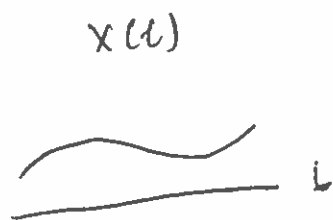


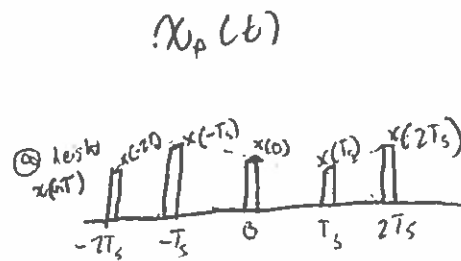
Alexander Crease

# Sig Sys Problem Set 08

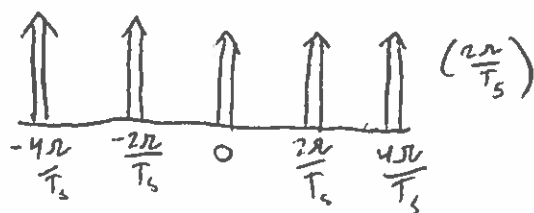
1. a)  $x_p(t)$



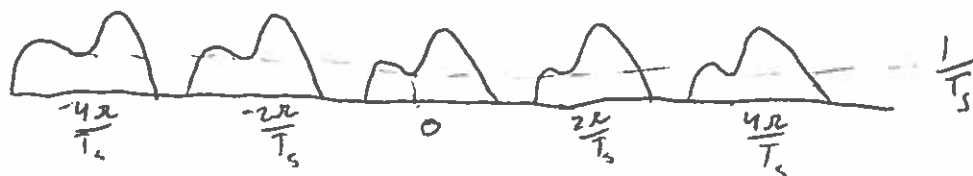
$$x(t) * \sum_{k=-\infty}^{\infty} \delta(t - kT_s) = x_p(t)$$



b)  $P(\omega)$



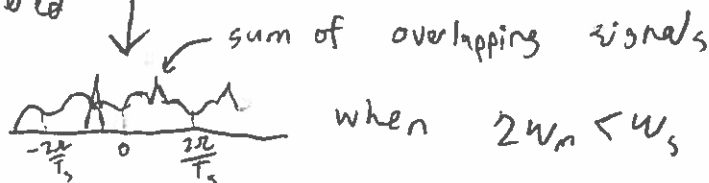
c)  $X_p(\omega)$



d)

$$\frac{2\pi}{T_s} > 2\omega_m \quad \frac{2\pi}{T_s} = \omega_s \rightarrow \text{sampling freq.}$$

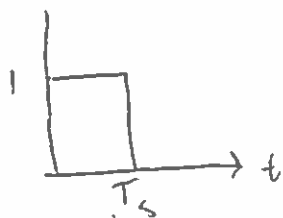
$\omega_s > 2\omega_m$  or else the signals will interfere and combine, and the information won't be able to be recovered



e) You run a ~~low~~ pass filter with a cutoff of  $\omega_m$  to recover the original signal.  
 f) To recover  $x(t)$  from  $x_p(t)$ , you need to infinitely decrease  $T_s$ , infinitely increasing our sample size.  

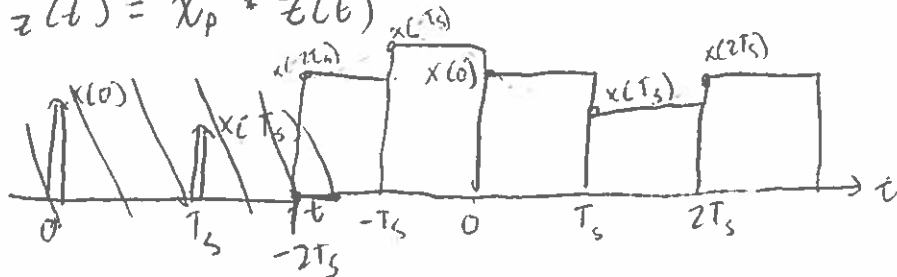
$$x(t) = \lim_{T_s \rightarrow 0} \sum_{k=-\infty}^{\infty} x(kT_s) \delta(t - kT_s)$$

f)  $z(t)$



This rectangular pulse has been considered.

