











AFE4490



SBAS602H - DECEMBER 2012 - REVISED OCTOBER 2014

AFE4490 Integrated Analog Front-End for Pulse Oximeters

Features

- Fully-Integrated Analog Front-End for Pulse Oximeter Applications:
 - Flexible Pulse Sequencing and Timing Control
- Transmit:
 - Integrated LED Driver (H-Bridge, Push, or Pull)
 - 110-dB Dynamic Range Across Full Range (Enables Low Noise at Low LED Current)
 - LED Current:
 - Programmable Ranges of 50 mA, 75 mA, 100 mA, 150 mA, and 200 mA. Each with 8-Bit Current Resolution
 - Low Power:
 - 100 μA + Average LED Current
 - LED On-Time Programmability from (50 µs + Settle Time) to 4 ms
 - Independent LED2, LED1 Current Reference
- Receive Channel with High Dynamic Range:
 - Input-Referred Noise: 50 pA_{RMS} (at 5-μA PD Current)
 - 13.5 Noise-Free Bits (at 5-µA PD Current)
 - Analog Ambient Cancellation Scheme with Selectable 1-µA to 10-µA Ambient Current
 - Low Power: < 2.3 mW at 3.0-V Supply
 - Rx Sample Time: 50 µs to 4 ms
 - I-V Amplifier with Seven Separate LED2 and LED1 Programmable Feedback R and C Settinas
 - Integrated Digital Ambient Estimation and Subtraction
- Integrated Fault Diagnostics:
 - Photodiode and LED Open and **Short Detection**
 - Cable On or Off Detection
- Supplies:
 - Rx = 2.0 V to 3.6 V
 - Tx = 3.0 V or 5.25 V
- Package: Compact VQFN-40 (6 mm x 6 mm)
- Specified Temperature Range: -40°C to 85°C

2 Applications

- Medical Pulse Oximeter Applications
- **Industrial Photometry Applications**

Description

The AFE4490 is a fully-integrated analog front-end (AFE) that is ideally suited for pulse-oximeter applications. The device consists of a low-noise receiver channel with a 22-bit analog-to-digital converter (ADC), an LED transmit section, and diagnostics for sensor and LED fault detection. The device is a very configurable timing controller. This flexibility enables the user to have complete control of the device timing characteristics. To ease clocking requirements and provide a low-jitter clock to the device, an oscillator is also integrated that functions from an external crystal. The device communicates to an external microcontroller or host processor using an SPI™ interface.

The device is a complete AFE solution packaged in a single, compact VQFN-40 package (6 mm × 6 mm) and is specified over the operating temperature range of -40°C to 85°C.

Device Information⁽¹⁾

PART NUMBER	PACKAGE	BODY SIZE (NOM)
AFE4490	VQFN (40)	6.00 mm × 6.00 mm

(1) For all available packages, see the orderable addendum at the end of the datasheet.

Simplified Schematic

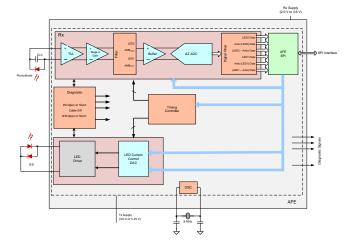




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4 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

Cr	nanges from Revision G (June 2014) to Revision H	Page
•	Changed V _(ESD) parameter specification values in Absolute Maximum Ratings table	8
<u>.</u>	Updated AFE Register Description section to current standards: added legend and bit settings to each bit register	<u>5</u> 9
Cł	nanges from Revision F (October 2013) to Revision G	Page
•	Added Applications and Implementation, Power Supply Recommendations, and Layout sections	1
•	Changed sub-bullet of <i>Transmit</i> Features bullet	1
•	Changed second sub-bullet of Integrated Fault Diagnostics Features bullet	1
•	Changed VCM row in Pin Functions table: changed INM to INN in VCM description	<mark>7</mark>
•	Changed Absolute Maximum Ratings table: changed first five rows and added TXP, TXN pins row	8
•	Added Handling Ratings table	8
•	Changed I-V Transimpedance Amplifier, $V_{O(shield)}$ parameter: changed test conditions and added minimum and maximum specifications	11
•	Changed Example value for rows t, t ₂ , t ₄ , t ₅ , t ₇ , t ₁₁ , t ₁₃ , t ₁₅ , t ₁₇ , t ₁₉ , t ₂₂ , t ₂₄ , t ₂₆ , and t ₂₈ in Table 2	
•	Added footnote 2 to Table 2	
•	Added footnote 2 to Figure 63	37
•	Added footnote 2 to Figure 64	38
•	Changed INN pin name in Figure 76	
•	Changed INM to INN throughout Table 3	51
•	Added STAGE2EN1 and STG2GAIN1[2:0] in TIAGAIN register	57
•	Changed STAGE2EN to STAGE2EN2 and STG2GAIN[2:0] to STG2GAIN2[2:0] in TIA_AMB_GAIN register	57
•	Added last two sentences to NUMAV[7:0] description in CONTROL1: Control Register 1	72



Cl	nanges from Revision E (October 2013) to Revision F	Page
•	Changed LED_DRV_SUP parameter in Recommended Operating Conditions table	9
•	Changed TXM to TXN in V _{LED} footnote of Recommended Operating Conditions table	9
•	Changed V _{LED} footnote and added V _{HR} footnote to Recommended Operating Conditions table	9
•	Changed Figure 77 (changed TXP and TXN pin names, deleted LED 1 and LED 2 pin names)	<u>5</u> 0
•	Changed Table 6 (added V _{HR} columns to table)	76
Cł	nanges from Revision D (May 2013) to Revision E	Page
•	Changed 2.3 mA to 2.3 mW in 4th sub-bullet and changed 250 µs to 4 ms in 5th sub-bullet of Receive Channel with High Dynamic Range Features bullet	
•	Changed Rx, Tx supplies and deleted 5-V supply from front-page graphic	1
•	Changed Tx Power Supply column in Family and Ordering Information table	6
•	Changed TX_REF description in Pin Descriptions table	<mark>7</mark>
•	Changed conditions of Electrical Characteristics table	10
•	Changed Performance, PRF parameter minimum specification in Electrical Characteristics table	10
•	Changed PRF = 1300 Hz to PRF = 1200 Hz in test conditions for the Performance, <i>Total integrated noise current</i> and N_{FB} parameters in Electrical Characteristics table	
•	Changed Ambient Cancellation Stage, <i>Gain</i> parameter in Electrical Characteristics table	11
•	Added last two Low-Pass Filter parameters to Electrical Characteristics table	
•	Added Diagnostics, <i>Diagnostics current</i> parameter to Electrical Characteristics table	12
•	Changed C _F to C and added TX_REF capacitor to Functional Block Diagram graphic	27
•	Updated Figure 55	
•	Changed second sentence in second paragraph of Receiver Front-End section	28
•	Changed third paragraph of Receiver Front-End section	28
•	Changed second paragraph of Ambient Cancellation Scheme section	30
•	Added last paragraph and Table 1 to Ambient Cancellation Scheme section	31
•	Updated Figure 58	32
•	Updated Figure 60	34
•	Added footnote 1 to Table 2 and changed Example column in Table 2	36
•	Changed corresponding register column description in rows t ₁₃ , t ₁₅ , t ₁₇ , and t ₁₉ and example column values for rows t ₂₂ , t ₂₄ , t ₂₆ , and t ₂₈ in Table 2	36
•	Updated Figure 63	
•	Updated Figure 64	38
•	Deleted supply voltage range from RX_ANA_SUP and RX_DIG_SUP in Figure 65	
•	Changed entire Transmit Section	
•	Deleted _5V from TX_CTRL_SUP and LED_DRV_SUP in Figure 68	
•	Changed second paragraph of ADC Operation and Averaging Module section	
•	Updated Equation 5 and Figure 71	
•	Updated Figure 72	
•	Added first paragraph of AFE Output Mode (ADC Bypass Mode) section	
•	Updated Figure 75	
•	Added last paragraph to the <i>Diagnostics Module</i> section	
•	Added first and last sentence to Writing Data section	
•	Changed second to last sentence in Writing Data section	
•	Added first and last sentence to Reading Data section	
•	Changed second to last sentence in Reading Data section	

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•	Added Multiple Data Reads and Writes section	54
•	Added last sentence to the AFE SPI Interface Design Considerations section	
•	Added Register Control column to Table 4	
•	Changed bits D16 and D10 in CONTROL2 row of Table 4	
•	Changed CONTROL0 paragraph description	
•	Added note to bit D2 description of CONTROL0 register	
•	Changed PRPCOUNT[15:0] (bits D[15:0]) description in PRPCOUNT register	
•	Changed note within CLKALMPIN[2:0] (bits D[11:9]) description of CONTROL1 register	
•	Changed second and third columns of Table 5	
•	Changed 001 and 011 bit settings for the STG2GAIN[2:0] bits (bits D[10:8]) in the TIA_AMB_GAIN register	
•	Changed description and name of bits D16 and D10 in CONTROL2 register	
Cł	nanges from Revision C (April 2013) to Revision D	Page
•	Changed descriptions of RX_ANA_SUP, RX_DIG_SUP, and TX_CTRL_SUP pins in Pin Descriptions table	
•	Added CMRR parameter to Electrical Characteristics table	10
•	Added External Clock, External clock input voltage and External clock input current parameters to Electrical Characteristics table	12
•	Changed TIMING, t _{RESET} parameter unit in Electrical Characteristics table	1 <mark>2</mark>
•	Added Pin Leakage Current section to Electrical Characteristics table	1 <mark>2</mark>
•	Added Supply Current, ADC bypass mode parameter to Electrical Characteristics table	13
•	Changed Serial Interface Timing section	15
•	Added Figure 14	18
•	Added Figure 20	19
•	Added Figure 26	20
•	Added Figure 32	21
•	Added Figure 54	25
•	Updated Functional Block Diagram graphic	27
•	Changed name of register 15h	56
•	Added note to descriptions of LED2-ALED2VAL and LED1-ALED1VAL registers	82
Cł	nanges from Revision B (February 2013) to Revision C	Page
•	Changed first two sub-bullets of Receive Channel with High Dynamic Range Features bullet	
•	Changed pin out figure	
•	Changed ESD ratings specification values in Absolute Maximum Ratings table	
•	Added Performance, PSRR parameter to Electrical Characteristics table	10
•	Changed Performance, Total integrated noise current and N_{FB} parameters in Electrical Characteristics table	10
•	Changed first row of Receiver Functional Block Level Specification, <i>Total integrated noise current</i> parameter in Electrical Characteristics table	
•	Changed Ambient Cancellation Stage, Gain parameter specifications in Electrical Characteristics table	11
•	Changed Transmitter, Transmitter noise dynamic range parameter in Electrical Characteristics table	11
•	Added External Clock, External clock input frequency parameter to Electrical Characteristics table	
•	Added Timing, Wake-up time from Rx power-down and Wake-up time from Tx power-down parameters to Electronic Characteristics table	ctrical
•	Changed Supply Current section of Electrical Characteristics table	
•	Changed typical specification in first row and unit in second row of Power Dissipation, $P_{D(q)}$ parameter in Elect Characteristics table	rical



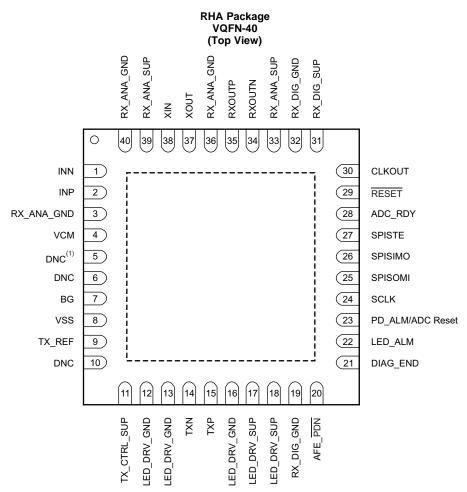
•	Changed Power Dissipation, After reset LED_DRV_SUP typical specification in Electrical Characteristics table	14
•	Changed Power Dissipation, <i>With stage 2 mode enabled</i> LED_DRV_SUP, TX_CTRL_SUP, and RX_DIG_SUP typical specifications in Electrical Characteristics table	14
•	Added Figure 13	18
•	Deleted Figure 11, Input-Referred Noise Current vs PLETH Current (BW = 5 Hz, PRF = 5000 Hz)	18
•	Deleted Figure 17, Input-Referred Noise Current vs PLETH Current (BW = 20 Hz, PRF = 5000 Hz)	19
•	Added Figure 24	20
•	Deleted Figure 23, Noise-Free Bits vs PLETH Current (BW = 5 Hz, PRF = 5000 Hz)	20
•	Added Figure 24	21
•	Deleted Figure 29, Noise-Free Bits vs PLETH Current (BW = 20 Hz, PRF = 5000 Hz)	21
•	Added Figure 38 through Figure 41	22
•	Added Figure 49 to Figure 53	24
•	Changed gain setting range in Receiver Front-End section	28
•	Changed corresponding register column description in rows t ₂₄ , t ₂₆ , and t ₂₈ in Table 2	36
•	Changed description of LED Power Reduction During Periods of Inactivity section	42
•	Changed last paragraph of AFE Analog Output Mode (ADC Bypass Mode) section	49
•	Updated Figure 76	49
•	Updated Figure 77	50
•	Changed LED2CONVEND register name in Table 4	56
•	Changed RESERVED1 and RESERVED2 register descriptions in Table 4	58
•	Changed description of bits D[15:0] in LED2STC register	60
•	Changed description of bits D[15:0] in LED2ENDC, LED2LEDSTC, and LED2LEDENDC registers	60
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•	Changed description of bits D[15:0] in LED1ENDC, LED1LEDSTC, and LED1LEDENDC registers	63
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•	Changed description of bits D[15:0] in LED1CONVST, LED1CONVEND, and ALED1CONVST registers	67
•	Changed description of bits D[15:0] in ALED1CONVEND register	68
•	Changed RESET to RESET in ADCRSTSTCT0 and ADCRSTENDCT0 registers	69
•	Changed RESET to RESET in ADCRSTSTCT1, ADCRSTENDCT1, and ADCRSTSTCT2 registers	69
•	Changed RESET to RESET in ADCRSTENDCT2, ADCRSTSTCT3, and ADCRSTENDCT3 registers	71
•	Added footnote to Table 6	76
•	Changed bits D18 and D17 names in CONTROL2 bit register	77
•	Added note to description of bits D[18:17] in CONTROL2 register	77
•	Changed RESERVED1 and RESERVED2 registers	79



5 Device Comparison Table

PRODUCT	PACKAGE-LEAD	LED DRIVE CONFIGURATION	LED DRIVE CURRENT (mA, max)	Tx POWER SUPPLY (V)	OPERATING TEMPERATURE RANGE
AFE4490	VQFN-40	Bridge, push-pull	50, 75, 100, 150, and 200	3 to 5.25	–40°C to 85°C
AFE4400	VQFN-40	Bridge, push-pull	50	3 to 5.25	0°C to 70°C
AFE4403	WCSP-36	Bridge, push-pull	100	3 to 5.25	–20°C to 70°C

6 Pin Configuration and Functions



Product Folder Links: AFE4490

(1) DNC = Do not connect.



Pin Functions

PIN	PIN FUNCTIONS					
NAME	NO.	FUNCTION	DESCRIPTION			
ADC_RDY	28	Digital	Output signal that indicates ADC conversion completion. Can be connected to the interrupt input pin of an external microcontroller.			
AFE_PDN	20	Digital	AFE-only power-down input; active low. Can be connected to the port pin of an external microcontroller.			
BG	7	Reference	Decoupling capacitor for internal band-gap voltage to ground (2.2-µF decoupling capacitor to ground, expected voltage = 1.0 V).			
CLKOUT	30	Digital	Buffered 4-MHz output clock output. Can be connected to the clock input pin of an external microcontroller.			
DIAG_END	21	Digital	Output signal that indicates completion of diagnostics. Can be connected to the port pin of an external microcontroller.			
DNC ⁽¹⁾	5, 6, 10	_	Do not connect these pins. Leave as open-circuit.			
INN	1	Analog	Receiver input pin. Connect to photodiode anode.			
INP	2	Analog	Receiver input pin. Connect to photodiode cathode.			
LED_DRV_GND	12, 13, 16	Supply	LED driver ground pin, H-bridge. Connect to common board ground.			
LED_DRV_SUP	17, 18	Supply	LED driver supply pin, H-bridge. Connect to an external power supply capable of supplying the large LED current, which is drawn by this supply pin.			
LED_ALM	22	Digital	Output signal that indicates an LED cable fault. Can be connected to the port pin of an external microcontroller.			
PD_ALM/ADC Reset	23	Digital	Output signal that indicates a PD sensor or cable fault. Can be connected to the port pin of an external microcontroller. In ADC bypass mode, the PD_ALM pin can be used to bring out the ADC reset signal.			
RESET	29	Digital	AFE-only reset input, active low. Can be connected to the port pin of an external microcontroller.			
RX_ANA_GND	3, 36, 40	Supply	Rx analog ground pin. Connect to common board ground.			
RX_ANA_SUP	33, 39	Supply	Rx analog supply pin; 0.1-µF decoupling capacitor to ground			
RX_DIG_GND	19, 32	Supply	Rx digital ground pin. Connect to common board ground.			
RX_DIG_SUP	31	Supply	Rx digital supply pin; 0.1-µF decoupling capacitor to ground			
RXOUTN	34	Analog	External ADC negative input when in ADC bypass mode			
RXOUTP	35	Analog	External ADC positive input when in ADC bypass mode			
SCLK	24	SPI	SPI clock pin			
SPISIMO	26	SPI	SPI serial in master out			
SPISOMI	25	SPI	SPI serial out master in			
SPISTE	27	SPI	SPI serial interface enable			
TX_CTRL_SUP	11	Supply	Transmit control supply pin (0.1-µF decoupling capacitor to ground)			
TX_REF	9	Reference	Transmitter reference voltage, 0.75 V default after reset. Connect a 2.2-μF decoupling capacitor to ground.			
TXN	14	Analog	LED driver out B, H-bridge output. Connect to LED.			
TXP	15	Analog	LED driver out B, H-bridge output. Connect to LED.			
VCM	4	Reference	Input common-mode voltage output. Connect a series resistor (1 k Ω) and a decoupling capacitor (10 nF) to ground. The voltage across the capacitor can be used to shield (guard) the INP, INN traces. Expected voltage = 0.9 V.			
VSS	8	Supply	Substrate ground. Connect to common board ground.			
XOUT	37	Digital	Crystal oscillator pins. Connect an external 8-MHz crystal between these pins with the correct load capacitor (as specified by vendor) to ground.			
XIN	38	Digital	Crystal oscillator pins. Connect an external 8-MHz crystal between these pins with the correct load capacitor (as specified by vendor) to ground.			

⁽¹⁾ Leave pins as open circuit. Do not connect.



7 Specifications

7.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted) (1)

		MIN	MAX	UNIT
RX_ANA_SUP, R	X_DIG_SUP to RX_ANA_GND, RX_DIG_GND	-0.3	4	V
TX_CTRL_SUP, I	_ED_DRV_SUP to LED_DRV_GND	-0.3	6	V
RX_ANA_GND, R	X_DIG_GND to LED_DRV_GND	-0.3	0.3	V
Analog inputs		RX_ANA_GND - 0.3	RX_ANA_SUP + 0.3	V
Digital inputs		RX_DIG_GND - 0.3	RX_DIG_SUP + 0.3	V
TXP, TXN pins		-0.3	Minimum [6, (LED_DRV_SUP + 0.3)]	V
Input current to any pin except supply pins (2)			±7	mA
lanut ourrant	Momentary		±50	mA
Input current	Continuous		±7	mA
Operating temperature range		-40	85	°C
Maximum junction temperature, T _J			125	°C

⁽¹⁾ Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, and do not imply functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

7.2 Handling Ratings

			MIN	MAX	UNIT
T _{stg}	T _{stg} Storage temperature range				°C
	Human body model (HBM), per ANSI/ESDA/JEDEC JS-001, all pins ⁽¹⁾	-1000 1000		V	
V(ESD)	V _(ESD) Electrostatic discharge	Charged device model (CDM), per JEDEC specification JESD22-C101, all pins (2)	-250	250	V

(1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

(2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

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⁽²⁾ Input pins are diode-clamped to the power-supply rails. Input signals that can swing beyond the supply rails must be current-limited to 10 mA or less.



7.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

		MIN	MAX	UNIT
SUPPLIES				
RX_ANA_SUP	AFE analog supply	2.0	3.6	V
RX_DIG_SUP	AFE digital supply	2.0	3.6	V
TX_CTRL_SUP	Transmit controller supply	3.0	5.25	V
LED_DRV_SUP	Transmit LED driver supply, H-bridge or common anode configuration	[3.0 or (V _{HR} + V _{LED} + V _{CABLE}) ⁽¹⁾⁽²⁾⁽³⁾ , whichever is greater]	5.25	V
	Difference between LED_DRV_SUP and TX_CTRL_SUP	-0.3	0.3	V
TEMPERATURE				
	Specified temperature range	-40	85	°C

⁽¹⁾ V_{HR} refers to the required voltage headroom necessary to drive the LEDs. See Table 6 for the appropriate V_{HR} value.

7.4 Thermal Information

		AFE4490	
	THERMAL METRIC ⁽¹⁾	RHA (VQFN)	UNIT
		40 PINS	
$R_{\theta JA}$	Junction-to-ambient thermal resistance	35	
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	31	
$R_{\theta JB}$	Junction-to-board thermal resistance	26	°C/W
ΨЈТ	Junction-to-top characterization parameter	0.1	C/VV
Ψ_{JB}	Junction-to-board characterization parameter	N/A	
$R_{\theta JC(bot)}$	Junction-to-case (bottom) thermal resistance	N/A	

(1) For more information about traditional and new thermal metrics, see the IC Package Thermal Metrics application report, SPRA953.

⁽²⁾ V_{LED} refers to the maximum voltage drop across the external LED (at maximum LED current) connected between the TXP and TXN pins (in H-bridge mode) and from the TXP and TXN pins to LED_DRV_SUP (in the common anode configuration).

⁽³⁾ V_{CABLE} refers to voltage drop across any cable, connector, or any other component in series with the LED.



7.5 Electrical Characteristics

Minimum and maximum specifications are at $T_A = -40^{\circ}\text{C}$ to 85°C. Typical specifications are at 25°C. All specifications are at RX_ANA_SUP = RX_DIG_SUP = 3 V, TX_CTRL_SUP = LED_DRV_SUP = 5 V, stage 2 amplifier disabled, and $f_{CLK} = 8$ MHz, unless otherwise noted.

