

EP2A04 Tutorial 1

TUTORIAL TAS:

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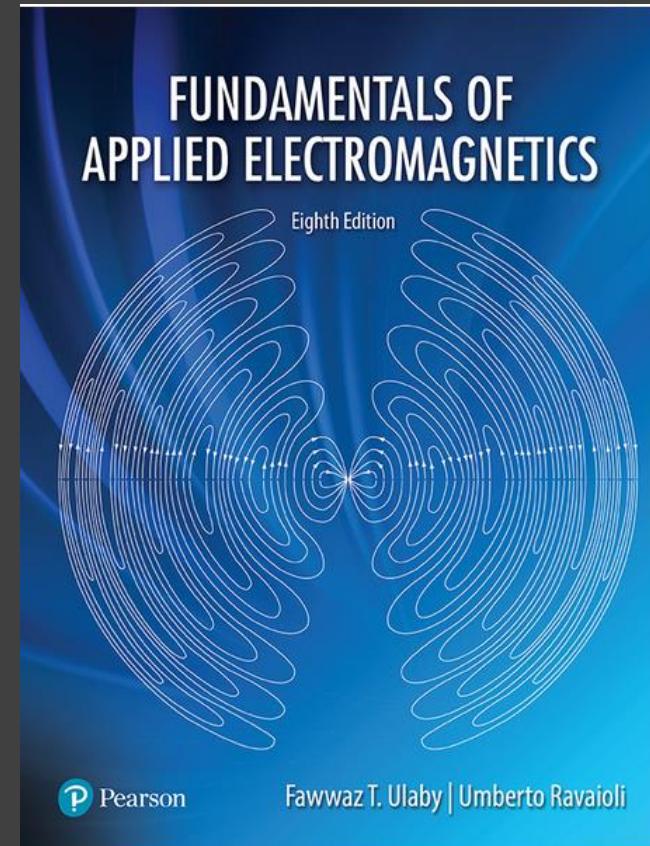
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Your Textbook

Fundamentals of Applied Electromagnetics,
Eighth Edition

Ulaby & Ravaioli

Seventh edition also acceptable, with some
inconsistencies



Tutorial Problem 1 (1.1, or 1.3 in 7th ed.)

A harmonic wave traveling along a string is generated by an oscillator that completes 180 vibrations per minute. If it is observed that a given crest, or maximum, travels 300 cm in 10 s, what is the wavelength?

Goal: **Wavelength**

Givens: **Frequency, Phase Velocity**

$$u_P = f\lambda \rightarrow \lambda = \frac{u_P}{f}$$

Convert to SI: $f = 180 \frac{1}{60s} = 3 \frac{1}{s} = 3 \text{ Hz}$ $u_P = \frac{300 \text{ cm}}{10 \text{ s}} = \frac{3 \text{ m}}{10 \text{ s}} = 0.3 \frac{\text{m}}{\text{s}}$

Solution: $\lambda = \frac{u_P}{f} = \frac{0.3}{3} = 0.1 \text{ m}$

Tutorial Problem 2 (1.2)

For the pressure wave described in Example 1-1, plot the following:

- A) $p(x, t)$ vs x at $t = 0$.
- B) $p(x, t)$ vs t at $x = 0$.

Be sure to use appropriate scales for and so that each of your plots covers at least two cycles.

The equation: $p(x, t) = 10 \cos\left(2\pi \times 10^3 t - \frac{4\pi}{3}x + \frac{\pi}{3}\right)$

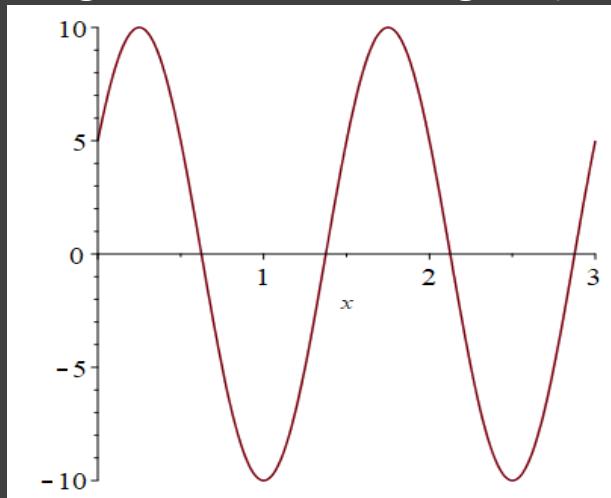
Tutorial Problem 2 (1.2) continued...

The equation: $p(x, t) = 10 \cos\left(2\pi \times 10^3 t - \frac{4\pi}{3}x + \frac{\pi}{3}\right)$

A) Substituting $t = 0$:

$$p(x, 0) = 10 \cos\left(-\frac{4\pi}{3}x + \frac{\pi}{3}\right)$$

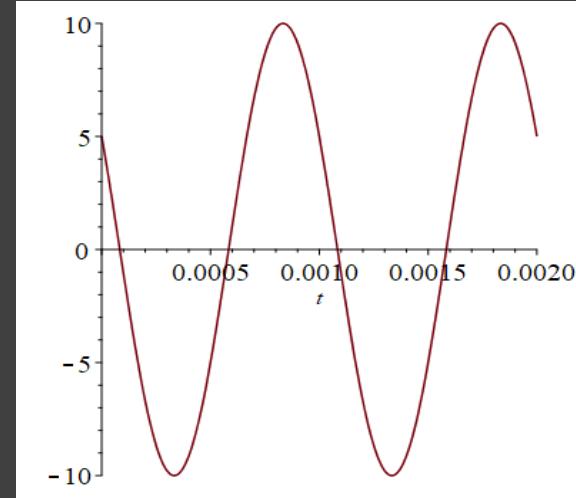
Plotting this for 2 wavelengths ($\lambda = \frac{3}{2}$):



B) Substituting $x = 0$:

$$p(0, t) = 10 \cos\left(2\pi \times 10^3 t + \frac{\pi}{3}\right)$$

Plotting this for 2 periods ($T = \frac{1}{f} = 0.001s$):



Tutorial Problem 3 (1.8)

Two waves on a string are given by the functions:

$$y_1(x, t) = 4 \cos(20t - 30x) \quad (\text{cm})$$

$$y_2(x, t) = -4 \cos(20t + 30x) \quad (\text{cm})$$

Where x is in centimeters. The waves are said to interfere constructively when their superposition $|y_S| = |y_1 + y_2|$ is a maximum, and they interfere destructively when $|y_S|$ is a minimum.

- A) What are the directions of propagation of waves $y_1(x, t)$ and $y_2(x, t)$?
- B) At $t = (\pi/50) \text{ s}$, what location x do the two waves interfere constructively, and what is the corresponding value of $|y_S|$?
- C) At $t = (\pi/50) \text{ s}$, what location x do the two waves interfere destructively, and what is the corresponding value of $|y_S|$?

Tutorial Problem 3 (1.8) *continued...*

A) What are the directions of propagation of waves $y_1(x, t)$ and $y_2(x, t)$?

$$y_1(x, t) = 4 \cos(20t - 30x) \quad (\text{cm})$$

The t and x terms have opposite signs, so the wave is propagating in the positive x -direction.

$$y_2(x, t) = -4 \cos(20t + 30x) \quad (\text{cm})$$

The t and x terms have the same sign, so the wave is propagating in the negative x -direction.

Tutorial Problem 3 (1.8) *continued...*

B) At $t = (\pi/50)$ s, what location x do the two waves interfere constructively, and what is the corresponding value of $|y_S|$?

First, substitute in $t = (\pi/50)$ s:

$$y_1(x, \pi/50) = 4 \cos\left(20\left(\frac{\pi}{50}\right) - 30x\right) = 4 \cos\left(\frac{2\pi}{5} - 30x\right)$$

$$y_2(x, \pi/50) = -4 \cos\left(20\left(\frac{\pi}{50}\right) + 30x\right) = -4 \cos\left(\frac{2\pi}{5} + 30x\right)$$

$$\left|y_S(x, \frac{\pi}{50})\right| = \left|4 \cos\left(\frac{2\pi}{5} - 30x\right) - 4 \cos\left(\frac{2\pi}{5} + 30x\right)\right| \quad \rightarrow \quad 2 \sin(\alpha) \sin(\beta) = \cos(\alpha - \beta) - \cos(\alpha + \beta)$$

$$\left|y_S\left(x, \frac{\pi}{50}\right)\right| = \left|4 \left(2 \sin\left(\frac{2\pi}{5}\right) \sin(30x)\right)\right| = \left|8 \sin\left(\frac{2\pi}{5}\right) \sin(30x)\right|$$

Tutorial Problem 3 (1.8) *continued...*

B) At $t = (\pi/50)$ s, what location x do the two waves interfere constructively, and what is the corresponding value of $|y_S|$?

$$\left|y_S\left(x, \frac{\pi}{50}\right)\right| = |7.61 \sin(30x)|$$

To find where they constructively interfere, we find the peaks of the superposition by differentiating and finding the zeros:

$$\frac{dy_S}{dx} = \frac{d}{dx}(7.61 \sin(30x)) = 30 * 7.61 \cos(30x) = 228.3 \cos(30x)$$

Find where $\frac{dy_S}{dx} = 0$. Cosine is 0 at $\theta = \left(n + \frac{1}{2}\right)\pi$, so:

$$x = \frac{\left(n + \frac{1}{2}\right)\pi}{30} = \frac{\pi}{60}, \frac{\pi}{20}, \frac{\pi}{12} \dots \text{(cm)}$$
$$x = 0.052\text{cm}, 0.157\text{cm}, 0.262\text{cm} \dots$$

The amplitude of the superposition at these values is:
$$\left|y_S\left(\frac{\pi}{60}, \frac{\pi}{50}\right)\right| = \left|7.61 \sin\left(30 \frac{\pi}{60}\right)\right| = 7.61 \text{ cm}$$

Tutorial Problem 3 (1.8) *continued...*

C) At $t = (\pi/50)$ s, what location x do the two waves interfere destructively, and what is the corresponding value of $|y_S|$?

$$\left|y_S\left(x, \frac{\pi}{50}\right)\right| = |7.61 \sin(30x)|$$

By a similar logic, we can find the points of destructive interference by finding the zeros of the superposition. \sin is zero at $\theta = n\pi$, so they interfere destructively at:

$$x = \frac{n\pi}{30} = 0, \frac{\pi}{30}, \frac{\pi}{15} \dots \text{ (cm)}$$

$$x = 0\text{cm}, 0.105\text{cm}, 0.209\text{cm} \dots$$

The amplitude of the superposition at these values is:

$$\left|y_S\left(0, \frac{\pi}{50}\right)\right| = |7.61 \sin(30(0))| = 0 \text{ cm}$$

Tutorial Problem 4 (1.12)

Given two waves characterized by

$$y_1(t) = 3 \cos(\omega t)$$

$$y_2(t) = 3 \sin(\omega t + 60^\circ),$$

Does $y_2(t)$ lead or lag $y_1(t)$ and by what phase angle?

Sin(x) lag cos(x) by 90° . Adding in a phase shift of $+60^\circ$ will create a lead of 60° . The total effect is $y_2(t)$ lagging $y_1(t)$ by 30° .

Good luck!

Assignment 1 is due 8AM on January 24.

No late submissions will be accepted – if your work is incomplete, submit whatever you have before the deadline for part marks!

Question 5 is bonus, to help us know where the class stands with circuits content.

Show all of your work for full marks.

ENG PHYS 2A04 Tutorial 2

Electricity and Magnetism

Your TAs today

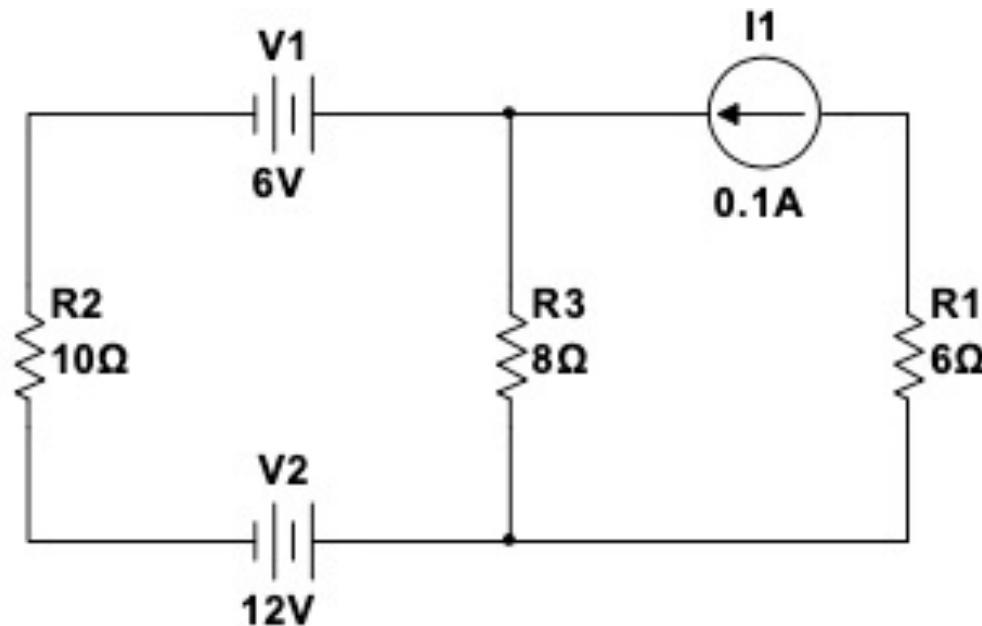
- Tommy Kean
keant1@mcmaster.ca
- Muhammad Munir
munirm6@mcmaster.ca

Chapter 28: DC Circuits

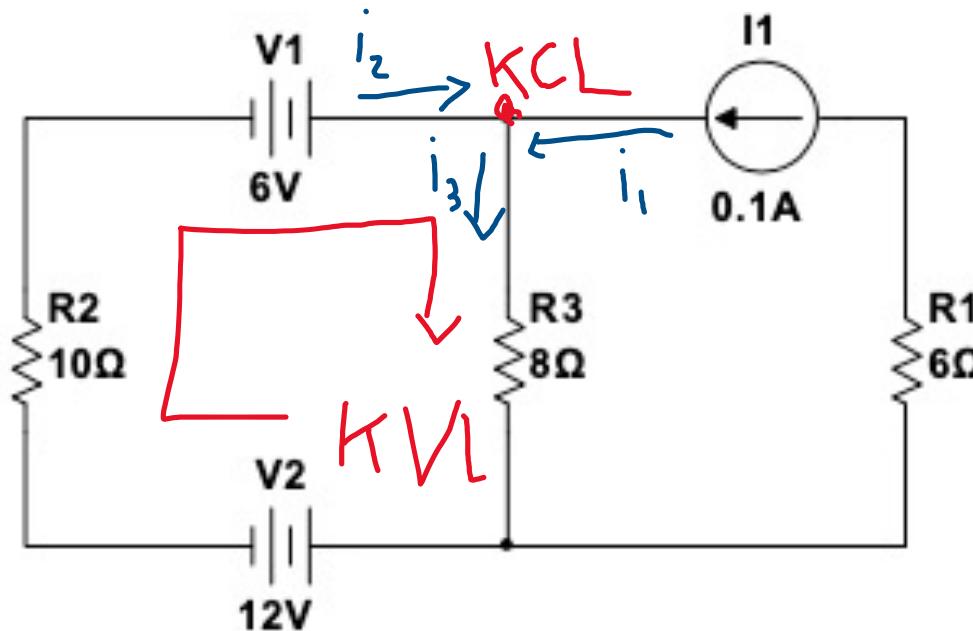
(Serway, 9th edition)

Problem 1 Example in Lecture

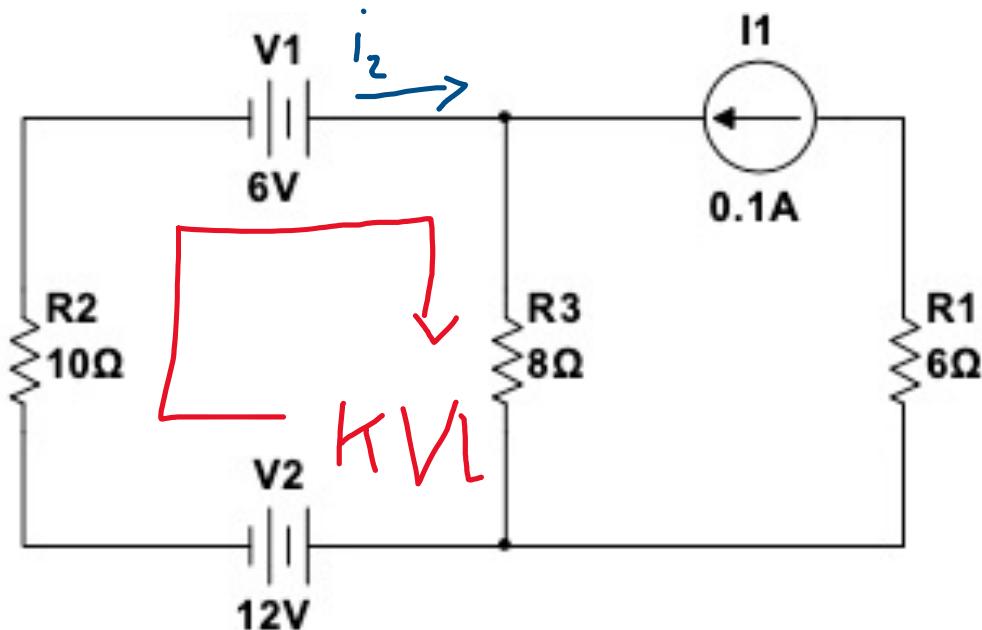
Find the current across R_1 , R_2 and R_3



Example in Lecture



Example in Lecture



Step 1 - Loop Rule (KCL) :

$$\sum_{closed\ loop} \Delta V = 0$$

$$V_1 + V_{R_3} + V_2 + V_{R_2} = 0$$

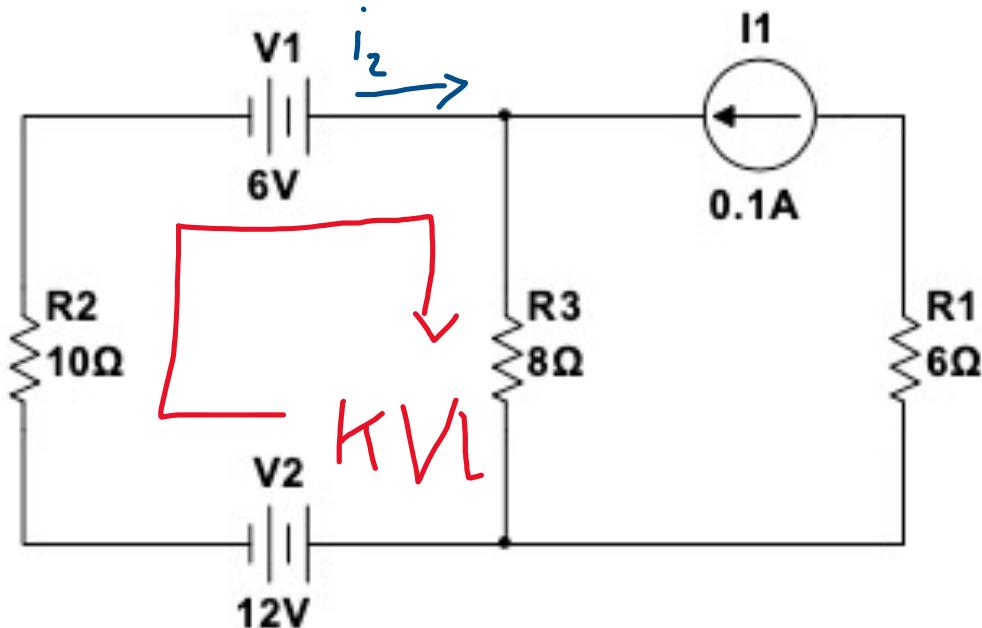
$$V_1 + (i_3 R_3) + V_2 + (i_2 R_2) = 0$$

$$V_1 + i_3 R_3 + V_2 + i_2 R_2 = 0$$

$$(-6) + (12) + i_3(8) + i_2(10) = 0$$
$$6 + 8i_3 + 10i_2 = 0$$

Example in Lecture

Step 2 - Junction Rule :



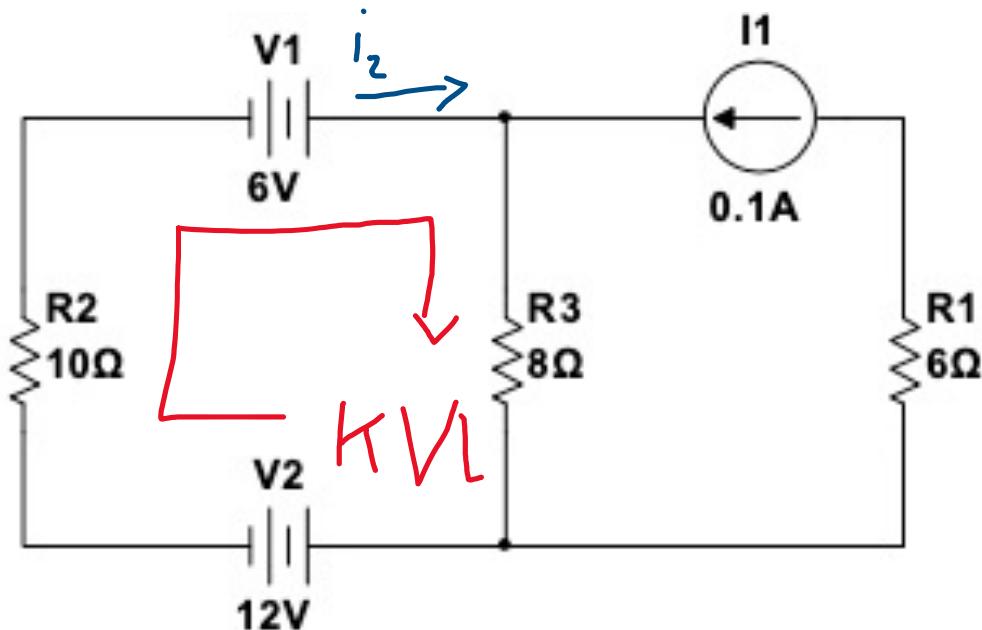
$$\sum_{junction} i = 0$$

$$i_1 + i_2 - i_3 = 0$$

$$0.1 + i_2 - i_3 = 0$$

$$i_3 = 0.1 + i_2$$

Example in Lecture



Step 3 - Substitute:

$$\begin{aligned}6 + 8i_3 + 10i_2 &= 0 \\6 + 8(0.1 + i_2) + 10i_2 &= 0 \\6 + 0.8 + 8i_2 + 10i_2 &= 0 \\6.8 + 18i_2 &= 0 \\i_2 &= \frac{-6.8}{18} \\i_2 &= -0.377\end{aligned}$$

$$\begin{aligned}i_3 &= 0.1 + i_2 \\i_3 &= 0.1 + (-0.377) \\i_3 &= -0.277\end{aligned}$$

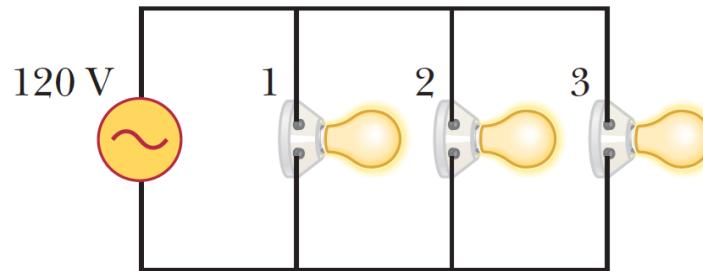
$$i_1 = 0.1 \text{ A}, i_2 = -0.377 \text{ A}, i_3 = -0.277 \text{ A}$$

Chapter 33: AC Circuits

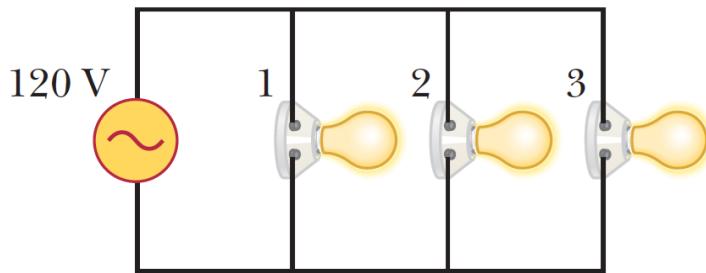
(Serway, 9th edition)

Problem 2

The figure below shows three lightbulbs connected to a 120-V AC (rms) household supply voltage. Bulb 1 has a power rating of 40 W, bulb 2 has a 75 W rating and bulb 3 has a 60 W rating. Find (a) the rms current in each bulb and (b) the resistance of each bulb. (c) What is the total resistance of the combination of the three lightbulbs:



Problem 2



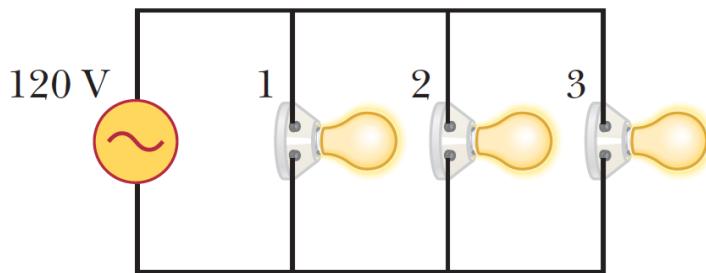
Find the RMS Current:

$$I_{1,rms} = \frac{P_1}{V} = \frac{40W}{120V} = 0.33A$$

$$I_{2,rms} = \frac{P_2}{V} = \frac{75W}{120V} = 0.625A$$

$$I_{3,rms} = \frac{P_3}{V} = \frac{60W}{120V} = 0.5A$$

Problem 2



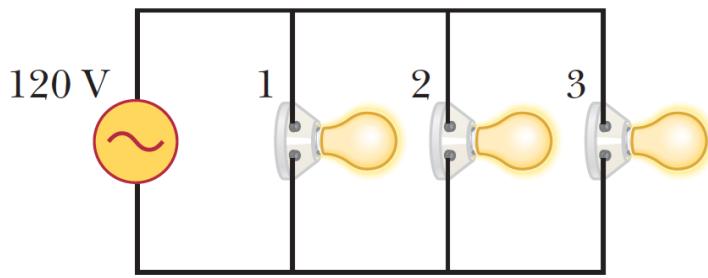
Find the resistance of each bulb :

$$R_1 = \frac{V^2}{P_1} = \frac{(120V)^2}{40W} = 360 \Omega$$

$$R_2 = \frac{V^2}{P_2} = \frac{(120V)^2}{75W} = 192 \Omega$$

$$R_3 = \frac{V^2}{P_3} = \frac{(120V)^2}{60W} = 240 \Omega$$

Problem 2



the total resistance of the combination of the three lightbulbs :

$$\frac{1}{R_{total}} = \frac{1}{R_1} + \frac{1}{R_2} + \frac{1}{R_3}$$

$$R_{total} = \left(\frac{1}{360} + \frac{1}{192} + \frac{1}{240} \right)^{-1} = 82.3 \Omega$$

Problem 3

What maximum current is delivered by an AC source M with $\Delta V_{max} = 48.0 \text{ V}$ and $f=90.0 \text{ Hz}$ when connected across a $3.70\text{-}\mu\text{F}$ capacitor?

Problem 3

What maximum current is delivered by an AC source with $\Delta V_{max} = 48.0 \text{ V}$ and $f=90.0 \text{ Hz}$ when connected across a $3.70\text{-}\mu\text{F}$ capacitor?

1. Find the V_{RMS}

$$\Delta V_{RMS} = \frac{\Delta V_{max}}{\sqrt{(2)}}$$

$$\Delta V_{RMS} = 33.94 \text{ V}$$

Problem 3

What maximum current is delivered by an AC source M with $\Delta V_{max} = 48.0 \text{ V}$ and $f=90.0 \text{ Hz}$ when connected across a $3.70\text{-}\mu\text{F}$ capacitor?

2. Find the reactance

$$x_c = \frac{1}{2\pi f C}$$

$$x_c = \frac{1}{2\pi(90 \text{ Hz})(3.7 \times 10^{-6})}$$

$$x_c = 477.94\Omega$$

Problem 3

What maximum current is delivered by an AC source M with $\Delta V_{max} = 48.0 \text{ V}$ and $f=90.0 \text{ Hz}$ when connected across a $3.70\text{-}\mu\text{F}$ capacitor?

3. Find the I_{rms}

$$I_{rms} = \frac{\Delta V_{RMS}}{x_c}$$

$$I_{rms} = \frac{33.94 \text{ V}}{477.94 \Omega}$$

$$I_{rms} = 0.071 \text{ A}$$

Problem 3

What maximum current is delivered by an AC source M with $\Delta V_{max} = 48.0 \text{ V}$ and $f=90.0 \text{ Hz}$ when connected across a $3.70\text{-}\mu\text{F}$ capacitor?

3. Find the I_{rms}

$$I_{peak} = I_{rms}\sqrt{2}$$

$$I_{peak} = (0.071 \text{ A})\sqrt{2}$$

$$I_{peak} = 0.100 \text{ A}$$

Section 1-7: Review of Phasors (Ulaby, 8th ed.)

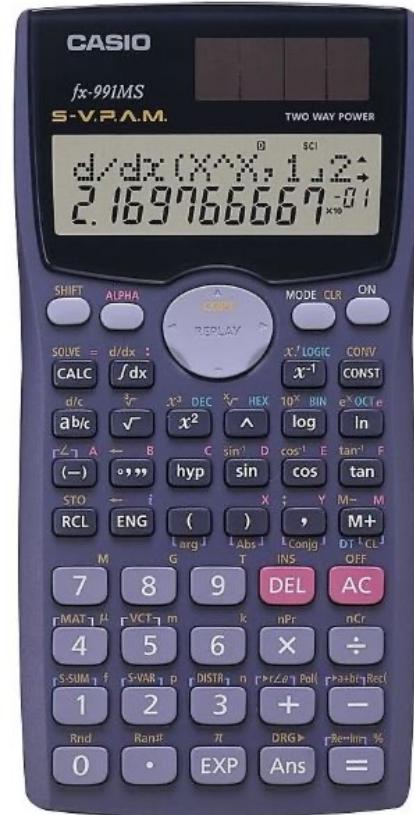
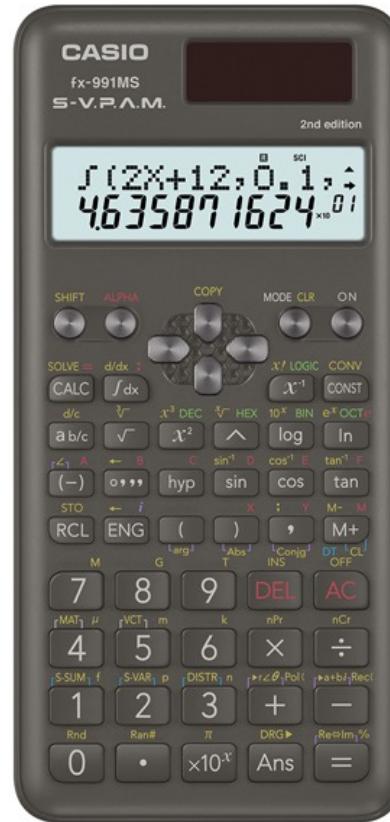
McMaster's Standard Calculator

- Casio FX-991MS

- 2nd Edition (left) available at McMaster Bookstore: \$21.99 (link below)

- Will need to solve phasor equations

<https://campusstore.mcmaster.ca/cgi-mcm/ws/gmdetail.pl?pwsPRODIDG1=2207018&sType=qm&prod esc=Casio%20fx%2D991%20Calculator&pwsGROUP=>



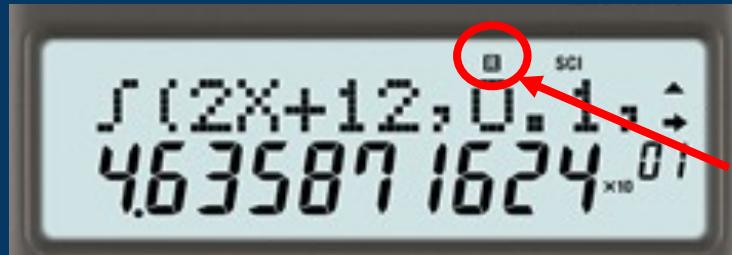
Phasor Conversion – Calculator Setup

Enable complex numbers:

- “CMPLX” appears at top of display

Ensure calculator is in degree mode

- For consistent units in polar form



BAD!



GOOD!