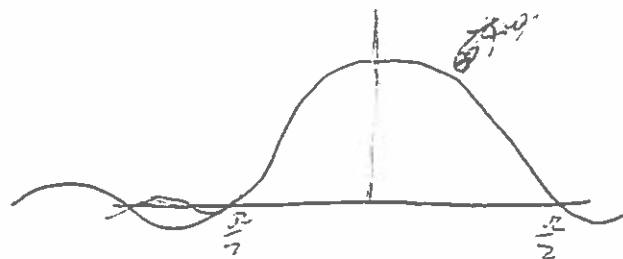
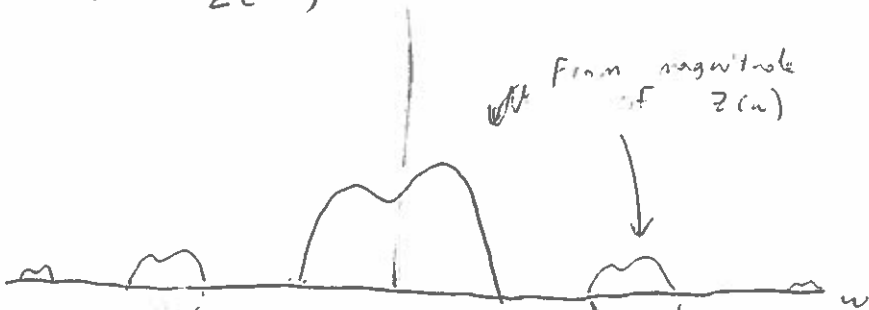
g)  $x_z(t) = x_p * z(t)$



$$Z(\omega) = e^{-j\omega \frac{T_s}{2}} \text{sinc}(\omega \frac{T_s}{2})$$

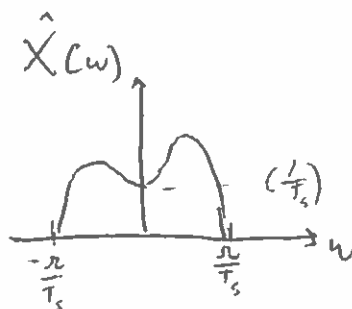
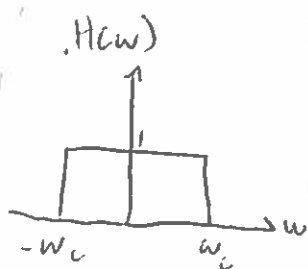
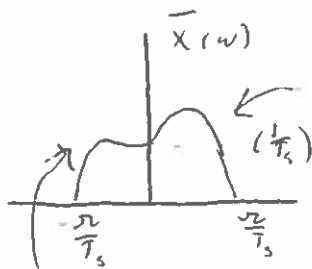
↓

h)  $X_z(\omega)$



i)  $\bar{X}(\omega) = X_z(\omega) H(\omega)$

$\hat{X}(\omega) = X_p(\omega) H(\omega)$



The sides are slightly different because of the sinc function.

