

# 3F4: Data Transmission

## Handout 11: Decoding Convolutional Codes

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[based on notes by Ramji Venkataramanan]

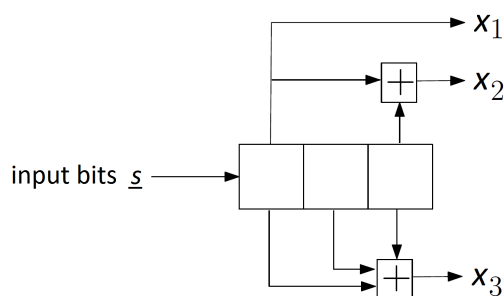
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1 / 21

## Convolutional codes

In convolutional codes, a stream of input bits is transformed into a stream of code bits using a shift register (filter)

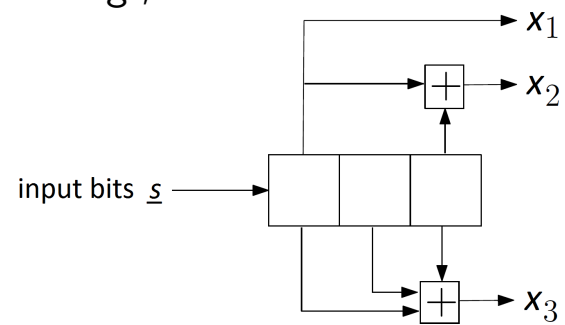


- Here, for every  $k = 1$  data bit, we have  $n = 3$  code bits
- Assuming the initial state of the shift register is (0 0 0) the code bits corresponding to the input  $\underline{s} = 1010$  are  
(111 001 100 001)
- Dependence between code bits is created via the shift register

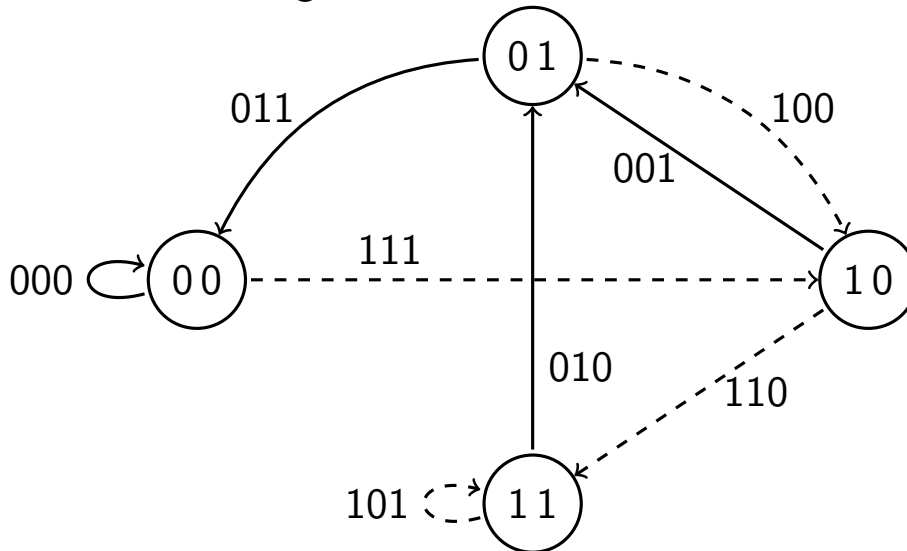
2 / 21

## Finite-state machine description

Convolutional codes can also be represented via *state diagrams* corresponding to *finite-state machines*. E.g., the code:



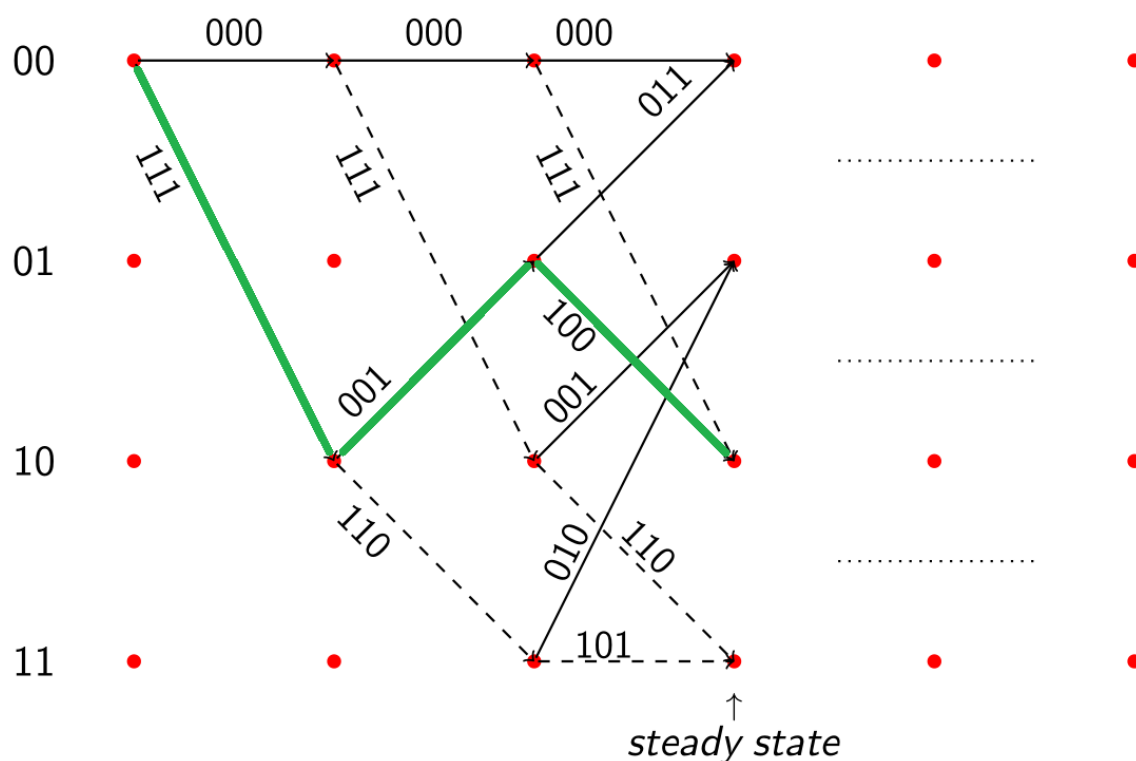
has the state diagram:



3 / 21

## The trellis representation

E.g.: The path traced by the input string  $\underline{s} = 101$  producing the code string  $\underline{x} = 111\ 001\ 100$

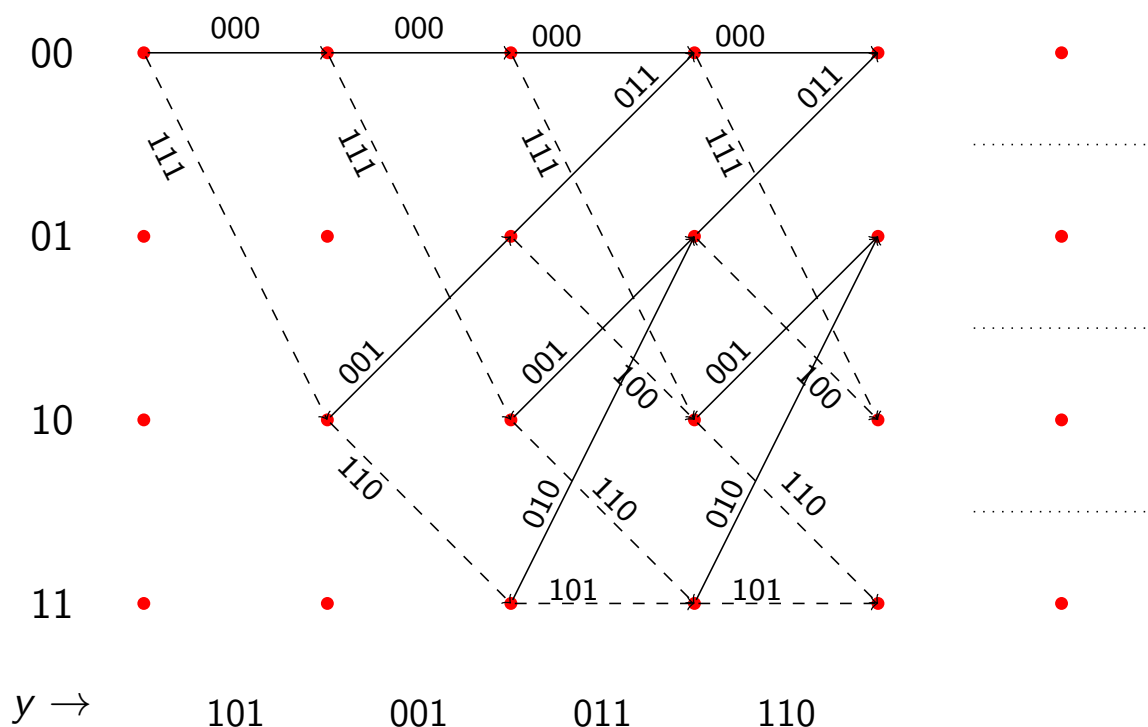


4 / 21

## Decoding convolutional codes

Recall: Optimal decoding is minimum distance decoding

E.g.: If we receive  $\underline{y} = 101\ 001\ 011\ 111$ , we need to find a path in the trellis which gives a code sequence  $\hat{\underline{x}}$  minimising  $d(\underline{y}, \hat{\underline{x}})$

