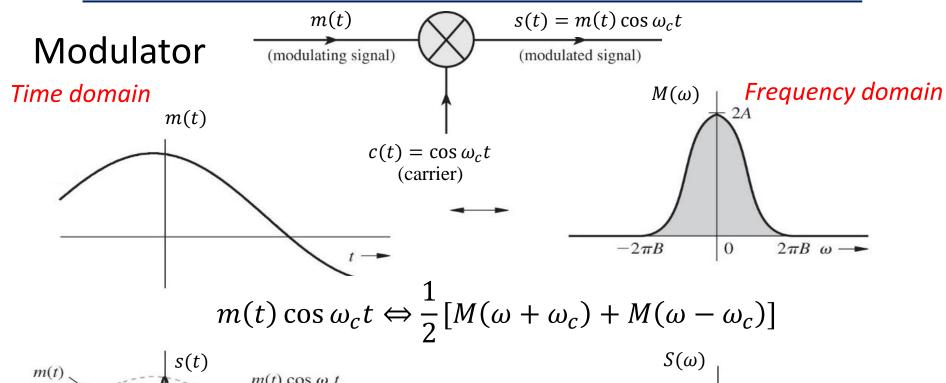
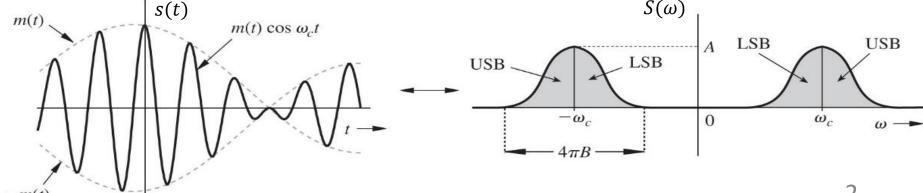
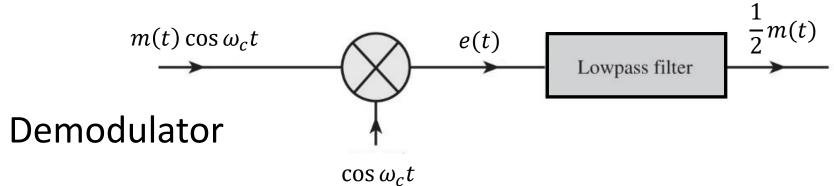
Lecture 3

Amplitude Modulation

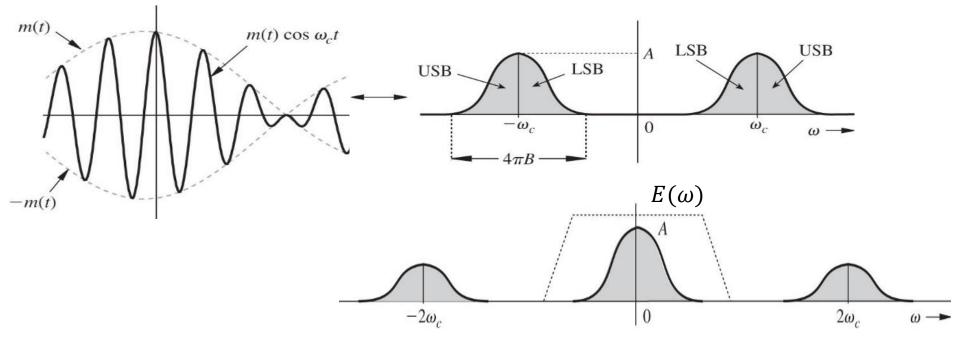
Review: 1. Double sideband suppressed carrier (DSB-SC)







Local carrier



$$e(t) = m(t)\cos^2 \omega_c t = \frac{1}{2}[m(t) + m(t)\cos 2\omega_c t]$$
 Eliminated by the LPF
$$E(\omega) = \frac{1}{2}M(\omega) + \frac{1}{4}[M(\omega + 2\omega_c) + M(\omega - 2\omega_c)]$$
 3

Notes on DSB-SC

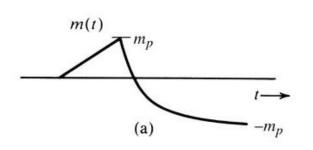
- Simple modulation
- Complex and expensive demodulation (requires synchronization)
- Waste of bandwidth (2B Hz)

