

Stability

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1

1.1. An op amp designed to have a low-frequency gain of 10^5 and a high-frequency response dominated by a single pole at 100 rad/s, acquires, through a manufacturing error, a pair of additional poles at 10,000 rad/s.

- At what frequency does the total phase shift reach 180° ?
- At this frequency, for what value of H , assumed to be frequency independent, does the loop gain reach a value of unity?
- What is the corresponding value of closed-loop gain at low frequencies?

Solution:

$$G(s) = \frac{G_0}{1 + \frac{s}{p}} \quad (1.1.1)$$

Considering manufacturing error

$$G(s) = \frac{G_0}{\left(1 + \frac{s}{p}\right)\left(1 + \frac{s}{p_{error}}\right)} \quad (1.1.2)$$

$$G_0 = \text{Low Frequency Gain} = 10^5 \quad (1.1.3)$$

$$p = 100 \quad (1.1.4)$$

$$p_{error} = 10^4 \quad (1.1.5)$$

$$G(s) = \frac{10^5}{\left(1 + \frac{s}{100}\right)\left(1 + \frac{s}{10^4}\right)^2} \quad (1.1.6)$$

$$\angle G(j\omega) = -\tan^{-1} \frac{\omega}{100} - 2 \tan^{-1} \frac{\omega}{10^4} \quad (1.1.7)$$

1.2. Calculating the frequency at which the total phase shift reach 180°

At ω_{180} , $\angle G(j\omega_{180}) = -180^\circ$

Also $\omega_{180} \gg 100$

$$180^\circ = 90^\circ + 2 \tan^{-1} \left(\frac{\omega_{180}}{10^4} \right) \quad (1.2.1)$$

$$\tan^{-1} \frac{\omega_{180}}{10^4} = 45^\circ \quad (1.2.2)$$

$$\frac{\omega_{180}}{10^4} = \tan 45^\circ = 1 \quad (1.2.3)$$

$$\omega_{180} = 10^4 \text{ rad/s} \quad (1.2.4)$$

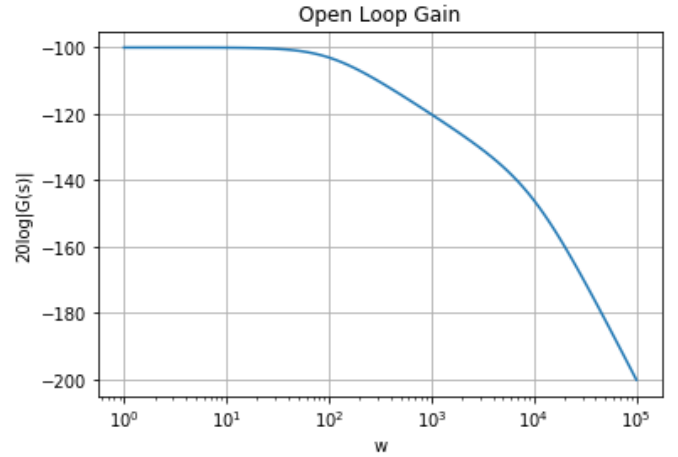


Fig. 1.2: Open Loop Gain

1.3. Calculating feedback factor H for which loop gain at ω_{180} is unity

$$\text{Loop Gain} = G(s)H = 1 \quad (1.3.1)$$

$$\frac{10^5 H}{\sqrt{1^2 + \left(\frac{\omega_{180}}{10^2}\right)^2} \sqrt{\left(1 + \frac{\omega_{180}}{10^4}\right)^2}} = 1 \quad (1.3.2)$$

$$H = 0.002 \quad (1.3.3)$$

1.4. Calculating the closed loop gain at low frequency Let $T(s)$ be the closed loop Transfer Function.

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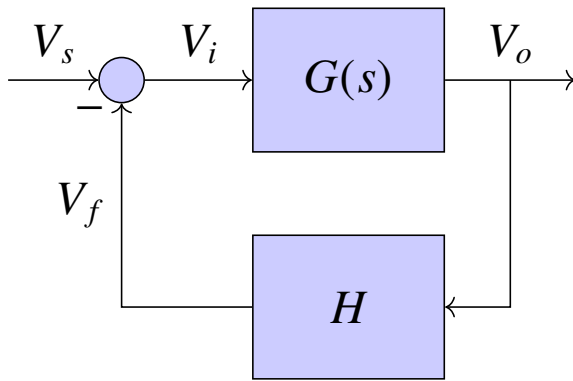