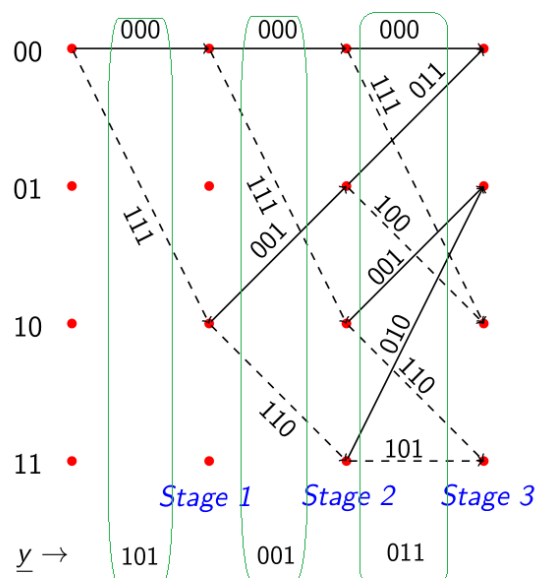
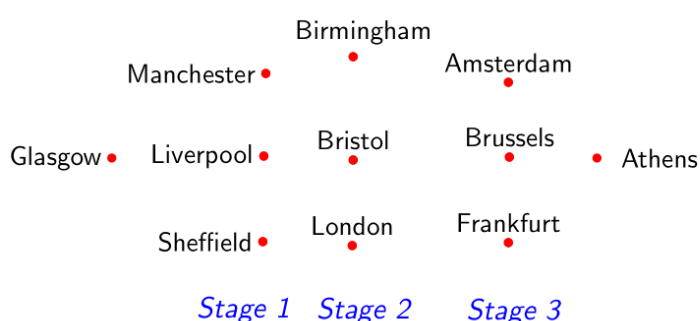


5 / 21

## The Viterbi algorithm

The Viterbi algorithm is simply dynamic programming on the trellis corresponding to a convolutional code

Except easier: Not all paths are allowed!



