



**System Reference document (SRdoc);
Technical characteristics for Low Power Wide Area Networks
Chirp Spread Spectrum (LPWAN-CSS)
operating in the UHF spectrum below 1 GHz**

Reference

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Contents

Intellectual Property Rights	5
Foreword.....	5
Modal verbs terminology	5
Executive summary	5
Introduction	5
1 Scope	6
2 References	6
2.1 Normative references	6
2.2 Informative references	6
3 Definitions, symbols and abbreviations	7
3.1 Definitions	7
3.2 Symbols	7
3.3 Abbreviations.....	7
4 Comments on the System Reference Document	9
4.1 Statements by ETSI Members	9
5 Presentation of the system or technology.....	12
5.1 Overview	12
5.2 The LoRaWAN™ protocol.....	13
5.2.1 Overview of the protocol.....	13
5.2.2 End Node.....	13
5.2.3 Gateway	14
5.2.4 Network Server	15
5.2.5 Adaptive Data Rate	15
5.2.6 Application Server	16
6 Market information	16
7 Technical information	18
7.1 Detailed technical description.....	18
7.1.1 LPWA-CSS signals.....	18
7.2 Technical parameters and implications on spectrum	20
7.2.1 General technical parameters	20
7.2.2 Status of technical parameters.....	23
7.2.2.1 Current ITU and European Common Allocations	23
7.2.2.2 Sharing and compatibility studies (if any) already available	23
7.2.2.3 Sharing and compatibility issues still to be considered	23
7.2.3 Transmitter parameters.....	23
7.2.3.1 Transmitter Output Power/Radiated Power.....	23
7.2.3.1.0 Introduction.....	23
7.2.3.1.1 Antenna Characteristics	24
7.2.3.2 Operating Frequency	24
7.2.3.3 Out of band emissions	24
7.2.3.4 Spurious emissions	25
7.2.4 Receiver parameters	27
7.2.4.0 Wanted signal level for degradation measurements	27
7.2.4.1 Sensitivity	27
7.2.4.2 Adjacent channel rejection	28
7.2.4.3 Blocking	28
7.2.4.4 Intermodulation response rejection.....	31
7.2.5 Channel access parameters.....	31
7.3 Information on relevant standard(s).....	34
8 Radio spectrum request and justification	34

9	Regulations.....	34
9.1	Current regulations	34
9.2	Proposed regulation and justification.....	36
Annex A:	Main use cases in different verticals.....	37
Annex B:	Interference Experimental setup	43
B.1	General requirements	43
B.2	Receiver's AWGN sensitivity.....	44
B.3	Continuous wave interference.....	45
B.4	GFSK modulated interferer.....	45
B.5	LPWAN-CSS modulated Interferer	47
Annex C:	An example of interference measurement with a single LPWAN CSS link and a single periodic pulsed CW interferer	49
Annex D:	Out of band emission measurements.....	50
Annex E:	Change History	53
History		54

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Foreword

This Technical Report (TR) has been produced by ETSI Technical Committee Electromagnetic compatibility and Radio spectrum Matters (ERM).

Modal verbs terminology

In the present document "**should**", "**should not**", "**may**", "**need not**", "**will**", "**will not**", "**can**" and "**cannot**" are to be interpreted as described in clause 3.2 of the [ETSI Drafting Rules](#) (Verbal forms for the expression of provisions).

"**must**" and "**must not**" are **NOT** allowed in ETSI deliverables except when used in direct citation.

Executive summary

The present document has been developed on request of CEPT to enable them to conduct compatibility studies on LPWAN-CSS (LoRaWAN™) systems.

The document contains information on the technical characteristics and parameters, as well as market relevant information on the LPWAN-CSS systems.

The document includes finally spectrum considerations to enable the market success of the LPWAN-CSS (LoRaWAN™) systems.

Introduction

The present document has been developed on request of CEPT WGFM to get a better description of LPWAN systems in the UHF frequency band.

1 Scope

The present document describes the LPWAN-CSS (Low Power Wide Area Networks - Chirp Spread Spectrum) system, and aims to respond a CEPT ECC Working Group Frequency Management request to better understand the LPWAN-CSS characteristics in view of allowing spectrum considerations for conventional SRDs and SRD networks healthy sharing. It includes in particular:

- Market information.
- Technical information (including expected sharing and compatibility issues).
- Regulatory considerations.

2 References

2.1 Normative references

Normative references are not applicable in the present document.

2.2 Informative references

References are either specific (identified by date of publication and/or edition number or version number) or non-specific. For specific references, only the cited version applies. For non-specific references, the latest version of the referenced document (including any amendments) applies.

NOTE: While any hyperlinks included in this clause were valid at the time of publication, ETSI cannot guarantee their long term validity.

The following referenced documents are not necessary for the application of the present document but they assist the user with regard to a particular subject area.

[i.1] "LoRaWAN™ 1.0.2 Specification".

NOTE: To obtain this document please contact admin@mail.lora-alliance.org.

[i.2] Hata, M. (August 1980): "Empirical Formula for Propagation Loss in Land Mobile Radio Services". IEEE Transactions on Vehicular Technology. VT-29 (3): 317-25.

[i.3] B. Reynders and S. Pollin: "Chirp spread spectrum as a modulation technique for long range communication", Symposium on Communications and Vehicular Technologies (SCVT), Mons, Belgium, November 2016, pp. 1-5.

[i.4] ETSI EN 300 220-1 (V3.1.1): "Short Range Devices (SRD) operating in the frequency range 25 MHz to 1 000 MHz; Part 1: Technical characteristics and methods of measurement".

[i.5] ETSI EN 300 220-2 (V3.1.1): "Short Range Devices (SRD) operating in the frequency range 25 MHz to 1 000 MHz; Part 2: Harmonised Standard covering the essential requirements of article 3.2 of Directive 2014/53/EU for non specific radio equipment".

[i.6] ERC Recommendation 70-03: "Relating to the Use of Short Range Devices (SRD)", Tromsø 1997 Subsequent amendments 19 May 2017.

[i.7] Commission Implementing Decision (EU) 2017/1483 of 8 August 2017 amending Decision 2006/771/EC on harmonisation of the radio spectrum for use by short-range devices and repealing Decision 2006/804/EC.

[i.8] ECC Report 261 (January 2017): "Short Range Devices in the frequency range 862-870 MHz range".

- [i.9] ECC Report 246 (January 2017): "Wideband and Higher DC Short Range Devices in 870-875.8 MHz and 915.2-920.8 MHz (companion to ECC Report 200)".
- [i.10] ETSI TR 103 435 (V1.1.1): "System Reference document (SRdoc); Short Range Devices (SRD); Technical characteristics for Ultra Narrow Band (UNB) SRDs operating in the UHF spectrum below 1 GHz".
- [i.11] Directive 2014/53/EU of the European Parliament and of the Council of 16 April 2014 on the harmonisation of the laws of the Member States relating to the making available on the market of radio equipment and repealing Directive 1999/5/EC Text with EEA relevance.
- [i.12] ERC Recommendation 74-01: "Unwanted Emissions in the Spurious Domain".

3 Definitions, symbols and abbreviations

3.1 Definitions

For the purposes of the present document, the terms and definitions given in ETSI EN 300 220-1 [i.4], ETSI EN 300 220-2 [i.5] and the following apply:

application server: terminates the application layer for the end devices connected to a Network Server; there can be multiple instances of the Application Server, each one serving a different application or a different group of applications

end-device or end-node: LoRaWAN™ client device communicating via a radio link with gateways; the correspondent ETSI/CEPT term is "Terminal Node (TN)"

gateway: radio system on the infrastructure-side. Communicates with end-devices and, via IP, with a network server; the correspondent ETSI/CEPT term is "Network Access Point (NAP)"

network server: termination entity for the LoRaWAN™ protocol for the end-devices connected to the network, it is the centre of the star topology

occupied bandwidth: width of a LPWAN CSS signal band such that, below the lower and above the upper frequency limits, the mean power emitted are each equal to 0,5 % of the total mean power of a given emission

time overhead: time taken to transmit everything else which is not payload

3.2 Symbols

For the purposes of the present document, the symbols given in ETSI EN 300 220-1 [i.4], ETSI EN 300 220-2 [i.5] and the following apply:

bps	bits per second
dB	decibel
dBc	decibels of the power referenced to the power of the carrier
dBd	decibels of antenna gain referenced to a half-wave dipole antenna
dBi	decibels of antenna gain referenced to a hypothetical isotropic antenna
dBm	decibels of the power referenced to one milliwatt

3.3 Abbreviations

For the purposes of the present document, the following abbreviations apply:

ACR	Adjacent Channel Rejection
ADR	Adaptive Data Rate
AFA	Adaptive Frequency Agility
ARM	Advanced Risc Machine (company)
AS	Application Server
AWGN	Additive White Gaussian Noise

BW	Bandwidth
CAPEX	Capital Expenditure
CCA	Clear Channel Assessment
CEPT	Conférence Européenne des administrations des Postes et des Télécommunications
CRC	Cyclic Redundancy Check
CSS	Chirp Spread Spectrum
CW	Continuous Wave
DR	Data Rate
E.R.P.	Effective Radiated Power
EC	European Commission
ECC	Electronic Communications Committee
ED	End Device
ERC	European Radiocommunication Committee
ER-GSM	Enhanced Railway Global System for Mobile communication
ERP	Effective Radiated Power
EU	European Union
FEC	Forward Error Correction
FSK	Frequency Shift Keying
GPS	Global Positioning system
GW	GateWay
HAL	Hardware Abstraction Layer
IoT	Internet of Things
IP	Internet Protocol
LBT	Listen Before Talk
LPWA	Low Power Wide Area
LPWAN	Low Power Wide Area Networks
LP-WAN	Low Power Wide Area Networks
LPWAN-CSS	LPWAN Chirp Spread Spectrum
LTE	Long Term Evolution
MAC	Medium Access Control
NF	Noise Figure
NS	Network Server
OOB	Out Of Band
PER	Paket Error Rate
PHY	Physical Layer
REC	Recommendation
RF	Radio Frequency
RX	Receiver
SA	Société Anonyme
SAW	Surface Acoustic Wave
SE	Spectrum Engineering
SF	Spreading Factor
SNR	Signal to Noise Ratio
SRD	Short Range Devices
SSN	Silver Spring Networks
TV	Television
UHF	Ultra High Frequency
UK	United Kingdom
UNB	Ultra Narrow Band
USD	USA Dollar
WG	Working Group
WGFM	Working Group Frequency Management
WGSE	Working Group Spectrum Engineering
WI	Work Item

4 Comments on the System Reference Document

4.1 Statements by ETSI Members

Source: Silver Spring Networks (UK) Ltd.

NOTE 1: For further information contact Dr. Simon Dunkley sdunkley@ssni.com.

SSN (UK), a Full ETSI Member, actively participated in the preparation of the present document and thanks Semtech for the depth and scope of the information contributed on the operation of the LoRa systems described. While not raising opposition to the completion of the present document, a number of points where SSN (UK) opinions disagree with the authors are noted here in this clause to be considered along with the rest of the present document material.

1) Relationship between LoRa™ and LoRaWAN™:

Throughout the present document preparation the Rapporteur and the contributors were asked to clarify the relationship between LoRa™ and LoRaWAN™. Satisfactory answers were never provided and the final draft provides no explanation of the relationship, instead simply removing (almost) all references to LoRa™ from the document. SSN believe this does not respond to the request from WGFM for the present document, is not helpful in understanding the products and market for these systems, and merely confuses the reader.

Instead of inventing a new term (LPWAN-CSS) and trying to use it to describe LPWAN systems built using the LoRaWAN™ protocols, SSN (UK) believes it would be much simpler and clearer to refer to published and freely available material from the LoRa Alliance™ website. The document entitled "LoRaWANTM What is it?", and subtitled "A technical overview of LoRa™ and LoRaWAN™", linked to via the home page of the LoRa Alliance™ website, clearly states that:

- "LoRa™ is the physical layer or modulation utilized to create the long range communication link".

It further states that:

- "LoRaWAN™ is the communications protocol and systems architecture for the network while the LoRa™ physical layer enables the long range communications link".

These statements are clear and unambiguous and match the LoRa™ Protocol Architecture depicted in Figure 2 where the LoRaWAN™ protocol is shown adapted via a HAL to the PHY layer shown exchanging LoRa™ or FSK modulated data with a peer PHY layer.

The attempt to replace LoRa™ with the term LPWAN-CSS while at the same time equating LPWAN-CSS to LoRaWAN™ is inconsistent within the present document and with the published technology descriptions by the LoRa Alliance™ and simply introduces confusion.

2) Relationship to studies carried out in ECC Report 261 [i.8]:

SSN (UK) actively participated in CEPT WI 42-2 compatibility studies on SRDs in the 862 MHz - 870 MHz frequency range. ECC Report 261 [i.8] included results of studies of interference into LPWAN base stations but using assumed parameters for the LPWAN receiver.

CEPT WGSE, at their 75th meeting in Berlin, January 2017, discussed ECC Report 261 [i.8] (among others) and minuted the decision to approve the report for publication. However, it was also noted that the studies on LPWAN used unrealistic system parameters. Section 9.1.1 of SE(17)035 Minutes of the 75th WGSE meeting, reporting the discussion on ECC Report 261 [i.8] states:

- "United Kingdom (SE(17)032) suggested further that the analysis of interference into LP-WAN technology should be revisited once the details of the LP-WAN technology have been established and a realistic set of victim system parameters agreed. United Kingdom is of the opinion that WG SE should commit to a new WI once the present document on LP-WAN is received from ETSI".

Requests by SSN (UK) to comment in the present document on the relationship between the presented system parameters and those reported in ECC Report 261 [i.8] assumed for LPWAN base stations were consistently refused on the basis that such comments were out of scope of the present document. SSN (UK) considers that, since WGMF specifically requested the present document because of a lack of technical detail for LoRaTM LPWAN systems to be used in compatibility studies, such comments would be fully in scope of the present document and be of considerable interest to CEPT.

3) Interference from LoRaTM Systems:

SSN (UK) notes that LPWAN systems are described as operating at both low and high power with Gateways often mounted at high elevations. SSN (UK) considers that, in addition to the usual compatibility investigations, studies of high power operation at high elevation, especially taking into account long continuous transmissions owing to low data rates, are required to determine the interference from LPWAN systems into other legitimate SRD devices.

4) LoRaTM System Waveform:

SSN (UK) repeatedly requested more details of the LoRaTM system waveform and its modulation but these requests were denied on the basis that sufficient information is presented for the needs of CEPT compatibility studies. SSN (UK) believes this is not the case and that when CEPT studies are launched the need for additional information will become apparent. Not providing information in the present document will simply lead to delays in the overall process.

SSN (UK) requests details of the modulation scheme used by LoRaTM systems so that when considering interference into LoRaTM devices the effects of the interferer signal on the LoRaTM receiver and demodulation mechanisms may be understood. Similarly, knowledge of the modulation process will assist in understanding the effects of LoRaTM waveforms as interferers into different victim devices.

In addition, it is currently impossible to verify some of the information presented since details of the modulation scheme are not available. For example, deriving the processing gain or even the bit rate from the Spreading Factor information is not possible.

5) Susceptibility to Interference:

The final figures presented for Adjacent Channel Rejection and Blocking are comparable to the performance reported for UNB devices, with ACR being worse by approximately 13 dB (47 dB vs 60 dB for UNB) while blocking is of the same order (approximately 90 dB). However, the co-channel performance is poor with at best a few dB of tolerance to in-band interferer signals.

Since the SRDs bands are by definition shared bands, with a wide range of waveforms permitted, such weak in-band rejection will lead to questionable real-world performance unless there are other means to recover the wanted signal which are not reported in the present document.

6) Time Domain Behaviour:

SSN (UK) notes that the LoRa AllianceTM, and LoRaTM device suppliers, stress the LoRaTM modulation resilience to interference, describing interleaving and FEC properties to mitigate typical bursty interference expected in the bands targeted for deployment, enabling data to be recovered provided a minimum number of symbols are received.

Unfortunately none of these properties are explored in the present document which only states a single fixed (5/4) Parity code for FEC (clause 7.1.1), and a one page Annex C referred to once in clause 7.2.4.2 on Adjacent Channel Rejection - even though the Annex C details tolerance to in-band pulsed interferer.

SSN (UK) considers it highly relevant to include receiver performance in real sharing scenarios when such features are present in the new system since CW interferers do not represent active transmissions of real SRDs sharing the spectrum. Discontinuous transmissions are realistic models for packet communications which form the vast majority of SRD systems in the concerned bands, and systems are designed to withstand multiple bit-errors characteristic of bursty interference in shared frequency bands.

Annex C provides a tantalising insight into potential behaviour of the LoRaTM modulation scheme but unfortunately, in the absence of details of the modulation/de-modulation process, it is difficult to extract more useful information pertinent to sharing studies which may need to include Time Domain behaviour in order to accurately represent the behaviour of these new innovative systems.

SSN (UK) notes that recent work in ETSI TG28 and relevant STFs has developed significantly improved mechanisms to specify short term and long term behaviour expressed as Duty Cycles which provide powerful tools for describing sharing scenarios deemed workable by CEPT compatibility studies.

By ETSI Member Great Circle Design

NOTE 2: Contact: Nicholas Long.

This comment on the present document relates to the following subject areas:

- 1) Lack of information about the signal.
- 2) Traffic Model.
- 3) Capacity constraints.
- 4) Long range performance in practice.
- 5) Compatibility studies.

Lack of information

Despite repeated requests, the authors of the present document have not included sufficient information about the signal transmitted. The missing information includes basic things such as:

- Number of bits per symbol.
- Means of modulating the data onto the signal.
- Spreading ratio (the Spreading Factor quoted is not the spreading ratio as normally understood in spread spectrum systems).

Also, the term LoRa is in general use and is heavily promoted. It would be useful if the document explained the relationship between LoRa and LoraWAN and between LoRa and LPWAN-CSS.

It is therefore unclear not only what the PHY layer is, but what is actually being described or claimed in the present document.

Traffic Model

The present document would benefit from an analysis of the predicted traffic expected on a LoRa/LoRaWAN channel.

There is a clue buried in Table 7, where it is shown that the duty cycle of an end node can be 0,016 % or 0,0045 %. These are remarkably precise figures and an explanation and analysis would be beneficial.

Capacity Constraints

Also missing is a discussion of the traffic capacity of a LoRa channel.

It appears that each signal in a 200 kHz channel has to be on a different SF (Spreading Factor). Since each SF corresponds to a different data rate and is optimized for a different range, it is expected that only one signal on the channel can be optimized and the others will have to run at lower than possible data rates.

It appears the maximum channel capacity can be found by adding up the payload data rates in Table 2 for SF=7 to SF=12:

$$5\,500 + 3\,100 + 1\,800 + 980 + 440 + 250 = 12,070 \text{ bits/sec}$$

This is a remarkably low capacity for a 200 kHz channel. Indeed, it would seem to make more sense to turn off the chirp spread spectrum option and just use the FSK mode that allows 50 kbps.

Long range performance in practice

One of the seeming advantages of LoRa/LorAWAN is the ability to function as both a long range and a short range system; the SF can be chosen to suit the distance required.

What is not clear is how well this will work in practice. Information that was in an earlier draft but appears to be missing from the current one was a table of isolation, or separation, between the various SFs. This showed quite low levels of isolation; indeed the theory of chirp spread spectrum shows that they are not necessarily orthogonal data streams.

LoRa/LoRaWAN therefore suffers from a severe case of the near-far problem, in that separate data streams cannot be received unless they are close in amplitude.

Once the traffic gets above a certain level, long range communication becomes impossible as it is blotted out by the short range traffic. With the current lack of information in the present document it is not possible to say at what level of usage this happens or whether this happens suddenly or gradually.

In this respect, however, the comment on p.30 in clause 7.2.1 of the present document about "network densification" is revealing: *...a network operator seeing an increasing number of connections will increase the gateways density....*

This would confirm the view that once a certain level of usage is reached, then long range operation has to be abandoned. Table 7 gives a figure for gateway density of 3,5 per sq. km, which would correspond with a maximum range of around 700 m.