MODE CLR



S-VAR p



MODE CLR

x4

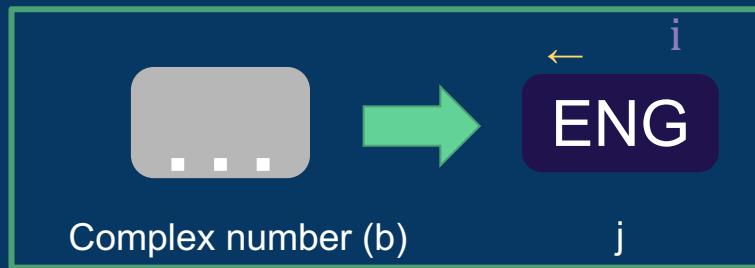


S-SUM f

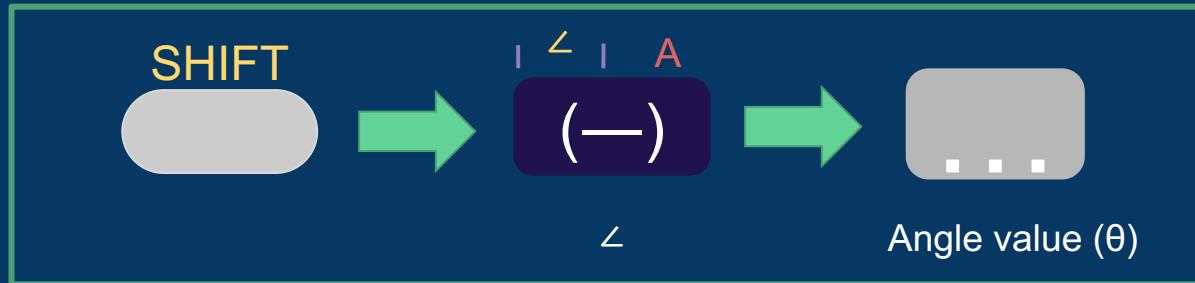


Phasor Conversion – Inputting Complex Values

To enter complex number in rectangular form:

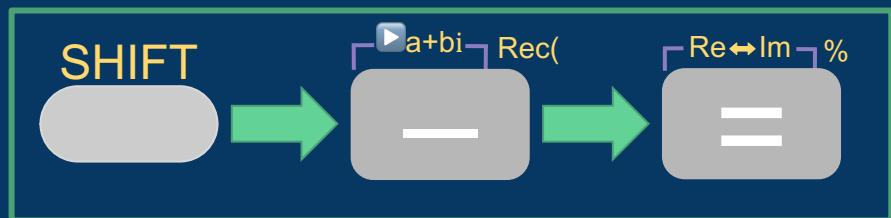


To enter angle value in polar form:

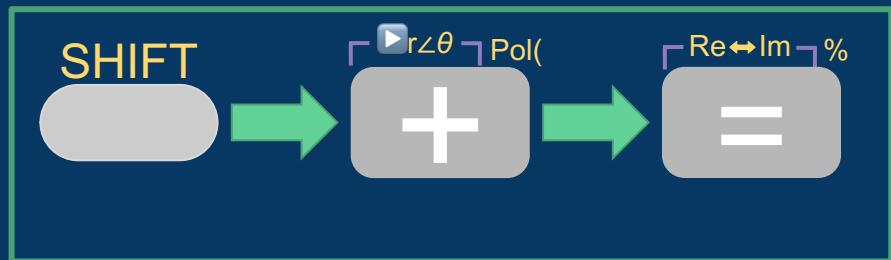


Phasor Conversion – Output Viewing

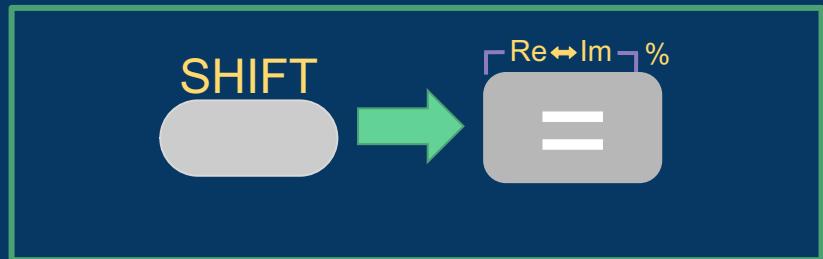
To view answer in rectangular form:



To view answer in polar form:



To change between real and imaginary (rectangular) or radius and angle (polar):



Phasor Conversion - Question

Convert the following phasors to rectangular form ($a + bj$):

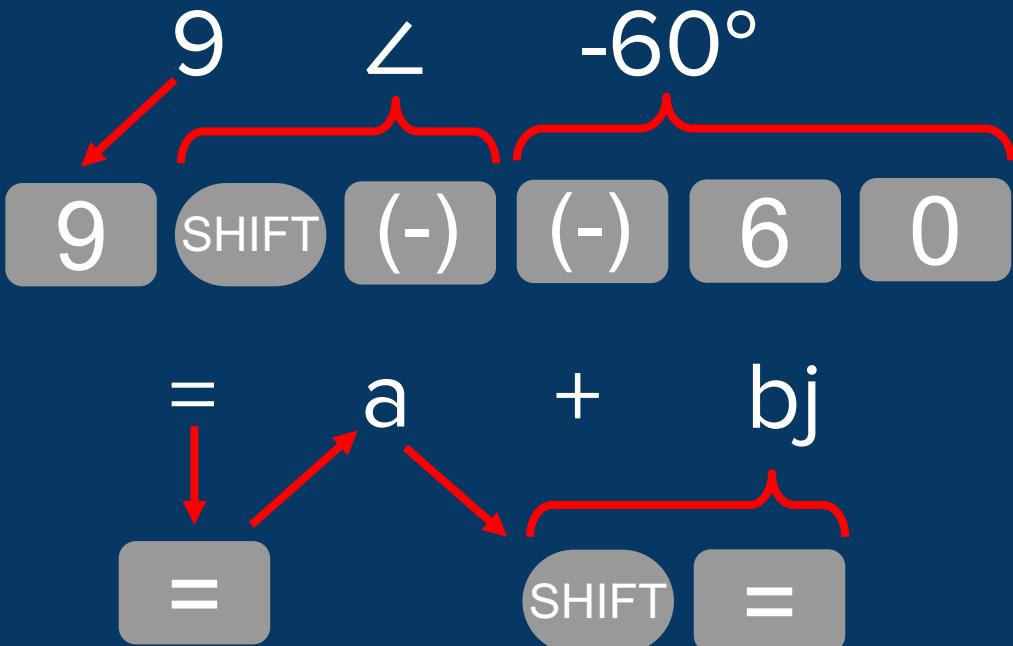
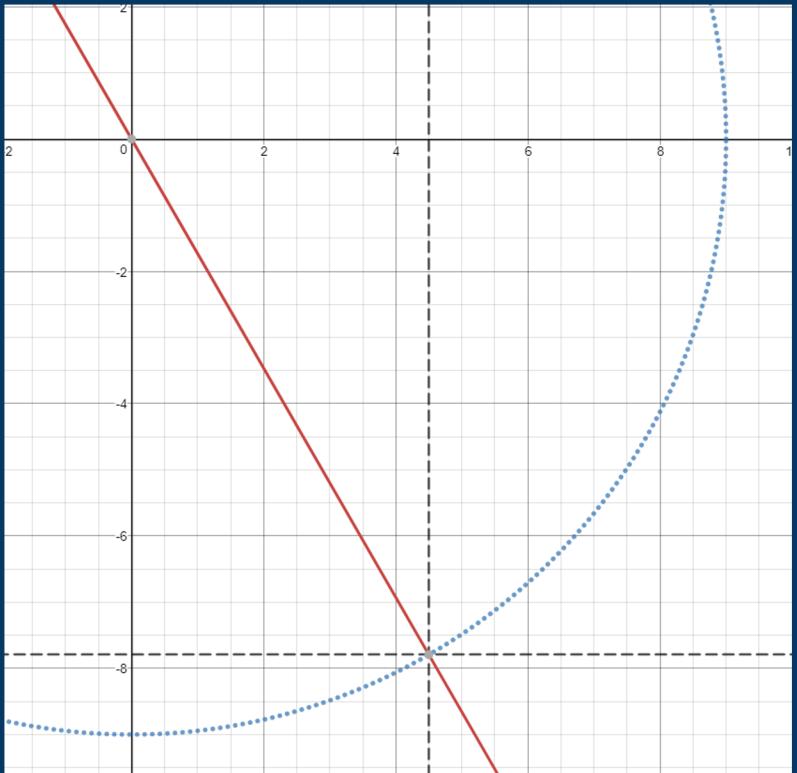
- a) $9 \angle -60^\circ$
- b) $12 \angle -45^\circ$

Convert the following phasors to polar form ($r \angle \theta$):

- a) $3 + 9j$
- b) $-0.4 + 0.3j$

Phasor Conversion - Solution

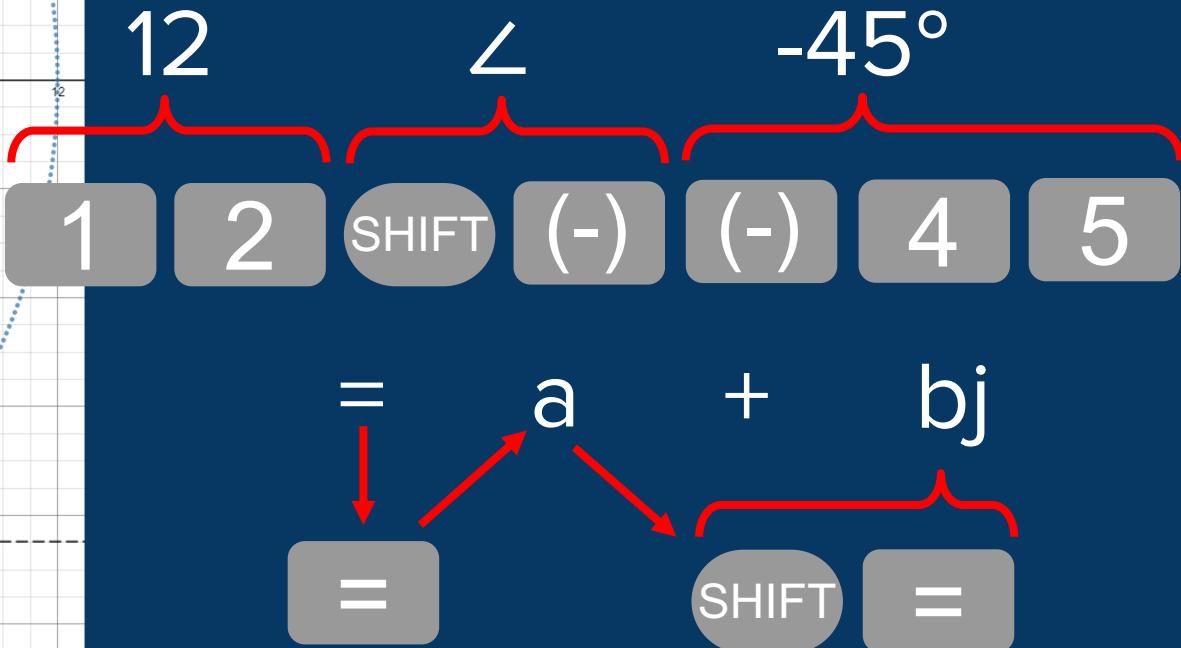
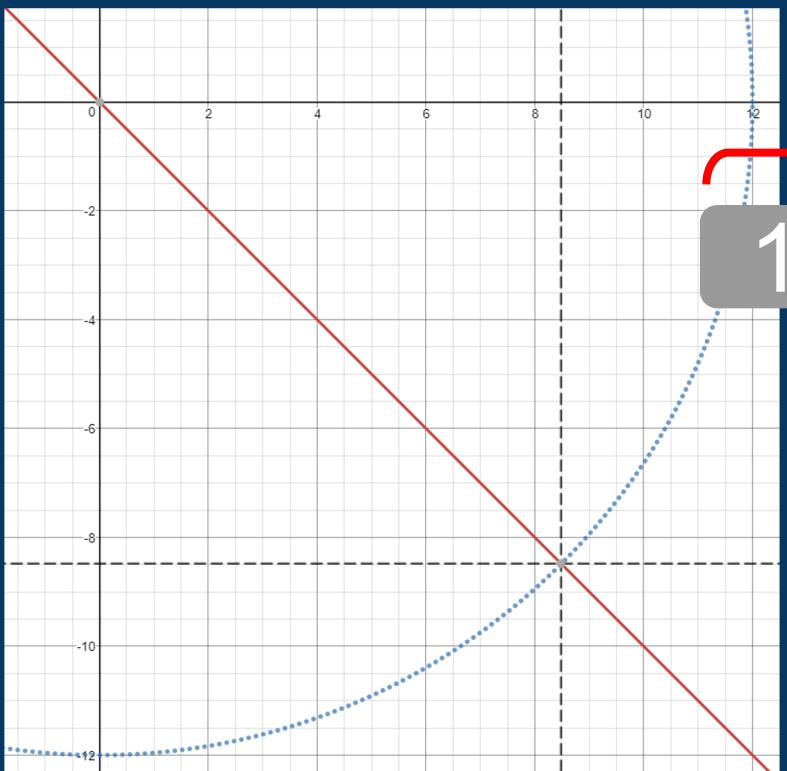
a) Convert the following to rectangular form ($a + bj$):



Answer: $4.5 - 7.79j$

Phasor Conversion - Solution

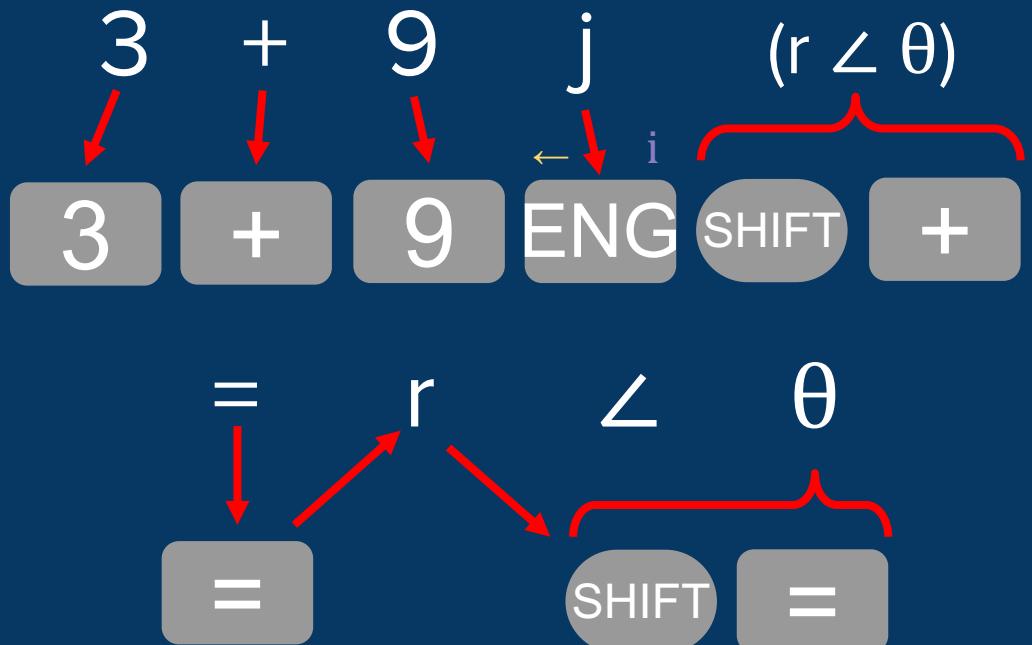
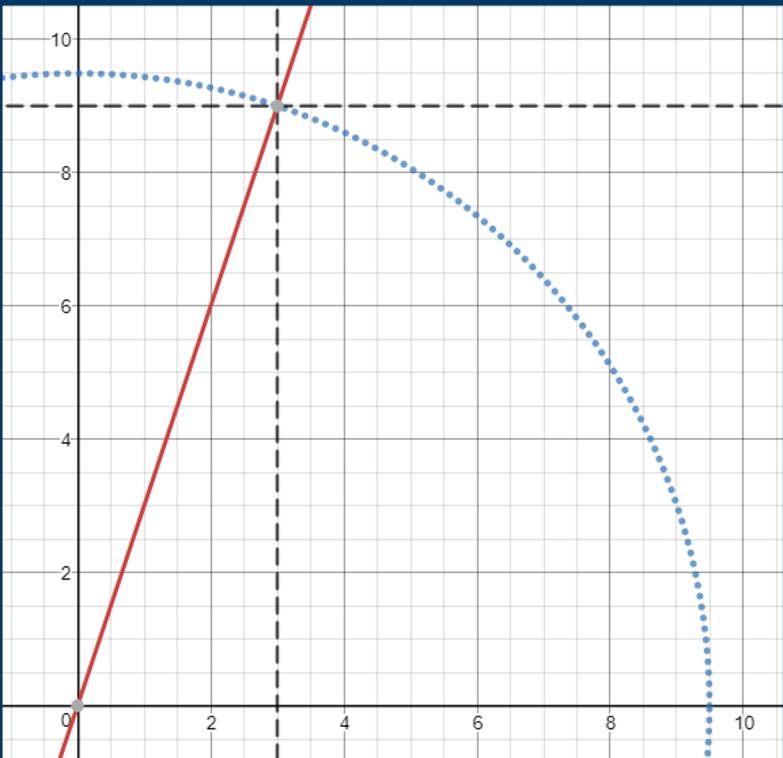
b) Convert the following to rectangular form ($a + bj$):



Answer: $8.48 - 8.48j$

Phasor Conversion - Solution

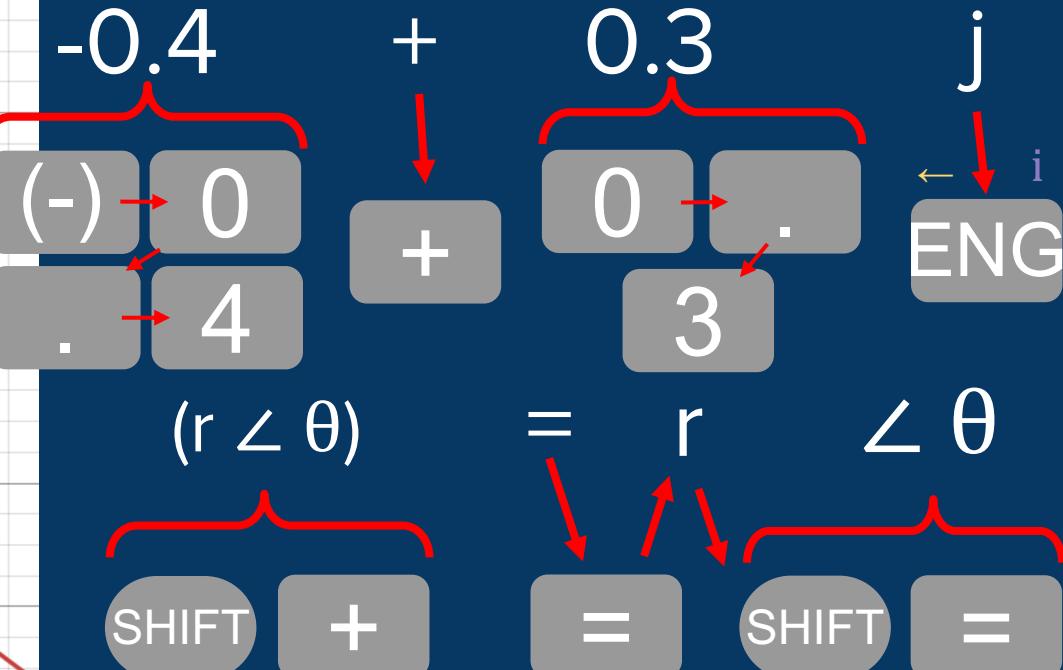
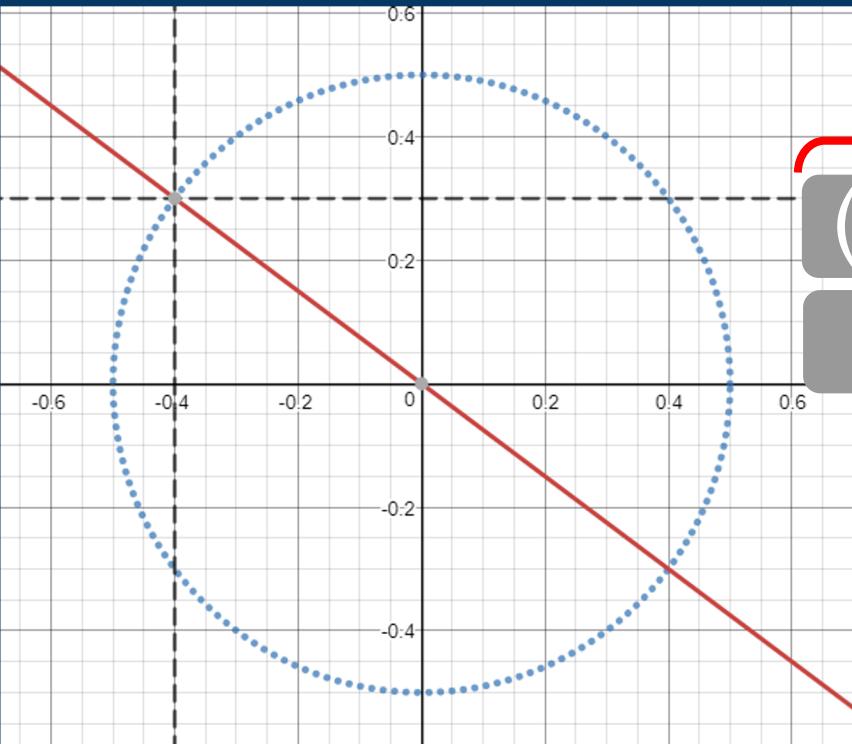
a) Convert the following to polar form ($r \angle \theta$):



Answer: $9.49 \angle 71.56^\circ$

Phasor Conversion - Solution

a) Convert the following to polar form ($r \angle \theta$):



Answer: $0.5 \angle 143.13^\circ$

Problem 1.28 – Question

Find the phasors of the following time functions:

$$a) \quad v(t) = 9 \cos\left(\omega t - \frac{\pi}{3}\right) \text{ (V)}$$

$$b) \quad v(t) = 12 \sin\left(\omega t + \frac{\pi}{4}\right) \text{ (V)}$$

$$c) \quad i(x, t) = 5e^{-3x} \sin\left(\omega t + \frac{\pi}{6}\right) \text{ (A)}$$

$$d) \quad i(t) = -2 \cos\left(\omega t + \frac{3\pi}{4}\right) \text{ (A)}$$

$$e) \quad i(t) = 4 \sin\left(\omega t + \frac{\pi}{3}\right) + 3 \cos\left(\omega t - \frac{\pi}{6}\right) \text{ (A)}$$

Problem 1.28 – Solution (a)

a) $v(t) = 9 \cos\left(\omega t - \frac{\pi}{3}\right)$ (V)

Adopt cosine reference ($v(t) = V_0 \cos(\omega t + \phi)$). Is this already done?

- Yes! $V_0 = 9$, $\phi = -\frac{\pi}{3}$

From sinusoidal time domain to phasor domain:

- $v(t) = V_0 \cos(\omega t + \phi) \rightarrow \mathbf{V} = V_0 e^{j\phi}$

$$\therefore v(t) = 9 \cos\left(\omega t - \frac{\pi}{3}\right) \text{ (V)} \rightarrow \mathbf{V}^{\sim} = 9e^{-j\frac{\pi}{3}} \text{ (V)}$$

Problem 1.28 – Solution (b)

b) $v(t) = 12 \sin\left(\omega t + \frac{\pi}{4}\right)$ (V)

Adopt cosine reference ($v(t) = V_0 \cos(\omega t + \phi)$). Is this already done?

- No! $\rightarrow v(t) = 12 \sin\left(\omega t + \frac{\pi}{4}\right) = 12 \cos\left(\omega t + \frac{\pi}{4} - \frac{\pi}{2}\right) = 12 \cos\left(\omega t - \frac{\pi}{4}\right)$
- $V_0 = 12$, $\phi = -\frac{\pi}{4}$

Sinusoidal time domain \rightarrow phasor domain: $v(t) = V_0 \cos(\omega t + \phi) \rightarrow \mathbf{V} = V_0 e^{j\phi}$

$$\therefore v(t) = 12 \sin\left(\omega t + \frac{\pi}{4}\right) \text{ (V)} = 12 \cos\left(\omega t - \frac{\pi}{4}\right) \text{ (V)} \rightarrow \mathbf{V}^{\sim} = \mathbf{12e}^{-j\frac{\pi}{4}} \text{ (V)}$$

Problem 1.28 – Solution (c)

c) $i(x, t) = 5e^{-3x} \sin\left(\omega t + \frac{\pi}{6}\right)$ (A)

Adopt cosine reference ($i(t) = V_0 \cos(\omega t + \phi)$). Is this already done?

- No! $\rightarrow i(x, t) = 5e^{-3x} \sin\left(\omega t + \frac{\pi}{6}\right) = 5e^{-3x} \cos\left(\omega t + \frac{\pi}{6} - \frac{\pi}{2}\right) = 5e^{-3x} \cos\left(\omega t - \frac{\pi}{3}\right)$
- $V_0 = 5e^{-3x}$, $\phi = -\frac{\pi}{3}$

Sinusoidal time domain \rightarrow phasor domain: $v(t) = V_0 \cos(\omega t + \phi) \rightarrow \mathbf{V} = V_0 e^{j\phi}$

$$\therefore i(x, t) = 5e^{-3x} \sin\left(\omega t + \frac{\pi}{6}\right) \text{ (A)} = 5e^{-3x} \cos\left(\omega t - \frac{\pi}{3}\right) \text{ (A)} \rightarrow \mathbf{V}^{\sim} = 5e^{-3x} e^{-j\frac{\pi}{3}} \text{ (V)}$$

Problem 1.28 – Solution (d)

d) $i(t) = -2 \cos\left(\omega t + \frac{3\pi}{4}\right)$ (A)

Adopt cosine reference ($i(t) = I_0 \cos(\omega t + \phi)$). Is this already done?

- Yes! $I_0 = -2$, $\phi = \frac{3\pi}{4}$
- One missing step: $-1 = e^{-j\pi} \rightarrow I_0 = -2 = 2(-1) = 2e^{-j\pi}$

Sinusoidal time domain \rightarrow phasor domain $i(t) = I_0 \cos(\omega t + \phi) \rightarrow \mathbf{I} = I_0 e^{j\phi}$

$$\therefore i(t) = -2 \cos\left(\omega t + \frac{3\pi}{4}\right) \text{ (A)} \rightarrow \mathbf{I} = 2e^{-j\pi} e^{j\frac{3\pi}{4}} \text{ (A)} \rightarrow \mathbf{I} = 2e^{-j\frac{\pi}{4}} \text{ (A)}$$

Problem 1.28 – Solution (e)

e) $i(t) = 4 \sin\left(\omega t + \frac{\pi}{3}\right) + 3 \cos\left(\omega t - \frac{\pi}{6}\right)$ (A)

Adopt cosine reference ($i(t) = I_0 \cos(\omega t + \phi)$). Is this already done?

- Kinda... $\rightarrow 4 \sin\left(\omega t + \frac{\pi}{3}\right) = 4 \cos\left(\omega t + \frac{\pi}{3} - \frac{\pi}{2}\right) = 4 \cos\left(\omega t - \frac{\pi}{6}\right)$

- $I_0 = 4, \phi = -\frac{\pi}{6}$ and $I_0 = 3, \phi = -\frac{\pi}{6}$

Sinusoidal time domain \rightarrow phasor domain $i(t) = I_0 \cos(\omega t + \phi) \rightarrow \mathbf{I} = I_0 e^{j\phi}$

$$\therefore i(t) = 4 \sin\left(\omega t + \frac{\pi}{3}\right) + 3 \cos\left(\omega t - \frac{\pi}{6}\right) \text{ (A)} \rightarrow \mathbf{I} = 4e^{-j\frac{\pi}{6}} + 3e^{-j\frac{\pi}{6}} \rightarrow \mathbf{I} = 7e^{-j\frac{\pi}{6}} \text{ (A)}$$

ENGPHYS 2A04

Tutorial 3

Electricity and Magnetism

Your TAs Today

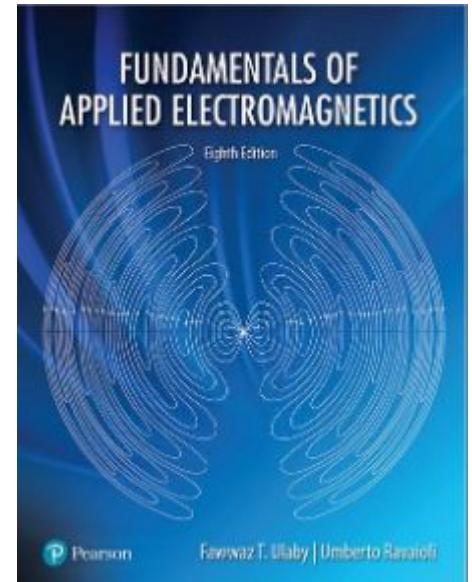
- Joanne Lee
leej298@mcmaster.ca
- Fatemeh Bakhshandeh
bakhshaf@mcmaster.ca

Your Textbook

Fundamentals of Applied Electromagnetics Eighth Edition.

Ulaby & Ravaioli

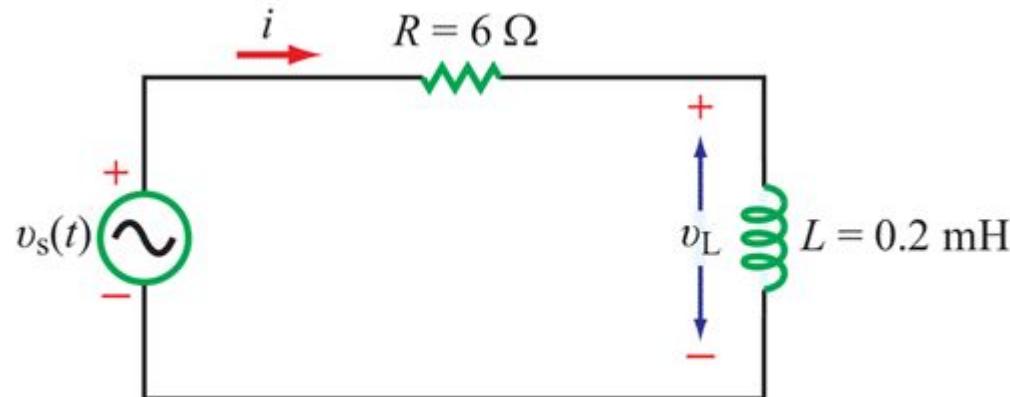
Seventh edition also acceptable, with some inconsistencies



Lecture Problem (Example 1-4 Page 40)

The voltage source of the RL circuit shown below is given by:

$V_s(t) = 5 \sin(4 \cdot 10^4 t - 30^\circ)$. Find an expression for the voltage across the inductor.



Lecture Problem Solution

Before converting to the phasor domain, we express the initial equation in terms of cosine.

$$\begin{aligned}v_s(t) &= 5 \sin(4 \times 10^4 t - 30^\circ) \\&= 5 \cos(4 \times 10^4 t - 120^\circ) \quad (\text{V}).\end{aligned}$$

The corresponding voltage phasor equation is

$$\tilde{V}_s = 5e^{-j120^\circ} \quad (\text{V}),$$

Lecture Problem Solution

The voltage loop equation of the RL circuit is

$$Ri + L \frac{di}{dt} = v_s(t).$$

The corresponding phasor equation is $R\tilde{I} + j\omega L\tilde{I} = \tilde{V}_s$.

Solving for current phasor I the equation becomes:

$$I = \frac{V_s}{R + j\omega L}$$

$$I = \frac{5e^{-j120^\circ}}{6 + j8}$$

$$I = \frac{5e^{-j120^\circ}}{6 + j4 * 10^4 * 2 * 10^{-4}}$$

$$I = 0.5e^{-j173.1^\circ}$$

$$I = \frac{5e^{-j120^\circ}}{10e^{j53.1^\circ}}$$

Lecture Problem Solution

The voltage phasor across the inductor is related to the current I by

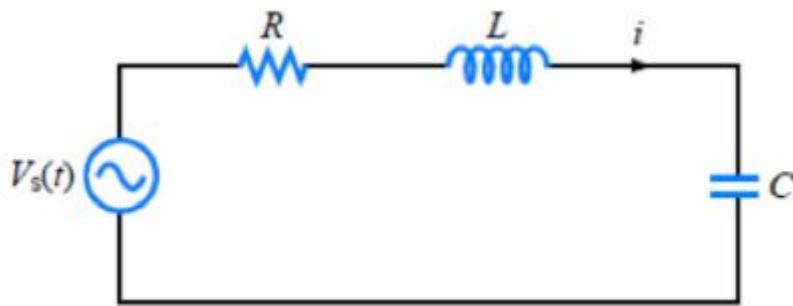
$$\begin{aligned}\tilde{V}_L &= j\omega L \tilde{I} \\ &= j4 \times 10^4 \times 2 \times 10^{-4} \times 0.5 e^{-j173.1^\circ} \\ &= 4e^{j(90^\circ - 173.1^\circ)} = 4e^{-j83.1^\circ} \quad (\text{V}),\end{aligned}$$

Corresponding Instantaneous Voltage $V_L(t)$ is:

$$\begin{aligned}v_L(t) &= \Re[\tilde{V}_L e^{j\omega t}] \\ &= \Re[4e^{-j83.1^\circ} e^{j4 \times 10^4 t}] \\ &= 4 \cos(4 \times 10^4 t - 83.1^\circ)\end{aligned}$$

Problem 33.4 - Phasor Method

A series RLC AC circuit has: $R = 425 \Omega$,
 $L = 1.25 \text{ H}$, $C = 3.50 \mu\text{F}$, $\omega = 377 \text{ s}^{-1}$, and
 $\Delta V_{\max} = 150 \text{ V}$.

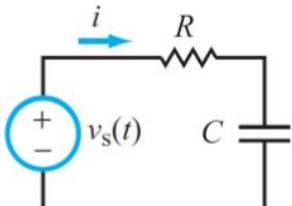


- Determine the inductive reactance and the capacitive reactance of the circuit.
- Find the impedance and phase angle between the current and voltage.
- Find the maximum current in the circuit.
- Find both the maximum voltage and the instantaneous voltage across each element.

AC Phasor Analysis

Step 1

Adopt Cosine Reference
(Time Domain)

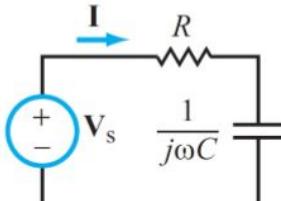


$$v_s(t) = 12 \sin(\omega t - 45^\circ) \text{ (V)}$$

Step 2

Transfer to Phasor Domain

- $i \rightarrow \mathbf{I}$
- $v \rightarrow \mathbf{V}$
- $R \rightarrow \mathbf{Z}_R = R$
- $L \rightarrow \mathbf{Z}_L = j\omega L$
- $C \rightarrow \mathbf{Z}_C = 1/j\omega C$



$$\mathbf{V}_s = 12e^{-j135^\circ} \text{ (V)}$$

Step 3

Cast Equations in
Phasor Form

$$\mathbf{I} \left(R + \frac{1}{j\omega C} \right) = \mathbf{V}_s$$

(apply Ohm's and Kirchoff's laws)

Step 4

Solve for Unknown Variable
(Phasor Domain)

$$\mathbf{I} = \frac{\mathbf{V}_s}{R + \frac{1}{j\omega C}}$$

Step 5

Transform Solution
Back to Time Domain

$$\begin{aligned} i(t) &= \Re[\mathbf{I} e^{j\omega t}] \\ &= I_0 \cos(\omega t - \phi_i) \text{ (A)} \end{aligned}$$

Example 33.4 Initial Steps

First create the time domain cosine equation for the given circuit using the information given.

$$V_s(t) = \Delta V_{max} \cos(\omega t + \phi)$$

$$V_s(t) = 150 \cos(377t + \phi)$$

Then transfer the equation into the phasor domain.

$$i \rightarrow \tilde{I}$$

$$L \rightarrow Z_L = j\omega L = jX_L$$

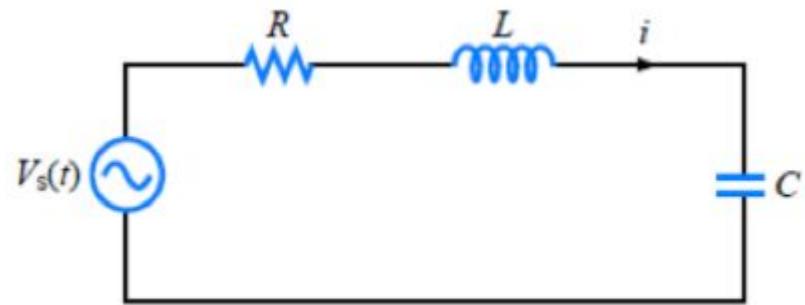
$$V_s(t) = \tilde{V} = \Delta V_{max} e^{j\phi}$$

$$R \rightarrow Z_R = R = 425 \Omega$$

$$C \rightarrow Z_C = \frac{1}{j\omega C} = -jX_C$$

Example 33.4 Solution to A)

A series RLC AC circuit has: $R = 425 \Omega$,
 $L = 1.25 \text{ H}$, $C = 3.50 \mu\text{F}$, $\omega = 377 \text{ s}^{-1}$, and
 $\Delta V_{\max} = 150 \text{ V}$.



