

Part IIA Paper 3C6 Ex-Sheet 2 Solutions

$$1. T = \frac{1}{2} m (\dot{x}^2 + \dot{y}^2) \quad \text{so } M = \begin{bmatrix} m & 0 \\ 0 & m \end{bmatrix}$$

$$V = \frac{1}{2} k_1 x^2 + \frac{1}{2} k_2 y^2 + \frac{1}{2} s (x-y)^2$$

$$\text{so } K = \begin{bmatrix} k_1+s & -s \\ -s & k_2+s \end{bmatrix}$$

$$\text{So eigenvalue equation is } \begin{vmatrix} k_1+s-\omega^2 m & -s \\ -s & k_2+s-\omega^2 m \end{vmatrix} = 0$$

$$\text{Define } \omega_1^2 = \frac{k_1+s}{m}, \quad \omega_2^2 = \frac{k_2+s}{m}, \quad \Omega^2 = \frac{s}{m},$$

$$\text{then } \begin{vmatrix} \omega_1^2 - \omega^2 & -\Omega^2 \\ -\Omega^2 & \omega_2^2 - \omega^2 \end{vmatrix} = 0$$

$$\text{So } (\omega^2 - \omega_1^2)(\omega^2 - \omega_2^2) - \Omega^4 = 0$$

$$\therefore \omega^4 - \omega^2(\omega_1^2 + \omega_2^2) + \omega_1^2 \omega_2^2 - \Omega^4 = 0$$

$$\therefore \omega^2 = \frac{1}{2} \left\{ \omega_1^2 + \omega_2^2 \pm \sqrt{(\omega_1^2 + \omega_2^2)^2 - 4\omega_1^2 \omega_2^2 + 4\Omega^4} \right\}$$

$$= \frac{1}{2} \left\{ \omega_1^2 + \omega_2^2 \pm \sqrt{(\omega_1^2 - \omega_2^2)^2 + 4\Omega^4} \right\}$$

$$\text{So (i) } \omega_1^2 = \omega_2^2 \Rightarrow \omega^2 = \omega_1^2 \pm \Omega^2$$

$$\text{(ii) } (\omega_1^2 - \omega_2^2)^2 \gg 4\Omega^4 \Rightarrow \omega^2 \approx \omega_1^2, \omega_2^2$$

$$\text{Mode shapes: } (k_1+s)x - sy = \omega^2 mx$$

$$\therefore \omega_1^2 x - \Omega^2 y = \omega^2 x$$

$$\therefore \frac{y}{x} = \frac{\omega_1^2 - \omega^2}{\Omega^2}$$

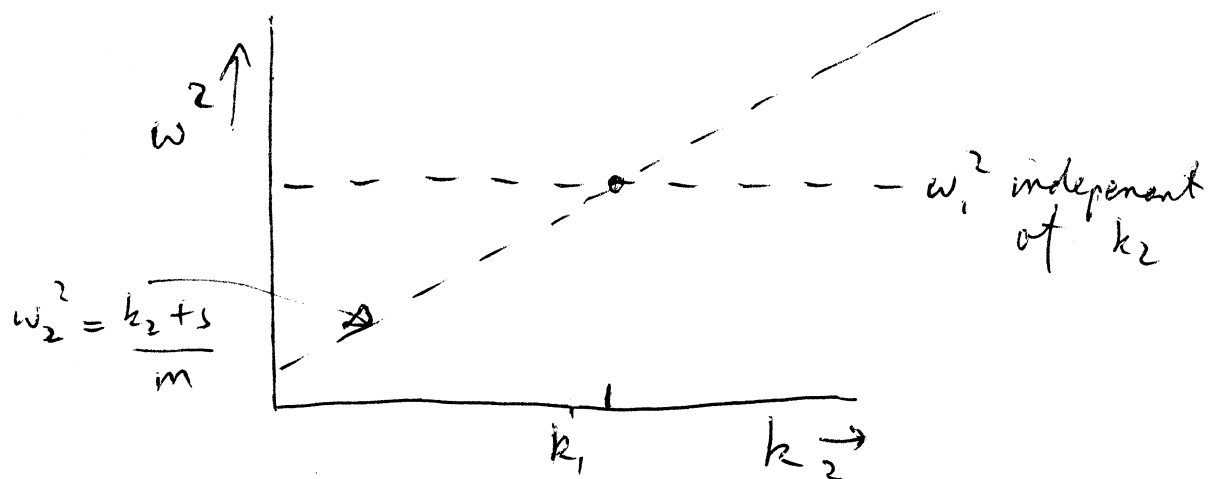
$$\text{So (i) modes are } \frac{y}{x} = \mp 1 \quad \text{as } \omega^2 = \omega_1^2 \pm \Omega^2$$

$$\text{(ii) } \omega^2 = \omega_1^2 \rightarrow \frac{y}{x} = 0, \quad \omega^2 = \omega_2^2 \rightarrow \frac{y}{x} \rightarrow \infty$$

$$\text{Now } \frac{\omega_1^2}{\omega_2^2} = \frac{k_1+s}{k_2+s} \approx \frac{k_1}{k_2}$$

