

# Tutorial 05 ACS Notes

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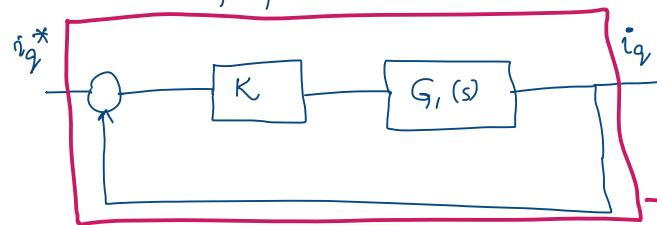
Monday, 11 May 2020 5:41 pm

## Question 0.1

Design a cascade control structure  
P+PI for the system given below

$$G_1(s) = \frac{1}{s+1} ; G_2(s) = \frac{0.1}{s+0.1}$$

In a cascade system, we design the inner loop first.



Block diagram for inner loop only

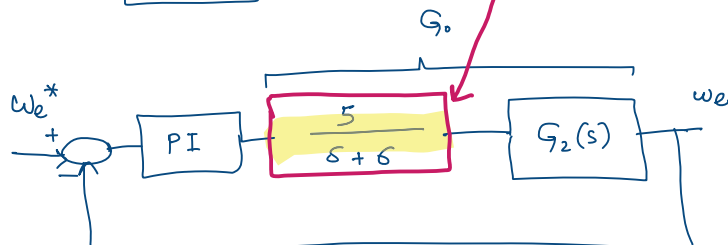
$$1 + KG = 0$$

$$\Rightarrow 1 + \frac{K}{s+1} = 0$$

$$s + 1 + K = 0$$

$$1 + K = 6$$

$$K = 5$$



$$G_o(s) = \left( \frac{5}{s+6} \right) \left( \frac{0.1}{s+0.1} \right)$$

$$G_o(s) = \frac{0.5}{(s+6)(s+0.1)}$$

Reduce to first-order

$$G_o(s) \approx \frac{0.5/6}{s+0.1}$$

$$1 + CG$$

$$1 + \left( \frac{C_1 s + C_0}{s} \right) \left( \frac{0.5/6}{s+0.1} \right) = 0$$

$$s^2 + 0.1s + \frac{0.5}{6} C_1 s + \frac{0.5}{6} C_0 = 0$$

$$s^2 + \left( 0.1 + \frac{0.5}{6} C_1 \right) s + \frac{0.5}{6} C_0 = 0$$

$$(s + \lambda)^2 = 0 \leftarrow \text{compare with}$$

$$0.1 + \frac{0.5}{6} C_1 = 2\lambda$$

$$\frac{0.5}{6} C_0 = \lambda^2$$

$$\Rightarrow C_1 = \left( \frac{2\lambda - 0.1}{0.5} \right) 6$$

$$C_1 = 12(2\lambda - 0.1)$$

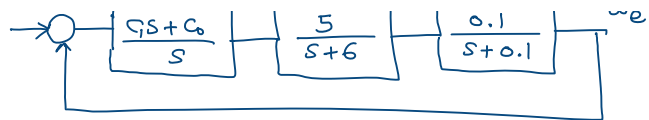
$$C_0 = 12\lambda^2$$

we haven't been given a value of " $\lambda$ "  
so, we just write the equations for now.

## Q 0.2

$w_e^*$

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$$\frac{w_e}{w_{e^*}} = \frac{Y(s)}{R(s)} \leftarrow \text{In conventional terms}$$

closed-loop polynomial for the whole system

$$\Rightarrow 1 + C(s)G_o(s)$$

$$1 + \left( \frac{0.5}{s} \right) \left( \frac{0.5}{(s+6)(s+0.1)} \right) = 0$$

$$s(s+6)(s+0.1) + 0.5c_1s + 0.5c_0 = 0$$

$$1s^3 + 6.1s^2 + (0.6 + 0.5c_1)s + 0.5c_0 = 0$$

Construct a RH-table

$s^3$	1	$0.6 + 0.5c_1$	0
$s^2$	6.1	$0.5c_0$	0
$s^1$	$\lambda$	0	0
$s^0$	$\gamma$	0	0

