



**Course: Digital Control**

**Date: 12/ 12 / 2020**

**Mid-Term Exam (1) Model Answer**

**Time: 1.00 H**

**Question (1): (10 marks):**

1. The Euler backward emulation technique is applied by using the substitution:

$$s = \frac{z - 1}{Tz}$$

Inserting this equation into the controller  $C(s)$  yields:

$$C(z) = \frac{2\frac{z-1}{Tz} + 1}{\frac{z-1}{Tz} + \alpha} = \frac{z(2 + T) - 2}{z(1 + \alpha T) - 1}$$

2.

The controller  $C(z)$  is stable if its pole  $z_p$  fulfills the condition  $|z_p| < 1$ .

The pole  $z_p$  is obtained from the equation  $z(1 + \alpha T) - 1 = 0$ .

$$z_p = \frac{1}{1 + \alpha T}$$

From the condition above,  $\alpha$  must satisfy:

$$\left| \frac{1}{1 + \alpha T} \right| < 1$$

$$|1 + \alpha T| > 1$$

$$1 + \alpha T > 1 \quad \text{or} \quad 1 + \alpha T < -1$$

These inequalities constraints lead to the solutions  $\alpha > 0$  or  $\alpha < -\frac{2}{T}$ .

3.

The digital controller is:  $C(z) = \frac{z(2+T)-2}{z(1+\alpha T)-1}$

We simplify the controller as:

$$C(z) = \left( \frac{2+T}{1+\alpha T} \right) \frac{z - \left( \frac{2}{2+T} \right)}{z - \left( \frac{1}{1+\alpha T} \right)} = K \frac{z-a}{z-b}$$

Where  $K = \frac{2+T}{1+\alpha T}$ ,  $a = \frac{2}{2+T}$ ,  $b = \frac{1}{1+\alpha T}$

Then,

$$\frac{U(z)}{E(z)} = K \frac{z-a}{z-b}$$

$$\begin{aligned} z.U(z) - b.U(z) &= z.K.E(z) - K.a.E(z) \\ u(k+1) - b.u(k) &= K.e(k+1) - K.a.e(k) \\ u(k) &= b.u(k-1) + K.e(k) - K.a.e(k-1) \end{aligned}$$

Code:

Parameters b, a, K

Variable e\_prev=0, e, u\_prev=0, u.

Loop:

Read e.

Compute  $u = b*u\_prev + K*e + K*a*e\_prev$ .

Output u.

$e\_prev = e$ .

$u\_prev = u$ .

End loop.

### **Question (2):**

$$G(s) = \frac{V(s)}{U(s)} = \frac{1}{Ms + b} = \frac{K/\tau}{s + 1/\tau}$$

where  $K = 1/b$  and  $\tau = M/b$

For a zero-order hold:

$$G(z) = (1 - z^{-1})Z\left[\frac{G(s)}{s}\right]$$

$$\frac{G(s)}{s} = \frac{K/\tau}{s(s+1/\tau)} = (K/\tau) \left[ \frac{\tau}{s} - \frac{\tau}{s+1/\tau} \right]$$

Thereby

$$\begin{aligned} G(z) &= (1-z^{-1})Z \left[ \left( \frac{K}{\tau} \right) \left[ \frac{\tau}{s} - \frac{\tau}{s+1/\tau} \right] \right] \\ &= (1-z^{-1})Z \left[ K \left[ \frac{1}{s} - \frac{1}{s+1/\tau} \right] \right] \\ \Rightarrow G(z) &= \frac{z-1}{z} K \left[ \frac{z}{z-1} - \frac{z}{z-e^{-T/\tau}} \right] = K \left[ 1 - \frac{z-1}{z-e^{-T/\tau}} \right] \end{aligned}$$

or

$$G(z) = K \left[ \frac{z - e^{-T/\tau} - z + 1}{z - e^{-T/\tau}} \right] = K \left[ \frac{1 - e^{-T/\tau}}{z - e^{-T/\tau}} \right]$$

