

ECE 3150 Homework #4 Solutions
(Farhan Rana)

4.1
a) charge stored in the N-side $= qA \int_{x_n}^{x_n+x_n} p'(x) dx = Q_d A$
 $= qA \frac{n_i^2}{N_d} \frac{W_n}{2} (e^{\frac{qV_D}{kT}} - 1)$

$$\Rightarrow C_d = \frac{q^2 A n_i^2}{kT N_d} \frac{W_n}{2} e^{\frac{qV_D}{kT}}$$

$$\text{current} = I_D = qA \frac{n_i^2}{N_d} \frac{D_p}{W_n} (e^{\frac{qV_D}{kT}} - 1)$$

$$\Rightarrow \bar{r}_d = \frac{q^2 A n_i^2}{kT N_d} \frac{D_p}{W_n} (e^{\frac{qV_D}{kT}} - 1)$$

b) $i_d(\omega) = (j\omega C_d + \frac{1}{\bar{r}_d}) V_d(\omega) \Rightarrow V_d(\omega) = \frac{\bar{r}_d i_d(\omega)}{1 + j\omega C_d \bar{r}_d}$

c) $\omega_{3dB} = \frac{1}{C_d \bar{r}_d}$

d) $\omega_{3dB} = \frac{2D_p}{W_n^2}$

e) 4 mice

f) $N = RT$

g) $J_p(x=x_n) = q \frac{n_i^2}{N_d} \frac{D_p}{W_n} (e^{\frac{qV_D}{kT}} - 1)$

$$Q_d = q \frac{n_i^2}{N_d} \frac{W_n}{2} (e^{\frac{qV_D}{kT}} - 1)$$

$$\tau_{p\text{-diff}} = \frac{Q_d}{J_p(x=x_n)} = \frac{W_n^2}{2D}$$

h) $\omega_{3dB} = \frac{1}{\tau_{p\text{-diff}}}$

4.2

a) $\Phi_B = \Phi_n - \Phi_p = \Phi_n - \Phi_p = kT \log \left(\frac{N_n N_d}{n_i^2} \right) = 0.89 \text{ V}$

b) $E_x(x=0^-) = \frac{\epsilon_s}{\epsilon_{ox}} E_x(x=0^+) = + q \frac{N_d X_{do}}{\epsilon_{ox}}$

$\frac{dE_x}{dx} = \frac{\rho_0}{\epsilon_{ox}} \Rightarrow E_x(x) = q \frac{N_d X_{do}}{\epsilon_{ox}} + \frac{\rho_0 x}{\epsilon_{ox}}$

d) $\frac{d\phi}{dx} = -E_x(x) = -q \frac{N_d X_{do}}{\epsilon_{ox}} - \frac{\rho_0 x}{\epsilon_{ox}}$

BC: $\phi(x=0) = \Phi_p + q \frac{N_d X_{do}^2}{2\epsilon_s}$

$\Rightarrow \phi(x) = \Phi_p + q \frac{N_d X_{do}^2}{2\epsilon_s} - q \frac{N_d X_{do}}{\epsilon_{ox}} x - \frac{\rho_0 x^2}{2\epsilon_{ox}}$

f) $\phi(x=-t_{ox}) = \Phi_m$

$\Rightarrow \Phi_p + q \frac{N_d X_{do}^2}{2\epsilon_s} + q \frac{N_d X_{do} t_{ox}}{\epsilon_{ox}} - \frac{\rho_0 t_{ox}^2}{2\epsilon_{ox}} = \Phi_m$

$\Rightarrow q \frac{N_d X_{do}^2}{2\epsilon_s} + q \frac{N_d X_{do} t_{ox}}{\epsilon_{ox}} = \Phi_B + \frac{\rho_0 t_{ox}^2}{2\epsilon_{ox}}$