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
PERFORM	MANCE (Full-Signal Chain)					
		$R_F = 10 \text{ k}\Omega$		50		μΑ
		R _F = 25 kΩ		20		μΑ
	Full-scale input current	$R_F = 50 \text{ k}\Omega$		10		μA
I _{IN FS}		$R_F = 100 \text{ k}\Omega$		5		μΑ
		$R_F = 250 \text{ k}\Omega$		2		μΑ
		$R_F = 500 \text{ k}\Omega$		1		μΑ
		$R_F = 1 M\Omega$		0.5		μΑ
PRF	Pulse repetition frequency		62.5		5000	SPS
DC _{PRF}	PRF duty cycle				25%	
OMPD	Common-mode rejection ratio	f_{CM} = 50 Hz and 60 Hz, LED1 and LED2 with R _{SERIES} = 500 kΩ, R _F = 500 kΩ		75		dB
CMRR		f_{CM} = 50 Hz and 60 Hz, LED1-AMB and LED2-AMB with R _{SERIES} = 500 kΩ, R _F = 500 kΩ		95		dB
2022		f _{PS} = 50 Hz, 60 Hz at PRF = 200 Hz		100		dB
PSRR	Power-supply rejection ratio	f _{CM} = 50 Hz, 60 Hz at PRF = 600 Hz		106		dB
PSRR _{LED}	PSRR, transmit LED driver	With respect to ripple on LED_DRV_SUP		75		dB
PSRR _{Tx}	PSRR, transmit control	With respect to ripple on TX_CTRL_SUP		60		dB
PSRR _{Rx}	PSRR, receiver	With respect to ripple on RX_ANA_SUP and RX_DIG_SUP		60		dB
	Total integrated noise current, input-	R_F = 100 kΩ with stage 2 gain disabled, PRF = 1200 Hz, duty cycle = 5%		36		pA _{RMS}
	referred (receiver with transmitter loop back, 0.1-Hz to 20-Hz bandwidth)	R_F = 500 kΩ with ambient cancellation enabled and stage 2 gain = 4, PRF = 1200 Hz, duty cycle = 25%		13		pA _{RM}
NI	Noise-free bits (receiver with transmitter	$R_F = 100 \text{ k}\Omega$, PRF = 1200 Hz, duty cycle = 5%		14.3		Bits
N _{FB}	loop back, 0.1-Hz to 20-Hz bandwidth) ⁽¹⁾	$R_F = 500 \text{ k}\Omega$, PRF = 1200 Hz, duty cycle = 25%		13.5		Bits
RECEIVER	R FUNCTIONAL BLOCK LEVEL SPECIFICA	TION				
	Total integrated noise current,	R_F = 500 k Ω , ambient cancellation enabled, stage 2 gain = 4, PRF = 1300 Hz, LED duty cycle = 25%		1.4		pA _{RM}
input-referred (receiver alone) over 0.1-Hz to 5-Hz bandwidth		R_F = 500 k Ω , ambient cancellation enabled, stage 2 gain = 4, PRF = 1300 Hz, LED duty cycle = 5%		5		pA _{RM}

(1) Noise-free bits (NFB) are defined as:

$$N_{FB} = log 2 \left[\frac{1}{6.6 \times I_{PD}} \right]$$

where: I_{PD} is the photodiode current, and I_{NOISE} is the input-referred RMS noise current.

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Minimum and maximum specifications are at $T_A = -40^{\circ}\text{C}$ to 85°C. Typical specifications are at 25°C. All specifications are at RX_ANA_SUP = RX_DIG_SUP = 3 V, TX_CTRL_SUP = LED_DRV_SUP = 5 V, stage 2 amplifier disabled, and $f_{CLK} = 8$ MHz, unless otherwise noted.

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
I-V TRANS	SIMPEDANCE AMPLIFIER					
G	Gain	$R_F = 10 \text{ k}\Omega$ to $R_F = 1 \text{ M}\Omega$	See the <i>Receiver Channel</i> section for details		section	V/µA
	Gain accuracy			±7%		
	Feedback resistance	R _F		0k, 100k, 25 x, and 1M	50k,	Ω
	Feedback resistor tolerance	R _F		±7%		
	Feedback capacitance	C _F	5, 10, 25, 5	0, 100, and	250	pF
	Feedback capacitor tolerance	C _F		±20%		
V _{OD(fs)}	Full-scale differential output voltage			1		V
	Common-mode voltage on input pins	Set internally		0.9		V
	External differential input capacitance	Includes equivalent capacitance of photodiode, cables, EMI filter, and so forth	10		1000	pF
V _{O(shield)}	Shield output voltage, V _{CM}	With a 1-kΩ series resistor and a 10-nF decoupling capacitor to ground, when loaded with a small current (for example, of a few μA or less)	0.8	0.9	1.0	V
AMBIENT	CANCELLATION STAGE				·	
G	Gain		0, 3.5, 6	, 9.5, and 12	2	dB
	Current DAC range		0		10	μΑ
	Current DAC step size			1		μA
LOW-PAS	S FILTER					
	Low-pass corner frequency	3-dB attenuation	0.	5 and 1		kHz
	5	Duty cycle = 25%		0.004		dB
	Pass-band attenuation, 2 Hz to 10 Hz	Duty cycle = 10%		0.041		dB
		After diagnostics mode with filter corner = 500 Hz		28		ms
	Filter settling time	After diagnostics mode with filter corner = 1000 Hz		16		ms
	ADC bypass outputs output impedance	RXOUTP and RXOUTN		1		kΩ
ANALOG-	TO-DIGITAL CONVERTER					
	Resolution				22	Bits
	Sample rate	See the ADC Operation and Averaging Module section	4	1 × PRF		SPS
	ADC full-scale voltage			±1.2		V
	ADC conversion time	See the ADC Operation and Averaging Module section	50		PRF / 4	μs
	ADC reset time		2			t _{CLK}
TRANSMI	TTER					
	Output current range		0, 50, 75, 10 (see the <i>LEDC</i> <i>Registe</i>		Control	mA
	LED current DAC error			±5%		
	Output current resolution			8		Bits
	Tono antitato antino de moneio antino	0.1-Hz to 20-Hz bandwidth, at 25-mA output current		110		dB
	Transmitter noise dynamic range	0.1-Hz to 20-Hz bandwidth, at 100-mA output current		110		dB
	Minimum sample time of LED1 and LED2 pulses			50		μs
	LED current DAC leakage current	LED_ON = 0		1		μA
	535 2.15 .5anago 5anon	LED_ON = 1		50		μA
	LED current DAC linearity	Percent of full-scale current		0.5%		
	Output current settling time	From zero current to 150 mA		7		μs
	(with resistive load)	From 150 mA to zero current		7		μs

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Minimum and maximum specifications are at $T_A = -40$ °C to 85°C. Typical specifications are at 25°C. All specifications are at RX_ANA_SUP = RX_DIG_SUP = 3 V, TX_CTRL_SUP = LED_DRV_SUP = 5 V, stage 2 amplifier disabled, and $f_{CLK} = 8$ MHz, unless otherwise noted.

	and f _{CLK} = 8 MHz, unless otherwise	TEST CONDITIONS	MIN TYP	MAX	UNIT
DIAGNOS.		TEST CONDITIONS	MIII III	WIAA	ONT
DIAGNOS		EN_SLOW_DIAG = 0 Start of diagnostics after the DIAG_EN register bit is set. End of diagnostic indicated by DIAG_END going high.	8		ms
	Duration of diagnostics state machine	EN_SLOW_DIAG = 1 Start of diagnostics after the DIAG_EN register bit is set. End of diagnostic indicated by DIAG_END going high.	16		ms
	Open fault resistance		> 100		kΩ
	Short fault resistance		< 10		kΩ
	Diagnostics current	During diagnostics mode	< 100		μΑ
INTERNAL	OSCILLATOR				
f _{CLKOUT}	CLKOUT frequency	With an 8-MHz crystal connected to the XIN and XOUT pins	4		MHz
DC _{CLKOUT}	CLKOUT duty cycle		50%		
	Crystal oscillator start-up time	With an 8-MHz crystal connected to the XIN and XOUT pins	200		μs
EXTERNA	L CLOCK				
	Maximum allowable external clock jitter			50	ps
	External clock input frequency	±10%	8		MHz
	External clock input voltage	Voltage input high (V _{IH})	0.75 x RX_DIG_SUP		V
	External clock input voltage	Voltage input low (V _{IL})	0.25 × RX_DIG_SUP		V
	External clock input current		1		μΑ
TIMING			T.		
	Wake-up time from complete power-down		1000		ms
	Wake-up time from Rx power-down		100		μs
	Wake-up time from Tx power-down		1000		ms
t _{RESET}	Active low RESET pulse duration		1		ms
t _{DIAGEND}	DIAG_END pulse duration at diagnostics completion		4		CLKOUT cycles
t _{ADCRDY}	ADC_RDY pulse duration		1		CLKOUT cycles
DIGITAL S	SIGNAL CHARACTERISTICS				
V_{IH}	Logic high input voltage	AFE_PDN, SPI CLK, SPI SIMO, SPI STE, RESET	0.75 × RX_DIG_SUP		V
V_{IL}	Logic low input voltage	AFE_PDN, SPI CLK, SPI SIMO, SPI STE, RESET	0.25 × RX_	DIG_SUP	V
I _{IN}	Logic input current	Digital inputs at V_{IH} or V_{IL}	0.1		μA
V_{OH}	Logic high output voltage	DIAG_END, LED_ALM, PD_ALM, SPI SOMI, ADC_RDY, CLKOUT	RX_DIG_SUP - 0.1		V
V _{OL}	Logic low output voltage	DIAG_END, LED_ALM, PD_ALM, SPI SOMI, ADC_RDY, CLKOUT		0.1	V
PIN LEAK	AGE CURRENT				
	Pin leakage current	To GND and supply	1		nA



Minimum and maximum specifications are at $T_A = -40^{\circ}\text{C}$ to 85°C. Typical specifications are at 25°C. All specifications are at RX_ANA_SUP = RX_DIG_SUP = 3 V, TX_CTRL_SUP = LED_DRV_SUP = 5 V, stage 2 amplifier disabled, and $f_{CLK} = 8$ MHz, unless otherwise noted.

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
SUPPLY C	URRENT					
	Descriper analog august august	RX_ANA_SUP = 3.0 V, with 8-MHz clock running, Rx stage 2 disabled		0.6		mA
	Receiver analog supply current	RX_ANA_SUP = 3.0 V, with 8-MHz clock running, Rx stage 2 enabled		0.7		mA
	Receiver digital supply current	RX_DIG_SUP = 3.0 V		0.27		mA
	ADC bypass mode	RX_ANA_SUP + RX_DIG_SUP (Excluding external ADC current)		1.8		mA
LED_DRV _SUP	LED driver supply current	With zero LED current setting		55		μΑ
TX_CTRL _SUP	Transmitter control supply current			15		μΑ
	Complete power-down (using AFE_PDN pin)	Receiver current only (RX_ANA_SUP)		3		μΑ
		Receiver current only (RX_DIG_SUP)		3		μΑ
		Transmitter current only (LED_DRV_SUP)		1		μΑ
		Transmitter current only (TX_CTRL_SUP)		1		μΑ
	Power-down Rx alone	Receiver current only (RX_ANA_SUP)		220		μΑ
	rower-down RX alone	Receiver current only (RX_DIG_SUP)		220		μΑ
	Power down Ty clone	Transmitter current only (LED_DRV_SUP)		2		μΑ
	Power-down Tx alone	Transmitter current only (TX_CTRL_SUP)		2		μΑ



Minimum and maximum specifications are at $T_A = -40^{\circ}\text{C}$ to 85°C. Typical specifications are at 25°C. All specifications are at RX_ANA_SUP = RX_DIG_SUP = 3 V, TX_CTRL_SUP = LED_DRV_SUP = 5 V, stage 2 amplifier disabled, and $f_{CLK} = 8$ MHz, unless otherwise noted.

PARAMETER			TEST CONDITIONS	MIN	TYP	MAX	UNIT
POWER	DISSIPATION			<u>'</u>			
			Normal operation (excluding LEDs)		2.84		mW
$P_{D(q)}$	Quiescent power dissipa	ition	Power-down		0.1		mW
		LED_DRV_SUP	LED_DRV_SUP current value. Does not include LED current.		1		μΑ
	Power-down with the	TX_CTRL_SUP			1		μA
	AFE_PDN pin	RX_ANA_SUP			5		μA
		RX_DIG_SUP			0.1		μA
		LED_DRV_SUP	LED_DRV_SUP current value. Does not include LED current.		1		μΑ
	Power-down with the	TX_CTRL_SUP			1		μA
	PDNAFE register bit	RX_ANA_SUP			15		μΑ
		RX_DIG_SUP			20		μA
	LED	LED_DRV_SUP	LED_DRV_SUP current value. Does not include LED current.		50		μΑ
	Power-down Rx	TX_CTRL_SUP			15		μΑ
		RX_ANA_SUP			220		μA
		RX_DIG_SUP			220		μΑ
		LED_DRV_SUP	LED_DRV_SUP current value. Does not include LED current.		2		μΑ
	Power-down Tx	TX_CTRL_SUP			2		μΑ
		RX_ANA_SUP			600		μΑ
		RX_DIG_SUP			230		μΑ
		LED_DRV_SUP	LED_DRV_SUP current value. Does not include LED current.		55		μΑ
	After reset, with 8-MHz	TX_CTRL_SUP			15		μΑ
	clock running	RX_ANA_SUP			600		μA
		RX_DIG_SUP			230		μA
		LED_DRV_SUP	LED_DRV_SUP current value. Does not include LED current.		55		μA
	With stage 2 mode enabled and 8-MHz	TX_CTRL_SUP			15		μA
	clock running	RX_ANA_SUP			700		μA
		RX_DIG_SUP			270		μA

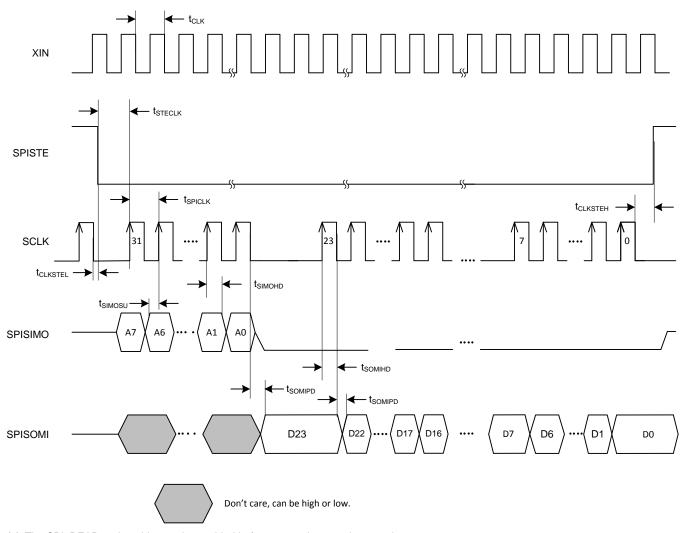
Product Folder Links: AFE4490

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7.6 Timing Requirements: Serial Interface

	9 9				
		MIN	TYP	MAX	UNIT
t _{CLK}	Clock frequency on XIN pin		8		MHz
t _{SCLK}	Serial shift clock period	62.5			ns
t _{STECLK}	STE low to SCLK rising edge, setup time	10			ns
t _{CLKSTEH,L}	SCLK transition to SPI STE high or low	10			ns
t _{SIMOSU}	SIMO data to SCLK rising edge, setup time	10			ns
t _{SIMOHD}	Valid SIMO data after SCLK rising edge, hold time	10			ns
t _{SOMIPD}	SCLK falling edge to valid SOMI, setup time	17			ns
t _{SOMIHD}	SCLK rising edge to invalid data, hold time	0.5			t _{SCLK}



- (1) The SPI_READ register bit must be enabled before attempting a register read.
- (2) Specify the register address whose contents must be read back on A[7:0].
- (3) The AFE outputs the contents of the specified register on the SOMI pin.

Figure 1. Serial Interface Timing Diagram, Read Operation (1)(2)(3)



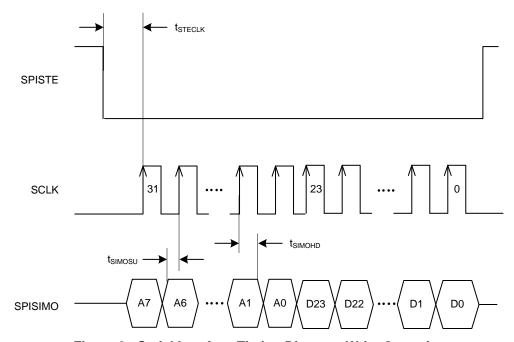


Figure 2. Serial Interface Timing Diagram, Write Operation



7.7 Supply Ramp and Power-Down Timing Requirements

		VALUE
1	Time between Rx and Tx supplies ramping up	Keep as small as possible (for example, ±10 ms)
2	Time between both supplies stabilizing and high-going edge of RESET	> 100 ms
3	RESET pulse width	> 0.5 ms
4	Time between RESET and SPI commands	> 1 µs
5	Time between SPI commands and the ADC_ $\overline{\rm RESET}$ which corresponds to valid data	> 3 ms of cumulative sampling time in each phase (1)(2)(3)
6	Time between RESET pulse and high-accuracy data coming out of the signal chain	> 1 s ⁽³⁾
7	Time from AFE_PDN high-going edge and RESET pulse (4)	> 100 ms
t ₈	Time from AFE_PDN high-going edge (or PDN_AFE bit reset) to high-accuracy data coming out of the signal chain	> 1 s ⁽³⁾

- (1) This time is required for each of the four switched RC filters to fully settle to the new settings. The same time is applicable whenever there is a change to any of the signal chain controls (for example, LED current setting, TIA gain, and so forth)
- (2) If the SPI commands involve a change in the value of TX_REF from its default, then there is additional wait time that is approximately 1 s (for a 2.2-µF decoupling capacitor on the TX_REF pin).
- (3) Dependent on the value of the capacitors on the BG and TX_REF pins. The 1-s wait time is necessary when the capacitors are 2.2 μF and scale down proportionate to the capacitor value. A very low capacitor (for example, 0.1 μF) on these pins causes the transmitter dynamic range to reduce to approximately 100 dB.
- (4) After an active power-down from AFE_PDN, reset the device by using a low-going pulse on RESET.

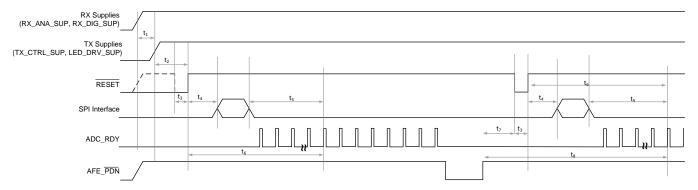


Figure 3. Supply Ramp and Hardware Power-Down Timing

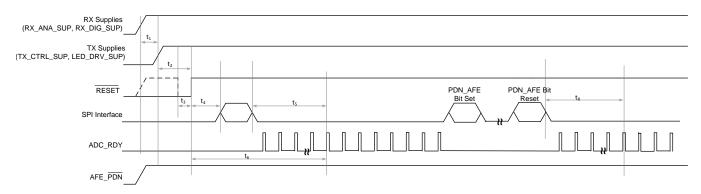
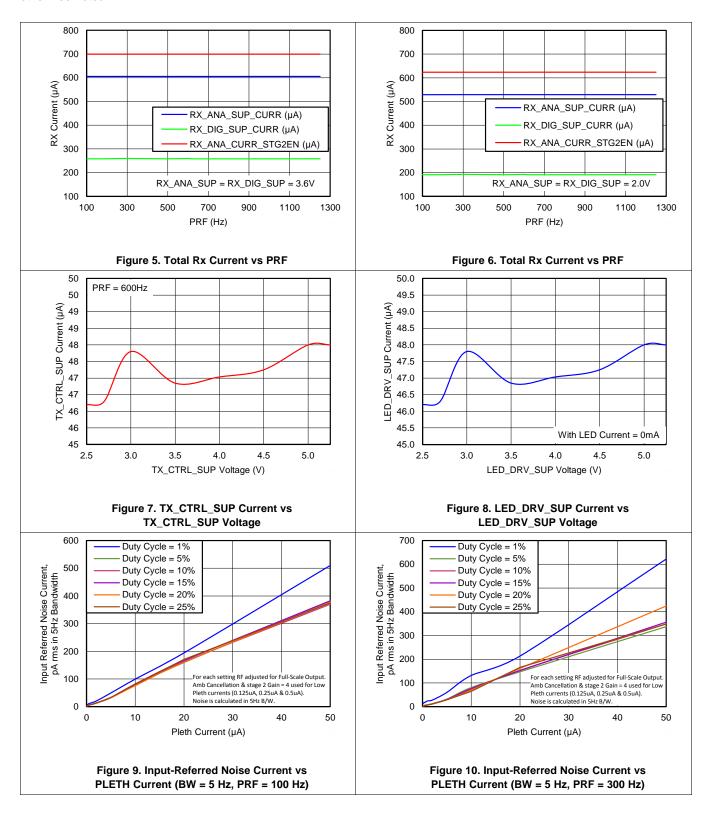


Figure 4. Supply Ramp and Software Power-Down Timing



7.8 Typical Characteristics

At $T_A = 25$ °C, RX_ANA_SUP = RX_DIG_SUP = 3.0 V, TX_CTRL_SUP = LED_DRV_SUP = 5 V, and $f_{CLK} = 8$ MHz, unless otherwise noted.

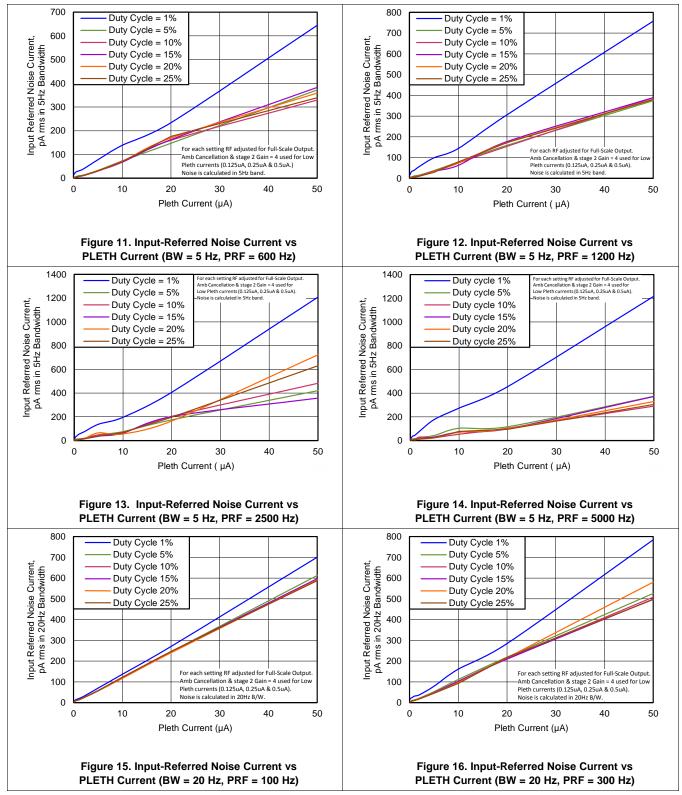


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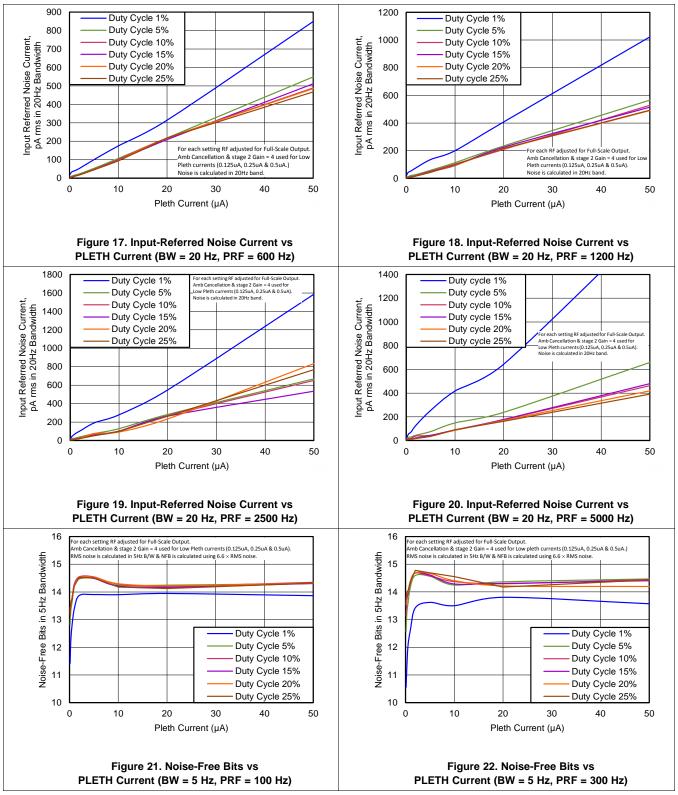
At T_A = 25°C, RX_ANA_SUP = RX_DIG_SUP = 3.0 V, TX_CTRL_SUP = LED_DRV_SUP = 5 V, and f_{CLK} = 8 MHz, unless otherwise noted.