$$\lambda = -\frac{1}{6.1} \begin{vmatrix} 1 & 0.6 + 0.5c_1 \\ 6.1 & 0.5c_0 \end{vmatrix}$$

$$\lambda = -\frac{1}{6.1} (0.5c_0 - 6.1(0.6 + 0.5c_1))$$

$$\lambda = 0.6 + 0.5c_1 - \frac{0.5}{6.1}c_0$$

To avoid an unstable pole,  $\lambda > 0$

$$\lambda = 0.6 + 5(2\lambda - 0.1) - \frac{6\lambda^2}{6.1}$$

$$0.6 + 12\lambda - 0.6 - \frac{6\lambda^2}{6.1} > 0$$

$$\left(12 - \frac{6\lambda}{6.1}\right)\lambda > 0$$

$$\lambda > 0 ; \quad 12 - \frac{6\lambda}{6.1} > 0$$

$$12 > \frac{6\lambda}{6.1}$$

$$12.2 > \lambda$$

$$\lambda < 12.2$$

$$0 < \lambda < 12.2 \quad \text{Condition from } \lambda > 0$$

$$\gamma = -\frac{1}{\lambda} \begin{vmatrix} 6.1 & 0.5c_0 \\ \lambda & 0 \end{vmatrix}$$

$$\gamma = 0.5c_0$$

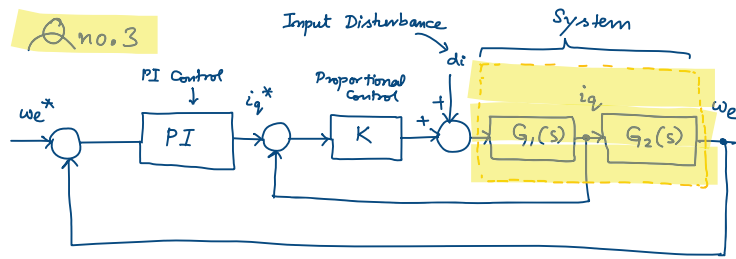
$$0.5 \times 12\lambda^2 > 0$$

$$6\lambda^2 > 0$$

$$\lambda > 0 \leftarrow \text{Condition from } \gamma > 0$$

Overall,  $\lambda$  must be between 0 & 12.2 in order for the system to be stable.

Q no. 3



$$i_q = G_1 (d_i + K(i_q^* - i_q))$$

Multiply both sides with " $G_2$ "

$$G_2 i_q = G_1 G_2 (d_i + K(i_q^* - i_q))$$

$w_e$

$$w_e = G_1 G_2 d_i + K G_1 G_2 i_q^* - K G_1 w_e$$

$$w_e [1 + K G_1] = G_1 G_2 d_i + K G_1 G_2 i_q^*$$

$$w_e = \frac{G_1 G_2}{1 + K G_1} d_i + \frac{K G_1 G_2}{1 + K G_1} i_q^*$$

$$i_q^* = C(w_e^* - w_e)$$

$$w_e = \frac{G_1 G_2}{1 + K G_1} d_i + \frac{K G_1 G_2 C}{1 + K G_1} w_e^* - \frac{K G_1 G_2 C w_e}{1 + K G_1}$$

$$w_e \left[ 1 + \frac{K G_1 G_2 C}{1 + K G_1} \right] = \frac{G_1 G_2}{1 + K G_1} d_i + \frac{K G_1 G_2 C}{1 + K G_1} w_e^*$$

$$w_e \left[ \frac{1 + K G_1 + K G_1 G_2 C}{1 + K G_2} \right] = \frac{G_1 G_2}{1 + K G_2} d_i + \frac{K G_1 G_2 C}{1 + K G_2} w_e^*$$

set all sources to zero except " $d_i$ "

$$\frac{w_e}{d_i} = \frac{G_1 G_2}{1 + K G_1 + K G_1 G_2 C}$$

$$\frac{w_e}{d_i} = \left( \frac{1}{s+1} \right) \left( \frac{0.1}{s+0.1} \right) \left( 1 + \frac{5}{s+1} + \frac{5}{s+1} \times \frac{0.1}{s+0.1} \times \frac{0.5s+0.5}{s} \right)$$