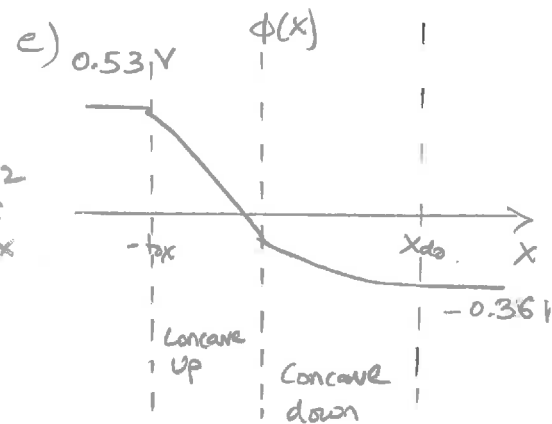
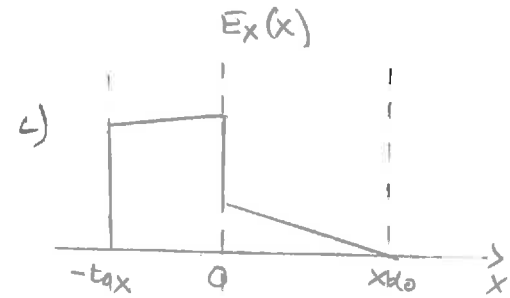
$\Rightarrow X_{do} = -\frac{\epsilon_s}{C_{ox}} + \sqrt{\left(\frac{\epsilon_s}{C_{ox}}\right)^2 + \left(\frac{2\epsilon_s}{qN_d}\right)\left(\Phi_B + \frac{\rho_0 t_{ox}^2}{2\epsilon_{ox}}\right)}$

g) when $V_{GB} \neq 0$ $X_{do} = -\frac{\epsilon_s}{C_{ox}} + \sqrt{\left(\frac{\epsilon_s}{C_{ox}}\right)^2 + \left(\frac{2\epsilon_s}{qN_d}\right)\left(\Phi_B + V_{GB} + \frac{\rho_0 t_{ox}^2}{2\epsilon_{ox}}\right)}$

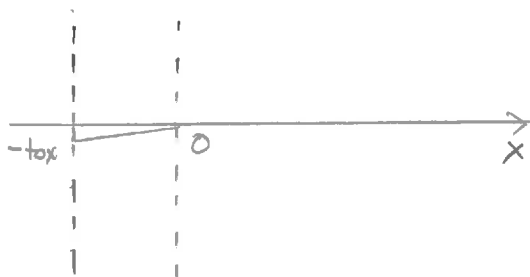
$X_{do} = 0$ when $V_{GB} = -\Phi_B - \frac{\rho_0 t_{ox}^2}{2\epsilon_{ox}} = V_{FB}$

h) In flatband, since there is no charge in the semiconductor, the gate charge must be equal and opposite to the

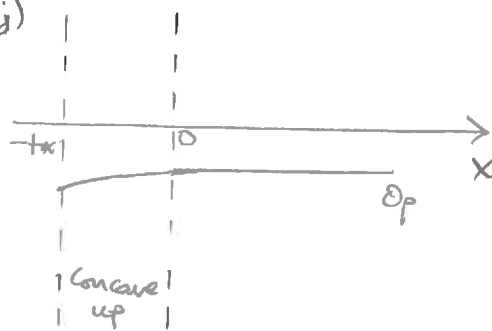
oxide charge $\Rightarrow Q_G = -\rho_0 t_{ox}$



i)



j)



$$k) Q_G = -\rho_0 t_{ox} + C_{ox} (V_{GB} - V_{FB})$$

$$l) X_{do} = -\frac{\epsilon_s}{C_{ox}} + \sqrt{\left(\frac{\epsilon_s}{C_{ox}}\right)^2 + \left(\frac{2\epsilon_s}{qNa}\right)(V_{GB} - V_{FB})}$$

$$m) Q_G = -\rho_0 t_{ox} + qNaX_{do}$$

$$n) \text{ From part (f) } V_{GB} - V_{FB} = \frac{qNa}{2\epsilon_s} X_{do}^2 + \frac{qNa}{\epsilon_0 x} X_{do} t_{ox}$$

$$= (\phi_s - \phi_p) + \frac{\sqrt{2q\epsilon_s Na} (\phi_s - \phi_p)}{C_{ox}}$$

Same as when $\rho_0 = 0$

$$\Rightarrow V_{TN} = V_{FB} - 2\phi_p + \frac{\sqrt{2q\epsilon_s Na} (-2\phi_p)}{C_{ox}}$$

depends
on ρ_0

$$o) V_{TN}(\rho_0) - V_{TN}(\rho_0=0) = -\frac{\rho_0 t_{ox}^2}{2\epsilon_0 x}$$

$$p) Q_G = \sqrt{2q\epsilon_s Na} (-2\phi_p) - \rho_0 t_{ox} + C_{ox} (V_{GB} - V_{TN})$$

4.3

a) $V_{FB} = +0.5 \text{ V}$

b) $V_{TN} = -0.6 \text{ V}$

c) $C_{ox} = \frac{\epsilon_{ox}}{t_{ox}} = 8 \times 10^{-15} \text{ F}/\mu\text{m}^2 = 8 \times 10^{-3} \text{ F}/\text{m}^2$

