

Simulation on Thursday 13/02/25

Quiz on Friday 14/02/25 in Tut slot

Practice questions on next Friday (21/02/25)

Midsem on 22/02/25

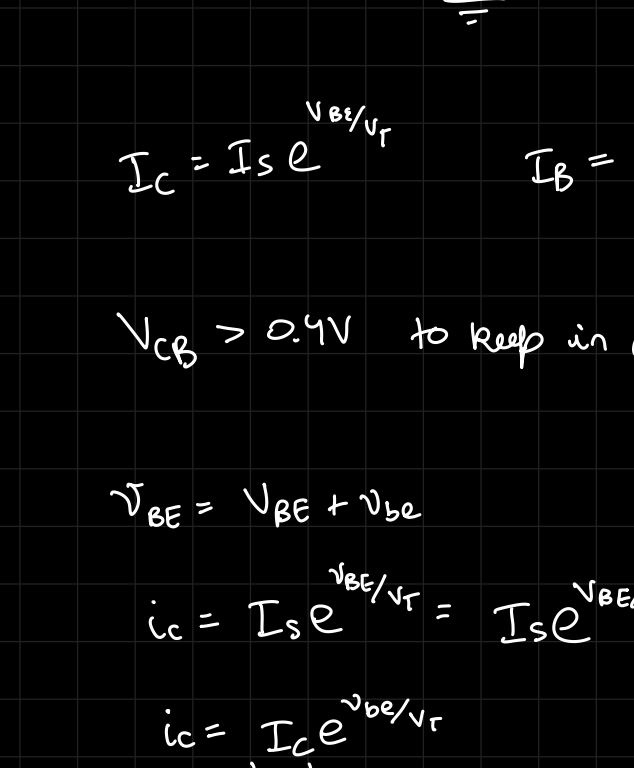
→ One A4 cheatsheet allowed

npr

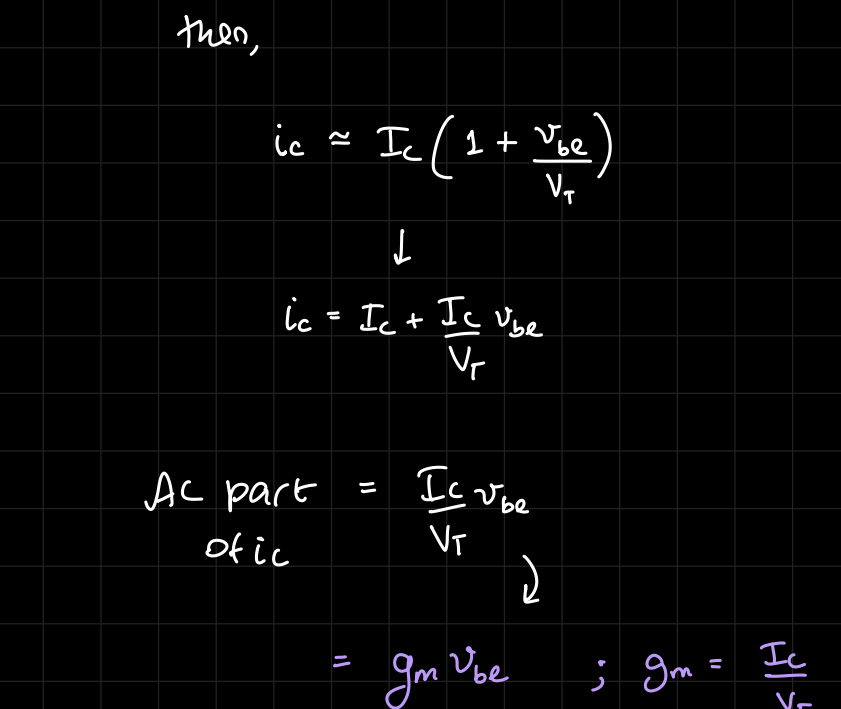
EBJ	CBJ	Mode	Application
FB	RB	Active	Amplifier
RB	RB	Cutoff	off-switch
FB	FB	Saturation	On-switch
RB	FB	Inverted	Nb! Used generally (amp but with very low gain)