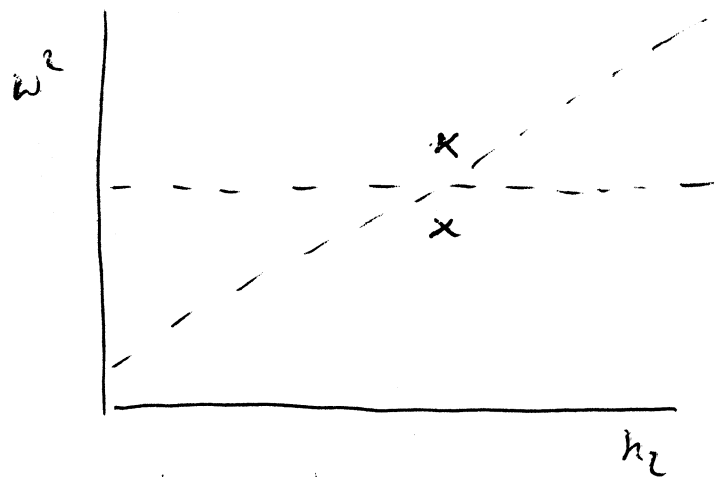
To sketch the curves:

First draw in the lines corresponding to case (ii)

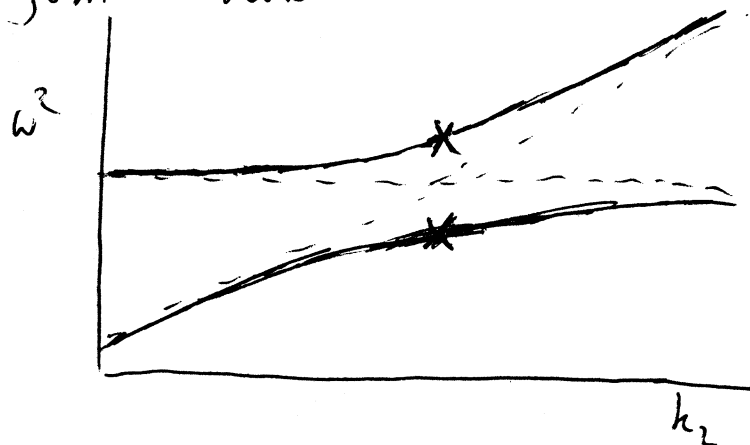


Expect the true graph to follow these lines except near the crossing point at $k_1 = k_2$.

The solution to case (i) shows that the true solutions lie above and below this crossing point:

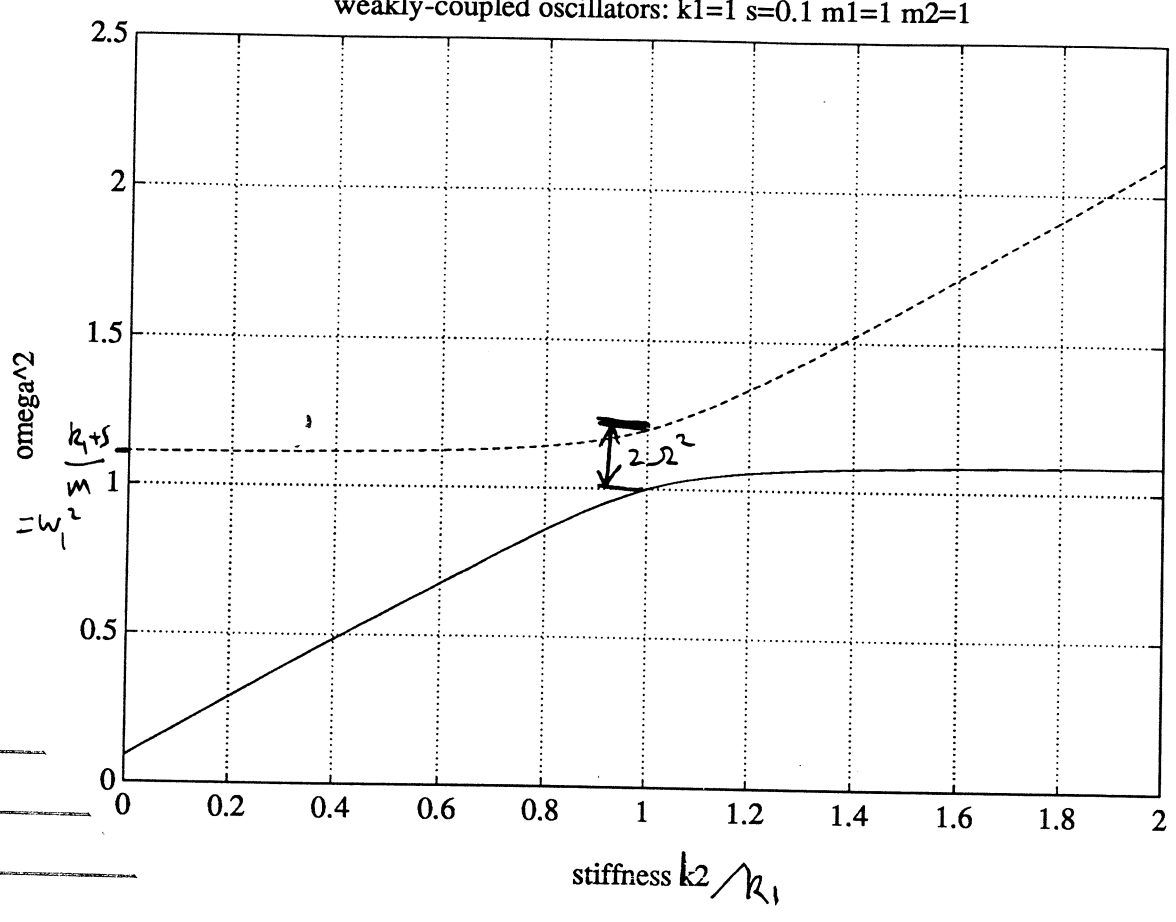


So now "join the dots"