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At T_A = 25°C, RX_ANA_SUP = RX_DIG_SUP = 3.0 V, TX_CTRL_SUP = LED_DRV_SUP = 5 V, and f_{CLK} = 8 MHz, unless otherwise noted.

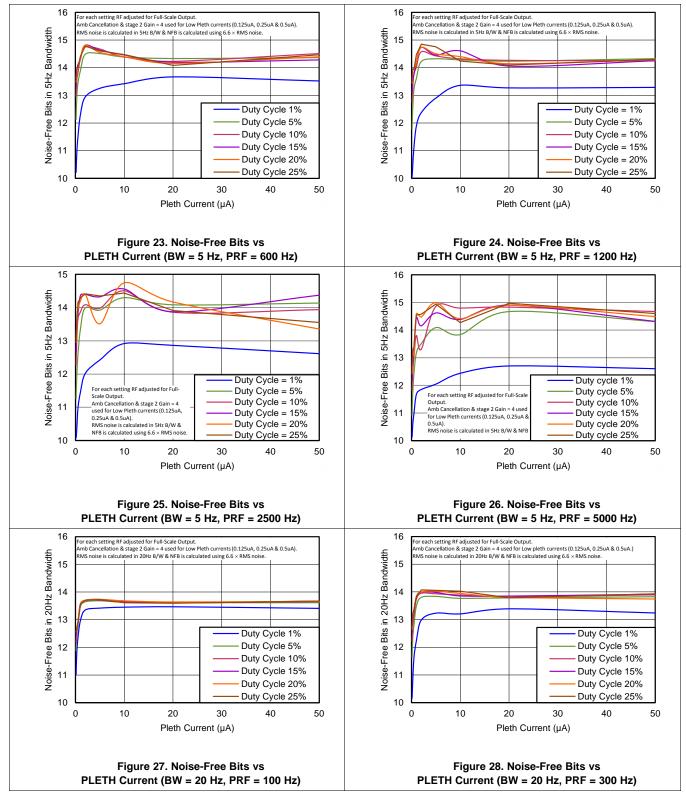


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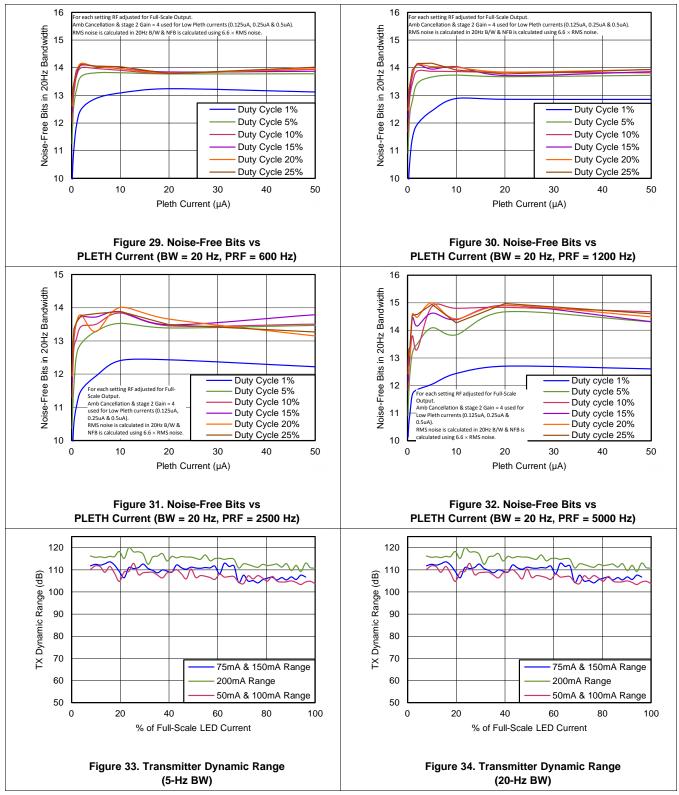
At T_A = 25°C, RX_ANA_SUP = RX_DIG_SUP = 3.0 V, TX_CTRL_SUP = LED_DRV_SUP = 5 V, and f_{CLK} = 8 MHz, unless otherwise noted.



TEXAS INSTRUMENTS

Typical Characteristics (continued)

At T_A = 25°C, RX_ANA_SUP = RX_DIG_SUP = 3.0 V, TX_CTRL_SUP = LED_DRV_SUP = 5 V, and f_{CLK} = 8 MHz, unless otherwise noted.

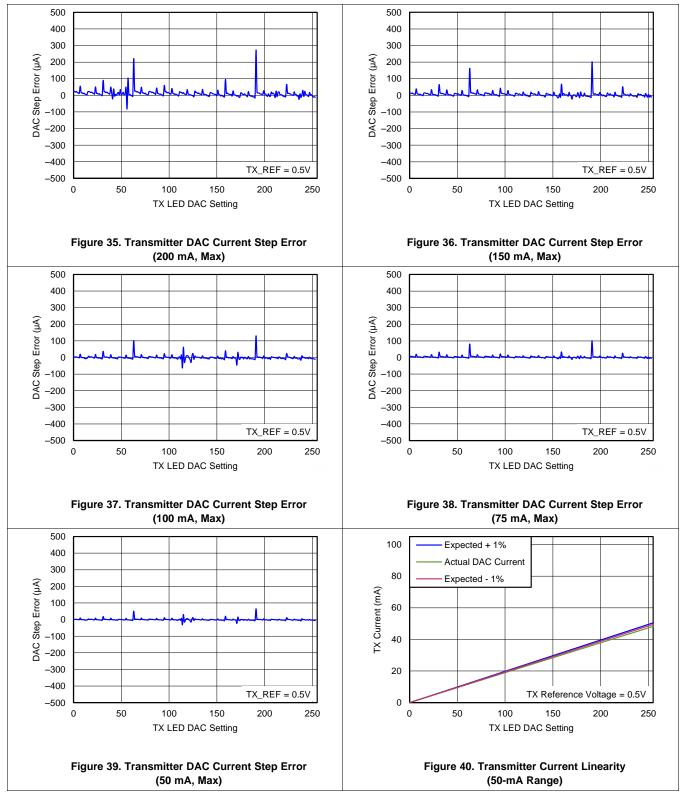


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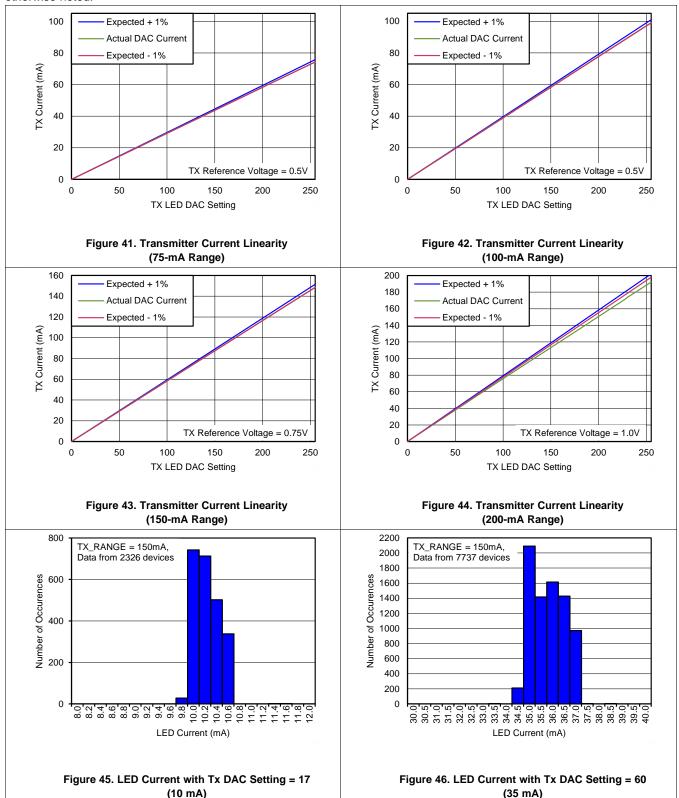
At T_A = 25°C, RX_ANA_SUP = RX_DIG_SUP = 3.0 V, TX_CTRL_SUP = LED_DRV_SUP = 5 V, and f_{CLK} = 8 MHz, unless otherwise noted.



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At T_A = 25°C, RX_ANA_SUP = RX_DIG_SUP = 3.0 V, TX_CTRL_SUP = LED_DRV_SUP = 5 V, and f_{CLK} = 8 MHz, unless otherwise noted.

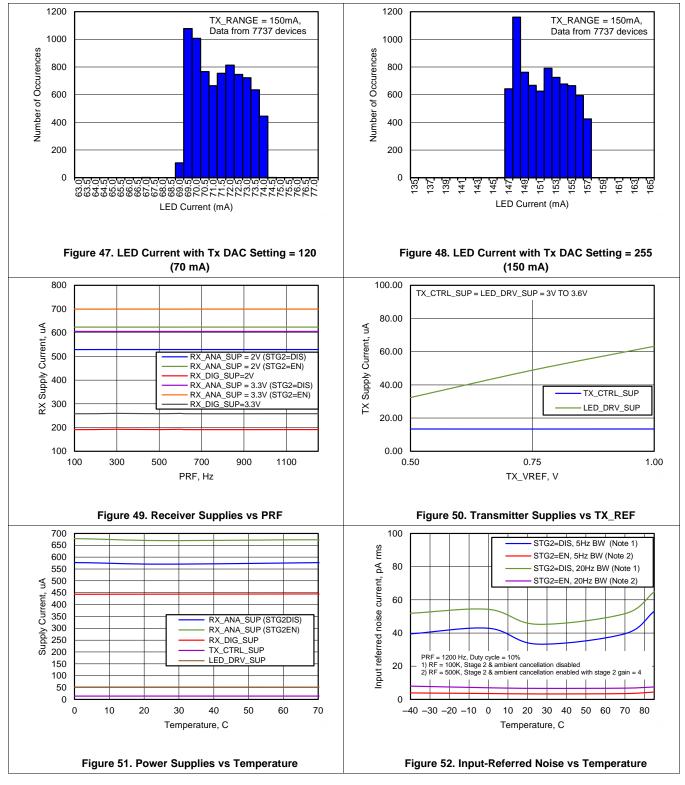


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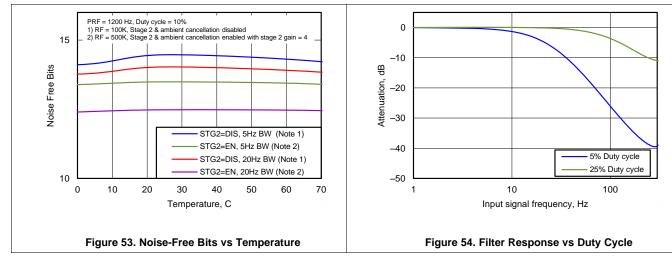
At T_A = 25°C, RX_ANA_SUP = RX_DIG_SUP = 3.0 V, TX_CTRL_SUP = LED_DRV_SUP = 5 V, and f_{CLK} = 8 MHz, unless otherwise noted.



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At T_A = 25°C, RX_ANA_SUP = RX_DIG_SUP = 3.0 V, TX_CTRL_SUP = LED_DRV_SUP = 5 V, and f_{CLK} = 8 MHz, unless otherwise noted.



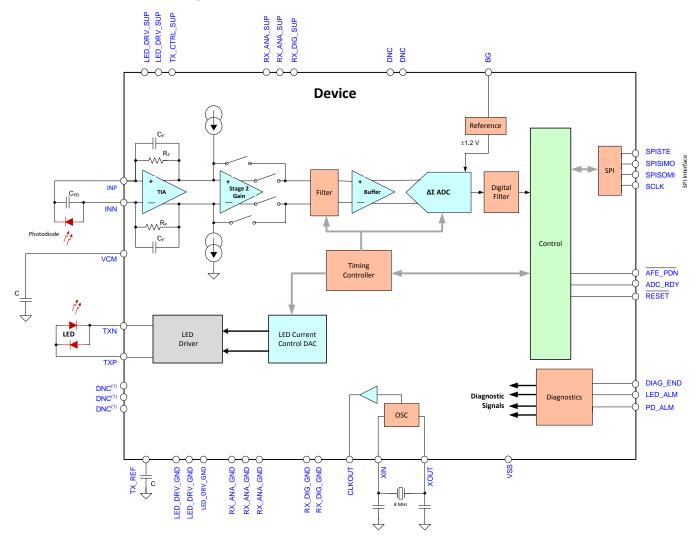


8 Detailed Description

8.1 Overview

The AFE4490 is a complete analog front-end (AFE) solution targeted for pulse-oximeter applications. The device consists of a low-noise receiver channel, an LED transmit section, and diagnostics for sensor and LED fault detection. To ease clocking requirements and provide the low-jitter clock to the AFE, an oscillator is also integrated that functions from an external crystal. The device communicates to an external microcontroller or host processor using an SPI interface. The *Functional Block Diagram* section provides a detailed block diagram for the device. The blocks are described in more detail in the following sections.

8.2 Functional Block Diagram





8.3 Feature Description

8.3.1 Receiver Channel

This section describes the receiver channel functionality.

8.3.1.1 Receiver Front-End

The receiver consists of a differential current-to-voltage (I-V) transimpedance amplifier that converts the input photodiode current into an appropriate voltage, as shown in Figure 55. The feedback resistor of the amplifier (R_F) is programmable to support a wide range of photodiode currents. Available R_F values include: 1 M Ω , 500 k Ω , 250 k Ω , 100 k Ω , 50 k Ω , 25 k Ω , and 10 k Ω .

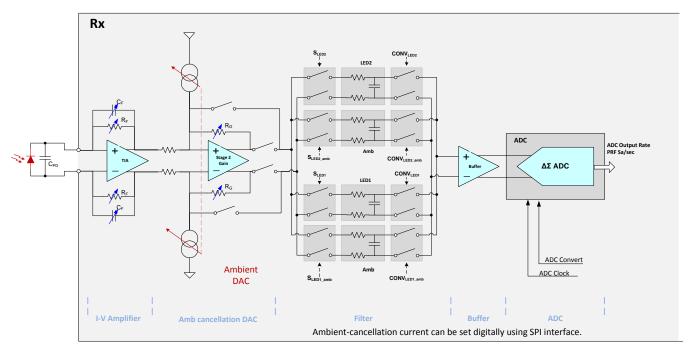


Figure 55. Receiver Front-End

The R_F amplifier and the feedback capacitor (C_F) form a low-pass filter for the input signal current. Always ensure that the low-pass filter RC time constant has sufficiently high bandwidth (as shown by Equation 1) because the input current consists of pulses. For this reason, the feedback capacitor is also programmable. Available C_F values include: 5 pF, 10 pF, 25 pF, 50 pF, 100 pF, and 250 pF. Any combination of these capacitors can also be used.

$$R_{F} \times C_{F} \leq \frac{\text{Rx Sample Time}}{10} \tag{1}$$

The output voltage of the I-V amplifier includes the pleth component (the desired signal) and a component resulting from the ambient light leakage. The I-V amplifier is followed by the second stage, which consists of a current digital-to-analog converter (DAC) that sources the cancellation current and an amplifier that gains up the pleth component alone. The amplifier has five programmable gain settings: 0 dB, 3.5 dB, 6 dB, 9.5 dB, and 12 dB. The gained-up pleth signal is then low-pass filtered (500-Hz bandwidth) and buffered before driving a 22-bit ADC. The current DAC has a cancellation current range of 10 μ A with 10 steps (1 μ A each). The DAC value can be digitally specified with the SPI interface. Using ambient compensation with the ambient DAC allows the dc-biased signal to be centered to near mid-point of the amplifier (±0.9 V). Using the gain of the second stage allows for more of the available ADC dynamic range to be used.

The output of the ambient cancellation amplifier is separated into LED2 and LED1 channels. When LED2 is on, the amplifier output is filtered and sampled on capacitor C_R . Similarly, the LED1 signal is sampled on the C_{LED1} capacitor when LED1 is ON. In between the LED2 and LED1 pulses, the idle amplifier output is sampled to estimate the ambient signal on capacitors C_{LED2} amb and C_{LED1} amb.



Feature Description (continued)

The sampling duration is termed the Rx sample time and is programmable for each signal, independently. Sampling can start after the I-V amplifier output is stable (to account for LED and cable settling times). The Rx sample time is used for all dynamic range calculations; the minimum time supported is 50 μ s.

A single, 22-bit ADC converts the sampled LED2, LED1, and ambient signals sequentially. Each conversion takes a maximum of 25% of the pulse repetition period (PRP) and provides a single digital code at the ADC output. As discussed in the *Receiver Timing* section, the conversions are staggered so that the LED2 conversion starts after the end of the LED2 sample phase, and so on. This configuration also means that the Rx sample time for each signal is no greater than 25% of the pulse repetition period.

Note that four data streams are available at the ADC output (LED2, LED1, ambient LED2, and ambient LED1) at the same rate as the pulse repetition frequency. The ADC is followed by a digital ambient subtraction block that additionally outputs the (LED2 – ambient LED2) and (LED1 – ambient LED1) data values.

8.3.1.2 Ambient Cancellation Scheme

The receiver provides digital samples corresponding to ambient duration. The host processor (external to the AFE) can use these ambient values to estimate the amount of ambient light leakage. The processor must then set the value of the ambient cancellation DAC using the SPI, as shown in Figure 56.

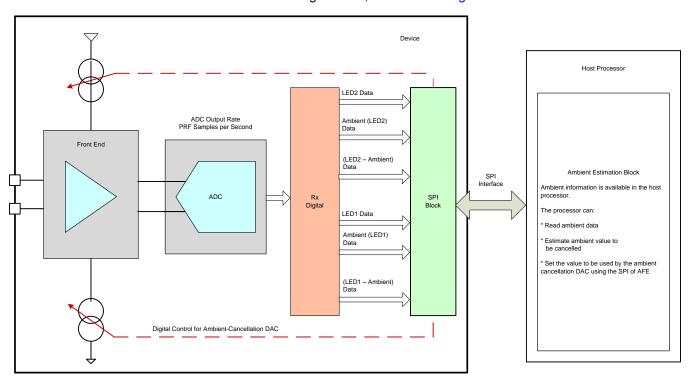


Figure 56. Ambient Cancellation Loop (Closed by the Host Processor)



Feature Description (continued)

Using the set value, the ambient cancellation stage subtracts the ambient component and gains up only the pleth component of the received signal, as shown in Figure 57. The amplifier gain is programmable to 0 dB, 3.5 dB, 6 dB, 9.5 dB, and 12 dB.

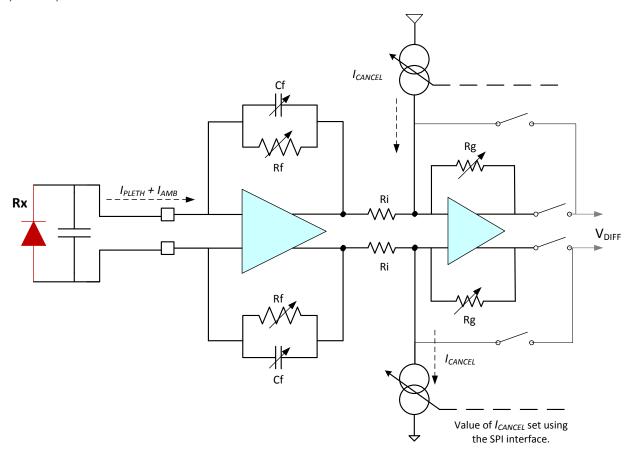


Figure 57. Front-End (I-V Amplifier and Cancellation Stage)

The differential output of the second stage is V_{DIFF}, as given by Equation 2:

$$V_{\text{DIFF}} = 2 \times \left[I_{\text{PLETH}} \times \frac{R_{\text{F}}}{R_{\text{I}}} + I_{\text{AMB}} \times \frac{R_{\text{F}}}{R_{\text{I}}} - I_{\text{CANCEL}} \right] \times R_{\text{G}}$$

where:

- $R_1 = 100 \text{ k}\Omega$,
- I_{PLETH} = photodiode current pleth component,
- I_{AMB} = photodiode current ambient component, and
- I_{CANCEL} = the cancellation current DAC value (as estimated by the host processor).

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(2)



Feature Description (continued)

R_G values with various gain settings are listed in Table 1.

Table 1. R_G Values

R _G (dB)	GAIN (kΩ)
0 (x1)	100
3.5 (x1.5)	150
6 (x2)	200
9.5 (x3)	300
12 (x4)	400

8.3.1.3 Receiver Control Signals

LED2 sample phase (S_{LED2}): When this signal is high, the amplifier output corresponds to the LED2 on-time. The amplifier output is filtered and sampled into capacitor C_{LED2} . To avoid settling effects resulting from the LED or cable, program S_{LED2} to start after the LED turns on. This settling delay is programmable.

Ambient sample phase (S_{LED2_amb}): When this signal is high, the amplifier output corresponds to the LED2 off-time and can be used to estimate the ambient signal (for the LED2 phase). The amplifier output is filtered and sampled into capacitor $C_{1 ED2_amb}$.

LED1 sample phase (S_{LED1}): When this signal is high, the amplifier output corresponds to the LED1 on-time. The amplifier output is filtered and sampled into capacitor C_{LED1} . To avoid settling effects resulting from the LED or cable, program S_{LED1} to start after the LED turns on. This settling delay is programmable.

Ambient sample phase (S_{LED1_amb}): When this signal is high, the amplifier output corresponds to the LED1 off-time and can be used to estimate the ambient signal (for the LED1 phase). The amplifier output is filtered and sampled into capacitor C_{LED1_amb} .

LED2 convert phase (CONV_{LED2}): When this signal is high, the voltage sampled on C_{LED2} is buffered and applied to the ADC for conversion. The conversion time duration is always 25% of the pulse repetition period. At the end of the conversion, the ADC provides a single digital code corresponding to the LED2 sample.