- a) Determine the inductive reactance and capacitive reactance of circuit.

$$X_L = \omega L = (377 \text{ s}^{-1})(1.25 \text{ H})$$

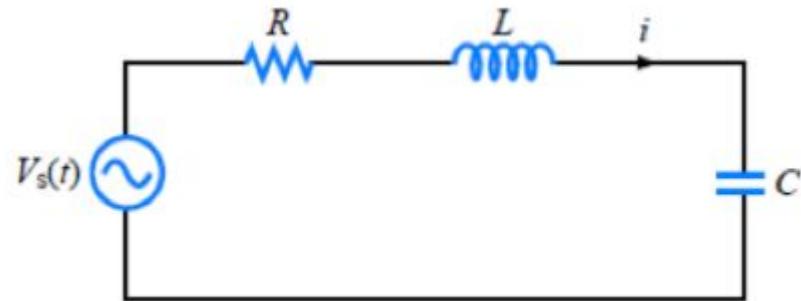
$$X_L = 471 \Omega$$

$$X_C = \frac{1}{\omega C} = \frac{1}{(377 \text{ s}^{-1})(3.50 * 10^{-6} \text{ F})}$$

$$X_C = 758 \Omega$$

Example 33.4 Solution to B)

A series RLC AC circuit has: $R = 425 \Omega$,
 $L = 1.25 \text{ H}$, $C = 3.50 \mu\text{F}$, $\omega = 377 \text{ s}^{-1}$, and
 $\Delta V_{\max} = 150 \text{ V}$.



- b) Find the impedance and phase angle between current and voltage

$$Z = Z_R + Z_L + Z_C = R + jX_L - jX_C$$

$$Z = 425 + 471j - 758j$$

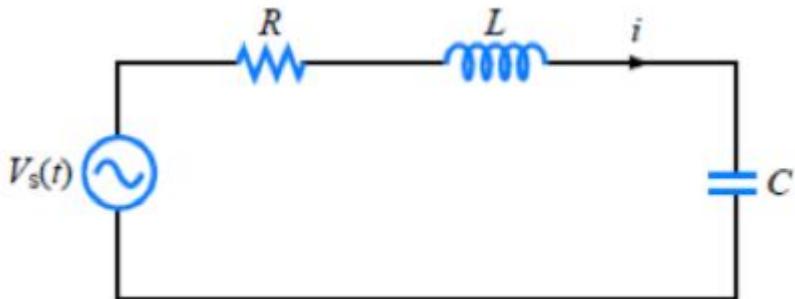
$$Z = 425 - 287j \Omega$$

Convert to phasor using method from before.

$$Z \cong 513 \Omega \angle -34.0^\circ$$

Example 33.4 Solution to C)

A series RLC AC circuit has: $R = 425 \Omega$,
 $L = 1.25 \text{ H}$, $C = 3.50 \mu\text{F}$, $\omega = 377 \text{ s}^{-1}$, and
 $\Delta V_{\max} = 150 \text{ V}$.



- c) Find the maximum current by applying Kirchoff's Rules in phasor form and solve for unknowns.

$$\tilde{V} = V_R + V_L + V_C$$

$$\tilde{I} = (Z_R + Z_L + Z_C) = \tilde{I}Z$$

$$\tilde{I} = \frac{\tilde{V}}{Z}$$

The max current occurs when: $\tilde{V} = \Delta V_{\max}$

$$\tilde{I}_{\max} = \frac{\Delta V_{\max}}{|Z|} = \frac{150 \text{ V}}{513 \Omega} = 0.292 \text{ A}$$

Example 33.4 Solution to D) Max Voltage

A series RLC AC circuit has: $R = 425 \Omega$,
 $L = 1.25 \text{ H}$, $C = 3.50 \mu\text{F}$, $\omega = 377 \text{ s}^{-1}$, and
 $\Delta V_{\max} = 150 \text{ V}$.

- c) Find both the maximum voltage and the instantaneous voltage across each element.

Max voltage occurs when: $\tilde{I} = \tilde{I}_{\max}$

$$\Delta V_R = \tilde{I}_{\max} Z_R = (0.292 \text{ A})(425 \Omega)$$

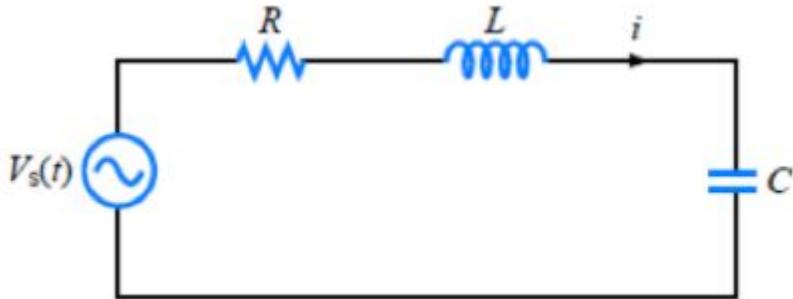
$$\Delta V_R = 124 \text{ V}$$

$$\Delta V_L = \tilde{I}_{\max} Z_L = (0.292 \text{ A})(471j \Omega)$$

$$\Delta V_L = 138j \text{ V}$$

$$\Delta V_C = \tilde{I}_{\max} Z_C = (0.292 \text{ A})(-758j \Omega)$$

$$\Delta V_C = -222j \text{ V}$$



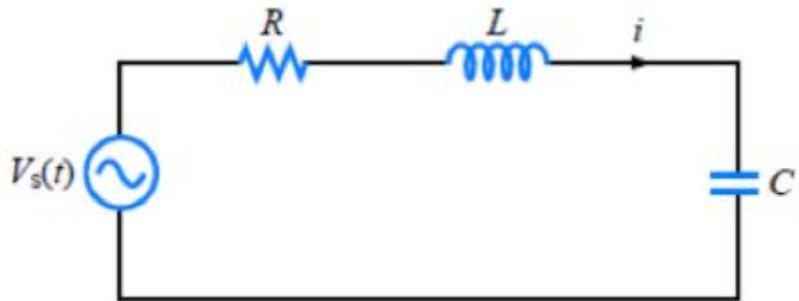
Example 33.4 Solution to D) Instantaneous Voltage

A series RLC AC circuit has: $R = 425 \Omega$,
 $L = 1.25 \text{ H}$, $C = 3.50 \mu\text{F}$, $\omega = 377 \text{ s}^{-1}$, and
 $\Delta V_{\max} = 150 \text{ V}$.

- c) Find both the maximum voltage and the instantaneous voltage across each element.

Transform all solutions back to time domain.

$$v(t) = \Re e [V_0 e^{j\phi} e^{j\omega t}] = V_0 \cos(\omega t + \phi)$$



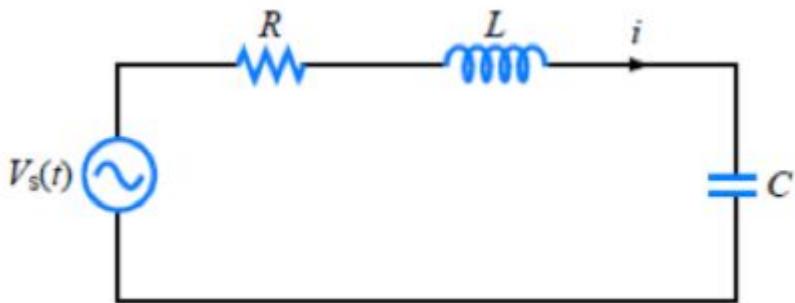
$$\Delta V_R = V_0 e^{j\phi} = 124 \text{ V}$$

$$\Delta V_L = V_0 e^{j\phi} = 138j = 138e^{\frac{j\pi}{2}} \text{ V}$$

$$\Delta V_C = V_0 e^{j\phi} = -222j = 222e^{-\frac{j\pi}{2}} \text{ V}$$

Example 33.4 Solution to D) Instantaneous Voltage

A series RLC AC circuit has: $R = 425 \Omega$, $L = 1.25 \text{ H}$, $C = 3.50 \mu\text{F}$, $\omega = 377 \text{ s}^{-1}$, and $\Delta V_{\max} = 150 \text{ V}$.



- c) Find both the maximum voltage and the instantaneous voltage across each element.

$$v(t) = \Re [V_0 e^{j\phi} e^{j\omega t}] = V_0 \cos(\omega t + \phi)$$

$$v_R(t) = \Re [\Delta V_R e^{j\omega t}] = \Re [124 e^{j377t}]$$

$$v_R(t) = 124 \cos 377t$$

$$v_L(t) = \Re [\Delta V_L e^{j\omega t}] = \Re \left[138 e^{\frac{j\pi}{2}} e^{j377t} \right]$$

$$v_L(t) = 138 \sin 377t$$

$$v_C(t) = \Re [\Delta V_C e^{j\omega t}] = \Re \left[222 e^{\frac{-j\pi}{2}} e^{j377t} \right]$$

$$v_C(t) = -222 \sin 377t$$

Example 2.1

A transmission line of length l connects to a sinusoidal voltage source with an oscillation frequency f . Assuming that the velocity of wave propagation on the line is c , for which of the following situations is it reasonable to ignore the presence of the transmission line in the solution of the circuit.

- a) $l = 20 \text{ cm}, f = 20 \text{ kHz}$,
- b) $l = 50 \text{ km}, f = 60 \text{ Hz}$
- c) $l = 20 \text{ cm}, f = 600 \text{ MHz}$,
- d) $l = 1 \text{ mm}, f = 100 \text{ GHz}$,

Example 2.1 Solution

Transmission line effects are negligible when $\frac{l}{\lambda} < 0.01$

It may be necessary to account for both the phase shift due to time delay and the presence of reflected signals that may have bounced back.

$$\frac{l}{\lambda} = \frac{lf}{u_p} = \frac{lf}{c} = \frac{f}{3 \times 10^8 \text{ m/s}}$$

The diagram illustrates the derivation of the formula $\frac{l}{\lambda} = \frac{lf}{u_p} = \frac{lf}{c} = \frac{f}{3 \times 10^8 \text{ m/s}}$. A central rectangular box contains the equation. Four arrows point from surrounding text labels to specific parts of the equation:

- An arrow from "Line length" points to the term l in the first fraction.
- An arrow from "Frequency of the oscillating wave" points to the term f in the last fraction.
- An arrow from "Wavelength" points to the term λ in the first fraction.
- An arrow from "Speed of light" points to the term c in the middle fraction.
- An arrow from "Velocity of wave propagation" points to the term u_p in the middle fraction.


$$\frac{l}{\lambda} = \frac{lf}{u_p} = \frac{lf}{c} = \frac{lf}{3 * 10^8 \text{ m/s}}$$

Example 2.1 Solution

a) $l = 20 \text{ cm}, f = 20 \text{ kHz}$ $= \frac{lf}{3 * 10^8 \text{ m/s}} = \frac{20 * 10^{-3} \text{ m} * 20 * 10^3 \text{ Hz}}{3 * 10^8 \text{ m/s}} = 1.33 * 10^{-5}$ \leftarrow negligible

b) $l = 50 \text{ km}, f = 60 \text{ Hz}$ $= \frac{lf}{3 * 10^8 \text{ m/s}} = \frac{50 * 10^3 \text{ m} * 60 \text{ Hz}}{3 * 10^8 \text{ m/s}} = 0.01$ \leftarrow included

c) $l = 20 \text{ cm}, f = 600 \text{ MHz}$, $= \frac{lf}{3 * 10^8 \text{ m/s}} = \frac{20 * 10^{-2} \text{ m} * 600 * 10^6 \text{ Hz}}{3 * 10^8 \text{ m/s}} = 0.4$ \leftarrow included

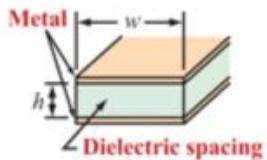
d) $l = 1 \text{ mm}, f = 100 \text{ GHz}$, $= \frac{lf}{3 * 10^8 \text{ m/s}} = \frac{1 * 10^{-3} \text{ m} * 20 * 10^9 \text{ Hz}}{3 * 10^8 \text{ m/s}} = 0.33$ \leftarrow included

Example 2.4

A 1 GHz parallel-plate transmission line consists of 1.2 cm wide copper strips separated by a 0.15 cm thick layer of polystyrene.

Appendix B gives $\mu_c = \mu_0 = 4\pi * 10^{-7} H/m$ and $\sigma_c = 5.8 * 10^7 S/m$ for copper, and $\epsilon_r = 2.6$

for polystyrene. Use Table 2-1 to determine the line parameters of the transmission line. Assume that $\mu_c = \mu_0$ and $\sigma \approx 0$ for polystyrene.



Example 2.4 Solution

From table 2-1, the following parameter formulas can be determined for a parallel-plate:

$$R' = \frac{2R_s}{w}$$

$$L' = \frac{\mu h}{w} \quad R_s = \sqrt{\frac{\pi f \mu_c}{\sigma_c}}$$

$$G' = \frac{\sigma w}{h}$$

$$C' = \frac{\epsilon w}{h}$$

For copper:

$$\begin{aligned}w &= 1.2 \text{ cm} = 0.012 \text{ m} \\ \mu_c &= \mu_0 = 4\pi * 10^{-7} \text{ H/m} \\ \sigma_c &= 5.8 * 10^8 \text{ S/m}\end{aligned}$$

Other values:

$$\begin{aligned}f &= 1 \text{ GHz} = 1 * 10^9 \text{ Hz} \\ \epsilon &= 8.854 * 10^{-12} \text{ F/m}\end{aligned}$$

For Polystyrene:

$$\begin{aligned}h &= 0.15 \text{ cm} = 0.0015 \text{ m} \\ \mu &= \mu_0 = 4\pi * 10^{-7} \text{ H/m} \\ \sigma &= 0 \\ \epsilon_r &= 2.6\end{aligned}$$

Example 2.4 Solution

From table 2-1, the following parameter formulas can be determined for a parallel-plate:

$$R' = \frac{2R_s}{w} \quad R' = \frac{2}{w} \sqrt{\frac{\pi f \mu_c}{\sigma_c}} = \frac{2}{1.2 * 10^{-2}} \sqrt{\frac{\pi * 1 * 10^9 * 4\pi * 10^{-7}}{5.8 * 10^7}} = 1.38 \Omega/m$$

$$L' = \frac{\mu h}{w} \quad L' = \frac{\mu h}{w} = \frac{4\pi * 10^{-7} * 0.0015}{0.012} = 1.57 * 10^{-7} H/m$$

$$G' = \frac{\sigma w}{h} \quad G' = \frac{\sigma w}{h} = 0 S/m$$

$$C' = \frac{\epsilon w}{h} \quad C' = \frac{\epsilon w}{h} = \frac{\epsilon_0 \epsilon_r w}{h} = \frac{8.854 * 10^{-12} * 2.6 * 0.012}{0.0015} = 1.84 * 10^{-10} F/m$$



EP2A04 TUTORIAL 4

FEBRUARY 7TH, 2022

TEACHING ASSISTANTS

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QUESTION 2.5

For the parallel-plate transmission line of Problem 2.4, the line parameters are given by $R' = 1 \Omega/m$, $L' = 167 \text{ nH}/m$, $G' = 0$ and $C' = 172 \text{ pF}/m$. Find α , β , u_P and Z_0 at 1 GHz.

Givens:

$$R' = 1 \Omega/m$$

$$L' = 167 \text{ nH}/m$$

$$G' = 0$$

$$C' = 172 \text{ pF}/m$$

$$f = 1 \text{ GHz}$$

Unknowns: α , β , u_P , Z_0

Plan:

- Find $\gamma = \alpha + j\beta$
- Isolate real and imaginary components
- Use β and ω to obtain u_P
- Find Z_0 using the transmission line parameters

Find $\gamma = \alpha + j\beta$

$$\gamma = \sqrt{(R' + j\omega L')(G' + j\omega C')}$$
$$\gamma = \sqrt{(1 + j(2\pi \times 10^9)(167 \times 10^{-9}))(j(2\pi \times 10^9)(172 \times 10^{-12}))}$$
$$\gamma = 0.016 + j33.7$$

QUESTION 2.5

For the parallel-plate transmission line of Problem 2.4, the line parameters are given by $R' = 1 \Omega/m$, $L' = 167 \text{ nH}/m$, $G' = 0$ and $C' = 172 \text{ pF}/m$. Find α , β , u_P and Z_0 at 1 GHz.

Isolate real and imaginary components

$$\alpha = \Re\{\gamma\} = \Re\{0.016 + j33.7\} = 0.016 \frac{Np}{m}$$

$$\beta = \Im\{\gamma\} = \Im\{0.016 + j33.7\} = 33.7 \frac{\text{rad}}{m}$$

Use β and ω to obtain u_P

$$u_P = \frac{\omega}{\beta} = \frac{2\pi f}{\beta} = \frac{2\pi(1 \times 10^9)}{33.7} = 1.86 \times 10^8 \text{ m/s}$$

Find Z_0 using the transmission line parameters

$$\begin{aligned} Z_0 &= \sqrt{\frac{R' + j\omega L'}{G' + j\omega C'}} \\ &= \sqrt{\frac{1 + j(2\pi \times 10^9)(167 \times 10^{-9})}{j(2\pi \times 10^9)(172 \times 10^{-12})}} \\ &= \sqrt{\frac{1 + j1049.3}{j1.081}} = \sqrt{970.7 - j0.9251} \\ &= \sqrt{970.7e^{-j9.53 \times 10^{-4}}} = 31.2e^{-j4.765 \times 10^{-4}} \\ &= (31.2 - j0.015) \Omega \end{aligned}$$

QUESTION 2.6

A coaxial line with inner and outer conductor diameters of 0.5 cm and 1 cm, respectively, is filled with an insulating material with $\epsilon_r = 4.5$ and $\sigma = 10^{-3} \text{ S/m}$. The conductors are made of copper. Calculate the line parameters at 1 GHz.

Givens:

$$d_{inner} = 0.5 \text{ cm}$$

$$d_{outer} = 1 \text{ cm}$$

$$\epsilon_r = 4.5$$

$$\sigma = 10^{-3} \frac{\text{S}}{\text{m}}$$

$$f = 1 \text{ GHz}$$

Hidden givens:

Copper parameters!

$$\sigma_{Cu} = 5.8 \times 10^7 \text{ S/m}$$

$$\mu_{r-Cu} = 1$$

Unknowns: R', L', G', C'

Plan:

- Compute R_S
- Use the coaxial line formulae to compute the line parameters

Compute R_S

$$R_S = \sqrt{\frac{\pi f \mu_c}{\sigma_c}} = \sqrt{\pi(1 \times 10^9)(4\pi \times 10^{-7})/(5.8 \times 10^7)}$$
$$= 8.2502 \times 10^{-3}$$

QUESTION 2.6

A coaxial line with inner and outer conductor diameters of 0.5 cm and 1 cm, respectively, is filled with an insulating material with $\epsilon_r = 4.5$ and $\sigma = 10^{-3} \text{ S/m}$. The conductors are made of copper. Calculate the line parameters at 1 GHz.

Use the coaxial line formulae to
compute the line parameters

$$R' = \frac{R_S}{2\pi} \left(\frac{1}{a} + \frac{1}{b} \right) = \frac{R_S}{2\pi} \left(\frac{2}{d_{inner}} + \frac{2}{d_{outer}} \right) = \frac{8.2502 \times 10^{-3}}{2\pi} \left(\frac{2}{0.5 \times 10^{-2}} + \frac{2}{1 \times 10^{-2}} \right) = 0.788 \Omega/m$$

$$L' = \frac{\mu}{2\pi} \ln \left(\frac{b}{a} \right) = \frac{\mu}{2\pi} \ln \left(\frac{d_{outer}}{d_{inner}} \right) = \frac{(4\pi \times 10^{-7})}{2\pi} \ln \left(\frac{1}{0.5} \right) = 1.39 \times 10^{-7} H/m$$

$$G' = \frac{2\pi\sigma}{\ln(b/a)} = \frac{2\pi \times 10^{-3}}{\ln(1/0.5)} = 9.06 \times 10^{-3} S/m$$

$$C' = \frac{2\pi\epsilon}{\ln(b/a)} = \frac{2\pi\epsilon_r\epsilon_0}{\ln(b/a)} = \frac{2\pi(4.5)(8.854 \times 10^{-12})}{\ln(1/0.5)} = 3.61 \times 10^{-10} F/m$$

QUESTION 2.18

Polyethylene with $\epsilon_r = 2.25$ is used as the insulating material in a lossless coaxial line with a characteristic impedance of 50Ω . The radius of the inner conductor is 1.2 mm . (A) What is the radius of the outer conductor? (B) What is the phase velocity of the line?

Givens:

$$\epsilon_r = 2.25$$

$$Z_0 = 50 \Omega$$

$$a = 1.2 \text{ mm}$$

Hidden givens:

Lossless! So...

$$R' = G' = 0$$

Plan:

- Write characteristic impedance equation using line parameters
- Solve for b
- Compute the phase velocity

Write characteristic impedance equation using line parameters

Since R' and G' are 0, the equation simplifies to:

$$Z_0 = \sqrt{\frac{R' + j\omega L'}{G' + j\omega C'}} = \sqrt{\frac{L'}{C'}} = \sqrt{\frac{\frac{\mu}{2\pi} \ln\left(\frac{b}{a}\right)}{\left(\frac{2\pi\epsilon}{\ln(b/a)}\right)}} = \sqrt{\frac{\mu}{4\pi^2\epsilon} \left(\ln\left(\frac{b}{a}\right)\right)^2} = \sqrt{\frac{\mu}{4\pi^2\epsilon}} \ln\left(\frac{b}{a}\right)$$

QUESTION 2.18

Polyethylene with $\epsilon_r = 2.25$ is used as the insulating material in a lossless coaxial line with a characteristic impedance of 50Ω . The radius of the inner conductor is 1.2 mm . (A) What is the radius of the outer conductor? (B) What is the phase velocity of the line?

Solve for b

$$Z_0 = \sqrt{\frac{\mu}{4\pi^2\epsilon}} \ln\left(\frac{b}{a}\right)$$

$$\frac{Z_0}{\sqrt{\frac{\mu}{4\pi^2\epsilon}}} = \ln\left(\frac{b}{a}\right)$$

$$b = a * \exp\left(\frac{Z_0}{\sqrt{\frac{\mu}{4\pi^2\epsilon}}}\right)$$

$$b = (1.2 \times 10^{-3}) * \exp\left(\frac{50}{\sqrt{\frac{4\pi \times 10^{-7}}{4\pi^2(2.25)(8.854 \times 10^{-12})}}}\right)$$

$$b = (1.2 \times 10^{-3}) * \exp\left(\frac{50}{\sqrt{\frac{10^5}{\pi(2.25)(8.854)}}}\right)$$

$$b = 4.2\text{ mm}$$

Compute the phase velocity

Lossless line, so...

$$u_P = \frac{c}{\sqrt{\epsilon_r}}$$

$$u_P = \frac{3 \times 10^8}{\sqrt{2.25}}$$

$$u_P = 2 \times 10^8\text{ m/s}$$

QUESTION 2.27

At an operating frequency of 300 MHz, a lossless 50 Ω air-spaced transmission line 2.5 m in length is terminated with an impedance $Z_L = (40 + j20) \Omega$. Find the input impedance.

Givens

$$f = 300 \text{ MHz}$$

$$Z_0 = 50 \Omega$$

$$l = 2.5 \text{ m}$$

$$Z_L = (40 + j20) \Omega$$

Unknowns: Z_i

Plan:

- Calculate the phase constant of the line
- Plug into the input impedance formula

Calculate the phase constant of the line

Since the line is air filled, $u_p = c$. Therefore...

$$\beta = \frac{\omega}{c} = \frac{2\pi(300 \times 10^6)}{3 \times 10^8} = 2\pi \text{ rad/s}$$

Plug into the input impedance formula

$$Z_i = Z_0 \frac{Z_L + jZ_0 \tan(\beta l)}{Z_0 + jZ_L \tan(\beta l)}$$

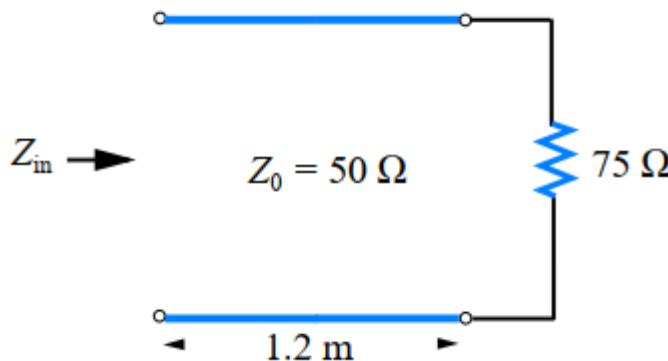
$$Z_i = 50 \frac{40 + j20 + j(50) \tan(2\pi(2.5))}{50 + j(40 + j20) \tan(2\pi(2.5))}$$

$$Z_i = 50 \frac{40 + j20}{50}$$

$$Z_i = (40 + j20) \Omega$$

QUESTION 2.35

For the lossless transmission line circuit shown in the figure, determine the equivalent series lumped-element circuit at 400 MHz at the input to the line. The line has a characteristic impedance of 50Ω and the insulating layer has $\epsilon_r = 2.25$.



Givens

$$f = 400 \text{ MHz}$$

$$Z_0 = 50 \Omega$$

$$l = 1.2 \text{ m}$$

$$Z_L = 75 \Omega$$

$$\epsilon_r = 2.25$$

Unknowns: Z_i

Plan:

- Compute phase constant of line
- Compute equivalent input impedance
- Design equivalent circuit based on impedance

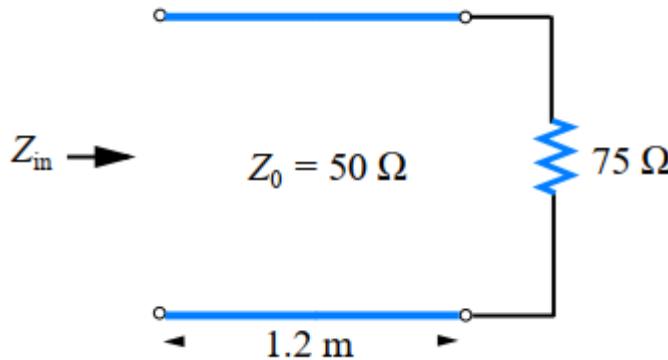
Compute phase constant of line

$$\text{Lossless line, so } u_P = \frac{c}{\sqrt{\epsilon_r}} = \frac{3 \times 10^8}{\sqrt{2.25}} = 2 \times 10^8 \text{ m/s}$$

$$\beta = \frac{\omega}{u_P} = \frac{2\pi(400 \times 10^6)}{2 \times 10^8} = 4\pi \text{ rad/m}$$

QUESTION 2.35

For the lossless transmission line circuit shown in the figure, determine the equivalent series lumped-element circuit at 400 MHz at the input to the line. The line has a characteristic impedance of 50Ω and the insulating layer has $\epsilon_r = 2.25$.



Compute equivalent input impedance

$$Z_i = Z_0 \frac{Z_L + jZ_0 \tan(\beta l)}{Z_0 + jZ_L \tan(\beta l)}$$

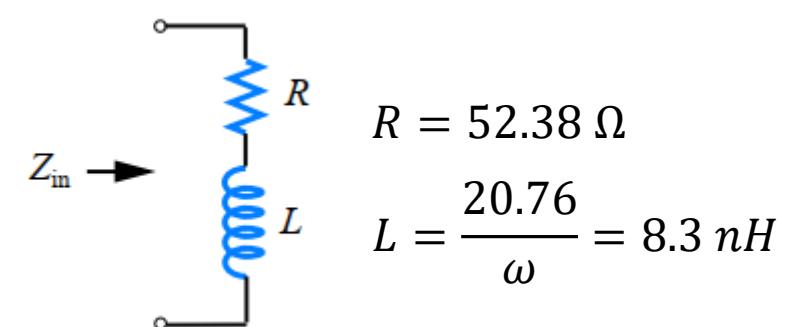
$$Z_i = 50 \frac{75 + j(50) \tan((4\pi)(1.2))}{50 + j(75) \tan((4\pi)(1.2))}$$

$$Z_i = 50 \frac{75 - j36.327}{50 - j54.491}$$

$$Z_i = (52.38 + j20.76) \Omega$$

Design equivalent circuit based on impedance

Imaginary part is positive, so if this is a series circuit, it must be a resistor and an inductor!



REMINDER

- Assignment 4 is out now, and is due at 8AM on February 14
- 4 calculation questions, 1 research question, 1 bonus derivation question
- Submit PDFs please!
- Good luck!

ENG PHYS 2A04 Tutorial 5

Electricity and Magnetism

Your TAs today

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Chapter 3

Problem 3.17 – Question

Find a vector **G** whose magnitude is 4 and whose direction is perpendicular to both vectors **E** and **F**, where:

$$\mathbf{E} = \hat{x} + \hat{y}2 - \hat{z}2 \text{ and } \mathbf{F} = \hat{y}3 - \hat{z}6$$

Problem 3.17 – Question

Find a vector **G** whose magnitude is 4 and whose direction is perpendicular to both vectors **E** and **F**, where:

$$\mathbf{E} = \hat{x} + \hat{y}2 - \hat{z}2 \text{ and } \mathbf{F} = \hat{y}3 - \hat{z}6$$

Eq. 3.1: $\mathbf{G} = G\hat{\mathbf{g}}$

Eq. 3.22: vector normal to the plane containing **E** and **F**: $\mathbf{E} \times \mathbf{F}$

Eq. 3.22: unit vector: $\frac{\mathbf{E} \times \mathbf{F}}{|\mathbf{E} \times \mathbf{F}|}$

Problem 3.17 – Question

Find a vector **G** whose magnitude is 4 and whose direction is perpendicular to both vectors **E** and **F**, where:

$$\mathbf{E} = \hat{x} + \hat{y}2 - \hat{z}2 \text{ and } \mathbf{F} = \hat{y}3 - \hat{z}6$$

$$\mathbf{G} = \pm 4 \frac{\mathbf{E} \times \mathbf{F}}{|\mathbf{E} \times \mathbf{F}|}$$

Problem 3.17 – Question

Find a vector \mathbf{G} whose magnitude is 4 and whose direction is perpendicular to both vectors \mathbf{E} and \mathbf{F} , where:

$$\mathbf{E} = \hat{x} + \hat{y}2 - \hat{z}2 \text{ and } \mathbf{F} = \hat{y}3 - \hat{z}6$$

$$\text{Eq.3.28: } \mathbf{E} \times \mathbf{F} = \begin{vmatrix} \hat{x} & \hat{y} & \hat{z} \\ E_x & E_y & E_z \\ F_x & F_y & F_z \end{vmatrix}$$

$$\begin{aligned} &= (E_y F_z - E_z F_y) \hat{x} + (E_z F_x - E_x F_z) \hat{y} + (E_x F_y - E_y F_x) \hat{z} \\ &= -\hat{x}6 + \hat{y}6 + \hat{z}3 \end{aligned}$$

Problem 3.17 – Question

Find a vector **G** whose magnitude is 4 and whose direction is perpendicular to both vectors **E** and **F**, where:

$$\mathbf{E} = \hat{x} + \hat{y}2 - \hat{z}2 \text{ and } \mathbf{F} = \hat{y}3 - \hat{z}6$$

Eq. 3.4: $|\mathbf{E} \times \mathbf{F}| = +\sqrt{(-6)^2 + (6)^2 + (3)^2} = 9$

$$\mathbf{G} = \pm 4 \frac{\mathbf{E} \times \mathbf{F}}{|\mathbf{E} \times \mathbf{F}|} = \pm 4 \frac{(-\hat{x}6 + \hat{y}6 + \hat{z}3)}{9} = \pm \left(-\hat{x}\frac{8}{3} + \hat{y}\frac{8}{3} + \hat{z}\frac{4}{3} \right)$$

Problem 3.26 – Question

Find the volume described by the following, Also sketch the outline of the volume:

$$2 \leq r \leq 5; \frac{\pi}{2} \leq \varphi \leq \pi; 0 \leq z \leq 2$$

Recognize that this is a cylindrical integration

Problem 3.26 – Question

Find the volume described by the following, Also sketch the outline of the volume:

$$2 \leq r \leq 5; \frac{\pi}{2} \leq \varphi \leq \pi; 0 \leq z \leq 2$$

Table 3-1: $V = \int_{z=0}^2 \int_{\varphi=\pi/2}^{\pi} \int_{r=2}^5 r dr d\varphi dz$

Problem 3.26 – Question

Find the volume described by the following, Also sketch the outline of the volume:

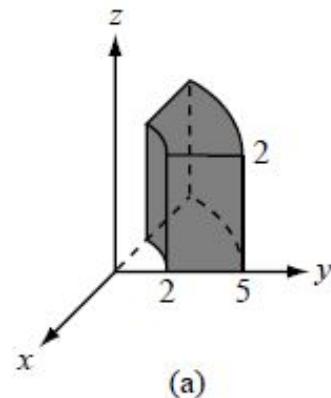
$$2 \leq r \leq 5; \frac{\pi}{2} \leq \varphi \leq \pi; 0 \leq z \leq 2$$

$$V = \left(\frac{1}{2} r^2 \Big|_{r=2}^5 \right) \left(\varphi \Big|_{\varphi=\pi/2}^{\pi} \right) \left(z \Big|_{z=0}^2 \right) = \frac{21\pi}{2}$$

Problem 3.26 – Question

Find the volume described by the following, Also sketch the outline of the volume:

$$2 \leq r \leq 5; \frac{\pi}{2} \leq \varphi \leq \pi; 0 \leq z \leq 2$$



Problem 3.34 – Question

Transform the following vectors into cylindrical coordinates and then evaluate them at the indicated points:

a) $\mathbf{A} = \hat{x}(x + y)$ at $P_1 = (1, 2, 3)$

b) $\mathbf{C} = \hat{x} \frac{y^2}{(x^2+y^2)} - \hat{y} \frac{x^2}{(x^2+y^2)} + \hat{z} 4$ at $P_2 = (1, -1, 2)$

Problem 3.34 – Question

$$\mathbf{A} = \hat{x}(x + y) \text{ at } P_1 = (1, 2, 3)$$

Convert to cylindrical coordinates

Table 3-2: $r = +\sqrt{x^2 + y^2} = +\sqrt{1^2 + 2^2} = \sqrt{5},$

$$\varphi = \tan^{-1} \left(\frac{y}{x} \right) = \tan^{-1} \left(\frac{2}{1} \right) = 63.4^\circ$$

$$z = z = 3$$

$$P_1 = (\sqrt{5}, 63.4^\circ, 3)$$

Problem 3.34 – Question

$$\mathbf{A} = \hat{x}(x + y) \text{ at } P_1 = (1, 2, 3)$$

Convert to cylindrical coordinates

Table 3-2: $\mathbf{A} = \hat{r}\mathbf{A}_r + \hat{\varphi}\mathbf{A}_\varphi + \hat{z}\mathbf{A}_z$

$$x = r \cos \varphi; y = r \sin \varphi$$

$$\mathbf{A}_r = A_x \cos \varphi + A_y \sin \varphi = (x + y) \cos \varphi = (r \cos \varphi + r \sin \varphi) \cos \varphi$$

$$\mathbf{A}_\varphi = -A_x \sin \varphi + A_y \cos \varphi = -(x + y) \sin \varphi = -(r \cos \varphi + r \sin \varphi) \sin \varphi$$

$$\mathbf{A}_z = 0$$

$$\mathbf{A} = \hat{r}r \cos \varphi (\cos \varphi + \sin \varphi) - \hat{\varphi}r \sin \varphi (\cos \varphi + \sin \varphi)$$

Problem 3.34 – Question

$\mathbf{A} = \hat{x}(x + y)$ at $P_1 = (1, 2, 3)$

$$\mathbf{A} = \hat{r}r \cos \varphi (\cos \varphi + \sin \varphi) - \hat{\varphi}r \sin \varphi (\cos \varphi + \sin \varphi)$$

$$P_1 = (\sqrt{5}, 63.4^\circ, 3)$$

$$\mathbf{A}(P_1) = \hat{r}1.34 - \hat{\varphi}2.68$$

Problem 3.34 – Question

$$\mathbf{C} = \hat{\mathbf{x}} \frac{y^2}{(x^2+y^2)} - \hat{\mathbf{y}} \frac{x^2}{(x^2+y^2)} + \hat{\mathbf{z}} 4 \text{ at } P_2 = (1, -1, 2)$$

Convert to cylindrical coordinates

Table 3-2: $r = +\sqrt{x^2 + y^2} = +\sqrt{1^2 + (-1)^2} = \sqrt{2},$

$$\varphi = \tan^{-1} \left(\frac{y}{x} \right) = \tan^{-1} \left(\frac{-1}{1} \right) = -45^\circ$$

$$z = z = 2$$

$$P_2 = (\sqrt{2}, -45^\circ, 2)$$

Problem 3.34 – Question

$$\mathbf{C} = \hat{\mathbf{x}} \frac{y^2}{(x^2+y^2)} - \hat{\mathbf{y}} \frac{x^2}{(x^2+y^2)} + \hat{\mathbf{z}} 4 \text{ at } P_2 = (1, -1, 2)$$

Table 3-2: $\mathbf{C} = \hat{\mathbf{r}} C_r + \hat{\varphi} C_\varphi + \hat{\mathbf{z}} C_z$

$$x = r \cos \varphi; y = r \sin \varphi$$

$$C_r = \sin \varphi \cos \varphi (\sin \varphi - \cos \varphi)$$

$$C_\varphi = -(\sin^3 \varphi + \cos^3 \varphi)$$

$$C_z = 4$$

$$\mathbf{C} = \hat{\mathbf{r}} \sin \varphi \cos \varphi (\sin \varphi - \cos \varphi) - \hat{\varphi} (\sin^3 \varphi + \cos^3 \varphi) + \hat{\mathbf{z}} 4$$

Problem 3.34 – Question

$$\mathbf{C} = \hat{\mathbf{x}} \frac{y^2}{(x^2+y^2)} - \hat{\mathbf{y}} \frac{x^2}{(x^2+y^2)} + \hat{\mathbf{z}} 4 \text{ at } P_2 = (1, -1, 2)$$

$$\mathbf{C} = \hat{\mathbf{r}} \sin \varphi \cos \varphi (\sin \varphi - \cos \varphi) - \hat{\boldsymbol{\varphi}} (\sin^3 \varphi + \cos^3 \varphi) + \hat{\mathbf{z}} 4$$

$$P_2 = (\sqrt{2}, -45^\circ, 2)$$

$$\mathbf{C}(P_2) = \hat{\mathbf{r}} 0.707 - \hat{\boldsymbol{\varphi}} 4$$

Problem 3.36-e – Question

Find the gradient of the following scalar function:

$$s = 4x^2e^{-z} + y^3$$

Problem 3.36-e – Solution

Find the gradient of the following scalar function:

$$S = 4x^2e^{-z} + y^3$$

$$\text{AS, } \nabla = \hat{x} \frac{\partial}{\partial x} + \hat{y} \frac{\partial}{\partial y} + \hat{z} \frac{\partial}{\partial z}$$

Problem 3.36-e – Solution

So

$$\nabla S = \hat{x} \frac{\partial S}{\partial x} + \hat{y} \frac{\partial S}{\partial y} + \hat{z} \frac{\partial S}{\partial z}$$

Putting values of function S

$$\nabla S = \hat{x} \frac{\partial(4x^2e^{-z} + y^3)}{\partial x} + \hat{y} \frac{\partial(4x^2e^{-z} + y^3)}{\partial y} + \hat{z} \frac{\partial(4x^2e^{-z} + y^3)}{\partial z}$$

$$\nabla S = \hat{x}8xe^{-z} + \hat{y}3y^2 - \hat{z}4x^2e^{-z}$$

Problem 3.48 – Question

For the vector field $E = \hat{x}xz - \hat{y}yz^2 - \hat{z}xy$, verify the divergence theorem by computing

- a) The total outward flux flowing through the surface of a cube centered at the origin and with sides equal to 2 units each and parallel to the Cartesian axes.
- b) The integral of $\nabla \cdot E$ over the cube's volume.