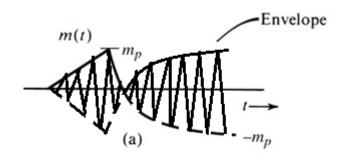
Lecture outline

 DSB-LC: Another type of AM with less complex demodulation (not requiring generation of a local carrier)

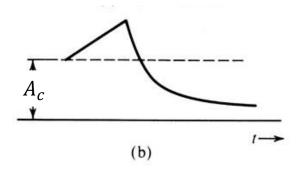
2. Double sideband large carrier (DSB-LC) Conventional AM (or simply AM)

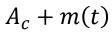
 $\pm m_p$: maximum and minimum values of m(t)

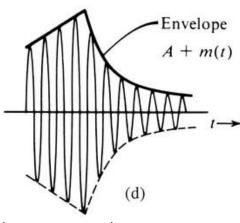




 $m(t)\cos\omega_c t$







$$(A_c + m(t)) \cos \omega_c t$$

2. Double sideband large carrier (DSB-LC) Conventional AM (or simply AM)

- Target: Simpler and cheaper demodulators (without generation of a local carrier)
- Unmodulated carrier:

$$c(t) = A_c \cos \omega_c t$$
, A_c : carrier amplitude f_c : carrier frequency ($\omega_c = 2\pi f_c$)

Modulated signal:

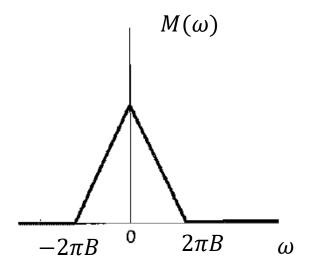
$$s(t) = (A_c + m(t)) \cos \omega_c t$$

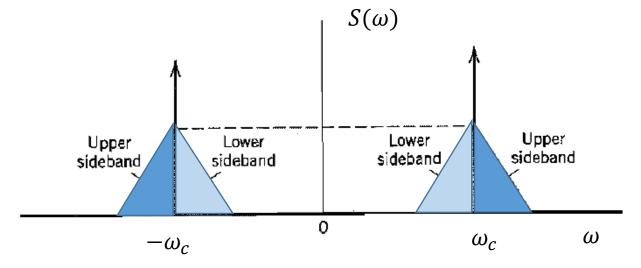
$$= A_c \cos \omega_c t + m(t) \cos \omega_c t$$

$$S(\omega) = \pi A_c [\delta(\omega + \omega_c) + \delta(\omega - \omega_c)]$$

$$+ \frac{1}{2} [M(\omega + \omega_c) + M(\omega - \omega_c)]$$

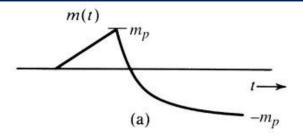
- s(t) = un-modulated carrier
 - + upper sideband (USB)
 - + lower sideband (LSB)

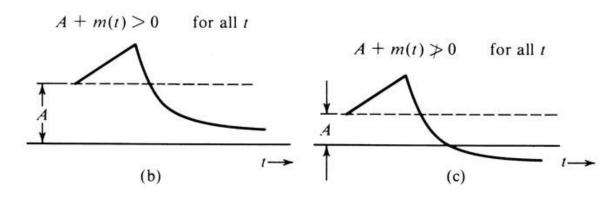


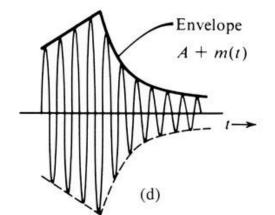


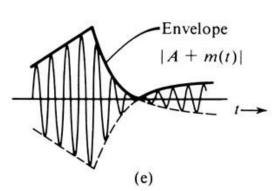
BW = B Hz(Baseband) BW of s(t) = 2B Hz = $4\pi B$ rad/s (Transmission BW)

 $\pm m_p$: maximum and minimum values of m(t)