Cont. Correct computed version:

weakly-coupled oscillators: $k_1=1$ $s=0.1$ $m_1=1$ $m_2=1$



Modes
Upper: $\begin{bmatrix} 1 \\ 0 \end{bmatrix}$
Lower: $\begin{bmatrix} 0 \\ 1 \end{bmatrix}$

Modes
 $\begin{bmatrix} 1 \\ -1 \end{bmatrix}$
 $\begin{bmatrix} 1 \\ 1 \end{bmatrix}$

Modes
 $\begin{bmatrix} 0 \\ 1 \end{bmatrix}$
 $\begin{bmatrix} 1 \\ 0 \end{bmatrix}$

Note they have swapped

If dissipation occurs in the spring s , the pattern of damping factors will be:

Upper: { Both the
Lower: { same

High
Zero

{ Both the
same

$$2. T = \frac{1}{2} m \dot{x}_1^2 + \frac{1}{2} m \dot{x}_2^2 + \dots + \frac{1}{2} m \dot{x}_N^2$$

$$\therefore M = \begin{bmatrix} m & & & \\ & m & & \\ & & m & \\ & & & n \end{bmatrix}$$

$$V = \frac{1}{2} k x_1^2 + \frac{1}{2} k x_2^2 + \dots + \frac{1}{2} s (x_1 - x_2)^2 + \frac{1}{2} s (x_2 - x_3)^2 + \dots$$

(leaf springs)
(coupling springs)

$$\therefore K = \begin{bmatrix} k+2s & -s & & -s \\ -s & k+2s & -s & \\ \vdots & & \ddots & \\ -s & & & k+2s \end{bmatrix}$$

Try $x_j = \cos \lambda j$.

The system is circular, so this only makes sense if $\cos \lambda N = 1 = \cos \lambda 0$ so that it joins up.
i.e. $\lambda = \frac{2n\pi}{N}$, $n = 0, 1, 2, \dots$ are possible values.

Now substitute in $K \underline{x} = \omega^2 M \underline{x}$.

Since the system is so symmetric, we only need to look at one typical row: e.g. j th row.

Then $-s \cos \lambda (j-1) + (k+2s) \cos \lambda j - s \cos \lambda (j+1) = \omega^2 m \cos \lambda j$

But $\cos \lambda (j-1) + \cos \lambda (j+1) = 2 \cos \lambda j \cos \lambda$ (D. Booth)

So a factor $\cos \lambda j$ cancels all through, leaving

$$-2s \cos \lambda + (k+2s) = \omega^2 m$$

$$\therefore \omega^2 = \frac{k}{m} + \frac{2s}{m} (1 - \cos \lambda) \quad (1)$$

Doesn't depend on j , so if this equation is satisfied, all rows of $K \underline{x} = \omega^2 M \underline{x}$ are satisfied.

$\therefore \underline{x}$ is a mode provided $\lambda = \frac{2n\pi}{N}$, with

natural frequency given by (1).

2 cont.

How many different modes can be found this way?

Since $\cos(\lambda_j) = -\cos(\lambda_j)$, and $\cos(\lambda_j + 2\pi) = \cos(\lambda_j)$,
the only distinct vectors arise from
 $\lambda = 0, \frac{2\pi}{N}, \frac{4\pi}{N}, \dots, \frac{2\pi}{N} \left(\frac{N}{2}\right)$

↑ or $\frac{N-1}{2}$ if N odd

This only gives roughly $N/2$ different modes. However, for nearly all these values of λ , as well as a mode $\cos \lambda_j$, there is also a mode $\sin \lambda_j$ having the same natural frequency. In total, there are always exactly N modes of this kind.

For N even:

$$\begin{cases} \lambda = 0 & 1 \text{ mode} \\ \lambda = \frac{2\pi}{N} & 2 \text{ modes} \\ \lambda = \frac{4\pi}{N} & 2 \text{ modes} \\ \vdots & \\ \lambda = \pi & 1 \text{ mode} \end{cases}$$

For N odd:

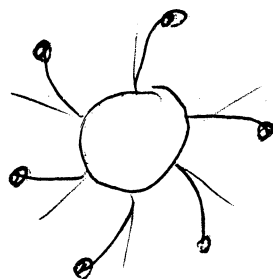
$$\begin{cases} \lambda = 0 & 1 \text{ mode} \\ \lambda = \frac{2\pi}{N} & 2 \text{ modes} \\ \vdots & \\ \lambda = \left(\frac{N-1}{N}\right)\pi & 2 \text{ modes} \end{cases}$$

So all natural frequencies satisfy ①, so they all lie between $\omega = \sqrt{\frac{k}{m}}$ and $\omega = \sqrt{\frac{k + k_s}{m}}$