Divergence Theorem!

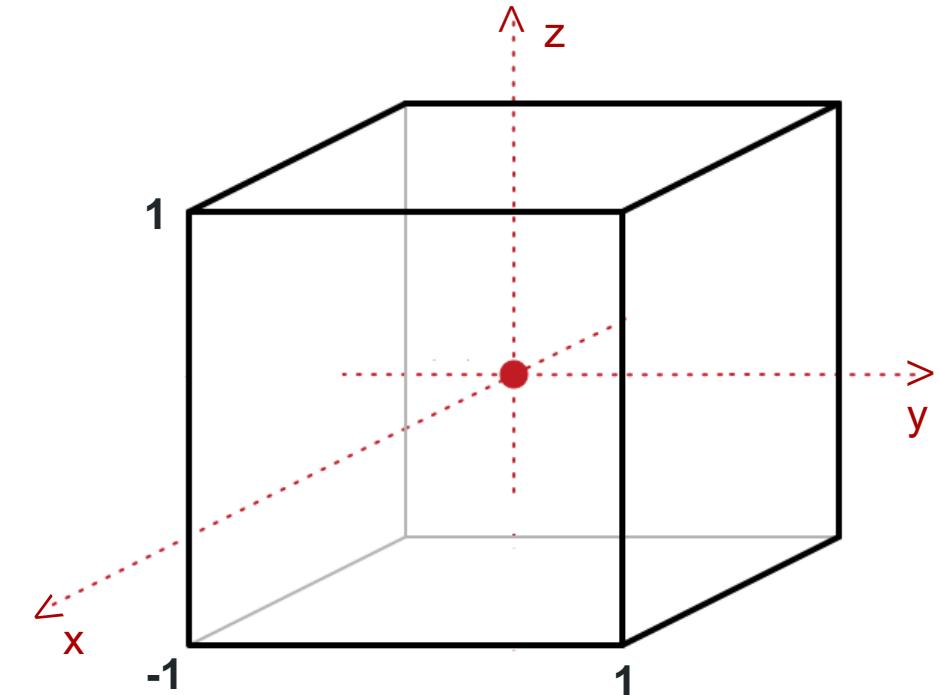
$$\oint_S \vec{E} \cdot d\vec{s} = \int_V (\nabla \cdot E) dV$$

Problem 3.48 – Solution

Part (a):

Goal → Calculate total outward flux

Given → $E = \hat{x}xz - \hat{y}yz^2 - \hat{z}xy$

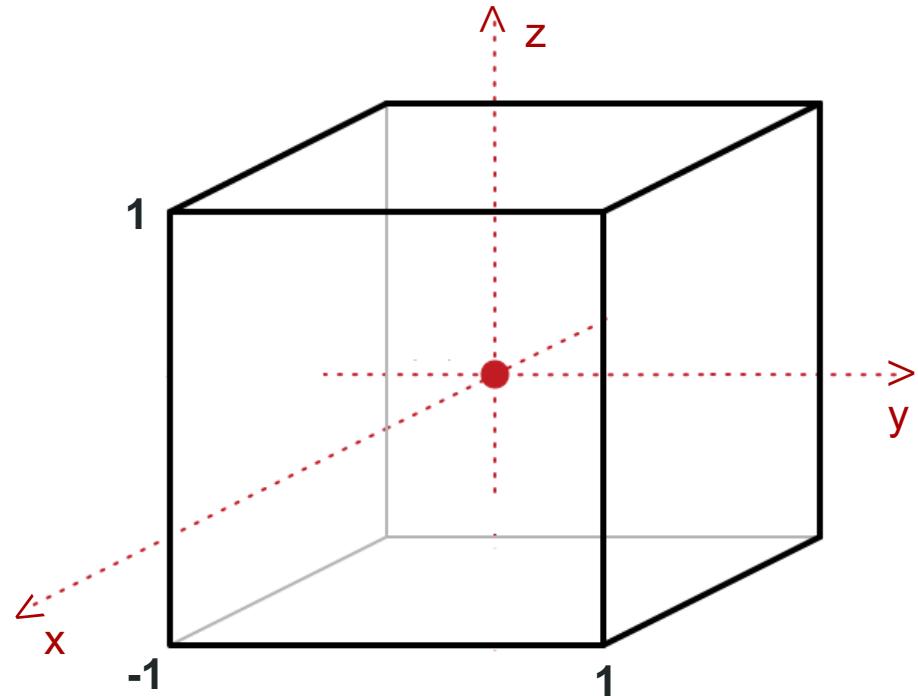


Problem 3.48 – Solution

Part (a):

1) How to calculate the total flux?

Flux through each face → Add them all



2) Flux: $F = \oint_S \vec{E} \cdot d\vec{s}$

$$F_{Total} = F_{top} + F_{bottom} + F_{right} + F_{left} + F_{front} + F_{back}$$

Problem 3.48 – Solution

Part (a):

$$F_{top} = \int \left(\vec{E} \Big|_{z=1} \cdot (\hat{z} dx dy) \right)$$

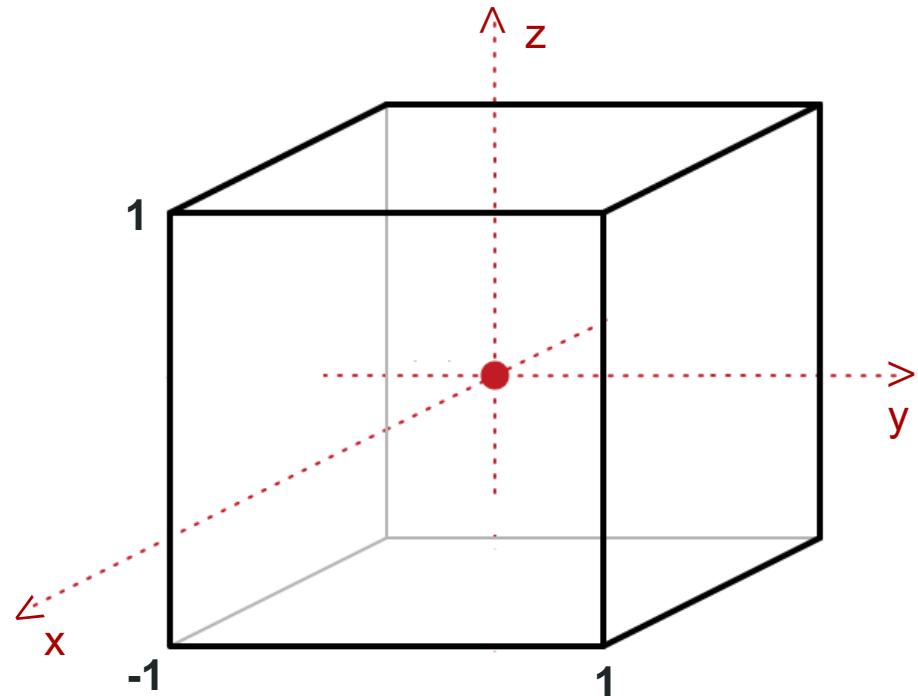
$$F_{bottom} = \int \left(\vec{E} \Big|_{z=-1} \cdot (-\hat{z} dx dy) \right)$$

$$F_{right} = \int \left(\vec{E} \Big|_{y=1} \cdot (\hat{y} dx dz) \right)$$

$$F_{left} = \int \left(\vec{E} \Big|_{y=-1} \cdot (-\hat{y} dx dz) \right)$$

$$F_{front} = \int \left(\vec{E} \Big|_{x=1} \cdot (\hat{x} dy dz) \right)$$

$$F_{back} = \int \left(\vec{E} \Big|_{x=-1} \cdot (-\hat{x} dy dz) \right)$$



Problem 3.48 – Solution

Part (a):

$$F_{top} = 0$$

$$F_{bottom} = 0$$

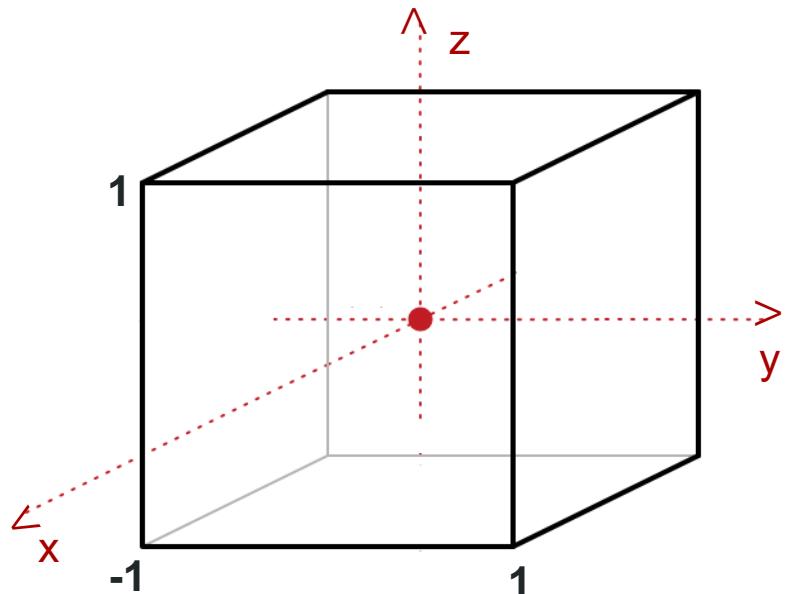
$$F_{right} = -\frac{4}{3}$$

$$F_{left} = -\frac{4}{3}$$

$$F_{front} = 0$$

$$F_{back} = 0$$

$$\mathbf{F}_{Total} = -\frac{8}{3}$$



Problem 3.48 – Solution

Part (a):

An Idea how to calculate F_{right} and F_{left}

$$F_{right} = \int_{x=1}^1 \int_{z=-1}^1 \hat{x}xz - \hat{y}yz^2 - \hat{z}xy \Big|_{y=1} \cdot (\hat{y} dx dz)$$

$$F_{right} = - \int_{x=1}^1 \int_{z=-1}^1 z^2 dz dx$$

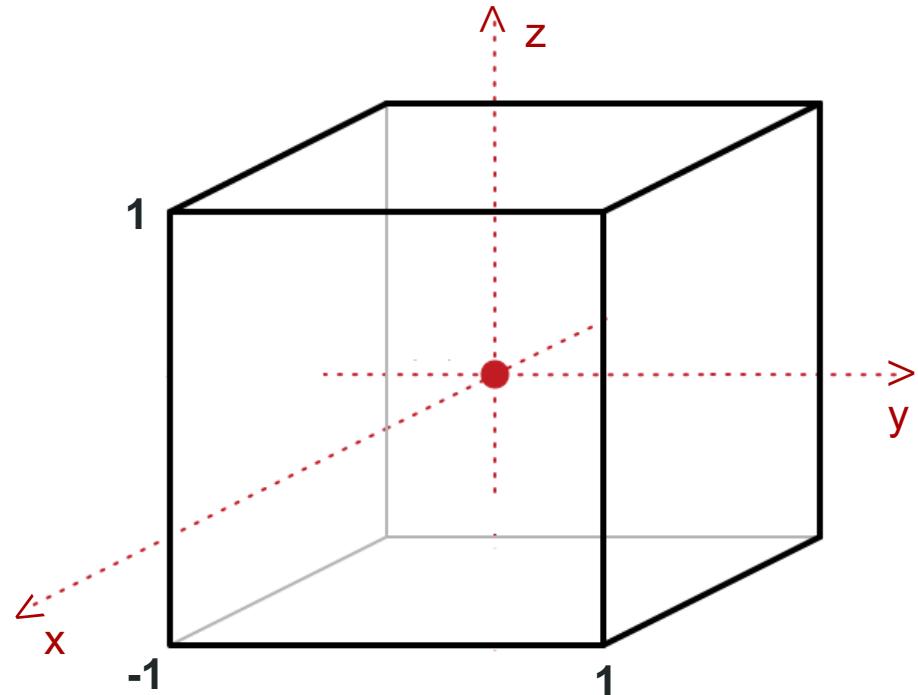
$$F_{right} = - \left(\left(\frac{xz^3}{3} \right) \Big|_{z=-1}^1 \Big|_{x=-1}^1 \right) = -\frac{4}{3}$$

Problem 3.48 – Solution

Part=b

The integral of $\nabla \cdot E$ over the cube's volume.

$$\nabla \cdot E = \frac{\partial E_x}{\partial x} + \frac{\partial E_y}{\partial y} + \frac{\partial E_z}{\partial z}$$



Problem 3.48 – Solution

Part=b

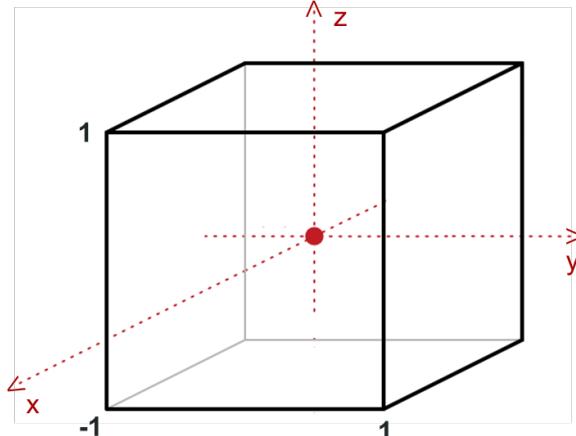
$$\nabla \cdot \mathbf{E} = \frac{\partial E_x}{\partial x} + \frac{\partial E_y}{\partial y} + \frac{\partial E_z}{\partial z}$$

Put

$$E_x = (\hat{x}xz)$$

$$E_y = (-\hat{y}yz^2)$$

$$E_z = (-\hat{z}xy)$$



$$\nabla \cdot \mathbf{E} = \frac{\partial(\hat{x}xz)}{\partial x} + \frac{\partial(-\hat{y}yz^2)}{\partial y} + \cancel{\frac{\partial(-\hat{z}xy)}{\partial z}}$$

Zero

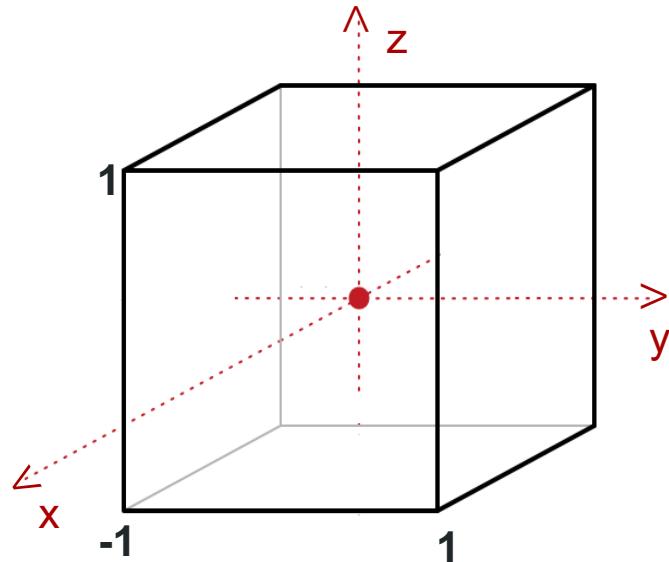
Problem 3.48 – Solution

Part=b

The integral of $\nabla \cdot E$ over the cube's volume.

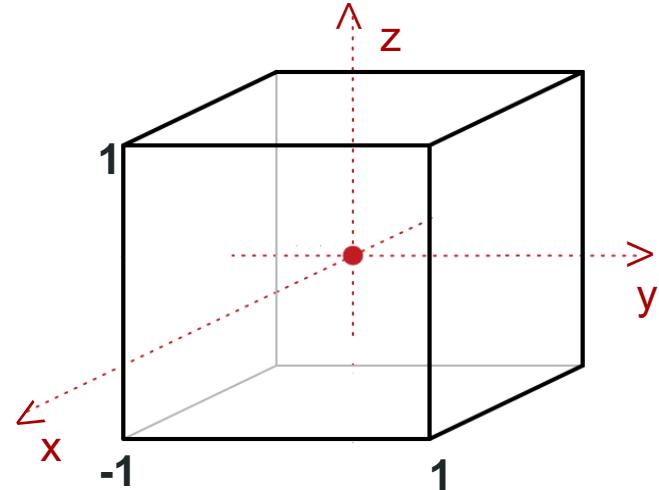
We get,

$$\nabla \cdot E = -z^2 + z$$



Problem 3.48 – Solution

Part=b



$$\iiint (\nabla \cdot \mathbf{E}) dV = \int_{x=-1}^1 \int_{y=-1}^1 \int_{z=-1}^1 \nabla \cdot (\hat{x}xz - \hat{y}yz^2 - \hat{z}xy) dx dy dz$$

$$\iiint (\nabla \cdot \mathbf{E}) dV = \int_{x=-1}^1 \int_{y=-1}^1 \int_{z=-1}^1 (z - z^2) dx dy dz$$

Problem 3.48 – Solution

Part=b

$$\int \int \int (\nabla \cdot E) dV = \left(\left(xy \left(\frac{z^2}{2} - \frac{z^3}{3} \right) \right) \Big|_{z=-1}^1 \Big|_{y=-1}^1 \Big|_{x=-1}^1 \right) = \boxed{-\frac{8}{3}}$$

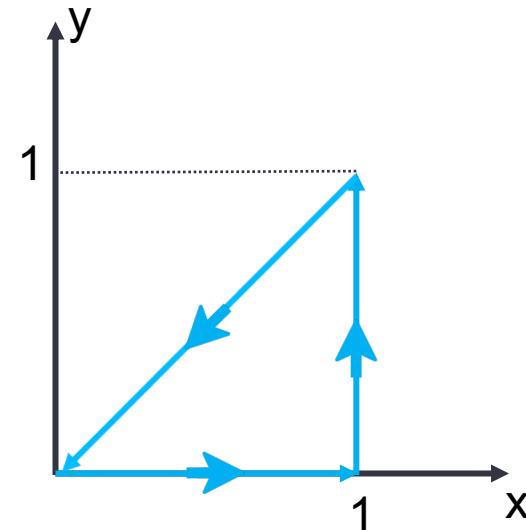
Divergence Theorem!

$$\oint_S \vec{E} \cdot d\vec{s} = \int_V (\nabla \cdot E) dV$$

Problem 3.52-Question

For the vector field $E = \hat{x}xy - \hat{y}(x^2 + 2y^2)$, calculate

- a) $\oint_C \vec{E} \cdot d\vec{l}$ around the triangular contour shown in the Figure
- b) $\int_S (\nabla \times E) \cdot ds$ over the area of the triangle



Stokes' Theorem!

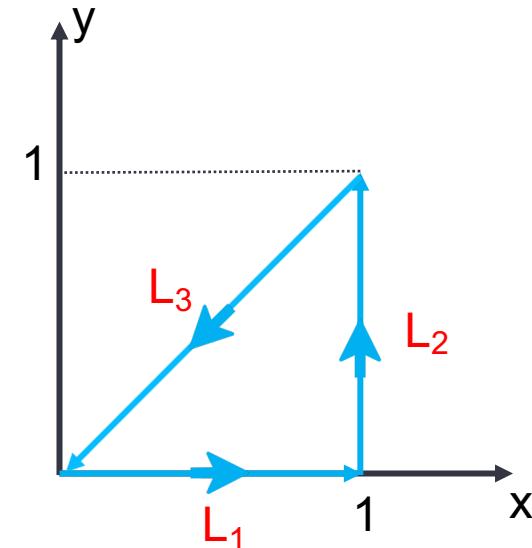
$$\oint_C \vec{E} \cdot d\vec{l} = \int_S (\nabla \times \vec{E}) \cdot d\vec{s}$$

Problem 3.52

Part (a)

Split the counter C in 3 sections:

$$\oint_C \vec{E} \cdot d\vec{l} = \int_{L_1} \vec{E} \cdot d\vec{l}_1 + \int_{L_2} \vec{E} \cdot d\vec{l}_2 + \int_{L_3} \vec{E} \cdot d\vec{l}_3$$



Problem 3.52

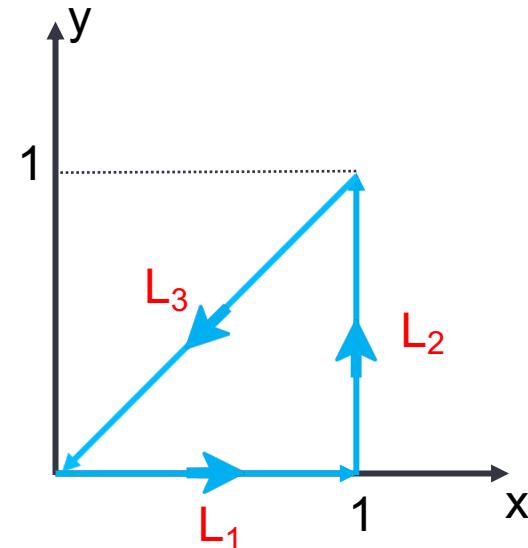
Part (a)

$$\mathbf{E} = \hat{x}xy - \hat{y}(x^2 + 2y^2)$$

$$\int_{L_1} \vec{E} \cdot d\vec{l}_1 = \int_0^1 [\hat{x}xy - \hat{y}(x^2 + 2y^2)] \Big|_{y=0} \cdot \hat{x} dx$$

$$\int_{L_2} \vec{E} \cdot d\vec{l}_2 = \int_1^1 [\hat{x}xy - \hat{y}(x^2 + 2y^2)] \Big|_{x=1} \cdot \hat{y} dy$$

$$\int_{L_3} \vec{E} \cdot d\vec{l}_3 = \int [\hat{x}xy - \hat{y}(x^2 + 2y^2)] \Big|_{y=x} \cdot (\hat{x}dx + \hat{y}dy)$$



Problem 3.52

Part (a)

Practice for solving L_2

$$L_2 = \int_{L_2} \vec{E} \cdot d\vec{l}_2 = \int [\hat{x}xy - \hat{y}(x^2 + 2y^2) \cdot (\hat{x}dx + \hat{y}dy + \hat{z}dz)]$$

$$L_2 = \int_{x=1}^1 [xy] \Big|_{z=0} dx - \int_{y=0}^1 (x^2 + 2y^2) \Big|_{x=1, z=0} dy \int_{z=0}^1 (0) \Big|_{x=1} dz$$

$$L_2 = 0 - (y + \frac{2y^2}{3}) \Big|_{y=1}^0 + 0 = -\frac{5}{3}$$

L_1 and L_3 can be solved by same process

Problem 3.52

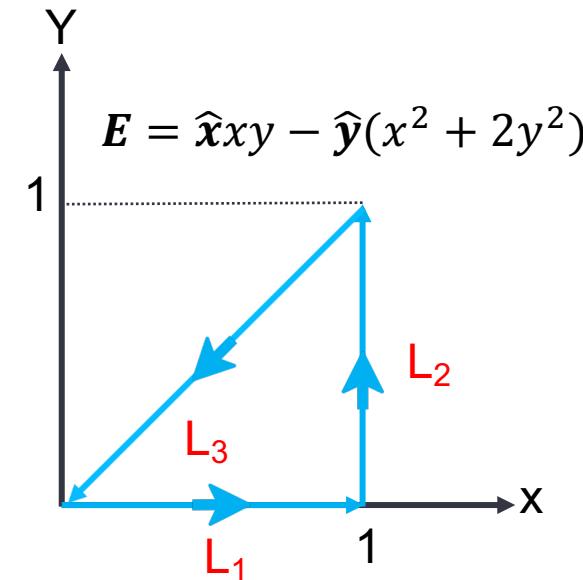
Part (a)

$$\int_{L_1} \vec{E} \cdot d\vec{l}_1 = 0$$

$$\int_{L_2} \vec{E} \cdot d\vec{l}_2 = -\frac{5}{3}$$

$$\int_{L_3} \vec{E} \cdot d\vec{l}_3 = \frac{2}{3}$$

$$\oint_C \vec{E} \cdot d\vec{l} = \int_{L_1} \vec{E} \cdot d\vec{l}_1 + \int_{L_2} \vec{E} \cdot d\vec{l}_2 + \int_{L_3} \vec{E} \cdot d\vec{l}_3$$



$$\oint_C \vec{E} \cdot d\vec{l} = 0 - \frac{5}{3} + \frac{2}{3} = -1$$

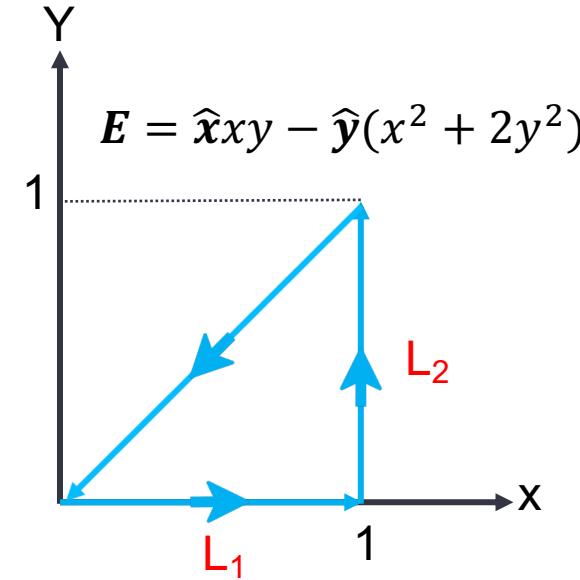
Problem 3.52

Part (b)

$$\int_S (\nabla \times \mathbf{E}) \cdot d\mathbf{s}$$

Calculate the curl:

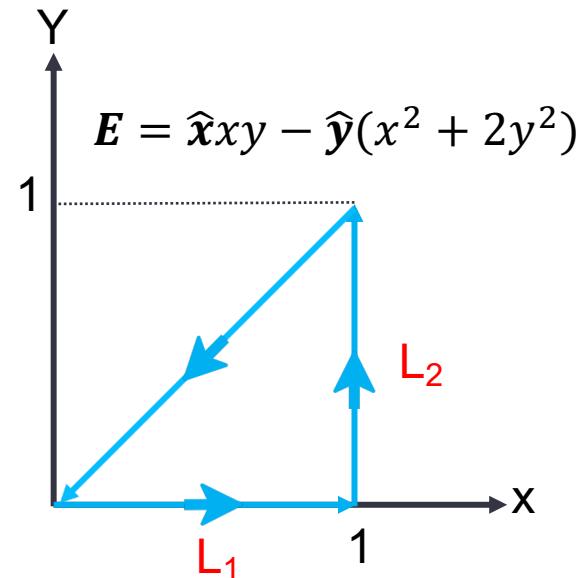
$$(\nabla \times \mathbf{E}) = \begin{vmatrix} \hat{x} & \hat{y} & \hat{z} \\ \frac{\partial}{\partial x} & \frac{\partial}{\partial y} & \frac{\partial}{\partial z} \\ E_x & E_y & E_z \end{vmatrix} = -3x\hat{z}$$



Problem 3.52

Part (b)

$$ds = dx dy \hat{z} \quad \text{and} \quad (\nabla \times E) = -3x \hat{z}$$



$$\iint (\nabla \times E) \cdot ds = \int_{x=0}^{x=1} \int_{y=0}^{y=x} ((-3x \hat{z}) \cdot (\hat{z} dy dx)) \Big|_{z=0}$$

$$\iint (\nabla \times E) \cdot ds = - \int_{x=0}^{x=1} \int_{y=0}^{y=x} 3x \, dy dx = - \int_{x=0}^1 3x(x-0) dx = -(x^3) \Big|_0^1 = \boxed{-1}$$

Stokes' Theorem!

$$\oint_C \vec{E} \cdot d\vec{l} = \int_S (\nabla \times \vec{E}) \cdot d\vec{s}$$

ENGPHYS 2A04 Tutorial 6

ELECTRICITY AND MAGNETISM

Your TAs Today

- Joanne Lee

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- Seung Il Lee (Alex)

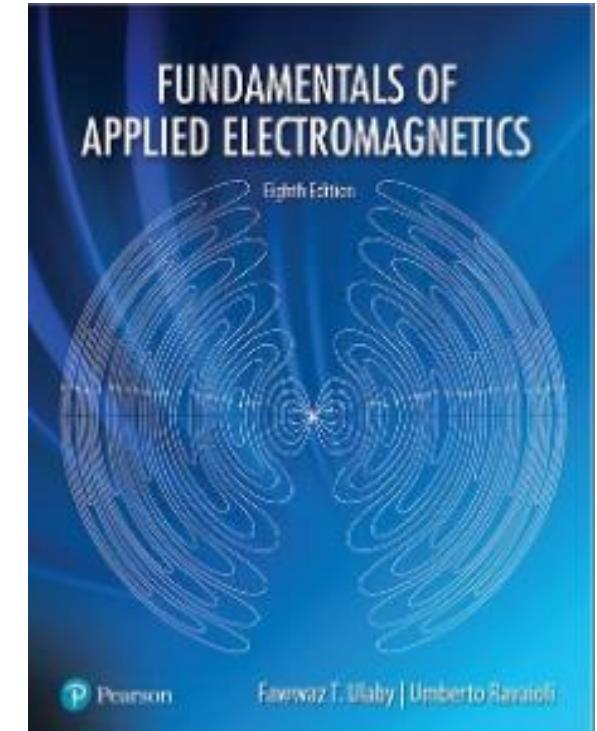
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Your Textbook

Fundamentals of Applied Electromagnetics Eighth Edition

Ulaby & Ravaioli

Seventh Edition also acceptable, with some inconsistencies



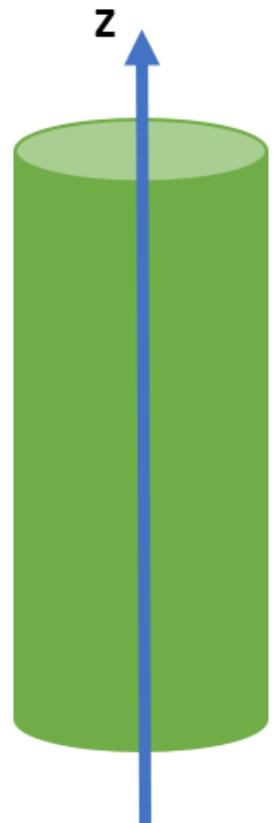
Problem 4.9

A circular beam of charge of radius a consists of electrons moving with a constant speed u along the $+z$ direction. The beam's axis is coincident with the z -axis and the electron charge density by,

$$\rho_v = -cr^2 \quad [c/m^3]$$

Where c is constant and r is the radial distance from the axis of the beam.

