the example design. See Chapter 6, "Status, Control, and the GTP/GTX Block Interface" for details.

FRAME_GEN

Framing TX Data Interface

Figure 10-2 shows the FRAME_GEN framing user interface of the Aurora 8B/10B core, with LocalLink compliant ports for TX data.

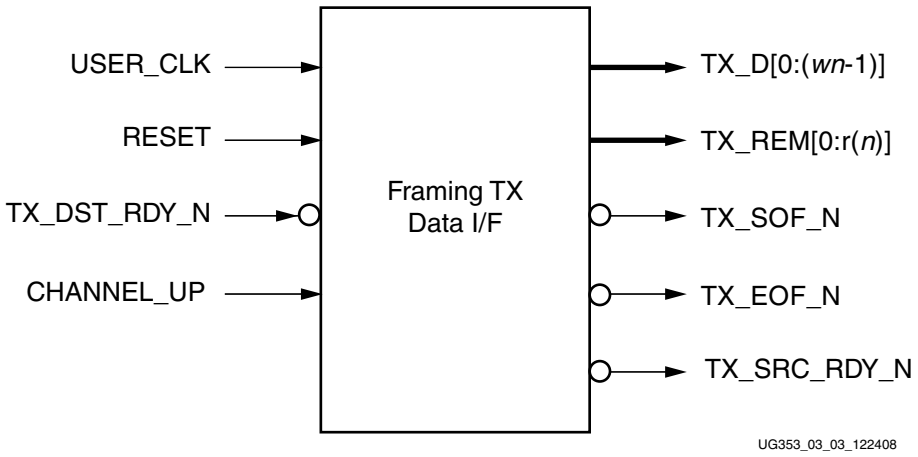

Figure 10-2: **Aurora 8B/10B Core Framing TX Data Interface (FRAME_GEN)**

Table 10-2 lists the FRAME_GEN framing TX data ports and their descriptions.

Table 10-2: **FRAME_GEN Framing User I/O Ports (TX)**

Name	Direction	Description
TX_D[0:(*wn*-1)]	Output	Outgoing data. Width is (*wn* -1) where *w* represents GTP/GTX transceiver internal data path width. • *w* takes values 16 or 32 depending on whether width is 2-byte or 4-byte • *n* is the number of lanes
TX_REM[0:r(*n*)]	Output	Specifies the number of valid bytes in the last data beat; valid only while TX_EOF_N is asserted. REM bus widths are given by [0:r(*n*)], where r(*n*) = ceiling {$\log_2 n$}-1
TX_SRC_RDY_N	Output	Asserted (Low) when LocalLink signals and/or data from the source are valid. Deasserted (High) when LocalLink control signals and/or data from the source should be ignored (active-Low).
TX_SOF_N	Output	Signals the start of the outgoing channel frame (active-Low).
TX_EOF_N	Output	Signals the end of the frame (active-Low).

Table 10-2: **FRAME_GEN Framing User I/O Ports (TX) *(Cont'd)***

Name	Direction	Description
TX_DST_RDY_N	Input	Asserted (Low) during clock edges when signals from the source will be accepted (if TX_SRC_RDY_N is also asserted). Deasserted (High) on clock edges when signals from the source will be ignored.
CHANNEL_UP	Input	Asserted when Aurora 8B/10B channel initialization is complete and channel is ready to transmit data.
USER_CLK	Input	Parallel clock shared by the Aurora 8B/10B core and the user application.
RESET	Input	Resets the Aurora 8B/10B core (active-High).

To transmit the user data, the FRAME_GEN user data state machine manipulates control signals to do the following:

- After the Aurora 8B/10B interface is out of RESET and reaches the CHANNEL_UP state, pseudo-random data is generated using user data linear feedback shift register (LFSR) and connected to TX_D bus.
- Generate the TX_SOF_N and TX_EOF_N for the current frame based on two counters. An 8-bit counter determines the size of the frame and another 8-bit counter keeps track of the number of user data bytes that has been sent. Frame size counter is initialized to zero and incremented by one for every frame.
- TX_REM bus is connected to lower bits of user data LFSR.
- TX_SRC_RDY_N is asserted according to LocalLink framing protocol specification.
- User data state machine state transitions are controlled by TX_DST_RDY_N provided by Aurora 8B/10B's LocalLink interface.
- Various kinds of frame traffic are generated including single cycle frame where in TX_SOF_N, TX_EOF_N are asserted in same cycle.

Streaming TX Data Interface

Figure 10-3 shows the FRAME_GEN streaming user interface of the Aurora 8B/10B core ports for TX data.

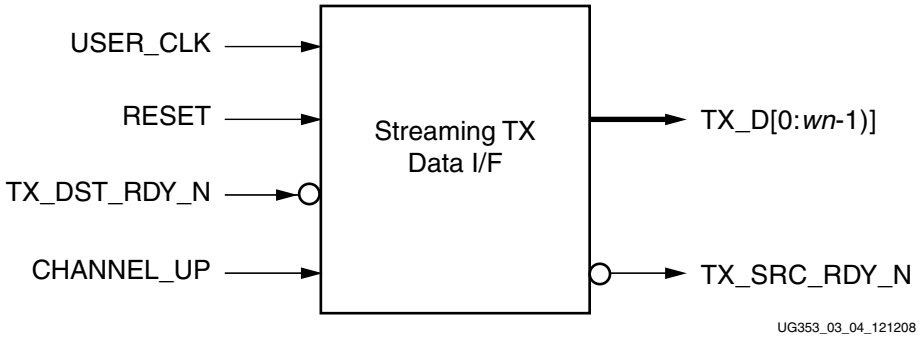

Figure 10-3: **Aurora 8B/10B Core Streaming TX Data Interface (FRAME_GEN)**

Table 10-3 lists the FRAME_GEN streaming TX data ports and their descriptions.

Table 10-3: **FRAME_GEN Streaming User I/O Ports (TX)**

Name	Direction	Description
TX_D[0:(*wn*-1)]	Output	Outgoing data. Width is *wn* where *w* represents GTP/GTX transceiver data path width. • *w* takes 16 or 32 depending on whether width is 2-byte or 4-byte. • *n* is the number of lanes
TX_SRC_RDY_N	Output	Asserted (Low) when LocalLink signals from the source are valid. Deasserted (High) when LocalLink control signals and/or data from the source should be ignored (active-Low).
TX_DST_RDY_N	Input	Asserted (Low) during clock edges when signals from the source will be accepted (if TX_SRC_RDY_N is also asserted). Deasserted (High) on clock edges when signals from the source will be ignored.
CHANNEL_UP	Input	Asserted when Aurora 8B/10B channel initialization is complete and channel is ready to transmit data.
USER_CLK	Input	Parallel clock shared by the Aurora 8B/10B core and the user application.
RESET	Input	Resets the Aurora 8B/10B core (active-High).

