

## Lecture #15

$$M^T = M \in \mathbb{R}^{n \times n}$$

Positive definite if  $\forall z \in \mathbb{R}^n$ , we get

semi

$$\underbrace{z^T M z}_{\text{---} \boxed{\quad} \text{---}} \geq 0 \Leftrightarrow M \geq 0$$

$$\text{eig}(M) = \lambda_1, \dots, \lambda_n \geq 0$$

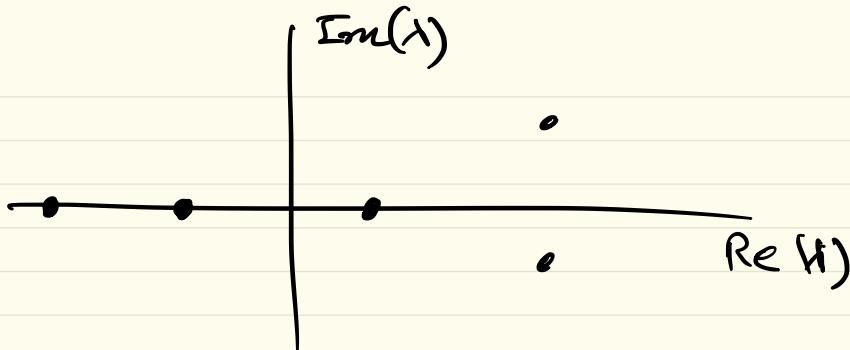
$$\begin{pmatrix} 1 \\ 2 \\ -3 \end{pmatrix}$$

$$\begin{matrix} M_1 & \xrightarrow{\quad} & M_2 \\ M_1 - M_2 & \xrightarrow{\quad} & 0 \end{matrix}$$

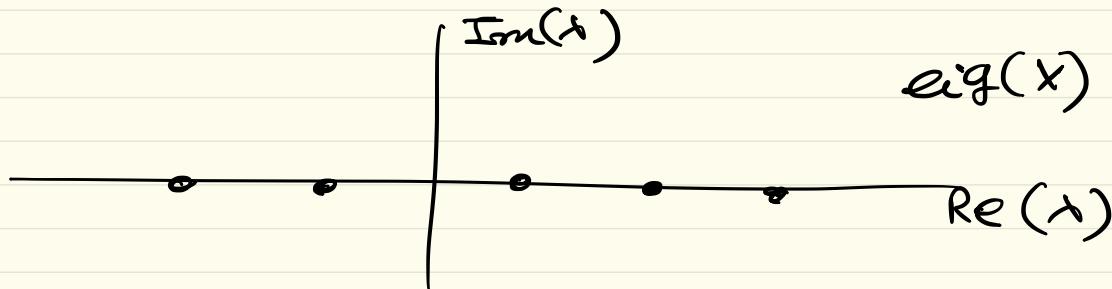
$$z^T (M_1 - M_2) z \geq 0$$

Transformation of matrices:

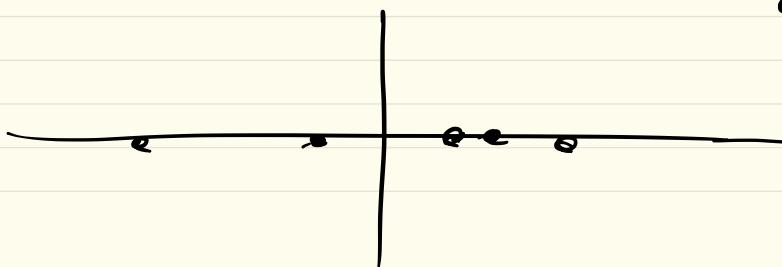
- $X \mapsto T X T^{-1}$ ,  $T$  is non-singular  
 $\text{eig}(X) = \text{eig}(T X T^{-1})$  : Similarity transformation
- $X \mapsto T X T^T$  : Congruence transformation



Symmetric  $X$ ,  $\text{eig}(X)$  are reals



$\text{eig}(TXT^T)$



# Sliding Mode Control / Variable Structure Control

Example:

$$\underline{x} \in \mathbb{R}^2$$

$$\begin{cases} \dot{x}_1 = x_2 \\ \dot{x}_2 = f(\underline{x}) + g(\underline{x}) u \end{cases}$$