2 Cont.

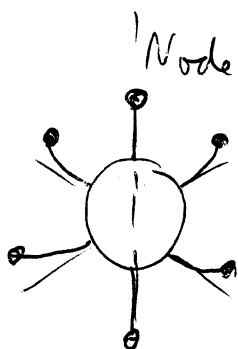
For $N=6$, the modes are:

No nodal
lines

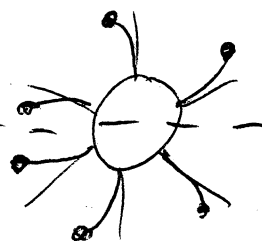


$$\lambda = 0$$

One
nodal
line

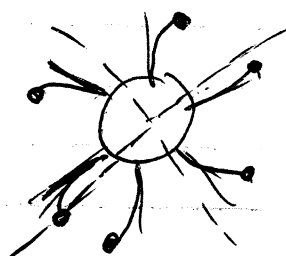
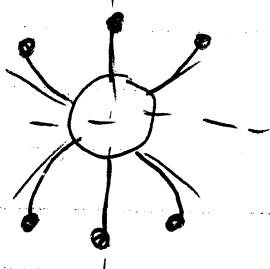


Node
line



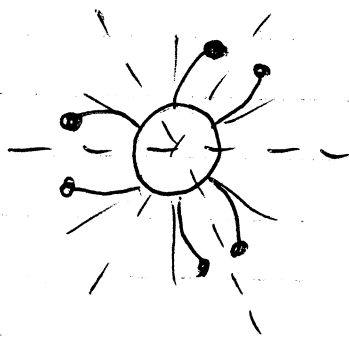
$$\lambda = \pi/3$$

2
nodal
lines



$$\lambda = 2\pi/3$$

3
nodal
lines



$$\lambda = \pi$$

$$3. \quad y = a_1 \frac{x}{L} + a_2 \frac{x^2}{L^2} + a_3 \frac{x^3}{L^3}$$

At $x=0$, $y=0$ already.

$$\text{At } x=L, y=0 \Rightarrow a_1 + a_2 + a_3 = 0$$

$$\text{So } y = a_1 \left(\frac{x}{L} - \frac{x^3}{L^3} \right) + a_2 \left(\frac{x^2}{L^2} - \frac{x^3}{L^3} \right)$$

$$\text{Now } V = \frac{1}{2} P \int_0^L y'^2 dx = \frac{1}{2} P \int_0^L \left[a_1 \left(1 - \frac{3x^2}{L^2} \right) + a_2 \left(\frac{2x}{L} - \frac{3x^2}{L^2} \right) \right]^2 dx$$

$$= \frac{1}{2} \frac{P}{L^2} \left[a_1^2 \int_0^L \left(1 - \frac{3x^2}{L^2} \right)^2 dx + 2a_1 a_2 \int_0^L \left(1 - \frac{3x^2}{L^2} \right) \left(\frac{2x}{L} - \frac{3x^2}{L^2} \right) dx + a_2^2 \int_0^L \left(\frac{2x}{L} - \frac{3x^2}{L^2} \right)^2 dx \right]$$

Multiply out tediously, do the integrals...

$$= \frac{1}{2} \frac{P}{L} \left[a_1^2 \cdot \frac{4}{5} + 2a_1 a_2 \cdot \frac{3}{10} + a_2^2 \cdot \frac{2}{15} \right]$$

$$\therefore K = \frac{P}{L} \begin{bmatrix} \frac{4}{5} & \frac{3}{10} \\ \frac{3}{10} & \frac{2}{15} \end{bmatrix}$$

$$T = \frac{1}{2} m \int_0^L \dot{y}^2 dx = \frac{1}{2} m \int_0^L \left[\dot{a}_1 \left(\frac{x}{L} - \frac{x^3}{L^3} \right)^2 + 2\dot{a}_1 \dot{a}_2 \left(\frac{x}{L} - \frac{x^3}{L^3} \right) \left(\frac{x^2}{L^2} - \frac{x^3}{L^3} \right) + \dot{a}_2^2 \left(\frac{x^2}{L^2} - \frac{x^3}{L^3} \right)^2 \right] dx$$

... more algebra ...

$$= \frac{1}{2} mL \left[\frac{8}{105} \dot{a}_1^2 + \frac{11}{420} 2\dot{a}_1 \dot{a}_2 + \frac{1}{105} \dot{a}_2^2 \right]$$

3. contd

$$\therefore M = \frac{mL}{420} \begin{bmatrix} 32 & 11 \\ 11 & 4 \end{bmatrix}$$

Now $K_u = \omega^2 M_u$

$$\rightarrow \frac{P}{30L} \begin{bmatrix} 24 & 9 \\ 9 & 4 \end{bmatrix} \begin{bmatrix} a_1 \\ a_2 \end{bmatrix} = \omega^2 \cdot \frac{mL}{420} \begin{bmatrix} 32 & 11 \\ 11 & 4 \end{bmatrix} \begin{bmatrix} a_1 \\ a_2 \end{bmatrix}$$

$$\therefore \frac{1}{30} \begin{bmatrix} 24 & 9 \\ 9 & 4 \end{bmatrix} \begin{bmatrix} a_1 \\ a_2 \end{bmatrix} = \Omega^2 \cdot \frac{1}{420} \begin{bmatrix} 32 & 11 \\ 11 & 4 \end{bmatrix} \begin{bmatrix} a_1 \\ a_2 \end{bmatrix}$$

with $\omega^2 = \frac{P}{mL^2} \Omega^2$

The Matlab code:

$$\begin{aligned} M &= (1/420) * [32 \ \ 11; 11 \ \ 4]; \\ K &= (1/30) * [24 \ \ 9; 9 \ \ 4]; \\ \text{eig}(K, M) \end{aligned}$$

will print the values of Ω^2 . Result is 10, 42
(Could also do this by hand, of course).