on

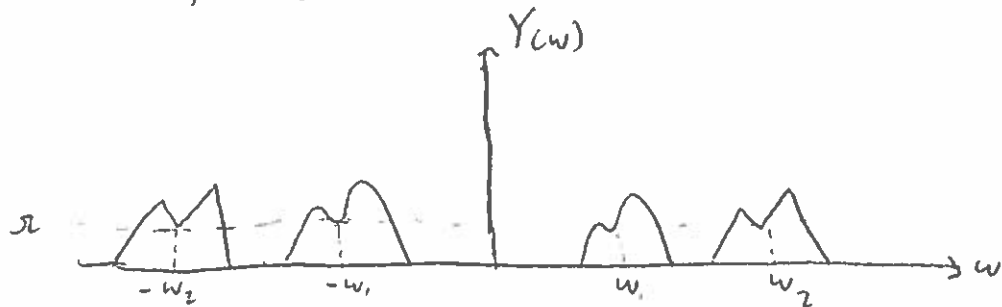
ii) when  $\omega_m = \frac{\pi}{T_s}$ ,  $\frac{\bar{X}(\omega_m)}{\hat{X}(\omega_m)} = e^{-j\omega_m \frac{T_s}{2}} = \boxed{e^{-j\frac{\pi}{2}}}$

2) a)  $y(t) = x_1(t) \cos(\omega_1 t) + x_2(t) \cos(\omega_2 t)$

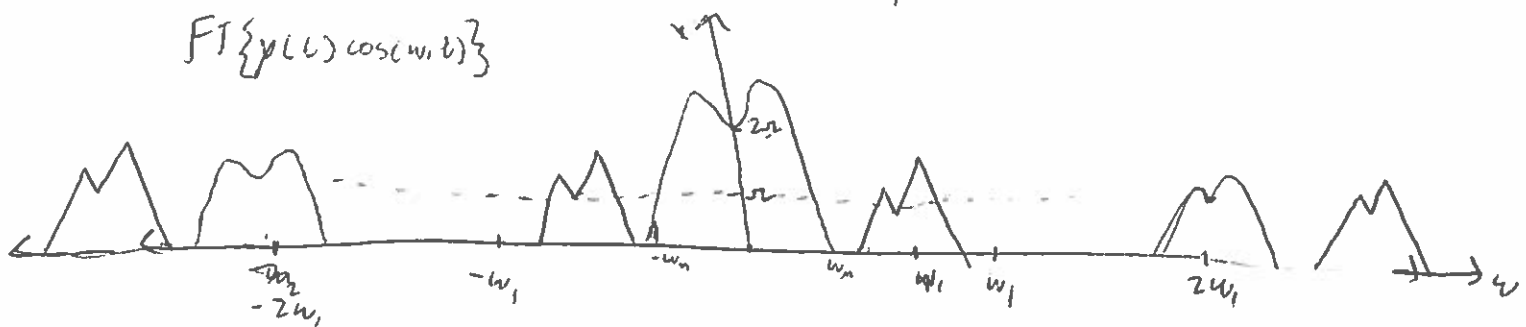
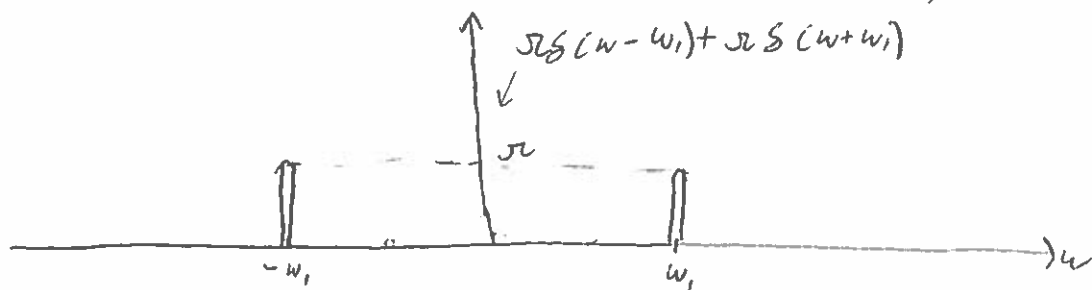
$$Y(\omega) = X_1(\omega) * (\pi \delta(\omega - \omega_1) + \pi \delta(\omega + \omega_1)) + X_2(\omega) * (\pi \delta(\omega - \omega_2) + \pi \delta(\omega + \omega_2))$$