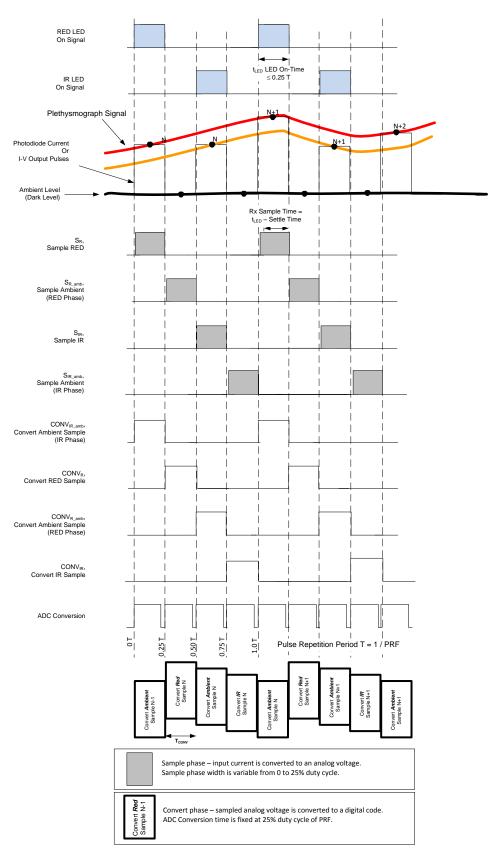
Ambient convert phases (CONV_{LED2_amb}, **CONV**_{LED1_amb}): When this signal is high, the voltage sampled on C_{LED2_amb} (or C_{LED1_amb}) is buffered and applied to the ADC for conversion. The conversion time duration is always 25% of the pulse repetition period. At the end of the conversion, the ADC provides a single digital code corresponding to the ambient sample.

LED1 convert phase (CONV_{LED1}): When this signal is high, the voltage sampled on C_{LED1} is buffered and applied to the ADC for conversion. The conversion time duration is always 25% of the pulse repetition period. At the end of the conversion, the ADC provides a single digital code corresponding to the LED1 sample.

8.3.1.4 Receiver Timing

See Figure 58 for a timing diagram detailing the control signals related to the LED on-time, Rx sample time, and the ADC conversion times for each channel.





NOTE: Relationship to the AFE4490EVM is: LED1 = IR and LED2 = RED.

Figure 58. Rx Timing Diagram



8.3.2 Clocking and Timing Signal Generation

The crystal oscillator generates a master clock signal using an external 8-MHz crystal. A divide-by-2 block converts the 8-MHz clock to 4 MHz, which is used by the AFE to operate the timer modules, ADC, and diagnostics. The 4-MHz clock is buffered and output from the AFE in order to clock an external microcontroller. The clocking functionality is shown in Figure 59.

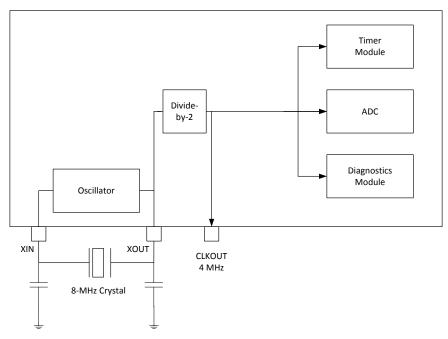


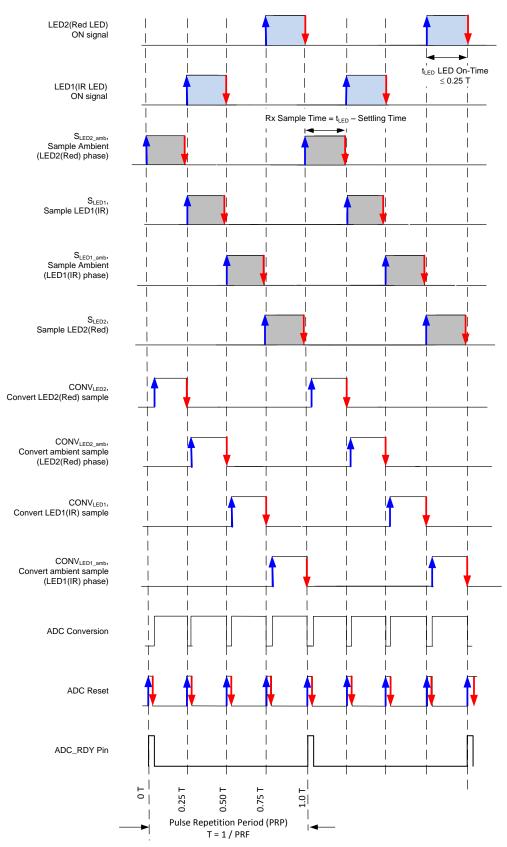
Figure 59. AFE Clocking

8.3.3 Timer Module

See Figure 60 for a timing diagram detailing the various timing edges that are programmable using the timer module. The rising and falling edge positions of 11 signals can be controlled. The module uses a single 16-bit counter (running off of the 4-MHz clock) to set the time-base.

All timing signals are set with reference to the pulse repetition period (PRP). Therefore, a dedicated compare register compares the 16-bit counter value with the reference value specified in the PRF register. Every time that the 16-bit counter value is equal to the reference value in the PRF register, the counter is reset to '0'.





NOTE: Programmable edges are shown in blue and red.

Figure 60. AFE Control Signals



For the 11 signals in Figure 58, the start and stop edge positions are programmable with respect to the PRF period. Each signal uses a separate timer compare module that compares the counter value with preprogrammed reference values for the start and stop edges. All reference values can be set using the SPI interface.

When the counter value equals the start reference value, the output signal is set. When the counter value equals the stop reference value, the output signal is reset. Figure 61 shows a diagram of the timer compare register. With a 4-MHz clock, the edge placement resolution is 0.25 µs. The ADC conversion signal requires four pulses in each PRF clock period. The 11th timer compare register uses four sets of start and stop registers to control the ADC conversion signal.

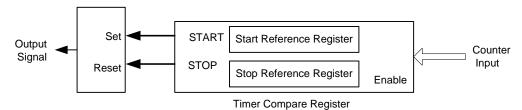


Figure 61. Compare Register

The ADC conversion signal requires four pulses in each PRF clock period. Timer compare register 11 uses four sets of start and stop registers to control the ADC conversion signal, as shown in Figure 62.

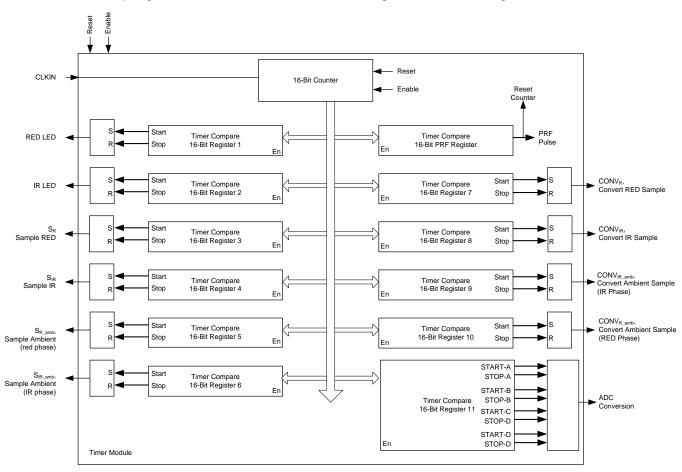


Figure 62. Timer Module



8.3.3.1 Using the Timer Module

The timer module registers can be used to program the start and end instants in units of 4-MHz clock cycles. These timing instants and the corresponding registers are listed in Table 2.

Note that the device does not restrict the values in these registers; thus, the start and end edges can be positioned anywhere within the pulse repetition period. Care must be taken by the user to program suitable values in these registers to avoid overlapping the signals and to make sure none of the edges exceed the value programmed in the PRP register. Writing the same value in the start and end registers results in a pulse duration of one clock cycle. The following steps describe the timer sequencing configuration:

- 1. With respect to the start of the PRP period (indicated by timing instant t₀ in Figure 63 and Figure 64), the sequence of conversions must be followed in order: convert LED2 → LED2 ambient → LED1 → LED1 ambient.
- 2. Also, starting from t_0 , the sequence of sampling instants must be staggered with respect to the respective conversions as follows: sample LED2 ambient \rightarrow LED1 \rightarrow LED1 ambient \rightarrow LED2.
- 3. Finally, align the edges for the two LED pulses with the respective sampling instants.

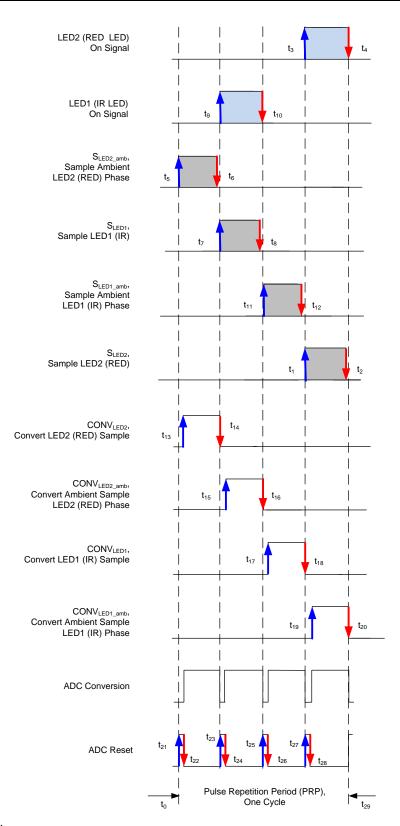
Table 2. Clock Edge Mapping to SPI Registers

TIME INSTANT (See Figure 63 and Figure 64)	DESCRIPTION	CORRESPONDING REGISTER ADDRESS AND REGISTER BITS	EXAMPLE ⁽¹⁾ (Decimal)	
t ₀	Start of pulse repetition period	No register control	_	
t ₁	Start of sample LED2 pulse	Sample LED2 start count (bits 15-0 of register 01h)	6050	
t ₂	End of sample LED2 pulse	Sample LED2 end count (bits 15-0 of register 02h)	7998	
t ₃	Start of LED2 pulse	LED2 start count (bits 15-0 of register 03h)	6000	
t ₄	End of LED2 pulse	LED2 end count (bits 15-0 of register 04h)	7999	
t ₅	Start of sample LED2 ambient pulse	Sample ambient LED2 start count (bits 15-0 of register 05h)	50	
t ₆	End of sample LED2 ambient pulse	Sample ambient LED2 end count (bits 15-0 of register 06h)	1998	
t ₇	Start of sample LED1 pulse	Sample LED1 start count (bits 15-0 of register 07h)	2050	
t ₈	End of sample LED1 pulse	Sample LED1 end count (bits 15-0 of register 08h)	3998	
t ₉	Start of LED1 pulse	LED1 start count (bits 15-0 of register 09h)	2000	
t ₁₀	End of LED1 pulse	LED1 end count (bits 15-0 of register 0Ah)	3999	
t ₁₁	Start of sample LED1 ambient pulse	Sample ambient LED1 start count (bits 15-0 of register 0Bh)	4050	
t ₁₂	End of sample LED1 ambient pulse	Sample ambient LED1 end count (bits 15-0 of register 0Ch)	5998	
t ₁₃	Start of convert LED2 pulse	LED2 convert start count (bits 15-0 of register 0Dh) Must start one AFE clock cycle after the ADC reset pulse ends.	4	
t ₁₄	End of convert LED2 pulse	LED2 convert end count (bits 15-0 of register 0Eh)	1999	
t ₁₅	Start of convert LED2 ambient pulse	LED2 ambient convert start count (bits 15-0 of register 0Fh) Must start one AFE clock cycle after the ADC reset pulse ends.	2004	
t ₁₆	End of convert LED2 ambient pulse	LED2 ambient convert end count (bits 15-0 of register 10h)	3999	
t ₁₇	Start of convert LED1 pulse	LED1 convert start count (bits 15-0 of register 11h) Must start one AFE clock cycle after the ADC reset pulse ends.	4004	
t ₁₈	End of convert LED1 pulse	LED1 convert end count (bits 15-0 of register 12h)	5999	
t ₁₉	Start of convert LED1 ambient pulse	LED1 ambient convert start count (bits 15-0 of register 13h) Must start one AFE clock cycle after the ADC reset pulse ends.	6004	
t ₂₀	End of convert LED1 ambient pulse	LED1 ambient convert end count (bits 15-0 of register 14h)	7999	
t ₂₁	Start of first ADC conversion reset pulse	ADC reset 0 start count (bits 15-0 of register 15h)	0	
t ₂₂	End of first ADC conversion reset pulse (2)	ADC reset 0 end count (bits 15-0 of register 16h)	3	
t ₂₃	Start of second ADC conversion reset pulse	ADC reset 1 start count (bits 15-0 of register 17h)	2000	
t ₂₄	End of second ADC conversion reset pulse (2)	ADC reset 1 end count (bits 15-0 of register 18h)	2003	
t ₂₅	Start of third ADC conversion reset pulse	ADC reset 2 start count (bits 15-0 of register 19h)	4000	
t ₂₆	End of third ADC conversion reset pulse (2)	ADC reset 2 end count (bits 15-0 of register 1Ah)	4003	
t ₂₇	Start of fourth ADC conversion reset pulse	ADC reset 3 start count (bits 15-0 of register 1Bh)	6000	
t ₂₈	End of fourth ADC conversion reset pulse ⁽²⁾	ADC reset 3 end count (bits 15-0 of register 1Ch)	6003	
t ₂₉	End of pulse repetition period	Pulse repetition period count (bits 15-0 of register 1Dh)	7999	

¹⁾ Values are based off of a pulse repetition frequency (PRF) = 500 Hz and duty cycle = 25%.

⁽²⁾ See Figure 64, note 2 for the affect of the ADC reset time crosstalk.



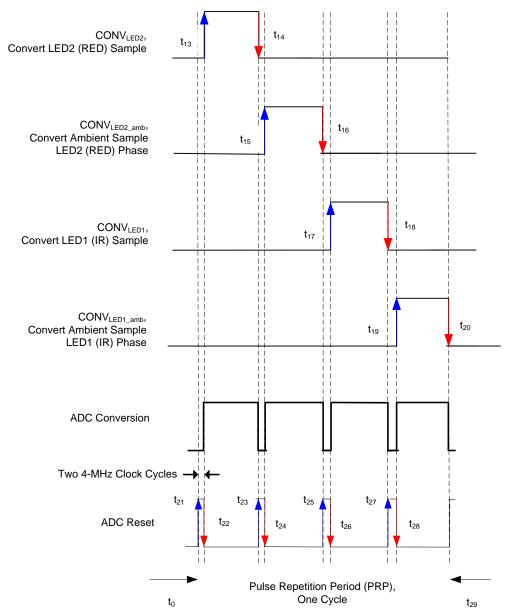


⁽¹⁾ RED = LED2, IR = LED1.

Figure 63. Programmable Clock Edges⁽¹⁾⁽²⁾

⁽²⁾ A low ADC reset time can result in a small component of the LED signal leaking into the ambient phase. With an ADC reset of two clock cycles, a -60-dB leakage is expected. In many cases, this leakage does not affect system performance. However, if this crosstalk must be completely eliminated, a longer ADC reset time of approximately six clock cycles is recommended for t_{22} , t_{24} , t_{26} , and t_{28} .





⁽¹⁾ RED = LED2, IR = LED1.

Figure 64. Relationship Between the ADC Reset and ADC Conversion Signals⁽¹⁾⁽²⁾

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⁽²⁾ A low ADC reset time can result in a small component of the LED signal leaking into the ambient phase. With an ADC reset of two clock cycles, a -60-dB leakage is expected. In many cases, this leakage does not affect system performance. However, if this crosstalk must be completely eliminated, a longer ADC reset time of approximately six clock cycles is recommended for t_{22} , t_{24} , t_{26} , and t_{28} .



8.3.4 Receiver Subsystem Power Path

The block diagram in Figure 65 shows the device Rx subsystem power routing.

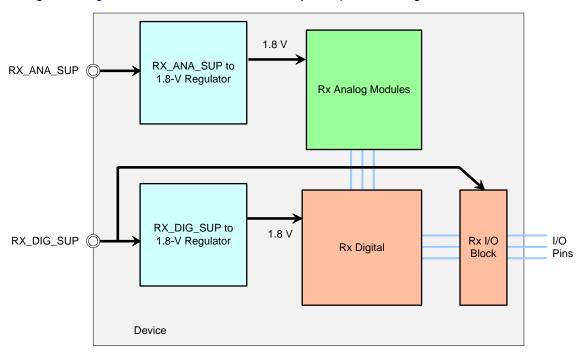


Figure 65. Receive Subsystem Power Routing

8.3.5 Transmit Section

The transmit section integrates the LED driver and the LED current control section with 8-bit resolution. This integration is designed to meet a dynamic range of better than 105 dB (based on a 1-sigma LED current noise).

The RED and IR LED reference currents can be independently set. The current source (I_{LED}) locally regulates and ensures that the actual LED current tracks the specified reference. The transmitter section uses a reference voltage for operation. This reference voltage is available on the REF_TX pin and must be decoupled to ground with a 2.2- μ F capacitor. The TX_REF voltage is derived from the TX_CTRL_SUP. The maximum LED current setting depends on the transmitter reference voltage. By default, after reset, this voltage is 0.75 V and supports up to a 150-mA LED current. For higher LED currents up to 200 mA, the reference can be programmed to 1.0 V (using the LED_RANGE[1:0] register bits).

The minimum LED_DRV_SUP voltage required for operation depends on the:

- Voltage drop across the LED (V_{LED}),
- Voltage drop across the external cable, connector, and any other component in series with the LED (V_{CABLE}), and
- Transmitter reference voltage.

Using the default reference voltage of 0.75 V, the minimum LED_DRV_SUP voltage can be as low as 3.25 V, provided that Equation 3 is met. Refer to the *Recommended Operating Conditions* table.

$$3.25 \text{ V} - (\text{V}_{\text{LED}} + \text{V}_{\text{CABLE}}) > 1.4 \text{ V}$$
 (3)

To lower the minimum LED_DRV_SUP voltage even further, the transmitter reference voltage can be programmed to 0.5 V. By doing so, the minimum LED_DRV_SUP voltage can be reduced to 3.0 V, provided that Equation 4 is met. Refer to the *Recommended Operating Conditions* table.

$$3.0 \text{ V} - (\text{V}_{\text{LED}} + \text{V}_{\text{CABLE}}) > 1.4 \text{ V}$$
 (4)

Note that with the 0.5-V transmitter reference voltage, the maximum LED current supported is 100 mA.



Two LED driver schemes are supported:

- An H-bridge drive for a two-pin, back-to-back LED package, as shown in Figure 66.
- A push-pull drive for a three-pin, common-anode LED package; see Figure 67.

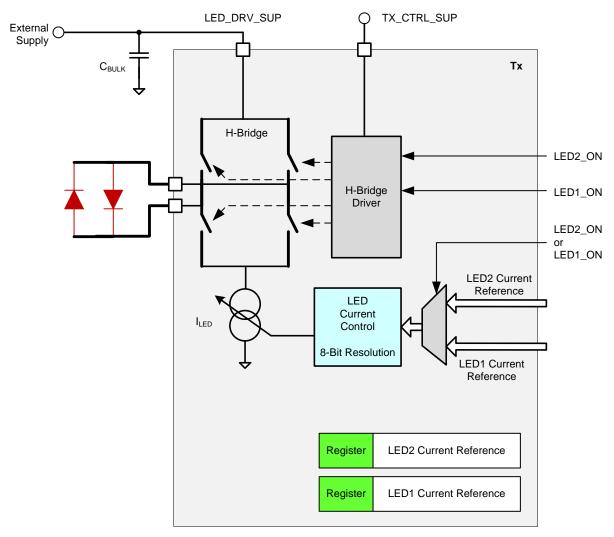


Figure 66. Transmit: H-Bridge Drive



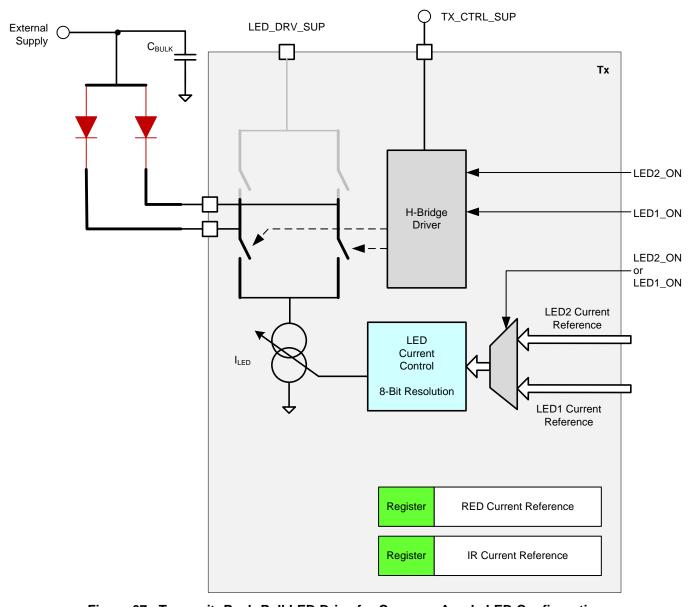


Figure 67. Transmit: Push-Pull LED Drive for Common Anode LED Configuration



8.3.5.1 Transmitter Power Path

The block diagram in Figure 68 shows the device Tx subsystem power routing.

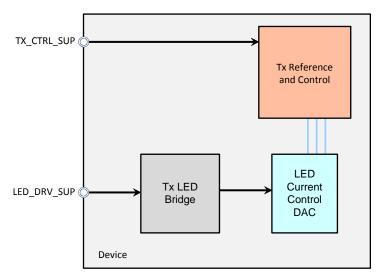


Figure 68. Transmit Subsystem Power Routing

8.3.5.2 LED Power Reduction During Periods of Inactivity

The diagram in Figure 69 shows how LED bias current passes 50 µA whenever LED_ON occurs. In order to minimize power consumption in periods of inactivity, the LED_ON control must be turned off. Furthermore, disable the TIMEREN bit in the CONTROL1 register by setting the value to '0'.

Note that depending on the LEDs used, the LED may sometimes appear dimly lit even when the LED current is set to 0 mA. This appearance is because of the switching leakage currents (as shown in Figure 69) inherent to the timer function. The dimmed appearance does not effect the ambient light level measurement because during the ambient cycle, LED_ON is turned off for the duration of the ambient measurement.

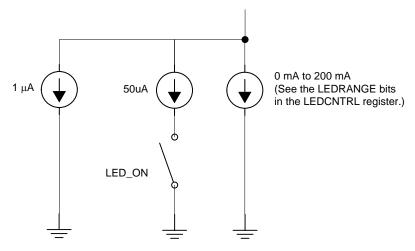


Figure 69. LED Bias Current



8.4 Device Functional Modes

8.4.1 ADC Operation and Averaging Module

After the falling edge of the ADC reset signal, the ADC conversion phase starts (refer to Figure 64). Each ADC conversion takes 50 µs.