Compatibility Studies

Compatibility studies with LoRa/LoRaWAN as victim and interferer are welcomed. It is essential, however, to take account of the following:

- 1) Studies with LoRa/LoRaWAN as victim should be representative of actual use. For instance, it would be incorrect to analyse a situation with a single LoRa/LoRaWAN signal against a thermal noise background. It will be necessary to consider the other LoRa/LoRaWAN signals on the same channel. Regrettably, this will require additional information to that in the present document.
- 2) The present document describes LoRa/LoRaWAN as a Short Range Device, operating under general authorization. The rules and conventions for SRDs should be respected. A prediction of potential interference between SRDs is not necessarily a reason for regulatory intervention.

5 Presentation of the system or technology

5.1 Overview

The LPWAN CSS system is mainly intended for applications which fall under the broad field of Internet of Things. There are many vertical applications which are part of this field and they are highlighted in Annex A.

The LPWAN CSS system comprises four entities: the end-device (also called end-node), the Gateway, the Network Server and the Application Server. The architecture describing the relationship between these entities is depicted in Figure 1. The Gateway, the Network Server and the Application Server are part of the infrastructure of the network.

The LoRaWAN™ protocol architecture is shown in Figure 2.

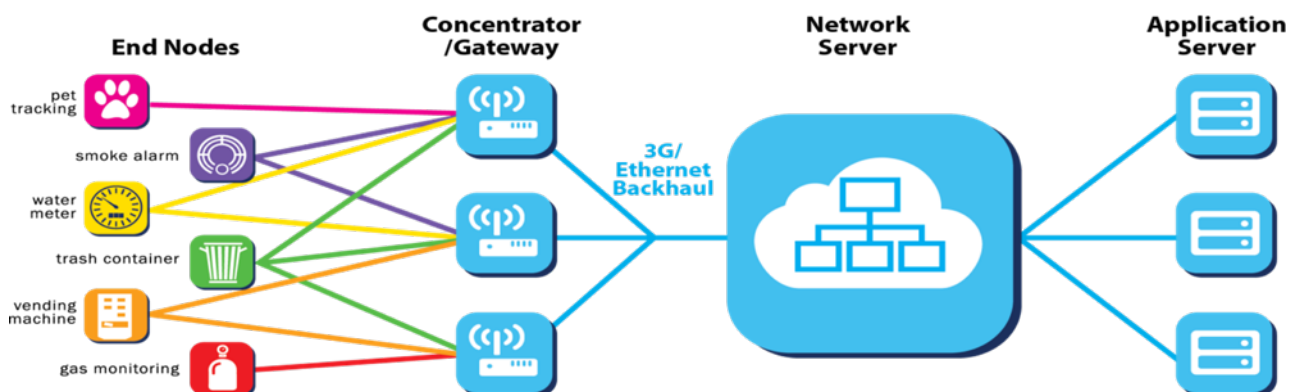


Figure 1: LoRaWAN™ architecture

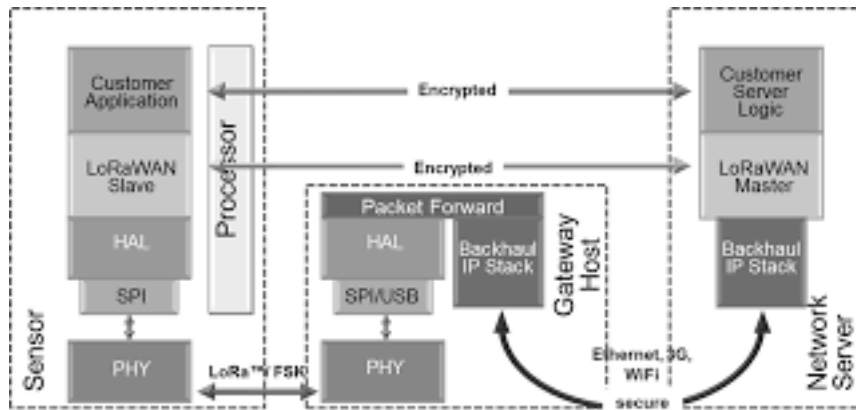


Figure 2: LoRaWAN™ Protocol Architecture

As far as the radio link is concerned, the LPWAN-CSS system is comprised of end nodes and gateways. LPWAN-CSS is a system comprised of end nodes and gateways. The end nodes are sensors and/or actuators equipped with a LPWAN-CSS radio transceiver. They are resources constrained in terms of processing power, memory and energy; in some cases, depending on the availability and the application, the end nodes can be attached to the main power and so the constraint on the energy is relaxed. Gateways are radio equipment able to receive and transmit LPWAN-CSS signals; gateway have powerful signal processing (analogue and digital) capabilities as well as medium-high computational capability (their processing unit may range from an ARM 9 to powerful dual or quad core ARM processors).

LPWAN-CSS end nodes communicate bi-directionally with one or more gateways, each gateway being connected with a central entity, the NS, controlling the network and connecting to external networks.

For the purposes of the present document, LPWAN-CSS is a system that relies on the LoRaWAN™ (see clause 5.2) protocols and radio access techniques, and uses the chirp spread spectrum (with optional fall back to FSK).

LoRaWAN™ is established and maintained by the Lo-Ra Alliance™ ([Lo-Ra Alliance website](http://www.lora-alliance.org)).

5.2 The LoRaWAN™ protocol

5.2.1 Overview of the protocol

LPWAN-CSS enables long-range, low power and low cost bidirectional communication and is deployed in a star-of-stars network architecture whereby end nodes are not associated with a specific gateway, but transmit data to multiple gateways within their range.

The gateway from a system and protocol description point of view is a combination of two functions: encapsulation/de-encapsulation of the LoRaWAN™ packets and packet forwarding, i.e. forwarding the IP packets to and from the Network Server. Each gateway is forwarding the received packets from/to the end-node to the network server via a backhaul IP connection over different possible media (fibre optic, Ethernet, Wi-Fi, Cellular data connections, satellite data connections, etc.).

5.2.2 End Node

The end node is a classical sensor or actuator equipped with LPWAN-CSS radio and a resource constrained microprocessor controlling the radio transceiver and carrying on the tasks of the LoRaWAN™ MAC as well as the application software.

In the LoRaWAN™ protocol there is such a concept as handover and related association of a node to a certain specific gateway. In the uplink (from the end node to the gateways and the network) the nodes transmit their packets which, in general, are collected by one or more gateways, each of which relays the packet to the Network Server, along with a link quality indicator. It is this link quality indicator that enables the Network Server to select (assuming reciprocity) for the downlink (from the Network Server to the end-node) what is usually called "best serving gateway", i.e. the gateway from which the packet has been received with the best link quality indicator.

The end-nodes, according to the LoRaWAN™ protocol are distinguished in 3 categories (classes) detailed below. It is pointed out that the category of a node is not, in general, fixed. For example, a node of the lowest category can only behave as the lowest class node; however, a node which implements the features of - for example - the second highest category starts, when switched on, as a node of the lowest class and then switch to the higher class after requesting and receiving the permission from the network server. The networks implemented according to the LoRaWAN™ protocol are not required to support all of the categories of nodes, but only the lowest one. Of course, supporting additional categories provides additional benefits for the network (which will be clear after the definition of the classes) and, as a matter of fact, existing networks are supporting more than only the lowest class:

- Class A: the communication can be initiated by the end node only; the end node does not make any sensing before initiating the communication; only after an uplink communication from the node there could be a downlink transmission from the network server (which might have cached information intended for the node coming from external networks); this downlink communication can be a control type information (e.g. an acknowledgement that the uplink packet has been received from the network server) and/or an actual payload.
- Class B: class B nodes can exist only in networks that support their features; LoRaWAN™ networks supporting the class B are configured to have all of the gateways emitting periodic beacon signals; Class B end nodes synchronize to this periodic beacon signal and get assigned a certain slot for the downlink traffic, if any, even without prior uplink transmission.
- Class C: class C end nodes have the radio receiver always on, in listening mode, ready to capture any packet the Network Server decides to send to them via a suitable gateway.

From the above definitions, one can easily recognize passing from Class A, to Class B and Class C the trade-off between the power consumption and the responsiveness in downlink of the nodes. For class A nodes the delivery of a downlink packet depends - apart from the network conditions in terms of congestion, interference, etc. - on the fact that an uplink transmission occurs, so it is not under control of the network; for a class B instead there might be a certain delay depending on how frequently the beacon is transmitted but the downlink packet can be scheduled by the network; eventually, with Class C end nodes the network can send the packet in downlink without any constraints, apart from what mentioned above in terms of interference, congestion, etc.

5.2.3 Gateway

The gateway is a radio transceiver, working in half duplex mode, much similar to a base station in cellular networks, except for a much lower complexity and power consumption. Its role in the uplink communication path is to "collect" the packets arriving on the air from end nodes and forwarding them, through the IP backhaul connection, to the network server, timestamping them (with high precision if equipped with a GPS receiver, otherwise getting the time information from the backhaul IP link) and attaching to them a link quality indicator. On the downlink, the gateway receives the packets to send to the nodes from the network server along with the time it should send them on air and transmits them on air on the radio channel and with the SF indicated once again by the network server (as described in clause 7.1.1). From the protocol point of view then, as already pointed out, the gateway is a "*simple*" relay of packets in uplink and downlink to/from the network server. However, it is pointed out that the gateway, as the end nodes, respects the regulation especially as concerns the duty cycle restrictions. So, on the control plane the gateway is reporting quite some information on the network status and its status (e.g. packets dropped because of the restriction on time on air, packets received with errors, e.g. because of a failed cyclic redundancy check, level of interference in the different radio channels) to the Network Server. The network server, based on this information can optimize the network and, for example, send suitable protocol commands to allocate the end nodes on the best (in terms of lower interference and traffic) radio channels.

Even if, from the protocol point of view, the gateway could seem a simple machine, it is actually a quite complex one from the signal processing point of view. As a matter of fact, the receiver in the gateway is a highly configurable parallel machine acting like (at least 8) different independent receivers, configurable on different channel frequencies (each of them can demodulate any SF index from 7 to 12 in parallel). The LoRaWAN™ protocol does not mandate a procedure on how to configure the gateway and it is left to the network server (and ultimately to the LoRaWAN™ network operator) to configure the gateways in the smartest possible way in order not to lose any uplink packet. This kind of situation may occur if an end node transmits on a certain radio channel with a certain SF and no one of the gateways that can receive the packets have any receiver allocated to that specific channel with that specific SF. It is an unlikely situation in properly designed real LoRaWAN™ networks but it is not a trivial task. It is part of the "*intelligence*" of the network server to ensure this proper allocation.

A final remark about the gateway is important: different types of gateway are already available in the market: they range from full-fledged outdoor gateways intended to be hosted on cellular network or TV towers to home gateways which use the home Wi-Fi connection to be connected to the Network server. The different types of gateway are intended for different application scenarios: the outdoor gateways are by far and large the most used type of gateway used in public open networks (i.e. by networks where an operator offers the LPWAN CSS connectivity to end users owning the end nodes sensors and/or actuators).

5.2.4 Network Server

As pointed out in the previous clauses, the network server is the central network element orchestrating the entire LoRaWAN™ network from the radio resource management point of view. As a matter of fact, the allocation of the nodes in different radio channels (uplink and downlink), the selection of the gateway for the downlink, the control of the timing of the downlink operation operations, etc. are all tasks for the Network server.

The Network Server is mainly the termination for the LoRaWAN™ Protocol on the network side. As such, according to the LoRaWAN™ specifications, it is to exchange protocol messages with all end nodes for Medium Access and Network Control purposes (e.g. assignment of the channel where to transmit, sending acknowledgments for message received if required, etc.). A specific LoRaWAN™ task, the Adaptive Data Rate, is explained in detail in clause 5.2.5.

Furthermore, the Network server carry on also different other tasks, among which:

- Authentication of the end nodes and admission control.
- Encryption of protocol commands and related data.
- Management of the gateways (e.g. in terms of the allocation the receivers).
- Localization of the end devices (when supported by the gateways).
- Network Operation Administration and Maintenance.

Regarding the localization task, the Network server can make use of the time stamping of the packets received from a single end node from multiple gateways and, knowing the exact location of the gateways, with a location solver find out the position of the end node. It should be remarked that the location capability does not involve any modification in end nodes, being a feature depending only on the accuracy of the time stamping of the received packets in the gateways and on the location solver of the Network server.

5.2.5 Adaptive Data Rate

The LoRaWAN™ protocol allows the end nodes to individually use any of the possible SF, whichever gives the highest data rate. This feature is used by the LoRaWAN™ network to adapt and optimize the data rate of static end-devices. This is referred to as Adaptive Data Rate (ADR). The use of the ADR is activated by the end node or by the network server.

An end node tries and estimate the highest data rate i.e. the lower SF it can use and be received correctly by the Network Server. Starting with that estimation (which, in any case, can be the highest data rate i.e. the lower SF) it initiates the transmissions. If no reply is received within the next expected downlink transmissions, the end node may try to establish connectivity by switching to the next lower data rate (i.e. the next higher SF) that provides a more robust connectivity. The end node will further lower its data rate (i.e. increase the SF) step by step until the communication with Network Server is established.

The adaptive data rate mechanism has an important benefit to the end-point: longer battery life. From an energy point of view, most of the consumption is spent on radio transmissions. Only a small part (20 %) of the node energy is spent on radio reception, data processing and idle times. Therefore, a LPWAN-CSS end-point is converting the energy of its battery mainly into radio waves.

Similarly, to end device, the LPWAN-CSS Gateway applies Adaptive Data Rate to Downlink emission based on the previous received uplink quality.

More than that, if the uplink is received by multiple gateways, the networks server will choose to perform a downlink on the gateway that has received the better quality signal, so that an higher data rate can be used (reducing downlink time on air).

This technique applies to both 25 mW and 500 mW emissions.

5.2.6 Application Server

Each end node, according to the LoRaWAN™ protocol, is equipped with an *AppKey*. The *AppKey* is intended to provide an end-to-end encryption of the application data of each sensor/actuator node between the node itself and the *Application Server* handling the data sent and received to/from the end node. The Application server is specific for a certain application which may be in common with many end nodes.

It is worthwhile to note that in the LoRaWAN™ protocol, the encryption of MAC payload is done independently of the application payload. This feature enables a clear separation between the control plane and the application plane. The network provider cannot inspect in any way the application data nor the applications can inspect the MAC payload (i.e. the protocol commands and their associated data).

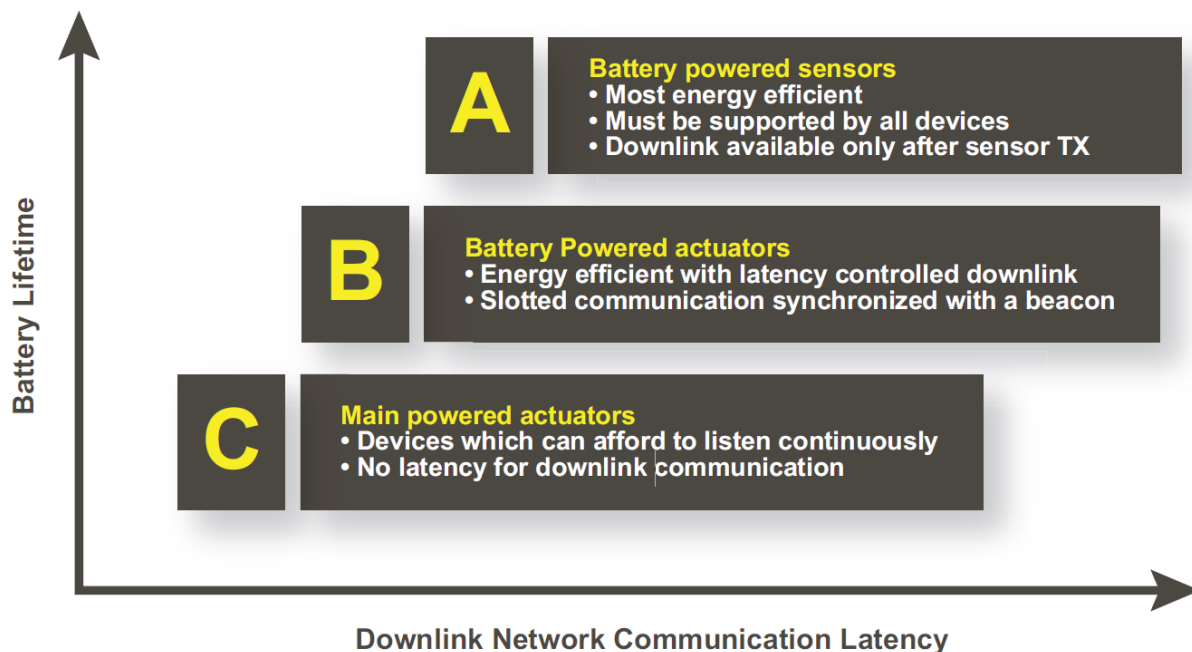


Figure 2a

6 Market information

The LoRa Alliance™, the consortium maintaining the LoRaWAN™ protocol, has attracted more than 450 members after two years of existence. Its members include mobile network operators, sensor and gateway manufacturers, chipset and module manufacturers, large enterprises, network management services, and application software providers. The LoRaWAN™ protocol is open in a twofold sense:

- Every developer, network provider, equipment manufacturer, etc. can get it by simply sending an email request to <mailto:admin@mail.lora-alliance.org>.
- Every developer, network provider, equipment manufacturer, etc. can become a member of the LoRa Alliance™ and contribute to the evolution of the LoRaWAN™ protocol (and related test and certification procedures).

While the LoRa Alliance™ defines the LoRaWAN™ specification and certifies products, it does not dictate how service providers should deploy LPWAN-CSS networks and price services. This open ecosystem approach creates flexibility for service providers and enables a variety of business models to flourish. For example, service providers currently offer LPWAN-CSS connectivity services based on monthly subscriptions for network server use, number of messages sent, number of devices connected, or according to time of day usage. Furthermore, the LoRaWAN™ protocol can be used to implement:

- a) A private (closed) network: an entity (e.g. a company in its own premises, a municipality, etc.) can deploy all of the elements of Figure 1 and run the network.
- b) A public (open) network: an entity (a network provider) can deploy all of the elements of Figure 1 except the end nodes and offer, as mentioned above, the connectivity to anyone intending to deploy sensor/actuator end nodes and related applications for different purposes.

Figure 3 represents only the announced open networks that are available for end users to benefit from the connectivity. For the purpose of the present document, one can see that in the vast of the countries members of CEPT there are already public (open) networks available.