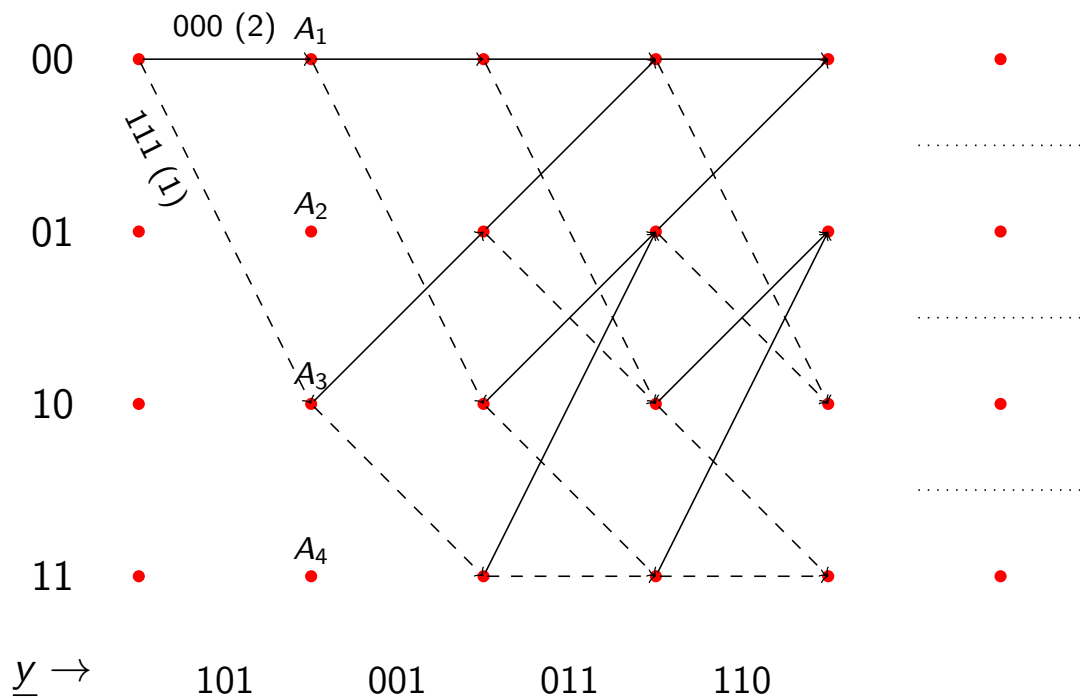
cities  $\mapsto$  states

distance  $\mapsto$  Hamming dist between produced and received code bits

6 / 21

# The Viterbi algorithm

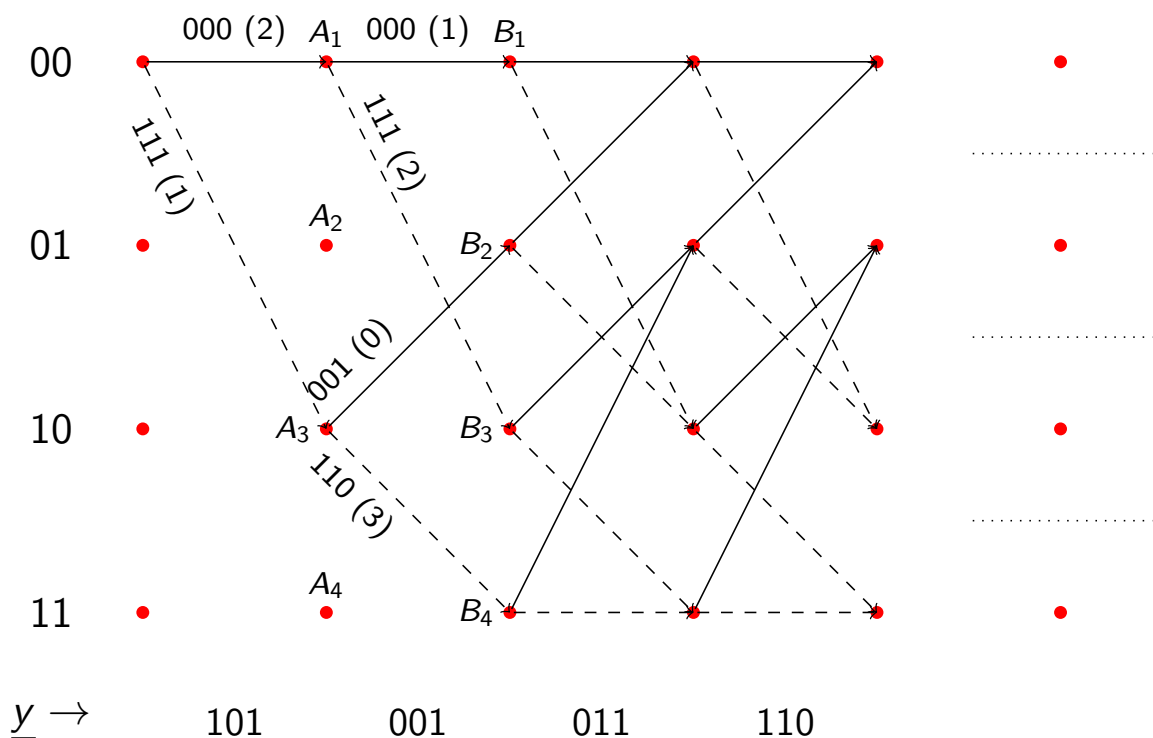
E.g., as before: To decode  $\underline{y} = 101\ 001\ 011\ 111$ , need to find the path in the trellis corresponding to that  $\hat{\underline{x}}$  which minimizes  $d(\underline{y}, \hat{\underline{x}})$ . First stage:



Observe: The number in parentheses is the *Hamming distance* between received and produced code bits

7 / 21

Second stage:



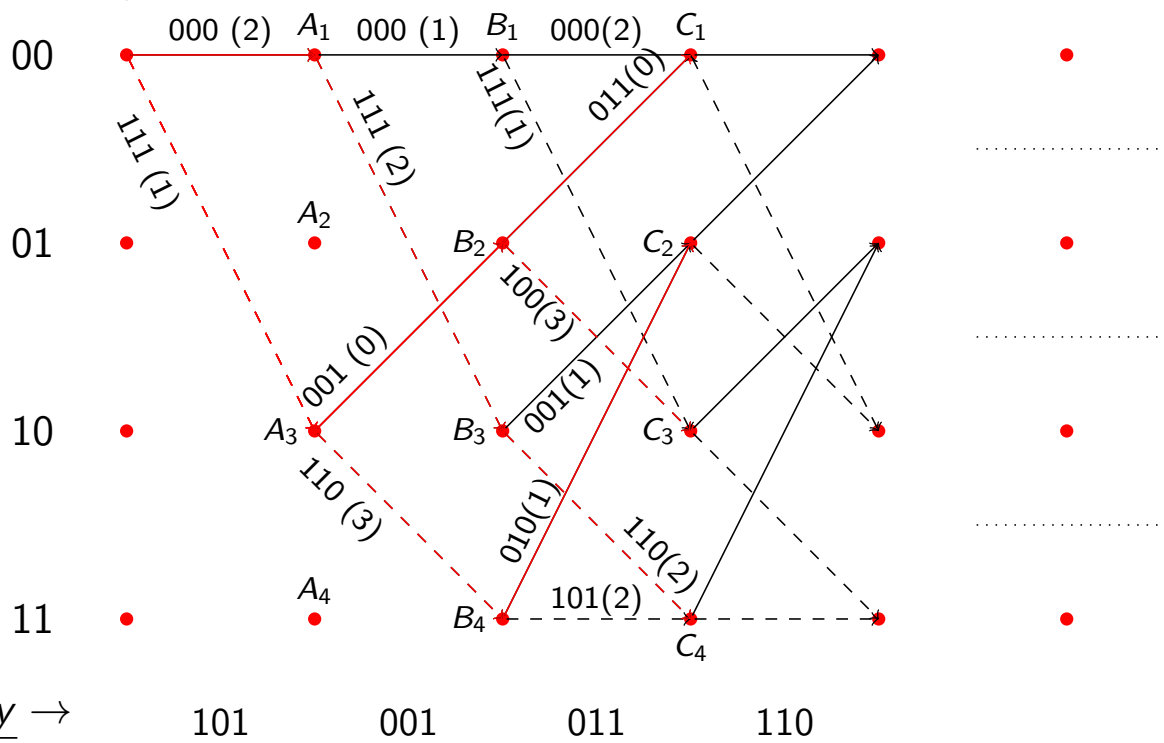
We only have one path from 00 to each of  $B_1, B_2, B_3, B_4$ .