After the Aurora 8B/10B interface is out of RESET and reaches the CHANNEL_UP state, pseudo-random data is generated using LFSR and connected to TX_D bus. LFSR generates new data for every assertion of TX_DST_RDY_N. TX_SRC_RDY_N is always asserted.

Framing UFC TX Interface

Figure 10-4 shows the FRAME_GEN framing UFC TX interface of the Aurora 8B/10B core for UFC TX data.

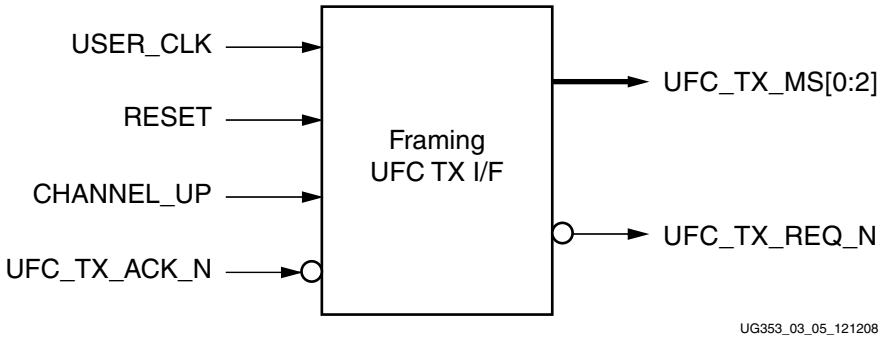

Figure 10-4: **Aurora 8B/10B Core Framing UFC TX Interface (FRAME_GEN)**

Table 10-4 lists the FRAME_GEN framing UFC TX data ports and their descriptions

Table 10-4: **FRAME_GEN Framing UFC User I/O Ports (TX)**

Name	Direction	Description
UFC_TX_REQ_N	Output	Asserted to request a UFC message to be sent to the channel partner (active- Low). Must be held until UFC_TX_ACK_N is asserted.
UFC_TX_MS[0:2]	Output	Specifies the size of the UFC message that will be sent. The size encoding is a value between 0 and 7.
CHANNEL_UP	Input	Asserted when Aurora 8B/10B channel initialization is complete and channel is ready to transmit data.
USER_CLK	Input	Parallel clock shared by the Aurora 8B/10B core and the user application.
RESET	Input	Resets the Aurora 8B/10B core (active-High).
UFC_TX_ACK_N	Output	Asserted when UFC message transmission starts.

To transmit the UFC data, the FRAME_GEN UFC state machine manipulates control signals to do the following:

- Asserts UFC_TX_REQ_N after CHANNEL_UP indication from the Aurora 8B/10B TX interface.
- UFC_TX_MS[0:2] is also transmitted along with UFC_TX_REQ_N. UFC_TX_MS is sent as zero initially for the first UFC frame and is incremented by one for the next UFC frame until it reaches 7.
- UFC data is transmitted immediately from UFC LFSR after receiving UFC_TX_ACK_N from the Aurora 8B/10B TX interface.
- UFC frame transmission frequency is controlled by UFC_IFG parameter.

Framing NFC TX Interface

Figure 10-5 shows the FRAME_GEN framing NFC TX interface of the Aurora 8B/10B core, with LocalLink compliant ports for NFC TX data.

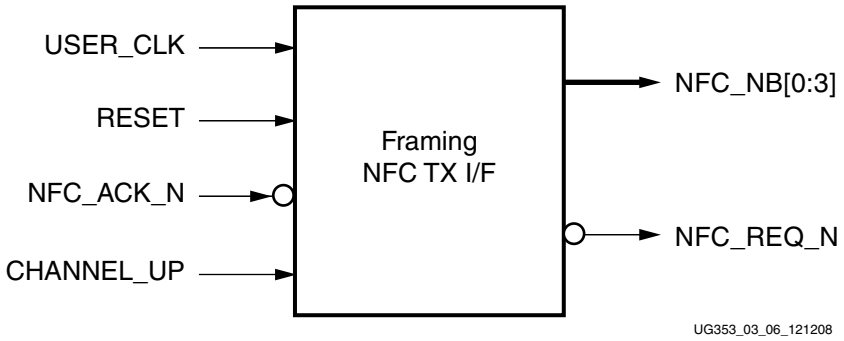

Figure 10-5: **Aurora 8B/10B Core Framing NFC TX Interface (FRAME_GEN)**

Table 10-5 lists the FRAME_GEN framing NFC TX data ports and their descriptions.

Table 10-5: **FRAME_GEN Framing NFC User I/O Ports (TX)**

Name	Direction	Description
NFC_REQ_N	Output	Asserted to request an NFC message to be sent to the channel partner (active-Low). Must be held until NFC_ACK_N is asserted.
NFC_NB[0:3]	Output	Indicates the number of PAUSE idles the channel partner must send when it receives the NFC message. Must be held until NFC_ACK_N is asserted.
NFC_ACK_N	Input	Asserted when an Aurora 8B/10B core accepts an NFC request (active-Low).
CHANNEL_UP	Input	Asserted when Aurora 8B/10B channel initialization is complete and channel is ready to send data.
USER_CLK	Input	Parallel clock shared by the Aurora 8B/10B core and the user application.
RESET	Input	Resets the Aurora 8B/10B core (active-High).

To transmit the NFC data, the FRAME_GEN NFC state machine manipulates control signals to do the following:

- NFC state machine waits until TX user data transmission and start transmitting first NFC frame after predefined period of time.
- NFC_NB[0:3] value is transmitted along with NFC_REQ_N.
- After a predefined period of time, the NFC state machine enters into NFC XOFF mode.
- NFC state machine enters into NFC XON mode after a predefined period of time.
- NFC frame transmission frequency is controlled by NFC_IFG parameter.
- NFC state transitions are governed by NFC_ACK_N.

FRAME_CHECK

Framing RX Data Interface

Figure 10-6 shows the FRAME_CHECK framing user interface of the Aurora 8B/10B core, with LocalLink compliant ports for RX data.