State

where  $f$  &  $g$  are unknown nonlinear functions

want to design a feedback  $u(\cdot)$  to stabilize origin of closed-loop // &  $g(x) \geq g_o > 0$  &  $\underline{x} \in \mathbb{R}^n$

Let  $x_2 = -a_1 x_1$

$$s := a_1 x_1 + x_2 = 0$$

manifold

On this manifold:

$$\dot{x}_1 = (x_2) = -a_1 x_1$$

$$\Leftrightarrow \begin{pmatrix} x^{(n)} \\ \dot{x}_1 \\ \dot{x}_2 \\ \vdots \\ \dot{x}_n \end{pmatrix} = \begin{pmatrix} x_2 \\ x_3 \\ \vdots \\ x_n \\ f(\underline{x}) + g(\underline{x}) u \end{pmatrix}$$

$$\text{Manifold } s := \{ \underline{x} \in \mathbb{R}^2 : a_1 x_1 + x_2 = 0 \}$$

Choosing  $a_1 > 0$  will ensure "sliding"

And choosing  $|a_1|$ , will allow us to vary the rate-of-convergence.

NOTE: The motion dynamics of  $\delta$  is independent of  $f$  &  $g$ .

Also: the variable " $\delta$ " satisfies

$$\dot{\delta} = a_1 \dot{x}_1 + \dot{x}_2 = a_1 x_2 + f(x) + g(x)$$

Suppose  $f$  &  $g$  satisfy the inequalities:

$$\left| \frac{a_1 x_2 + f(x)}{g(x)} \right| \leq P(x) \quad \forall x \in \mathbb{R}^2$$

for some known function  $P(\cdot)$

Now choose  $V = \frac{1}{2} \delta^2$  for dynamics

$$\dot{\delta} = a_1 x_2 + f(x) + g(x)$$

$$\begin{aligned}
 \text{Then } \dot{V} &= \dot{s}\dot{s} \\
 &= s(a_1 x_2 + f(x)) + g(x)s u \\
 &\leq |g(x)|s|P(x)| + g(x)s u
 \end{aligned}$$

(assumption  
inequality)

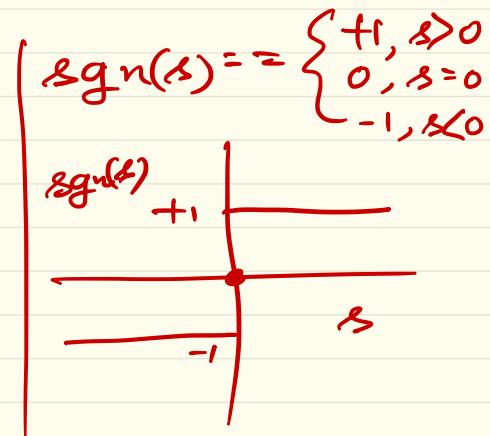
Take  $u = -\beta(x) \operatorname{sgn}(s)$

where

$$\beta(x) \geq P(x) + \beta_0, \quad \beta_0 > 0$$

This gives

$$\begin{aligned}
 \dot{V} &\leq \underbrace{|g(x)|s|P(x)| - g(x)(P(x) + \beta_0)s \cdot \operatorname{sgn}(s)}_{= -g(x)\beta_0|s|} \\
 &\leq -g_0\beta_0|s| \quad (\because g(x) \geq g_0 > 0)
 \end{aligned}$$



$\Rightarrow \dot{s} \leq -g_0 \beta_0 |s|$  (Differential inequality)

$\Rightarrow D^+ |s| \leq -g_0 \beta_0$

upper right hand derivative.