The distances from the corresponding received bits are

$$d(00, B_1) = 3, \quad d(00, B_2) = 1, \quad d(00, B_3) = 4, \quad d(00, B_4) = 4$$

8 / 21

Third stage:



Surviving paths:

$$d(00, C_1) = \min\{d(00, B_1)+2, d(00, B_2)+0\} = 1; \text{ Path : } 00 - A_3 - B_2 - C_1$$

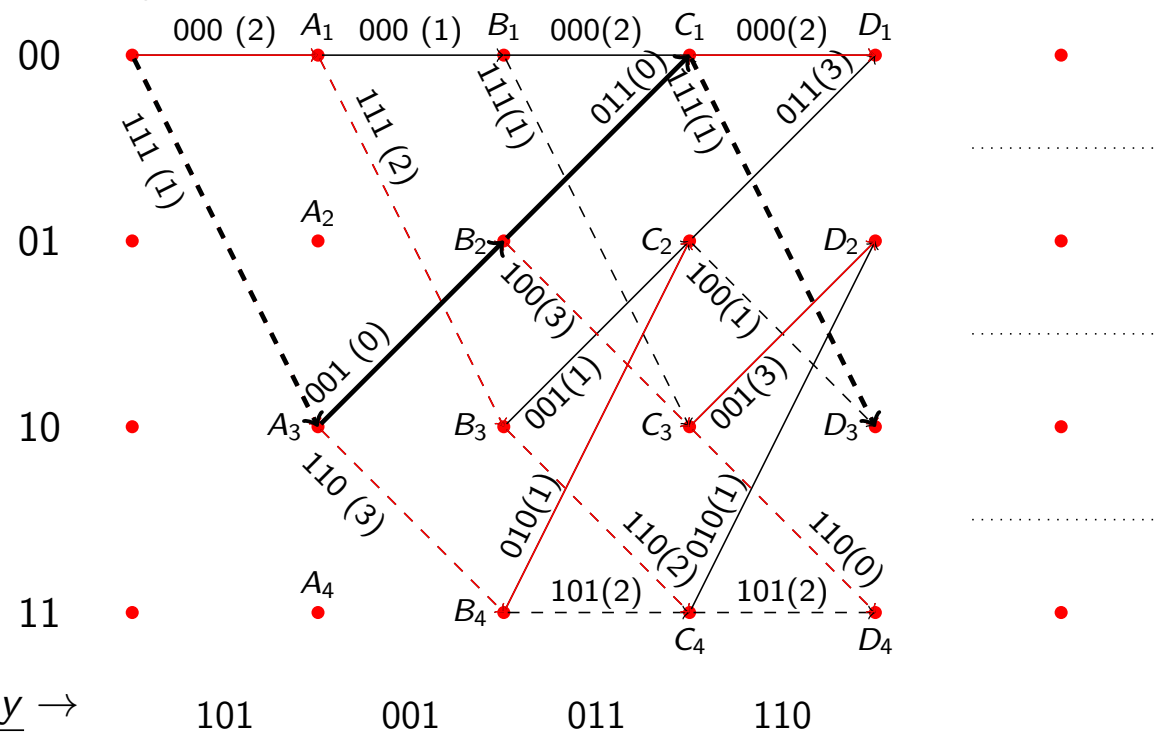
$$d(00, C_2) = \min\{d(00, B_3)+1, d(00, B_4)+1\} = 5; \text{ Path : } 00 - A_3 - B_4 - C_2$$

$$d(00, C_3) = \min\{d(00, B_2)+3, d(00, B_1)+1\} = 4; \text{ Path : } 00 - A_3 - B_2 - C_3$$

$$d(00, C_4) = \min\{d(00, B_3)+2, d(00, B_4)+2\} = 6; \text{ Path : } 00 - A_1 - B_3 - C_4$$

9/21

Fourth stage:



Surviving paths:

$$d(00, D_1) = \min\{d(00, C_1)+2, d(00, C_2)+3\} = 3; \text{ Path : } 00 - A_3 - B_2 - C_1 - D_1$$