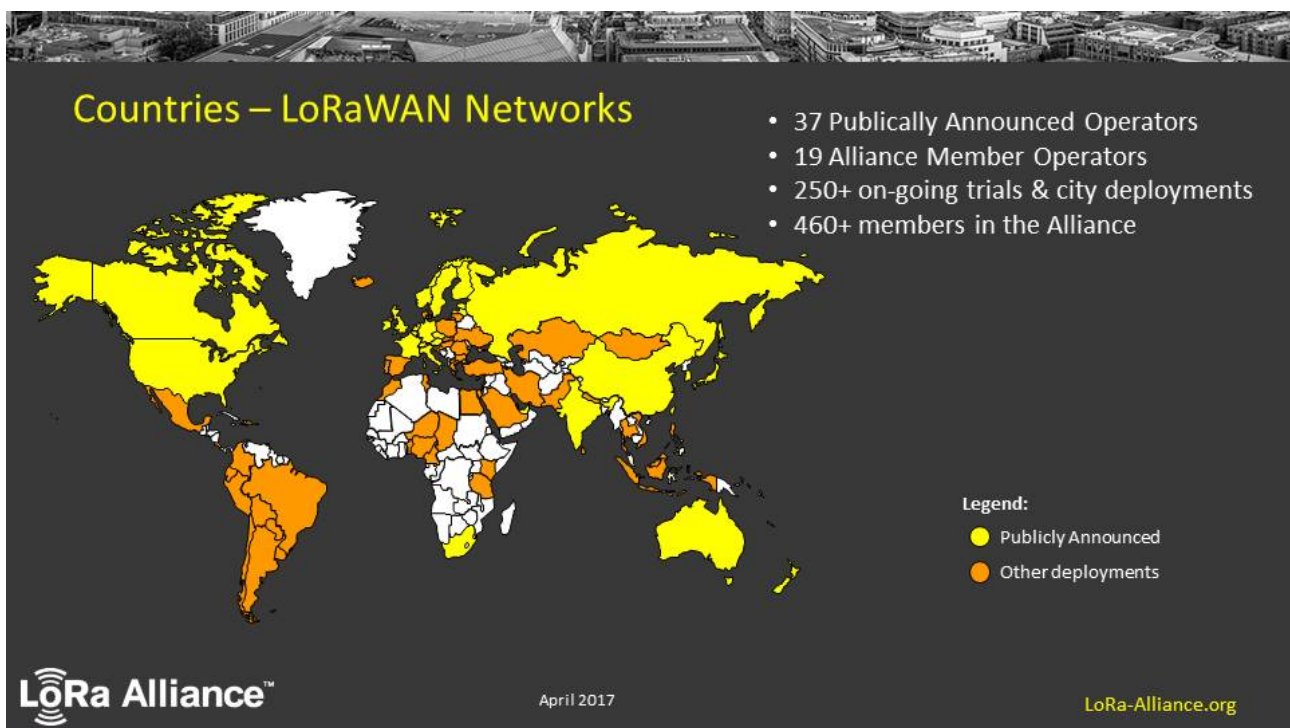


Figure 3: LPWAN-CSS networks worldwide coverage map, April 2017 (source LoRa Alliance™)

LPWAN-CSS technology is used for variety of applications that can be classified in different verticals. Figure 4 is a list of the main verticals and applications that are currently foreseen for these networks.



Figure 4: Main verticals and use cases (source LoRa Alliance™)

Overall the proportion of indoor devices varies from 30 % to 70 %, depending on the specific market. Indoor meters and smart building devices represent a large number of use cases. And outdoor parking sensors or tracking devices are also used for a large portion of the applications.

After having described the main applications of LPWAN-CSS technology, the market forecasts are extremely positive for the LPWAN-CSS technology:

In 2021 the following is predicted:

- 1,4 billion USD revenues from LPWA-CSS module shipping, the majority of which, with equal sharing, in Europe, North America and Asia-Pacific.
- The sales of LPWA modules is projected to have a volume almost equal to the sum of the module's sales from other LPWA technologies (including LTE CATM1, LTE CATNB1 and SigFox).

7 Technical information

7.1 Detailed technical description

7.1.1 LPWA-CSS signals

The radio signals used by both the end nodes and the gateway employ the same modulation and spreading. They are chirp spread spectrum signals parametrized by the Spreading Factor (SF), a parameter ranging from 7 to 12 controlling the slope and the length in time of the signals. As a useful convention in the context of LPWAN-CSS systems and networks, the term SF index represents the $\log_2()$ of the actual spreading factor computed according to the LoRaWAN Specifications (see [i.1]). So, the actual spreading factor, according to the LoRaWAN Specifications range from 128 (SF=7) to 4 096 (SF=12).

In Figure 5 an example of the instantaneous frequency of a LPWAN-CSS radio signal with SF=7 in a bandwidth of 125 kHz is given: it can be observed that the frequency is sweeping with respect to the centre of the operating channel (assumed to 0 for simplicity) from -62,5 kHz to 62,5 kHz. The channels are separated by 200 kHz.

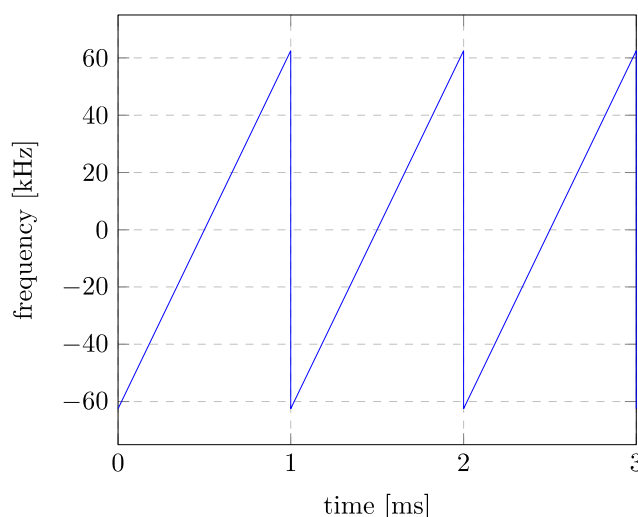


Figure 5: An example of a LPWAN-CSS signal

In Table 1, one can see the influence of the SF parameter which is basically stretching in the time domain the signal represented in Figure 5, doubling the duration of the each sweep at every step increase of the SF.

Table 1: Chirp duration as a function of the Spreading Factor

Spreading Factor index	7	8	9	10	11	12
Chirp duration	1 ms	2 ms	4 ms	8 ms	16 ms	32 ms

The LPWAN-CSS signals are used in asynchronous manner over the air, i.e. the end nodes transmit uncoordinated between themselves and the gateways. One then may think that the collisions on air might lead to the impossibility of correct reception of most of the signals. However, a fundamental feature of the LPWAN-CSS signals is that any two LPWAN-CSS signals have a certain isolation factor if they have different spreading factors. Two LPWAN-CSS signals with different spreading factors, if superimposed in time and frequency, can be demodulated provided there is a sufficient power margin between the wanted and unwanted signals at each receiver.

The LPWAN-CSS packet of data consists of a preamble, a PHY (Physical Layer) header and an actual payload. The preamble is used for detection and synchronization purposes, the PHY header describes the payload length which ranges from 13 to 255 bytes. Indeed, the minimum size of a LoRaWAN™ physical payload is 13 bytes. A 16 bits CRC is also transmitted. The channel encoding for the packet is a (5,4) parity code. In Table 2 the time overhead (i.e. the time used for the transmission of the preamble, the PHY header and the CRC) and the payload data rate (parity code included) are shown as function of the spreading factor.

Table 2: Payload Data Rate and Time overhead (per packet, excluding data) as a function of the SF index

SF index	Payload Data rate	Time overhead
7	5,5 kbps	40 ms
8	3,1 kbps	80 ms
9	1,8 kbps	150 ms
10	0,98 kbps	280 ms
11	0,44 kbps	570 ms
12	0,25 kbps	1 100 ms

The LPWAN-CSS modulation belongs to the spread-spectrum class. As such it encodes a low data rate bit stream onto a comparatively wide occupied bandwidth. Similarly, there is no direct relationship between occupied bandwidth and actual bit rate, the bit rate can be changed over a wide range without modifying the spectrum shape. The transmitted LPWAN-CSS signal is constant envelope. This means that it can be transmitted using a simple and very power efficient radio architecture where the RF Phase locked loop output directly drives a saturated power amplifier. See the Annex D for resulting modulated spectrum.

7.2 Technical parameters and implications on spectrum

7.2.1 General technical parameters

In the deployment in Europe of a LPWAN-CSS network (although the network operator can in principle choose a different channel line-up) a set of channels are defined by the LoRaWAN™ protocol both for the end devices and the gateways. As far as the terminology is concerned, in the LoRaWAN™ protocol a channel is a triplet [center frequency, occupied bandwidth, SF for LPWAN-CSS or FSK] while the Data Rate (DR) designator is reference to a combination of two values [SF for LPWAN-CSS or FSK, occupied bandwidth]. It is remarked that the LoRaWAN™ protocol allows devices to make use of the plain FSK modulation with bit rate of 50 kbits/s.

For the European region, the following set of data rates are allowed by the LoRaWAN™ protocol.

Table 3: data rates allowed by the LoRaWAN™ protocol

DR	Configuration (SF for LPWAN-CSS or FSK, occupied bandwidth)	bit rate (bit/s)
0	LPWAN-CSS: SF12 / 125 kHz	250
1	LPWAN-CSS: SF11 / 125 kHz	440
2	LPWAN-CSS: SF10 / 125 kHz	980
3	LPWAN-CSS: SF9 / 125 kHz	1 760
4	LPWAN-CSS: SF8 / 125 kHz	3 125
5	LPWAN-CSS: SF7 / 125 kHz	5 470
6	LPWAN-CSS: SF7 / 250 kHz	11 000
7	FSK	50 000

The central frequencies of the LPWAN-CSS (LoRaWAN™) signals are in Europe as following (as an example), the occupied bandwidth being 125 kHz:

- 1) 867,1 MHz (1 % duty cycle, uplink and downlink)
- 2) 867,3 MHz (1 % duty cycle, uplink and downlink)
- 3) 867,5 MHz (1 % duty cycle, uplink and downlink)
- 4) 867,7 MHz (1 % duty cycle, uplink and downlink)
- 5) 867,9 MHz (1 % duty cycle, uplink and downlink)
- 6) 868,1 MHz (1 % duty cycle, uplink and downlink)
- 7) 868,3 MHz (1 % duty cycle, uplink and downlink)
- 8) 868,5 MHz (1 % duty cycle, uplink and downlink)
- 9) 869,525 MHz (downlink only, 10 % duty cycle)

Three channels (channels 6, 7 and 8, see Table 4) are the so called "*default channels*" and are to be implemented by every end node.

Table 4: 863 MHz - 870 MHz Default Channel List

Modulation	Bandwidth [kHz]	Central Channel Frequency [MHz]	LPWAN-CSS DR/Bitrate
LPWAN-CSS	125	868,10 868,30 868,50	DR0 - DR5 / 0,3 - 5 kbps

The use of the *default channel* is mandatory for end nodes and gateway, as said before, and the use of the different Data Rates (DR) is regulated by the Adaptive Data Rate described above.

Furthermore, apart from the 3 default channels, the network server can instruct the gateways and the end nodes to use up to 5 additional channels through a specific protocol command. For each additional channel the bandwidth is 125 kHz and the data rates that can be employed are DR0 to DR5.

Table 5 shows the typical configuration for antenna heights for gateways and end nodes.

Table 5: typical antenna heights; in brackets the percentage of Devices and Gateways at the respective antenna height

Equipment	Typical antenna height (indoor)	Typical antenna height (Outdoor)
End nodes (all with integrated antenna)	<ul style="list-style-type: none"> • 1,5 m (80 %) • 9 m (9 %) • 30 m (1 %) 	<ul style="list-style-type: none"> • 1,5 m (10 %)
Gateway	<ul style="list-style-type: none"> • 1,5 m (10 %) • 9 m (10 %) (indoor gateways have integrated internal antenna)	<ul style="list-style-type: none"> • 25 m (75 %) • 40 m (5 %)

Table 6 presents the range of coverage radius for LPWAN-CSS outdoor gateways. This radius strongly depends on a number of factors such as antenna height, terrain, buildings height and density, location of the devices.

Table 6: typical coverage radius

Environment	Typical coverage radius
Urban	1 km
Rural	10 km

In Figure 6 is a simulation result showing the impact of gateway density on the realistic distribution of used data rates among end nodes which populate an LPWAN-CSS network. The assumptions used were a square grid gateways layout, small/medium city Hata model [i.2] for attenuation, Rayleigh fading, and uniformly distributed additional shadow margin between 0 dB and 40 dB to account for end-nodes positions from outdoor to deep indoor.

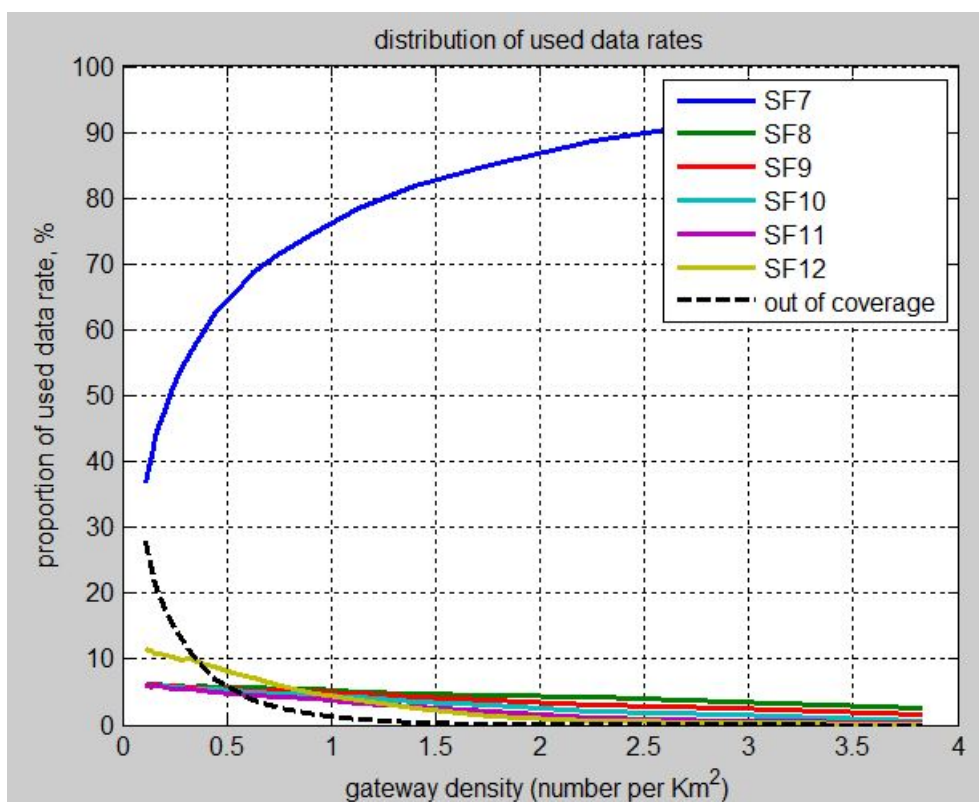


Figure 6: Data rates distribution as a function of gateway density

The SF12 is more used than SF11 because this is the lowest available data rate: this shows the ADR algorithm uses some margin.

The adaptive data rate here assumes that the devices are static so that after a few messages they can be controlled to the target data rate. SF7/8/9/10/11/12 have to be considered as index of data rate, with a fixed signal bandwidth of 125 kHz. SF7 data rate is roughly 20 times faster than SF12 data rate, which explains why the average duty cycle reduces a lot as network density increases.

According to the above assumptions, the results show that most of the end nodes are using SF7. This is because in the above-mentioned simulation scenario, there is a mix of deep indoor, indoor and outdoor nodes, due to the fact that it is assumed a uniform distribution of the attenuation due to building penetration losses ranging from 0 dB to 40 dB. This would change if indoor and deep indoor only nodes are considered: in this case, the end nodes using SF7 would only become dominant when the density of gateways is high.

It may look like a strong assumption that devices are static, and that adaptive data rate is not possible for moving devices. However, the vast majority of moving devices will be outdoor, so that - due to better propagation conditions with respect to the underground and indoor end nodes - they are very likely to use the smallest SF, i.e. SF 7. This is another reason why the big proportion of SF7 end nodes: outdoor end nodes which do not use adaptive data rate but they do not need it.

As the density of gateways is varied, the resulting average message duration is computed and reported below on Figure 7.

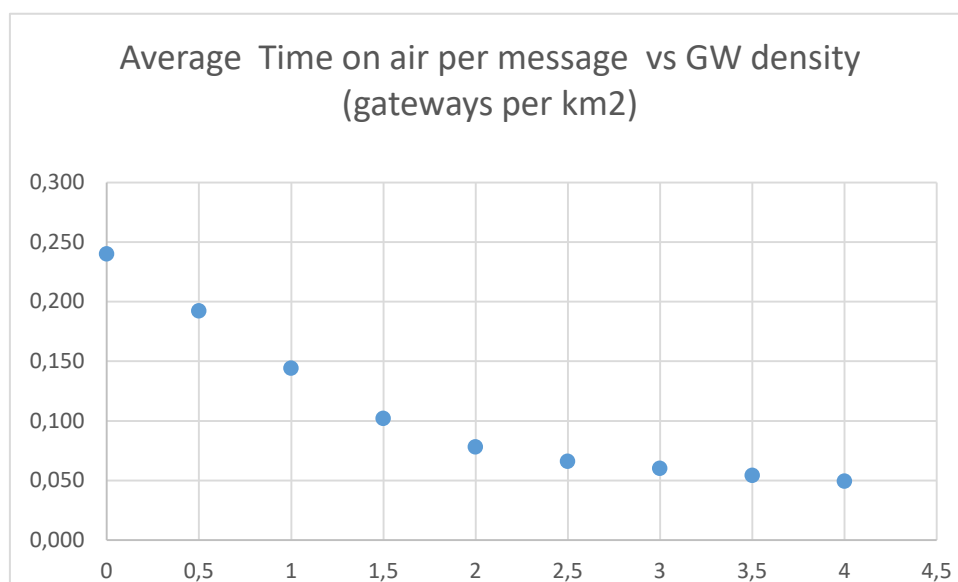


Figure 7: Average time on air [s] per message as a function of the number of gateways per square kilometre

Table 7 shows the $T_{on} / (T_{on} + T_{off})$ for end nodes and gateways, for typical applications.

Table 7: $T_{on} / (T_{on} + T_{off})$ per square kilometre LPWAN-CSS , for typical applications

Year	End nodes $T_{on} / (T_{on} + T_{off})$	End node density per square kilometre	Gateway $T_{on} / (T_{on} + T_{off})$ per channel (there are at least 3+1 channels)	Gateway density per square kilometre
2017	0,0160 %	360	0,7 %	0,5
2023	0,0045 %	5 582	0,5 %	3,5

We base our derivations on the peculiarity of the LPWAN CSS Systems called "network densification": a network operator seeing an increasing number of connections will increase the gateways density and this is counterbalancing the global time on air since, as a consequence of an increase of gateways, the time on air is decreasing (because the devices use a lower spreading factor).

As mentioned above the number of channels used to deploy a LPWAN CSS system are usually 9 (8 uplink/downlink and 1 downlink); actually a gateway can demodulate 8 channels. The minimum number of channels needed to deploy a LPWAN CSS system is 4 (3 uplink/downlink and 1 downlink). The uplink communications are only low power; the downlink can use both low power and high power channels; the network defines the power based on the channel frequency (according the regulations).

If more than one operator (suppose for simplicity two) is operating in the same geographical area two situations can arise:

- (competing networks) 4 out of 9 channels are in common (including the high power one in downlink only) between e.g. two networks, the others channels are chosen in an independent way by the two networks;
- (cooperative networks): all channels are the same between the two networks, roaming between network servers handle the packets to/from different networks: if a packet is received by a network to which it is not pertaining, the packet is forwarded to the network server of the other network.

For sake of clarity, no exclusivity/protection of any channels is needed; this means no specific frequency planning or specific re-use principles are applied.

The above mentioned results of Figure 7 and Table 7 already include the traffic of potential multiple networks running in parallel.

7.2.2 Status of technical parameters

7.2.2.1 Current ITU and European Common Allocations

See clause 9.1.

7.2.2.2 Sharing and compatibility studies (if any) already available

In clause 7.3.1 of ETSI TR 103 435 [i.10] an overview of existing studies are provided. Since the publication of ETSI TR 103 435 [i.10] the ECC Report 246 [i.9] and ECC Report 261 [i.8] were published.

7.2.2.3 Sharing and compatibility issues still to be considered

It is expected that CEPT will perform the usual compatibility and sharing studies as for any new technology.

Annex B provides some input.

7.2.3 Transmitter parameters

7.2.3.1 Transmitter Output Power/Radiated Power

7.2.3.1.0 Introduction

The end nodes transmit (always according to the limitations of the bands in which they work) with a power of 5 mW or 25 mW; the gateways transmit with power equal to 5 mW, 25 mW or 500 mW (always according to the limitations of the bands in which they work). The end nodes operate according to the duty cycle restriction in each band or according to the LBT/AFA (see clause 7.2.5) policy, while the gateways can operate only according to the duty cycle limitations.