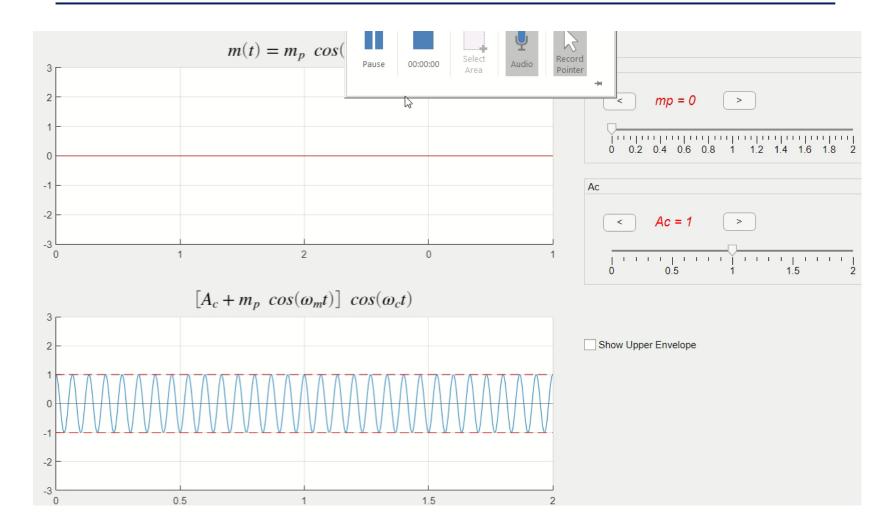




• Condition for using envelop detection in demodulator:

$$A_c + m(t) > 0$$
 for all t
 $A_c \ge m_p \ (A_c \ge \text{absolute -ve peak amplitude, } |m(t)_{min}|)$
Thus, the envelope has the same shape as $m(t)$

- This makes the demodulator simpler (no need for generation of local carrier)
- Modulation index: $\mu = \frac{m_p}{A_c}$ (Generally, $\mu = \frac{|m(t)_{min}|}{A_c}$)
- $0 \le \mu \le 1$ since $A_c \ge m_p$ and there is no upper bound on A_c



- $\mu \times 100$: percentage modulation
- $\mu > 1$: over-modulation (envelope detection not viable)
- Coherent detection can be used for both DSB-SC and DSB-LC (for any μ)

- We transmit higher power level than DSB-SC (higher cost for transmitter)
- Trade-off: higher power at transmitter, cheaper less complex receiver.
- Suitable for broadcast systems (1 transmitter, many receivers)
 - More economical to have one expensive high power transmitter, and many cheaper receivers.

- $s(t) = A_c \cos \omega_c t + m(t) \cos \omega_c t$ carrier sidebands (signal)
- Carrier power doesn't carry any information and is a waste of power
- Carrier power $P_c = A_c^2/2$
- Sideband power $P_S = \lim_{T \to \infty} \frac{1}{T} \int_T [m(t) cos \omega_c]^2 dt$

$$= \frac{1}{2} \lim_{T \to \infty} \frac{1}{T} \int_{T} m^{2}(t) dt = \frac{1}{2} P_{m}$$

(1/2 baseband signal power)

- Total transmitted power = $P_c + P_s$
- Useful power (containing information) = P_s

• Efficiency
$$\eta = \frac{\text{useful power}}{\text{total power}} = \frac{P_S}{P_C + P_S} = \frac{P_m}{A_C^2 + P_m}$$

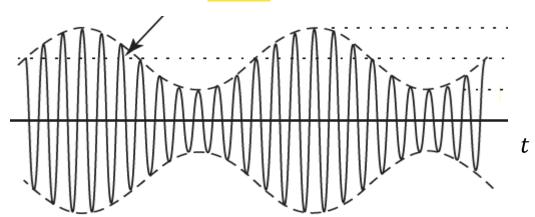
Example (tone modulation):

```
Let m(t) = A_m \cos \omega_m t (single tone signal)
```

