



40 years of sensor fusion for orientation tracking via magnetic and inertial measurement units: Methods, lessons learned, and future challenges

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ABSTRACT

Technological developments over the past two decades have resulted in the development of more accurate and lightweight low-cost magnetic and inertial measurement units (MIMUs). These developments have allowed the extensive application of MIMUs in various fields, specifically tracking the 3D orientation of a rigid body. Despite recent technological improvements, measurements from a tri-axial gyroscope, accelerometer, and/or magnetometer inside the MIMU are characterized by uncertainties. Numerous studies have been conducted to address these uncertainties and develop sensor fusion algorithms (SFAs) to estimate the 3D orientation accurately and robustly. This paper contributes to these efforts by providing a survey of the state-of-the-art SFAs for orientation estimation. We surveyed +250 publications, categorized the SFAs with various structures, identified the modifications proposed to improve their performance, and discussed the strengths and weaknesses of these approaches. We found that, while early SFAs were mostly a vector observation algorithm or an extended Kalman filter, to improve the computational efficiency, more recent works have developed SFAs with a complementary filter or complementary Kalman filter structure. At the same time, to improve the performance of the SFAs, several research teams have proposed various modifications to the basic structure of these filters, such as adaptive gain tuning or imperfect measurement rejection. We also provided an outlook on the lessons learned as well as the main challenges related to SFAs and discussed the practical steps toward developing an effective SFA. We have identified the need for benchmarking studies as the main challenge at the moment. This paper is among the first surveys which provide such breadth of coverage across different SFAs for tracking orientation with MIMUs.

1. Introduction

Magnetic and inertial measurement units (MIMUs), comprised of a tri-axial accelerometer, gyroscope, and magnetometer, are used to track the displacement and orientation of a rigid body in real-time. Because of their lightweight, small size, and long battery life [1,2], MIMUs have been used extensively as an ideal tool in aerospace, unmanned vehicle navigation, robotics, and human motion tracking. However, in almost all applications, whether the 3D joint angle measurement in an ambulatory human motion tracking system [3] or displacement estimation in a dead-reckoning system [4], the MIMU orientation must be first calculated using a sensor fusion algorithm (SFA), also known as attitude and heading reference system (AHRS).

Raw MIMU data can be used to estimate orientation under specific conditions. In particular, the MIMU's accelerometer measures the

gravitational acceleration and can be used to compute the attitude. Also, in a magnetically neutral environment, the magnetometer measures the geomagnetic field, which can be used to estimate the yaw angle. Furthermore, the gyroscope measures the angular velocity (rate of change of orientation), which can be used to calculate the change in orientation using numerical strap-down integration (SDI).

However, using the accelerometer, gyroscope, and magnetometer alone may yield poor estimations in terms of accuracy or robustness due to various sources of error [5]. For example, accelerometers are not suitable for orientation estimation during dynamic tasks as they measure the external non-gravitational acceleration (due to motion) in addition to the gravitational acceleration. Also, the geomagnetic field could be distorted by the presence of ferrous materials, specifically during indoor motion tracking, which makes magnetometer-based estimations inaccurate. Moreover, because the cumulative error of the SDI increases unbounded over time, MEMS gyroscopes are not suitable for orientation

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Nomenclature

AHRS	Attitude and heading reference system.
DCM	Direction cosine matrix.
FQA	Factored quaternion algorithm.
GDA	Gradient descent algorithm.
GNA	Gauss-Newton algorithm.
L/E/CKF	Linear/extended/complementary Kalman filter.
L/NCF	Linear/non-linear complementary filter.
MIMU	Magnetic and inertial measurement unit.
PSO	Particle swarm optimization.
QUEST	QUaternion ESTimator.
SDI	Numerical strap-down integration.
SFA	Sensor fusion algorithm.
TRIAD	TRi-axial attitude determination.
VO	Vector observation.
w.r.t.	With respect to

estimation during long-duration tasks. Thus, various SFAs have been proposed in the literature to achieve an accurate and reliable estimation.

1.1. Orientation parametrization

Rigid body orientation can be expressed via one of the following parametrizations: (i) Euler angles, i.e., roll and pitch (also known as the attitude) and yaw (also known as the heading); (ii) quaternions; or (iii) direction cosine matrix (DCM) [6,7]. Euler angles can be used when an intuitive physical meaning of the estimated orientation is needed. For example, in controlling an unmanned vehicle [8], measuring human joint angles, or tracking in an augmented reality device [9]. However, under certain configurations, there are singularities associated with this parametrization that make Euler angles unsuitable for most control or tracking applications.

Quaternion or DCM parametrization could be used to express/calculate the orientation without singularity in a computationally-efficient manner [10,11]. However, the unconstrained estimation of these parametrizations would lead to an ambiguous and meaningless representation. In particular, a *quaternion parametrization of orientation* is an element of the quaternion group, i.e., homeomorphic to rotation group $SO(3)$, with S^3 as its domain [12]. Therefore, to reduce its degrees of freedom from four to three, consistent with the dimension of the $SO(3)$, a constrained quaternion estimation is required, such as enforcing the unit norm constraint after estimation. See more on constrained estimation in [12,13]. A similar hard constraint must be used when using DCM parametrization [14,15].

1.2. Previous surveys on SFAs

Previous surveys have reviewed the literature related to MIMUs with a focus on applications [16–21], technical developments [4,17,22,23], and/or experimental comparison. However, in line with the aim of the present paper, in this section, we review the previous surveys of SFAs for MIMU orientation estimation only. Table 1 summarizes the surveys with a focus on the experimental comparison. Table 1 identifies the filter with the best performance in each survey and provides the details of the SFAs tested, parameter tuning strategies, and experimental validations.

Cavallo et al. [24] compared the estimation accuracy of three SFAs, their proposed Extended Kalman Filter (EKF) and SFAs in [25,26], using a MIMU mounted on KUKA Youbot robot while the accelerometer and magnetometer were calibrated using a 3D-ellipsoid fitting method [27]. The robot measured the reference (true) orientation during slow (18 deg/s) and fast (45 deg/s) trajectory tracking motions. Filippeschi et al. [10] performed an experimental study to compare five SFAs [28–32] for

Table 1

Review of the previous literature surveys of the sensor fusion algorithms, including the tested algorithms, the gain selection procedure, the used reference system for validation, the type experiments, and the algorithm(s) concluded to have the best performance. Abbreviations used in the table are described in the table footnotes.

Study	Sensor Fusion Algorithms	Filter Parameters	Reference	Experiments (duration in seconds)	Selected Filter
[24]	PEKF, [25, 26]	EC	Robot [182]	Slow & fast motions of robotic arm (≤ 45 s)	PEKF
[10]	[28–32]	EC	MCS	Elbow, forearm & shoulder functional tasks (≤ 15 s)	Acc.: [28] Corr.: [31]
[33]	[29,34–36]	EC	MCS	Random motions (≤ 20), Treadmill walking (≤ 30 s)	[36]
[37]	PEKF, KF	EC	MCS	Random head motions (≤ 35 s)	PEKF
[38]	[25,39–41]	EC	MCS	Static and dynamic tests in magnetically clean & disturbed environments (≤ 60 s)	[41]
[42]	SDI, [25, 37]	EC	MCS	Daily routine tasks (≤ 60 s), Walking (≤ 180 s)	[37]
[183]	[25,184]	EC	None	Single axis rotations (≤ 250 s)	[184]
[39]	[122,129]	EC	MCS	Daily routine tasks (≤ 60 s), Walking (≤ 180 s)	[129]
[43]	[25,37, 44], XKF	OP	MCS	Slow, medium & fast random rotations (≤ 70 s)	Slow/medium: XKF, Fast: [37]
[185]	[25,26, 143]	OP	MCS	Quadcopter motion adopted from [186]	[26]
[45]	[25,26, 46–49]	EC	MCS	Smartphone data during various activities (≤ 180 s)	[49]
[187]	[25,26]	EC	MCS	Various walking tests, adopted from [188]	[25]
[189]	[25,26,80, 190]	OP	Synthetic	-	[25,26]
[191]	[192], modified [127]	EC	MCS	Material handling task (≤ 60 s)	Modified [127]

- **Sensor Fusion Algorithms:** PEKF: Proposed Extended Kalman Filter by the study; XKF: Xesne proprietary Kalman Filter; SDI: Strap-down integration.

- **Filter Parameters:** EC: Selecting the sensor fusion algorithm's gain(s) by trial-and-error; OP: Selecting the sensor fusion algorithm's gain(s) by a rigorous search.

- **Reference:** MCS: Gold-standard camera motion-capture system. Synthetic: used synthetic MIMU data.

- **Selected Filter:** Acc: Selected based on accuracy; Corr: Selected based on correlation.

wrist position tracking. They used orientation and length of the trunk, upper arm and forearm were used to estimate the wrist position. However, the accuracy of the SFAs was not directly comparable for the following reasons: (i) one method required visual reference for position tracking; (ii) methods were different in terms of constraints of the kinematic chain; and (iii) parameters of the methods were not selected systematically.

Young [33] compared the EKF proposed in [29] with different Complementary Filters (CFs) presented in [34–36]. After calibrating sensors for offset and scale errors, the accuracy of the SFAs was evaluated during gentle random motions and over treadmill walking at a normal pace. They concluded that in general, CFs with lower computational complexity could outperform KFs for human motion tracking, as CFs have no assumption about the process dynamics. Also, they showed that because of the lower computational complexity and minimal loss of accuracy, the use of non-optimal vector observation (VO) algorithms, such as TRi-axial Attitude Determination (TRIAD), is preferred compared to optimal ones, such as the QUaternion ESTimator (QUEST). More recently, Sabitini [37] provided a general survey of SFAs, including algorithms based on VOs and KFs, with great emphasis on practical aspects such as SFA gain selection.

Fan et al. [38] performed a systematic review of standard strategies for reducing the effect of magnetic disturbance on the accuracy and robustness of SFAs. They also compared the accuracy of five SFAs, [25, 39–41], and concluded that a “good” filter must have the following three features: (i) gyroscope drift compensation; (ii) decoupling attitude from yaw estimation; and (iii) an adaptive strategy to reject magnetic distortions. Similarly, Ligorio and Sabatini [39] reviewed the main magnetic disturbance compensation strategies used in KFs and showed that model-based approaches (estimating the magnetic disturbance at each iteration using a stochastic model) had the best performance.

Bergamini et al. [42] used routine manual tasks such as teeth/hair brushing and walking along a ∞ -shaped pathway to compare the accuracy of SDI with the CF proposed by Madgwick et al. [25] and the KF proposed by Sabatini [37]. This study showed that the tested CF and KF were significantly more accurate than SDI for yaw estimation and that the two SFAs achieved similar accuracy. Using the optimal SFAs gains, Caruso et al. [43] analyzed the performance of four SFAs ([25, 37, 44], and Xesne proprietary KF) and concluded that the performance of SFAs depends on the experimental conditions, such as the rate of rotations during experiments. Michel et al. [45] compared the accuracy and robustness of six SFAs ([25, 26, 46–49]) using smartphone data during a variety of daily routine activities. They showed that the overall performance of the nonlinear CF proposed by Martin and Salaun [49] was best, but SFAs with a simpler structure and lower computational complexity, such as [25] and [26], could be highly beneficial for saving the battery life of the smartphone. See [50] for a more detailed study on the energy characterization of CF and KF.

