



**BITS Pilani**  
Pilani Campus

# **Mechatronics** (Merged - DEZG516/DMZG511/ESZG511)

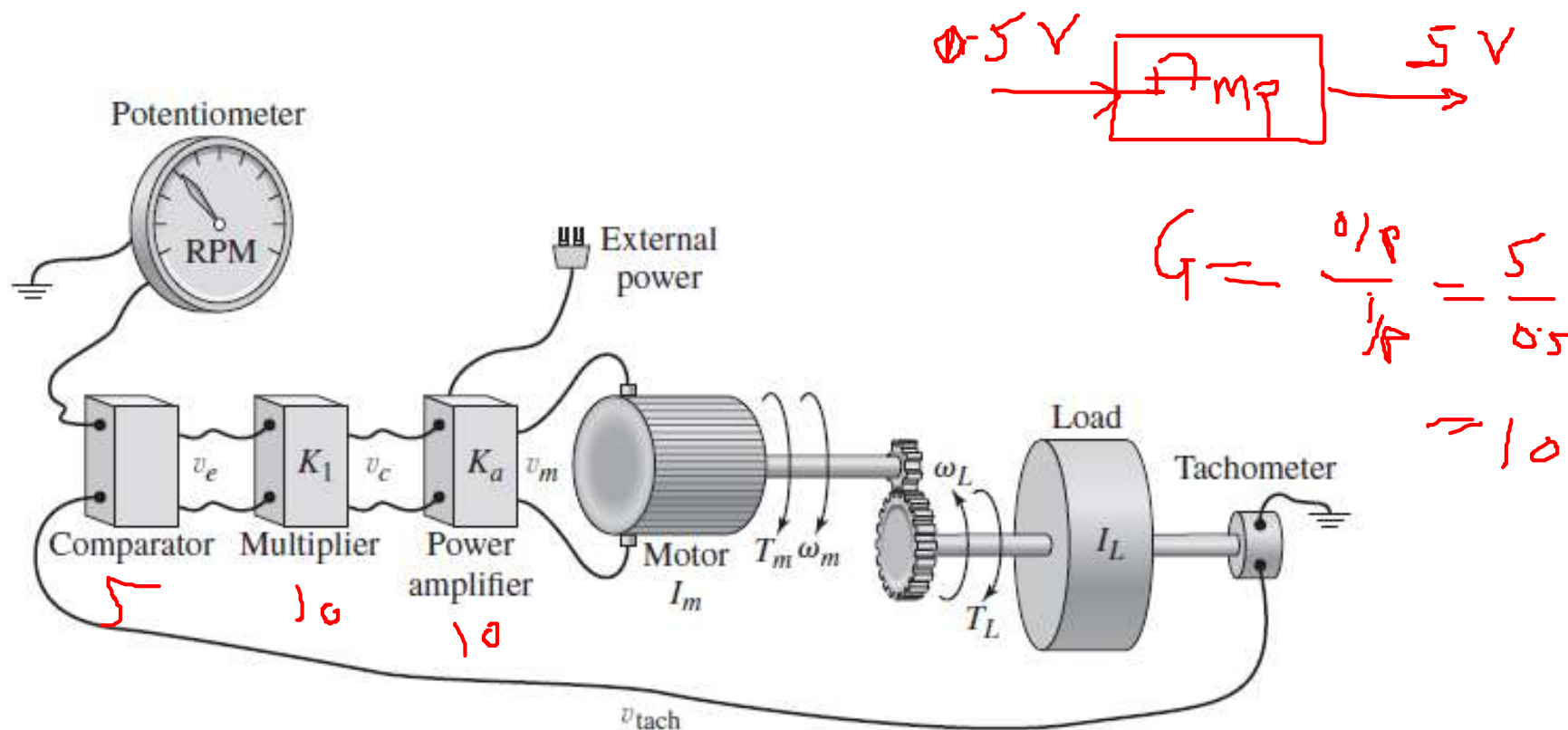
**Lecture**



# Session 11

Type	Content Ref.	Topic Title	Study/HW Resource Reference
Pre CH			
During CH	T1, R4	Transfer Function, Effect of pole location, Frequency response of systems	T1: Chapter 14,15, R4
Post CH			Chapter end problems

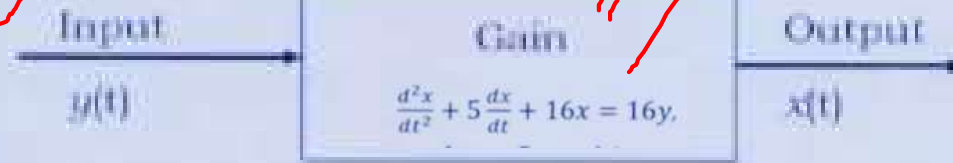
# Control systems



# Transfer functions

- For an amplifier we define gain as

- Gain =  $\frac{\text{Output}}{\text{Input}}$   $\Rightarrow \eta$



- The physical meaning of this equation is that if the gain of an amplifier is 5, then for an input of 4mV, the output will be 20 mV.
- For many systems the relationship between output and input is not in algebraic form as above.