$$\frac{w_e}{d_i} = \frac{0.1s}{s(s+0.1)(s+1) + 5s(s+0.1) + 0.5s(s+0.5)}$$

$$\frac{w_e}{d_i} = \frac{0.1s}{s^3 + 6.5s^2 + s(0.6+0.5s) + 0.5s}$$

Apply final value theorem

$$\lim_{s \rightarrow 0} s w_e(s) = 0 \quad \text{due to disturbance}$$

Consider LHS

$$\lim_{s \rightarrow 0} s \frac{w_e(s)}{d_i(s)} \times d_i(s)$$

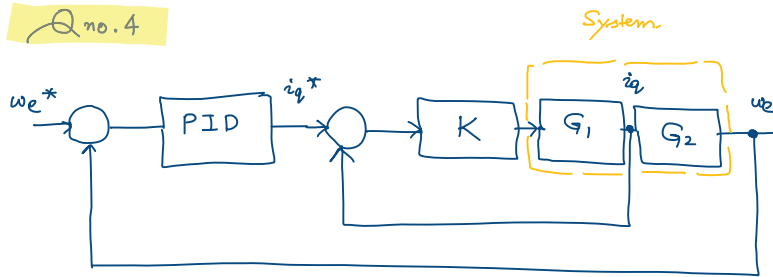
$$\lim_{s \rightarrow 0} s \left( \frac{0.1s}{s^3 + \square s^2 + \square s + \square} \right) \times \frac{1}{s}$$

Apply limit  $s=0$

$$0 \left( \frac{1}{\square} \right) = 0 \quad \text{as a unit step disturbance}$$

will be rejected successfully.

Q no. 4



$$G_1 = \frac{-s+10}{(s+10)(s+3)} \quad \bigg| \quad G_2 = \frac{0.1}{s(s+2)}$$

Consider inner loop only

$$1 + K G_1 = 0$$

$$1 - \frac{(Ks - 10K)}{(s+10)(s+3)} = 0$$

$$(s+10)(s+3) - Ks + 10K = 0$$

$$s^2 + 13s + 30 - Ks + 10K = 0$$

Find "K" such that the closed loop system has identical real poles.

$$s^2 + (13-K)s + 30 + 10K = 0$$

$$s_{1,2} = \frac{-(13-K) \pm \sqrt{(13-K)^2 - 4(30+10K)}}{2}$$

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this should be zero to give a pair of identical real poles.

So,

$$(13-K)^2 - 4(30+10K) = 0$$

$$169 + K^2 - 26K - 120 - 40K = 0$$

$$K^2 - 66K + 49 = 0$$

find "K"

$$K_1, K_2 = \frac{+66 \pm \sqrt{66^2 - 4(49)}}{2}$$

$$K_1, K_2 = \frac{66 \pm 64.5}{2}$$

$$K_1 = 65.25 \quad ; \quad K_2 = 0.75$$

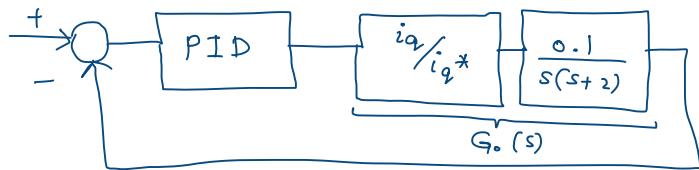
Putting  $K = K_1$  gives unstable poles so we will take  $K = K_2$

$$s_1, s_2 = \frac{K-13}{2} = \frac{0.75-13}{2} = \frac{-12.25}{2}$$

$$s_1, s_2 = -6.125$$

So, K for inner loop will be taken as 0.75

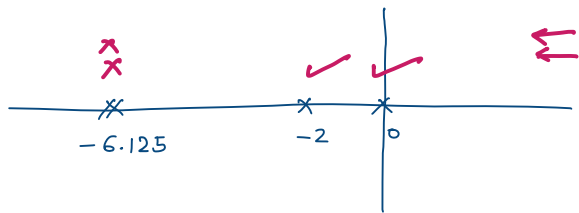
$$\frac{i_q}{i_q^*} = \frac{K G_1}{1 + K G_1} = \frac{0.75(-s+10)}{(s+6.125)^2}$$