There are two modes of operation: without averaging and with averaging. The averaging mode can average multiple ADC samples and reduce noise to improve dynamic range because the ADC conversion time is usually shorter than 25% of the pulse repetition period. Figure 70 shows a diagram of the averaging module. The ADC output format is in 22-bit twos complement. The two MSB bits of the 24-bit data can be ignored.

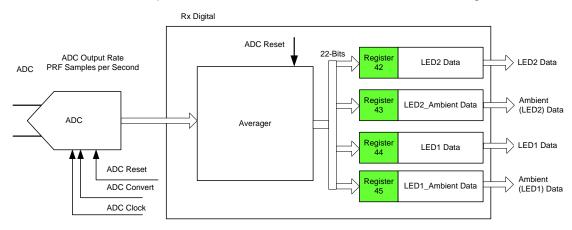


Figure 70. Averaging Module

8.4.1.1 Operation Without Averaging

In this mode, the ADC outputs a digital sample one time for every 50 µs. At the next rising edge of the ADC reset signal, the first 22-bit conversion value is written into the result registers sequentially as follows (see Figure 71):

- At the 25% reset signal, the first 22-bit ADC sample is written to register 2Ah.
- At the 50% reset signal, the first 22-bit ADC sample is written to register 2Bh.
- At the 75% reset signal, the first 22-bit ADC sample is written to register 2Ch.
- At the next 0% reset signal, the first 22-bit ADC sample is written to register 2Dh. The contents of registers 2Ah and 2Bh are written to register 2Eh and the contents of registers 2Ch and 2Dh are written to register 2Fh.

At the rising edge of the ADC_RDY signal, the contents of all six result registers can be read out.

8.4.1.2 Operation With Averaging

In this mode, all ADC digital samples are accumulated and averaged after every 50 μ s. At the next rising edge of the ADC reset signal, the average value (22-bit) is written into the output registers sequentially as follows (see Figure 72):

- At the 25% reset signal, the averaged 22-bit word is written to register 2Ah.
- At the 50% reset signal, the averaged 22-bit word is written to register 2Bh.
- At the 75% reset signal, the averaged 22-bit word is written to register 2Ch.
- At the next 0% reset signal, the averaged 22-bit word is written to register 2Dh. The contents of registers 2Ah and 2Bh are written to register 2Eh and the contents of registers 2Ch and 2Dh are written to register 2Fh.

At the rising edge of the ADC_RDY signal, the contents of all six result registers can be read out.



The number of samples to be used per conversion phase is specified in the CONTROL1 register (NUMAV[7:0]). The user must specify the correct value for the number of averages, as described in Equation 5:

When the number of averages is '0', the averaging is disabled and only one ADC sample is written to the result registers.

Note that he number of average conversions is limited by 25% of the PRF. For example, eight samples can be averaged with PRF = 625 Hz, and four samples can be averaged with PRF = 1250 Hz.



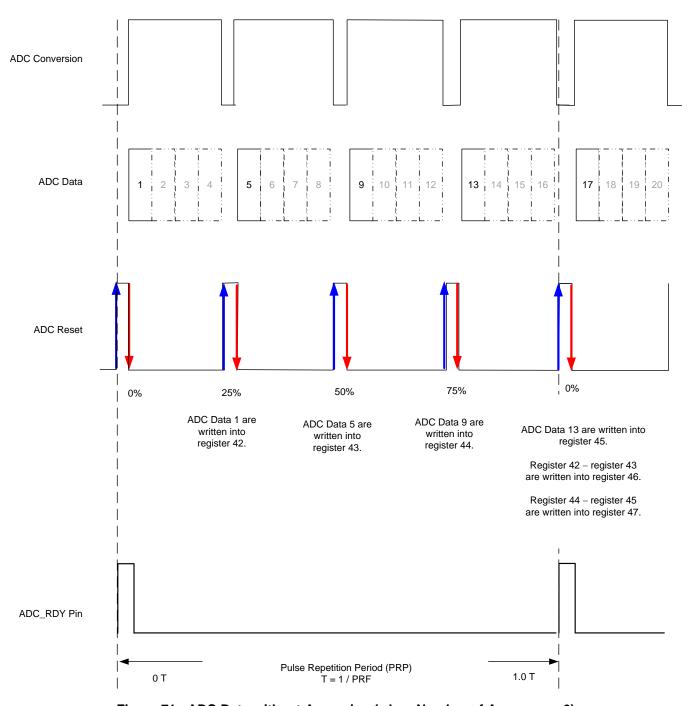
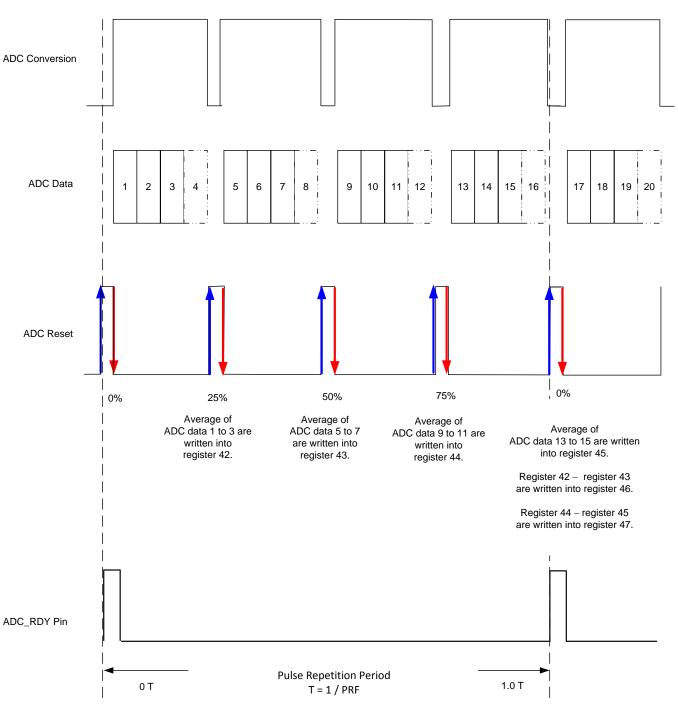


Figure 71. ADC Data without Averaging (when Number of Averages = 0)





NOTE: Example is with three averages. The value of the NUMAVG[7:0] register bits = 2.

Figure 72. ADC Data with Averaging Enabled



8.4.2 AFE Analog Output Mode (ADC Bypass Mode)

This mode is only intended for use in system debug. Note that this function is not recommended for production use because of the minimal device production testing performed on this function.

The ADC bypass mode brings out the analog output voltage of the receiver front-end on two pins (RXOUTP, RXOUTN), around a common-mode voltage of approximately 0.9 V. In this mode, the internal ADC of the AFE4490 is disabled. Figure 73 shows a block diagram of this mode.

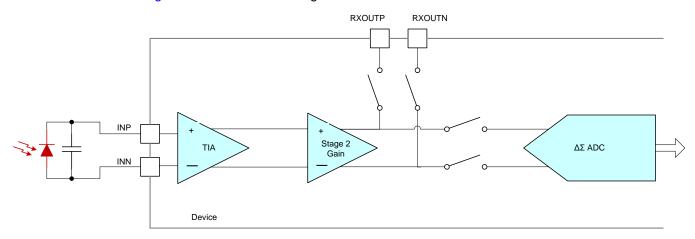


Figure 73. Device Set to ADC Bypass Mode

In ADC bypass mode, one of the internal clocks (ADC_Reset) can be brought out on the PD_ALM pin, as shown in Figure 74. This signal can be used to convert each of the four phases (within every pulse repetition period). Additionally, the ADC_RDY signal can be used to synchronize the external ADC with the AFE. See Figure 75 for the timing of this mode.

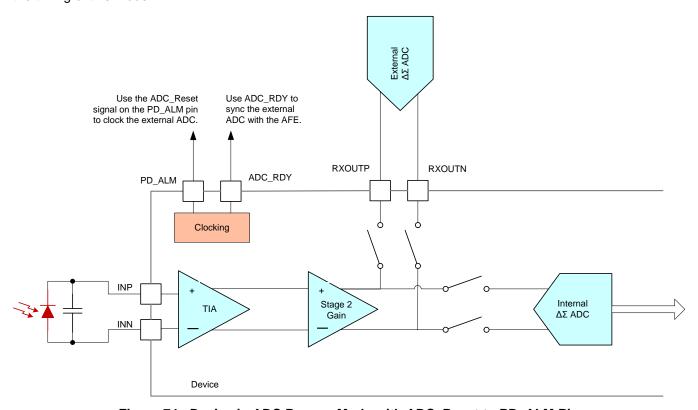
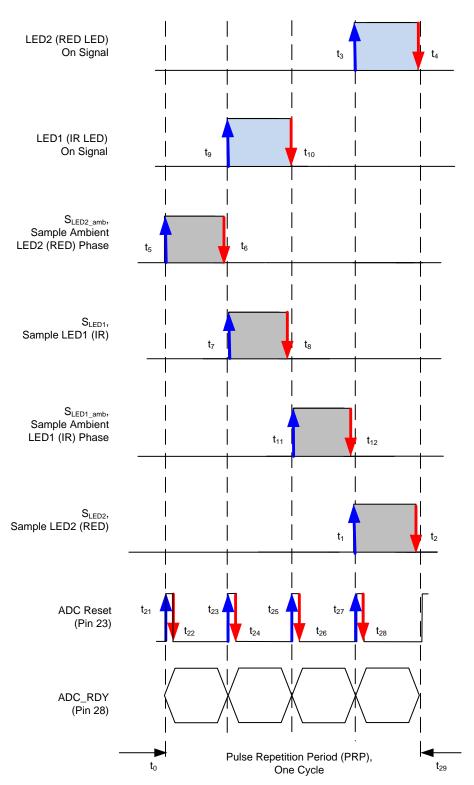


Figure 74. Device in ADC Bypass Mode with ADC_Reset to PD_ALM Pin





NOTE: RED = LED2, IR = LED1.

Figure 75. Device Analog Output Mode (ADC Bypass) Timing Diagram



In ADC bypass mode, the ADC reset signal can be used to start conversions with the external ADC. Use registers 15h through 1Ch to position the ADC reset signal edges appropriately. Also, use the CLKALMPIN[2:0] bits on the PD_ALM pin register bit to bring out the ADC reset signal to the PD_ALM pin. ADC_RDY can be used to indicate the start of the pulse repetition period to the external ADC.

8.4.3 Diagnostics

The device includes diagnostics to detect open or short conditions of the LED and photosensor, LED current profile feedback, and cable on or off detection.

8.4.3.1 Photodiode-Side Fault Detection

Figure 76 shows the diagnostic for the photodiode-side fault detection.

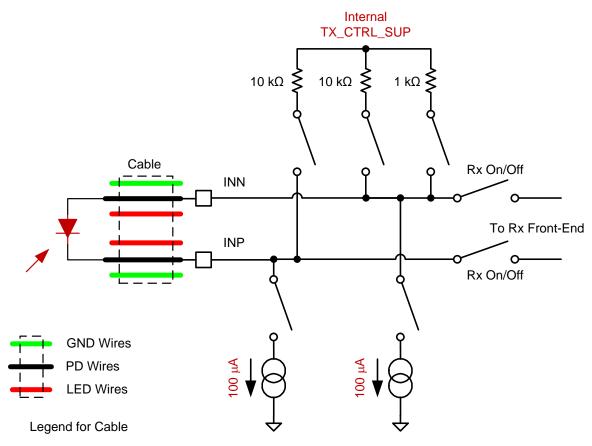


Figure 76. Photodiode Diagnostic



8.4.3.2 Transmitter-Side Fault Detection

Figure 77 shows the diagnostic for the transmitter-side fault detection.

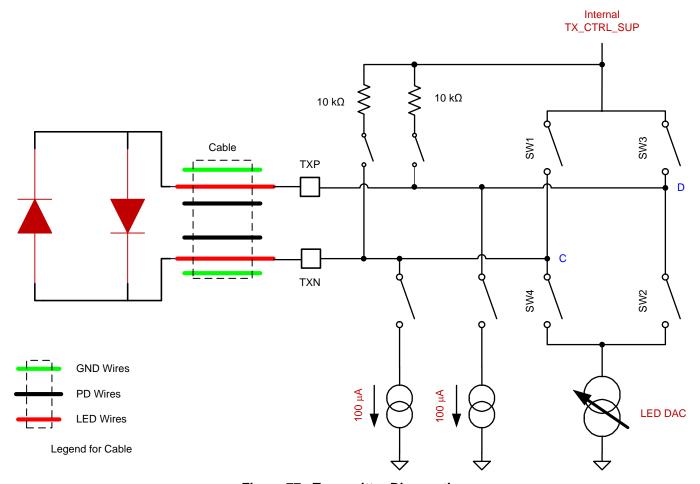


Figure 77. Transmitter Diagnostic



8.4.3.3 Diagnostics Module

The diagnostics module, when enabled, checks for nine types of faults sequentially. The results of all faults are latched in 11 separate flags. At the end of the sequence, the state of the 11 flags are combined to generate two interrupt signals: PD_ALM for photodiode-related faults and LED_ALM for transmit-related faults. The status of all flags can also be read using the SPI interface. Table 3 details each fault and flag used. Note that the diagnostics module requires all AFE blocks to be enabled in order to function reliably.

Table 3. Fault and Flag Diagnostics (1)

MODULE	SEQ.	FAULT	FLAG1	FLAG2	FLAG3	FLAG4	FLAG5	FLAG6	FLAG7	FLAG8	FLAG9	FLAG10	FLAG11
_	_	No fault	0	0	0	0	0	0	0	0	0	0	0
	1	Rx INP cable shorted to LED cable	1										
	2	Rx INN cable shorted to LED cable		1									
PD	3	Rx INP cable shorted to GND cable			1								
	4	Rx INN cable shorted to GND cable				1							
	5	PD open or shorted					1	1					
	6	Tx OUTM line shorted to GND cable							1				
LED	7	Tx OUTP line shorted to GND cable								1			
	8	LED open or shorted									1	1	
	9	LED open or shorted											1

(1) Resistances below 10 $k\Omega$ are considered to be shorted.

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Figure 78 shows the timing for the diagnostic function.

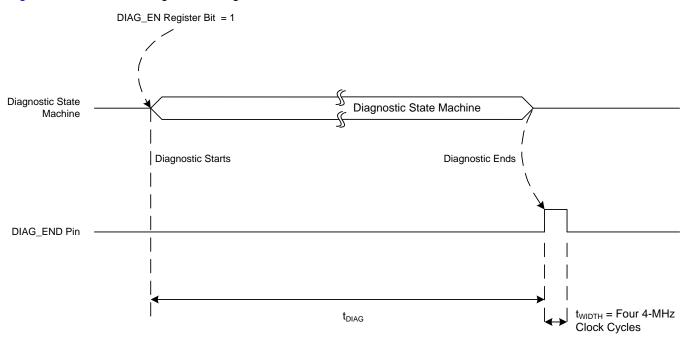


Figure 78. Diagnostic Timing Diagram

By default, the diagnostic function takes $t_{DIAG} = 8$ ms to complete after the DIAG_EN register bit is enabled. By setting the EN_SLOW_DIAG register bit (CONTROL2 register, bit D8) the diagnostic time can be increased to 16 ms.



After completion of the diagnostics function, time must be allowed for the device filter to settle. See the *Electrical Characteristics* for the filter settling time. The slow diagnostics feature is provided for use in systems where high-capacitance sensors (such as photodiodes, capacitors, cables, and so forth) are connected to the INP, INN, TXP, or TXN pins.

8.5 Programming

8.5.1 Serial Programming Interface

The SPI-compatible serial interface consists of four signals: SCLK (serial clock), SPISOMI (serial interface data output), SPISIMO (serial interface data input), and SPISTE (serial interface enable).

The serial clock (SCLK) is the serial peripheral interface (SPI) serial clock. SCLK shifts in commands and shifts out data from the device. SCLK features a Schmitt-triggered input and clocks data out on SPISOMI. Data are clocked in on the SPISIMO pin. Even though the input has hysteresis, TI recommends keeping SCLK as clean as possible to prevent glitches from accidentally shifting the data. When the serial interface is idle, hold SCLK low.

The SPISOMI (SPI serial out master in) pin is used with SCLK to clock out device data. The SPISIMO (SPI serial in master out) pin is used with SCLK to clock in data to the device. The SPISTE (SPI serial interface enable) pin enables the serial interface to clock data on the SPISIMO pin in to the device.

8.5.2 Reading and Writing Data

The device has a set of internal registers that can be accessed by the serial programming interface formed by the SPISTE, SCLK, SPISIMO, and SPISOMI pins.

8.5.2.1 Writing Data

The SPI_READ register bit must be first set to '0' before writing to a register. When SPISTE is low,

- · Serially shifting bits into the device is enabled.
- Serial data (on the SPISIMO pin) are latched at every SCLK rising edge.
- The serial data are loaded into the register at every 32nd SCLK rising edge.

In case the word length exceeds a multiple of 32 bits, the excess bits are ignored. Data can be loaded in multiples of 32-bit words within a single active SPISTE pulse. The first eight bits form the register address and the remaining 24 bits form the register data. Figure 79 shows an SPI timing diagram for a single write operation. For multiple read and write cycles, refer to the *Multiple Data Reads and Writes* section.

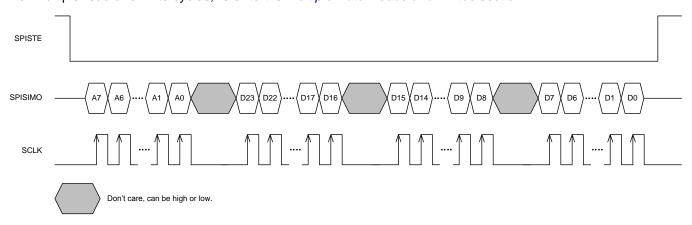


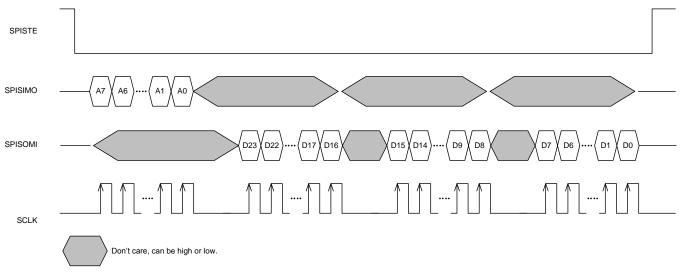
Figure 79. AFE SPI Write Timing Diagram



Programming (continued)

8.5.2.2 Reading Data

The SPI_READ register bit must be first set to '1' before reading from a register. The device includes a mode where the contents of the internal registers can be read back on the SPISOMI pin. This mode may be useful as a diagnostic check to verify the serial interface communication between the external controller and the AFE. To enable this mode, first set the SPI_READ register bit using the SPI write command, as described in the *Writing Data* section. In the next command, specify the SPI register address with the desired content to be read. Within the same SPI command sequence, the AFE outputs the contents of the specified register on the SPISOMI pin. Figure 80 shows an SPI timing diagram for a single read operation. For multiple read and write cycles, refer to the *Multiple Data Reads and Writes* section.



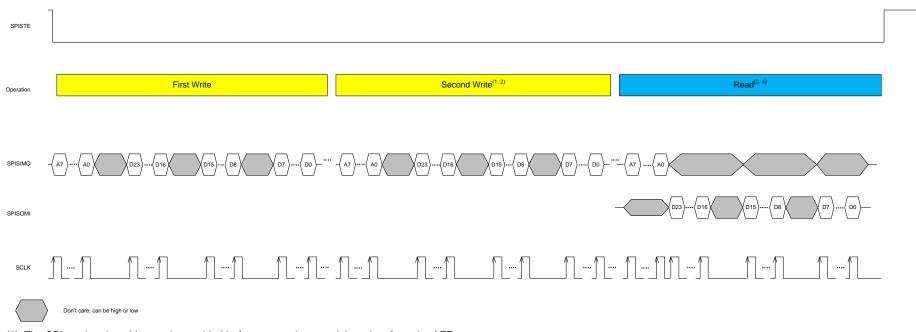
- (1) The SPI READ register bit must be enabled before attempting a serial readout from the AFE.
- (2) Specify the register address of the content that must be readback on bits A[7:0].
- (3) The AFE outputs the contents of the specified register on the SPISOMI pin.

Figure 80. AFE SPI Read Timing Diagram⁽¹⁾⁽²⁾⁽³⁾



8.5.2.3 Multiple Data Reads and Writes

The device includes functionality where multiple read and write operations can be performed during a single SPISTE event. To enable this functionality, the first eight bits determine the register address to be written and the remaining 24 bits determine the register data. Perform two writes with the SPI read bit enabled during the second write operation in order to prepare for the read operation, as described in the *Writing Data* section. In the next command, specify the SPI register address with the desired content to be read. Within the same SPI command sequence, the AFE outputs the contents of the specified register on the SPISOMI pin. This functionality is described in the *Writing Data* and *Reading Data* sections. Figure 81 shows a timing diagram for the SPI multiple read and write operations.



- (1) The SPI read register bit must be enabled before attempting a serial readout from the AFE.
- (2) The second write operation must be configured for register 0 with data 000001h.
- (3) Specify the register address whose contents must be read back on A[7:0].
- (4) The AFE outputs the contents of the specified register on the SOMI pin.

Figure 81. Serial Multiple Read and Write Operations

Product Folder Links: AFE4490

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8.5.2.4 Register Initialization

After power-up, the internal registers must be initialized to the default values. This initialization can be done in one of two ways:

- Through a hardware reset by applying a low-going pulse on the RESET pin, or
- By applying a software reset. Using the serial interface, set SW_RESET (bit D3 in register 00h) high. This
 setting initializes the internal registers to the default values and then self-resets to '0'. In this case, the RESET
 pin is kept high (inactive).

8.5.2.5 AFE SPI Interface Design Considerations

Note that when the device is deselected, the SPISOMI, CLKOUT, ADC_RDY, PD_ALM, LED_ALM, and DIAG_END digital output pins do not enter a 3-state mode. This condition, therefore, must be taken into account when connecting multiple devices to the SPI port and for power-management considerations. In order to avoid loading the SPI bus when multiple devices are connected, the DIGOUT_TRISTATE register bit must be to '1' whenever the AFE SPI is inactive.



8.6 Register Maps

The AFE consists of a set of registers that can be used to configure it, such as receiver timings, I-V amplifier settings, transmit LED currents, and so forth. The registers and their contents are listed in Table 4. These registers can be accessed using the AFE SPI interface.