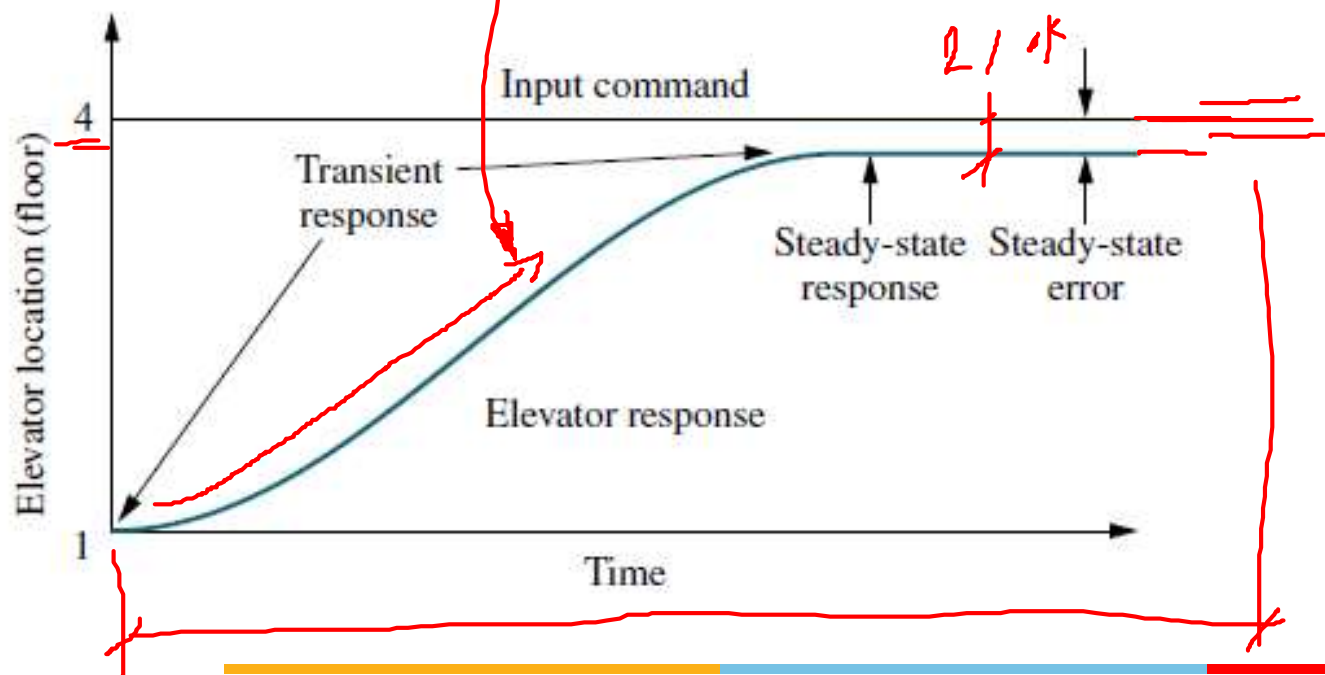
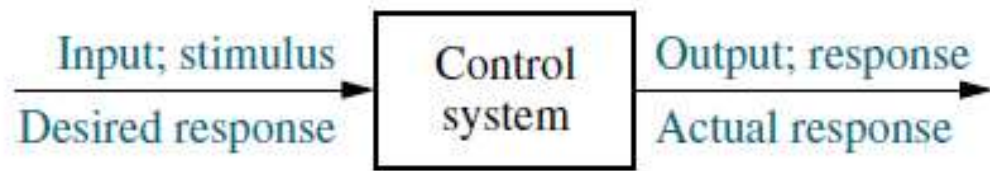
$$\text{gain} = \frac{\text{output}}{\text{input}} = \frac{x(t)}{y(t)}$$



# Transfer functions

- Rather it is in form of differential equation, due to which above formula can not be applied as differential equation describes how system behaves with time.
- This differential equation can be converted into an algebraic equation using Laplace transform.
- We say that conversion has been taken place from time domain to s domain.