Adaptive Power Control can also be applied to the transmitted signal power of the end nodes in Uplink and the gateways in Downlink, based on network indications. Similarly to the Adaptive Data Rate, the Network server can impose to an end device to reduce its output power (through a Downlink control command) based on the signal quality information of previous received packets in Uplink. Same is valid for the Gateways, where the Network Server can reduce the Downlink signal power based on the signal quality of previous packets received from an end node in Uplink. Moreover for Uplink frames received by multiple GW, the network server will choose to perform a downlink transmission using the gateway that has received the signal with the best quality (called "assigned downlink gateway"), so that a lower output power can be used.

For the uplink, based on previous emissions signal quality (i.e. signal level computed by the assigned downlink gateway) the network server reviews the transmitted output power of the end node via a protocol command: if the resulting margin with respect to the sensitivity of the gateway is higher than 10 dB, then the output power of the end nodes is requested to be reduced, with a 2 dB granularity, ranging from 14 dBm to 0 dBm.

For the downlink, based on previous emissions signal quality (i.e. signal level computed by the assigned downlink gateway) the network server reviews the transmitted output power of the assigned gateway: if the resulting margin with respect to the sensitivity of the gateway is higher than 10 dB, then the output power of the assigned gateway is reduced, with a 3 dB granularity, ranging from 27 dBm to 3 dBm on 500 mW downlink and from 14 dBm to 3 dBm on 25 mW downlink.

It is noted that the RX signal level is based on the uplink signal received between 1s and 5 s before the transmission.

7.2.3.1.1 Antenna Characteristics

The end-device uses an omnidirectional antenna which gain is 0 dBi, equivalent to -2,15 dBd.

The gateway uses an omnidirectional antenna which gain is 5,5 dBi, equivalent to 3,35 dBd.

7.2.3.2 Operating Frequency

The bands in which LPWAN CSS systems could operate are for example the following mostly harmonised SRD bands, see ERC Recommendation 70-03 (ref [i.6]):

- a) 863 MHz - 870 MHz (h1.3).
- b) 868 MHz - 868,6 MHz (h1.4).
- c) 868,7 MHz - 869,2 MHz (h1.5).
- d) 869,4 MHz - 869,65 MHz (h1.6).
- e) 869.7 MHz - 870 MHz (h1.7).
- f) 870 MHz - 876 MHz (h2).
- g) 915 MHz - 921 MHz (h3).

LoRaWAN™ system can withstand frequency tolerances of typically $\pm 25\%$ of the LPWAN-CSS occupied bandwidth and still maintain a 10 % PER link.

7.2.3.3 Out of band emissions

Table 8 and Table 9 are showing the out of band emissions. The centre frequencies are not specified, actually the centre frequency can be any frequency compatible with the spectrum requests in Paragraph 8.

Table 8: Downlink (gateway) and Uplink (End-node) emission level for 25 mW ERP

F-fc (kHz)	Level e.r.p
$\pm 62,5$	14 dBm/125 kHz
± 125	-45 dBm/1 kHz
± 250	-55 dBm/1 kHz
± 500	-52 dBm/100 kHz
$\pm 6\ 000$	-52 dBm/100 kHz

Table 9: Downlink (gateway) emission level for 500mW ERP

F-fc (kHz)	Level e.r.p
$\pm 62,5$	27 dBm/125 kHz
± 125	-32 dBm/1 kHz
± 250	-42 dBm/1 kHz
± 500	-39 dBm/100 kHz
$\pm 6\ 000$	-39 dBm/100 kHz

The results of OOB measurements are shown in Annex D.

7.2.3.4 Spurious emissions

Measurements of the spurious emissions for a LPWAN-CSS signal of 125 kHz bandwidth, centered at 869,525 MHz with a power of 27 dBm (worst case) are shown below (the limits of ERC Recommendation 74-01 [i.12] are provided with red lines in Figure 8, Figure 9 and Figure 10).

First the emissions below 1 GHz are addressed in Figure 8 and Figure 9 (the measurements are done with a notch filter tuned to the carrier frequency for any frequency below 1 GHz).

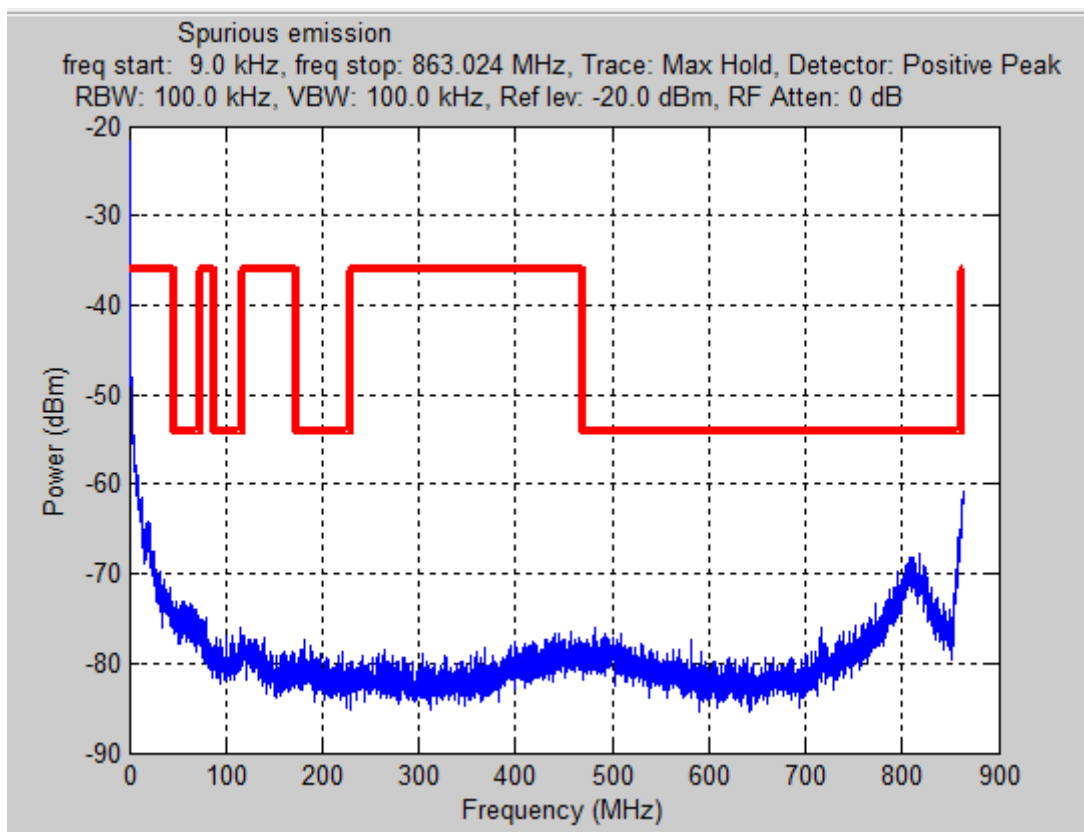
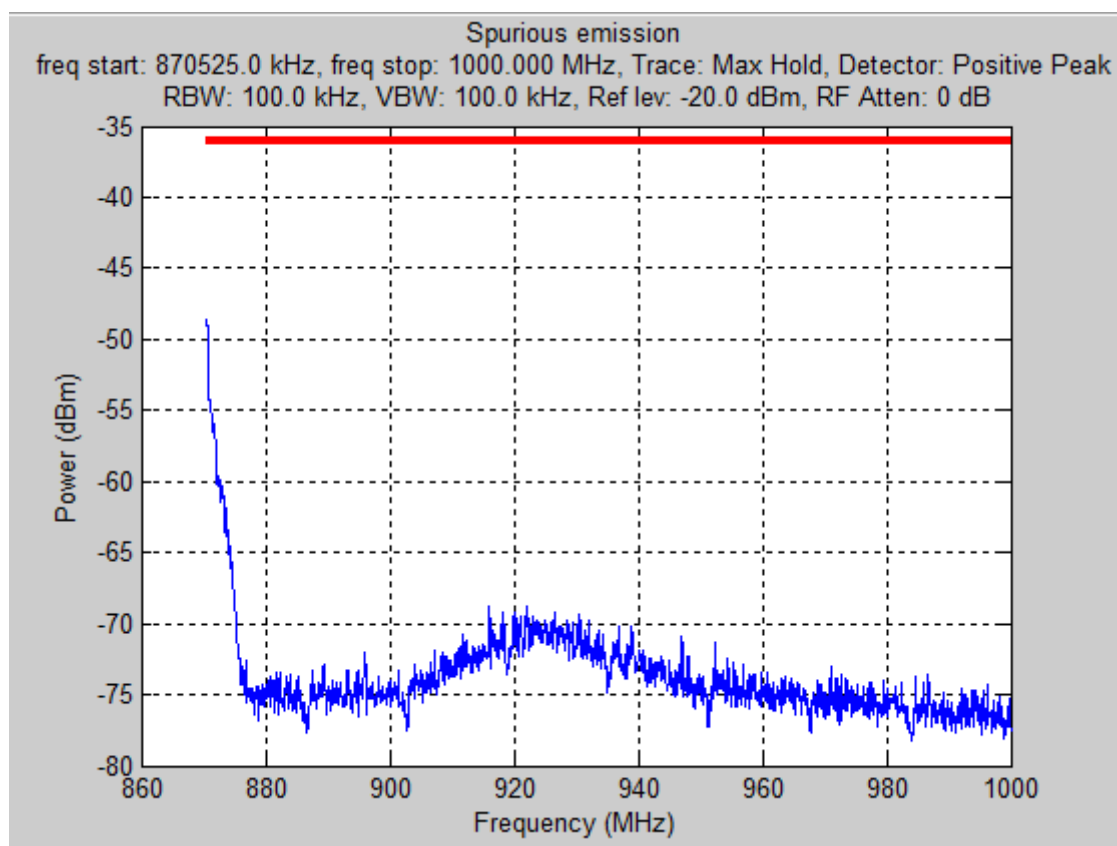


Figure 8: 125 kHz, 869,525 MHz, 27 dBm Spurious emission (conducted),
 Frequency below 1 GHz - 1st part



**Figure 9: 125 kHz, 869,525 MHz, 27 dBm Spurious emission (conducted),
 Frequency below 1 GHz - 2nd part**

In Figure 10 spurious emission measurements done in the region above 1 GHz are shown; the measurements has been done using a high pass filter ($F_c = 1,2$ GHz) for frequency above 1 GHz.

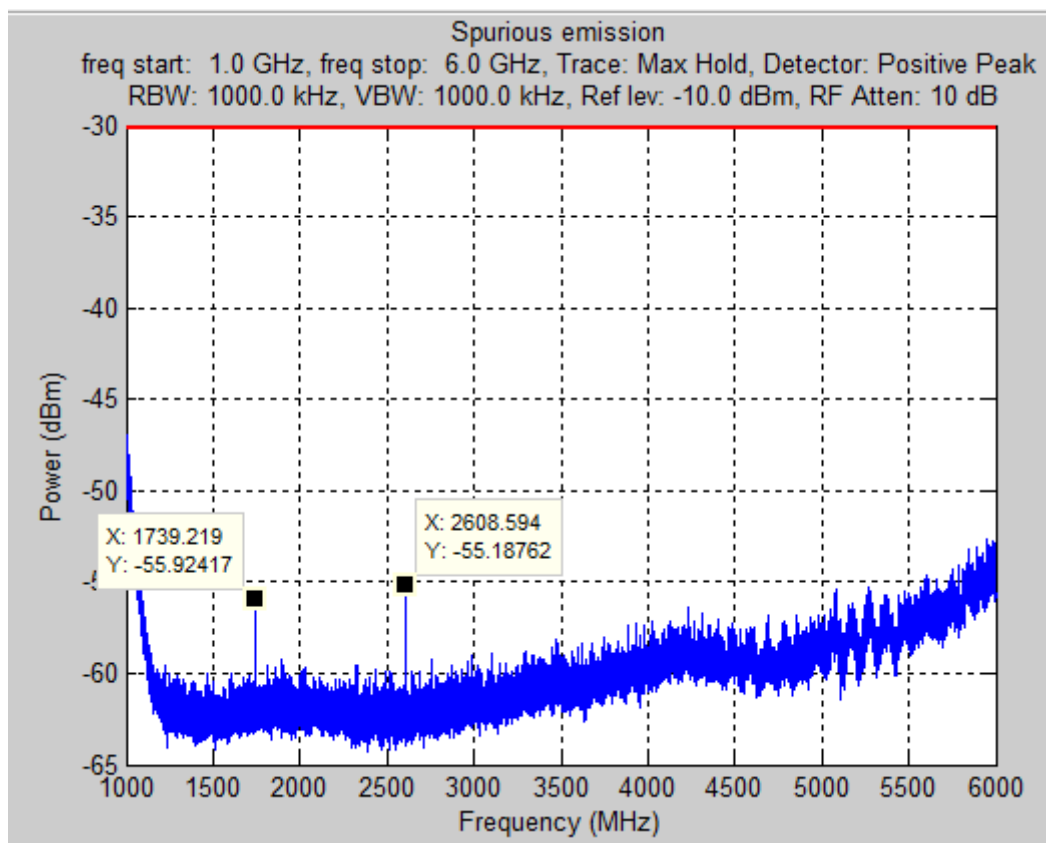


Figure 10: 125 kHz, 869,525 MHz, 27 dBm Spurious emission (conducted), Frequency above 1 GHz

7.2.4 Receiver parameters

7.2.4.0 Wanted signal level for degradation measurements

In general, in the present document, the wanted signal level is equivalent to the sensitivity per Spreading Factor (as reported in Table 10, both the gateways and the end nodes) plus 2 dB of margin. As an example, for the end device SF=7 case the wanted signal level considered is $-124 + 2 = -122$ dBm.

7.2.4.1 Sensitivity

Table 10 gives the in-band required SNR $\left(\frac{C}{I+N}\right)$ to achieve the demodulation with a 10 % Packet error rate with coding rate of 4/5, for 20 bytes payload and different SF for LPWAN-CSS signals i.e. both for the end node and the gateway.

Table 10 indicates as well the sensitivity of a LPWAN-CSS receiver, both for the end node and the gateway, taking into account the equipment noise factor (NF).

We assume, as usual, the thermal noise power spectral density (kTB) to be -174 dBm/Hz in a 50 Ω load.

In that case for a given SF a LPWAN-CSS receiver would have a sensitivity of:

$$sensi(SF) = -174 + NF + \log_{10}(BW) + SNR(SF)$$

where $sensi(SF)$ is the sensitivity at the spreading factor SF , BW is the bandwidth (125 or 250 kHz) and $SNR(SF)$ is the SNR to achieve a demodulation with a 10 % Packet error rate with coding rate of 4/5, for 20 bytes payload at spreading factor SF , according to Table 10.

For example, for a LPWAN-CSS receiver which exhibits a noise factor of 7 dB, for a bandwidth of 125 kHz, the sensitivity using SF12 will be:

$$sensi(12) = -174 + 7 + 10\log_{10}(125 \cdot 10^3) - 21,9 = -137,9 \text{ dBm}$$

A low power end node chip typically exhibits a noise factor of 5 dB to 7 dB depending on the external impedance matching components. A gateway front-end typically exhibits a noise factor of 3 dB to 7 dB.

The sensitivity of a receiver is normally taken as the minimum input signal (S_{min}) required to produce an output signal with a specific signal-to-noise (S/N) ratio. S/N is a required minimum ratio, if N is increased, then S is also be increased to maintain the S/N ratio. Considering the fact that S/N is negative for this system, the most appropriate protection factor is the C/(I+N); indeed, when the interferer is below the thermal noise level the receiver sensitivity is not significantly degraded, while when the interferer is above the noise floor, it becomes predominant and it degrades the receiver sensitivity.

Table 10: Receiver sensitivity in a 125 kHz bandwidth (considering an AWGN channel) for different SF and NF

SF index	Spreading factor	Data rate	C/(I+N) (dB)	Sensitivity @3 dB NF w/ N = -120 dBm	Sensitivity @7 dB NF w/ N = -116 dBm
7	128	5,5 Kbps	-8,0	-128,0 dBm	-124,0 dBm
8	256	3,1 Kbps	-10,8	-130,8 dBm	-126,8 dBm
9	512	1,8 Kbps	-13,6	-133,6 dBm	-129,6 dBm
10	1 024	0,98 Kbps	-16,3	-136,3 dBm	-132,3 dBm
11	2 048	0,44 Kbps	-19,2	-139,2 dBm	-135,2 dBm
12	4 096	0,25 Kbps	-21,9	-141,9 dBm	-137,9 dBm

7.2.4.2 Adjacent channel rejection

The adjacent channel rejection (victim LoRaWAN signal at -122 dBm versus adjacent LoRaWAN channel) at 200 kHz offset is -75 dBm, at 400 kHz is -62 dBm. It is noted that the center frequencies of the different LPWAN-CSS channels are spaced of 200 kHz. Annex C is providing additional information on pulsed interference.

7.2.4.3 Blocking

In the following some measurement result examples are reported, made by Semtech SA Neuchatel on actual production devices and gateways, to provide insight on the blocking performance of some real gateways and end nodes. The measurement setup uses two signal generators: one for the wanted signal, the other one for the interferer. Useful and interferer signals are summed in a power combiner to be injected in the device under test. Attenuators inserted on the combiner ports allow reducing the mutual interference between both signal generators.

Figure 11 shows the blocking robustness of a gateway reference design to a continuous carrier wave from -10 MHz to +10 MHz for all the spreading factors. A refinement is performed around the carrier frequency to evaluate the in-band robustness (see Figure 12). The results of this measurement is the blocking level which correspond to the interferer level causing a PER of 50 %.

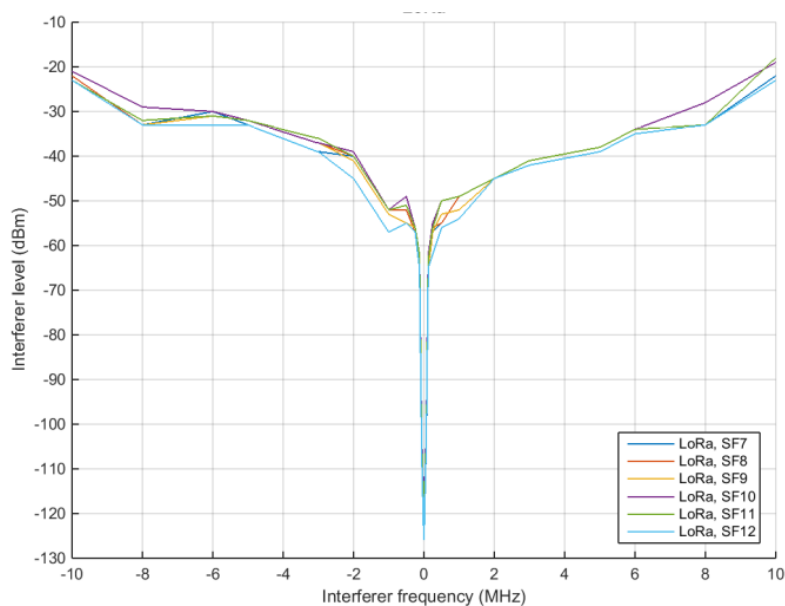


Figure 11: LPWAN-CSS gateway, blocking performances for all the spreading factors

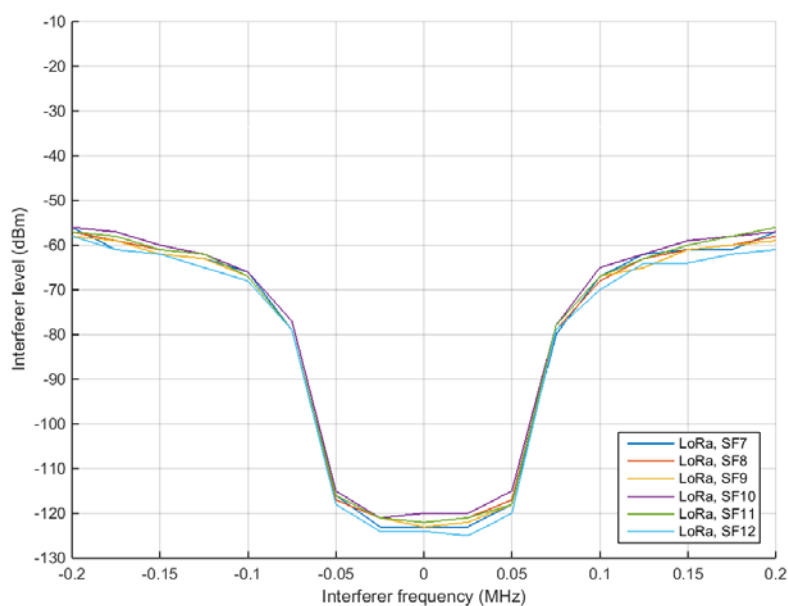


Figure 12: LPWAN-CSS gateway, In-band blocking performances

The measurement is reproduced on an End-Device evaluation board; the results are shown in Figure 13 and Figure 14. It is remarked that similar figures apply to end nodes and that the transmission and reception of the LPWAN-CSS signal is symmetrical between gateways and end nodes. As a matter of fact, the uplink and downlink link budgets are made symmetrical by a proper choice of the settings like e.g. the antenna gain.