Table 4. AFE Register Map

	REGISTER	ADDI	RESS												REGIS	TER DA	TA										
NAME	CONTROL ⁽¹⁾	Hex	Dec	D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
CONTROLO	W	00	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	SW_RST	DIAG_EN	TIM_COUNT_RST	SPI_READ
LED2STC	R/W	01	1	0	0	0	0	0	0	0	0								LED2S	STC[15:0	0]						
LED2ENDC	R/W	02	2	0	0	0	0	0	0	0	0								LED2E	NDC[15:	:0]						
LED2LEDSTC	R/W	03	3	0	0	0	0	0	0	0	0								LED2LE	DSTC[1	5:0]						
LED2LEDENDC	R/W	04	4	0	0	0	0	0	0	0	0							L	ED2LED	DENDC[15:0]						
ALED2STC	R/W	05	5	0	0	0	0	0	0	0	0								ALED2	STC[15:	:0]						
ALED2ENDC	R/W	06	6	0	0	0	0	0	0	0	0								ALED2E	NDC[15	5:0]						
LED1STC	R/W	07	7	0	0	0	0	0	0	0	0								LED18	STC[15:0	0]						
LED1ENDC	R/W	08	8	0	0	0	0	0	0	0	0								LED1E	NDC[15	:0]						
LED1LEDSTC	R/W	09	9	0	0	0	0	0	0	0	0								LED1LE	DSTC[1	5:0]						
LED1LEDENDC	R/W	0A	10	0	0	0	0	0	0	0	0							L	ED1LED	DENDC[15:0]						
ALED1STC	R/W	0B	11	0	0	0	0	0	0	0	0								ALED1	STC[15:	:0]						
ALED1ENDC	R/W	0C	12	0	0	0	0	0	0	0	0								ALED1E	NDC[15	5:0]						
LED2CONVST	R/W	0D	13	0	0	0	0	0	0	0	0								LED2CO	NVST[1	5:0]						
LED2CONVEND	R/W	0E	14	0	0	0	0	0	0	0	0							L	ED2CON	NVEND[15:0]						
ALED2CONVST	R/W	0F	15	0	0	0	0	0	0	0	0							P	LED2C0)TSVNC	15:0]						
ALED2CONVEND	R/W	10	16	0	0	0	0	0	0	0	0							Al	ED2CO	NVEND	[15:0]						
LED1CONVST	R/W	11	17	0	0	0	0	0	0	0	0								LED1CO	NVST[1	5:0]						
LED1CONVEND	R/W	12	18	0	0	0	0	0	0	0	0							L	ED1CON	NVEND[15:0]						
ALED1CONVST	R/W	13	19	0	0	0	0	0	0	0	0							A	LED1C0	ONVST[15:0]						
ALED1CONVEND	R/W	14	20	0	0	0	0	0	0	0	0							Al	ED1CO	NVEND	[15:0]						
ADCRSTSTCT0	R/W	15	21	0	0	0	0	0	0	0	0								ADCRS	TCT0[15	5:0]						
ADCRSTENDCT0	R/W	16	22	0	0	0	0	0	0	0	0								ADCREN	NDCT0[1	5:0]						
ADCRSTSTCT1	R/W	17	23	0	0	0	0	0	0	0	0								ADCRS	TCT1[15	5:0]						
ADCRSTENDCT1	R/W	18	24	0	0	0	0	0	0	0	0								ADCREN	NDCT1[1	5:0]						
ADCRSTSTCT2	R/W	19	25	0	0	0	0	0	0	0	0								ADCRS	TCT2[15	5:0]						
ADCRSTENDCT2	R/W	1A	26	0	0	0	0	0	0	0	0								ADCREN	NDCT2[1	5:0]						
ADCRSTSTCT3	R/W	1B	27	0	0	0	0	0	0	0	0								ADCRS	TCT3[15	5:0]						
ADCRSTENDCT3	R/W	1C	28	0	0	0	0	0	0	0	0								ADCREN	NDCT3[1	5:0]						

⁽¹⁾ R = read only, R/W = read or write, N/A = not available, and W = write only.



Register Maps (continued)

Table 4. AFE Register Map (continued)

	REGISTER	ADDI	RESS												REGIST	TER DA	TA										
NAME	CONTROL ⁽¹⁾	Hex	Dec	D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
PRPCOUNT	R/W	1D	29	0	0	0	0	0	0	0	0								PRP	CT[15:0]							
CONTROL1	R/W	1E	30	0	0	0	0	0	0	0	0	0	0	0	0	CLK	KALMPIN	[2:0]	TIMEREN				NUM	IAV[7:0]			
SPARE1	N/A	1F	31	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
TIAGAIN	R/W	20	32	0	0	0	0	0	0	0	0	ENSEPGAN	STAGE2EN1	0	0	0	STG	2GAIN1	[2:0]		CF	LED1[4:0]		F	F_LED1	[2:0]
TIA_AMB_GAIN	R/W	21	33	0	0	0	0		AMBD	AC[3:0]		FLTRCNRSEL	STAGE2EN2	0	0	0	STG	32GAIN2	2[2:0]		CF	LED2[4:0]		F	F_LED2	[2:0]
LEDCNTRL	R/W	22	34	0	0	0	0	0	0	LE RANG	ED SE[1:0]			•	LED	1[7:0]							LE	02[7:0]			
CONTROL2	R/W	23	35	0	0	0	0	0	TX_REF1	TX_REF0	RST_CLK_ON_PD_ALM	EN_ADC_BYP	0	0	0	TXBRGMOD	DIGOUT_TRISTATE	XTALDIS	EN_SLOW_DIAG	0	0	0	0	0	PDNTX	PDNRX	PDNAFE
SPARE2	N/A	24	36	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
SPARE3	N/A	25	37	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
SPARE4	N/A	26	38	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
RESERVED1	N/A	27	39	Х	Х	Х	Х	Х	Х	Х	Х	Х	Х	Х	Х	Х	Х	Х	Х	Х	Х	Х	Х	Х	Х	Х	Х
RESERVED2	N/A	28	40	Х	Х	Х	Х	Х	Х	Х	Х	Х	Х	Х	Х	Х	Х	Х	Х	Х	Х	Х	Х	Х	Х	Х	Х
ALARM	R/W	29	41	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	ALMPINCLKEN	0	0	0	0	0	0	0



Register Maps (continued)

Table 4. AFE Register Map (continued)

NAME	REGISTER	ADD	RESS												REGIS	TER DA	TA										
NAME	CONTROL ⁽¹⁾	Hex	Dec	D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
LED2VAL	R	2A	42												LED2	VAL[23:0	0]										
ALED2VAL	R	2B	43												ALED2	2VAL[23:	0]										
LED1VAL	R	2C	44												LED1	VAL[23:0	0]										
ALED1VAL	R	2D	45												ALED1	VAL[23:	0]										
LED2-ALED2VAL	R	2E	46											LE	D2-ALE	D2VAL[23:0]										
LED1-ALED1VAL	R	2F	47											LE	D1-ALE	D1VAL[[23:0]										
DIAG	R	30	48	0	0	0	0	0	0	0	0	0	0	0	PD_ALM	LED_ALM	LED10PEN	LEDZOPEN	LEDSC	OUTPSHGN	OUTNSHGND	PDOC	PDSC	INNSCGND	INPSCGND	INNSCLED	INPSCLED



8.6.1 AFE Register Description

Figure 82. CONTROL0: Control Register 0 (Address = 00h, Reset Value = 0000h)

D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12
0	0	0	0	0	0	0	0	0	0	0	0
W-0h	W-0h	W-0h	W-0h								
D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
0	0	0	0	0	0	0	0	SW_RST	DIAG_EN	TIM_ COUNT_ RST	SPI_ READ
W-0h	W-0h	W-0h	W-0h								

LEGEND: W = Write only; -n = value after reset

This register is write-only. CONTROL0 is used for AFE software and count timer reset, diagnostics enable, and SPI read functions.

Bits D[23:4] Must be '0'

Bit D3 SW_RST: Software reset

0 = No action (default after reset)

1 = Software reset applied; resets all internal registers to the default values and self-clears

to '0'

Bit D2 DIAG EN: Diagnostic enable

0 = No Action (default after reset)

1 = Diagnostic mode is enabled and the diagnostics sequence starts when this bit is set.

At the end of the sequence, all fault statuses are stored in the *DIAG: Diagnostics Flag*

Register. Afterwards, the DIAG_EN register bit self-clears to '0'.

Note that the diagnostics enable bit is automatically reset after the diagnostics completes (slow =16 ms, fast = 8 ms). During diagnostics mode, the ADC data are invalid because of

toggling diagnostics switches.

Bit D1 TIM CNT RST: Timer counter reset

0 = Disables timer counter reset, required for normal timer operation (default after reset)

1 = Timer counters are in reset state

Bit D0 SPI READ: SPI read

0 = SPI read is disabled (default after reset)

1 = SPI read is enabled

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Figure 83. LED2STC: Sample LED2 Start Count Register (Address = 01h, Reset Value = 0000h)

D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12			
0	0	0	0	0	0	0	0		LED2S1	ΓC[15:0]				
R/W-0h	R/W-0h		R/W-0h											
D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0			
					LED2S	TC[15:0]								
					R/W	V-0h								

This register sets the start timing value for the LED2 signal sample.

Bits D[23:16] Must be '0'

Bits D[15:0] LED2STC[15:0]: Sample LED2 start count

The contents of this register can be used to position the start of the sample LED2 signal with respect to the pulse repetition period (PRP), as specified in the PRPCOUNT register. The count is specified as the number of 4-MHz clock cycles. Refer to the *Using the Timer Module* section for details.

Figure 84. LED2ENDC: Sample LED2 End Count Register (Address = 02h, Reset Value = 0000h)

D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12			
0	0	0	0	0	0	0	0		LED2EN	DC[15:0]				
R/W-0h	R/W-0h		R/W-0h											
D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0			
					LED2EN	DC[15:0]								
					R/W	/-0h								

LEGEND: R/W = Read/Write; -n = value after reset

This register sets the end timing value for the LED2 signal sample.

Bits D[23:16] Must be '0'

Bits D[15:0] LED2ENDC[15:0]: Sample LED2 end count

The contents of this register can be used to position the end of the sample LED2 signal with respect to the PRP, as specified in the PRPCOUNT register. The count is specified as the number of 4-MHz clock cycles. Refer to the *Using the Timer Module* section for details.

Figure 85. LED2LEDSTC: LED2 LED Start Count Register (Address = 03h, Reset Value = 0000h)

D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12				
0	0	0	0	0	0	0	0		LED2LED	STC[15:0]					
R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h		R/W-0h						
D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0				
					LED2LED	STC[15:0]									
	<u>'</u>		<u>'</u>	<u>'</u>	R/W	V-0h	<u>'</u>	·		<u>'</u>					

LEGEND: R/W = Read/Write; -n = value after reset

This register sets the start timing value for when the LED2 signal turns on.

Bits D[23:16] Must be '0'

Bits D[15:0] LED2LEDSTC[15:0]: LED2 start count

The contents of this register can be used to position the start of the LED2 with respect to the PRP, as specified in the PRPCOUNT register. The count is specified as the number of 4-MHz clock cycles. Refer to the *Using the Timer Module* section for details.



Figure 86. LED2LEDENDC: LED2 LED End Count Register (Address = 04h, Reset Value = 0000h)

D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12			
0	0	0	0	0	0	0	0		LED2LEDE	NDC[15:0]				
R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h		R/W-0h					
D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0			
					LED2LEDE	NDC[15:0]								
					R/W	/-0h								

This register sets the end timing value for when the LED2 signal turns off.

Bits D[23:16] Must be '0'

Bits D[15:0] LED2LEDENDC[15:0]: LED2 end count

The contents of this register can be used to position the end of the LED2 signal with respect to the PRP, as specified in the PRPCOUNT register. The count is specified as the number of 4-MHz clock cycles. Refer to the *Using the Timer Module* section for details.

Figure 87. ALED2STC: Sample Ambient LED2 Start Count Register (Address = 05h, Reset Value = 0000h)

D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12			
0	0	0	0	0	0	0	0		ALED2S	TC[15:0]				
R/W-0h	R/W-0h		R/W-0h											
D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0			
					ALED2S	TC[15:0]								
					R/W	V-0h								

LEGEND: R/W = Read/Write; -n = value after reset

This register sets the start timing value for the ambient LED2 signal sample.

Bits D[23:16] Must be '0'

Bits D[15:0] ALED2STC[15:0]: Sample ambient LED2 start count

The contents of this register can be used to position the start of the sample ambient LED2 signal with respect to the PRP, as specified in the PRPCOUNT register. The count is specified as the number of 4-MHz clock cycles. Refer to the *Using the Timer Module* section for details.



Figure 88. ALED2ENDC: Sample Ambient LED2 End Count Register (Address = 06h, Reset Value = 0000h)

D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12			
0	0	0	0	0	0	0	0		ALED2EN	NDC[15:0]				
R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h		R/W-0h					
D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0			
					ALED2EN	NDC[15:0]								
					R/V	V-0h								

This register sets the end timing value for the ambient LED2 signal sample.

Bits D[23:16] Must be '0'

Bits D[15:0] ALED2ENDC[15:0]: Sample ambient LED2 end count

The contents of this register can be used to position the end of the sample ambient LED2 signal with respect to the PRP, as specified in the PRPCOUNT register. The count is specified as the number of 4-MHz clock cycles. Refer to the *Using the Timer Module* section for details.

Figure 89. LED1STC: Sample LED1 Start Count Register (Address = 07h, Reset Value = 0000h)

D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12			
0	0	0	0	0	0	0	0		LED1S	ΓC[15:0]				
R/W-0h	R/W-0h		R/W-0h											
D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0			
					LED1S	ΓC[15:0]								
					R/W	V-0h								

LEGEND: R/W = Read/Write; -n = value after reset

This register sets the start timing value for the LED1 signal sample.

Bits D[23:17] Must be '0'

Bits D[16:0] LED1STC[15:0]: Sample LED1 start count

The contents of this register can be used to position the start of the sample LED1 signal with respect to the PRP, as specified in the PRPCOUNT register. The count is specified as the number of 4-MHz clock cycles. Refer to the *Using the Timer Module* section for details.



Figure 90. LED1ENDC: Sample LED1 End Count (Address = 08h, Reset Value = 0000h)

D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12	
0	0	0	0	0	0	0	0		LED1EN	DC[15:0]		
R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h				
D11	D10	D9	D8	D7	D6	D5	D4	D3 D2 D1 D0				
	LED1ENDC[15:0]											
	R/W-0h											

This register sets the end timing value for the LED1 signal sample.

Bits D[23:17] Must be '0'

Bits D[16:0] LED1ENDC[15:0]: Sample LED1 end count

> The contents of this register can be used to position the end of the sample LED1 signal with respect to the PRP, as specified in the PRPCOUNT register. The count is specified as the number of 4-MHz clock cycles. Refer to the *Using the Timer Module* section for details.

Figure 91. LED1LEDSTC: LED1 LED Start Count Register (Address = 09h, Reset Value = 0000h)

D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12		
0	0	0	0	0	0	0	0		LED1LED	STC[15:0]			
R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h		R/W-0h				
D11	D10	D9	D8	D7	D6	D5	D4	D3 D2 D1 D					
	LED1LEDSTC[15:0]												
	R/W-0h												

LEGEND: R/W = Read/Write: -n = value after reset

This register sets the start timing value for when the LED1 signal turns on.

Bits D[23:16] Must be '0'

Bits D[15:0] LED1LEDSTC[15:0]: LED1 start count

> The contents of this register can be used to position the start of the LED1 signal with respect to the PRP, as specified in the PRPCOUNT register. The count is specified as the number of 4-MHz clock cycles. Refer to the *Using the Timer Module* section for details.

Figure 92. LED1LEDENDC: LED1 LED End Count Register (Address = 0Ah, Reset Value = 0000h)

D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12		
0	0	0	0	0	0	0	0		LED1LEDE	NDC[15:0]			
R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h		R/W-0h				
D11	D10	D9	D8	D7	D6	D5	D4	D3	D2 D1 D0				
					LED1LEDE	NDC[15:0]							
	R/W-0h												

LEGEND: R/W = Read/Write; -n = value after reset

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This register sets the end timing value for when the LED1 signal turns off.

Bits D[23:16] Must be '0'

Bits D[15:0] LED1LEDENDC[15:0]: LED1 end count

> The contents of this register can be used to position the end of the LED1 signal with respect to the PRP, as specified in the PRPCOUNT register. The count is specified as the number of 4-MHz clock cycles. Refer to the *Using the Timer Module* section for details.



Figure 93. ALED1STC: Sample Ambient LED1 Start Count Register (Address = 0Bh, Reset Value = 0000h)

D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12	
0	0	0	0	0	0	0	0		ALED1S	TC[15:0]		
R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h		R/W-0h			
D11	D10	D9	D8	D7	D6	D5	D4	D3 D2 D1 D0				
	ALED1STC[15:0]											
	R/W-0h											

This register sets the start timing value for the ambient LED1 signal sample.

Bits D[23:16] Must be '0'

Bits D[15:0] ALED1STC[15:0]: Sample ambient LED1 start count

The contents of this register can be used to position the start of the sample ambient LED1 signal with respect to the PRP, as specified in the PRPCOUNT register. The count is specified as the number of 4-MHz clock cycles. Refer to the *Using the Timer Module* section for details.

Figure 94. ALED1ENDC: Sample Ambient LED1 End Count Register (Address = 0Ch, Reset Value = 0000h)

D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12		
0	0	0	0	0	0	0	0		ALED1EN	NDC[15:0]			
R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h		R/W-0h				
D11	D10	D9	D8	D7	D6	D5	D4	D3 D2 D1 D0					
	ALED1ENDC[15:0]												
	R/W-0h												

LEGEND: R/W = Read/Write; -n = value after reset

This register sets the end timing value for the ambient LED1 signal sample.

Bits D[23:16] Must be '0'

Bits D[15:0] ALED1ENDC[15:0]: Sample ambient LED1 end count

The contents of this register can be used to position the end of the sample ambient LED1 signal with respect to the PRP, as specified in the PRPCOUNT register. The count is specified as the number of 4-MHz clock cycles. Refer to the *Using the Timer Module* section for details.



Figure 95. LED2CONVST: LED2 Convert Start Count Register (Address = 0Dh, Reset Value = 0000h)

D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12		
0	0	0	0	0	0	0	0		LED2CON	IVST[15:0]			
R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h		R/W-0h				
D11	D10	D9	D8	D7	D6	D5	D4	D3 D2 D1 D0					
LED2CONVST[15:0]													
	R/W-0h												

This register sets the start timing value for the LED2 conversion.

Bits D[23:16] Must be '0'

Bits D[15:0] LED2CONVST[15:0]: LED2 convert start count

The contents of this register can be used to position the start of the LED2 conversion signal with respect to the PRP, as specified in the PRPCOUNT register. The count is specified as the number of 4-MHz clock cycles. Refer to the *Using the Timer Module* section for details.

Figure 96. LED2CONVEND: LED2 Convert End Count Register (Address = 0Eh, Reset Value = 0000h)

D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12		
0	0	0	0	0	0	0	0		LED2CON\	VEND[15:0]			
R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h		R/W-0h				
D11	D10	D9	D8	D7	D6	D5	D4	D3	D2 D1 D0				
					LED2CON\	VEND[15:0]							
	R/W-0h												

LEGEND: R/W = Read/Write: -n = value after reset

This register sets the end timing value for the LED2 conversion.

Bits D[23:16] Must be '0'

Bits D[15:0] LED2CONVEND[15:0]: LED2 convert end count

The contents of this register can be used to position the end of the LED2 conversion signal with respect to the PRP, as specified in the PRPCOUNT register. The count is specified as the number of 4-MHz clock cycles. Refer to the *Using the Timer Module* section for details.



Figure 97. ALED2CONVST: LED2 Ambient Convert Start Count Register (Address = 0Fh, Reset Value = 0000h)

D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12		
0	0	0	0	0	0	0	0		ALED2COI	NVST[15:0]			
R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h		R/W-0h				
D11	D10	D9	D8	D7	D6	D5	D4	D3	D2 D1 D0				
					ALED2COI	NVST[15:0]							
	R/W-0h												

This register sets the start timing value for the ambient LED2 conversion.

Bits D[23:16] Must be '0'

Bits D[15:0] ALED2CONVST[15:0]: LED2 ambient convert start count

The contents of this register can be used to position the start of the LED2 ambient conversion signal with respect to the PRP, as specified in the PRPCOUNT register. The count is specified as the number of 4-MHz clock cycles. Refer to the *Using the Timer Module* section for details.

Figure 98. ALED2CONVEND: LED2 Ambient Convert End Count Register (Address = 10h, Reset Value = 0000h)

D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12		
0	0	0	0	0	0	0	0		ALED2CON	VEND[15:0]			
R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h		R/W-0h				
D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D0			
					ALED2CON	IVEND[15:0]							
	R/W-0h												

LEGEND: R/W = Read/Write; -n = value after reset

This register sets the end timing value for the ambient LED2 conversion.

Bits D[23:16] Must be '0'

Bits D[15:0] ALED2CONVEND[15:0]: LED2 ambient convert end count

The contents of this register can be used to position the end of the LED2 ambient conversion signal with respect to the PRP. The count is specified as the number of 4-MHz clock cycles. Refer to the *Using the Timer Module* section for details.



Figure 99. LED1CONVST: LED1 Convert Start Count Register (Address = 11h, Reset Value = 0000h)

D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12	
0	0	0	0	0	0	0	0		LED1CON	IVST[15:0]		
R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h		R/W-0h D2 D1 D0			
D11	D10	D9	D8	D7	D6	D5	D4	D3	<u> </u>			
					LED1CON	IVST[15:0]						
	R/W-0h											

This register sets the start timing value for the LED1 conversion.

Bits D[23:16] Must be '0'

Bits D[15:0] LED1CONVST[15:0]: LED1 convert start count

The contents of this register can be used to position the start of the LED1 conversion signal with respect to the PRP, as specified in the PRPCOUNT register. The count is specified as the number of 4-MHz clock cycles. Refer to the *Using the Timer Module* section for details.

Figure 100. LED1CONVEND: LED1 Convert End Count Register (Address = 12h, Reset Value = 0000h)

D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12	
0	0	0	0	0	0	0	0		LED1CON\	VEND[15:0]		
R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h		R/W-0h			
D11	D10	D9	D8	D7	D6	D5	D4	D3	D2 D1 D0			
					LED1CON	VEND[15:0]						
R/W-0h												

LEGEND: R/W = Read/Write: -n = value after reset

This register sets the end timing value for the LED1 conversion.

Bits D[23:16] Must be '0'

Bits D[15:0] LED1CONVEND[15:0]: LED1 convert end count

The contents of this register can be used to position the end of the LED1 conversion signal with respect to the PRP. The count is specified as the number of 4-MHz clock cycles. Refer to the *Using the Timer Module* section for details.



Figure 101. ALED1CONVST: LED1 Ambient Convert Start Count Register (Address = 13h, Reset Value = 0000h)

D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12		
0	0	0	0	0	0	0	0		ALED1CO	NVST[15:0]			
R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h		R/W-0h				
D11	D10	D9	D8	D7	D6	D5	D4	D3	D2 D1 D0				
					ALED1CO	NVST[15:0]							
	R/W-0h												

This register sets the start timing value for the ambient LED1 conversion.

Bits D[23:16] Must be '0'

Bits D[15:0] ALED1CONVST[15:0]: LED1 ambient convert start count

The contents of this register can be used to position the start of the LED1 ambient conversion signal with respect to the PRP, as specified in the PRPCOUNT register. The count is specified as the number of 4-MHz clock cycles. Refer to the *Using the Timer Module* section for details.