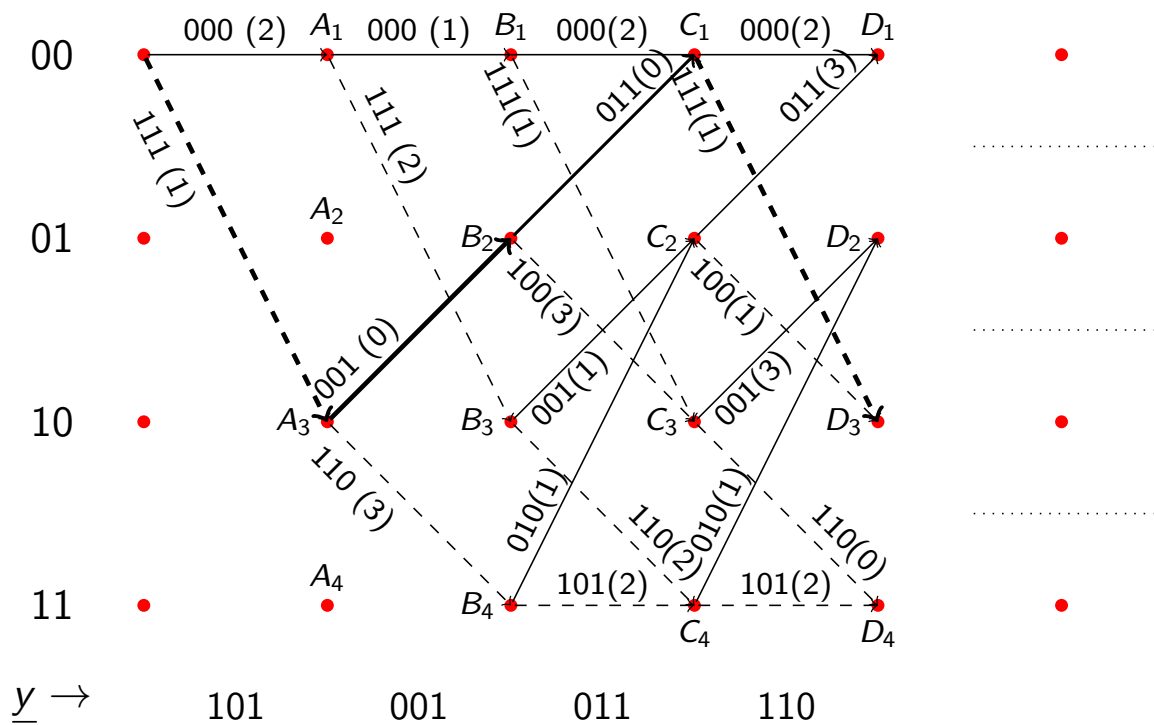
$$d(00, D_2) = \min\{d(00, C_3)+3, d(00, C_4)+1\} = 7; \text{ Path : } 00 - A_3 - B_2 - C_3 - D_2$$

$$d(00, D_3) = \min\{d(00, C_2)+1, d(00, C_1)+1\} = 2; \text{ Path : } 00 - A_3 - B_2 - C_1 - D_3$$

$$d(00, D_4) = \min\{d(00, C_3)+0, d(00, C_4)+2\} = 4; \text{ Path : } 00 - A_3 - B_2 - C_3 - D_4$$

10/21

Finally: Decoded codeword  $\hat{x}$ :



The min-distance path is shown in bold, corresponding to:

$$\hat{x} = 111\ 001\ 011\ 111$$

And the corresponding source sequence, easily determined from the state transitions, is:  $\hat{s} = 1001$

11 / 21

## Aside: Yet another representation of a convolutional code

Input sequence:  $\underline{s} = (s_1, \dots, s_m)$

Generators:  $g_1 = (100)$ ,  $g_2 = (101)$ ,  $g_3 = (111)$

Code bits:  $\underline{x} = (x_1^{(1)}, x_2^{(1)}, x_3^{(1)}, x_1^{(2)}, x_2^{(2)}, x_3^{(2)}, \dots, x_1^{(m)}, x_2^{(m)}, x_3^{(m)})$

Observe: We obtain  $\underline{x}$  simply by interleaving the convolutions of  $\underline{s}$  with each of the generators:  $\underline{x}_1 = \underline{s} * g_1$

$$\underline{x}_2 = \underline{s} * g_2$$

$$\underline{x}_3 = \underline{s} * g_3$$

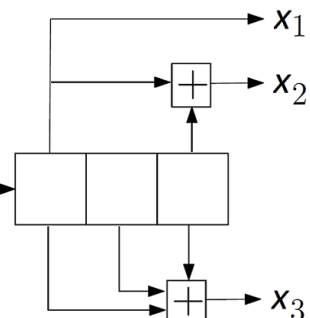
Alternatively: We can express this in terms of the Z-transform (with  $Z = z^{-1}$ ):