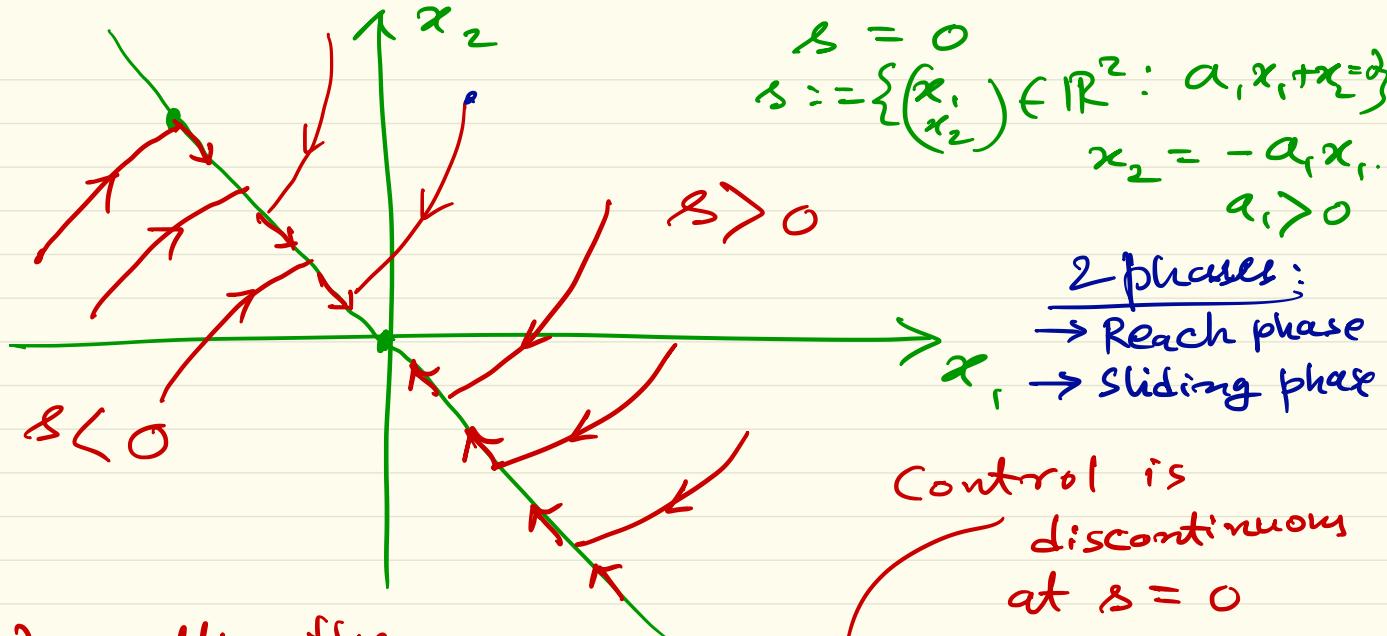
$$D^+ v(t) := \limsup_{h \rightarrow 0^+} \frac{v(t+h) - v(t)}{h}$$

If  $v(t)$  is differentiable,  $D^+ = \frac{dv}{dt}$

$\Rightarrow$  (Comparison Lemma) ↗

$\Rightarrow$  Trajectory  $s(t)$  reaches  $s=0$  (our manifold)

Already shown,  
once on  $s=0$ , then cannot leave if ( $\dot{V} \leq -g_0 \beta_0 |s|$ )  
in finite time



We call this

control  $u = -\beta(x) \operatorname{sgn}(s)$  as

"sliding mode control"

Difficulty/Con: Chattering

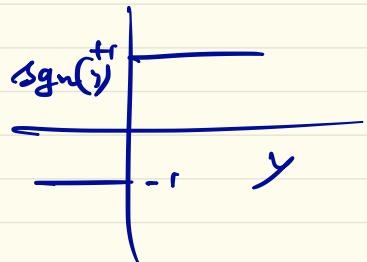
(Discretization, Delay)

Pro:  
Sliding phase is indep. of  $f$  &  $g$ .

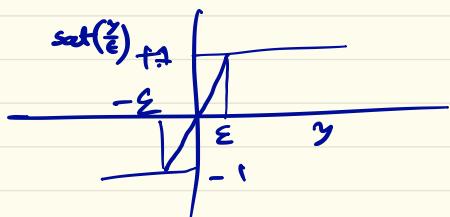


approx. control law  
(sliding mode)

$$u = -\beta(x) \operatorname{sgn}(s)$$
$$\approx -\beta(x) \operatorname{sat}\left(\frac{s}{\epsilon}\right)$$



where  $\operatorname{sat}(y) := \begin{cases} y & \text{if } y \leq 1 \\ \operatorname{sgn}(y) & \text{if } |y| > 1 \end{cases}$



## Backstepping

### Integrator Backstepping (Single Input)

$$\begin{aligned} \dot{\eta} &= f(\eta) + g(\eta) \xi \\ \dot{\xi} &= u \end{aligned} \quad \left. \begin{array}{l} \eta \in \mathbb{R}^n \\ \xi \in \mathbb{R} \\ u \in \mathbb{R} \end{array} \right\} \quad \left. \begin{array}{l} \text{state} \\ \{\eta\} \in \mathbb{R}^{n+1} \\ \{\xi\} \in \mathbb{R} \end{array} \right\}$$