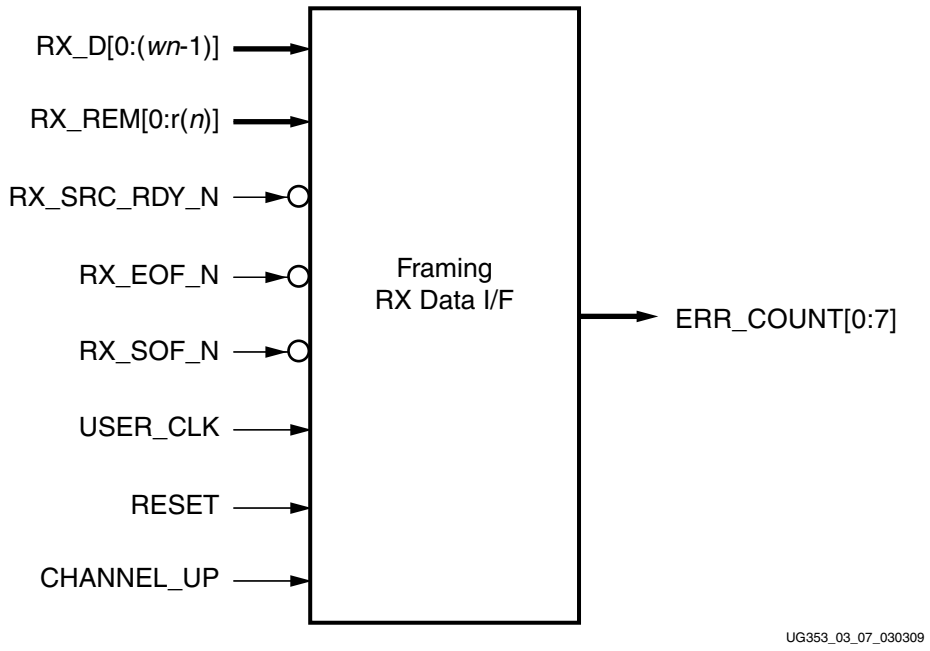

Figure 10-6: **Aurora 8B/10B Core Framing RX Data Interface (FRAME_CHECK)**

Table 10-6 lists the FRAME_CHECK framing RX data ports and their descriptions.

Table 10-6: **FRAME_CHECK Framing User I/O Ports (RX)**

Name	Direction	Description
RX_D[0:(*wn*-1)]	Input	Incoming data from channel partner (Ascending bit order).
RX_REM[0:r(*n*)]	Input	Specifies the number of valid bytes in the last data beat; valid only when RX_EOF_N is asserted. REM bus widths are given by [0:r(*n*)], where r(*n*) = ceiling [{$\log_2$ (n)}-1].
RX_SRC_RDY_N	Input	Asserted (Low) when data and control signals from an Aurora 8B/10B core are valid. Deasserted (High) when data and/or control signals from an Aurora 8B/10B core should be ignored (active-Low).
RX_SOF_N	Input	Signals the start of the incoming frame (active-Low, asserted for a single USER_CLK cycle).

Table 10-6: **FRAME_CHECK Framing User I/O Ports (RX) *(Cont'd)***

Name	Direction	Description
RX_EOF_N	Input	Signals the end of the incoming frame (active-Low, asserted for a single USER_CLK cycle).
ERR_COUNT[0:7]	Output	Count of the number of RX data words received by the frame checker that did not match with the expected value
CHANNEL_UP	Input	Asserted when Aurora 8B/10B channel initialization is complete and channel is ready to transmit data.
USER_CLK	Input	Parallel clock shared by the Aurora 8B/10B core and the user application.
RESET	Input	Resets the Aurora 8B/10B core (active-High).

Expected frame RX data is computed by LFSR. Received user data is validated by checking against the following LocalLink protocol rules:

1. Start the frame when RX_SOF_N is asserted
2. RX_REM bus is valid during RX_EOF_N assertion
3. RX_SRC_RDY_N should be asserted for all of the valid user data

The incoming RX data through RX_D port is registered and compared with the calculated RX data internal to FRAME_CHECK. If the incoming RX data does not match the expected RX data, an 8-bit counter is incremented. This error counter is indicated to the user through ERR_COUNT port. This counter stops counting and count value is latched when it reaches to 255.

Streaming RX Data Interface

Figure 10-7 shows the FRAME_CHECK streaming user interface of the Aurora 8B/10B core, ports for RX data.

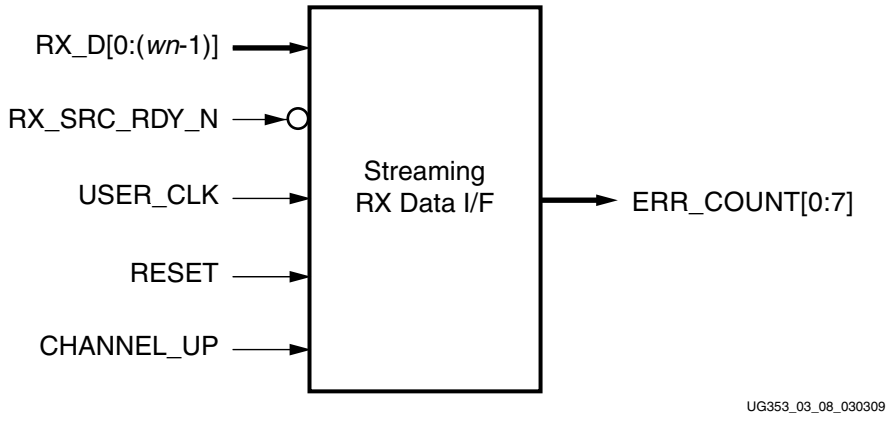

Figure 10-7: **Aurora 8B/10B Core Streaming RX Data Interface (FRAME_CHECK)**

Table 10-7 lists the FRAME_CHECK streaming RX data ports and their descriptions.

Table 10-7: **FRAME_CHECK Streaming User I/O Ports (RX)**

Name	Direction	Description
RX_D[0:(*wn*-1)]	Input	Incoming data from channel partner (Ascending bit order).
RX_SRC_RDY_N	Input	Asserted (Low) when data and control signals from an Aurora 8B/10B core are valid. Deasserted (High) when data and/or control signals from an Aurora 8B/10B core should be ignored (active-Low)
ERR_COUNT[0:7]	Output	Count of the number of RX data words received by the frame checker that did not match with the expected value
CHANNEL_UP	Input	Asserted when Aurora 8B/10B channel initialization is complete and channel is ready to send data.
USER_CLK	Input	Parallel clock shared by the Aurora 8B/10B core and the user application.
RESET	Input	Resets the Aurora 8B/10B core (active-High).

In streaming mode, the incoming RX Data is compared against calculated RX data.