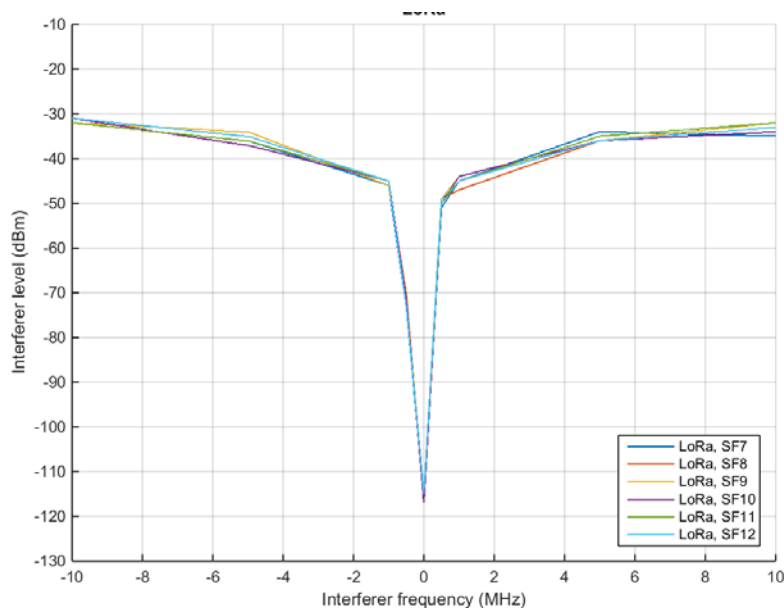


Figure 13: End-device, blocking performances for all the SF

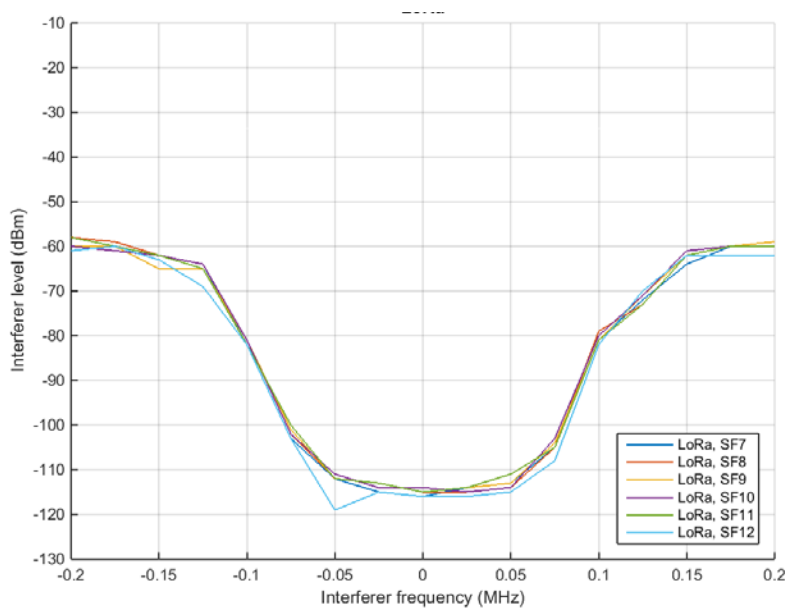


Figure 14: End-device, In-band blocking performances

The typical performance is provided in Figure 15 for the gateway, in order to facilitate the reader's use of the above measures. In Table 11 the results are reported in a tabular form for both the End Device and the Gateway.

Note that the figures provided in this and the above paragraph do not take into account possible external filters (e.g. SAW or cavity) which are typically applied to gateways and optionally to end nodes.

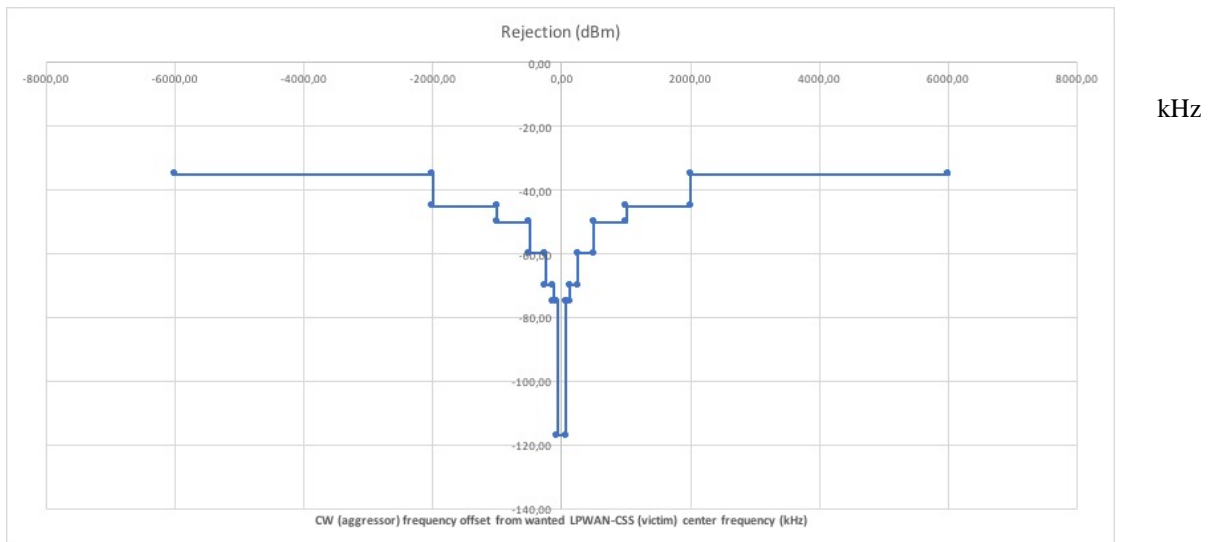


Figure 15: Continuous wave interferer rejection for the gateway; the wanted signal in the figure is an example of the case of an SF7 signal at -122 dBm received by the end node; packet error rate less or equal to 50 %

Table 11: Interferer rejection

CW (aggressor) frequency offset from wanted LoraWAN (victim) center frequency (kHz)	Rejection ED (dBm)	Rejection GW (dBm)
±62,5	-117	-121
±125	-75	-65
±250	-70	-62
±500	-60	-60
±1 000	-50	-57
±2 000	-45	-45
±5 000	-35	-35
±6 000	-35	-35

7.2.4.4 Intermodulation response rejection

The Input third-order Intercept Point or IIP3 is a commonly accepted parameter allowing to measure the receiver linearity. The LoRaWAN™ gateway front-end provides a typical IIP3 of -15 dBm. For the end nodes, the typical IIP3 value is -12,5 dBm. These measurements consider an offset up to 1 MHz between the wanted and the closer interferer signal.

7.2.5 Channel access parameters

The end nodes perform a Polite Spectrum Access by adopting, as mandated by LoRaWAN™, only the relevant (according to the European standards) duty cycle or LBT/AFA. It is left to the network operator to select which policy to adopt and the detailed arrangements.

The gateway, as mandated by LoRaWAN™, uses a limited duty cycle policy.

The channel access for the end nodes is described in LoRaWAN™ [i.1] and is basically an Aloha access scheme. A retransmission back-off scheme is mandatory for all end nodes; it is detailed in the LoRaWAN™ protocol [i.1], paragraph 7. It is adopted for the usual reasons, i.e. to avoid catastrophic re-transmissions in case of acknowledged messages or unexpected events (including a network bootstrap after an outage).

In the following the timing for the uplink and downlink transmissions are presented. The reader is invited to refer to clause 5.2 for the definitions of Class A, B and C end nodes and their behaviour. It is noted that, referring to Figure 16, in the RX1 receive window gateway's transmissions are limited to 25 mW channels only, while - when necessary - in the RX2 gateway's transmissions can occur in 500 mW channels. In general, in 20 % of the cases RX1 window only is used, while in the remaining downlink transmission RX2 window is used.

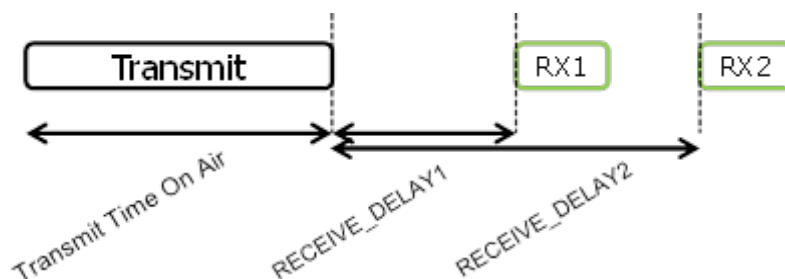


Figure 16: Class A end nodes timing for uplink and downlink transmission

For class A the timings of the transmission and reception operation are shown in Figure 16. After an asynchronous transmission (box "Transmit") with a certain spreading factor (the choice of the spreading factor is explained in clause 5.2.5) in a certain radio channel (the available spectrum is divided in different channels 125 kHz, or 250 kHz wide and the LoRaWAN™ protocol provides for a unique identification of them) the receiver opens a first window (the minimum time for the receiver to stay on from the beginning of the first window is set by the LoRaWAN™ protocol to be at least the time required by the radio of the end node to detect if a preamble is present) in a channel and using a SF to decode the received signal which are a function of the channel and SF used in the uplink. A second receive window is mandated by the LoRaWAN™ protocol with a completely similar behaviour for the end node, except from the fact that the SF and the channel are not a function of the uplink SF and channel but are configurable via LoRaWAN™ protocol commands. At the very first transmission a default configuration depending on the region (Europe has its own i.e. the one based on the default channels) is used.

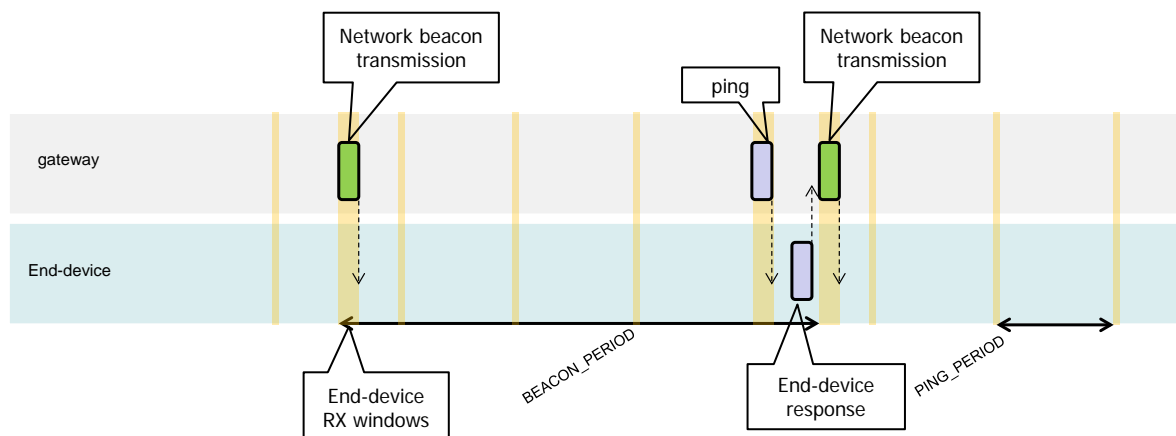


Figure 17: Class B end nodes timing for uplink and downlink transmissions

For Class B operation (see Figure 17) all of the gateways need be time synchronized and each one emits a beacon signal used by each end device to synchronize with the network. The beacon may include other information intended for upper layers (e.g. MAC layer) of the end nodes. According to the LoRaWAN™ protocol, the MAC layers - both at the end node side and the Network server side - select a period and an offset for the so-called "ping" slots i.e. the time slots where the end node is opening the radio receiver for 30 ms in order to see if there is a preamble of a data packet coming from the network. In that case, the node continues for the time needed to the complete decoding of the packet sent from the network. These "ping" slots are periodic, the period being *PING_PERIOD* and their offset from the beacon is called *PingOffset*. At the end of each *PING_PERIOD* a new *PingOffset* is computed by both the end node and the network server, in order to randomize the allocation of the end nodes and avoid e.g. periodic interference. Although the additional power consumption for a class B node is limited with respect to a Class A node to the time needed to detect the possible preambles in the ping slots, there are some additional tasks to be performed by an end node when in class B mode: the end node in this case keeps the network server informed about the gateway from which it is hearing the beacon with best quality. That will be the Gateway involved in the downlink transmission in case a packet needs to be sent from the Network Server. This gateway may of course change because of changing propagation conditions or because the end-node is moving.

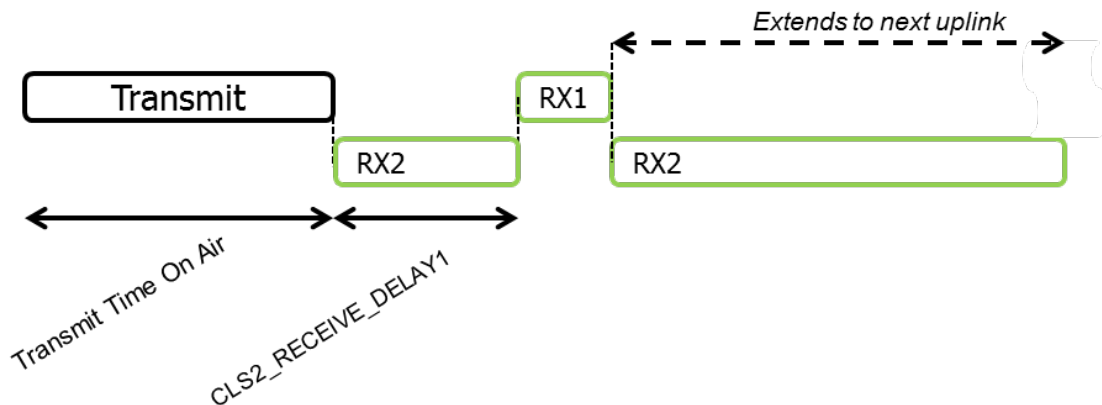


Figure 18: Class C end nodes timing for uplink and downlink transmission

The timing of the operation of class C is depicted in Figure 18. A node operating in Class C mode is always listening to a radio channel with a certain SF for packets coming from the network server, which sends the packets in a totally asynchronous manner. When a node operating in class C needs to send a packet to the network, it operates in way similar to the class A end end-nodes: it send the packets ("transmit" in Figure 18) and then continue listening in the usual channel and SF ("RX2" in Figure 18), switching to the equivalent of the first receive window of class A after *CLS2_RECEIVE_DELAY1* seconds. At the end of this receive window the end node returns to listening to the same radio channel and SF as before, until the transmission of the next packet. As one can see the power consumption of a Class C node is higher than both Class A and Class B nodes. Class C nodes are ideally suited for sensors and - even more - for actuators nodes attached to the main power.

Two categories of end nodes can coexist on a LPWAN network:

- Device Duty cycle limited.
- Device that use LBT/AFA algorithm.

Below are LBT/AFA algorithm details:

- Clear Channel Assessment (CCA) threshold = -83 dBm.
- The minimum CCA interval employed: 5 ms.
- The maximum dead time: 5 ms.
- The minimum transmission off time on the same operating frequency, *Toff* min: 2 s.

Adaptive Frequency Agility is used, i.e. when a channel is detected as occupied a retransmission is tried on a different channel (one of the 8 channels defined per end device) in a random manner. A CCA is performed at any new retry. After 8 retries the communication is aborted waiting for next event.

7.3 Information on relevant standard(s)

For the MAC layer and for many aspects of the upper layers of the LPWAN-CSS systems the reference is the LoRaWAN™ Specification [i.1].

8 Radio spectrum request and justification

The frequency bands in which LPWAN CSS systems are able to operate are the following mostly harmonised SRD bands:

- a) 863 MHz - 870 MHz (h1.3 of ERC Recommendation 70-03 [i.6]).
- b) 868 MHz - 868,6 MHz (h1.4 of ERC Recommendation 70-03 [i.6]).
- c) 868,7 MHz - 869,2 MHz (h1.5 of ERC Recommendation 70-03 [i.6] and sub-band 48 of EC decision 2017/1483 [i.7]).
- d) 869,4 MHz - 869,65 MHz (h1.6 of ERC Recommendation 70-03 [i.6] and sub-band 54 of EC decision 2017/1483 [i.7]).
- e) 869,7 MHz - 870 MHz (h1.7 of ERC Recommendation 70-03 [i.6] and sub-band 56b of EC decision 2017/1483 [i.7]).
- f) 870 MHz - 876 MHz (h2 of ERC Recommendation 70-03 [i.6]).
- g) 915 MHz - 921 MHz (h3 of ERC Recommendation 70-03 [i.6]).
- h) 863 MHz - 868 MHz (Sub-band 84 of EC decision 2017/1483 [i.7]).
- i) 865 MHz - 868MHz (Sub-band 47 of EC decision 2017/1483 [i.7]).

9 Regulations

9.1 Current regulations

Table 12 is extracted from the ERC Recommendation 70-03, Annex 1, regarding the regulation for non-specific SRDs operating in 800 MHz and 900 MHz bands.

Table 12: Extract from the ERC Recommendation 70-03, Annex 1

Frequency Band		Power/Magnetic Field	Spectrum access and mitigation requirements
h1.1	863 MHz - 870 MHz (notes 3 and 4)	25 mW e.r.p.	$\leq 0,1$ % duty cycle or LBT (notes 1 and 5)
h1.2	863 MHz - 870 MHz (notes 3 and 4)	25 mW e.r.p. Power density - 4,5 dBm/100 kHz (note 7)	$\leq 0,1$ % duty cycle or LBT+AFA (notes 1, 5 and 6)
h1.3	863 MHz - 870 MHz (notes 3 and 4)	25 mW e.r.p.	$\leq 0,1$ % duty cycle or LBT+AFA (notes 1 and 5)
h1.4	868,000 MHz - 868,600 MHz (note 4)	25 mW e.r.p.	≤ 1 % duty cycle or LBT+AFA (note 1)
h1.5	868,700 MHz - 869,200 MHz (note 4)	25 mW e.r.p.	$\leq 0,1$ % duty cycle or LBT+AFA (note 1)
h1.6	869,400 MHz - 869,650 MHz	500 mW e.r.p.	≤ 10 % duty cycle or LBT+AFA (note 1)
h1.7	869,700 MHz - 870,000 MHz (note 11)	5 mW e.r.p. 25 mW e.r.p.	No requirement ≤ 1 % duty cycle or LBT+AFA (note 1)
h2	870 MHz - 876 MHz	25 mW e.r.p.	$\leq 0,1$ % duty cycle For ER-GSM protection (873 MHz - 876 MHz, where applicable), the duty cycle is limited to $\leq 0,01$ % and limited to a maximum transmit on-time of 5 ms/1 s
h3	915 MHz - 921 MHz	25 mW e.r.p.	$\leq 0,1$ % duty cycle For ER-GSM protection (918 MHz - 921 MHz, where applicable), the duty cycle is limited to $\leq 0,01$ % and limited to a maximum transmit on-time of 5 ms/1 s

Table 13 is extracted from the Commission Implementing Decision (EU) 2017/1483 [i.7], Official Journal of the European Union 18-8-2017.

