

# 3F4: Data Transmission

## Handout 3: Power spectral density of the PAM signal

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Recall the transmitted baseband PAM signal is  $x(t) = \sum_k X_k p(t - kT)$ .

- This is a *random signal*, so its power spectral density (PSD) describes the average power in any frequency band
- Also recall from 3F1 that integrating the PSD gives the average power of the signal
- We may want the PSD to be close zero outside the frequency band supported by the channel. Sometimes regulations also impose that the PSD lie below a given frequency-domain “mask”

To calculate the PSD, we will consider a slightly modified PAM signal:

$$x(t) = \sum_{k=-\infty}^{\infty} X_k p(t - kT - \Theta),$$

where  $\Theta$  is a random dither (delay) uniformly distributed in  $[0, T)$ .

- The dither models the fact that an observer measuring the PSD has no information about when the transmitter’s clock was switched on.
- Importantly, we will see that the dither makes  $x(t)$  a wide-sense stationary (WSS) process, which greatly simplifies our PSD derivation.

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## Assumptions on the $\{X_k\}$ symbols

$$x(t) = \sum_{k=-\infty}^{\infty} X_k p(t - kT - \Theta),$$

- We assume that the  $X_k$ 's are drawn from a constellation with zero mean, e.g.,  $\{-3A, -A, A, 3A\}$
- Furthermore, we also assume that the random process  $\{X_k\}_{k=-\infty}^{\infty}$  is a WSS discrete-time process. What this means is that the autocovariance between  $X_j$  and  $X_{j+i}$  depends only on the shift  $i$ :

$$\mathbb{E}[(X_{j+i} - \mathbb{E}[X_{j+i}]) (X_j - \mathbb{E}[X_j])] = \mathbb{E}[X_{j+i} X_j] =: R_X[i]$$

- In fact, if we assume that the input bits being transmitted are uniformly random, then the  $X_k$ 's will also be independent, i.e.,

$$R_X[i] = \begin{cases} \mathbb{E}X_k^2 \text{ (avg. energy of a constellation symbol)}, & i = 0 \\ 0, & i \neq 0 \end{cases}$$

- But our derivation will be for general  $R_X[i]$ . In Ex. paper 1, you will see a transmission scheme where  $X_k$ 's are not independent

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## Autocovariance of $x(t)$

$$x(t) = \sum_{k=-\infty}^{\infty} X_k p(t - kT - \Theta),$$

Since  $\{X_k\}$  have zero mean, so does  $x(t)$ . Therefore the *autocovariance function* of  $x(t)$  is

$$\begin{aligned} R_x(t + \tau, t) &:= \mathbb{E}[x(t + \tau)x(t)] \\ &= \mathbb{E}\left[\sum_{k=-\infty}^{\infty} X_k p(t + \tau - kT - \Theta) \sum_{\ell=-\infty}^{\infty} X_\ell p(t - \ell T - \Theta)\right] \\ &= \sum_{k=-\infty}^{\infty} \sum_{\ell=-\infty}^{\infty} \mathbb{E}[X_k X_\ell p(t + \tau - kT - \Theta)p(t - \ell T - \Theta)] \end{aligned}$$

Note that the expectation is over  $(X_k, X_\ell)$  as well as over  $\Theta \sim \text{unif}[0, T)$ . Since  $\theta$  and the symbols  $\{X_k\}$  are independent,

$$\begin{aligned} R_x(t + \tau, t) &= \sum_{k=-\infty}^{\infty} \sum_{\ell=-\infty}^{\infty} \mathbb{E}[X_k X_\ell] \mathbb{E}[p(t + \tau - kT - \Theta)p(t - \ell T - \Theta)] \\ &= \sum_{k=-\infty}^{\infty} \sum_{\ell=-\infty}^{\infty} R_X[k - \ell] \frac{1}{T} \int_0^T p(t + \tau - kT - \theta)p(t - \ell T - \theta)d\theta \end{aligned}$$

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Making the change of variable  $k - \ell = j$ , we obtain

$$\begin{aligned} R_x(t + \tau, t) &= \sum_{j=-\infty}^{\infty} R_X[j] \sum_{k=-\infty}^{\infty} \frac{1}{T} \int_0^T p(t + \tau - kT - \theta) p(t - kT + jT - \theta) d\theta \\ &= \sum_{j=-\infty}^{\infty} R_X[j] \frac{1}{T} \int_{-\infty}^{\infty} p(t + \tau - \theta) p(t + jT - \theta) d\theta, \end{aligned}$$

where the last equality above is obtained using the following fact: For any function  $f(y)$  and any interval  $[0, T]$ ,

$$\sum_{k=-\infty}^{\infty} \int_0^T f(y - kT) dy = \int_{-\infty}^{\infty} f(y) dy.$$

(The LHS integrates the entire function by adding up the integrals over disjoint length  $T$  intervals.)