Then
$$s(t) = (A_c + A_m \cos \omega_m t) \cos \omega_c t$$

Modulation index:
$$\mu = A_m/A_c$$

 $s(t) = A_c \cos \omega_c t + \mu A_c \cos \omega_m t \cos \omega_c t$



•
$$m(t) = A_m \cos \omega_m t = \mu A_c \cos \omega_m t$$

•
$$\overline{m^2(t)} = \frac{(\mu A_c)^2}{2}$$

• Efficiency
$$\eta = \frac{\text{useful power}}{\text{total power}} = \frac{P_S}{P_C + P_S} = \frac{P_m}{A_C^2 + P_m}$$

$$= \frac{\mu^2}{2 + \mu^2} 100\%$$
 Only for tone modulation

• For $0 \le \mu \le 1$, $0\% \le \eta \le 33.3\%$

- For tone modulation, under the best conditions $(\mu = 1)$, only one-third of the transmitted power is carrying messages. (useful information)
- For practical signals, the efficiency is even worse, on the order of 25% or lower.
- Smaller values of μ , degrade efficiency further.

- 1. Sketch the modulated signal for $\mu=0.5$ and 1 in time domain. Assume $A_m=1$. Find the efficiency in each case.
- 2. Sketch the spectrum of the modulated signal. Find the bandwidth
- 3. Find the carrier power, USB and LSB power and total sideband power.

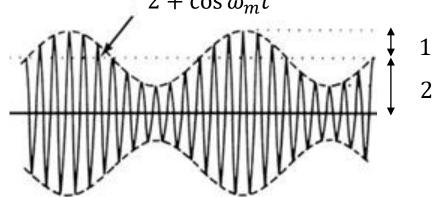
1.
$$s(t) = (A_c + A_m \cos \omega_m t) \cos \omega_c t$$

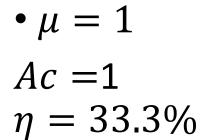
 $A_m = 1$ $2 + \cos \omega_m t$

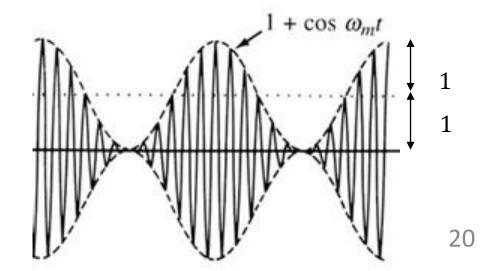
•
$$\mu = 0.5$$

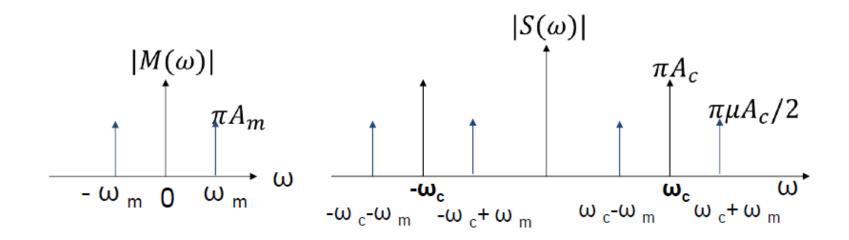
$$Ac = 2$$

$$\eta = 11.1\%$$









- $s(t) = A_c \cos \omega_c t + (A_m \cos \omega_m t) \cos \omega_c t$
- $BW = 2f_m \text{ Hz} = 2\omega_m \text{ rad/s}$

$$S(\omega) = \pi A_c [\delta(\omega - \omega_c) + \delta(\omega + \omega_c)]$$
$$+ \frac{1}{2} [M(\omega - \omega_c) + M(\omega + \omega_c)]$$

$$= \pi A_{c} \left[\delta(\omega - \omega_{c}) + \delta(\omega + \omega_{c}) \right]$$

$$+ (\pi \mu A_{c}/2) \left[\delta(\omega + (\omega_{c} + \omega_{m})) \right]$$

$$+ (\pi \mu A_{c}/2) \left[\delta(\omega - (\omega_{c} + \omega_{m})) \right]$$

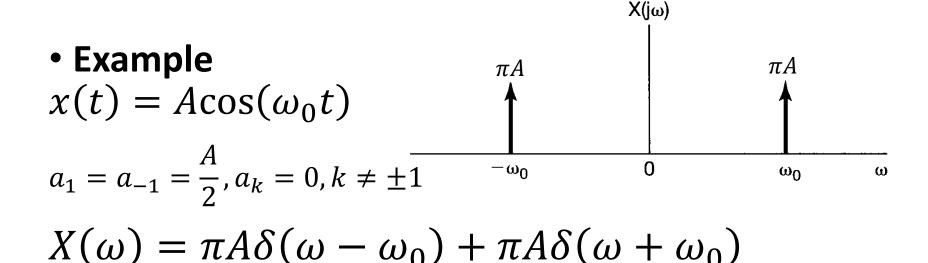
$$+ (\pi \mu A_{c}/2) \left[\delta(\omega - (\omega_{c} - \omega_{m})) \right]$$