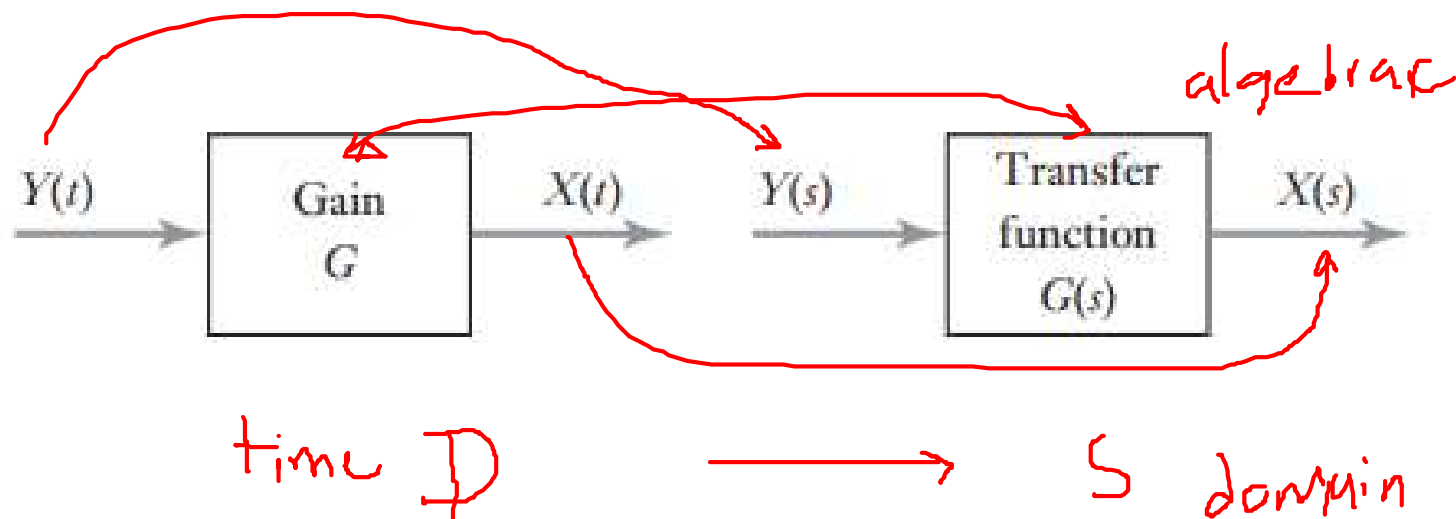
# Control systems



# Transfer function

transfer function =  $\frac{\text{Laplace transform of output}}{\text{Laplace transform of input}}$

$$G(s) = \frac{X(s)}{Y(s)}$$





# Laplace Transforms

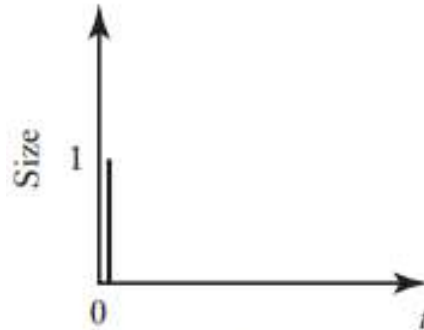
## Advantage of Laplace Transform Method

- It allows use of graphical technique for predicting the system performance without solving system differential equation.
- When we solve differential equation using this method transients and steady state components of the solution can be obtained simultaneously.