Fig. 1.4: Closed loop circuit

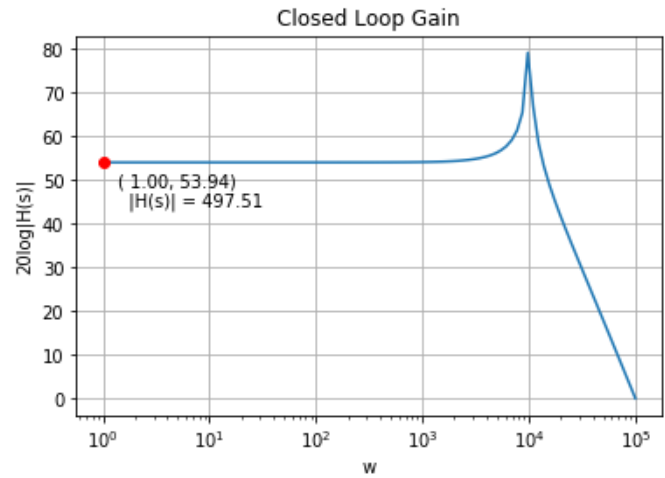


Fig. 1.4: Closed Loop Gain

$$T(s) = \frac{G(s)}{1 + HG(s)} \quad (1.4.1)$$

$$T(s) = \frac{10^5}{H10^5 + \left(1 + \frac{s}{100}\right)\left(1 + \frac{s}{1000}\right)^2} \quad (1.4.2)$$

$$T(s) = \frac{10^5}{10^{-10}s^3 + (2 \times 10^{-6})s^2 + 10^{-2}s + 1 + 10^5H} \quad (1.4.3)$$

At low frequencies

$$T(0) = \frac{10^5}{1 + 10^5H} \quad (1.4.4)$$

$$|T(0)| = \frac{10^5}{\sqrt{(200)^2}} \quad (1.4.5)$$

$$|T(0)| = 500V/V \quad (1.4.6)$$

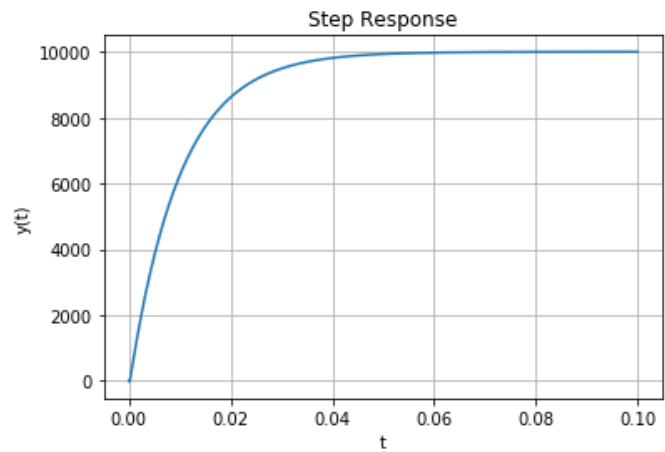


Fig. 1.4: Step Response

Parameter	Value
ω_{180}	10^4 rad/s
H	0.002
$ T(0) $	500V/V

TABLE 1.4: Obtained Parameters

The following code performs all the calculations of above equations

```
codes/code1.py
```

The following code plots the open loop gains, closed loop gains and step response to the system

```
codes/code2.py
```

1.5. Designing the circuit for transfer function $T(s)$

1) Designing $G(s)$

Let us assume Op-Amp to be ideal. So this means $V_1 = 0$ Applying KCL at node V_1

$$I_{in} = I_{C_1} + I_{R_2} \quad (1.5.1)$$

$$\frac{V_{in}}{R_1} = \frac{V_{o1}}{R_2} + C_1 \frac{dV_{out}}{dt} \quad (1.5.2)$$