However, the mentioned works only surveyed a limited number of SFAs, i.e., in total, 30 filters were tested in the abovementioned works, and only 4 of them were used in more than two studies. Therefore, in this paper, we reviewed a wide range of SFAs for orientation tracking with MIMUs, including VOs, CFs, Linear KFs (LKFs), EKFs, and Complementary KFs (CKFs). We did not, however, review fusing strategies such as the Unscented KF, Cubature KF, and Particle filter, or the use of other technologies for orientation tracking, whether alone or together with MIMUs. In summary, this review sought to answer the following question: how can MIMU signals be fused for orientation tracking? To answer this question, we provided a survey on state-of-the-art strategies, including SFAs based on VO, CF, and LKF/EKF/CKF families.

2. Literature survey

2.1. MIMU model

In this section, we present the MIMU model commonly used in

developing SFAs [51, 52]. Gyroscope readout, y_G , can be modelled as the summation of the true angular velocity, ω , the bias b_G , and a white noise term, v_G , as in (1a),

$$y_{G, k} = K_G \omega_k + b_{G, k} + v_{G, k} \quad (1a)$$

$$b_{G, k} = b_{G, k-1} + w_{b, k} \quad (1b)$$

where K_G is the scale factor matrix, and (1b) models b_G as a first-order Markov process driven by white Gaussian noise, w_b . Accelerometer readout, y_A , can be modelled as the summation of the external non-gravitational acceleration, α , the gravitational acceleration, g , the bias b_A , and a white noise term, v_A , as in (2a),

$$y_{A, k} = K_A [\alpha_k + g_k] + b_{A, k} + v_{A, k} \quad (2a)$$

$$\alpha_k = c_a \alpha_{k-1} + w_{\alpha, k} \quad (2b)$$

where K_A is the scale factor matrix, and (1b) models α as a first-order low-pass filtered ($0 \leq c_a < 1$ is the cut-off frequency of the filter) white Gaussian noise process. The bias term b_A was commonly obtained through a calibration procedure [27, 53] or estimated during orientation tracking as part of the KF state vector [54, 55]. Finally, the magnetometer readout, y_M , can be modelled as the summation of the true geomagnetic field, m , magnetic distortion, δ , and a white noise term, v_M , as in (3a),

$$y_{M, k} = K_M m_k + \delta_k + v_{M, k} \quad (3a)$$

$$\delta_k = c_d \delta_{k-1} + w_{\delta, k} \quad (3b)$$

where K_M is the scale factor matrix, and (3b) models δ as a first-order low-pass filtered ($0 \leq c_d < 1$ is the cut-off frequency of the filter) white Gaussian noise process.

2.2. Strap-down integration (SDI)

While SDI is not an SFA on its own, we dedicate a separate section to it as it is the core of almost all CFs and KFs. The numerical SDI of gyroscope readout, y_G , can be used to update the orientation with respect to (w.r.t.) a known orientation as in (4a) [56],

$$\begin{cases} q_{k+1} = \exp(\Omega(y_{G, k}) T_s) q_k \\ q_0 = q(0) \end{cases} \quad (4a)$$

$$\Omega(y_G) = \frac{1}{2} \begin{bmatrix} 0 & -y_{G,x} & -y_{G,y} & -y_{G,z} \\ y_{G,x} & 0 & y_{G,z} & -y_{G,y} \\ y_{G,y} & -y_{G,z} & 0 & y_{G,x} \\ y_{G,z} & y_{G,y} & -y_{G,x} & 0 \end{bmatrix} \quad (4b)$$

where q is the quaternion parametrization of orientation, $\Omega(y_G)$ is a 4×4 skew-symmetric matrix shown in (4b), $\exp(\cdot)$ is the matrix exponential operator, which can be estimated using the Taylor series or Padé approximation, T_s is the sampling period of the MIMU, and $q(0)$ is the known initial orientation. Lee and Choi [57] showed that when using the Taylor series to calculate $\exp(\cdot)$, the approximation order highly depends on (i) the sampling rate of the MIMU and (ii) the magnitude of the y_G . Equation (4a) can also be extended to account for the relative angular velocity of (i) the Earth's frame w.r.t. the inertial frame and (ii) the navigation frame w.r.t. the Earth's frame [58].

Nevertheless, SDI has two disadvantages: (i) the initial orientation must be known; and (ii) the gyroscope bias, b_G , results in an increasing cumulative error in the estimated orientation due to the numerical integration (see [59] for comparison of the drifts obtained from the SDI and a KF). To address the former, y_A and y_M , along with a VO algorithm (described in Section 2.3), can be used to calculate $q(0)$. To address the latter, various SFAs have been developed to use y_A and/or y_M to correct the SDI drift over time. Also, using SFAs, b_G can be estimated at each time instant based on a stochastic model, as in (1b), to correct the y_G

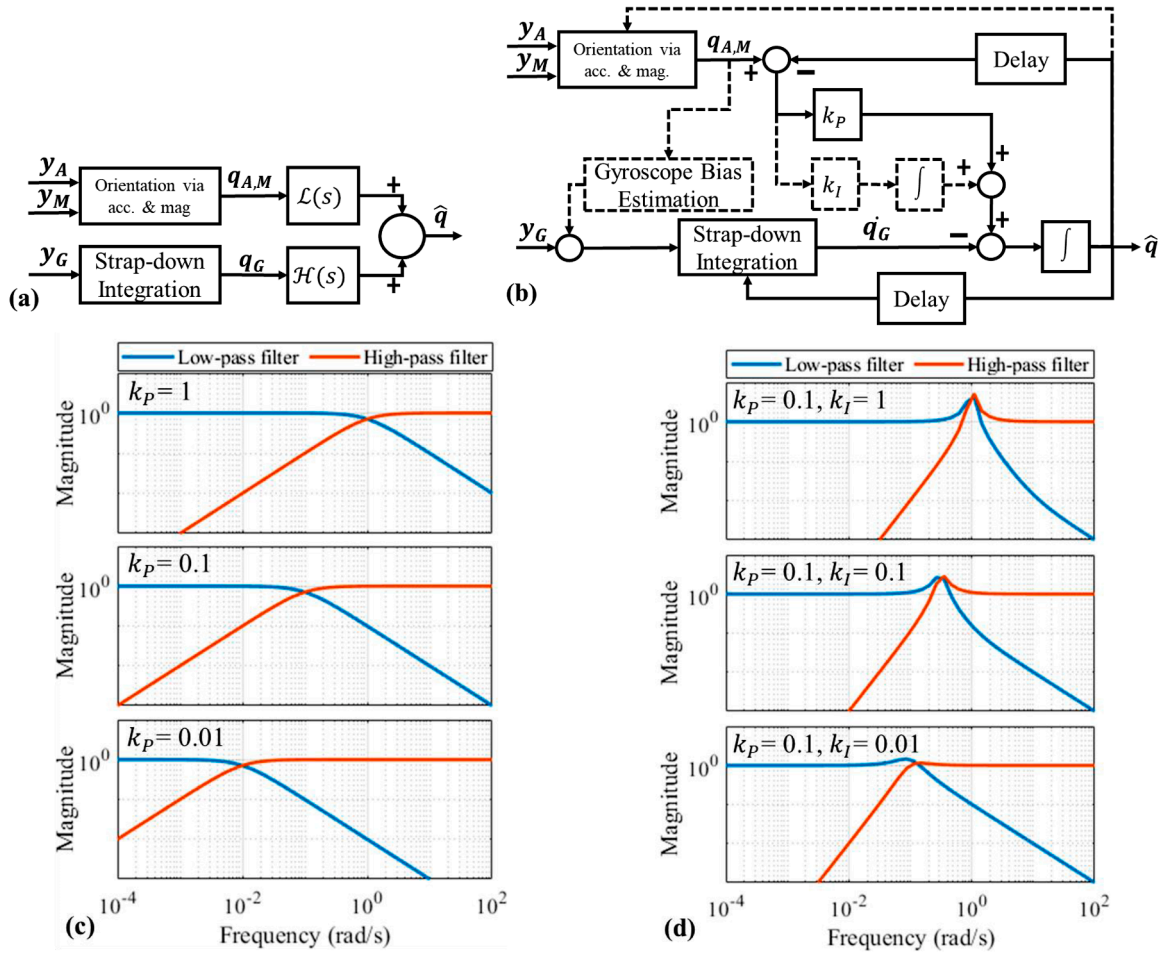


Fig. 1. Flowchart and frequency response of a Complimentary Filter (CF). (a) The general structure of a CF; (b) Flowchart of the CF with low-pass filter transfer function defined as $\mathcal{L}(s) = \frac{C(s)}{C(s)+s}$ where $C(s) = k_p + \frac{k_I}{s}$, and k_p and k_I are the proportional and integral gains, respectively; (c) Frequency response of the proportional CF, (d) Frequency response of the proportional-integral CF.

before the SDI. However, one must note that b_G is a function of environmental conditions, such as ambient temperature, and extra caution must be exercised when environmental conditions change drastically during data acquisition [60].

2.3. Vector observation (VO) algorithms

VO algorithms, commonly used in spacecraft AHRs, provide an estimate of the rigid body absolute orientation w.r.t. to a reference frame. To this end, two or more vectors measured in the local rigid body frame, as well as their counterpart in the desired reference frame, are required. In this setting, almost all VO algorithms are based on minimizing the cost function $\mathcal{J}(A) = \frac{1}{2} \sum_i a_i |b_i - A r_i|^2$, known as the Wahba's problem [61], where A is the DCM representing the rigid body orientation (to be estimated), b_i is a unit vector measured in local rigid body frame, r_i is the corresponding unit vector in the reference frame, and a_i is the weight associated with each VO. Markley and Mortari [62] provided an overview of the most popular algorithms for solving the Wahba's problem and compared their accuracy and speed.

For orientation tracking with MIMUs, VOs (b_i) are the normalized y_A and y_M , and their reference counterparts (r_i) are the gravitational acceleration, g , and the Earth's geomagnetic field vector, m , respectively, where G represents the Earth's reference frame. In this setting, the TRIAD algorithm calculates the DCM A using (5) [34,63],

$$w_1 = b_1/|b_1|, w_2 = (w_1 \times b_2)/|w_1 \times b_2|, w_3 = w_1 \times w_2 \quad (5a)$$

$$v_1 = r_1/|r_1|, v_2 = (v_1 \times r_2)/|v_1 \times r_2|, v_3 = v_1 \times v_2 \quad (5b)$$

$$A = w_1 v_1^T + w_2 v_2^T + w_3 v_3^T \quad (5c)$$

In TRIAD, attitude is calculated using g and is immune to magnetic disturbance. In contrast to TRIAD, QUEST [64] can accommodate more than two VOs by selecting proper weights a_i for sensors with different accuracies. However, in QUEST, magnetic disturbance can affect the attitude estimation as well as yaw angle.