$$g_1(Z) = 1$$

$$g_2(Z) = 1 + Z^2$$

$$g_3(Z) = 1 + Z + Z^2$$

so that:  $x(Z) = s(Z^3)g_1(Z^3) + Zs(Z^3)g_2(Z^3) + Z^2s(Z^3)g_3(Z^3)$



12 / 21

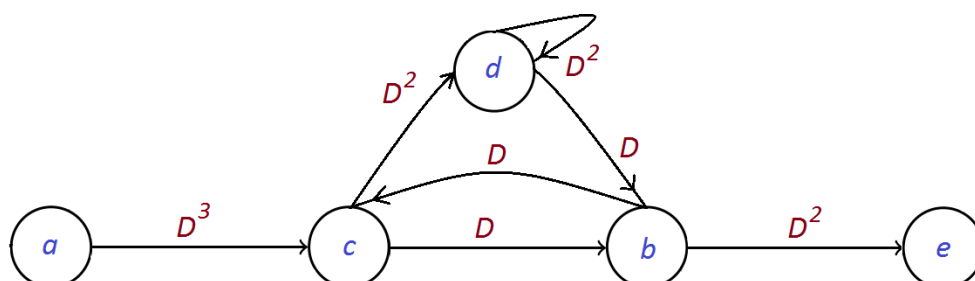
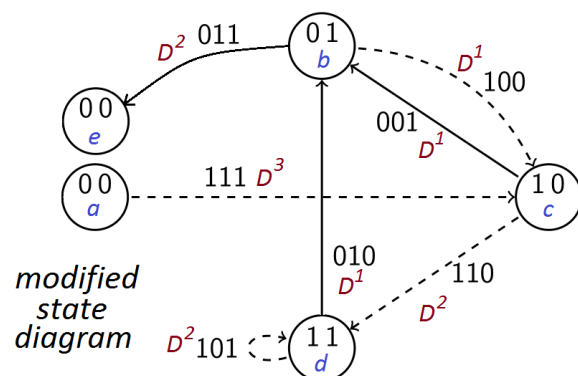
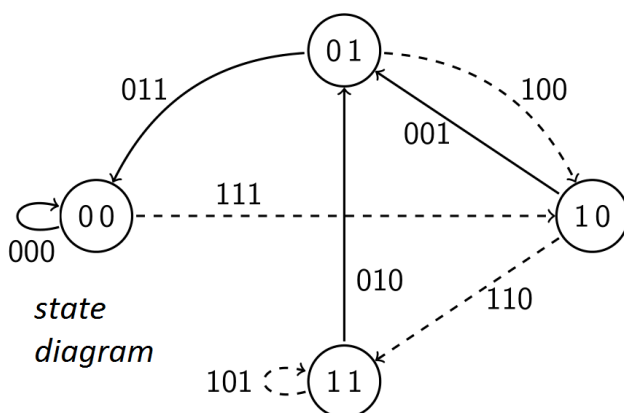
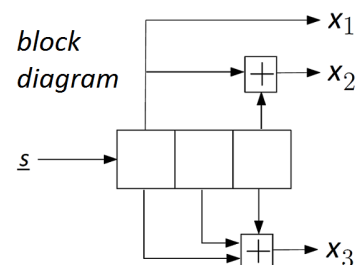
## Performance of convolutional codes

- Since convolutional codes are linear codes  
 $d_{min}$  = the minimum weight among nonzero codewords  
 = minimum distance between any codeword  $\underline{x}$  and  $\underline{0}$
- But since convolutional codes do not have a fixed blocklength we do not examine their minimum distance  $d_{min}$
- Instead we will consider the **free distance  $d_{free}$**  defined as *the min weight among all codewords generated with the code starting and ending in the all-zero state*
- *Interpretation.* If  $d_{free} = d$ , then any collection of  $\leq \lfloor \frac{d-1}{2} \rfloor$  errors can be corrected, as long as these “error bursts” are not “too close” to one another
- Simplest to compute  $d_{free}$  via the code's **transfer function**
- To define the transfer function, first modify the state diagram:
  - (i) simplify state labels
  - (ii) duplicate the all-zero state (as start and end)
  - (iii) label each transition by  $D$  raised to the Hamming weight of the sequence it produces

13 / 21

## Modified state diagram

- (i) simplify state labels, (ii) duplicate all-zero state
- (iii) label each transition by  $D$  raised to the Hamming weight of the sequence it produces

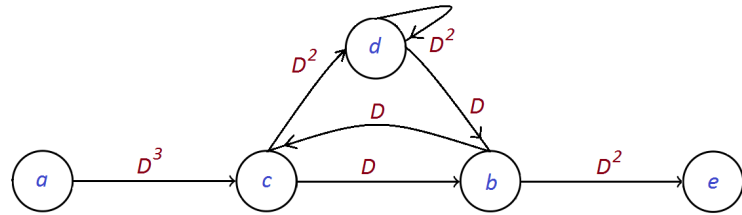


14 / 21

## State equations and transfer function

From the diagram we obtain the **state equations**

$$\begin{aligned} X_b &= DX_c + DX_d \\ X_c &= D^3X_a + DX_b \\ X_d &= D^2X_c + DX_d \\ X_e &= D^2X_b \end{aligned}$$



We define the **transfer function**:  $T(D) = X_e/X_a$

Solving this system:

$$T(D) = D^6/(1 - 2D^2) = D^6 + 2D^8 + 4D^{10} + 8D^{12} + \dots$$

*Interpretation*

The first term in  $T(D) \Rightarrow$  there is a single path of weight  $d = 6$  starting and ending in state 00; from the state diagram we see it is  $acbe$

Second term  $\Rightarrow$  there are exactly two paths from 00 to 00 of weight  $d = 8$ :  $acdbe$  and  $acbcbe$

Third term  $\Rightarrow$  there are four paths of weight  $d = 10$ , etc.

- In particular,  $d_{\text{free}} = 6$

15 / 21

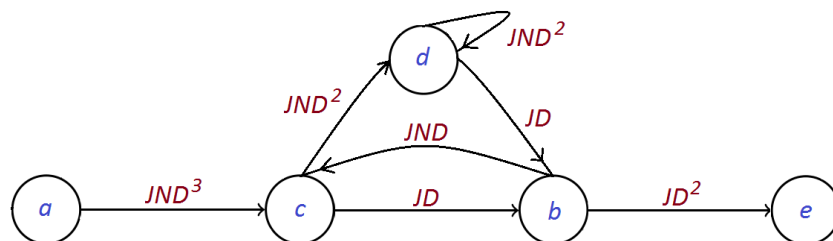
## The extended transfer function

In fact, it is easy to similarly get more information than only on the Hamming weight of paths