Table 13: Extract from the Commission Implementing Decision (EU) 2017/1483 [i.7]

Band no	Frequency band [i]	Category of short-range devices [ii]	Transmit power limit/field strength limit/power density limit [iii]	Additional parameters (channelling and/or channel access and occupation rules) [iv]
84	863 MHz - 868 MHz	Wideband data transmission devices [16]	25 mW e.r.p.	Techniques to access spectrum and mitigate interference that provide at least equivalent performance to the techniques described in harmonised standards adopted under Directive 2014/53/EU [i.11] should be used. Bandwidth: ≤ 1 MHz. Duty cycle [vi]: ≤ 10 % for network access points [26] Duty cycle [vi]: $\leq 2,8$ % otherwise
47	865 MHz - 868 MHz	Non-specific short-range devices [3]	25 mW e.r.p.	Techniques to access spectrum and mitigate interference that provide at least equivalent performance to the techniques described in harmonised standards adopted under Directive 2014/53/EU [i.11] should be used. Alternatively a duty cycle limit [vi] of 1 % may also be used.

9.2 Proposed regulation and justification

For the LPWAN CSS system to be implemented no changes in the regulations are needed.

The bare minimum spectrum for a LPWAN CSS system to be implemented is made of 4 channels, the channel bandwidth being 125 kHz and channel spacing 200 kHz, each as follows:

- 1) 868,1 MHz (uplink and downlink).
- 2) 868,3 MHz (uplink and downlink).
- 3) 868,5 MHz (uplink and downlink).
- 4) 869,525 MHz (downlink only).

However, not only an explosive growth is expected for LPWAN CSS systems but also more requirements are coming, which technically can be satisfied by LPWAN CSS systems if more bandwidth is available. These requirements include:

- a) Dependability: some applications have strict requirements on being sure that the LPWAN CSS messages actually reach their destination with very high probability and are acknowledged quickly.
- b) Low delay applications: in industry 4.0 application the control of production plants and machineries needs low delay communications.

Furthermore, with more bandwidth available the need to densify the network will diminish and so, also due to the competition of more and more operators attracted by the available bands, there will be the possibility to lower the price of the LPWAN CSS connectivity (due to lower CAPEX investments).

For these reasons the new included opportunities in EC Decision 2017/1483 [i.7] are welcomed (e.g. sub-band 84).

Annex A:

Main use cases in different verticals

One of the main applications where LPWAN-CSS technology benefits are key is water metering and flow monitoring, illustrated in Figure A.1.

In this case the communicating devices are mostly located indoors and can even be positioned underground.

The main benefits of this application are:

- Monitoring the water network efficiency between the flows that are going in and out of the network.
- The detection of discrepancies or anomalies.
- The calibration of the hydraulic network model.

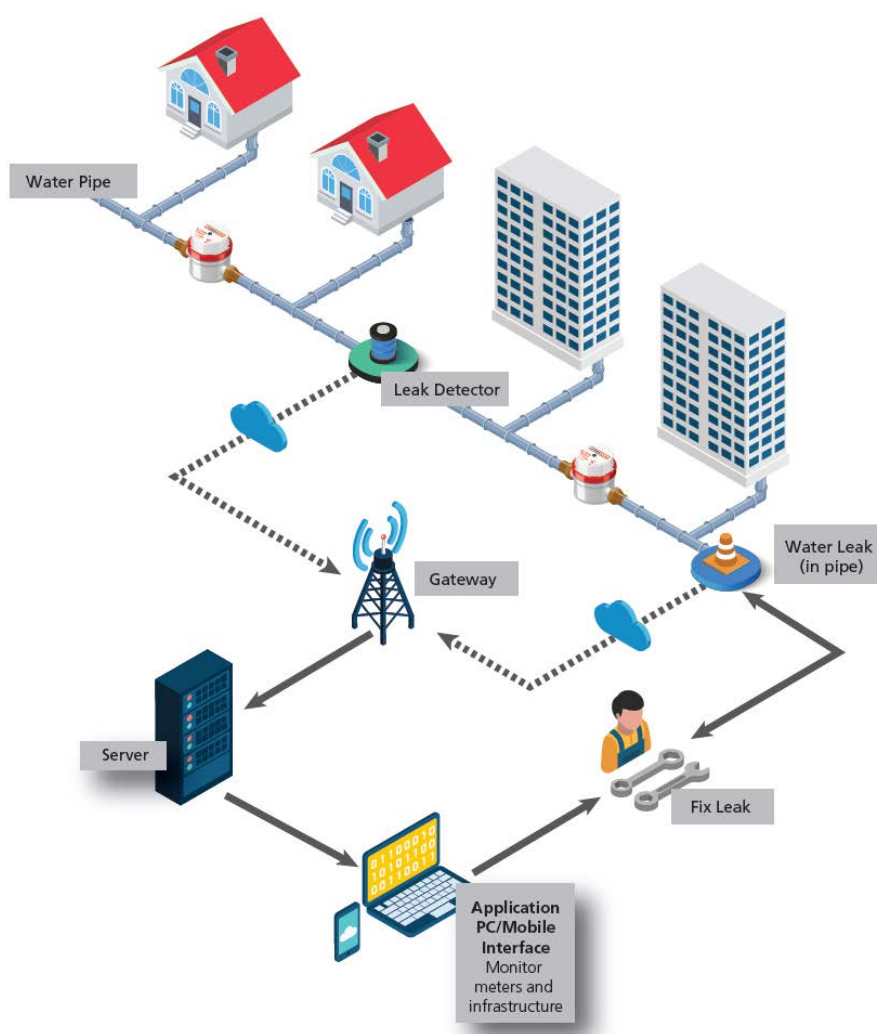


Figure A.1: Water flow monitoring

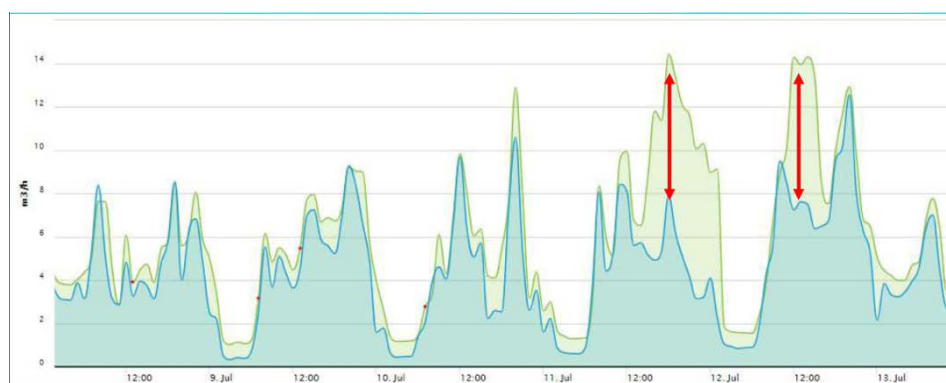


Figure A.2: Water network efficiency and anomaly detection

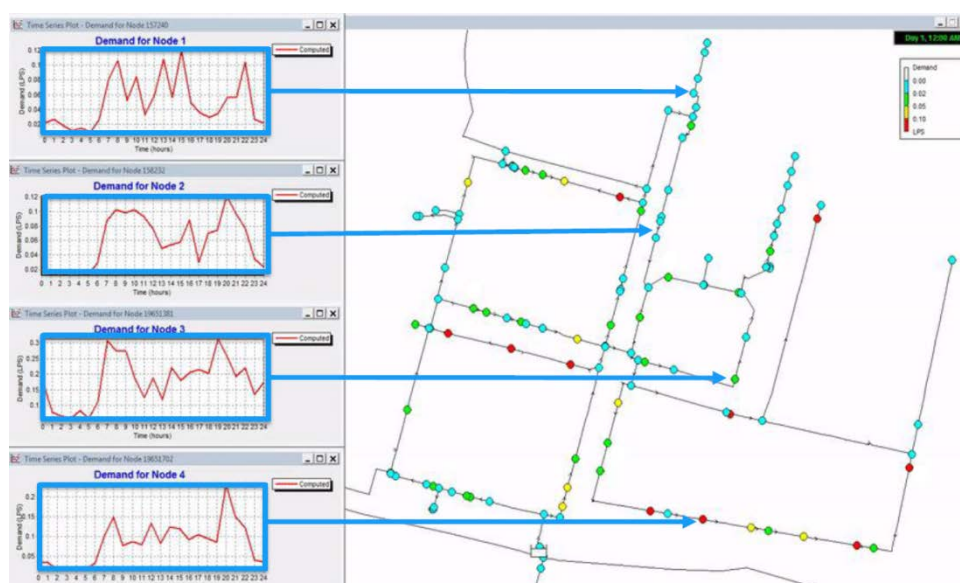


Figure A.3: Hydraulic model calibration

Following water metering, below figures are presenting some of the main vertical applications for which LPWAN-CSS technology is being used.

Within the smart building vertical the main applications are energy savings and occupancy optimization through measurement of values like temperature, humidity, CO₂, presence, door or window opening, light intensity or noise level.

Hospitals are particular buildings where several categories of people and critical processes are on-going. Therefore healthcare is also an important area of application for LPWAN-CSS technology. In addition to the values mentioned above, inventory, gas cylinder level and tracking are very relevant to this application.

Facility Management

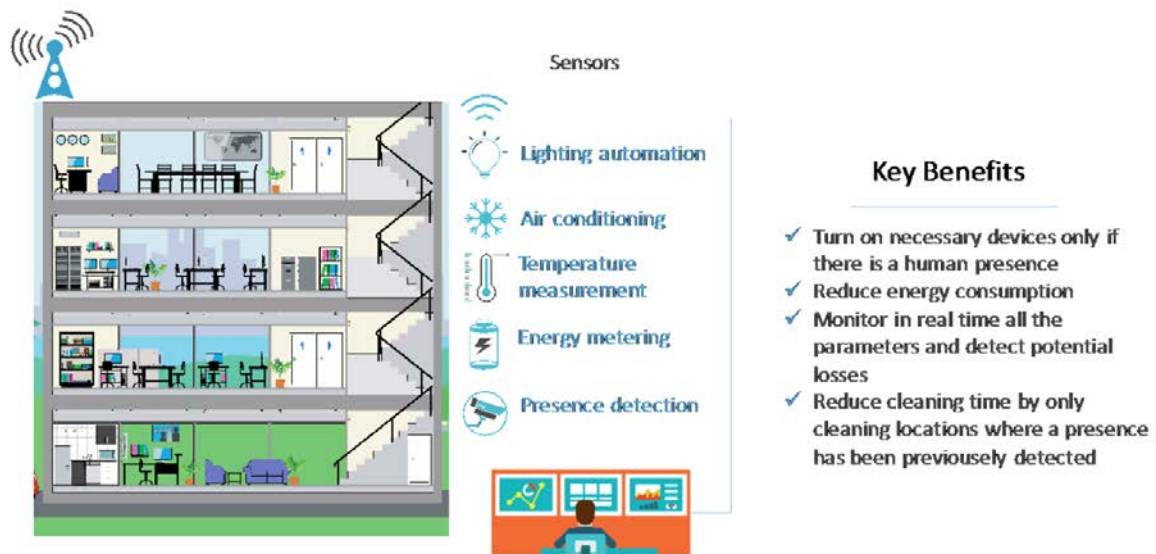


Figure A.4: Smart building vertical

Healthcare Applications

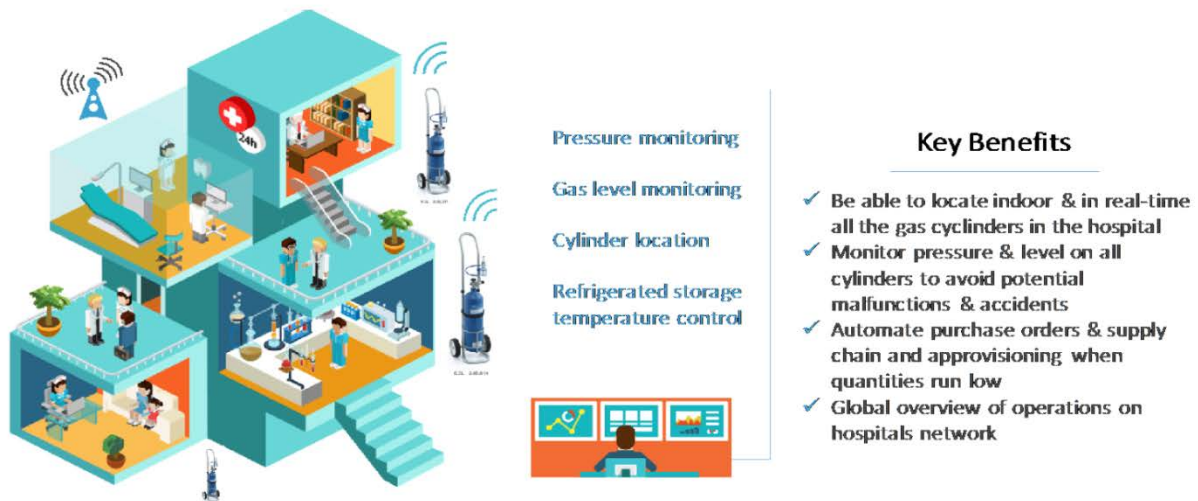


Figure A.5: Healthcare vertical

The other area where LPWAN-CSS technology has a great potential is Smart Industry. Maybe more than for other verticals the flexibility between private and public networks is particularly important here. For a usage that would be limited to a large facility, a private network would certainly be an attractive option while for smaller scattered sites the wide coverage of public networks would be more cost effective. Of course a large industrial company having both kinds of configurations is able to using the roaming capability to benefit from both coverage options at the same time. Among the numerous use cases of this vertical asset tracking is probably the more straightforward.

Factories and Industrial Applications

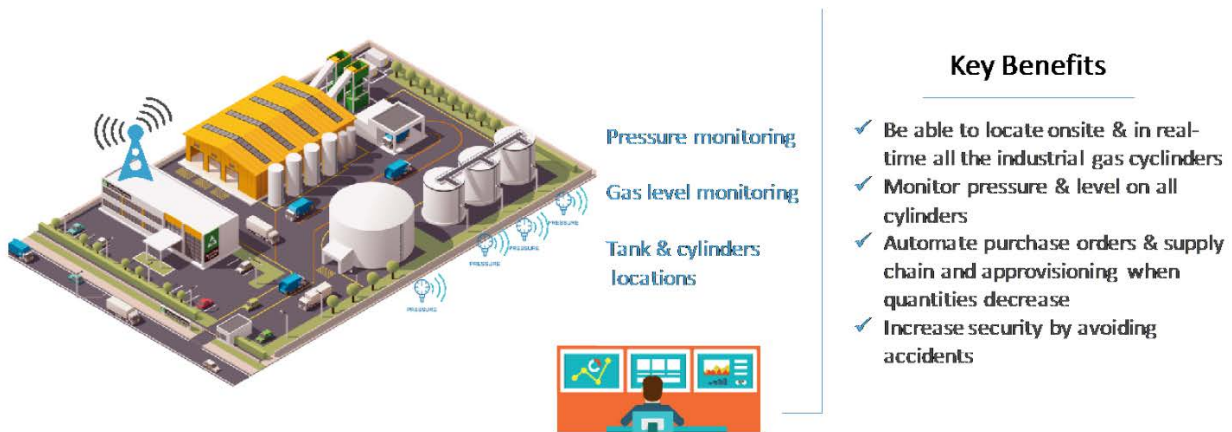


Figure A.6: Industry vertical

Airports are wider than a building and smaller than a city. In terms of use cases they are mixing both categories with some flavour of industrial applications as far as supply chain is concerned. Most of the smart building features apply to airports, from energy savings to passengers flow management. The main smart cities use cases like parking optimization, waste management and lighting control are also relevant to airports. Finally containers or even luggage tracking also helps minimizing the airplane time in the airport which is the overall target.

Airport Services Management

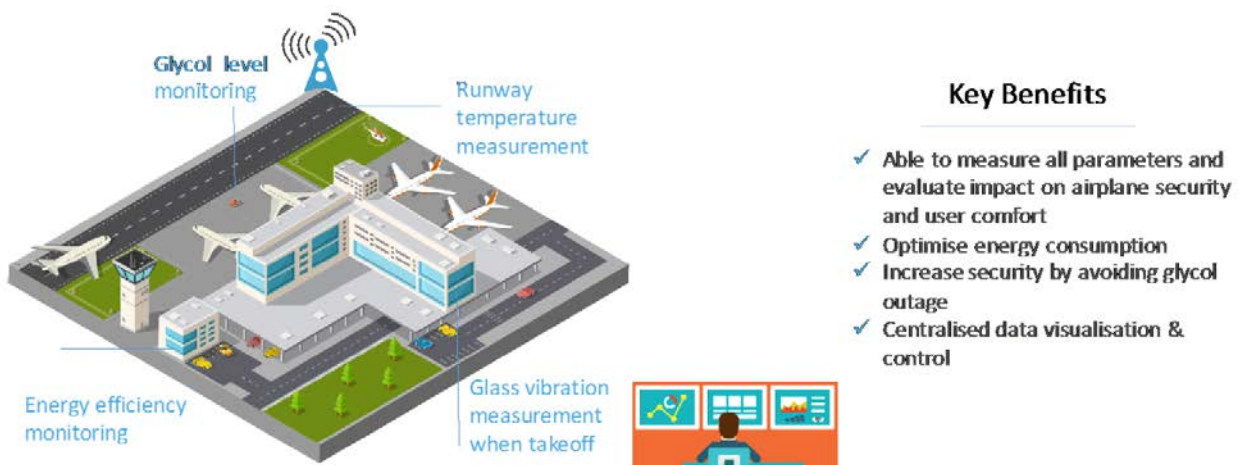


Figure A.7: Airport vertical

Figure A.8 describe the main smart city applications which are smart parking and street lighting.

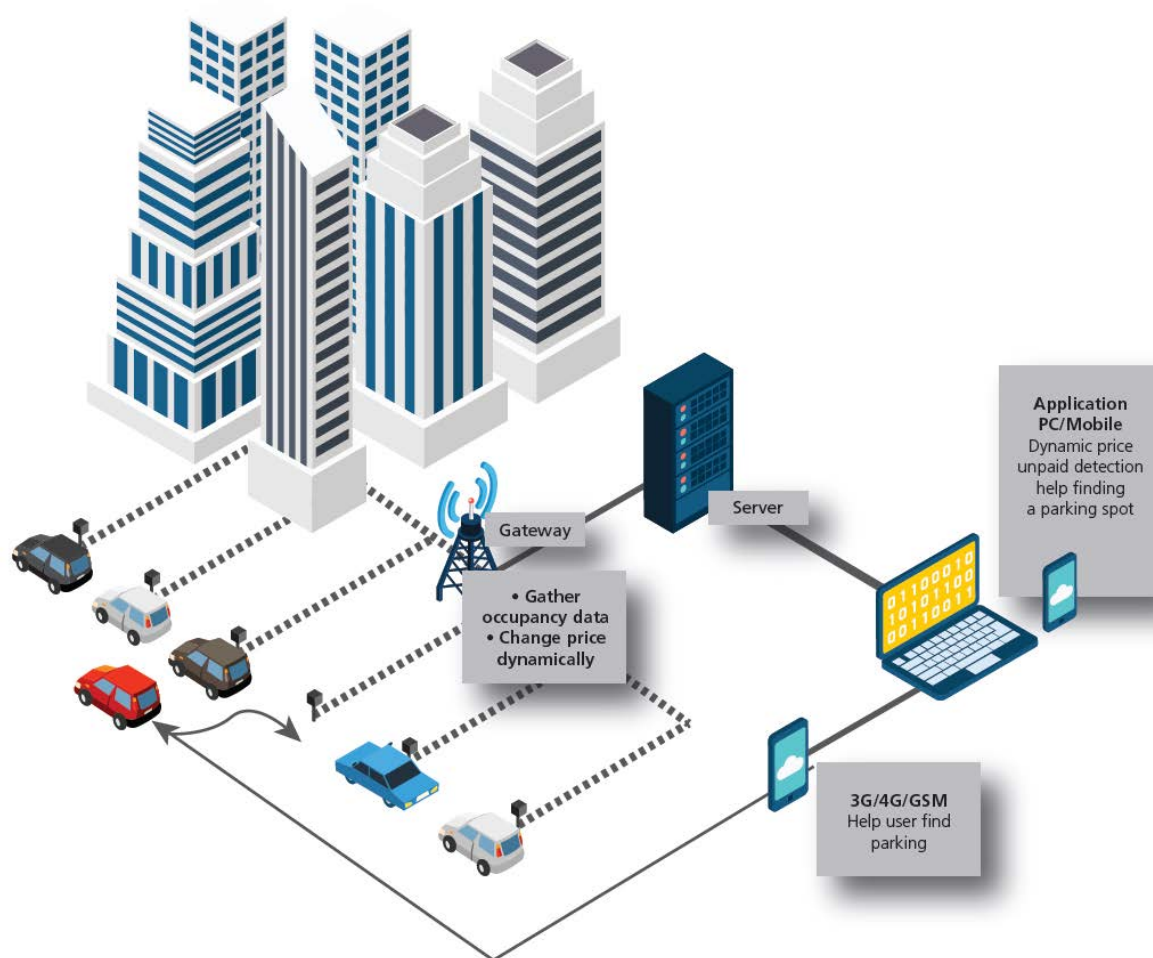


Figure A.8: Smart Parking use case

In this case, the communicating devices are located outdoor and their height can vary from the ground level to several meters high when they are mounted on street lamp posts.