Along with TRIAD, the Factored Quaternion Algorithm (FQA) [36] was designed to decouple y_A and y_M and cancel the effect of magnetic disturbance on attitude calculation. TRIAD and FQA produce similar solutions to the Wahba's problem, with the exception that the former produces a DCM and the latter produces a quaternion [36]. Other variants of VO algorithms include: Davenport q -method [65]; fast optimal attitude matrix [66]; singular value decomposition-based method [67]; filter QUEST [68]; recursive QUEST [69]; fast linear quaternion attitude estimator [70]; recursive linear continuous quaternion attitude estimator [71]; quaternion-based iterated least-square [72]; and algebraic quaternion algorithm [73,74]. The main advantage of the VO algorithms is that they can estimate the absolute orientation. However, as the VOs (y_A and y_M) could be corrupted by α and δ , these algorithms are not suitable for indoor orientation tracking during highly dynamic tasks.

2.4. Complementary filters

2.4.1. Foundations

For an accurate estimate of orientation, the accelerometer and magnetometer should be used during static conditions (low-frequency), while the gyroscope should be used during dynamic conditions (high-frequency). To do this, the CF structure includes a low-pass filter $\mathcal{L}(s)$ to remove the high-frequency estimations obtained by the accelerometer and magnetometer and a high-pass filter $\mathcal{H}(s)$ to remove the low-frequency estimations obtained by the gyroscope, as shown in Fig. 1 (a). Commonly, the two filters' structures have been selected such that $\mathcal{L}(s) = \frac{\mathcal{C}(s)}{\mathcal{C}(s)+s}$ and $\mathcal{L}(s) + \mathcal{H}(s) = 1$, where $\mathcal{C}(s)$ is a transfer function.

The simplest choice for $\mathcal{L}(s)$ and $\mathcal{H}(s)$ is a first-order low-pass and high-pass filter, respectively, with a cut-off frequency of k_p . To achieve this, we can set $\mathcal{C}(s) = k_p$ in $\mathcal{L}(s)$, which results in the basic CF structure shown in Fig. 1(b) with solid lines in the time domain. Fig 1(c) shows the frequency response of the basic CF for three different values of k_p . The optimal value of k_p depends on the dynamics of motion. That is, for relatively slow motions, a larger k_p is preferred while for high dynamics, a smaller k_p should be chosen [75]. As shown in Fig. 1(c), the slope of the frequency response plot of the low- and high-pass filters is not steep, and thus, the filter output contains components of both filters. To address this issue, and to achieve steeper slopes with frequency responses in Fig. 1(d), we can add an integrator with gain k_I to the basic CF structure, i.e., $\mathcal{C}(s) = k_p + \frac{k_I}{s}$, as shown with the dashed line in Fig. 1(b).

2.4.2. Literature survey

As the SDI was used in almost all CFs developed for MIMU orientation tracking, this section is focused on various approaches for (i) estimating the orientation via accelerometer and magnetometer and (ii) low and high-pass filters' structure. See Table 2 for a comparison of CFs in the literature. Bachmann et al. [76] used the Gauss-Newton Algorithm (GNA) to modify the SDI output based on a corrective term that minimized the error between the measured y_A and y_M and the projection of the Gg and Gm in MIMU's sensor frame, SF, as in (6),

$$q_{A,M} \left({}^{SF}\hat{q}_k \right) = [y_A, y_M] - \left[{}^{SF}\hat{q}_k^* \otimes {}^Gg \otimes {}^{SF}\hat{q}_k, {}^{SF}\hat{q}_k^* \otimes {}^Gm \otimes {}^{SF}\hat{q}_k \right] \quad (6)$$

where ${}^{SF}\hat{q}$ is the orientation of the MIMU sensor frame w.r.t. the Earth's reference frame, $*$ and \otimes are the quaternion conjugate and multiplication operations, respectively. Close variants of the same approach have been used in other works, including fast quaternion-based orientation optimizer [77]; Gradient Descent Algorithm (GDA)-based CF [25]; Levenberg-Marquardt algorithm-based CF [48,78,79]; GNA-based adaptive-gain CF [80]; geometrically-intuitive CF [81]; adaptive quaternion-based CF [82]; generalized linear quaternion-based CF [83]; and others [84–90].

The CF structure shown in Fig. 1(b) has been extensively revisited in the literature [26,75,91–95]. For instance, Calusdian et al. [75] used the proportional gain k_p to correct the SDI output using the weighted error between the estimated orientations by CF at the previous time instant and the FQA. In CF proposed by Lai et al. [91], first, attitude was estimated using y_A as in (7),

$$roll_{acc} = \tan^{-1}(y_{Ay} / y_{Az}), \quad pitch_{acc} = \tan^{-1}(y_{Ax} / \sqrt{y_{Ay}^2 + y_{Az}^2}) \quad (7)$$

then, the proportional-integral CF in Fig. 1(b) was employed to correct the SDI output. Similarly, Liu and Zhu [111] proposed a CF based on the proportional-integral and multi-sample rotation vector concepts to eliminate the cumulative and noncommutativity errors of SDI in high dynamics.

Mahony et al. [26] formulated the CF as a deterministic observer posed on the special orthogonal group $SO(3)$. The proposed explicit CF does not require an algebraic reconstruction of the attitude, as in VO

algorithms, and due to its low computational complexity, is ideal for embedded hardware platforms. Also, the explicit CF can work with only one VO (y_A or y_M). Similarly, Khosravian and Namvar [96] presented a nonlinear observer with asymptotic convergence based on a single VO (y_A or y_M). In general, a single VO can be used along with a thresholding scheme as in [88,93] (also known as vector selection [51,77]), or with fuzzy logic [97,98], to reject imperfect measurements, such as rejecting y_A during high dynamic tasks or y_M during magnetic disturbance.

The stability properties of the explicit CF proposed in [26] were proven when the reference vectors are (i) stationary or (ii) time-varying but the y_G is bias-free. Later, Grip et al. [99] proposed an improved explicit CF for cases when the reference vectors are time-varying and y_G is corrupted by bias. Also, Jensen [100] proposed a generalized version of the explicit CF by replacing the constant scalar gains by time-varying matrix gains. Because the gain tuning procedure for the explicit CF was based on trial-and-error, to provide a Kalman-like gain tuning capability for the filter, De Silva et al. [101] proposed the right-invariant formulation for the explicit CF with an intuitive gain tuning procedure based on the system's noise parameters.

Martin and Salaun [49] and Hua et al. [102] developed nonlinear observers by decoupling the yaw angle estimation from attitude estimation to minimize the effect of δ on attitude. Fan et al. [103] and Wu et al. [104] proposed quaternion-based two-step CFs (step 1: estimate attitude by fusing y_G and y_A , step 2: estimate yaw using y_M). This allowed them to make attitude estimations immune to δ and enable the filter to use two separate gains to minimize the effect of α and δ in steps 1 and 2, respectively. Valenti et al. [73] introduced a VO algorithm for solving the Wahba's problem without singularity and, accordingly, developed a quaternion-based CF with attitude estimation immune to δ . Marantos et al. [105] developed a CF based on a new analytical solution for Wahba's problem complemented with a gyroscope propagation and an adaptive gain regulation scheme to provide smooth orientation tracking in high dynamics. Yang et al. [106] formulated a linearly discrete dynamic model to relate the attitude to y_A and then developed a computationally efficient adaptive gain CF to combine the estimated attitude with SDI output.

Chang et al. [107] first developed a smooth sliding-mode observer for attitude estimation using y_A , and then improved the observer estimations by including measurement noise and biases [108]. El Hadri and Benallegue [109] proposed a nonlinear sliding-mode CF to achieve robust orientation tracking and b_G estimation under parametric uncertainties and modelling errors. Vasconcelos et al. [110] introduced a CF using Euler angle parametrization (stable for non-singular configurations) and b_G compensation and tuned the filter gains in the frequency domain to account for unmodeled disturbances found in the experimental setup. By estimating a virtual angular velocity for the rigid body using y_A and y_M , Tayebi et al. [111] developed a quaternion-based CF to track the orientation as well as an asymptotic estimate of b_G . Sheng and Zhang [112] proposed the application of a neural network-based proportional-integral-derivative controller for calculation of b_G based on the error between estimated orientation by SDI and a VO algorithm. In the nonlinear CF introduced by Wu et al. [113], the error between y_A and the projection of Gg in MIMU sensor frame was used to estimate the b_G and correct the angular velocity before SDI. However, this method does not provide an estimate of b_G when the MIMU is motionless or y_A is parallel to Gg .

2.4.3. Modified complementary filters

Because the performance of a CF highly depends on its gains, Alves Neto et al. [114] proposed a CF using Euler angle parametrization and adaptive gains with values proportional to $\exp(-||{}^Gg - ||y_A||)$. Kottath et al. [115] presented the application of an ensemble of linear CFs with different gains to minimize the effect of gain value on the estimated orientation. In the proposed ensemble, the weighting factor associated with each CF was adaptively calculated based on the difference between

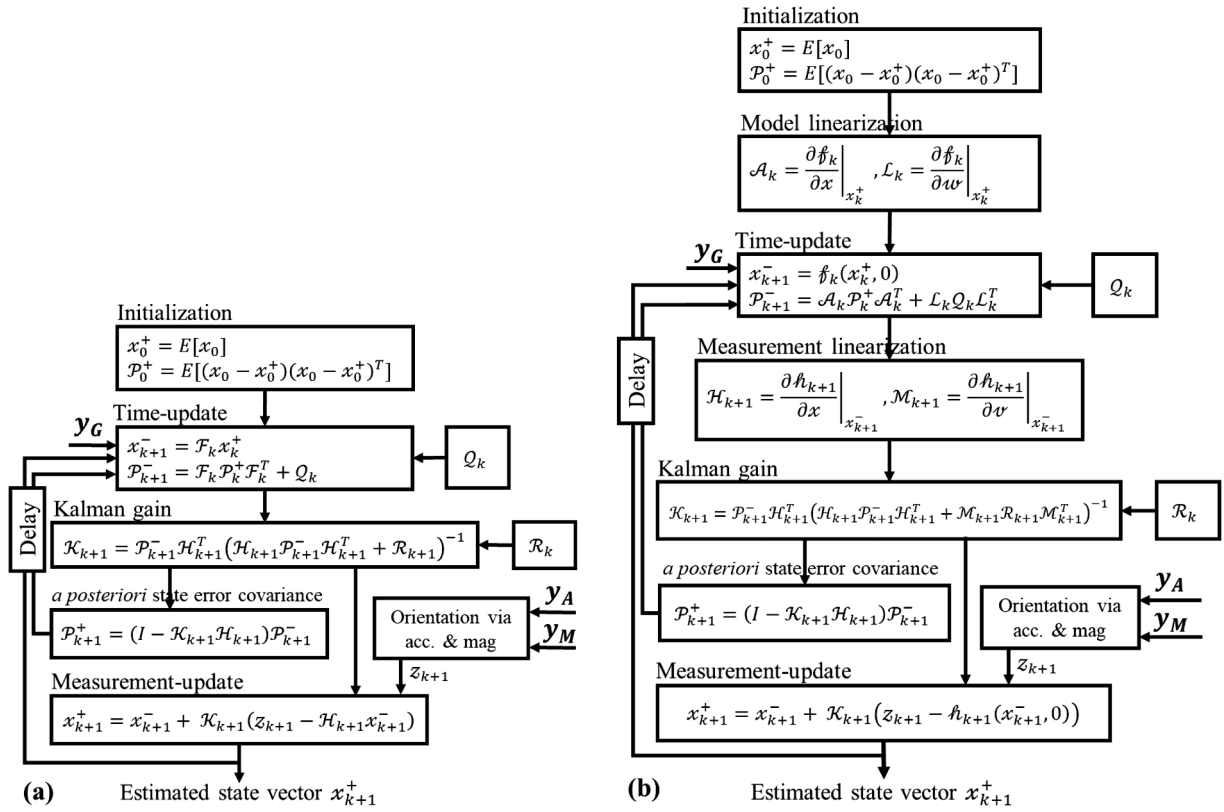


Fig. 2. Flowchart of a general (a) Linear Kalman Filter and (b) Extended Kalman Filter. \mathbf{x} is the state vector, \mathbf{y} is the measurement vector. \mathbf{P} is the state error covariance matrix. \mathcal{F} and $\mathbf{f}(\cdot)$ are the state transition matrix and equation, respectively. \mathcal{H} and $\mathbf{h}(\cdot)$ are the measurement prediction matrix and equation, respectively. \mathbf{Q} and \mathcal{R} are the system and measurement model covariance matrices, respectively. \mathcal{K} is the Kalman gain.