- A) Determine the charge density per unit length.
- B) Determine the current crossing the z -plane.



Problem 4.9

Solution to A)

$$\rho_v = \frac{dq}{dv} \quad \rightarrow \quad dv = dl \, ds \quad \rightarrow \quad \frac{dq}{dl} = \rho_l \quad \rightarrow \quad \rho_v = \frac{dq}{dv} = \frac{dq}{dl \, ds} = \frac{\rho_l}{ds}$$

$$\rho_l = \rho_v ds \rightarrow \rho_l = \iint \rho_v ds \quad \rightarrow \quad ds = r dr d\theta \quad (\text{polar coordinates})$$

$$\rightarrow \quad \rho_l = \int_{r=0}^a \int_{\theta=0}^{2\pi} -cr^2 * r \, dr \, d\theta$$

Problem 4.9

Solution to A)

$$\rho_l = \int_{r=0}^a \int_{\theta=0}^{2\pi} -cr^2 * r \, dr \, d\theta = -2\pi c \frac{r^4}{4}$$

$$\rho_l = \frac{-\pi c a^4}{2} \left[\frac{C}{m} \right]$$

Problem 4.9

Solution to B)

$$I = \frac{\text{charge}}{\text{time}} = \left[\frac{\text{Coulomb}}{\text{Second}} \right] = [A]$$

$$I = \int \mathbf{J} \cdot d\mathbf{s} = \int_{r=0}^a \int_{\phi=0}^{2\pi} (-c u r^2 \hat{\mathbf{z}}) \cdot (\hat{\mathbf{z}}) r \, dr \, d\phi \quad \text{as } \mathbf{J} = \rho_v \mathbf{u} = -c r^2 \cdot \mathbf{u} \hat{\mathbf{z}}$$

$$I = -2\pi c u \int_{r=0}^a r^3 \, dr = -\frac{\pi c a^4 u}{2} = \rho_l u$$

$$\text{Units for } \rho_l u \rightarrow \left[\frac{C}{m} \right] \left[\frac{m}{s} \right] = \left[\frac{C}{s} \right] = [A]$$

Problem 4.15

Electric charge is distributed along an arc located in the x-y plane and defined by $r = 2 \text{ cm}$ and $0 \leq \phi \leq \frac{\pi}{4}$

If $\rho_l = 5 \frac{\mu\text{C}}{\text{m}}$, find \mathbf{E} at $(0,0,z)$ and then evaluate it at:

- A) The origin
- B) $z = 5 \text{ cm}$
- C) $z = -5 \text{ cm}$

$$\mathbf{R}' = -0.02\hat{r} + z\hat{z}$$

$$dl' = r d\phi = 0.02 d\phi$$

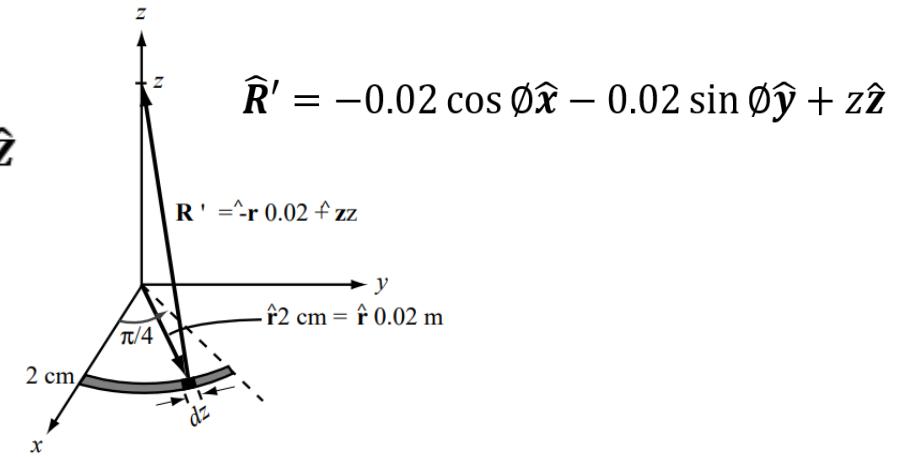


Figure P4.13: Line charge along an arc.

Problem 4.15

Solution:

$$\mathbf{E} = \frac{1}{4\pi\epsilon_0} \int_{l'} \hat{\mathbf{R}}' \frac{\rho_l dl'}{R'^2} = \frac{1}{4\pi\epsilon_0} \int_{\phi=0}^{\frac{\pi}{4}} \rho_l \frac{-0.02 \cos \phi \hat{x} - 0.02 \sin \phi \hat{y} + z \hat{z}}{((0.02)^2 + (z^2)^{3/2})} 0.02 d\phi$$


$$\mathbf{E} = \frac{898.8}{((0.02)^2 + (z^2)^{3/2})} [-0.014 \hat{x} - 0.006 \hat{y} + 0.78 z \hat{z}] \left[\frac{V}{m} \right]$$

Problem 4.15

Solution

A) At origin ($z = 0$)

$$\mathbf{E} = [-1.6\hat{x} - 0.66\hat{y}] \left[\frac{MV}{m} \right]$$

B) $z = 5 \text{ cm}$

$$\mathbf{E} = [-81.4\hat{x} - 33.7\hat{y} + 226\hat{z}] \left[\frac{kV}{m} \right]$$

C) $z = -5 \text{ cm}$

$$\mathbf{E} = [-81.4\hat{x} - 33.7\hat{y} - 226\hat{z}] \left[\frac{kV}{m} \right]$$

Problem 4.27

An infinitely long cylindrical shell extending between $r = 1 \text{ m}$ and $r = 3 \text{ m}$ contains a uniform charge density ρ_{v_0}

Apply Gauss' law to find D in all regions.

Gauss' Law:

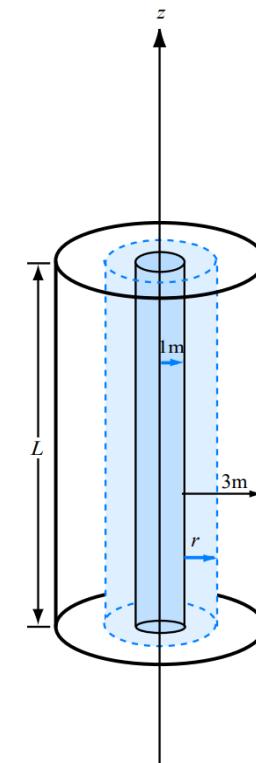
$$\oint_S (D_r \hat{r}) \cdot d\mathbf{s} = Q$$

Ignore axial direction



Out of symmetry

$$D_r \cdot 2\pi r = Q$$



Problem 4.27

Solution:

$$D_r \cdot 2\pi r = Q$$

For $r < 1 \text{ m}$, $Q = 0, D = 0$

For $1 \leq r \leq 3 \text{ m}$

$$Q = \rho_{v_0} \cdot \pi L(r^2 - 1)$$

$$D_r \cdot 2\pi r L = \rho_{v_0} \cdot \pi L(r^2 - 1)$$

$$D = D_r \hat{\mathbf{r}} = \frac{\rho_{v_0} \cdot \pi(r^2 - 1)}{2\pi r} \hat{\mathbf{r}} \quad \left[\frac{\text{C}}{\text{m}^2} \right]$$

Problem 4.27

Solution:

For $r \geq 3$ m,

$$Q = \rho_{v_0} \cdot \pi(3^2 - 1^2) = 8\rho_{v_0} \cdot \pi$$

$$D_r \cdot 2\pi r L = 8\rho_{v_0} \cdot \pi L$$

$$\mathbf{D} = D_r \hat{\mathbf{r}} = \frac{4\rho_{v_0}}{r} \hat{\mathbf{r}} \quad \left[\frac{\text{C}}{\text{m}^2} \right]$$

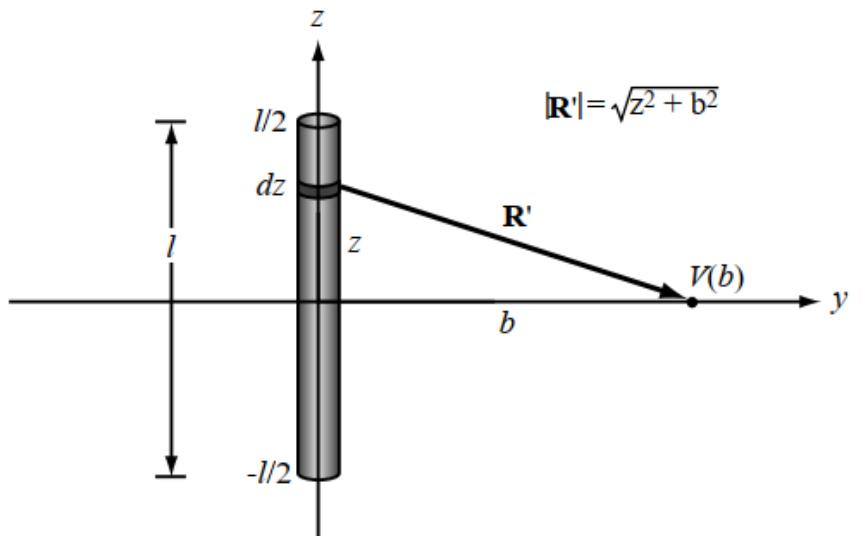
2A04 Tutorial 7

March 7th, 2022

Fraser McCauley & Jacob Saunders

1. Problem 4.34

Find the electric potential V at a location a distance b from the origin in the $x - y$ plane due to a line charge with charge density ρ_l and of length l . The line charge is coincident with the z axis and extends from $z = -l/2$ to $z = l/2$.



For a line charge, we have the formula:

$$V = \frac{1}{4\pi\epsilon} \int_{l'} \frac{\rho_l}{R'} dl'$$

In this specific case, we can integrate along z with the following limits:

$$V = \frac{\rho_l}{4\pi\epsilon} \int_{-l/2}^{l/2} \frac{dz}{R'}$$

$$V = \frac{\rho_l}{4\pi\epsilon} \int_{-l/2}^{l/2} \frac{dz}{\sqrt{z^2 + b^2}}$$

$$V = \frac{\rho_l}{4\pi\epsilon} \ln \left(\frac{\sqrt{l^2 + 4b^2} + l}{\sqrt{l^2 + 4b^2} - l} \right)$$

$$V = \frac{\rho_0}{4\pi\epsilon} \int_{-\ell/2}^{\ell/2} \frac{dz}{\sqrt{z^2 + b^2}}$$

$$\int \frac{1}{\sqrt{U^2 + a^2}} dU = \ln(U + \sqrt{U^2 + a^2}) / + C$$

$$\text{Let } C_1 = \frac{\rho_0}{4\pi\epsilon}$$

$$V = C_1 \ln(z + \sqrt{z^2 + b^2}) \Big|_{-\ell/2}^{\ell/2} = C_1 \left(\ln\left(\frac{\ell/2 + \sqrt{\ell^2/4 + b^2}}{-\ell/2 + \sqrt{\ell^2/4 + b^2}}\right) \right)$$

$$V = C_1 \ln\left(\frac{\frac{\ell/2 + \sqrt{\ell^2/4 + b^2}}{1}}{-\ell/2 + \sqrt{\ell^2/4 + b^2}}\right)$$

$$V = C_1 \ln\left(\frac{\ell + \sqrt{4(\ell^2/4 + b^2)}}{-\ell + \sqrt{4(\ell^2/4 + b^2)}}\right)$$

$$V = \frac{\rho_0}{4\pi\epsilon} \ln\left(\frac{\sqrt{\ell^2 + 4b^2} + \ell}{\sqrt{\ell^2 + 4b^2} - \ell}\right)$$

2. Problem 4.41

A cylindrical bar of silicon has a radius of 4mm and a length of 8cm. If a voltage of 5V is applied between the ends of the bar and $\mu_e = 0.13 \left(\frac{m^2}{Vs} \right)$, $\mu_h = 0.05 \left(\frac{m^2}{Vs} \right)$, $N_e = 1.5 \times 10^{16} \text{ electrons/m}^3$, and $N_h = N_e$, find the following:

- The conductivity of silicon,
- The current I flowing in the bar,
- The drift velocities \overrightarrow{u}_e and \overrightarrow{u}_h ,
- The resistance of the bar, and
- The power dissipated in the bar.

Givens

$$r = 4\text{mm}$$

$$l = 8\text{cm}$$

$$V = 5V$$

$$\mu_e = 0.13 \left(\frac{m^2}{Vs} \right)$$

$$\mu_h = 0.05 \left(\frac{m^2}{Vs} \right)$$

$$N_e = N_h = 1.5 \times 10^{16} \text{ electrons/m}^3$$

What does each parameter mean???

2. Problem 4.41a

A cylindrical bar of silicon has a radius of 4mm and a length of 8cm. If a voltage of 5V is applied between the ends of the bar and $\mu_e = 0.13 \left(\frac{m^2}{Vs} \right)$, $\mu_h = 0.05 \left(\frac{m^2}{Vs} \right)$, $N_e = 1.5 \times 10^{16} \text{ electrons/m}^3$, and $N_h = N_e$, find the following:

The conductivity of silicon

How can we calculate conductivity?

Givens

$$r = 4\text{mm}$$

$$l = 8\text{cm}$$

$$V = 5\text{V}$$

$$\mu_e = 0.13 \left(\frac{m^2}{Vs} \right)$$

$$\mu_h = 0.05 \left(\frac{m^2}{Vs} \right)$$

$$N_e = N_h = 1.5 \times 10^{16} \text{ electrons/m}^3$$

$$\sigma = (N_e \mu_e + N_h \mu_h) e \left(\frac{S}{m} \right)$$

$$\sigma = (1.5 \times 10^{16}(0.13 + 0.05))(1.602 \times 10^{-19})$$

$$\sigma = 4.33 \times 10^{-4} \left(\frac{S}{m} \right)$$

2. Problem 4.41a

A cylindrical bar of silicon has a radius of 4mm and a length of 8cm. If a voltage of 5V is applied between the ends of the bar and $\mu_e = 0.13 \left(\frac{m^2}{Vs} \right)$, $\mu_h = 0.05 \left(\frac{m^2}{Vs} \right)$, $N_e = 1.5 \times 10^{16} \text{ electrons/m}^3$, and $N_h = N_e$, find the following:

The conductivity of silicon

$$\sigma = 4.33 \times 10^{-4} \left(\frac{S}{m} \right)$$

Table B-2 Conductivity σ of Some Common Materials^a

Givens

$$r = 4\text{mm}$$

$$l = 8\text{cm}$$

$$V = 5\text{V}$$

$$\mu_e = 0.13 \left(\frac{m^2}{Vs} \right)$$

$$\mu_h = 0.05 \left(\frac{m^2}{Vs} \right)$$

$$N_e = N_h = 1.5 \times 10^{16} \text{ electrons/m}^3$$

| Material | Conductivity σ (S/m) | Material | Conductivity σ (S/m) | | |
|-----------------------|--------------------------------|----------------|--------------------------------|--|--|
| Conductors | | | Semiconductors | | |
| Silver | 6.2×10^7 | Pure germanium | 2.2 | | |
| Copper | 5.8×10^7 | Pure silicon | 4.4×10^{-4} | | |
| Gold | 4.1×10^7 | | | | |
| Aluminum | 3.5×10^7 | | | | |
| Tungsten | 1.8×10^7 | | | | |
| Zinc | 1.7×10^7 | | | | |
| Brass | 1.5×10^7 | | | | |
| Iron | 10^7 | | | | |
| Bronze | 10^7 | | | | |
| Tin | 9×10^6 | | | | |
| Lead | 5×10^6 | | | | |
| Mercury | 10^6 | | | | |
| Carbon | 3×10^4 | | | | |
| Seawater | 4 | | | | |
| Animal body (average) | 0.3 (poor cond.) | | | | |

^aThese are low-frequency values at room temperature (20 °C).

2. Problem 4.41b

A cylindrical bar of silicon has a radius of 4mm and a length of 8cm. If a voltage of 5V is applied between the ends of the bar and $\mu_e = 0.13 \left(\frac{m^2}{Vs} \right)$, $\mu_h = 0.05 \left(\frac{m^2}{Vs} \right)$, $N_e = 1.5 \times 10^{16} \text{ electrons/m}^3$, and $N_h = N_e$, find the following:

The current I flowing in the bar

How can we compute the current?

Givens

$$r = 4\text{mm}$$

$$l = 8\text{cm}$$

$$V = 5\text{V}$$

$$\mu_e = 0.13 \left(\frac{m^2}{Vs} \right)$$

$$\mu_h = 0.05 \left(\frac{m^2}{Vs} \right)$$

$$N_e = N_h = 1.5 \times 10^{16} \text{ electrons/m}^3$$

$$I = JA$$

$$I = \sigma EA$$

$$I = \sigma \frac{V}{l} \pi r^2$$

$$I = (4.33 \times 10^{-4}) \frac{5}{0.08} \pi 0.004^2$$

$$\mathbf{I = 1.36 \mu A}$$

2. Problem 4.41c

A cylindrical bar of silicon has a radius of 4mm and a length of 8cm. If a voltage of 5V is applied between the ends of the bar and $\mu_e = 0.13 \left(\frac{m^2}{Vs} \right)$, $\mu_h = 0.05 \left(\frac{m^2}{Vs} \right)$, $N_e = 1.5 \times 10^{16} \text{ electrons/m}^3$, and $N_h = N_e$, find the following:

The drift velocities $\overrightarrow{u_e}$ and $\overrightarrow{u_h}$

Givens

$$r = 4\text{mm}$$

$$l = 8\text{cm}$$

$$V = 5\text{V}$$

$$\mu_e = 0.13 \left(\frac{m^2}{Vs} \right)$$

$$\mu_h = 0.05 \left(\frac{m^2}{Vs} \right)$$

$$N_e = N_h = 1.5 \times 10^{16} \text{ electrons/m}^3$$

From the definition of the mobility:

$$\overrightarrow{u_e} = -\mu_e \vec{E}, \quad \overrightarrow{u_h} = \mu_h \vec{E}$$

$$\overrightarrow{u_e} = -\mu_e E \frac{\vec{E}}{E}, \quad \overrightarrow{u_h} = \mu_h E \frac{\vec{E}}{E}$$

What is the fraction at the end there? Why do we need it?

$$\overrightarrow{u_e} = -\mu_e \frac{V}{l} \hat{E}, \quad \overrightarrow{u_h} = \mu_h \frac{V}{l} \hat{E}$$

$$\overrightarrow{u_e} = -0.13 \frac{5}{0.08} \hat{E}, \quad \overrightarrow{u_h} = 0.05 \frac{5}{0.08} \hat{E}$$

$$\overrightarrow{u_e} = -8.125 \hat{E}, \quad \overrightarrow{u_h} = 3.125 \hat{E}$$

2. Problem 4.41d,e

A cylindrical bar of silicon has a radius of 4mm and a length of 8cm. If a voltage of 5V is applied between the ends of the bar and $\mu_e = 0.13 \left(\frac{m^2}{Vs} \right)$, $\mu_h = 0.05 \left(\frac{m^2}{Vs} \right)$, $N_e = 1.5 \times 10^{16} \text{ electrons/m}^3$, and $N_h = N_e$, find the following:

Resistance of, and power dissipated across, the bar.

Givens

$$r = 4\text{mm}$$

$$l = 8\text{cm}$$

$$V = 5\text{V}$$

$$\mu_e = 0.13 \left(\frac{m^2}{Vs} \right)$$

$$\mu_h = 0.05 \left(\frac{m^2}{Vs} \right)$$

$$N_e = N_h = 1.5 \times 10^{16} \text{ electrons/m}^3$$

We already found $I = 1.36\mu\text{A}$. How can we find the resistance?

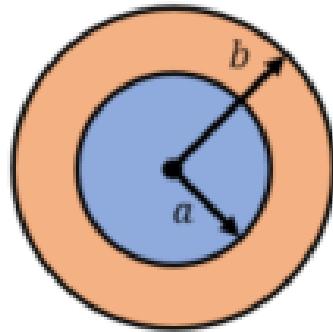
$$\text{Ohm's Law: } R = \frac{V}{I} = \frac{5}{1.36 \times 10^{-6}} = \mathbf{3.68 \text{ M}\Omega}$$

With the same parameters, we can find the power dissipated:

$$P = VI = 5(1.36 \times 10^{-6}) = \mathbf{6.8 \mu\text{W}}$$

3. Problem 4.44

A coaxial resistor of length l consists of two concentric cylinders. The inner cylinder has radius a and is made of a material with conductivity σ_1 , and the outer cylinder, extending between $r = a$ and $r = b$, is made of a material with conductivity σ_2 . If the two ends of the resistor are capped with conducting plates, show that the resistance between the two ends is



These are basically parallel resistors!

$$R = \left(\frac{1}{R_1} + \frac{1}{R_2} \right)^{-1}$$

$$R = \frac{l}{\pi(\sigma_1 a^2 + \sigma_2(b^2 - a^2))}$$

So how can we find R_1 and R_2 ?

Each R will follow $R = \frac{l}{\sigma A}$ where A is the surface area.

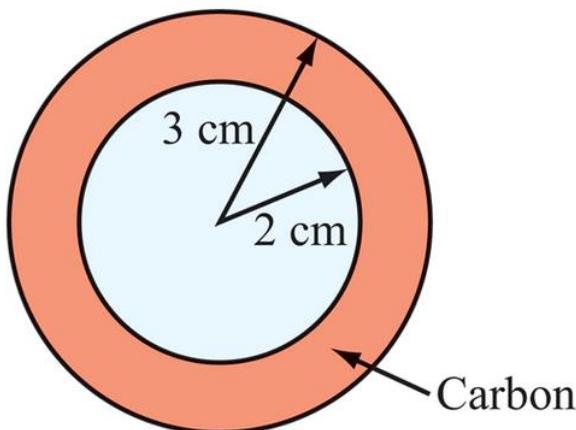
$$R = \left(\frac{1}{R_1} + \frac{1}{R_2} \right)^{-1} = \left(\frac{\sigma_1 A_1}{l} + \frac{\sigma_2 A_2}{l} \right)^{-1} \dots \text{what are the areas?}$$

$$R = \left(\frac{\sigma_1 \pi a^2}{l} + \frac{\sigma_2 \pi (b^2 - a^2)}{l} \right)^{-1} = \left(\frac{\sigma_1 \pi a^2 + \sigma_2 \pi (b^2 - a^2)}{l} \right)^{-1}$$

$$R = \frac{l}{\sigma_1 \pi a^2 + \sigma_2 \pi (b^2 - a^2)} = \frac{l}{\pi(\sigma_1 a^2 + \sigma_2(b^2 - a^2))}$$

4. Problem 4.45

Apply the result of Problem 4.44 (*that's the one we just did!*) to find the resistance of a 20 cm long hollow cylinder made of carbon with $\sigma = 3 \times 10^4 \text{ S/m}$.



$$R = \frac{l}{\pi(\sigma_1 a^2 + \sigma_2(b^2 - a^2))}.$$

Givens

$$a = 2\text{cm}$$

$$b = 3\text{cm}$$

$$\sigma = 3 \times 10^4 \text{ S/m}$$

$$l = 20\text{cm}$$

Which σ is the given value? And what's the other value?

Cylinder is hollow... so take $\sigma_1 = 0$ and $\sigma_2 = 3 \times 10^4 \text{ S/m}$

$$\begin{aligned} R &= \frac{0.2}{\pi((0)(0.02)^2 + (3 \times 10^4)(0.03^2 - 0.02^2))} \\ &= \frac{0.2}{\pi((3 \times 10^4)(5 \times 10^{-4}))} = 4.2 \text{ m}\Omega \end{aligned}$$

Reminders

- Assignment 7 is out, and is due at 8AM on March 14.
- Complete the mid-semester survey for a 5% bonus on Assignment 7!
Survey must be completed by Sunday March 13th.
- Have a great week!

ENG PHYS 2A04 Tutorial 8

Electricity and Magnetism

Your TAs today

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Chapter 4

Problem 4.48 - Question

Find \mathbf{E}_1 if $\mathbf{E}_2 = \hat{x}3 - \hat{y}2 + \hat{z}2 (\frac{V}{m})$, $\epsilon_1 = 2\epsilon_0$, $\epsilon_2 = 18\epsilon_0$ and the boundary has a surface charge density $\rho_s = 3.54 \times 10^{-11} (\frac{C}{m^2})$. What angle does \mathbf{E}_2 make with the z-axis?

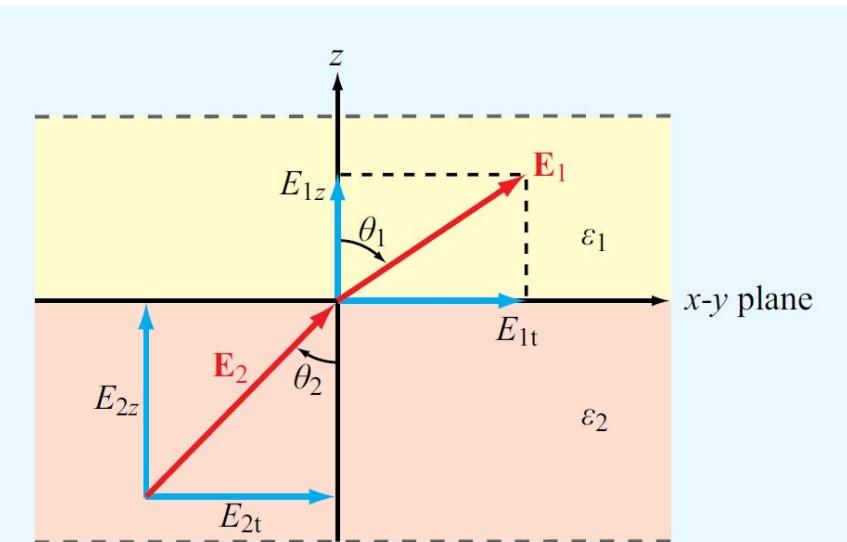


Figure 4-19 Application of boundary conditions at the interface between two dielectric media (Example 4-10).

Problem 4.48 - Question

Find \mathbf{E}_1 if $\mathbf{E}_2 = \hat{x}3 - \hat{y}2 + \hat{z}2 (\frac{V}{m})$, $\epsilon_1 = 2\epsilon_0$, $\epsilon_2 = 18\epsilon_0$ and the boundary has a surface charge density $\rho_s = 3.54 \times 10^{-11} (\frac{C}{m^2})$. What angle does \mathbf{E}_2 make with the z-axis?

- a) Find \mathbf{E}_1
- b) Find θ_2

\mathbf{E}_1 is comprised of 3 component vectors: \hat{x} , \hat{y} , \hat{z}

E_{1t} constitutes the \hat{x} and \hat{y} components

E_{1z} constitutes the \hat{z} component only

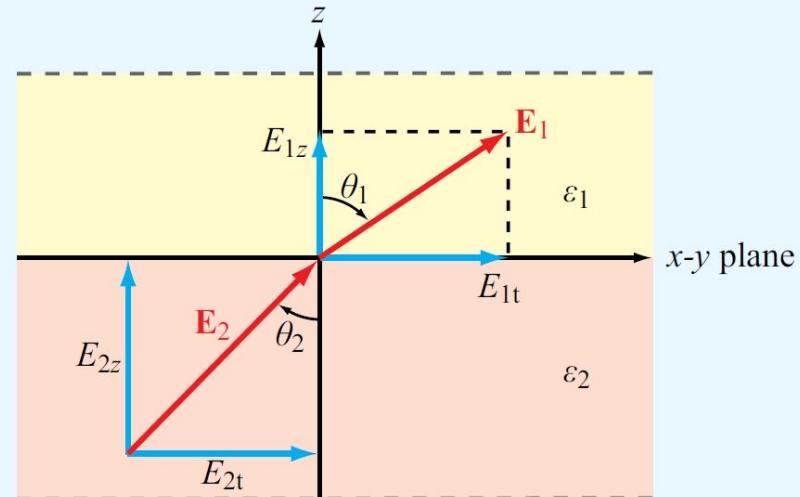


Figure 4-19 Application of boundary conditions at the interface between two dielectric media (Example 4-10).

Problem 4.48 – Work (a)

$$E_1 = E_{1t} + E_{1z}$$

$$E_{1t} = E_{2t} = \hat{x}3 - \hat{y}2 \left(\frac{V}{m}\right)$$

Table 4-3 Boundary conditions for the electric fields.

| Field Component | Any Two Media | Medium 1 Dielectric ϵ_1 | Medium 2 Conductor |
|-----------------|--|-------------------------------------|-----------------------|
| Tangential E | $E_{1t} = E_{2t}$ | $E_{1t} = E_{2t} = 0$ | |
| Tangential D | $D_{1t}/\epsilon_1 = D_{2t}/\epsilon_2$ | $D_{1t} = D_{2t} = 0$ | |
| Normal E | $\epsilon_1 E_{1n} - \epsilon_2 E_{2n} = \rho_s$ | $E_{1n} = \rho_s/\epsilon_1$ | $E_{2n} = 0$ |
| Normal D | $D_{1n} - D_{2n} = \rho_s$ | $D_{1n} = \rho_s$ | $D_{2n} = 0$ |

Notes: (1) ρ_s is the surface charge density at the boundary; (2) normal components of \mathbf{E}_1 , \mathbf{D}_1 , \mathbf{E}_2 , and \mathbf{D}_2 are along $\hat{\mathbf{n}}_2$, the outward normal unit vector of medium 2.

Problem 4.48 – Work (a)

$$E_1 = E_{1t} + E_{1z}$$

$$E_{1t} = E_{2t} = \hat{x}3 - \hat{y}2\left(\frac{V}{m}\right)$$

$$\begin{aligned}\epsilon_1(E_1 \cdot \hat{n}) - \epsilon_2(E_2 \cdot \hat{n}) &= \rho_s \\ \epsilon_1 E_{1z} - \epsilon_2 E_{2z} &= \rho_s\end{aligned}$$

$$\begin{aligned}E_{1z} &= \frac{\rho_s + \epsilon_2 E_{2z}}{\epsilon_1} \\ &= \frac{3.54 \times 10^{-11} + 18\epsilon_0 \cdot 2}{2\epsilon_0}\end{aligned}$$

$$E_{1z} = 20\left(\frac{V}{m}\right)$$

Table 4-3 Boundary conditions for the electric fields.

| Field Component | Any Two Media | Medium 1 Dielectric ϵ_1 | Medium 2 Conductor |
|-----------------|--|-------------------------------------|-----------------------|
| Tangential E | $E_{1t} = E_{2t}$ | $E_{1t} = E_{2t} = 0$ | |
| Tangential D | $D_{1t}/\epsilon_1 = D_{2t}/\epsilon_2$ | $D_{1t} = D_{2t} = 0$ | |
| Normal E | $\epsilon_1 E_{1n} - \epsilon_2 E_{2n} = \rho_s$ | $E_{1n} = \rho_s/\epsilon_1$ | $E_{2n} = 0$ |
| Normal D | $D_{1n} - D_{2n} = \rho_s$ | $D_{1n} = \rho_s$ | $D_{2n} = 0$ |

Notes: (1) ρ_s is the surface charge density at the boundary; (2) normal components of \mathbf{E}_1 , \mathbf{D}_1 , \mathbf{E}_2 , and \mathbf{D}_2 are along \hat{n}_2 , the outward normal unit vector of medium 2.

Problem 4.48 – Work (a)

$$E_1 = E_{1t} + E_{1z}$$

$$E_{1t} = E_{2t} = \hat{x}3 - \hat{y}2\left(\frac{V}{m}\right)$$

$$E_{1z} = 20\left(\frac{V}{m}\right)$$

**Therefore, this is
the answer for E_1**

$$E_1 = E_{1t} + E_{1z}$$

$$E_1 = \hat{x}3 - \hat{y}2 + \hat{z}20\left(\frac{V}{m}\right)$$

Table 4-3 Boundary conditions for the electric fields.

| Field Component | Any Two Media | Medium 1 Dielectric ϵ_1 | Medium 2 Conductor |
|-----------------|--|-------------------------------------|-----------------------|
| Tangential E | $E_{1t} = E_{2t}$ | $E_{1t} = E_{2t} = 0$ | |
| Tangential D | $D_{1t}/\epsilon_1 = D_{2t}/\epsilon_2$ | $D_{1t} = D_{2t} = 0$ | |
| Normal E | $\epsilon_1 E_{1n} - \epsilon_2 E_{2n} = \rho_s$ | $E_{1n} = \rho_s/\epsilon_1$ | $E_{2n} = 0$ |
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Notes: (1) ρ_s is the surface charge density at the boundary; (2) normal components of \mathbf{E}_1 , \mathbf{D}_1 , \mathbf{E}_2 , and \mathbf{D}_2 are along $\hat{\mathbf{n}}_2$, the outward normal unit vector of medium 2.

Problem 4.48 - Question

Find \mathbf{E}_1 if $\mathbf{E}_2 = \hat{x}3 - \hat{y}2 + \hat{z}2 (\frac{V}{m})$, $\epsilon_1 = 2\epsilon_0$, $\epsilon_2 = 18\epsilon_0$ and the boundary has a surface charge density $\rho_s = 3.54 \times 10^{-11} (\frac{C}{m^2})$. What angle does \mathbf{E}_2 make with the z-axis?

a) Find \mathbf{E}_1 $E_1 = \hat{x}3 - \hat{y}2 + \hat{z}20 (\frac{V}{m})$

b) Find θ_2

Use trig!

$$E_{2z} = |\mathbf{E}_2| \cos \theta_2$$

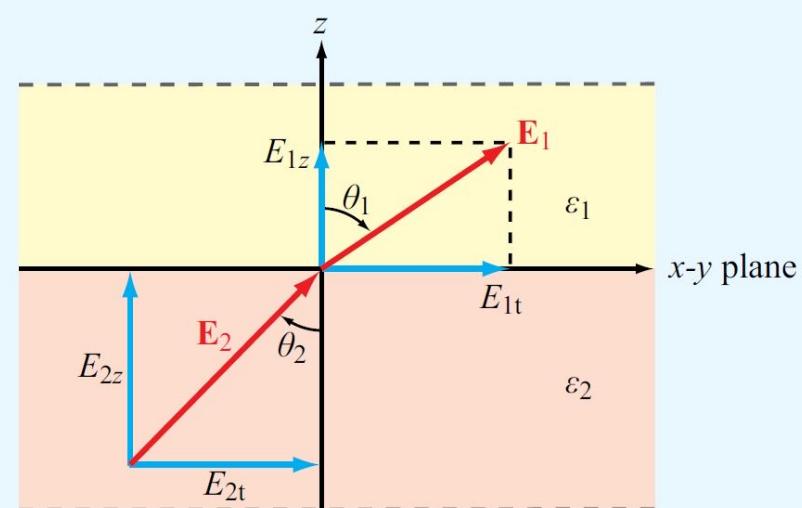


Figure 4-19 Application of boundary conditions at the interface between two dielectric media (Example 4-10).