Street Lighting



Key Benefits

- ✓ Lora sensor detects if someone passes by and adapts the light level accordingly
- ✓ Optimized power consumption
- ✓ Monitor in real time the status of all the street lights
- ✓ Easy installation

Figure A.9: Street lighting use case

Leaving the city space, wide area long range coverage is also the opportunity to address agriculture use cases. The two main applications are:

- Water and fertilizer optimization through soil sensors.
- Cattle tracking and health monitoring with ear tags or implants, including calving risks reduction.

Of course, replication from smart building use cases to breeding buildings and industrial supply chain to food production process are also very relevant.

HOW A LoRaWAN-BASED AGRICULTURE AND FOOD PROCESSING SYSTEM WORKS

Semtech LoRa Technology enables connectivity, real-time analytics, reporting, and additional functions such as geolocation.

- 1 LoRa-equipped sensors located in the dirt and in the air on farms can track crop maturation and can help determine harvest times
- 2 Once crops are harvested, LoRa-equipped packing box sensors track transportation timing and environment ensuring that food is kept at the optimal temperature
- 3 These same packing box sensors track environment in storage as the crops wait to be processed and then as they are processed. LoRa-equipped sensors on the production equipment track production volumes and maintenance needs
- 4 Data from all sensors are periodically sent to a LoRa-based gateway
- 5 Gateway sends information to network server where the data is analyzed by an application server
- 6 Application server sends alerts on the crop readiness to farmers and sends environmental and production equipment information to producers via mobile device

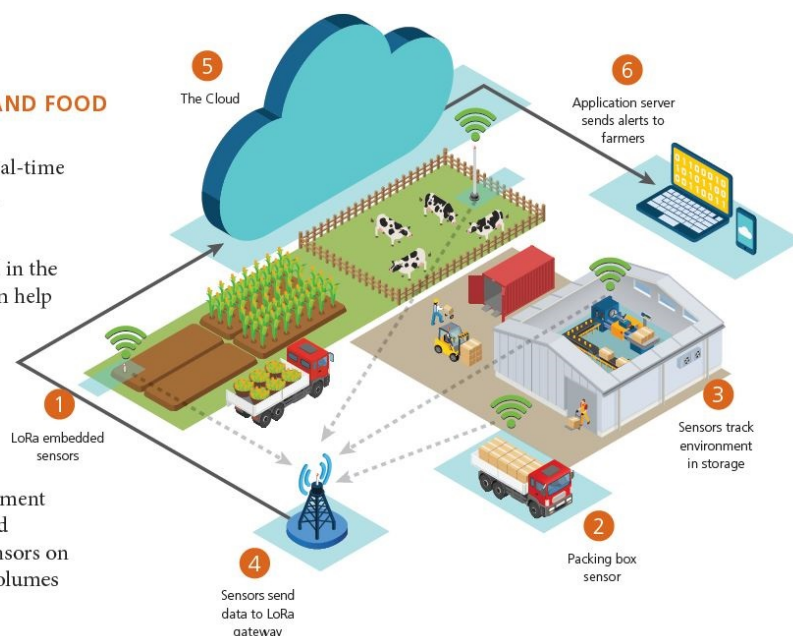


Figure A.10: Agriculture vertical

Annex B: Interference Experimental setup

B.1 General requirements

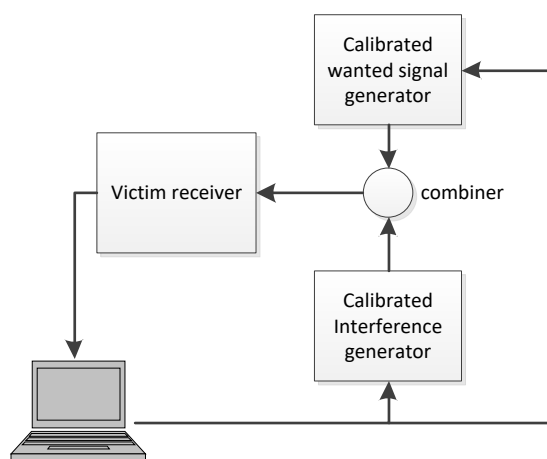


Figure B.1: interference experimental setup

The victim receiver used consists of a Semtech SX1272 evaluation board configured as an FSK receiver. This FSK demodulation performance of this chip is within the industry average. It does not include any specific interference rejection circuitry or algorithm.

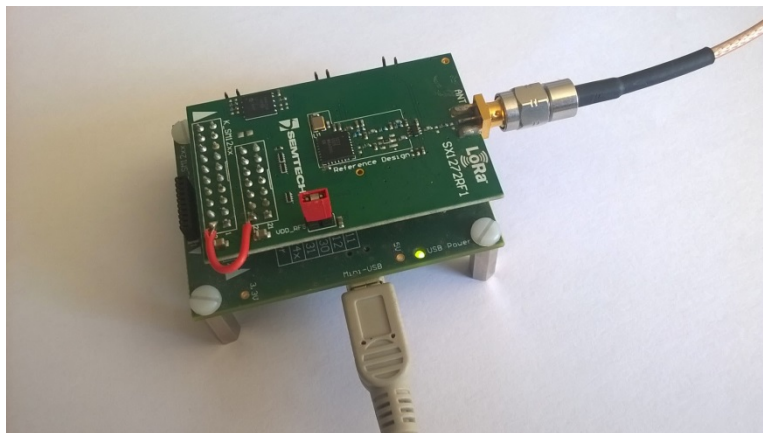


Figure B.2: FSK receiver

That receiver is connected via USB to a laptop to collect packet reception statistics. The interferer and the wanted signal are generated using calibrated professional signal generators (SMBV100A from R&S). Those instruments are also controlled by the laptop. The interferer and wanted signal power are calibrated at the input connector of the evaluation board.

For all measurements in this chapter the victim receiver is configured as follow.

Table B.1

Parameters	Value
Demodulator	FSK
Bit rate	100 kBit/s
Demodulator bandwidth	200 kHz
Center frequency (Fcenter)	860 MHz
Sync word	24 bits C194C1
Receiver sensitivity (@0,1%BER)	-105 dBm

The wanted signal generator is configured as follow.

Table B.2

Parameters	Value
Modulation	GFSK, BT = 0,5
Bit rate	100 kBit/s
Frequency deviation	50 kHz
Center frequency	860 MHz
Sync word	24 bits C194C1
Payload size	16 bytes (random values)
Signal power	-102 dBm (sensi + 3 dB)

The frequency of the interfering signal is swept in the range $F_{center} \pm 300$ kHz with 10 kHz step. For each frequency point the level of the interfering signal is varied until the Packet Error Rate of the victim receiver is 10 %.

All measurement are done with a $\pm 1,5$ dB precision.

B.2 Receiver's AWGN sensitivity

First the receiver's sensitivity without interference is characterized. The wanted signal power is swept, using 500 packets at each step.

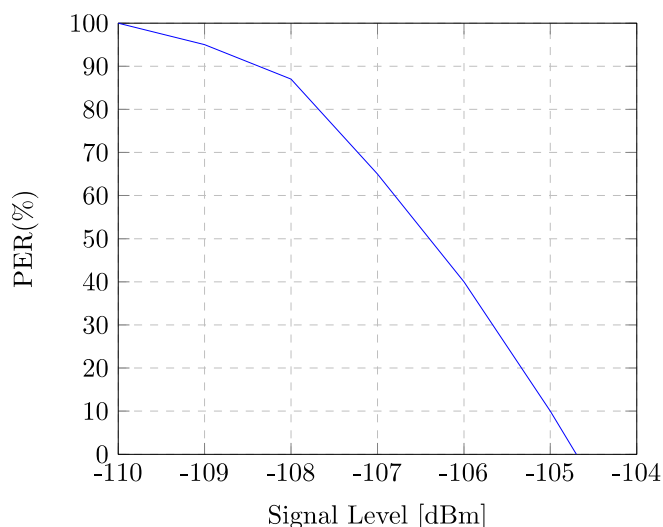


Figure B.3: Packet error rate as a function of the signal level

The receiver exhibits 10 % Packet Error Rate at **-105 dBm** corresponding roughly to 0,1 % Bit Error Rate generally accepted as the definition of a receiver's sensitivity level. It is noted that for a packet of 16 bytes 0,1 % Bit Error Probability corresponds actually to a 13 % Packet error probability:

$$(1 - 10^{-3})^{16 \times 8} = 0,87$$

For all subsequent measurements, the wanted signal power is set to:

$$\text{sensitivity} + 3 \text{ dB} = -102 \text{ dBm}$$

B.3 Continuous wave interference

The interferer is an un-modulated continuous signal with 100 % duty cycle.

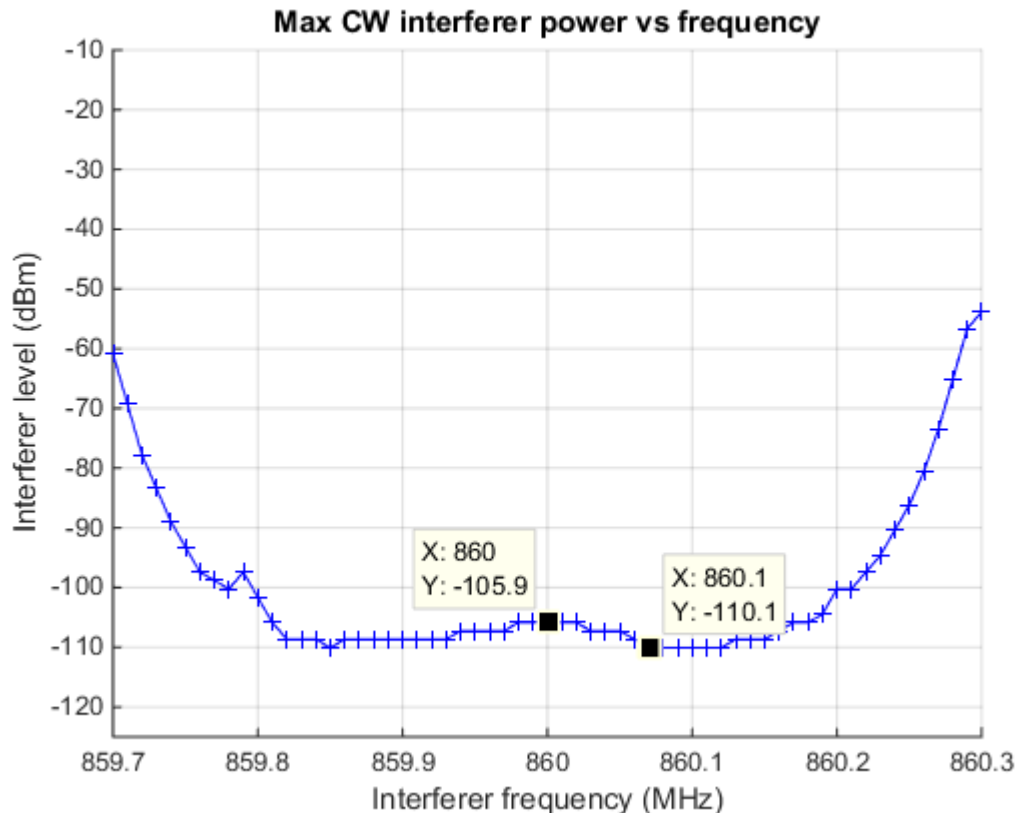


Figure B.4: CW interferer

In the bandwidth of the receiver (from 859,9 MHz to 860,1 MHz), the interferer level tolerated varies between -110 dBm to -106 dBm, for a wanted signal power of -102 dBm. This indicates that the FSK demodulator operates with Signal To Interferer ratio of 4 dB to 8 dB. At 50 % PER and given the modulation parameters (specifically the high modulation index) the theoretical performance of the chip FSK demodulator is close 6 dB SIR. Therefore this measurement confirms the expected theoretical result.

The slight asymmetry of the curve is due to a few kHz frequency offset of the receiver caused by an offset of its crystal timing reference.

B.4 GFSK modulated interferer

Two measurements were performed considering a GFSK modulated interferer:

- 1) 5 kbit/s/25 kHz frequency deviation modulated interferer.
- 2) 100 kbit/s/50 kHz frequency deviation modulated interferer.

In both case the Gaussian filter has a BT of 0,5. The interferer duty cycle is 100 %.

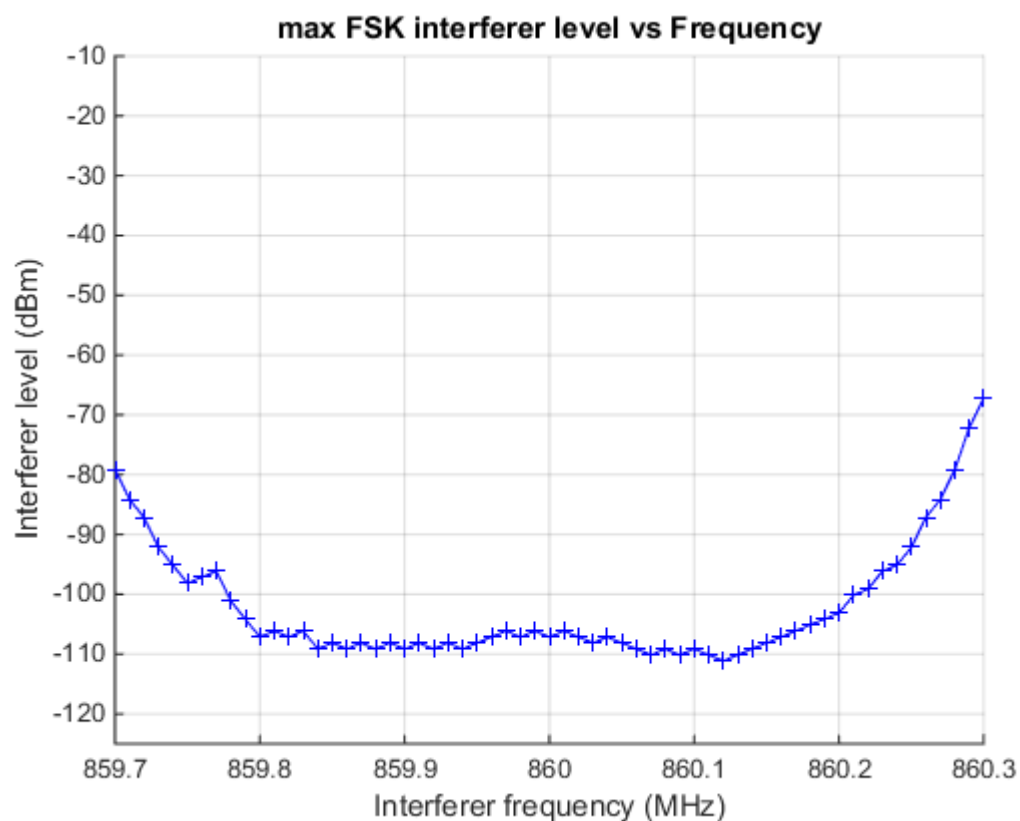


Figure B.5: GFSK modulated interferer, bit rate 5 kbit/s, Fdev=25kHz

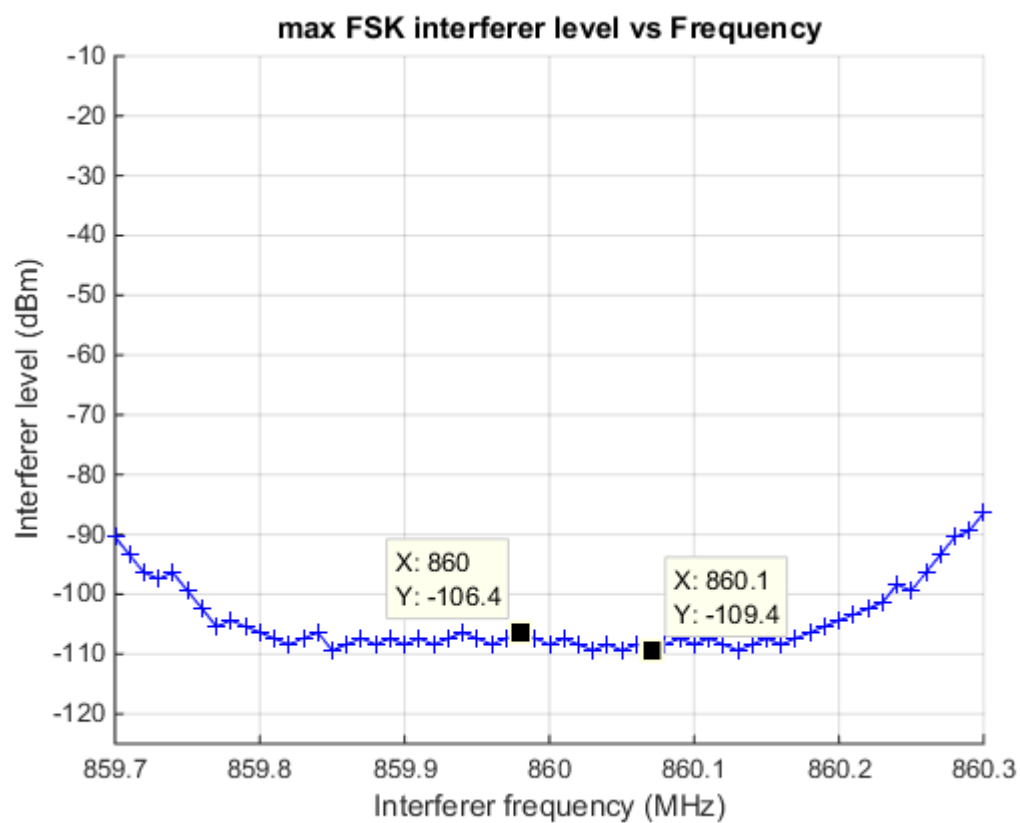


Figure B.6: interferer GFSK 100 kbit/s, Fdev=50kHz

Conclusion: The tolerable interferer level in the victim's receiver bandwidth is nearly identical to the CW case for both modulated FSK interferers (-106 dBm to -109 dBm).

Outside the bandwidth of the victim's receiver (more specifically at ± 300 kHz offset) the receiver tolerance is reduced when the interferer is modulated as summarized in Table B.3.

Table B.3

Interference level @ ± 300 kHz (average of -300 kHz and +300 kHz measured interferer level)	Modulation
-58 dBm	CW
-74 dBm	GFSK 5 kbit/s: Fdev 25 kHz
-89 dBm	GFSK 100 kbit/s: Fdev 50 kHz

This looks very logical. As the modulation bandwidth of the interferer increases the amount of energy spilled inside the victim's receiver bandwidth increases. Therefore one would expect the CW interferer to have the minimum impact on the victim's receiver outside the receiver bandwidth, which is experimentally verified.