$$+ (\pi \mu A_{c}/2) \left[\delta(\omega + (\omega_{c} - \omega_{m})) \right]$$

Review Fourier Transform for Periodic Signals

For a periodic x(t) with FS coefficients $\{a_k\}$

$$X(\omega) = \sum_{k=-\infty}^{\infty} 2\pi a_k \, \delta(\omega - k\omega_0)$$



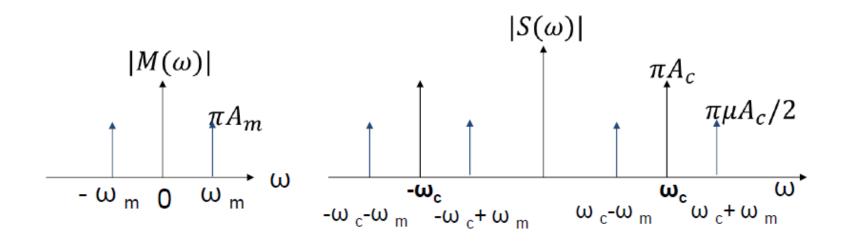
Review Power of sinusoidal signals

Parseval's theorem

Periodic
$$P = \frac{1}{T} \int_{T} |x(t)|^{2} dt = \sum_{k=-\infty}^{\infty} |a_{k}|^{2}$$

• For
$$x(t) = A \cos \omega t$$

$$P = \frac{A^2}{2}$$

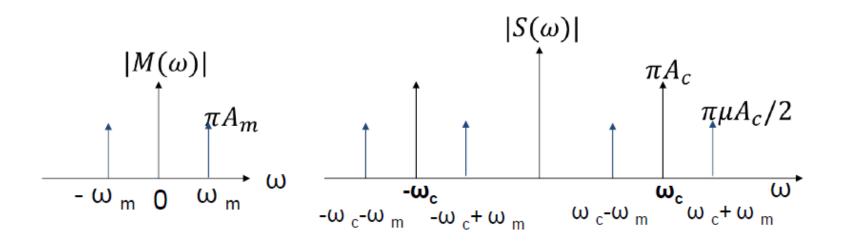


•
$$s(t) = A_c \cos \omega_c t + (A_m \cos \omega_m t) \cos \omega_c t$$

$$= A_c \cos \omega_c t + \frac{A_m}{2} \cos(\omega_c + \omega_m) t$$

$$+ \frac{A_m}{2} \cos(\omega_c - \omega_m) t$$

$$X(\omega) = \sum_{k=0}^{\infty} 2\pi a_k \, \delta(\omega - k\omega_0)$$



•
$$s(t) = A_c \cos \omega_c t + (A_m \cos \omega_m t) \cos \omega_c t$$

• To get the power: divide $S(\omega)$ by 2π (to get F.S. coeff), square the magnitude and add all terms

$$(P = \sum_{k=-\infty}^{\infty} |a_k|^2)$$

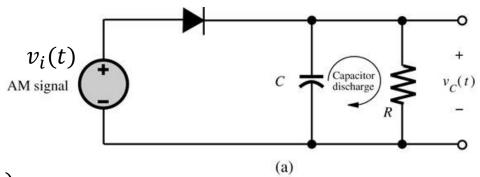
- Carrier power = $Ac^2/2$
- U.S.B power = L.S.B. power = $\mu^2 Ac^2 / 8$
- Total sidebands power = $\mu^2 Ac^2 / 4$