$$\omega_1 + 2\omega_m < \omega_2$$

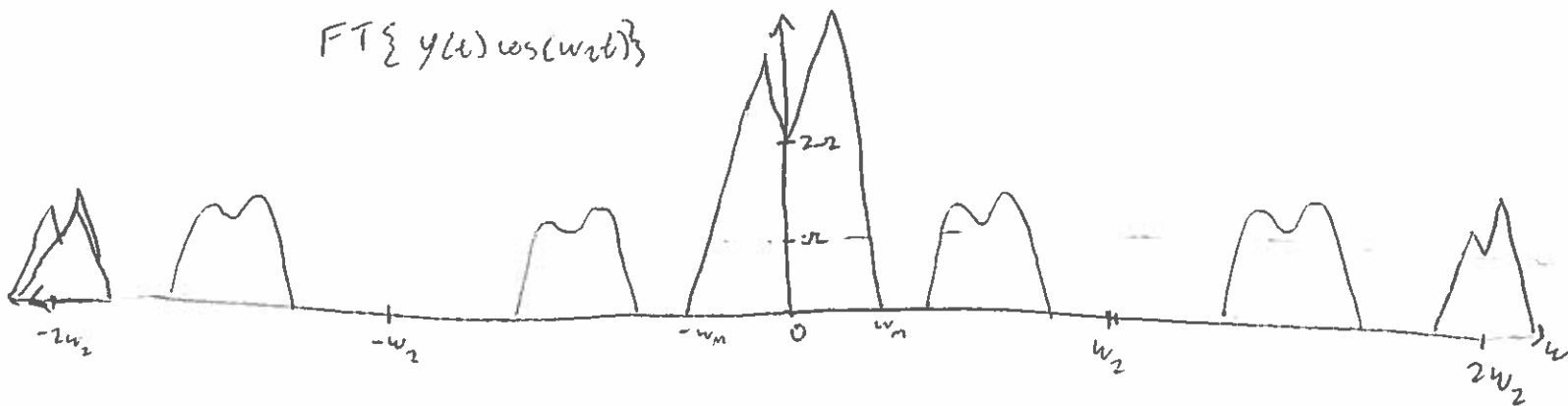
$$\omega_1 - \omega_2 < -2\omega_m$$



b)  $y(t) \cos(\omega_1 t) = Y(\omega) * (\pi \delta(\omega - \omega_1) + \pi \delta(\omega + \omega_1))$