the CF's and the ensemble's estimates. To adaptively compute the CF gains based on the error between the orientation obtained from \mathbf{y}_A and \mathbf{y}_M and CF output, Poddar and Narkhede [116] developed a non-linear CF aided with a Particle Swarm Optimization (PSO) gain tuning scheme. It should be noted that the application of an adaptive gain tuning scheme or thresholding method for rejecting α and \mathbf{b} cannot guarantee an accurate and robust estimation. For example, if gain tuning is used for an extended period of time to cancel the estimation errors from \mathbf{y}_A during high dynamics, drift in attitude estimations will eventually occur because of \mathbf{b}_G .

2.5. Kalman filters

2.5.1. Foundations

Suppose that we describe the orientation tracking problem using linear or nonlinear discrete-time systems in (8a) and (8b), respectively,

$$\mathbf{x}_{k+1} = \mathcal{F}_k \mathbf{x}_k + \mathcal{B}_k \mathbf{u}_k + \mathbf{w}_k$$

$$\mathbf{z}_{k+1} = \mathcal{H}_{k+1} \mathbf{x}_{k+1} + \mathbf{v}_{k+1} \quad (8a)$$

$$\mathbf{x}_{k+1} = \mathbf{f}_k(\mathbf{x}_k, \mathbf{u}_k, \mathbf{w}_k)$$

$$\mathbf{z}_{k+1} = \mathbf{h}_{k+1}(\mathbf{x}_{k+1}, \mathbf{v}_{k+1}) \quad (8b)$$

where \mathbf{x} , \mathbf{u} , and \mathbf{z} are states of the system (commonly the MIMU orientation), input to the system (commonly \mathbf{y}_G), and sensor measurements (commonly \mathbf{y}_A and \mathbf{y}_M), respectively; \mathbf{w} and \mathbf{v} are white Gaussian noise processes associated with system and measurement models with covariances \mathbf{Q} and \mathcal{R} , respectively; \mathcal{F} and $\mathbf{f}(\cdot)$ are the state transition matrix and equation, respectively (also known as state propagation); and \mathcal{H} and $\mathbf{h}(\cdot)$ are the measurement prediction matrix and equation, respectively. Now, suppose that we show the estimation error, i.e., the

error between the true state and the estimated state, of the linear system in (8a) by $\tilde{\mathbf{x}}$ and our aim is to find the estimator that minimizes the weighted 2-norm of the expected value of $\tilde{\mathbf{x}}$, i.e., $\min E[\tilde{\mathbf{x}}^T \mathcal{T} \tilde{\mathbf{x}}]$, where \mathcal{T} is a positive definite weighting matrix. Then:

- If \mathbf{w}_k and \mathbf{v}_k are uncorrelated zero-mean white Gaussian processes, KF is the solution.
- If \mathbf{w}_k and \mathbf{v}_k are uncorrelated zero-mean white processes, KF is the best linear solution.
- For other cases, such as correlated/coloured noise or nonlinear system in (8b), the KF can be modified to answer the above problem [117].

In other words, the aim of using a KF is to estimate the state vector, \mathbf{x} , based on our knowledge of the system model and the availability of the noisy input, \mathbf{u} , and measurements, \mathbf{z} . To this end, we can use LKF and EKF, as shown in Fig. 2(a) and (b), to estimate the orientation using the linear and nonlinear systems in (8a) and (8b), respectively [117].

For orientation tracking applications, the rigid body orientation is commonly considered to be the state vector. Also, (i) the SDI is used to model the time-update equation, i.e., \mathcal{F} or $\mathbf{f}(\cdot)$; and (ii) the difference between the measured and estimated acceleration and geomagnetic field is used to model the measurement-update equation, i.e., \mathcal{H} or $\mathbf{h}(\cdot)$. More complex KFs have been introduced by including α , \mathbf{b}_G , and/or \mathbf{b} in the state vector (see Table 3 for a comparison of various KFs introduced in the literature).

2.5.2. Linear Kalman filters (LKF)

Barshan and Durrant-Whyte [118] used \mathbf{y}_G to build an LKF for estimating the yaw angle of a planar mobile robot. By inspection, they found an exponential model for \mathbf{b}_G during the warm-up period and showed that the inclusion of this model reduced the estimation error by a factor of 5.

Qi and Moore [8] proposed an LKF for GPS/MIMU data fusion, where the GPS readouts were taken as the LKF measurements, and MIMU readouts were taken as the additional information required for state propagation. Yun et al. [119] and Yean et al. [120] introduced LKFs where SDI was used to build the time-update equation and the error between y_A/y_M and the projection of ${}^Gg/{}^Gm$ in MIMU sensor frame, as in (6), along with GNA and GDA were used to construct the measurement-update equation. Whole and Gebhard [121] extended this approach by switching between calculating the gradient of y_A and y_M under homogenous magnetic condition to calculating the gradient of y_A only under magnetic disturbance.

Lee and Park [122] used the optimal two-observation quaternion estimation method proposed in [123] along with a vector selection technique to estimate the orientation from y_A and y_M under quasi-static motions in magnetically homogeneous conditions and correct the predicted orientation by SDI in an LKF structure. Seo et al. [124] proposed the application of the FQA to calculate the orientation using y_A and y_M and correct the SDI output in a standard LKF. They also suggested a correction scheme using quaternion linear interpolation to combine the LKF and FQA outputs during static conditions which would eliminate the effect of b_G . Similarly, Valenti et al. [74], Guo et al. [125], Feng et al. [40], and Wu et al. [71] developed new algebraic quaternion estimators using y_A and y_M to formulate the measurement-update equation of a standard LKF. Some of these algebraic quaternion estimators made the attitude estimation immune to magnetic disturbance while providing higher computational efficiency compared to FQA.

Kim and Golnaraghi [126] proposed a magnetometer-free LKF in which the angular velocity used in SDI and its bias were modelled as first-order Markov processes and included in the state vector. Also, the measurement-update equation was constructed using the error between (i) y_G and the estimated angular velocity, $\hat{\omega}$, plus bias and (ii) y_A and the projection of Gg in MIMU sensor frame. Also, Ligorio and Sabatini [127] developed a magnetometer-free LKF to estimate the attitude from the estimated gravitational acceleration, ${}^G\hat{g}$. However, these magnetometer-free LKFs limit the tracking to attitude only. Thus, other works [128–130] proposed two separate LKFs, or dual filtering, for estimating gravitational acceleration and geomagnetic field, to compute the 3D orientation while eliminating the errors in yaw estimation from those of attitude estimation. For instance, Zhu and Zhou [28] and Zhu et al. [131] used an LKF to estimate the gravitational acceleration, ${}^G\hat{g}$, and geomagnetic field, ${}^G\hat{m}$, by propagating the state vector using y_G in time-update step and then comparing the state vector with y_A and y_M during measurement-update. Batista et al. [132,133] extended the previous LKF structure by including b_G in the state vector and correcting the y_G accordingly during the time-update step.

Jurman et al. [134] constructed an adaptive LKF with Euler angles in its state vector. In this LKF, the state vector was propagated in time using SDI and the Euler angles calculated from y_A (using (7) for attitude) and y_M (using $\tan^{-1}(y_{M,y}/y_{M,x})$ for yaw) were used as the measurements. Sun et al. [135] included α with a first-order Markov process model in the state vector of their proposed LKF and used the modified acceleration (${}^Gg + \text{estimated } \alpha$) in (6) to build the measurement-update equation. In another approach, instead of an explicit modeling of α , Makni et al. [136,137] proposed an adaptive LKF where the part of \mathcal{R} associated with the accelerometer was estimated adaptively in real-time based on the LKFs accelerometer residual. Also, the proposed LKF was structured with energy consumption considerations in mind such that the gyroscope was turned-off and re-activated alternately while the \mathcal{Q} was adaptively tuned to compensate for the errors. Also, Rehbinder and Hu [138] proposed an SFA by switching between two LKFs similar to the high gain observer previously proposed by the same authors in [139], i. e., one for tracking under low accelerations and one for high accelerations.

2.5.3. Extended Kalman filters (EKF)

Lefferts et al. [140] investigated the effect of constructing an EKF with different variations of quaternions as the state vector and concluded that including all elements of the quaternions and b_G could lead to singularity in calculating \mathcal{Q} . Koifman and Merhav [141] proposed the application of an EKF for real-time attitude estimation using y_G and y_M , investigated the effect of piecewise constant modelling of b_G on estimation error, and then devised a method for updating the corresponding elements of the covariance matrix when required. Vaganay and Aldon [142] proposed an EKF in which the state vector included roll and pitch angles as well as their drift rates, while the SDI was performed out of EKF and as a part of the measurement-update.

Marins et al. [143] and Yun and Bachmann [29,144] investigated two approaches for EKF design using different measurement vectors: (i) using MIMU readouts as the measurement; and (ii) using the orientation (in quaternion) obtained by applying GNA or QUEST to y_A and y_M . They showed that in the former, the measurement-update equations were highly nonlinear functions of the state vector while in the latter, linear equations could be used to relate measurements to the state vector, which made the EKF suitable for real-time applications due to lower computational complexity. Also, Mazza et al. [145] introduced a similar magnetometer-free EKF but included b_G in the state vector. Sabatini [51] and Zhang et al. [146] included the accelerometer and magnetometer bias terms in the state vector of their proposed EKF and used their estimated values to correct the projection of Gg and Gm in MIMU sensor frame during calculation of the predicted measurements.

Also, Sabatini [147] developed a similar EKF by substituting the bias terms in the previous EKF by b_G and δ . Later, Sabatini [148] extended this work by introducing a variable-state-dimension EKF. This EKF switched between an EKF where a first-order Markov process was used to model δ and a higher-order EKF where a second-order Markov process modelled the time-rate of change of the magnetic field. This approach could make the propagation of state vector and calculation of the predicted measurements more realistic, particularly for long-duration tracking under magnetic disturbances. Following the same concept, Xu et al. [149] used a decision tree-based switching technique to develop an EKF capable of changing its measurement model between three modes (high y_G , high y_A , moderate y_A) to achieve robust tracking performance under different conditions.