So the approximate natural frequencies satisfy

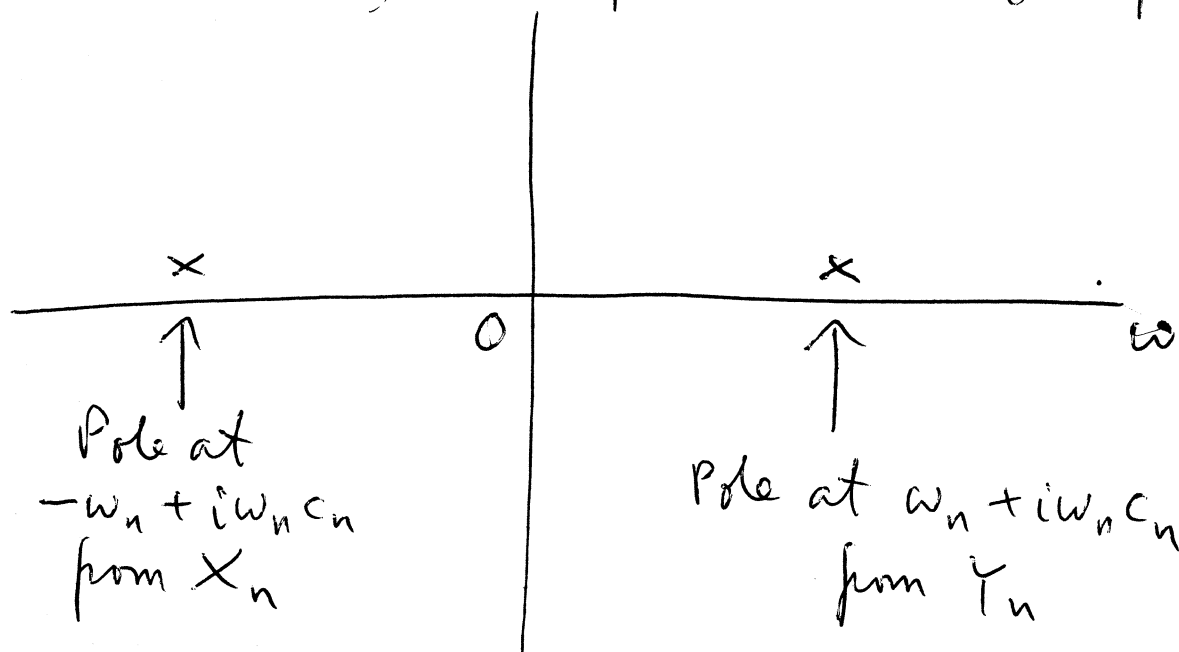
$$\omega^2 = \frac{P}{mL^2} \times \begin{cases} 10 \\ 42 \end{cases}$$

Exact answers are $\omega^2 = \frac{P}{mL^2} \begin{cases} \pi^2 \\ 4\pi^2 \end{cases} = \begin{cases} 9.87 \\ 39.5 \end{cases}$

So pretty close for a 2DOF approximation.

$$\begin{aligned}
 4(a) \quad Y_n + X_n &= \frac{b_n}{\omega - \omega_n(1 + i c_n)} - \frac{b_n}{\omega - \omega_n(-1 + i c_n)} \\
 &= \frac{2b_n \omega_n}{(\omega - \omega_n - i\omega_n c_n)(\omega + \omega_n - i\omega_n c_n)} \\
 &= \frac{2b_n \omega_n}{(\omega - i\omega_n c_n)^2 - \omega_n^2} \approx \frac{2b_n \omega_n}{\omega^2 - 2i\omega \omega_n c_n - \omega_n^2} \quad (|c_n|^2 \ll 1) \\
 &= H_n \quad \text{if} \quad b_n = -a_n / 2\omega_n
 \end{aligned}$$

Each of X_n, Y_n represents one pole of the transfer function. This division shows that occur in symmetric pairs in the ω Argand plane



For approximate behavior near the positive frequency ω_n , it will be OK to take Y_n only, and ignore X_n . The pole at negative frequency is further away than the other positive-frequency poles of the system, corresponding to other modes.