Figure 102. ALED1CONVEND: LED1 Ambient Convert End Count Register (Address = 14h, Reset Value = 0000h)

D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12	
0	0	0	0	0	0	0	0		ALED1CON	VEND[15:0]		
R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h		R/W-0h			
D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0	
					ALED1CON	IVEND[15:0]						
					R/W	V-0h						

LEGEND: R/W = Read/Write; -n = value after reset

This register sets the end timing value for the ambient LED1 conversion.

Bits D[23:16] Must be '0'

Bits D[15:0] ALED1CONVEND[15:0]: LED1 ambient convert end count

The contents of this register can be used to position the end of the LED1 ambient conversion signal with respect to the PRP. The count is specified as the number of 4-MHz clock cycles. Refer to the *Using the Timer Module* section for details.



Figure 103. ADCRSTSTCT0: ADC Reset 0 Start Count Register (Address = 15h, Reset Value = 0000h)

D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12
0	0	0	0	0	0	0	0		ADCRSTS	TCT0[15:0]	
R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h		R/W	/-0h	
D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
					ADCRSTS	TCT0[15:0]					
					R/W	/-0h					

This register sets the start position of the ADC0 reset conversion signal.

Bits D[23:16] Must be '0'

Bits D[15:0] ADCRSTSTCT0[15:0]: ADC RESET 0 start count

The contents of this register can be used to position the start of the ADC reset conversion signal (default value after reset is 0000h). Refer to the *Using the Timer Module* section for details.

Figure 104. ADCRSTENDCT0: ADC Reset 0 End Count Register (Address = 16h, Reset Value = 0000h)

D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12
0	0	0	0	0	0	0	0		ADCRSTEN	IDCT0[15:0]	
R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h		R/W	/-0h	
D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
					ADCRSTEN	IDCT0[15:0]					
					R/W	/-0h					

LEGEND: R/W = Read/Write; -n = value after reset

This register sets the end position of the ADC0 reset conversion signal.

Bits D[23:16] Must be '0'

Bits D[15:0] ADCRSTENDCT0[15:0]: ADC RESET 0 end count

The contents of this register can be used to position the end of the ADC reset conversion signal (default value after reset is 0000h). Refer to the *Using the Timer Module* section for details.

Figure 105. ADCRSTSTCT1: ADC Reset 1 Start Count Register (Address = 17h, Reset Value = 0000h)

D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12
0	0	0	0	0	0	0	0		ADCRSTS	TCT1[15:0]	
R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h		R/W	/-0h	
D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
					ADCRSTS	TCT1[15:0]					
R/W-0h											

LEGEND: R/W = Read/Write; -n = value after reset

This register sets the start position of the ADC1 reset conversion signal.

Bits D[23:16] Must be '0'

Bits D[15:0] ADCRSTSTCT1[15:0]: ADC RESET 1 start count

The contents of this register can be used to position the start of the ADC reset conversion. Refer to the *Using the Timer Module* section for details.



Figure 106. ADCRSTENDCT1: ADC Reset 1 End Count Register (Address = 18h, Reset Value = 0000h)

D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12
0	0	0	0	0	0	0	0		ADCRSTEN	IDCT1[15:0]	
R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h		R/W	/-0h	
D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
					ADCRSTEN	IDCT1[15:0]					
					R/W	/-0h					

This register sets the end position of the ADC1 reset conversion signal.

Bits D[23:16] Must be '0'

Bits D[15:0] ADCRSTENDCT1[15:0]: ADC RESET 1 end count

The contents of this register can be used to position the end of the ADC reset conversion. Refer to the *Using the Timer Module* section for details.

Figure 107. ADCRSTSTCT2: ADC Reset 2 Start Count Register (Address = 19h, Reset Value = 0000h)

D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12	
0	0	0	0	0	0	0	0		ADCRSTS	TCT2[15:0]		
R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h		R/W-0h			
D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0	
					ADCRSTS	TCT2[15:0]						
R/W-0h												

LEGEND: R/W = Read/Write; -n = value after reset

This register sets the start position of the ADC2 reset conversion signal.

Bits D[23:16] Must be '0'

Bits D[15:0] ADCRSTSTCT2[15:0]: ADC RESET 2 start count

The contents of this register can be used to position the start of the ADC reset conversion. Refer to the *Using the Timer Module* section for details.



Figure 108. ADCRSTENDCT2: ADC Reset 2 End Count Register (Address = 1Ah, Reset Value = 0000h)

D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12		
0	0	0	0	0	0	0	0		ADCRSTENDCT2[15:0]				
R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h		R/W	/-0h			
D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0		
					ADCRSTEN	NDCT2[15:0]							
R/W-0h													

This register sets the end position of the ADC2 reset conversion signal.

Bits D[23:16] Must be '0'

Bits D[15:0] ADCRSTENDCT2[15:0]: ADC RESET 2 end count

> The contents of this register can be used to position the end of the ADC reset conversion. Refer to the *Using the Timer Module* section for details.

Figure 109. ADCRSTSTCT3: ADC Reset 3 Start Count Register (Address = 1Bh, Reset Value = 0000h)

D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12
0	0	0	0	0	0	0	0		ADCRSTS	TCT3[15:0]	
R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h		R/W	/-0h	
D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
					ADCRSTS	TCT3[15:0]					
	R/W-0h										

LEGEND: R/W = Read/Write: -n = value after reset

This register sets the start position of the ADC3 reset conversion signal.

Bits D[23:16] Must be '0'

Bits D[15:0] ADCRSTSTCT3[15:0]: ADC RESET 3 start count

> The contents of this register can be used to position the start of the ADC reset conversion. Refer to the *Using the Timer Module* section for details.

Figure 110. ADCRSTENDCT3: ADC Reset 3 End Count Register (Address = 1Ch, Reset Value = 0000h)

D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12
0	0	0	0	0	0	0	0		ADCRSTEN	IDCT3[15:0]	
R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h		R/W	/-0h	
D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
					ADCRSTEN	IDCT3[15:0]					
	R/W-0h										

LEGEND: R/W = Read/Write; -n = value after reset

This register sets the end position of the ADC3 reset conversion signal.

Bits D[23:16] Must be '0'

Bits D[15:0] ADCRSTENDCT3[15:0]: ADC RESET 3 end count

> The contents of this register can be used to position the end of the ADC reset conversion signal (default value after reset is 0000h). Refer to the *Using the Timer Module* section for details.

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Figure 111. PRPCOUNT: Pulse Repetition Period Count Register (Address = 1Dh, Reset Value = 0000h)

D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12
0	0	0	0	0	0	0	0		PRPCOL	JNT[15:0]	
R/W-0h	R/W-0h		R/W	/-0h							
D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
					PRPCOL	JNT[15:0]					
	R/W-0h										

This register sets the device pulse repetition period count.

Bits D[23:16] Must be '0'

Bits D[15:0] PRPCOUNT[15:0]: Pulse repetition period count

The contents of this register can be used to set the pulse repetition period (in number of clock cycles of the 4-MHz clock). The PRPCOUNT value must be set in the range of 800 to 64000. Values below 800 do not allow sufficient sample time for the four samples; see the *Electrical Characteristics* table.

Figure 112. CONTROL1: Control Register 1 (Address = 1Eh, Reset Value = 0000h)

D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12
0	0	0	0	0	0	0	0	0	0	0	0
R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h
D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
CL	.KALMPIN[2	2:0]	TIMEREN	NUMAV[7:0]							
	R/W-0h		R/W-0h	V-0h R/W-0h							

LEGEND: R/W = Read/Write; -n = value after reset

This register configures the clock alarm pin, timer, and number of averages.

Bits D[23:12] Must be '0'

Bits D[11:9] CLKALMPIN[2:0]: Clocks on ALM pins

Internal clocks can be brought to the PD_ALM and LED_ALM pins for monitoring.

Note that the ALMPINCLKEN register bit must be set before using this register bit. Table 5

defines the settings for the two alarm pins.

Bit D8 TIMEREN: Timer enable

0 = Timer module is disabled and all internal clocks are off (default after reset)

1 = Timer module is enabled

Bits D[7:0] NUMAV[7:0]: Number of averages

Specify an 8-bit value corresponding to the number of ADC samples to be averaged – 1.

For example, to average four ADC samples, set NUMAV[7:0] equal to 3.

The maximum number of averages is 16. Any NUMAV[7:0] setting greater than or equal to a decimal value of 15 results in the number of averages being set to 16.



Table 5. PD_ALM and LED_ALM Pin Settings

CLKALMPIN[2:0]	PD_ALM PIN SIGNAL	LED_ALM PIN SIGNAL
000	Sample LED2 pulse	Sample LED1 pulse
001	LED2 LED pulse	LED1 LED pulse
010	Sample LED2 ambient pulse	Sample LED1 ambient pulse
011	LED2 convert	LED1 convert
100	LED2 ambient convert	LED1 ambient convert
101	No output	No output
110	No output	No output
111	No output	No output

Figure 113. SPARE1: SPARE1 Register For Future Use (Address = 1Fh, Reset Value = 0000h)

D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12
0	0	0	0	0	0	0	0	0	0	0	0
R/W-0h											
D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
0	0	0	0	0	0	0	0	0	0	0	0
R/W-0h											

LEGEND: R/W = Read/Write; -n = value after reset

This register is a spare register and is reserved for future use.

Bits D[23:0] Must be '0'



Figure 114. TIAGAIN: Transimpedance Amplifier Gain Setting Register (Address = 20h, Reset Value = 0000h)

D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12
0	0	0	0	0	0	0	0	ENSEP GAIN	STAGE2 EN1	0	0
R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h
D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
0	ST	G2GAIN1[2	2:0]	CF_LED1[4:0]					R	F_LED1[2:0	[0
R/W-0h		R/W-0h		R/W-0h					R/W-0h		

This register sets the device transimpedance amplifier gain mode and feedback resistor and capacitor values.

Bits D[23:16] Must be '0'

Bit D15 **ENSEPGAIN: Enable separate gain mode**

> 0 = The R_F, C_F values and stage 2 gain settings are the same for both the LED2 and LED1 signals; the values are specified by the RF_LED2, CF_LED2, STAGE2EN2, and

STG2GAIN2 bits in the TIA AMB GAIN register (default after reset)

1 = The R_F, C_F values and stage 2 gain settings can be independently set for the LED2 and

LED1 signals. The values for LED1 are specified using the RF LED1, CF LED1,

STAGE2EN1, and STG2GAIN1 bits in the TIAGAIN register, whereas the values for LED2

are specified using the corresponding bits in the TIA AMB GAIN register.

Bit D14 STAGE2EN1: Enable Stage 2 for LED 1

0 = Stage 2 is bypassed (default after reset)

1 = Stage 2 is enabled with the gain value specified by the STG2GAIN1[2:0] bits

Bits D[13:11] Must be '0'

Bits D[10:8] STG2GAIN1[2:0]: Program Stage 2 gain for LED1

> 000 = 0 dB, or linear gain of 1 (default after 100 = 12 dB, or linear gain of 4

reset)

101 = Do not use $001 = 3.5 \, dB$, or linear gain of 1.5 110 = Do not use 010 = 6 dB, or linear gain of 2 111 = Do not use

 $011 = 9.5 \, dB$, or linear gain of 3

Bits D[7:3] CF_LED1[4:0]: Program C_F for LED1

> 00000 = 5 pF (default after reset) 00100 = 25 pF + 5 pF01000 = 50 pF + 5 pF00001 = 5 pF + 5 pF00010 = 15 pF + 5 pF10000 = 150 pF + 5 pF

Note that any combination of these C_F settings is also supported by setting multiple bits to

'1'. For example, to obtain $C_F = 100 \text{ pF}$, set D[7:3] = 01111.

Bits D[2:0] RF_LED1[2:0]: Program R_F for LED1

> $000 = 500 \text{ k}\Omega \text{ (default after reset)}$ $100 = 25 \text{ k}\Omega$ $001 = 250 \text{ k}\Omega$ $101 = 10 \text{ k}\Omega$ $010 = 100 \text{ k}\Omega$ $110 = 1 M\Omega$ 011 = 50 kO111 = None



Figure 115. TIA_AMB_GAIN: Transimpedance Amplifier and Ambient Cancellation Stage Gain Register (Address = 21h, Reset Value = 0000h)

D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12
0	0	0	0		AMBD	AC[3:0]		FLTR CNRSEL	STAGE2 EN2	0	0
R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h				R/W-0h	R/W-0h	R/W-0h	R/W-0h
D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
0	ST	G2GAIN2[2	2:0]	CF_LED2[4:0]					RF_LED2[2:0]		
R/W-0h		R/W-0h		R/W-0h						R/W-0h	

This register configures the ambient light cancellation amplifier gain, cancellation current, and filter corner frequency.

Bits D[23:20] Must be '0'

Bits D[19:16] AMBDAC[3:0]: Ambient DAC value

These bits set the value of the cancellation current.

0000 = 0 μA (default after reset)	1000 = 8 μA
$0001 = 1 \mu A$	1001 = 9 μA
$0010 = 2 \mu A$	$1010 = 10 \mu A$
$0011 = 3 \mu A$	1011 = Do not use
$0100 = 4 \mu A$	1100 = Do not use
$0101 = 5 \mu A$	1101 = Do not use
$0110 = 6 \mu A$	1110 = Do not use
$0111 = 7 \mu A$	1111 = Do not use

Bit D15 FLTRCNRSEL: Filter corner selection

0 = 500-Hz filter corner (default after reset)

1 = 1000-Hz filter corner

Bit D14 STAGE2EN2: Stage 2 enable for LED 2

0 = Stage 2 is bypassed (default after reset)

1 = Stage 2 is enabled with the gain value specified by the STG2GAIN2[2:0] bits

Bits D[13:11] Must be '0'

Bits D[10:8] STG2GAIN2[2:0]: Stage 2 gain setting for LED 2

000 = 0 dB, or linear gain of 1 (default after reset)

100 = 12 dB, or linear gain of 4 101 = Do not use

100 = 12 dB, or linear gain of 4 101 = Do not use

110 = Do not use

111 = Do not use

111 = Do not use

Bits D[7:3] CF_LED2[4:0]: Program C_F for LED2

00000 = 5 pF (default after reset) 00100 = 25 pF + 5 pF 00001 = 5 pF + 5 pF 01000 = 50 pF + 5 pF 00010 = 15 pF + 5 pF 10000 = 150 pF + 5 pF

Note that any combination of these C_F settings is also supported by setting multiple bits to '1'. For example, to obtain $C_F = 100$ pF, set D[7:3] = 01111.

Bits D[2:0] RF_LED2[2:0]: Program R_F for LED2

 $\begin{array}{lll} 000 = 500 \; k\Omega & 100 = 25 \; k\Omega \\ 001 = 250 \; k\Omega & 101 = 10 \; k\Omega \\ 010 = 100 \; k\Omega & 110 = 1 \; M\Omega \\ 011 = 50 \; k\Omega & 111 = None \end{array}$



Figure 116. LEDCNTRL: LED Control Register (Address = 22h, Reset Value = 0000h)

D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12
0	0	0	0	0	0	LED_RA	NGE[1:0]	LED1[7:0]			
R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/V	/-0h	R/W-0h			
D11	D10	D9	D8	D7	D6	D5	D4	D3 D2 D1			
	LED1[7:0]						LED2	2[7:0]			
	R/W	/-0h		R/W-0h							

This register sets the LED current range and the LED1 and LED2 drive current.

Bits D[23:18] Must be '0'

Bits D[17:16] LED_RANGE[1:0]: LED range

These bits program the full-scale LED current range for Tx. Table 6 details the settings.

Bits D[15:8] LED1[7:0]: Program LED current for LED1 signal

Use these register bits to specify the LED current setting for LED1 (default after reset is 00h).

The nominal value of the LED current is given by Equation 6,

where the full-scale LED current is either 0 mA, 50 mA, 75 mA, 100 mA, 150 mA, or 200 mA (as specified by the LED_RANGE[1:0] register bits).

Bits D[7:0] LED2[7:0]: Program LED current for LED2 signal

Use these register bits to specify the LED current setting for LED2 (default after reset is 00h).

The nominal value of LED current is given by Equation 7,

where the full-scale LED current is either 0 mA, 50 mA, 75 mA, 100 mA, 150 mA, or 200 mA (as specified by the LED_RANGE[1:0] register bits).

Table 6. Full-Scale LED Current across Tx Reference Voltage Settings⁽¹⁾

LED DANCE(4:01	0.75 V (TX_R	EF[1:0] = 00)	0.5 V (TX_RE	EF[1:0] = 01)	1.0 V (TX_REF[1:0] = 10)		
LED_RANGE[1:0]	I _{MAX}	V_{HR}	I _{MAX}	V_{HR}	I _{MAX}	V _{HR}	
00 (default after reset)	150 mA	1.4 V	100 mA	1.1 V	200 mA	1.7 V	
01	75 mA	1.3 V	50 mA	1.0 V	100 mA	1.6 V	
10	150 mA	1.4 V	100 mA	1.1 V	200 mA	1.7 V	
11	Tx is off	_	Tx is off	1	Tx is off	_	

(1) For a 3-V to 3.6-V supply, use TX_REF = 0.5 V. For a 4.75-V to 5.25-V supply, use TX_REF = 0.75 V or 1.0 V.

 $\frac{256}{256} \times \text{Full-Scale Current} \tag{6}$

× Full-Scale Current

LED1[7:0]

256 (7)



Figure 117. CONTROL2: Control Register 2 (Address = 23h, Reset Value = 0000h)

D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12
0	0	0	0	0	TX_REF1	TX_REF0	RST_ CLK_ON _PD_ ALM	EN_ADC _BYP	0	0	0
R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h
D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
TXBRG MOD	DIGOUT_ TRI STATE	XTAL DIS	EN_ SLOW_ DIAG	0	0	0	0	0	PDNTX	PDNRX	PDNAFE
R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h

This register controls the LED transmitter, crystal, and the AFE, transmitter, and receiver power modes.

Bits D[23:19] Must be '0'

Bits D[18:17] TX REF[1:0]: Tx reference voltage

These bits set the transmitter reference voltage. This Tx reference voltage is available on the device TX_REF pin.

00 = 0.75-V Tx reference voltage (default value after reset)

01 = 0.5-V Tx reference voltage 10 = 1.0-V Tx reference voltage

11 = 0.75-V Tx reference voltage

NOTE: For best results, use TX_REF = 0.5 V for 3-V operation. Use TX_REF = 0.75V and

 $TX_REF = 1.0 V \text{ for } 5-V \text{ operation.}$

Bit D16 RST CLK ON PD ALM: Reset clock onto PD ALM pin

0 = Normal mode; no reset clock signal is connected to the PD ALM pin

1 = Reset clock signal is connected to the PD ALM pin

Bit D15 EN_ADC_BYP: ADC bypass mode enable

0 = Normal mode, the internal ADC is active (default after reset)

1 = ADC bypass mode, the analog signal is output to the ADC_BYPP and ADC_BYPN pins

Bits D[14:12] Must be '0'

Bit D11 TXBRGMOD: Tx bridge mode

0 = LED driver is configured as an H-bridge (default after reset)

1 = LED driver is configured as a push-pull

Bit D10 DIGOUT_TRISTATE: Digital output 3-state mode

This bit determines the state of the device digital output pins, including the clock output pin and SPI output pins. In order to avoid loading the SPI bus when multiple devices are connected, this bit must be set to '1' (3-state mode) whenever the device SPI is inactive.

0 = Normal operation (default)

1 = 3-state mode

Bit D9 XTALDIS: Crystal disable mode

0 = The crystal module is enabled; the 8-MHz crystal must be connected to the XIN and XOUT pins

1 = The crystal module is disabled; an external 8-MHz clock must be applied to the XIN pin

Bit D8 EN_SLOW_DIAG: Fast diagnostics mode enable

0 = Fast diagnostics mode, 8 ms (default value after reset)

1 = Slow diagnostics mode, 16 ms

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Bits D[7:3] Must be '0'

Bit D2 PDN_TX: Tx power-down

0 = The Tx is powered up (default after reset)1 = Only the Tx module is powered down

Bit D1 PDN_RX: Rx power-down

0 = The Rx is powered up (default after reset)1 = Only the Rx module is powered down

Bit D0 PDN_AFE: AFE power-down

0 = The AFE is powered up (default after reset)

1 = The entire AFE is powered down (including the Tx, Rx, and diagnostics blocks)

Figure 118. SPARE2: SPARE2 Register For Future Use (Address = 24h, Reset Value = 0000h)

D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12
0	0	0	0	0	0	0	0	0	0	0	0
R/W-0h											
D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
0	0	0	0	0	0	0	0	0	0	0	0
R/W-0h											

LEGEND: R/W = Read/Write; -n = value after reset

This register is a spare register and is reserved for future use.

Bits D[23:0] Must be '0'

Figure 119. SPARE3: SPARE3 Register For Future Use (Address = 25h, Reset Value = 0000h)

D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12
0	0	0	0	0	0	0	0	0	0	0	0
R/W-0h											
D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
0	0	0	0	0	0	0	0	0	0	0	0
R/W-0h											

LEGEND: R/W = Read/Write; -n = value after reset

This register is a spare register and is reserved for future use.

Bits D[23:0] Must be '0'

Figure 120. SPARE4: SPARE4 Register For Future Use (Address = 26h, Reset Value = 0000h)

D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12
0	0	0	0	0	0	0	0	0	0	0	0
R/W-0h											
D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
0	0	0	0	0	0	0	0	0	0	0	0
R/W-0h											

LEGEND: R/W = Read/Write; -n = value after reset

This register is a spare register and is reserved for future use.

Bits D[23:0] Must be '0'

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Figure 121. RESERVED1: RESERVED1 Register For Factory Use Only (Address = 27h, Reset Value = XXXXh)

D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12
X ⁽¹⁾	Х	Х	Х	Х	Х	Х	Х	Х	Х	Х	Х
R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h
D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
Х	Х	Χ	Х	Х	Х	Х	Х	Х	Х	Х	Х
R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h

(1) X = don't care.

This register is reserved for factory use. Readback values vary between devices.

Bits D[23:0] Must be '0'

Figure 122. RESERVED2: RESERVED2 Register For Factory Use Only (Address = 28h, Reset Value = XXXXh)

D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12
X ⁽¹⁾	Χ	Χ	Х	Х	Х	Х	Х	Х	Х	Х	Х
R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h
D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
Χ	Χ	Χ	Х	Х	Х	Х	Х	Х	Х	Х	Χ
R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h

LEGEND: R = Read only; -n = value after reset

(1) X = don't care.

This register is reserved for factory use. Readback values vary between devices.

Must be '0' Bits D[23:0]

Figure 123. ALARM: Alarm Register (Address = 29h, Reset Value = 0000h)

D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12
0	0	0	0	0	0	0	0	0	0	0	0
R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h
D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
0	0	0	0	ALMPIN CLKEN	0	0	0	0	0	0	0
R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h

LEGEND: R = Read only; -n = value after reset

This register controls the Alarm pin functionality.

Bits D[23:8] Must be '0'

Bit D7 ALMPINCLKEN: Alarm pin clock enable

> 0 = Disables the monitoring of internal clocks; the PD_ALM and LED_ALM pins function as diagnostic fault alarm output pins (default after reset)

> 1 = Enables the monitoring of internal clocks; these clocks can be brought out on PD ALM and LED_ALM selectively (depending on the value of the CLKALMPIN[2:0] register bits).

