

$$b) p(t) = \sum_{k=-\infty}^{\infty} C_k e^{j \frac{2\pi k t}{T}}$$

$$C_k = \frac{1}{T} \int_{-T/2}^{T/2} p(t) e^{-j \frac{2\pi}{T} k t} dt = \frac{1}{T} (e^{-j \frac{2\pi}{T} 0}) =$$

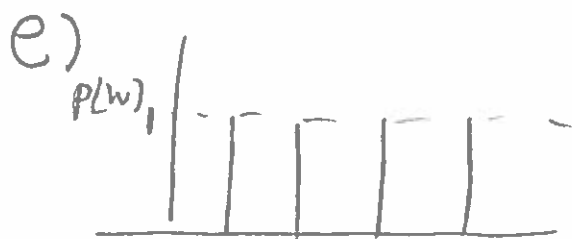
$$\frac{1}{T} \rightarrow p(t) = \frac{1}{T} \sum_{k=-\infty}^{\infty} e^{j \frac{2\pi k t}{T}}$$

$$c) x(t) = \sum_{k=-\infty}^{\infty} C_k e^{j \frac{2\pi}{T} k t}$$

$$X(\omega) = \int_{-\infty}^{\infty} x(t) e^{-j \omega t} dt \rightarrow C_k = \frac{1}{T} X(\omega)$$

$$\rightarrow X(\omega) = T C_k \rightarrow X(\omega) = \int_{-\infty}^{\infty} x(t) e^{-j \omega t} dt$$

$$d) P(\omega) = T C_k = P(\omega) = 1$$



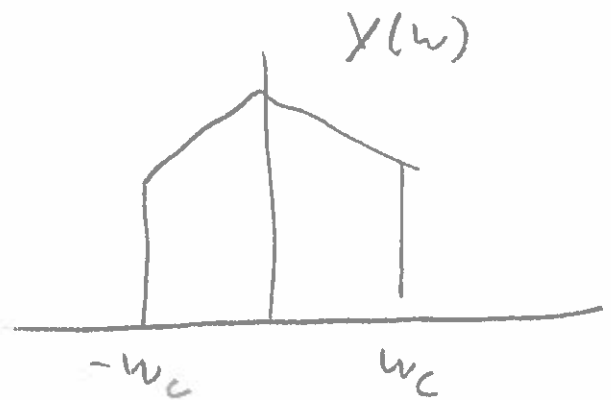
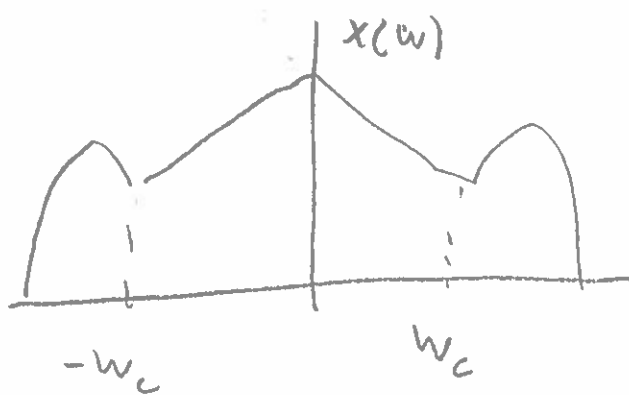
A change in T will change the spacing between the impulses
 As $T \uparrow$ $p(t) \downarrow$ and as $T \downarrow$ $p(t) \uparrow$

2. a) $h(t) = \frac{1}{2\pi} \int_{-w_c}^{w_c} e^{j\omega t} d\omega =$

$$h(t) = \frac{1}{2\pi} \int_{-w_c}^{w_c} e^{j\omega t} d\omega = \frac{1}{2\pi j t} [e^{j\omega_c t} - e^{-j\omega_c t}]$$

$$= \frac{1}{\pi t} \sin(\omega_c t)$$

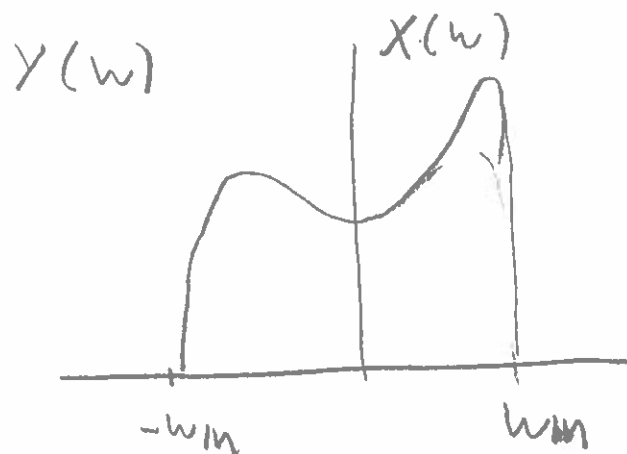
b) convolution in the frequency domain is multiplication



c) This LTI is an ideal low pass filter because it cuts off all of the frequencies above the cut off. There are no partial cut offs

d) attached

3 Let $y(t) = x(t) \cos(\omega_c t)$ where $\omega_c \gg \omega_m$



$$Y(\omega) = X \cdot h(\omega)$$

$$h(\omega) = \mathcal{F}(\cos(\omega_c t))$$

$$= \pi \delta(\omega - \omega_c) + \pi \delta(\omega + \omega_c)$$