$\Rightarrow t_{ox} = 4.3 \text{ nm} = 43 \text{ \AA}$

d) N-type (PMOS capacitor)

e) Just before threshold; $C \approx 2 \text{ fF}/\mu\text{m}^2$

$\frac{1}{C} = \frac{1}{C_{ox}} + \frac{1}{C_b} \Rightarrow C_b = \frac{8}{3} \text{ fF}/\mu\text{m}^2 \quad C_b = \frac{\epsilon_s}{x_{dmax}}$

$x_{dmax} = \sqrt{\frac{2 \epsilon_s}{q N_d} (+2\phi_n)} \Rightarrow \sqrt{\frac{q N_d \epsilon_s}{2 (2\phi_n)}} = C_b = \frac{8}{3} \text{ fF}/\mu\text{m}^2$

$\Rightarrow N_d \approx 8 \times 10^{17} \text{ /cm}^3$

f) Analysis similar to that in problem 4.3 gives the

result: $V_{TP}(p_0) - V_{TP}(p_0=0) = -\frac{p_0 t_{ox}^2}{2 \epsilon_{ox}} = -0.25 \text{ V}$

$\Rightarrow p_0 = +9.33 \times 10^{11} \text{ C/cm}^3$

4.4

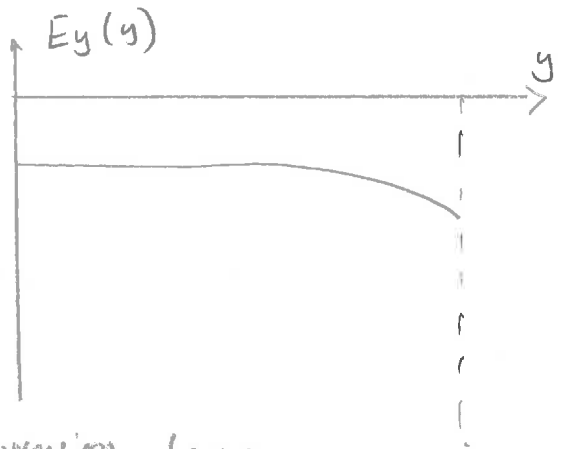
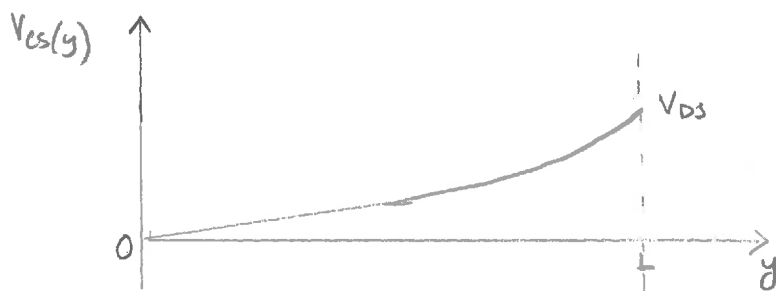
$$a) I_D = W \mu_n C_{ox} (V_{GS} - V_{TN} - V_{cs}(y)) \frac{dV_{cs}(y)}{dy}$$

$$\Rightarrow \int_0^y I_D dy = \int_0^{V_{cs}} W \mu_n C_{ox} (V_{GS} - V_{TN} - V_{cs}) dV_{cs}$$

$$I_D y = W \mu_n C_{ox} \left[V_{GS} - V_{TN} - \frac{V_{cs}(y)}{2} \right] V_{cs}(y)$$

$$\Rightarrow V_{cs}(y) = (V_{GS} - V_{TN}) - \sqrt{(V_{GS} - V_{TN})^2 - \frac{2y I_D}{W \mu_n C_{ox}}}$$

$$b) E_y = -\frac{dV_{cs}}{dy} = -\frac{\frac{I_D}{W \mu_n C_{ox}}}{\sqrt{(V_{GS} - V_{TN})^2 - \frac{2y I_D}{W \mu_n C_{ox}}}}$$



The field is not uniform because the inversion layer charge density is not uniform along the channel. As the inversion layer charge density decreases going towards the drain end, the field increases, but the product gives the constant current (i.e. independent of position).

c) As $V_{DS} \rightarrow V_{GS} - V_{TN}$ and $I_D \rightarrow \frac{W}{L} \mu_n C_{ox} (V_{GS} - V_{TN})^2$ the field at $y=L$, $E_y(y=L)$, approaches infinity — because the inversion layer charge density at $y=L$ is approaching zero.