BJT Small signal model

AC variation super imposed on the DC



↓ DC analysis



$$I_C = I_S e^{\frac{V_{BE}}{V_T}} \quad I_B = \frac{I_C}{\beta}$$

$V_{CE} > 0.4V$ to keep in active region

$$V_{BE} = V_{BE} + v_{be}$$

$$i_C = I_S e^{\frac{V_{BE}}{V_T}} = I_S e^{\frac{V_{BE}}{V_T}} e^{\frac{v_{be}}{V_T}}$$

$$i_C = I_C e^{\frac{v_{be}}{V_T}}$$

DC

if $v_{be} \ll V_T$ (i.e. peak of $v_{be} \ll 10mV$) then,

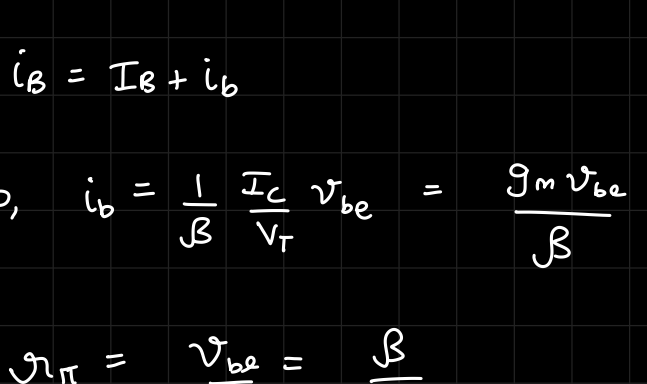
$$i_C \approx I_C \left(1 + \frac{v_{be}}{V_T} \right)$$

$$i_C = I_C + \frac{I_C}{V_T} v_{be}$$

$$\text{AC part of } i_C = \frac{I_C}{V_T} v_{be} = g_m v_{be} \quad ; \quad g_m = \frac{I_C}{V_T}$$

trans conductance

$$i_C = I_C + \frac{I_C}{V_T} v_{be}$$



① $v_{be} = 0$ ② $v_{be} \neq 0$

$$V_{BE} = V_{BE} \quad V_{BE} = V_{BE} + v_{be}$$

$$i_C = I_C = I_S e^{\frac{V_{BE}}{V_T}} \quad i_C = I_C + \frac{I_C}{V_T} v_{be}$$

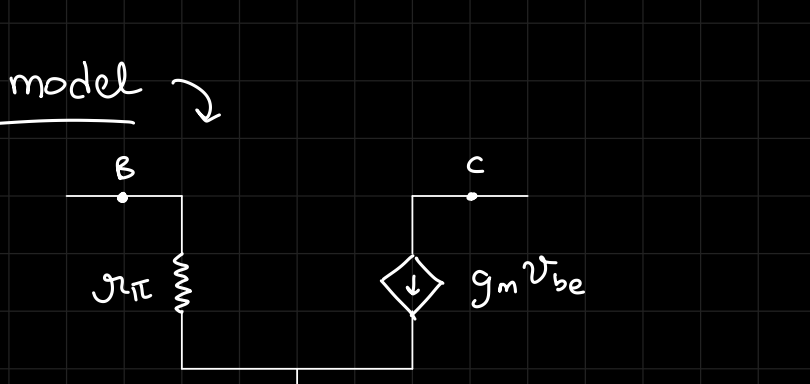
$$\text{Slope of } \bullet \rightarrow \frac{I_C + \frac{I_C}{V_T} v_{be} - I_C - \frac{I_C}{V_T} v_{be}}{V_{BE} + v_{be} - V_{BE} + v_{be}}$$

$$m = \frac{I_C}{V_T} = g_m$$

trans conductance

g_m has a direct effect on the output swings

if $g_m \uparrow$, more amplification, more power consumption, higher V_{BE} required.