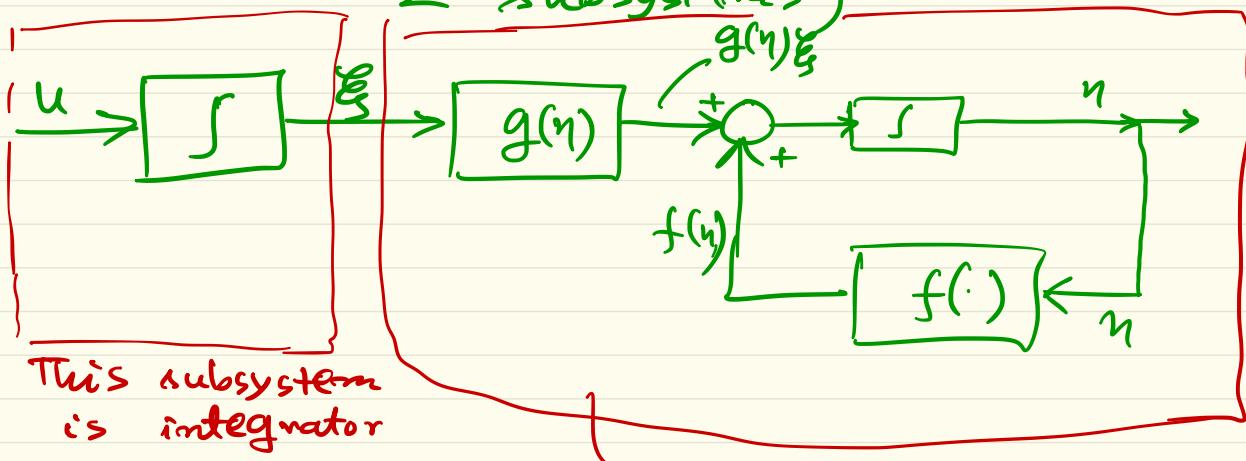
$$f: \mathcal{D} \mapsto \mathbb{R}^n, f(0) = 0$$

$$g: \mathcal{D} \mapsto \mathbb{R}$$

Objective: Design state feedback law  
to stabilize origin in  $\mathbb{R}^{n+1}$ ,  $\begin{pmatrix} \eta \\ \xi \end{pmatrix} = \begin{pmatrix} 0 \\ 0 \end{pmatrix}$

Assumption: Both  $f$  &  $g$  are known.

Diagram: (Can be thought of as cascade of  
2 subsystems)



This subsystem  
is integrator

$$\xi = u$$

This subsystem "thinks"  
 $\xi$  is its input

$$(\underline{\eta} = \underline{f}(\underline{\eta}) + \underline{g}(\underline{\eta})\xi)$$

Suppose, I can stabilize this subsystem  
by some feedback law  $\xi = \phi(\eta)$

$$\underline{\eta} = \underline{f}(\underline{\eta}) + \underline{g}(\underline{\eta})\phi(\underline{\eta}) \text{ with } \phi(0) = 0$$

$\eta = f(\eta) + g(\eta)\phi(\eta)$  is A.S.

Suppose further we know  $V(\eta)$  which is  $C^1$ , p. d. s.t.

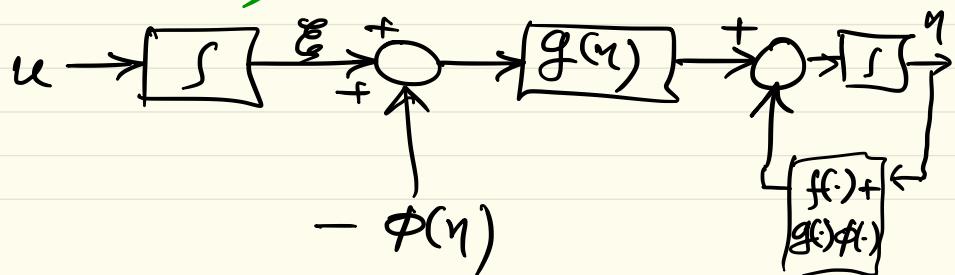
$$\left\langle \underbrace{\nabla_{\underline{\eta}} V}_{\substack{= \\ \frac{\partial V}{\partial \underline{\eta}}}} , f(\underline{\eta}) + g(\underline{\eta}) \phi(\underline{\eta}) \right\rangle \leq -W(\underline{\eta}) \quad \forall \underline{\eta} \in \partial$$

p. d.