Problem 4.48 - Question

Find \mathbf{E}_1 if $\mathbf{E}_2 = \hat{x}3 - \hat{y}2 + \hat{z}2 (\frac{V}{m})$, $\epsilon_1 = 2\epsilon_0$, $\epsilon_2 = 18\epsilon_0$ and the boundary has a surface charge density $\rho_s = 3.54 \times 10^{-11} (\frac{C}{m^2})$. What angle does \mathbf{E}_2 make with the z-axis?

a) Find \mathbf{E}_1 $E_1 = \hat{x}3 - \hat{y}2 + \hat{z}20 (\frac{V}{m})$

b) Find θ_2

Use trig!

$$E_{2z} = |\mathbf{E}_2| \cos \theta_2$$

$$\theta_2 = \cos^{-1} \left(\frac{E_{2z}}{|\mathbf{E}_2|} \right)$$

$$= \cos^{-1} \left(\frac{2}{\sqrt{3^2 + 2^2 + 2^2}} \right)$$

$$= 61^\circ$$

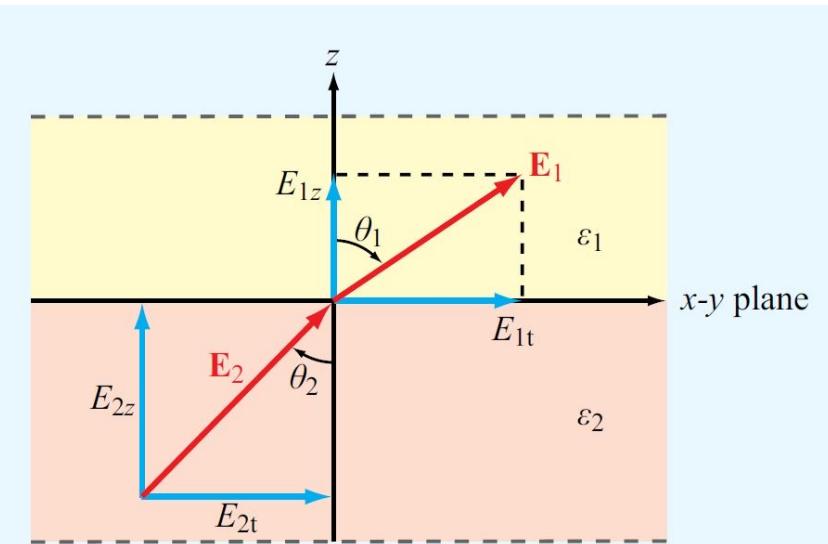


Figure 4-19 Application of boundary conditions at the interface between two dielectric media (Example 4-10).

Problem 4.48 - Question

Find \mathbf{E}_1 if $\mathbf{E}_2 = \hat{x}3 - \hat{y}2 + \hat{z}2 (\frac{V}{m})$, $\epsilon_1 = 2\epsilon_0$, $\epsilon_2 = 18\epsilon_0$ and the boundary has a surface charge density $\rho_s = 3.54 \times 10^{-11} (\frac{C}{m^2})$. What angle does \mathbf{E}_2 make with the z-axis?

a) Find \mathbf{E}_1 $E_1 = \hat{x}3 - \hat{y}2 + \hat{z}20 (\frac{V}{m})$

b) Find θ_2 $\theta_2 = 61^\circ$

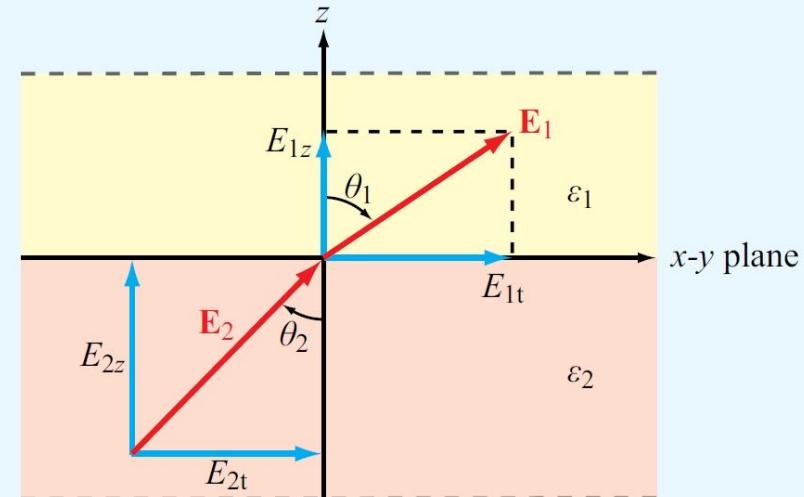


Figure 4-19 Application of boundary conditions at the interface between two dielectric media (Example 4-10).

***4.50** If $\mathbf{E} = \hat{\mathbf{R}}150$ (V/m) at the surface of a 5-cm conducting sphere centered at the origin, what is the total charge Q on the sphere's surface?

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Solution: From Table 4-3, $\hat{\mathbf{n}} \cdot (\mathbf{D}_1 - \mathbf{D}_2) = \rho_s$. \mathbf{E}_2 inside the sphere is zero, since we assume it is a perfect conductor. Hence, for a sphere with surface area $S = 4\pi a^2$,

$$D_{1R} = \rho_s, \quad E_{1R} = \frac{\rho_s}{\epsilon_0} = \frac{Q}{S\epsilon_0},$$

Table 4-3 Boundary conditions for the electric fields.

| Field Component | Any Two Media | Medium 1 Dielectric ϵ_1 | Medium 2 Conductor |
|-------------------------|---|---|-----------------------|
| Tangential \mathbf{E} | $E_{1t} = E_{2t}$ | $E_{1t} = E_{2t} = 0$ | |
| Tangential \mathbf{D} | $\mathbf{D}_{1t}/\epsilon_1 = \mathbf{D}_{2t}/\epsilon_2$ | $\mathbf{D}_{1t} = \mathbf{D}_{2t} = 0$ | |
| Normal \mathbf{E} | $\epsilon_1 E_{1n} - \epsilon_2 E_{2n} = \rho_s$ | $E_{1n} = \rho_s/\epsilon_1$ | $E_{2n} = 0$ |
| Normal \mathbf{D} | $D_{1n} - D_{2n} = \rho_s$ | $D_{1n} = \rho_s$ | $D_{2n} = 0$ |

Notes: (1) ρ_s is the surface charge density at the boundary; (2) normal components of \mathbf{E}_1 , \mathbf{D}_1 , \mathbf{E}_2 , and \mathbf{D}_2 are along $\hat{\mathbf{n}}_2$, the outward normal unit vector of medium 2.

***4.50** If $\mathbf{E} = \hat{\mathbf{R}}150$ (V/m) at the surface of a 5-cm conducting sphere centered at the origin, what is the total charge Q on the sphere's surface?

Solution: From Table 4-3, $\hat{\mathbf{n}} \cdot (\mathbf{D}_1 - \mathbf{D}_2) = \rho_s$. \mathbf{E}_2 inside the sphere is zero, since we assume it is a perfect conductor. Hence, for a sphere with surface area $S = 4\pi a^2$,

$$D_{1R} = \rho_s, \quad E_{1R} = \frac{\rho_s}{\epsilon_0} = \frac{Q}{S\epsilon_0},$$

$$Q = E_R S \epsilon_0 = (150) 4\pi (0.05)^2 \epsilon_0 = \frac{3\pi \epsilon_0}{2} \quad (\text{C}).$$

Table 4-3 Boundary conditions for the electric fields.

| Field Component | Any Two Media | Medium 1 Dielectric ϵ_1 | Medium 2 Conductor |
|---------------------|---|---|-----------------------|
| Tangential E | $E_{1t} = E_{2t}$ | $E_{1t} = E_{2t} = 0$ | |
| Tangential D | $\mathbf{D}_{1t}/\epsilon_1 = \mathbf{D}_{2t}/\epsilon_2$ | $\mathbf{D}_{1t} = \mathbf{D}_{2t} = 0$ | |
| Normal E | $\epsilon_1 E_{1n} - \epsilon_2 E_{2n} = \rho_s$ | $E_{1n} = \rho_s/\epsilon_1$ | $E_{2n} = 0$ |
| Normal D | $D_{1n} - D_{2n} = \rho_s$ | $D_{1n} = \rho_s$ | $D_{2n} = 0$ |

Notes: (1) ρ_s is the surface charge density at the boundary; (2) normal components of \mathbf{E}_1 , \mathbf{D}_1 , \mathbf{E}_2 , and \mathbf{D}_2 are along $\hat{\mathbf{n}}_2$, the outward normal unit vector of medium 2.

Problem 4.54 - Question

4.54 An electron with charge $Q_e = -1.6 \times 10^{-19} \text{ C}$ and mass $m_e = 9.1 \times 10^{-31} \text{ kg}$ is injected at a point adjacent to the negatively charged plate in the region between the plates of an air-filled parallel-plate capacitor with separation of 1 cm and rectangular plates each 10 cm^2 in area (**Fig. P4.54**). If the voltage across the capacitor is 10 V, find the following:

- (a) The force acting on the electron.
- (b) The acceleration of the electron.
- (c) The time it takes the electron to reach the positively charged plate, assuming that it starts from rest.

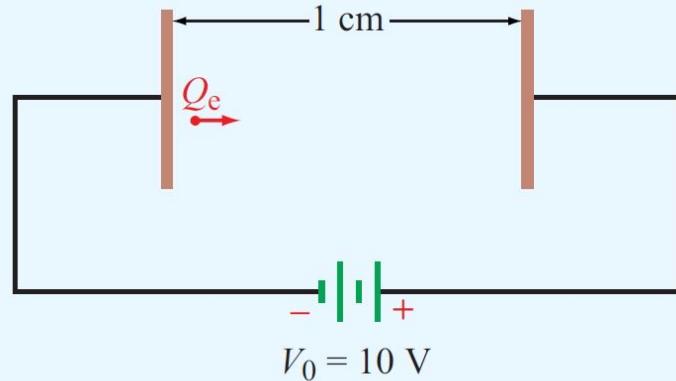


Figure P4.54 Electron between charged plates of Problem 4.54.

Problem 4.54 - Solution

4.54 An electron with charge $Q_e = -1.6 \times 10^{-19} \text{ C}$ and mass $m_e = 9.1 \times 10^{-31} \text{ kg}$ is injected at a point adjacent to the negatively charged plate in the region between the plates of an air-filled parallel-plate capacitor with separation of 1 cm and rectangular plates each 10 cm^2 in area (Fig. P4.54). If the voltage across the capacitor is 10 V, find the following:

- (a) The force acting on the electron.
- (b) The acceleration of the electron.
- (c) The time it takes the electron to reach the positively charged plate, assuming that it starts from rest.

Part = a

$$F = Q_e E = Q_e \frac{V}{d}$$

Putting Values in above equation we get

$$F = -1.6 \times \frac{10}{0.01} = -1.6 \times 10^{-16} \text{ N}$$

Part = b

$$a = \frac{F}{m} = \frac{1.6 \times 10^{-16}}{9.1 \times 10^{-31}} = 1.76 \times 10^{14} \text{ m/s}^2$$

Part = c

$$t = \sqrt{\frac{2d}{a}} = \left(\frac{2 \times 0.01}{1.76 \times 10^{14}} \right)^{1/2}$$

$$t = 10.7 \times 10^{-9} \text{ sec} = 10 \cdot 7(ns)$$

Problem 4.58 - Question

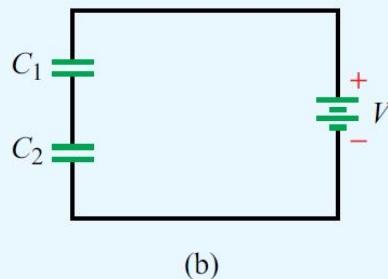
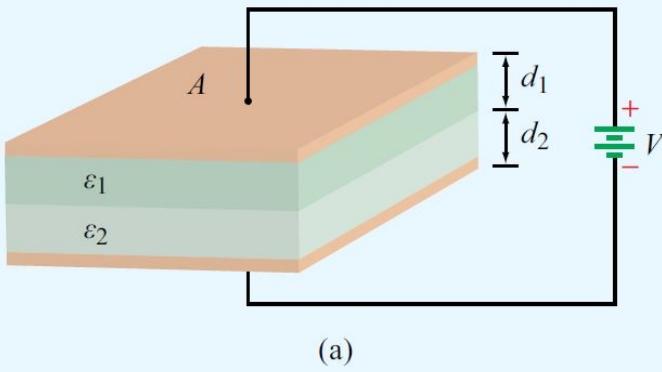


Figure P4.58 (a) Capacitor with parallel dielectric layers, and (b) equivalent circuit (Problem 4.58).

4.58 The capacitor shown in [Fig. P4.58](#) consists of two parallel dielectric layers. Use energy considerations to show that the equivalent capacitance of the overall capacitor, C , is equal to the series combination of the capacitances of the individual layers, C_1 and C_2 , namely

$$C = \frac{C_1 C_2}{C_1 + C_2} \quad (4.136)$$

where

$$C_1 = \epsilon_1 \frac{A}{d_1}, \quad C_2 = \epsilon_2 \frac{A}{d_2}.$$

- Let V_1 and V_2 be the electric potentials across the upper and lower dielectrics, respectively. What are the corresponding electric fields E_1 and E_2 ? By applying the appropriate boundary condition at the interface between the two dielectrics, obtain explicit expressions for E_1 and E_2 in terms of ϵ_1 , ϵ_2 , V , and the indicated dimensions of the capacitor.
- Calculate the energy stored in each of the dielectric layers and then use the sum to obtain an expression for C .
- Show that C is given by [Eq. \(4.136\)](#).

Part = a (Solution)

If V_1 is the voltage across the top layer and V_2 across the bottom layer, then

$$V = V_1 + V_2$$

and,

$$V_1 = E_1 d_1 \quad \text{and} \quad V_2 = E_2 d_2$$

So,

$$V = E_1 d_1 + E_2 d_2 \longrightarrow (\text{A})$$

According to boundary conditions, the normal component of \mathbf{D} is continuous across the boundary

$$D_{1n} = D_{2n}$$

From Equation 4.15

$$D = \epsilon E$$

So,

$$\epsilon_1 E_1 = \epsilon_2 E_2 \longrightarrow E_2 = \frac{\epsilon_1 E_1}{\epsilon_2}$$

$$(\text{A}) \longrightarrow V = E_1 d_1 + \frac{\epsilon_1 E_1}{\epsilon_2} d_2 \longrightarrow E_1 = \frac{V}{d_1 + \frac{\epsilon_1}{\epsilon_2} d_2} \quad \text{and} \quad E_2 = \frac{V}{d_2 + \frac{\epsilon_2}{\epsilon_1} d_1}$$

Part = b (Solution)

From Eq= 4.122, stored potential energy is given by,

$$W_e = \frac{1}{2} \varepsilon E^2 (Ad)$$

So,

$$W_{e1} = \frac{1}{2} \varepsilon_1 E_1^2 (Ad_1)$$

Replacing E_1 by

$$E_1 = \frac{V}{d_1 + \frac{\varepsilon_1}{\varepsilon_2} d_2} \quad \longrightarrow$$

$$W_{e1} = \frac{1}{2} \varepsilon_1 \left(\frac{V}{d_1 + \frac{\varepsilon_1}{\varepsilon_2} d_2} \right)^2 (Ad_1) \quad \longrightarrow$$

$$W_{e1} = \frac{1}{2} V^2 \left(\frac{\varepsilon_1 \varepsilon_2^2 A d_1}{(\varepsilon_2 d_1 + \varepsilon_1 d_2)^2} \right)$$

$$W_{e1} = \frac{1}{2} V^2 \left(\frac{\varepsilon_1 \varepsilon_2^2 A d_1}{(\varepsilon_2 d_1 + \varepsilon_1 d_2)^2} \right)$$

$$\text{And} \quad W_{e2} = \frac{1}{2} V^2 \left(\frac{\varepsilon_1^2 \varepsilon_2 A d_2}{(\varepsilon_1 d_2 + \varepsilon_2 d_1)^2} \right)$$

$$W_e = W_1 + W_2 = \frac{1}{2} V^2 \left(\frac{\varepsilon_1 \varepsilon_2^2 A d_1 + \varepsilon_1^2 \varepsilon_2 A d_2}{(\varepsilon_1 d_2 + \varepsilon_2 d_1)^2} \right)$$

But,

$$w_e = \frac{1}{2} C V^2$$

$$C = \frac{\varepsilon_1 \varepsilon_2^2 A d_1 + \varepsilon_1^2 \varepsilon_2 A d_2}{(\varepsilon_2 d_1 + \varepsilon_1 d_2)^2}$$

$$\varepsilon_1 \varepsilon_2 A \frac{(\varepsilon_2 d_1 + \varepsilon_1 d_2)}{(\varepsilon_2 d_1 + \varepsilon_1 d_2)^2}$$

$$\frac{\varepsilon_1 \varepsilon_2 A}{(\varepsilon_2 d_1 + \varepsilon_1 d_2)}$$

Part = c (Solution)

From solution of part b we have,

$$C = \frac{\varepsilon_1 \varepsilon_2 A}{(\varepsilon_2 d_1 + \varepsilon_1 d_2)}$$

Multiply numerator and denominator by $A/d_1 d_2$

$$C = \frac{\varepsilon_1 \varepsilon_2 A (A/d_1 d_2)}{(\varepsilon_2 d_1 + \varepsilon_1 d_2)(A/d_1 d_2)}$$

$$C = \frac{\frac{\varepsilon_1 A}{d_1} \cdot \frac{\varepsilon_2 A}{d_2}}{\frac{A \varepsilon_2 d_1}{d_1 d_2} + \frac{A \varepsilon_1 d_2}{d_1 d_2}} = \frac{\frac{\varepsilon_1 A}{d_1} \cdot \frac{\varepsilon_2 A}{d_2}}{\frac{A \varepsilon_2}{d_2} + \frac{A \varepsilon_1}{d_1}} = \frac{C_1 C_2}{C_1 + C_2}$$

$$C_2 = \frac{\varepsilon_2 A}{d_2} \quad C_1 = \frac{\varepsilon_1 A}{d_1}$$

ENGPHYS 2A04 TUTORIAL 9

ELECTRICITY & MAGNETISM

Your TAs Today

- Joanne Lee

leej298@mcmaster.ca

- Muhammad Munir

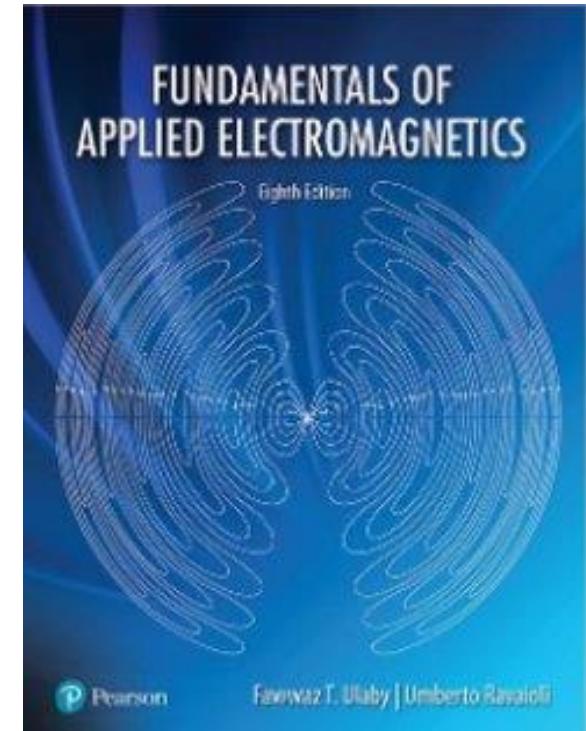
munirm6@mcmaster.ca

Your Textbook

Fundamentals of Applied Electromagnetics Eighth Edition

Ulaby & Ravaioli

Seventh Edition also acceptable, with
some inconsistencies



Problem 5.1

An electron with a speed of 8×10^6 m/s is projected along the positive x direction into a medium containing a uniform magnetic flux density $\mathbf{B} = (\hat{x}4 - \hat{z}3)T$

Given that $e = 1.6 \times 10^{19}$ C and the mass of an electron is $m_e = 9.1 \times 10^{-31}$ kg, determine the initial acceleration vector of the electron (at the moment it is projected into the medium).

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Given that $e = 1.6 \times 10^{19}$ C and the mass of an electron is $m_e = 9.1 \times 10^{-31}$ kg, determine the initial acceleration vector of the electron (at the moment it is projected into the medium).

Known Values:

$$\text{Electron speed: } u = 8 \times 10^6 \frac{m}{s}$$

$$\text{Magnetic Flux Density: } \mathbf{B} = (\hat{x}4 - \hat{z}3)T$$

$$\text{Elementary Charge: } e = 1.6 \times 10^{19} C$$

$$\text{Electron Mass: } m_e = 9.1 \times 10^{-31} kg$$

Problem 5.1

Particle of a charge q moving with velocity \mathbf{u} in a magnetic field experiences magnetic force \mathbf{F}_m given by:

$$\text{Electron speed: } u = 8 * 10^6 \frac{\text{m}}{\text{s}}$$

$$\text{Magnetic Flux Density: } \mathbf{B} = (\hat{x}4 - \hat{z}3)\text{T}$$

$$\text{Elementary Charge: } e = 1.6 * 10^{19} \text{ C}$$

$$\text{Electron Mass: } m_e = 9.1 * 10^{-31} \text{ kg}$$

$$\mathbf{F}_m = q\mathbf{u} \times \mathbf{B} \quad (\text{N})$$

Use Newton's Second Law: $F = m * a$

Rearrange equation and substitute \mathbf{F} for equation above.

Problem 5.1

$$\mathbf{a} = \frac{\mathbf{F}_m}{m_e} = \frac{q\mathbf{u} \times \mathbf{B}}{m_e}$$

Assuming $q = -e$

$$= \frac{-1.6 * 10^{-19}}{9.1 * 10^{-31}} (\hat{x}8 * 10^6) \times (\hat{x}4 - \hat{z}3)$$

$$= -\hat{y}4.22 * 10^{18} \quad (m/s^2)$$

$$\text{Electron speed: } u = 8 * 10^6 \frac{m}{s}$$

$$\text{Magnetic Flux Density: } \mathbf{B} = (\hat{x}4 - \hat{z}3)\mathbf{T}$$

$$\text{Elementary Charge: } e = 1.6 * 10^{19} C$$

$$\text{Electron Mass: } m_e = 9.1 * 10^{-31} kg$$

$$\begin{aligned}\overline{\mathbf{a}} \times \overline{\mathbf{b}} &= \begin{vmatrix} \mathbf{i} & \mathbf{j} & \mathbf{k} \\ a_x & a_y & a_z \\ b_x & b_y & b_z \end{vmatrix} = \begin{vmatrix} \mathbf{i} & \mathbf{j} & \mathbf{k} \\ 8000000 & 0 & 0 \\ 4 & 0 & -3 \end{vmatrix} = \mathbf{i}(0 \cdot (-3) - 0 \cdot 0) - \mathbf{j}(8000000 \cdot (-3) - 0 \cdot 4) + \mathbf{k}(8000000 \cdot 0 - 0 \cdot 4) = \\ &= \mathbf{i}(0 - 0) - \mathbf{j}(-24000000 - 0) + \mathbf{k}(0 - 0) = \{0; 24000000; 0\}\end{aligned}$$

Problem 5.4

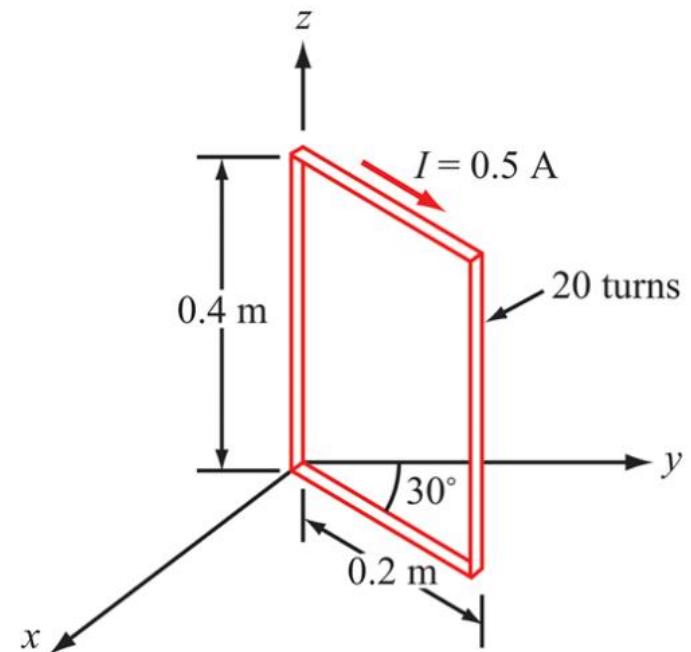
The rectangular loop shown in Fig. P5.4 consists of 20 closely wrapped turns and is hinged along the z axis. The plane of the loop makes an angle of 30° with the y axis, and the current in the windings is 0.5 A. What is the magnitude of the torque exerted on the loop in the presence of a uniform field $\mathbf{B} = \hat{\mathbf{y}}2.4$ T? When viewed from above, is the expected direction of rotation clockwise or counterclockwise?

(5.19)

$$\mathbf{m} = \hat{\mathbf{n}}NIA = \hat{\mathbf{n}}m \quad (\text{A} \cdot \text{m}^2),$$

(5.20)

$$\mathbf{T} = \mathbf{m} \times \mathbf{B} \quad (\text{N} \cdot \text{m}).$$



Problem 5.4

Magnetic Moment m :

$$\mathbf{m} = \hat{\mathbf{n}}NIA = \hat{\mathbf{n}}m (A \cdot m^2)$$

where $\hat{\mathbf{n}}$ is the surface normal of the loop and governed by the following *right-hand rule*: *When the four fingers of the right-hand advance in the direction of the current I , the direction of the thumb specifies the direction of $\hat{\mathbf{n}}$.*

$$N = 20$$

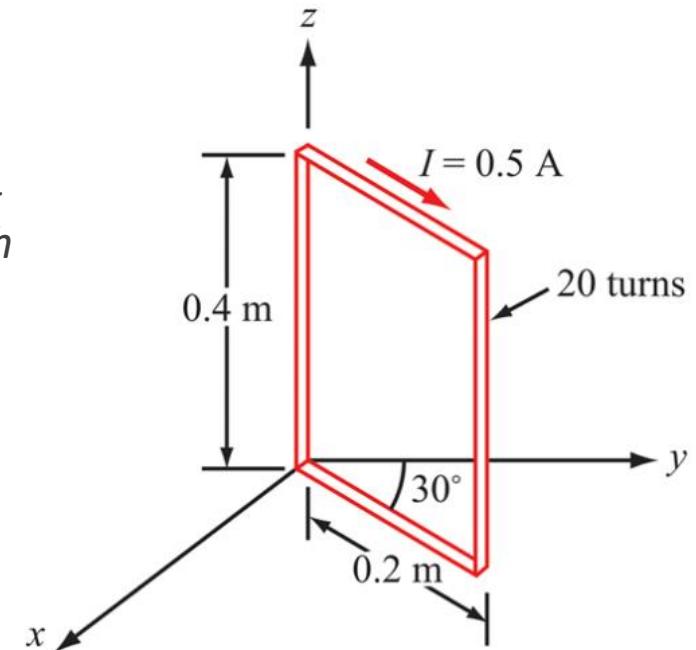
$$I = 0.5 \text{ A}$$

$$A = l \times w = 0.2 \text{ m} \times 0.4 \text{ m} = 0.08 \text{ m}^2$$

$$\therefore m = NIA = (20)(0.5 \text{ A})(0.08 \text{ m}^2) = 0.8 \text{ A} \cdot \text{m}^2$$

(5.19)

$$\mathbf{m} = \hat{\mathbf{n}}NIA = \hat{\mathbf{n}}m (A \cdot m^2),$$



Problem 5.4

Magnetic Moment m :

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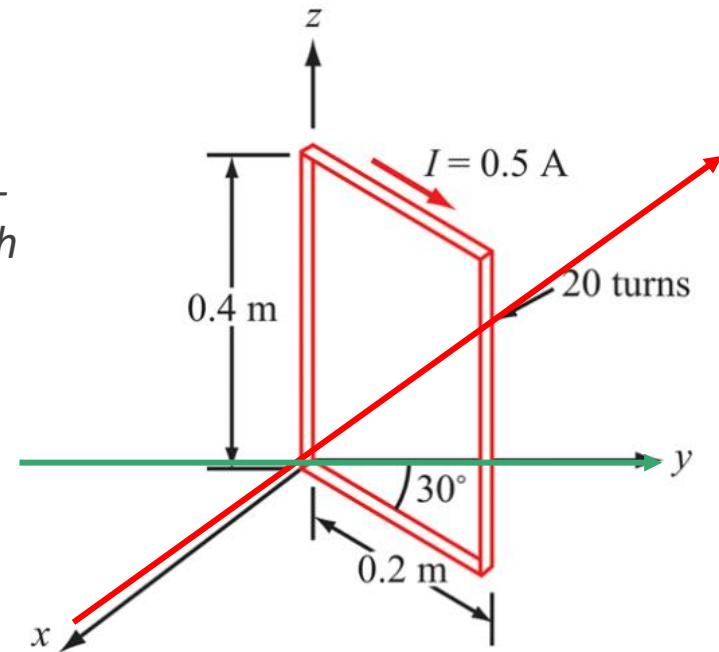
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(5.19)

$$\mathbf{m} = \hat{\mathbf{n}}NIA = \hat{\mathbf{n}}m (A \cdot m^2),$$



Problem 5.4

Magnetic Moment m:

$$\mathbf{m} = \hat{\mathbf{n}}NIA = \hat{\mathbf{n}}m (A \cdot m^2)$$

$$N = 20$$

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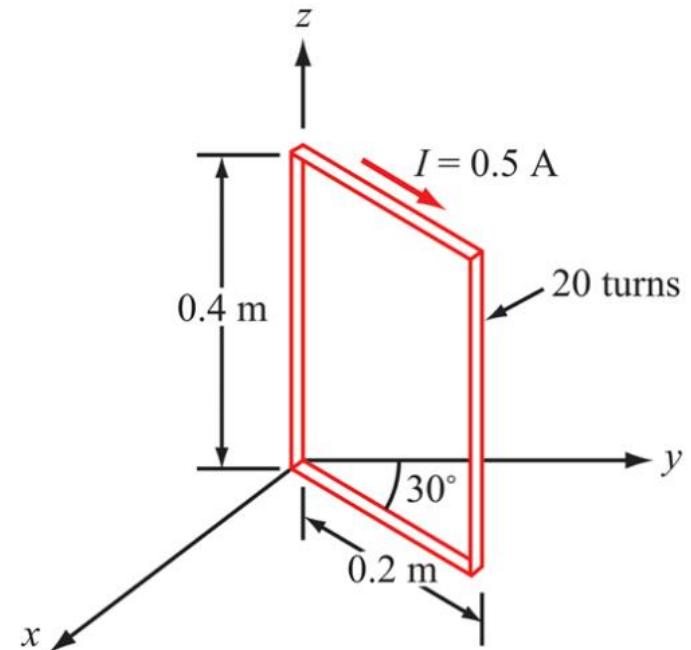
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$$\therefore m = NIA = (20)(0.5 \text{ A})(0.08 \text{ m}^2) = 0.8 \text{ A} \cdot \text{m}^2$$

$$\hat{\mathbf{n}} = -\hat{\mathbf{x}} \cos 30^\circ + \hat{\mathbf{y}} \sin 30^\circ$$

(5.19)

$$\mathbf{m} = \hat{\mathbf{n}}NIA = \hat{\mathbf{n}}m \quad (\text{A} \cdot \text{m}^2),$$



Problem 5.4

Magnetic Moment m:

$$\mathbf{m} = \hat{\mathbf{n}}NIA = \hat{\mathbf{n}}m (A \cdot m^2)$$

$$N = 20$$

$$I = 0.5 \text{ A}$$

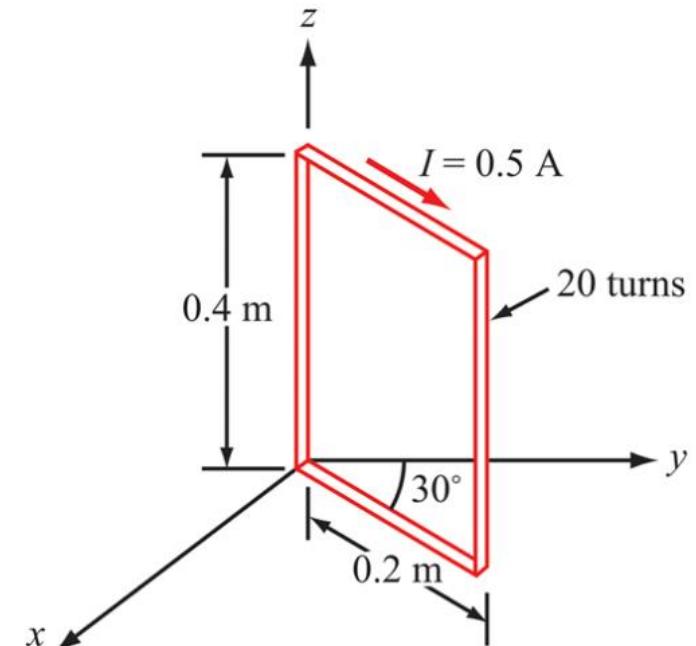
$$A = l \times w = 0.2 \text{ m} \times 0.4 \text{ m} = 0.08 \text{ m}^2$$

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$$\hat{\mathbf{n}} = -\hat{\mathbf{x}} \cos 30^\circ + \hat{\mathbf{y}} \sin 30^\circ$$

Torque is defined as:

$$\begin{aligned}\mathbf{T} &= \mathbf{m} \times \mathbf{B} = \hat{\mathbf{n}}m \times \mathbf{B} \\ &= (-\hat{\mathbf{x}} \cos 30^\circ + \hat{\mathbf{y}} \sin 30^\circ)0.8 \times \hat{\mathbf{y}}2.4 \cong -1.66 \hat{\mathbf{z}} \text{ N} \cdot \text{m}\end{aligned}$$



Negative torque indicates clockwise.