Extend the state diagram further:

Add a “ $J$ ” to every branch, counting total path length

Add an “ $N$ ” to each branch corresponding to an input bit “1”



Obtain from the diagram the **extended state equations**

$$\begin{aligned} X_b &= JDX_c + JDX_d \\ X_c &= JND^3X_a + JNDX_b \\ X_d &= JND^2X_c + JND^2X_d \\ X_e &= JD^2X_b \end{aligned}$$

Define the **extended transfer function**  $T(J, N, D) = X_e/X_a$

16 / 21



## Extended transfer function: Interpretation

Again, solving the above system:

$$\begin{aligned} T(J, N, D) &= \frac{J^3 N D^6}{1 - J N D^2 (1 + J)} \\ &= J^3 N D^6 + [J^4 N^2 D^8 + J^5 N^2 D^8] \\ &\quad + [J^5 N^3 D^{10} + 2J^6 N^3 D^{10} + J^7 N^3 D^{10}] + \dots \end{aligned}$$

- First term: There is exactly one path of Hamming weight  $d = 6$ , it has length 3, & of its 3 source bits exactly one is a 1
- Second and third terms: There are two paths of Hamming weight  $d = 8$ , one of length 4, and one of length 5, and the information sequence of each path contains two 1s
- Note: If we are transmitting a length- $m$  source sequence  $\underline{s} = (s_1, \dots, s_m)$ , we can truncate the expansion of the extended transfer function after all the terms of order  $J^m$

17 / 21

## Warning: Catastrophic encoders

Consider the encoder shown, with input

$$\underline{s} = 11111111 \dots$$

The code sequence generated is

$$\underline{x} = 11 \text{ 01 00 00 00 00 00 00 } \dots$$

Suppose the received sequence is

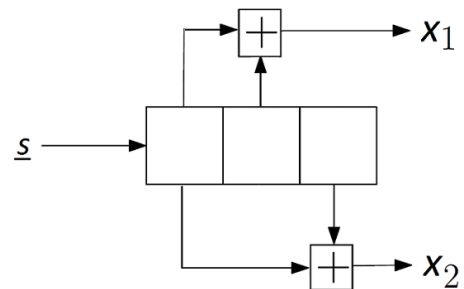
$$\underline{y} = 11 \text{ 10 01 00 00 00 00 00 } \dots$$

This is itself a codeword, corresponding to source string

$$\underline{s}' = 100000 \dots$$

⇒ A finite number of channel errors (only three, in fact) led to infinitely many decoding errors!

- Encoders with this very undesirable property are called *catastrophic*
  - The root of the problem is that we can go from state 11 to state 11, with an input bit of weight 1 and an output string 00 of weight 0



18 / 21

## Avoiding catastrophic encoders

Suppose an encoder of an  $S$ -stage shift register  
has generators  $\{g_1, g_2, \dots, g_n\}$   
with corresponding  $Z$ -transforms (again with  $Z = z^1$ ):  
 $\{g_1(Z), g_2(Z), \dots, g_n(Z)\}$

### Theorem [Massey & Sain, 1968]

If the greatest common divisor of the polynomials  
 $\{g_1(Z), g_2(Z), \dots, g_n(Z)\}$  is of the form  $Z^\ell$   
for some integer  $\ell \geq 0$ , then the encoder defined by the generators  
 $\{g_1, g_2, \dots, g_n\}$  is *not* catastrophic

*Exercise.* Show that the encoder of the previous slide  
does not satisfy the conditions of the theorem

19 / 21

## Summary: Convolutional codes

- Information bits are fed into an  $S$ -stage shift register
- Code bits are linear combinations of the contents of the register
- The code can be represented by a block diagram, a state diagram, or a trellis diagram
- In each representation there are  $2^{S-1}$  states, each state representing the bits in the first  $S - 1$  cells of the register
- Each state transition caused by an input bit is labeled with a sequence of code bits
- Efficient minimum distance decoding via Viterbi algorithm: Complexity scales *linearly* with number of input bits
- Code properties are summarized in the transfer function
- Error correcting ability described by the free distance instead of the minimum distance of linear codes

20 / 21

## Convolutional codes: Applications and extensions

- Convolutional codes are very widely used, e.g., in digital video, wireless communications including today's popular wireless standards (such as 802.11), and satellite links
- *Turbo codes*, in which two convolutional codes cleverly interact with one another, are the state-of-the-art in mobile communication systems
- The same ideas easily generalize to non-binary codes

- \* The exact same process of Viterbi decoding generalizes for transmission over Gaussian noise channels:

Map  $x$ 's as  $1 \mapsto +1$  and  $0 \mapsto -1$ , so that the channel output is  $Y_i = x_i + Z_i$ , where  $Z_i \sim N(0, N)$  is independent noise.

Then maximum likelihood decoding becomes minimum distance decoding wrt Euclidean instead of Hamming distance

**Ex.** In this case, repeat Exercise 3 ii) in Examples Paper 3, with  $\underline{Y} = (0.2, -1, 1.9, 1.35, -0.1, 0.4, -1.7, 0.6, 0.8, 3, 2.1, -0.4)$ .