Now let's add & subtract  $g(\eta) \phi(\eta)$  to the RHS of  $\dot{\underline{\eta}} = f(\underline{\eta}) + g(\underline{\eta}) \underline{\xi}$  to get

$$\dot{\underline{\eta}} = (f(\underline{\eta}) + g(\underline{\eta}) \phi(\underline{\eta})) + g(\underline{\eta})(\underline{\xi} - \phi(\underline{\eta}))$$

$$\dot{\underline{\xi}} = u$$



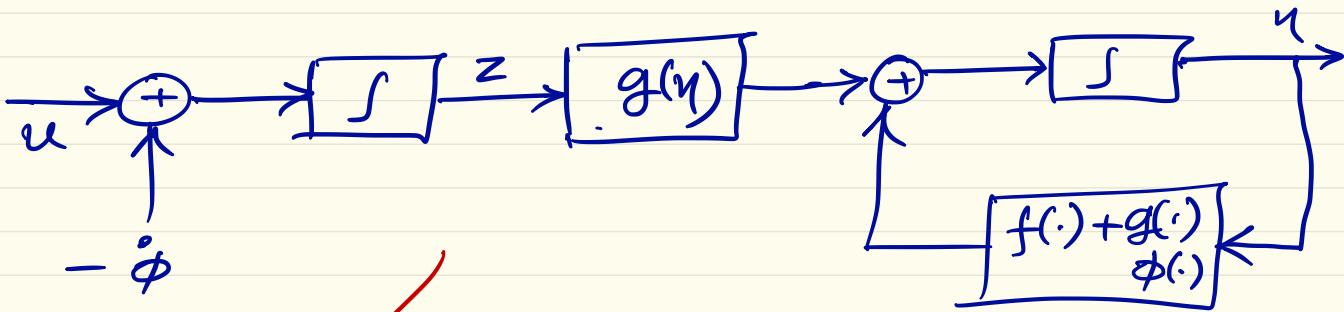
Introduce a change-of-variable:

$$z := \xi - \phi(\eta)$$

This gives

$$\dot{\eta} = (f(\eta) + g(\eta)\phi(\eta)) + g(\eta)z$$

$$\dot{z} = u - \dot{\phi}$$



prev picture  $\rightarrow$  This picture is "backstepping"  
 $-\dot{\phi}(\eta)$  through the integrator

We know  $f, g, \phi$ :

$\therefore$  we can compute  $\dot{\phi} = \frac{\partial \phi}{\partial \eta} (f(\eta) + g(\eta) \xi)$

Let  $v := u - \dot{\phi}$ , then the system reduces to

$$\begin{cases} \dot{\eta} = (f(\eta) + g(\eta) \phi(\eta)) + g(\eta) z \\ \dot{z} = v \end{cases}$$

Similar to what we started with,  
except the first component now has  
A.S. origin when input = 0

Idea: Exploit this to design  $\mathcal{D}$  to stabilize  
overall system.

we:  $V_e(\eta, \xi) = V(\eta) + \frac{1}{2} z^2$

$$\text{Then } \dot{V}_c = \frac{\partial V}{\partial \eta} \left( f(\eta) + g(\eta) \phi(\eta) \right) + \frac{\partial V}{\partial \eta} g(\eta) z + z v$$

$$\leq -W(\eta) + \frac{\partial V}{\partial \eta} g(\eta) z + z v$$

$$\text{Choose: } v = -\frac{\partial V}{\partial \eta} g(\eta) - kz, \quad k > 0$$

$$\text{This makes: } \dot{V}_c \leq -W(\eta) - kz^2$$

$$\Rightarrow \begin{pmatrix} \eta \\ \xi \end{pmatrix} = \begin{pmatrix} 0 \\ 0 \end{pmatrix} \text{ is A.S. } (\because \phi(0) = 0 \text{ LaSalle etc.})$$

Sub-back  $v, z, \dot{\phi}, +_o$  get original  
feedback control:

$$u = \frac{\partial \phi}{\partial \eta} (f(\eta) + g(\eta) \xi) - \frac{\partial V}{\partial \eta} g(\eta) - k(\xi - \phi(\eta))$$