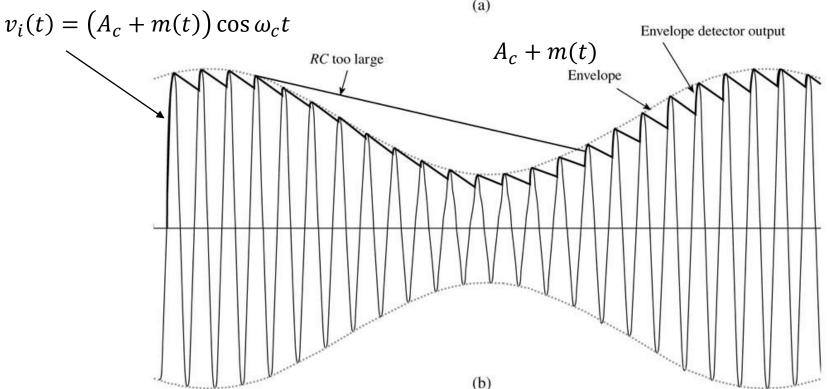
•
$$P_t = Pc + Ps = (Ac^2/2) + (\mu^2 Ac^2/4)$$

= $Pc [1 + (\mu^2/2)]$

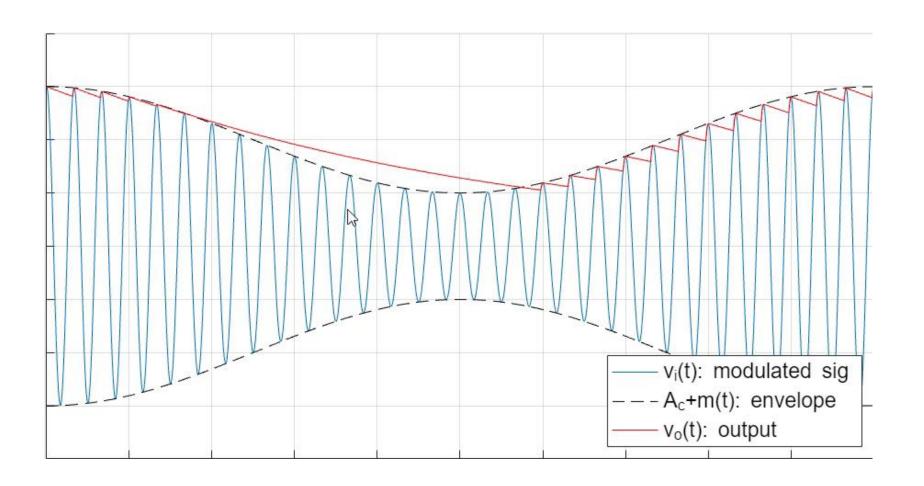
Efficiency
$$\eta = \frac{\text{useful power}}{\text{total power}} = \frac{\mu^2}{2+\mu^2}$$
 (only for tone modulation)

DSB-LC Demodulator: Envelope detector (Non-coherent/asynchronous)





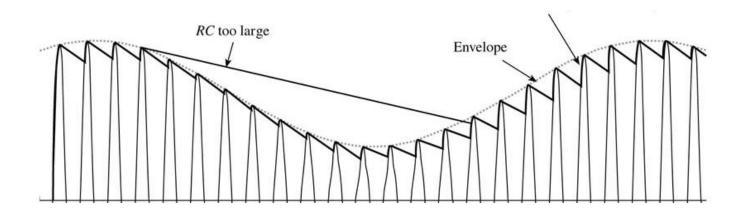
DSB-LC Demodulator: Envelope detector (Non-coherent/asynchronous)



DSB-LC Demodulator: Envelope detector (Non-coherent/asynchronous)

$$\frac{1}{W} > RC > (\frac{1}{f_c} = T_c)$$
 signal BW carrier freq.

- The dc term A_c can be blocked by a capacitor
- The ripple may be reduced further by a LPF



Notes on DSB-LC (conventional AM)

Advantages: Ease of Modulation and demodulation (cheap to build the system)

- Disadvantages
 - Waste of power.
 - Waste of B.W.

So Far ...

- DSC-SC
 - More complicated demodulator (synchronization issues)
 - Excellent power efficiency
 - Need BW = 2W to send a signal of BW=W
- DSB-LC (Conventional AM)
 - Easy to demodulate (envelope detector)
 - Low power efficiency
 - Need BW = 2W to send a signal of BW=W
- Can we improve the usage of the BW?