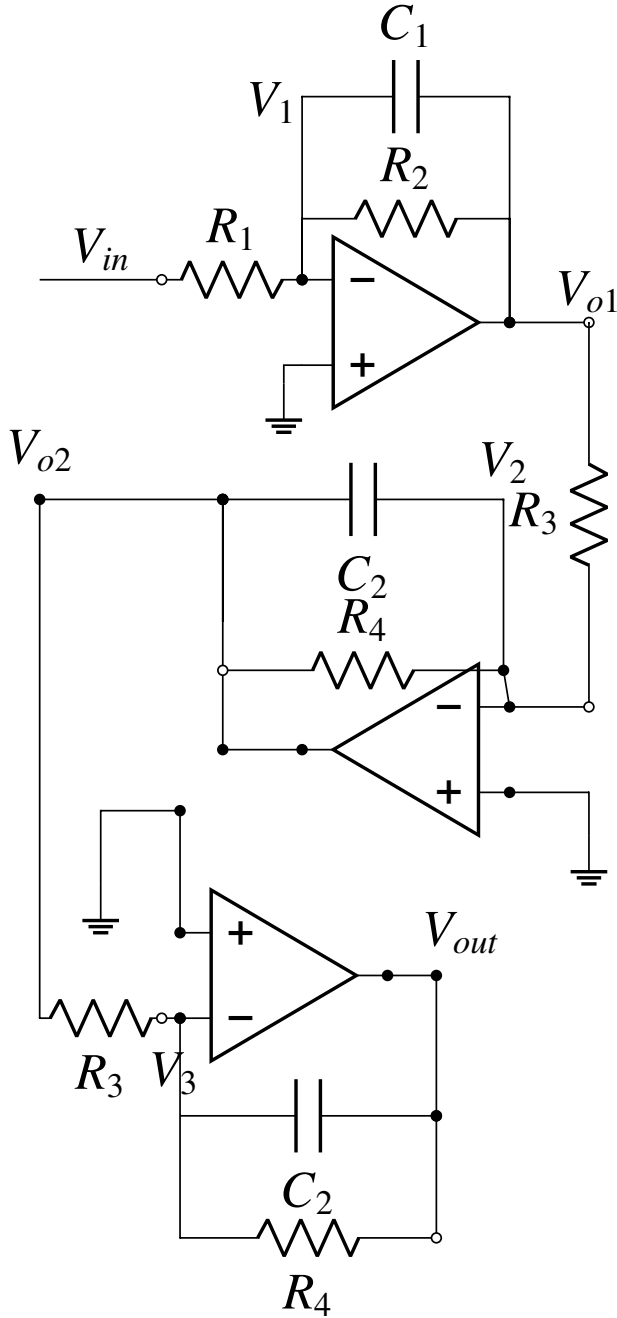


Fig. 1.5: Open Loop design

In Laplace domain

$$\frac{V_{in}(s)}{R_1} = \frac{V_{o1}(s)}{R_2} + C_1 s V_{out}(s) \quad (1.5.3)$$

$$\frac{V_{o1}}{V_{in}} = \frac{R_2/R_1}{1 + sR_2C_1} \quad (1.5.4)$$

$$\frac{V_{o1}}{V_{in}} = \frac{10^5}{1 + \frac{s}{100}} \quad (1.5.5)$$

$$\frac{R_2}{R_1} = 10^5 \quad (1.5.6)$$

$$R_2C_1 = \frac{1}{100} \quad (1.5.7)$$

As shown in 1.5 for the two identical poles at 10000 rad/sec we place similar op amp circuits twice.

Solving the circuit for second pole

$$\frac{V_{o1}(s)}{R_3} = \frac{V_{o2}(s)}{R_4} + C_2 s V_{out}(s) \quad (1.5.8)$$

$$\frac{V_{o2}}{V_{o1}} = \frac{R_4/R_3}{1 + sR_4C_2} \quad (1.5.9)$$

$$\frac{V_{o2}}{V_{o1}} = \frac{1}{1 + \frac{s}{10000}} \quad (1.5.10)$$

$$\frac{R_4}{R_3} = 1 \quad (1.5.11)$$

$$R_4C_2 = \frac{1}{10000} \quad (1.5.12)$$

2) Designing $H(s) = H$

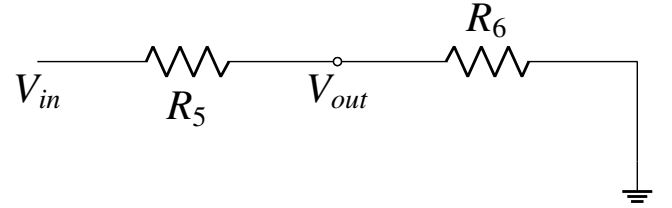


Fig. 1.5: Loop Gain

$$V_{out} = V_{in} \left(\frac{R_6}{R_5 + R_6} \right) \quad (1.5.13)$$

$$\frac{R_6}{R_5 + R_6} = 0.002 \quad (1.5.14)$$

$$R_6 = 0.002R_5 \quad (1.5.15)$$

3) Closed loop design

Figure 1.5 is the final closed loop design for transfer function $T(s)$

The table 1.5 provides the parameters for our circuit design.