**Question (3): (10 marks).**

The maximum frequency can be determined from Nyquist frequency:

$$f < \frac{f_s}{2}$$

$$\text{Therefore, } \omega < \frac{2\pi}{2T} = \frac{\pi}{0.1} = 31.4159 \text{ rad/sec}$$

$$G(z) = Z \left\{ \frac{1 - e^{-sT}}{s} \frac{5}{s+5} \right\} = \frac{1 - e^{-0.5}}{z - e^{-0.5}},$$

$$G(z) = \frac{0.393}{z - 0.606}.$$

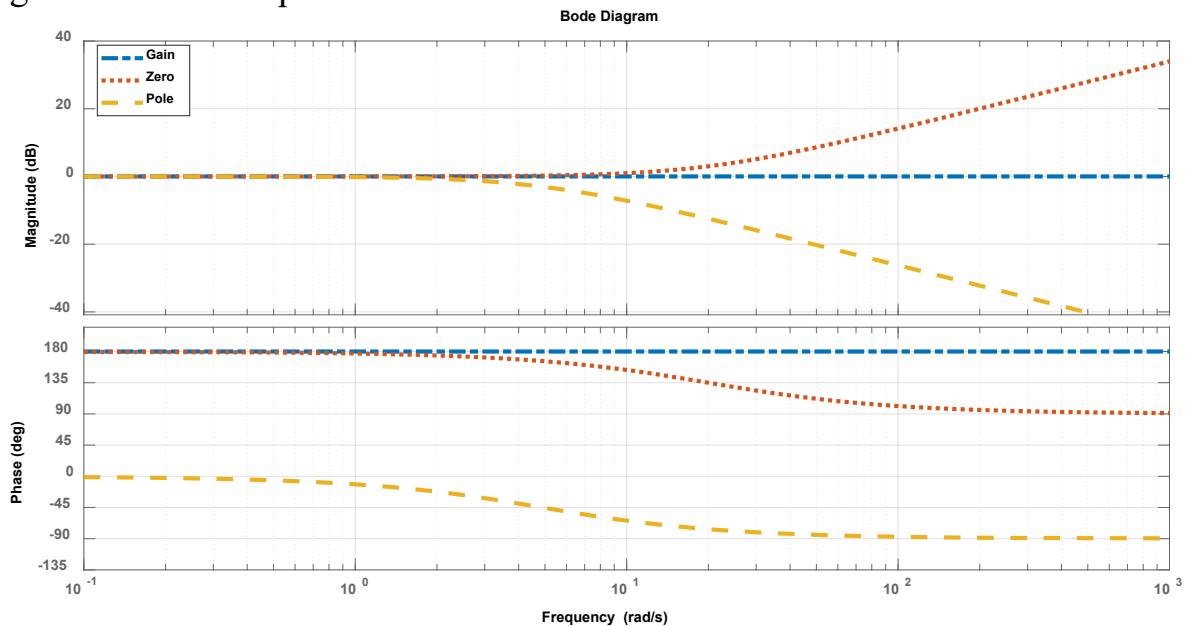
Applying bilinear transform:

$$z = \frac{1 + \left(\frac{T}{2}\right)w}{1 - \left(\frac{T}{2}\right)w} = \frac{1 + 0.05w}{1 - 0.05w}$$

$$G(w) = \frac{0.3935}{\frac{1 + 0.05w}{1 - 0.05w} - 0.6065} = \frac{-0.01968w + 0.3935}{0.08033w + 0.3935}$$

$$G(w) = \frac{-0.24494(w - 20)}{(w + 4.899)} = \frac{-1.0020\left(\frac{j\omega}{20} - 1\right)}{\left(\frac{j\omega}{4.899} + 1\right)}$$

Bode diagram of each component



Complete bode diagram:

