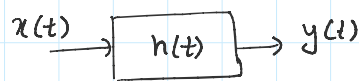


Last class:

\* LTI systems - time domain characterization using impulse response

\* Frequency analysis of LTI systems - Frequency response



$$h(t) \xleftrightarrow{FT} H(\omega)$$

$$* e^{j\omega t} \rightarrow H(\omega) e^{j\omega t}$$

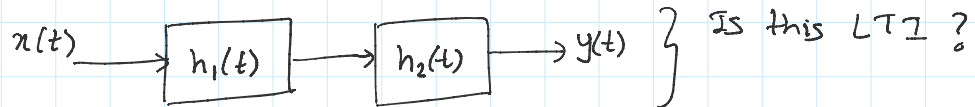
eigen functions

\* Convolution property:  $x(t) * h(t) \xleftrightarrow{FT} X(\omega) H(\omega)$

\* Use convolution property to find convolution (via FT)

Today's Class:

Cascade of LTI systems:



} Is this LTI?

Equivalent system?

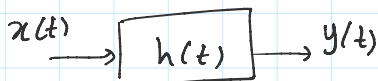


$$\begin{aligned} h(t) &= h_1(t) * h_2(t) \\ \xrightarrow{FT} H(\omega) &= H_1(\omega) H_2(\omega) \\ H(\omega) &= H_2(\omega) H_1(\omega) \\ h(t) &= h_2(t) * h_1(t) \end{aligned}$$



remark: order of cascade does not matter for LTI.

Ex.



$$x(t) = \cos(\omega_0 t)$$

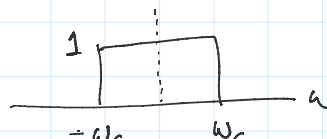
$$\omega_0, \omega_c > 0$$

$$h(t) = \frac{\sin(\omega_c t)}{\pi t}$$

Find  $y(t) = x(t) * h(t)$

$$\rightarrow x(t) \xleftrightarrow{FT} X(\omega) = \pi [\delta(\omega - \omega_0) + \delta(\omega + \omega_0)]$$

$$h(t) \xleftrightarrow{FT} H(\omega) = \begin{cases} 1, & |\omega| \leq \omega_c \\ 0, & |\omega| > \omega_c \end{cases}$$



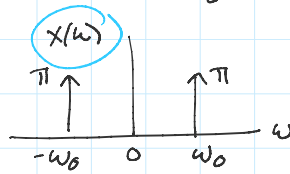
$$h(t) \xrightarrow{FT} H(\omega) = \begin{cases} 1 & |\omega| < \omega_c \\ 0 & |\omega| > \omega_c \end{cases}$$

$$Y(\omega) = X(\omega) H(\omega)$$

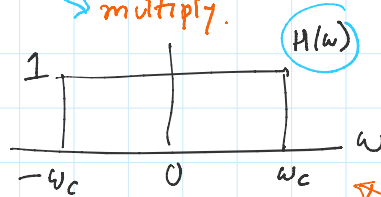
If  $\omega_0 < \omega_c$

$$Y(\omega) = \begin{cases} \pi & \omega = -\omega_0 \\ \pi & \omega = \omega_0 \end{cases} = X(\omega)$$

$$\Rightarrow y(t) = \cos(\omega_0 t)$$



multiply.



If  $\omega_0 > \omega_c$

$$Y(\omega) = 0$$

$$\Rightarrow y(t) = 0$$

$$y(t) = \begin{cases} \cos(\omega_0 t) & \omega_0 < \omega_c \\ 0 & \omega_0 > \omega_c \end{cases}$$

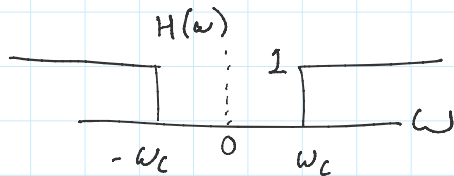
Ideal Low pass filter with cutoff  $\omega_c$

remark: new frequencies cannot be added (in the output) by an LTI system

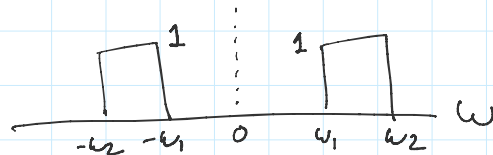
LTI systems can be represented by  $h(t) \xrightarrow{FT} H(\omega)$   
 $\hookrightarrow$  can be thought of as filter

Ideal LPF - not practical.