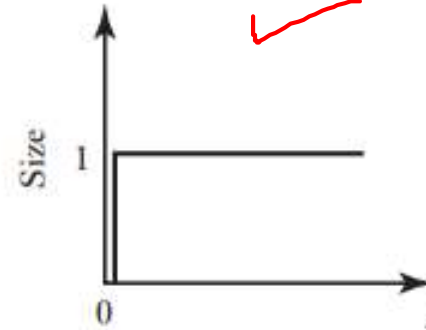


# Laplace Transform

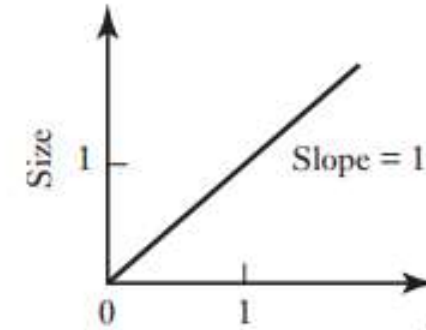
types of  $y_p$



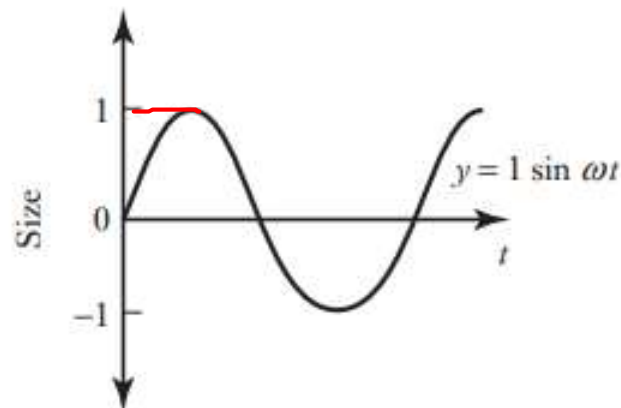
Unit impulse at zero time  
has the transform of 1



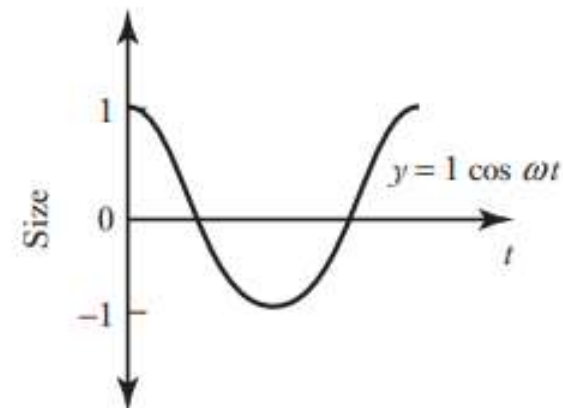
Unit step at zero time  
has the transform of  $1/s$



Unit ramp at zero time  
has the transform of  $1/s^2$



Unit amplitude sine wave has  
the transform of  $\omega/(s^2 + \omega^2)$



Unit amplitude cosine wave has  
the transform of  $s/(s^2 + \omega^2)$



# Laplace transform of derivatives

$$\mathcal{L}\left\{\frac{d}{dt}f(t)\right\} = sF(s) - f(0)$$

$$\mathcal{L}\left\{\frac{d^2}{dt^2}f(t)\right\} = \underline{s^2} \underline{F(s)} - \underline{s} \underline{f(0)} - \frac{d}{dt} \underline{f(0)}$$

# Laplace transform rules

$af(t)$  has the transform of  $\underline{a}F(s)$

$f(t) + g(t)$  has the transform  $F(s) + G(s)$

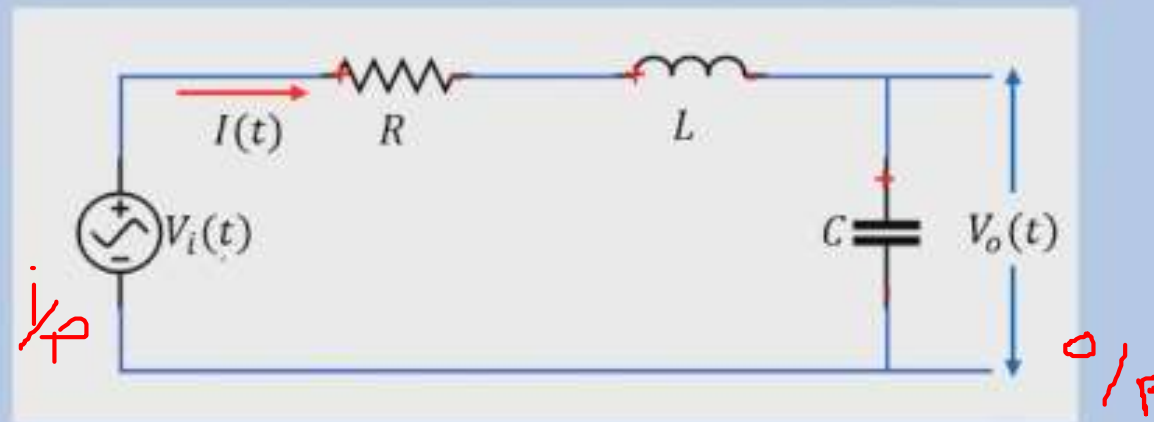
$$\mathcal{L}\left\{\frac{d}{dt}f(t)\right\} = \underline{sF(s)} - f(0)$$

$$\mathcal{L}\left\{\frac{d^2}{dt^2}f(t)\right\} = s^2F(s) - \underbrace{sf(0)}_{=0} - \frac{d}{dt}f(0)$$