4 cont

10

(b) $|Y_n|$ at $\omega = \omega_n$ is $\frac{|b_n|}{\omega_n C_n}$

So want ω where $|Y_n| = \frac{|b_n|}{\sqrt{2} \omega_n C_n}$

$$\text{i.e. } \frac{b_n^2}{2\omega_n^2 C_n^2} = \frac{b_n^2}{(\omega - \omega_n)^2 + \omega_n^2 C_n^2}$$

$$\therefore (\omega - \omega_n)^2 = \omega_n^2 C_n^2$$

$$\therefore \omega = \omega_n (1 \pm C_n) \quad //$$

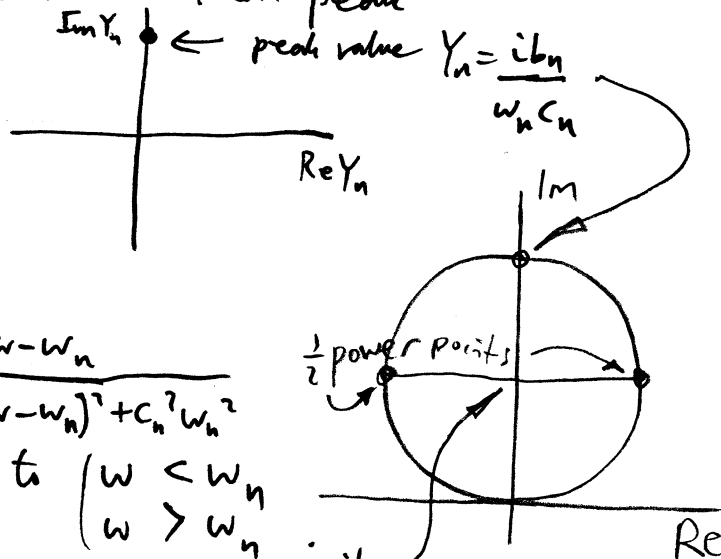
(c) Phase at peak: Y_n is pure imaginary, so phase is $\pm \pi/2$ depending on whether $b_n < 0$ or $b_n > 0$.

Phase of $Y_n = \pm \frac{\pi}{2} \pm \frac{\pi}{4}$ requires $\frac{\text{Re } Y_n}{\text{Im } Y_n} = \pm 1$

$$\text{i.e. } \frac{\omega - \omega_n}{\omega_n C_n} = \pm 1$$

$\rightarrow \omega = \omega_n \pm \omega_n C_n$, as in (a).

(d) Think of $b_n > 0$, for definiteness. Then peak value has phase $\frac{\pi}{2}$:



As $\omega \rightarrow \infty$, $Y_n \rightarrow 0$.

$$\text{Expression for } \text{Re } Y_n = \frac{\omega - \omega_n}{(\omega - \omega_n)^2 + C_n^2 \omega_n^2}$$

is antisymmetric with respect to $\begin{cases} \omega < \omega_n \\ \omega > \omega_n \end{cases}$.

So expect centre of circle to be at $\frac{1}{2} \frac{ib_n}{\omega_n C_n}$.

$$\text{So want } [\text{Re } Y_n]^2 + \left[\text{Im } Y_n - \frac{b_n}{2\omega_n C_n} \right]^2 = \frac{1}{4\omega_n^2 C_n^2}.$$

Substitute the expressions, and it comes out immediately.

5 Begin with the general solution for a freely vibrating beam : $y(z, t) = Y(z) e^{i\omega t}$
 $Y(z) = A \cos \alpha z + B \sin \alpha z + C \cosh \alpha z + D \sinh \alpha z$
 where $\alpha^4 = \frac{m\omega^2}{EI}$ (1)

The usual boundary conditions are :

pinned end : $Y=0$ and $\frac{d^2 Y}{dz^2} = 0$ (Moment)

clamped end : $Y=0$ and $\frac{dY}{dz} = 0$ (slope)

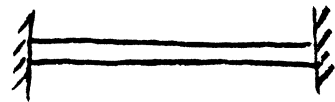
free end : $\frac{d^2 Y}{dz^2} = 0$ and $\frac{d^3 Y}{dz^3} = 0$ (shear)

$$Y'(z) = \alpha (-A \sin + B \cos + C \sinh + D \cosh)$$

$$Y''(z) = \alpha^2 (-A \cos - B \sin + C \cosh + D \sinh)$$

$$Y'''(z) = \alpha^3 (A \sin - B \cos + C \sinh + D \cosh)$$

Now solve for the special cases :

(i)  $z=L$
 $Y=0 \therefore A \cos + B \sin - A \cosh - B \sinh = 0$
 $Y'=0 \therefore -A \sin + B \cos - A \sinh - B \cosh = 0$
 $A = -C$
 $B = -D$
 (as for (i)) $\therefore \begin{vmatrix} \cos - \cosh & \sin - \sinh \\ -(\sin + \sinh) & \cos - \cosh \end{vmatrix} = 0$
 $\therefore (\cos - \cosh)^2 + (\sin + \sinh)(\sin - \sinh) = 0$

$$\therefore \cos^2 + \sin^2 - 2 \cos \cosh + \cosh^2 - \sinh^2 = 0$$

$$\therefore \cos \cosh = 1$$

$$\therefore \cos \alpha L = \operatorname{sech} \alpha L$$

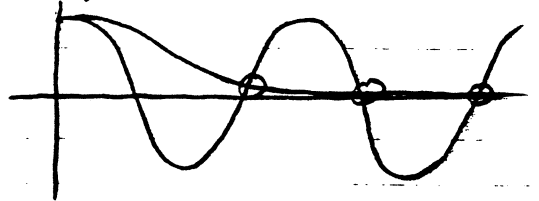
Approx solns

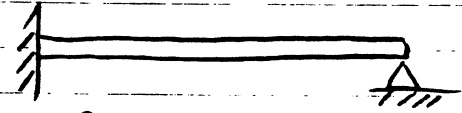
$$\alpha L = \frac{3\pi}{2}, \frac{5\pi}{2}, \frac{7\pi}{2} \dots$$