$$g_m = \left. \frac{\partial i_C}{\partial V_{BE}} \right|_{i_C = I_C}$$

* To determine resistance seen by V_{BE} , we calculate the total base current i_B using:

$$i_B = \frac{i_C}{\beta} = \frac{I_C}{\beta} + \frac{1}{\beta} \frac{I_C}{V_T} v_{be}$$

$$i_B = I_B + i_b$$

$$\text{So, } i_b = \frac{1}{\beta} \frac{I_C}{V_T} v_{be} = \frac{g_m v_{be}}{\beta}$$

$$r_{\pi} = \frac{v_{be}}{i_b} = \frac{\beta}{g_m}$$

$$r_{\pi} = \frac{\beta}{g_m} = \frac{\beta V_T}{I_C} = \frac{V_T}{I_B}$$

* Emitter Current i_E

$$\alpha = \frac{i_C}{i_E} \rightarrow i_E = \frac{i_C}{\alpha}$$

$$r_E = \frac{v_{be}}{i_E}$$

$$r_E = \frac{v_{be}}{i_C} \cdot \alpha = \frac{\alpha}{g_m} = \frac{V_T}{I_E}$$

$$i_E = \frac{i_C}{\alpha} = \frac{1}{\alpha} I_C \frac{v_{be}}{V_T} = \frac{I_E}{V_T} v_{be}$$

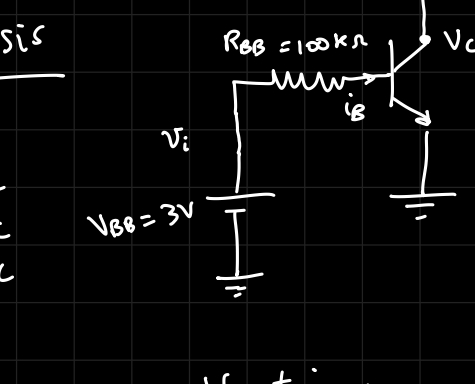
generally, $r_{\pi} \gg r_E$

$$v_{be} = i_b r_{\pi} = i_E r_E$$

$$r_{\pi} = \left(\frac{i_E}{i_C} \right) r_E \quad \left. \begin{array}{l} i_E = \alpha = \frac{\beta+1}{\beta} \\ i_C = i_b = \beta+1 \end{array} \right\}$$

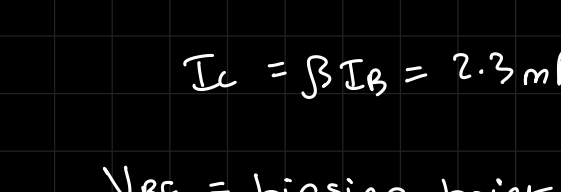
$$r_{\pi} = (\beta+1) r_E \quad \frac{i_E}{i_b} = \beta+1$$

π -model



Voltage gain:

$$v_{CE} = V_{CC} - i_C R_C$$



$$V_{CE} = V_{CC} - (I_C + i_c) R_C$$

$$v_{ce} = (V_{CC} - I_C R_C) - i_C R_C$$

$$V_{CE} = V_{CE} - \frac{i_C R_C}{DC} - \frac{i_C R_C}{AC}$$

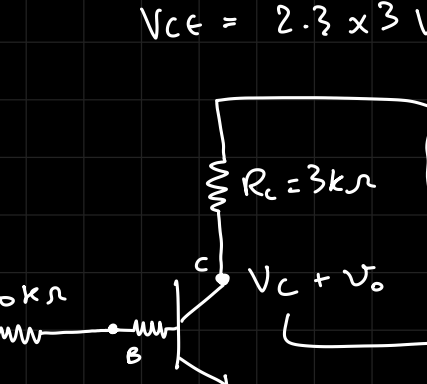
$$v_{ce} = -i_C R_C = -g_m v_{be} R_C = (-g_m R_C) v_{be}$$

$$A_v = \frac{-I_C R_C}{V_T}$$

every volt/current in the amplifier has
↳ dc component
↳ signal component

MODELS

π -MODEL



$$i_E = \frac{v_{be}}{r_{\pi}} + g_m v_{be} = \frac{v_{be}}{r_{\pi}} (1 + g_m r_{\pi}) = \frac{v_{be}}{r_{\pi}} (1 + \beta)$$