Bits D[6:0] Must be '0'

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Figure 124. LED2VAL: LED2 Digital Sample Value Register (Address = 2Ah, Reset Value = 0000h)

D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12
					LED2V	AL[23:0]					
					R-	0h					
D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
					LED2V	AL[23:0]					
R-0h											

This register contains the digital value of the latest LED2 sample converted by the ADC. The ADC_RDY signal goes high each time that the contents of this register are updated. The host processor must readout this register before the next sample is converted by the AFE.

Bits D[23:0] LED2VAL[23:0]: LED2 digital value

This register contains the digital value of the latest LED2 sample converted by the ADC. The ADC_RDY signal goes high each time that the contents of this register are updated. The host processor must readout this register before the next sample is converted by the AFE.

Figure 125. ALED2VAL: Ambient LED2 Digital Sample Value Register (Address = 2Bh, Reset Value = 0000h)

_		_	_	_		_		_	_	_	_
D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12
					ALED2V	'AL[23:0]					
					R-	0h					
D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
					ALED2V	'AL[23:0]					
R-0h											

LEGEND: R = Read only; -n = value after reset

This register contains the digital value of the latest LED2 ambient sample converted by the ADC. The ADC_RDY signal goes high each time that the contents of this register are updated. The host processor must readout this register before the next sample is converted by the AFE.

Bits D[23:0] ALED2VAL[23:0]: LED2 ambient digital value

This register contains the digital value of the latest LED2 ambient sample converted by the ADC. The ADC_RDY signal goes high each time that the contents of this register are updated. The host processor must readout this register before the next sample is converted by the AFE.



Figure 126. LED1VAL: LED1 Digital Sample Value Register (Address = 2Ch, Reset Value = 0000h)

D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12
					LED1V	AL[23:0]					
					R-	0h					
D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
					LED1V	AL[23:0]					
R-0h											

This register contains the digital value of the latest LED1 sample converted by the ADC. The ADC_RDY signal goes high each time that the contents of this register are updated. The host processor must readout this register before the next sample is converted by the AFE.

Bits D[23:0] LED1VAL[23:0]: LED1 digital value

This register contains the digital value of the latest LED1 sample converted by the ADC. The ADC_RDY signal goes high each time that the contents of this register are updated. The host processor must readout this register before the next sample is converted by the AFE.

Figure 127. ALED1VAL: Ambient LED1 Digital Sample Value Register (Address = 2Dh, Reset Value = 0000h)

D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12
					ALED1V	'AL[23:0]					
					R-	0h					
D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
					ALED1V	'AL[23:0]					
R-0h											

LEGEND: R = Read only; -n = value after reset

This register contains the digital value of the latest LED1 ambient sample converted by the ADC. The ADC_RDY signal goes high each time that the contents of this register are updated. The host processor must readout this register before the next sample is converted by the AFE.

Bits D[23:0] ALED1VAL[23:0]: LED1 ambient digital value

This register contains the digital value of the latest LED1 ambient sample converted by the ADC. The ADC_RDY signal goes high each time that the contents of this register are updated. The host processor must readout this register before the next sample is converted by the AFE.



Figure 128. LED2-ALED2VAL: LED2-Ambient LED2 Digital Sample Value Register (Address = 2Eh, Reset Value = 0000h)

D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12
					LED2-ALE	D2VAL[23:0]					
					R-	0h					
D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
					LED2-ALE	D2VAL[23:0]					
R-0h											

This register contains the digital value of the LED2 sample after the LED2 ambient is subtracted. The host processor must readout this register before the next sample is converted by the AFE.

Bits D[23:0] LED2-ALED2VAL[23:0]: (LED2 – LED2 ambient) digital value

This register contains the digital value of the LED2 sample after the LED2 ambient is subtracted. The host processor must readout this register before the next sample is converted by the AFE.

Note that this value is inverted when compared to waveforms shown in many publications.

Figure 129. LED1-ALED1VAL: LED1-Ambient LED1 Digital Sample Value Register (Address = 2Fh, Reset Value = 0000h)

D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12
					LED1-ALE	01VAL[23:0]					
					R-	0h					
D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
					LED1-ALE	01VAL[23:0]					
R-0h											

LEGEND: R = Read only; -n = value after reset

This register contains the digital value of the LED1 sample after the LED1 ambient is subtracted. The host processor must readout this register before the next sample is converted by the AFE.

Bits D[23:0] LED1-ALED1VAL[23:0]: (LED1 – LED1 ambient) digital value

This register contains the digital value of the LED1 sample after the LED1 ambient is subtracted from it. The host processor must readout this register before the next sample is converted by the AFE.

Note that this value is inverted when compared to waveforms shown in many publications.



Figure 130. DIAG: Diagnostics Flag Register (Address = 30h, Reset Value = 0000h)

D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12
0	0	0	0	0	0	0	0	0	0	0	PD_ALM
R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h
D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
LED_ ALM	LED1 OPEN	LED2 OPEN	LEDSC	OUTPSH GND	OUTNSH GND	PDOC	PDSC	INNSC GND	INPSC GND	INNSC LED	INPSC LED
R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h

LEGEND: R = Read only; -n = value after reset

This register is read only. This register contains the status of all diagnostic flags at the end of the diagnostics sequence. The end of the diagnostics sequence is indicated by the signal going high on DIAG_END pin.

Bits D[23:13] Read only

Bit D12 PD_ALM: Power-down alarm status diagnostic flag

This bit indicates the status of PD_ALM (and the PD_ALM pin).

0 = No fault (default after reset)

1 = Fault present

Bit D11 LED_ALM: LED alarm status diagnostic flag

This bit indicates the status of LED_ALM (and the LED_ALM pin).

0 = No fault (default after reset)

1 = Fault present

Bit D10 LED1OPEN: LED1 open diagnostic flag

This bit indicates that LED1 is open. 0 = No fault (default after reset)

1 = Fault present

Bit D9 LED2OPEN: LED2 open diagnostic flag

This bit indicates that LED2 is open. 0 = No fault (default after reset)

1 = Fault present

Bit D8 LEDSC: LED short diagnostic flag

This bit indicates an LED short. 0 = No fault (default after reset)

1 = Fault present

Bit D7 OUTPSHGND: OUTP to GND diagnostic flag

This bit indicates that OUTP is shorted to the GND cable.

0 = No fault (default after reset)

1 = Fault present

Bit D6 OUTNSHGND: OUTN to GND diagnostic flag

This bit indicates that OUTN is shorted to the GND cable.

0 = No fault (default after reset)

1 = Fault present

Bit D5 PDOC: PD open diagnostic flag

This bit indicates that PD is open. 0 = No fault (default after reset)

1 = Fault present



Bit D4 PDSC: PD short diagnostic flag

This bit indicates a PD short. 0 = No fault (default after reset)

1 = Fault present

Bit D3 INNSCGND: INN to GND diagnostic flag

This bit indicates a short from the INN pin to the GND cable.

0 = No fault (default after reset)

1 = Fault present

Bit D2 INPSCGND: INP to GND diagnostic flag

This bit indicates a short from the INP pin to the GND cable.

0 = No fault (default after reset)

1 = Fault present

Bit D1 INNSCLED: INN to LED diagnostic flag

This bit indicates a short from the INN pin to the LED cable.

0 = No fault (default after reset)

1 = Fault present

Bit D0 INPSCLED: INP to LED diagnostic flag

This bit indicates a short from the INP pin to the LED cable.

0 = No fault (default after reset)

1 = Fault present

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9 Applications and Implementation

NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

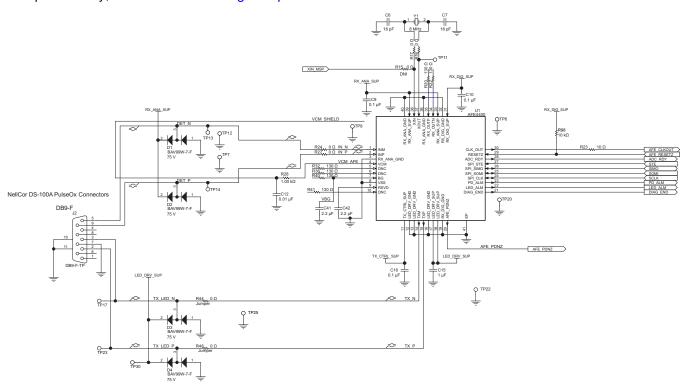
9.1 Application Information

The AFE4490 can be used for measuring SPO2 and for monitoring heart rate. The high dynamic range of the device enables measuring SPO2 with a high degree of accuracy even under low-perfusion (ac-to-dc ratio) conditions. An SPO2 measurement system involves two different wavelength LEDs—usually Red and IR. By computing the ratio of the ac to dc at the two different wavelengths, the SPO2 can be calculated. Heart rate monitoring systems can also benefit from the high dynamic range of the device, which enables capturing a high-fidelity pulsating signal even in cases where the signal strength is low.

For more information on application guidelines, refer to the AFE44x0SP02EVM User's Guide (SLAU480).

9.2 Typical Application

Device connections in a typical application are shown in Figure 131. Refer to the *AFE44x0SP02EVM User's Guide* (SLAU480) for more details. The schematic in Figure 131 is a part of the AFE44x0SP02EVM and shows a cabled application in which the LEDs and photodiode are connected to the AFE4490 through a cable. However, in an application without cables, the LEDs and photodiode can be directly connected to the TXP, TXN and INP, INN pins directly, as shown in the *Design Requirements* section.



NOTE: The following signals must be considered as two sets of differential pains and routed as adjacent signals within each pair: TXM, TXP and INM, INP.

INM and INP must be guarded with VCM_SHIELD the signal. Run the VCM_SHIELD signal to the DB9 connector and back to the device.

Figure 131. AFE44x0SPO2EVM: Connections to the AFE4490



Typical Application (continued)

9.2.1 Design Requirements

An SPO2 application usually involves a Red LED and an IR LED. These LEDs can be connected either in the common anode configuration or H-bridge configuration to the TXP, TXN pins. Figure 132 shows common anode configuration and Figure 133 shows H-bridge configuration.

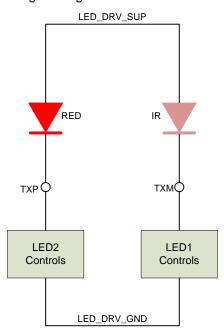


Figure 132. LEDs in Common Anode Configuration

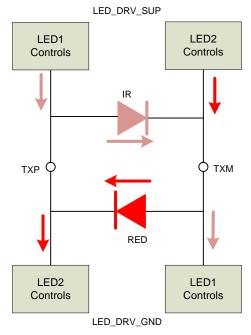


Figure 133. LEDs in H-Bridge Configuration



Typical Application (continued)

9.2.2 Detailed Design Procedure

The photodiode receives the light from both the Red and IR phases and usually has good sensitivities at both these wavelengths.

The photodiode connected in this manner operates in zero bias because of the negative feedback from the transimpedance amplifier. The connections of the photodiode to the AFE inputs are shown in Figure 134.



Figure 134. Photodiode Connection

The signal current generated by the photodiode is converted into a voltage by the transimpedance amplifier, which has a programmable transimpedance gain. The rest of the signal chain then presents a voltage to the ADC. The full-scale output of the transimpedance amplifier is ±1 V and the full-scale input to the ADC is ±1.2 V. An automatic gain control loop can be used to set the target dc voltage at the ADC input to approximately 50% of full scale. This type of AGC loop can control a combination of LED current and TIA gain to achieve this target value; see Figure 135.

TEXAS INSTRUMENTS

Typical Application (continued)

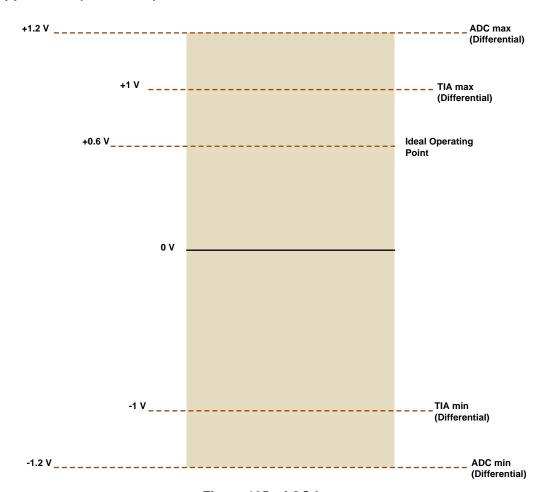


Figure 135. AGC Loop

The ADC output is a 22-bit code that is obtained by discarding the two MSBs of the 24-bit registers (for example the register with address 2Ah).

D23	D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12	
Ign	ore	22-Bit ADC Code, MSB to LSB										
R/W	/-0h	R/W-0h										
D11	D10	D9 D8 D7 D6 D5 D4 D3 D2 D1 D0										
			22-Bit ADC Code, MSB to LSB									
		R/W-0h (TBD register correct?)										

LEGEND: R/W = Read/Write; R = Read only; -n = value after reset

Table 7 shows the mapping of the input voltage to the ADC output code.

Table 7. Input Voltage Mapping

DIFFERENTIAL INPUT VOLTAGE AT ADC INPUT	22-BIT ADC OUTPUT CODE
–1.2 V	10000000000000000000000000000
(-1.2 / 2 ²¹) V	11111111111111111111
0	000000000000000000000000000000000000000
(1.2 / 2 ²¹) V	000000000000000000000000000000000000000
1.2 V	0111111111111111111



The data format is binary twos complement format, MSB first. TI recommends that the input to the ADC does not exceed ± 1 V (which is approximately 80% full-scale) because the TIA has a full-scale range of ± 1 V.

9.2.3 Application Curve

The dc component of the current from the PPG signal is referred to as *Pleth* (short for photoplethysmography) current. The input-referred noise current (referred differentially to the INP, INN inputs) as a function of the Pleth current is shown in Figure 136 at a PRF of 600 Hz and for various duty cycles of LED pulsing. For example, a duty cycle of 25% refers to a case where the LED is pulsed for 25% of the pulse repetition period and the receiver samples the photodiode current for the same period of time. The noise shown in Figure 136 is the integrated noise over a 20-Hz bandwidth from dc.

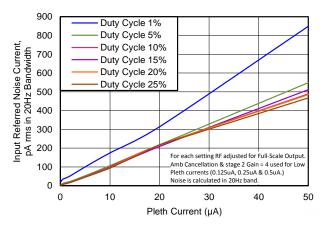


Figure 136. Input-Referred Noise Current vs Pleth Current (BW = 20Hz, PRF = 600 Hz)



10 Power-Supply Recommendations

The AFE4490 has two sets of supplies: the receiver supplies (RX_ANA_SUP, RX_DIG_SUP) and the transmitter supplies (TX_CTRL_SUP, LED_DRV_SUP). The receiver supplies can be between 2.0 V to 3.6 V whereas the transmitter supplies can be between 3.0 V to 5.25 V. Another consideration that determines the minimum allowed value of the transmitter supplies is the forward voltage of the LEDs being driven. The current source and switches inside the AFE require voltage headroom that mandates the transmitter supply to be a few hundred millivolts higher than the LED forward voltage. TX_REF is the voltage that governs the generation of the LED current from the internal reference voltage. Choosing the lowest allowed TX_REF setting reduces the additional headroom required but results in higher transmitter noise. Other than for the highest end clinical SPO2 applications, this extra noise resulting from a lower TX_REF setting might be acceptable.

The LED_DRV_SUP and TX_CTRL_SUP are recommended to be tied together to the same supply (between 3.0 V and 5.25 V). The external supply (connected to the common anode of the two LEDs) must be high enough to account for the forward drop of the LEDs as well as the voltage headroom required by the current source and switches inside the AFE. In most cases, this voltage is expected to fall below 5.25 V; thus the external supply can be the same as the LED_DRV_SUP. However, there might be cases (for instance when two LEDs are connected in series) where the voltage required on the external supply is higher than 5.25 V. Such a case must be handled with care to ensure that the voltage on the TXP and TXN pins stays less than 5.25 V and also never exceeds the supply voltage of LED_DRV_SUP, TX_CTRL_SUP by more than 0.3 V.

Many scenarios of power management are possible.

Case 1: LED forward voltage is such that a voltage of 3.3 V (for example) is acceptable on LED_DRV_SUP. In that case, a single 3.3-V supply can be used to drive all four pins (RX_ANA_SUP, RX_DIG_SUP, TX_CTRL_SUP, LED_DRV_SUP). Care must be taken to provide some isolation between the transmit and receive supplies because the LED_DRV_SUP carries the high switching current from the LEDs.

Case 2: A low-voltage supply (2.2 V for instance) is available in the system. In this case, a boost converter can be used to derive the voltage for the LED_DRV_SUP, as shown in Figure 137.

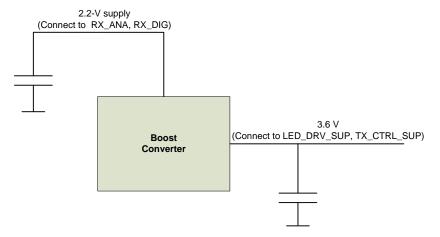


Figure 137. Boost Converter

The boost converter requires a clock (usually in the megahertz range) and there is usually a ripple at the boost converter output at this switching frequency. While this frequency is much higher than the signal frequency of interest (which is at maximum a few 10s of hertz around dc), a small fraction of this switching noise might possibly alias to the low-frequency band. Therefore, TI strongly recommends that the switching frequency of the boost converter be offset from every multiple of the PRF by at least 20 Hz, which can be ensured by choosing the appropriate PRF.



Case 3: In cases where a high voltage supply is available in the system, a buck converter or an LDO can be used to derive the voltage levels required to drive RX_ANA and RX_DIG. Such a scenario is shown in Figure 138.

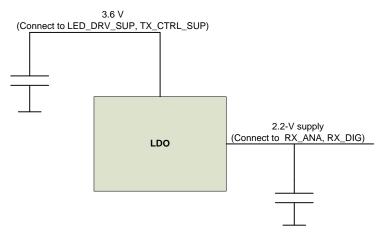


Figure 138. Buck Converter or an LDO

For more information on power-supply recommendations, see the AFE44x0SP02EVM User's Guide (SLAU480).



11 Layout

11.1 Layout Guidelines

Some key layout guidelines are:

- 1. TXP, TXN are fast switching lines and must be routed away from sensitive reference lines as well as from the INP, INN inputs.
- 2. If required to route long, TI recommends that the VCM be used as a shield for the INP, INN lines.
- 3. The device can draw high switching currents from the LED_DRV_SUP pin. Therefore, having a decoupling capacitor electrically close to the pin is recommended.

11.2 Layout Example

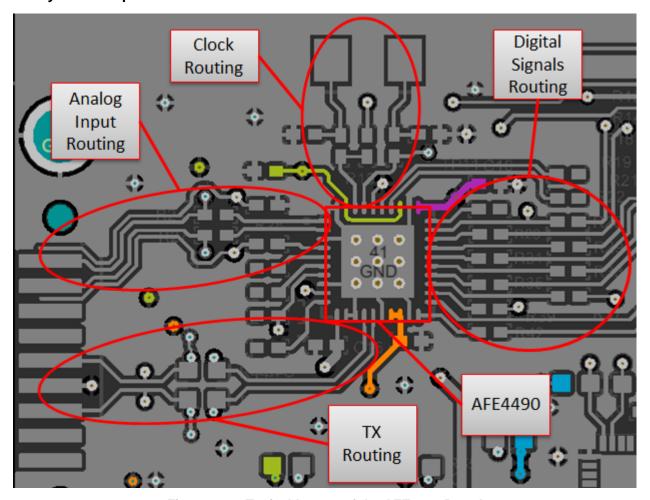


Figure 139. Typical Layout of the AFE4490 Board

Submit Documentation Feedback



12 Device and Documentation Support

12.1 Documentation Support

12.1.1 Related Documentation

AFE44x0SPO2EVM User's Guide, SLAU480

SpO Pulse Ox Wrist Oximeter Reference Design, TIDU124

12.2 Trademarks

SPI is a trademark of Motorola.

All other trademarks are the property of their respective owners.

12.3 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

12.4 Glossary

SLYZ022 — TI Glossary.

This glossary lists and explains terms, acronyms, and definitions.

13 Mechanical, Packaging, and Orderable Information

The following pages include mechanical packaging and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.



PACKAGE OPTION ADDENDUM

24-Jun-2014

PACKAGING INFORMATION

Orderable Device	Status	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead/Ball Finish	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
AFE4490RHAR	ACTIVE	VQFN	RHA	40	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	-40 to 85	AFE4490	Samples
AFE4490RHAT	ACTIVE	VQFN	RHA	40	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	-40 to 85	AFE4490	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

- (3) MSL, Peak Temp. The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.
- (4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.
- (5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.
- (6) Lead/Ball Finish Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.

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PACKAGE OPTION ADDENDUM

24-Jun-2014

n no event shall TI's liabili	tv arising out of such information	exceed the total purchase	price of the TI part(s) at issue in this document sold by	y TI to Customer on an annual basis.

PACKAGE MATERIALS INFORMATION

www.ti.com 24-Jun-2014

TAPE AND REEL INFORMATION





	Dimension designed to accommodate the component width
	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal

Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
AFE4490RHAR	VQFN	RHA	40	2500	330.0	16.4	6.3	6.3	1.5	12.0	16.0	Q2
AFE4490RHAT	VQFN	RHA	40	250	180.0	16.4	6.3	6.3	1.5	12.0	16.0	Q2

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*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
AFE4490RHAR	VQFN	RHA	40	2500	367.0	367.0	38.0
AFE4490RHAT	VQFN	RHA	40	250	210.0	185.0	35.0



- NOTES: A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M—1994.
 - B. This drawing is subject to change without notice.
 - C. QFN (Quad Flatpack No-Lead) Package configuration.
 - D. The package thermal pad must be soldered to the board for thermal and mechanical performance.
 - E. See the additional figure in the Product Data Sheet for details regarding the exposed thermal pad features and dimensions.
 - F. Package complies to JEDEC MO-220 variation VJJD-2.



RHA (S-PVQFN-N40)

PLASTIC QUAD FLATPACK NO-LEAD

THERMAL INFORMATION

This package incorporates an exposed thermal pad that is designed to be attached directly to an external heatsink. The thermal pad must be soldered directly to the printed circuit board (PCB). After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For information on the Quad Flatpack No—Lead (QFN) package and its advantages, refer to Application Report, QFN/SON PCB Attachment, Texas Instruments Literature No. SLUA271. This document is available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.



NOTES: A. All linear dimensions are in millimeters

B. The Pin 1 Identification mark is an optional feature that may be present on some devices In addition, this Pin 1 feature if present is electrically connected to the center thermal pad and therefore should be considered when routing the board layout.



RHA (S-PVQFN-N40)

PLASTIC QUAD FLATPACK NO-LEAD



NOTES:

- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- C. Publication IPC-7351 is recommended for alternate designs.
- D. This package is designed to be soldered to a thermal pad on the board. Refer to Application Note, Quad Flat—Pack Packages, Texas Instruments Literature No. SLUA271, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at www.ti.com http://www.ti.com>.
- E. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC 7525 for stencil design considerations.
- F. Customers should contact their board fabrication site for recommended solder mask tolerances and via tenting recommendations for vias placed in the thermal pad.



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