2.5.4. Complementary Kalman filters (CKF)

While it is common to use the KF structure to estimate the primary states directly, some works proposed CKF structure, also known as the error-state KF [150,151], to estimate the errors in the primary states [6, 152]. By defining the state vector as an error process, the *a priori* estimate of the state vector is always zero, and thus, the state transition, \mathcal{F} or $\hat{f}(\cdot)$, is zero, which simplifies the standard KF formulation. For instance, Foxlin [153], Setoodeh et al. [154], and Gebre-egziabher et al. [155] proposed CKFs by estimating the errors in gyroscope bias and Euler angles to correct y_G and the SDI output before and after the SDI, respectively. Gebre-egziabher et al. [155] also used the quaternion errors to develop a similar CKF and showed that the quaternion parametrization has advantages such as more efficient gain tuning process and computations as compared to Euler angle parametrization. To achieve more robust estimates for indoor orientation tracking, Zhang and Reindl [156] added a magnetic disturbance rejection step to the CKF by Foxlin [153].

Roumeliotis et al. [150] introduced a smoothing filter, comprised of two CKFs, where the first CKF propagated the attitude estimate forward and then updated it via the absolute orientation obtained from the accelerometer and sun sensor, while the second CKF propagated the most recent estimate back in time to lower the uncertainty of the estimation. To estimate the errors in the attitude of a mobile robot moving on uneven terrains using y_A and the SDI, Fuke and Krotkov [157] proposed the application of a CKF. Luinge and Veltink [54] developed an

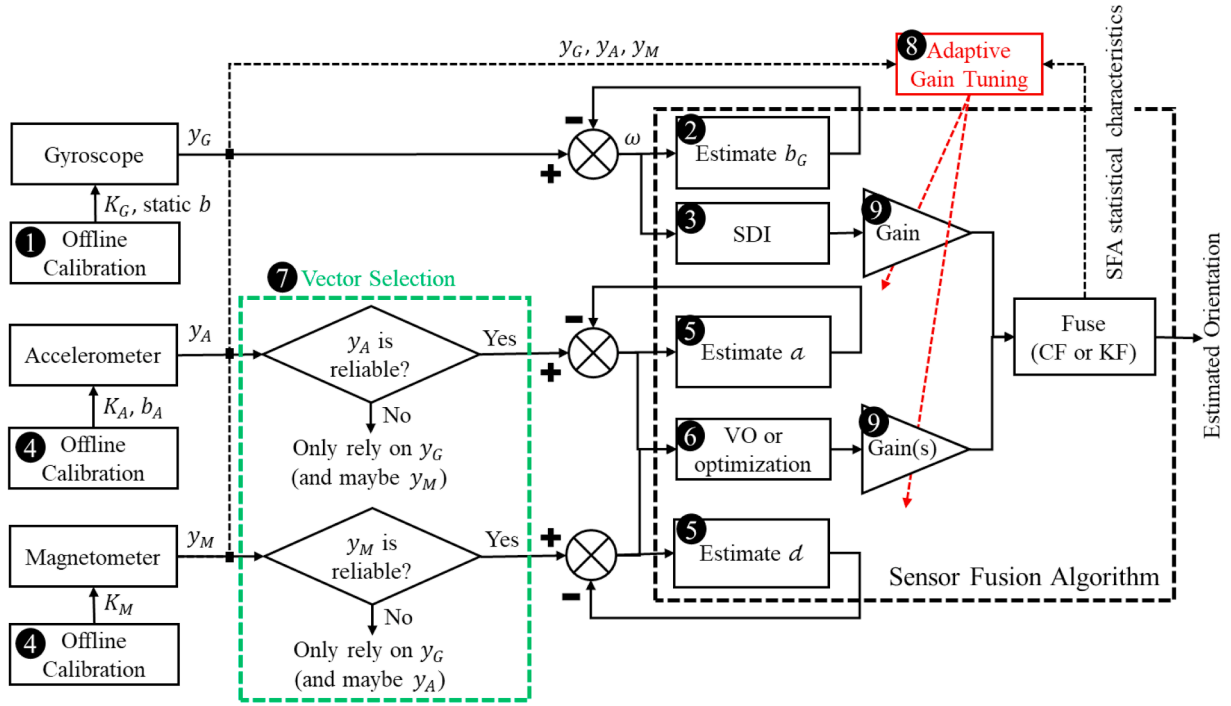


Fig. 3. Flowchart of a general sensor fusion algorithm (SFA) containing offline calibration of sensors, vector selection for imperfect measurement rejection, and adaptive gain tuning.

accelerometer-based CKF to estimate the error in gravitational acceleration and accelerometer offset by deriving the relationship between y_A and the state vector as well as calculating the angular velocity based on the direction of the y_A at each two successive time instants. Luinge and Veltink [59] also used both accelerometer and gyroscope to estimate the orientation error and gyroscope bias using the CKF structure. Later, the magnetometer was included in the CKF to prevent drift in the yaw angle, even under magnetic disturbances [52].

Hall et al. [158] proposed a CKF where y_G was used to propagate the error in orientation while y_A and GPS were used to perform the measurement-update. However, as the accelerometer and GPS had different sampling rates and reference frames, the proposed CKF took advantage of the incremental update concept [159], where the measurement-update could be performed independently for each sensor. Kang et al. [160] developed a CKF to fuse a gyroscope and a vision system where a fading factor, tuned using fuzzy logic, has been applied to the Kalman gain, K , to tune the tracking performance of the filter in response to the quality of the vision system data. Kannan [161] proposed a linear CKF where the compensation of the drift in the estimated orientation was performed via the time-update equation (using the orientation obtained by TRAID from y_A and y_G) instead of the measurement-update equation. This approach resulted in a substantial improvement of the response and settling times of the proposed CKF.

2.5.5. Modified Kalman filters

Ren and Kazanzides [162,163] proposed a two-layer filtering scheme such that two LKFs in the first layer were designed to estimate $^G\hat{g}$ and $^G\hat{m}$ from y_A and y_M , respectively, and an EKF in the second layer was used to correct the predicted orientation by SDI using the orientation calculated by VOs from the first layer. Music et al. [164] introduced a two-layer filter where in the first layer, a standard EKF estimated orientation using SDI and direct Euler angle calculation from y_A and y_M , while the estimated orientation in the second layer (calculated using a Particle Filter) was used to present extra information to EKF and improve the overall performance. Sabatelli et al. [165] developed a two-layer filter where the first and second layers used y_A and y_M , respectively, to correct

the propagated orientation by the SDI. Also, Dang and Nguyen [166] proposed a two-layer filter where in the first layer, an EKF estimated the third column of DCM (associated with attitude) and α while in the second layer, an EKF with similar structure estimated the first column of DCM (associated with yaw angle) and δ . Such structures have the benefit of eliminating the effect of magnetic disturbance from the attitude estimations.

2.5.6. Adaptive gain tuning of Kalman filters

KF performance highly relies on an accurate definition of the system and measurement models, as well as noise covariance matrices (commonly, Q and R are defined before the estimation starts and remain fixed during the tracking process). However, in reality, Q and R could change over time due to the time-varying nature of the errors, such as orientation tracking during slow (quasi-static) and fast motions. Therefore, adaptive KFs have been proposed to respond to the changes in the nature of the error by tuning Q and R in real-time [167].

For example, in the covariance scaling technique, Q or R are scaled by the factor $S_k > 1$ to put more weight on the *a priori* state vector x^- , or measurements, respectively [168]. S_k can be tuned using the magnitude of y_A , y_G , and/or y_M at some predefined levels [51,145], or the magnitude of the predicted residuals [168]. For example, Suh [55] and Suh et al. [169] customized the application of the covariance scaling technique for a CKF and EKF, respectively, such that instead of applying a small S_k to all elements of Q or R to guard against α (and rely more on y_G), first, the direction of the α was estimated, and small weights were supplied only for the affected axes. Sun et al. [170] included an adaptive scheme in their standard CKF to weight R based on the error between magnitude and direction of y_A (y_M) with its reference value recorded during static (magnetically undisturbed) condition. Later, this method was extended by Johnson and Sathyan [171] to include b_G estimation. Tong et al. [172] proposed the application of a hidden Markov Model recognizer to identify the measurement disturbances and adjust R of their proposed CKF adaptively. Also, Jamil et al. [173] used an Artificial Neural Network to eliminate the errors in y_A and y_G by adjusting R .

In the innovation-based adaptive estimation technique, the innova-

tion sequence, that is, the difference between the real measurement and its predicted value by the KF, is used in an algorithm known as the covariance matching to tune the value of Q , R , or K such that the actual value of the innovation sequence covariance matrix matches its theoretical value [174,175]. Finally, in the multiple-model adaptive estimation technique, a bank of KFs (each using a different Q or R) runs in parallel to calculate the *a posteriori* state estimate \hat{x}^+ based on the weighted combination of the predicted state by each KF (KF weights could be obtained based on the probability density function of the innovation sequence covariance matrix) [176].

3. Discussion

Over the past forty years, researchers have developed various SFAs for orientation tracking with MIMUs. KFs and CFs were used to compensate for the limitations associated with the SDI or VO algorithms. In both KF and CF approaches, the SDI of y_G was used to propagate the orientation in time, and then, the propagated orientation was corrected with the orientation estimated by a VO algorithm or a corrective term obtained by an optimization rule such as GDA. However, technical differences were observed amongst these SFAs, even amongst those in the same family. Therefore, we conducted a survey study to (i) review and categorize publications in this field to help researchers acquire an inclusive understanding of the technical developments; (ii) aggregate the lessons learned from the previous works to help researchers developing novel SFAs; and (iii) summarize the main challenges that are needed to be addressed in future. The first task was done in Section 2 of the current survey, while the second and third tasks were addressed in Sections 3.1 and 3.2, respectively.

3.1. Lessons learned

Numerous works proved the efficiency of SFAs in compensating the limitations associated with the SDI or VO algorithms for orientation tracking with MIMUs. In this section, we listed the lessons learned during the literature survey. The general flowchart of an SFA in Fig. 3 includes features that could improve the accuracy and robustness of the orientation estimation.

3.1.1. Gyroscope and strap-down integration (SDI)

SDI of y_G is the core of almost all CFs and KFs. Therefore, the accuracy of the propagated orientation in time via SDI is vital to the accuracy and robustness of any SFA, especially when aiding sensors' recordings, i. e., y_A and y_M , are not reliable, such as tracking under high dynamics or magnetically disturbed environments. To improve the accuracy of SDI, the followings must be considered.