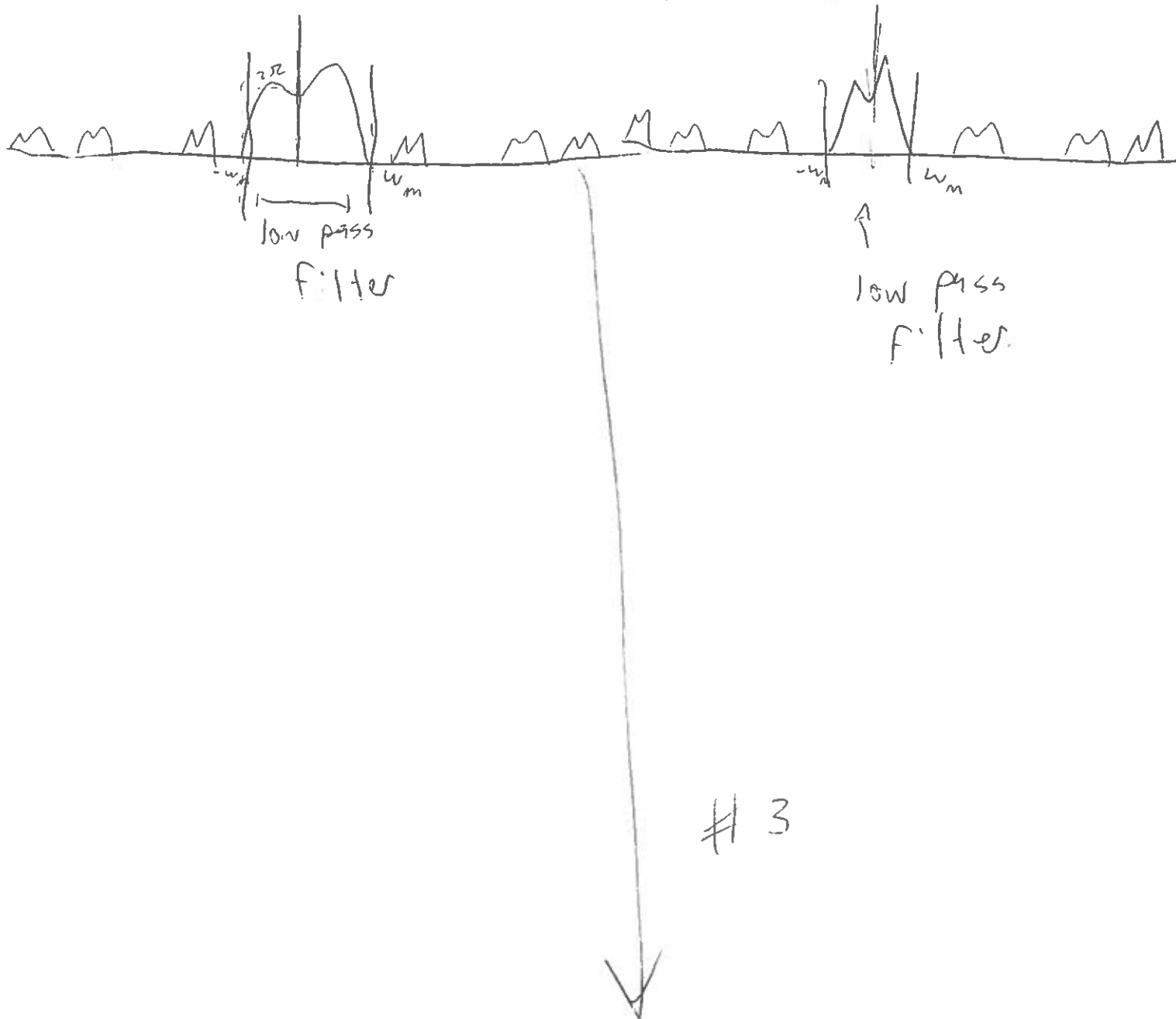


$$FT\{y(t) \cos(\omega_2 t)\}$$



c) You would obtain  $x_1(t)$  using a ~~low~~ <sup>low</sup> pass filter at  $\omega_m$  ~~from  $-\omega_m$  to  $\omega_m$~~  on  $X_1(\omega)$ , and a ~~low~~ <sup>low</sup> pass filter similarly ~~from  $-\omega_m$  to  $\omega_m$~~  on  $X_2(\omega)$ , and then do an IFT to convert back to the time domain.

You then need to divide by  $2\pi$ .



$$3) \quad i(t) = C \frac{d}{dt} V_{out}(t)$$

$$V_L(t) = L \frac{d}{dt} i(t)$$

$$a) \quad V_{in} = V_R + V_L + V_{out}$$

$$V_{in} = i(\omega)R + L \frac{d}{dt} i(t) + V_{out}$$

$$V_{in} = RC \frac{d}{dt} V_{out}(t) + LC \frac{d^2}{dt^2} V_{out} + V_{out}$$

$$\boxed{V_{in} = LC V_{out}'' + RC V_{out}' + V_{out}}$$

$$b) \quad V_{in}(\omega) = j^2 \omega^2 LC V_{out}(\omega) + j\omega RC V_{out}(\omega) + V_{out}(\omega)$$

$$V_{in}(\omega) = V_{out}(\omega) [j^2 \omega^2 LC + j\omega RC + 1]$$

$$j^2 = -1$$

$$\frac{V_{in}(\omega)}{V_{out}(\omega)} = H(\omega) = \frac{1}{-j\omega^2 LC + j\omega RC + 1}$$