$$= v_{be} \left(\frac{r_{\pi}}{1 + \beta} \right) = \frac{v_{be}}{r_E}$$

$$i_E = \frac{v_{be}}{r_E}$$

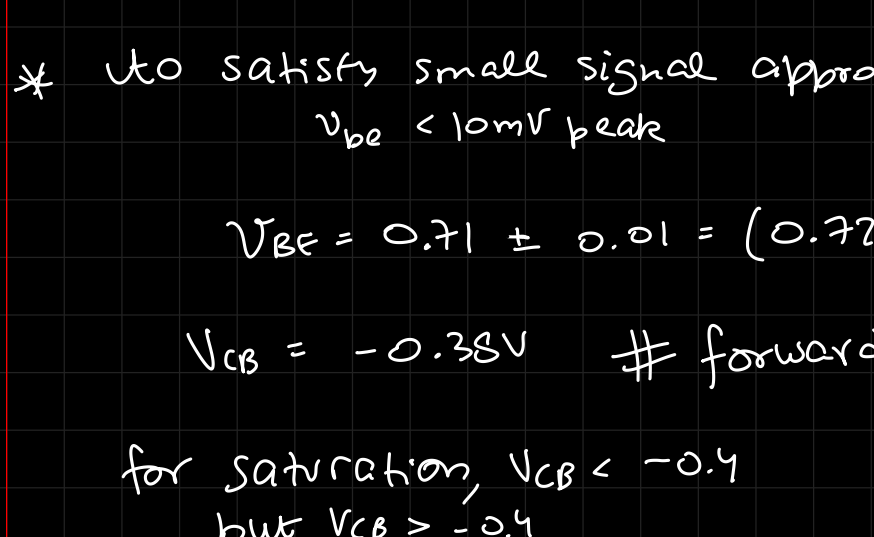
Voltage controlled current source

hybrid- π model

↳ used to carry out small signal analysis of any transistor circuit

Sometimes, the T model is more convenient.

eg1) bias points?



DC analysis

$V_{BE} \rightarrow DC$
 $A_{ac} \rightarrow AC$
 $L \rightarrow DC$
 $C \rightarrow AC$

$$-V_{BE} + i_B \cdot 100k + V_{BE} = 0$$

$$-10 + V_{CE} + I_C \cdot 3k = 0$$

$$V_{CE} + I_C \cdot 3k = 10$$

$$I_B = \frac{3 - 0.7}{100} \text{ mA}$$

$$I_C = \beta I_B = 2.3 \text{ mA}$$

V_{BE} = biasing point

$$g_m = \frac{I_C}{V_T} = \frac{2.3 \text{ mA}}{25 \text{ mV}} = 92 \text{ mA/V}$$

$$r_{\pi} = \frac{\beta}{g_m} = 1.09 \text{ k}\Omega$$

$$V_C = V_{CC} - I_C R_C = 10 - 2.3 \times 3$$

$$V_C = 3.1V$$

π -model

DC sources in AC model

AC equivalent circuit

$$-V_i + i_B \cdot 100k + V_{BE} = 0$$

$$i_B \cdot 100k + V_{BE} = V_i$$

$$i_C \cdot 3k + V_{CE} = 0$$

$$V_{CE} = -i_C \cdot 3k$$

$$V_{CE} = 2.3 \times 3 \text{ V}$$

π model

$$V_o = -i_C R_C = -g_m v_{be} R_C$$

$$v_{be} = \frac{V_i}{r_{\pi} + R_{BB}} = \frac{V_i}{1.09 + 10.1} = 0.011 V_i$$

$$V_o = -92 \cdot 0.011 V_i \cdot 3 = -3.04 V_i$$

$$\text{gain: } A_v = \frac{V_o}{V_i} = -3.04 \text{ V/V}$$

V_{CE} should be $> -0.4V$

$$V_{be} \approx 0.011 V_i$$

$$V_i = \frac{V_{be}}{0.011} = \frac{10}{0.011} = 0.91V$$

$$V_{CE} = V_{CE} + |V_o|$$

$$= 3.1 \pm 2.77 = (5.83, 0.33)$$

* To satisfy small signal approximation, $v_{be} < 10mV$ peak

$$V_{BE} = 0.71 \pm 0.01 = (0.72, 0.7)$$

$$V_{CB} = -0.38V \quad \# \text{ forward Bias}$$

for saturation, $V_{CB} < -0.4$

but $V_{CB} > -0.4$

So, amp is in active mode