B.5 LPWAN-CSS modulated Interferer

In this experiment the interferer is a LPWAN-CSS modulated signal with 125 kHz bandwidth. Both the minimum (SF12) and maximum (SF7) data rate are measured. The interferer duty cycle is 100 %.

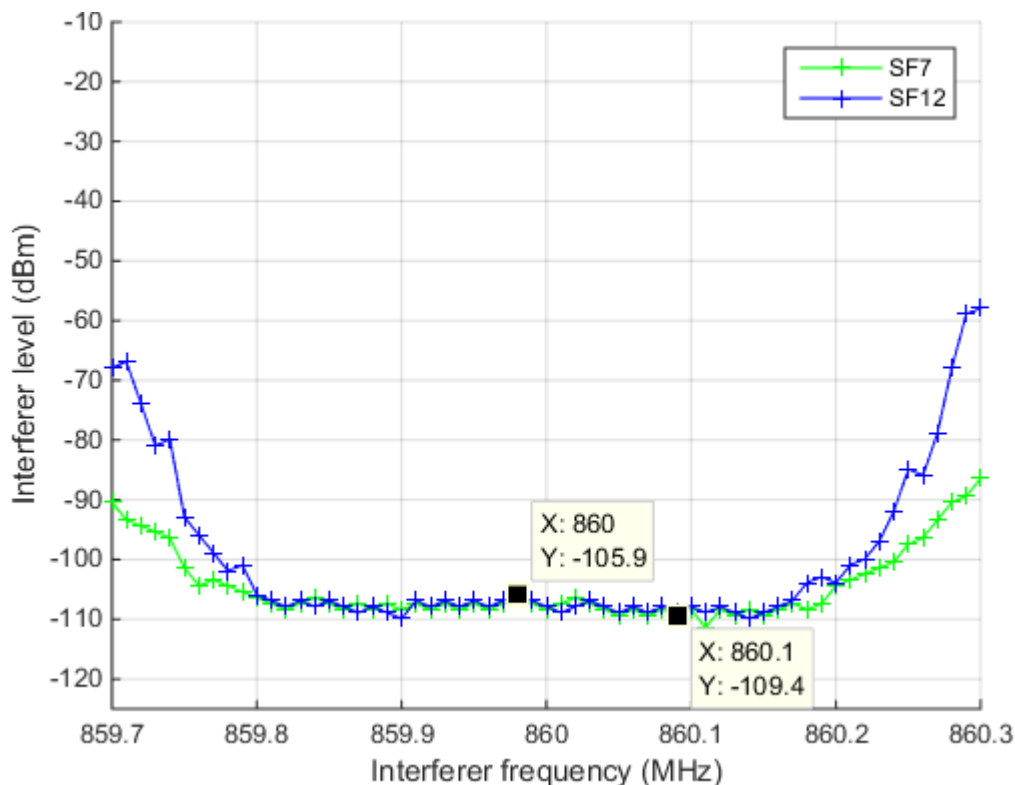


Figure B.7: LPWAN-CSS interferer

Conclusion: once more, inside the receiver's bandwidth the maximum tolerable interference level is identical to FSK or CW modulated interferers (-106 dBm to -109,5 dBm). It is worth noting that the interference tolerance does not vary with the LPWAN-CSS chirp rate used.

Outside the victim's receiver bandwidth one can notice a difference between the two LPWAN-CSS modulated interferers. The lowest data rate interferer (SF12) corresponding to the slowest chirping modulation is better tolerated by the victim receiver's at ± 300 kHz offset. This is because the amount of energy spilled by the LPWAN-CSS modulator outside the 6 dB modulation bandwidth (the 125 kHz modulation bandwidth) increases with the data rate.

The SF7 modulated interferer as a higher energy content at 200 kHz offset and therefore "leaks" inside the victim's receiver bandwidth. The SF12 modulated interferer nearly looks like a CW tone for the victim's receiver at 300 kHz offset. Whereas the impact of the SF7 modulated one is more comparable to the 100 kbit/s GFSK modulated interferer previously measured.

The reason why the LPWAN-CSS modulation with spreading factor 7 (SF7) has a higher energy content outside the 200 kHz offset is that the change in the frequency occurs at a higher speed with respect to the case with spreading factor 12 (SF12): the sweep in the first case (SF7) occurs in 1 ms while in the second case (SF12) occurs in 32 ms and, furthermore, there are significant signal changes every 1 ms instead of 32 ms. That's why the LPWAN-CSS modulation with spreading factor 12 (SF12) looks more like a continuous wave for the victim receiver.

Annex C:

An example of interference measurement with a single LPWAN CSS link and a single periodic pulsed CW interferer

Some co-channel interference measurements are reported in Figure C.1, using as a victim a LPWAN-CSS signal with an occupied bandwidth of 125 kHz bandwidth and SF12. The LPWAN-CSS signal with SF12 is the most robust against interference among the possible LPWAN-CSS signals. The interference which is considered is a pulsed tone, with a constant $T_{\text{on}} / (T_{\text{on}} + T_{\text{off}})$ of 10 %. While on, the interference level is -20 dBm, therefore the average interference level is -30 dBm. For a given measure, T_{on} and T_{off} times are constant, with $T_{\text{off}} = 9 \cdot T_{\text{on}}$. The period $T_{\text{on}} + T_{\text{off}}$ is varied from 0,01 ms to 1s. The results are shown below on Figure C.1. On Figure C.1, the sensitivity level without interferer is also reported, which is almost 20 dB below the noise floor.

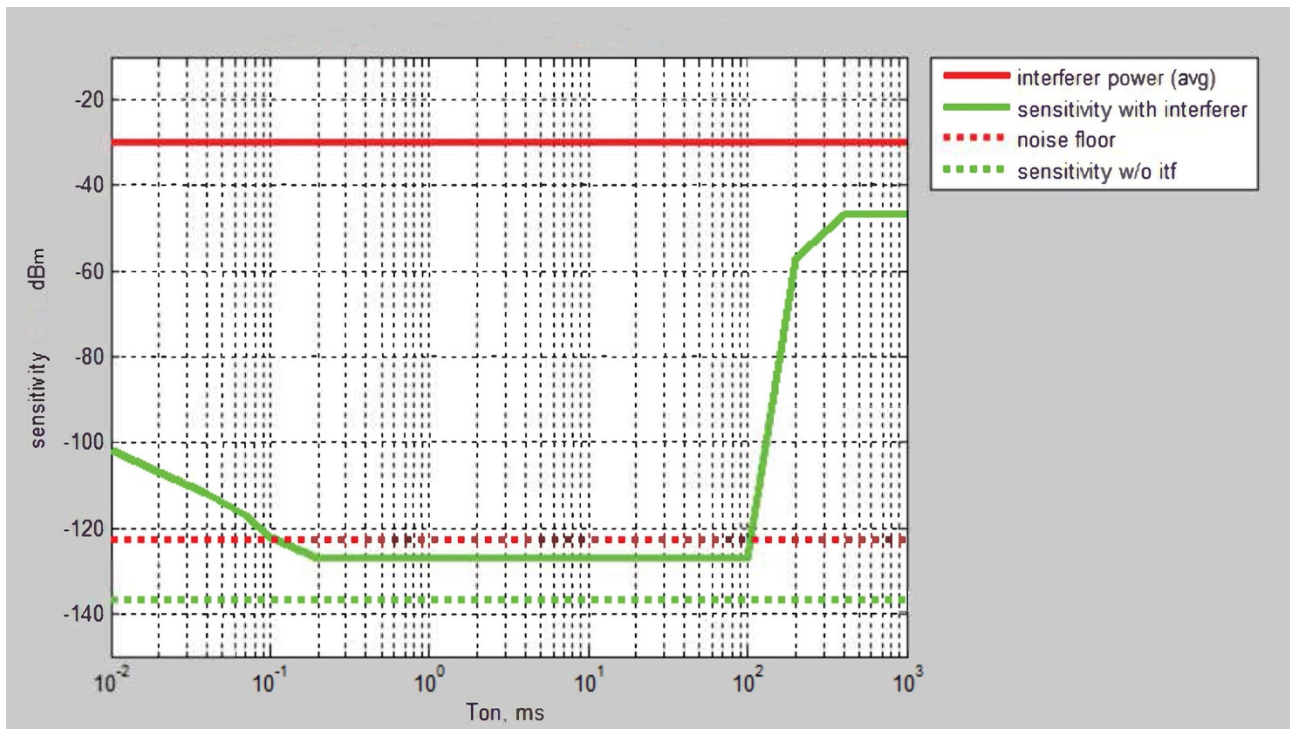


Figure C.1: Pulsed co-channel interferer rejection, SF12

The main thing to notice is that the receiver can tolerate up to -100 dB SIR: the interference can be 100 dB higher than the wanted signal, and still the packet error rate is kept below 10 %. 100 dB higher means that while the interference is present, absolutely no useful signal can be received. This happens when the T_{on} time of the interference is shorter than 100 ms, and longer than 0,1 ms. For higher data rates, these numbers are expected to scale linearly with spreading factor. Here 100 ms correspond roughly to 3 symbols ($4096 \times 3 / 125e3 = 98$ ms). It has to be notice that for SF11 in the same configuration it is assumed that the interference T_{on} should be shorter than 50 ms, 24 ms for SF10, 12 ms for SF9, 6 ms for SF8 and 3 ms for SF7.

When the interference lasts longer than this threshold, its impact is close to a continuous interference, so the SNR table applies.

Annex D: Out of band emission measurements

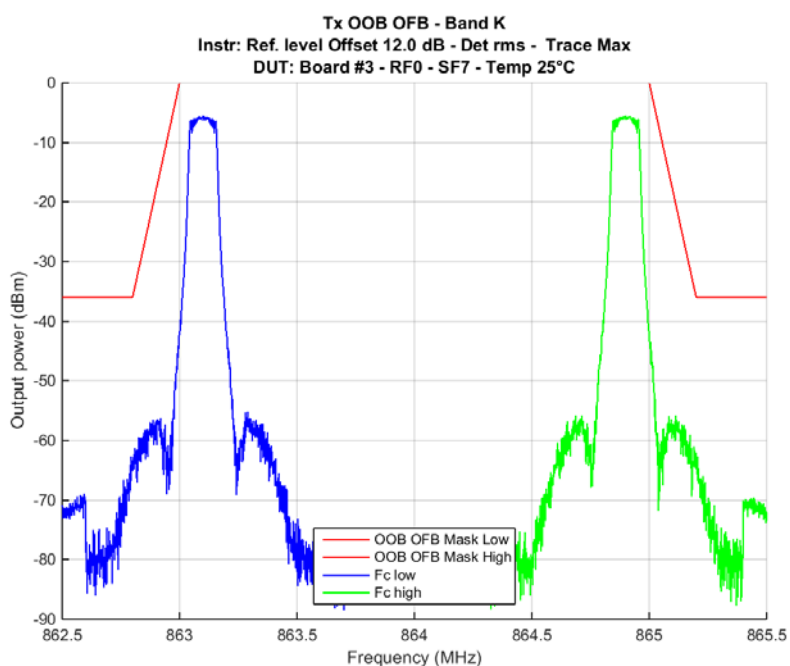


Figure D.1: Tx Out Of Band Emissions - Band K

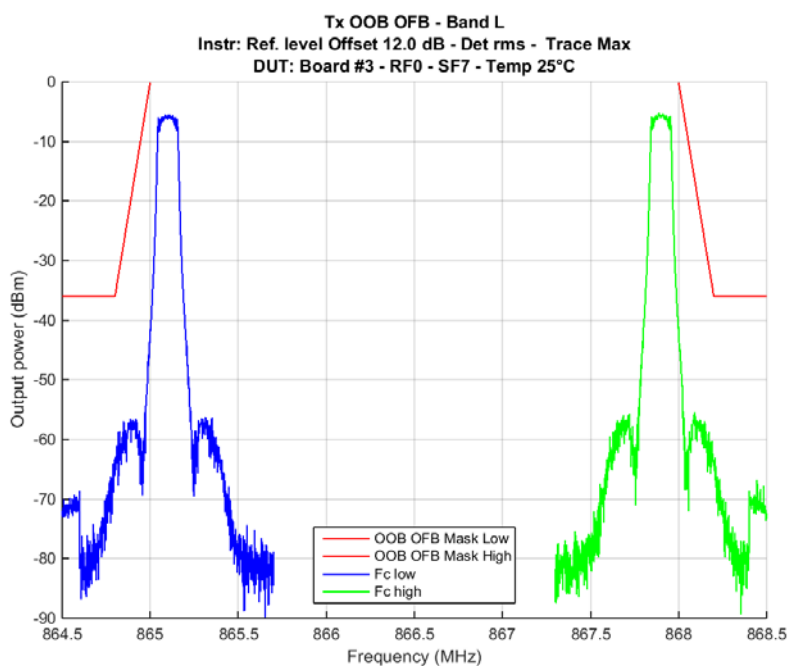


Figure D.2: Tx Out Of Band Emissions - Band L

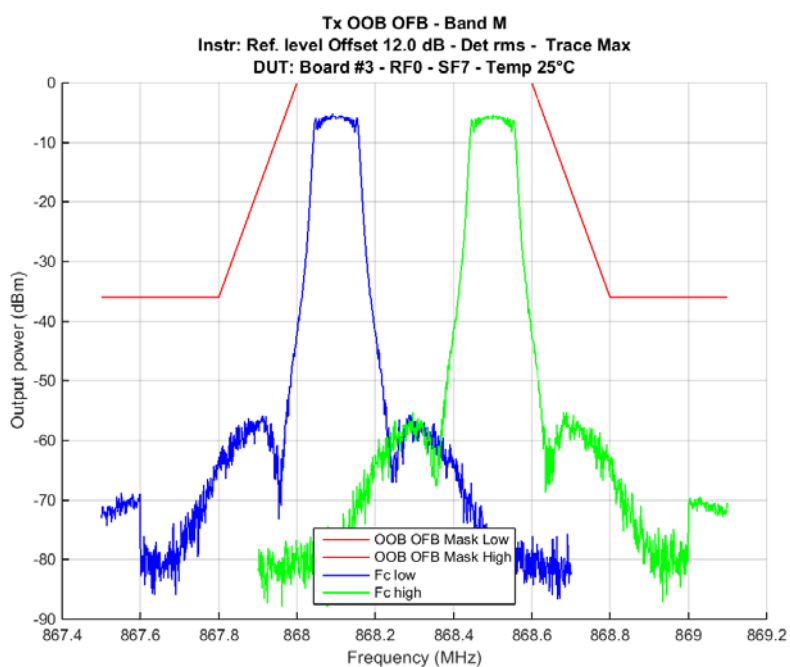


Figure D.3: Tx Out Of Band Emissions - Band M

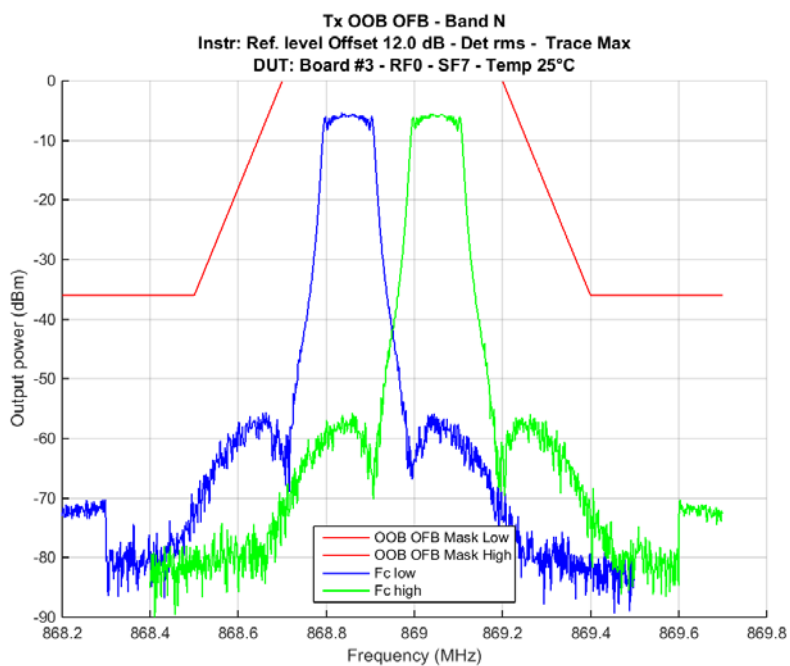


Figure D.4: Tx Out Of Band Emissions - Band N

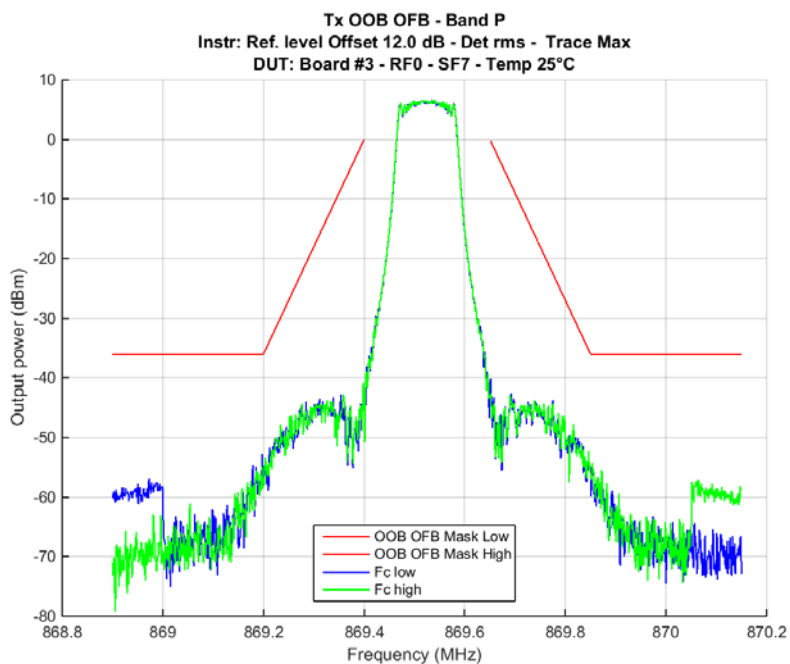


Figure D.5: Tx Out Of Band Emissions - Band P

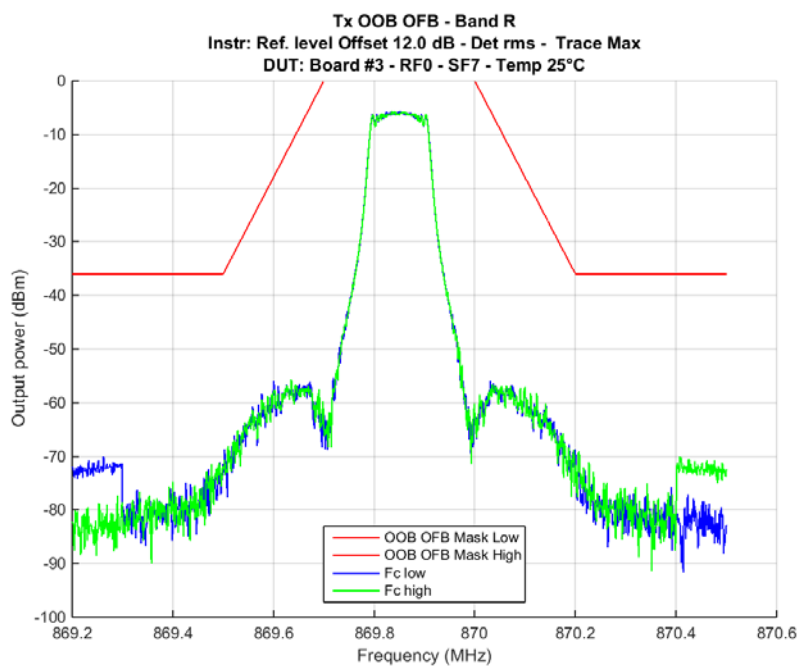


Figure D.6: Tx Out Of Band Emissions - Band R

Annex E: Change History

date	Version	Information about changes
March 2018	1.1.1	first release

History

Document history		
V1.1.1	April 2018	Publication