$$\mathcal{L}\left\{\int_0^t f(t) dt\right\} = \frac{1}{s}F(s)$$

# Transfer functions

## Transfer Function : Example 1



Model Equations:

$$\underline{V_i(t)} = RI(t) + L \frac{dI}{dt} + \frac{1}{C} \int I dt \quad \text{i/p}$$

$$\underline{V_o(t)} = \frac{1}{C} \int I dt \quad \text{o/p}$$

$$TF = \frac{L(o/p)}{L(i/p)}$$

# Transfer functions

## 2. Input and Output Variables:

- Input:  $V_i(t)$
- Output:  $V_o(t)$

## 3. Laplace Transform: (assuming initial conditions to be zero)

$$V_i(s) = RI(s) + sLI(s) + \frac{1}{sC}I(s) \quad \text{--- I}$$

$$V_o(s) = \frac{1}{sC}I(s) \quad \text{--- II}$$

## 4. Transfer Function:

$$G(s) = \frac{V_o(s)}{V_i(s)} = \frac{\frac{1}{sC}I(s)}{\left(R + sL + \frac{1}{sC}\right)I(s)} = \frac{\frac{1}{sC}}{\left(R + sL + \frac{1}{sC}\right)} = \frac{1}{s^2LC + sRC + 1}$$

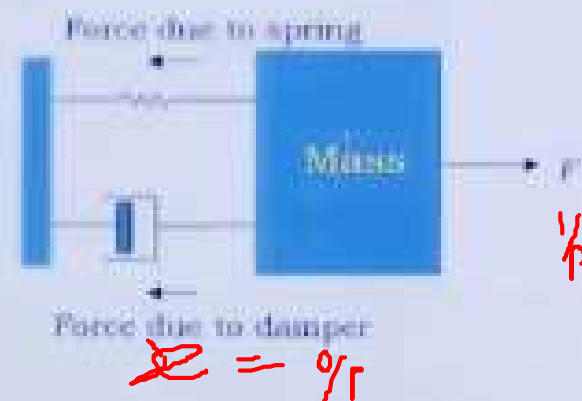


# Transfer functions

$$= \frac{L\{y(t)\}}{L\{x(t)\}} = \frac{Y(s)}{X(s)}$$

## Example

- Consider mass spring damper system shown.
- Dynamics of system is described by



- $m \frac{d^2x}{dt^2} + c \frac{dx}{dt} + kx = F$
- Taking Laplace transform
- $ms^2X(s) + csX(s) + kX(s) = F(s)$

$$G(s) = \frac{X(s)}{F(s)} = \frac{1}{ms^2 + cs + k}$$

$$X(s) [ms^2 + cs + k] = F(s)$$



# Laplace Transforms

- Let  $f(t)$  be a function of time  $t$  such that  $f(t)=0$ , for  $t<0$
- $s$  be a complex variable
- $\mathcal{L}$  be an operational symbol indicating the Laplace transform of function
- Laplace transform of function  $f(t)$  is given by
- $\mathcal{L}[f(t)] = F(s) = \int_0^{\infty} e^{-st} dt [f(t)] = \int_0^{\infty} f(t) e^{-st} dt$



# Inverse Laplace Transforms

- The process of finding the function  $f(t)$  from the Laplace transform  $F(s)$  is called inverse Laplace transformation.
- Inverse Laplace transform is represented as  $\mathcal{L}^{-1}$
- Inverse Laplace transform can be found from  $F(s)$  by inversion integral as
- $$\mathcal{L}^{-1}[F(s)] = f(t) = \frac{1}{2\pi j} \int_{c-j\infty}^{c+j\infty} F(s)e^{st} ds, \text{ for } t > 0$$
- Here  $c$  is the abscissa of convergence, is a real constant and is chosen larger than the real parts of all singular points of  $F(s)$



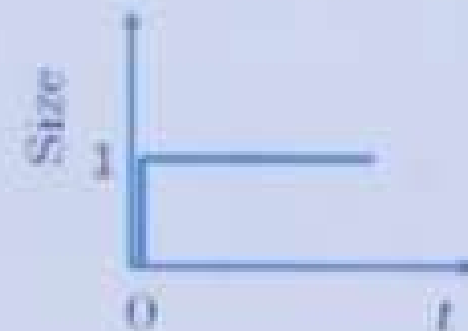
# Laplace Transforms

## Basic Laplace Transforms for Common Input

- Unit impulse at time  $t=0$  has a transform of 1
- A unit step signal defined by

$$\begin{aligned} 1(t) &= 0, \text{ for } t < 0 \\ &= 1, \text{ for } t > 0 \end{aligned}$$

