

Digital Signal Processing for Music

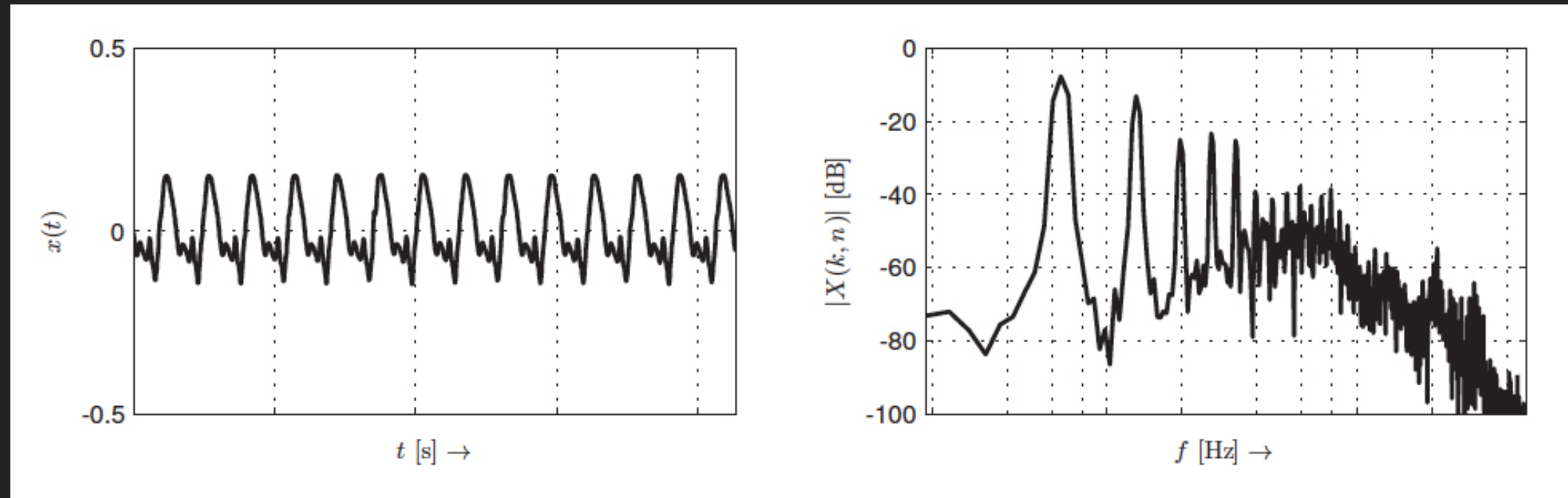
Part 7: Fourier Transform, Part 1

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Fourier Transform: Overview

- » Fourier series to Fourier Transform
- » Properties of the Fourier Transform
- » Windowed Fourier Transform (STFT)
- » Transform of sampled time signals
- » Discrete Fourier Transform

Fourier Transform: Introduction



Fourier series is a brilliant insight, but:

- » Works only for periodic signals
- » Difficult to use for real-world analysis as it requires knowledge of fundamental frequency

→ **Fourier Transform**

Fourier series, revisited

$$c_k = \frac{1}{T_0} \int_{-\frac{T_0}{2}}^{\frac{T_0}{2}} x(t) e^{-j\omega_0 kt} dt$$

- » Fourier series coefficient can be interpreted as **correlation coefficient** between signal and sinusoids of different frequencies
- » Only frequencies $k\omega_0$ are used (ω_0 has to be known)
- » Fourier series produces a '**line spectrum**'

Fourier series, revisited

$$c_k = \frac{1}{T_0} \int_{-\frac{T_0}{2}}^{\frac{T_0}{2}} x(t) e^{-j\omega_0 kt} dt$$

- » Distance between frequency components decreases as T_0 increases
- » Aperiodic functions could be analyzed by increasing $T_0 \rightarrow \infty$

Fourier series, revisited

$$c_k = \frac{1}{T_0} \int_{-\frac{T_0}{2}}^{\frac{T_0}{2}} x(t) e^{-j\omega_0 kt} dt$$

$$\rightarrow T_0 \rightarrow \infty$$

$$\rightarrow k\omega_0 \rightarrow \omega$$

$$\rightarrow \frac{1}{T_0} \rightarrow 0$$

To avoid zero result, multiply with T_0

Definition of Fourier Transform (Continuous)

$$X(j\omega) = \mathcal{F}[x(t)] = \int_{-\infty}^{\infty} x(t)e^{-j\omega t} dt$$

Example 1: Rect Window

$$w_R(t) = \begin{cases} 1, & -\frac{1}{2} \leq t \leq \frac{1}{2} \\ 0, & \text{otherwise} \end{cases}$$

$$W_R(\omega) = \int_{-\infty}^{\infty} w_R(t) e^{-j\omega t} dt$$

$$= \int_{-\frac{1}{2}}^{\frac{1}{2}} e^{-j\omega t} dt$$

$$= \frac{1}{-j\omega} \underbrace{\left(e^{-j\frac{\omega}{2}} - e^{j\frac{\omega}{2}} \right)}_{= -2j \sin\left(\frac{\omega}{2}\right)}$$

$$= \frac{\sin\left(\frac{\omega}{2}\right)}{\frac{\omega}{2}} = \text{sinc}\left(\frac{\omega}{2}\right)$$