Framing UFC RX Interface

Table 10-8 shows the FRAME_CHECK framing UFC RX interface of the Aurora 8B/10B core, with LocalLink compliant ports for UFC RX data.

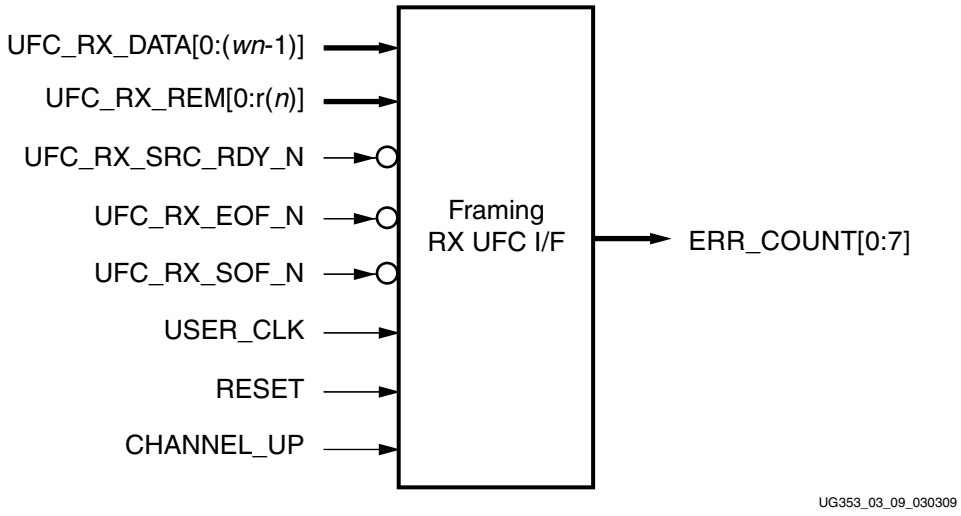

Figure 10-8: **Aurora 8B/10B Core Framing UFC RX Interface (FRAME_CHECK)**

Table 10-8 lists the FRAME_CHECK framing UFC RX data ports and their descriptions.

Table 10-8: **FRAME_CHECK Framing UFC User I/O Ports (RX)**

Name	Direction	Description
UFC_RX_DATA [0:(*wn*-1)]	Input	Incoming UFC message data from the channel partner.
UFC_RX_REM [0:r(*n*)]	Input	Specifies the number of valid bytes of data presented on the UFC_RX_DATA port on the last word of a UFC message. Valid only when UFC_RX_EOF_N is asserted. n = 16 bytes max. REM bus widths are given by [0:r(n)], where r(n) = ceiling [{$\log_2(n)$}-1].
UFC_RX_SRC_RDY_N	Input	Asserted when the values on the UFC_RX_DATA port is valid. When this signal is not asserted, all values on the UFC_RX_DATA port should be ignored (active-Low).
UFC_RX_SOF_N	Input	Signals the start of the incoming UFC message (active-Low).
UFC_RX_EOF_N	Input	Signals the end of the incoming UFC message (active-Low).
ERR_COUNT[0:7]	Output	Count of the number of RX UFC data words received by the frame checker that did not match with the expected value
CHANNEL_UP	Input	Asserted when Aurora 8B/10B channel initialization is complete and channel is ready to send data.
USER_CLK	Input	Parallel clock shared by the Aurora 8B/10B core and the user application.
RESET	Input	Resets the Aurora 8B/10B core (active-High).

The expected UFC RX data is computed by LFSR. Error checking and counter logic is similar to that of "Framing RX Data Interface," page 102. If the incoming RX UFC data does not match the expected RX UFC data, an 8-bit counter is incremented. This error counter is indicated to the user through ERR_COUNT port.

Framing NFC RX Interface

Figure 10-9 shows the framing NFC RX interface of the Aurora 8B/10B core with ports.

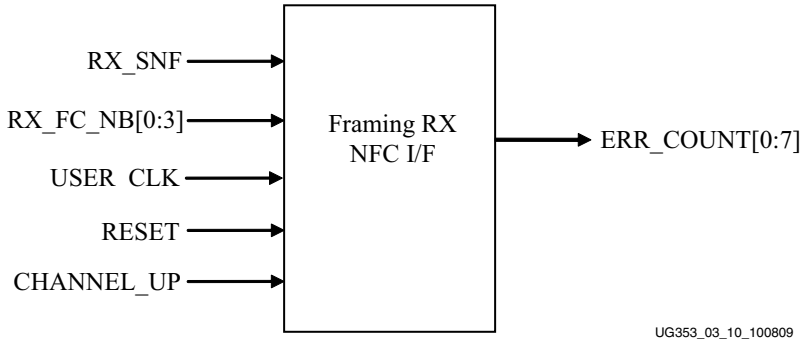

Figure 10-9: **Aurora 8B/10B Core Framing NFC RX Interface (FRAME_CHECK)**

Table 10-9 lists the FRAME_CHECK framing NFC RX ports and their descriptions.

Table 10-9: **Aurora 8B/10B Core Framing NFC RX Interface (FRAME_CHECK**

Name	Direction	Description
RX_SNF	Input	Signals the start of the incoming NFC message.
RX_FC_NB[0:3]	Input	Received NFC message PAUSE value
CHANNEL_UP	Input	Asserted when Aurora channel initialization is complete and channel is ready to send data.
USER_CLK	Input	Parallel clock shared by the Aurora 8B/10B core and the user application.
RESET	Input	Resets the Aurora core (active-High).
ERR_COUNT[0:7]	Output	Incremented when mismatch detected between the expected PAUSE value and the PAUSE value received by the FRAME_CHECK.

FRAME_GEN transmits NFC message with predefined PAUSE value each time. This PAUSE value is compared with PAUSE value received by FRAME_CHECK. If the incoming PAUSE value does not match with expected PAUSE value, an 8-bit counter will be incremented. This error counter is indicated to the user through ERR_COUNT port.

Because cores are generated one at a time, simulating simplex cores requires additional steps (except for RX/TX Simplex cores, which can be connected to themselves). To simulate a simplex TX or simplex RX core, perform the following steps:

1. Generate the core for simulation.
2. Generate a complimentary simplex core. Go to the implement directory of the first core generated.
3. Set the environment variable SIMPLEX_PARTNER to point to the directory for the complementary core.
4. Run the script according to the instructions in Chapter 9, "Quick Start Example Design."

Note: The top-level module name of the simplex design and simplex partner design should be similar. For example, if the top-level module name of the TX simplex design is simplex_201_tx, then the top-level module name of the simplex partner should be rx_simplex_201_tx.