The arbitrary parameters can be selected based on practical availability.

1.6. Verification of closed loop circuit design through SPICE

A SPICE simulation of circuit 1.5 is done by providing a DC input(Unit Step Input).

The obtained plot (figure 1.6) is similar to the Step response of the feedback system in figure 1.4. Hence we verify our design is correct.

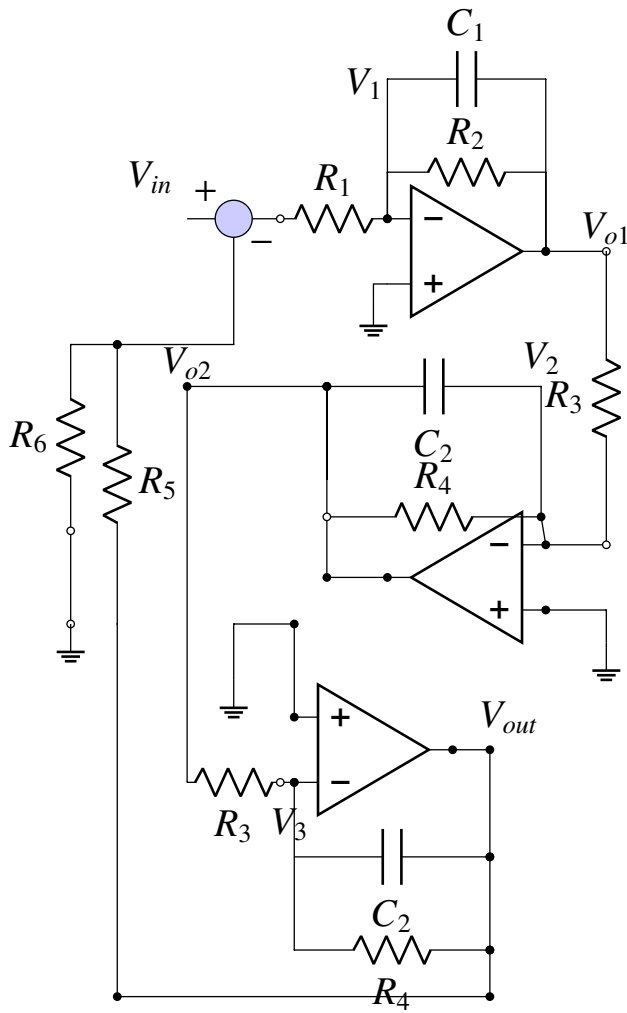


Fig. 1.5: Closed Loop Circuit

Parameter	Value
R_1	R
R_2	$10^5 R$
C_1	$10^{-7}/R$
R_3	R'
R_4	R'
C_2	$10^{-4}/R'$
R_5	R''
R_6	$0.002R''$

TABLE 1.5: Circuit Parameters

The following code plots the SPICE simulation results from SPICE plot .dat file and spice .net file

```
codes/spice/plotter.py
```

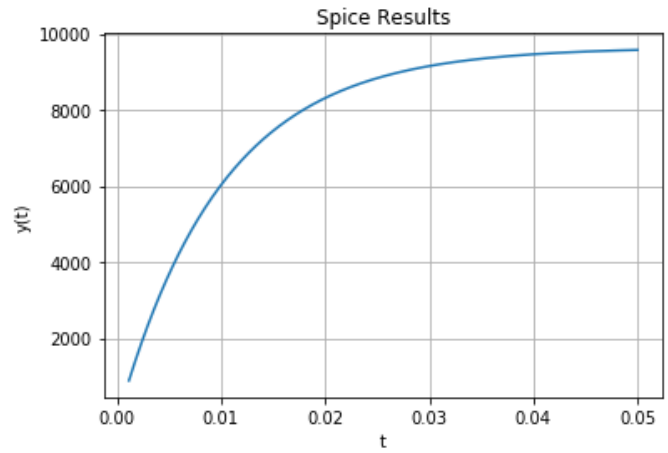


Fig. 1.6: SPICE simulation of circuit 1.5

The following is the spice simulation file

```
codes/spice/spice.net
```

For instructions to run the spice simulation please refer

```
codes/spice/readme.md
```