Problem 5.7

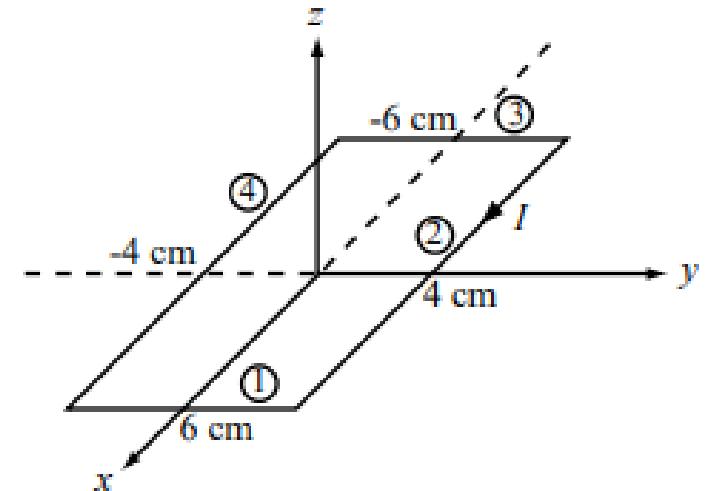
An 8 cm x 12 cm rectangular loop of wire is situation in the x-y plane with the center of the loop at the origin and its long sides parallel to the x-axis. The loop has a current of 50 flowing with clockwise direction (when viewed from above). Determine the magnetic field at the center of the loop.

$$\text{Biot Savart Law: } d\mathbf{H} = \frac{1}{4\pi} \frac{dl \times \hat{\mathbf{R}}}{R^2} \left[\frac{A}{m} \right]$$

Where $d\mathbf{H}$ = differential magnetic field intensity

dl = differential length vector

$\hat{\mathbf{R}}$ = distance vector between dl and the observation point



Problem 5.7

Biot Savart Law: $d\mathbf{H} = \frac{1}{4\pi} \frac{dl \times \hat{\mathbf{R}}}{R^2} \left[\frac{A}{m} \right] \rightarrow \mathbf{H} = \int_l d\mathbf{H}$

$$\mathbf{B} = \mu \mathbf{H}$$

Break conductor into 4 segments and calculate each segment's contribution to total magnetic field.

Segment 1 (blue circle):

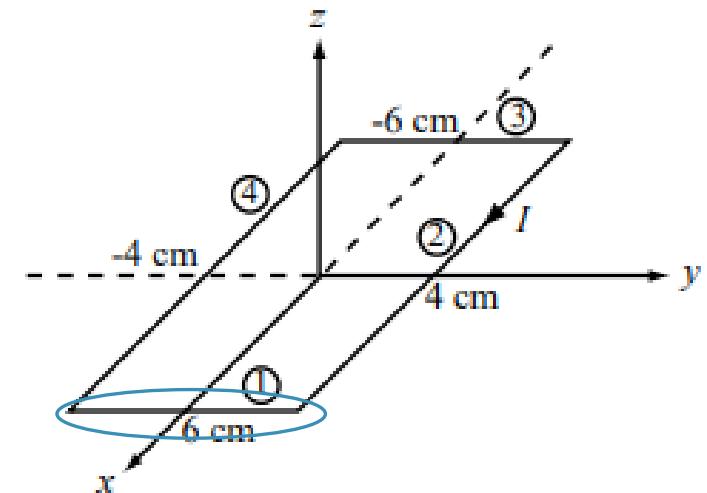
Right hand Rule gives direction of magnetic field, $\therefore \mathbf{B}_1$ is along $-z$ direction.

Using equation for wire of finite length.

$$|\mathbf{B}| = \mu |\mathbf{H}| = \mu_0 \frac{Il}{2\pi r \sqrt{4r^2 + l^2}} [T]$$

Using:

$$\begin{aligned}\mu_0 &= \mu = 4\pi \times 10^{-7} N/A^2 \\ I &= 50 A \\ r &= 6 cm \\ l &= 8 cm\end{aligned}$$



Problem 5.7

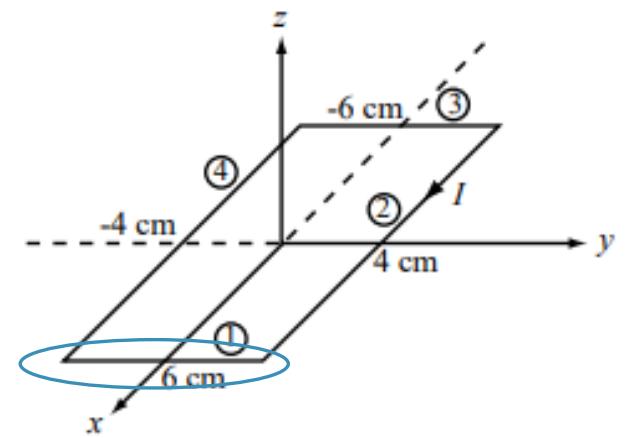
$$|\mathbf{B}| = \mu |\mathbf{H}| = \mu_0 \frac{Il}{2\pi r \sqrt{4r^2+l^2}} \quad [T]$$

Segment 1 (blue circle):

$$|\mathbf{B}_1| = \mu |\mathbf{H}| = \mu_0 \frac{Il}{2\pi r \sqrt{4r^2+l^2}} (-\mathbf{z})$$

$$= 4\pi * 10^{-7} NA^{-2} \frac{50 A * 0.08 m}{2\pi(0.06 m) \sqrt{4(0.06 m)^2 + (0.08 m)^2}} (-\mathbf{z})$$
$$= -9.24 * 10^{-5} \mathbf{z} \quad [T]$$

$$\mathbf{B}_1 = -9.24 * 10^{-5} \mathbf{z} \quad [T]$$



Problem 5.7

$$|\mathbf{B}| = \mu |\mathbf{H}| = \mu_0 \frac{Il}{2\pi r \sqrt{4r^2 + l^2}} \quad [T]$$

Segment 2 (blue circle):

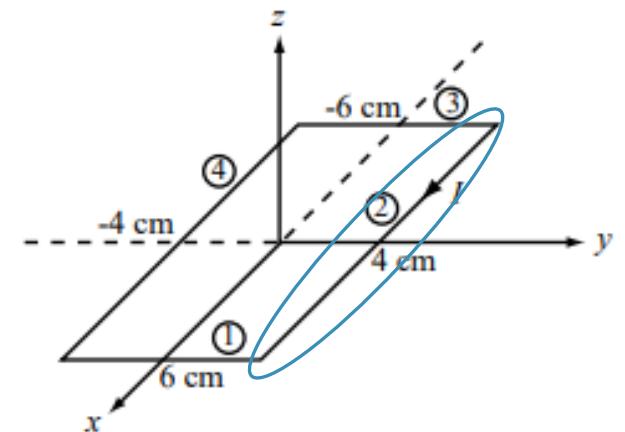
\mathbf{B}_2 is along $-z$ direction.

$$|\mathbf{B}_2| = \mu |\mathbf{H}| = \mu_0 \frac{Il}{2\pi r \sqrt{4r^2 + l^2}} (-\mathbf{z})$$

$$= 4\pi * 10^{-7} NA^{-2} \frac{50 A * 0.12 m}{2\pi(0.04 m) \sqrt{4(0.04 m)^2 + (0.12 m)^2}} (-\mathbf{z})$$

$$= -20.80 * 10^{-5} \mathbf{z} \quad [T]$$

$$\mathbf{B}_2 = -20.80 * 10^{-5} \mathbf{z} \quad [T]$$



Problem 5.7

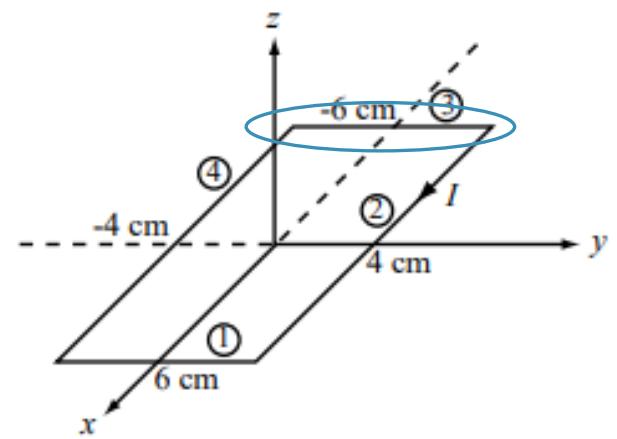
$$|\mathbf{B}| = \mu |\mathbf{H}| = \mu_0 \frac{Il}{2\pi r \sqrt{4r^2+l^2}} \quad [T]$$

Segment 3 (blue circle):

$$|\mathbf{B}_3| = \mu |\mathbf{H}| = \mu_0 \frac{Il}{2\pi r \sqrt{4r^2+l^2}} (-\mathbf{z})$$

$$= 4\pi * 10^{-7} NA^{-2} \frac{50 A * 0.08 m}{2\pi(0.06 m) \sqrt{4(0.06 m)^2 + (0.08 m)^2}} (-\mathbf{z})$$
$$= -9.24 * 10^{-5} \mathbf{z} \quad [T]$$

$$\mathbf{B}_3 = -9.24 * 10^{-5} \mathbf{z} \quad [T]$$



Problem 5.7

$$|\mathbf{B}| = \mu |\mathbf{H}| = \mu_0 \frac{Il}{2\pi r \sqrt{4r^2 + l^2}} \quad [T]$$

Segment 4 (blue circle):

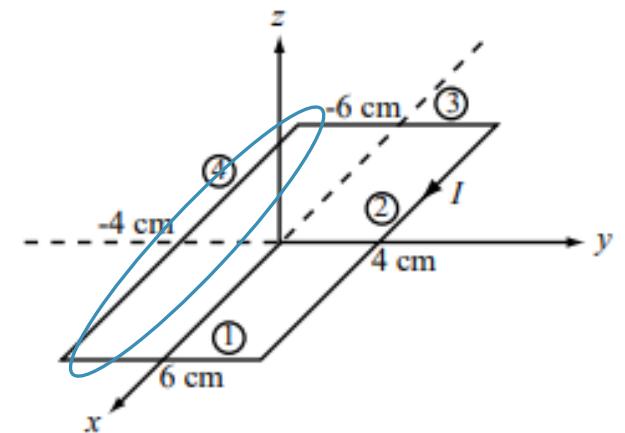
\mathbf{B}_4 is along $-z$ direction.

$$|\mathbf{B}_4| = \mu |\mathbf{H}| = \mu_0 \frac{Il}{2\pi r \sqrt{4r^2 + l^2}} (-\mathbf{z})$$

$$= 4\pi * 10^{-7} NA^{-2} \frac{50 A * 0.12 m}{2\pi(0.04 m) \sqrt{4(0.04 m)^2 + (0.12 m)^2}} (-\mathbf{z})$$

$$= -20.80 * 10^{-5} \mathbf{z} \quad [T]$$

$$\mathbf{B}_4 = -20.80 * 10^{-5} \mathbf{z} \quad [T]$$

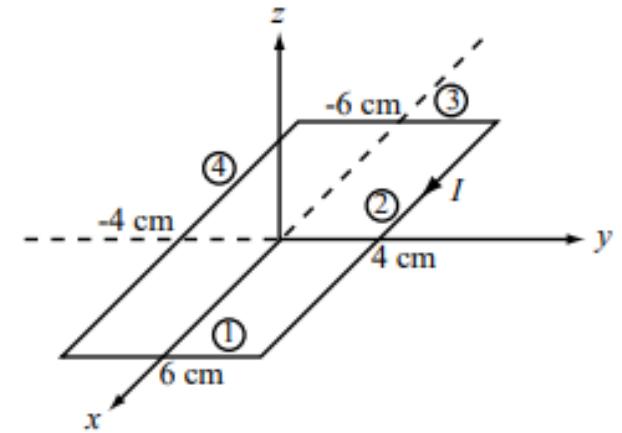


Problem 5.7

$$B = B_1 + B_2 + B_3 + B_4$$

$$= -9.24 * 10^{-5}z + -20.80 * 10^{-5}z + -9.24 * 10^{-5}z + -20.80 * 10^{-5}z$$

$$B = -0.60 * 10^{-3}z [T]$$



2A04 Tutorial 10

March 28th, 2022

Fraser McCauley & Alex Lee

1. Problem 5.24

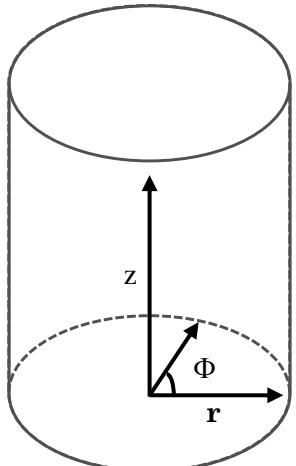
In a certain conducting region, the magnetic field is given in cylindrical coordinates by

$$\mathbf{H} = \hat{\Phi} \frac{4}{r} [1 - (1 + 3r)e^{-3r}]$$

Find the current density \mathbf{J} .

What equation relates \mathbf{J} to \mathbf{H} ?

$$\mathbf{J} = \nabla \times \mathbf{H}$$



Curl of a system
in cylindrical
coordinates:

$$\begin{aligned}\nabla \times \mathbf{A} &= \frac{1}{r} \begin{vmatrix} \hat{\mathbf{r}} & \hat{\phi}r & \hat{\mathbf{z}} \\ \frac{\partial}{\partial r} & \frac{\partial}{\partial \phi} & \frac{\partial}{\partial z} \\ A_r & rA_\phi & A_z \end{vmatrix} \\ &= \hat{\mathbf{r}} \left(\frac{1}{r} \frac{\partial A_z}{\partial \phi} - \frac{\partial A_\phi}{\partial z} \right) \\ &\quad + \hat{\phi} \left(\frac{\partial A_r}{\partial z} - \frac{\partial A_z}{\partial r} \right) \\ &\quad + \hat{\mathbf{z}} \frac{1}{r} \left[\frac{\partial}{\partial r} (rA_\phi) - \frac{\partial A_r}{\partial \phi} \right]\end{aligned}$$

H_r and H_z are zero, so the curl simplifies to:

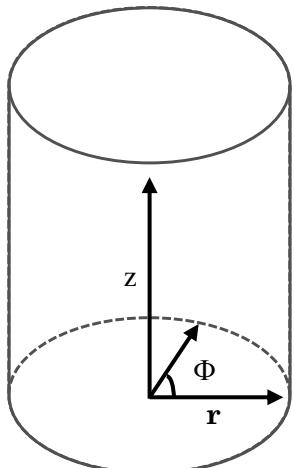
$$\begin{aligned}\mathbf{J} &= \hat{\mathbf{z}} \frac{1}{r} \left[\frac{\partial}{\partial r} (4[1 - (1 + 3r)e^{-3r}]) \right] \\ \mathbf{J} &= \hat{\mathbf{z}} \frac{1}{r} \left[\frac{\partial}{\partial r} [4 - 4(1 + 3r)e^{-3r}] \right]\end{aligned}$$

1. Problem 5.24

In a certain conducting region, the magnetic field is given in cylindrical coordinates by

$$\mathbf{H} = \hat{\Phi} \frac{4}{r} [1 - (1 + 3r)e^{-3r}]$$

Find the current density \mathbf{J} .



H_r and H_z are zero, so the curl simplifies to:

$$\mathbf{J} = \hat{\mathbf{z}} \frac{1}{r} \left[\frac{\partial}{\partial r} (r H_\phi) \right]$$

$$\mathbf{J} = \hat{\mathbf{z}} \frac{1}{r} \left[\frac{\partial}{\partial r} (4[1 - (1 + 3r)e^{-3r}]) \right]$$

$$\mathbf{J} = \hat{\mathbf{z}} \frac{1}{r} \left[\frac{\partial}{\partial r} [4 - 4(1 + 3r)e^{-3r}] \right]$$

$$\mathbf{J} = \hat{\mathbf{z}} \frac{-4}{r} \left[\frac{\partial}{\partial r} [(1 + 3r)e^{-3r}] \right]$$

$$\mathbf{J} = \hat{\mathbf{z}} \frac{-4}{r} [3e^{-3r} - 3e^{-3r}(1 + 3r)]$$

$$\mathbf{J} = \hat{\mathbf{z}} \frac{-4}{r} [3e^{-3r}(1 - (1 + 3r))] \quad J = \hat{\mathbf{z}} \frac{-4}{r} [3e^{-3r}(-3r)]$$

$$J = \hat{\mathbf{z}} 36e^{-3r} \text{ A/m}^2$$

2. Problem 5.27

In a given region of space, the vector magnetic potential is given by

$$\mathbf{A} = \hat{x}5 \cos \pi y + \hat{z}(2 + \sin \pi x) \quad (\text{Wb/m})$$

- a) Determine \mathbf{B} .
- b) Use Eq 5.66 to calculate the magnetic flux passing through a square loop with 0.25-m long edges if the loop is in the x-y plane, its center is at the origin, and its edges are parallel to the x and y axes.
- c) Calculate Φ again using Eq 5.67

2. Problem 5.27a

In a given region of space, the vector magnetic potential is given by

$$\mathbf{A} = \hat{\mathbf{x}}5 \cos \pi y + \hat{\mathbf{z}}(2 + \sin \pi x) \quad (\text{Wb/m})$$

a) Determine \mathbf{B} .

What equation relates \mathbf{A} to \mathbf{B} ?

$$\mathbf{B} = \nabla \times \mathbf{A}$$

$$\begin{aligned}\nabla \times \mathbf{A} &= \begin{vmatrix} \hat{\mathbf{x}} & \hat{\mathbf{y}} & \hat{\mathbf{z}} \\ \frac{\partial}{\partial x} & \frac{\partial}{\partial y} & \frac{\partial}{\partial z} \\ A_x & A_y & A_z \end{vmatrix} \\ &= \hat{\mathbf{x}} \left(\frac{\partial A_z}{\partial y} - \frac{\partial A_y}{\partial z} \right) \\ &\quad + \hat{\mathbf{y}} \left(\frac{\partial A_x}{\partial z} - \frac{\partial A_z}{\partial x} \right) \\ &\quad + \hat{\mathbf{z}} \left(\frac{\partial A_y}{\partial x} - \frac{\partial A_x}{\partial y} \right)\end{aligned}$$

- There is no component in y ($A_y = 0, \frac{\partial A_y}{\partial \text{anything}} = 0$)
- The x component only depends on y ($\frac{\partial A_x}{\partial z} = 0$)
- The z component only depends on x ($\frac{\partial A_z}{\partial y} = 0$)

$$\mathbf{B} = \hat{\mathbf{x}}(0 - 0) + \hat{\mathbf{y}} \left(0 - \frac{\partial A_z}{\partial x} \right) + \hat{\mathbf{z}} \left(0 - \frac{\partial A_x}{\partial y} \right)$$

$$\mathbf{B} = -\hat{\mathbf{y}} \frac{\partial A_z}{\partial x} - \hat{\mathbf{z}} \frac{\partial A_x}{\partial y} = -\hat{\mathbf{y}} \frac{\partial}{\partial x} (2 + \sin \pi x) - \hat{\mathbf{z}} \frac{\partial}{\partial y} 5 \cos \pi y$$

$$\mathbf{B} = -\hat{\mathbf{y}} \pi \cos \pi x + \hat{\mathbf{z}} 5 \pi \sin \pi y$$

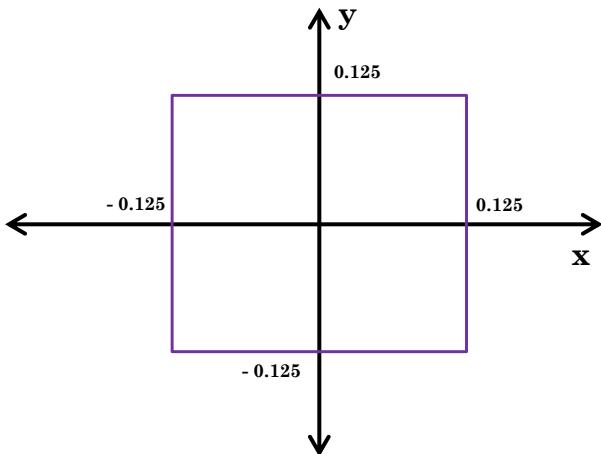
2. Problem 5.27b

In a given region of space, the vector magnetic potential is given by

$$\mathbf{A} = \hat{x}5 \cos \pi y + \hat{z}(2 + \sin \pi x) \quad (\text{Wb}/\text{m})$$

- b) Use Eq 5.66 to calculate the magnetic flux passing through a square loop with 0.25-m long edges if the loop is in the x-y plane, its center is at the origin, and its edges are parallel to the x and y axes.

$$\Phi = \int_S \mathbf{B} \cdot d\mathbf{s} \quad (\text{Wb})$$



$$\Phi = \int_{-0.125}^{0.125} \int_{-0.125}^{0.125} (-\hat{y}\pi \cos \pi x + \hat{z}5\pi \sin \pi y) \hat{z} dx dy$$

$$\Phi = \int_{-0.125}^{0.125} \int_{-0.125}^{0.125} (5\pi \sin \pi y) \hat{z} dx dy = \hat{z} \int_{-0.125}^{0.125} x 5\pi \sin \pi y \Big|_{x=-0.125}^{x=0.125} dy$$

$$\Phi = -\hat{z}x5\pi \frac{\cos \pi y}{\pi} \Big|_{x=-0.125}^{x=0.125} \Big|_{y=-0.125}^{y=0.125} = -\hat{z}5(0.25) \left(\cos \frac{\pi}{8} - \cos \frac{-\pi}{8} \right)$$

$$\Phi = 0$$

2. Problem 5.27c

In a given region of space, the vector magnetic potential is given by

$$\mathbf{A} = \hat{x}5 \cos \pi y + \hat{z}(2 + \sin \pi x) \quad (\text{Wb/m})$$

c) Calculate Φ again using Eq 5.67

$$\Phi = \oint_C \mathbf{A} \cdot d\mathbf{l}$$

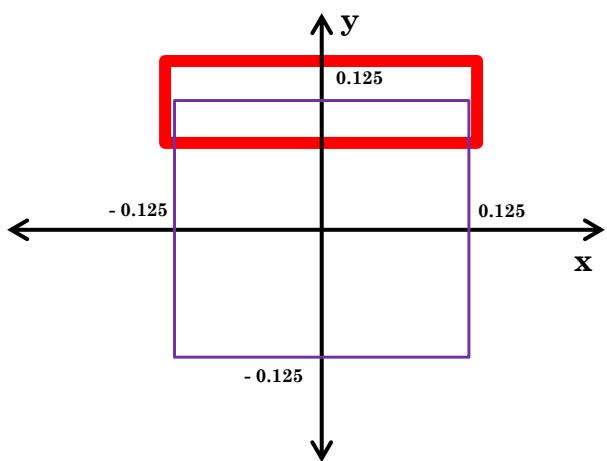
$$\Phi = S_{top} + S_{right} + S_{bottom} + S_{left}$$

Along y=0.125:

$$S_{top} = \int_{-0.125}^{0.125} \hat{x}5 \cos \pi y + \hat{z}(2 + \sin \pi x) \cdot \hat{x} dx$$

$$S_{top} = \int_{-0.125}^{0.125} 5 \cos(0.125\pi) dx = 4.6194x \Big|_{x=-0.125}^{x=0.125}$$

$$S_{top} = 1.15485$$



2. Problem 5.27c

In a given region of space, the vector magnetic potential is given by

$$\mathbf{A} = \hat{x}5 \cos \pi y + \hat{z}(2 + \sin \pi x) \quad (\text{Wb/m})$$

c) Calculate Φ again using Eq 5.67

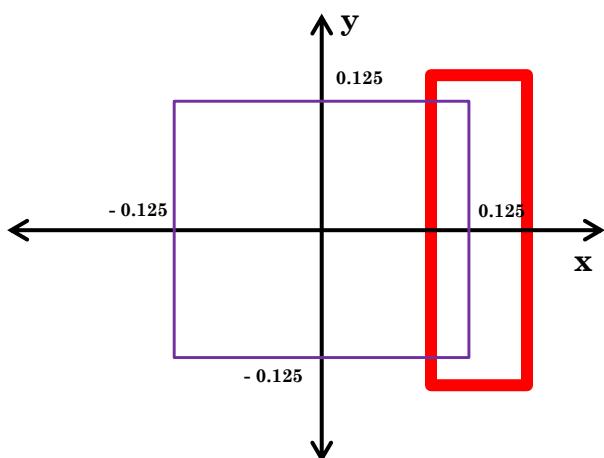
$$\Phi = \oint_C \mathbf{A} \cdot d\mathbf{l}$$

$$\Phi = S_{top} + S_{right} + S_{bottom} + S_{left}$$

Along x=0.125:

$$S_{right} = \int_{0.125}^{-0.125} \hat{x}5 \cos \pi y + \hat{z}(2 + \sin \pi x) \cdot \hat{y} dy$$

$$S_{right} = \int_{0.125}^{-0.125} 0 dx = 0$$



2. Problem 5.27c

In a given region of space, the vector magnetic potential is given by

$$\mathbf{A} = \hat{x}5 \cos \pi y + \hat{z}(2 + \sin \pi x) \quad (\text{Wb/m})$$

c) Calculate Φ again using Eq 5.67

$$\Phi = \oint_C \mathbf{A} \cdot d\mathbf{l}$$

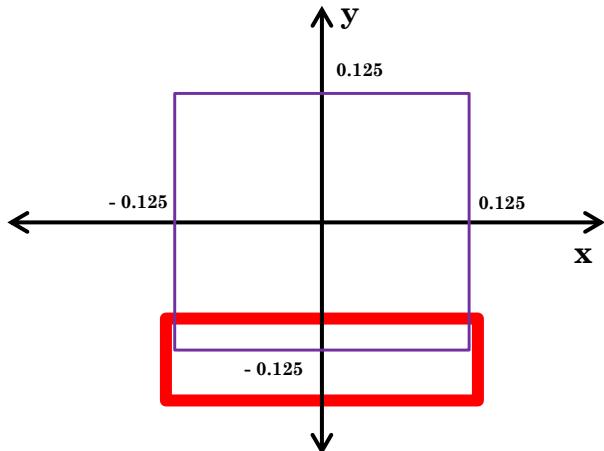
$$\Phi = S_{top} + S_{right} + S_{bottom} + S_{left}$$

Along y=-0.125:

$$S_{bottom} = \int_{0.125}^{-0.125} \hat{x}5 \cos \pi y + \hat{z}(2 + \sin \pi x) \cdot \hat{x} dx$$

$$S_{bottom} = \int_{0.125}^{-0.125} 5 \cos(0.125\pi) dx = 4.6194x \Big|_{x=0.125}^{x=-0.125}$$

$$S_{bottom} = -1.15485$$



2. Problem 5.27c

In a given region of space, the vector magnetic potential is given by

$$\mathbf{A} = \hat{x}5 \cos \pi y + \hat{z}(2 + \sin \pi x) \quad (\text{Wb/m})$$

c) Calculate Φ again using Eq 5.67

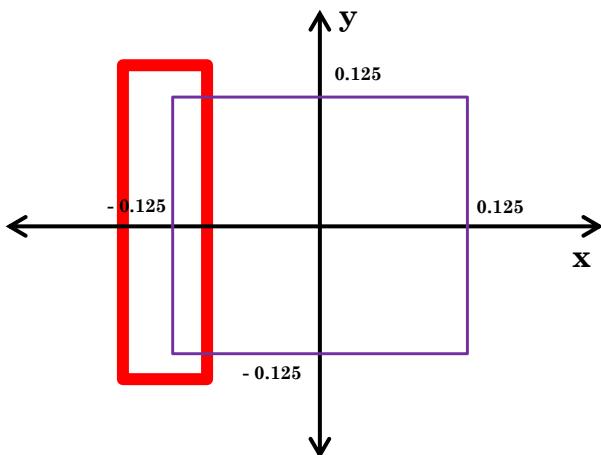
$$\Phi = \oint_C \mathbf{A} \cdot d\mathbf{l}$$

$$\Phi = S_{top} + S_{right} + S_{bottom} + S_{left}$$

Along x=-0.125:

$$S_{left} = \int_{-0.125}^{0.125} \hat{x}5 \cos \pi y + \hat{z}(2 + \sin \pi x) \cdot \hat{y} dy$$

$$S_{left} = \int_{0.125}^{-0.125} 0 dx = 0$$



$$\Phi = 1.15485 + 0 - 1.15485 + 0$$

$$\Phi = 0$$

3. Problem 5.31

Iron contains $8.5 \times 10^{28} \text{ atoms/m}^3$. At saturation, the alignment of the electrons' spin magnetic moments in iron can contribute 1.5T to the total magnetic flux density **B**. If the spin magnetic moment of a single electron is $9.27 \times 10^{-24} (\text{A} \cdot \text{m}^2)$, how many electrons per atom contribute to the saturated field?

The number of electrons contributing to the bulk magnetization *per atom* is n_e , the total number of contributing electrons is N_e , and the number of atoms is N_{atoms} .

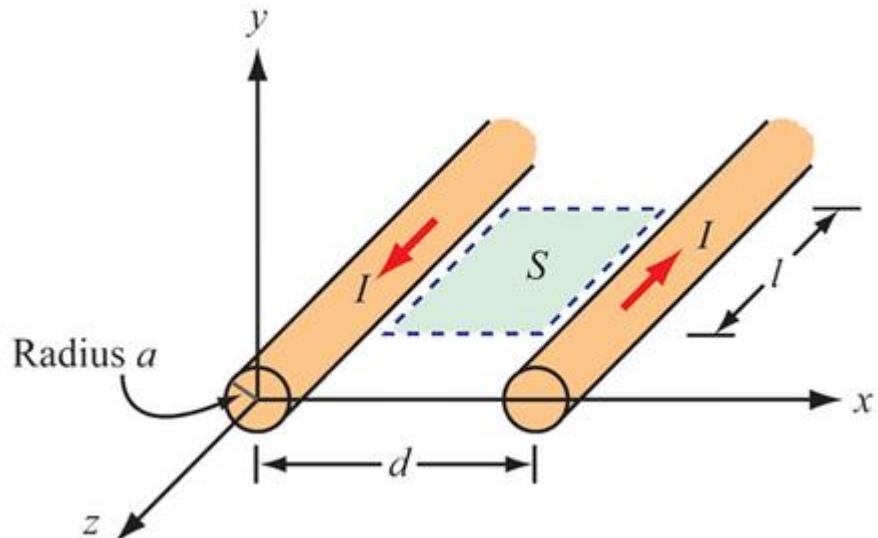
$$\text{So, } n_e = \frac{N_e}{N_{\text{atoms}}}$$

$$\text{From } B_m = \mu_0 M = \mu_0 N_e m_s \rightarrow N_e = \frac{B_m}{\mu_0 m_s}$$

$$n_e = \frac{N_e}{N_{\text{atoms}}} = \frac{B_m}{\mu_0 m_s N_{\text{atoms}}} = \frac{1.5}{\mu_0 (9.27 \times 10^{-24})(8.5 \times 10^{28})} = 1.5 \text{ electrons/atom}$$

4. Problem 5.37

Obtain an expression for the self-inductance per unit length for the parallel wire transmission line of Fig. 5-28a in terms of a , d , and μ , where a is the radius of the wires, d is the axis-to-axis distance between the wires, and μ is the permeability of the medium in which they reside.



Notes:

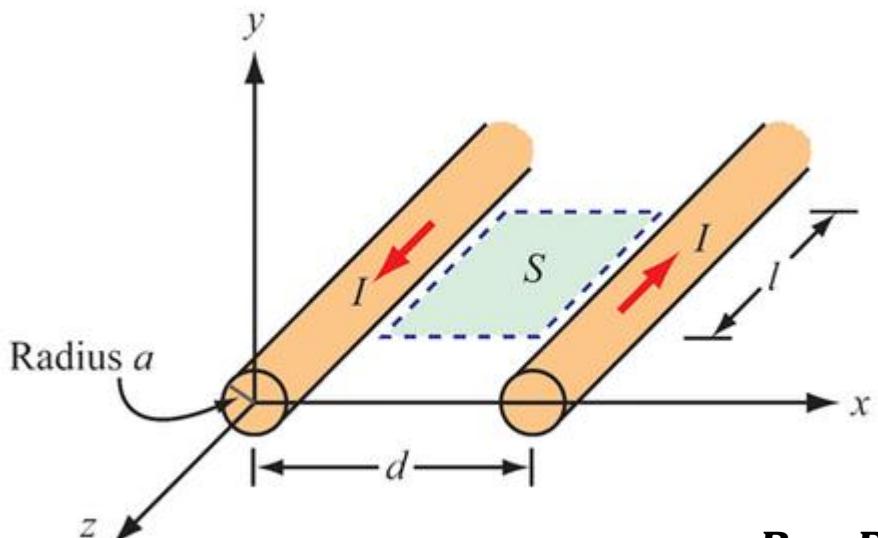
- Self-inductance is based on magnetic flux
- Magnetic flux is based on magnetic field strength
- ASSUMPTION: radii are much smaller than the separation distance

Plan:

- Calculate the magnetic field caused by each wire
- Calculate the magnetic flux
- Calculate the inductance

4. Problem 5.37 – step 1

Obtain an expression for the self-inductance per unit length for the parallel wire transmission line of Fig. 5-28a in terms of a , d , and μ , where a is the radius of the wires, d is the axis-to-axis distance between the wires, and μ is the permeability of the medium in which they reside.



Calculate the magnetic field caused by each wire

Currents are both I , the magnetic field at a position x for long wires is:

$$\mathbf{B}_1 = \hat{\mathbf{y}} \frac{\mu I}{2\pi x} \quad \mathbf{B}_2 = \hat{\mathbf{y}} \frac{\mu I}{2\pi(d-x)}$$

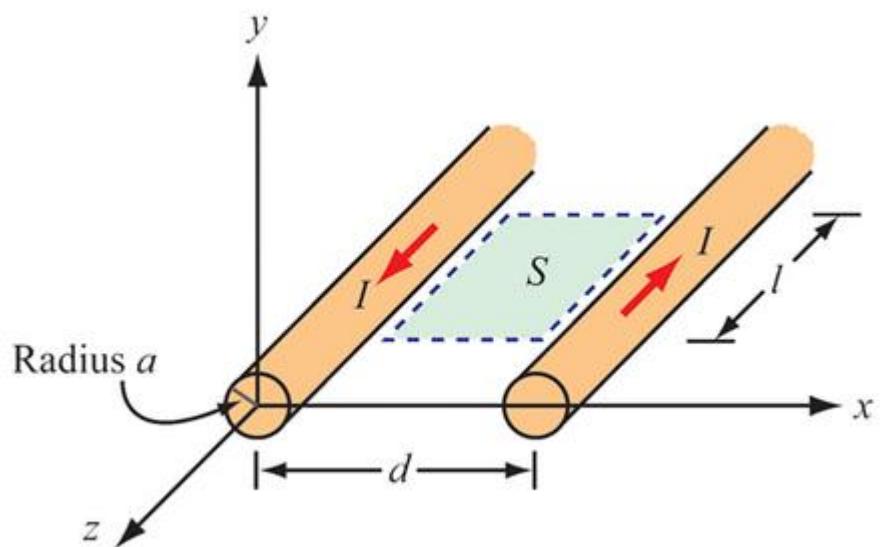
They're both positive, because the currents are in opposite directions.

$$\mathbf{B} = \mathbf{B}_1 + \mathbf{B}_2 = \hat{\mathbf{y}} \frac{\mu I}{2\pi} \left(\frac{1}{x} + \frac{1}{d-x} \right) = \hat{\mathbf{y}} \frac{\mu I}{2\pi} \left(\frac{d-x}{x(d-x)} + \frac{x}{x(d-x)} \right)$$

$$\mathbf{B} = \hat{\mathbf{y}} \frac{\mu I d}{2\pi x(d-x)}$$

4. Problem 5.37 – step 2

Obtain an expression for the self-inductance per unit length for the parallel wire transmission line of Fig. 5-28a in terms of a , d , and μ , where a is the radius of the wires, d is the axis-to-axis distance between the wires, and μ is the permeability of the medium in which they reside.