- $\mathcal{L}[1(t)] = \int_0^{\infty} e^{-st} dt = \frac{1}{s}$



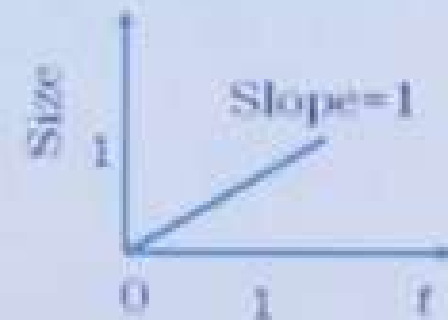
# Laplace Transforms

- Unit ramp function

$$f(t) = 0, \text{ for } t < 0,$$

$$= t, \text{ for } t \geq 0$$

$$\mathcal{L}[t] = \int_0^{\infty} t e^{-st} dt = \frac{1}{s^2}$$



- Unit amplitude sine wave signal

$$f(t) = 0, \text{ for } t < 0,$$

$$= \sin \omega t, \text{ for } t \geq 0$$

$$\mathcal{L}[\sin \omega t] = F(s) = \int_0^{\infty} \sin \omega t e^{-st} dt = \frac{\omega}{s^2 + \omega^2}$$

# Laplace Transforms

- Unit amplitude cosine wave signal

$$f(t) = 0, \text{ for } t < 0,$$

$$= \cos \omega t, \text{ for } t \geq 0$$

$$\mathcal{L}[\cos \omega t] = F(s) = \int_0^{\infty} \cos \omega t e^{-st} dt = \frac{s}{s^2 + \omega^2}$$

# Laplace Transforms

## Basic rules in working with Laplace transform

- $\mathcal{L}[Af(t)] = AF(s)$
- $\mathcal{L}[f_1(t) + f_2(t)] = F_1(s) + F_2(s)$
- Laplace transform of 1<sup>st</sup> derivative of a function is
- $\mathcal{L}\left[\frac{df(t)}{dt}\right] = sF(s) - f(0)$ , however with transfer function all initial values are taken to be zero.
- $\mathcal{L}\left[\frac{d^2f(t)}{dt^2}\right] = s^2F(s) - sf(0) - \frac{df(0)}{dt}$
- Laplace transform of an integral of a function is
- $\mathcal{L}\left[\int_0^t f(t)dt\right] = \frac{1}{s}F(s)$

# Inverse Laplace Transforms

<i>Laplace transform</i>	<i>Function of time</i>
1 $\frac{1}{s + a}$	$e^{-at}$
2 $\frac{a}{s(s + a)}$	$(1 - e^{-at})$
3 $\frac{b - a}{(s + a)(s + b)}$	$e^{-at} - e^{-bt}$
4 $\frac{s}{(s + a)^2}$	$(1 - at)e^{-at}$
5 $\frac{a}{s^2(s + a)}$	$t - \frac{1 - e^{-at}}{a}$



# Transfer function – 1<sup>st</sup> order system

$$a_1 \frac{dx}{dt} + a_0 x = b_0 y$$

$T_{\text{domain}} \Rightarrow S\text{-domain}$

where  $a_1, a_0, b_0$  are constants,  $y$  is the input and  $x$  the output, both being functions of time. The Laplace transform of this, with all initial conditions zero, is

$$a_1 s X(s) + a_0 X(s) = b_0 Y(s)$$

and so we can write the transfer function  $G(s)$  as

$$G(s) = \frac{X(s)}{Y(s)} = \frac{b_0}{a_1 s + a_0}$$

This can be rearranged to give

$$G(s) = \frac{b_0/a_0}{(a_1/a_0)s + 1} = \frac{G}{\tau s + 1}$$



$$TF = \frac{X(s)}{Y(s)}$$

$$\frac{b_0/a_0}{\frac{a_1}{a_0}s + \frac{a_0}{a_0}}$$



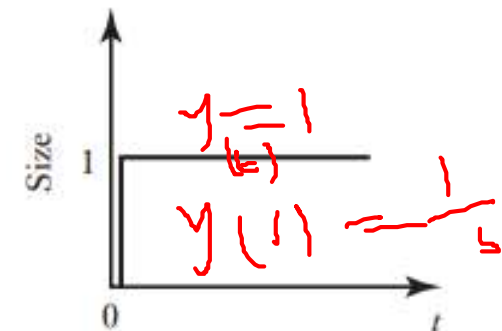
# Transfer function – 1<sup>st</sup> order system