Exact

$$\alpha L = 4.730, 7.854, 10.996 \dots$$

Which gives natural frequencies as before

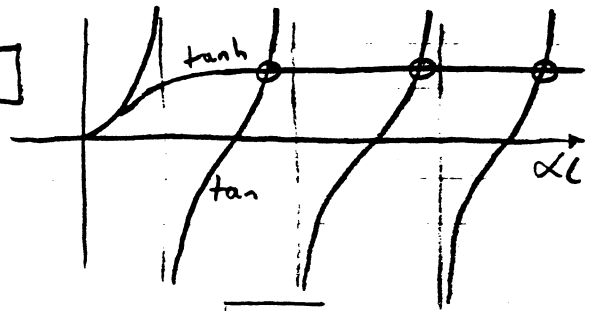


(ii)  $z=L$
 $Y=0 \therefore A \cos + B \sin - A \cosh - B \sinh = 0$
 $Y''=0 \therefore -A \cos - B \sin - A \cosh - B \sinh = 0$
 $A = -C$
 $B = -D$
 (as for (i)) $\therefore \begin{vmatrix} \cos - \cosh & \sin - \sinh \\ \cos + \cosh & \sin + \sinh \end{vmatrix} = 0$
 $\therefore \cos \sin - \cosh \sinh + \cos \sinh - \cosh \sin$
 $- (\cos \sin + \cosh \sinh - \cos \sinh - \cosh \sin) = 0$

5 contd:

$$\therefore \cosh \sin = \cos \sinh$$

$$\therefore \boxed{\tan \alpha L = \tanh \alpha L}$$



Approx:

$$\alpha L = \frac{5\pi}{4}, \frac{9\pi}{4}, \frac{13\pi}{4}$$

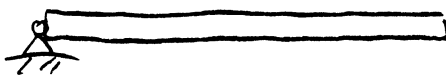
Exact

$$\alpha L = 3.927, 7.069, 10.210$$

and frequencies from

$$\omega = (\alpha L)^2 \sqrt{\frac{EI}{mL^4}}$$

(iv)



$$z=L$$

$$Y''=0 \therefore -B \sin + D \sinh = 0$$

$$Y'''=0 \therefore -B \cos + D \cosh = 0$$

$$\therefore \begin{vmatrix} \sin & -\sinh \\ \cos & -\cosh \end{vmatrix} = 0$$

$$\therefore \sin \cosh = \cos \sinh$$

$$\therefore \boxed{\tan \alpha L = \tanh \alpha L} \text{ as in (iv)}$$

$$\begin{aligned} Y=0 \quad Y''=0 \\ \therefore A=-C \\ \& \quad A=C \end{aligned} \therefore A=C=0$$

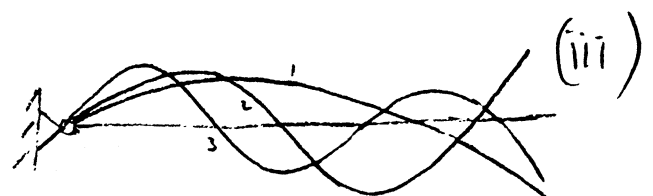
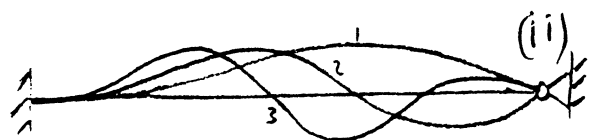
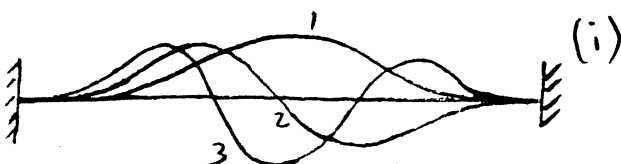
Note: in case (iii) a zero natural frequency is valid and it describes rigid body motions. This should be deduced by "common sense". Good students might check that the maths works this way too if α is not factorised in the early stages.

The free-free problem satisfies the DE, and $y''=y'''=0$ at both ends. Define now $w = \frac{\partial^2 y}{\partial z^2}$.

This satisfies the same DE (just take $\frac{\partial^2}{\partial z^2}$ (DE)),

but satisfies boundary conditions $w=w'=0$ so it is the solution of the clamped-clamped problem. So these have the same natural frequencies, and the clamped modes are $\frac{d^2}{dz^2}$ (free modes).