$$G_o = \frac{0.75(-s+10)}{(s+6.125)^2} \times \frac{0.1}{s(s+2)}$$

$$G_o = \frac{0.075(-s+10)}{s(s+2)(s+6.125)(s+6.125)}$$

for PID, we want a second-order system



$$G_o = \frac{0.075(-s+10)}{s(s+2)(6.125)^2}$$

$$G_o(s) = \frac{\boxed{\frac{0.075}{(6.125)^2}} (-s+10)}{s(s+2)}$$

$$G_o(s) = \frac{\alpha(-s+10)}{s(s+2)}$$

$$s(s+2)$$

$$C(s) = \frac{c_2(s+\gamma_1)(s+\gamma_2)}{s}$$

Use pole-zero cancellation

$$C(s) = \frac{c_2(s+\gamma_1)(s+2)}{s}$$

$$1 + C G_o = 0$$

$$1 + \frac{\alpha c_2(s+\gamma_1)(-s+10)}{s^2} = 0$$

$$s^2 + \alpha c_2(s+\gamma_1)(-s+10) = 0$$

$$s^2 + \alpha c_2(-s^2 + (10-\gamma_1)s + 10\gamma_1) = 0$$

$$s^2 [1 - \alpha c_2] + \alpha c_2(10-\gamma_1)s + 10\gamma_1\alpha c_2 = 0$$

$$s^2 + \frac{\alpha c_2(10-\gamma_1)s}{1 - \alpha c_2} + \frac{10\gamma_1\alpha c_2}{1 - \alpha c_2} = 0$$

$$(s+\lambda)^2 \Rightarrow (s+1)^2 = s^2 + 2s + 1$$

$$\frac{\alpha c_2(10-\gamma_1)}{1 - \alpha c_2} = 2 \quad \left| \quad \frac{10\gamma_1\alpha c_2}{1 - \alpha c_2} = 1 \right.$$

$$10\gamma_1\alpha c_2 = 1 - \alpha c_2$$

$$\alpha c_2(10-\gamma_1) = 20\gamma_1\alpha c_2$$

$$10\alpha C_2 - r_1 \alpha C_2 = 20r_1 \alpha C_2$$

$$10\cancel{\alpha} C_2 = 21r_1 \cancel{\alpha} C_2$$

$$r_1 = \frac{10}{21}$$

$$10r_1 \alpha C_2 = 1 - \alpha C_2$$

$$C_2 [10r_1 \alpha + \alpha] = 1$$

$$C_2 = \frac{1}{\alpha [10r_1 + 1]}$$

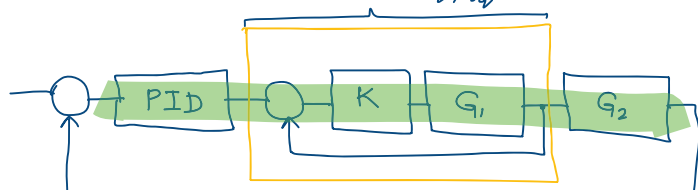
$$C_2 = \frac{1}{\alpha \left[ \frac{100}{21} + 1 \right]}$$

$$C_2 = \frac{21}{121\alpha}$$

$$C(s) = \frac{21}{121\alpha} \frac{(s + \frac{10}{21})(s+2)}{s}$$

### Nyquist plot

find forward path gain  
 $i_2 / i_2^*$



The multiplication of the green highlighted transfer functions is forward path gain

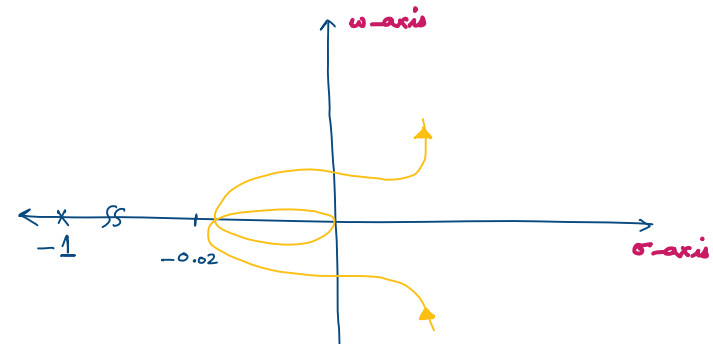
$$F(s) = \frac{21}{121\alpha} \frac{(s + \frac{10}{21})(s+2)}{s} \times \frac{0.75(-s+10)}{(s+6.125)^2} \times \frac{0.1}{s(s+2)}$$