When a first-order system is subject to a unit step input, then  $Y(s) = 1/s$  and the output transform  $X(s)$  is

$$\underline{X(s)} = G(s) Y(s) = \frac{G}{s(\tau s + 1)} = G \frac{(1/\tau)}{s(s + 1/\tau)}$$

Hence, since we have the transform in the form  $a/s(s + a)$ , using the inverse transformation listed as item 2 in the previous section gives

$$x = G(1 - e^{-t/\tau})$$



Unit step at zero time  
has the transform of  $1/s$

# First order system example-1

$$v = RC \frac{dv_C}{dt} + v_C$$

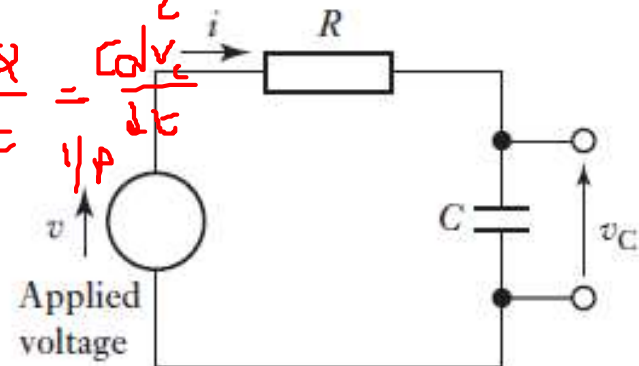
The Laplace transform is:

$$V(s) = RCsV_C(s) + V_C(s)$$

Thus  $V(s)$  is the Laplace transform of the input voltage  $v$  and  $V_C(s)$  is the Laplace transform of the output voltage  $v_C$ . Rearranging gives:

$$\frac{V_C(s)}{V(s)} = \frac{1}{RCs + 1}$$

The above equation thus describes the relationship between the input and output of the system when described as s functions.



$$IR + v_C + \frac{1}{C} \int I dt$$

$$Q = CV_C$$

$$I = \frac{dQ}{dt} = C \frac{dv_C}{dt}$$

$v_O$

$$(RCs + 1)V_C(s)$$





# Example 2 – Transfer Function

## ■ Problem

Obtain the transfer functions  $X(s)/F(s)$  and  $X(s)/G(s)$  for the following equation.

$$5\ddot{x} + 30\dot{x} + 40x = 6f(t) - 20g(t)$$

→ 1<sup>nd</sup>

## ■ Solution

Using the derivative property with zero initial conditions, we can immediately write the equation as

$$5s^2 X(s) + 30s X(s) + 40X(s) = 6F(s) - 20G(s)$$

Solve for  $X(s)$ .

$$X(s) = \frac{6}{5s^2 + 30s + 40} F(s) - \frac{20}{5s^2 + 30s + 40} G(s)$$

When there is more than one input, the transfer function for a specific input can be obtained by temporarily setting the other inputs equal to zero (this is another aspect of the superposition property of linear equations). Thus, we obtain

$$\frac{X(s)}{F(s)} = \frac{6}{5s^2 + 30s + 40} \quad \frac{X(s)}{G(s)} = -\frac{20}{5s^2 + 30s + 40}$$

# First order system problem

## Example

A thermocouple which has a transfer function linking its voltage output  $V$  and temperature input of:

$$G(s) = \frac{30 \times 10^{-6}}{10s + 1} \text{ V/}^{\circ}\text{C}$$

Determine the response of the system when it is suddenly immersed in a water bath at 100°C.

The output as an  $s$  function is:

$$V(s) = G(s) \times \text{input}(s)$$

The sudden immersion of the thermometer gives a step input of size 100°C and so the input as an  $s$  function is  $100/s$ . Thus:

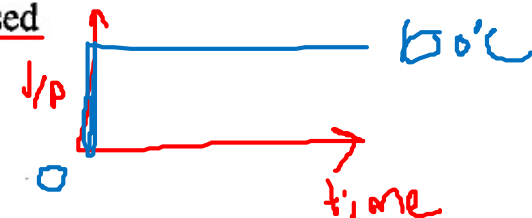
$$V(s) = \frac{30 \times 10^{-6}}{10s + 1} \times \frac{100}{s} = \frac{30 \times 10^{-4}}{10s(s + 0.1)} = 30 \times 10^{-4} \frac{0.1}{s(s + 0.1)}$$