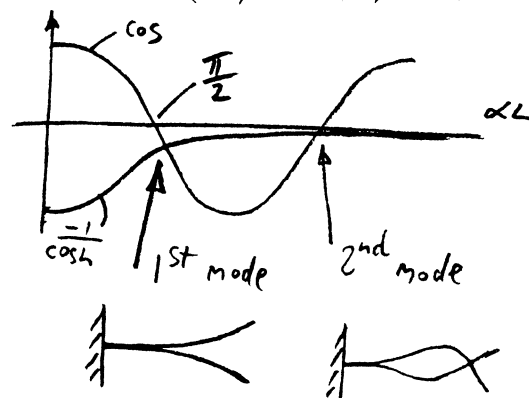
Mode shapes:



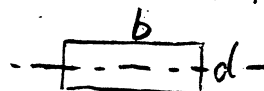
6. (i) $\omega_j = (\alpha L)_j^2 \sqrt{\frac{EI}{mL^4}}$

For 1st mode, the solution for αL is approximately $\frac{\pi}{2}$ but more accurately solved by iteration to give $\alpha L = 1.875$

where $\cos(\alpha L) \cosh(\alpha L) = -1$



and for a rectangular section

$I = \frac{bd^3}{12}$ 

and $m = \rho b d$

so $f_1 = \frac{1}{2\pi} \omega_1 = \frac{1}{2\pi} (1.875)^2 \sqrt{\frac{E b d^3}{12 \rho b d L^4}} = \underline{\underline{0.162 \frac{dc}{L^2}}}$

where $c = \sqrt{\frac{E}{\rho}}$

(ii) For all steels, $E \approx 200 \text{ GPa}$ $\rho = 7800 \text{ kg/m}^3$
 $\therefore c = \sqrt{\frac{200 \times 10^9}{7800}} = 5064 \text{ m/s}$ which is at the top end of the range

We have $f_1 = 230 \text{ Hz}$ and $L = 0.03 \text{ m}$

$\therefore d = \frac{230 \times .03^2}{.162 \times 5064} = 0.25 \text{ mm}$

(iii) Each semitone corresponds to a frequency ratio of $\sqrt[12]{2}$ (since 12 semitones is an octave)
 $= 1.0595$

and $\frac{230}{220} = 1.045$ which is just less than one semitone so the tweezers sound a slightly flat Bb.

7(a)



From structure data book, equivalent stiffness at centre of beam is $k = \frac{48EI}{L^3}$

$$\text{So } \omega^2 = \frac{k}{M} = \frac{48EI}{ML^3}$$

From lecture notes, answer for continuous beam is $\omega^2 = \frac{EI}{m} \left(\frac{\pi}{L}\right)^4$

So to get same ω , need $M = \frac{48}{\pi^4} mL = 0.493 mL$

(b)



Stiffness here is $k = \frac{3EI}{L^3}$

$$\text{So } \omega^2 = \frac{k}{M} = \frac{3EI}{ML^3}$$

Continuous beam has $\omega^2 = \frac{EI}{m} \alpha^4$ with $\alpha = \frac{1.875}{L}$

So to get the same frequency, need

$$M = \frac{3}{(1.875)^4} mL = 0.243 mL$$

8. (a) For simply support beam, mode shapes are
 $u_j(x) = \sin \frac{j\pi x}{L}$

So want $\int_0^L \rho A \sin \frac{j\pi x}{L} \sin \frac{k\pi x}{L} dx$

$$= \frac{\rho A}{2} \int_0^L \left[\cos \frac{(j-k)\pi x}{L} - \cos \frac{(j+k)\pi x}{L} \right] dx$$

$$= \frac{\rho A}{2} \left[\frac{L}{\pi(j-k)} \sin \frac{(j-k)\pi x}{L} - \frac{L}{\pi(j+k)} \sin \frac{(j+k)\pi x}{L} \right]_0^L$$

$$= 0 \quad \text{if } j \neq k.$$

(Fourier series orthogonality: see 1A Maths notes)

(b) Equations are
$$\begin{cases} EI \frac{d^4 u_j}{dx^4} = m\omega_j^2 u_j \\ EI \frac{d^4 u_k}{dx^4} = m\omega_k^2 u_k \end{cases}$$

So
$$\begin{cases} EI \int \frac{d^4 u_j}{dx^4} u_k dx = m\omega_j^2 \int u_j u_k dx \\ EI \int \frac{d^4 u_k}{dx^4} u_j dx = m\omega_k^2 \int u_j u_k dx \end{cases}$$

Integrate by parts twice:

$$EI \left\{ \left[\frac{d^3 u_j}{dx^3} u_k \right]_0^L - \left[\frac{d^2 u_j}{dx^2} \frac{du_k}{dx} \right]_0^L + \int_0^L \frac{d^2 u_j}{dx^2} \frac{d^2 u_k}{dx^2} dx \right\} = m\omega_j^2 \int u_j u_k dx$$

plus the corresponding result with $j \leftrightarrow k$.

Subtract:

$$m(\omega_j^2 - \omega_k^2) \int u_j u_k dx = EI \left\{ [u_j''' u_k] - [u_j'' u_k'] - [u_k''' u_j] + [u_k'' u_j'] \right\}$$

8 cont.

Boundary conditions:

Clamped

$u = 0$

$u' = 0$

Pinned

$u = 0$

$u'' = 0$

Free

$u''' = 0$

$u'' = 0$

So always, either $u = 0$ or $u''' = 0$
 and also either $u' = 0$ or $u'' = 0$

So for any combination of these boundary conditions, all four $\left[\text{---} \right]_0^L$ terms

vanish, and so $\int_0^L u_j u_k dx = 0$ as required.