- (i) Use a gyroscope with low bias.
- (ii) Correct the gyroscope's static bias and scale factor (block 1 in Fig. 3). For the static bias, before any data acquisition, the MIMU must be turned on and put at rest for a while (commonly specified by the manufacturer, and if not, at least 20 minutes according to [42]). This will ensure that the gyroscope has reached ambient temperature, and the estimated static bias will not change significantly due to temperature change. Then, y_G must be measured when the MIMU is at rest to obtain an estimate of the static bias. Finally, the estimated static bias must be removed from y_G recorded during data acquisition.
- (iii) Use stochastic models [52,147] or heuristics [41] to estimate the time-varying b_G in real-time (block 2 in Fig. 3). A first-order Markov process, as in (1b), can be used for this purpose. According to the principle of pseudo-noise injection, estimating b_G can improve the accuracy and robustness of the orientation tracking, even for short trials where b_G does not have time to change drastically [127].

- (iv) Use an accurate estimation of $\exp(\cdot)$ for the SDI in (4a) (block 3 in Fig. 3). For example, using a first-order Taylor series showed inaccurate results under low sampling frequencies [57].

3.1.2. Accelerometer and magnetometer

During long-duration trials, the use of accelerometer and magnetometer are crucial to an SFA, as they can correct the cumulative error of the SDI. Thus, consider the following for these aiding sensors.

- (i) Evaluate the calibration of these sensors (blocks 4 in Fig. 3). Various procedures have been introduced in the literature to verify and possibly re-calibrate accelerometers [27,53,177] and magnetometers [178]. These procedures use *ad hoc* tests to estimate the scale factors K_A and K_M (in (2a) and (3a)), as well as the bias terms of these sensors, which can be later used to correct y_A and y_M .
- (ii) Use online stochastic models, e.g., first-order Markov processes as in (2b) and (3b), to estimate α and δ , respectively, and correct y_A and y_M (blocks 5 in Fig. 3). This correction is effective for motions with high dynamics or in magnetically disturbed environments [51,52,127].

3.1.3. Dealing with magnetic disturbance

Magnetic disturbance can significantly affect the yaw angle estimation, which in turn may affect the attitude estimation. Thus, consider the following when formulating an SFA.

- (i) Decouple the attitude estimation from y_M or yaw angle (block 6 in Fig. 3). For example, use TRIAD, instead of QUEST, as the VO algorithm in CF or KF, or use two-layer filters as in [79,104,130,179,180].
- (ii) Use vector selection to detect and reject imperfect measurements (block 7 in Fig. 3). When the geomagnetic field is disturbed, the field strength and dip angle change significantly. Therefore, magnetic disturbance can be identified by applying thresholds to the field strength and/or dip angle. In such cases, the propagated orientation by SDI can be declared as the estimated orientation without any correction, see [38] for more details. A similar approach can be applied to y_A to reduce the effect of non-gravitational acceleration.

3.1.4. Adaptive gain tuning

Gain tuning plays a significant role in the performance of any SFA; if gains are selected improperly, the SFA could even diverge. Therefore, consider the following regarding SFA gains.

- (i) Use an online gain turning strategy to adaptively put more weight on the most reliable source of information (block 8 in Fig. 3). For instance, to reduce the effect of correction via the magnetometer and rely more on the gyroscope, when the magnetic disturbance is detected using a vector selection, the CF gain or KF measurement covariance matrix should be adjusted accordingly.
- (ii) Decouple the gains associated with gyroscope, accelerometer, and magnetometer (blocks 9 in Fig. 3). Generally, developing a KF with a different gain for each sensor is easy. However, in most well-recognized CFs, such as [25,26], the same gain is used to weight accelerometer and magnetometer. In such cases, both sensors will be declared reliable or unreliable at the same time, while one can be more reliable than the other one.
- (iii) Select the filter gains rigorously. Although filter gains found by trial-and-error might result in acceptable estimations under certain working conditions, the filter performance will not be guaranteed for other situations. Therefore, filter gains must be evaluated under various motion patterns, intensities, and durations.

Table 2

Review of the state-of-the-art sensor fusion algorithms with a Complimentary Filter (CF) structure, including linear CF (LCK) and nonlinear CF (NCF). Abbreviations used in the table are described in the table footnotes.

Study	Year	Application	Method	Parametrization	b_G, α, δ compensation	Notes (gain tuning or thresholding, etc.)
[76]	1999	HMT	GNA+CF	Q	-	-
[193]	2003	STC	NCF	Q	b_G	-
[184]	2006	AVTC	NCF	DCM	b_G	-
[111]	2007	-	NCF	Q	b_G	-
[26]	2008	RATC	NCF	DCM	b_G	-
[94]	2008	AVTC	NCF	Q	b_G	-
[114]	2009	MRTC	LCK	EA	-	$k_p \propto \exp(- g - \ y_A\)$
[109]	2009	-	Sliding mode observer	EA	b_G	-
[77]	2009	HMT	GNA+CF	Q	-	α & δ rejection with thresholding
[87]	2009	-	Azimuth-level detector CF	Q & DCM	-	-
[49]	2010	AVTC	NCF	Q	b_G	-
[91]	2010	-	NCF	EA	-	-
[25]	2011	HMT	GDA+CF	Q	-	-
[75]	2011	-	FQA+CF	Q	-	-
[78]	2011	Bio-logging	LMA+CF	Q	-	-
[110]	2011	SVTC	LCK	EA	-	-
[100]	2011	-	NCF	DCM	b_G	Time-varying gains
[92]	2011	MRTC	NCF	EA	b_G	-
[93]	2011	SVTC	NCF	EA	b_G	CF gain $\propto g - \ y_A\ $
[96]	2012	STC	NCF	DCM	-	-
[99]	2012	AVTC	NCF	DCM	b_G & b_A	-
[194]	2012	-	LCK	DCM	-	Using inverse sensor models to widen the frequency range
[195]	2012	-	TRIAD+LCK	DCM	b_G	-
[196]	2012	HMT	GDA+CF	Q	b_G	-
[197]	2012	HMT	LCK	DCM	-	CF gain \propto duration of the experiment
[198]	2012	MRTC	NCF	Q	-	-
[97]	2015	-	TRIAD+NCK	DCM	-	Adaptive gain tuning using fuzzy logic
[80]	2013	HMT	GDA+CF	Q	b_G	α & δ rejection with thresholding
[79]	2014	HMT	Two-layer LMA+CF	Q	-	-
[41]	2011	HMT	GDA+CF	Q	b_G	-
[199]	2014	AVTC	PI controller+CF	EA	-	-
[102]	2014	AVTC	NCF	DCM	b_G	-
[98]	2014	HMT	LCK	EA	-	Adaptive gain tuning using fuzzy logic
[200]	2014	HMT	LCK	DCM	b_G	-
[89]	2015	Bicycle crank angle tracking	NCF + vertical acceleration update	Q	-	-
[112]	2015	AVTC	LCK	DCM	b_G estimation by neural network PID controller	-
[107]	2015	AVTC	Smooth 2 nd order sliding mode observer	EA	-	-
[88]	2015	HMT	GDA+CF	Q	-	δ rejection with thresholding
[73]	2015	AVTC	AQA+CF	Q	-	α rejection using LERP or SLERP + adaptive gain tuning
[201]	2015	AVTC	NCF	DCM	-	-
[81]	2016	HMT with smartphone	Geometrically-intuitive CF	Q	-	-
[104]	2016	-	Two-layer LCK	Q	-	δ rejection with thresholding
[105]	2016	AVTC	Nonlinear SVD+CF	DCM	-	-
[113]	2016	RATC	NCF	DCM	b_G	-
[82]	2017	-	GDA+CF	Q	-	δ rejection with thresholding
[115]	2017	AVTC	LCK	EA	b_G	Combining output of multiple fixed-gain CFs with MMAE
[116]	2017	AVTC	NCF	EA	-	Adaptive gain turning via PSO
[84]	2018	-	GDA+CF	Q	-	δ rejection with thresholding
[103]	2018	HMT	LCK	Q	-	α & δ rejection with finite state machine
[85]	2018	RATC	FQA+CF	Q	-	Adaptive gain turning via GDA
[108]	2018	AVTC	Smooth 2 nd order sliding mode observer	EA	b_G	-
[101]	2018	AVTC	Right invariant NCF	Q & DCM	b_G	CF gain \propto system noise parameters
[106]	2018	AVTC	LCK	Q	-	CF gain $\propto \ y_G\ $
[202]	2018	-	MSRVA+CF	Q	b_G	-
[86]	2018	-	GDA+CF	Q	-	-
[83]	2019	AVTC	GDA+CF	Q	-	α & δ rejection with thresholding
[90]	2019	-	GDA+CF	DCM	b_G	-

(continued on next page)

Table 2 (continued)

Study	Year	Application	Method	Parametrization	b_G , α , δ compensation	Notes (gain tuning or thresholding, etc.)
[203]	2019	Robot teleoperation	GDA+CF	Q	b_G	-
[95]	2019	AVTC	NCF	EA	b_G	Identified reliable y_G based on interquartile range of y_G previous samples
[204]	2020	HMT	GDA+CF	Q	-	δ rejection
[205]	2020	HMT	GDA+CF	Q	-	y_G rejection during “no motion” state
[206]	2020	HMT	Two-layer LCK	Q	-	-
[207]	2020	-	LCF	Q	-	-
[208]	2020	AVTC	NCF	Q	-	α compensation by thresholding
[209]	2020	HMT	Extended CF	Q	b_G	δ rejection with thresholding

- **Application:** STC: Spacecraft (satellite) tracking/control; HMT: Human motion tracking; MRTC: Mobile robot tracking/control; RATC: Robotic arm tracking/control; AVTC: Aerial vehicle tracking/control; SVTC: Surface vehicle tracking/control.

- **Method:** GNA: Gauss-Newton algorithm; GDA: Gradient descent algorithm; LMA: Levenberg-Marquardt algorithm; FQA: Factored quaternion algorithm; AQA: algebraic quaternion algorithm; LERP: Linear interpolation; SLERP: Spherical Linear interpolation; SVD: Singular value decomposition; PSO: Particle Swarm Optimization; MSRVA: Multi-sample rotation vector algorithm; MMAE: Multiple-model adaptive estimation.

- **Parametrization:** Q: Quaternion parametrization of orientation; EA: Euler angle parametrization of orientation; DCM: Direction cosine matrix parametrization of orientation.

- **Compensation:** b_G : Gyroscope bias; b_A : Accelerometer bias; α : external non-gravitational acceleration; δ : magnetic disturbance.

3.1.5. Beyond EKF

Violation of the assumptions that were used to formulate a filter can degrade its performance. Thus, consider the following two extensions for KFs.

- (i) When execution time is not a consideration, use unscented KF instead of EKF. As EKF uses linearization to propagate the mean and covariance of the state vector, its performance can deteriorate for highly nonlinear systems.
- (ii) Use robust filtering techniques when adaptive gain tuning is not possible. For example, the $-H_\infty$ filter formulates an estimation strategy that bounds the worst-case estimation error by adaptively weighting the state error covariance matrix [117]. Also, the $-H_\infty$ filter assumes that the process and measurement noises are energy bounded signals, and drops the Gaussian distribution assumption for noises in formulating the standard KF.