Making a change of variable in the integral  $v = t + \tau - \theta$ , we get

$$R_x(t + \tau, t) = \sum_{k=-\infty}^{\infty} R_X[j] \frac{1}{T} \int_{-\infty}^{\infty} p(v) p(v + jT - \tau) dv$$

Defining  $R_p(\tau) = \int_{-\infty}^{\infty} p(v) p(v - \tau) dv$ , we can summarize our result as

$$R_x(t + \tau, t) := \mathbb{E}[x(t + \tau)x(t)] = \frac{1}{T} \sum_{j=-\infty}^{\infty} R_X[j] R_p(\tau - jT)$$

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## PSD of $x(t)$

Observe that the RHS of  $R_x(t + \tau, t)$  depends only on  $\tau$ , not on  $t$ . Hence  $x(t)$  is WSS, and we write

$$R_x(t + \tau, t) = R_x(\tau) = \frac{1}{T} \sum_{j=-\infty}^{\infty} R_X[k] R_p(\tau - kT) \quad (1)$$

Fourier transform of  $R_x(\tau)$  will give us power spectral density of  $x(t)$ .

Note:

- $R_p(\tau) = \int_{-\infty}^{\infty} p(v) p(v - \tau) dv = p(\tau) \star p(-\tau)$ .
- Therefore,  $\mathcal{F}[R_p(\tau)] = P(f)P(-f) = |P(f)|^2$  because  $p(t)$  being real implies  $P^*(f) = P(-f)$

Using these, we take Fourier transforms in (1) to get the PSD:

$$S_x(f) = \frac{1}{T} \sum_{k=-\infty}^{\infty} R_X[k] \mathcal{F}[R_p(\tau - kT)] = \frac{1}{T} \sum_{k=-\infty}^{\infty} R_X[k] |P(f)|^2 e^{-j2\pi kfT}$$

Therefore the PSD of the transmitted PAM signal  $x(t)$  is

$$S_x(f) = \frac{|P(f)|^2}{T} \sum_{k=-\infty}^{\infty} R_X[k] e^{-j2\pi kfT}$$

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## When the $\{X_k\}$ are independent

In most (not all) applications, the  $\{X_k\}$  symbols are independent as independent sets of bits are mapped to successive constellation symbols. In this case,

$$R_X[k] = \begin{cases} \mathbb{E}[X_j^2] = \mathcal{E}_s, & k = 0 \\ 0, & k \neq 0 \end{cases}$$

where  $\mathcal{E}_s$  denotes the average energy per constellation symbol.

Then the formulas for autocovariance and PSD of  $x(t)$  simplify to

$$R_x(\tau) = \frac{\mathcal{E}_s}{T} R_p(\tau), \quad S_x(f) = \frac{\mathcal{E}_s}{T} |P(f)|^2$$

The average power of the PAM waveform is then calculated as

$$\frac{\mathcal{E}_s}{T} \int_{-\infty}^{\infty} |P(f)|^2 df = \frac{\mathcal{E}_s}{T} \int_{-\infty}^{\infty} |p(t)|^2 dt, \quad (2)$$

where we have used Parseval's theorem (see IB Signal & Data Analysis).

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Thus, when the  $\{X_k\}$  are independent:

- The expression in (2) is just what we would expect: in each symbol period of time  $T$ , we transmit a shift of pulse  $p(t)$  modulated by a symbol whose average squared value is  $\mathcal{E}_s$ .
- If the pulse  $p(t)$  has unit energy, then power of PAM signal is just  $\mathcal{E}_s/T$ .
- If the matched receive filter is chosen as  $q(t) = p(-t)$ , then the overall filter frequency response is

$$G(f) = P(f)P(-f) = |P(f)|^2.$$

Then the PSD on the previous slide can be written as

$$S_x(f) = \frac{\mathcal{E}_s}{T} G(f).$$

When the  $\{X_k\}$  are not independent, the PSD is given by the general expression on slide 7.

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