Example 2: Dirac (Impulse)

$$\delta(t) = \begin{cases} 1, & t = 0 \\ 0, & t \neq 0 \end{cases} \quad \int_{-\infty}^{\infty} \delta(t) dt = 1$$

$$\Delta(j\omega) = \int_{-\infty}^{\infty} \delta(t) e^{-j\omega t} dt = e^{-j\omega \cdot 0} = 1$$

Shifted Dirac $\delta(t - \tau_0)$

Properties of Fourier Transform

Property 1: Invertibility

$$\begin{aligned}x(t) &= \mathfrak{F}^{-1}[X(j\omega)] \\&= \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) e^{j\omega t} d\omega\end{aligned}$$

Reminder: Signal reconstruction with Fourier series coefficients

$$x(t) = \sum_{k=-\infty}^{\infty} c_k e^{j\omega_0 kt}$$

» Comments:

» Invertibility: No signal can be reconstructed with only the

Property 2: Superposition

$$y(t) = c_1 \cdot x_1(t) + c_2 \cdot x_2(t)$$

$$\mapsto Y(j\omega) = c_1 \cdot X_1(j\omega) + c_2 \cdot X_2(j\omega)$$

$$Y(j\omega) = \int_{-\infty}^{\infty} (c_1 \cdot x_1(t) + c_2 \cdot x_2(t)) \cdot e^{-j\omega t} dt$$

$$= c_1 \cdot \int_{-\infty}^{\infty} x_1(t) e^{-j\omega t} dt + c_2 \cdot \int_{-\infty}^{\infty} x_2(t) e^{-j\omega t} dt$$

$$= c_1 \cdot X_1(j\omega) + c_2 \cdot X_2(j\omega)$$

Property 3: Convolution and Multiplication

$$y(t) = \int_{-\infty}^{\infty} h(\tau) \cdot x(t - \tau) d\tau$$

$$\mapsto Y(j\omega) = H(j\omega) \cdot X(j\omega)$$

$$Y(j\omega) = \int_{-\infty}^{\infty} y(t) e^{-j\omega t} dt$$

$$= \int_{-\infty}^{\infty} \left(\int_{-\infty}^{\infty} h(\tau) \cdot x(t - \tau) d\tau \right) e^{-j\omega t} dt$$

$$= \int_{-\infty}^{\infty} h(\tau) \int_{-\infty}^{\infty} x(t - \tau) e^{-j\omega t} dt d\tau$$

$$\begin{aligned}
 y(t) &= \int_{-\infty}^{\infty} h(\tau) \cdot x(t - \tau) d\tau & Y(j\omega) &= \int_{-\infty}^{\infty} y(t) e^{-j\omega t} dt \\
 \mapsto Y(j\omega) &= H(j\omega) \cdot X(j\omega) & &= \int_{-\infty}^{\infty} \left(\int_{-\infty}^{\infty} h(\tau) \cdot x(t - \tau) d\tau \right) e^{-j\omega t} dt \\
 & & &= \int_{-\infty}^{\infty} h(\tau) \int_{-\infty}^{\infty} x(t - \tau) e^{-j\omega t} dt d\tau \\
 & & &= \int_{-\infty}^{\infty} h(\tau) e^{-j\omega \tau} \underbrace{\int_{-\infty}^{\infty} x(t - \tau) e^{-j\omega(t - \tau)} d(t - \tau)}_{X(j\omega)} d\tau \\
 & & &= \int_{-\infty}^{\infty} h(\tau) e^{-j\omega \tau} d\tau \cdot X(j\omega) \\
 & & &= H(j\omega) \cdot X(j\omega)
 \end{aligned}$$

Property 4: Parseval's Theorem

$$\int_{-\infty}^{\infty} x^2(t) dt = \frac{1}{2\pi} \int_{-\infty}^{\infty} |X(j\omega)|^2 d\omega$$

$$\int_{-\infty}^{\infty} h(\tau) \cdot x(t - \tau) d\tau = \frac{1}{2\pi} \int_{-\infty}^{\infty} H(j\omega) \cdot X(j\omega) e^{j\omega t} d\omega$$

$$H(j\omega) \longrightarrow X^*(j\omega)/h(\tau) \longrightarrow x(-\tau), t = 0$$

$$\int_{-\infty}^{\infty} x(-\tau) \cdot x(-\tau) d\tau = \frac{1}{2\pi} \int_{-\infty}^{\infty} X^*(j\omega) \cdot X(j\omega) d\omega$$

$$\int_{-\infty}^{\infty} x^2(t) dt = \frac{1}{2\pi} \int_{-\infty}^{\infty} |X(j\omega)|^2 d\omega$$