3.2. Future research challenges

Various VO/CF/KF formulations with major or minor differences have been developed for orientation tracking with MIMUs. Thus, one could argue that technical contribution to this field is not the primary research challenge at the moment. Instead, we suggest that test platforms [181] and benchmarking studies are required to identify the most effective SFAs, as well as techniques that could improve the accuracy and robustness of SFAs. The reason is that each publication assessed the performance of its proposed SFA using limited test scenarios in terms of motion duration, pattern, and intensity. Also, other contributing factors, such as the manufacturing quality of the MIMU or optimal SFA gains, were not considered in assessments. As a result, contradictory results on the performance of SFAs have been reported in various studies.

Therefore, developing benchmarks where SFAs can be assessed against each other under similar testing conditions remains the main challenge in the field of orientation tracking with MIMUs. Such a benchmark must have the following features.

- (i) Various test scenarios are recorded. Short- and long-duration trials are required to evaluate the effect of b_G . Static, medium and high dynamic (with static periods in the middle) trials are required to evaluate convergence and tracking ability of the SFAs. Extreme testing scenarios, such as MIMU impact, are required to reveal the robustness of an SFA.
- (ii) The true orientation is recorded synchronously with MIMUs using a reference system such as the camera motion-capture system.

This information is not only useful for performance assessment but also for tuning the SFA's gain(s).

- (iii) Data of multiple MIMUs, preferably with different manufacturing qualities, are recorded. Such trials can demonstrate the effectiveness of the SFA correction strategy using the aiding sensors, specifically for long-duration trials where SDI faces a significant drift.

Using a benchmark with these features, optimal gains for each SFA and MIMU should be obtained using a portion of the data and by comparing the estimated and reference orientations. The rest of the data can be used for validation.

Finding the most effective techniques for improving the performance of the existing SFAs is the second research challenge. Using a benchmark, the following should be evaluated.

- (i) Various adaptive gain tuning strategies must be assessed. While some strategies, such as the covariance scaling technique, can only be applied to KFs, vector selection for gain turning can be applied to almost any SFA.
- (ii) Various stochastic and heuristic models for estimating the unmeasurable time-varying terms in the MIMU model must be investigated. For example, applying the first- vs second-order Markov process for estimating the non-gravitational acceleration or magnetic disturbance.
- (iii) Uncommon modelling techniques, such as two-layer filters or variable state KF, must be further evaluated.

4. Conclusion

Due to the number of publications in the field, a comprehensive survey of all publications related to SFAs for orientation tracking with MIMUs is virtually impossible. Nevertheless, in this paper, the three main strategies, including VO, CF, and KF, were surveyed and categorized, and references to more than 250 publications are provided. Specifically, for CFs and KFs, details of the filter structure, compensation strategies for correcting the raw MIMU signals, and gain tuning strategies are tabulated. Therefore, the current survey can be regarded as a source for details on the state-of-the-art SFAs, all described with a unified notation.

Moreover, we aggregated the lessons learned from the previous works to help researchers develop effective SFAs (Tables 1, 2 and 3). In summary, as shown in Fig. 3, an SFA should include the following elements: (i) offline calibration of gyroscope, accelerometer, and magnetometer; (ii) online estimation of gyroscope bias, non-gravitational

Table 3

Review of the state-of-the-art sensor fusion algorithms with a Kalman Filter (KF) structure, including linear KF (LKF), extended KF (EKF), and Complimentary KF (CKF). Abbreviations used in the table are described in the table footnotes.

Study	Year	Application	Method	State vector components	Measurement-update	Notes (gain tuning or thresholding, etc.)
[140]	1982	STC	EKF	$[Q, b_G]$	[Line-of-sight attitude sensor]	-
[141]	1991	SVTC	EKF	[Velocity, ω , EA, altitude, wind gust velocity]	$[y_G, y_M]$, airspeed sensor, barometric altimeter]	-
[142]	1994	MRTC	EKF	[Attitude]	$[y_A, y_M]$, gyrometric attitude]	-
[118]	1995	MRTC	LKF	[Yaw]	$[y_G]$	-
[153]	1996	HMT	CKF	$[e EA, eb_G]$	$[y_G, y_M]$	$Q \propto \max(y_G) \mathcal{R} \propto \text{slosh in fluid acc.}$
[157]	1996	MRTC	CKF	[e Attitude]	$[y_A]$	-
[150]	1999	MRTC	CKF	$[e EA, eb_G]$	$[y_A, \text{sun sensor}]$	-
[143]	2001	-	EKF	$[Q, \hat{\omega}]$	$[y_G, y_A, y_M]$	-
[143]	2001	-	GNA+EKF	$[Q, \hat{\omega}]$	$[\hat{q}, y_G, \hat{q}]$ from GNA(y_A, y_M)	-
[8]	2002	SVTC	LKF	[GPS position, GPS receiver's clock range bias, b_G, b_A]	[Position, velocity, clock] errors from the algebraic GPS equations	-
[119]	2003	-	GNA+LKF	$[Q, \hat{\omega}]$	$[\hat{q}, y_G, \hat{q}]$ from GNA(y_A, y_M)	-
[126]	2004	-	LKF	$[Q, \hat{\omega}, b_G]$	$[y_G, y_A]$	-
[28]	2004	HMT	LKF	$[\hat{g}, \hat{m}]$	$[y_A, y_M]$	-
[138]	2004	Walking robot motion tracking	Ensemble of two LKFs	[DCM(3)]	$[y_A]$	Switching between LKFs by $ y_A $
[47]	2004	-	LKF	$[Q, b_G]$	$[y_A]$ or $[y_M]$	Adaptive Q
[54]	2004	HMT	CKF	$[e\hat{g}, eb_A]$	$[y_A]$	-
[210]	2004	-	LKF	[1D angle, ω]	$[1D-y_G]$	-
[154]	2004	-	CKF	$[e EA, eb_G]$	[EA from y_A & y_M]	-
[155]	2004	SVTC	CKF	$[e EA, eb_G]$ OR $[e Q, eb_G]$	[EA from y_A & y_M]	Switching Q & \mathcal{R} between two levels
[59]	2005	HMT	CKF	$[e \text{ Helical angle/axis, } eb_G]$	[Attitude from y_A & y_G]	-
[6]	2005	-	CKF	$[e Q, eb_G]$	[Sun sensor]	-
[52]	2005	HMT	CKF	$[e \text{ Helical angle/axis, } eb_G, eb]$	[Attitude from y_A & y_G , magnetic vector from y_M & y_G]	-
[51]	2006	HMT	EKF	$[Q, b_A, b_M]$	$[y_A, y_M]$	Switching \mathcal{R} between two levels
[29]	2006	HMT	QUEST+EKF	$[Q, \hat{\omega}]$	$[\hat{q}, y_G, \hat{q}]$ from QUEST(y_A, y_M)	-
[169]	2006	-	EKF	[Attitude, $\hat{\omega}$]	$[y_{Ax}, y_{Ay}, y_G]$	Adaptive acc. measurement equation
[131]	2007	-	LKF	$[\hat{g}, \hat{m}]$	$[y_A, y_M]$	Abandoning past estimations by forget factor
[134]	2007	-	LKF	EA	[EA from y_A & y_M]	Switching \mathcal{R} between two levels
[151]	2007	-	CKF	$[e Q, eb_G]$	$[y_A, y_M]$	-
[158]	2008	AVTC	CKF	$[e Q]$	$[e \hat{q}]$ from y_A & GPS]	Tuning \mathcal{R} based on $\ y_A - \hat{g}\ $
[211]	2008	SVTC	CKF	$[e Q, eb_G]$	$[y_A, y_M]$	-
[212]	2008	HMT	EKF	$[e EA, b_G, b_A, b]$	$[y_A, y_M]$	α & β rejection using vector selection + tuning Q & \mathcal{R}
[163]	2009	Surgical tool motion tracking	EKF	$[Q, b_G]$	$[y_A, y_M]$	-
[122]	2009	HMT	LKF	$[Q]$	$[\hat{q}]$ from y_A & y_M using [123]	α & β rejection using vector selection
[9]	2009	Augmented Reality	-	$[EA, \hat{\omega}, \hat{\omega}, b_G]$	$[\hat{q}, y_G, \hat{q}]$ from y_A & y_M	-
[213]	2009	HMT	EKF	[Velocity, $\hat{y}_A, \hat{\omega}, b_G$, Attitude]	$[y_G, y_A]$	-
[132]	2009	-	LKF	$[\hat{g}, \hat{m}, b_G]$	$[y_A, y_M]$	-
[55]	2010	-	CKF	$[e Q, b_G, b_A]$	$[y_A, y_M]$	Tuning \mathcal{R} using residual in acc. measurement-update
[164]	2010	HMT	EKF+Particle Filter	[EA, b_G]	[EA from y_A & y_M]	-
[170]	2010	HMT	CKF	$[e Q]$	$[y_A, y_M]$	Tuning \mathcal{R} based on $\ y_A - \hat{g}\ $ & $\ y_M - \hat{g}_m\ $
[175]	2011	-	CKF	$[e EA, e \text{ velocity, } eb_G, eb_A]$	Velocity from GPS	\mathcal{K} tuning by innovation-based adaptive estimation technique
[124]	2011	-	LKF	$[Q]$	$[\hat{q}]$ from FQA(y_A, y_M)	-
[135]	2011	HMT	LKF	$[Q, \alpha]$	$[y_M, y_A, \ y_A\]$	Switching Q between two levels
[214]	2011	HMT	EKF	$[Q, b_G]$	$[y_A, y_M]$	-
[215]	2011	HMT	CKF	$[e EA, e \text{ position, } e \text{ velocity, } eb_G, b_A]$	[Yaw from y_M, eb_G , Velocity from ZUPT]	Heuristic heading reduction [216]
[147]	2011	HMT	EKF	$[Q, b_G, b]$	$[y_A, y_M]$	-
[217]	2011	-	EKF	$[Q]$	$[y_A]$	-
[218]	2011	-	CKF	$[e Q, eb_G, eK_G]$	[e EA from y_A & y_M]	-
[219]	2011	-	EKF	$[Q, b_G]$	$[y_A, y_M]$	Tuning \mathcal{R} based on $\ y_A - \hat{g}\ $
[220]	2011	-	EKF	[DCM(1), DCM(3), b_G]	$[y_A, y_M]$	-

(continued on next page)

Table 3 (continued)

Study	Year	Application	Method	State vector components	Measurement-update	Notes (gain tuning or thresholding, etc.)
[221]	2011	HMT	CKF	[e EA, eb _G]	[y _M]	Quasi-static geomagnetic field detection
[162]	2012	Surgical tool motion tracking	Two LKFs + EKF	LKF1: [\hat{g}], LKF2: [\hat{m}], EKF: [Q, b _G]	LKF1: [y _A], LKF2: [y _M], EKF: [\hat{q} from y _A & y _M]	-
[133]	2012	-	LKF	[\hat{g} , \hat{m} , b _G]	[y _A , y _M]	-
[145]	2012	HMT	EKF	[Q, b _G]	[\hat{q} from y _A & y _M]	Switching \mathcal{R} a piecewise linear function
[222]	2012	HMT/ SVTC	LKF	[\hat{g}]	[y _A]	-
[223]	2012	AVTC	EKF	[DCM(3), $\hat{\omega}$, b _G]	[y _A , y _G]	Switching \mathcal{R} between two levels
[224]	2012	HMT	LKF	[Q]	[\hat{q} from y _A & y _M]	α & δ rejection using vector selection
[148]	2012	-	Variable state EKF	EKF1: [Q, b _G , δ] EKF2: [Q, b _G , δ , $\dot{\delta}$]	[y _A , y _G]	-
[225]	2013	-	CKF	[e EA, eb _G , K _G]	[e EA from y _A & y _M]	Switching \mathcal{R} between three levels
[166]	2013	-	EKF	EKF 1: [DCM(3), α] EKF 2: [DCM(1), δ]	EKF1: [y _A], EKF2: [y _M]	Resetting α and δ by thresholding
[226]	2013	SVTC	CKF	[e position, e velocity, e EA, b _G , b _A]	[Position, velocity] from GNSS	-
[146]	2013	HMT	EKF	[Q, b _A , b _M]	[y _A , y _M]	Switching \mathcal{R} between two levels
[165]	2013	-	EKF	[Q]	[y _A , y _M]	-
[136]	2014	-	LKF	[Q]	[y _A , y _M]	Tuning \mathcal{R} based on acc. residuals
[227]	2014	HMT	Two-layer LKF	EKF1: [\hat{g}], EKF2: [Yaw]	EKF1: [y _A], EKF2: [y _M]	Innovation-based adaptive estimation tuning of \mathcal{R} using fuzzy logic
[46]	2014	HMT with smartphones	EKF	[Q, b _G , α]	[y _A , y _M]	-
[228]	2014	AVTC	EKF	[Q, $\hat{\omega}$, b _G]	[\hat{q} from y _A & y _M]	-
[229]	2014	HMT	Two-layer LKF	EKF1: [DCM(3)], EKF2: [DCM(1)]	EKF1: [y _A], EKF2: [y _M]	Switching \mathcal{R} between two levels
[230]	2014	AVTC	CKF	[e EA, eb _G , e α , e δ]	[y _A , y _M]	$\mathcal{P}_A \propto \text{Error}(\alpha)$, $\mathcal{P}_M \propto \text{Error}(\delta)$
[231]	2015	Marine Satellite Tracking Antenna	LKF	[Q]	[y _A , y _M]	$\mathcal{Q} = \alpha \mathcal{Q}_0$, $\mathcal{R} = (1 - \alpha) \mathcal{R}_0$
[127]	2015	HMT	LKF	[\hat{g} , α]	[y _A]	-
[129]	2015	HMT	LKF+LKF	LKF1: [\hat{g} , α], LKF2: [\hat{m} , δ]	LKF1: [y _A], LKF2: [y _M]	-
[232]	2015	AVTC	GDA+LKF	[Q]	[y _A]	GDA step size $\propto y_G T_s$
[233]	2015	HMT	EKF	[Q]	[y _A]	Switching \mathcal{R} between three levels
[234]	2015	HMT	LKF	[e Yaw, eK _G , b _G]	[Yaw from y _M]	-
[161]	2015	HMT	LKF	[e EA, eb _G]	-	-
[14]	2015	-	Constrained-LKF	[$\hat{\omega}$, DCM, α , δ]	[y _G , y _A , y _M]	-
[15]	2015	-	Constrained-EKF	[DCM(3), b _G , k]	[y _A]	Tuning \mathcal{R} based on $\ y_A\ $
[137]	2016	-	LKF	[Q]	[y _A , y _M]	Tuning \mathcal{R} based on accelerometer residuals
[235]	2016	AVTC	CF-EKF	[e position, e velocity, e EA, eb _G , eb _A , e δ]	[Position, velocity] from GNSS	-
[74]	2016	-	LKF	[Q]	[Position, velocity] from GNSS	-
[180]	2016	HMT/ SVTC	Two-layer LKF	EKF 1: [\hat{g}], EKF 2: [\hat{m}]	EKF1: [y _A], EKF2: [y _M]	Switching \mathcal{R} between three levels
[160]	2016	HMT	CKF	[e EA, eb _G]	[EA from vision system]	Adaptive fading factor by fuzzy logic
[236]	2016	HMT	LKF	[EA, b _G]	[Attitude from y _A]	$\mathcal{R} = \mathcal{R}_{nominal} + \ \alpha\ ^2$
[237]	2016	HMT	CKF	[e EA, eb _G]	[y _A , y _M]	Tuning \mathcal{R} using Hidden Markov Model
[149]	2017	AVTC	EKF	[Q]	[y _A]	-
[238]	2017	SVTC	CKF	[EA, b _G]	[y _A , velocity from GPS]	-
[239]	2017	AVTC	EKF	[Q, b _G]	[y _A , y _M]	-
[125]	2017	AVTC	LKF	[Q]	[\hat{q} from y _A & y _M]	-
[40]	2017	SVTC	LKF	[Q]	[\hat{q} from y _A & y _M]	δ rejection using vector selection
[130]	2017	-	Two-layer LKF	LKF1: [\hat{g}], LKF2: [\hat{m}]	LKF1: [y _A], LKF2: [y _M]	$\mathcal{R}_1 \propto \ \alpha\ ^2$, $\mathcal{R}_2 \propto \ \delta\ ^2$
[240]	2017	-	EKF	[Q, $\hat{\omega}$]	[y _A , y _G]	-
[241]	2017	Marine satellite tracking antennas	CKF	[e Q, eb _G]	[y _A , y _M]	-
[242]	2017	Wearable robotic systems	EKF	[EA, $\hat{\omega}$, $\dot{\omega}$, \hat{Q} , \ddot{Q} , d, d]	[y _G , y _A , y _M]	-
[120]	2018	HMT with smartphones	GDA+LKF	[Q]	[\hat{q}], \hat{q} from GDA(y _A , y _M)	-
[172]	2018	-	CKF	[e EA, eb _G]	[y _A , y _G]	Tuning \mathcal{R} using Hidden Markov Model
[71]	2018	-	LKF	[Q]	[\hat{q} from y _A & y _M]	-
[121]	2018	-	GDA+LKF	[Q]	[\hat{q}], \hat{q} from GDA(y _A , y _M)	$\mathcal{R} = \mathcal{R}_{nominal} + \alpha_k$ δ rejection using vector selection

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Table 3 (continued)

Study	Year	Application	Method	State vector components	Measurement-update	Notes (gain tuning or thresholding, etc.)
[243]	2018	HMT	Two-layer CKF	CKF1: [e Attitude], CKF2: [e Yaw]	CKF1: [e \hat{q} from y_A], CKF2: [e \hat{q} from y_M]	Tuning \mathcal{R} by variances of errors
[244]	2018	SVTC	EKF	[Position, velocity, Q , b_G , b_A]	[Position from GPS, \hat{q} from y_A]	-
[179]	2018	SVTC	Two-step LKF	[Q , b_G]	Step1: [y_A , $y_{G,z}$], Step2: [y_M]	$\mathcal{R}_A \propto \exp(\ y_A - \hat{g}_g\ ^2)$ $\mathcal{R}_M \propto \exp(\ y_M - \hat{g}_m\ ^2)$
[128]	2019	-	LKF+LKF	LKF1: [\hat{g}], LKF2: [\hat{m}]	LKF1: [y_A], LKF2: [y_M]	-
[245]	2019	HMT	CKF	[e EA, b_G , b]	[y_A , y_M]	-
[246]	2019	SVTC	CKF	[e EA, e velocity, e position]	[Velocity, Position] from GPS	Combined Sage-Husa [247] and Strong Tracking [248] KFs
[249]	2019	HMT	EKF	[Q]	[y_A , y_M]	$Q \propto \cos nt + \max(\ \omega\ - thr, 0)$ $\mathcal{R}_A \propto \cos nt + \max(\ y_A\ - \ \hat{g}_g\ - thr, 0)$ $\mathcal{R}_M \propto \cos nt + \max(\ y_M\ - \ \hat{g}_m\ - thr, 0)$
[250]	2020	HMT	EKF	[EA]	[EA] obtained by NCF	Innovation-based adaptive estimation tuning of \mathcal{R}
[251]	2020	MRTC	EKF	[Q , b_G]	[\hat{q}], \hat{q} from TRIAD(y_A , y_M)	Tuning \mathcal{R} by fuzzy logic
[252]	2020	HMT	LKF	[Q]	[y_A]	-
[253]	2020	HMT	Cascade LKF	LKF1: [DCM(3)], LKF2: [b_G]	LKF1: [y_A], LKF2: [y_A]	$\mathcal{R}_A \propto \alpha\ ^2 + \ y_A - \hat{g}_g\ ^2$
[254]	2020	HMT	GDA+EKF	[Q , b_G]	[\hat{q}], \hat{q} from GDA(y_A , y_M)	GDA step size $\propto \ y_G\ $, $error(y_A, y_M)$
[255]	2020	SVTC	EKF	[Q]	[y_A , y_M]	Post-EKF error reduction by proportional-integral controller
[256]	2020	AVTC	EKF	[Position, velocity, Q , b_G , b_A , LiDAR bias]	[Position, velocity, y_A]	-
[257]	2020	RATC	CKF	[e EA, e b_G]	[y_A , y_M]	Adaptive tuning of \mathcal{R} using ellipsoidal method

- **Application:** STC: Spacecraft (satellite) tracking/control; HMT: Human motion tracking; MRTC: Mobile robot tracking/control; RATC: Robotic arm tracking/control; AVTC: Aerial vehicle tracking/control; SVTC: Surface vehicle tracking/control; GNSS: Global Navigation Satellite System.

- **Method:** GNA: Gauss-Newton algorithm; GDA: Gradient descent algorithm; FQA: Factored quaternion algorithm; QUEST: Quaternion ESTimator; TRIAD: Tri-axial Attitude Determination; ZUPT: Zero-Velocity-Update strategy.

- **State vector components:** Q : Quaternion parametrization of orientation; EA: Euler angle parametrization of orientation; DCM: Direction cosine matrix parametrization of orientation; e: error of the component.

- **Measurement-update:** b_G : Gyroscope bias; b_A : Accelerometer bias; b_M : Magnetometer bias; α : external non-gravitational acceleration; b : magnetic disturbance; \hat{q} : Orientation used in measurement-update of KF; $\hat{\omega}$: Estimated angular velocity; \hat{g} : Estimated gravitational acceleration; \hat{m} : Estimated geomagnetic field.

- **Notes:** Q : System model covariance matrix; \mathcal{R} : Measurement model covariance matrix; \mathcal{P} : State error covariance matrix; \mathcal{K} : Kalman filter gain.

external acceleration, and magnetic disturbance; (iii) proper implementation of the SDI as well as an algorithm for orientation estimation via the corrected accelerometer and magnetometer readouts; (iv) vector selection technique for imperfect measurement rejection; and (v) adaptive gain tuning for intelligent selection of SFA gains in real-time. Finally, the accuracy and reliability of the newly developed SFAs must be evaluated against others using a publicly available data set. This could help researchers identify the strengths and weaknesses of each SFA in practice as well as gaps to be addressed.

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CRedit author statement

Milad Nazarahari: Conceptualization, Methodology, Investigation, Writing - Original Draft, Visualization.

Hossein Rouhani: Conceptualization, Resources, Writing - Review & Editing, Supervision, Project administration.

Declaration of Competing Interest

None.

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