Ideal HPF



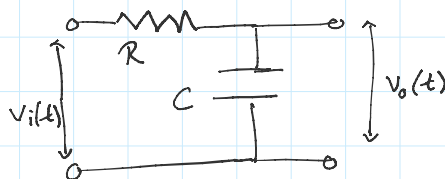
Ideal BPF



RC - low pass filter.

System.

Input  $V_i(t)$   
 output  $V_o(t)$



Low pass filter.

$$V_i(t) = V_o(t) + R \cdot C \frac{dV_o(t)}{dt}$$

Take FT  
 both sides.

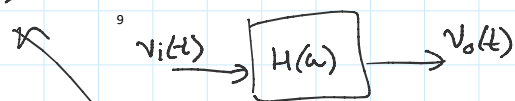
$$V_i(\omega) = V_o(\omega) + R \cdot C j\omega V_o(\omega) = (1 + j\omega RC) V_o(\omega)$$

Take FT  
on both sides.

$$V_i(\omega) = V_o(\omega) + R \cdot C j\omega V_o(\omega) = (1 + j\omega RC) V_o(\omega)$$

$$V_o(\omega) = \frac{1}{1 + j\omega RC} \cdot V_i(\omega)$$

compare with



Freq. response

$$\Rightarrow H(\omega) = \frac{1}{1 + j\omega RC}$$

$$V_o(\omega) = H(\omega) V_i(\omega)$$

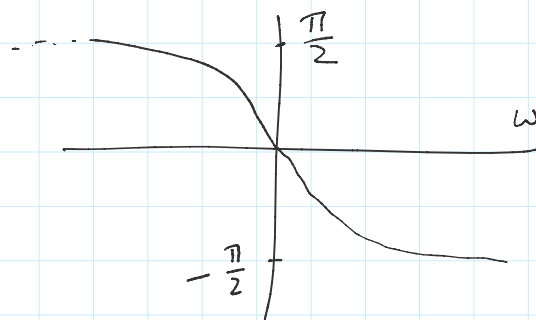
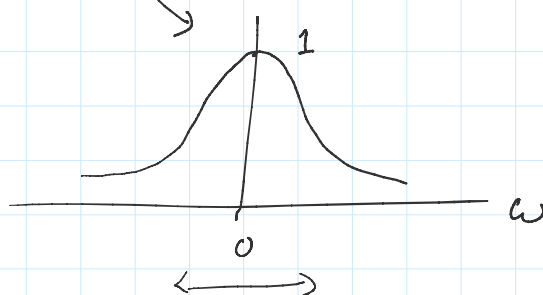
why is this low pass?

plot  $|H(\omega)|$  &  $\angle H(\omega)$

$$|H(\omega)| = \frac{1}{\sqrt{1 + \omega^2 R^2 C^2}}$$

$$\angle H(\omega) = -\tan^{-1}(\omega RC)$$

Analog non-ideal  
Low-pass filter.



what is impulse response  $h(t) \xleftrightarrow{FT} H(\omega)$  of this RC system.

$$e^{-at} u(t) \xleftrightarrow{FT} \frac{1}{a + j\omega} \Rightarrow \frac{1}{1 + j\omega RC} \xleftrightarrow{FT} ? \quad \frac{1}{RC} \exp\left(\frac{-t}{RC}\right) u(t)$$

Complex freq. response = phase change (in freq. domain) } F.T.  
= time-delay or shift in time.

HW. For above RC-LPF, with input  $x(t) = v_i(t) = \cos(\omega_0 t)$   
find the output  $y(t) = v_o(t)$  using FT.

Ex. Find Convolution:  $x(t) = \frac{\sin(\omega_0 t)}{\pi t}$  &  $h(t) = \frac{\sin(\omega_c t)}{\pi t}$

Ex. Find convolution:  $x(t) = \frac{\sin(\omega_0 t)}{\pi t}$   $\Delta$   $h(t) = \frac{\sin(\omega_c t)}{\pi t}$

find  $y(t) = x(t) * h(t)$

$\rightarrow$   $X(\omega) = \begin{cases} 1, & |\omega| \leq \omega_0 \\ 0, & |\omega| > \omega_0 \end{cases}$   $\Delta$   $H(\omega) = \begin{cases} 1, & |\omega| \leq \omega_c \\ 0, & |\omega| > \omega_c \end{cases}$