Calculate the magnetic flux

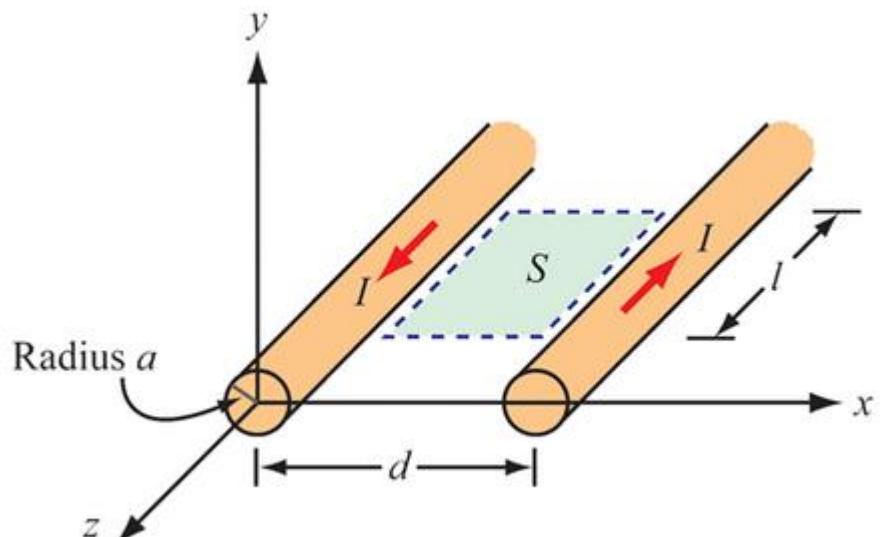
$$\Phi = \int \int \mathbf{B} \cdot d\mathbf{s}$$

$$\Phi = \int_{z=0}^l \int_{x=a}^{d-a} \hat{\mathbf{y}} \frac{\mu Id}{2\pi x(d-x)} \cdot \hat{\mathbf{y}} dx dz$$

$$\Phi = \frac{\mu Id}{2\pi} \int_{z=0}^l \int_{x=a}^{d-a} \frac{1}{x(d-x)} dx dz$$

4. Problem 5.37 – step 2

Obtain an expression for the self-inductance per unit length for the parallel wire transmission line of Fig. 5-28a in terms of a , d , and μ , where a is the radius of the wires, d is the axis-to-axis distance between the wires, and μ is the permeability of the medium in which they reside.



Calculate the magnetic flux

$$\Phi = \frac{-\mu I d}{2\pi} \int_{z=0}^l \int_{x=a}^{d-a} \frac{1}{x(x-d)} dx dz$$

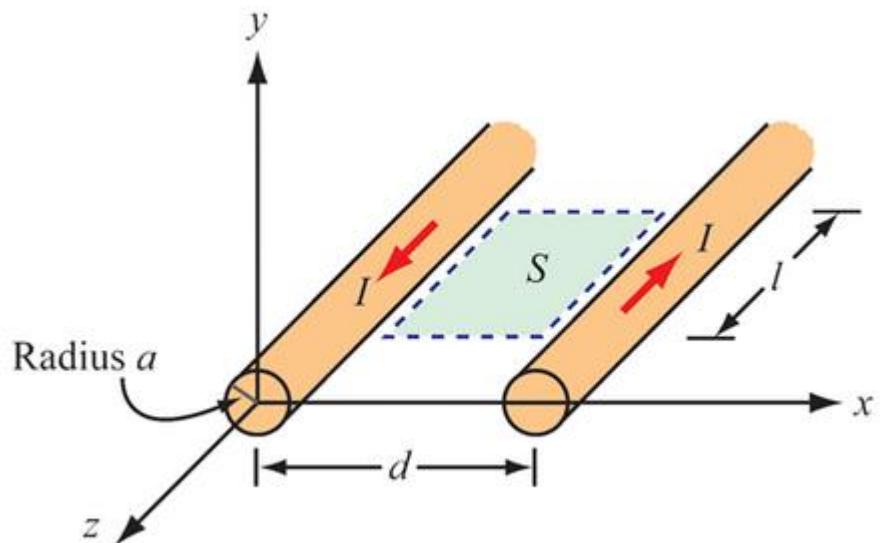
Use common integral:

$$\int \frac{1}{(x+a)(x+b)} dx = \frac{1}{b-a} \ln \frac{a+x}{b+x}, \quad a \neq b$$

$$\Phi = \frac{-\mu I d}{2\pi} \int_{z=0}^l \left(\frac{1}{-d-0} \ln \frac{x}{x-d} \Big|_{x=a}^{d-a} \right) dz = \frac{\mu I l}{2\pi} \left(\ln \frac{d-a}{-a} - \ln \frac{a}{a-d} \right)$$

4. Problem 5.37 – step 2

Obtain an expression for the self-inductance per unit length for the parallel wire transmission line of Fig. 5-28a in terms of a , d , and μ , where a is the radius of the wires, d is the axis-to-axis distance between the wires, and μ is the permeability of the medium in which they reside.



Calculate the magnetic flux

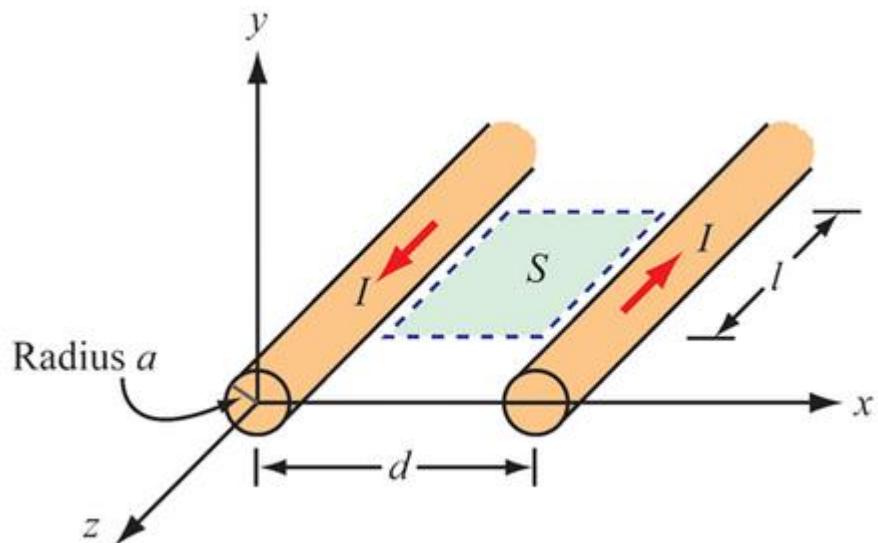
$$\Phi = \frac{\mu Il}{2\pi} \left(\ln \frac{d-a}{-a} - \ln \frac{a}{a-d} \right)$$

$$\Phi = \frac{\mu Il}{2\pi} \left(\ln \frac{(d-a)(a-d)}{-a^2} \right) = \frac{\mu Il}{2\pi} \left(\ln \frac{(d-a)^2}{a^2} \right)$$

$$\Phi = \frac{\mu Il}{\pi} \ln \frac{d-a}{a}$$

4. Problem 5.37

Obtain an expression for the self-inductance per unit length for the parallel wire transmission line of Fig. 5-28a in terms of a , d , and μ , where a is the radius of the wires, d is the axis-to-axis distance between the wires, and μ is the permeability of the medium in which they reside.



Calculate the inductance

There is only one 'loop', so inductance is:

$$L = \frac{\Phi}{I} = \frac{\mu I l}{\pi} \ln \frac{d-a}{a} / I = \frac{\mu l}{\pi} \ln \frac{d-a}{a}$$

Inductance per unit length:

$$L' = \frac{L}{l} = \frac{\mu l}{\pi} \ln \frac{d-a}{a}$$

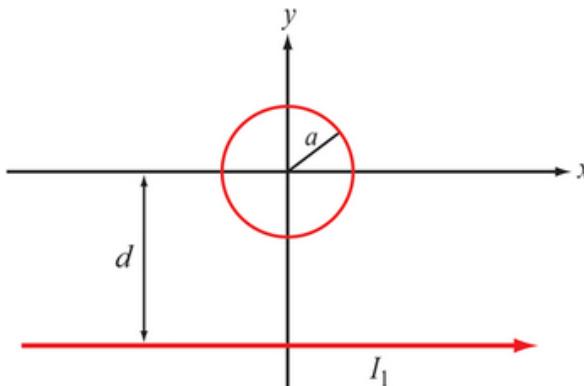
Assuming a is very small relative to d :

$$L' \approx \frac{\mu l}{\pi} \ln \frac{d}{a}$$

5. Problem 5.40

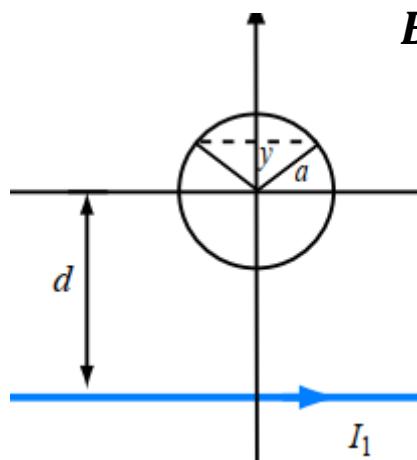
Determine the mutual inductance between the circular loop and the linear current shown in Fig P5.40.

Figure P5.40 Linear conductor with current I_1 next to a circular loop of radius a at distance d (Problem 40).



Find the magnetic field induced by the current on the linear conductor, for any line of constant y :

$$\mathbf{B} = \hat{\mathbf{z}} \frac{\mu_0 I}{2\pi(d + y)}$$



Find the flux through each infinitesimal strip of constant y inside the loop.

$$d\Phi = \mathbf{B}(y) \cdot d\mathbf{s} = \hat{\mathbf{z}} \frac{\mu_0 I}{2\pi(d + y)} \cdot \hat{\mathbf{z}} 2\sqrt{(a^2 - y^2)} dy$$

5. Problem 5.40

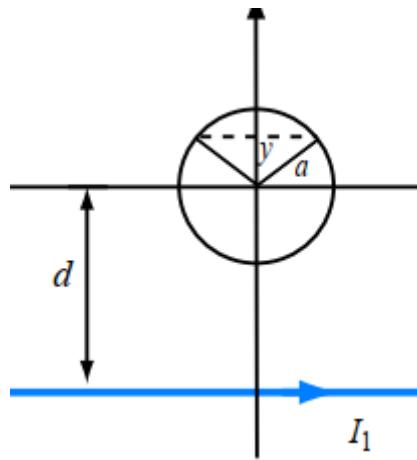
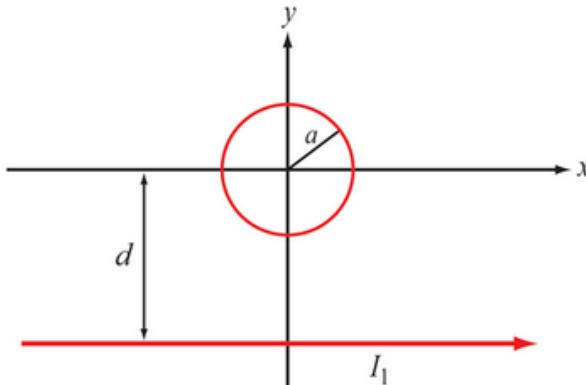
Determine the mutual inductance between the circular loop and the linear current shown in Fig P5.40.

$$d\Phi = \mathbf{B}(y) \cdot d\mathbf{s} = \hat{\mathbf{z}} \frac{\mu_0 I}{2\pi(d+y)} \cdot \hat{\mathbf{z}} 2\sqrt{(a^2 - y^2)} dy$$

Figure P5.40 Linear conductor

with current I_1 next to a circular loop of radius a at distance d

(Problem 40 ).



$$d\Phi = \frac{\mu_0 I \sqrt{(a^2 - y^2)}}{\pi(d+y)} dy$$

$$L = \frac{\Phi}{I} = \frac{1}{I} \int_S d\Phi = \frac{1}{I} \int_{y=-a}^a \frac{\mu_0 I \sqrt{(a^2 - y^2)}}{\pi(d+y)} dy$$

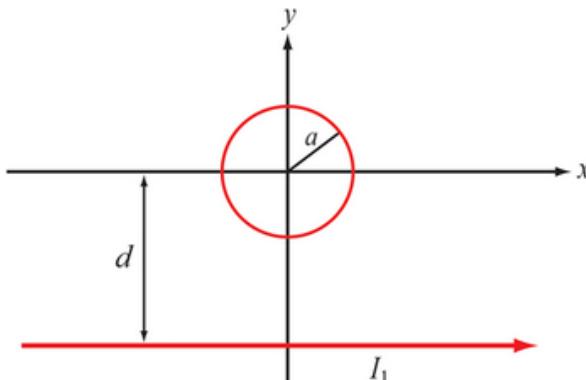
$$L = \frac{\mu_0}{\pi} \int_{y=-a}^a \frac{\sqrt{(a^2 - y^2)}}{(d+y)} dy$$

5. Problem 5.40

Determine the mutual inductance between the circular loop and the linear current shown in Fig P5.40.

Solving the following integral:

Figure P5.40 Linear conductor with current I_1 next to a circular loop of radius a at distance d (Problem 40).



$$L = \frac{\mu_0}{\pi} \int_{y=-a}^a \frac{\sqrt{(a^2 - y^2)}}{(d + y)} dy$$

Let $z = d + y$: $L = \frac{\mu_0}{\pi} \int_{z=d-a}^{d+a} \frac{\sqrt{(a^2 - (z - d)^2)}}{z} dz$

$$L = \frac{\mu_0}{\pi} \int_{z=d-a}^{d+a} \frac{\sqrt{(a^2 - d^2) + 2dz - z^2}}{z} dz$$

Let $R = a_0 + b_0z + c_0z^2$ be the polynomial $(a^2 - d^2) + 2dz - z^2$:

From a table of integrals: $\int \frac{\sqrt{R}}{z} dz = \sqrt{R} + a_0 \int \frac{dz}{z\sqrt{R}} + \frac{b_0}{z} \int \frac{dz}{\sqrt{R}}$

5. Problem 5.40

Determine the mutual inductance between the circular loop and the linear current shown in Fig P5.40.

So now we're solving *these* integrals:

Let $R = a_0 + b_0z + c_0z^2$ be the polynomial $(a^2 - d^2) + 2dz - z^2$:

$$\int \frac{\sqrt{R}}{z} dz = \sqrt{R} + a_0 \int \frac{dz}{z\sqrt{R}} + \frac{b_0}{z} \int \frac{dz}{\sqrt{R}}$$

Evaluating the first term

over $d - a$ to $d + a$:

$$\sqrt{R} \Big|_{d-a}^{d+a} = \sqrt{(a^2 - d^2) + 2d(d + a) - (d + a)^2} - \sqrt{(a^2 - d^2) + 2d(d - a) - (d - a)^2} = 0$$

So our integral is now $a_0 \int \frac{dz}{z\sqrt{R}} + \frac{b_0}{z} \int \frac{dz}{\sqrt{R}}$

5. Problem 5.40

Determine the mutual inductance between the circular loop and the linear current shown in Fig P5.40.

$$a_0 \int \frac{dz}{z\sqrt{R}} + \frac{b_0}{z} \int \frac{dz}{\sqrt{R}}$$

From a table of integrals:

$$\begin{aligned} a_0 \int \frac{dz}{z\sqrt{R}} &= a_0 \left[\frac{1}{\sqrt{-a_0}} \sin^{-1} \left(\frac{2a_0 + b_0 z}{z \sqrt{b_0^2 - 4a_0 c_0}} \right) \right]_{z=d-a}^{d+a} \\ &= -\sqrt{d^2 - a^2} \left[\sin^{-1} \left(\frac{a^2 - d^2 + dz}{az} \right) \right]_{z=d-a}^{d+a} \\ &= -\pi \sqrt{d^2 - a^2} \end{aligned}$$

5. Problem 5.40

Determine the mutual inductance between the circular loop and the linear current shown in Fig P5.40.

$$a_0 \int \frac{dz}{z\sqrt{R}} + \frac{b_0}{z} \int \frac{dz}{\sqrt{R}}$$

From a table of integrals:

$$\begin{aligned}\frac{b_0}{z} \int \frac{dz}{\sqrt{R}} &= \frac{b_0}{2} \left[\frac{-1}{\sqrt{-c_0}} \sin^{-1} \frac{2c_0 z + b_0}{\sqrt{-\Delta}} \right]_{z=d-a}^{d+a} \\ &= -d \left[\sin^{-1} \left(\frac{d-z}{a} \right) \right]_{z=d-a}^{d+a} = \pi d\end{aligned}$$

5. Problem 5.40

Determine the mutual inductance between the circular loop and the linear current shown in Fig P5.40.

Last step!

$$L = \frac{\mu_0}{\pi} \int_{y=-a}^a \frac{\sqrt{(a^2 - y^2)}}{(d + y)} dy$$

$$L = \frac{\mu_0}{\pi} (\pi d - \pi \sqrt{d^2 - a^2})$$

$$L = \mu_0 (d - \sqrt{d^2 - a^2})$$

Reminders

- Assignment 10 is out, and is due at 8AM on April 4th.
- Good luck!

ENG PHYS 2A04 Tutorial 11

Electricity and Magnetism

Your TAs today

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Chapter 6

Problems

Problem 6.3

Problem 6.4

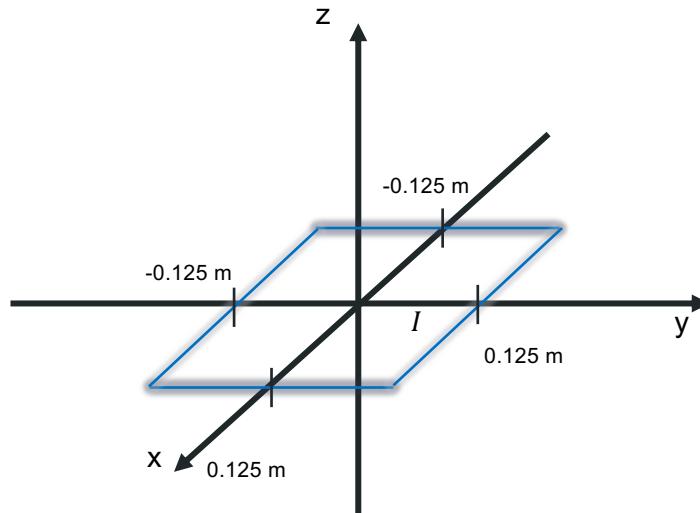
Problem 6.5

Problem 6.7

Problem 6.3 – Question

A coil consists of 100 turns of wire wrapped around a square frame of sides 0.25 m. The coil is centered at the origin with each of its sides parallel to the x- or y-axis. Find the induced emf across the open-circuited ends of the coil if the magnetic field is given by

- (a) $\mathbf{B} = \hat{\mathbf{z}} 20e^{-3t}$ (T)
- (b) $\mathbf{B} = \hat{\mathbf{z}} 20 \cos x \cos 10^3 t$ (T)
- (c) $\mathbf{B} = \hat{\mathbf{z}} 20 \cos x \sin 2y \cos 10^3 t$ (T)



Problem 6.3 – Details

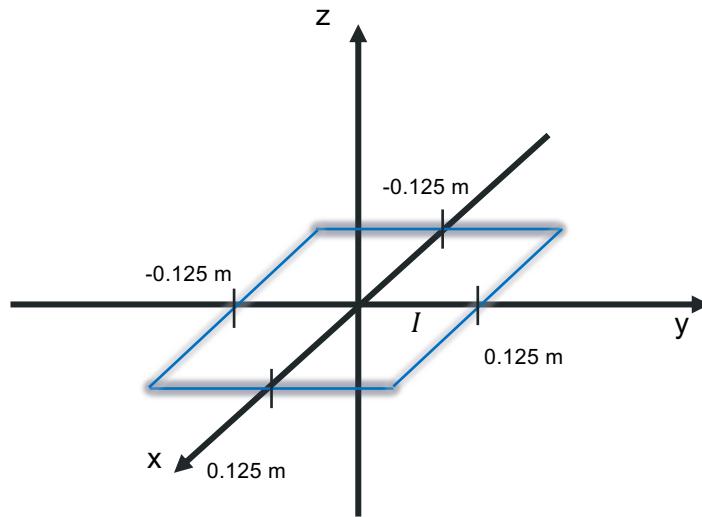
A coil consists of 100 turns of wire wrapped around a square frame of sides 0.25 m. The coil is centered at the origin with each of its sides parallel to the x- or y-axis. Find the induced emf across the open-circuited ends of the coil if the magnetic field is given by

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(c) $\mathbf{B} = \hat{\mathbf{z}} 20 \cos x \sin 2y \cos 10^3 t$ (T)

Solution? → Apply Faraday's Law!

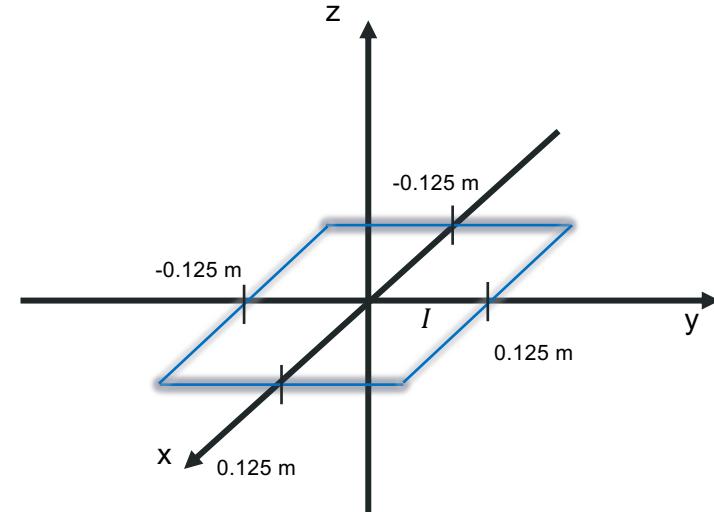


Problem 6.3 – Solution (a)

Faraday's Law states: $V_{emf} = -N \frac{d}{dt} \int_S \mathbf{B} \cdot d\mathbf{s}$

$N = 100$ turns, $d\mathbf{s} = \hat{\mathbf{z}} dx dy$

$$\mathbf{B} = \hat{\mathbf{z}} 20e^{-3t} \text{ (T)}$$



$$\rightarrow V_{emf} = -N \frac{d}{dt} \int_S \mathbf{B} \cdot d\mathbf{s} = -N \frac{d}{dt} \left[\int_{-0.125}^{0.125} \int_{-0.125}^{0.125} \hat{\mathbf{z}} 20e^{-3t} \cdot \hat{\mathbf{z}} dx dy \right]$$

$$= -100 \frac{d}{dt} [(20e^{-3t})(0.25)^2] = -100(0.0625)(20) \frac{d}{dt} [e^{-3t}] = -125(-3)e^{-3t}$$

$$\therefore V_{emf} = 375e^{-3t} \text{ (V)}$$

Problem 6.3 – Solution (b)

$$V_{emf} = -N \frac{d}{dt} \int_S \mathbf{B} \cdot d\mathbf{s}, N = 100 \text{ turns}, d\mathbf{s} = \hat{\mathbf{z}} dx dy$$

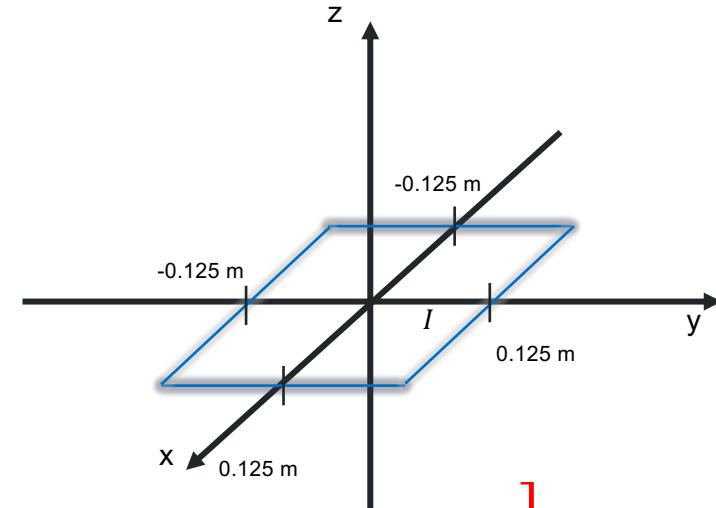
$$\mathbf{B} = \hat{\mathbf{z}} 20 \cos x \cos 10^3 t \text{ (T)}$$

$$\rightarrow V_{emf} = -N \frac{d}{dt} \int_S \mathbf{B} \cdot d\mathbf{s} = -N \frac{d}{dt} \left[\int_{-0.125}^{0.125} \int_{-0.125}^{0.125} \hat{\mathbf{z}} 20 \cos x \cos 10^3 t \cdot \hat{\mathbf{z}} dx dy \right]$$

$$= -100 \frac{d}{dt} [20 \cos 10^3 t (\sin 0.125 - \sin -0.125)(0.25)]$$

$$= -100(20)(0.25)(0.125 - (-0.125)) \frac{d}{dt} [\cos 10^3 t] = -125(-1000 \sin 10^3 t)$$

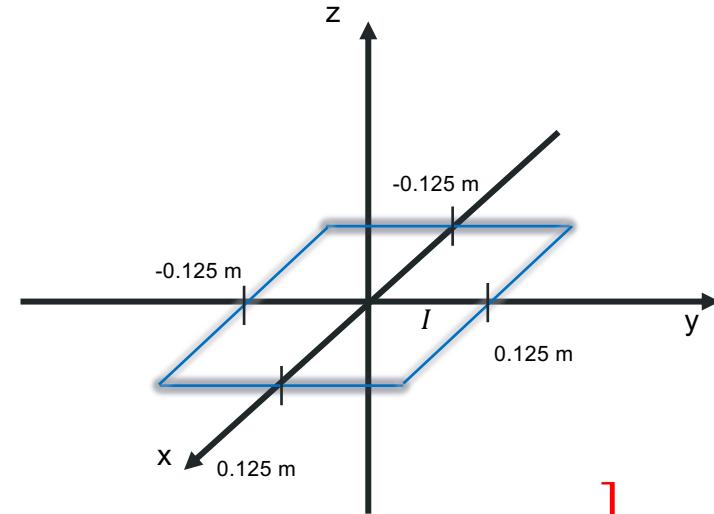
$$\therefore V_{emf} = 125 \sin 10^3 t \text{ (kV)}$$



Problem 6.3 – Solution (c)

$$V_{emf} = -N \frac{d}{dt} \int_S \mathbf{B} \cdot d\mathbf{s}, N = 100 \text{ turns}, d\mathbf{s} = \hat{\mathbf{z}} dx dy$$

$$\mathbf{B} = \hat{\mathbf{z}} 20 \cos x \sin 2y \cos 10^3 t \text{ (T)}$$



$$\begin{aligned} \rightarrow V_{emf} &= -N \frac{d}{dt} \int_S \mathbf{B} \cdot d\mathbf{s} = -N \frac{d}{dt} \left[\int_{-0.125}^{0.125} \int_{-0.125}^{0.125} \hat{\mathbf{z}} 20 \cos x \sin 2y \cos 10^3 t \cdot \hat{\mathbf{z}} dx dy \right] \\ &= -100 \frac{d}{dt} \left[20 \cos 10^3 t \left(\frac{-\cos 2y + \cos(-2y)}{2} \right) (\sin 0.125 - \sin -0.125) \right] \end{aligned}$$

$$\rightarrow \therefore V_{emf} = 0 \text{ (V)}$$

Problem 6.4 – Question

A stationary conducting loop with internal resistance of 0.5Ω is placed in a time-varying magnetic field. When the loop is closed, a current of 5 A flows through it. What will the current be if the loop is opened to create a small gap and a 2Ω resistor is connected across its open ends?

Problem 6.4 – Details

A stationary conducting loop with internal resistance of 0.5Ω is placed in a time-varying magnetic field. When the loop is closed, a current of 5 A flows through it. What will the current be if the loop is opened to create a small gap and a 2Ω resistor is connected across its open ends?

Goal → find current in closed loop

Let:

- V_{emf} = induced emf
- $I = 5 \text{ A}$ = current of stationary conducting loop
- $R = 0.5 \Omega$ = internal resistance of stationary conducting loop
- $I' = ?$ = current of new loop with 2Ω resistor inserted
- $R' = 2 \Omega$ = equivalent resistance

Problem 6.4 – Solution

A stationary conducting loop with internal resistance of 0.5Ω is placed in a time-varying magnetic field. When the loop is closed, a current of 5 A flows through it. What will the current be if the loop is opened to create a small gap and a 2Ω resistor is connected across its open ends?

Does V_{emf} change with the resistance? → NO 

$$V_{emf} = I \cdot R = I' \cdot R'$$
$$\rightarrow I' = \frac{I \cdot R}{R'} = \frac{5 \text{ A} \cdot 0.5 \Omega}{2 \Omega + 0.5 \Omega} \rightarrow \therefore I' = 1 \text{ A}$$

Let:

- V_{emf} = induced emf
- $I = 5 \text{ A}$ = current of stationary conducting loop
- $R = 0.5 \Omega$ = internal resistance of stationary conducting loop
- $I' = ?$ = current of new loop with 2Ω resistor inserted
- $R' = 2 \Omega$ = equivalent resistance

Problem 6.5

A circular-loop TV antenna with 0.02 m^2 area is in the presence of a uniform-amplitude 300-MHz signal. When oriented for maximum response, the loop develops an emf with a peak value of 30 (mV). What is the peak magnitude of B of the incident wave?

Problem 6.5

Solution:

TV loop antennas have **one turn**.

At maximum orientation with the loop area of A and Uniform magnetic field of $B=|B|$,

$$F=300 \text{ MHz}, \omega = 2\pi \times 300 \times 10^6 = 6\pi \times 10^8 \text{ rad/s}$$

$$\Phi = \int B \cdot ds = \pm BA$$

$$V_{emf} = -N \frac{d\Phi}{dt} = -A \frac{d}{dt} [B_0 \cos(\omega t + \alpha_0)] = AB_0 \omega \sin(\omega t + \alpha_0)$$

$$-1 \leq \sin(\omega t + \alpha_0) \leq 1$$

$$V_{emf} = AB_0 \omega$$

$$B_0 = \frac{V_{emf}}{A\omega} = \frac{0.03V}{0.02m^2(6\pi \times 10^8 \text{ rad/s})}$$

$$B_0 = 0.8nA/m$$

Problem 6.7

The rectangular conducting loop shown in Fig. P6.7 rotates at 6,000 revolutions per minute in a uniform magnetic flux density given by

$$\mathbf{B} = \hat{\mathbf{y}}50 \text{ (mT)}.$$

Determine the current induced in the loop if its internal resistance is 0.5Ω .

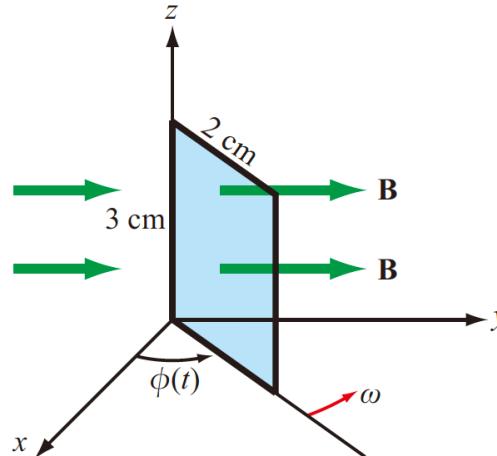


Figure P6.7: Rotating loop in a magnetic field (Problem 6.7).

Problem 6.7

The rectangular conducting loop shown in Fig. P6.7 rotates at 6,000 revolutions per minute in a uniform magnetic flux density given by

$$\mathbf{B} = \hat{\mathbf{y}}50 \text{ (mT)}.$$

Determine the current induced in the loop if its internal resistance is 0.5 Ω .

Analysis:

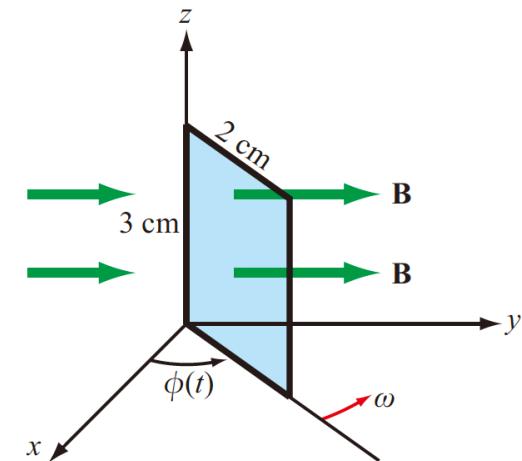
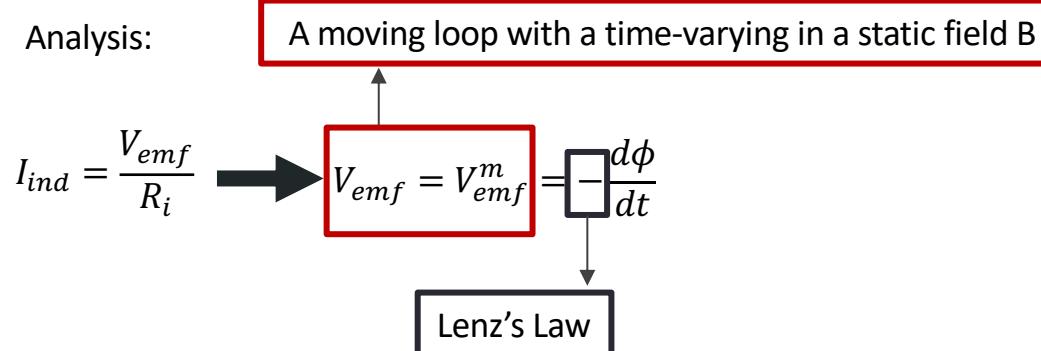


Figure P6.7: Rotating loop in a magnetic field (Problem 6.7).

Solution:

$$\Phi = \int_S \mathbf{B} \cdot d\mathbf{S} = \hat{\mathbf{y}} 50 \times 10^{-3} \cdot \hat{\mathbf{y}} (2 \times 3 \times 10^{-4}) \cos \phi(t) = 3 \times 10^{-5} \cos \phi(t),$$

$$\phi(t) = \omega t = \frac{2\pi \times 6 \times 10^3}{60} t = 200\pi t \quad (\text{rad/s}),$$

$$\Phi = 3 \times 10^{-5} \cos(200\pi t) \quad (\text{Wb}),$$

$$V_{\text{emf}} = -\frac{d\Phi}{dt} = 3 \times 10^{-5} \times 200\pi \sin(200\pi t) = 18.85 \times 10^{-3} \sin(200\pi t) \quad (\text{V}),$$

$$I_{\text{ind}} = \frac{V_{\text{emf}}}{0.5} = 37.7 \sin(200\pi t) \quad (\text{mA}).$$

Don't forget the direction of current, because current is a vector.

The direction of the current is CW (if looking at it along $-\hat{\mathbf{x}}$ -direction) when the loop is in the first quadrant ($0 \leq \phi \leq \pi/2$). The current reverses direction in the second quadrant, and reverses again every quadrant.