The fraction element is of the form  $a/s(s + a)$  and so the output as a function of time is:

$$V = 30 \times 10^{-4} (1 - e^{-0.1t}) \text{ V}$$

$$\begin{aligned} 1/P &= T \\ 1/P &= \Delta V \end{aligned}$$

Hot



$$G(s) = \frac{V(s)}{T(s)}$$

$$V(s) = G(s) T(s)$$



# Second Order system

- For a 2<sup>nd</sup> order system, the relationship between input  $y$  and output  $x$  is given by differential equation of the form
- $a_2 \frac{d^2x}{dt^2} + a_1 \frac{dx}{dt} + a_0x = b_0y,$
- Here  $a_2, a_1, a_0$  and  $b_0$  are constants.
- Taking the Laplace transform of equation with all initial conditions as zero, gives
- $a_2s^2X(s) + a_1sX(s) + a_0X(s) = b_0Y(s),$

# Second Order system

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$$a_2 \frac{d^2x}{dt^2} + a_1 \frac{dx}{dt} + a_0x = b_0y$$

Taking Laplace transform,

$$a_2s^2X(s) + a_1sX(s) + a_0X(s) = b_0Y(s)$$

Hence

$$G(s) = \frac{X(s)}{Y(s)} = \frac{b_0}{a_2s^2 + a_1s + a_0}$$



# Second Order system

$$a_2 \frac{d^2x}{dt^2} + a_1 \frac{dx}{dt} + a_0x = b_0y \quad \text{And}$$

$$\omega_n^2 = \frac{a_0}{a_2} \text{ and } \zeta^2 = \frac{a_1^2}{4a_2a_0}$$

$$\frac{d^2x}{dt^2} + 2\zeta\omega_n \frac{dx}{dt} + \omega_n^2x = b_0\omega_n^2y$$

where  $\omega_n$  is the natural angular frequency with which the system oscillates and  $\zeta$  is the damping ratio. The Laplace transform of this equation gives

$$G(s) = \frac{X(s)}{Y(s)} = \frac{b_0\omega_n^2}{s^2 + 2\zeta\omega_ns + \omega_n^2}$$

The above are the general forms taken by the transfer function for a second-order system.



# Second Order system

When a second-order system is subject to a unit step input, i.e.  $Y(s) = 1/s$ , then the output transform is

$$X(s) = G(s)Y(s) = \frac{b_0\omega_n^2}{s(s^2 + 2\zeta\omega_n s + \omega_n^2)}$$

This can be rearranged as

$$X(s) = \frac{b_0\omega_n^2}{s(s + p_1)(s + p_2)}$$

where  $p_1$  and  $p_2$  are the roots of the equation

$$s^2 + 2\zeta\omega_n s + \omega_n^2 = 0$$

Hence, using the equation for the roots of a quadratic equation,

$$p = \frac{-2\zeta\omega_n \pm \sqrt{4\zeta^2\omega_n^2 - 4\omega_n^2}}{2}$$

and so the two roots  $p_1$  and  $p_2$  are

$$p_1 = -\zeta\omega_n + \omega_n\sqrt{\zeta^2 - 1} \quad p_2 = -\zeta\omega_n - \omega_n\sqrt{\zeta^2 - 1}$$

# Second Order system

$\zeta > 1$  **Over damped**

$$x = \frac{b_0 \omega_n^2}{p_1 p_2} \left[ 1 - \frac{p_2}{p_2 - p_1} e^{-p_2 t} + \frac{p_1}{p_2 - p_1} e^{-p_1 t} \right]$$

$\zeta = 1$  **Critically damped**

$$x = b_0 [1 - e^{-\omega_n t} - \omega_n t e^{-\omega_n t}]$$

$\zeta < 1$  **Under damped**

With  $\zeta < 1$ , then

$$x = b_0 \left[ 1 - \frac{e^{-\zeta \omega_n t}}{\sqrt{1 - \zeta^2}} \sin(\omega_n \sqrt{1 - \zeta^2} t + \phi) \right]$$

where  $\cos \phi = \zeta$ . This is an under-damped oscillation.





# Example- Second Order system

What will be the state of damping of a system having the following transfer function and subject to a unit step input?

$$G(s) = \frac{1}{s^2 + 8s + 16}$$

The output  $Y(s)$  from such a system is given by:

$$Y(s) = G(s)X(s)$$

For a unit step input  $X(s) = 1/s$  and so the output is given by:

$$Y(s) = \frac{1}{s(s^2 + 8s + 16)} = \frac{1}{s(s + 4)(s + 4)}$$

The roots of  $s^2 + 8s + 16$  are  $p_1 = p_2 = -4$ . Both the roots are real and the same and so the system is critically damped.





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# Thank you