Property 5: Time & Frequency Shift

$$y(t) = x(t - t_0) \rightarrow Y(j\omega) = X(j\omega)e^{-j\omega t_0}$$

$$\begin{aligned}\int_{-\infty}^{\infty} x(t - t_0)e^{-j\omega t} dt &= \int_{-\infty}^{\infty} x(\tau)e^{-j\omega(\tau+t_0)} d\tau \\ &= e^{-j\omega t_0} \int_{-\infty}^{\infty} x(\tau)e^{-j\omega\tau} d\tau \\ &= e^{-j\omega t_0} \cdot X(j\omega)\end{aligned}$$

$$\frac{1}{2\pi} \int_{-\infty}^{\infty} X(j(\omega - \omega_0))e^{j\omega t} d\omega = e^{j\omega_0 t} \cdot x(t)$$

Property 6: Symmetry 1/2

$$|X(j\omega)| = |X(-j\omega)|$$

$$\Phi_X(\omega) = -\Phi_X(-\omega)$$

Time signal sum of even and odd component $x_e(t)$, $x_o(t)$

$$x(t) = \underbrace{\frac{1}{2}(x(t) + x(-t))}_{x_e(t)} + \underbrace{\frac{1}{2}(x(t) - x(-t))}_{x_o(t)}$$

$$X_e(j\omega) = \int_{-\infty}^{\infty} x_e(t) \cos(\omega t) dt - j \underbrace{\int_{-\infty}^{\infty} x_e(t) \sin(\omega t) dt}_{=0}$$

$X_e(j\omega)$ is real

$X_e(j\omega) = X_e(-j\omega)$ (substitute $x(t)$ with $x(-t)$)

Property 6: Symmetry 2/2

$$|X(j\omega)| = |X(-j\omega)|$$

$$\Phi_X(\omega) = -\Phi_X(-\omega)$$

Time signal sum of even and odd component $x_e(t)$, $x_o(t)$

$$x(t) = \underbrace{\frac{1}{2}(x(t) + x(-t))}_{x_e(t)} + \underbrace{\frac{1}{2}(x(t) - x(-t))}_{x_o(t)}$$

$$X_o(j\omega) = \underbrace{\int_{-\infty}^{\infty} x_o(t) \cos(\omega t) dt}_{=0} - j \int_{-\infty}^{\infty} x_o(t) \sin(\omega t) dt$$

$X_o(j\omega)$ is imaginary

$X_o(j\omega) = X_o(-j\omega)$ (substitute $x(t)$ with $-x(-t)$)

Property 7: Time & Frequency Scaling

$$y(t) = x(c \cdot t)$$

$$\mapsto Y(j\omega) = \frac{1}{|c|} X\left(j\frac{\omega}{c}\right)$$

$$Y(j\omega) = \int_{-\infty}^{\infty} x(c \cdot t) e^{-j\omega t} dt$$

$$= \int_{-\infty}^{\infty} x(\tau) e^{-j\omega \frac{\tau}{c}} d\frac{\tau}{c}$$

$$= \frac{1}{c} \int_{-\infty}^{\infty} x(\tau) e^{-j\frac{\omega}{c}\tau} d\tau$$

$$= \frac{1}{c} X\left(j\frac{\omega}{c}\right)$$

Verifying Fourier Transform Implementation

- » *Property 1: **Invertibility***: Running IFT returns the EXACT original signal

$$x(t) = \mathfrak{F}^{-1}[X(j\omega)]$$

- » *Property 2: **Superposition***: Scaled addition in time domain maps to linear scale in magnitudes in frequency domain

$$\begin{aligned} y(t) &= c_1 \cdot x_1(t) + c_2 \cdot x_2(t) \\ \mapsto Y(j\omega) &= c_1 \cdot X_1(j\omega) + c_2 \cdot X_2(j\omega) \end{aligned}$$

Verifying Fourier Transform Implementation

- » *Property 4: **Parseval`s Theorem***: Energy conservation between time domain and frequency domain

$$\int_{-\infty}^{\infty} x^2(t) dt = \frac{1}{2\pi} \int_{-\infty}^{\infty} |X(j\omega)|^2 d\omega$$

- » *Property 6: **Symmetry***: Frequency domain is symmetric across zero frequency (and for DFT across windowSize)

$$\begin{aligned} |X(j\omega)| &= |X(-j\omega)| \\ \Phi_X(\omega) &= -\Phi_X(-\omega) \end{aligned}$$

Key Properties for Future Topics

- » **Property 3: Convolution and Multiplication:** Convolution in the time domain is multiplication in the frequency domain

$$y(t) = \int_{-\infty}^{\infty} h(\tau) \cdot x(t - \tau) d\tau$$

$$\mapsto Y(j\omega) = H(j\omega) \cdot X(j\omega)$$

- » **Property 5: Time & Frequency Shift:** Time shift in time domain is phase shift in frequency domain

$$y(t) = x(t - t_0) \rightarrow Y(j\omega) = X(j\omega)e^{-j\omega t_0}$$