XILINX®

Chapter 11

Project Directory Structure

The customized Aurora 8B/10B core is delivered as a set of HDL source modules in the language selected in the CORE Generator™ software project with supporting script and documentation files. These files are arranged in a predetermined directory structure under the project directory name provided to the CORE Generator software when the project is created as shown in this section.

Directory and File Structure

- **<project directory>**
  Top-level project directory; name is user-defined.
  - <project directory>/<component name>
    Top-level core directory; contains the core deliverables and the readme file
    - <component name>/doc
      Product documentation
    - <component name>/example_design
      Example design and user constraints files
      - /example_design/cc_manager
        Verilog/VHDL design files for the clock compensation block
      - /example_design/clock_module
        Verilog/VHDL design files for the clocking blocks
      - /example_design/gt
        Verilog/VHDL wrapper files for the GTP/GTX transceivers
      - /example_design/traffic_gen_check
        Verilog/VHDL design files for the frame generator and checker
    - <component name>/implement
      Implementation scripts and support files
    - <component name>/simulation
      Simulation test bench and simulation script files
      - /simulation/functional
        Functional simulation files
      - /simulation/timing
        Timing simulation file
    - <component name>/src
      Verilog/VHDL files for the core

Directory and File Contents

The Aurora 8B/10B core directories and their associated files are defined in the following sections.

<project directory>

The project directory contains the CORE Generator project files.

Table 11-1: **project Directory**

Name	Description
<project directory>	
<coregen project filename>.cgp	CORE Generator project file

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<project directory>/<component name>

This top level core directory contains the core deliverables and the readme file.

Table 11-2: **component name Directory**

Name	Description
<project directory>/<component name>	
aurora_8b10b_readme.txt	Release notes file

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<component name>/doc

The doc directory contains the product documentation.

Table 11-3: **doc Directory**

Name	Description
<component name>/doc	
aurora_8b10b_ds637.pdf	*LogiCORE IP Aurora 8B/10B v5.2 Data Sheet*
aurora_8b10b_ug353.pdf	*LogiCORE IP Aurora 8B/10B v5.2 User Guide*

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<component name>/example_design

The example_design directory contains the example design and user constraints files provided with the core.

Table 11-4: **example_design Directory**

Name	Description
<component name>/example_design	
`<component name>_example_design.v[hd]`	Example design source file
`<component_name>_example_design.ucf`	Aurora 8B/10B example design constraints
`<component_name>_reset_logic.v[hd]`	Aurora 8B/10B reset logic

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/example_design/cc_manager

The cc_manager directory contains the clock compensation source file.

Table 11-5: **cc_manager Directory**

Name	Description
<component name>/example_design/cc_manager	
`<component name>_standard_cc_module.v[hd]`	Clock compensation module source file

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/example_design/clock_module

The clock_module directory contains the clock module source file.

Table 11-6: **clock_module Directory**

Name	Description
<component name>/example_design/clock_module	
`<component name>_clock_module.v[hd]`	Clock module source file

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/example_design/gt

The gt directory contains the Verilog/VHDL wrapper files for the GTP/GTX transceivers.

Table 11-7: **gt Directory**

Name	Description
<component name>/example_design/gt	
<component name>_transceiver_tile.v[hd] <component name>_transceiver_wrapper.v[hd]	Verilog/VHDL wrapper files for the transceiver

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/example_design/traffic_gen_check

The traffic_gen_check directory contains frame generator and frame checker modules for Aurora 8B/10B core.

Table 11-8: **traffic_gen_check Directory**

Name	Description
<component name>/example_design/traffic_gen_check	
<component name>_frame_check.v[hd] <component name>_frame_gen.v[hd]	Example design traffic generation and checker files

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<component name>/implement

The implement directory contains scripts and support files for both Linux and Windows operating systems. These scripts automate the process of synthesizing and implementing the files needed for the example design.

Table 11-9: **implement Directory**

Name	Description
<component name>/implement	
implement.bat	Windows batch file that processes the example design through the Xilinx tool flow
implement.sh	Linux shell script that processes the example design through the Xilinx tool flow
xst.scr	XST script file for the example design
xst.prj	XST project file for the example design
icon.ngc vio.ngc	Virtex-5 family NGC files for the debug cores compatible with the ChipScope Pro Analyzer tool
v6_icon.ngc v6_vio.ngc	Virtex-6 family NGC files for the debug cores compatible with the ChipScope Pro Analyzer tool
s6_icon.ngc s6_vio.ngc	Spartan-6 family NGC files for the debug cores compatible with the ChipScope Pro Analyzer tool

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<component name>/simulation

The simulation directory contains the test bench files for the example design.

Table 11-10: **simulation Directory**

Name	Description
<component name>/simulation	
demo_tb.v[hd]	Test bench file for simulating the example design

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/simulation/functional

The functional directory contains functional simulation scripts provided with the core.

Table 11-11: **functional Directory**

Name	Description
<component name>/simulation/functional	
simulate_mti.do	ModelSim macro file that compiles the example design sources, the structural simulation model, and the demonstration test bench then runs the functional simulation to completion
mti_wave.do	ModelSim macro file that opens a Wave window

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/simulation/timing

The timing directory contains the timing simulation scripts provided with the core.

Table 11-12: **timing Directory**

Name	Description
<component name>/simulation/timing	
simulate_mti.do	Modelsim macro file that compiles the post place and route netlist of the example design along with standard delay format (SDF) back annotation then runs timing simulation to completion

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<component name>/src

The src directory contains the source files related to the Aurora 8B/10B example design.

Table 11-13: **src Directory**

Name	Description
<component name>/src	
<component name>_aurora_lane.v[hd] <component name>_aurora_pkg.vhd (VHDL Only) <component name>_channel_err_detect.v[hd] <component name>_channel_init_sm.v[hd] <component name>_chbond_count_dec.v[hd] <component name>_err_detect.v[hd] <component name>_global_logic.v[hd] <component name>_idle_and_ver_gen.v[hd] <component name>_lane_init_sm.v[hd] <component name>_rx_ll.v[hd] <component name>_rx_ll_pdu_datapath.v[hd] <component name>_sym_dec.v[hd] <component name>_sym_gen.v[hd] <component name>_tx_ll.v[hd] <component name>_tx_ll_control.v[hd] <component name>_tx_ll_datapath.v[hd]	Aurora 8B/10B source files

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Appendix A

Two Aurora 8B/10B Cores in the Virtex-5 Family Sharing a High-Speed Serial GTX Transceiver Tile

The high-speed serial GTX transceivers in the Virtex®-5 family have dual architecture in which two transceivers in each tile share a common PLL (see the *Virtex-5 FPGA RocketIO GTX Transceiver User Guide*). The Aurora 8B/10B core requires one transceiver per lane and the transceiver can be mapped to GTX0 or GTX1 in a tile. The unused transceiver is powered down during core generation. To use the powered down transceiver in another Aurora 8B/10B core sharing a common reference clock source, the user needs to bring out the unused transceiver ports to the Aurora 8B/10B GTX transceiver wrapper file.

The Aurora 8B/10B core instantiates the GTX_DUAL primitive in <component name>_tile.v[hd] file located in <component name>/example_design/gt/. This file contains GTX0 and GTX1 specific ports together with shared tile ports. This file is hierarchically called in <component name>_gtx_wrapper.v[hd]. The user needs to bring out unused transceiver ports to the <component name>_gtx_wrapper.v[hd] file that is instantiated in the Aurora 8B/10B core's top level file. See Figure A-1, page 115 for the file level hierarchy of the Aurora 8B/10B core.

Shared Ports and Attributes

The Aurora 8B/10B core automatically sets shared PLL ports and attributes based on the CORE Generator™ software options. The reference clock input and the GTX transceiver reset to a tile is driven from the example design. The Aurora 8B/10B core derives the clock from TXOUTCLK0/1 and based on GTX0 or GTX1 selection, TXOUTCLK0 or TXOUTCLK1 is fed to the PLL input for generating the core's clock. The user can use the same TXOUTCLK source used in core 1 (See Figure A-1, page 115), or optionally can use the other TXOUCLK port for core 2. Because the reference clock and GTX transceiver reset ports are common for GTX0 and GTX1, application of the GTX transceiver reset resets both cores.

Steps to Modify Transceiver Specific Ports and Attributes

When the user shares a tile across two Aurora 8B/10B cores, the shared PLL settings are not user modifiable. Do the following steps to modify GTX0 or GTX1 specific ports and attributes:

1. Generate an Aurora 8B/10B v5.2 core (core 1) using the CORE Generator software. Select transceiver locations such that GTX0 or GTX1 in a tile is used for the Aurora 8B/10B core.
2. Generate another Aurora 8B/10B core (core 2) such that the unused transceiver (GTX1 or GTX0) in core 1 is used for core 2.
3. Go to the directory <component name>/example_design/gt/ of core 1.
4. Open the <component name>_tile.v[hd] file.

This procedure assumes Line rate and GT reference clock are same for both generated Aurora 8B/10B cores (core1 and core2).

Steps to Bring Out Unused Transceiver Ports to the Aurora 8B/10B GTX Transceiver Wrapper File

Do the following steps to bring out unused transceiver ports to the Aurora 8B/10B <component name>_gtx_wrapper.v[hd] file:

1. Go to the directory <component name>/example_design/gt/ of core 1.
2. Open the <component name>_gtx_wrapper.v[hd] file.
3. Search for GTX_TILE_INST in <component name>_gtx_wrapper.v[hd]
   a. For a multi-lane Aurora 8B/10B core, search for <component name>_TILE_INST_LANEn (*n*=1, 2, 3, . . .)
   b. Select the instance in which one transceiver is used for the Aurora 8B/10B core and the other is powered down
4. If GTX0 is used for the Aurora 8B/10B core, attributes and ports with a "1" appended are hard coded to the default values.

   If GTX1 is used for the Aurora 8B/10B core, attributes and ports with a "0" appended are hard coded to the default values.

   Port map all such hard-coded ports and attributes to new ports and attributes and declare them in the module list (Verilog) or entity list (VHDL).

Steps to Instantiate Two Aurora 8B/10B Cores in the Top Level File

Use these steps to instantiate two Aurora 8B/10B cores in the top level file:

1. Go to the directory <component name>/example_design of core 1.
2. Open <component name>.v[hd].
3. Search for gtx_wrapper_i instance. Add all new ports and attributes in step 4, page 114 in the port map list.
4. Map these new ports to another Aurora 8B/10B core (see core 2 in Figure A-1).

Figure A-1 shows two single lane Aurora 8B/10B cores sharing a tile in the Virtex-5 family.

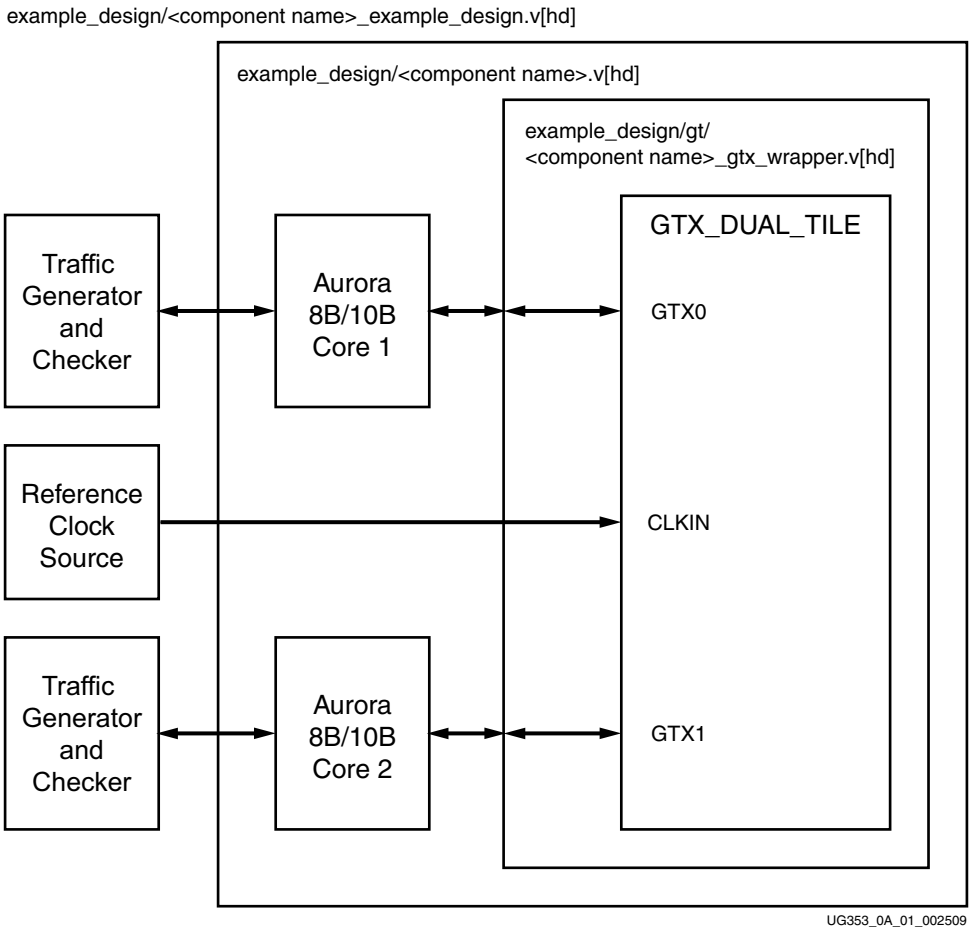

Figure A-1: **Example Showing Two Single Lane Aurora 8B/10B Cores Sharing a Transceiver Tile in the Virtex-5 Family**

Note: The same procedure can be followed when a GTP transceiver device is selected. But instead of GTX transceiver references, GTP transceiver should be used in appropriate places.

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Appendix B

Handling Timing Errors Due To Far Apart Transceiver Selection

The Aurora 8B/10B core allows the user to select any combination of transceiver(s) during core generation. The design parameters that affect the timing performance are:

- Line rate
- Transceiver data path width (2/4 bytes) and
- Number of unused transceivers between two selected transceivers

As a result of one or more of the above parameters, timing errors can occur because:

- CHBONDO does not meet timing
- RXCHARISCOMMA, RXCHARISK, and RXCHANISALIGNED do not meet timing

Example Solutions

The following suggestions can be attempted to meet timing.

1. Select the transceivers consecutively.
   a. Use the Lane Assignment in the Aurora 8B/10B GUI to select the transceivers during core generation.

   Note: Most of the timing errors are due to unused transceivers and channel bonding signals connections among transceivers.
2. Use the smartexplorer tool provided with the ISE® software.
   a. Refer to the ISE software documentation for instructions to use smartexplorer.

Example Solution

This example solution describes how to meet timing for a 2-lane Aurora 8B/10B design using the following Virtex®-5 GTX transceivers that were selected in the GUI:

- GTX_DUAL_X0Y0
- GTX_DUAL_X0Y5

  Note: The unused transceivers (GTX_DUAL_X0Y1 to GTX_DUAL_X0Y4) cause timing issues between channel bonding signals.

The example solution uses the following Virtex-5 GTX transceiver channel bonding attributes:

- CHAN_BOND_LEVEL_0
- CHAN_BOND_LEVEL_1

These two attributes define the internal pipeline stages used inside the GTX transceiver. The channel bond levels must be continuous in normal working conditions from MASTER to SLAVE transceiver(s). Transceiver channel bonding mode is defined by CHAN_BOND_MODE_0/CHAN_BOND_MODE_1.

For details about enabling channel bonding and daisy chaining, see the *Virtex-5 FPGA RocketIO GTX Transceiver User Guide*. The Aurora 8B/10B core sets the channel bonding attributes as recommended in *Virtex-5 FPGA RocketIO GTX Transceiver User Guide*.

The channel bonding attributes are available as parameters in <component_name>_transceiver_wrapper.v[hd] file in example_design/gt directory. Figure B-1 shows the typical channel bonding attribute settings for the 2-lane Aurora 8B/10B design.

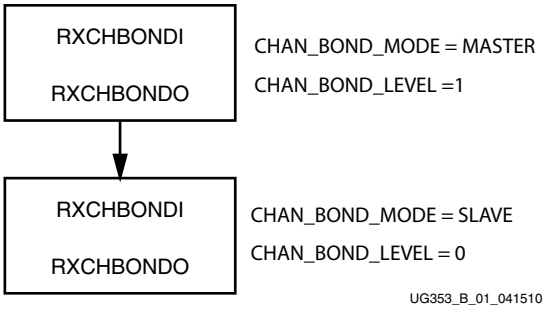

Figure B-1: **Example of Typical Channel Bonding Daisy Chaining**

Channel bond levels are adjusted and flip-flop(s) are introduced to meet timing in the proposed solution. The flip-flop is clocked by RXUSRCLK. Figure B-2 shows the modifications in channel bonding port connections. Note that channel bond level of MASTER has been updated to 2 due to introduction of the new flip-flop in the FPGA fabric. This new flip-flop helps for timing closure.

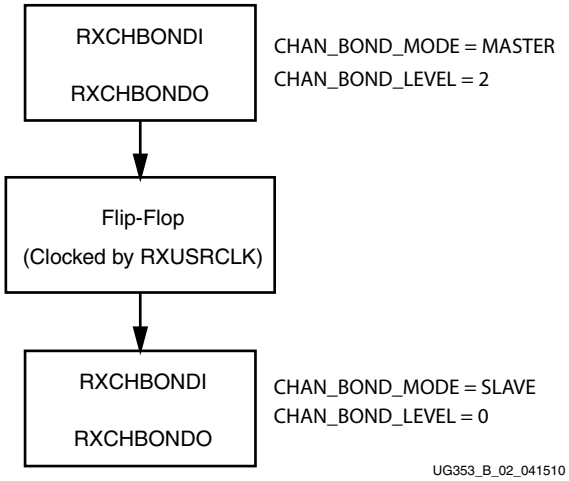

Figure B-2:

One flip-flop stage is shown for demonstration purpose and six flip-flop stages can be added at maximum. The number of flip-flop stages varies from one to six and depends on following parameters: device, speed grade, line rate, lane width, and unused transceivers between transceiver.

This solution can be extended for the Aurora 8B/10B design that is having many unused transceivers intermittently. In that case, channel bonding daisy chaining must be handcrafted. This solution should be attempted only when timing errors are a result of unused transceiver(s) in between the channel bonding signals.

Appendix C

Performance and Core Latency

Introduction

Latency through an Aurora 8B/10B core is caused by pipeline delays through the protocol engine (PE) and through the GTP/GTX transceivers. The PE pipeline delay increases as the LocalLink interface width increases. The GTP/GTX transceivers delays are fixed per the features of the GTP/GTX transceivers.

This section outlines expected latency for the Aurora 8B/10B core's LocalLink user interface in terms of USER_CLK cycles for 2-byte-per-lane and 4-byte-per-lane designs. For the purposes of illustrating latency, the Aurora 8B/10B modules partitioned into GTP/GTX transceivers logic and protocol engine (PE) logic implemented in the FPGA fabric.

Note: These figures do not include the latency incurred due to the length of the serial connection between each side of the Aurora 8B/10B channel.

Latency of the Frame path

Figure C-1 illustrates the latency of the frame path.