$$\boxed{\frac{V_{out}(\omega)}{V_{in}(\omega)} = H(\omega) = \frac{1}{-j\omega^2 LC + j\omega RC + 1}}$$

$$c) \quad |H(\omega)| = \frac{1}{\sqrt{(j\omega RC)^2 + (-\omega^2 LC + 1)^2}} = \frac{1}{\sqrt{j^2 \omega^2 (RC)^2 + (\omega^4 (LC)^2 - 2\omega^2 LC + 1)}}$$

$$\boxed{|H(\omega)| = \frac{1}{\sqrt{-\omega^2 R^2 C^2 + \omega^4 L^2 C^2 - 2\omega^2 LC + 1}}}$$

$$\frac{1}{\sqrt{\omega^2 (-R^2 C^2 + \omega^2 L^2 C^2 - 2LC) + 1}}$$

$$d) |H(\omega)| = \frac{1}{\sqrt{-\omega^2 R^2 C^2 + \omega^4 L^2 C^2 - 2\omega^2 LC + 1}}$$

$$|H(\omega)| = (-\omega^2 R^2 C^2 + \omega^4 L^2 C^2 - 2\omega^2 LC + 1)^{-1/2}$$

$$\frac{d|H(\omega)|}{d\omega} = - \frac{(-\omega^2 R^2 C^2 + \omega^4 L^2 C^2 - 2\omega^2 LC + 1)^{-3/2}}{2} (-2\omega R^2 C^2 + 4\omega^3 L^2 C^2 - 2LC)$$

$$\frac{d|H(\omega)|}{d\omega} = \frac{(2\omega R^2 C^2 - 4\omega^3 L^2 C^2 + 2\omega LC)}{2(-\omega^2 R^2 C^2 + \omega^4 L^2 C^2 - 2\omega^2 LC + 1)^{3/2}}$$

$$= \frac{\omega R^2 C^2 - 2\omega^3 L^2 C^2 + \omega LC}{(-\omega^2 R^2 C^2 + \omega^4 L^2 C^2 - 2\omega^2 LC + 1)^{3/2}}$$

$$\frac{\omega R^2 C^2 - 2\omega^3 L^2 C^2 + \omega LC}{\omega} = 0$$

$$R^2 C^2 - 2\omega^2 L^2 C^2 + 2LC = 0$$

$$\frac{R^2 C^2 + 2LC}{2L^2 C^2} = \omega^2$$

$$\frac{R^2 C + 2L}{2L^2} = \omega^2$$

Because of the  $\omega^2$  &  $\omega^4$  terms in the original magnitude function, both the + & - of this solution give equal values of  $|H(\omega)|$

$$\omega = \pm \sqrt{\frac{R^2 C + 2L}{2L^2}}$$

$$\begin{aligned}
 2) H(\omega) &= \frac{1}{- \omega^2 LC + 1 + j\omega RC} = \frac{1}{1 - \omega^2 LC + j\omega RC} \\
 &= \frac{1}{(1 - \omega^2 LC) + j\omega RC} \left( \frac{(1 - \omega^2 LC) - j\omega RC}{(1 - \omega^2 LC) - j\omega RC} \right)
 \end{aligned}$$

$$\frac{1 - \omega^2 LC - j\omega RC}{1 - 2\omega^2 LC + \omega^4 L^2 C^2 + \omega^2 R^2 C^2} = \frac{1 - \omega^2 LC - j\omega RC}{1 - 2\omega^2 LC + \omega^4 L^2 C^2 + \omega^2 R^2 C^2}$$

$$\phi = \arctan \left( \frac{j\omega RC}{1 - \omega^2 LC} \right)$$

graphs  
down  
here

e) **Matlab Code:**

```
C = 10^-7;  
L = 10^-2;  
R = 400;  
J = sqrt(-1);  
H = tf([1],[-L*C J*R*C 1]);  
bode(H)
```