$Y(\omega) = X(\omega)H(\omega) = \begin{cases} 1, & |\omega| \leq \min(\omega_0, \omega_c) \\ 0, & |\omega| > \min(\omega_0, \omega_c) \end{cases}$

$\Rightarrow y(t) = \frac{\sin(\omega_1 t)}{\pi t}, \quad \omega_1 = \min(\omega_0, \omega_c).$

(11) Multiplication property:

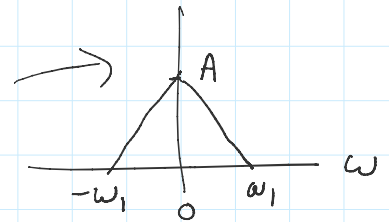
$x(t) \xleftrightarrow{FT} X(\omega) \quad y(t) \xleftrightarrow{FT} Y(\omega)$

$\Rightarrow x(t)y(t) \xleftrightarrow{FT} \frac{1}{2\pi} [X(\omega) * Y(\omega)]$

proof: HW

Ex.  $s(t) = \cos(\omega_0 t)$   $\Delta$   $p(t) \xleftrightarrow{FT} P(\omega)$

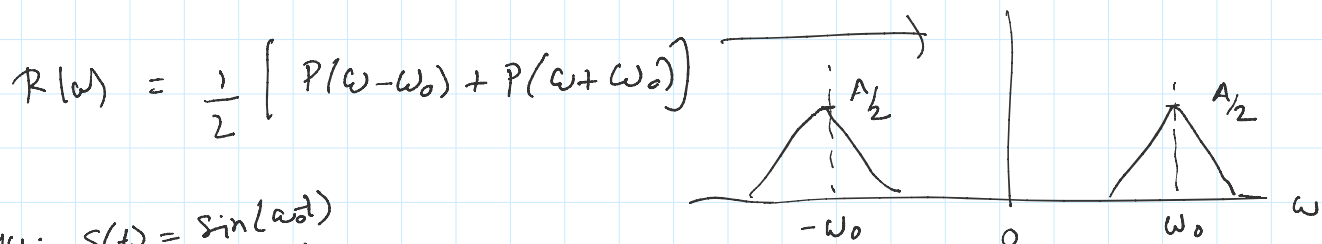
find FT of  $s(t)p(t) = \underline{r(t)}$



$\rightarrow S(\omega) = \pi [\delta(\omega - \omega_0) + \delta(\omega + \omega_0)]$

$\omega_0 > \omega_1$

$R(\omega) = \frac{1}{2\pi} [S(\omega) * P(\omega)] = \frac{1}{2\pi} \cdot \pi [\delta(\omega - \omega_0) + \delta(\omega + \omega_0)] * P(\omega)$

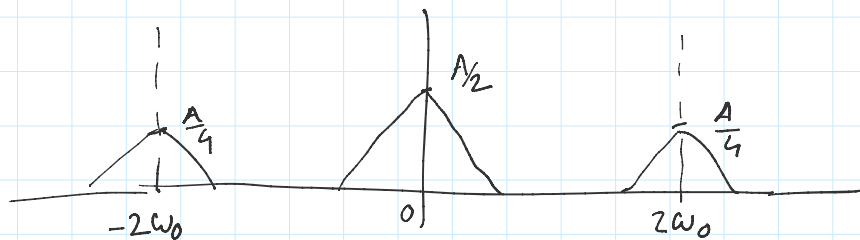


HW:  $s(t) = \sin(\omega_0 t)$   
 $s(t) = e^{j\omega_0 t}$

Ex. find FT. of  $s(t) \cdot \underline{x(t)}$ ,  $s(t) = \cos(\omega_0 t)$

$$S(\omega) \cdot X(\omega) \longleftrightarrow \frac{1}{2\pi} \pi [S(\omega - \omega_0) + S(\omega + \omega_0)] * R(\omega)$$

$$= \frac{1}{2} [R(\omega - \omega_0) + R(\omega + \omega_0)]$$



useful in CT for modulation & demodulation.

Ex. Find FT of  $x(t) = t \left( \frac{\sin t}{\pi t} \right)^2$  using FT properties

Ex. Find FT of  $\underline{\cos^2(\omega_0 t)}$  @ using multiplication property

(b) using  $\cos^2(\omega_0 t) = \frac{1 + \cos(2\omega_0 t)}{2}$