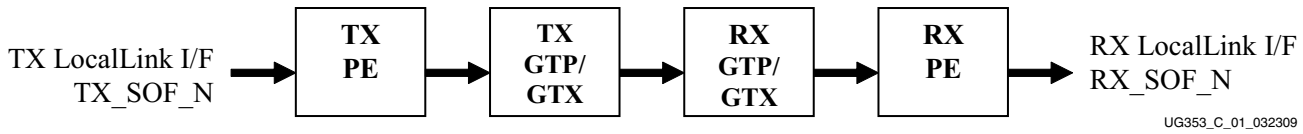

Figure C-1: **Latency of the Frame Path**

Maximum latency for a 2-byte designs from TX_SOF_N to RX_SOF_N is approximately 52 GTP/GTX transceivers clock cycles in simulation.

Maximum latency for a 4-byte designs from TX_SOF_N to RX_SOF_N is approximately 61 GTP/GTX transceivers clock cycles in simulation.

The pipeline delays are designed to maintain the clock speed.

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Appendix D

Generating a Wrapper File from Its Respective GTP/GTX Transceiver Wizard

The transceiver attributes play a vital role in the functionality of the Aurora 8B/10B core. It is important to have the updated transceiver wrapper file. One way to achieve this is to generate the transceiver wrapper file from the latest RocketIO™ Transceiver Wizard.

This appendix provides instructions to generate the following transceiver wrapper files:

Case 1: Virtex-5 FPGA GTP Wrapper

Use the following steps to generate the transceiver wrapper file using the Virtex®-5 FPGA RocketIO GTP Transceiver Wizard:

1. Using CORE Generator™ tool, run the latest version of the Virtex-5 FPGA RocketIO GTP Transceiver Wizard. Make sure the Component Name of the transceiver wizard matches the Component Name of the Aurora 8B/10B core.
2. Select transceiver(s) and the clock source(s) based on application requirement.
3. Make sure Internal Data Width is 10.
4. Select the protocol template to either *aurora single lane* or *aurora multi lane* based on the number of lane(s).
5. Change the Target Line Rate in Gbps based on the application requirement.
6. Select the Reference Clock from the drop-down box menu based on the application requirement.
7. Select RXCHARISCOMMA port in 8b/10b Optional Ports, if not selected by default.
8. Select TXBUFSTATUS port in Latency, Buffering and Clocking, if not selected by default.
9. Keep all other settings as default.
10. Generate the core.
11. Replace the <component name>_tile.v[hd] file in the example_design/gt directory available in the Aurora 8B/10B core with the generated <component name>_tile.v[hd] file generated from the Virtex-5 FPGA RocketIO GTP Transceiver Wizard.

The transceiver tile settings for Aurora 8B/10B core are up to date now.

Case 2: Virtex-5 FPGA GTX Wrapper

Use the following steps to generate the transceiver wrapper file using the Virtex-5 FPGA RocketIO GTX Transceiver Wizard.

1. Using CORE Generator tool, run the latest version of the Virtex-5 FPGA RocketIO GTX Transceiver Wizard. Make sure the 'Component Name' of the transceiver wizard matches the Component Name of the Aurora 8B/10B core
2. Select transceiver(s) and the clock source(s) based on the application requirement
3. Make sure Internal Data Width as 20
4. Select the protocol template to either *aurora single lane* or *aurora multi lane* based on the number of lane(s)
5. Change the Target Line Rate in Gbps based on the application requirement
6. Select the Reference Clock from the drop-down box menu based on the application requirement
7. Select Data Path Width to 16 in both TX Settings and RX Settings if Aurora 8B/10B core's Lane Width is selected as 2.
8. Select Data Path Width to 32 in both TX Settings and RX Settings if Aurora 8B/10B core's Lane Width is selected as 4.
9. Select TXBUFSTATUS port in Latency, Buffering and Clocking, if not selected by default
10. Keep all other settings as default.
11. Generate the core.
12. Replace the <component name>_tile.v[hd] file in the example_design/gt directory available in the Aurora 8B/10B core with the generated <component name>_tile.v[hd] file generated from the Virtex-5 FPGA RocketIO GTX Transceiver Wizard.

The transceiver tile settings for Aurora 8B/10B core are up to date now.

Case 3: Virtex-6 FPGA GTX Wrapper

Use the following steps to generate the transceiver wrapper file using the Virtex-6 FPGA GTX Transceiver Wizard.

1. Using CORE Generator tool, run the latest version of the Virtex-6 FPGA GTX Transceiver Wizard. Make sure the Component Name of the transceiver wizard matches the Component Name of the Aurora 8B/10B core.
2. Select the protocol template from the following based on number of lane(s) and Lane Width:
   - Aurora 2byte single lane
   - aurora 4byte single lane
   - aurora 2byte multi lane
   - aurora 4byte multi lane
3. Change the Line Rate in both TX and RX based on the application requirement
4. Select the Reference Clock from the drop-down box menu in both TX and RX based on the application requirement
5. Select transceiver(s) and the clock source(s) based on the application requirement
6. Keep all other settings as default.
7. Generate the core.
8. Replace the <component name>_tile.v[hd] file in the example_design/gt directory available in the Aurora 8B/10B core with the generated <component name>_tile.v[hd] file generated from the Virtex-6 FPGA GTX Transceiver Wizard.

The transceiver tile settings for Aurora 8B/10B core are up to date now.

Case 4: Spartan-6 FPGA GTP Wrapper

Use the following steps to generate the transceiver wrapper file using the Spartan®-6 FPGA GTP Transceiver Wizard.

1. Using CORE Generator tool, run the latest version of Spartan-6 FPGA GTP Transceiver Wizard. Make sure the Component Name of the transceiver wizard matches the Component Name of the Aurora 8B/10B core
2. Select transceiver(s) and the clock source(s) based on application requirement
3. Select the protocol template to either *aurora single lane* or *aurora multi lane* based on the number of lane(s)
4. Change the Target Line Rate in Gbps based on application requirement
5. Select the Reference Clock from the drop-down box menu based on application requirement
6. Select RXCHARISCOMMA and RXCHARISK ports in 8B/10B Optional Ports, if not selected by default
7. Select TXBUFSTATUS port in Synchronization and Clocking, if not selected by default
8. Keep all other settings as default.
9. Generate the core.
10. Replace the <component name>_tile.v[hd] file in the example_design/gt directory available in the Aurora 8B/10B core with the generated <component name>_tile.v[hd] file generated from the Spartan-6 FPGA GTP Transceiver Wizard.

The transceiver tile settings for Aurora 8B/10B core are up to date now.