put  $s = j\omega$  in  $F$

$$F(j\omega) = \sigma + j\omega$$

where

$\sigma$  = real part of  $F(j\omega)$  &  $\omega$  = imaginary part to  $F(j\omega)$





if the curve does not encircle  $(1,0)$  point,  
the system is stable!

Question no. 5

$$G_1(s) = \frac{1}{s+1} ; G_2 = \frac{0.1}{s+0.1}$$

PI for inner loop & PI for outer loop

$$1 + C_1 G_1 = 0$$

$$1 + \left( \frac{C_1 s + C_0}{s} \right) \left( \frac{1}{s+1} \right) = 0$$

$$s(s+1) + C_1 s + C_0 = 0$$

$$s^2 + (1+C_1)s + C_0 = 0$$

$$(s+6)^2 = s^2 + 12s + 36 \text{ place poles at } -6$$

$$1 + C_1 = 12$$

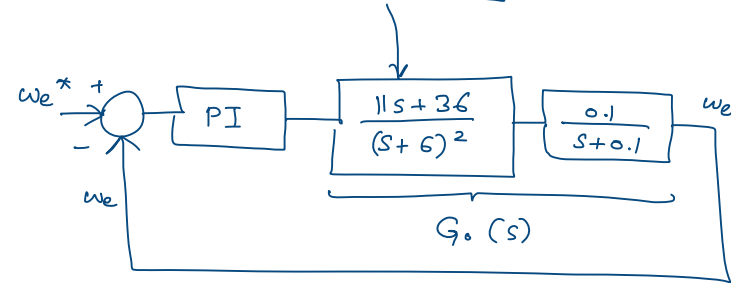
$$C_1 = 11$$

$$C_0 = 36$$

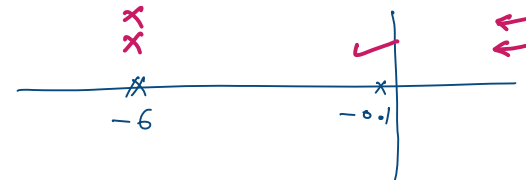
$$\underline{i_q} = \frac{C_1 G_1}{1 + C_1 G_1}$$

$$i_q^* \quad 1 + C_1 G_1$$

$$\frac{i_q}{i_q^*} = \frac{11s+36}{(s+6)^2}$$



$$G_o(s) = \frac{11s+36}{(s+6)^2} \times \frac{0.1}{s+0.1}$$



$$G_o(s) \approx \frac{(11s+36) 0.1}{6^2 (s+0.1)}$$

$$G_o(s) \approx \frac{\alpha (11s+36)}{s+0.1}$$

$$\alpha = \frac{0.1}{6^2}$$

$$C_2 = \frac{C_1 s + C_0}{s}$$

$$1 + C_2 G_o = 0$$

$$1 + \alpha (C_1 s + C_0) (11s+36)$$

$$\frac{1}{s(s+0.1)} = 0$$

$$s^2 + 0.1s + \alpha(c_1 s + c_0)(11s + 36) = 0$$

$$s^2 + 0.1s + \alpha(11c_1 s^2 + (11c_0 + 36c_1)s + 36c_0) = 0$$

$$s^2 + 0.1s + 11c_1 \alpha s^2 + \alpha(11c_0 + 36c_1)s + 36\alpha c_0 = 0$$

$$s^2 + \frac{0.1 + \alpha(11c_0 + 36c_1)}{1 + 11c_1 \alpha} s + \frac{36\alpha c_0}{1 + 11c_1 \alpha} = 0$$

place poles at  $-0.6$

$$(s + 0.6)^2 = s^2 + 1.2s - 0.036$$

Compare coeffs & solve for  $c_1$  &  $c_0$